

## 9A Peak Low-Side MOSFET Drivers

#### **Features**

- High Output Current: 9A Peak (Typical)
- Wide Operating Range: 4.5V to 18V (Typical)
- · Minimum Pulse Width: 50 ns
- Latch-Up Proof: Fully Isolated Process is Inherently Immune to Any Latch-Up
- Input Will Withstand Negative Swing of Up to 5V
- · High Capacitive Load Drive: 47,000 pF
- · Low Delay Time: 15 ns (Typical)
- Logic High Input for Any Voltage from 2.4V to  $\ensuremath{V_S}$
- Low Equivalent Input Capacitance: 7 pF (typical)
- Low Supply Current: 500 μA (Typical)
- Output Voltage Swing to Within 25 mV of GND or  $\rm V_{\rm S}$

## **Applications**

- · Switch Mode Power Supplies
- · Motor Controls
- · Pulse Transformer Driver
- · Class-D Switching Amplifiers
- · Line Drivers
- Driving MOSFET or IGBT Parallel Chip Modules
- · Local Power On/Off Switch
- Pulse Generators

#### **General Description**

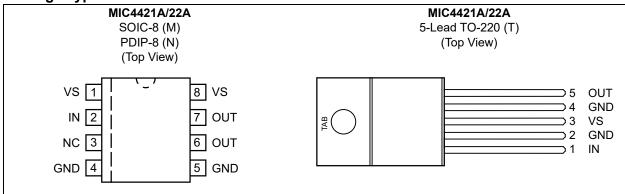
The MIC4421A and MIC4422A MOSFET drivers are rugged, efficient, and easy to use. The MIC4421A is an inverting driver, while the MIC4422A is a non-inverting driver

Both versions are capable of 9A (peak) output and can drive the largest MOSFETs with an improved safe operating margin. The MIC4421A/4422A accepts any logic input from 2.4V to  $V_{\rm S}$  without external speed-up capacitors or resistor networks. Proprietary circuits allow the input to swing negative by as much as 5V without damaging the part. Additional circuits protect against damage from electrostatic discharge.

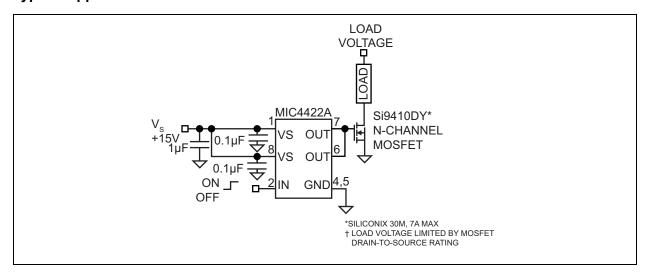
MIC4421A/22A drivers can replace three or more discrete components, reducing PCB area requirements, simplifying product design, and reducing assembly cost.

Modern Bipolar/CMOS/DMOS construction guarantees freedom from latch-up. The rail-to-rail swing capability of CMOS/DMOS ensures adequate gate voltage to the MOSFET during power up/down sequencing. Because these devices are fabricated on a self-aligned process, they have very low crossover current, run cool, use little power, and are easy to drive.

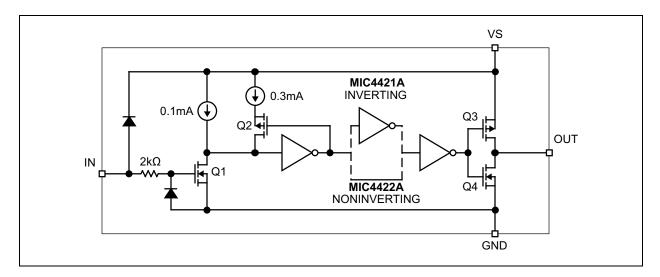
#### Package Types



## **Typical Application Circuit**



## **Functional Block Diagram**



#### 1.0 ELECTRICAL CHARACTERISTICS

## **Absolute Maximum Ratings †**

Supply Voltage (V <sub>S</sub> )	+20V
Control Input Voltage (V <sub>IN</sub> )	V <sub>S</sub> + 0.3V to GND – 5V
Control Input Current (V <sub>IN</sub> > V <sub>S</sub> )	50 mA
Power Dissipation (T <sub>A</sub> ≤ +25°C, Note 1)	
PDIP ( $\theta_{JA}$ )	1478 mW
SOIC (θ <sub>JA</sub> )	
TO-220 (θ <sub>JA</sub> )	1756W
ESD Rating (Note 2)	
Operating Ratings ††	
Supply Voltage (V <sub>S</sub> )	+4.5V to +18V

**† Notice:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

**†† Notice:** The device is not guaranteed to function outside its operating ratings.

Note 1: Minimum footprint.

2: Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5 k $\Omega$  in series with 100 pF.

## **ELECTRICAL CHARACTERISTICS**

**Electrical Characteristics:**  $T_A = +25^{\circ}C$  with  $4.5V \le V_S \le 18V$ , **bold** values valid for X Version:  $-55^{\circ}C \le T_A \le +125^{\circ}C$ , for Y Version:  $-40^{\circ}C \le T_A \le +85^{\circ}C$ , and for Z Version:  $0^{\circ}C \le T_A \le +70^{\circ}C$ , unless noted.

Parameter	Sym.	Min.	Тур.	Max.	Units	Conditions		
Power Supply								
Operating Input Voltage	V <sub>S</sub>	4.5	_	18	V	_		
High Output Quiescent		_	0.5	1.5		V <sub>IN</sub> = 3V (MIC4422A), V <sub>IN</sub> = 0		
Current		_	_	3	mA	(MIC4421A)		
Low Output Quiescent	I <sub>S</sub>	_	50	150		V <sub>IN</sub> = 0V (MIC4422A), V <sub>IN</sub> = 3V		
Current		_	_	200	μA	(MIC4421A)		
Input								
Logic 1 Input Voltage	V <sub>IH</sub>	3.0	2.1	_	V	See Figure 1-3		
Logic 0 Input Voltage	V <sub>IL</sub>	_	1.5	0.8	V	See Figure 1-3		
Input Voltage Range	V <sub>IN</sub>	<b>-</b> 5	_	V <sub>S</sub> + 0.3	V			
Input Current	I <sub>IN</sub>	-10	_	10	μA	$0V \le V_{IN} \le V_{S}$		
Output								
High Output Voltage	V <sub>OH</sub>	V <sub>S</sub> + .025	_	_	V	See Figure 1-1		
Low Output Voltage	V <sub>OL</sub>	_	_	0.025	V	See Figure 1-1		
Output Resistance, Output High	1	_	0.6	1.0		10 10 \		
		_	_	3.6	Ω	$I_{OUT} = 10 \text{ mA}, V_{S} = 18V$		
Output Resistance,	R <sub>O</sub>		0.8	1.7		_ 10 10 10 /		
Output Low		_	_	2.7	Ω	$I_{OUT}$ = 10 mA, $V_S$ = 18V		

Note 1: Guaranteed by design.

