5-Pin SOT23-5 Package

Inverts Input Supply Voltage

95% Voltage Conversion Efficiency

60µA Quiescent Current (MAX828)

+1.5V to +5.5V Input Voltage Range

Requires Only Two Capacitors

25mA Output Current

# **General Description**

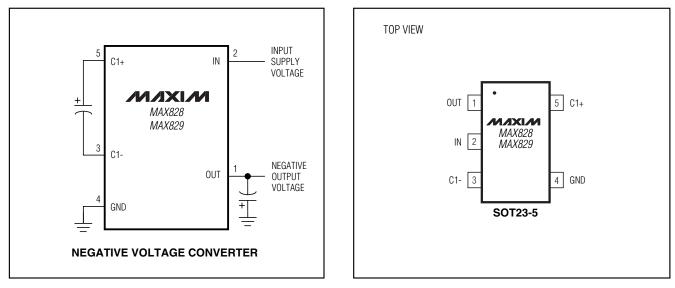
The ultra-small MAX828/MAX829 monolithic, CMOS charge-pump inverters accept input voltages ranging from +1.5V to +5.5V. The MAX828 operates at 12kHz, and the MAX829 operates at 35kHz. Their high efficiency (greater than 90% over most of the load-current range) and low operating current (60µA for the MAX828) make these devices ideal for both battery-powered and board-level voltage-conversion applications.

The MAX828/MAX829 combine low quiescent current and high efficiency. Oscillator control circuitry and four power MOSFET switches are included on-chip. Applications include generating a -5V supply from a +5V logic supply to power analog circuitry. Both parts come in a 5-pin SOT23-5 package and can deliver 25mA with a voltage drop of 500mV.

For a similar device with logic-controlled shutdown, refer to the MAX1719/MAX1720/MAX1721. For applications requiring more power, the MAX860 delivers up to 50mA with a voltage drop of 600mV, in a space-saving  $\mu$ MAX package.

# Applications

Small LCD Panels Cell Phones Medical Instruments Handy-Terminals, PDAs Battery-Operated Equipment



# **Typical Operating Circuit**

# 

Maxim Integrated Products 1

Features MAX828/MAX829

# **Ordering Information**

**Pin Configuration** 

PART	TEMP. RANGE	PIN- PACKAGE	SOT TOP MARK
MAX828EUK	-40°C to +85°C	5 SOT23-5	AABI
MAX829EUK	-40°C to +85°C	5 SOT23-5	AABJ

For price, delivery, and to place orders, please contact Maxim Distribution at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

# **ABSOLUTE MAXIMUM RATINGS**

IN to GND	+6.0V, -0.3V
OUT to GND	6.0V, +0.3V
OUT Output Current	50mA
OUT Short-Circuit to GND	Indefinite
Continuous Power Dissipation ( $T_A = +70^{\circ}C$ )	
SOT23-5 (derate 7.1mW/°C above +70°C)	571mW

Operating Temperature Range	
MAX828EUK/MAX829EUK	40°C to +85°C
Storage Temperature Range	65°C to +160°C
Lead Temperature (soldering, 10s).	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

# **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = +5V, C1 = C2 = 10\mu F (MAX828), C1 = C2 = 3.3\mu F (MAX829), T_A = 0^{\circ}C$  to  $+85^{\circ}C$ , unless otherwise noted. Typical values are at T\_A =  $+25^{\circ}C$ .)

