

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

General Description

Features

The MAX98314 mono 3.2W Class D amplifier provides Class AB audio performance with Class D efficiency. This device offers five selectable gain settings (0dB, 3dB, 6dB, 9dB, and 12dB) set by a single gain-select input (GAIN).

Active emissions limiting (AEL) edge rate and overshoot control circuitry and a filterless spread-spectrum modulation (SSM) scheme greatly reduce EMI and eliminate the need for output filtering found in traditional Class D devices.

The IC's low 0.95mA at 3.7V, 1.2mA at 5.0V guiescent current extends battery life in portable applications.

Highly linear, integrated input coupling capacitors (C_{IN}) reduce solution size and provide excellent THD+N, PSRR, and CMRR performance at low frequencies vs. standard Class D amplifiers using external input capacitors.

The IC is available in a small 9-bump, 0.3mm pitch WLP (1.0mm x 1.0mm x 0.80mm) package and is specified over the -40°C to +85°C extended temperature range.

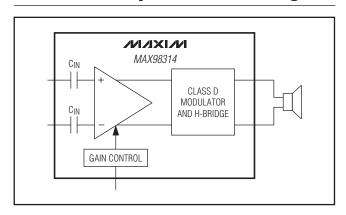
Applications

Mobile Phones Portable Audio Notebook Computers MP3 Players **Netbook Computers VoIP Phones**

Ordering Information appears at end of data sheet.

- ♦ Integrated Input Coupling Capacitors with **Excellent Linearity**
 - \Leftrightarrow f_C = 100Hz (6dB)
 - \Leftrightarrow f_C = 200Hz (12dB)
- **♦ Low Quiescent Current**
 - ♦ 0.95mA at 3.7V
 - ♦ 1.2mA at 5.0V
- ♦ Delivers High Output Power at 10% THD+N
 - \Rightarrow 3.2W into 4Ω , $V_{PVDD} = 5V$
 - \Rightarrow 960mW into 8 Ω , V_{PVDD} = 3.7V
- ♦ Ultra-Low Noise: 19µV
- **♦** Eliminates Output Filtering Requirement
 - ♦ Spread Spectrum and Active Emissions Limiting
- ♦ Click-and-Pop Suppression
- **♦ Thermal and Overcurrent Protection**
- **♦ Low Current Shutdown Mode**
- ♦ Small, Space-Saving Package

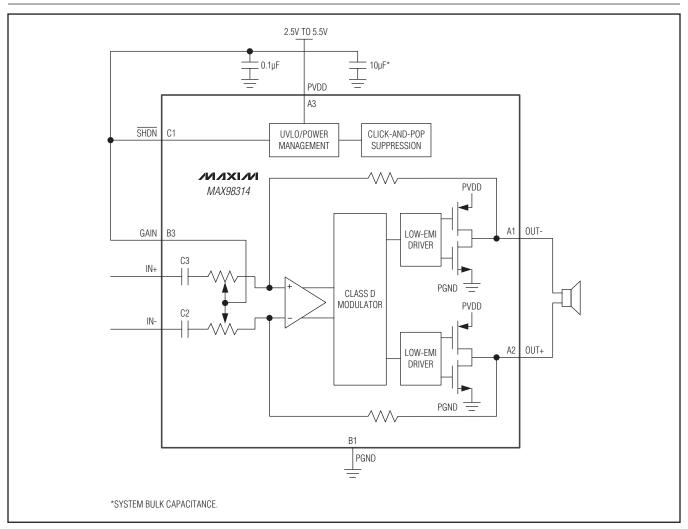
Simplified Block Diagram



For related parts and recommended products to use with this part, refer to: www.maxim-ic.com/MAX98314.related

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Functional Diagram/Typical Application Circuit



Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

ABSOLUTE MAXIMUM RATINGS

PVDD, IN+, IN-, SHDN, GAIN to PGND	0.3V to +6V
OUT+, OUT- to PGND	
Continuous Current In/Out of PVDD, PGND, OU	T750mA
Continuous Input Current (all other pins)	±20mA
Duration of Short Circuit Between	
OUT_ to PVDD, PGND	Continuous
Between OUT+ and OUT- Pins	Continuous

