

# MCP6281/1R/2/3/4/5

## 450 μA, 5 MHz Rail-to-Rail Op Amp

#### **Features**

- · Gain Bandwidth Product: 5 MHz (typical)
- Supply Current: I<sub>O</sub> = 450 μA (typical)
- Supply Voltage: 2.2V to 6.0V
- · Rail-to-Rail Input/Output
- Extended Temperature Range: -40°C to +125°C
- · Available in Single, Dual, and Quad Packages
- Single with CS (MCP6283)
- Dual with CS (MCP6285)

## **Applications**

- Automotive
- · Portable Equipment
- · Photodiode Amplifier
- · Analog Filters
- · Notebooks and PDAs
- · Battery-Powered Systems

### **Design Aids**

- SPICE Macro Models
- FilterLab<sup>®</sup> Software
- Mindi™ Circuit Designer & Simulator
- · MAPS (Microchip Advanced Part Selector)
- · Analog Demonstration and Evaluation Boards
- · Application Notes

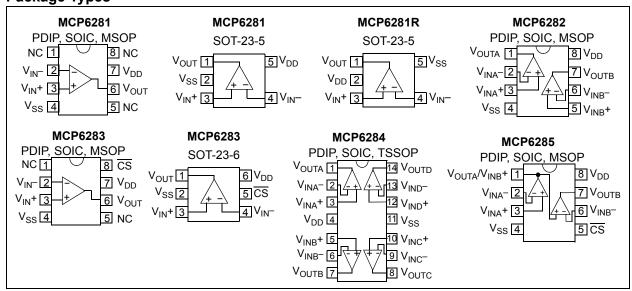
### **Description**

The Microchip Technology Inc. MCP6281/1R/2/3/4/5 family of operational amplifiers (op amps) provide wide bandwidth for the current. This family has a 5 MHz Gain Bandwidth Product (GBWP) and a 65° phase margin. This family also operates from a single supply voltage as low as 2.2V, while drawing 450  $\mu\text{A}$  (typical) quiescent current. Additionally, the MCP6281/1R/2/3/4/5 supports rail-to-rail input and output swing, with a common mode input voltage range of  $V_{DD}$  + 300 mV to  $V_{SS}$  – 300 mV. This family of operational amplifiers is designed with Microchip's advanced CMOS process.

The MCP6285 has a Chip Select  $(\overline{CS})$  input for dual op amps in an 8-pin package. This device is manufactured by cascading the two op amps (the output of op amp A connected to the non-inverting input of op amp B). The  $\overline{CS}$  input puts the device in Low-power mode.

The MCP6281/1R/2/3/4/5 family operates over the Extended Temperature Range of -40°C to +125°C. It also has a power supply range of 2.2V to 6.0V.

## Package Types



## 1.0 ELECTRICAL CHARACTERISTICS

### **Absolute Maximum Ratings †**

V <sub>DD</sub> – V <sub>SS</sub>	7.0V
Current at Input Pins	±2 mA
Analog Inputs ( $V_{IN}$ +, $V_{IN}$ -) †† $V_{SS}$ -	1.0V to V <sub>DD</sub> + 1.0V
All Other Inputs and Outputs V <sub>SS</sub> -	0.3V to V <sub>DD</sub> + 0.3V
Difference Input Voltage	V <sub>DD</sub> – V <sub>SS</sub>
Output Short Circuit Current	Continuous
Current at Output and Supply Pins	±30 mA
Storage Temperature	65°C to +150°C
Maximum Junction Temperature (T <sub>J</sub> )	+150°C
ESD Protection On All Pins (HBM; MM)	≥ 4 kV; 400V

† Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

†† See Section 4.1.2 "Input Voltage and Current Limits".

#### DC ELECTRICAL SPECIFICATIONS

Parameters	Sym	Min	Тур	Max	Units	Conditions
Input Offset						
Input Offset Voltage	Vos	-3.0	_	+3.0	mV	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Input Offset Voltage (Extended Temperature)	V <sub>OS</sub>	-5.0	_	+5.0	mV	T <sub>A</sub> = -40°C to +125°C, V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Input Offset Temperature Drift	$\Delta V_{OS}/\Delta T_{A}$	_	±1.7	_	μV/°C	T <sub>A</sub> = -40°C to +125°C, V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Power Supply Rejection Ratio	PSRR	70	90	_	dB	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Input Bias, Input Offset Current and I	npedance					
Input Bias Current	I <sub>B</sub>	_	±1.0	_	pА	Note 2
At Temperature	I <sub>B</sub>	_	50	200	pА	T <sub>A</sub> = +85°C (Note 2)
At Temperature	Ι <sub>Β</sub>	_	2	5	nA	T <sub>A</sub> = +125°C (Note 2)
Input Offset Current	los	_	±1.0	_	pА	Note 3
Common Mode Input Impedance	Z <sub>CM</sub>	_	10 <sup>13</sup>   6	_	$\Omega    pF$	Note 3
Differential Input Impedance	Z <sub>DIFF</sub>	_	10 <sup>13</sup>   3	_	$\Omega    pF$	Note 3
Common Mode (Note 4)						
Common Mode Input Range	$V_{CMR}$	$V_{SS} - 0.3$	_	V <sub>DD</sub> + 0.3	V	
Common Mode Rejection Ratio	CMRR	70	85	_	dB	$V_{CM} = -0.3V$ to 2.5V, $V_{DD} = 5V$
Common Mode Rejection Ratio	CMRR	65	80	_	dB	$V_{CM} = -0.3V$ to 5.3V, $V_{DD} = 5V$
Open-Loop Gain						
DC Open-Loop Gain (Large Signal)	A <sub>OL</sub>	90	110		dB	$V_{OUT} = 0.2V \text{ to } V_{DD} - 0.2V,$ $V_{CM} = V_{SS} \text{ (Note 1)}$
Output						
Maximum Output Voltage Swing	V <sub>OL</sub> , V <sub>OH</sub>	V <sub>SS</sub> + 15		V <sub>DD</sub> – 15	mV	0.5V input overdrive
Output Short Circuit Current	I <sub>SC</sub>		±25		mA	
Power Supply						
Supply Voltage	$V_{DD}$	2.2	_	6.0	V	(Note 5)
Quiescent Current per Amplifier	IQ	300	450	570	μA	I <sub>O</sub> = 0

- Note 1: The MCP6285's  $V_{CM}$  for op amp B (pins  $V_{OUTA}/V_{INB}$ + and  $V_{INB}$ -) is  $V_{SS}$  + 100 mV.
  - 2: The current at the MCP6285's V<sub>INB</sub>- pin is specified by I<sub>B</sub> only.
  - 3: This specification does not apply to the MCP6285's  $V_{OUTA}/V_{INB}$ + pin.
  - 4: The MCP6285's V<sub>INB</sub>- pin (op amp B) has a common mode range (V<sub>CMR</sub>) of V<sub>SS</sub> + 100 mV to V<sub>DD</sub> 100 mV. The MCP6285's V<sub>OUTA</sub>/V<sub>INB</sub>+ pin (op amp B) has a voltage range specified by V<sub>OH</sub> and V<sub>OL</sub>.
  - 5: All parts with date codes November 2007 and later have been screened to ensure operation at V<sub>DD</sub> = 6.0V. However, the other minimum and maximum specifications are measured at 2.4V and/or 5.5V.

#### **AC ELECTRICAL SPECIFICATIONS**

**Electrical Characteristics:** Unless otherwise indicated,  $T_A = \pm 25^{\circ}C$ ,  $V_{DD} = \pm 2.2V$  to  $\pm 5.5V$ ,  $V_{SS} = GND$ ,  $V_{OUT} \approx V_{DD}/2$ ,  $V_{CM} = V_{DD}/2$ ,  $V_L =$ 

Parameters	Sym	Min	Тур	Max	Units	Conditions
AC Response						
Gain Bandwidth Product	GBWP	_	5.0	_	MHz	
Phase Margin at Unity-Gain	PM	_	65	_	0	G = +1 V/V
Slew Rate	SR	_	2.5	_	V/µs	
Noise						
Input Noise Voltage	E <sub>ni</sub>	_	5.2	_	$\mu V_{P-P}$	f = 0.1 Hz to 10 Hz
Input Noise Voltage Density	e <sub>ni</sub>	_	16	_	nV/√Hz	f = 1 kHz
Input Noise Current Density	i <sub>ni</sub>	_	3	_	fA/√Hz	f = 1 kHz

## MCP6283/MCP6285 CHIP SELECT (CS) SPECIFICATIONS

Electrical Characteristics: Unless otherwise indicated,  $T_A = \pm 25^{\circ}C$ ,  $V_{DD} = +2.2V$  to +5.5V,  $V_{SS} = GND$ ,  $V_{OUT} \approx V_{DD}/2$ ,  $V_{CM} = V_{DD}/2$ ,  $V_L = V_{DD}/2$ ,  $R_L = 10 \text{ k}\Omega$  to  $V_L$ ,  $C_L = 60 \text{ pF}$  and  $\overline{CS}$  is tied low. (refer to Figure 1-2 and Figure 1-3). **Parameters** Sym Max Units Conditions Min Тур **CS Low Specifications** CS Logic Threshold, Low  $V_{IL}$  $V_{SS}$  $0.2 V_{DD}$ ٧ CS Input Current, Low 0.01  $\overline{CS} = V_{SS}$ μΑ I<sub>CSL</sub> **CS** High Specifications CS Logic Threshold, High  $V_{IH}$  $0.8 V_{DD}$ ٧  $V_{DD}$ CS Input Current, High 0.7 μΑ  $\overline{CS} = V_{DD}$ I<sub>CSH</sub> μΑ **GND** Current per Amplifier -0.7  $\overline{CS} = V_{DD}$  $I_{SS}$  $\overline{\text{CS}} = V_{\text{DD}}$ Amplifier Output Leakage 0.01 μΑ **Dynamic Specifications (Note 1)** CS Low to Valid Amplifier  $\overline{\text{CS}} \text{ Low} \leq 0.2 \text{ V}_{\text{DD}}, \text{ G} = +1 \text{ V/V},$ 4 10 μs  $t_{ON}$ Output, Turn-on Time  $V_{IN} = V_{DD}/2, V_{OUT} = 0.9 V_{DD}/2,$  $V_{DD} = 5.0V$ CS High to Amplifier Output High-Z  $\overline{\text{CS}}$  High  $\geq$  0.8 V<sub>DD</sub>, G = +1 V/V, 0.01 toff μs  $V_{IN} = V_{DD}/2, V_{OUT} = 0.1 V_{DD}/2$ Hysteresis 0.6 V  $V_{HYST}$ 

Note 1: The input condition (V<sub>IN</sub>) specified applies to both op amp A and B of the MCP6285. The dynamic specification is tested at the output of op amp B (V<sub>OUTB</sub>).

