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## AD8552ARZ-MS/AD8554ARZ-MS

### Product specification

**Ultra Low Noise Rail-to-Rail I/O CMOS  
Precision OPERATIONAL AMPLIFIERS**

**GENERAL DESCRIPTION**

The AD8552ARZ-MS/AD8554ARZ-MS family represents a new generation of low-noise operational amplifiers, offering outstanding dc precision and a performance. Rail-to-Rail input and output, low offset (2μV), low noise (6nV/√Hz), quiescent current of 600 μA, and a 6-MHz bandwidth make this part very attractive for a variety of precision and portable applications.

In addition, this device has a reasonably wide supply range (2V to 5.5V) with excellent PSRR making it attractive for applications that run directly from batteries without regulation.

The AD8552ARZ-MS (dual), AD8554ARZ-MS (quad) families of operational amplifiers are specified for operation from -25°C to +125°C.


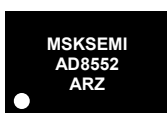


**FEATURES**

- Input Offset Voltage: 2μV (Typical)
- Zero Drift: 0.03μV/°C (Typical)
- Ultra Low Noise: 6nV/√Hz at 1kHz
- Supply Range: 2V to 5.5V
- Gain Bandwidth: 6 MHz
- Slew rate: 5V/μs
- Quiescent current: 600μA (Vs=5V)
- Rail-to-Rail Input and Output
- Micro size Packages:  
AD8552ARZ-MS: SOP-8  
AD8554ARZ-MS: SOP-14

**APPLICATIONS**

- ADC Buffer
- Audio Equipment
- Medical Instrumentation
- Handheld Test Equipment
- Active Filtering
- Sensor Signal Conditioning

**Reference News**

MODEL	Op Temp(°C)	PACKAGE OUTLINE		Marking	Minimum packaging (PCS)
AD8552ARZ-MS	-25°C~125°C	SOP-8			2500
AD8554ARZ-MS	-25°C~125°C	SOP-14			2500

**TYPICAL APPLICATION**

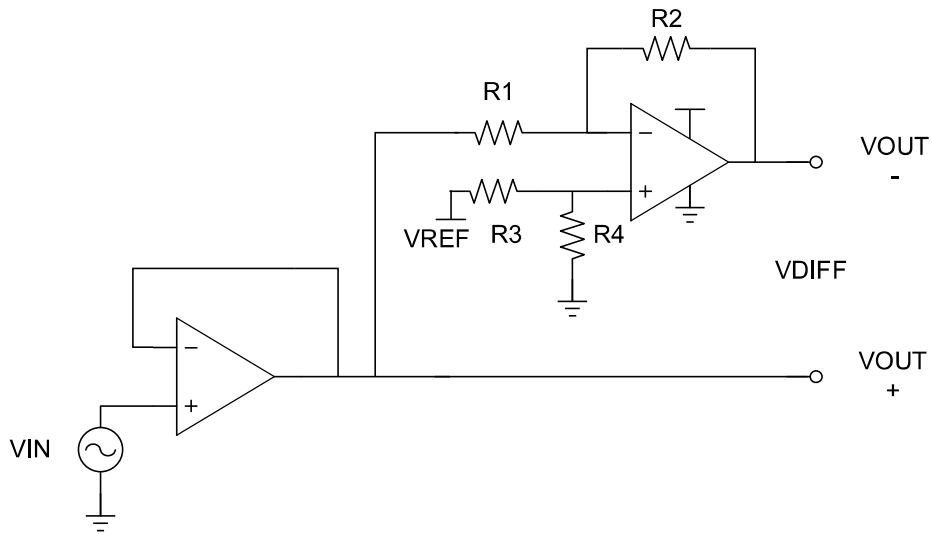
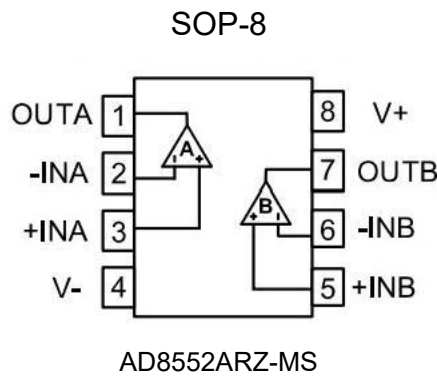


