



USB Power Controller

General Description

The MIC2010 is a dual channel USB power switch designed to support the power distribution requirements for USB Wakeup from the ACPI S3 state. The MIC2010 will directly switch its two outputs between a 5V main supply and a 5V auxiliary supply normally provided in ATX style power supplies.

The MIC2010 will adjust its current-limit threshold according to the ACPI state it is in. In the normal active ACPI S0 state the current-limit is set at 500mA minimum per channel satisfying the USB continuous output current specification. In the S3 state the current-limit can be reduced to a value determined by an external resistor to minimize the current that is supplied by the auxiliary supply, thereby ensuring that voltage regulation is maintained even during fault conditions.

The MIC2010 provides make-before-break switching to ensure glitch-free transitions between the S3 and S0 states. Each channel is also thermally isolated from the other so that a fault in one channel does not effect the other. FAULT status output signals are also provided indicating overcurrent and thermal shutdown conditions.

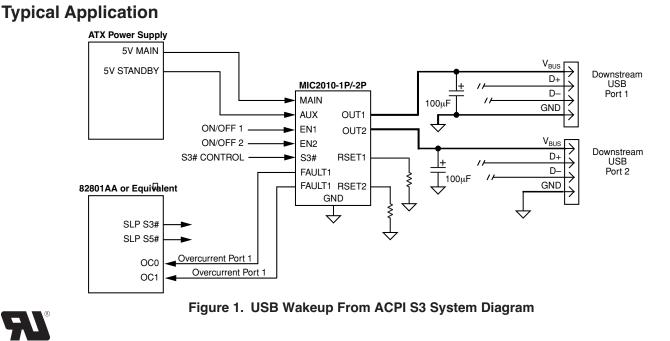
The MIC2070 option latches the output off upon detecting an overcurrent condition lasting for more than 5ms minimum. The output can be reset by either toggling the EN inputs of the MIC2070-1/-2 or by removing the load. Latching the output off provides a circuit breaker mode of operation which reduces power consumption during fault conditions.

Features

- · Compliant to USB power distribution specifications
- Two completely independent switches
- Integrated switching matrix supports ACPI S0/S3 state transitions without external FET circuits
- Make-before-break switching ensures glitch-free transitions
- No back-feed of auxiliary supply onto main supply during standby mode
- Bi-level current-limit preserves auxiliary supply voltage regulation in standby mode
- Thermally isolated channels
- Thermal shutdown protection
- Fault status outputs with filter prevents false assertions during hot-plug events
- · Circuit breaker options with auto-reset (MIC2070)
- Undervoltage lockout
- UL File #179633

Applications

- Desktop PCs
- Notebook PCs
- Notebook docking stations
- LAN Servers
- PC motherboard



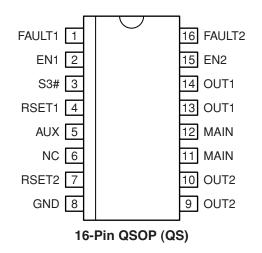
UL Recognized Component

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Ordering Information

Part Number			Circuit Breaker	Temperature		
Standard	Pb-Free	Enable	Function	Range	Package	
MIC2010-1PCQS	MIC2010-1PZQS	Active High		0°C to +70°C	16-lead QSOP	
MIC2010-2PCQS	MIC2010-1PZQS	Active Low		0°C to +70°C	16-lead QSOP	
MIC2070-1PCQS	MIC2070-1PZQS	Active High	✓	0°C to +70°C	16-lead QSOP	
MIC2070-2PCQS	MIC2070-2PZQS	Active Low	1	0°C to +70°C	16-lead QSOP	

