

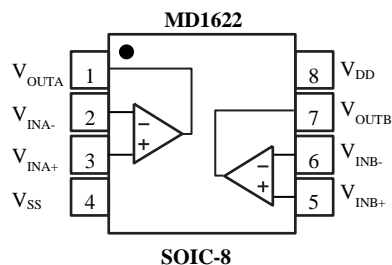
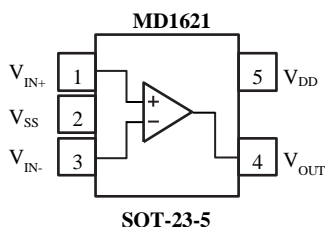
## FEATURES

- Low Cost
- Rail-to-Rail Input and Output  
0.8mV Typical  $V_{OS}$
- Unity Gain Stable
- Gain-Bandwidth Product: 3MHz
- Very Low Input Bias Current: 10pA
- Supply Voltage Range: 2.1V to 5.5V
- Input Voltage Range:  
-0.1V to +5.6V with  $(V_{DD} - V_{SS}) = 5.5V$
- Low Supply Current: 200 $\mu$ A/Amplifier
- Small Packaging  
MD1621 Available in SOT-23-5  
MD1622 Available in SOIC-8

## APPLICATIONS

- ASIC Input or Output Amplifier
- Sensor Interface
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Mobile Communication
- Audio Output
- Portable Systems
- Smoke Detectors
- Notebook PC
- PCMCIA Cards
- Battery-Powered Equipment
- DSP Interface

## PIN CONFIGURATIONS (TOP VIEW)



## PRODUCT DESCRIPTION

The MD1621 (single), MD1622 (dual) are low cost, rail-to-rail input and output voltage feedback amplifiers. They have a wide input common mode voltage range and output voltage swing, and take the minimum operating supply voltage down to 2.1V. The maximum recommended supply voltage is 5.5V. It is specified over the extended -40 $^{\circ}$ C to +85 $^{\circ}$ C temperature range.

The MD1621/2 provide 3MHz bandwidth at a low current consumption of 200 $\mu$ A per amplifier. Very low input bias currents of 10pA enable MD1621/2 to be used for integrators, photodiode amplifiers, and piezoelectric sensors. Rail-to-rail input and output are useful to designers for buffering ASIC in single-supply systems.

Applications for this series of amplifiers include safety monitoring, portable equipment, battery and power supply control, and signal conditioning and interfacing for transducers in very low power systems.

The MD1621 is available in the Green SOT-23-5 Package. The MD1622 comes in the Green SOIC-8 package.

**ORDER INFORMATION**

MODEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
MD1621	--	SOT23-5	Tape and Reel, 2500	--
MD1622	--	SOIC-8	Tape and Reel, 2500	--

**ABSOLUTE MAXIMUM RATINGS**

Supply Voltage, $V_{DD}$ to $V_{SS}$ .....	6V	Package Thermal Resistance @ $T_A = +25^\circ\text{C}$	
Common Mode Input Voltage.....	$V_{SS} - 0.3\text{V}$ to $V_{DD} + 0.3\text{V}$	SOIC-8, $\theta_{JA}$ .....	125 $^\circ\text{C}/\text{W}$
Storage Temperature Range.....	-65 $^\circ\text{C}$ to +150 $^\circ\text{C}$	Lead Temperature (Soldering 10sec).....	260 $^\circ\text{C}$
Junction Temperature.....	150 $^\circ\text{C}$		
Operating Temperature Range.....	-40 $^\circ\text{C}$ to +85 $^\circ\text{C}$		

**NOTE:**

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**ESD, Electrostatic Discharge Protection**

Symbol	Parameter	Condition	Minimum Level	Unit
HBM	Human Body Model ESD		4	kV
MM	Machine Model ESD		300	V

**ELECTRICAL CHARACTERISTICS**

The • denotes the specifications which apply over the full operating temperature range, otherwise specifications are At  $T_A=25^{\circ}\text{C}$ ,  $V_{DD} = +5\text{V}$ ,  $V_{SS} = \text{GND}$ ,  $R_L = 100\text{k}\Omega$  connected to  $V_{DD}/2$ , and  $V_{OUT} = V_{DD}/2$ .