Figure 1. Typical Application

**Pin Configuration and Functions (Top View)**

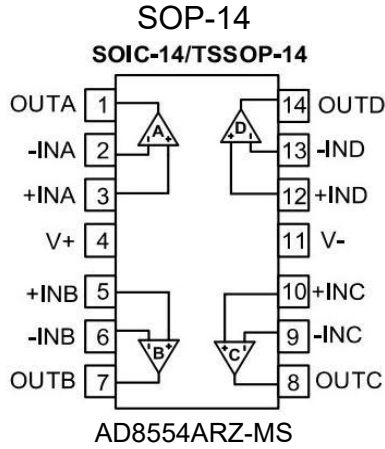
**Pin Description**



PIN		I/O	DESCRIPTION
NAME	Number		
+INA	3	I	Noninverting input, channel A
+INB	5	I	Noninverting input, channel B
-INA	2	I	Inverting input, channel A
-INB	6	I	Inverting input, channel B
OUTA	1	O	Output, channel A
OUTB	7	O	Output, channel B
V-	4	-	Negative (lowest) power supply
V+	8	-	Positive (highest) power supply

**Pin Configuration and Functions (Top View)**

**Pin Description**



PIN		I/O	DESCRIPTION
NAME	Number		
+INA	3	I	Noninverting input, channel A
+INB	5	I	Noninverting input, channel B
+INC	10	I	Noninverting input, channel C
+IND	12	I	Noninverting input, channel D
-INA	2	I	Inverting input, channel A
-INB	6	I	Inverting input, channel B
-INC	9	I	Inverting input, channel C
-IND	13	I	Inverting input, channel D
OUTA	1	O	Output, channel A
OUTB	7	O	Output, channel B
OUTC	8	O	Output, channel C
OUTD	14	O	Output, channel D
V-	4	-	Negative (lowest) power supply
V+	11	-	Positive (highest) power supply

## SPECIFICATIONS

### Absolute Maximum Ratings<sup>(1)</sup>

		MIN	MAX	UNIT
Voltage	Supply Voltage		6	V
	Signal Input Terminals Voltage <sup>(2)</sup>	(V-) - 0.5	(V+) + 0.5	V
	Signal Input Terminals Voltage <sup>(3)</sup>	(V-) - 0.5	(V+) + 0.5	V
Current	Signal Input Terminals Current <sup>(2)</sup>	-10	10	mA
	Signal output Terminals Current <sup>(3)</sup>	-200	200	mA
	Output Short-Circuit <sup>(4)</sup>	Continuous		
$\theta_{JA}$	Operating Temperature Range	-25	125	°C
	Storage Temperature Range	-65	150	°C
	Junction Temperature	-40	150	°C

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

(2) Input terminals are diode clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current limited to 10mA or less.

(3) Output terminals are diode-clamped to the power-supply rails. Output signals that can swing more than 0.5V beyond the supply rails should be current-limited to  $\pm 200$ mA or less.

(4) Short-circuit to ground, one amplifier per package.

### ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-Body Model (HBM)	$\pm 4000$	V
		Charged-Device Model (CDM)	$\pm 500$	V
		Machine Model	100	V

### Recommended Operating Conditions

		MIN	MAX	UNIT
Supply voltage, $V_s = (V+) - (V-)$	Single-supply	2	5.5	V
	Dual-supply	$\pm 1$	$\pm 2.75$	V

**ELECTRICAL CHARACTERISTICS(V<sub>S</sub> = +5V)**

 At T<sub>A</sub> = 25°C, V<sub>CM</sub>=V<sub>OUT</sub>= V<sub>S</sub> /2, unless otherwise noted.