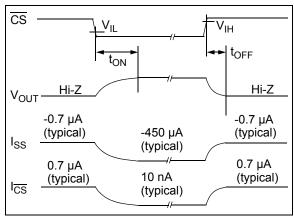


FIGURE 1-1: Timing Diagram for the Chip Select (CS) pin on the MCP6283 and MCP6285.

#### **TEMPERATURE SPECIFICATIONS**

<b>Electrical Characteristics:</b> Unless otherwise indicated, $V_{DD}$ = +2.2V to +5.5V and $V_{SS}$ = GND.								
Parameters	Sym	Min	Тур	Max	Units	Conditions		
Temperature Ranges								
Operating Temperature Range	T <sub>A</sub>	-40	_	+125	°C	Note		
Storage Temperature Range	T <sub>A</sub>	-65	_	+150	°C			
Thermal Package Resistances								
Thermal Resistance, 5L-SOT-23	$\theta_{\sf JA}$	_	256	_	°C/W			
Thermal Resistance, 6L-SOT-23	$\theta_{JA}$	_	230	_	°C/W			
Thermal Resistance, 8L-PDIP	$\theta_{JA}$	_	85	_	°C/W			
Thermal Resistance, 8L-SOIC	$\theta_{\sf JA}$	_	163	_	°C/W			
Thermal Resistance, 8L-MSOP	$\theta_{JA}$	_	206	_	°C/W			
Thermal Resistance, 14L-PDIP	$\theta_{JA}$	_	70	_	°C/W			
Thermal Resistance, 14L-SOIC	$\theta_{JA}$	_	120	_	°C/W			
Thermal Resistance, 14L-TSSOP	$\theta_{JA}$	_	100	_	°C/W			

Note: The Junction Temperature (T<sub>J</sub>) must not exceed the Absolute Maximum specification of +150°C.

#### 1.1 Test Circuits

The test circuits used for the DC and AC tests are shown in Figure 1-2 and Figure 1-2. The bypass capacitors are laid out according to the rules discussed in **Section 4.6 "Supply Bypass"**.

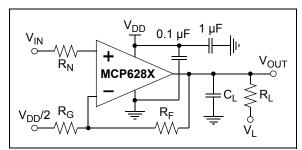


FIGURE 1-2: AC and DC Test Circuit for Most Non-Inverting Gain Conditions.

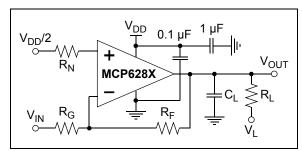


FIGURE 1-3: AC and DC Test Circuit for Most Inverting Gain Conditions.

#### 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

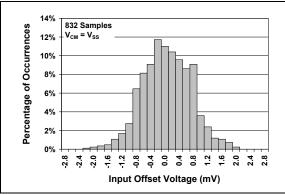
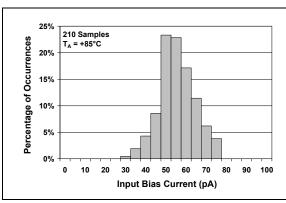
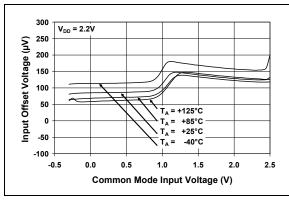


FIGURE 2-1: Input Offset Voltage.



**FIGURE 2-2:** Input Bias Current at  $T_A = +85$  °C.



**FIGURE 2-3:** Input Offset Voltage vs. Common Mode Input Voltage at  $V_{DD} = 2.2V$ .

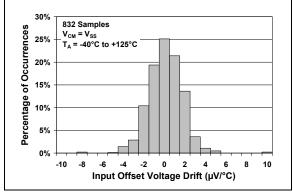
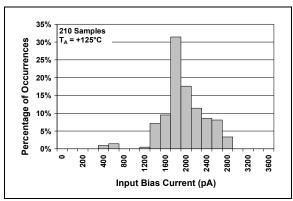
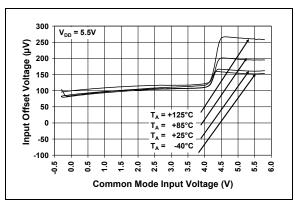


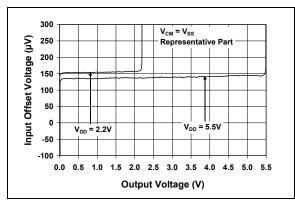
FIGURE 2-4: Input Offset Voltage Drift.



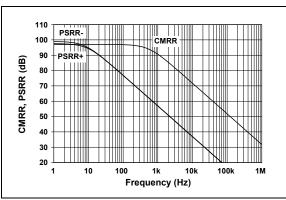
**FIGURE 2-5:** Input Bias Current at  $T_A = +125$  °C.



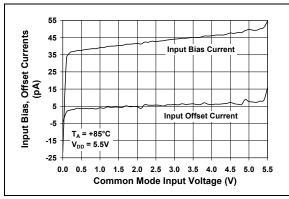
**FIGURE 2-6:** Input Offset Voltage vs. Common Mode Input Voltage at  $V_{DD} = 5.5V$ .



**FIGURE 2-7:** Input Offset Voltage vs. Output Voltage.



**FIGURE 2-8:** CMRR, PSRR vs. Frequency.



**FIGURE 2-9:** Input Bias, Offset Currents vs. Common Mode Input Voltage at  $T_A = +85$ °C.

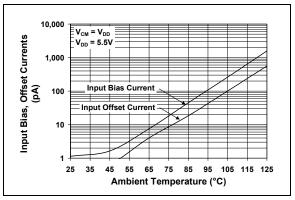
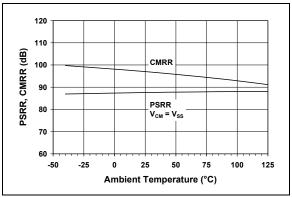
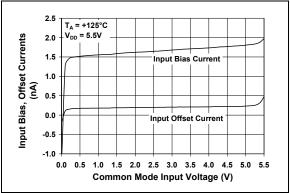


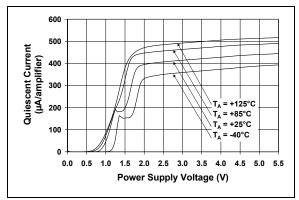
FIGURE 2-10: Input Bias, Input Offset Currents vs. Ambient Temperature.



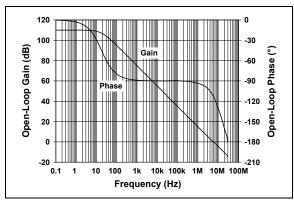
**FIGURE 2-11:** CMRR, PSRR vs. Ambient Temperature.



**FIGURE 2-12:** Input Bias, Offset Currents vs. Common Mode Input Voltage at  $T_A = +125$ °C.



**FIGURE 2-13:** Quiescent Current vs. Power Supply Voltage.



**FIGURE 2-14:** Open-Loop Gain, Phase vs. Frequency.

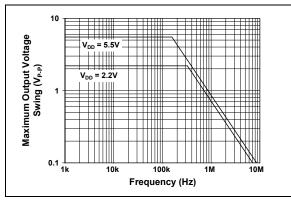
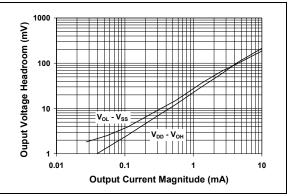
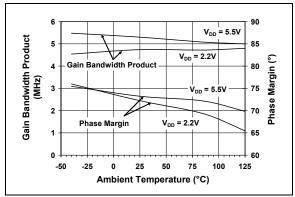


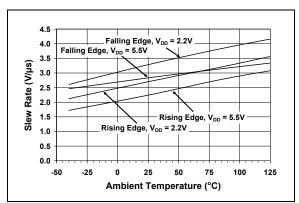
FIGURE 2-15: Maximum Output Voltage Swing vs. Frequency.



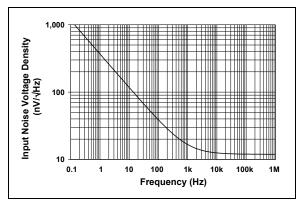
**FIGURE 2-16:** Output Voltage Headroom vs. Output Current Magnitude.



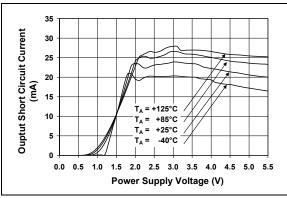
**FIGURE 2-17:** Gain Bandwidth Product, Phase Margin vs. Ambient Temperature.



**FIGURE 2-18:** Slew Rate vs. Ambient Temperature.



**FIGURE 2-19:** Input Noise Voltage Density vs. Frequency.



**FIGURE 2-20:** Output Short Circuit Current vs. Power Supply Voltage.

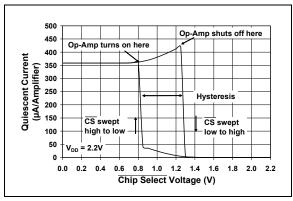
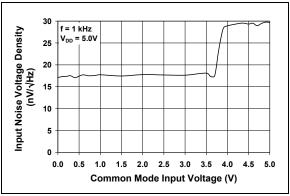


FIGURE 2-21: Quiescent Current vs. Chip Select (CS) Voltage at  $V_{DD}$  = 2.2V (MCP6283 and MCP6285 only).



**FIGURE 2-22:** Input Noise Voltage Density vs. Common Mode Input Voltage at 1 kHz.

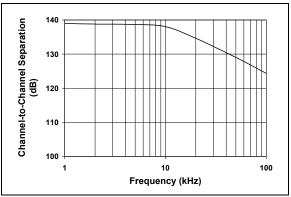


FIGURE 2-23: Channel-to-Channel Separation vs. Frequency (MCP6282 and MCP6284 only).