Continuous Power Dissipation ($T_A = +70^{\circ}$ C) for	or Multilayer Board
WLP (derate 10.64mW/°C above +70°C)	851mW
Junction Temperature	+150°C
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Soldering Temperature (reflow)	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

PACKAGE THERMAL CHARACTERISTICS (Note 1)

WI P

Junction-to-Ambient Thermal Resistance (θ,JA) 102°C/W Junction-to-Case Thermal Resistance (θ_{JC})......47°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

ELECTRICAL CHARACTERISTICS

 $(V_{PVDD} = V_{\overline{SHDN}} = V_{GAIN} = 5V, V_{PGND} = 0V, A_V = 6dB (GAIN = PVDD), R_L = \infty, R_L connected between OUT+ to OUT-, AC measure$ ment bandwidth 20Hz to 22kHz, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25$ °C.) (Note 2, 3)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS		
AMPLIFIER CHARACTERISTIC	S								
Supply Voltage Range	V _{PVDD}	Guarantee	Guaranteed by PSRR test		2.5		5.5	V	
Undervoltage Lockout	UVLO	PVDD falli	ng			1.8	2.2	V	
Quiescent Current	1	V _{PVDD} = 5	5V			1.2	1.8	m A	
Quiescent Current	I _{PVDD}	V _{PVDD} = 3	3.7V			0.95		mA	
Shutdown Supply Current	ISHDN	V _{SHDN} =	$V_{A} = +25^{\circ}$	C		< 0.1	10	μA	
Turn-On Time	t _{ON}					3.7	10	ms	
Bias Voltage	V _{BIAS}					V _{PVDD} /2		V	
			GAIN conne	cted to PGND	11.75	12	12.25		
		A _V f = 1kHz		cted to PGND kΩ ±5% resistor	8.75	9	9.25		
Voltage Gain	A _V		GAIN conne	AIN connected to PVDD		6	6.25	dB	
					cted to PVDD kΩ ±5% resistor	2.75	3	3.25	
			GAIN uncon	nected	-0.25	0	+0.25		
Input Capacitance	C _{IN}	All gains				0.011		μF	
				$A_V = 12dB$		199			
Highpass Corner Frequency	f _C	-3dB down		$A_V = 9dB$		139]	
			n	$A_V = 6dB$	63	100	189	Hz	
				$A_V = 3dB$		70			
				$A_V = 0dB$		50		1	

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{PVDD} = V_{\overline{SHDN}} = V_{GAIN} = 5V, V_{PGND} = 0V, A_V = 6dB (GAIN = PVDD), R_L = \infty, R_L connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, <math>T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 2, 3)