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>INPUT CHARACTERISTICS</b>						
Input Offset Voltage	$V_{OS}$	$V_{CM} = V_{DD}/2$ $V_{CM} = V_{DD}/2$		0.8	5 6.6	mV
Input Bias Current	$I_B$			10		pA
Input Offset Current	$I_{OS}$			10		pA
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta T$			2		$\mu\text{V}/^{\circ}\text{C}$
Input Common Mode Voltage Range	$V_{CM}$	$V_{DD} = 5.5\text{V}$		-0.1-5.6		V
Common Mode Rejection Ratio	CMRR	$V_{DD} = 5.5\text{V}$ , $V_{CM} = -0.1\text{V to } 4\text{V}$		76		dB
Open-Loop Voltage Gain	$A_{OL}$	$R_L = 100\text{k}\Omega$ , $V_{OUT} = 2.5\text{V}$		100		dB
		$R_L = 100\text{k}\Omega$ , $V_{OUT} = +0.2\text{V to } +4.8\text{V}$		98		dB
<b>OUTPUT CHARACTERISTICS</b>						
Output Voltage Swing from Rail	$V_{OH}$	$R_L = 100\text{k}\Omega$		4.980	4.995	V
	$V_{OL}$	$R_L = 100\text{k}\Omega$		25	5	mV
	$V_{OH}$	$R_L = 10\text{k}\Omega$		4.970	4.994	V
	$V_{OL}$	$R_L = 10\text{k}\Omega$		35	6	mV
Output Current	$I_{SOURCE}$	$R_L = 10\Omega$ to $V_{DD}/2$		40		mA
	$I_{SINK}$			40		
<b>POWER SUPPLY</b>						
Operating Voltage Range				2.1		V
				2.5	5.5	V
Power Supply Rejection Ratio	PSRR	$V_{DD} = +2.5\text{V to } +5.5\text{V}$ , $V_{CM} = +0.5\text{V}$		85		dB
Quiescent Current/Amplifier	$I_Q$			200		$\mu\text{A}$
				120	270	
<b>DYNAMIC PERFORMANCE (<math>C_L = 100\text{pF}</math>)</b>						
Gain-Bandwidth Product	GBP			3		MHz
Phase Margin	PM	$R_L = 100\text{k}\Omega$ , $C_L = 100\text{pF}$		45		$^{\circ}$
Harmonic Distortion	HD2	$f = 10\text{kHz}$ , $G = +1$ , $R_L=100\text{k}$ , $V_{OUT}=2V_{PP}$		>80		dBc
	HD3	$f = 10\text{kHz}$ , $G = +1$ , $R_L=100\text{k}$ , $V_{OUT}=2V_{PP}$		>80		
Slew Rate	SR	$G = +1$ , 2V Output Step		1.74		V/ $\mu\text{s}$
Settling Time to 0.1%	$t_s$	$G = +1$ , 2V Output Step		6		$\mu\text{s}$
Overload Recovery Time		$V_{IN} \cdot G = V_{DD}$		2.5		$\mu\text{s}$
<b>NOISE PERFORMANCE</b>						
Voltage Noise Density	$e_n$	$f = 1\text{kHz}$		20		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 10\text{kHz}$		12		$\text{nV}/\sqrt{\text{Hz}}$

TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = +5\text{V}$ ,  $V_{SS} = \text{GND}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_{DD}/2$ , unless otherwise specified.