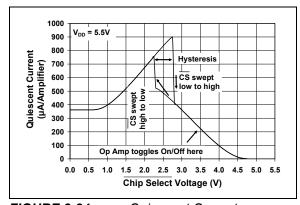
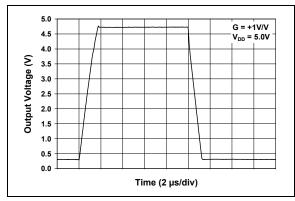
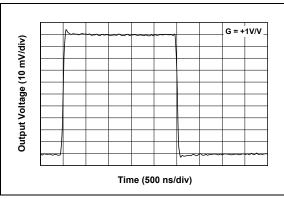


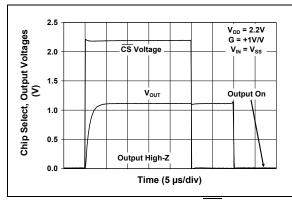
FIGURE 2-24: Quiescent Current vs. Chip Select (CS) Voltage at  $V_{DD}$  = 5.5V (MCP6283 and MCP6285 only).



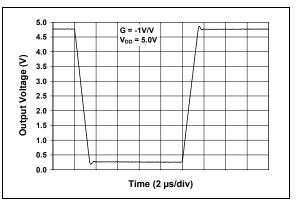
**FIGURE 2-25:** Large-Signal, Non-inverting Pulse Response.



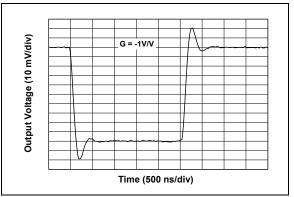
**FIGURE 2-26:** Small-Signal, Non-inverting Pulse Response.



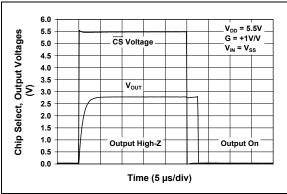
**FIGURE 2-27:** Chip Select (CS) to Amplifier Output Response Time at  $V_{DD}$  = 2.2V (MCP6283 and MCP6285 only).



**FIGURE 2-28:** Large-Signal, Inverting Pulse Response.

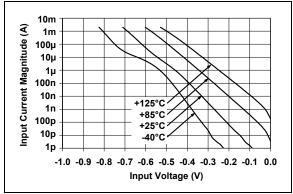


**FIGURE 2-29:** Small-Signal, Inverting Pulse Response.



**FIGURE 2-30:** Chip Select  $(\overline{CS})$  to Amplifier Output Response Time at  $V_{DD} = 5.5V$  (MCP6283 and MCP6285 only).

## MCP6281/1R/2/3/4/5



**FIGURE 2-31:** Measured Input Current vs. Input Voltage (below  $V_{SS}$ ).

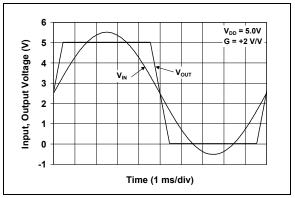


FIGURE 2-32: The MCP6281/1R/2/3/4/5 Show No Phase Reversal.

#### 3.0 PIN DESCRIPTIONS

Descriptions of the pins are listed in Table 3-1 (single op amps) and Table 3-2 (dual and quad op amps).

TABLE 3-1: PIN FUNCTION TABLE FOR SINGLE OP AMPS

МСР6	281	MCP6281R	MCP6	283		
PDIP, SOIC, MSOP	SOT-23-5	SOT-23-5	PDIP, SOIC, MSOP	SOT-23-6	Symbol	Description
6	1	1	6	1	V <sub>OUT</sub>	Analog Output
2	4	4	2	4	V <sub>IN</sub> -	Inverting Input
3	3	3	3	3	V <sub>IN</sub> +	Non-inverting Input
7	5	2	7	6	$V_{DD}$	Positive Power Supply
4	2	5	4	2	$V_{SS}$	Negative Power Supply
_	_	_	8	5	CS	Chip Select
1,5,8	_	_	1,5		NC	No Internal Connection

TABLE 3-2: PIN FUNCTION TABLE FOR DUAL AND QUAD OP AMPS

TABLE 3-2. FIN FUNCTION TABLE FOR DUAL AND QUAD OF AMIFS								
MCP6282	MCP6284	MCP6285	Symbol	Description				
PDIP, SOIC, MSOP	PDIP, SOIC, TSSOP	PDIP, SOIC, MSOP	Symbol	Description				
1	1	_	V <sub>OUTA</sub>	Analog Output (op amp A)				
2	2	2	V <sub>INA</sub> -	Inverting Input (op amp A)				
3	3	3	V <sub>INA</sub> +	Non-inverting Input (op amp A)				
8	4	8	$V_{DD}$	Positive Power Supply				
5	5	_	V <sub>INB</sub> +	Non-inverting Input (op amp B)				
6	6	6	V <sub>INB</sub> -	Inverting Input (op amp B)				
7	7	7	V <sub>OUTB</sub>	Analog Output (op amp B)				
_	8	_	V <sub>OUTC</sub>	Analog Output (op amp C)				
_	9	_	V <sub>INC</sub> -	Inverting Input (op amp C)				
_	10	_	V <sub>INC</sub> +	Non-inverting Input (op amp C)				
4	11	4	V <sub>SS</sub>	Negative Power Supply				
_	12	_	V <sub>IND</sub> +	Non-inverting Input (op amp D)				
_	13	_	V <sub>IND</sub> -	Inverting Input (op amp D)				
_	14	_	V <sub>OUTD</sub>	Analog Output (op amp D)				
_	_	1	V <sub>OUTA</sub> /	Analog Output (op amp A)/Non-				
			V <sub>INB</sub> +	inverting Input (op amp B)				
_	_	5	CS	Chip Select				

#### 3.1 Analog Outputs

The output pins are low-impedance voltage sources.

#### 3.2 Analog Inputs

The non-inverting and inverting inputs are highimpedance CMOS inputs with low bias currents.

### 3.3 MCP6285's V<sub>OUTA</sub>/V<sub>INB</sub>+ Pin

For the MCP6285 only, the output of op amp A is connected directly to the non-inverting input of op amp B; this is the  $V_{OUTA}/V_{INB}$ + pin. This connection makes it possible to provide a Chip Select pin for duals in 8-pin packages.

## 3.4 Chip Select Digital Input (CS)

This is a CMOS, Schmitt-triggered input that places the part into a low-power mode of operation.

#### 3.5 Power Supply Pins

The positive power supply ( $V_{DD}$ ) is 2.2V to 6.0V higher than the negative power supply ( $V_{SS}$ ). For normal operation, the other pins are between  $V_{SS}$  and  $V_{DD}$ .

Typically, these parts are used in a single (positive) supply configuration. In this case,  $V_{SS}$  is connected to ground and  $V_{DD}$  is connected to the supply.  $V_{DD}$  will need bypass capacitors.

#### 4.0 APPLICATION INFORMATION

The MCP6281/1R/2/3/4/5 family of op amps is manufactured using Microchip's state-of-the-art CMOS process. This family is specifically designed for low-cost, low-power and general purpose applications. The low supply voltage, low quiescent current and wide bandwidth makes the MCP6281/1R/2/3/4/5 ideal for battery-powered applications.

#### 4.1 Rail-to-Rail Inputs

#### 4.1.1 PHASE REVERSAL

The MCP6281/1R/2/3/4/5 op amp is designed to prevent phase reversal when the input pins exceed the supply voltages. Figure 2-32 shows the input voltage exceeding the supply voltage without any phase reversal.

## 4.1.2 INPUT VOLTAGE AND CURRENT LIMITS

The ESD protection on the inputs can be depicted as shown in Figure 4-1. This structure was chosen to protect the input transistors, and to minimize input bias current ( $I_B$ ). The input ESD diodes clamp the inputs when they try to go more than one diode drop below  $V_{SS}$ . They also clamp any voltages that go too far above  $V_{DD}$ ; their breakdown voltage is high enough to allow normal operation, and low enough to bypass quick ESD events within the specified limits.

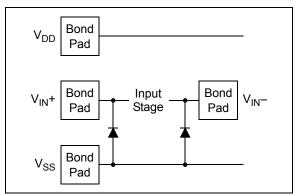


FIGURE 4-1: Simplified Analog Input ESD Structures.

In order to prevent damage and/or improper operation of these op amps, the circuit they are in must limit the currents and voltages at the  $V_{IN}+$  and  $V_{IN}-$  pins (see **Absolute Maximum Ratings †** at the beginning of **Section 1.0 "Electrical Characteristics"**). Figure 4-2 shows the recommended approach to protecting these inputs. The internal ESD diodes prevent the input pins  $(V_{IN}+$  and  $V_{IN}-)$  from going too far below ground, and the resistors  $R_1$  and  $R_2$  limit the possible current drawn out of the input pins. Diodes  $D_1$  and  $D_2$  prevent the input pins  $(V_{IN}+$  and  $V_{IN}-)$  from going too far above

 $V_{DD}$ , and dump any currents onto  $V_{DD}$ . When implemented as shown, resistors  $R_1$  and  $R_2$  also limit the current through  $D_1$  and  $D_2$ .

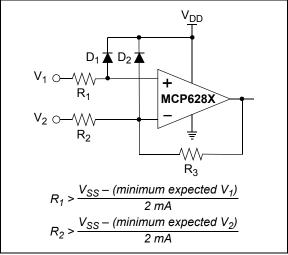


FIGURE 4-2: Protecting the Analog Inputs.

It is also possible to connect the diodes to the left of resistors  $\mathsf{R}_1$  and  $\mathsf{R}_2.$  In this case, current through the diodes  $\mathsf{D}_1$  and  $\mathsf{D}_2$  needs to be limited by some other mechanism. The resistors then serve as in-rush current limiters; the DC current into the input pins (V<sub>IN</sub>+ and V<sub>IN</sub>-) should be very small.

A significant amount of current can flow out of the inputs when the common mode voltage ( $V_{CM}$ ) is below ground ( $V_{SS}$ ); see Figure 2-31. Applications that are high impedance may need to limit the usable voltage range.