Output Offset Voltage Vos Ta = +25°C (Note 4) ±1 ±3 mV	PARAMETER	SYMBOL	CONDITIO	ONS	MIN	TYP	MAX	UNITS
Figure 1	Common-Mode Rejection Ratio	CMRR	f _{IN} = 1kHz, input referre	d		67		dB
Voltage, T _A = +25°C, A-weighted, 32 samples per second, T _A = +25°C (Notes 4, 5) Out of shutdown -82 Out of shutdown -82	Output Offset Voltage	Vos	$T_A = +25^{\circ}C \text{ (Note 4)}$			±1	±3	mV
Samples per second, T _A = +25°C (Notes 4, 5) Out of shutdown -82	Click-and-Pop Level	IZ.	voltage, $T_A = +25$ °C,	Into shutdown		-59		dBV
Power-Supply Rejection Ratio (Note 4) PSRR Power-Supply Rejection Ratio (Note 4) F = 217Hz		NCP	samples per second,	Out of shutdown		-82		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Power-Supply Rejection Ratio		$V_{PVDD} = 2.5V \text{ to } 5.5V, T$	A = +25°C	70	90		
VRIPPLE = 200mVp-p f = 1kHz 72		PSRR		f = 217Hz		74		dB
$ \text{Output Power} \\ \text{Output Power} \\ \text{Output Power} \\ \text{Pout} \\ \\ \text{Fil} = 4\Omega + 33\mu \\ \\ \text{Fil} = 4\Omega + 32\mu \\$	(Note 4)	1 01111	$V_{RIPPLE} = 200 \text{mV}_{P-P}$			72		J GB
$ \text{Output Power} \\ \text{Output Power} \\ \text{Pout} \\ \text{Pout} \\ \\ \text{Pout} \\$								
$POUT = \begin{array}{c ccccccccccccccccccccccccccccccccccc$			7 1					-
$ \text{Output Power} \\ \text{Output Power} \\ \text{Pout} \\ \begin{tabular}{l l l l l l l l l l l l l l l l l l l $						2.2		
Output Power $ POUT = $			$R_L = 4\Omega + 33\mu H$	$V_{PVDD} = 3.7V$		1.7		
Output Power $ P_{OUT} = $			f = 1kHz	$V_{PVDD} = 5.0V$		2.6		W
Output Power POUT THD+N = 10% f = 1kHz VPVDD = 5.0V 1.8 W THD+N = 10% f = 1kHz VPVDD = 4.2V 1.2 VPVDD = 3.7V 0.96 VPVDD = 5.0V 1.4 VPVDD = 5.0V 1.4 VPVDD = 4.2V 1 VPVDD = 4.2V 1 VPVDD = 3.7V 0.8 N <td< td=""><td rowspan="3">Output Power</td><td>$V_{PVDD} = 4.2V$</td><td></td><td>1.8</td><td></td></td<>	Output Power			$V_{PVDD} = 4.2V$		1.8		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$		D		$V_{PVDD} = 3.7V$		1.4		
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		POUT	f = 1kHz	$V_{PVDD} = 5.0V$		1.8		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$				$V_{PVDD} = 4.2V$		1.2		
				$V_{PVDD} = 3.7V$		0.96		
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$			1115111 = 170	$V_{PVDD} = 5.0V$		1.4		
Total Harmonic Distortion Plus Noise $ \begin{array}{c ccccccccccccccccccccccccccccccccccc$				$V_{PVDD} = 4.2V$		1		1
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$			$R_L = 8\Omega + 68\mu H$	$V_{PVDD} = 3.7V$		0.8]
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	Total Harmonic Distortion Plus	TUD: N	£ 41.11-	_		0.03	0.1	0/
Output Noise $V_{N} = A \text{-weighted (Note 4)} = A_{V} = 9 \text{dB} = 26$ $A_{V} = 6 \text{dB} = 23$ $A_{V} = 3 \text{dB} = 21$ $A_{V} = 0 \text{dB} = 19$ $A_{V} = 0 \text{dB} = 1$	Noise	THD+N	TIN = 1KHZ			0.03		- %
Output Noise $\begin{array}{c ccccccccccccccccccccccccccccccccccc$				$A_V = 12dB$		31		
$A_V = 3dB $				$A_V = 9dB$		26		μV _{RMS}
$A_V = 3dB $	Output Noise	V _N	A-weighted (Note 4)	$A_V = 6dB$		23		
Efficiency η $R_L = 8\Omega$, $P_{OUT} = 1.8W$, $f = 1kHz$ 93 % Oscillator Frequency f_{OSC} 300 kHz Spread-Spectrum Bandwidth 20 kHz Current Limit 2.8 A				$A_V = 3dB$		21		
Efficiency η $R_L = 8\Omega$, $P_{OUT} = 1.8W$, $f = 1kHz$ 93 % Oscillator Frequency f_{OSC} 300 kHz Spread-Spectrum Bandwidth 20 kHz Current Limit 2.8 A				$A_V = 0dB$		19		1
Oscillator Frequency fosc 300 kHz Spread-Spectrum Bandwidth 20 kHz Current Limit 2.8 A	Efficiency	η	$R_L = 8\Omega$, $P_{OLIT} = 1.8W$,					%
Spread-Spectrum Bandwidth 20 kHz Current Limit 2.8 A	Oscillator Frequency	+				300		kHz
Current Limit 2.8 A	Spread-Spectrum Bandwidth	300						
	Current Limit					2.8		
	Thermal Shutdown Level							°C

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{PVDD} = V_{\overline{SHDN}} = V_{GAIN} = 5V, \ V_{PGND} = 0V, \ A_{V} = 6dB \ (GAIN = PVDD), \ R_{L} = \infty, \ R_{L} \ connected \ between \ OUT+ \ to \ OUT-, \ AC \ measure-pvd \ AC \ meas$ ment bandwidth 20Hz to 22kHz, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25$ °C.) (Note 2, 3)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Thermal Hysteresis				15		°C
DIGITAL INPUT (SHDN)						
Input Voltage High	V _{INH}	V _{PVDD} = 2.5V to 5.5V	1.4			V
Input Voltage Low	V _{INL}	V _{PVDD} = 2.5V to 5.5V			0.4	V
Input Leakage Current		$T_A = +25$ °C			±1	μΑ

Note 2: All devices are 100% production tested at $T_A = +25^{\circ}C$. Specifications over temperature limits are guaranteed by design.