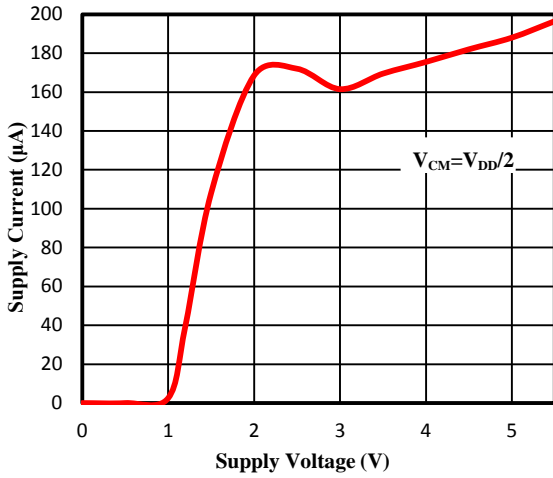


Figure 1. Supply Current vs. Supply Voltage

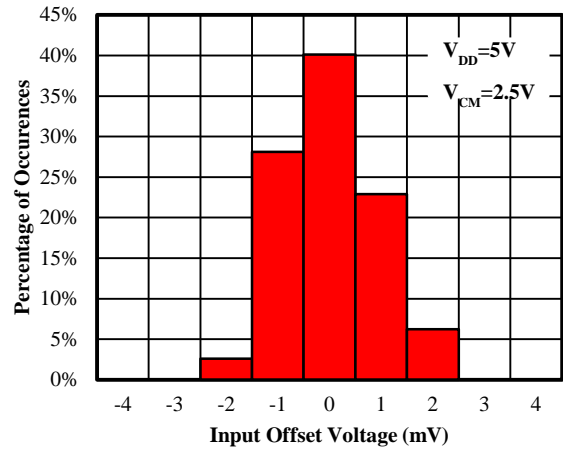


Figure 2. Input Offset Voltage Distribution

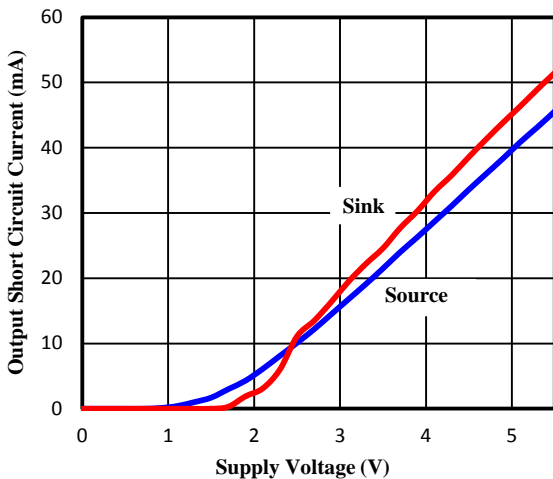


Figure 3. Output Short Circuit Current vs. Supply Voltage

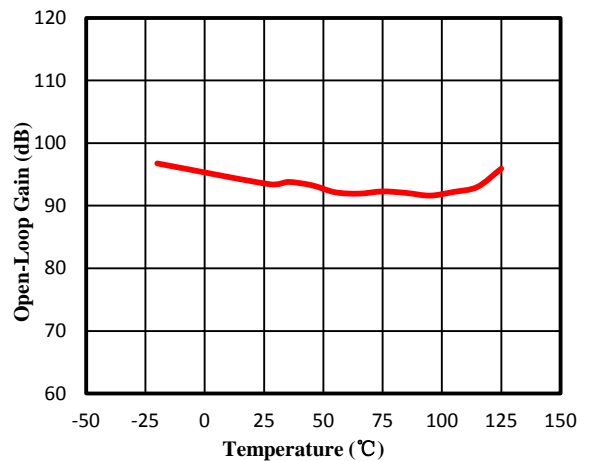


Figure 4. Open-Loop Gain vs. Temperature

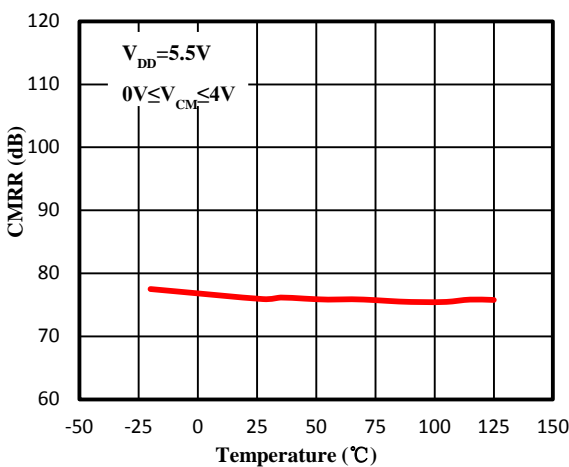


Figure 5. CMRR vs. Temperature

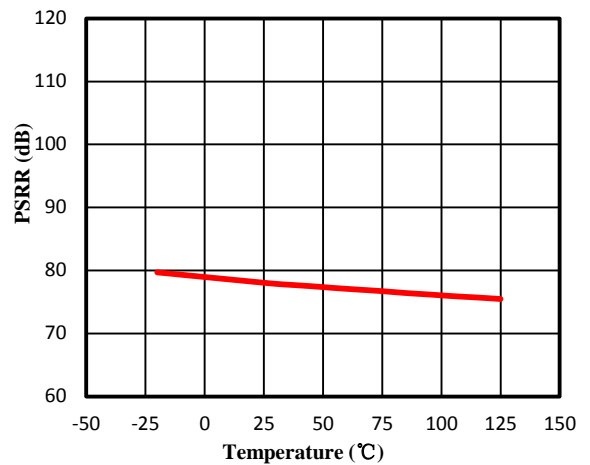


Figure 6. PSRR vs. Temperature

TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = +5\text{V}$ ,  $V_{SS} = \text{GND}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_{DD}/2$ , unless otherwise specified.