Pin Configuration



Pin Description

Pin Number	Pin Name	Pin Function				
1	FAULT1	Fault Status (Output): Channel 1, active-low; weak pull-up to AUX. FAULT1 is asserted LOW when channel 1 is in a thermal shutdown state or overcurrent condition for more than 5ms. MIC2070 latches this output in its asserted state upon overcurrent condition. Toggling EN1 or removing the load will reset the circuit breaker latch, and deassert FAULT1.				
2	EN1	Enable (Input): Channel 1, active-high (-1) or active-low (-2). Toggling this input also resets the latched output of the MIC2070.				
3	S3#	Control (Input): When this input is high, the MAIN inputs are connected to OUT1 and OUT2 via 100m Ω , 500mA MOSFET switches. When this input is LOW, the AUX inputs are connected to OUT1 and OUT2 via 500m Ω MOSFET switches with a current-limit threshold specified by external resistors, RSET1 and RSET2.				
4	RSET1	Current-Limt Set Resistor (Input): Channel 1. A resistor connected to this input sets the current-limit threshold in AUX mode (S3# asserted). The current-limit threshold is determined by approximately 18/RSET1.				
5	AUX	Auxiliary 5V Supply (Input): Also used as power supply for internal circuitry.				
6	NC	No Connection: This pin may be connected to other pins without restriction.				
7	RSET2	Current-Limit Set Resistor (Input): Channel 2. A resistor connected to this input sets the current-limit threshold in AUX mode (S3# asserted). The current-limit threshold is determined by approximately 18/RSET2.				
8	GND	Ground				
9, 10	OUT2	Channel 2 (Output): Both pins must be connected together externally.				
11, 12	MAIN	5V Main Supply (Input): All MAIN inputs must be connected together externally.				
13, 14	OUT1	Channel 1 (Output): Both pins must be connected together externally.				
15	EN2	Enable (Input): Channel 2, active-high (-1) or active-low (-2). Toggling this input also resets the latched output for the MIC2070.				
16	FAULT2	Fault Status (Output): Channel 2, active-low; weak pull-up to AUX. FAULT2 is asserted LOW when channel 2 is in a thermal shutdown state or overcur- rent condition for more than 5ms. MIC2070 latches this output in it's asserted state upon overcurrent condition. Toggling EN2 or removing load will reset the circuit breaker latch, and deassert FAULT2.				

Absolute Maximum Ratings (1, 4)

Supply Voltage (VIN)	–0.3V to 6V
FAULT#, OUT1, OUT2 Output Pins	–0.3V to 6V
FAULT1, FAULT2, Output Current	25mA
ESD Rating ⁽³⁾	2kV

Operating Ratings⁽²⁾

Supply Voltage (V _{MAIN} , V _{AUX})	+4.5V to +5.5V
Continuous Output Current (AUX I	Mode) 50mA to 300mA
Ambient Temperature (T _A)	–0°C to +70°C
Junction Temperature (T _J)	Internally Limited
Package Thermal Resistance	
$QSOP(\theta_{JA})$	163°C/W

Electrical Characteristics

 $V_{MAIN} = 5V$; AUX = 5V; $R_{SET} = 125\Omega$, $T_A = 25^{\circ}C$.

Symbol	Parameter	Condition	Min	Тур	Max	Units
V _{MAIN}	MAIN Supply Voltage		4.5	5.0	5.5	V
I _{MAIN}	MAIN Supply Current ⁽⁵⁾	S3# = 1, both switches ON, no load S3# = 1, both switches OFF, no load		16	22 5	μΑ μΑ
I _{LEAK}	MAIN Reverse Leakage Current ⁽⁵⁾	S3# = 0, both switches ON, $V_{MAIN} = 0V$	-10		+10	μA
V _{AUX}	AUX Supply Voltage		4.5	5.0	5.5	V
I _{AUX}	AUX Supply Current ⁽⁵⁾	S3# = 0; No load, both switches ON S3# = 0; No load, both switches OFF		.6	1 5	mA μA
V _{UV/AUX}	AUX Undervoltage Lockout Threshold	V _{AUX} increasing V _{AUX} decreasing	3.5 3.3	3.7 3.5	4.0 3.8	V V
V _{HYS}	AUX Undervoltage Lockout Hysteresis			200		mV
R _{DSMAIN}	MAIN On-Resistance, Each Output	S3# = 1, I _{OUT} = 500mA		100	140	mΩ
R _{DSAUX}	AUX On-Resistance, Each Output	S3# = 0, I _{OUT} = 100mA		500	700	mΩ
I _{LIMIT}	MAIN Current-Limit Threshold,	S3# = 1, V _{OUT} = 4.0V, ramped load	0.5		1.25	A
	MAIN Short-Circuit Current-Limit	V _{OUT} = 0V	UT = 0V 0.5		1.25	A
	Current-Limit Factor (AUX Supply) ⁽⁶⁾	S3# = 0, I _{OUT} = 50mA to 300mA V _{OUT} = 4V	14.4	18	21.6	A•Ω
V _{TH}	S3#, EN1, EN2 Input Threshold Voltage	High-to-Low transition Low-to-High transition	.8	1.5 1.7	2.0	V V
I _{IN}	S3#, EN1, EN2 Input Current	V _{S3/EN} =5V, 0V	-1		1	μΑ
V _{HYS}	EN1, EN2 and S3# Input Hysteresis			200		mV
I _{OFF}	OUT1, 2 Leakage Current	Outputs are off, V _{OUT} = 0	-10		10	μA
	Pull-Up Current During Latched Output State (MIC2070-x only)	Outputs latched off	TBD	1	5 +10 5.5 1 5 4.0 3.8 140 700 1.25 1.25 21.6 2.0 1	mA
V _{TH LATCH}	Latch Reset Threshold	V _{OUT} rising		1.95		V
	Minimum Output Slew Rate to Reset Latch (MIC2070-x only) ⁽⁷⁾	Output rising		.4	5 +10 5.5 1 5 4.0 3.8 140 700 1.25 1.25 21.6 2.0 1 1 10 TBD	V/s
	Overtemperature Threshold	T _J increasing, single channel T _J decreasing, single channel T _J increasing, both channels T _J decreasing, both channels		140 120 160 125		0° 0° 0° 0°
V _{OL}	FAULT1, 2 Output Low Voltage	I _{FAULT} = 5mA			0.2	V
	FAULT1, 2 Output Off Current (Not Applicable for 'P' versions)	V _{FAULT} = 5V		0.2	10	μΑ