## **ELECTRICAL CHARACTERISTICS (CONTINUED)**

**Electrical Characteristics:**  $T_A = +25^{\circ}C$  with  $4.5V \le V_S \le 18V$ , **bold** values valid for X Version:  $-55^{\circ}C \le T_A \le +125^{\circ}C$ , for Y Version:  $-40^{\circ}C \le T_A \le +85^{\circ}C$ , and for Z Version:  $0^{\circ}C \le T_A \le +70^{\circ}C$ , unless noted.

Parameter	Sym.	Min.	Тур.	Max.	Units	Conditions	
Peak Output Current	I <sub>PK</sub>	_	9	_	Α	V <sub>S</sub> = 18V, See Figure 5-2	
Continuous Output Current	I <sub>DC</sub>		2		Α	_	
Latch-Up Protection Withstand Reverse Current	I <sub>R</sub>	>1500	_	_	mA	Duty Cycle ≤ 2%, t ≤ 300 μs, Note 1	
Switching Time, Note 1							
Diag Time	t <sub>r</sub>	1	20	75	20	Figure 1-1, C <sub>L</sub> = 10,000 pF	
Rise Time			_	120	ns		
E " T	t <sub>f</sub>	-	24	75	20	Figure 1.1. C. = 10.000 pE	
Fall Time			_	120	ns	Figure 1-1, C <sub>L</sub> = 10,000 pF	
	4	1	15	68	20	Figure 4.4	
Dalay Time	t <sub>D1</sub>	1	_	80	ns	Figure 1-1	
Delay Time	t <sub>D2</sub>	-	35	60		Firmer 4.4	
		1	_	80	ns	Figure 1-1	
Minimum Input Pulse Width	t <sub>PW</sub>		50		ns	See Figure 1-1 and Figure 1-2.	
Maximum Input Frequency	f <sub>MAX</sub>	_	1	_	MHz	See Figure 1-1 and Figure 1-2.	

Note 1: Guaranteed by design.

#### **Test Circuits**

Note that the output pulse width may increase with input pulse widths less than 50 ns.

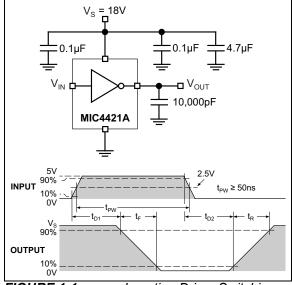


FIGURE 1-1: Inverting Driver Switching Time.

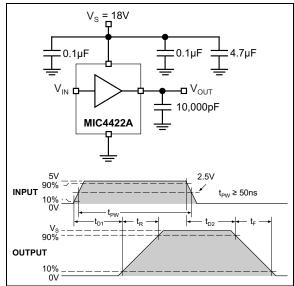


FIGURE 1-2: Non-Inverting Driver Switching Time.

## **Control Input Behavior**

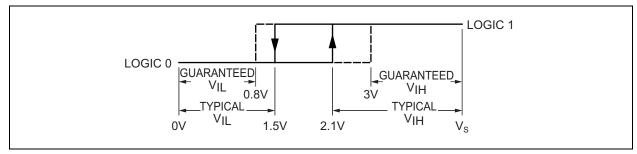


FIGURE 1-3: Input Hysteresis.

## **TEMPERATURE SPECIFICATIONS**

Parameters	Sym.	Min.	Тур.	Max.	Units	Conditions	
Temperature Ranges							
Maximum Junction Temperature	TJ	_	_	+150	°C	_	
Storage Temperature Range	T <sub>S</sub>	-65	_	+150	°C	_	
Lead Temperature	_	_	_	+300	°C	Soldering, 10 sec.	
		0	_	+70	°C	Z option	
Ambient Operating Temperature Range	$T_A$	-40	_	+85	°C	Y option	
		-55	_	+125		X option	
Package Thermal Resistances							
Thermal Resistance, PDIP 8-Ld	$\theta_{JA}$	_	84.6	_	°C/W	_	
Thermal Resistance, SOIC 8-Ld	$\theta_{JA}$	_	163	_	°C/W	_	
Thermal Resistance, TO-220 5-Ld	$\theta_{JA}$	_	71.2	_	°C/W	_	
Thermal Resistance, PDIP 8-Ld	$\theta_{\sf JC}$	_	41.2	_	°C/W	_	
Thermal Resistance, SOIC 8-Ld	$\theta_{\sf JC}$	_	38.8	_	°C/W	_	
Thermal Resistance, TO-220 5-Ld	$\theta_{\sf JC}$	_	6.5	_	°C/W	_	

Note 1: The maximum allowable power dissipation is a function of ambient temperature, the maximum allowable junction temperature and the thermal resistance from junction to air (i.e., T<sub>A</sub>, T<sub>J</sub>, θ<sub>JA</sub>). Exceeding the maximum allowable power dissipation will cause the device operating junction temperature to exceed the maximum +125°C rating. Sustained junction temperatures above +125°C can impact the device reliability.

#### 2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

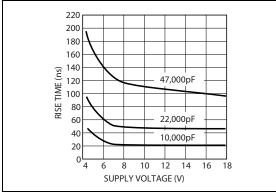
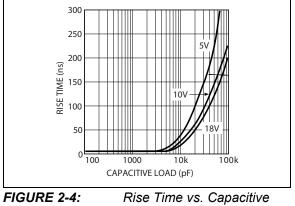


FIGURE 2-1: Rise Time vs. Supply Voltage.



Load.

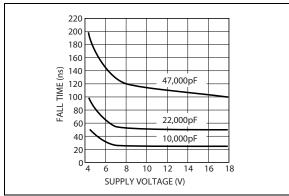


FIGURE 2-2: Fall Time vs. Supply Voltage.

