## LAN8810/LAN8810i

# GMII 10/100/1000 Ethernet Transceiver with HP Auto-MDIX Support 

## Highlights

- Single-Chip Ethernet Physical Layer Transceiver (PHY)
- Compliant with IEEE 802.3ab (1000BASE-T), IEEE 802.3u (Fast Ethernet), and ISO 802-3/IEEE 802.3 (10BASE-T)
- HP Auto-MDIX support in accordance with IEEE 802.3ab specification at 10/100/1000 Mbps operation
- Small footprint 72-pin QFN lead-free RoHS compliant package with GMII ( $10 \times 10 \times 0.9 \mathrm{~mm}$ height)
- Flexible configurations for LED status indicators
- Implements Reduced Power Operating Modes


## Target Applications

- Set-Top Boxes
- Networked Printers and Servers
- Test Instrumentation
- LAN on Motherboard
- Embedded Telecom Applications
- Video Record/Playback Systems
- Cable Modems/Routers
- DSL Modems/Routers
- Digital Video Recorders
- IP and Video Phones
- Wireless Access Points
- Digital Televisions
- Digital Media Adapters/Servers
- Gaming Consoles
- POE Applications


## Key Benefits

- High-Performance 10/100/1000 Ethernet Transceiver
- Compliant with IEEE 802.3ab (1000BASE-T)
- Compliant with IEEE 802.3/802.3u (Fast Ethernet)
- Compliant with ISO 802-3/IEEE 802.3 (10BASE-T)
- 10BASE-T, 100BASE-TX and 1000BASE-T support
- Loop-back modes
- Auto-negotiation (NEXT page support)
- Automatic polarity detection and correction
- Link status change wake-up detection
- Vendor specific register functions
- Supports GMII interface
- Controlled impedance outputs
- Four status LED outputs and configurable LED modes with support for tricolor operation
- Compliant with IEEE 802.3-2005 standards - GMII pins tolerant to 3.6 V
- Integrated DSP implements adaptive equalizer, echo cancelers, and crosstalk cancelers
- Efficient digital baseline wander correction
- Power and I/Os
- Configurable LED outputs
- Various low power modes
- Variable voltage I/O supply (2.5V/3.3V)
- Miscellaneous Features
- IEEE 1149.1 (JTAG) boundary scan
- Multiple clock options - 25MHz crystal or 25 MHz single-ended clock
- Packaging
- 72-pin QFN (10x10 mm) RoHS compliant package with GMII
- Environmental
- Commercial temperature range $\left(0^{\circ} \mathrm{C}\right.$ to $\left.+70^{\circ} \mathrm{C}\right)$
- Industrial temperature range $\left(-40^{\circ} \mathrm{C}\right.$ to $+85^{\circ} \mathrm{C}$ )


## LAN8810/LAN8810i

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## LAN8810/LAN8810i

### 1.0 INTRODUCTION

The LAN8810/LAN8810i is a low-power 10BASE-T/100BASE-TX/1000BASE-T Gigabit Ethernet physical layer (PHY) transceiver with variable I/O voltage that is fully compliant with the IEEE 802.3 and 802.3 ab standards.
The LAN8810/LAN8810i can be configured to communicate with an Ethernet MAC via the standard MII(IEEE 802.3u)/ GMII(IEEE 802.3z) interfaces. It contains a full-duplex transceiver for 1000 Mbps operation on four pairs of category 5 or better balanced twisted pair cable. Per IEEE 802.3-2005 standards, all digital interface pins are tolerant to 3.6V.
The LAN8810/LAN8810i is configurable via hardware and software, supporting both IEEE 802.3-2005 compliant and vendor-specific register functions via SMI. The LAN8810/LAN8810i implements Auto-Negotiation to automatically determine the best possible speed and duplex mode of operation. HP Auto-MDIX support allows the use of direct connect or crossover cables.
An internal block diagram of the LAN8810/LAN8810i is shown in Figure 1-1. A typical system-level diagram is shown in Figure 1-2.

FIGURE 1-1: INTERNAL BLOCK DIAGRAM


FIGURE 1-2: SYSTEM LEVEL BLOCK DIAGRAM


### 2.0 PIN DESCRIPTION AND CONFIGURATION

FIGURE 2-1: 72-QFN PIN ASSIGNMENTS (TOP VIEW)


## LAN8810/LAN8810i

## TABLE 2-1: GMII INTERFACE PINS

| Num <br> Pins | Name | Symbols | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :--- |
| 1 | Transmit Data 0 | TXD0 | VIS <br> (PD) | The MAC transmits data to the PHY using this <br> signal. |
| 1 | Transmit Data 1 | TXD1 | VIS <br> (PD) | The MAC transmits data to the PHY using this <br> signal. |
| 1 | Transmit Data 2 | TXD2 | VIS <br> (PD) | The MAC transmits data to the PHY using this <br> signal. |
| 1 | Transmit Data 3 | TXD3 | VIS <br> (PD) | The MAC transmits data to the PHY using this <br> signal. |
| 1 | Transmit Data 4 | TXD4 | VIS <br> (PD) | The MAC transmits data to the PHY using this <br> signal. |
| 1 | Transmit Data 6 | TXD6 | Receive Data 5 | RXD5 |

## TABLE 2-1: GMII INTERFACE PINS (CONTINUED)

| $\begin{array}{c}\text { Num } \\ \text { Pins }\end{array}$ | Name | Symbols | $\begin{array}{c}\text { Buffer } \\ \text { Type }\end{array}$ | Description |
| :---: | :---: | :---: | :---: | :--- |
| 1 | Receive Data 6 | RXD6 | VO6 | $\begin{array}{l}\text { The PHY transfers data to the MAC using this } \\ \text { signal. }\end{array}$ |
| 1 | Receive Data 7 | RXD7 | VO6 | $\begin{array}{l}\text { The PHY transfers data to the MAC using this } \\ \text { signal. }\end{array}$ |
| 1 | $\begin{array}{c}\text { Receive Data } \\ \text { Valid }\end{array}$ | RXDV | VO6 | $\begin{array}{l}\text { Indicates that recovered and decoded data is being } \\ \text { presented on the receive data pins. }\end{array}$ |
| 1 | Receive Error | RXER | VO6 | $\begin{array}{l}\text { Asserted to indicate an error has been detected in } \\ \text { the frame presently being transferred from the } \\ \text { PHY. }\end{array}$ |
| 1 | Receive Clock | RXCLK | VO6 | $\begin{array}{l}\text { Used to transfer data to the MAC. } \\ \text { GMII (1000BASE-T): 125 MHz }\end{array}$ |
| 1 | Collision Detect | COL | VO6 | $\begin{array}{l}\text { MII (100BASE-TX): 25 MHz } \\ \text { MII (10BASE-T): 2.5 MHz }\end{array}$ |
| condition. (used in half-duplex mode only) |  |  |  |  |$]$

Note 2-1 Configuration strap values are latched on hardware reset. Configuration straps are identified by an underlined symbol name. Signals that function as configuration straps must be augmented with an external resistor when connected to a load. Refer to Section 3.8, "Configuration," on page 22 for additional information.

## TABLE 2-2: SERIAL MANAGEMENT INTERFACE (SMI) PINS

| Num <br> Pins | Name | Symbols | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :--- |
| 1 | SMI Clock | MDC | VIS <br> (PD) | Serial Management Interface clock. |
| 1 | SMI Data Input/ <br> Output | MDIO | VIS/VO8 <br> (PU) | Serial Management Interface data input/output. |

## TABLE 2-3: LED \& CONFIGURATION PINS

| Num Pins | Name | Symbols | Buffer Type | Description |
| :---: | :---: | :---: | :---: | :---: |
| 1 | 10BASE-T Link LED Indicator | 10_LED | VO8 | 10BASE-T LED link indication. Refer to Section 3.9.1, "LEDs," on page 26 for additional information. |
|  | $\begin{aligned} & \text { 100BASE-TX } \\ & \text { Link LED } \\ & \text { Indicator } \end{aligned}$ | 100_LED | VO8 | 100BASE-TX LED link indication. Refer to Section 3.9.1, "LEDs," on page 26 for additional information. |
| 1 | Hardware Power Down (HPD) Mode Configuration Strap | HPD MODE | VIS (PD) | This configuration strap is used to select the Hardware Power Down (HPD) mode. When pulledup, the PLL is not disabled when HPD is asserted. When pulled-down, the PLL is disabled when HPD is asserted. <br> Refer to Section 3.7.3, "Hardware Power-Down," on page 22 for additional information. <br> See Note 2-2 for more information on configuration straps. |
| 1 | 1000BASE-T Link LED Indicator | 1000_LED | VO8 | 1000BASE-T LED link indication. Refer to Section 3.9.1, "LEDs," on page 26 for additional information. |
| 1 | Link Activity LED Indicator | ACT_LED | VO8 | Link activity LED indication. Refer to Section 3.9.1, "LEDs," on page 26 for additional information. |
| 1 | Configuration Input 0 | CONFIG0 | VIS (PD) | This pin sets the PHYADD[1:0] bits of the 10/100 Special Modes Register on reset or power-up. It must be connected to VSS, 100_LED, 1000_LED, or VDDVARIO. Refer to Section 3.8.1.2, "CONFIG[3:0] Configuration Pins," on page 23 for additional information. |
| 1 | Configuration Input 1 | CONFIG1 | VIS (PD) | This pin sets the PAUSE bit of the Auto Negotiation Advertisement Register and PHYADD [2] bit of the 10/100 Special Modes Register on reset or powerup. It must be connected to VSS, 100_LED, 1000_LED, or VDDVARIO. Refer to Seection 3.8.1.2, "CONFIG[3:0] Configuration Pins," on page 23 for additional information. |
| 1 | Configuration Input 2 | CONFIG2 | VIS (PD) | This pin sets the MOD[1:0] bits of the Extended Mode Control/Status Register on reset or powerup. It must be connected to VSS, 100_LED, 1000 LED, or VDDVARIO. Refer to Seection 3.8.1.2, "CONFIG[3:0] Configuration Pins," on page 23 for additional information. |
| 1 | Configuration Input 3 | CONFIG3 | VIS (PD) | This pin sets the MOD[3] bit of the Extended Mode Control/Status Register on reset or power-up. It must be connected to 1000 LED or VDDVARIO. Refer to Section 3.8.1.2, "CŌNFIG[3:0] Configuration Pins," on page 23 for additional information. |

Note 2-2 Configuration strap values are latched on hardware reset. Configuration straps are identified by an underlined symbol name. Signals that function as configuration straps must be augmented with an external resistor when connected to a load. Refer to Section 3.8, "Configuration," on page 22 for additional information.

## TABLE 2-4: ETHERNET PINS

| Num Pins | Name | Symbol | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :--- |
| 1 | Ethernet TX/ <br> RX Positive <br> Channel 0 | TROP | AIO | Transmit/Receive Positive Channel 0. |
| 1 | Ethernet TX/ <br> RX Negative <br> Channel 0 | TR0N | AIO | Transmit/Receive Negative Channel 0. |
| 1 | Ethernet TX/ <br> RX Positive <br> Channel 1 | TR1P | AIO | Transmit/Receive Positive Channel 1. |
| 1 | Ethernet TX/ <br> RX Negative <br> Channel 1 | TR1N | AIO | Transmit/Receive Negative Channel 1. |
| 1 | Ethernet TX/ <br> RX Positive <br> Channel 2 | TR2P | AIO | Transmit/Receive Positive Channel 2. |
| 1 | Ethernet TX/ <br> RX Negative <br> Channel 2 | TR2N | AIO | Transmit/Receive Negative Channel 2. |
| 1 | Ethernet TXI <br> RX Positive <br> Channel 3 | TR3P | AIO | Transmit/Receive Positive Channel 3. |
| 1 | Ethernet TX/ <br> RX Negative <br> Channel 3 | TR3N | AIO | Transmit/Receive Negative Channel 3. |
| External PHY |  |  |  |  |
| Bias Resistor |  |  |  |  |$\quad$ ETHRBIAS $\quad$ AI $\quad$| Used for the internal bias circuits. Connect to an |
| :--- |
| external 8.06K 1.0\% resistor to ground. |

## TABLE 2-5: JTAG PINS

| Num Pins | Name | Symbol | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :--- |
| 1 | JTAG Test <br> Data Out | TDO | VO8 | JTAG (IEEE 1149.1) data output. |
| 1 | JTAG Test <br> Data Input | TDI | VIS <br> (PU) | JTAG (IEEE 1149.1) data input. <br> Note: <br> When not used, tie this pin to <br> VDDVARIO. |
| 1 | JTAG Test <br> Clock | TCK | VIS <br> (PD) | JTAG (IEEE 1149.1) test clock. <br> Note: When not used, tie this pin to VSS. |
| 1 | JTAG Test <br> Mode Select | TMS | VIS <br> (PU) | JTAG (IEEE 1149.1) test mode select. <br> Note: |

## TABLE 2-6: MISCELLANEOUS PINS

| Num Pins | Name | Symbol | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :--- |

## TABLE 2-7: POWER PINS

| Num Pins | Name | Symbol | Buffer <br> Type | Description |
| :---: | :---: | :---: | :---: | :--- |
| 6 | +3.3V/+2.5V <br> I/O Power <br> Supply Input | VDDVARIO | P | +2.5V/+3.3V variable I/O power. <br> Refer to Section 3.10, "Application Diagrams," on <br> page 33 and the LAN8810/LAN8810i reference <br> schematics for connection information. |
| 6 | Digital Core <br> $+1.2 V ~ P o w e r ~$ <br> Supply Input | VDD12CORE | P | Refer to Section 3.10, "Application Diagrams," on <br> page 33 and the LAN8810/LAN8810i reference <br> schematics for connection information. |
| 4 | Ethernet <br> $+1.2 V ~ P o r t ~$ <br> Power Supply <br> Input For <br> Channels 0-3 | VDD12A | P | Refer to Section 3.10, "Application Diagrams," on <br> page 33 and the LAN8810/LAN8810i reference <br> schematics for connection information. |
| 1 | Ethernet <br> +1.2V Bias <br> Power Supply <br> Input | VDD12BIAS | P | Refer to Section 3.10, "Application Diagrams," on <br> page 33 and the LAN8810/LAN8810i reference <br> schematics for connection information. |
| 1 | Ethernet PLL <br> $+1.2 V ~ P o w e r ~$ <br> Supply Input | VDD12PLL | P | Refer to Section 3.10, "Application Diagrams," on <br> page 33 and the LAN8810/LAN8810i reference <br> schematics for connection information. |
| Note 2-3 | Ground | VSS | P | Common Ground |

Note 2-3 Exposed pad on package bottom (Figure 2-1).

## TABLE 2-8: $\quad$ 72-QFN PIN ASSIGNMENTS

| Pin Num | Pin Name | Pin Num | Pin Name | Pin Num | Pin Name | Pin Num | Pin Name |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | TCK | 19 | RXER | 37 | TXER | 55 | nRESET |
| 2 | TMS | 20 | CRS | 38 | TXCLK | 56 | HPD |
| 3 | TDO | 21 | COL | 39 | GTXCLK | 57 | TRON |
| 4 | XI | 22 | RXCLK | 40 | VDD12CORE | 58 | TROP |
| 5 | XO | 23 | VDDVARIO | 41 | VDDVARIO | 59 | VDD12A |
| 6 | VDDVARIO | 24 | VDD12CORE | 42 | ACT_LED | 60 | TR1N |
| 7 | VDD12CORE | 25 | NC | 43 | 1000_LED | 61 | TR1P |
| 8 | RXD0 | 26 | TXD7 | 44 | $\begin{gathered} 100 \text { LED/ } \\ \text { HPD_MODE } \\ \hline \end{gathered}$ | 62 | VDD12A |
| 9 | RXD1 | 27 | TXD6 | 45 | 10_LED | 63 | VDD12BIAS |
| 10 | RXD2 | 28 | TXD5 | 46 | CONFIG3 | 64 | VDD12PLL |
| 11 | RXD3 | 29 | TXD4 | 47 | CONFIG2 | 65 | TR2N |
| 12 | RXD4 | 30 | TXD3 | 48 | VDD12CORE | 66 | TR2P |
| 13 | RXD5 | 31 | VDD12CORE | 49 | VDDVARIO | 67 | VDD12A |
| 14 | VDDVARIO | 32 | VDDVARIO | 50 | CONFIG1 | 68 | TR3N |
| 15 | VDD12CORE | 33 | TXD2 | 51 | CONFIG0 | 69 | TR3P |
| 16 | RXD6 | 34 | TXD1 | 52 | MDC | 70 | VDD12A |
| 17 | RXD7 | 35 | TXD0 | 53 | MDIO | 71 | IRQ |
| 18 | RXDV | 36 | TXEN | 54 | ETHRBIAS | 72 | TDI |
| EXPOSED PAD <br> MUST BE CONNECTED TO VSS |  |  |  |  |  |  |  |

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### 2.1 Buffer Types

## TABLE 2-9: BUFFER TYPES

| Buffer Type | Description |
| :---: | :--- |
| VIS | Variable voltage Schmitt-triggered input. |
| VO6 | Variable voltage output with 6 mA sink and 6 mA source. |
| VO8 | Variable voltage output with 8 mA sink and 8 mA source. |
| PU | 50uA (typical) internal pull-up. Unless otherwise noted in the pin description, internal pull- <br> ups are always enabled. <br> Note: $\quad$Internal pull-up resistors prevent unconnected inputs from floating. Do not rely on <br> internal resistors to drive signals external to the device. When connected to a load <br> that must be pulled high, an external resistor must be added. <br> PD50uA (typical) internal pull-down. Unless otherwise noted in the pin description, internal <br> pull-downs are always enabled. <br> Note: Internal pull-down resistors prevent unconnected inputs from floating. Do not rely <br> on internal resistors to drive signals external to the device. When connected to a <br> load that must be pulled low, an external resistor must be added. |
| AI | Analog input. |
| AIO | Analog bi-directional. |
| ICLK | Crystal oscillator input pin. |
| OCLK | Crystal oscillator output pin. |
| P | Power pin. |

Note 1: The digital signals are not 5 V tolerant. Refer to Section 5.1, "Absolute Maximum Ratings*," on page 60 for additional buffer information.
2: Sink and source capabilities are dependent on the VDDVARIO voltage. Refer to Section 5.1, "Absolute Maximum Ratings*," on page 60 for additional information.