Symbol	Parameter	Condition	Min	Тур	Max	Units
V _{OH}	FAULT1, 2 Output High Voltage (MIC2010-1P, 2P), (MIC2070-1P, 2P)	I _{FAULT} = -20μA	4			V
T _H	MAIN to S3# Hold Time, Note 7	Figure 5	5			ms
Τ _S	MAIN to S3# Set-up Time, Note 7	Figure 5	0			ms
t _{DLY}	FAULT Delay Filter Response Time (Overcurrent only), Note 8	Output shorted to ground, Figure 4	5	10	20	ms
t _{oc}	Overcurrent Response Time	Output shorted to ground, Figure 4 MAIN output AUX output		2 2		μs μs
t _{ON(MAIN)}	MAIN Output Turn-On Time	$R_L = 10\Omega$, $C_L = 1\mu F$, Figure 3		2		ms
t _{OFF(MAIN)}	MAIN Output Turn-Off Time	$R_L = 10\Omega$, $C_L = 1\mu F$, Figure 3		35		μS
t _{r(MAIN)}	MAIN Output Rise Time	$R_L = 10\Omega$, $C_L = 1\mu F$, Figure 3		2		ms
t _{f(MAIN)}	MAIN Output Fall Time	$R_L = 10\Omega$, $C_L = 1\mu F$, Figure 3		32		μS
t _{ON(AUX)}	AUX Output Turn-On Time	$R_L = 50\Omega$, $C_L = 1\mu F$, Figure 3		0.6		ms
t _{OFF(AUX)}	AUX Output Turn-Off Time	$R_L = 50\Omega$, $C_L = 1\mu F$, Figure 3		120		μS
t _{r(AUX)}	AUX Output Rise Time	$R_L = 50\Omega$, $C_L = 1\mu F$, Figure 3		0.5		ms
t _{f(AUX)}	AUX Output Fall Time	$R_L = 50\Omega$, $C_L = 1\mu F$, Figure 3		115		μS
t _{XMA}	MAIN to AUX Cross Conduction Time, Note 9	S3# transition to 0		5	7.5	ms
t _{XAM}	AUX to MAIN Cross Conduction Time, Note 9	S3# transition to 1		5	7.5	ms

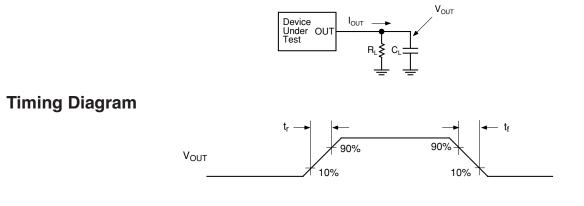
Notes:

1. Exceeding the absolute maximum rating may damage the device.

2. The device is not guaranteed to function outside its operating rating.

- 3. Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5k in series with 100pF.
- 4. All voltages are referenced to ground.
- 5. For MIC2070-1(P) OFF occurs when $V_{EN} < 0.8V$ and ON occurs when $V_{EN} > 2.4V$. For MIC2070-2(P) OFF occurs when $V_{EN} > 2.4V$ and ON occurs when $V_{EN} < 0.8V$.
- 6. Current-limit threshold is defined by the current-limit factor divided by R_{SET}.
- 7. Guaranteed by design. Not production tested.
- Assumes only one channel in current-limit. Delay circuitry is shared among channels so it is possible for t_{DLY} to be 40ms max if one channel enters current-limit as the other is about to time-out.
- 9. Cross conduction time is the duration in which both MAIN and AUX internal switches are on subsequent to S3# transitioning.