PARAMETER	CONE	CONDITIONS		TYP	MAX	UNITS
Supply Current	TA = +25°C	MAX828		60	90	- μΑ
	IA = +25 C	MAX829		150	260	
Minimum Supply Voltage	$R_{LOAD} = 10k\Omega$	$T_A = +25^{\circ}C$	1.25	1.0		- V
	$\Pi_{\text{LOAD}} = 10 \text{ksz}$	$T_A = 0^{\circ}C \text{ to } + 85^{\circ}C$	1.5			
Maximum Supply Voltage	$R_{LOAD} = 10k\Omega$	$R_{LOAD} = 10k\Omega$			5.5	V
Oscillator Frequency	TA = +25°C	MAX828	8.4	12	15.6	- kHz
	IA = +25 C	MAX829	24.5	35	45.5	
Power Efficiency	$R_{LOAD} = 1k\Omega, T_A = +25^{\circ}$	$R_{LOAD} = 1k\Omega$ , $T_A = +25^{\circ}C$		94		%
Voltage Conversion Efficiency	RLOAD = ∞		95	99.9		%
Output Resistance	$I_{OUT} = 5mA$	$T_{A} = +25^{\circ}C$ 20	20	50	Ω	
		$T_A = 0^{\circ}C \text{ to } + 85^{\circ}C$			65	

Note 1: Capacitor contribution is approximately 20% of the output impedance [ESR + 1 / (pump frequency x capacitance)].

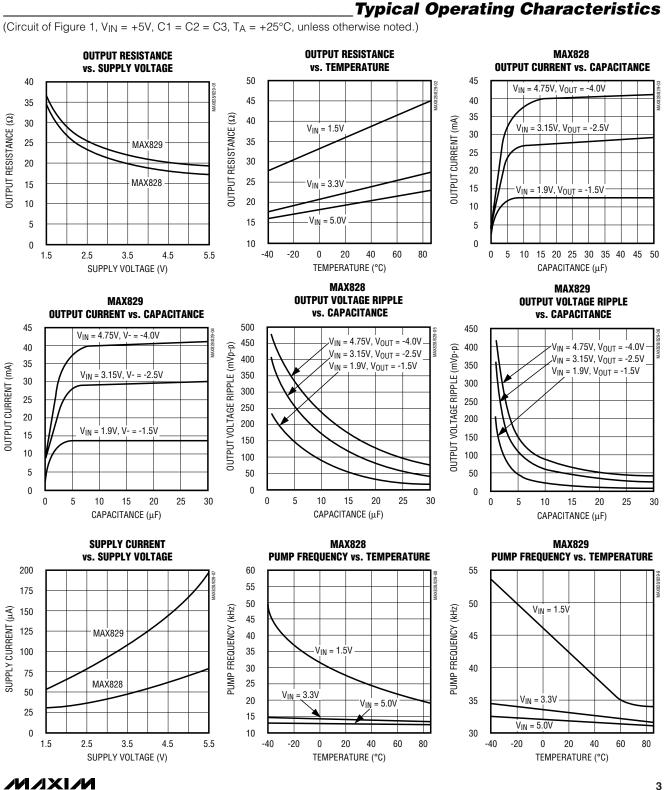
# **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = +5V, C1 = C2 = 10\mu F (MAX828), C1 = C2 = 3.3\mu F (MAX829), T_A = -40^{\circ}C$  to  $+85^{\circ}C$ , unless otherwise noted. Typical values are at T\_A =  $+25^{\circ}C$ .) (Note 2)

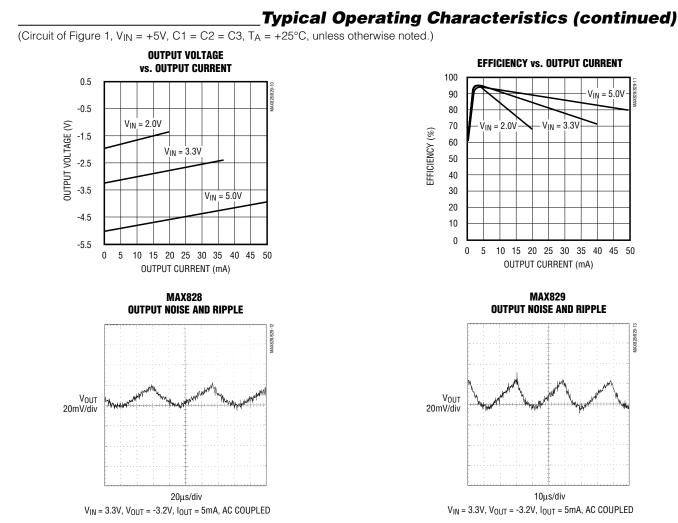
PARAMETER	CONDITIONS	MIN	TYP MAX	UNITS	
Supply Current	MAX828		115		
	MAX829		325	μΑ	
Supply Voltage Range	$R_{LOAD} = 10k\Omega$	1.5	5.5	V	
Oscillator Frequency	MAX828	6	20	- kHz	
	MAX829	19	54.3		
Output Resistance	I <sub>OUT</sub> = 5mA		65	Ω	

