# Mono/Stereo High-Power Class D Amplifier

#### **General Description**

The MAX9736A/B/D Class D amplifiers provide highperformance, thermally efficient amplifier solutions. The MAX9736A delivers 2 x 15W into 8 $\Omega$  loads, or 1 x 30W into a 4 $\Omega$  load. The MAX9736B/D deliver 2 x 6W into 8 $\Omega$ loads or 1 x 12W into a 4 $\Omega$  load.

These devices operate from 8V to 28V and provide a high PSRR, eliminating the need for a regulated power supply. The MAX9736 offers up to 88% efficiency at 12V supply.

Pin-selectable modulation schemes select between filterless modulation and classic PWM modulation. Filterless modulation allows the MAX9736 to pass CE EMI limits with 1m cables using only a low-cost ferrite bead and capacitor on each output. Classic PWM modulation is optimized for best audio performance when using a full LC filter.

A pin-selectable stereo/mono mode allows stereo operation into  $8\Omega$  loads or mono operation into  $4\Omega$  loads. In mono mode, the right input op amp becomes available as a spare device, allowing flexibility in system design.

Comprehensive click-and-pop reduction circuitry minimizes noise coming into and out of shutdown or mute.

Input op amps allow the user to create summing amplifiers, lowpass or highpass filters, and select an optimal gain.

The MAX9736A/B/D are available in 32-pin TQFN packages and specified over the -40 $^{\circ}$ C to +85 $^{\circ}$ C temperature range.

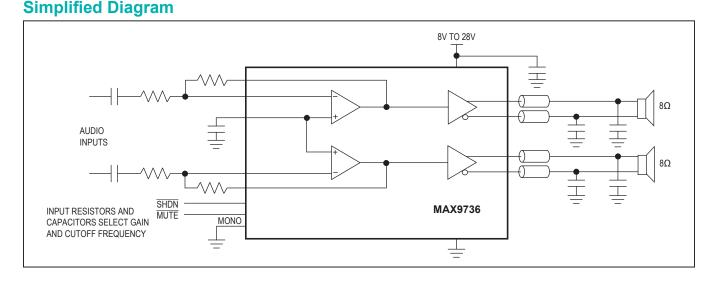
#### **Features**

- Independently Controlled Dual Paths
- Wide 8V to 28V Supply Voltage Range
- Spread-Spectrum Modulation Enables Low EMI Solution
- Passes CE EMI Limits with Low-Cost Ferrite Bead/Capacitor Filter
- Low BOM Cost
- High 67dB PSRR at 1kHz Reduces Supply Cost
- 88% Efficiency Eliminates Heatsink
- Thermal and Output Current Protection
- < 1µA Shutdown Mode
- Mute Function
- Space-Saving, 32-Pin TQFN Packages, 7mm x 7mm x 0.8mm (MAX9736A/B), 5mm x 5mm x 0.8mm (MAX9736D)

#### **Applications**

- LCD/PDP/CRT Monitors
- LCD/PDP/CRT TVs
- MP3 Docking Stations
- Notebook PCs
- PC Speakers
- All-in-One PCs

Ordering Information appears at end of data sheet.



# Mono/Stereo High-Power Class D Amplifier

#### **Absolute Maximum Ratings**

PVDD to PGND	0.3V to +30V
AGND to PGND	0.3V to +0.3V
INL, INR, FBL, FBR, COM to AGND	0.3V to (V <sub>RFG</sub> + 0.3V)
MUTE, SHDN, MONO, MOD, REGEN	to AGND0.3V to +6V
REG to AGND	0.3V to (VS + 0.3V)
VS to AGND (Note 1)	0.3V to +6V
OUTL+, OUTL-, OUTR+,	
OUTR-, to PGND	0.3V to (PVDD + 0.3V)
C1N to PGND	0.3V to (PVDD + 0.3V)
C1P to PGND(PVDD	- 0.3V) to (V <sub>BOOT</sub> + 0.3V)
BOOT to PGND(VC	
OUTL+, OUTL-, OUTR+, OUTR-,	
Short Circuit to PGND or PVDD	Continuous
Thermal Limits (Notes 2, 3)	

Continuous Power Dissipation (T <sub>A</sub> = +70°C) 32-Pin TQFN 5mm x 5mm Multiple Layer PCB	
	0.7014
(derate 34.5mW/°C above +70°C)	
θ <sub>JA</sub>	
θ <sub>JC</sub>	2°C/W
Continuous Power Dissipation (T <sub>A</sub> = +70°C)	
32-Pin TQFN 7mm x 7mm Multiple Layer PCB	
(derate 37mW/°C above +70°C)	2.96W
θ <sub>JA</sub>	27°C/W
θ <sub>JC</sub>	
Operating Temperature Range4	0°C to +85°C
Storage Temperature Range65	°C to +150°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C

Note 1: VS cannot exceed PVDD + 0.3V. See the Power-Supply Sequencing section.

Note 2: Thermal performance of this device is highly dependant on PCB layout. See the Applications Information section for more details.

Note 3: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **Package Information**

#### 32 TQFN-EP (7mm x 7mm)

Package Code	T3277-3
Outline Number	21-0144

#### 32 TQFN-EP (5mm x 5mm)

Package Code	T3255-4
Outline Number	21-0140

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to <u>www.maximintegrated.com/thermal-tutorial</u>.

# Mono/Stereo High-Power Class D Amplifier

### **Electrical Characteristics**

 $(V_{PVDD} = 20V, V_{VS} = 5V, V_{AGND} = V_{PGND} = 0V, V_{MOD} = V_{\overline{SHDN}} = V_{\overline{MUTE}} = 5V, REGEN = MONO = AGND, C1 = 0.1\muF, C2 = 1\muF, R_{IN} = 20k\Omega$  and  $R_{FB} = 20k\Omega$ ,  $R_L = \infty$ , AC measurement bandwidth 22Hz to 22kHz,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .) (Notes 4, 5)

