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DESCRIPTION

The MP1924 is a high-frequency, 100V, halfbridge, N-channel, power MOSFET driver. Its low-side and high-side driver channels are independently controlled and matched with less than 5ns in time delay. Under-voltage lockout on both high-side and low-side supplies force their outputs low in case of insufficient supply. The integrated bootstrap diode reduces external component count.

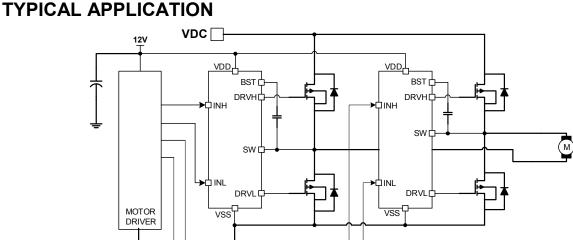
FEATURES

- Drives an N-Channel MOSFET Half Bridge
- 118V V_{BST} Voltage Range
- On-Chip Bootstrap Diode
- Typical Propagation Delay of 20ns
- Gate Drive Matching of Less than 5ns
- Drives a 2.2nF Load with 15ns Rise Time and 12ns Fall Time at 12V VDD
- TTL-Compatible Input
- Quiescent Current of Less than 150µA
- UVLO for Both High Side and Low Side
- QFN-10 (4mmx4mm) and SOIC-8 Packages

APPLICATIONS

- Motor Drivers
- Telecom Half-Bridge Power Supplies
- Avionics DC-DC Converters
- Two-Switch Forward Converters
- Active-Clamp Forward Converters

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MP1924 Rev. 1.0 1/14/2015

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ORDERING INFORMATION

Part Number	Package	Top Marking		
MP1924HR*	QFN-10 (4x4mm)	See Below		
MP1924HS**	SOIC-8	See Below		

* For Tape & Reel, add suffix –Z (e.g. MP1924HR–Z) For RoHS compliant packaging, add suffix –LF (e.g. MP1924HR–LF–Z) ** For Tape & Reel, add suffix –Z (e.g. MP1924HS–Z) For RoHS compliant packaging, add suffix –LF (e.g. MP1924HS–LF–Z)

TOP MARKING (MP1924HR)

MPSYWW

MP1924

LLLLLL

MPS: MPS prefix; Y: year code; WW: week code; MP1924: product code of MP1924HR; LLLLLLL: lot number;

TOP MARKING (MP1924HS)

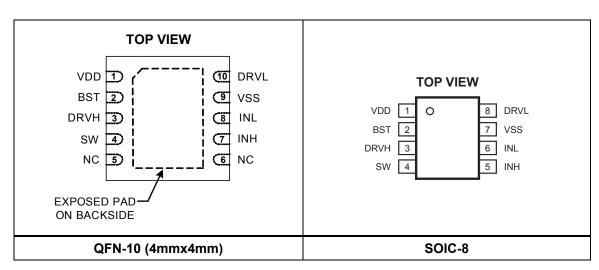
MP1924

LLLLLLL

MPSYWW

MP1924: product code of MP1924HS; LLLLLLL: lot number; MPS: MPS prefix; Y: year code; WW: week code;





PACKAGE REFERENCE

ABSOLUTE MAXIMUM RATINGS (1)

Supply Voltage (V _{DD})0.3V to 18V SW Voltage (V _{SW})5.0V to 105V
BST Voltage (V _{BST})0.3V to 118V
BST to SW0.3V to 18V
DRVH to SW0.3V to (BST-SW) + 0.3V
DRVL to VSS0.3V to (VDD + 0.3V)
All Other Pins $0.3V$ to (V _{DD} + $0.3V$)
Continuous Power Dissipation $(T_A = 25^{\circ}C)^{(2)}$
QFN-10 (4mmx4mm)2.66W
SOIC-81.3W
Junction Temperature
Lead Temperature
Storage Temperature65°C to 150°C

Recommended Operating Conditions ⁽³⁾

Supply Voltage V _{DD}	9.0V to 16.0V
SW Voltage (V _{SW})	1.0V to 100V
SW Slew Rate	<50V/ns
Operating Junction Temp. (T _J).	40°C to 125°C

Thermal Resistance θJA θJC QFN-10 (4mmx4mm) 47 7 °C/W SOIC-8 96 45 °C/W

Notes:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T_J(MAX), the junction-to-ambient thermal resistance θ_{JA}, and the ambient temperature T_A. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P_D(MAX)=(T_J(MAX)-T_A)/ θ_{JA}. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- 3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