Test Circuit



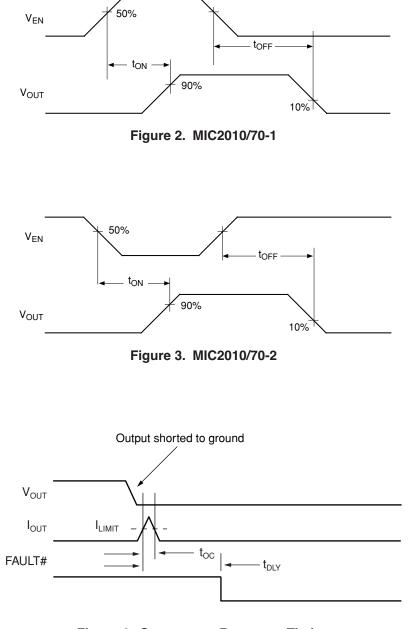


Figure 4. Overcurrent Response Timing

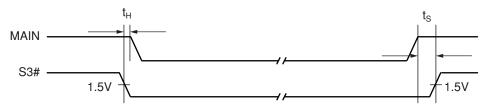
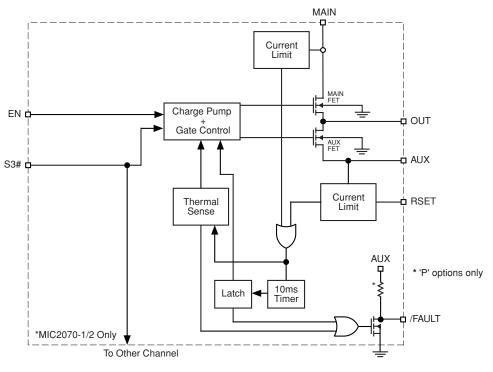


Figure 5. MAIN to S3# Timing

Functional Diagram



Functional Description

The MIC2010/2070 are designed to support the power distribution requirements for USB wakeup from the ACPI S3 state. It integrates two independent channels under control of input S3#. When S3# is asserted LOW (ACPI S3 state) the MIC2010/2070 will switch a 500m Ω MOSFET switch from the AUX input to each of its two outputs. In addition the current-limit threshold will be set to a value specified by a resistor connected to the RSET inputs. Conversely when the S3# input is HIGH the MIC2010/2070 will switch a 100m Ω MOSFET switch from the MAIN inputs to each of its two outputs. The current-limit threshold is preset to 500mA in this state. The lower current limit during the ACPI S3 state helps to ensure that the standby supply maintains regulation even during fault conditions.

Thermal Shutdown

Thermal shutdown is employed to protect the device from damage should the die temperature exceed safe margins due mainly to short circuit faults. Thermal shutdown shuts off the output MOSFET and asserts the FAULT output if the die temperature reaches 140°C and the overheated channel is in current limit. The other channel is not affected. If however, the die temperature exceeds 160°C, both channels will be shut off even if neither channel is in current limit.

Power Dissipation

The device's junction temperature depends on several factors such as the load, PCB layout, ambient temperature and package type. The power dissipated in each channel is

 $P_D = R_{DS(on)} \times I_{OUT}^2$ where $R_{DS(on)}$ is the on-resistance of the internal MOSFETs and I_{OUT} is the continuous output current.

Total power dissipation of the device will be the summation of P_D for both channels. To relate this to junction temperature, the following equation can be used:

$$TJ = P_{D} \times \theta_{JA} + T_{A}$$

where:

 $T_J = junction temperature$

T_A = ambient temperature

 θ_{JA} = is the thermal resistance of the package

Current Sensing and Limiting

The current-limit threshold of each channel is preset internally at 500mA when S3# is deasserted. When S3# is asserted the current-limit threshold is specified by a resistor connected to the RSET input. The value of the current-limit threshold is determined by the equation $18/R_{SET}$ ohms where R_{SET} is the resistance connected between the RSET pin and ground. The current-limit threshold should be set at 1.2X of the applications continous output current requirement.