PARAMETER	SYMBOL		CONDI	TIONS	MIN	TYP	MAX	UNITS
AMPLIFIER DC CHARACTERIS	TICS	1						1
Speaker Supply Voltage Range	PVDD	Inferred from P	SRR test		8		28	V
Preamplifier Supply Voltage Range	VS	(Notes 1 and 7)	)		4.5		5.5	V
Undervoltage Lockout	UVLO					7		V
	1	R <sub>L</sub> = ∞, V <sub>REGE</sub>	N = 5V,	T <sub>A</sub> = +25°C		30	45	m 4
Quiescent Supply Current	I <sub>PVDD</sub>	V <sub>VS</sub> = open		$T_A = T_{MIN}$ to $T_{MAX}$			50	mA
	h in	R <sub>L</sub> = ∞, V <sub>REGE</sub>	N = 0V,	T <sub>A</sub> = +25°C		14	20	mA
	IVS	$V_{VS} = 5V$		$T_A = T_{MIN}$ to $T_{MAX}$			22	
Shutdown Supply Current				I <sub>PVDD</sub>		1	10	μA
Shuldown Supply Current	ISHDN	V <sub>SHDN</sub> = 0V		I <sub>VS</sub>			10	μΑ
REG Voltage	V <sub>REG</sub>					4.2		V
Preregulator Voltage	VS	Internal regulat	ed 5V, V <sub>F</sub>	REGEN = 5V		4.8		V
COM Voltage	V <sub>COM</sub>				1.9	2.05	2.2	V
INPUT AMPLIFIER CHARACTE	RISTICS							
Capacitive Drive	CL					30		pF
Output Swing (Note 6)		Sinking ±1mA			±2		V	
Open-Loop Gain	A <sub>VO</sub>	V <sub>FB</sub> = V <sub>COM</sub> =	±500mV, I		88		dB	
Input Offset Voltage	V <sub>OS</sub>					±1		mV
Input Amplifier Slew Rate						2.5		V/µs
Input Amplifier Unity-Gain Bandwidth						3.5		MHz
AMPLIFIER CHARACTERISTIC	S	L						,
Output Amplifier Cain (Nate 9)	Δ	MAX9736A, MA	AX9736D		16.5	17	17.5	dD
Output Amplifier Gain (Note 8)	A <sub>V</sub>	MAX9736B			13.1	13.6	14.1	dB
Output Current Limit					3.3	4.6		A
Output Offset	V <sub>OS</sub>	OUT_+ to OUT				±2	±10	mV
Power-Supply Rejection Ratio	PSRR	PVDD = 8V to 2			65	80		dB
		f = 1kHz, 100m	V <sub>P-P</sub> ripp		-	67		
		N = 10V	Stereo	$R_L = 8\Omega$		8		-
		V <sub>PVDD</sub> = 12V	Mana	$R_L = 4\Omega$		13		-
MAX9736A Output Power	Davis		Mono	$R_L = 4\Omega$		15.5		W
(THD+N = 1%)	POUT_1%	V <sub>PVDD</sub> = 18V	Stereo Mono	$R_{L} = 8\Omega$ $R_{L} = 4\Omega$	-	13.5 27		v
			Stereo	$R_L = 4\Omega$ $R_L = 8\Omega$	-	13.5		-
		V <sub>PVDD</sub> = 24V	Mono	$R_L = 6\Omega$	-	27		-
				'\L - 412		21		

### **Electrical Characteristics (continued)**

 $(V_{PVDD} = 20V, V_{VS} = 5V, V_{AGND} = V_{PGND} = 0V, V_{MOD} = V_{\overline{SHDN}} = V_{\overline{MUTE}} = 5V, REGEN = MONO = AGND, C1 = 0.1\muF, C2 = 1\muF, R_{IN} = 20k\Omega$  and  $R_{FB} = 20k\Omega$ ,  $R_L = \infty$ , AC measurement bandwidth 22Hz to 22kHz,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .) (Notes 4, 5)

PARAMETER	SYMBOL		CONDITIC	NS	MIN	TYP	MAX	UNITS	
			Ctores	R <sub>L</sub> = 8Ω		6			
		V <sub>PVDD</sub> = 12V	Stereo	R <sub>L</sub> = 4Ω		11		]	
			Mono	R <sub>L</sub> = 4Ω		12			
MAX9736B Output Power (THD+N = 1%)	POUT_1%	1/-1 = -101/	Stereo	R <sub>L</sub> = 8Ω		6		W	
		V <sub>PVDD</sub> = 18V	Mono	$R_L = 4\Omega$		12			
		1/-1 = -241/	Stereo	R <sub>L</sub> = 8Ω		6			
		V <sub>PVDD</sub> = 24V	Mono	$R_L = 4\Omega$		12			
			Stereo	R <sub>L</sub> = 8Ω		10			
		V <sub>PVDD</sub> = 12V	Stereo	R <sub>L</sub> = 4Ω		16			
			Mono	$R_L = 4\Omega$		19.5			
MAX9736A Output Power (THD+N = 10%)	POUT_10%	V <sub>PVDD</sub> = 18V	Stereo	R <sub>L</sub> = 8Ω		17.5		W	
		vpvDD = 10v	Mono	$R_L = 4\Omega$		35			
		V <sub>PVDD</sub> = 24V	Stereo	R <sub>L</sub> = 8Ω		17.5			
		v PVDD - 24v	Mono	R <sub>L</sub> = 4Ω		35			
		V <sub>PVDD</sub> = 12V	Stereo	R <sub>L</sub> = 8Ω		7.5			
			Stereo	R <sub>L</sub> = 4Ω		14			
			Mono	$R_L = 4\Omega$		15			
MAX9736B Output Power (THD+N = 10%)	P <sub>OUT_10%</sub>	V <sub>PVDD</sub> = 18V	Stereo	R <sub>L</sub> = 8Ω		7.5		W	
(		VPVDD = 10V	Mono	$R_L = 4\Omega$		15			
		V <sub>PVDD</sub> = 24V	Stereo	R <sub>L</sub> = 8Ω		7.5			
		V PVDD - 24V	Mono	R <sub>L</sub> = 4Ω		15			
		1/-1 = -12/	Stereo	R <sub>L</sub> = 8Ω		8.5			
		V <sub>PVDD</sub> = 12V	Mono	R <sub>L</sub> = 4Ω		17			
MAX9736D Output Power		- 401/	Stereo	R <sub>L</sub> = 8Ω		5.5			
(Thermally Limited)		V <sub>PVDD</sub> = 18V	Mono	R <sub>L</sub> = 4Ω		11		W	
			Stereo	R <sub>L</sub> = 8Ω		3.5		1	
		V <sub>PVDD</sub> = 24V	Mono	R <sub>L</sub> = 4Ω		7		1	
Total Harmonic Distortion	TUDIN	MAX9736A, P <sub>0</sub> PWM modulati	DUT = 4W, f = on mode, R <sub>L</sub>	= 1kHz, = 8Ω		0.04		0/	
Plus Noise	THD+N	MAX9736B/D, PWM modulati				0.04		- %	
Signal to Noise Datio		A weighted		MAX9736A/D, P <sub>OUT</sub> = 8W, R <sub>L</sub> = 8Ω		96.5		dD	
Signal-to-Noise Ratio	SNR	A-weighted	MAX9 P <sub>OUT</sub>	736B, = 6W, R <sub>L</sub> = 8Ω		97		dB	
Nutra		A-weighted	MAX9	736A/D		120			
Noise	V <sub>N</sub>	(Note 9)	MAX9	736B		100		μV <sub>RMS</sub>	