ELECTRICAL CHARACTERISTICS

$V_{DD} = V_{BST}-V_{SW} = 12V$, $V_{SS} = V_{SW} = 0V$, No load at DRVH and DRVL, $T_A = +25^{\circ}C$, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Supply Currents						
VDD quiescent current	I _{DDQ}	INL = INH = 0		100	150	μA
VDD operating current	I _{DDO}	fsw = 500kHz		9		mA
Floating driver quiescent current	I _{BSTQ}	INL = INH = 0		60	90	μA
Floating driver operating current	I _{BSTO}	fsw = 500kHz		7.5		mA
Leakage current	I _{LK}	BST = SW = 100V		0.05	1	μA
Inputs						
INL/INH High				2	2.4	V
INL/INH Low			1	1.4		V
INL/INH internal pull-down resistance	R _{IN}			185		kΩ
Under Voltage Protection			·			
VDD rising threshold	V _{DDR}		8.1	8.4	8.8	V
VDD hysteresis	V _{DDH}			0.5		V
(BST-SW) rising threshold	V _{BSTR}		6.9	7.3	7.7	V
(BST-SW) hysteresis	V _{BSTH}			0.55		V
Bootstrap Diode			·			
Bootstrap diode VF @ 100µA	V_{F1}			0.5		V
Bootstrap diode VF @ 100mA	V _{F2}			0.95		V
Bootstrap diode dynamic R	R _D	@ 100mA		2		Ω
Low Side Gate Driver						
Low level output voltage	V _{OLL}	I _o = 100mA		0.08		V
High level output voltage to rail	V _{OHL}	I _o = -100mA		0.23		V
Source Current ⁽⁵⁾	I _{OHL}	$V_{DRVL} = 0V, V_{DD} = 12V$		3		А
Source Current		$V_{DRVL} = 0V, V_{DD} = 16V$		4.7		Α
Sink Current ⁽⁵⁾	1	$V_{DRVL} = V_{DD} = 12V$		4.5		Α
	I _{OLL}	$V_{DRVL} = V_{DD} = 16V$		6		А
Floating Gate Driver						
Low level output voltage Vol		I _O = 100mA		0.08		V
High level output voltage to rail		I _O = -100mA		0.23		V
Source Current ⁽⁵⁾	I _{ОНН}	V_{DRVH} = 0V, V_{DD} = 12V		2.6		А
		$V_{DRVH} = 0V, V_{DD} = 16V$		4		Α
Sink Current ⁽⁵⁾	1	$V_{DRVH} = V_{DD} = 12V$		4.5		Α
	I _{OLH}	$V_{DRVH} = V_{DD} = 16V$		5.9		Α



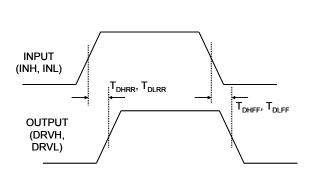
ELECTRICAL CHARACTERISTICS (continued)

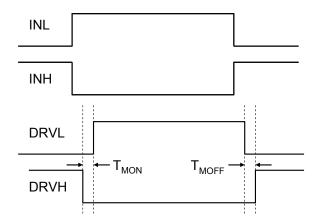
 $V_{DD} = V_{BST}-V_{SW} = 12V$, $V_{SS} = V_{SW} = 0V$, No load at DRVH and DRVL, $T_A = +25^{\circ}C$, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Switching Spec Low Side Gate Driver						
Turn-off propagation delay INL falling to DRVL falling	T_{DLFF}			20		ns
Turn-on propagation delay INL rising to DRVL rising	T _{DLRR}			20		
DRVL rise time		C _L = 2.2nF		15		ns
DRVL fall time		C _L = 2.2nF		9		ns
Switching Spec Floating Gate	e Driver					
Turn-off propagation delay INH falling to DRVH falling	T_{DHFF}			20		ns
Turn-on propagation delay INH rising to DRVH rising	T _{DHRR}			20		ns
DRVH rise time		C _L = 2.2nF		15		ns
DRVH fall time		C _L = 2.2nF		12		ns
Switching Spec Matching					-	-
Floating driver turn-off to low side drive turn-on ⁽⁵⁾	T _{MON}			1	5	ns
Low side driver turn-off to floating driver turn-on ⁽⁵⁾	T_{MOFF}			1	5	ns
Minimum input pulse width that changes the output ⁽⁵⁾	T_PW				50	ns
Bootstrap diode turn-on or turn-off time $^{(5)}$	T_{BS}			10		ns
Thermal shutdown				150		°C
Thermal shutdown hysteresis				25		°C

Note:

5) Guaranteed by design.