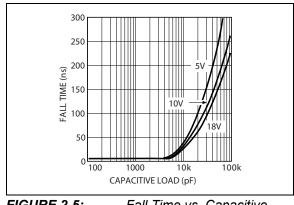


FIGURE 2-5: Fall Time vs. Capacitive Load.

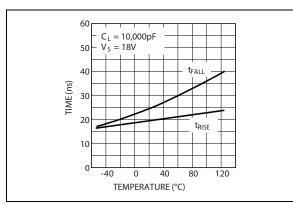


FIGURE 2-3: Rise and Fall Times vs. Temperature.

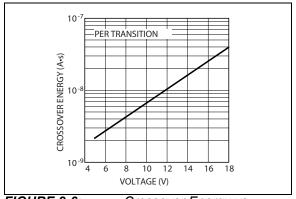
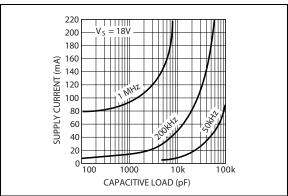
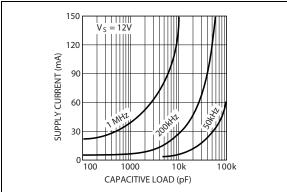


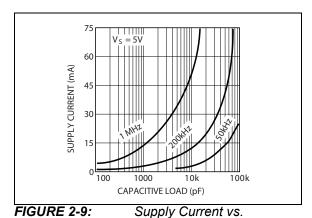
FIGURE 2-6: Crossover Energy vs. Supply Voltage.

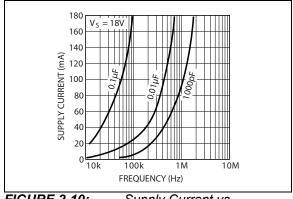


**FIGURE 2-7:** Supply Current vs. Capacitive Load ( $V_S = 18V$ ).

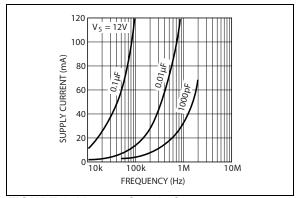


**FIGURE 2-8:** Supply Current vs. Capacitive Load ( $V_S = 12V$ ).

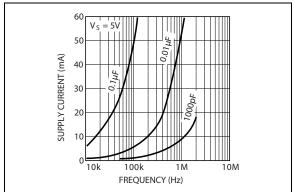




**FIGURE 2-10:** Supply Current vs. Frequency  $(V_S = 18V)$ .



**FIGURE 2-11:** Supply Current vs. Frequency  $(V_S = 12V)$ .



**FIGURE 2-12:** Supply Current vs. Frequency  $(V_S = 5V)$ .

Capacitive Load ( $V_S = 5V$ )

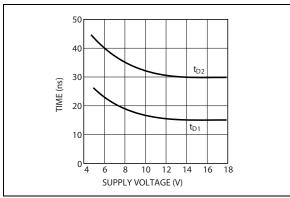


FIGURE 2-13: Supply Voltage.

Propagation Delay vs.

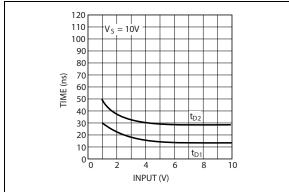


FIGURE 2-14: Amplitude.

Propagation Delay vs. Input

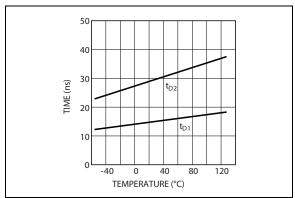


FIGURE 2-15: Temperature.

Propagation Delay vs.

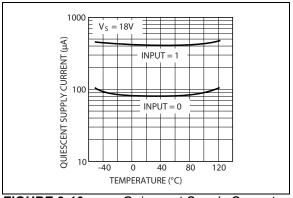


FIGURE 2-16: vs. Temperature.

Quiescent Supply Current



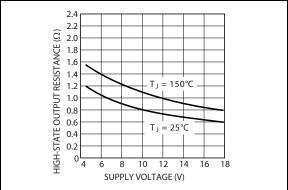


FIGURE 2-17: High-State Output Resistance vs. Supply Voltage.

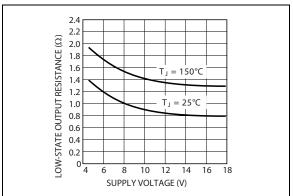


FIGURE 2-18: Low-State Output Resistance vs. Supply Voltage.

## 3.0 PIN DESCRIPTIONS

The descriptions of the pins are listed in Table 3-1.

TABLE 3-1: PIN FUNCTION TABLE

Pin Number PDIP/SOIC	Pin Number TO-220	Pin Name	Description	
2	1	IN	Control Input.	
4, 5	2, 4	GND	Ground: Duplicate pins must be externally connected.	
1, 8	3, TAB	VS	Supply Input: Duplicate pins must be externally connected.	
6, 7	5	OUT	Output: Duplicate pins must be externally connected.	
3	_	NC	Not connected.	

#### 4.0 FUNCTIONAL DESCRIPTION

Refer to the Functional Block Diagram.

The MIC4422A is a non-inverting driver. A logic high on the IN produces gate drive output. The MIC4421A is an inverting driver. A logic low on the IN produces gate drive output. The output is used to turn on an external N-channel MOSFET.

### 4.1 Supply

VS (supply) is rated for +4.5V to +18V. External capacitors are recommended to decouple noise.

## 4.2 Input

IN (control) is a TTL-compatible input. IN must be forced high or low by an external signal. A floating input will cause unpredictable operation.

A high input turns on Q1, which sinks the output of the 0.1 mA and the 0.3 mA current source, forcing the input of the first inverter low.

### 4.3 Hysteresis

The control threshold voltage, when IN is rising, is slightly higher than the control threshold voltage when CTL is falling.

When IN is low, Q2 is on, which applies the additional 0.3 mA current source to Q1. Forcing IN high turns on Q1 which must sink 0.4 mA from the two current sources. The higher current through Q1 causes a larger drain-to-source voltage drop across Q1. A slightly higher control voltage is required to pull the input of the first inverter down to its threshold.