### **Electrical Characteristics (continued)**

 $(V_{PVDD} = 20V, V_{VS} = 5V, V_{AGND} = V_{PGND} = 0V, V_{MOD} = V_{\overline{SHDN}} = V_{\overline{MUTE}} = 5V, REGEN = MONO = AGND, C1 = 0.1\muF, C2 = 1\muF, R_{IN} = 20k\Omega$  and  $R_{FB} = 20k\Omega$ ,  $R_L = \infty$ , AC measurement bandwidth 22Hz to 22kHz,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .) (Notes 4, 5)

PARAMETER	SYMBOL	со	NDITIONS	MIN	TYP	MAX	UNITS
Crosstalk		L to R, R to L, POUT	- = 1W, f = 1kHz, R <sub>L</sub> = 8Ω		100		dB
Efficiency	η	P <sub>OUT</sub> = 8W, MAX973	36A, PVDD = 12V, R <sub>L</sub> = 8Ω		88		%
Click and Dan Laval	K	Peak voltage, 32 samples/second,	Into mute		36		dBV
Click-and-Pop Level	K <sub>CP</sub>	A-weighted (Notes 9 and 10)	Out of mute		36		UDV
Switching Frequency				270	300	330	kHz
Spread-Spectrum Bandwidth					±4		kHz
Thermal Shutdown Level					160		°C
Thermal Shutdown Hysteresis					30		°C
Turn-On Time	t <sub>ON</sub>				110		ms
DIGITAL INTERFACE							
Input Voltage High	V <sub>INH</sub>			2			V
Input Voltage Low	V <sub>INL</sub>					0.8	V
Input Voltage Hysteresis					50		mV
Input Leakage Current						±10	μA

**Note 4:** All devices are 100% production tested at +25°C. All temperature limits are guaranteed by design.

**Note 5:** Stereo mode (MONO = GND) specified with  $8\Omega$  resistive load in series with a  $68\mu$ H inductive load connected across BTL outputs. Mono mode (MONO = 5V) specified with a  $4\Omega$  resistive load in series with a  $33\mu$ H inductive load connected across BTL outputs.

**Note 6:** Output swing is specified with respect to  $V_{COM}$ .

- **Note 7:** For typical applications, an external 5V supply is not required. Therefore, set REGEN = 5V. If thermal performance is a concern, set REGEN = 0V and provide an external regulated 5V supply.
- **Note 8:** Output amplifier gain is defined as:

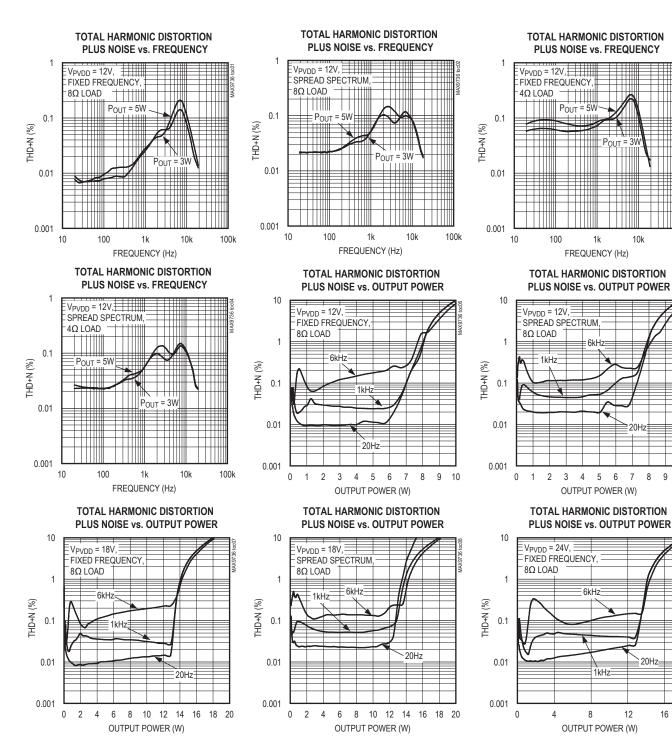
$$20 \times \text{log} \left( \frac{|(V_{\text{OUT}\_+}) - (V_{\text{OUT}\_-})|}{|V_{\text{FB}\_}|} \right)$$

Note 9: Amplifier inputs AC-coupled to GND.

Note 10: Specified at room temperature with an 8Ω resistive load in series with a 68µH inductive load connected across BTL outputs. Mode transitions controlled by SHDN control pin.

### **Typical Operating Characteristics**

(MAX9736A, V<sub>PVDD</sub> = 12V, MOD = high, spread-spectrum modulation mode, V<sub>GND</sub> = V<sub>PGND</sub> = 0V, V<sub>SHDN</sub> = V<sub>MUTE</sub> = 5V, unless otherwise noted.)