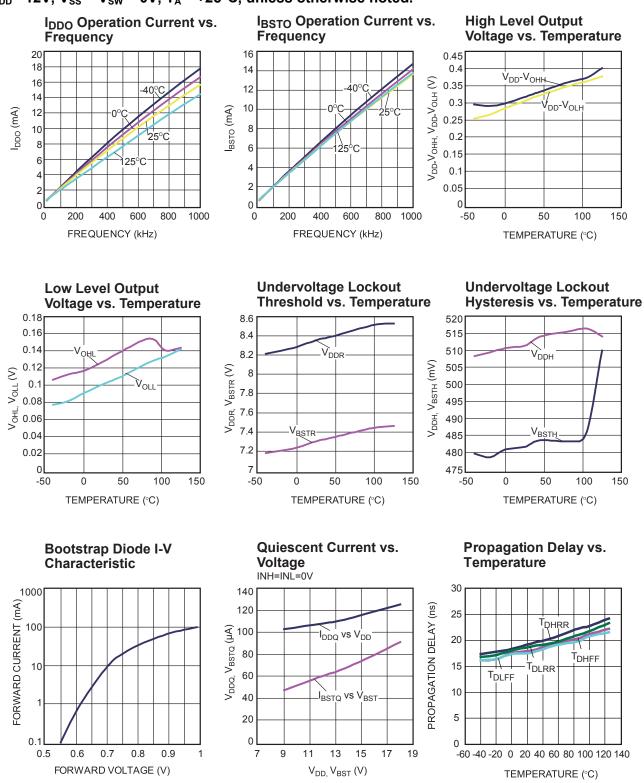
PIN FUNCTIONS

QFN4x4-10 Pin #	SOIC-8 Pin #	Name	Description
1	1	VDD	Supply input. This pin supplies power to all the internal circuitry. Place a decoupling capacitor to ground close to this pin to ensure stable and clean supply.
2	2	BST	Bootstrap. This is the positive power supply for the internal floating high-side MOSFET driver. Connect a bypass capacitor between this pin and SW pin.
3	3	DRVH	Floating driver output.
4	4	SW	Switching node.
5, 6		NC	No connection.
7	5	INH	Control signal input for the floating driver.
8	6	INL	Control signal input for the low side driver.
9	7	VSS, exposed pad	Chip ground. Connect exposed pad to VSS for proper thermal operation.
10	8	DRVL	Low side driver output.



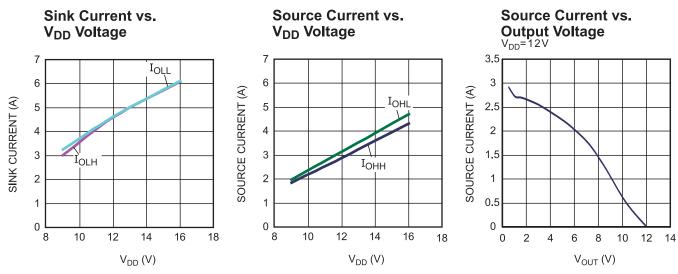
TYPICAL PERFORMANCE CHARACTERISTICS

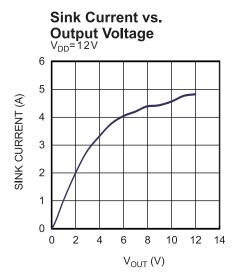
 V_{DD} =12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.



TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 V_{DD} =12V, V_{SS} = V_{SW} = 0V, T_A = +25°C, unless otherwise noted.