### 3.0 FUNCTIONAL DESCRIPTION

This chapter provides functional descriptions of the various device features. These features have been categorized into the following sections:

- Auto-negotiation
- HP Auto-MDIX
- GMII Interface
- Serial Management Interface (SMI)
- Interrupt Management
- Resets
- Power-Down modes
- Configuration
- Miscellaneous Functions
- Application Diagrams


### 3.1 Auto-negotiation

The purpose of the auto-negotiation function is to automatically configure the PHY to the optimum link parameters based on the capabilities of its link partner. Auto-negotiation is a mechanism for exchanging configuration information between two link-partners and automatically selecting the highest performance mode of operation supported by both sides. Autonegotiation is fully defined in clause 28 and clause 40 of the IEEE 802.3 specification.
Once auto-negotiation has completed, information about the resolved link can be passed back to the controller via the integrated Serial Management Interface (SMI). The results of the negotiation process are reflected in the Speed Indication field of the PHY Special Control / Status Register as well as the Auto Negotiation Link Partner Ability Register.
The advertised capabilities of the PHY are stored in Auto Negotiation Advertisement Register. The defaults advertised by the device are determined as described in Section 3.8.1.2.2, "Configuring the Mode of Operation (CONFIG[3:2])," on page 24.
The auto-negotiation protocol is a purely physical layer activity and proceeds independently of the MAC controller. When enabled, auto-negotiation is started by the occurrence of one of the following events:

- Hardware reset
- Software reset
- Power-down reset
- Link status down
- Setting the Restart Auto-Negotiate bit of the Basic Control Register

On detection of one of these events, the device begins auto-negotiation by transmitting bursts of Fast Link Pulses (FLP). The data transmitted by an FLP burst is known as a "Link Code Word." This exchange of information allows link partners to determine the Highest Common Ability (HCD).

Once a capability match has been determined, the link code words are repeated with the acknowledge bit set. Any difference in the main content of the link code words at this time will cause auto-negotiation to re-start. Auto-negotiation will also re-start if all of the required FLP bursts are not received.

Writing the 100BASE-TX Full Duplex, 100BASE-TX, 10BASE-T Full Duplex, and 10BASE-T bits of the Auto Negotiation Advertisement Register allows software control of the advertised capabilities. However, writing the Auto Negotiation Advertisement Register does not automatically re-start auto-negotiation. The Restart Auto-Negotiate bit of the Basic Control Register must be set before the new abilities will be advertised. Auto-negotiation can also be disabled via software by clearing the Auto-Negotiation Enable bit of the Basic Control Register.
Auto-Negotiation also resolves the Master/Slave clocking relationship between two PHYs for a 1000BASE-T link. Refer to Section 3.1.4, "Master/Slave," on page 14 for additional information.

### 3.1.1 RESTARTING AUTO-NEGOTIATION

Auto-negotiation can be restarted at any time by using the Restart Auto-Negotiate bit of the Basic Control Register. Auto-negotiation will also re-start if the link is broken at any time. A broken link is caused by signal loss. This may occur because of a cable break, or because of an interruption in the signal transmitted by the Link Partner. Auto-negotiation resumes in an attempt to determine the new link configuration.

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If the management entity restarts Auto-negotiation by writing to the Restart Auto-Negotiate bit, the device will respond by stopping all transmission/receiving operations. Auto-negotiation will restart after approximately 1200 mS . The Link Partner will have also dropped the link and will resume auto-negotiation.

### 3.1.2 DISABLING AUTO-NEGOTIATION

Auto-negotiation can be disabled via software by clearing the Auto-Negotiation Enable bit of the Basic Control Register. The device will then force its speed of operation to reflect the information in the Speed Select[1], Speed Select[0], and Duplex Mode bits of the Basic Control Register. These bits are ignored when auto-negotiation is enabled.

### 3.1.3 PARALLEL DETECTION

If the LAN8810/LAN8810i is connected to a device lacking the ability to auto-negotiate (for example, no FLPs are detected), it is able to determine the speed of the link based on either 100M MLT-3 symbols or 10M Normal Link Pulses. In this case, the link is presumed to be half-duplex per the IEEE standard. This ability is known as "Parallel Detection". This feature ensures inter operability with legacy link partners.
The Ethernet MAC has access to information regarding parallel detect via the Auto Negotiation Expansion Register. If a link is formed via parallel detection, the Link Partner Auto-Negotiation Able bit of the Auto Negotiation Expansion Register is cleared to indicate that the Link Partner is not capable of auto-negotiation. If a fault occurs during parallel detection, the Parallel Detection Fault bit of this register is set.
The Auto Negotiation Link Partner Ability Register is updated with information from the link partner which is coded in the received FLPs. If the Link Partner is not auto-negotiation capable, then the Auto Negotiation Link Partner Ability Register is updated after completion of parallel detection to reflect the speed capability of the Link Partner.
Parallel detect cannot be used to establish Gigabit Ethernet links because echo cancellation and signal recovery on a Gigabit Ethernet link requires resolution of the Master/Slave clock relationship, which requires the exchange of FLPs.

### 3.1.4 MASTER/SLAVE

In 1000BASE-T, one of the two link partner devices must be configured as Master and the other as Slave. The Master device transmits data using the local clock, while the Slave device uses the clock recovered from incoming data.

The Master and Slave assignments are set using the configuration pins as described in Section 3.8.1.2.2, "Configuring the Mode of Operation (CONFIG[3:2])," on page 24 or by using the Master/Slave Manual Config Enable and Master/ Slave Manual Config Value bits of the Master/Slave Control Register. If both the link partner and the local device are manually given the same Master/Slave assignment, an error will be indicated in the Master/Slave Configuration Fault bit of the Master/Slave Status Register.
Depending on the link partner configuration, the manual Master/Slave mode can be resolved to sixteen possible outcomes, as shown in Table 3-1.

## TABLE 3-1: MASTER/SLAVE RESOLUTION FOR 1000BASE-T

| LAN8810/LAN8810i <br> Advertisement | Link Partner <br> Advertisement | LAN8810/LAN8810i Result | Link Partner Result |
| :--- | :--- | :--- | :--- |
| Single-Port | Single-Port | M/S resolved by random seed | M/S resolved by random seed |
| Single-Port | Multi-Port | Slave | Master |
| Single-Port | Manual Master | Slave | Master |
| Single-Port | Manual Slave | Master | Slave |
| Multi-Port | Single-Port | Master | Slave |
| Multi-Port | Multi-Port | M/S resolved by random seed | M/S resolved by random seed |
| Multi-Port | Manual Master | Slave | Master |
| Multi-Port | Manual Slave | Master | Slave |
| Manual Master | Single-Port | Master | Slave |

## TABLE 3-1: MASTER/SLAVE RESOLUTION FOR 1000BASE-T (CONTINUED)

| LAN8810/LAN8810i <br> Advertisement | Link Partner <br> Advertisement | LAN8810/LAN8810i Result | Link Partner Result |
| :--- | :--- | :--- | :--- |
| Manual Master | Multi-Port | Master | Slave |
| Manual Master | Manual Master | No Link | No Link |
| Manual Master | Manual Slave | Master | Slave |
| Manual Slave | Single-Port | Slave | Master |
| Manual Slave | Multi-Port | Slave | Master |
| Manual Slave | Manual Master | Slave | Master |
| Manual Slave | Manual Slave | No Link | No Link |

### 3.1.5 MANUAL OPERATION

The device supports a manual (forced) operation for test purposes. In manual operation, the user sets the link speed (10Mbps or 100 Mbps ) and the duplex state (full or half).
Auto-negotiation must be disabled in order to manually configure the speed and the duplex. This may be accomplished using the configuration pins, as described in Section 3.8.1.2.2, "Configuring the Mode of Operation (CONFIG[3:2])," on page 24, or by using the Basic Control Register as described in Section 3.1.2, "Disabling Auto-negotiation," on page 14. For 10BASE-T and 100BASE-TX, the link state of the device is determined by the Speed Select[1], Speed Select[0], and Duplex Mode bits of the Basic Control Register. Manual operation at a link speed of 1000Mbps is not supported.

### 3.1.6 HALF VS. FULL-DUPLEX

Half-duplex operation relies on the CSMA/CD (Carrier Sense Multiple Access / Collision Detect) protocol to handle network traffic and collisions. In this mode, the internal carrier sense signal, CRS, responds to both transmit and receive activity. If data is received while the PHY is transmitting, a collision results.
In full-duplex mode, the PHY is able to transmit and receive data simultaneously and collision detection is disabled. In this mode, the internal CRS responds only to receive activity. In 10BASE-T and 100BASE-T mode, CRS is redefined to respond only to received activity. In 1000BASE-T, CRS is disabled.
Table 3-2 describes the behavior of the internal CRS bit under all receive/transmit conditions.

## TABLE 3-2: CRS BEHAVIOR

| Mode | Speed | Duplex | Activity | CRS Behavior <br> (Note 3-1) |
| :---: | :---: | :---: | :---: | :---: |
| Manual | 10 Mbps | Half-Duplex | Transmitting | Active |
| Manual | 10 Mbps | Half-Duplex | Receiving | Active |
| Manual | 10 Mbps | Full-Duplex | Transmitting | Low |
| Manual | 10 Mbps | Full-Duplex | Receiving | Active |
| Manual | 100 Mbps | Half-Duplex | Transmitting | Active |
| Manual | 100 Mbps | Half-Duplex | Receiving | Active |
| Manual | 100 Mbps | Full-Duplex | Transmitting | Low |
| Manual | 100 Mbps | Full-Duplex | Receiving | Active |
| Auto-Negotiation | 10 Mbps | Half-Duplex | Transmitting | Active |

## TABLE 3-2: CRS BEHAVIOR (CONTINUED)

| Mode | Speed | Duplex | Activity | CRS Behavior <br> (Note 3-1) |
| :---: | :---: | :---: | :---: | :---: |
| Auto-Negotiation | 10 Mbps | Half-Duplex | Receiving | Active |
| Auto-Negotiation | 10 Mbps | Full-Duplex | Transmitting | Low |
| Auto-Negotiation | 10 Mbps | Full-Duplex | Receiving | Active |
| Auto-Negotiation | 100 Mbps | Half-Duplex | Transmitting | Active |
| Auto-Negotiation | 100 Mbps | Half-Duplex | Receiving | Active |
| Auto-Negotiation | 100 Mbps | Full-Duplex | Transmitting | Low |
| Auto-Negotiation | 100 Mbps | Full-Duplex | Receiving | Active |

Note 3-1 The internal CRS signal operates in two modes: Active and Low. When in Active mode, the internal CRS will transition high and low upon line activity, where a high value indicates a carrier has been detected. In Low mode, the internal CRS stays low and does not indicate carrier detection.

### 3.2 HP Auto-MDIX

HP Auto-MDIX facilitates the use of CAT-5 (100BASE-T) media UTP interconnect cable without consideration of interface wiring scheme. If a user plugs in either a direct connect LAN cable, or a crossover patch cable, as shown in Figure 3-1, the Auto-MDIX PHY is capable of configuring the twisted pair pins for correct transceiver operation.
The internal logic of the device detects the TX and RX pins of the connecting device. It can automatically re-assign channel 0 and 1 if required to establish a link. In 1000BASE-T mode, it can re-assign channel 2 and 3 . Crossover resolution precedes the actual auto-negotiation process that involves exchange of FLPs to advertise capabilities. Automatic MDI/ MDIX is described in IEEE 802.3ab Clause 40, section 40.8.2. Since the RX and TX line pairs are interchangeable, special PCB design considerations are needed to accommodate the symmetrical magnetics and termination of an AutoMDIX design.
Auto-MDIX is enabled by default, and can be disabled by the Auto MDIX Disable bit in the 10/100 Mode Control/Status Register. When Auto-MDIX is disabled, the TX and RX pins can be configured manually by the MDI/MDI-X 0:1 and MDI/ MDI-X 2:3 bits in the Extended Mode Control/Status Register.
The device includes an advanced crossover resolution capability called Semi Crossover. This is an extension to HP Auto-MDIX that corrects for a cable with only two pairs crossed. If Semi Crossover is enabled, after the device has attempted to establish a link with all four signal pairs normal or crossed, it will attempt to establish a link with pairs $2 / 3$ switched and $0 / 1$ straight, and then with pairs $0 / 1$ switched and pairs $2 / 3$ straight. The Semi Crossover is enabled by default, and can be disabled by the Semi Crossover Enable bit in the 10/100 Mode Control/Status Register.
After resolution of crossed pairs is complete, using either HP Auto-MDIX or the Semi Crossover function, the MDI/MDIX status is reported through the XOVER Resolution 0:1 and XOVER Resolution 2:3 bits of the User Status 2 Register.

FIGURE 3-1: CABLE CONNECTION TYPES: STRAIGHT-THROUGH, CROSSOVER, SEMI CROSSOVER


### 3.2.1 REQUIRED ETHERNET MAGNETICS

The magnetics selected for use with the device should be an Auto-MDIX style magnetic available from several vendors. Refer to Application Note 8.13 "Suggested Magnetics" for the latest qualified and suggested magnetics. Vendors and part numbers are provided in this application note.

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### 3.3 GMII Interface

The device communicates with an external MAC using the Gigabit Media Independent Interface (GMII). The GMII is compliant with the IEEE 802.3 standard, and provides support for 1000BASE-T, 100BASE-TX, or 10BASE-T operation.
For 1000BASE-T, the MAC generated 125 MHz GTXCLK transmit clock is used to synchronize the TXD, TXEN and TXER signals. For 100BASE-TX and 10BASE-T operation, the device generated TXCLK transmit clock is used to synchronize the TXD, TXEN and TXER signals. TXCLK is 25 MHz or 2.5 MHz for 100BASE-TX and 10BASE-T operation, respectively. TXER and TXEN are both driven by the MAC and indicate a transmit error and valid transmit data, respectively. On the receiver side, the device generated RXCLK is used to synchronize the RXD, RXDV, RXER, COL and CRS signals for all modes of operation.
The GMII provides backwards compatibility with the legacy MII. Table 3-3, "GMII/MII Signal Mapping" describes which pins are used in each mode.

TABLE 3-3: GMII/MII SIGNAL MAPPING

| GMII Mode <br> (IEEE 802.3 Clause 35) | MII Mode <br> (IEEE 802.3 Clause 22) |
| :---: | :---: |
| TXD[3:0] | TXD[3:0] |
| TXD[7:4] |  |
| TXEN | TXEN |
| TXER | TXER |
| TXCLK (10/100Mbps operation) | TXCLK |
| GTXCLK (1000Mbps operation) | CRS |
| COL | RXD[3:0] |
| CRS |  |
| RXD[3:0] | RXDV |
| RXD[7:4] | RXER |
| RXDV | RXCLK |
| RXER |  |
| RXCLK |  |

Timing information for the GMII/MII interface is provided in Section 5.5, "AC Specifications," on page 65. For additional information on the GMII/MII interface, refer to the IEEE 802.3 specification.

### 3.3.1 MII ISOLATE MODE

The device may be configured to electrically isolate the GMII pins by setting the Isolate bit of the Basic Control Register. In this mode, all MAC data interface output pins are HIGH and all MAC data interface input pins are ignored. In this mode, the SMI interface is kept active, allowing the MAC to access the SMI registers and generate interrupts. All MDI operations are halted while in isolate mode.

### 3.4 Serial Management Interface (SMI)

The Serial Management Interface is used to control the device and obtain its status. This interface supports the standard PHY registers required by Clause 22 of the 802.3 standard, as well as "vendor-specific" registers allowed by the specification. Non-supported registers (such as 11 to 14) will be read as hexadecimal "FFFF". Device registers are detailed in Section 4.0, "Register Descriptions," on page 35.
At the system level, SMI provides 2 signals: MDIO and MDC. The MDC signal is an aperiodic clock provided by the station management controller (SMC). MDIO is a bi-directional data SMI input/output signal that receives serial data (commands) from the controller SMC and sends serial data (status) to the SMC. The minimum time between edges of the MDC is 160 ns . There is no maximum time between edges. The minimum cycle time (time between two consecutive rising or two consecutive falling edges) is 400 ns . These modest timing requirements allow this interface to be easily driven by the I/O port of a microcontroller.
The data on the MDIO line is latched on the rising edge of the MDC. The frame structure and timing of the data is shown in Figure 1-1 and Figure 1-2. The timing relationships of the MDIO signals are further described in Section 5.5.7, "SMI Timing," on page 73.

FIGURE 3-2: MDIO TIMING AND FRAME STRUCTURE - READ CYCLE


FIGURE 3-3: MDIO TIMING AND FRAME STRUCTURE - WRITE CYCLE


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### 3.5 Interrupt Management

The device supports multiple interrupt capabilities which are not a part of the IEEE 802.3 specification. An active low asynchronous interrupt signal may be generated on the IRQ pin when selected events are detected, as configured by the Interrupt Mask Register.
To set an interrupt, the corresponding mask bit in the Interrupt Mask Register must be set (see Table 3-4). When the associated event occurs, the IRQ pin will be asserted. When the corresponding event to deassert IRQ is true, the IRQ pin will be deasserted. All interrupts are masked following a reset.

Note: Table 3-4 utilizes register index and bit number referencing in lieu of individual names. For example, " 30.10 " is used to reference bit 10 (transmitter elastic buffer overflow interrupt enable) of the Interrupt Mask Register (register index 30).

TABLE 3-4: INTERRUPT MANAGEMENT TABLE

| Mask | Interrupt Source Flag |  | Interrupt Source |  | Event to Assert IRQ | Event to Deassert IRQ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 30.15:11 | 29.15:11 | RESERVED | -NA- | -NA- | -NA- | -NA- |
| 30.10 | 29.10 | Transmitter Elastic Buffer Overflow | -NA- | -NA- <br> (Note 3-3) | Transmitter Elastic Buffer Overflow | Overflow condition resolved |
| 30.9 | 29.9 | Transmitter Elastic Buffer Underflow | -NA- | -NA- <br> (Note 3-3) | Transmitter Elastic Buffer Underflow | Underflow condition resolved |
| 30.8 | 29.8 | Idle Error Count Overflow | 10.7:0 | Idle Error Count | Idle Error Count Overflow | Reading register 10 |
| 30.7 | 29.7 | ENERGYON | 17.1 | ENERGYON | Rising 17.1 <br> (Note 3-2) | Falling 17.1 or Reading register 29 |
| 30.6 | 29.6 | Auto-Negotiation complete | 1.5 | Auto-Negotiate Complete | Rising 1.5 | Falling 1.5 or Reading register 29 |
| 30.5 | 29.5 | Remote Fault Detected | 1.4 | Remote Fault | Rising 1.4 | Falling 1.4, or Reading register 1 or Reading register 29 |
| 30.4 | 29.4 | Link Down | 1.2 | Link Status | Falling 1.2 | Reading register 1 or Reading register 29 |
| 30.3 | 29.3 | RESERVED | -NA- | -NA- | -NA- | -NA- |
| 30.2 | 29.2 | Parallel Detection Fault | 6.4 | Parallel <br> Detection Fault | Rising 6.4 | Falling 6.4 or Reading register 6, or Reading register 29 or Re-AutoNegotiate or Link down |
| 30.1 | 29.1 | Auto-Negotiation <br> Page Received | 6.1 | Page Received | Rising 6.1 | Falling of 6.1 or Reading register 6, or Reading register 29 Re-auto-negotiate, or Link Down. |

Note 3-2 The ENERGYON bit of the 10/100 Mode Control/Status Register (17.1) defaults to "1" after a hardware reset. If no energy is detected before 256 mS , the ENERGYON bit will be cleared. When ENERGYON is " 0 " and energy is detected, due to the establishment of a valid link or the PHY autonegotiation moving past the ability detect state, the ENERGYON bit will be set and the INT7 bit of the Interrupt Source Flags Register will assert. If ENERGYON is set and the energy is removed, the INT7 bit will assert. The ENERGYON bit will clear 256 mS after the interrupt. If the PHY is in manual mode, INT7 will be asserted 256 mS after the link is broken. If the PHY is auto-negotiating, INT7 will be asserted 256 mS after the PHY returns to the ability detect state (maximum of 1.5 S after the link
is broken). To prevent an unexpected assertion of IRQ, the ENERGYON interrupt mask (INT7_EN) should always be cleared as part of the ENERGYON interrupt service routine.
Note 3-3 The transmitter FIFO depth can be adjusted via the Transmitter FIFO Depth field of the Extended Mode Control/Status Register (19.10:9).