Q2 turns off after the first inverter output goes high. This reduces the current through Q1 to 0.1 mA. The lower current reduces the drain-to-source voltage drop across Q1. A slightly lower control voltage will pull the input of the first inverter up to its threshold.

#### 4.4 Drivers

The second (optional) inverter permits the driver to be manufactured in inverting and non-inverting versions.

The last inverter functions as a driver for the output MOSFETs Q3 and Q4.

#### 4.5 Output

OUT is designed to drive a capacitive load.  $V_{OUT}$  (output voltage) is either approximately the supply voltage or approximately ground, depending on the logic state applied to IN.

If IN is high, and  $V_S$  (supply) drops to zero, the output will be floating (unpredictable).

### 5.0 APPLICATION INFORMATION

## 5.1 Supply Bypassing

Charging and discharging large capacitive loads quickly requires large currents. For example, charging a 10,000 pF load to 18V in 50 ns requires 3.6A.

The MIC4421A/22A has double bonding on the supply pins, the ground pins, and output pins. This reduces parasitic lead inductance. Low inductance enables large currents to be switched rapidly. It also reduces internal ringing that can cause voltage breakdown when the driver is operated at or near the maximum rated voltage.

Internal ringing can also cause output oscillation due to feedback. This feedback is added to the input signal because it is referenced to the same ground.

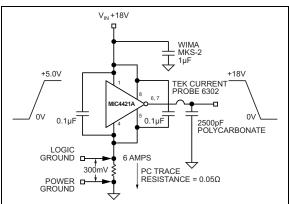


FIGURE 5-1: Switching Time Due to Negative Feedback.

TABLE 5-1: MIC4421A MAX. OPERATING FREQUENCY

INLQULINGI						
V <sub>S</sub>	Maximum Frequency					
18V	220 kHz					
15V	300 kHz					
10V	640 kHz					
5V	2 MHz					
<b>Conditions:</b> $\theta_{JA} = 150^{\circ}\text{C/W}$ , $T_A = 25^{\circ}\text{C}$ , $C_L = 10,000 \text{ pF}$ .						

To guarantee low supply impedance over a wide frequency range, a parallel capacitor combination is recommended for supply bypassing. Low inductance ceramic disk capacitor with short lead lengths (<0.5 inch) should be used. A 1  $\mu F$  low-ESR film capacitor in parallel with two 0.1  $\mu F$  low-ESR ceramic capacitors, (such as AVX RAM Guard  $^{(8)}$ ), provides adequate bypassing. Connect one ceramic capacitor directly between pins 1 and 4. Connect the second ceramic capacitor directly between pins 8 and 5.

#### 5.2 Grounding

The high current capability of the MIC4421A/22A demands careful PCB layout for best performance. Because the MIC4421A is an inverting driver, any ground lead impedance will appear as negative feedback which can degrade switching speed. Feedback is especially noticeable with slow rise time inputs. The MIC4421A input structure includes about 600 mV of hysteresis to ensure clean transitions and freedom from oscillation, but attention to layout is still recommended.

Figure 5-1 shows the feedback effect in detail. As the MIC4421A input begins to go positive, the output goes negative and several amperes of current flow in the ground lead. As little as  $0.05\Omega$  of PC trace resistance can produce hundreds of millivolts at the MIC4421A ground pins. If the driving logic is referenced to power ground, the effective logic input level is reduced and oscillation may result.

To ensure optimum performance, separate ground traces should be provided for the logic and power connections. Connecting the logic ground directly to the MIC4421A GND pins will ensure full logic drive to the input and ensure fast output switching. Both of the MIC4421A GND pins should, however, still be connected to power ground.

## 5.3 Input Stage

The input voltage level of the MIC4421A changes the quiescent supply current. The N-Channel MOSFET input stage transistor drives a 320  $\mu$ A current source load. With a logic "1" input, the quiescent supply current is typically 500  $\mu$ A. Logic "0" input level signals reduce quiescent current to 80  $\mu$ A typical.

The MIC4421A/22A input is designed to provide 600 mV of hysteresis. This provides clean transitions, reduces noise sensitivity, and minimizes output stage current spiking when changing states. Input voltage threshold level is approximately 1.5V, making the device TTL-compatible over the full temperature and operating supply voltage ranges. Input current is less than  $\pm 10~\mu A$ .

The MIC4421A can be directly driven by the TL494, SG1526/1527, SG1524, TSC170, MIC38C42, and similar switch mode power supply integrated circuits. By off-loading the power-driving duties to the MIC4421A/22A, the power supply controller can operate at lower dissipation. This can improve performance and reliability.

The input can be greater than the  $V_S$  supply, however, current will flow into the input lead. The input currents can be as high as 30 mA<sub>PP</sub> (6.4 mA<sub>RMS</sub>) with the input. No damage will occur to the MIC4421A/22A, however, and it will not latch.

The input appears as a 7 pF capacitance and does not change even if the input is driven from an AC source.

#### 5.4 Power Dissipation

CMOS circuits usually permit the user to ignore power dissipation. Logic families such as 4000 and 74C have outputs that can only supply a few milliamperes of current, and even shorting outputs to ground will not force enough current to destroy the device. The MIC4421A/22A on the other hand, can source or sink several amperes and drive large capacitive loads at high frequency. The package power dissipation limit can easily be exceeded. Therefore, some attention should be given to power dissipation when driving low impedance loads and/or operating at high frequency.

The supply current vs. frequency and supply current vs. capacitive load characteristic curves aid in determining power dissipation calculations. Table 5-1 lists the maximum safe operating frequency for several power supply voltages when driving a 10,000 pF load. More accurate power dissipation figures can be obtained by summing the three dissipation sources.

Given the power dissipation in the device and the thermal resistance of the package, junction operating temperature for any ambient is easy to calculate. For example, the thermal resistance of the 8-lead plastic DIP package, from the data sheet, is 84.6°C/W. In a 25°C ambient, then, using a maximum junction temperature of 150°C, this package will dissipate 1478 mW.

Accurate power dissipation numbers can be obtained by summing the three sources of power dissipation in the device:

- · Load Power Dissipation (PL)
- · Quiescent power dissipation (PQ)
- Transition power dissipation (PT)

Calculation of load power dissipation differs depending on whether the load is capacitive, resistive or inductive.

### 5.5 Resistive Load Power Dissipation

Dissipation caused by a resistive load can be calculated as:

#### **EQUATION 5-1:**

$$P_I = I^2 \times R_O \times D$$

Where:

I = The current drawn by the load.