12

20Hz

16

100k

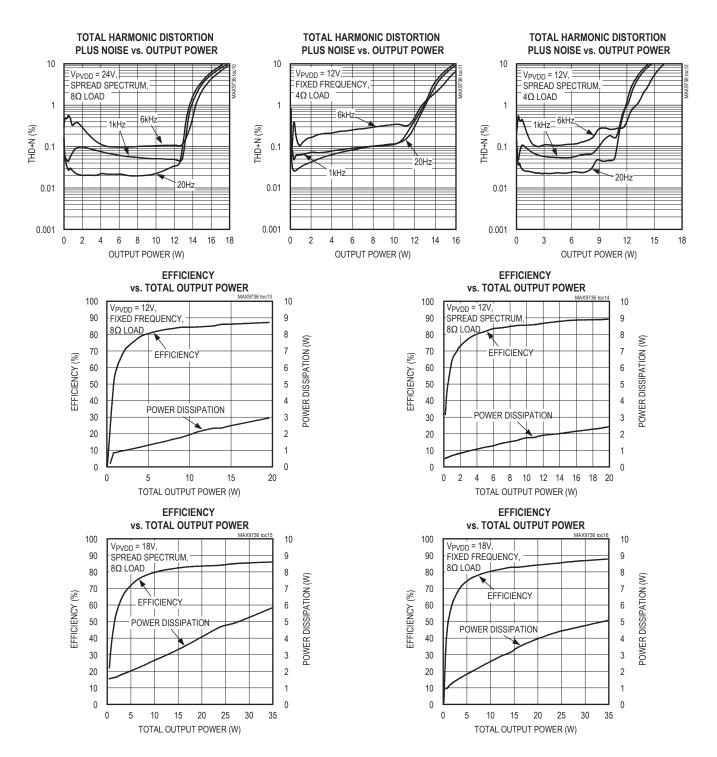
10k

20Hz

7 8 9 10

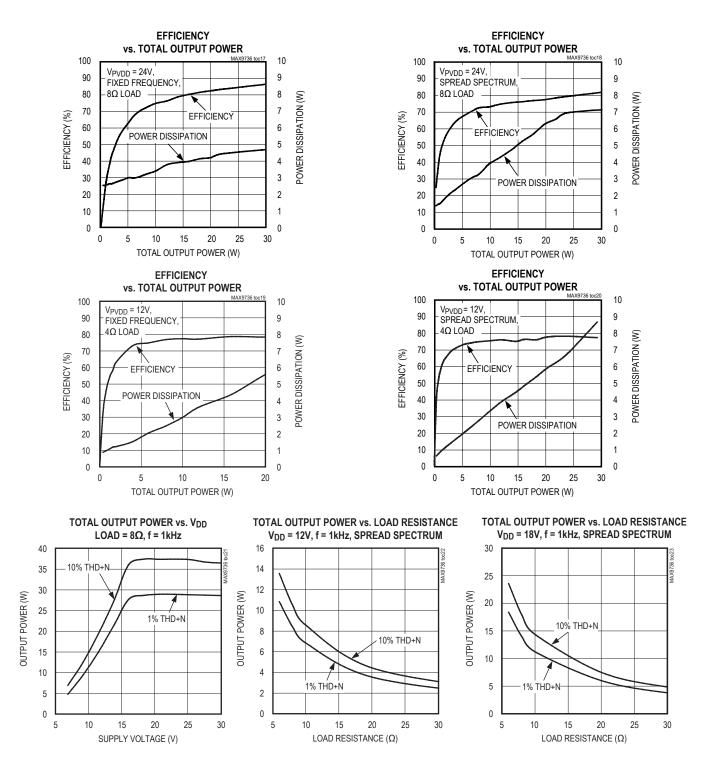
## Mono/Stereo High-Power Class D Amplifier

### **Typical Operating Characteristics (continued)**



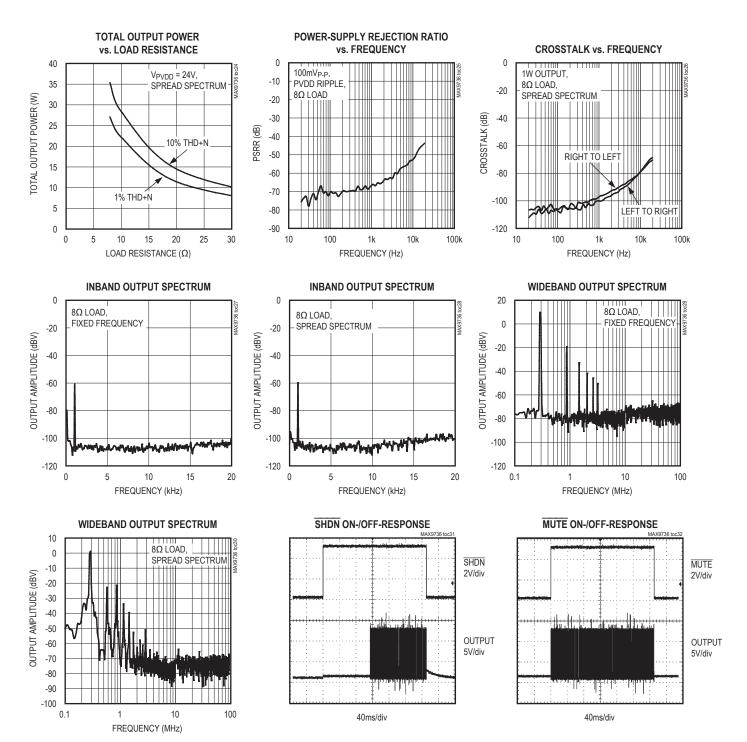
## Mono/Stereo High-Power Class D Amplifier

### **Typical Operating Characteristics (continued)**



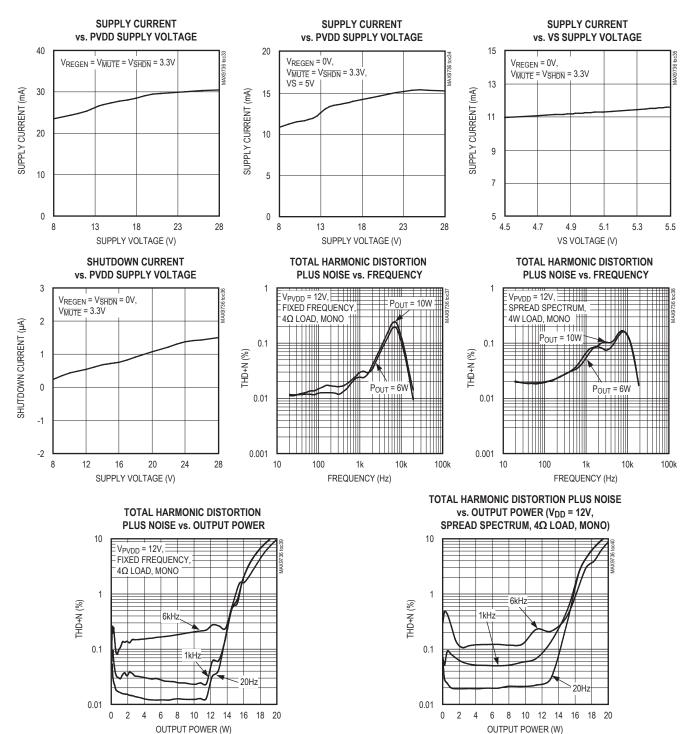
## Mono/Stereo High-Power Class D Amplifier

### **Typical Operating Characteristics (continued)**



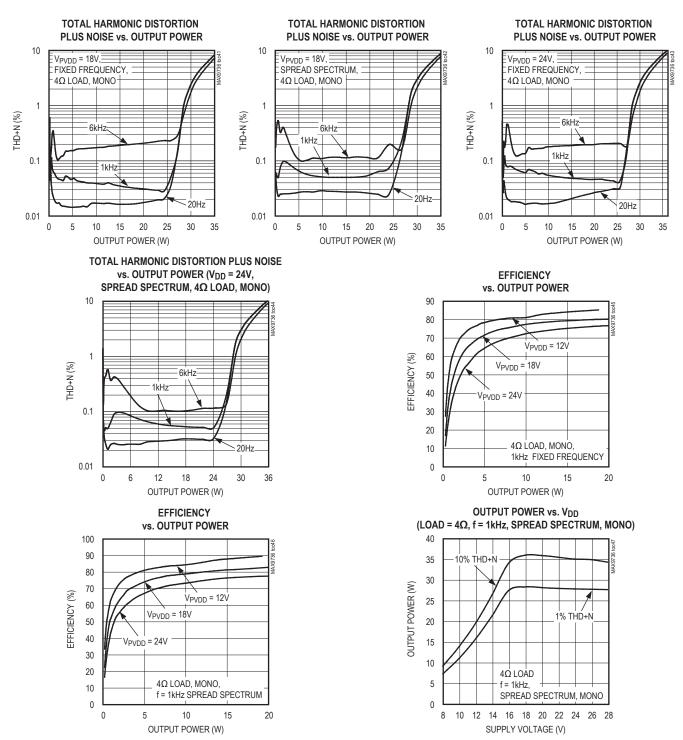
# Mono/Stereo High-Power Class D Amplifier