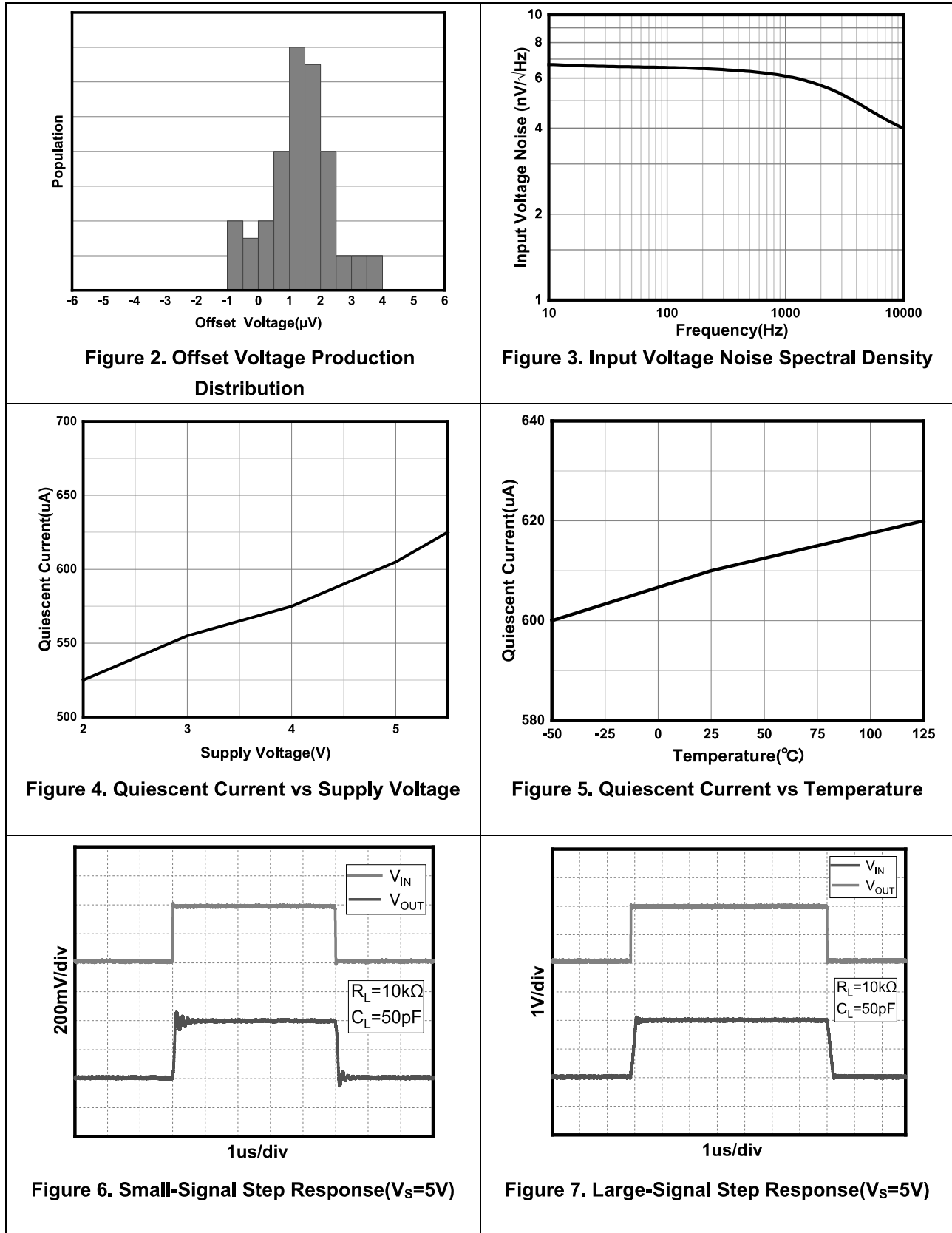
PARAMETER		CONDITIONS	MIN	TYP	MAX	UNIT
<b>OFFSET VOLTAGE</b>						
V <sub>OS</sub>	Input Offset Voltage			2	10	μV
dV <sub>OS</sub> /dT	Input Offset Voltage Average Drift	T <sub>A</sub> = -25°C to 125°C		0.03		μV/°C
<b>INPUT CURRENT</b>						
I <sub>B</sub>	Input Bias Current			500		pA
I <sub>OS</sub>	Input Offset Current			50		pA
<b>NOISE</b>						
V <sub>N</sub>	Input Voltage Noise	f=0.1Hz to 10Hz		0.3		μV <sub>PP</sub>
e <sub>n</sub>	Input Voltage Noise Density	f=1kHz		6		nV/√Hz
<b>INPUT VOLTAGE</b>						
V <sub>CM</sub>	Common-Mode Voltage Range		V <sub>S</sub> -0.1		V <sub>S</sub> +0.1	V
CMRR	Common-Mode Rejection Ratio	V <sub>CM</sub> =0.1V to 4V	110	130		dB
<b>FREQUENCY RESPONSE</b>						
GBW	Gain-Bandwidth Product	C <sub>L</sub> =100pF		6		MHz
SR	Slew Rate	G = +1, V <sub>IN</sub> =2V Step		5		V/μs
t <sub>s</sub>	Settling Time to 0.1%	G = +1, V <sub>IN</sub> =2V Step		0.7		μs
THD+N	Total Harmonic Distortion +Noise	G=1, V <sub>O</sub> =1V <sub>RMS</sub> , f=1kHz, R <sub>L</sub> =10kΩ		0.0004		%
<b>OUTPUT</b>						
A <sub>v</sub>	Open-Loop Voltage Gain	V <sub>OUT</sub> =0.1V to 4.9V R <sub>L</sub> =10kΩ	135	150		dB
V <sub>OH</sub>	High output voltage swing	R <sub>L</sub> =10kΩ		10	20	mV
		R <sub>L</sub> =2kΩ		50	60	mV