 $R_O$  = The output resistance of the driver when the output is high, at the power supply voltage used.

D = Fraction of time the load is conducting (duty cycle).

# 5.6 Capacitive Load Power Dissipation

Dissipation caused by a capacitive load is simply the energy placed in, or removed from, the load capacitance by the driver. The energy stored in a capacitor is described by the equation:

#### **EQUATION 5-2:**

$$E = 1/2 \times C \times V^2$$

As this energy is lost in the driver each time the load is charged or discharged, for power dissipation calculations, the 1/2 is removed. This equation also shows that it is good practice not to place more voltage in the capacitor than is necessary, because dissipation increases as the square of the voltage applied to the capacitor. For a driver with a capacitive load:

#### **EQUATION 5-3:**

$$P_L = f \times C \times {V_S}^2$$

Where:

f = Operating frequency.

C = Load capacitance.

 $V_S$  = Driver supply voltage.

## 5.7 Inductive Load Power Dissipation

For inductive loads, the situation is more complicated. For the part of the cycle in which the driver is actively forcing current into the inductor, the situation is the same as it is in the resistive case:

#### **EQUATION 5-4:**

$$P_{L1} = I^2 \times R_O \times D$$

However, in this instance, the  $R_{\rm O}$  required may be either the on-resistance of the driver when its output is in the high state, or its on-resistance when the driver is in the low state. This depends on how the inductor is connected, and this is still only half the story. For the part of the cycle when the inductor is forcing current through the driver, dissipation is best described as:

#### **EQUATION 5-5:**

$$P_{L2} = I \times V_D \times (1 - D)$$

Where:

 $V_D$  = The forward voltage drop of the clamp diode in the driver (generally around 0.7V).

The two parts of the load dissipation must be summed in to produce  $\mathsf{P}_\mathsf{L}$ :

#### **EQUATION 5-6:**

$$P_L = P_{L1} + P_{L2}$$

#### 5.8 Quiescent Power Dissipation

Quiescent power dissipation ( $P_Q$ , as described in the input section) depends on whether the input is high or low. A low input will result in a maximum current drain (per driver) of  $\leq$ 0.2 mA; logic high will result in a current drain of  $\leq$ 3.0 mA.

Quiescent power can therefore be found from:

#### **EQUATION 5-7:**

$$P_O = V_S \times (D \times I_H + (1 - D) \times I_L)$$

Where:

I<sub>H</sub> = Quiescent current with input high.

I<sub>L</sub> = Quiescent current with input low.

D = Fraction of time input is high (duty cycle).

 $V_S$  = Power supply voltage.

## 5.9 Transition Power Dissipation

Transition power is dissipated in the driver each time its output changes state, because during the transition, for a very brief interval, both the N- and P-Channel MOSFETs in the output totem-pole are ON simultaneously, and a current is conducted through them from  $V_S$  to ground. The transition power dissipation is approximately:

#### **EQUATION 5-8:**

$$P_T = 2 \times f \times V_S \times (A \bullet s)$$

A•s is a time-current factor derived from the typical operating characteristic curve Figure 2-6.

Total power (P<sub>D</sub>) then, as previously described, is just:

#### **EQUATION 5-9:**

$$P_D = P_L + P_Q + P_T$$

## 5.10 Definitions

C<sub>L</sub> = Load Capacitance in Farads.

D = Duty Cycle expressed as the fraction of time the input to the driver is high.

f = Operating Frequency of the driver in Hertz.

I<sub>H</sub> = Power supply current drawn by a driver when both inputs are high and neither output is loaded.

 $I_L$  = Power supply current drawn by a driver when both inputs are low and neither output is loaded.

I<sub>D</sub> = Output current from a driver in Amps.

P<sub>D</sub> = Total power dissipated in a driver in Watts.

 $\mathbf{P}_{L}$  = Power dissipated in the driver due to the driver's load in Watts.

P<sub>O</sub> = Power dissipated in a quiescent driver in Watts.

P<sub>T</sub> = Power dissipated in a driver when the output changes states ("shoot-through current") in Watts.

R<sub>O</sub> = Output resistance of a driver in Ohms.

V<sub>S</sub> = Power supply voltage to the IC in Volts.

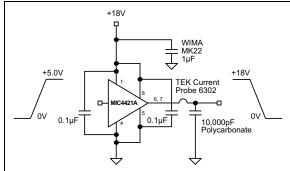


FIGURE 5-2: Circuit.

Peak Output Current Test

### 6.0 PACKAGING INFORMATION

## 6.1 Package Marking Information

8-Lead SOIC\*

XXX XXXXXX WNNN

8-Lead PDIP\*

XXX XXXXXX WNNN XXX

5-Lead TO-220\*

XXX XXXXXXX WNNNP Example

MIC 4421AXM 8469

Example

MIC 4422AYN 9771 USA

Example

MIC 4421AZT 4287P

**Legend:** XX...X Product code or customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

e3 Pb-free JEDEC® designator for Matte Tin (Sn)

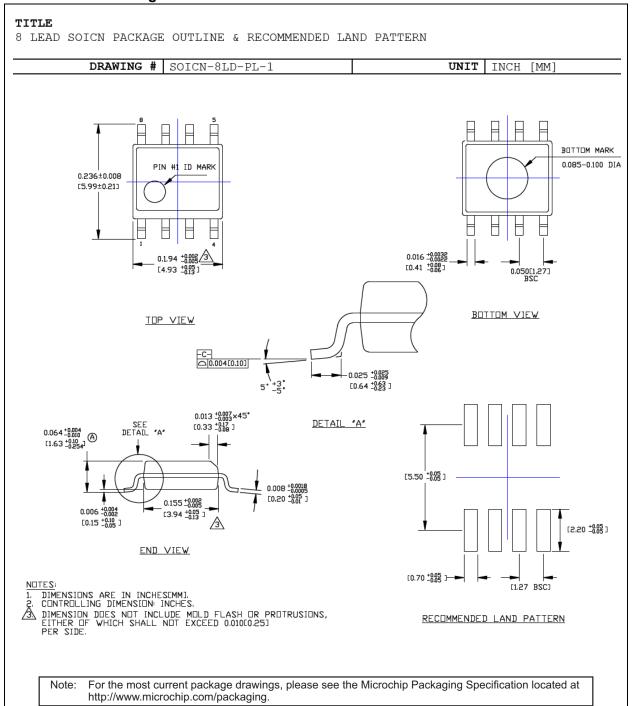
This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

•, ▲, ▼ Pin one index is identified by a dot, delta up, or delta down (triangle mark).

ote: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information. Package may or may not include the corporate logo.