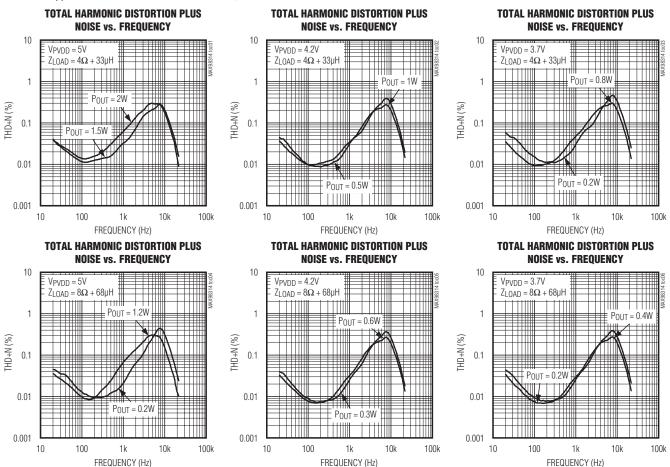
Note 3: Testing performed with a resistive load in series with an inductor to simulate an actual speaker load. For $R_1 = 4\Omega$, $L = 33\mu H$. For $R_1 = 8\Omega$, $L = 68\mu H$.

Note 4: Amplifier inputs AC-coupled to ground.

Note 5: Mode transitions controlled by SHDN control pin.

Typical Operating Characteristics

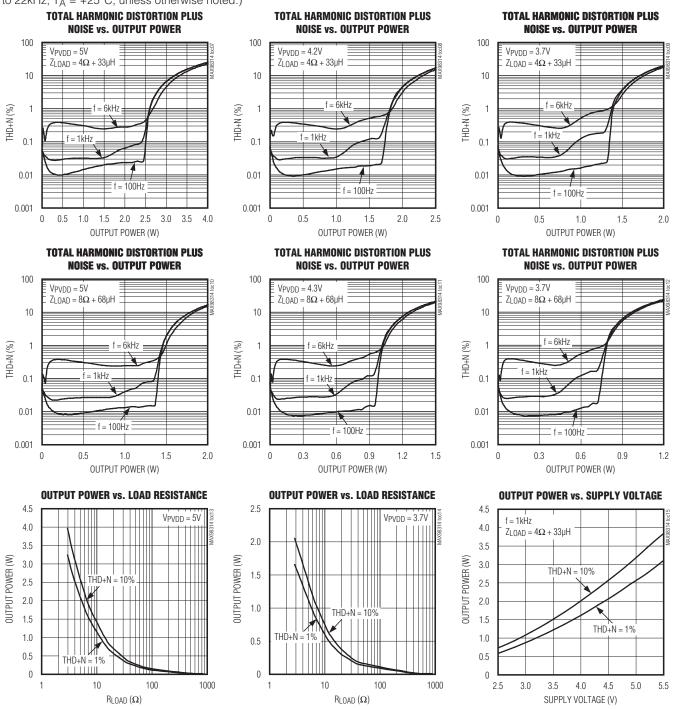
 $(V_{PVDD} = V_{\overline{SHDN}} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = \infty, R_L$ connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = +25^{\circ}C$, unless otherwise noted.)



Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Typical Operating Characteristics (continued)

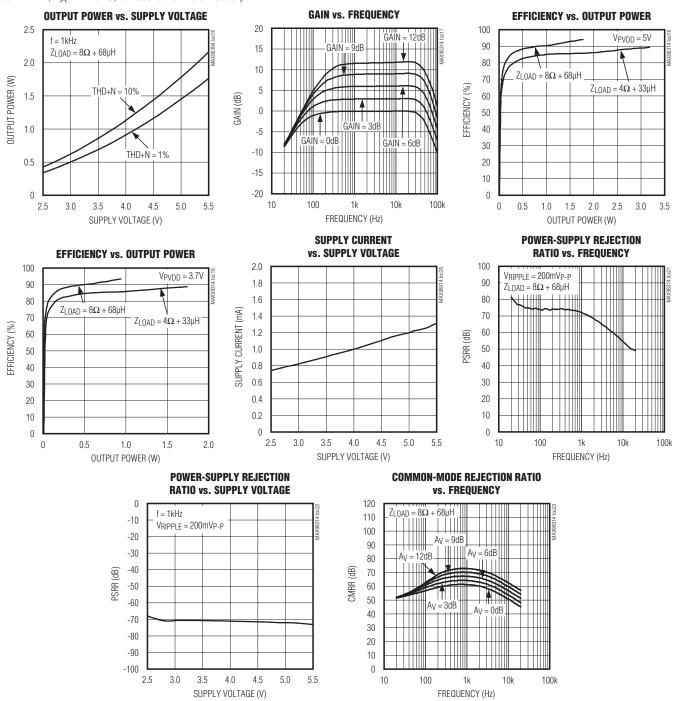
 $(V_{PVDD} = V_{\overline{SHDN}} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = \infty, R_L$ connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = +25$ °C, unless otherwise noted.)



Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Typical Operating Characteristics (continued)

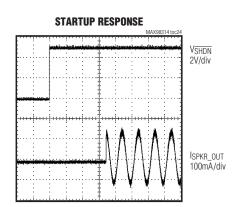
 $(V_{PVDD} = V_{\overline{SHDN}} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = \infty, R_L$ connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = +25$ °C, unless otherwise noted.)

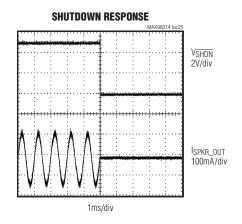


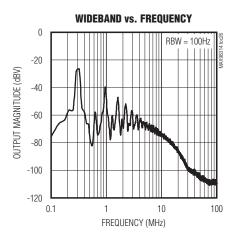
Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

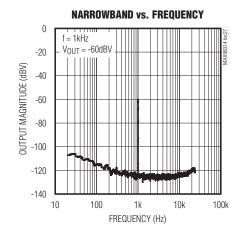
Typical Operating Characteristics (continued)

 $(V_{PVDD} = V_{\overline{SHDN}} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = \infty, R_L$ connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = +25$ °C, unless otherwise noted.)



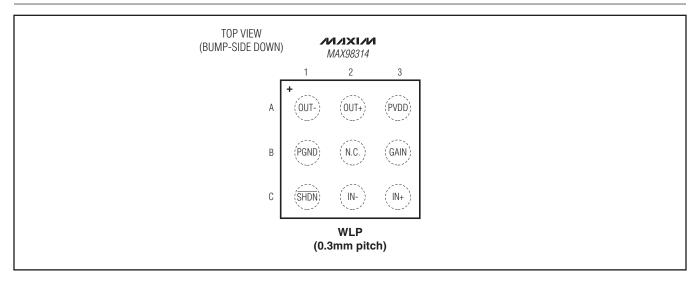






Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Pin Configuration



Pin Description

BUMP	NAME	FUNCTION
A1	OUT-	Negative Speaker Output
A2	OUT+	Positive Speaker Output
А3	PVDD	Power Supply. Bypass PVDD with a 0.1µF and 10µF capacitor to PGND.
B1	PGND	Power Ground
B2	N.C.	No Connection. Can be left unconnected or connected to PGND.
В3	GAIN	Gain Select. See Table 1 for GAIN settings.
C1	SHDN	Active-Low Shutdown Input. Drive SHDN low to place the device in shutdown.
C2	IN-	Inverting Audio Input
C3	IN+	Noninverting Audio Input

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Detailed Description

The MAX98314 features low quiescent current, a lowpower shutdown mode, comprehensive click-and-pop suppression, and excellent RF immunity.