### 3.6 Resets

The device provides the following chip-level reset sources:

- Hardware Reset (nRESET)
- Software Reset
- Power-Down Reset


### 3.6.1 HARDWARE RESET (NRESET)

Note: System implementers should connect the nRESET input pin to an output pin from the respective MAC or microcontroller, so that the required power-up sequence can be performed without causing a full system reset event.

A hardware reset will occur when the system reset nRESET input pin is driven low. Anytime nRESET is asserted, it must be held low for the minimum time specified in Section 5.5.4, "Reset Timing," on page 68 to ensure proper reset to the PHY. Following a hardware reset, the device resets the device registers and relatches the configuration straps and CONFIG[3:0] pins.
On first power-up of the device, the sequence below must be also be followed to ensure the device exits reset in the correct operational state:

1. Perform a hardware reset on power-up as per Section 5.5.3, "Power-On Hardware Reset Timing," on page 67.
2. Wait a minimum of 250 mS
3. Write SMI Register 0 (Basic Control Register) $=0 \times 4040$
4. Wait a minimum of 1 second
5. Assert the nRESET input pin (nRESET = 0)
6. Wait a minimum of 50 mS
7. Deassert the nRESET input pin (nRESET = 1)

After completing this sequence, the LAN8810/LAN8810i will be in the default states and ready for any initialization or configuration and allow operation.

Note: A hardware reset (nRESET assertion) is required following power-up. Refer to Section 5.5.3, "Power-On Hardware Reset Timing," on page 67 for additional information.

### 3.6.2 SOFTWARE RESET

A software reset is initiated by writing a ' 1 ' to the PHY Soft Reset (RESET) bit of the Basic Control Register. This selfclearing bit will return to ' 0 ' after approximately $256 \mu \mathrm{~s}$, at which time the PHY reset is complete. This reset initializes the logic within the PHY, with the exception of register bits marked as "NASR" (Not Affected by Software Reset).

Following a software reset, the device configuration is reloaded from the register bit values, and not from the configuration straps and CONFIG[3:0] pins. The device does not relatch the hardware configuration settings. For example, if the device is powered up and a configuration strap is changed from its initial power up state, a software reset will not load the new strap setting.

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### 3.6.3 POWER-DOWN RESET

A power-down reset is automatically activated when the device comes out of the power-down mode. During powerdown, the registers are not reset. Configuration straps and CONFIG[3:0] pins are not latched as a result of a powerdown reset. The power-down reset is internally extended by $256 \mu$ s after exiting the power-down mode to allow the PLLs to stabilize before the logic is released from reset. Refer to Section 3.7, "Power-Down modes," on page 22 for details on the various power-down modes.

### 3.7 Power-Down modes

The device supports 3 power-down modes:

- General Power-Down
- Energy Detect Power-Down
- Hardware Power-Down


### 3.7.1 GENERAL POWER-DOWN

This power-down mode is controlled by the Power Down bit of the Basic Control Register. In this mode, the entire device is powered-down except for the serial management interface. The device remains in the general power-down mode while Power Down is set. When Power Down is cleared, the device powers up and is automatically reset (via a PowerDown Reset). For maximum power savings, auto-negotiation should be disabled before enabling the general powerdown mode.

### 3.7.2 ENERGY DETECT POWER-DOWN

This power-down mode is controlled by the EDPWRDOWN bit of the 10/100 Mode Control/Status Register. In this mode, when no energy is present on the line, nothing is transmitted and the device is powered-down except for the management interface, the SQUELCH circuit and the ENERGYON logic.

The ENERGYON bit in the 10/100 Mode Control/Status Register is asserted when there is valid energy from the line (100BASE-TX, 10BASE-T, or Auto-Negotiation signals) and the PHY powers-up. It automatically resets itself into the previous state prior to power-down, and stays in active mode as long as energy exists on the line. If the ENGERGYON interrupt is enabled (INT7_EN of the Interrupt Mask Register), IRQ is asserted.

## Note: The first and possibly second packet to activate ENERGYON may be lost.

### 3.7.3 HARDWARE POWER-DOWN

This power-down mode is controlled by the HPD pin. In this mode, the entire device is powered-down except for the serial management interface. The HPD MODE configuration strap selects whether the PLL will be shut down when in hardware power-down mode. To exit the hardware power-down mode, the HPD pin must be deasserted, followed by the deassertion of the Power Down bit in the Basic Control Register. If the hardware power-down mode is set to shut down the PLL, a software reset must also be issued.
Note 1: The device will wake-up in the hardware power-down mode if the HPD pin is asserted during hardware reset.
2: For additional information on the HPD MODE configuration strap, refer to Section 3.8.1.1, "Configuration Straps," on page 23.

### 3.8 Configuration

The device mode of operation may be controlled by hardware and software (register-selectable) configuration options. The initial configuration may be selected in hardware as described in Section 3.8.1. In addition, register-selectable software configuration options may be used to further define the functionality of the transceiver as described in Section 3.8.2. The device supports both IEEE 802.3-2005 compliant and vendor-specific register functions.

### 3.8.1 HARDWARE CONFIGURATION

Hardware configuration is controlled via multiple configuration straps and the CONFIG[3:0] configuration pins. These items are detailed in the following sub-sections.

### 3.8.1.1 Configuration Straps

Configuration straps are multi-function pins that are driven as outputs during normal operation. During a Hardware Reset (nRESET), these outputs are tri-stated. The high or low state of the signal is latched following de-assertion of the reset and is used to determine the default configuration of a particular feature. Table 3-5 details the configuration straps. Configuration straps are also listed as part of Section 2.0, "Pin Description and Configuration," on page 5 with underlined names.
Configuration straps include internal resistors in order to prevent the signal from floating when unconnected. If a particular configuration strap is connected to a load, an external pull-up or pull-down should be used to augment the internal resistor to ensure that it reaches the required voltage level prior to latching. The internal resistor can also be overridden by the addition of an external resistor.

Note 1: The system designer must guarantee that configuration straps meet the timing requirements specified in Section 5.5.3, "Power-On Hardware Reset Timing," on page 67. If configuration straps are not at the correct voltage level prior to being latched, the device may capture incorrect strap values.
2: Configuration straps must never be driven as inputs. If required, configuration straps can be augmented, or overridden with external resistors.

## TABLE 3-5: CONFIGURATION STRAPS

| Configuration <br> Strap | Description | Logic 0 <br> (PD) | Logic $\mathbf{1}$ <br> (PU) |
| :--- | :--- | :---: | :---: |
| HPD MODE | Selects the hardware power-down (HPD) mode | HPD with PLL <br> disabled (Default) | HPD with PLL <br> enabled |

### 3.8.1.2 CONFIG[3:0] Configuration Pins

The device provides 4 dedicated configuration pins, CONFIG[3:0], which are used to select the default SMI address and mode of operation. The CONFIG[3:0] configuration pins differ from configuration straps in that they are single-purpose pins and have different latch timing requirements. The high or low states of the CONFIG[3:0] pins are latched following deassertion of a Hardware Reset (nRESET). Refer to Section 5.5.3, "Power-On Hardware Reset Timing," on page 67 for additional CONFIG[3:0] timing information.
Each CONFIG[3:0] configuration pin can be connected in one of four ways. The Configuration Pin Value (CPV) represented by each connection option is shown in Table 3-6.

TABLE 3-6: HARDWARE CONNECTION DETERMINES CONFIGURATION PIN VALUE (CPV)

| CONFIG[X] <br> Connects to: | Value |
| :--- | :--- |
| GND | $\mathrm{CPV}(0)$ |
| 100_LED | $\mathrm{CPV}(1)$ |
| 1000_LED | $\mathrm{CPV}(2)$ |
| VDD | $\mathrm{CPV}(3)$ |

Using the CPV nomenclature for each CONFIG[3:0] pin, Section 3.8.1.2.1 describes how to configure the SMI address and Section 3.8.1.2.2 describes how to configure the initial mode of operation.

Note: The HPD pin is also a dedicated configuration pin. HPD forces the entire device to power down except for the management interface. The Hardware Power-Down mode is described in Section 3.7.3, "Hardware Power-Down," on page 22.

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### 3.8.1.2.1 Configuring the SMI Address (CONFIG[1:0])

The SMI address may be configured via hardware to any value between 0 and 7 . If an address greater than 7 is required, the user can configure the PHY address using Software Configuration via the PHYADD[4:0] field of the 10/100 Special Modes Register (after SMI communication at an address is established).
The CONFIG1 pin is used to configure both the SMI address and the value of the Pause Operation (PAUSE) bit in the Auto Negotiation Advertisement Register. The user must first determine the desired PAUSE value. The configuration pin values for CONFIG1 and CONFIG0 should then be selected using Table 3-7 (PAUSE=0) or Table 3-8 (PAUSE=1), respectively.

TABLE 3-7: SMI ADDRESS CONFIGURATION WITH PAUSE=0

| PHYADD[2:0] | CONFIG1 | CONFIG0 |
| :---: | :---: | :---: |
| 000 | $\operatorname{CPV}(0)$ | $\mathrm{CPV}(0)$ |
| 001 | $\mathrm{CPV}(0)$ | $\mathrm{CPV}(1)$ |
| 010 | $\mathrm{CPV}(0)$ | $\mathrm{CPV}(2)$ |
| 011 | $\mathrm{CPV}(0)$ | $\mathrm{CPV}(3)$ |
| 100 | $\mathrm{CPV}(1)$ | $\mathrm{CPV}(0)$ |
| 101 | $\mathrm{CPV}(1)$ | $\mathrm{CPV}(1)$ |
| 110 | $\mathrm{CPV}(1)$ | $\mathrm{CPV}(2)$ |
| 111 |  | $\mathrm{CPV}(3)$ |

## TABLE 3-8: SMI ADDRESS CONFIGURATION WITH PAUSE=1

| PHYADD[2:0] | CONFIG1 | CONFIG0 |
| :---: | :---: | :---: |
| 000 | $\mathrm{CPV}(2)$ | $\mathrm{CPV}(0)$ |
| 001 | $\mathrm{CPV}(2)$ | $\mathrm{CPV}(1)$ |
| 010 | $\mathrm{CPV}(2)$ | $\mathrm{CPV}(2)$ |
| 011 | $\mathrm{CPV}(2)$ | $\mathrm{CPV}(3)$ |
| 100 | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(0)$ |
| 101 | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(1)$ |
| 110 | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(2)$ |
| 111 |  | $\mathrm{CPV}(3)$ |

### 3.8.1.2.2 Configuring the Mode of Operation (CONFIG[3:2])

This section describes the initial modes of operation that are available using the CONFIG[3:2] configuration pins. The user may configure additional modes using Software Configuration when the CONFIG[3:2] options do not include the desired mode.
The CONFIG3 pin is used to configure the values of the MOD field (19.15:11) The configuration pin values for CONFIG3 and CONFIG2 should be selected using Table 3-9. These tables also detail how the MOD field of the Extended Mode Control/Status Register will be configured.

Section 3.8.1.2.3 describes how the MOD field controls other configuration bits in the device. When a soft reset is issued via the PHY Soft Reset (RESET) bit of the Basic Control Register, configuration is controlled by the register bit values and the CONFIG[3:0] pins have no affect. Likewise, changing the MOD field of the Extended Mode Control/Status Register bits does not change the configuration of the device in this case.

Note: Table 3-9 utilizes register index and bit number referencing in lieu of individual names.

## TABLE 3-9: CONFIGURING THE MODE OF OPERATION

| Mode Definitions | CONFIG3 | CONFIG2 | Reg 19 <br> [15:11] |
| :--- | :---: | :---: | :---: |
| 10BASE-T Half Duplex. Auto-negotiation disabled. | CPV(2) | CPV(0) | 00000 |
| 10BASE-T Full Duplex. Auto-negotiation disabled. | CPV(2) | CPV(1) | 00001 |
| 100BASE-TX Half Duplex. Auto-negotiation disabled. <br> CRS is active during Transmit \& Receive. | $\mathrm{CPV}(2)$ | $\mathrm{CPV}(2)$ | 00010 |
| 100BASE-TX Full Duplex. Auto-negotiation disabled. <br> CRS is active during Receive. | $\mathrm{CPV}(2)$ | $\mathrm{CPV}(3)$ | 00011 |
| All mode capable (10/100/1000). Auto-negotiation enabled. <br> Auto master/slave resolution single port. | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(0)$ | 00111 |
| 10BASE-T/100BASE-TX capable. Auto-negotiation enabled. | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(1)$ | 00100 |
| All mode capable (10/100/1000). Auto-negotiation enabled. <br> Manual master/slave resolution slave port. | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(2)$ | 01001 |
| All mode capable (10/100/1000). Auto-negotiation enabled. <br> Manual master/slave resolution master port. | $\mathrm{CPV}(3)$ | $\mathrm{CPV}(3)$ | 01010 |

### 3.8.1.2.3 Configuration Bits Impacted by the Mode of Operation

Immediately after a reset, the MOD field of the Extended Mode Control/Status Register will be set dependent on the configuration pin values of the CONFIG3 and CONFIG2 pins, as described in Section 3.8.1.2.2. Table 3-10 details how the MOD field effects other device configuration register bits.

Note: Table 3-10 utilizes register index and bit number referencing in lieu of individual names

TABLE 3-10: REGISTER BITS IMPACTED BY THE MODE OF OPERATION (MOD)

| $\begin{aligned} & \text { Reg } 19 \\ & {[15: 11]} \end{aligned}$ | Mode Definitions | $\begin{gathered} \text { Reg } 0 \\ {[13,12,8,6]} \end{gathered}$ | $\begin{gathered} \operatorname{Reg} 4 \\ {[8,7,6,5]} \end{gathered}$ | $\begin{gathered} \text { Reg } 9 \\ {[12,11,10,9,8]} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: |
| 00000 | 10BASE-T Half Duplex. Auto-negotiation disabled. | 0000 | XXXX | XXX00 |
| 00001 | 10BASE-T Full Duplex. Auto-negotiation disabled. | 0010 | XXXX | XXX00 |
| 00010 | 100BASE-TX Half Duplex. Auto-negotiation disabled. CRS is active during Transmit \& Receive. | 1000 | XXXX | XXX00 |
| 00011 | 100BASE-TX Full Duplex. Auto-negotiation disabled. CRS is active during Receive. | 1010 | XXXX | XXX00 |
| 00100 | 10BASE-T/100BASE-TX capable. Auto-negotiation enabled. | 11X0 | 1111 | XXX00 |
| 00101 | RESERVED | - | - | - |
| 00110 | RESERVED | - | - | - |
| 00111 | All capable. Auto-negotiation enabled. Auto master/slave resolution single port. | 01X1 | 1111 | 00011 |
| 01000 | RESERVED | - | - | - |
| 01001 | All capable. Auto-negotiation enabled. Manual master/slave resolution slave port. | 01X1 | 1111 | 10011 |
| 01010 | All capable. Auto-negotiation enabled. Manual master/slave resolution master port. | 01X1 | 1111 | 11111 |
| $\begin{aligned} & 01011 \text { - } \\ & 11111 \end{aligned}$ | RESERVED | - | - | - |

### 3.8.2 SOFTWARE CONFIGURATION

The Serial Management Interface (SMI) allows for the configuration and control of multiple transceivers. Several 16-bit status and control registers are accessible through the management interface pins MDC and MDIO for 10/100/ 1000 Mbps operation. The device implements all the required MII registers and optional registers as described in Section 4.0, "Register Descriptions". Configuring the SMI address is described in Section 3.8.1.2.1. Refer to Section 3.4, "Serial Management Interface (SMI)," on page 19 for additional information on the SMI.

### 3.9 Miscellaneous Functions

### 3.9.1 LEDS

The device provides the following LED signals to enable visual indication of status:

- 1000_LED
- 100_LED
- 10_LED
- ACT_LED

In all modes, the Full Duplex LED (ACT_LED) is driven immediately after auto-negotiation has resolved the duplex state and before the link is established. The Speed LEDs (1000_LED, 100_LED, 10_LED) are driven after a link is established. Three LED modes of operation are available:

- LED Mode 1 (Table 3-11)
- LED Mode 2 (Table 3-12)
- LED Mode 3 (Default Mode - Table 3-13)


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The LED mode is configured via the LED_MODE[1:0] field of the Extended Mode Control/Status Register. The functional operation of the LEDs in each mode is described in Table 3-11, Table 3-12 and Table 3-13.

Note 1: LED Mode 3 is the default mode of operation and requires four LEDs for full operation.
2: LED Modes 1 and 2 require two dual LED packages (two LEDs in each package). LED Mode 2 provides collision detection, while LED Mode 1 does not.

TABLE 3-11: LED MODE 1 OPERATION - LED_MODE[1:0]=01B

| LED | STATUS | DESCRIPTION |
| :--- | :--- | :--- |
| 100_LED | On Color1 | 100BASE-T link |
| 1000_LED | On Color2 | 1000BASE-T link |
|  <br> $100 \overline{0} \_$LED | On Color3 <br> Note 3-4 | 10BASE-T link |
| ACT_LED | On Color1 <br> Note 3-5 | On Color2 |
|  | Off | Transmit/receive activity |
|  <br> ACT_LED | On Color1 | Half duplex, no activity |
|  | Off/Color2 | Full duplex, no activity |
|  | Color1/Color2 | Full duplex, transmit/receive activity |

Note 3-4 Switching at $50 \%$ duty cycle creates a third color (for example, If Color1 is orange and Color2 is green, Color 3 will be yellow (Orange + Green)).
Note 3-5 Activity will cause duplex LED color to blink.
TABLE 3-12: LED MODE 2 OPERATION - LED_MODE[1:0]=10B

| LED | STATUS | DESCRIPTION |
| :--- | :--- | :--- |
| 1000_LED | On Color1 | 1000BASE-T link |
| 100_LED | On Color2 | 100BASE-T link |
|  <br> 100 | On Color3 <br> Note 3-4 | On Color1 <br> Note 3-7 |
| ACT_LED | On Color2 | Link operates at full duplex |
|  | On Color3 <br>  | Tote 3-4 |

[^0]
## TABLE 3-13: LED MODE 3 OPERATION - LED_MODE[1:0]=11B

| LED | STATUS | DESCRIPTION |
| :--- | :--- | :--- |
| 1000_LED | On | 1000 BASE-T link |
|  | Blinking | Transmit/receive activity |
| $100 \_$LED | On | $100 B A S E-T$ link |
|  | Blinking | Transmit/receive activity |
| $10 \_$LED | On | $10 B A S E-T$ link |
|  | Blinking | Transmit/receive activity |
| ACT_LED | On | Link operates at full duplex |
|  | Off | Link operates at half duplex |

### 3.9.2 VARIABLE VOLTAGE I/O

The device's digital I/O pins are variable voltage, allowing them to take advantage of low power savings from shrinking technologies. These pins can operate at +2.5 V or 3.3 V . The applied I/O voltage must maintain its value within the tolerances specified in Section 5.0, "Operational Characteristics," on page 60. Varying the voltage up or down after the transceiver has completely powered-on can cause errors in the transceiver operation. Refer to Section 5.0, "Operational Characteristics," on page 60 for additional information.