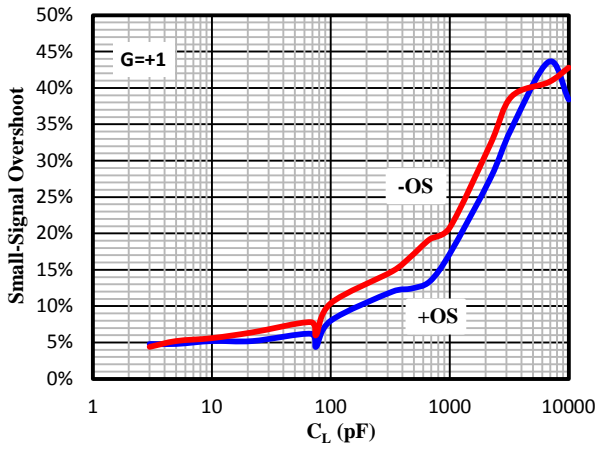


Figure 7. Small-Signal Overshoot vs. Load Capacitance

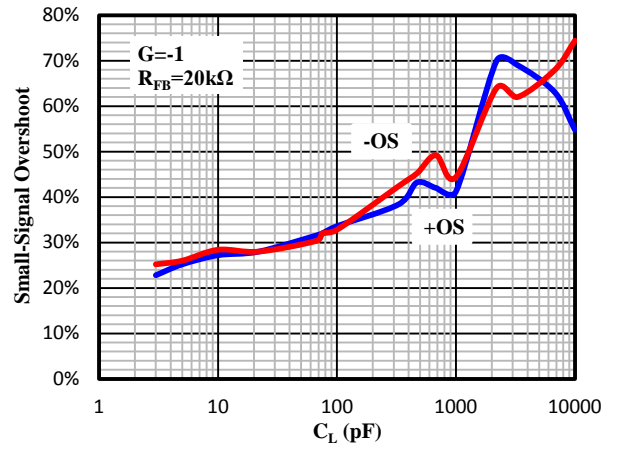


Figure 8. Small-Signal Overshoot vs. Load Capacitance

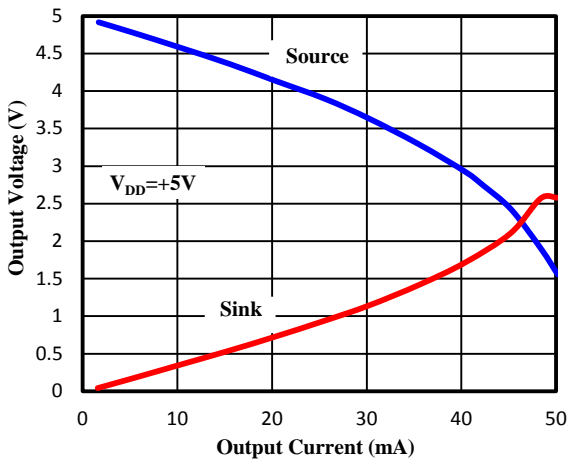


Figure 9. Output Voltage vs. Output Current

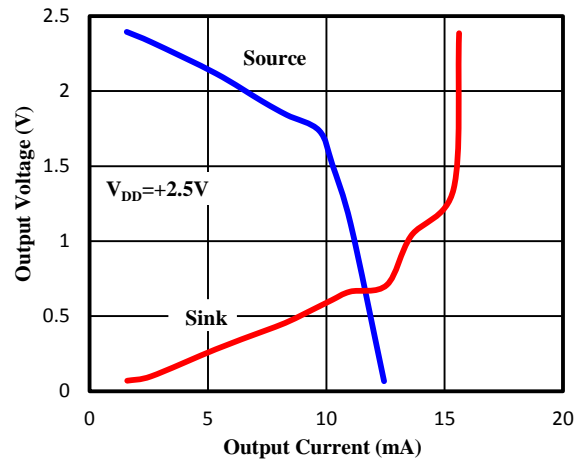


Figure 10. Output Voltage vs. Output Current

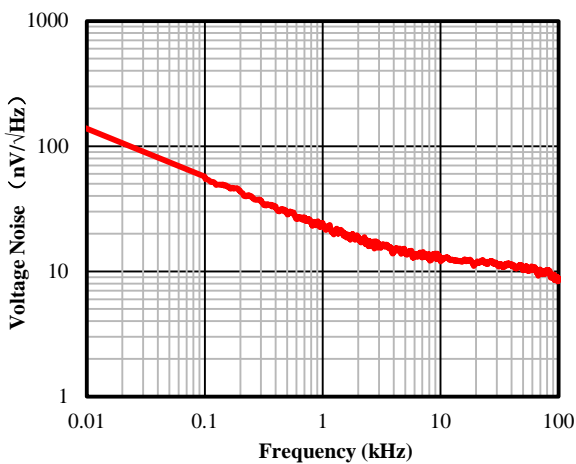


Figure 11. Input Voltage Noise Spectral Density vs. Frequency

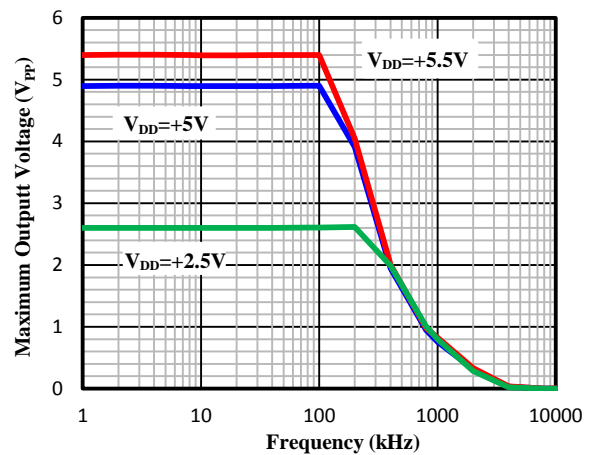


Figure 12. Maximum Output Voltage vs. Frequency

TYPICAL PERFORMANCE CHARACTERISTICS

At  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = +5\text{V}$ ,  $V_{SS} = \text{GND}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_{DD}/2$ , unless otherwise specified