### **Typical Operating Characteristics (continued)**



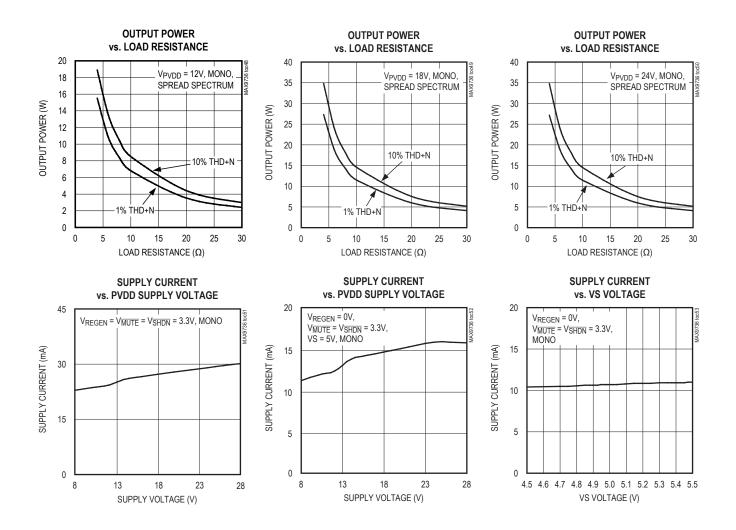
# Mono/Stereo High-Power Class D Amplifier

### **Typical Operating Characteristics (continued)**



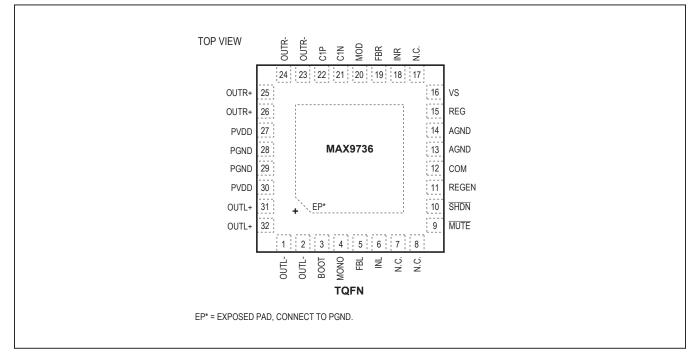
## Mono/Stereo High-Power Class D Amplifier

### **Typical Operating Characteristics (continued)**



# Mono/Stereo High-Power Class D Amplifier

# **Pin Configuration**



### **Pin Description**

PIN	NAME	FUNCTION
1, 2	OUTL-	Left-Channel Negative Speaker Output
3	BOOT	Charge-Pump Output. Connect a 1µF charge-pump holding capacitor from BOOT to PVDD.
4	MONO	Mono Select. Set MONO high for mono mode, low for stereo mode.
5	FBL	Left-Channel Feedback. Connect feedback resistor between FBL and INL to set amplifier gain.
6	INL	Stereo Left-Channel Inverting Input. In mono mode, INL is the inverting audio input for the mono amplifier.
7, 8, 17	N.C.	No Connection. Not internally connected. OK to connect to PGND.
9	MUTE	Mute Input. Drive MUTE low to place the device in mute mode.
10	SHDN	Shutdown Input. Drive SHDN low to place the device in shutdown mode.
11	REGEN	Internal Regulator Enable Input. Connect REGEN to SHDN to enable the internal regulator. Drive REGEN low to disable the internal regulator, and supply the device with an external 5V supply on VS. See the <i>Power-Supply Sequencing</i> section.
12	COM	Internal 2V Bias. Bypass COM to AGND with a 1µF capacitor.
13, 14	AGND	Analog Ground
15	REG	Internal Regulator Output. Bypass REG to AGND with a 1µF capacitor.
16	VS	5V Regulator Supply. Bypass VS to AGND with a 1µF capacitor. If REGEN is low, the internal regulator is disabled, and an external 5V supply must be connected to VS. See the <i>Power-Supply Sequencing</i> section.

PIN	NAME	FUNCTION
18	INR	Stereo Right-Channel Inverting Audio Input. In mono mode, INR is the inverting audio input for the uncommitted preamplifier (see the <i>Mono Configuration</i> section for more details).
19	FBR	Right-Channel Feedback. Connect feedback resistor between FBR and INR to set amplifier gain.
20	MOD	Output Modulation Select. Sets the output modulation scheme: V <sub>MOD</sub> = Low, classic PWM/fixed-frequency mode V <sub>MOD</sub> = High, filterless modulation/spread-spectrum mode
21	C1N	Charge-Pump Flying-Capacitor Negative Terminal
22	C1P	Charge-Pump Flying-Capacitor Positive Terminal
23, 24	OUTR-	Right-Channel Negative Speaker Output
25, 26	OUTR+	Right-Channel Positive Speaker Output
27, 30	PVDD	Power Supply. Bypass each PVDD pin to ground with 0.1 $\mu$ F capacitors. Also, use a single 220 $\mu$ F capacitor between PVDD and PGND.
28, 29	PGND	Power Ground
31, 32	OUTL+	Left-Channel Positive Speaker Output
	EP	Exposed Pad. Must be externally connected to PGND.

### **Pin Description (continued)**

### **Detailed Description**

The MAX9736A/B/D filterless, stereo Class D audio power amplifiers offer Class AB performance and Class D efficiency with minimal board space. The MAX9736A outputs 2x15W in stereo mode and 30W in mono mode. The MAX9736B/D output 2x6W in stereo mode and 12W in mono mode. These devices operate from an 8V to 28V supply range.

The MAX9736 features a filterless, spread-spectrum switching mode (MOD = high) or a classic PWM fixed-frequency switching mode (MOD = low).

The MAX9736 features externally set gain and a low-power shutdown mode that reduces supply current to less than 1 $\mu$ A. Comprehensive click-and-pop circuitry minimizes noise into and out of shutdown or mute.

### **Operating Modes**

#### **Filterless Modulation/PWM Modulation**

The MAX9736 features two output modulation schemes, filterless modulation (MOD = high) or classic PWM (MOD = low). Maxim's unique, filterless modulation scheme eliminates the LC filter required by traditional Class D amplifiers, reducing component count, conserving board space, and reducing system cost. Configure for classic PWM output when using a full LC filter.

Click-and-pop protection does not apply when the output is switching between modulation schemes. To maintain

click-and-pop protection when switching between output schemes the device must enter shutdown mode and be configured to the new output scheme before the startup sequence is finished.

#### Spread-Spectrum Mode

The MAX9736 features a unique spread-spectrum mode that flattens the wideband spectral components, improving EMI radiated by the speaker and cables. The switching frequency of the Class D amplifier varies randomly by ±6kHz around the 300kHz center frequency. Instead of a large amount of spectral energy present at multiples of the switching frequency, the energy is spread over a bandwidth that increases with frequency. Above a few megahertz, the wideband spectrum looks like white noise for EMI purposes. A proprietary amplifier topology ensures this white noise does not corrupt the noise floor in the audio bandwidth. The spread-spectrum mode is enabled only with filterless modulation.