#### 4.1.3 NORMAL OPERATION

The input stage of the MCP6281/1R/2/3/4/5 op amps use two differential CMOS input stages in parallel. One operates at low common mode input voltage (V $_{\rm CM}$ ), while the other operates at high V $_{\rm CM}$ . With this topology, the device operates with V $_{\rm CM}$  up to 0.3V above V $_{\rm DD}$  and 0.3V below V $_{\rm SS}$ .

There is a transition in input behavior as  $V_{CM}$  is changed. It occurs when  $V_{CM}$  is near  $V_{DD}-1.2V$  (see Figure 2-3 and Figure 2-6). For the best distortion performance with non-inverting gains, avoid these regions of operation.

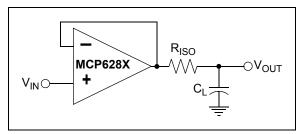
### 4.2 Rail-to-Rail Output

The output voltage range of the MCP6281/1R/2/3/4/5 op amp is V $_{DD}$  – 15 mV (min.) and V $_{SS}$  + 15 mV (max.) when R $_{L}$  = 10 k $\Omega$  is connected to V $_{DD}$ /2 and V $_{DD}$  = 5.5V. Refer to Figure 2-16 for more information.

#### 4.3 Capacitive Loads

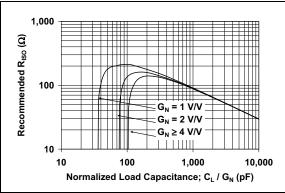
Driving large capacitive loads can cause stability problems for voltage feedback op amps. As the load capacitance increases, the feedback loop's phase margin decreases and the closed-loop bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. A unity-gain buffer (G = +1) is the most sensitive to capacitive loads, though all gains show the same general behavior.

When driving large capacitive loads with these op amps (e.g., > 100 pF when G = +1), a small series resistor at the output ( $R_{ISO}$  in Figure 4-3) improves the feedback loop's phase margin (stability) by making the output load resistive at higher frequencies. The bandwidth will generally be lower than the bandwidth with no capacitive load.



**FIGURE 4-3:** Output Resistor,  $R_{ISO}$  stabilizes large capacitive loads.

Figure 4-4 gives recommended  $R_{ISO}$  values for different capacitive loads and gains. The x-axis is the normalized load capacitance  $(C_L/G_N)$ , where  $G_N$  is the circuit's noise gain. For non-inverting gains,  $G_N$  and the Signal Gain are equal. For inverting gains,  $G_N$  is 1+|Signal Gain| (e.g., -1 V/V gives  $G_N$  = +2 V/V).



**FIGURE 4-4:** Recommended R<sub>ISO</sub> Values for Capacitive Loads.

After selecting  $R_{\rm ISO}$  for your circuit, double-check the resulting frequency response peaking and step response overshoot. Modify  $R_{\rm ISO}$ 's value until the response is reasonable. Bench evaluation and simulations with the MCP6281/1R/2/3/4/5 SPICE macro model are helpful.

## 4.4 MCP628X Chip Select (CS)

The MCP6283 and MCP6285 are single and dual op amps with Chip Select ( $\overline{CS}$ ), respectively. When  $\overline{CS}$  is pulled high, the supply current drops to 0.7  $\mu$ A (typical) and flows through the  $\overline{CS}$  pin to V<sub>SS</sub>. When this happens, the amplifier output is put into a high-impedance state. By pulling  $\overline{CS}$  low, the amplifier is enabled. The  $\overline{CS}$  pin has an internal 5 M $\Omega$  (typical) pull-down resistor connected to V<sub>SS</sub>, so it will go low if the  $\overline{CS}$  pin is left floating. Figure 1-1 shows the output voltage and supply current response to a  $\overline{CS}$  pulse.

## 4.5 Cascaded Dual Op Amps (MCP6285)

The MCP6285 is a dual op amp with Chip Select ( $\overline{\text{CS}}$ ). The Chip Select input is available on what would be the non-inverting input of a standard dual op amp (pin 5). This pin is available because the output of op amp A connects to the non-inverting input of op amp B, as shown in Figure 4-5. The Chip Select input, which can be connected to a microcontroller I/O line, puts the device in Low-power mode. Refer to Section 4.4 "MCP628X Chip Select (CS)".

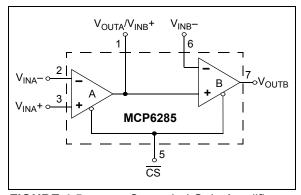


FIGURE 4-5: Cascaded Gain Amplifier.

The output of op amp A is loaded by the input impedance of op amp B, which is typically  $10^{13}\Omega||6$  pF, as specified in the DC specification table (Refer to **Section 4.3 "Capacitive Loads"** for further details regarding capacitive loads).

The common mode input range of these op amps is specified in the data sheet as  $V_{SS}-300~\text{mV}$  and  $V_{DD}+300~\text{mV}$ . However, since the output of op amp A is limited to  $V_{OL}$  and  $V_{OH}$  (20 mV from the rails with a 10 k $\Omega$  load), the non-inverting input range of op amp B is limited to the common mode input range of  $V_{SS}+20~\text{mV}$  and  $V_{DD}-20~\text{mV}$ .

#### 4.6 Supply Bypass

With this family of operational amplifiers, the power supply pin ( $V_{DD}$  for single-supply) should have a local bypass capacitor (i.e., 0.01  $\mu$ F to 0.1  $\mu$ F) within 2 mm for good, high-frequency performance. It also needs a bulk capacitor (i.e., 1  $\mu$ F or larger) within 100 mm to provide large, slow currents. This bulk capacitor can be shared with nearby analog parts.

#### 4.7 Unused Op Amps

An unused op amp in a quad package (MCP6284) should be configured as shown in Figure 4-6. These circuits prevent the output from toggling and causing crosstalk. Circuits A sets the op amp at its minimum noise gain. The resistor divider produces any desired reference voltage within the output voltage range of the op amp; the op amp buffers that reference voltage. Circuit B uses the minimum number of components and operates as a comparator, but it may draw more current.

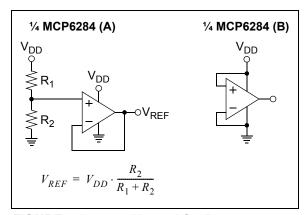
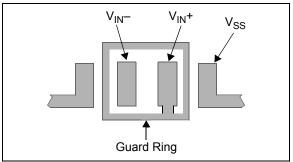


FIGURE 4-6: Unused Op Amps.

#### 4.8 PCB Surface Leakage

In applications where low input bias current is critical, Printed Circuit Board (PCB) surface-leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is  $10^{12}\Omega$ . A 5V difference would cause 5 pA of current to flow, which is greater than the MCP6281/1R/2/3/4/5 family's bias current at +25°C (1 pA, typical).

The easiest way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 4-7.



**FIGURE 4-7:** Example Guard Ring Layout for Inverting Gain.

- For Inverting Gain and Transimpedance Amplifiers (convert current to voltage, such as photo detectors):
  - a.Connect the guard ring to the non-inverting input pin  $(V_{IN}^+)$ . This biases the guard ring to the same reference voltage as the op amp (e.g.,  $V_{DD}/2$  or ground).
  - b.Connect the inverting pin (V<sub>IN</sub>-) to the input with a wire that does not touch the PCB surface.
- 2. Non-inverting Gain and Unity-Gain Buffer:
  - a.Connect the non-inverting pin  $(V_{\rm IN}^+)$  to the input with a wire that does not touch the PCB surface.
  - b.Connect the guard ring to the inverting input pin (V<sub>IN</sub>-). This biases the guard ring to the common mode input voltage.

### 4.9 Application Circuits

#### 4.9.1 SALLEN-KEY HIGH-PASS FILTER

The MCP6281/1R/2/3/4/5 op amps can be used in active-filter applications. Figure 4-8 shows a second-order Sallen-Key high-pass filter with a gain of 1. The output bias voltage is set by the  $V_{\rm DD}/2$  reference, which can be changed to any voltage within the output voltage range.

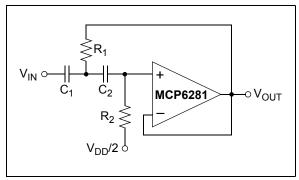


FIGURE 4-8: Sallen-Key High-Pass Filter.

This filter, and others, can be designed using Microchip's Design Aids; see Section 5.2 "FilterLab® Software" and Section 5.3 "Mindi™ Circuit Designer & Simulator".

#### 4.9.2 INVERTING MILLER INTEGRATOR

Analog integrators are used in filters, control loops and measurement circuits. Figure 4-9 shows the most common implementation, the inverting Miller integrator. The non-inverting input is at  $V_{DD}/2$  so that the op amp properly biases up. The switch (SW) is used to zero the output in some applications. Other applications use a feedback loop to keep the output within its linear range of operation.

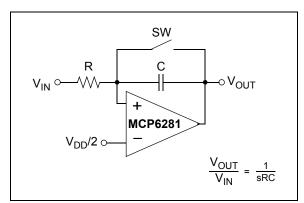


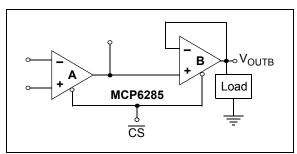
FIGURE 4-9: Miller Integrator.

## 4.9.3 CASCADED OP AMP APPLICATIONS

The MCP6285 provides the flexibility of Low-power mode for dual op amps in an 8-pin package. The MCP6285 eliminates the added cost and space in battery-powered applications by using two single op amps with Chip Select lines or a 10-pin device with one Chip Select line for both op amps. Since the two op amps are internally cascaded, this device cannot be used in circuits that require active or passive elements between the two op amps. However, there are several applications where this op amp configuration with Chip Select line becomes suitable. The circuits below show possible applications for this device.

#### 4.9.3.1 Load Isolation

With the cascaded op amp configuration, op amp B can be used to isolate the load from op amp A. In applications where op amp A is driving capacitive or low resistance loads in the feedback loop (such as an integrator circuit or filter circuit), the op amp may not have sufficient source current to drive the load. In this case, op amp B can be used as a buffer.