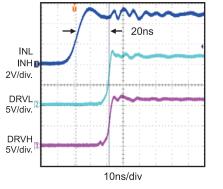
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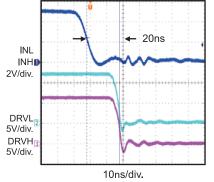
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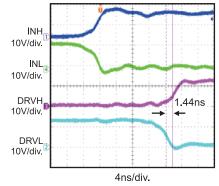
Turn-on Propagation Delay

Turn-off Propagation Delay

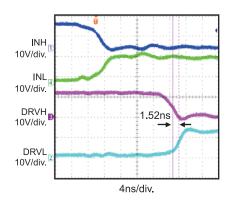
Gate Drive Matching T_{MOFF}

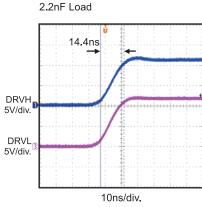




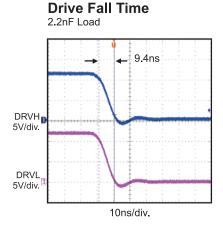


Gate Drive Matching T_{MON}





Drive Rise Time





BLOCK DIAGRAM

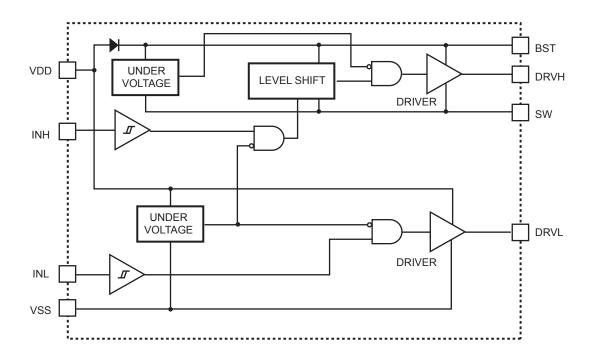


Figure 2: Function Block Diagram



APPLICATION

The input signals of INH and INL can be controlled independently. If both INH and INL control the high-side MOSFET and low-side MOSFET of the same bridge, then users must avoid shoot through by setting sufficient dead time between INH and INL low, and vice versa. See Figure 3 below. Dead time is defined as the time interval between INH low and INL low.

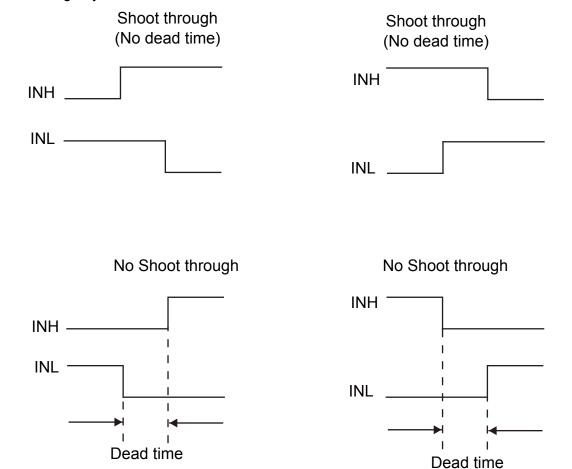


Figure 3: Shoot-Through Timing Diagram



REFERENCE DESIGN CIRCUITS

Half Bridge Converter

The MP1924 drives the MOSFETS with alternating signals (with dead time) in half-bridge converter topology. Therefore, from the PWM

controller drives INH and INL with alternating signals the input voltage can go up to 100V.

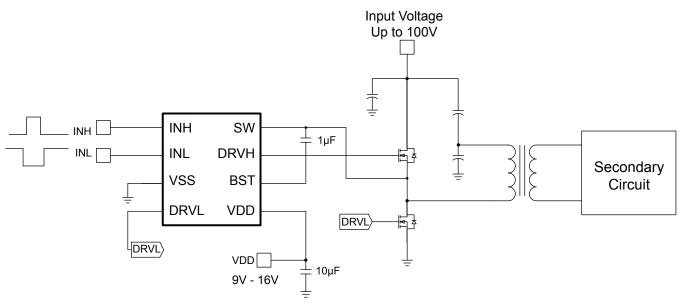


Figure 4: Half Bridge Converter

Two-Switch Forward Converter

In two-switch forward converter topology, both MOSFETs are turned on and off simultaneously. The input signal (INH and INL) comes from a PWM controller that senses the output voltage (and output current during current-mode control).

The Schottky diodes clamp the reverse swing of the power transformer and must be rated for the input voltage. The input voltage can go up to 100V.