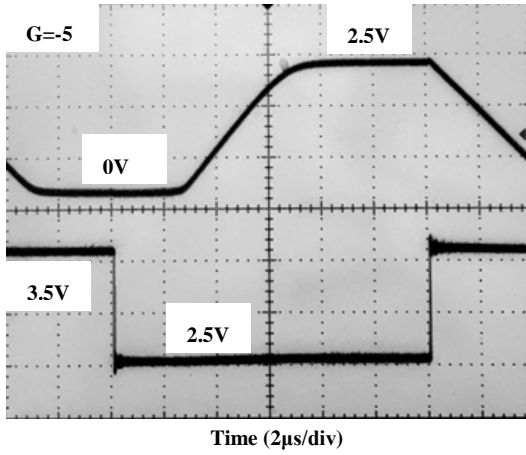


Figure 13. Positive Overload Recovery Time

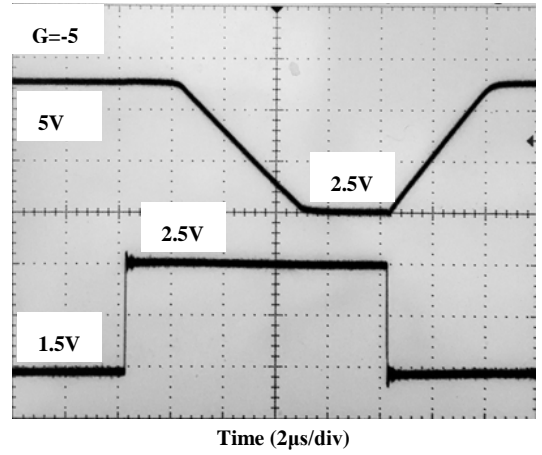


Figure 14. Negative Overload Recovery Time

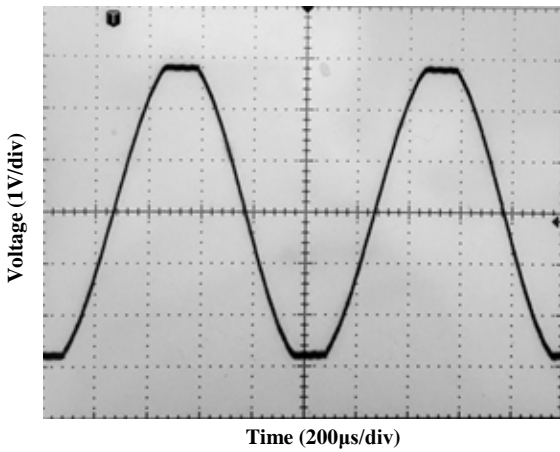


Figure 15. Phase Reversal

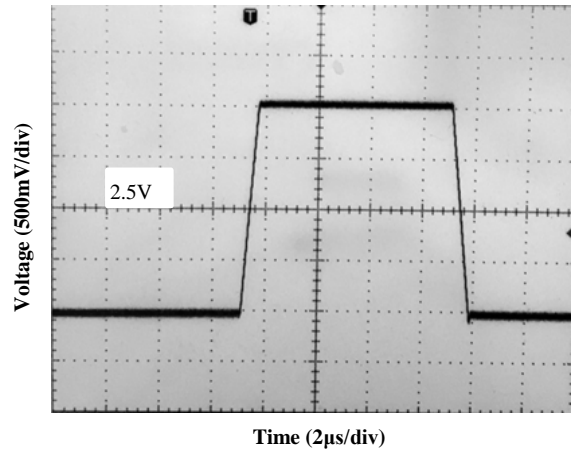


Figure 16. Large-Signal Step Response

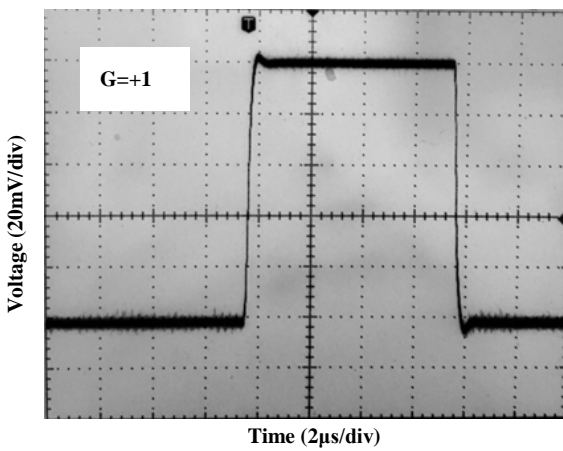


Figure 17. Small-Signal Step Response

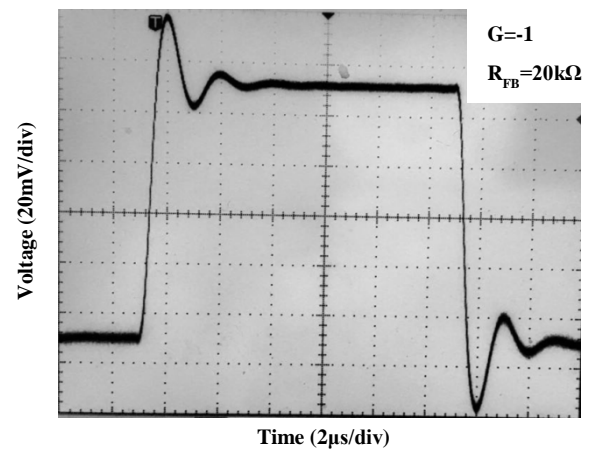


Figure 18. Small-Signal Step Response

**TYPICAL PERFORMANCE CHARACTERISTICS**

At  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = +5\text{V}$ ,  $V_{SS} = \text{GND}$ , and  $R_L = 100\text{k}\Omega$  connected to  $V_{DD}/2$ , unless otherwise specified.