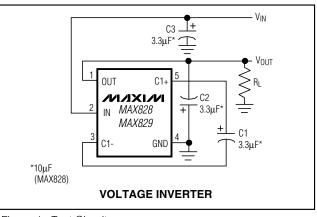
Note 2: All -40°C to +85°C specifications above are guaranteed by design.

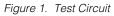




# **MAX828/MAX829**









MAX828/MAX829

# **Pin Description**

PIN	NAME	FUNCTION	
1	OUT	Inverting Charge-Pump Output	
2	IN	Positive Power-Supply Input	
3	C1-	Flying Capacitor's Negative Terminal	
4	GND	Ground	
5	C1+	Flying Capacitor's Positive Terminal	

# **Detailed Description**

The MAX828/MAX829 capacitive charge pumps invert the voltage applied to their input. For highest performance, use low equivalent series resistance (ESR) capacitors.

During the first half-cycle, switches S2 and S4 open, switches S1 and S3 close, and capacitor C1 charges to the voltage at IN (Figure 2). During the second half-cycle, S1 and S3 open, S2 and S4 close, and C1 is level shifted downward by  $V_{IN}$  volts. This connects C1 in parallel with the reservoir capacitor C2. If the voltage across C2 is smaller than the voltage across C1, then charge flows from C1 to C2 until the voltage across C2 reaches -  $V_{IN}$ . The actual voltage at the output is more positive than - $V_{IN}$ , since switches S1–S4 have resistance and the load drains charge from C2.

# **Charge-Pump Output**

The MAX828/MAX829 are not voltage regulators: the charge pump's output source resistance is approximately  $20\Omega$  at room temperature (with VIN = +5V), and VOUT approaches -5V when lightly loaded. VOUT will droop toward GND as load current increases. The droop of the negative supply (VDROOP-) equals the current draw from OUT (IOUT) times the negative converter's source resistance (RS-):

VDROOP- = IOUT x RS-

The negative output voltage will be:

VOUT = -(VIN - VDROOP-)

# **Efficiency Considerations**

The efficiency of the MAX828/MAX829 is dominated by its quiescent supply current (I<sub>Q</sub>) at low output current and by its output impedance ( $R_{OUT}$ ) at higher output current; it is given by:

$$\eta \cong \frac{I_{OUT}}{I_{OUT} + I_Q} \left( 1 - \frac{I_{OUT} \times R_{OUT}}{V_{IN}} \right)$$

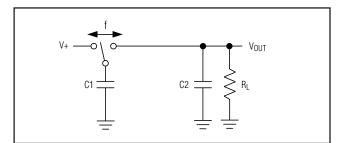


Figure 3a. Switched-Capacitor Model



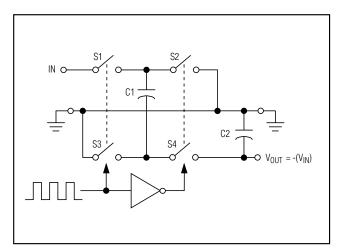


Figure 2. Ideal Voltage Inverter

where the output impedance is roughly approximated by:

$$R_{OUT} \cong \frac{1}{(f_{OSC}) \times C1} + 2R_{SW} + 4ESR_{C1} + ESR_{C2}$$

The first term is the effective resistance of an ideal switched-capacitor circuit (Figures 3a and 3b), and R<sub>SW</sub> is the sum of the charge pump's internal switch resistances (typically  $8\Omega$  to  $9\Omega$  at V<sub>IN</sub> = +5V). The typical output impedance is more accurately determined from the *Typical Operating Characteristics*.