The IC offers Class AB audio performance with Class D efficiency in a minimal board-space solution. The Class D amplifier features spread-spectrum modulation, edgerate, and overshoot control circuitry that offers significant improvements to switch-mode amplifier radiated emissions.

The amplifier features click-and-pop suppression that reduces audible transients on startup and shutdown. The amplifier additionally includes thermal overload and short-circuit protection.

Highly linear, integrated input coupling capacitors (C_{IN}) reduce solution size and provide excellent THD+N, PSRR, and CMRR performance at low frequencies vs. standard Class D amplifiers using external input capacitors.

Class D Speaker Amplifier

The IC's filterless Class D amplifier offers much higher efficiency than Class AB amplifiers. The high efficiency of a Class D amplifier is due to the switching operation of the output stage transistors. Any power loss associated with the Class D output stage is mostly due to the I²R loss of the MOSFET on-resistance and quiescent switching current overhead.

Ultra-Low EMI Filterless Output Stage

Traditional Class D amplifiers require the use of external LC filters, or shielding, to meet electromagnetic interference (EMI) regulation standards. Maxim's patented active emissions limiting edge-rate control circuitry and spread-spectrum modulation reduces EMI emissions, while maintaining up to 93% efficiency.

The spread-spectrum modulation mode flattens wideband spectral components, while proprietary techniques ensure that the cycle-to-cycle variation of the switching period does not degrade audio reproduction or efficiency. The IC's spread-spectrum modulator randomly varies the switching frequency by ±20kHz around the center frequency (300kHz). Above 10MHz, the wideband spectrum looks like noise for EMI purposes (Figure 1).

Amplifier Current Limit

If the output current of the speaker amplifier exceeds the current limit (2.8A typ), the IC disables the outputs for approximately 100µs. At the end of 100µs, the outputs are reenabled. If the fault condition still exists, the IC continues to disable and reenable the outputs until the fault condition is removed.

Selectable Amplifier Gain

The IC offers five programmable gain settings, selectable by a single gain input (GAIN).

Table 1. GAIN Selection

GAIN PIN	MAXIMUM GAIN (dB)
Connect to PGND	12
Connect to PGND through 100kΩ ±5%	9
Connect to PVDD	6
Connect to PVDD through 100kΩ ±5%	3
Unconnected	0

Integrated Input Coupling Capacitors (CIN)

The IC integrates two 0.011µF input coupling capacitors, CIN. The input coupling capacitors, in conjunction with the amplifier's internal input resistance (R_{IN}), form a firstorder highpass filter that removes the DC bias from the incoming signal. These capacitors allow the amplifier to bias the signal to an optimum DC level.

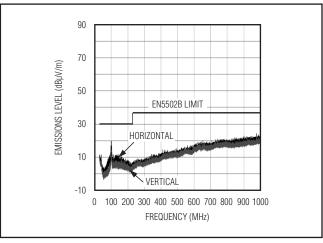


Figure 1. EMI Performance with 60cm of Speaker Cable, No Output Filter

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Assuming zero source impedance, the -3dB corner frequency, f-3dB, is:

$f_{-3dB} = 1/2\pi R_{IN}C_{IN}$ [Hz]

The 100ppm/V voltage coefficient of the integrated input coupling capacitor results in excellent low-frequency THD+N performance. Figure 2 illustrates the superior linearity of the IC's integrated input coupling capacitors compared to a similar amplifier with external 0.01µF X7R and X5R 0402 input coupling capacitors.

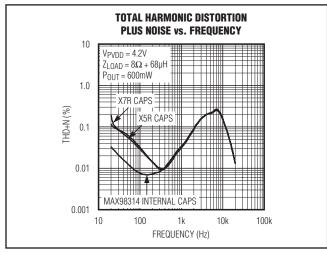


Figure 2. Low-Frequency THD+N Performance

Shutdown

The IC features a low-power shutdown mode, drawing < 0.1µA (typ) of supply current. Drive SHDN low to put the IC into shutdown.

Click-and-Pop Suppression

The speaker amplifier features Maxim's comprehensive click-and-pop suppression. During startup, the clickand-pop suppression circuitry reduces any audible transient sources internal to the device. When entering shutdown, the differential speaker outputs ramp down to PGND quickly and simultaneously.