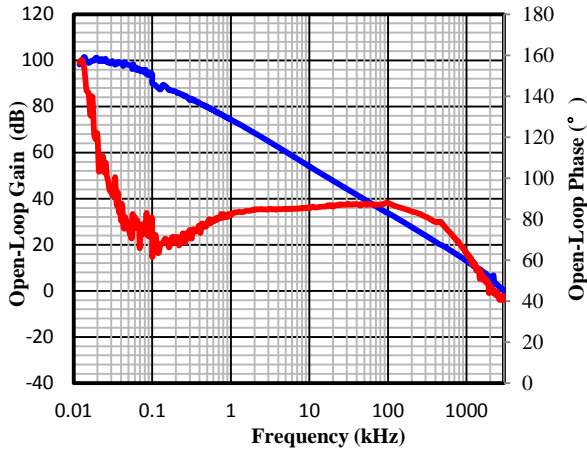


Figure 19. Gain and Phase vs. Frequency

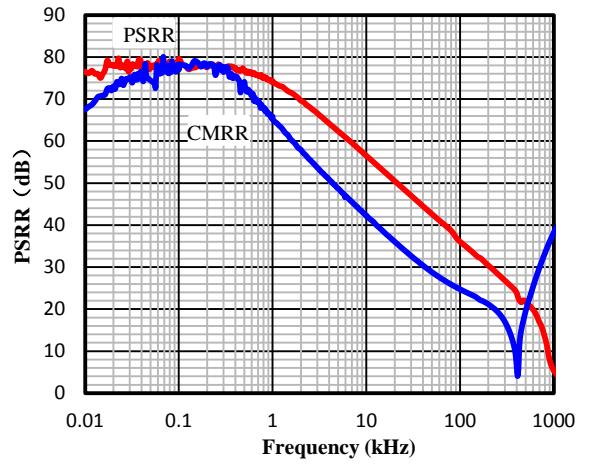


Figure 20. CMRR and PSRR vs. Frequency

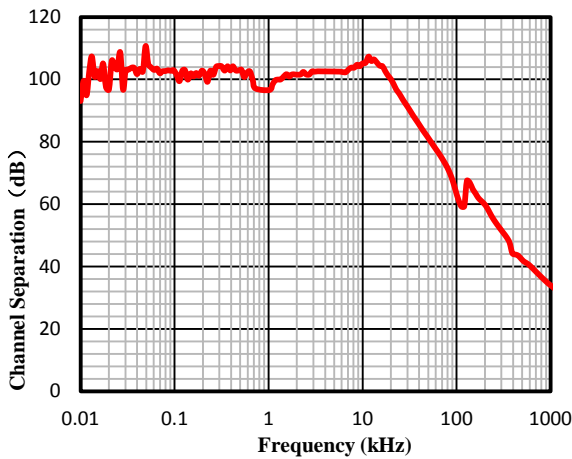


Figure 21. Channel Separation vs. Frequency

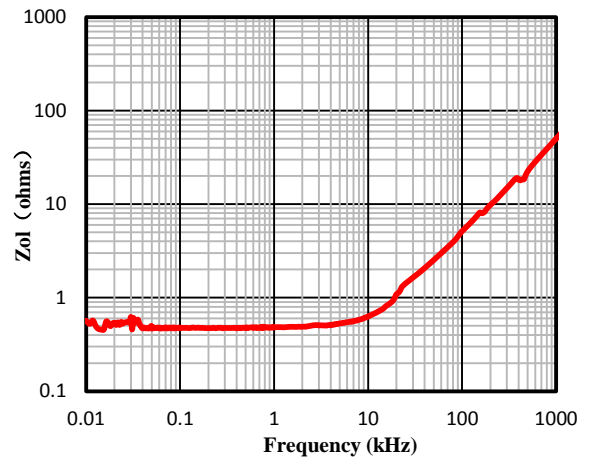


Figure 22. Zol vs. Frequency

MD1621/2

APPLICATION INFORMATION

MD1621/2 are CMOS, rail-to-rail input and output voltage feedback amplifiers designed for general purpose applications.

**Operating Voltage**

The MD1621/2 are specified over a power-supply range of +2.1V to +5.5V ( $\pm 1.05V$  to  $\pm 2.75V$ ), Supply voltages higher than 6V (absolute maximum) can permanently damage the amplifier.

Parameters that vary over supply voltage or temperature are shown in the typical characteristics section of this datasheet.

**Rail-to-Rail Input**

The input stage of the amplifiers is a true rail-to-rail architecture, allowing the input common-mode voltage range of the op amp to extend to both positive and negative supply rails. This maximizes the usable voltage range of the amplifier, an important feature for single-supply and low voltage applications. This rail-to-rail input range is achieved with a complementary input stage—an NMOS input differential pair in parallel with a PMOS differential pair. The NMOS pair is active at the upper end of the common-mode voltage range, typically  $V_{DD} - 1.2V$  to 100mV above the positive supply, while the PMOS pair is active for inputs from 100mV below the negative supply to approximately  $V_{DD} - 1.2V$ .