Underbar (\_) and/or Overbar (\_) symbol may not be to scale.

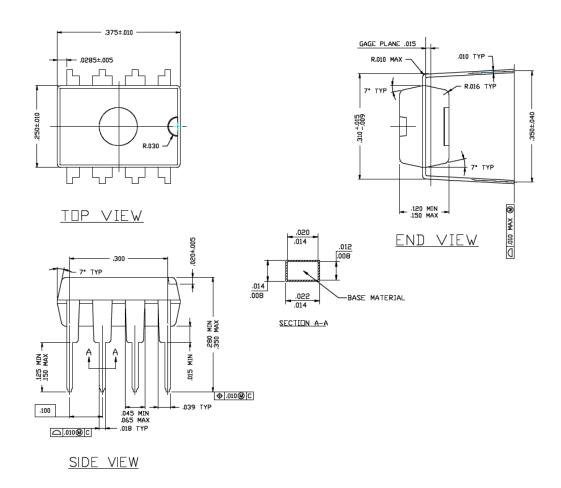
## 8-Lead SOICN Package Outline & Recommended Land Pattern



## 8-Lead PDIP Package Outline and Recommended Land Pattern

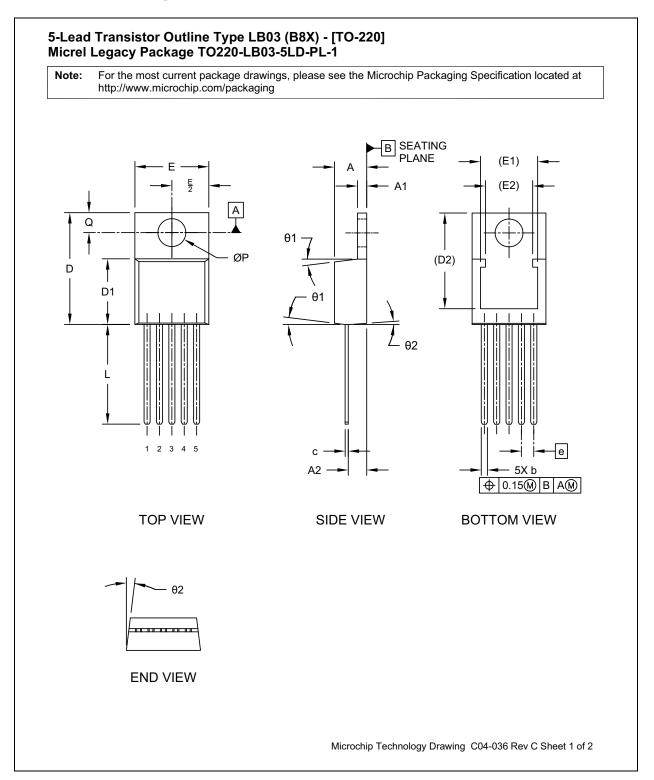
## TITLE 8 LEAD PDIP PACKAGE OUTLINE & RECOMMENDED LAND PATTERN

DRAWING #	PDIP-8LD-PL-1	UNIT	INCH
Lead Frame	Copper	Lead Finish	Matte Tin



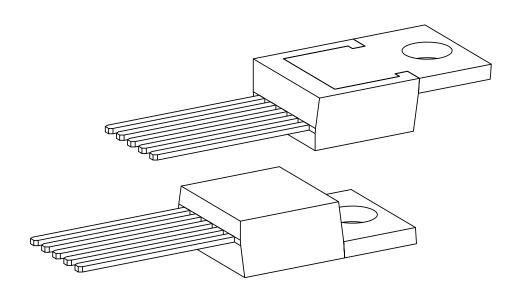
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging.

## 5-Lead TO-220 Package Outline and Recommended Land Pattern



## 5-Lead Transistor Outline Type LB03 (B8X) - [TO-220] Micrel Legacy Package TO220-LB03-5LD-PL-1

For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	INCHES			
Dimension	Min	Nom	Max	
Number of Leads	N		5	
Pitch	е		.067 BSC	
Overall Height	Α	.160	.175	.190
Tab Height	A1	.045	.050	.055
Seating Plane to Lead	A2	.080	.098	.115
Lead Width	b	.025	.033	.040
Lead Thickness	С	.012 .016 .020		
Lead Length	L	.500 .540 .580		
Total Body Length Including Tab	D	.542 .580 .619		
Molded Body Length	D1	.348 .354 .360		
Total Width	Е	.380 .400 .420		
Pad Width	E1	0.256 REF		
Pad Length	D2	0.486 REF		
Hole Diameter	ØP	.146 .151 .156		
Hole Center to Tab Edge	Q	.103 .108 .113		
Molded Body Draft Angle	θ2	3 7 10		
Molded Body Draft Angle	θ2	1 4 7		

#### Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
   Dimensioning and tolerancing per ASME Y14.5M
   BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-036 Rev C Sheet 2 of 2

NOTES:

## APPENDIX A: REVISION HISTORY

## Revision A (March 2021)

- Converted Micrel document MIC4421A/22A to Microchip data sheet template DS20006511A.
- Minor grammatical text changes throughout.