#### Efficiency

The high efficiency of a Class D amplifier is due to the switching operation of the output stage transistors. In a Class D amplifier, the output transistors act as switches and consume negligible power. Power loss associated with the Class D output stage is due to the I<sup>2</sup>R loss of the MOSFET on-resistance, various switching losses, and quiescent current overhead.

The theoretical best efficiency of a linear amplifier is 78% at peak output power. Under typical music reproduction levels, the efficiency falls below 30%, whereas the MAX9736 exhibits > 80% efficiency under the same conditions (Figure 1).

#### Shutdown

The MAX9736 features a shutdown mode that reduces power consumption and extends battery life in portable applications. The shutdown mode reduces supply current to 1 $\mu$ A (typ). Drive SHDN high for normal operation. Drive SHDN low to place the device in low-power shutdown mode. In shutdown mode, the outputs are high impedance; and the common-mode voltage at the output decays to zero. In shutdown mode, connect REGEN low to minimize current consumption.

#### **Mute Function**

The MAX9736 features a clickless-and-popless mute mode. When the device is muted, the signal is attenuated at the speaker and the outputs stop switching. To mute the MAX9736, drive MUTE low. Hold MUTE low during system power-up and power-down to ensure that clicks and pops caused by circuits before the MAX9736 are suppressed.

#### **Click-and-Pop Suppression**

The MAX9736 features comprehensive click-and-pop suppression that minimizes audible transients on startup and shutdown. While in shutdown, the H-bridge is in a high-impedance state.

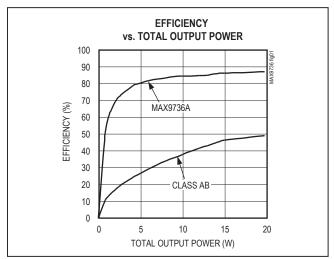


Figure 1. MAX9736A Efficiency vs. Class AB Efficiency

#### **Mono Configuration**

The MAX9736 features a mono mode that allows the right and left channels to operate in parallel, achieving up to 30W (MAX9736A) of output power. Apply a logic-high to MONO to enable mono mode. In mono mode, an audio signal applied to the left channel (INL) is routed to the H-bridges of both channels. Also in mono mode, the rightchannel preamplifier becomes an uncommitted operational amplifier, allowing for flexibility in system design. Connect OUTL+ to OUTR+ and OUTL- to OUTR- using heavy PCB traces as close as possible to the device. Driving MONO low (stereo mode) while the outputs are wired together in mono mode can trigger the short-circuit or thermal-overload protection or both.

#### **Current Limit**

When the output current reaches the current limit, 4.6A (typ), the MAX9736 disables the outputs and initiates a 450µs startup sequence. The shutdown and startup sequence is repeated until the output fault is removed. Properly designed applications do not enter current-limit mode unless the output is short circuited or connected incorrectly.

#### **Thermal Shutdown**

When the die temperature reaches the thermal shutdown threshold, +160°C (typ), the MAX9736 outputs are disabled. When the die temperature decreases by 30°C, normal operation resumes. Some causes of thermal shutdown are excessively low load impedance, poor thermal contact between the MAX9736's exposed pad and the PCB, elevated ambient temperature, or poor PCB layout and assembly.

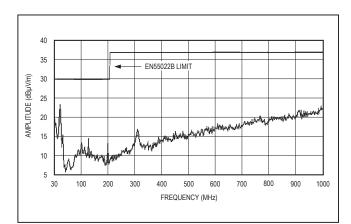


Figure 2. EMI Performance

### **Applications Information**

#### **Filterless Class D Operation**

The MAX9736 meets EN55022B EMC radiation limits with an inexpensive ferrite bead and capacitor filter when the speaker leads are less than or equal to 1m. Select a ferrite bead with  $100\Omega$  to  $600\Omega$  impedance and rated for at least 2A. The capacitor value varies based on the ferrite bead chosen and the speaker lead length. See Figure 3 for the correct connections of these components.

When evaluating the MAX9736 with a ferrite bead filter and resistive load, include a series inductor ( $68\mu$ H for  $8\Omega$  load and  $33\mu$ H for  $4\Omega$  load) to model the actual loudspeaker's behavior. Omitting the series inductor reduces the efficiency, the THD+N performance, and the output power of the MAX9736. When evaluating with a load speaker, no series inductor is required.

#### **Inductor-Based Output Filters**

Some applications use the MAX9736 with a full inductor-/ capacitor-based (LC) output filter. Select the PWM output mode for best audio performance. See <u>Figure 4</u> for the correct connections of these components.

The load impedance of the speaker determines the filter component selection (see <u>Table 1</u>).

Inductors L1 and L2, and capacitor C1 form the primary output filter. Capacitors C2 and C3 provide common-mode filtering to reduce radiated emissions. Capacitors C4 and C5, plus resistors R1 and R2, form a Zobel at the output. A Zobel corrects the output loading to compensate for the rising impedance of the loudspeaker. Without a Zobel the filter exhibits a peak response near the cutoff frequency.

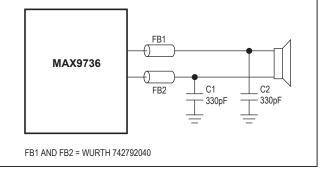


Figure 3. Ferrite Bead Filter

### Table 1. Suggested Values for LC Filter

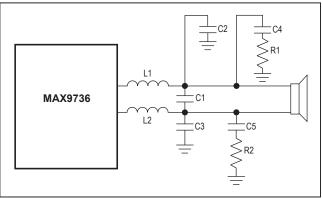
#### **Component Selection**

#### **Gain-Setting Resistors**

External feedback resistors set the gain of the MAX9736. The output stage provides a fixed internal gain in addition to the externally set input stage gain. For the MAX9736A/D, the fixed output-stage gain is set at 17dB (7V/V). For the MAX9736B, the fixed output-stage gain is set at 13.6dB (4.8V/V). Set overall gain by using resistors  $R_F$  and  $R_{IN}$  (Figure 5) as follows:

MAX9736A : A<sub>V</sub> = -7.1
$$\left(\frac{R_F}{R_{IN}}\right)$$
V / V  
MAX9736B : A<sub>V</sub> = -4.8 $\left(\frac{R_F}{R_{IN}}\right)$ V / V

where  $A_V$  is the desired voltage gain. Choose  $R_F$  between  $10 k \Omega$  and  $50 k \Omega.$ 





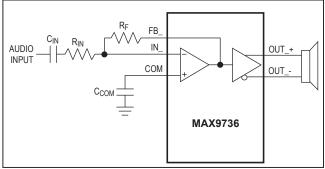


Figure 5. Setting Gain

•••					
R <sub>L</sub> (Ω)	L1, L2 (µH)	C1 (µF)	C2, C3 (µF)	C4, C5 (µF)	R1, R2 (Ω)
4	10	0.47	0.10	0.22	10
8	15	0.15	0.15	0.15	15
16	33	0.10	0.10	0.10	33

The FB terminal is an op amp output and the IN terminal is the op amp inverting input, allowing the MAX9736 to be configured as a summing amplifier, a filter, or an equalizer.