### 3.9.3 ISOLATE MODE

The device data paths may be electrically isolated from the GMII interface by setting the Isolate bit of the Basic Control Register to " 1 ". In isolation mode, the transceiver does not respond to the TXD, TXEN and TXER inputs, but does respond to management transactions.
Isolation provides a means for multiple transceivers to be connected to the same GMII interface without contention. By default, the transceiver is not isolated (on power-up, Isolate $=0$ ).

### 3.9.4 CARRIER SENSE

The carrier sense signal is output on CRS. Carrier sense operation is defined in the IEEE 802.3u standard. The PHY asserts carrier sense based only on receive activity whenever the PHY is either in repeater mode or full-duplex mode. Otherwise, the PHY asserts carrier sense based on either transmit or receive activity.
The carrier sense logic uses the encoded, unscrambled data to determine carrier activity status. It activates carrier sense with the detection of 2 non-contiguous zeros within any 10 bit span. Carrier sense terminates if a span of 10 consecutive ones is detected before a $/ \mathrm{J} / \mathrm{K} /$ Start-of Stream Delimiter pair. If an SSD pair is detected, carrier sense is asserted until either /T/R/ End-of-Stream Delimiter pair or a pair of IDLE symbols is detected. Carrier is negated after the /T/ symbol or the first IDLE. If/T/ is not followed by /R/, then carrier is maintained. Carrier is treated similarly for IDLE followed by some non-IDLE symbol.

### 3.9.5 COLLISION DETECT

A collision is the occurrence of simultaneous transmit and receive operations. The COL output is asserted to indicate that a collision has been detected. COL remains active for the duration of the collision. COL is changed asynchronously to both RXCLK and TXCLK. A collision can only occur in half-duplex mode. The COL output becomes inactive during full duplex mode.
COL may be tested by setting the Collision Test bit of the Basic Control Register. This enables the collision test. COL will be asserted within 512 bit times of TXEN rising and will be de-asserted within 4 bit times of TXEN falling.
In 10BASE-T mode, COL pulses for approximately 10 bit times (1us), 2us after each transmitted packet (de-assertion of TXEN). This serves as the Signal Quality Error (SQE) signal and indicates that the transmission was successful.

### 3.9.6 LINK INTEGRITY

This section details the establishment, maintenance and removal of links between the device and a link partner in 1000BASE-T, 100BASE-TX and 10BASE-T modes. Link status is reported in the Link Status bit of the Basic Status Register. The link status is also used to drive the device LEDs as described in Section 3.9.1, "LEDs," on page 26.

### 3.9.6.1 Establishing and Maintaining a Link

Once a link state is determined via auto-negotiation, parallel detection, or forced operation, the device and the link partner establish a link.
The completion of the auto-negotiation process is reported via the Auto-Negotiate Complete bit of the Basic Status Register and issues an interrupt as described in Section 3.5, "Interrupt Management," on page 20. The speed of the link is reported in the Speed Indication field of the PHY Special Control / Status Register. The speed is also reported on the LED pins for any link.
Failure to complete the auto-negotiation process is reported through the following status bits:

- Parallel Detection Fault reported in the Auto Negotiation Expansion Register while operating in 10BASE-T or 100BASE-TX modes.
- Master/Slave Configuration Fault reported in the Master/Slave Status Register while operating in 1000BASE-T mode.
A fault occurs if the Master/Slave configuration conditions do not allow master/slave resolution, as defined in the Master/ Slave Manual Config Enable and Master/Slave Manual Config Value bits in the Master/Slave Control Register of the local and remote link partners.


### 3.9.6.2 1000BASE-T

For 1000BASE-T links, the device and its link partner enter a training phase after completion of the auto-negotiation process. The links exchange idle symbols and use the information obtained from receiving this signal to set their adaptive filter coefficients.
These coefficients are used in the receiver to equalize the incoming signal, as well as eliminate signal impairments such as echo and cross-talk. Each side indicates completion of the training phase to its link partner by changing the encoding of the idle-symbols it transmits.
The link is established after both sides indicate completion of the training phase. Each side continues to send idle symbols whenever it has no data to transmit. The link is maintained as long as valid idle, data, or carrier extension symbols are received.
Status of both local and remote receivers is reported in the Local Receiver Status and Remote Receiver Status bits of the Master/Slave Status Register.
The device also provides an advanced Auto Link Breaker feature (only for 1000BASE-T links). Using this feature, the link can be taken down if the bit error rate (BER) exceeds the threshold defined in Link Break Threshold field of the Link Control Register. The error counting occurs during the idle time for a period commensurate with the specified BER. The Auto Link Breaker feature is enabled via the Link Break Enable bit of the Link Control Register.

### 3.9.6.3 100BASE-TX

For 100BASE-TX links, the device and its link partner begin transmitting idle symbols after completion of the auto-negotiation process. Each side continues sending idle symbols whenever it has no data to transmit. The link is maintained as long as valid idle symbols or data are received.

### 3.9.6.4 10BASE-T

For 10BASE-T links, the device and its link partner begin exchanging normal link pulses (NLPs) after completion of the auto-negotiation process. The device transmits an NLP every 16 ms and expects to receive an NLP every 8 to 24 ms . A link is maintained as long as NLPs are received.

### 3.9.6.5 Taking Down a Link

The device takes down an established link when the required conditions are no longer met. When a link is down, data transmission stops. For 10BASE-T links, the link is taken down after NLPs are no longer received. For 100BASE-TX and 1000BASE-T links, the link is taken down after valid idle codes are no longer received.
After a link is down, the device does the following:

- If auto-negotiation is enabled, the device re-enters the auto-negotiation phase and begins transmitting FLPs


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- If auto-negotiation is not enabled, the device transmits NLPs in 10BASE-T mode, and MLT-3s in 100BASE-TX mode.


### 3.9.7 SPEED OPTIMIZER

The Speed Optimize function is designed to resolve the issue of using auto-negotiation to establish a link on impaired cable plants

Examples of impaired cable plants for 1000BASE-T (Gigabit) connections include:

- Channel 2 twisted pair cable plant is broken
- Channel 3 twisted pair cable plant is broken
- Channel 2 and 3 twisted pair cable plants are broken
- Cable plant is too long

Examples of impaired cable plants for 100BASE-TX connections include:

- Cable plant is too long
- Using wrong cable plant (such as CAT-3)

The Speed Optimize function requires the MAC to support 1000/100/10 Mbps speeds, 1000/100 Mbps speeds, 1000/ 10 Mbps or 100/10 Mbps speeds.
If a link fails to establish after the link partners go through auto-negotiation several times at the HCD (Highest Common Denominator), the device advertises the next highest-allowable speed (as set in the Auto Negotiation Advertisement Register) and restarts auto-negotiation with the new speed.

When 1000BASE-T is advertised, the Speed Optimize function can change its advertised speed from 1000BASE-T to 100BASE-TX and from 100BASE-TX to 10BASE-T. When 100BASE-TX is advertised, the Speed Optimize function can change its advertised speed from 100BASE-TX to 10BASE-T. If a previous link has used the Speed Optimize function to establish a link, when the link goes down, the device begins advertising with all capable speeds.
The Speed Optimize function resets itself to advertise HCD/all speed capabilities after any of the following occurrences:

- Hardware reset
- Software reset
- While link partners exchange link pulses through the Speed Optimize process, the device does not receive link pulses for a period of few seconds
- After an established link goes down

The Speed Optimize function is enabled via Speed Optimize Enable bit in the 10/100 Mode Control/Status Register. When a link (with a speed slower than HCD) is being established through the Speed Optimize process, it is reported via the Speed Optimize Status bit in the User Status 2 Register.

### 3.9.8 LOOPBACK OPERATION

The local loopback mode is enabled by setting the Loopback bit of the Basic Control Register. In this mode, the scrambled transmit data (output of the scrambler) is looped into the receive logic (input of the descrambler). This mode is useful as a board diagnostic and serves as a quick functional verification of the device. The COL signal will be inactive in this mode, unless the Collision Test bit of the Basic Control Register is set.

Note: During transmission in local loopback mode, nothing is transmitted to the line and the transmitters are powered down.

### 3.9.9 IEEE 1149.1 (JTAG) BOUNDARY SCAN

The device includes an integrated JTAG boundary-scan test port for board-level testing. The interface consists of four pins (TDO, TDI, TCK and TMS) and includes a state machine, data register array, and an instruction register. The JTAG pins are described in Table 2-5, "JTAG Pins". The JTAG interface conforms to the IEEE Standard 1149.1-1990 Standard Test Access Port (TAP) and Boundary-Scan Architecture.
All input and output data is synchronous to the TCK test clock input. TAP input signals TMS and TDI are clocked into the test logic on the rising edge of TCK, while the output signal TDO is clocked on the falling edge.
The JTAG logic is reset via a hardware reset or when the TMS and TDI pins are high for five TCK periods.

The implemented IEEE 1149.1 instructions and their op codes are shown in Table 3-14.

## TABLE 3-14: IEEE 1149.1 OP CODES

| INSTRUCTION | OP CODE | COMMENT |
| :--- | :---: | :--- |
| Bypass | 111 | Mandatory Instruction |
| Sample/Preload | 010 | Mandatory Instruction |
| EXTEST | 000 | Mandatory Instruction |
| Clamp | 011 | Optional Instruction |
| HIGHZ | 100 | Optional Instruction |
| IDCODE | 001 | Optional Instruction |

$$
\begin{array}{ll}
\text { Note: } & \begin{array}{l}
\text { All digital I/O pins support IEEE } 1149.1 \text { operation. Analog pins and the XO pin do not support IEEE } 1149.1 \\
\text { operation. }
\end{array}
\end{array}
$$

### 3.9.10 ADVANCED FEATURES

The device implements several advanced features to enhance manageability of the Ethernet link. These features are detailed in the following sub-sections.

### 3.9.10.1 Crossover Indicators

The device reports crossed channels in the XOVER Resolution 0:1 and XOVER Resolution 2:3 fields of the User Status 2 Register. This feature is useful for trouble-shooting problems during network installation.

### 3.9.10.2 Polarity Inversion Indicators

The device automatically detects and corrects inverted signal polarity. This is reported in the polarity inversion bits (POLARITY_INV_3, POLARITY_INV_2, POLARITY_INV_1 and POLARITY_INV_0) of the User Status 1 Register.
The polarity inversion bit for Channel 1 (POLARITY_INV_1) is valid after auto-negotiation is complete as indicated by the Auto-Negotiate Complete bit of the Basic Status Register. The polarity inversion bits for Channels 0,2 and 3 (POLARITY_INV_0, POLARITY_INV_2, POLARITY_INV_3) are valid only after the link is up as indicated by the Link Status bit of the Basic Status Register.

### 3.9.10.3 Receive Error-Free Packets Counter

The quality of a link can be monitored by using the Receive Error-Free Packets Counter. The device counts the number of good packets received and reports a 48-bit value across 3 advanced registers: Receive Error-Free Packets Counter Low Register, Receive Error-Free Packets Counter Mid Register, and Receive Error-Free Packets Counter High Register. The Receive Error-Free Packets Counter Low Register latches the two other related counter registers and must always be read first. The Receive Error-Free Packets Counter High Register register must be read last, and will automatically clear the counter.

### 3.9.10.4 CRC Error Counter

This 48-bit counter counts the number of CRC errors detected. It's value can be read across 3 advanced registers: CRC Error Counter Low Register, CRC Error Counter Mid Register, and CRC Error Counter High Register. The CRC Error Counter Low Register latches the two other related counter registers and must always be read first. The CRC Error Counter High Register must be read last, and will automatically clear the counter.

### 3.9.10.5 Receive Error During Data Counter

This 16-bit counter counts the number of errors that occurred while data was being received. The value is read from the Receive Error During Data Counter Register.

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### 3.9.10.6 Receive Error During Idle Counter

This 16 -bit counter counts the number of errors that occurred during idle. The value is read from the Receive Error During Idle Counter Register register.

### 3.9.10.7 Transmitted Packets Counter

This 48-bit counter counts the number of packets that were transmitted. It's value can be read across 3 advanced registers: Transmit Packet Counter Low Register, Transmit Packet Counter Mid Register, and Transmit Packet Counter High Register. The Transmit Packet Counter Low Register latches the two other related counter registers and must always be read first. The Transmit Packet Counter High Register must be read last, and it will automatically clear the counter.

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### 3.10 Application Diagrams

This section provides typical application diagrams for the following:

- Simplified Application Diagram
- Power Supply \& Twisted Pair Interface Diagram


### 3.10.1 SIMPLIFIED APPLICATION DIAGRAM

## FIGURE 3-4: SIMPLIFIED APPLICATION DIAGRAM


3.10.2

POWER SUPPLY \& TWISTED PAIR INTERFACE DIAGRAM

FIGURE 3-5: POWER SUPPLY \& TWISTED PAIR INTERFACE DIAGRAM


### 4.0 REGISTER DESCRIPTIONS

This chapter describes the various control and status registers (CSRs). All registers follow the IEEE 802.3 (clause 22.2.4) management register set. All functionality and bit definitions comply with these standards. The IEEE 802.3 specified register index (in decimal) is included with each register definition, allowing for addressing of these registers via the Serial Management Interface (SMI) protocol.
The device registers are categorized into following groups:

- Primary PHY Registers
- Advanced PHY Registers


### 4.1 Register Nomenclature

Table 4-1 describes the register bit attributes used throughout this document.

## TABLE 4-1: REGISTER BIT TYPES

| Register Bit Type <br> Notation | Register Bit Description |
| :---: | :--- |
| R | Read: A register or bit with this attribute can be read. |
| W | Write: A register or bit with this attribute can be written. |
| RO | Read only: Read only. Writes have no effect. |
| WO | Write only: If a register or bit is write-only, reads will return unspecified data. |
| WC | Write One to Clear: writing a one clears the value. Writing a zero has no effect. |
| WAC | Write Anything to Clear: writing anything clears the value. |
| RC | Read to Clear: Contents is cleared after the read. Writes have no effect. |
| LL | Latch Low: Clear on read of register. |
| LH | Latch High: Clear on read of register. |
| SC | Self-Clearing: Contents is self-cleared after the being set. Writes of zero have no effect. <br> Contents can be read. |
| RO/LH | Read Only, Latch High: This mode is used by the Ethernet PHY registers. Bits with this <br> attribute will stay high until the bit is read. After it a read, the bit will remain high, but will <br> change to low if the condition that caused the bit to go high is removed. If the bit has not <br> been read the bit will remain high regardless of if its cause has been removed. |
| NASR | Not Affected by Software Reset: The state of NASR bits does not change on assertion <br> of a software reset. |
| X | Either a 1 or 0. |
| RESERVED | Reserved Field: Reserved fields must be written with zeros, unless otherwise indicated, <br> to ensure future compatibility. The value of reserved bits is not guaranteed on a read. |

Many of these register bit notations can be combined. Some examples of this are shown below:

- R/W: Can be written. Will return current setting on a read.
- R/WAC: Will return current setting on a read. Writing anything clears the bit.


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### 4.2 Primary PHY Registers

The primary PHY registers are accessed via the SMI bus. An index is used to access individual primary registers. Primary PHY register indexes are shown in Table 4-2, "PHY Control and Status Registers". Additional read-only advanced registers are indirectly accessible via the Advanced Register Address Port and Advanced Register Read Data Port. Section 4.3, "Advanced PHY Registers," on page 53 provides detailed information regarding the advanced registers.
Note 1: All unlisted register index values are not supported and should not be addressed.
2: The NASR (Not Affected by Software Reset) designation is only applicable when the PHY Soft Reset (RESET) bit of the Basic Control Register is set.

## TABLE 4-2: PHY CONTROL AND STATUS REGISTERS

| Index <br> (In Decimal) | Register Name |
| :---: | :---: |
| 0 | Basic Control Register |
| 1 | Basic Status Register |
| 2 | PHY Identifier 1 Register |
| 3 | PHY Identifier 2 Register |
| 4 | Auto Negotiation Advertisement Register |
| 5 | Auto Negotiation Link Partner Ability Register |
| 6 | Auto Negotiation Expansion Register |
| 7 | Auto Negotiation Next Page TX Register |
| 8 | Auto Negotiation Next Page RX Register |
| 9 | Master/Slave Control Register |
| 10 | Master/Slave Status Register |
| 15 | Extended Status Register |
| 16 | Link Control Register |
| 17 | 10/100 Mode Control/Status Register |
| 18 | 10/100 Special Modes Register |
| 19 | Extended Mode Control/Status Register |
| 20 | Advanced Register Address Port |
| 21 | Advanced Register Read Data Port |
| 27 | Control / Status Indications Register |
| 29 | Interrupt Source Flags Register |
| 30 | Interrupt Mask Register |
| 31 | PHY Special Control / Status Register |

### 4.2.1 BASIC CONTROL REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15 | PHY Soft Reset (RESET) <br> $1=$ PHY software reset. This bit is self-clearing. When setting this bit, do not set other bits in this register. The configuration is set from the register bit values as described in Section 3.6.2, "Software Reset," on page 21. <br> Note: The PHY will be in the normal mode after a PHY software reset. | $\begin{aligned} & \mathrm{R} / \mathrm{W} \\ & \mathrm{SC} \end{aligned}$ | Ob |
| 14 | Loopback <br> 0 = normal operation <br> 1 = loopback mode | R/W | Ob |
| 13 | Speed Select[0] <br> Together with Speed Select[1], sets speed per the following table: <br> [Speed Select1][Speed Select 0] $00=10 \mathrm{Mbps}$ <br> $01=100 \mathrm{Mbps}$ <br> $10=1000 \mathrm{Mbps}$ <br> 11 = Reserved <br> Note: Ignored if the Auto-Negotiation Enable bit of this register is 1. | R/W | Note 4-1 |
| 12 | Auto-Negotiation Enable <br> 0 = disable auto-negotiate process <br> 1 = enable auto-negotiate process (overrides the Speed Select[0], Speed Select[1], and Duplex Mode bits of this register) | R/W | Note 4-1 |
| 11 | Power Down <br> 0 = normal operation <br> 1 = General power down mode <br> Note: Auto-Negotiation Enable must be cleared before setting this bit. | R/W | Ob |
| 10 | Isolate <br> 0 = normal operation <br> 1 = electrical isolation of PHY from GMII | R/W | Ob |
| 9 | Restart Auto-Negotiate $0=$ normal operation 1 = restart auto-negotiate process Note: Bit is self-clearing. | $\mathrm{R} / \mathrm{W}$ | Ob |
| 8 | Duplex Mode <br> 0 = half duplex <br> 1 = full duplex <br> Note: Ignored if the Auto-Negotiation Enable bit of this register is 1. | R/W | Note 4-1 |
| 7 | Collision Test $0=$ disable COL test 1 = enable COL test | R/W | Ob |
| 6 | Speed Select[1] <br> See description for Speed Select[0] for details. | RO | Note 4-1 |
| 5:0 | RESERVED | RO | - |