**Rail-to-Rail Output**

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. The maximum output voltage swing is proportional to the output current, and larger currents will limit how close the output voltage can get to the proximity of the output voltage to the supply rail. This is a characteristic of all rail-to-rail output amplifiers. See the typical performance characteristic Figure 9, Output Voltage Swing vs. Output Current.

**Capacitive Loads**

The MD1621/2 op amps can directly drive large capacitive loads. As the load capacitance increases, the feedback loop's phase margin decreases and the closed-loop's bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. While a op amp in unity gain configuration ( $G = +1 V/V$ ) is most susceptible to the effects of capacitive loading.

When driving large capacitive loads with the MD1621/2 amplifiers (e.g.,  $> 100pF$  when  $G = +1 V/V$ ), a small series resistor at the output ( $R_{ISO}$  in Figure 23) improves the feedback

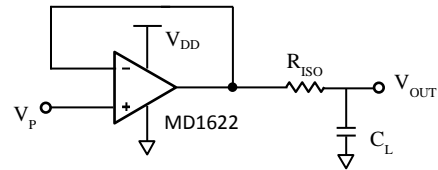


Figure 23. Driving Large Capacitive Loads

loop's phase margin (stability) by making the output load resistive at higher frequencies.

**PCB Surface Leakage**

In Applications where low input bias current is critical, PCB board surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is  $10^{12} \Omega$ . A 5V difference would cause 5pA of current to flow; which is similar to the MD1621/2 op amps' bias current at +25 $^{\circ}C$  ( $\pm 10pA$ , typical).

The best way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 24.

1. Non-inverting Gain and Unity-Gain Buffer:
  - a) Connect the non-inverting pin ( $V_{IN+}$ ) to the input with a wire that does not touch the PCB surface.
  - b) Connect the guard ring to the inverting input pin ( $V_{IN-}$ ). This biases the guard ring to the Common Mode input voltage.
2. Inverting Gain and Transimpedance Gain Amplifiers (convert current to voltage, such as photo detectors):
  - a) Connect the guard ring to the non-inverting input pin ( $V_{IN+}$ ). This biases the guard ring to the same reference voltage as the op amp (e.g.,  $V_{DD}/2$  or ground).
  - b) Connect the inverting pin ( $V_{IN-}$ ) to the input with a wire that does not touch the PCB surface.

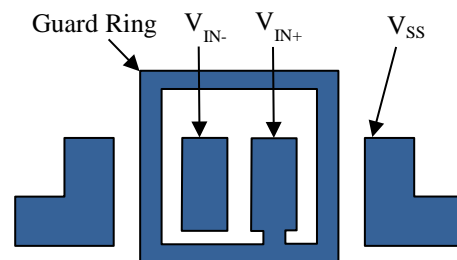


Figure 24. Example Guard Ring Layout for Inverting Gain



**TYPICAL APPLICATION**

**Differential Amplifier**

The circuit shown in Figure 25 performs the difference function. If the resistor ratios are equal to ( $R_4 / R_3 = R_2 / R_1$ ), then  $V_{OUT} = (V_P - V_N) \times R_2 / R_1 + V_{REF}$ .

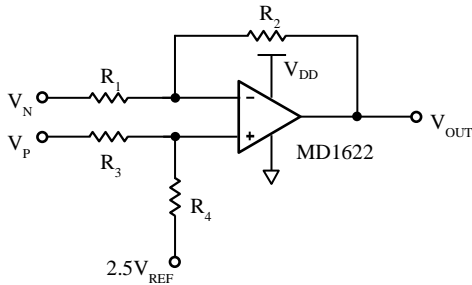


Figure 25. Differential Amplifier

**Photodiode Application**

The MD1621/2 have very high impedance with an input bias current typically around 10 pA. This characteristic allows the MD1621/2 op amp to be used in photodiode applications and other applications that require high input impedance. Note that the MD1621/2 have significant voltage offset that can be removed by capacitive coupling or software calibration.

Figure 26 illustrates a photodiode or current measurement application. The feedback resistor is limited to 10 MΩ to avoid

excessive output offset. In addition, a resistor is not needed on the noninverting input to cancel bias current offset because the bias current-related output offset is not significant when compared to the voltage offset contribution. For best performance, follow the standard high impedance layout techniques, which include the following:

- Shielding the circuit.
- Cleaning the circuit board.
- Putting a trace connected to the noninverting input around the inverting input.
- Using separate analog and digital power supplies.