**FIGURE 4-10:** Isolating the Load with a Buffer.

#### 4.9.3.2 Cascaded Gain

Figure 4-11 shows a cascaded gain circuit configuration with Chip Select. Op amps A and B are configured in a non-inverting amplifier configuration. In this configuration, it is important to note that the input offset voltage of op amp A is amplified by the gain of op amp A and B, as shown below:

$$V_{OUT} = V_{IN}G_AG_B + V_{OSA}G_AG_B + V_{OSB}G_B$$
 Where: 
$$G_A = \text{ op amp A gain}$$
 
$$G_B = \text{ op amp B gain}$$
 
$$V_{OSA} = \text{ op amp A input offset voltage}$$
 
$$V_{OSB} = \text{ op amp B input offset voltage}$$

Therefore, it is recommended to set most of the gain with op amp A and use op amp B with relatively small gain (e.g., a unity-gain buffer).

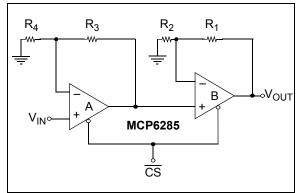


FIGURE 4-11: Cascaded Gain Circuit Configuration.

#### 4.9.3.3 Difference Amplifier

Figure 4-12 shows op amp A configured as a difference amplifier with Chip Select. In this configuration, it is recommended to use well-matched resistors (e.g., 0.1%) to increase the Common Mode Rejection Ratio (CMRR). Op amp B can be used to provide additional gain and isolate the load from the difference amplifier.

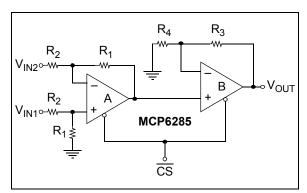


FIGURE 4-12: Difference Amplifier Circuit.

### 4.9.3.4 Buffered Non-inverting Integrator

Figure 4-13 shows a lossy non-inverting integrator that is buffered and has a Chip Select input. Op amp A is configured as a non-inverting integrator. In this configuration, matching the impedance at each input is recommended.  $R_{\text{F}}$  is used to provide a feedback loop at frequencies << 1/( $2\pi R_1 C_1$ ) and makes this a lossy integrator (it has a finite gain at DC). Op amp B is used to isolate the load from the integrator.

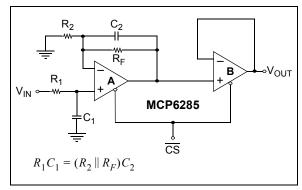
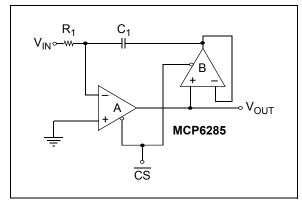


FIGURE 4-13: Buffered Non-inverting Integrator with Chip Select.

## 4.9.3.5 Inverting Integrator with Active Compensation and Chip Select

Figure 4-14 uses an active compensator (op amp B) to compensate for the non-ideal op amp characteristics introduced at higher frequencies. This circuit uses op amp B as a unity-gain buffer to isolate the integration capacitor  $C_1$  from op amp A and drives the capacitor with low-impedance source. Since both op amps are matched very well, they provide a higher quality integrator.



**FIGURE 4-14:** Integrator Circuit with Active Compensation.

## 4.9.3.6 Second-Order MFB Low-Pass Filter with an Extra Pole-Zero Pair

Figure 4-15 is a second-order multiple feedback low-pass filter with Chip Select. Use the FilterLab software from Microchip to determine the R and C values for the op amp A's second-order filter. Op amp B can be used to add a pole-zero pair using  $C_3$ ,  $R_6$ , and  $R_7$ .

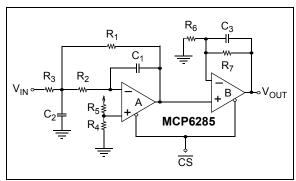


FIGURE 4-15: Second-Order Multiple Feedback Low-Pass Filter with an Extra Pole-Zero Pair.

## 4.9.3.7 Second-Order Sallen-Key Low-Pass Filter with an Extra Pole-Zero Pair

Figure 4-16 is a second-order Sallen-Key low-pass filter with Chip Select. Use the FilterLab software from Microchip to determine the R and C values for the op amp A's second-order filter. Op amp B can be used to add a pole-zero pair using  $C_3$ ,  $R_5$  and  $R_6$ .

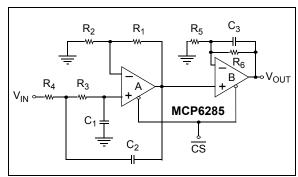


FIGURE 4-16: Second-Order Sallen-Key Low-Pass Filter with an Extra Pole-Zero Pair and Chip Select.

## 4.9.3.8 Capacitorless Second-Order Low-Pass filter with Chip Select

The low-pass filter shown in Figure 4-17 does not require external capacitors and uses only three external resistors; the op amp's GBWP sets the corner frequency.  $R_1$  and  $R_2$  are used to set the circuit gain and  $R_3$  is used to set the Q. To avoid gain peaking in the frequency response, Q needs to be low (lower values need to be selected for  $R_3$ ). Note that the amplifier bandwidth varies greatly over temperature and process. However, this configuration provides a low-cost solution for applications with high bandwidth requirements.

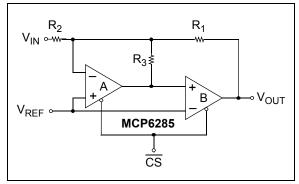


FIGURE 4-17: Capacitorless Second-Order Low-Pass Filter with Chip Select.

#### 5.0 DESIGN AIDS

Microchip provides the basic design tools needed for the MCP6281/1R/2/3/4/5 family of op amps.

#### 5.1 SPICE Macro Model

The latest SPICE macro model for the MCP6281/1R/2/3/4/5 op amps is available on the Microchip web site at www.microchip.com. This model is intended to be an initial design tool that works well in the op amp's linear region of operation over the temperature range. See the model file for information on its capabilities.

Bench testing is a very important part of any design and cannot be replaced with simulations. Also, simulation results using this macro model need to be validated by comparing them to the data sheet specifications and characteristic curves.

## 5.2 FilterLab<sup>®</sup> Software

Microchip's FilterLab<sup>®</sup> software is an innovative software tool that simplifies analog active filter (using op amps) design. Available at no cost from the Microchip web site at www.microchip.com/filterlab, the FilterLab design tool provides full schematic diagrams of the filter circuit with component values. It also outputs the filter circuit in SPICE format, which can be used with the macro model to simulate actual filter performance.

## 5.3 Mindi™ Circuit Designer & Simulator

Microchip's Mindi™ Circuit Designer & Simulator aids in the design of various circuits useful for active filter, amplifier and power-management applications. It is a free online circuit designer & simulator available from the Microchip web site at www.microchip.com/mindi. This interactive circuit designer & simulator enables designers to quickly generate circuit diagrams, simulate circuits. Circuits developed using the Mindi Circuit Designer & Simulator can be downloaded to a personal computer or workstation.

## 5.4 MAPS (Microchip Advanced Part Selector)

MAPS is a software tool that helps semiconductor professionals efficiently identify Microchip devices that fit a particular design requirement. Available at no cost from the Microchip web site at www.microchip.com/maps, the MAPS is an overall selection tool for Microchip's product portfolio that includes Analog, Memory, MCUs and DSCs. Using this tool you can define a filter to sort features for a parametric search of devices and export side-by-side technical comparison reports. Helpful links are also provided for Data sheets, Purchase, and Sampling of Microchip parts.

## 5.5 Analog Demonstration and Evaluation Boards

Microchip offers a broad spectrum of Analog Demonstration and Evaluation Boards that are designed to help you achieve faster time to market. For a complete listing of these boards and their corresponding user's guides and technical information, visit the Microchip web site at www.microchip.com/analogtools.

Two of our boards that are especially useful are:

- P/N SOIC8EV: 8-Pin SOIC/MSOP/TSSOP/DIP Evaluation Board
- P/N SOIC14EV: 14-Pin SOIC/TSSOP/DIP Evaluation Board

#### 5.6 Application Notes

The following Microchip Application Notes are available on the Microchip web site at www.microchip.com/appnotes and are recommended as supplemental reference resources.

**ADN003:** "Select the Right Operational Amplifier for your Filtering Circuits," DS21821

**AN722:** "Operational Amplifier Topologies and DC Specifications," DS00722

**AN723:** "Operational Amplifier AC Specifications and Applications," DS00723

AN884: "Driving Capacitive Loads With Op Amps," DS00884

**AN990:** "Analog Sensor Conditioning Circuits – An Overview," DS00990

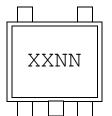
These application notes and others are listed in the design guide:

"Signal Chain Design Guide," DS21825

#### 6.0 PACKAGING INFORMATION

### 6.1 Package Marking Information

5-Lead SOT-23 (MCP6281 and MCP6281R)



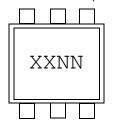
Device	Code			
MCP6281	CHNN			
MCP6281R	EUNN			
Note: Applies to Fill and COT 22				

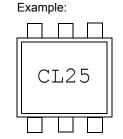
**Note:** Applies to 5-Lead SOT-23.





6-Lead SOT-23 (MCP6283)





8-Lead MSOP

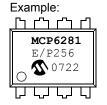






8-Lead PDIP (300 mil)







Legend: XX...X Customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')
NNN Alphanumeric traceability code

e3 Pb-free JEDEC designator for Matte Tin (Sn)

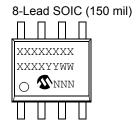
This package is Pb-free. The Pb-free JEDEC designator (e3)

can be found on the outer packaging for this package.