Note 4-1 The default is determined by the CONFIG[3:2] pins as described in Section 3.8.1.2.3, "Configuration Bits Impacted by the Mode of Operation," on page 25"

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### 4.2.2 BASIC STATUS REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15 | $\begin{aligned} & \text { 100BASE-T4 } \\ & 0=\text { no T4 ability } \\ & 1=\mathrm{T} 4 \text { able } \end{aligned}$ | RO | Ob |
| 14 | 100BASE-TX Full Duplex <br> $0=$ no TX full duplex ability <br> 1 = TX with full duplex | RO | 1b |
| 13 | 100BASE-TX Half Duplex <br> $0=$ no TX half duplex ability <br> $1=$ TX with half duplex | RO | 1b |
| 12 | 10BASE-T Full Duplex <br> $0=$ no 10Mbps with full duplex ability <br> $1=10 \mathrm{Mbps}$ with full duplex | RO | 1b |
| 11 | 10BASE-T Half Duplex <br> $0=$ no 10 Mbps with half duplex ability <br> $1=10 \mathrm{Mbps}$ with half duplex | RO | 1b |
| 10 | 100BASE-T2 Full Duplex <br> $0=$ PHY not able to perform full duplex 100BASE-T2 <br> $1=$ PHY able to perform full duplex 100BASE-T2 | RO | Ob |
| 9 | 100BASE-T2 Half Duplex <br> $0=$ PHY not able to perform half duplex 100BASE-T2 <br> $1=$ PHY able to perform half duplex 100BASE-T2 | RO | Ob |
| 8 | Extended Status <br> $0=$ no extended status information in register 15 <br> $1=$ extended status information in register 15 | RO | 1b |
| 7:6 | RESERVED | RO | - |
| 5 | Auto-Negotiate Complete <br> $0=$ auto-negotiate process not completed <br> 1 = auto-negotiate process completed | RO | Ob |
| 4 | Remote Fault <br> $0=$ no remote fault <br> 1 = remote fault condition detected | RO/LH | Ob |
| 3 | Auto-Negotiate Ability <br> $0=$ unable to perform auto-negotiation function <br> 1 = able to perform auto-negotiation function | RO | 1b |
| 2 | Link Status <br> $0=$ link is down <br> 1 = link is up | RO/LL | Ob |
| 1 | Jabber Detect <br> $0=$ no jabber condition detected <br> 1 = jabber condition detected | RO/LH | Ob |
| 0 | Extended Capabilities <br> $0=$ does not support extended capabilities registers <br> 1 = supports extended capabilities registers | RO | 1b |

### 4.2.3 PHY IDENTIFIER 1 REGISTER

Index (In Decimal): $2 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15:0 | PHY ID Number <br> Assigned to the 3rd through 18th bits of the Organizationally Unique <br> Identifier (OUI), respectively. OUI=00800Fh | R/W | 0007h |

### 4.2.4 PHY IDENTIFIER 2 REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15:10 | PHY ID Number <br> Assigned to the 19th through 24th bits of the OUI. | R/W | C0h |
| $9: 4$ | Model Number <br> Six-bit manufacturer's model number. | R/W | 0Eh |
| $3: 0$ | Revision Number <br> Four-bit manufacturer's revision number. | R/W | Note 4-2 |

Note 4-2 The default value of this field will vary dependent on the silicon revision number.

### 4.2.5 AUTO NEGOTIATION ADVERTISEMENT REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | Next Page <br> $0=$ no next page ability <br> $1=$ next page capable <br> Note: This device does not support next page ability. | $\mathrm{R} / \mathrm{W}$ | Ob |
| 14 | RESERVED | RO | - |
| 13 | Remote Fault <br> $0=$ no remote fault <br> $1=$ remote fault detected | $\mathrm{R} / \mathrm{W}$ | 0 b |
| 12 | RESERVED | RO | - |

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| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 11 | Asymmetric Pause <br> $0=$ Asymmetrical pause direction is not supported by MAC <br> $1=$ Asymmetrical pause direction is supported by MAC | $\mathrm{R} / \mathrm{W}$ | Ob |
| 10 | Pause Operation (PAUSE) <br> $0=$ Pause operation is not supported by MAC <br> $1=$ Pause operation is supported by MAC | R/W | Note 4-3 |
| 9 | RESERVED | RO |  |
| 8 | $100 B A S E-T X ~ F u l l ~ D u p l e x ~$ <br> $0=$ no TX full duplex ability <br> $1=$ TX with full duplex | R/W | Note 4-4 |
| 7 | $100 B A S E-T X$ <br> $0=$ no TX ability <br> $1=$ TX able | R/W | Note 4-4 |
| 6 | $10 B A S E-T$ Full Duplex <br> $0=$ no 10Mbps with full duplex ability <br> $1=10 M b p s ~ w i t h ~ f u l l ~ d u p l e x ~$ | Note 4-4 |  |
| 5 | $10 B A S E-T$ <br> $0=$ no 10Mbps ability <br> $1=10 M b p s ~ a b l e$ | R/W | 00001b |
| $4: 0$ | Selector Field <br> $00001=$ IEEE 802.3 | Note 4-4 |  |

Note 4-3 The default is determined by the CONFIG1 pin as described in Section 3.8.1.2.1, "Configuring the SMI Address (CONFIG[1:0])," on page 24
Note 4-4 The default is determined by the CONFIG[3:2] pins as described in Section 3.8.1.2.3, "Configuration Bits Impacted by the Mode of Operation," on page 25.

### 4.2.6 AUTO NEGOTIATION LINK PARTNER ABILITY REGISTER

Index (In Decimal): 5
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | Next Page <br> $0=$ no next page ability <br> $1=$ next page capable | RO | Ob |
| 14 | Acknowledge <br> $0=$ link code word not yet received <br> $1=$ link code word received from partner | RO | Ob |
| 13 | Remote Fault <br> $0=$ no remote fault <br> $1=$ remote fault detected | RO | Ob |
| $12: 11$ | RESERVED | RO | - |
| 10 | Pause Operation (PAUSE) <br> $0=$ Pause Operation is not supported by remote MAC <br> $1=$ Pause Operation is supported by remote MAC | Ob |  |


| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 9 | 100BASE-T4 $0=\text { no T4 ability }$ $1 \text { = T4 able }$ <br> Note: This PHY does not support T4 ability. | RO | Ob |
| 8 | 100BASE-TX Full Duplex <br> $0=$ no TX full duplex ability <br> $1=$ TX with full duplex | RO | Ob |
| 7 | $\begin{aligned} & \text { 100BASE-TX } \\ & 0=\text { no TX ability } \\ & 1=\text { TX able } \end{aligned}$ | RO | Ob |
| 6 | 10BASE-T Full Duplex <br> $0=$ no 10Mbps with full duplex ability <br> $1=10 \mathrm{Mbps}$ with full duplex | RO | Ob |
| 5 | 10BASE-T <br> $0=$ no 10Mbps ability <br> $1=10 \mathrm{Mbps}$ able | RO | Ob |
| 4:0 | Selector Field $00001=$ IEEE 802.3 | RO | 00001b |

### 4.2.7 AUTO NEGOTIATION EXPANSION REGISTER

Index (In Decimal): 6
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 5$ | RESERVED | RO | - |
| 4 | Parallel Detection Fault <br> $0=$ no fault detected by parallel detection logic <br> $1=$ fault detected by parallel detection logic | $\mathrm{RO} / \mathrm{LH}$ | Ob |
| 3 | Link Partner Next Page Able <br> $0=$ link partner does not have next page ability <br> $1=$ link partner has next page ability | RO | Ob |
| 2 | Next Page Able <br> $0=$ local device does not have next page ability <br> $1=$ local device has next page ability | RO | RO |
| 1 | Page Received <br> $0=$ new page not yet received <br> $1=$ new page received | RO | Ob |
| 0 | Link Partner Auto-Negotiation Able <br> $0=$ link partner does not have auto-negotiation ability <br> $1=$ link partner has auto-negotiation ability | Ob |  |

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### 4.2.8 AUTO NEGOTIATION NEXT PAGE TX REGISTER

Index (In Decimal): $7 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15 | Next Page <br> $0=$ no next page ability <br> 1 = next page capable | R/W | Ob |
| 14 | RESERVED | RO | - |
| 13 | Message Page <br> $0=$ unformatted page <br> 1 = message page | R/W | 1b |
| 12 | Acknowledge 2 <br> $0=$ device cannot comply with message <br> 1 = device will comply with message | R/W | Ob |
| 11 | Toggle <br> $0=$ previous value was HIGH <br> 1 = previous value was LOW | RO | Ob |
| 10:0 | Message Code Message/Unformatted Code Field | RW | $\begin{gathered} 00 \\ 0000 \\ 0001 \mathrm{~b} \end{gathered}$ |

### 4.2.9 AUTO NEGOTIATION NEXT PAGE RX REGISTER

Index (In Decimal): $8 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | Next Page <br> $0=$ no next page ability <br> $1=$ next page capable | RO | Ob |
| 14 | Acknowledge <br> $0=$ Link code word not yet received from partner <br> $1=$ Link code word received from partner | RO | Ob |
| 13 | Message Page <br> $0=$ unformatted page <br> $1=$ message page | RO | 1 b |
| 12 | Acknowledge 2 <br> $0=$ device cannot comply with message <br> $1=$ device will comply with message | RO | 0 RO |
| 11 | Toggle <br> $0=$ previous value was HIGH <br> $1=$ previous value was LOW | RO | 0 Ob |
| $10: 0$ | Message Code <br> Message/Unformatted Code Field |  |  |

4.2.10 MASTER/SLAVE CONTROL REGISTER

$$
\text { Index (In Decimal): } 9 \quad \text { Size: } 16 \text { bits }
$$

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15:13 | Test Mode <br> $000=$ Normal mode <br> 001 = Test Mode 1 - Transmit waveform test <br> $010=$ Test Mode 2 - Transmit jitter test in Master mode <br> 011 = Test Mode 3 - Transmit jitter test in Slave mode <br> $100=$ Test Mode 4 - Transmitter distortion test <br> 101 = Reserved <br> 110 = Reserved <br> 111 = Reserved <br> Note: Setting these bits may prevent correct link partner connection if both the device PHY and link partner PHY are set as masters. | R/W | 000b |
| 12 | Master/Slave Manual Config Enable <br> 0 = disable MASTER-SLAVE manual configuration value <br> 1 = enable MASTER-SLAVE manual configuration value | R/W | Note 4-5 |
| 11 | Master/Slave Manual Config Value <br> Active only when the Master/Slave Manual Config Enable bit of this register is 1 . $\begin{aligned} & 0=\text { Slave } \\ & 1=\text { Master } \end{aligned}$ | R/W | Note 4-5 |
| 10 | Port Type <br> Active only when the Master/Slave Manual Config Enable bit of this register is 0 . <br> $0=$ Single port device <br> 1 = Multiport device | R/W | Note 4-5 |
| 9 | 1000BASE-T Full Duplex <br> $0=$ advertise PHY is not 1000BASE-T full duplex capable <br> 1 = advertise PHY is 1000BASE-T full duplex capable | R/W | Note 4-5 |
| 8 | 1000BASE-T Half Duplex <br> 0 = advertise PHY is not 1000BASE-T half duplex capable <br> 1 = advertise PHY is 1000BASE-T half duplex capable | R/W | Note 4-5 |
| 7:0 | RESERVED | RO | - |

Note 4-5 The default is determined by the CONFIG[3:2] pins as described in Section 3.8.1.2.3, "Configuration Bits Impacted by the Mode of Operation," on page 25.

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### 4.2.11 MASTER/SLAVE STATUS REGISTER

Index (In Decimal): 10 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | Master/Slave Configuration Fault <br> $0=$ No MASTER-SLAVE configuration fault detected <br> $1=$ MASTER-SLAVE configuration fault detected | $\mathrm{RO} / \mathrm{LH}$ | Ob |
| 14 | Master/Slave Configuration Resolution <br> $0=$ Local PHY configuration resolved to SLAVE <br> $1=$ Local PHY configuration resolved to MASTER | RO | Ob |
| 13 | Local Receiver Status <br> $0=$ Local Receiver not OK <br> $1=$ Local Receiver OK | RO | Ob |
| 12 | Remote Receiver Status <br> $0=$ Remote Receiver not OK <br> $1=$ Remote Receiver OK | RO | Ob |
| 11 | LP 1000T FD <br> This bit is valid only when the Page Received bit of the Auto Negotiation <br> Expansion Register is 1. <br> $0=$ Link Partner is not capable of 1000BASE-T full duplex <br> $1=$ Link Partner is capable of 1000BASE-T full duplex | RO | Ob |
| 10 | LP 1000T HD <br> This bit is valid only when the Page Received bit of the Auto Negotiation <br> Expansion Register is 1. <br> $0=$ Link Partner is not capable of 1000BASE-T half duplex <br> $1=$ Link Partner is capable of 1000BASE-T half duplex | Ob |  |
| $9: 8$ | RESERVED | RO | RO |
| $7: 0$ | Idle Error Count <br> Cumulative count of the errors detected when the receiver is receiving idles. <br> These bits are reset to all zeros when the error count is read by the <br> management function or upon execution of a hardware reset, software reset, <br> or logical reset. This field is held at all ones in case of over-flow. <br> This field can be used to trigger an interrupt upon overflow. Refer to Section <br> $3.5, ~ " I n t e r r u p t ~ M a n a g e m e n t, " ~ o n ~ p a g e ~ 20 ~ f o r ~ a d d i t i o n a l ~ i n f o r m a t i o n . ~$ | 00h |  |

### 4.2.12 EXTENDED STATUS REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | 1000 BASE-X Full Duplex <br> $0=$ PHY not able to perform full duplex 1000BASE-X <br> $1=$ PHY able to perform full duplex 1000BASE-X | RO | Ob |
| 14 | 1000 BASE-X Half Duplex <br> $0=$ PHY not able to perform half duplex 1000BASE-X <br> $1=$ PHY able to perform half duplex 1000BASE-X | RO | Ob |
| 13 | 1000 BASE-T Full Duplex <br> $0=$ PHY not able to perform full duplex 1000BASE-T <br> $1=$ PHY able to perform full duplex 1000BASE-T | RO | 1 b |
| 12 | 1000 BASE-T Half Duplex <br> $0=$ PHY not able to perform half duplex 1000BASE-T <br> $1=$ PHY able to perform half duplex 1000BASE-T | RO | 1 b |
| $11: 0$ | RESERVED | RO | - |

### 4.2.13 LINK CONTROL REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 10$ | RESERVED | RO | - |
| $9: 8$ | Speed Optimize Control <br> This register sets the number of Auto Negotiation attempts before the Speed <br> Optimize mechanism reduces the advertised speed. <br> $00=7$ attempts <br> $01=5$ attempts <br> $10=4$ attempts <br> $11=3$ attempts <br> Note: $\quad$Refer to Section 3.9.7, "Speed Optimizer," on page 30 for additional <br> information. | $\mathrm{R} / \mathrm{W}$ | OOb |
| $7: 6$ | RESERVED | RO |  |
| $5: 4$ | Link Break Threshold <br> Idle error threshold for failing the link, if Link break in enabled. <br> $00=$ link break threshold is 10E-8. <br> $01=$ link break threshold is 10E-9. <br> $10=$ link break threshold is 10E-10. <br> $11=$ link break threshold is 10E-11 | - |  |
| 3 | Link Break Enable <br> $0=$ link break is disabled <br> $1=$ link break is enabled | $\mathrm{R} / \mathrm{W}$ | Ob |

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| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 2 | Power Optimization Disable <br> $0=$ Automatic power optimization is enabled <br> $1=$ Automatic power optimization is disabled (power consumption is <br> maximum) | R/W | Ob |
| 1 | RESERVED | RO | - |
| 0 | LRST <br> Logic reset. This bit generates a reset that put all the logic into a known <br> state, but DOES NOT affect the register sets and 10/100 circuits. This bit is <br> NOT a self-clearing bit. Writing "1" to this bit generates synchronous reset. | RO | - |

### 4.2.14 10/100 MODE CONTROL/STATUS REGISTER

Index (In Decimal): $17 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15 | EDSHORT <br> Energy Detect Short detection mode $0=$ Normal detect mode <br> 1 = Short detect mode | R/W | Ob |
| 14 | FASTRIP <br> 10BASE-T fast mode <br> 0 = normal operation <br> 1 = activates PHYT_10 test mode | R/W | Ob |
| 13 | EDPWRDOWN <br> Enable the Energy Detect Power-Down mode $0=$ Energy Detect Power-Down is disabled 1 = Energy Detect Power-Down is enabled | R/W | 0b |
| 12 | ED Power Down Mode <br> Select energy detect power down mode $0=$ ED power down mode without NLP transmission 1 = ED power down mode with NLP transmission | R/W | Ob |
| 11:8 | RESERVED | RO | - |
| 7 | Speed Optimize Enable <br> 0 = Disable Speed Optimize <br> 1 = Enable Speed Optimize <br> Note: Refer to Section 3.9.7, "Speed Optimizer," on page 30 for additional information. | R/W | Ob |
| 6 | AutoNeg NP Enable <br> $0=$ Next page is disabled in the auto-negotiation process <br> 1 = Next page is enabled in the auto-negotiation process | R/W | 1b |
| 5 | Auto MDIX Disable <br> $0=$ Auto Xover is enabled <br> 1 = Auto Xover is disabled selection is done manually | R/W | Ob |


| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 4 | Auto Next Page Disable <br> Setting this bit disables automatic next page exchange in 1000BASE-T. <br> Advertising of next pages then depends on the value of the Next Page bit of <br> the Auto Negotiation Advertisement Register. In this case, if Next Page is <br> cleared, only the base page is sent. <br> $0=$ Normal operation <br> $1=$ Automatic next page is disabled | R/W | Ob |
| $3: 2$ | RESERVED | RO | - RO |
| 1 | ENERGYON <br> This bit indicates whether energy is detected on the line. It is reset to "1" by <br> a hardware reset. When a software reset is asserted, this bit is cleared. If <br> this bit was set prior to a software reset, it will cause the INT7 bit of the <br> Interrupt Source Flags Register to be set. Therefore, after a software or <br> hardware reset, the INT7 bit should be cleared by writing a "1" to it. <br> Refer to Section 3.5, "Interrupt Management," on page 20 for additional <br> ENERGYON information. | 1b |  |
| 0 | Semi Crossover Enable <br> Setting this register enables semi cross over. <br> $0=$ Disable Semi cross over <br> $1=$ Enable Semi cross over <br> Note:Refer to Section 3.2, "HP Auto-MDIX," on page 16 for additional <br> information. | R/W | 0b |