NOTES:

### PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, contact your local Microchip representative or sales office.

Device <u>X</u> XX-<u>XX</u> Part No. Junction Package Media Type Temp. Range MIC4421A: Inverting, 9A-Peak Low-Side MOSFET Driver Device: Non-Inverting, 9A-Peak Low-Side MIC4422A: MOSFET Driver Junction -55°C to +125°C, RoHS-Compliant Temperature = -40°C to +85°C, RoHS-Compliant Range: 7 = 0°C to +70°C, RoHS-Compliant 8-Lead PDIP Package: 8-Lead SOIC Μ 5-Lead TO-220 95/Tube (SOIC only) <black>= Media Type: 50/Tube (PDIP, TO-220 only) <black>= 2,500/Reel (SOIC only)

#### Examples:

a) MIC4421A: Inverting, 9A-Peak Low-Side MOSFET Driver, -55°C to +125°C Temperature Range MIC4421AXM-TR 8-Lead SOIC 2,500/Reel b) MIC4421A: Inverting, 9A-Peak Low-Side MOSFET Driver, -40°C to +85°C Temperature Range MIC4421AYN 8-Lead PDIP 50/Tube MIC4421AYM 8-Lead SOIC 95/Tube MIC4421AYM-TR 8-Lead SOIC 2,500/Reel c) MIC4421A: Inverting, 9A-Peak Low-Side MOSFET Driver, 0°C to +70°C Temperature Range MIC4421AZN 8-Lead PDIP 50/Tube MIC4421AZM 8-Lead SOIC 95/Tube MIC4421AZM-TR 8-Lead SOIC 2,500/Reel MIC4421AZT 5-Lead TO-220 50/Tube d) MIC4422A: Non-Inverting, 9A-Peak Low-Side MOSFET Driver, -40°C to +85°C Temperature Range 8-Lead PDIP 50/Tube MIC4422AYN MIC4422AYM 8-Lead SOIC 95/Tube MIC4422AYM-TR 8-Lead SOIC 2,500/Reel e) MIC4422A: Non-Inverting, 9A-Peak Low-Side MOSFET Driver, 0°C to +70°C Temperature Range MIC4422AZN 8-Lead PDIP 50/Tube MIC4422AYM 8-Lead SOIC 95/Tube 2,500/Reel MIC4422AYM-TR 8-Lead SOIC MIC4422AYT 50/Tube 5-Lead TO-220

Note 1: Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option.

NOTES:

#### Note the following details of the code protection feature on Microchip devices:

- Microchip products meet the specifications contained in their particular Microchip Data Sheet.
- · Microchip believes that its family of products is secure when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods being used in attempts to breach the code protection features of the Microchip devices. We believe that these methods require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Attempts to breach these code protection features, most likely, cannot be accomplished without violating Microchip's intellectual property rights.
- Microchip is willing to work with any customer who is concerned about the integrity of its code.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of its code. Code protection does not
  mean that we are guaranteeing the product is "unbreakable." Code protection is constantly evolving. We at Microchip are
  committed to continuously improving the code protection features of our products. Attempts to break Microchip's code protection
  feature may be a violation of the Digital Millennium Copyright Act. If such acts allow unauthorized access to your software or
  other copyrighted work, you may have a right to sue for relief under that Act.

Information contained in this publication is provided for the sole purpose of designing with and using Microchip products. Information regarding device applications and the like is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications.

THIS INFORMATION IS PROVIDED BY MICROCHIP "AS IS". MICROCHIP MAKES NO REPRESENTATIONS OR WARRANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION INCLUDING BUT NOT LIMITED TO ANY IMPLIED WARRANTIES OF NON-INFRINGEMENT, MERCHANTABILITY, AND FITNESS FOR A PARTICULAR PURPOSE OR WARRANTIES RELATED TO ITS CONDITION, QUALITY, OR PERFORMANCE.

IN NO EVENT WILL MICROCHIP BE LIABLE FOR ANY INDI-RECT, SPECIAL, PUNITIVE, INCIDENTAL OR CONSEQUEN-TIAL LOSS, DAMAGE, COST OR EXPENSE OF ANY KIND WHATSOEVER RELATED TO THE INFORMATION OR ITS USE, HOWEVER CAUSED, EVEN IF MICROCHIP HAS BEEN ADVISED OF THE POSSIBILITY OR THE DAMAGES ARE FORESEEABLE. TO THE FULLEST EXTENT ALLOWED BY LAW, MICROCHIP'S TOTAL LIABILITY ON ALL CLAIMS IN ANY WAY RELATED TO THE INFORMATION OR ITS USE WILL NOT EXCEED THE AMOUNT OF FEES, IF ANY, THAT YOU HAVE PAID DIRECTLY TO MICROCHIP FOR THE INFORMATION. Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights unless otherwise stated.

For information regarding Microchip's Quality Management Systems, please visit www.microchip.com/quality.

#### **Trademarks**

The Microchip name and logo, the Microchip logo, Adaptec, AnyRate, AVR, AVR logo, AVR Freaks, BesTime, BitCloud, chipKIT, chipKIT logo, CryptoMemory, CryptoRF, dsPIC, FlashFlex, flexPWR, HELDO, IGLOO, JukeBlox, KeeLoq, Kleer, LANCheck, LinkMD, maXStylus, maXTouch, MediaLB, megaAVR, Microsemi, Microsemi logo, MOST, MOST logo, MPLAB, OptoLyzer, PackeTime, PIC, picoPower, PICSTART, PIC32 logo, PolarFire, Prochip Designer, QTouch, SAM-BA, SenGenuity, SpyNIC, SST, SST Logo, SuperFlash, Symmetricom, SyncServer, Tachyon, TimeSource, tinyAVR, UNI/O, Vectron, and XMEGA are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AgileSwitch, APT, ClockWorks, The Embedded Control Solutions Company, EtherSynch, FlashTec, Hyper Speed Control, HyperLight Load, IntelliMOS, Libero, motorBench, mTouch, Powermite 3, Precision Edge, ProASIC, ProASIC Plus, ProASIC Plus logo, Quiet-Wire, SmartFusion, SyncWorld, Temux, TimeCesium, TimeHub, TimePictra, TimeProvider, WinPath, and ZL are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Adjacent Key Suppression, AKS, Analog-for-the-Digital Age, Any Capacitor, AnyIn, AnyOut, Augmented Switching, BlueSky, BodyCom, CodeGuard, CryptoAuthentication, CryptoAutomotive, CryptoCompanion, CryptoController, dsPICDEM, dsPICDEM.net, Dynamic Average Matching, DAM, ECAN, Espresso T1S, EtherGREEN, IdealBridge, In-Circuit Serial Programming, ICSP, INICnet, Intelligent Paralleling, Inter-Chip Connectivity, JitterBlocker, maxCrypto, maxView, memBrain, Mindi, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, PowerSmart, PureSilicon, QMatrix, REAL ICE, Ripple Blocker, RTAX, RTG4, SAM-ICE, Serial Quad I/O, simpleMAP, SimpliPHY, SmartBuffer, SMART-I.S., storClad, SQI, SuperSwitcher, SuperSwitcher II, Switchtec, SynchroPHY, Total Endurance, TSHARC, USBCheck, VariSense, VectorBlox, VeriPHY, ViewSpan, WiperLock, XpressConnect, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

The Adaptec logo, Frequency on Demand, Silicon Storage Technology, and Symmcom are registered trademarks of Microchip Technology Inc. in other countries.