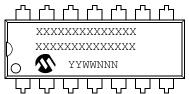
**Note**: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

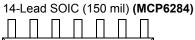
## MCP6281/1R/2/3/4/5

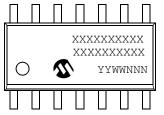
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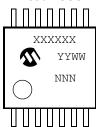
### 14-Lead PDIP (300 mil) (MCP6284)

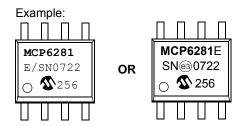




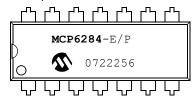


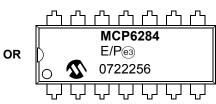
#### 14-Lead TSSOP (MCP6284)



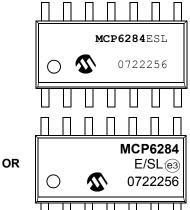


## Example:







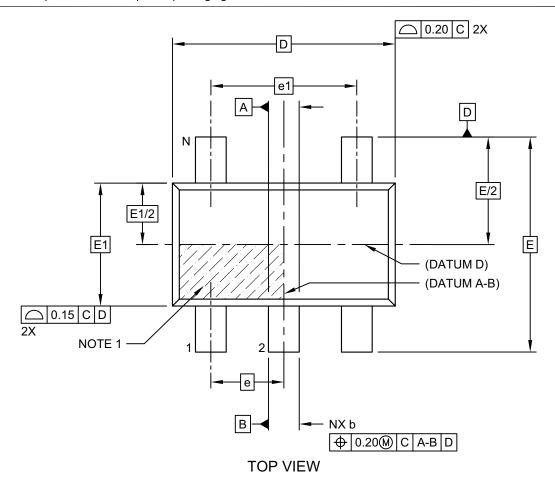


## Example:



## 5-Lead Plastic Small Outline Transistor (OT) [SOT23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



A A2
A A2
A A2
SEE SHEET 2
A1

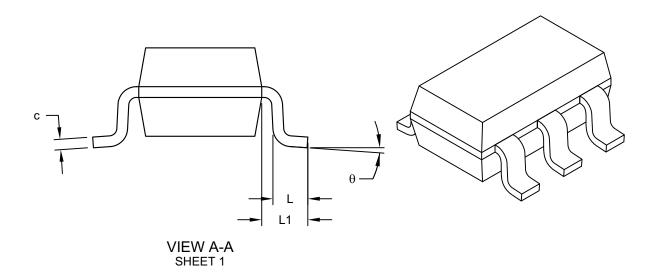
SEATING PLANE

SIDE VIEW

Microchip Technology Drawing C04-091-OT Rev E Sheet 1 of 2

## 5-Lead Plastic Small Outline Transistor (OT) [SOT23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension	Limits	MIN	NOM	MAX	
Number of Pins	N		5		
Pitch	е		0.95 BSC		
Outside lead pitch	e1		1.90 BSC		
Overall Height	Α	0.90	ı	1.45	
Molded Package Thickness	A2	0.89	ı	1.30	
Standoff	A1	-	-	0.15	
Overall Width	Е		2.80 BSC		
Molded Package Width	E1		1.60 BSC		
Overall Length	D		2.90 BSC		
Foot Length	L	0.30	-	0.60	
Footprint	L1	0.60 REF			
Foot Angle	ф	0°	-	10°	
Lead Thickness	С	0.08	-	0.26	
Lead Width	b	0.20	-	0.51	

#### Notes:

- 1. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25mm per side.
- 2. Dimensioning and tolerancing per ASME Y14.5M

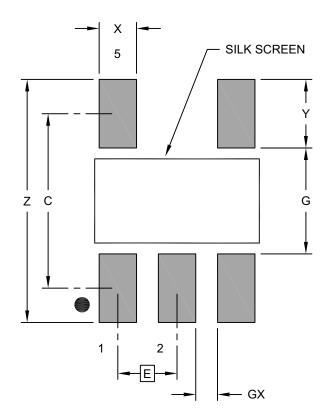
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-091-OT Rev E Sheet 2 of 2

## 5-Lead Plastic Small Outline Transistor (OT) [SOT23]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E	0.95 BSC		
Contact Pad Spacing	С		2.80	
Contact Pad Width (X5)	Х			0.60
Contact Pad Length (X5)	Υ			1.10
Distance Between Pads	G	1.70		
Distance Between Pads	GX	0.35		
Overall Width	Z			3.90

#### Notes:

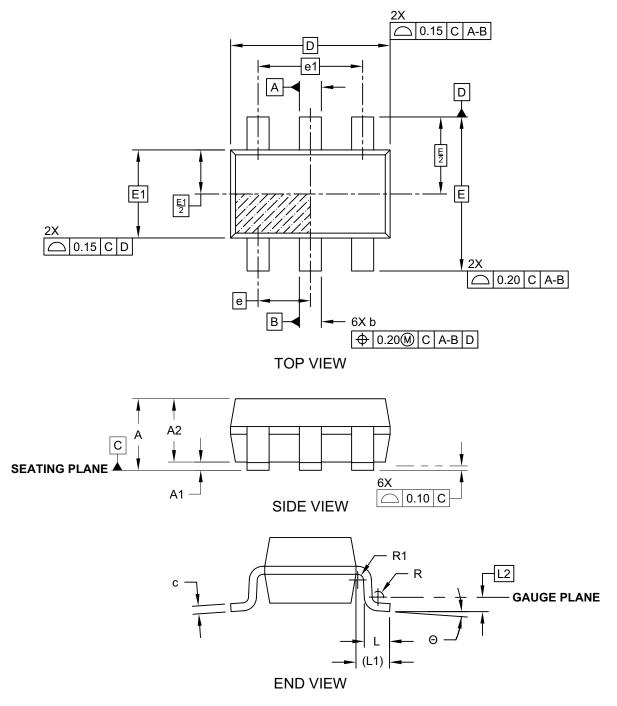
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2091B [OT]

## 6-Lead Plastic Small Outline Transistor (CH, CHY) [SOT-23]

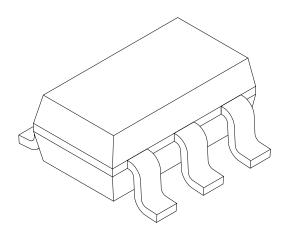
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-028C (CH) Sheet 1 of 2

## 6-Lead Plastic Small Outline Transistor (CH, CHY) [SOT-23]

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS			
Dimension I	Limits	MIN	NOM	MAX	
Number of Leads	N		6		
Pitch	е		0.95 BSC		
Outside lead pitch	e1		1.90 BSC		
Overall Height	Α	0.90	-	1.45	
Molded Package Thickness	A2	0.89	1.15	1.30	
Standoff	A1	0.00	-	0.15	
Overall Width	Е	2.80 BSC			
Molded Package Width	E1		1.60 BSC		
Overall Length	D		2.90 BSC		
Foot Length	L	0.30	0.45	0.60	
Footprint	L1		0.60 REF		
Seating Plane to Gauge Plane	L1	0.25 BSC			
Foot Angle	ф	0°	ı	10°	
Lead Thickness	С	0.08	-	0.26	
Lead Width	b	0.20	-	0.51	

#### Notes:

Note:

- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.25mm per side.
- 2. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

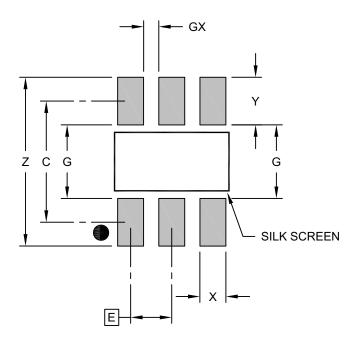
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-028C (CH) Sheet 2 of 2

Note:

## 6-Lead Plastic Small Outline Transistor (CH, CHY) [SOT-23]

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E 0.95 BSC			
Contact Pad Spacing	С		2.80	
Contact Pad Width (X3)	Х			0.60
Contact Pad Length (X3)	Υ			1.10
Distance Between Pads	G	1.70		
Distance Between Pads	GX	0.35		
Overall Width	Z			3.90

#### Notes:

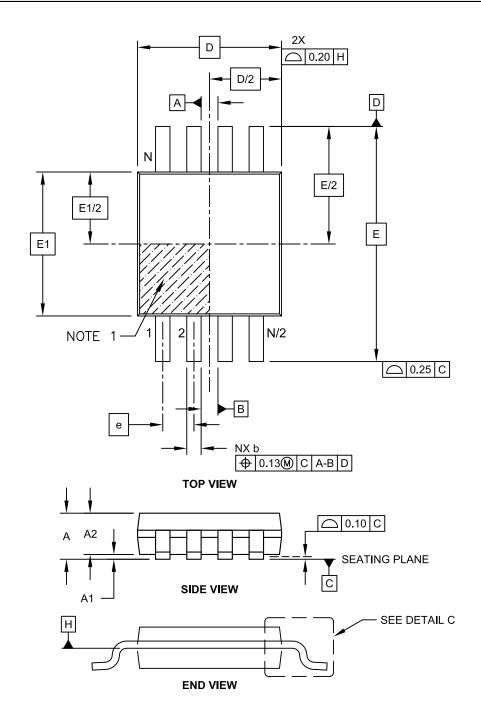
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2028B (CH)

### 8-Lead Plastic Micro Small Outline Package (MS) [MSOP]

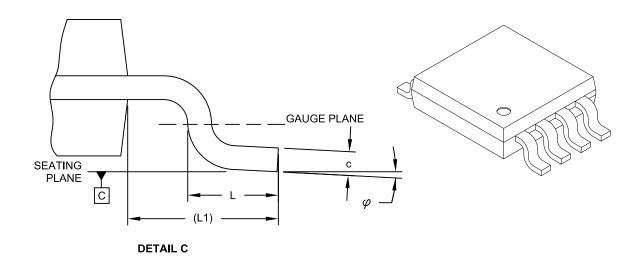
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing C04-111C Sheet 1 of 2

## 8-Lead Plastic Micro Small Outline Package (MS) [MSOP]

For the most current package drawings, please see the Microchip Packaging Specification located at Note: http://www.microchip.com/packaging



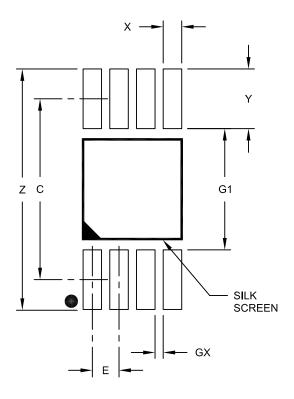
	MILLIMETERS				
Dimension	Limits	MIN	NOM	MAX	
Number of Pins	N		8		
Pitch	е		0.65 BSC		
Overall Height	Α	-	1	1.10	
Molded Package Thickness	A2	0.75	0.85	0.95	
Standoff	A1	0.00	-	0.15	
Overall Width	E	4.90 BSC			
Molded Package Width	E1		3.00 BSC		
Overall Length	D		3.00 BSC		
Foot Length	L	0.40	0.60	0.80	
Footprint	L1	0.95 REF			
Foot Angle	φ	0°	-	8°	
Lead Thickness	С	0.08	-	0.23	
Lead Width	b	0.22	-	0.40	