#### **Input Capacitor**

An input capacitor,  $C_{IN}$ , in conjunction with the input resistor,  $R_{IN}$ , of the MAX9736 forms a highpass filter that removes the DC bias from an incoming signal. The AC-coupling capacitor allows the amplifier to automatically bias the signal to an optimum DC level. Assuming zero-source impedance, the -3dB point of the highpass filter is given by:

$$f_{-3dB} = \frac{1}{2\pi R_{IN}C_{IN}}$$

Choose  $C_{IN}$  so that  $f_{-3dB}$  is well below the lowest frequency of interest. Use capacitors whose dielectrics have low voltage coefficients. Capacitors with high-voltage coefficients cause increased distortion close to  $f_{-3dB}$ .

#### **COM Capacitor**

COM is the output of the internally generated DC bias voltage. Bypass COM with a  $1\mu$ F capacitor to AGND.

#### **Power Supplies**

The MAX9736 features separate supplies for signal and power portions of the device, allowing for the optimum combination of headroom, power dissipation, and noise immunity. The speaker amplifiers are powered from PVDD and can range from 8V to 28V. The remainder of the MAX9736 is powered by VS.

#### **Power-Supply Sequencing**

During power-up and power-down, VS must not exceed PVDD. VS greater than PVDD will damage the device.

#### **Internal Regulator**

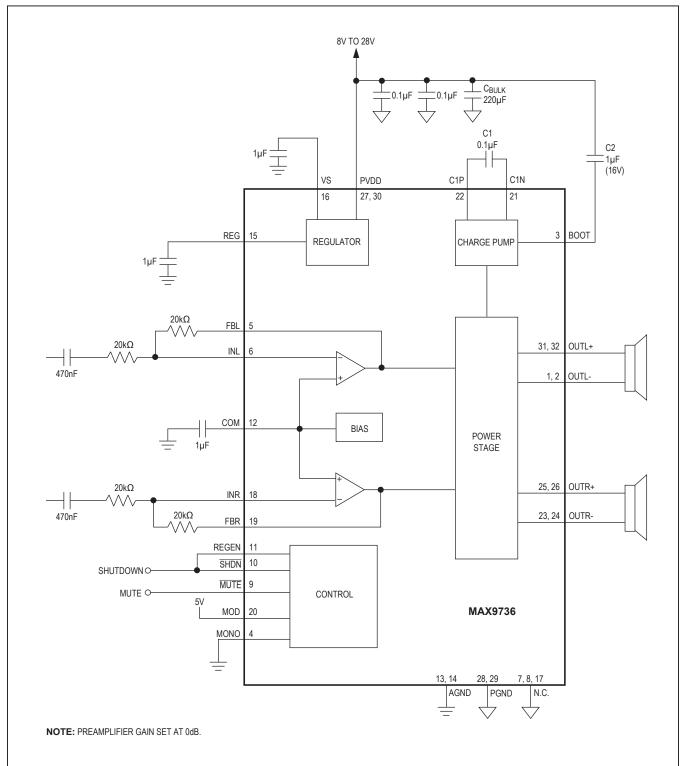
The MAX9736 features an internal 5V regulator, VS, powered from PVDD. Connect REGEN to SHDN so that the internal 5V regulator is enabled/disabled when the MAX9736 is enabled/disabled. If an external 5V supply is available, drive REGEN low and connect external 5V supply to VS to minimize the power dissipation of the MAX9736.

#### Supply Bypassing, Layout, and Grounding

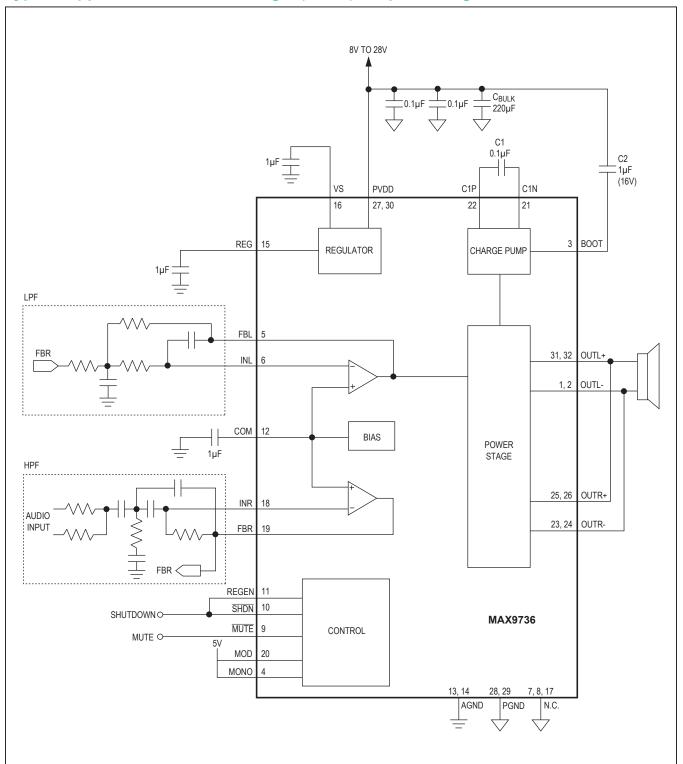
Proper layout and grounding are essential for optimum performance. Use wide traces for the power-supply inputs and amplifier outputs to minimize losses due to parasitic trace resistance. Proper grounding improves audio performance, minimizes crosstalk between channels, and prevents switching noise from coupling into the audio signal. Connect PGND and AGND together at a single point on the PCB. Route all traces that carry switching transients away from AGND and the traces/components in the audio signal path.

Bypass each PVDD pin with a  $0.1\mu$ F capacitor to PGND. Place the bypass capacitors as close as possible to the MAX9736. Place a 220 $\mu$ F capacitor between PVDD and PGND. Bypass VS with a  $1\mu$ F capacitor to AGND.

Use wide, low-resistance output traces. Current drawn from the outputs increases as load impedance decreases. High-output trace resistance decreases the power delivered to the load. The MAX9736 TQFN package features an exposed thermal paddle on its underside. This paddle lowers the package's thermal resistance by providing a heat conduction path from the die to the PCB. Connect the exposed thermal pad to PGND by using a large pad and multiple vias to the PGND plane.