# **Applications Information**

# **Capacitor Selection**

To maintain the lowest output resistance, use capacitors with low ESR (Table 1). The charge-pump output resistance is a function of C1's and C2's ESR. Therefore, minimizing the charge-pump capacitor's ESR minimizes the total output resistance.

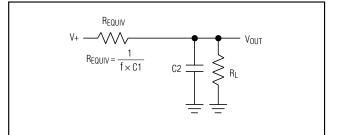


Figure 3b. Equivalent Circuit

**MAX828/MAX829** 

# Flying Capacitor (C1)

Increasing the flying capacitor's size reduces the output resistance. Small C1 values increase the output resistance. Above a certain point, increasing C1's capacitance has a negligible effect, because the output resistance becomes dominated by the internal switch resistance and capacitor ESR.

### **Output Capacitor (C2)**

Increasing the output capacitor's size reduces the output ripple voltage. Decreasing its ESR reduces both output resistance and ripple. Smaller capacitance values can be used with light loads if higher output ripple can be tolerated. Use the following equation to calculate the peak-to-peak ripple:

$$V_{\text{RIPPLE}} = \frac{I_{\text{OUT}}}{f_{\text{OSC}} \times C2} + 2 \times I_{\text{OUT}} \times \text{ESR}_{C2}$$

### Input Bypass Capacitor

Bypass the incoming supply to reduce its AC impedance and the impact of the MAX828/MAX829's switching noise. The recommended bypassing depends on the circuit configuration and on where the load is connected.

When the inverter is loaded from OUT to GND, current from the supply switches between 2 x IOUT and zero. Therefore, use a large bypass capacitor (e.g., equal to the value of C1) if the supply has a high AC impedance.

When the inverter is loaded from IN to OUT, the circuit draws  $2 \times I_{OUT}$  constantly, except for short switching spikes. A 0.1µF bypass capacitor is sufficient.

### **Voltage Inverter**

The most common application for these devices is a charge-pump voltage inverter (Figure 1). This application requires only two external components—capacitors C1 and C2—plus a bypass capacitor, if necessary. Refer to

the *Capacitor Selection* section for suggested capacitor types and values.

### **Cascading Devices**

Two devices can be cascaded to produce an even larger negative voltage (Figure 4). The unloaded output voltage is normally -2 x VIN, but this is reduced slightly by the output resistance of the first device multiplied by the quiescent current of the second. When cascading more than two devices, the output resistance rises dramatically. For applications requiring larger negative voltages, see the MAX864 and MAX865 data sheets.

### **Paralleling Devices**

Paralleling multiple MAX828s or MAX829s reduces the output resistance. Each device requires its own pump capacitor (C1), but the reservoir capacitor (C2) serves all devices (Figure 5). Increase C2's value by a factor of n, where n is the number of parallel devices. The equation for calculating output resistance is also shown in Figure 5.

# **Combined Doubler/Inverter**

In the circuit of Figure 6, capacitors C1 and C2 form the inverter, while C3 and C4 form the doubler. C1 and C3 are the pump capacitors; C2 and C4 are the reservoir capacitors. Because both the inverter and doubler use part of the charge-pump circuit, loading either output causes both outputs to decline toward GND. Make sure the sum of the currents drawn from the two outputs does not exceed 40mA.