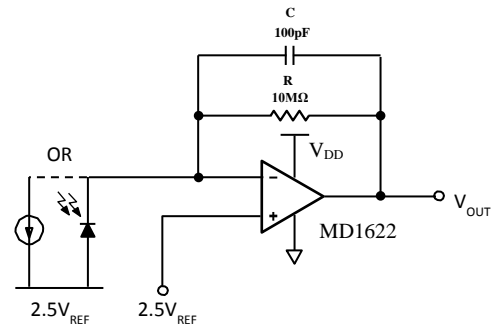
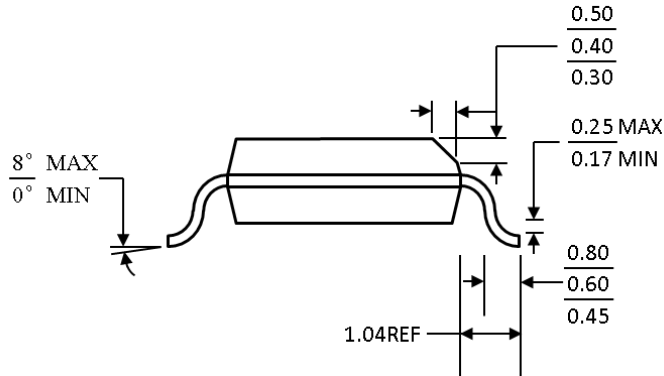
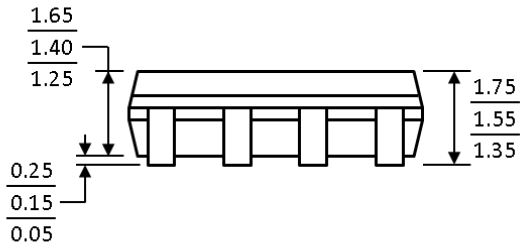
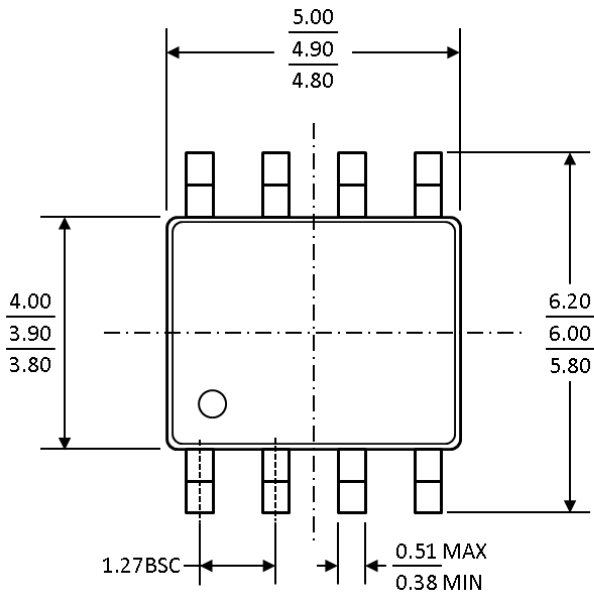


Figure 26. High Input Impedance Application—Photodiode Amplifier

PACKAGE OUTLINE DIMENSIONS

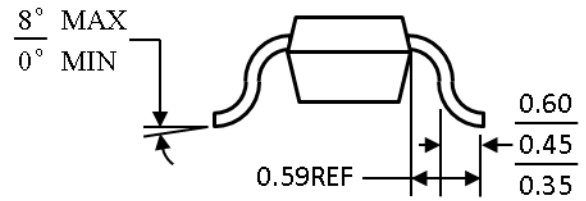
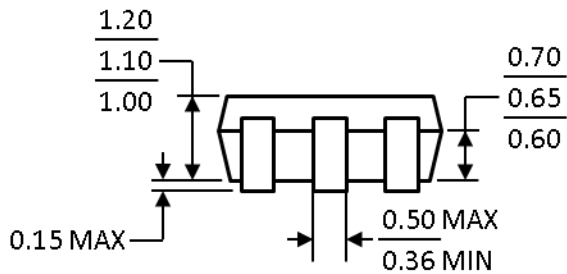
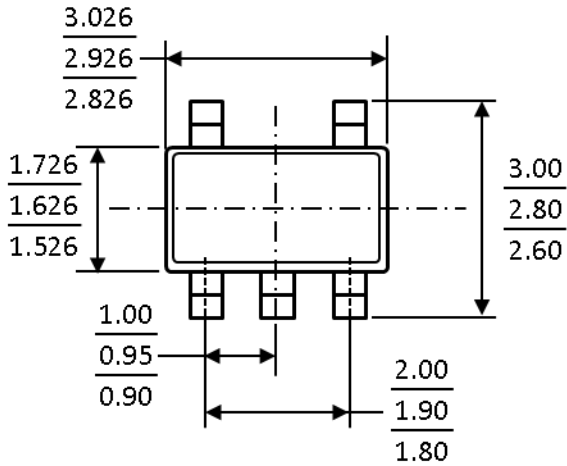


COMPLIANT TO JEDEC STANDARD MS-012-AA

Figure 27 8-Lead Small Outline Package [SOIC]

Dimensions shown in millimeters

PACKAGE OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARD MO-178-AA

Figure 28 5-Lead Small Outline Transistor Package [SOT-23]

Dimensions shown in millimeters

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