# **Typical Application Circuit for Stereo Output Configuration**



## Typical Application Circuit for Single (Mono) Output Configuration

# Mono/Stereo High-Power Class D Amplifier

# **Ordering Information**

PART	STEREO/MONO OUTPUT POWER	PIN-PACKAGE				
MAX9736AETJ+	2 x 15W/ 1 x 30W	32 TQFN-EP* 7mm x 7mm				
MAX9736AETJ/V+	2 x 15W/ 1 x 30W	32 TQFN-EP* 7mm x 7mm				
MAX9736BETJ+	2 x 6W/ 1 x 12W	32 TQFN-EP* 7mm x 7mm				
MAX9736BETJ/V+	2 x 6W/ 1 x 12W	32 TQFN-EP* 7mm x 7mm				
MAX9736DETJ+	2 x 6W/ 1 x 12W	32 TQFN-EP* 5mm x 5mm				

**Note:** All devices are specified over the -40°C to +85°C operating temperature range.

+Denotes a lead(Pb)-free/RoHS-compliant package.

\*EP = Exposed pad.

N denotes an automotive qualified part.

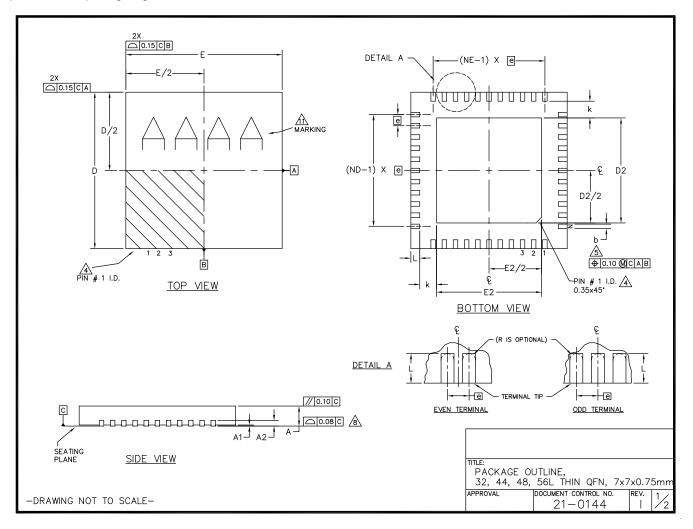
# **Chip Information**

PROCESS: BICMOS

# Mono/Stereo High-Power Class D Amplifier

### **Package Information**

For the latest package outline information and land patterns (footprints), go to <u>www.maximintegrated.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.



### **Package Information (continued)**

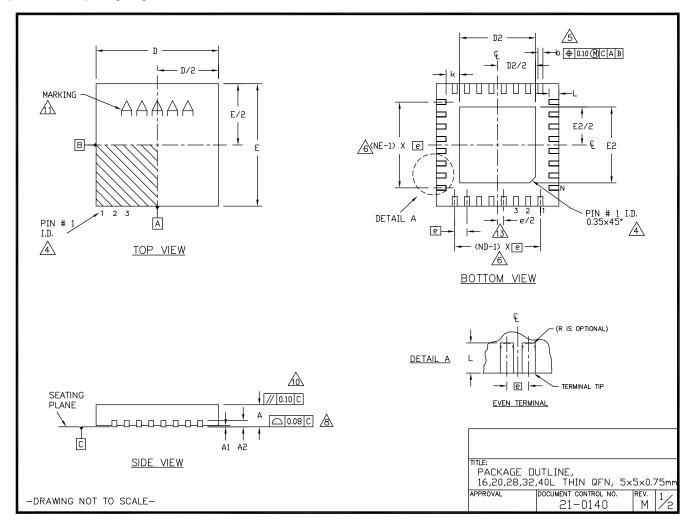
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					COM	IMON [	MENSI	ONS										EXPOS	ed pai	) vari	ATIONS			
									CUSTOM PKG.		PKG. DEPO			POPULATED D2				E2	JEDEC MO220					
5.40										``	T4877-	'				CODES	LEADS	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	REV. C
PKG		32L 7x7			4L 7x7			-8L 7x7			8L 7x7			6L 7x7		T3277-2	-	4.55	4.70	4.85	4.55	4.70	4.85	-
SYMBOL	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	T3277-3	-	4.55	4.70	4.85	4.55	4.70	4.85	-
A	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	T4477-2	-	4.55	4.70	4.85	4.55	4.70	4.85	WKKD-1
A1	0	0.02	0.05	0	0.02	0.05	0	0.02	0.05	0	0.02	0.05	0	-	0.05	T4477-3	-	4.55	4.70	4.85	4.55	4.70	4.85	WKKD-1
A2	0.	.20 RE	F.	С	.20 R	EF.	C	.20 R	EF.	0	.20 R	EF.	c	.20 R	EF.	T4877-3	-	4.95	5.10	5.25	4.95	5.10	5.25	-
b	0.25	0.30	0.35	0.20	0.25	0.30	0.20	0.25	0.30	0.20	0.25	0.30	0.15	0.20	0.25	T4877-4	-		-		-		5.60	-
D	6.90	7.00	7.10	6.90	7.00	7.10	6.90	7.00	7.10	6.90	7.00	7.10	6.90	7.00	7.10	T4877-6	-						5.60	-
Е	6.90	7.00	7.10	6.90	7.00	7.10	6.90	7.00	7.10	6.90	7.00	7.10	6.90	7.00	7.10	T4877-7	-		5.10					-
е	0.	.65 BS	c.	0	.50 B	SC.	C	.50 B	SC.	0	.50 B	SC.	0	.40 B	SC.	T4877M-1	-		5.50		-	-		_
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L	0.45	0.55	0.65	0.45	0.55	0.65	0.30	0.40	0.50	0.45	0.55	0.65	0.30	0.40	0.50	T4877MN-8	-		5.50		-	-		-
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# Mono/Stereo High-Power Class D Amplifier

### **Package Information (continued)**

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A	0.70			0.70 0.7	-	-			-				0.75			T1	655-2	3.00	3.10	3.20	3.00	3.10	3.20
A1	0	0.02	0.05		2 0.05		0.02		0	0.02	0.05	0	0.02	0.05		T1	655-3	3.00	3.10	3.20	3.00	3.10	3.20
A2	-	20 RE		0.20 F		-	20 RE			20 REF			20 RE			T1	655-4	2.19	2.29	2.39	2.19	2.29	2.39
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# Mono/Stereo High-Power Class D Amplifier

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	1/08	Initial release	—
1	12/08	Corrected various errors	1–15, 17–21
2	8/09	Added MAX9736D and automotive parts numbers and updated the <i>Absolute Maximum Ratings</i> section	1, 2, 3, 4,13, 15, 19, 20, 21
3	9/09	Corrected error in Absolute Maximum Ratings, Pin Description, Typical Application Circuit for Stereo Output Configuration, and Typical Application Circuit for Single (Mono) Output Configuration	2, 12, 17, 18
4	11/19	Updated Ordering Information table	20

For pricing, delivery, and ordering information, please visit Maxim Integrated's online storefront at https://www.maximintegrated.com/en/storefront/storefront.html.

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