4.2.15 10/100 SPECIAL MODES REGISTER
Index (In Decimal): $18 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | Enable RXDV Early Assertion <br> Setting this bit enables early assertion of RXDV in 10BASE-T. RXDV is <br> asserted before the SFD. <br> $0=$ Disable <br> $1=$ Enable | R/W | Ob |
| 14 | 10BT HD Loopback Disable <br> Setting this bit disables MII loopback in 10BASE-T half duplex mode. <br> $0=$ Enable operational loopback <br> $1=$ Disable operational loopback* <br> *Modern implementations of the MAC require this bit be set to 1b for <br> 10BASE-T half duplex communication. Legacy implementation of the MAC <br> may require this bit be set to 0b for 10BASE-T half duplex communication. <br> This bit affects only 10BASE-T half duplex PHY mode. This bit may be set <br> to either Ob or 1b with no affect on 10BASE-T full duplex, 100BASE-TX, or <br> 1000BASE-T PHY operation. | NASR | Ob |
| $13: 8$ | RESERVED | RO |  |
| 7 | CRC Error Counter Data Source <br> Setting this bit changes the data source of the 1000BASE-T CRC error <br> counter. <br> $0=$ Data source in 1000BASE-T received data <br> 1 = Data source in 1000BASE-T transmitted data | R/W | Ob |

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| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 6 | MCLK25EN <br> Enable an 25Mhz MAC clock output. <br> $0=125 \mathrm{MHz}$ <br> $1=25 \mathrm{MHz}$ | RO | Ob |
| 5 | RESERVED | RO | - |
| $4: 0$ | PHYADD[4:0] <br> The PHY Address is used for the SMI address and for the initialization of the <br> Cipher (Scrambler) key. | R/W <br> NASR | Note 4-6 |

Note 4-6 The default is determined by the CONFIG[1:0] pins as described in Section 3.8.1.2.1, "Configuring the SMI Address (CONFIG[1:0])," on page 24.
4.2.16 EXTENDED MODE CONTROL/STATUS REGISTER

Index (In Decimal): $19 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15:11 | MOD <br> Configures mode of operation. Refer to Section 3.8.1.2.2, "Configuring the Mode of Operation (CONFIG[3:2])," on page 24 for details. <br> Note: The MOD bits should not be modified and must be preserved when writing to this register. | $\begin{gathered} \text { R/W } \\ \text { NASR } \end{gathered}$ | Note 4-7 |
| 10:9 | Transmitter FIFO Depth $00=4$ bytes $01=5$ bytes $10=6$ bytes $11=7$ bytes | R/W | 00b |
| 8:6 | RESERVED <br> These bits must be written as 000b. | RO | - |
| 5:4 | LED_MODE[1:0] <br> $00=$-Reserved <br> 01 = LED mode 1 <br> $10=$ LED mode 2 <br> 11 = LED mode 3 <br> Refer to Section 3.9.1, "LEDs," on page 26 for additional information. | R/W | 11b |
| 3 | RESERVED <br> This bit must be written as 1b | R/W | 1b |
| 2 | MDI/MDI-X 0:1 <br> Selects between MDI and MDI-X for channel 0 and channel 1 only if the Auto MDIX Disable bit of the 10/100 Mode Control/Status Register is 1. $\begin{aligned} & 0=\mathrm{MDI} \\ & 1=\mathrm{MDI}-\mathrm{X} \end{aligned}$ | RW | Ob |
| 1 | MDI/MDI-X 2:3 <br> Selects between MDI and MDI-X for channel 2 and channel 3 only if the Auto MDIX Disable bit of the 10/100 Mode Control/Status Register is 1. $\begin{aligned} & 0=\mathrm{MDI} \\ & 1=\mathrm{MDI}-X \end{aligned}$ | RW | Ob |


| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 0 | CONDITIONAL PARALLEL DETECT <br> $0=$ Parallel detect. (Auto Negotiation Advertisement Register is ignored.) <br> $1=$ Conditional Parallel Detect only at the speed advertised in the Auto <br> Negotiation Advertisement Register. <br> 10BASE-T half duplex (10BASE-T bit =1) <br> 100BASE-TX half duplex (100BASE-TX bit =1) | RW | 0b |

Note 4-7 The default mode is determined by the CONFIG[3:2] pins as described in Section 3.8.1.2.2, "Configuring the Mode of Operation (CONFIG[3:2])," on page 24

### 4.2.17 ADVANCED REGISTER ADDRESS PORT

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | Read <br> When this bit is set to 1, the contents of the advanced register selected by <br> the Register Address field are latched to the Advanced Register Read Data <br> Port. This bit is self-cleared. | SC | Ob |
| $14: 7$ | RESERVED <br> Must be written with 00000011b for proper operation. <br> The values of RESERVED bits are not guaranteed on a read. | R/W | - |
| $6: 0$ | Register Address <br> The address of the Advanced Register being accessed (0-12). | RO | 0000000 b |

Note: $\quad$ Refer to Section 4.3, "Advanced PHY Registers," on page 53 for additional information on the advanced register set.

### 4.2.18 ADVANCED REGISTER READ DATA PORT

Index (In Decimal): 21 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | Read <br> Data read from the Advanced Register selected via the Advanced Register <br> Address Port. | RO | 0000h |

Note: Refer to Section 4.3, "Advanced PHY Registers," on page 53 for additional information on the advanced register set.

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4.2.19 CONTROL / STATUS INDICATIONS REGISTER

Index (In Decimal): $27 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 5$ | RESERVED | RO | - |
| 4 | XPOL <br> Polarity state of the 10BASE-T: <br> $0=$ Normal polarity <br> $1=$ Reversed polarity | RO | Ob |
| $3: 0$ | RESERVED | RO | - |

4.2.20 INTERRUPT SOURCE FLAGS REGISTER

Index (In Decimal): 29
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 11$ | RESERVED | RO | - |
| 10 | INT10 <br> $0=$ Not source of interrupt <br> $1=$ Transmitter Elastic Buffer Overflow | LH/WC | Ob |
| 9 | INT9 <br> $0=$ Not source of interrupt <br> $1=$ Transmitter Elastic Buffer Underflow | LH/WC | Ob |
| 8 | INT8 <br> $0=$ Not source of interrupt <br> $1=$ Idle Error Count Overflow | LH/WC | Ob |
| 7 | INT7 <br> $0=$ Not source of interrupt <br> $1=$ ENERGYON generated <br> This bit is set when there is a "0" to "1" transition of the ENERGYON bit in <br> the 10/100 Mode Control/Status Register. This occurs when transitioning <br> from no energy detected to energy detected, or vice versa. | Ob |  |
| 6 | INT6 <br> $0=$ Not source of interrupt <br> $1=$ Auto-Negotiation complete | LH/WC | Ob |
| 5 | INT5 <br> $0=$ Not source of interrupt <br> $1=$ Remote Fault Detected | LH/WC | Ob |
| 4 | INT4 <br> $0=$ Not source of interrupt <br> $1=$ Link Down (link status negated) | Ob |  |


| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 3 | INT3 <br> $0=$ Not source of interrupt <br> $1=$ Auto-Negotiation LP Acknowledged | LH/WC | Ob |
| 2 | INT2 <br> $0=$ Not source of interrupt <br> $1=$ Parallel Detection Fault or Master/Slave Configuration Fault | LH/WC | Ob |
| 1 | INT1 <br> $0=$ Not source of interrupt <br> $1=$ Auto-Negotiation Page Received | LH/WC | Ob |
| 0 | RESERVED | RO | - |

Note: Refer to Section 3.5, "Interrupt Management," on page 20 for additional information.

### 4.2.21 INTERRUPT MASK REGISTER

Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :---: | :---: | :---: |
| 15:11 | RESERVED | RO | - |
| 10 | INT10_EN <br> Transmitter Elastic Buffer Overflow interrupt enable. <br> $0=$ interrupt source is masked <br> $1=$ interrupt source is enabled | R/W | Ob |
| 9 | INT9_EN <br> Transmitter Elastic Buffer Underflow interrupt enable. $0=$ interrupt source is masked <br> 1 = interrupt source is enabled | R/W | Ob |
| 8 | INT8_EN <br> Idle Error Count Overflow interrupt enable. <br> 0 = interrupt source is masked <br> 1 = interrupt source is enabled | R/W | Ob |
| 7 | INT7_EN <br> ENE $\bar{R} G Y O N$ interrupt enable <br> $0=$ interrupt source is masked <br> $1=$ interrupt source is enabled | R/W | Ob |
| 6 | INT6_EN <br> Auto-Negotiation interrupt enable $0=$ interrupt source is masked. <br> 1 = interrupt source is enabled. | R/W | Ob |
| 5 | INT5_EN <br> Remōte Fault Detected interrupt enable. <br> $0=$ interrupt source is masked. <br> 1 = interrupt source is enabled. | R/W | Ob |
| 4 | INT4_EN <br> Link Down (Link status negated) interrupt enable. <br> $0=$ interrupt source is masked. <br> $1=$ interrupt source is enabled. | R/W | Ob |

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| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 3 | INT3_EN <br> Auto-Negotiation LP Acknowledge interrupt enable. <br> $0=$ interrupt source is masked. <br> $1=$ interrupt source is enabled. | R/W | Ob |
| 2 | INT2_EN <br> Parallel Detection Fault or Master/Slave Configuration Fault interrupt enable. <br> $0=$ interrupt source is masked. <br> $1=$ interrupt source is enabled. | R/W | Ob |
| 1 | INT1_EN <br> Auto-Negotiation Page Received interrupt enable. <br> $0=$ interrupt source is masked. <br> $1=$ interrupt source is enabled. | R/W | Ob |
| 0 | RESERVED | RO | - |

Note: Refer to Section 3.5, "Interrupt Management," on page 20 for additional information.
4.2.22 PHY SPECIAL CONTROL / STATUS REGISTER

Index (In Decimal): $31 \quad$ Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 13$ | RESERVED | RO | - |
| 12 | Auto-negotiation done indication <br> $0=$ Auto-negotiation is not done or disabled (or not active) <br> $1=$ Auto-negotiation is done <br> Note: $\quad$ This is a duplicate of register 1.5, however reads to register 31 do <br> not clear status bits. | RO | Ob |
| $11: 5$ | RESERVED | RO | - |
| $4: 2$ | Speed Indication <br> HCDSPEED value: <br> [001]=10Mbps Half-duplex <br> $[101]=10 M b p s ~ F u l l-d u p l e x ~$ <br> $[010]=100 B A S E-T X ~ H a l f-d u p l e x ~$ <br> $[110]=100 B A S E-T X ~ F u l l-d u p l e x ~$ | Note 4-8 |  |
| $1: 0$ | RESERVED RO |  |  |

Note 4-8 Set according to the results of Auto-Negotiation.

### 4.3 Advanced PHY Registers

The advanced PHY registers are accessed using the following procedure:

1. Write to the Advanced Register Address Port with the Read bit set high, and the address of the desired advanced register in the Register Address field.
2. Read the contents of the selected register from Advanced Register Read Data Port.

Note: The advanced registers cannot be written. All advanced registers are comprised of read-only (RO), or read-to-clear (RC) bits.

### 4.3.1 ADVANCED REGISTER MAPPING

TABLE 4-3: ADVANCED REGISTER MAPPING

| REGISTER <br> NUMBER | REGISTER NAME |
| :---: | :--- |
| U0 | User Status 1 Register |
| U1 | User Status 2 Register |
| U2 | Receive Error-Free Packets Counter High Register |
| U3 | Receive Error-Free Packets Counter Mid Register |
| U4 | Receive Error-Free Packets Counter Low Register |
| U5 | CRC Error Counter High Register |
| U6 | CRC Error Counter Mid Register |
| U7 |  |
| U8 | Receive Error During Data Counter Register |
| U9 | Receive Error During Idle Counter Register |
| U10 | Transmit Packet Counter High Register |
| U11 | Transmit Packet Counter Mid Register |
| U12 | Transmit Packet Counter Low Register |

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### 4.3.2 USER STATUS 1 REGISTER

Index: U0 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | PLLREADY <br> $0=$ PLL is not locked <br> $1=$ PLL is locked | RO | Ob |
| 14 | POLARITY INV_3 <br> This bit indicates reverse polarity on channel 3 when operating in <br> 1000BASE-T mode. <br> $0=$ Channel 3 polarity is correct <br> $1=$ Channel 3 polarity is reversed | RO | Ob |
| 13 | POLARITY_INV_2 <br> This bit indicates reverse polarity on channel 2 when operating in <br> $1000 B A S E-T ~ m o d e . ~$ |  |  |
| $0=$ Channel 2 polarity is correct <br> $1=$ Channel 2 polarity is reversed | RO | Ob |  |
| 12 | POLARITY_INV_1 <br> This bit indicates reverse polarity on channel 1 when operating in <br> $1000 B A S E-T$ mode. <br> $0=$ Channel 1 polarity is correct <br> $1=$ Channel 1 polarity is reversed | RO | Ob |
| 11 | POLARITY_INV_0 <br> This bit indicates reverse polarity on channel 0 when operating in <br> $1000 B A S E-T ~ m o d e . ~$ <br> $0=$ Channel 0 polarity is correct <br> $1=$ Channel 0 polarity is reversed | RO | Ob |
| $10: 0$ | RESERVED | RO |  |

### 4.3.3 USER STATUS 2 REGISTER

Index:
U1
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15 | XOVER Resolution 0:1 <br> $0=$ Channel 0 and Channel 1 resolved as MDI. <br> $1=$ Channel 0 and Channel 1 resolved as MDI-X. | RO | Ob |
| 14 | XOVER Resolution 2:3 <br> $0=$ Channel 2 and Channel 3 resolved as MDI. <br> $1=$ Channel 2 and Channel 3 resolved as MDI-X. | RO | Ob |
| 13 | Speed Optimize Status <br> When set, indicates the link was established using the Speed Optimize <br> mechanism. <br> Note:Refer to Section 3.9.7, "Speed Optimizer," on page 30 for additional <br> information. | RO | Ob |
| $12: 0$ | RESERVED | RO | - |

### 4.3.4 RECEIVE ERROR-FREE PACKETS COUNTER HIGH REGISTER

Index:
U2
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | RCVGPKT[47:32] <br> Counts the received error-free packets. <br> Contains the 16 upper bits of the 48-bit counter. <br> Reading this register resets all bits in the Receive Error-Free Packets <br> Counter. | RO/ <br> RC | 0000 h |

Note: The 48-bit receive error-free packets counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: Receive Error-Free Packets Counter Low Register, Receive Error-Free Packets Counter Mid Register, Receive Error-Free Packets Counter High Register. After reading the high register, the counter will be automatically cleared.

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### 4.3.5 RECEIVE ERROR-FREE PACKETS COUNTER MID REGISTER

Index: U3 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | RCVGPKT[31:16] <br> Counts the received error-free packets. <br> Contains the 16 middle bits of the 48-bit counter. | RO | 0000h |

Note: The 48-bit receive error-free packets counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: Receive Error-Free Packets Counter Low Register, Receive Error-Free Packets Counter Mid Register, Receive Error-Free Packets Counter High Register. After reading the high register, the counter will be automatically cleared.

### 4.3.6 RECEIVE ERROR-FREE PACKETS COUNTER LOW REGISTER

Index: U4 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | RCVGPKT[15:0] <br> Counts the received error-free packets. <br> Contains the 16 low-order bits of the 48-bit counter. | RO | 0000h |

Note: The 48-bit receive error-free packets counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: Receive Error-Free Packets Counter Low Register, Receive Error-Free Packets Counter Mid Register, Receive Error-Free Packets Counter High Register. After reading the high register, the counter will be automatically cleared.

### 4.3.7 CRC ERROR COUNTER HIGH REGISTER

Index: U5 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| 15:0 | CRCERR[47:32] <br> Counts the CRC errors, which are generated by the CRC checker circuit. <br> Contains the 16 upper bits of the 48-bit counter. <br> Reading this register resets all bits in the CRC Error Counter. | RO/ <br> RC | 0000 h |

Note: The 48-bit CRC error counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: CRC Error Counter Mid Register, CRC Error Counter High Register. After reading the high register, the counter will be automatically cleared.

### 4.3.8 CRC ERROR COUNTER MID REGISTER

Index:
U6
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | CRCERR[31:16] <br> Counts the CRC errors, which are generated by the CRC checker circuit. <br> Contains the 16 middle bits of the 48-bit counter. | RO | 0000h |

Note: The 48-bit CRC error counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: CRC Error Counter Mid Register, CRC Error Counter High Register. After reading the high register, the counter will be automatically cleared.

### 4.3.9 CRC ERROR COUNTER LOW REGISTER

Index: U7 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | CRCERR[15:0] <br> Counts the CRC errors, which are generated by the CRC checker circuit. <br> Contains the 16 low-order bits of the 48-bit counter. | RO | 0000h |

Note: The 48-bit CRC error counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: CRC Error Counter Mid Register, CRC Error Counter High Register. After reading the high register, the counter will be automatically cleared.

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4.3.10 RECEIVE ERROR DURING DATA COUNTER REGISTER
Index:
U8
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | RXERIND_DATA[15:0] <br> Counts the assertions of RXER (going from low to high) when RXDV is high. | RO/ <br> $R C$ | 0000h |

### 4.3.11 RECEIVE ERROR DURING IDLE COUNTER REGISTER

Index: U9 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | RXERIND_IDLE[15:0] <br> Counts the assertions of RXER (going from low to high) when RXDV is low. | RO/ <br> RC | 0000h |

### 4.3.12 TRANSMIT PACKET COUNTER HIGH REGISTER

Index: U10 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | TXPKT[47:32] <br> Counts the number of transmitted packets. <br> Contains the 16 upper bits of the 48-bit counter. <br> Reading this register resets all bits in the Transmit Packet Counter. | RO/ <br> RC | 00000h |

Note: The 48-bit transmit packet counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: Transmit Packet Counter Low Register, Transmit Packet Counter Mid Register, Transmit Packet Counter High Register. After reading the high register, the counter will be automatically cleared.

### 4.3.13 TRANSMIT PACKET COUNTER MID REGISTER

Index:
Size:
16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | TXPKT[31:16] <br> Counts the number of transmitted packets. <br> Contains the 16 middle bits of the 48-bit counter. | RO | 0000h |

Note: $\quad$ The 48-bit transmit packet counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: Transmit Packet Counter Low Register, Transmit Packet Counter Mid Register, Transmit Packet Counter High Register. After reading the high register, the counter will be automatically cleared.

### 4.3.14 TRANSMIT PACKET COUNTER LOW REGISTER

Index: U12 Size: 16 bits

| Bits | Description | Type | Default |
| :---: | :--- | :---: | :---: |
| $15: 0$ | TXPKT[15:0] <br> Counts the number of transmitted packets. <br> Contains the 16 low-order bits of the 48-bit counter. | RO | 0000h |

Note: The 48-bit transmit packet counter is split across 3 registers. In order to read the counter correctly, the registers must be read in the following order: Transmit Packet Counter Low Register, Transmit Packet Counter Mid Register, Transmit Packet Counter High Register. After reading the high register, the counter will be automatically cleared.