When an overcurrent condition lasts longer than t_{DLY} the MIC2070 will activate an internal circuit breaker that will latch the output off and assert FAULT. The output will remain off until either the load is removed or EN is toggled. When the MIC2070 enters a latched output condition a 1mA pull-up current source is activated. This provides a way to automatically reset the output once the load is removed without the need to toggle the enable input. Please refer to Figure 7 for timing details.

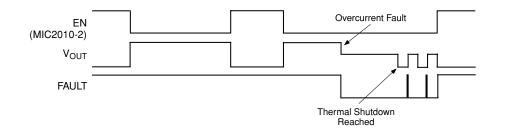
The MIC2010 will automatically reset its output when the die temperature cools down to 120°C. The MIC2010 output and FAULT signal will continue to cycle on and off until the device is disabled or the fault is removed. Figure 6 depicts typical timing. Depending on PCB layout, package, ambient temperature, etc., it may take several hundred milliseconds from the incidence of the fault to the output MOSFET being shut off. This time duration will be shortest in the case of a dead short on the output.

Fault Status Output

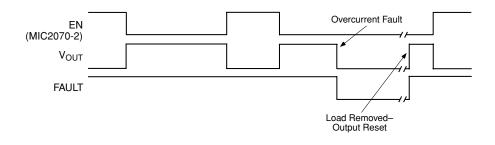
The FAULT signal is an active-low output with a weak pullup. FAULT is asserted when either an overcurrent or thermal shutdown condition occurs. In the case of an overcurrent condition, FAULT will be asserted only after the flag response delay time, t_{DLY} , has elapsed. This ensures that FAULT is asserted only upon valid overcurrent conditions and that erroneous error reporting is eliminated. The FAULT response delay time t_{DLY} is typically 10ms. Since the delay timer is shared between both channels it is possible for t_{DLY} to be extended by an amount equal to the difference between the occurrence of the over-current event in both channels.

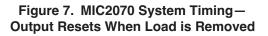
Undervoltage Lockout

Undervoltage lockout (UVLO) prevents the output MOSFET from turning on until the AUX input exceeds approximately 3.5V. UVLO ensures that the output MOSFETs remain off to prevent high transient inrush current due to stray or bulk load capacitance. This helps to ensure that the power supply voltage regulation is preserved and also prevents possible damage to sensitive components.