Applications Information

Filterless Class D Operation

Traditional Class D amplifiers require an output filter. The filter adds cost and size, and decreases efficiency and THD+N performance. The IC's filterless modulation scheme does not require an output filter.

Because the switching frequency of the IC is well beyond the bandwidth of most speakers, voice coil movement due to the switching frequency is very small. Use a speaker with a series inductance > 10 μ H. Typical 8 Ω speakers exhibit series inductances in the 20µH to 100µH range.

Speaker Amplifier Power-Supply Input (PVDD)

PVDD powers the speaker amplifier and ranges from 2.5V to 5.5V. Bypass PVDD with a 0.1µF and 10µF capacitor to PGND. Apply additional bulk capacitance at the device if long input traces between PVDD and the power source are used.

Layout and Grounding

Proper layout and grounding are essential for optimum performance. Good grounding improves audio performance and prevents switching noise from coupling into the audio signal.

Use wide, low-resistance output traces. As the load impedance decreases, the current drawn from the device increases. At higher current, the resistance of the output traces decrease the power delivered to the load. For example, if 2W is delivered from the device output to a 4Ω load through $100m\Omega$ of total speaker trace, 1.904W is delivered to the speaker. If power is delivered through $10m\Omega$ of total speaker trace, 1.99W is delivered to the speaker. Wide output, supply, and ground traces also improve the power dissipation of the device.

The IC is inherently designed for excellent RF immunity. For best performance, add ground fills around all signal traces on top or bottom PCB layers.

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

WLP Applications Information

For the latest application details on WLP construction, dimensions, tape carrier information, PCB techniques, bump-pad layout, and recommended reflow temperature profile, as well as the latest information on reliability testing results, refer to Application Note 1891: Wafer-Level Packaging (WLP) and Its Applications. Figure 3 shows the dimensions of the WLP balls used on the IC.

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX98314EWL+	-40°C to +85°C	9 WLP

⁺Denotes a lead(Pb)-free/RoHS-compliant package.

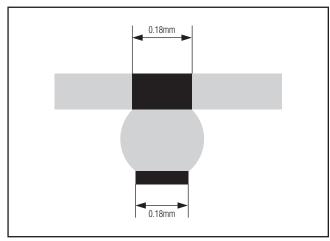


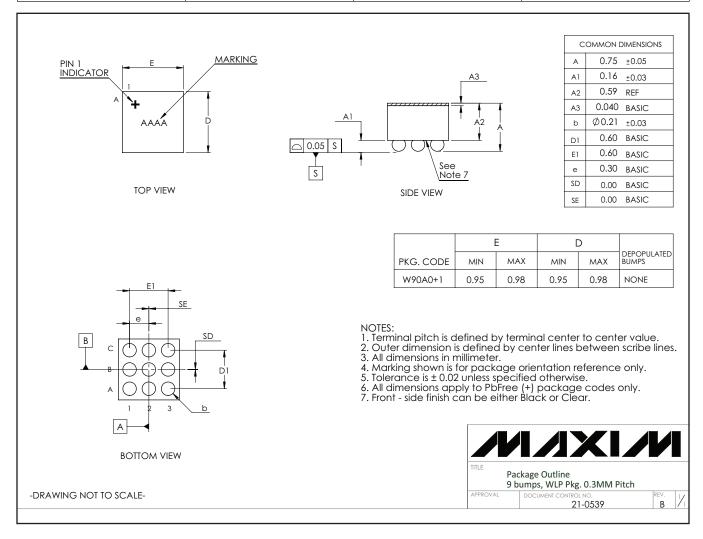
Figure 3. WLP Ball Dimensions

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Package Information

For the latest package outline information and land patterns (footprints), go to www.maxim-ic.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
9 WLP (0.3mm pitch)	W90A0+1	<u>21-0539</u>	Refer to Application Note 1891



Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	11/11	Initial release	_

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.

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IS31AP4996-GRLS2-TR STPA002OD-4WX NCP2823BFCT1G MAX9717DETA+T MAX9717CETA+T MAX9724AEBC+TG45

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NCP2823AFCT2G NCS2211MNTXG CPA2233CQ16-A1 OPA1604AIPWR OPA1612AQDRQ1 TDA7492 SSM2519ACBZ-R7

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