# Table 1. Low-ESR Capacitor Manufacturers

MANUFACTURER		PHONE	FAX	DEVICE TYPE	
AVX		(803) 946-0690 (800) 282-4975	(803) 626-3123	Surface-mount, TPS series	
Matsuo		(714) 969-2491	(714) 960-6492	Surface-mount, 267 series	
Sanya	USA	(619) 661-6835	(619) 661-1055	Through hole, OS CON aprice	
Sanyo	Japan	81-7-2070-6306	81-7-2070-1174	Through-hole, OS-CON series	
Sprague	·	(603) 224-1961	(603) 224-1430	Surface-mount, 595D series	



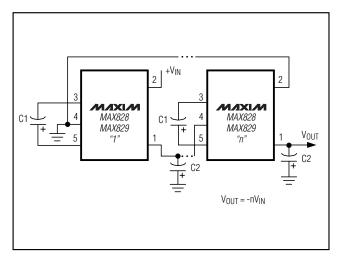


Figure 4. Cascading MAX828s or MAX829s to Increase Output Voltage

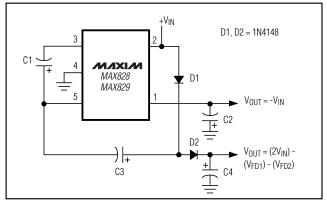


Figure 6. Combined Doubler and Inverter

# **Heavy Output Current Loads**

When under heavy loads, where higher supply is sourcing current into OUT, the OUT supply must not be pulled above ground. Applications that sink heavy current into OUT require a Schottky diode (1N5817) between GND and OUT, with the anode connected to OUT (Figure 7).

# Layout and Grounding

Good layout is important, primarily for good noise performance. To ensure good layout, mount all components as close together as possible, keep traces short to minimize parasitic inductance and capacitance, and use a ground plane.

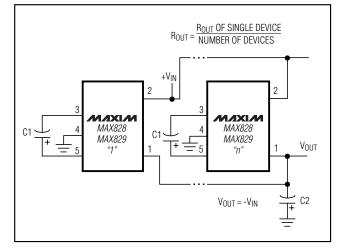


Figure 5. Paralleling MAX828s or MAX829s to Reduce Output Resistance

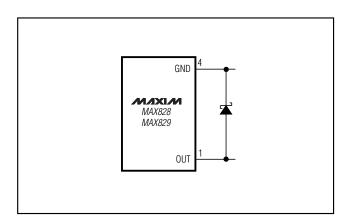


Figure 7. High V-Load Current

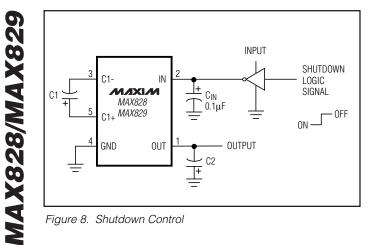
### Shutting Down the MAX828/MAX829

For a similar device with logic-controlled shutdown, please refer to the MAX1719/MAX1720/MAX1721. To add manual shutdown control to the MAX828/MAX829, use the circuit in Figure 8. The output resistance of the MAX828/MAX829 will typically be  $20\Omega$  plus two times the output resistance of the buffer driving IN. The 0.1µF capacitor at the IN pin absorbs the transient input currents of the MAX828/MAX829.

The output resistance of the buffer driving the IN pin can be reduced by connecting multiple buffers in parallel. The polarity of the SHUTDOWN signal can also be changed by using a noninverting buffer to drive IN.

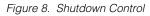


**MAX828/MAX829** 



# **Chip Information**

**TRANSISTOR COUNT: 58** SUBSTRATE CONNECTED TO IN



### OT5L 0.20 SYMBOL MIN MAX "∀" 0.90 1.45 Α 0.00 A1 0.15 DATUM b 0.90 1.30 A2 b 0.35 0.50 С 0.08 0.20 D 2.80 3.00 2.60 1.50 3.00 1.75 Ε E1 0.35 0.55 L q 0.95 REF q-E е 1.90 REF e1 0° α ρĺ -C D NDTE: ALL DIMENSIONS ARE IN MILLIMETERS. FODT LENGTH MEASURED AT INTERCEPT POINT BETWEEN DATUM A & LEAD SURFACE. PACKAGE DUTLINE EXCLUSIVE OF MOLD FLASH & METAL q Â 3. BURR 4. PACKAGE DUTLINE INCLUSIVE DF SOLDER PLATING. A1 PACKAGE DUTLINE, SDT23, B 1/1 21-0057

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