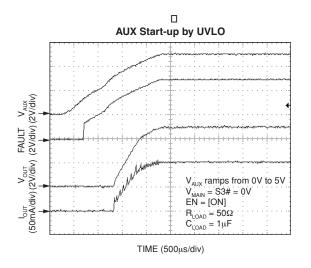


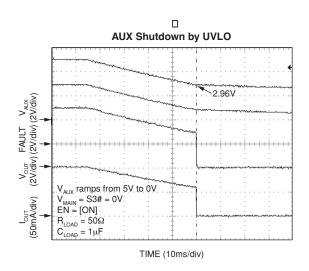


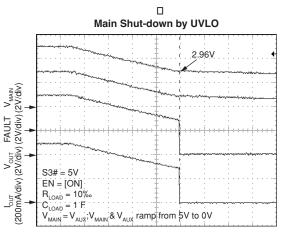




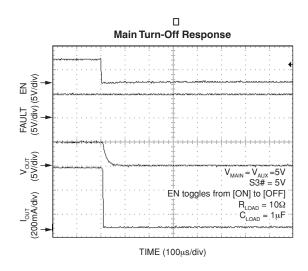
Functional Characteristics



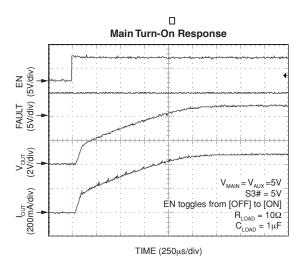


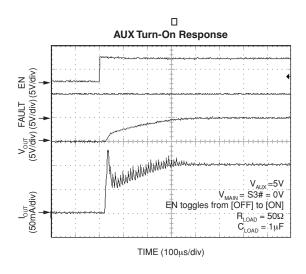


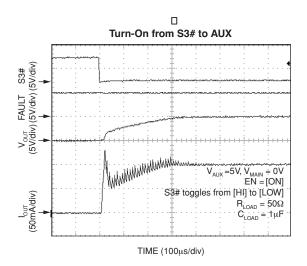


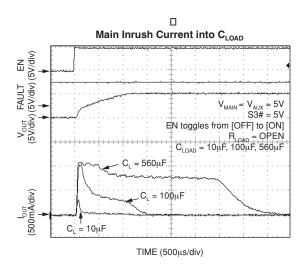


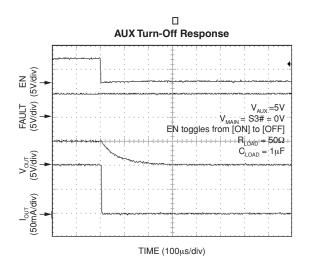
 \square **Main Start-up by UVLO** $V_{MAIN} = V_{AUX}, V_{MAIN} & V_{AUX} ramp from 0V to 5V$ S3# = 5V EN = [ON] $R_{LOAD} = 10\Omega$ $C_{LOAD} = 1\mu F$ $TIME (500\mu s/div)$

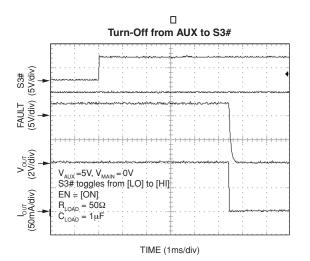


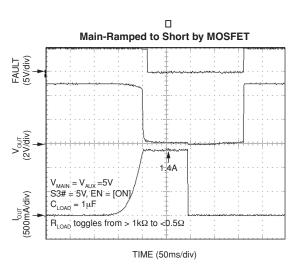


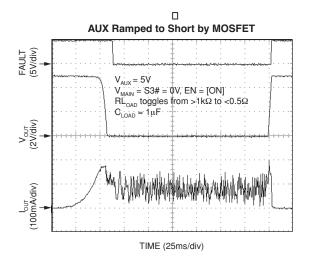


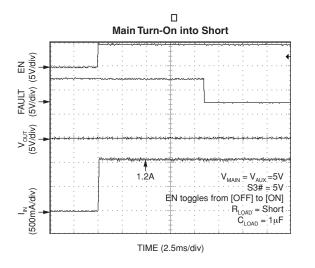


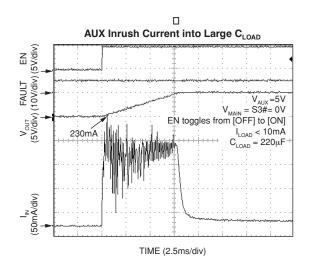


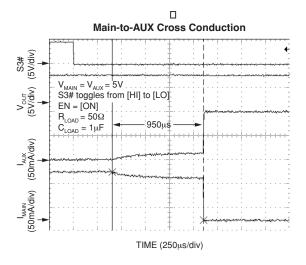


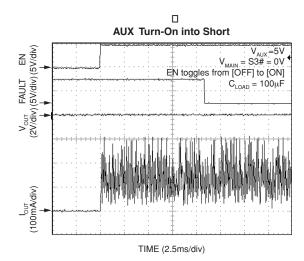


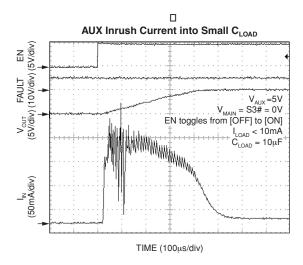


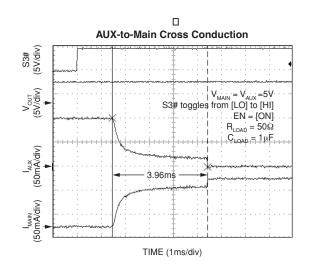




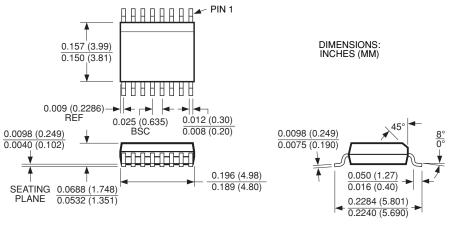








Package Information



16-Pin QSOP (QS)

MICREL, INC. 2180 FORTUNE DRIVE SAN JOSE, CA 95131 USA

TEL + 1 (408) 944-0800 FAX + 1 (408) 474-1000 WEB http://www.micrel.com

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