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### 5.0 OPERATIONAL CHARACTERISTICS

### 5.1 Absolute Maximum Ratings*

Supply Voltage (VDDVARIO) (Note 5-1) ..... -0.5 V to +3.6 V
Analog Supply Voltage (VDD12A) (Note 5-1) 0.5 V to +1.5 V
Digital Core Supply Voltage (VDD12CORE) (Note 5-1) ..... -0.5 V to +1.5 V
Ethernet Magnetics Supply Voltage -0.5 V to +3.6 V
Positive voltage on signal pins, with respect to ground (Note 5-2) ..... $+6.0 \mathrm{~V}$
Negative voltage on signal pins, with respect to ground (Note 5-3) ..... -0.5V
Positive voltage on XI, with respect to ground ..... +3.6V
Positive voltage on XO , with respect to ground ..... $+2.5 \mathrm{~V}$
Ambient Operating Temperature in Still Air $\left(T_{A}\right)$ ..... Note 5-4
Junction to Ambient ( $\theta_{\mathrm{JA}}$ ) ..... $22.5^{\circ} \mathrm{C} / \mathrm{W}$
Junction to Top of Package ( $\Psi_{J T}$ ). ..... $.0 .1^{\circ} \mathrm{C} / \mathrm{W}$
Storage Temperature ..... $-55^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature Range. Refer to JEDEC Spec. J-STD-020Latch-up Performance per EIA/JESD 78.+/-150mA

Note 5-1 When powering this device from laboratory or system power supplies, it is important that the absolute maximum ratings not be exceeded or device failure can result. Some power supplies exhibit voltage spikes on their outputs when AC power is switched on or off. In addition, voltage transients on the AC power line may appear on the DC output. If this possibility exists, it is suggested that a clamp circuit be used.

Note 5-2 This rating does not apply to the following pins: XI, XO, ETHRBIAS.
Note 5-3 This rating does not apply to the following pins: ETHRBIAS.
Note 5-4 $\quad 0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ for commercial version, $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ for industrial version.
*Stresses exceeding those listed in this section could cause permanent damage to the device. This is a stress rating only. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Functional operation of the device at any condition exceeding those indicated in Section 5.2, "Operating Conditions**", Section 5.4, "DC Specifications", or any other applicable section of this specification is not implied. Note, device signals are NOT 5 volt tolerant unless specified otherwise.

### 5.2 Operating Conditions**

Supply Voltage (VDDVARIO) . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . +2.25 V to +3.6V
Supply Voltage (VDD12A) . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . +1.14V to +1.26V
Digital Core Supply Voltage (VDD12CORE) . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . +1.14 V to +1.26V
Ethernet Magnetics Supply Voltage . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . +2.25 V to +3.6V
Ambient Operating Temperature in Still Air ( $\mathrm{T}_{\mathrm{A}}$ ) . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . Note 5-4
**Proper operation of the device is guaranteed only within the ranges specified in this section. After the device has completed power-up, VDDVARIO and the magnetics power supply must maintain their voltage level within $+/-10 \%$. Varying the voltage greater than $+/-10 \%$ after the device has completed power-up can cause errors in device operation.

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### 5.3 Power Consumption

This section details the power consumption of the device as measured during various modes of operation. Power consumption values are provided for both the device-only, and for the device plus Ethernet components. Power dissipation is impacted by temperature, supply voltage, and external source/sink requirements. All measurements were taken at $+25^{\circ} \mathrm{C}$.

TABLE 5-1: POWER CONSUMPTION - 1000BASE-T LINKED

| Parameter | Typical | Unit |
| :--- | :---: | :---: |
| Supply Current (VDDVARIO) (@+3.3V) | 36 | mA |
| Supply Current (VDD12CORE, VDD12BIAS, VDD12PLL, VDD12A) (@+1.2V) | 454 | mA |
| External Magnetics Current (@ +3.3V) | 202 | mA |
| Total Power Dissipation (Device Only) (Note 5-5) | 665 | mW |
| Total Power Dissipation (Device and Ethernet components) (Note 5-5) | 1331 | mW |

Note 5-5 When the external magnetics operate at +2.5 V , current consumption remains the same. However, Ethernet component power consumption will be reduced accordingly.

## TABLE 5-2: POWER CONSUMPTION - 100BASE-TX LINKED

| Parameter | Typical | Unit |
| :--- | :---: | :---: |
| Supply Current (VDDVARIO) (@ +3.3V) | 24 | mA |
| Supply Current (VDD12CORE, VDD12BIAS, VDD12PLL, VDD12A) (@+1.2V) | 82 | mA |
| External Magnetics Current (@ +3.3V) | 63 | mA |
| Total Power Dissipation (Device Only) (Note 5-6) | 177 | mW |
| Total Power Dissipation (Device and Ethernet components) (Note 5-6) | 385 | mW |

Note 5-6 When the external magnetics operate at +2.5 V , current consumption remains the same. However, Ethernet component power consumption will be reduced accordingly.

## TABLE 5-3: POWER CONSUMPTION - 10BASE-T LINKED

| Parameter | Typical | Unit |
| :--- | :---: | :---: |
| Supply Current (VDDVARIO) (@ +3.3V) | 21 | mA |
| Supply Current (VDD12CORE, VDD12BIAS, VDD12PLL, VDD12A) (@+1.2V) | 32 | mA |
| External Magnetics Current (@ +3.3V) | 123 | mA |
| Total Power Dissipation (Device Only) (Note 5-7) | 107 | mW |
| Total Power Dissipation (Device and Ethernet components) (Note 5-7) | 513 | mW |

Note 5-7 When the external magnetics operate at +2.5 V , current consumption remains the same. However, Ethernet component power consumption will be reduced accordingly.

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## TABLE 5-4: POWER CONSUMPTION - ENERGY DETECT

| Parameter | Typical | Unit |
| :--- | :---: | :---: |
| Supply Current (VDDVARIO) (@ +3.3V) | 30 | mA |
| Supply Current (VDD12CORE, VDD12BIAS, VDD12PLL, VDD12A) (@ +1.2V) | 28 | mA |
| External Magnetics Current (@ +3.3V) | 21 | mA |
| Total Power Dissipation (Device Only) (Note 5-8) | 133 | mW |
| Total Power Dissipation (Device and Ethernet components) (Note 5-8) | 203 | mW |

Note 5-8 When the external magnetics operate at +2.5 V , current consumption remains the same. However, Ethernet component power consumption will be reduced accordingly.

## TABLE 5-5: POWER CONSUMPTION - HARDWARE POWER DOWN (PLL ENABLED)

| Parameter | Typical | Unit |
| :--- | :---: | :---: |
| Supply Current (VDDVARIO) (@ +3.3V) | 0.12 | mA |
| Supply Current (VDD12CORE, VDD12BIAS, VDD12PLL, VDD12A) (@+1.2V) | 17.29 | mA |
| External Magnetics Current (@ +3.3V) | 7.00 | mA |
| Total Power Dissipation (Device Only) (Note 5-9) | 21.16 | mW |
| Total Power Dissipation (Device and Ethernet components) (Note 5-9) | 44.26 | mW |

Note 5-9 When the external magnetics operate at +2.5 V , current consumption remains the same. However, Ethernet component power consumption will be reduced accordingly.

## TABLE 5-6: POWER CONSUMPTION - HARDWARE POWER DOWN (PLL DISABLED)

| Parameter | Typical | Unit |
| :--- | :---: | :---: |
| Supply Current (VDDVARIO) (@ +3.3V) | 0.12 | mA |
| Supply Current (VDD12CORE, VDD12BIAS, VDD12PLL, VDD12A) (@ +1.2V) | 4.39 | mA |
| External Magnetics Current (@ +3.3V) | 0.02 | mA |
| Total Power Dissipation (Device Only) (Note 5-10) | 5.68 | mW |
| Total Power Dissipation (Device and Ethernet components) (Note 5-10) | 5.73 | mW |

Note 5-10 When the external magnetics operate at +2.5 V , current consumption remains the same. However, Ethernet component power consumption will be reduced accordingly.

### 5.4 DC Specifications

## TABLE 5-7: I/O BUFFER CHARACTERISTICS

| Parameter | Symbol | Min | $\begin{aligned} & 2.5 \mathrm{~V} \\ & \text { TYP } \end{aligned}$ | $\begin{aligned} & 3.3 \mathrm{~V} \\ & \text { typ } \end{aligned}$ | Max | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VIS Type Input Buffer |  |  |  |  |  |  |  |
| Low Input Level | $\mathrm{V}_{\text {ILI }}$ | -0.3 |  |  |  | V |  |
| High Input Level | $\mathrm{V}_{\mathrm{IHI}}$ |  |  |  | 3.6 | V |  |
| Negative-Going Threshold | $V_{\text {ILT }}$ | 0.64 | 1.15 | 1.41 | 1.76 | V | Schmitt trigger |
| Positive-Going Threshold | $\mathrm{V}_{\text {IHT }}$ | 0.81 | 1.29 | 1.65 | 1.90 | V | Schmitt trigger |
| SchmittTrigger Hysteresis $\left(V_{I H T}-V_{I L T}\right)$ | $\mathrm{V}_{\mathrm{HYS}}$ | 102 | 136 | 138 | 288 | mV |  |
| Input Leakage <br> ( $\mathrm{V}_{\text {IN }}=$ VSS or VDDVARIO) | $\mathrm{IIH}^{\text {H }}$ | -10 |  |  | 10 | uA | Note 5-11 |
| Input Capacitance | $\mathrm{C}_{\text {IN }}$ |  |  |  | 3 | pF |  |
| VO6 Type Buffers |  |  |  |  |  |  |  |
| Low Output Level | $\mathrm{V}_{\mathrm{OL}}$ |  |  |  | 0.4 | V | $\mathrm{l}_{\mathrm{OL}}=6 \mathrm{~mA}$ |
| High Output Level | $\mathrm{V}_{\mathrm{OH}}$ | VDDVARIO <br> - 0.4 |  |  |  | V | $\mathrm{l}_{\mathrm{OH}}=-6 \mathrm{~mA}$ |
| VO8 Type Buffers |  |  |  |  |  |  |  |
| Low Output Level | $\mathrm{V}_{\mathrm{OL}}$ |  |  |  | 0.4 | V | $\mathrm{l}_{\mathrm{OL}}=8 \mathrm{~mA}$ |
| High Output Level | $\mathrm{V}_{\mathrm{OH}}$ | VDDVARIO <br> - 0.4 |  |  |  | V | $\mathrm{l}_{\mathrm{OH}}=-8 \mathrm{~mA}$ |
| ICLK Type Buffer (XI Input) |  |  |  |  |  |  | Note 5-12 |
| Low Input Level | $\mathrm{V}_{\text {ILI }}$ | -0.3 |  |  | 0.5 | V |  |
| High Input Level | $\mathrm{V}_{\mathrm{IHI}}$ | 0.9 |  |  | 3.6 | V |  |

Note 5-11 This specification applies to all inputs and tri-stated bi-directional pins. Internal pull-down and pull-up resistors add +/-50uA per-pin (typical).
Note 5-12 XI can optionally be driven from a 25 MHz single-ended clock oscillator.

TABLE 5-8: 1000BASE-T TRANSCEIVER CHARACTERISTICS

| Parameter | Symbol | Min | Typ | Max | Units | Notes |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Peak Differential Output Voltage | $\mathrm{V}_{\mathrm{OP}}$ | 670 |  | 820 | mV | Note 5-13 |
| Signal Amplitude Symmetry | $\mathrm{V}_{\mathrm{SS}}$ |  |  | 1 | $\%$ | Note $5-13$ |
| Signal Scaling | $\mathrm{V}_{\mathrm{SC}}$ |  |  | 2 | $\%$ | Note $5-14$ |
| Output Droop | $\mathrm{V}_{\mathrm{OD}}$ | 73.1 |  |  | $\%$ | Note $5-13$ |
| Transmission Distortion |  |  |  | 10 | mV | Note $5-15$ |

Note 5-13 IEEE 802.ab Test Mode 1

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Note 5-14 From $1 / 2$ of average $\mathrm{V}_{\mathrm{OP}}$, Test Mode 1
Note 5-15 IEEE 802.ab distortion processing

TABLE 5-9: 100BASE-TX TRANSCEIVER CHARACTERISTICS

| Parameter | Symbol | Min | Typ | max | units | notes |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Peak Differential Output Voltage High | $\mathrm{V}_{\mathrm{PPH}}$ | 950 | - | 1050 | mVpk | Note 5-16 |
| Peak Differential Output Voltage Low | $\mathrm{V}_{\mathrm{PPL}}$ | -950 | - | -1050 | mVpk | Note 5-16 |
| Signal Amplitude Symmetry | $\mathrm{V}_{\mathrm{SS}}$ | 98 | - | 102 | $\%$ | Note 5-16 |
| Signal Rise and Fall Time | $\mathrm{T}_{\mathrm{RF}}$ | 3.0 | - | 5.0 | nS | Note 5-16 |
| Rise and Fall Symmetry | $\mathrm{T}_{\mathrm{RFS}}$ | - | - | 0.5 | nS | Note 5-16 |
| Duty Cycle Distortion | $\mathrm{D}_{\mathrm{CD}}$ | 35 | 50 | 65 | $\%$ | Note 5-17 |
| Overshoot and Undershoot | $\mathrm{V}_{\mathrm{OS}}$ | - | - | 5 | $\%$ |  |
| Jitter |  |  |  | 1.4 | nS | Note 5-18 |

Note 5-16 Measured at line side of transformer, line replaced by $100 \Omega(+/-1 \%)$ resistor.
Note 5-17 Offset from 16 nS pulse width at $50 \%$ of pulse peak.
Note 5-18 Measured differentially.

TABLE 5-10: 10BASE-T TRANSCEIVER CHARACTERISTICS

| Parameter | Symbol | Min | Typ | Max | Units | Notes |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Transmitter Peak Differential Output Voltage | $\mathrm{V}_{\text {OUT }}$ | 2.2 | 2.5 | 2.8 | V | Note 5-19 |
| Receiver Differential Squelch Threshold | $\mathrm{V}_{\mathrm{DS}}$ | 300 | 420 | 585 | mV |  |

Note 5-19 Min/max voltages guaranteed as measured with $100 \Omega$ resistive load.
5.5 AC Specifications

This section details the various AC timing specifications of the device.

Note: $\quad$ The GMII/MII timing adheres to the IEEE 802.3 specification. Refer to the IEEE 802.3 specification for additional GMII/MII timing information.

Note: The Ethernet TX/RX pin timing adheres to the IEEE 802.3 specification. Refer to the IEEE 802.3 specification for detailed Ethernet timing information.
5.5.1 EQUIVALENT TEST LOAD

Output timing specifications assume a 25 pF equivalent test load, unless otherwise noted, as illustrated in Figure 5-1.

FIGURE 5-1:
OUTPUT EQUIVALENT TEST LOAD

OUTPUT



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### 5.5.2 POWER SEQUENCE TIMING

Power supplies must adhere to the following rules:

- All power supplies of the same voltage must be powered up/down together.
- There is no power-up sequencing requirement, however all power supplies must reach operational levels within the time periods specified in Table 5-11.
- There is no power-down sequencing or timing requirement, however the device must not be powered for an extended period of time without all supplies at operational levels.
- Following initial power-on, or if a power supply brownout occurs (for example, one or more supplies drops below operational limits), a power-on reset must be executed once all power supplies reach operational levels. Refer to Section 5.5.3, "Power-On Hardware Reset Timing," on page 67 for power-on reset requirements.
- Do not drive input signals without power supplied to the device.

Note: Violation of these specifications may damage the device.

## FIGURE 5-2: POWER SEQUENCE TIMING

All 3.3V Power Supply Pins

All 2.5V Power Supply Pins

All 1.2V Power Supply Pins


TABLE 5-11: POWER SEQUENCE TIMING VALUES

| Symbol | Description | Min | Typ | Max | Units |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {pon }}$ | Power supply turn on time | 0 |  | 25 | mS |

Note 1: The VDDVARIO power supply can be run at 2.5 V or 3.3 V .
2: The magnetics power supply can be run at 2.5 V or 3.3 V .

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### 5.5.3 POWER-ON HARDWARE RESET TIMING

Figure 5-3 illustrates the nRESET, configuration strap/pin, and CONFIG[3:0] timing requirements in relation to poweron. A hardware reset (nRESET assertion) is required following power-up. For proper operation, nRESET must be asserted for no less than $t_{\text {rstia }}$. The nRESET pin can be asserted at any time, but must not be deasserted before $t_{\text {purstd }}$ after all external power supplies have reached operational levels. In order for valid configuration strap values to be read at power-up, the $t_{\text {css }}$ and $t_{\text {csh }}$ timing constraints must be followed. In order for CONFIG[3:0] values to be read at powerup, the $t_{c s}$ and $t_{c h}$ timing constraints must be followed. Refer to Section 3.6.1, "Hardware Reset (nRESET)," on page 21 for additional information.

FIGURE 5-3: HARDWARE RESET TIMING


TABLE 5-12: HARDWARE RESET TIMING VALUES

| Symbol | Description | Min | Typ | Max | Units |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {purstd }}$ | External power supplies at operational level to nRESET <br> deassertion | 25 |  |  | mS |
| $\mathrm{t}_{\text {purstv }}$ | External power supplies at operational level to nRESET <br> valid | 0 |  |  | nS |
| $\mathrm{t}_{\text {rstid }}$ | nRESET input deassertion time | 10 |  |  | $\mu \mathrm{~S}$ |
| $\mathrm{t}_{\text {rstia }}$ | nRESET input assertion time | 100 |  |  | $\mu \mathrm{~S}$ |
| $\mathrm{t}_{\text {css }}$ | Configuration strap pins setup to nRESET deassertion | 200 |  |  | nS |
| $\mathrm{t}_{\text {csh }}$ | Configuration strap pins hold after nRESET deassertion | 10 |  | nS |  |
| $\mathrm{t}_{\text {otaa }}$ | Output tri-state after nRESET assertion |  |  | 50 | nS |
| $\mathrm{t}_{\text {odad }}$ | Output drive after deassertion | 40 |  | 800 | nS |
| $\mathrm{t}_{\text {cs }}$ | CONFIG[3:0] setup to nRESET deassertion | 0 |  | nS |  |
| $\mathrm{t}_{\mathrm{ch}}$ | CONFIG[3:0] hold after nRESET deassertion | 1 |  | uS |  |

Note 1: Device configuration straps are latched as a result of nRESET assertion. Refer to Section 3.8.1.1, "Configuration Straps," on page 23 details. Configuration straps must only be pulled high or low and must not be driven as inputs.
2: nRESET deassertion must be monotonic.

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### 5.5.4 RESET TIMING

Figure 5-4 illustrates the nRESET pin timing requirements. For proper operation, nRESET must be asserted for no less than $t_{\text {rstia. }}$. In order for valid configuration strap values to be read upon a nRESET assertion, the $t_{\text {css }}$ and $t_{\text {csh }}$ timing constraints must be followed. In order for CONFIG[3:0] values to be read at power-up, the $\mathrm{t}_{\mathrm{cs}}$ and $\mathrm{t}_{\mathrm{ch}}$ timing constraints must be followed. Refer to Section 3.6.1, "Hardware Reset (nRESET)," on page 21 for additional information.