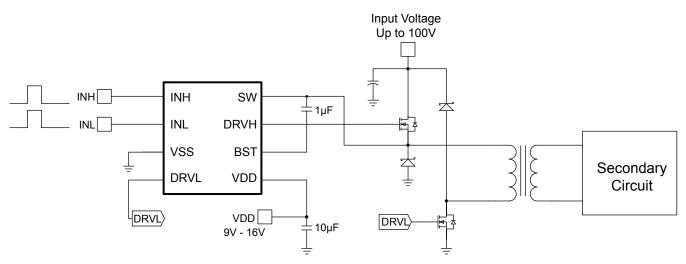


Figure 5: Two-Switch Forward Converter



Active-Clamp Forward Converter

In active-clamp forward converter topology, the MP1924 drives the MOSFETs with alternating signals. The high-side MOSFET, in conjunction with C_{reset} , is used to reset the power transformer in a lossless manner.

This topology lends itself well to run at duty cycles exceeding 50%. The device may not be able to run at 100V under this topology.

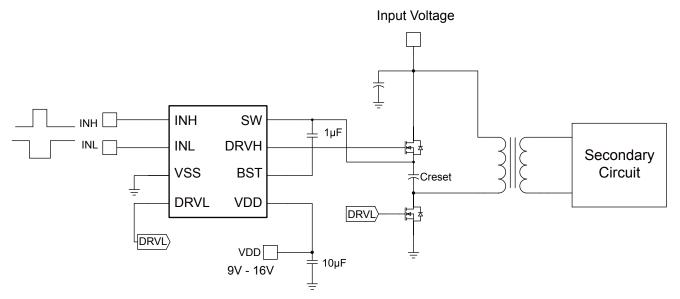
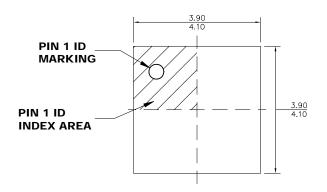


Figure 6 Active-Clamp Forward Converter

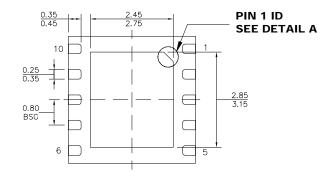


PACKAGE INFORMATION

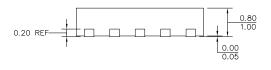
QFN-10 (4mm×4mm)



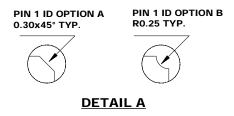
TOP VIEW

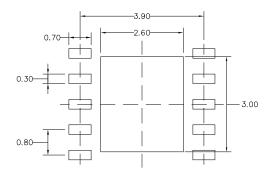


BOTTOM VIEW



SIDE VIEW





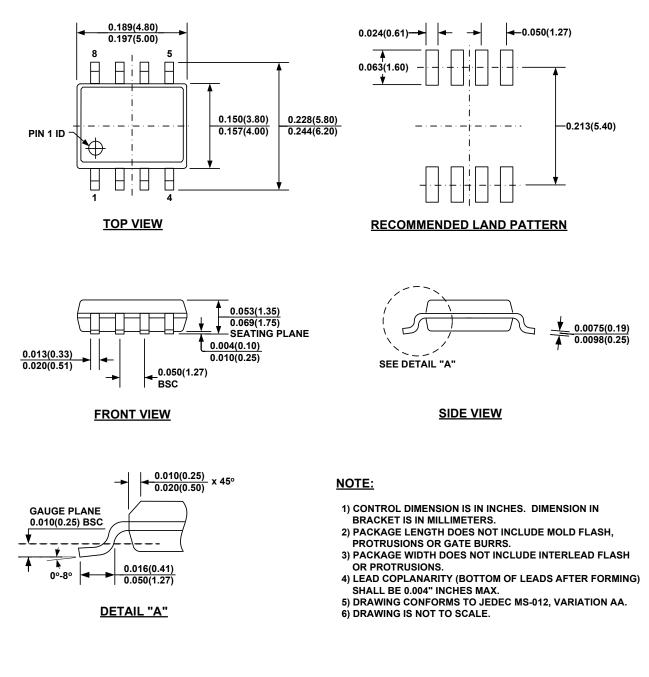
RECOMMENDED LAND PATTERN

NOTE:

 ALL DIMENSIONS ARE IN MILLIMETERS.
 EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
 LEAD COPLANARITY SHALL BE 0.10 MILLIMETERS MAX.
 JEDEC REFERENCE IS MO-220.
 DRAWING IS NOT TO SCALE.



SOIC-8



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