GestIC is a registered trademark of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

© 2021, Microchip Technology Incorporated, All Rights Reserved.

ISBN: 978-1-5224-7912-3



## Worldwide Sales and Service

#### **AMERICAS**

Corporate Office 2355 West Chandler Blvd. Chandler, AZ 85224-6199

Tel: 480-792-7200 Fax: 480-792-7277 Technical Support:

http://www.microchip.com/ support

Web Address:

www.microchip.com

Atlanta Duluth, GA

Tel: 678-957-9614 Fax: 678-957-1455

**Austin, TX** Tel: 512-257-3370

**Boston** 

Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL

Tel: 630-285-0071 Fax: 630-285-0075

Dallas

Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

**Detroit** Novi, MI

Tel: 248-848-4000

Houston, TX Tel: 281-894-5983

Tel: 281-894-598 Indianapolis Noblesville, IN

Tel: 317-773-8323 Fax: 317-773-5453 Tel: 317-536-2380

Los Angeles

Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608 Tel: 951-273-7800

Raleigh, NC Tel: 919-844-7510

New York, NY Tel: 631-435-6000

**San Jose, CA** Tel: 408-735-9110 Tel: 408-436-4270

Canada - Toronto Tel: 905-695-1980 Fax: 905-695-2078

#### ASIA/PACIFIC

Australia - Sydney Tel: 61-2-9868-6733

**China - Beijing** Tel: 86-10-8569-7000

China - Chengdu Tel: 86-28-8665-5511

China - Chongqing Tel: 86-23-8980-9588

**China - Dongguan** Tel: 86-769-8702-9880

**China - Guangzhou** Tel: 86-20-8755-8029

China - Hangzhou Tel: 86-571-8792-8115

China - Hong Kong SAR Tel: 852-2943-5100

China - Nanjing Tel: 86-25-8473-2460

China - Qingdao Tel: 86-532-8502-7355

**China - Shanghai** Tel: 86-21-3326-8000

**China - Shenyang** Tel: 86-24-2334-2829

**China - Shenzhen** Tel: 86-755-8864-2200

China - Suzhou Tel: 86-186-6233-1526

China - Wuhan

Tel: 86-27-5980-5300 China - Xian

Tel: 86-29-8833-7252

**China - Xiamen** Tel: 86-592-2388138

**China - Zhuhai** Tel: 86-756-3210040

#### ASIA/PACIFIC

India - Bangalore Tel: 91-80-3090-4444

India - New Delhi Tel: 91-11-4160-8631

India - Pune Tel: 91-20-4121-0141

Japan - Osaka

Tel: 81-6-6152-7160 Japan - Tokyo

Tel: 81-3-6880- 3770

Korea - Daegu Tel: 82-53-744-4301

Korea - Seoul Tel: 82-2-554-7200

Malaysia - Kuala Lumpur Tel: 60-3-7651-7906

Malaysia - Penang Tel: 60-4-227-8870

Philippines - Manila Tel: 63-2-634-9065

**Singapore** Tel: 65-6334-8870

**Taiwan - Hsin Chu** Tel: 886-3-577-8366

Taiwan - Kaohsiung Tel: 886-7-213-7830

**Taiwan - Taipei** Tel: 886-2-2508-8600

Thailand - Bangkok Tel: 66-2-694-1351

Vietnam - Ho Chi Minh Tel: 84-28-5448-2100

#### **EUROPE**

**Austria - Wels** Tel: 43-7242-2244-39

Fax: 43-7242-2244-393

**Denmark - Copenhagen** Tel: 45-4485-5910 Fax: 45-4485-2829

Finland - Espoo Tel: 358-9-4520-820

France - Paris
Tel: 33-1-69-53-63-20
Fax: 33-1-69-30-90-79

Germany - Garching Tel: 49-8931-9700

**Germany - Haan** Tel: 49-2129-3766400

Germany - Heilbronn Tel: 49-7131-72400

Germany - Karlsruhe Tel: 49-721-625370

**Germany - Munich** Tel: 49-89-627-144-0 Fax: 49-89-627-144-44

Germany - Rosenheim Tel: 49-8031-354-560

**Israel - Ra'anana** Tel: 972-9-744-7705

**Italy - Milan** Tel: 39-0331-742611

Fax: 39-0331-466781

**Italy - Padova** Tel: 39-049-7625286

**Netherlands - Drunen** Tel: 31-416-690399 Fax: 31-416-690340

Norway - Trondheim Tel: 47-7288-4388

**Poland - Warsaw** Tel: 48-22-3325737

**Romania - Bucharest** Tel: 40-21-407-87-50

**Spain - Madrid** Tel: 34-91-708-08-90 Fax: 34-91-708-08-91

**Sweden - Gothenberg** Tel: 46-31-704-60-40

Sweden - Stockholm Tel: 46-8-5090-4654

**UK - Wokingham** Tel: 44-118-921-5800 Fax: 44-118-921-5820

## **X-ON Electronics**

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Gate Drivers category:

Click to view products by Microchip manufacturer:

Other Similar products are found below:

89076GBEST 00053P0231 56956 57.404.7355.5 LT4936 57.904.0755.0 5882900001 00600P0005 00-9050-LRPP 00-9090-RDPP
5951900000 01-1003W-10/32-15 0131700000 00-2240 LTP70N06 LVP640 5J0-1000LG-SIL LY1D-2-5S-AC120 LY2-US-AC240 LY3UA-DC24 00576P0020 00600P0010 LZN4-UA-DC12 LZNQ2M-US-DC5 LZNQ2-US-DC12 LZP40N10 00-8196-RDPP 00-8274-RDPP
00-8275-RDNP 00-8722-RDPP 00-8728-WHPP 00-8869-RDPP 00-9051-RDPP 00-9091-LRPP 00-9291-RDPP 0207100000 0207400000
01312 0134220000 60713816 M15730061 61161-90 61278-0020 6131-204-23149P 6131-205-17149P 6131-209-15149P 6131-218-17149P
6131-220-21149P 6131-260-2358P 6131-265-11149P