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
  3. Dimensioning and tolerancing per ASME Y14.5M.
- - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-111C Sheet 2 of 2

## 8-Lead Plastic Micro Small Outline Package (MS) [MSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	Units	N	MILLIMETERS	
Dimension	Dimension Limits		NOM	MAX
Contact Pitch	Е	0.65 BSC		
Contact Pad Spacing	С		4.40	
Overall Width	Z			5.85
Contact Pad Width (X8)	X1			0.45
Contact Pad Length (X8)	Y1			1.45
Distance Between Pads	G1	2.95		
Distance Between Pads	GX	0.20	·	·

#### Notes:

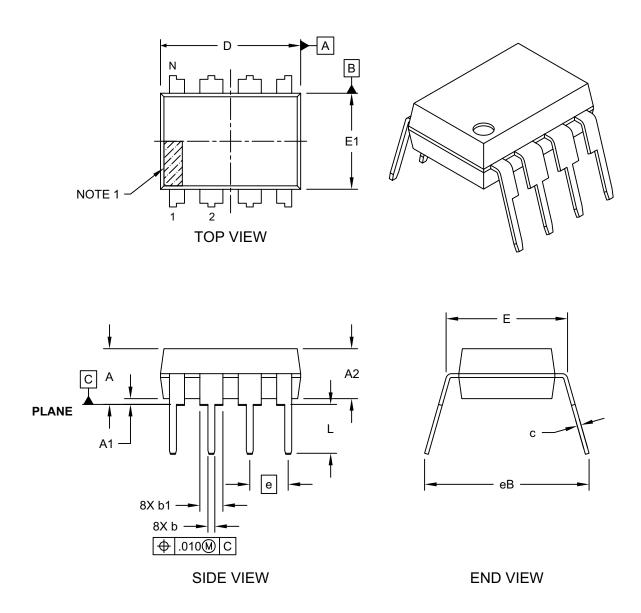
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2111A

## 8-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

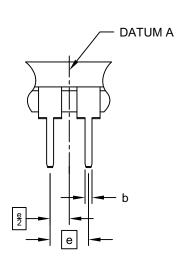
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



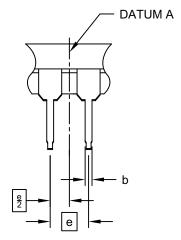
Microchip Technology Drawing No. C04-018-P Rev E Sheet 1 of 2

## 8-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



## ALTERNATE LEAD DESIGN (NOTE 5)



Units		INCHES		
Dimension	Limits	MIN	NOM	MAX
Number of Pins	N	8		
Pitch	е	.100 BSC		
Top to Seating Plane	Α	•	-	.210
Molded Package Thickness	A2	.115	.130	.195
Base to Seating Plane	A1	.015	-	-
Shoulder to Shoulder Width	Е	.290	.310	.325
Molded Package Width	E1	.240	.250	.280
Overall Length	D	.348	.365	.400
Tip to Seating Plane	L	.115	.130	.150
Lead Thickness	С	.008	.010	.015
Upper Lead Width	b1	.040	.060	.070
Lower Lead Width	b	.014	.018	.022
Overall Row Spacing §	eВ	-	-	.430

#### Notes:

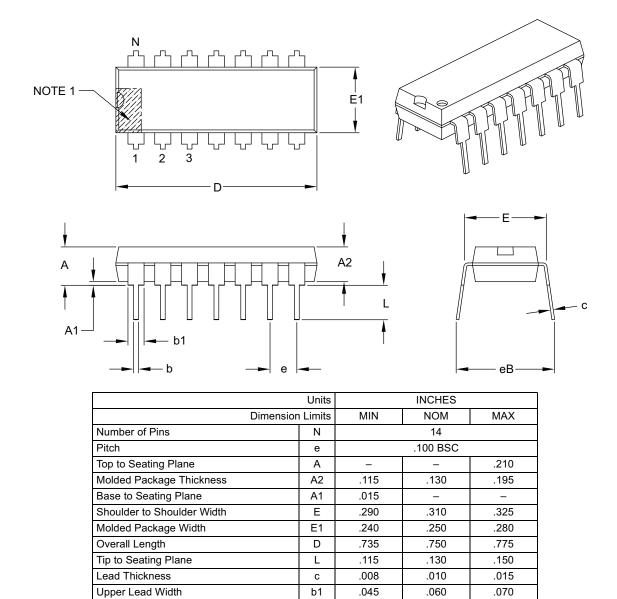
Note:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
- 5. Lead design above seating plane may vary, based on assembly vendor.

Microchip Technology Drawing No. C04-018-P Rev E Sheet 2 of 2

## 14-Lead Plastic Dual In-Line (P) - 300 mil Body [PDIP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located with the hatched area.
- 2. § Significant Characteristic.
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.

b

eВ

.014

.018

4. Dimensioning and tolerancing per ASME Y14.5M.

Lower Lead Width

Overall Row Spacing §

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

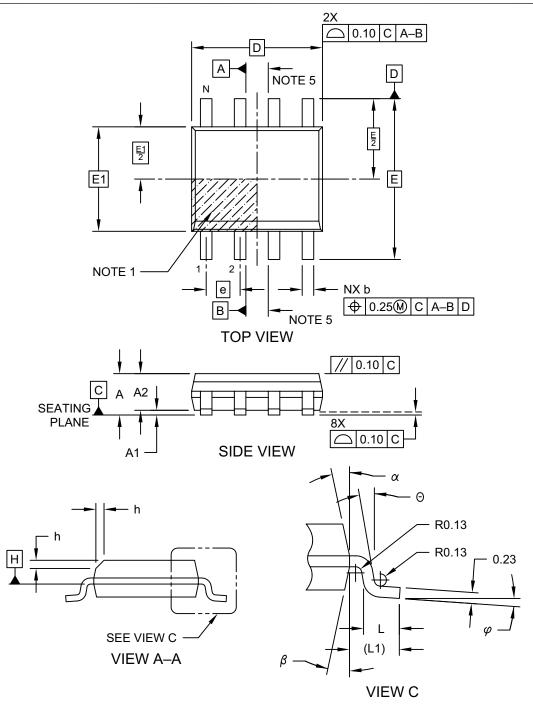
Microchip Technology Drawing C04-005B

.022

.430

## 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm (.150 ln.) Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging

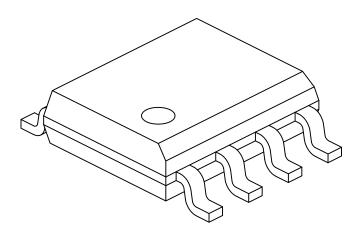


Microchip Technology Drawing No. C04-057-SN Rev E Sheet 1 of 2

Note:

### 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm (.150 ln.) Body [SOIC]

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N	8		
Pitch	е	1.27 BSC		
Overall Height	Α	ı	ı	1.75
Molded Package Thickness	A2	1.25	ı	-
Standoff §	A1	0.10	ı	0.25
Overall Width	Е	6.00 BSC		
Molded Package Width	E1	3.90 BSC		
Overall Length	D	4.90 BSC		
Chamfer (Optional)	h	0.25	ı	0.50
Foot Length	L	0.40	ı	1.27
Footprint	L1	1.04 REF		
Foot Angle	$\varphi$	0°	ı	8°
Lead Thickness	С	0.17	ı	0.25
Lead Width	b	0.31	-	0.51
Mold Draft Angle Top	α	5°	- 1	15°
Mold Draft Angle Bottom	β	5°	-	15°

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

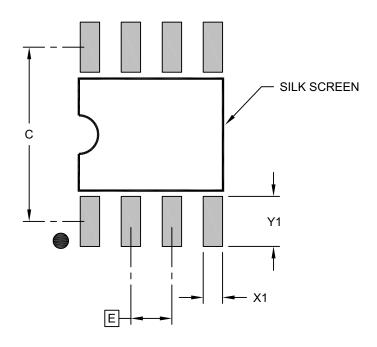
REF: Reference Dimension, usually without tolerance, for information purposes only.

5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-057-SN Rev E Sheet 2 of 2

## 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm Body [SOIC]

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



### RECOMMENDED LAND PATTERN

	Units MILLIMETERS			S
Dimension	Dimension Limits		NOM	MAX
Contact Pitch	Е	1.27 BSC		
Contact Pad Spacing	C		5.40	
Contact Pad Width (X8)	X1			0.60
Contact Pad Length (X8)	Y1			1.55

#### Notes:

Note:

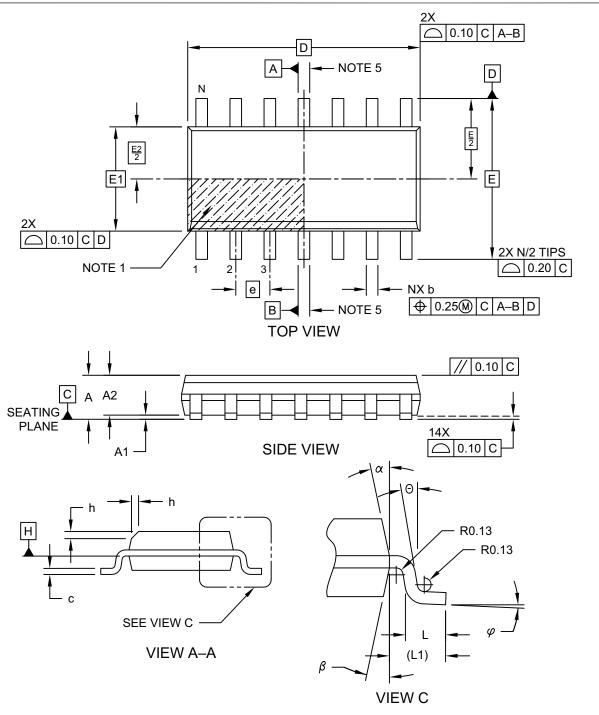
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-2057-SN Rev E

## 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

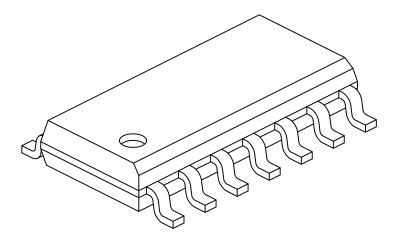
**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Microchip Technology Drawing No. C04-065-SL Rev D Sheet 1 of 2