Note: $\quad$ A hardware reset (nRESET assertion) is required following power-up. Refer to Section 5.5.3, "Power-On Hardware Reset Timing," on page 67 for additional information.

FIGURE 5-4: RESET TIMING


TABLE 5-13: RESET TIMING VALUES

| Symbol | Description | Min | Typ | Max | Units |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\text {rstia }}$ | nRESET input assertion time | 1 |  |  | $\mu \mathrm{~S}$ |
| $\mathrm{t}_{\mathrm{css}}$ | Configuration strap pins setup to nRESET deassertion | 200 |  |  | nS |
| $\mathrm{t}_{\mathrm{csh}}$ | Configuration strap pins hold after nRESET deassertion | 10 |  |  | nS |
| $\mathrm{t}_{\text {otaa }}$ | Output tri-state after nRESET assertion |  |  | 50 | nS |
| $\mathrm{t}_{\text {odad }}$ | Output drive after deassertion | 40 |  | 800 | nS |
| $\mathrm{t}_{\mathrm{cs}}$ | CONFIG[3:0] setup to nRESET deassertion | 0 |  |  | nS |
| $\mathrm{t}_{\mathrm{ch}}$ | CONFIG[3:0] hold after nRESET deassertion | 1 |  |  | uS |

Note: Device configuration straps are latched as a result of nRESET assertion. Refer to Section 3.8.1.1, "Configuration Straps," on page 23 details. Configuration straps must only be pulled high or low and must not be driven as inputs.

### 5.5.5 GMII TIMING (1000BASE-T)

This section specifies the GMII interface transmit and receive timing. Please refer to Section 3.3, "GMII Interface," on page 18 for additional details.

Note: All GMII timing specifications assume a point-to-point test circuit as defined in Section 35.4.2.2 of the IEEE 802.3-2005 specification.

FIGURE 5-5: GMII TRANSMIT TIMING


TABLE 5-14: GMII TRANSMIT TIMING VALUES

| Symbol | Description | Min | Max | Units | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{gtxclk}}$ | GTXCLK Frequency | $125-$ <br> 100 ppm | $125+$ <br> 100 ppm | MHz |  |
| $\mathrm{t}_{\mathrm{clkp}}$ | GTXCLK period | 7.5 | 8.5 | ns | Note $5-1$ |
| $\mathrm{t}_{\mathrm{clkh}}$ | GTXCLK high time | 2.5 |  | ns |  |
| $\mathrm{t}_{\mathrm{clkl}}$ | GTXCLK low time | 2.5 |  | ns |  |
| $\mathrm{t}_{\text {val }}$ | TXD[7:0], TXEN, TXER setup time to rising <br> edge of GTXCLK | 2.0 |  | ns |  |
| $\mathrm{t}_{\text {hold }}$ | TXD[7:0], TXEN, TXER hold time after rising <br> edge of GTXCLK | 0.0 |  | ns |  |
| $\mathrm{t}_{\mathrm{r}}$ | GTXCLK rise time |  | 1 | ns | Note $5-2$ |
| $\mathrm{t}_{\mathrm{f}}$ | GTXCLK fall time | 1 | ns | Note $5-2$ |  |

Note 5-1 $\quad \mathrm{Min} / \mathrm{max}$ limits are non-sustainable long term.
Note 5-2 $\quad t_{r}$ and $t_{f}$ are measured from $\mathrm{V}_{\mathrm{IL} \_A C(\operatorname{Max})}=0.7 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{IH}_{-} A C(\operatorname{Min})}=1.9 \mathrm{~V}$.

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FIGURE 5-6:
GMII RECEIVE TIMING


TABLE 5-15: GMII RECEIVE TIMING VALUES

| Symbol | Description | Min | Max | Units | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{clkp}}$ | RXCLK period | 7.5 |  | ns |  |
| $\mathrm{t}_{\mathrm{clkh}}$ | RXCLK high time | 2.5 |  | ns |  |
| $\mathrm{t}_{\mathrm{clkl}}$ | RXCLK low time | 2.5 |  | ns |  |
| $\mathrm{t}_{\mathrm{val}}$ | RXD[7:0], RXDV, RXER output valid from rising <br> edge of RXCLK |  | 5.0 | ns |  |
| $\mathrm{t}_{\text {hold }}$ | RXD[7:0], RXDV, RXER output hold from rising <br> edge of RXCLK | 0.5 |  | ns |  |
| $\mathrm{t}_{\mathrm{r}}$ | RXCLK rise time |  | 1 | ns | Note 5-1 |
| $\mathrm{t}_{\mathrm{f}}$ | RXCLK fall time |  | 1 | ns | Note 5-1 |

Note 5-1 $\quad t_{r}$ and $t_{f}$ are measured from $\mathrm{V}_{\mathrm{IL} \_A C(\operatorname{Max})}=0.7 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{IH}_{-} A C(\operatorname{Min})}=1.9 \mathrm{~V}$.

Note 5-2

### 5.5.6 MII TIMING (100BASE-TX, 10BASE-T)

This section specifies the MII interface transmit and receive timing. Please refer to Section 3.3, "GMII Interface," on page 18 for additional details.

FIGURE 5-7:
MII RECEIVE TIMING


TABLE 5-16: MII RECEIVE TIMING VALUES

| Symbol | Description | Min | Max | Units | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{clkp}}$ | RXCLK period | Note 5-3 |  | ns |  |
| $\mathrm{t}_{\mathrm{clkh}}$ | RXCLK high time | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.4$ | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.6$ | ns |  |
| $\mathrm{t}_{\mathrm{clkl}}$ | RXCLK low time | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.4$ | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.6$ | ns |  |
| $\mathrm{t}_{\mathrm{val}}$ | RXD[3:0], RXDV output valid from rising edge of <br> RXCLK |  | 28.0 | ns | Note $5-4$ |
| $\mathrm{t}_{\text {hold }}$ | RXD[3:0], RXDV output hold from rising edge of <br> RXCLK | 10.0 | ns | Note $5-4$ |  |

Note 5-3 40ns for 100BASE-TX operation, 400ns for 10BASE-T operation.
Note 5-4 Timing was designed for system load between 10 pf and 25 pf.

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FIGURE 5-8:
MII TRANSMIT TIMING


TABLE 5-17: MII TRANSMIT TIMING VALUES

| Symbol | Description | Min | Max | Units | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{clkp}}$ | TXCLK period | Note 5-5 |  | ns |  |
| $\mathrm{t}_{\mathrm{clkh}}$ | TXCLK high time | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.4$ | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.6$ | ns |  |
| $\mathrm{t}_{\mathrm{clkl}}$ | TXCLK low time | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.4$ | $\mathrm{t}_{\mathrm{clkp}}{ }^{*} 0.6$ | ns |  |
| $\mathrm{t}_{\mathrm{su}}$ | TXD[3:0], TXEN setup time to rising edge of <br> TXCLK | 12.0 |  | ns | Note 5-6 |
| $\mathrm{t}_{\mathrm{hold}}$ | TXD[3:0], TXEN hold time after rising edge of <br> TXCLK | 0 | ns | Note 5-6 |  |

Note 5-5 40ns for 100BASE-TX operation, 400ns for 10BASE-T operation.
Note 5-6 Timing was designed for system load between 10 pf and 25 pf.

### 5.5.7 SMI TIMING

This section specifies the SMI timing of the device. Please refer to Section 3.4, "Serial Management Interface (SMI)," on page 19 for additional details.

FIGURE 5-9: SMI TIMING


TABLE 5-18: $\quad$ SMI TIMING VALUES

| Symbol | Description | Min | Max | Units | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{clkp}}$ | MDC period | 400 |  | ns |  |
| $\mathrm{t}_{\mathrm{clkh}}$ | MDC high time | $160(80 \%)$ | $160(80 \%)$ | ns |  |
| $\mathrm{t}_{\mathrm{clkl}}$ | MDC low time | 300 | ns |  |  |
| $\mathrm{t}_{\mathrm{val}}$ | MDIO (read from PHY) output valid from rising <br> edge of MDC |  | ns |  |  |
| $\mathrm{t}_{\text {ohold }}$ | MDIO (read from PHY) output hold from rising <br> edge of MDC | 0 | ns |  |  |
| $\mathrm{t}_{\text {su }}$ | MDIO (write to PHY) setup time to rising edge <br> of MDC | 10 |  | ns |  |
| $\mathrm{t}_{\text {ihold }}$ | MDIO (write to PHY) input hold time after rising <br> edge of MDC | 10 |  |  |  |

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### 5.5.8 JTAG TIMING

This section specifies the JTAG timing of the device. Please refer to Section 3.9.9, "IEEE 1149.1 (JTAG) Boundary Scan," on page 30 for additional details.

FIGURE 5-10: JTAG TIMING


TABLE 5-19: JTAG TIMING VALUES

| Symbol | Description | Min | Max | Units | Notes |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{tckp}}$ | TCK clock period | 66.67 |  | ns |  |
| $\mathrm{t}_{\text {tckhl }}$ | TCK clock high/low time | $\mathrm{t}_{\text {tckp }}{ }^{*} 0.4$ | $\mathrm{t}_{\text {tckp }}{ }^{*} 0.6$ | ns |  |
| $\mathrm{t}_{\mathrm{su}}$ | TDI, TMS setup to TCK rising edge | 10 |  | ns |  |
| $\mathrm{t}_{\mathrm{h}}$ | TDI, TMS hold from TCK rising edge | 10 |  | ns |  |
| $\mathrm{t}_{\text {dov }}$ | TDO output valid from TCK falling edge |  | 16 | ns |  |
| $\mathrm{t}_{\text {dohinvld }}$ | TDO output invalid from TCK falling edge | 0 |  | ns |  |

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### 5.6 Clock Circuit

The device can accept either a 25 MHz crystal (preferred) or a 25 MHz single-ended clock oscillator ( $+/-50 \mathrm{ppm}$ ) input. If the single-ended clock oscillator method is implemented, XO should be left unconnected and XI should be driven with a nominal OV-VDDVARIO clock signal. The input clock duty cycle is $40 \%$ minimum, $50 \%$ typical and $60 \%$ maximum.
It is recommended that a crystal utilizing matching parallel load capacitors be used for the crystal input/output signals (XI/XO). See Table 5-20 for the recommended crystal specifications.

## TABLE 5-20: CRYSTAL SPECIFICATIONS

| Parameter | Symbol | Min | Nom | Max | Units | Notes |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Crystal Cut | AT, typ |  |  |  |  |  |
| Crystal Oscillation Mode | Fundamental Mode |  |  |  |  |  |
| Crystal Calibration Mode | Parallel Resonant Mode |  |  |  |  |  |
| Frequency | $F_{\text {fund }}$ | - | 25.000 | - | MHz |  |
| Frequency Tolerance @ $25^{\circ} \mathrm{C}$ | $F_{\text {tol }}$ | - | - | +/-50 | PPM | Note 5-7 |
| Frequency Stability Over Temp | $\mathrm{F}_{\text {temp }}$ | - | - | +/-50 | PPM | Note 5-7 |
| Frequency Deviation Over Time | $\mathrm{F}_{\text {age }}$ | - | +/-3 to 5 | - | PPM | Note 5-8 |
| Total Allowable PPM Budget |  | - | - | +/-50 | PPM | Note 5-9 |
| Shunt Capacitance | $\mathrm{C}_{0}$ | - |  | 7 | pF |  |
| Load Capacitance | $\mathrm{C}_{\mathrm{L}}$ | - |  | 18 | pF |  |
| Drive Level | $\mathrm{P}_{\mathrm{W}}$ | 300 | - | - | uW |  |
| Equivalent Series Resistance | $\mathrm{R}_{1}$ | - | - | 50 | Ohm |  |
| Operating Temperature Range |  | Note 5-10 | - | Note 5-11 | ${ }^{\circ} \mathrm{C}$ |  |
| XI Pin Capacitance |  | - | 3 typ | - | pF | Note 5-12 |
| XO Pin Capacitance |  | - | 3 typ | - | pF | Note 5-12 |

Note 5-7 The maximum allowable values for Frequency Tolerance and Frequency Stability are application dependent. Since any particular application must meet the IEEE +/-50 PPM Total PPM Budget, the combination of these two values must be approximately $+/-45$ PPM (allowing for aging).
Note 5-8 Frequency Deviation Over Time is also referred to as Aging.
Note 5-9 The total deviation for the Transmitter Clock Frequency is specified by IEEE 802.3 u as +/- 50 PPM.
Note 5-10 $\quad 0^{\circ} \mathrm{C}$ for commercial version, $-40^{\circ} \mathrm{C}$ for industrial version.
Note 5-11 $+70^{\circ} \mathrm{C}$ for commercial version, $+85^{\circ} \mathrm{C}$ for industrial version.
Note 5-12 This number includes the pad, the bond wire and the lead frame. PCB capacitance is not included in this value. The XO/XI pin and PCB capacitance values are required to accurately calculate the value of the two external load capacitors. These two external load capacitors determine the accuracy of the 25.000 MHz frequency.

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### 6.0 PACKAGE OUTLINE

### 6.1 72-QFN Package

Note: For the most current package drawings, see the Microchip Packaging Specification at: http://www.microchip.com/packaging.

FIGURE 6-1:
72-QFN PACKAGE


TABLE 6-1: 72-QFN DIMENSIONS

|  | Min | Nominal | Max | Remarks |
| :---: | :---: | :---: | :---: | :---: |
| A | 0.80 | 0.85 | 1.00 | Overall Package Height |
| A1 | 0.00 | 0.02 | 0.05 | Standoff |
| A2 | - | 0.65 | 0.80 | Mold Cap Thickness |
| D/E | 9.90 | 10.00 | 10.10 | X/Y Body Size |
| D1/E1 | 9.65 | 9.75 | 9.85 | X/Y Mold Cap Size |
| D2/E2 | 5.90 | 6.00 | 6.10 | X/Y Exposed Pad Size |
| L | 0.30 | 0.40 | 0.50 | Terminal Length |
| b | 0.18 | 0.25 | 0.30 | Terminal Width |
| K | 1.50 | - | - | Center Pad to Pin Clearance |
| e | 0.50 BSC |  |  | Terminal Pitch |

Note 1: All dimensions are in millimeters unless otherwise noted.
2: Dimension "b" applies to plated terminals and is measured between 0.15 and 0.30 mm from the terminal tip.
3: The pin 1 identifier may vary, but is always located within the zone indicated.

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FIGURE 6-2: 72-QFN RECOMMENDED PCB LAND PATTERN


## APPENDIX A: DATA SHEET REVISION HISTORY

## TABLE A-1: REVISION HISTORY



## TABLE A-1: REVISION HISTORY (CONTINUED)

| Revision Level \& Date | Section/Figure/Entry | Correction |
| :---: | :---: | :---: |
|  | Table 3-9 | Corrected typos in mode definitions of rows 3 and 4. |
|  | Table 3-10 | Corrected typos in row 5 (00100) mode definitions, Reg 0, and Reg 4 columns. |
|  | Section 5.6 | Updated second sentence to "If the single-ended clock oscillator method is implemented, XO should be left unconnected and XI should be driven with a nominal OV-VDDVARIO clock signal." |
| DS00001870A (12-09-14) | Section 3.6.1, "Hardware Reset (nRESET)," on page 21 | Updated section with additional details on powerup sequencing requirements. |
|  | Section 5.5.3, "Power-On Hardware Reset Timing," on page 67 | Updated section with additional details on powerup sequencing requirements. Section title and included figures/tables updated. |
| Rev. 1.1 (06-03-13) | Section 4.2.4, "PHY Identifier 2 Register," on page 39 | Corrected bits 9:4 default values from 0Ch to 0Eh |
|  | Table 3-11, "LED Mode 1 Operation - <br> LED_MODE[1:0]=01b," on page 27 and Table 3-12, "LED Mode 2 Operation LED_MODE[1:0]=10b," on page 27 | Updated LED function definitions |
|  | Table 2-3, "LED \& Configuration Pins," on page 8, Table 2-6, "Miscellaneous Pins," on page 10, Table 3-5, "Configuration Straps," on page 23, Section 5.6, "Clock Circuit," on page 75 | Updated XI and REFCLK_SEL definitions/info for clarity. |
|  | FIGURE 2-1: 72-QFN Pin Assignments (TOP VIEW) on page 5 , Table 2-1, "GMI Interface Pins," on page 6. Table 2-6, "Miscellaneous Pins," on page 10, Table 28, "72-QFN Pin <br> Assignments," on page 11 | Updated pin 25 definition to no connect (NC). |

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## TABLE A-1: REVISION HISTORY (CONTINUED)

| Revision Level \& Date | Section/Figure/Entry | Correction |
| :--- | :--- | :--- |
|  | Section 3.7.1, "General <br> Power-Down," on page 22, <br> Section 3.7.2, "Energy <br> Detect Power-Down," on <br> page 22, Table 3-5, <br> "Configuration Straps," on <br> page 23, Section 3.8.1.2.2, <br> "Configuring the Mode of <br> Operation (CONFIG[3:2])," <br> on page 24, Table 3-9, <br> "Configuring the Mode of <br> Operation," on page 25, <br> FIGURE 3-4: Simplified <br> Application Diagram on <br> page 33, Section 4.2.16, <br> "Extended Mode Control/ <br> Status Register," on page 48 | Removed references to MACCLK and updated <br> definition of bit 3 of the Extended Mode Control/ <br> Status Register. |
| Rev. 1.0 (08-02-12) | Section 5.3, "Power <br> Consumption," on page 61 | Updated power numbers with latest lab <br> measurements. |
|  | All | Added commercial version. |
| Rev. 1.0 (06-29-11) | All | Table 5-15, "GMII Receive <br> Timing Values," on page 70 |
| Removed t val min value and added t ${ }^{\text {(0al }}$ max value <br> of 5.Ons. |  |  |
|  | Initial release |  |

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BCM56960B1KFSBG EZX557AT2 S LKVX BCM56842A1KFTBG BCM56450B1KFSBG EZX557AT S LKW4 RTL8153-VC-CG
CH395L BCM56864A1IFSBG WGI219LM SLKJ2 KSZ8462FHLI KSZ8841-16MVLI KSZ9897STXC KSZ8842-16MVLI KSZ8893MQL
VSC8244XHG ADIN2111BCPZ FIDO2100BGA128IR0 FIDO5210BBCZ FIDO5200CBCZ ADIN1110BCPZ ADIN1110CCPZ
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[^0]:    Note 3-6 Switching at $50 \%$ duty cycle creates a third color [for example, If Color1 is orange and Color2 is green, Color 3 will be yellow (Orange + Green)]
    Note 3-7 Activity will cause duplex LED color to blink.

[^1]:    Microchip received ISO/TS-16949:2009 certification for its worldwide headquarters, design and wafer fabrication facilities in Chandler and Tempe, Arizona; Gresham, Oregon and design centers in California and India. The Company's quality system processes and procedures are for its $P I C^{\circledR}$ MCUs and dsPIC® DSCs, KEELOQ ${ }^{\circledR}$ code hopping devices, Serial EEPROMs, microperipherals, nonvolatile memory and analog products. In addition, Microchip's quality system for the design and manufacture of development systems is ISO 9001:2000 certified.