## 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension	Limits	MIN	NOM	MAX	
Number of Pins	N		14		
Pitch	е	1.27 BSC			
Overall Height	Α	-	-	1.75	
Molded Package Thickness	A2	1.25	-	-	
Standoff §	A1	0.10	-	0.25	
Overall Width	Е	6.00 BSC			
Molded Package Width	E1	3.90 BSC			
Overall Length	D	8.65 BSC			
Chamfer (Optional)	h	0.25	-	0.50	
Foot Length	L	0.40	-	1.27	
Footprint	L1				
Lead Angle	Θ	0°	-	-	
Foot Angle	$\varphi$	0°	-	8°	
Lead Thickness	С	0.10	-	0.25	
Lead Width	b	0.31	-	0.51	
Mold Draft Angle Top	α	5°	_	15°	
Mold Draft Angle Bottom	β	5° - 15°			

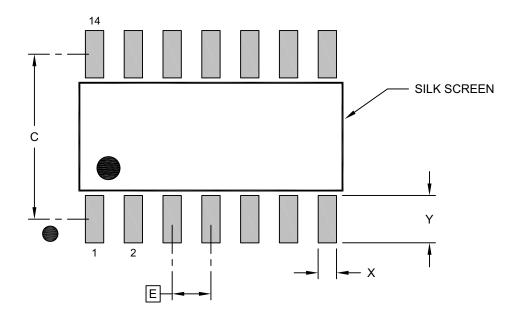
#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.
- 5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-065-SL Rev D Sheet 2 of 2  $\,$ 

## 14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



### RECOMMENDED LAND PATTERN

	Units	MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX
Contact Pitch	Е		1.27 BSC	
Contact Pad Spacing	С		5.40	
Contact Pad Width (X14)	Х			0.60
Contact Pad Length (X14)	Υ			1.55

#### Notes:

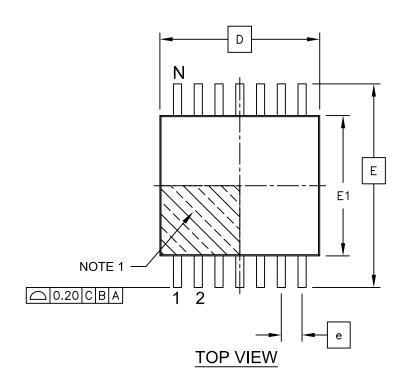
1. Dimensioning and tolerancing per ASME Y14.5M

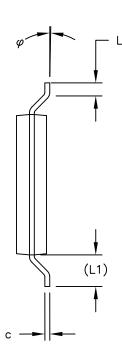
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

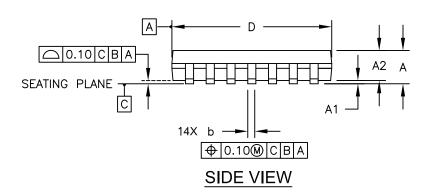
Microchip Technology Drawing No. C04-2065-SL Rev D

## 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



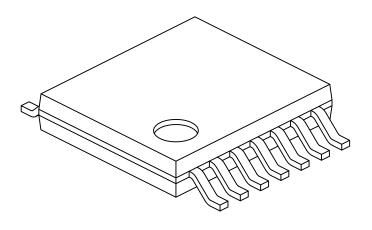




Microchip Technology Drawing C04-087C Sheet 1 of 2

## 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	MILLIMETERS		
Dimension Li		MIN	NOM	MAX
Number of Pins	Ν	14		
Pitch	е		0.65 BSC	
Overall Height	Α	ı	-	1.20
Molded Package Thickness	A2	0.80	1.00	1.05
Standoff	A1	0.05	-	0.15
Overall Width	Е	6.40 BSC		
Molded Package Width	E1	4.30	4.40	4.50
Molded Package Length	D	4.90	5.00	5.10
Foot Length	L	0.45	0.60	0.75
Footprint	(L1)	1.00 REF		
Foot Angle	φ	0°	-	8°
Lead Thickness	С	0.09	-	0.20
Lead Width	b	0.19	-	0.30

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 3. Dimensioning and tolerancing per ASME Y14.5M

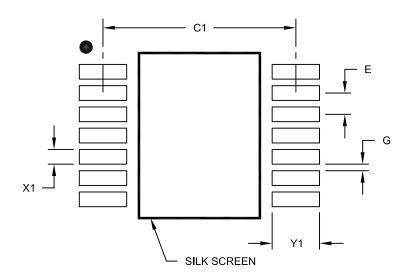
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-087C Sheet 2 of 2

## 14-Lead Plastic Thin Shrink Small Outline (ST) - 4.4 mm Body [TSSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



### RECOMMENDED LAND PATTERN

	Units	Units MILLIMETERS		
Dimension	Limits MIN		NOM	MAX
Contact Pitch	E			
Contact Pad Spacing	C1		5.90	
Contact Pad Width (X14)	X1			0.45
Contact Pad Length (X14)	Y1			1.45
Distance Between Pads	G	0.20		

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2087A

# MCP6281/1R/2/3/4/5

NOTES:

#### APPENDIX A: REVISION HISTORY

### **Revision F (December 2019)**

The following is the list of modifications:

 Updated Section 6.0 "Packaging Information".

### Revision E (February 2008)

The following is the list of modifications:

- Updated notes to Section 1.0 "Electrical Characteristics".
- Increased absolute maximum voltage range of input pins. Increased maximum operating supply voltage (V<sub>DD</sub>).
- 3. Added Section 1.1 "Test Circuits".
- 4. Added Figure 2-32.
- Updated Table 3-1 and Table 3-2 in Section 3.0 "Pin Descriptions".
- Added Section 4.1.1 "Phase Reversal", Section 4.1.2 "Input Voltage and Current Limits", and Section 4.1.3 "Normal Operation".
- 7. Added Section 4.7 "Unused Op Amps".
- 8. Updated Section 5.0 "Design AIDS".
- Updated package outline drawings in Section 6.0 "Packaging Information".

#### **Revision D (December 2004)**

The following is the list of modifications:

- Added SOT-23-5 packages for the MCP6281 and MCP6281R single op amps.
- Added SOT-23-6 package for the MCP6283 single op amp.
- 3. Added Section 3.0 "Pin Descriptions".
- 4. Corrected application circuits (Section 4.9 "Application Circuits").
- Added SOT-23-5 and SOT-23-6 packages and corrected package marking information (Section 6.0 "Packaging Information").
- 6. Added Appendix A: Revision History.

## Revision C (June 2004)

The following is the list of modifications:

1. Undocumented changes.

#### **Revision B (October 2003)**

The following is the list of modifications:

1. Undocumented changes.

#### Revision A (June 2003)

· Original data sheet release.

# MCP6281/1R/2/3/4/5

NOTES:

## PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO	X	/XX	Exa	mples:	
Device Ten	 nperature P	ackage	a)	MCP6281-E/SN:	Extended Temperature, 8LD SOIC package.
F	Range	·	b)	MCP6281-E/MS:	Extended Temperature, 8LD MSOP package.
			c)	MCP6281-E/P:	Extended Temperature, 8LD PDIP package.
Device:	MCP6281: MCP6281T:	Single Op Amp Single Op Amp (Tape and Reel)	d)	MCP6281T-E/OT:	Tape and Reel, Extended Temperature, 5LD SOT-23 package.
	MCP6281RT:	(SOIC, MSOP, SOT-23-5) Single Op Amp (Tape and Reel) (SOT-23-5)	e)	MCP6281RT-E/OT:	
	MCP6282: MCP6282T:	Dual Op Amp Dual Op Amp (Tape and Reel) (SOIC, MSOP)	a)	MCP6282-E/SN:	Extended Temperature, 8LD SOIC package.
	MCP6283: MCP6283T:	Single Op Amp with CS Single Op Amp with CS	b)	MCP6282-E/MS:	Extended Temperature, 8LD MSOP package.
		(Tape and Reel) (SOIC, MSOP, SOT-23-6)	c)	MCP6282-E/P:	Extended Temperature, 8LD PDIP package.
	MCP6284: MCP6284T:	Quad Op Amp Quad Op Amp (Tape and Reel) (SOIC, TSSOP)	d)	MCP6282T-E/SN:	Tape and Reel, Extended Temperature, 8LD SOIC package.
	MCP6285: MCP6285T:	Dual Op Amp with CS Dual Op Amp with CS	a)	MCP6283-E/SN:	Extended Temperature,
		(Tape and Reel) (SOIC, MSOP)	b)	MCP6283-E/MS:	8LD SOIC package. Extended Temperature, 8LD MSOP package.
Temperature Range:	E = -40°C	to +125°C	c)	MCP6283-E/P:	Extended Temperature, 8LD PDIP package.
Package:		Small Outline Transistor (SOT-23), 6-lead	d)	MCP6283T-E/CH:	Tape and Reel, Extended Temperature, 6LD SOT-23 package.
	MS = Plastic MSOP, 8-lead P = Plastic DIP (300 mil body), 8-lead, 14-lead		a)	MCP6284-E/P:	Extended Temperature, 14LD PDIP package.
OT = Plastic Small Outline Transistor (SOT-23), 5-lead (MCP6281, MCP6281R only) SL = Plastic SOIC (3.90 mm body), 14-lead		b)	MCP6284T-E/SL:	Tape and Reel, Extended Temperature,	
		SOIC, (3.90 mm body), 8-lead TSSOP (4.4 mm body), 14-lead	c)	MCP6284-E/SL:	14LD SOIC package. Extended Temperature, 14LD SOIC package.
			d)	MCP6284-E/ST:	Extended Temperature, 14LD TSSOP package.
			a)	MCP6285-E/SN:	Extended Temperature, 8LD SOIC package.
			b)	MCP6285-E/MS:	Extended Temperature, 8LD MSOP package.
			c)	MCP6285-E/P:	Extended Temperature, 8LD PDIP package.
			d)	MCP6285T-E/SN:	Tape and Reel, Extended Temperature, 8LD SOIC package.

# MCP6281/1R/2/3/4/5

NOTES:

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ISBN: 978-1-5224-5356-7

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