$V_{OL}$	Low output voltage swing	$R_L=10k\Omega$		10	20	mV
		$R_L=2k\Omega$		35	45	mV
$I_{SC}$	Output Short-Circuit Current	Source current		30		mA
		Sink current		65		mA
$C_L^{(1)}$	Capacitive Load Drive	$G = +1,$ $V_{IN}=0.2V$ Step			560	pF
<b>POWER SUPPLY</b>						
PSRR	Power-Supply Rejection Ratio	$V_S=1.5V$ to 5.5V	110	130		dB
$V_S$	Operating Voltage Range		2		5.5	V
$I_Q$	Quiescent Current/Amplifier	$I_O=0A$		600	700	$\mu A$

(1) Capacitive load drive means that above a given maximum value, the output waveform will oscillate under the step response.

**TYPICAL CHARACTERISTICS**

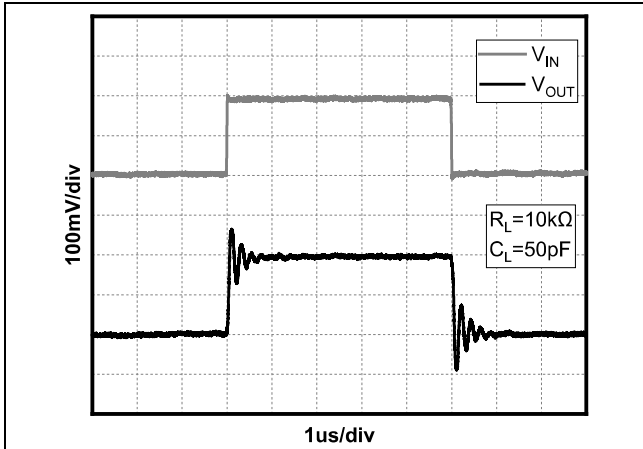
At  $T_A = 25^\circ\text{C}$ ,  $V_S = +5\text{V}$ ,  $G=+1$ ,  $V_{IN}=V_{OUT}= V_S / 2$ , unless otherwise noted.



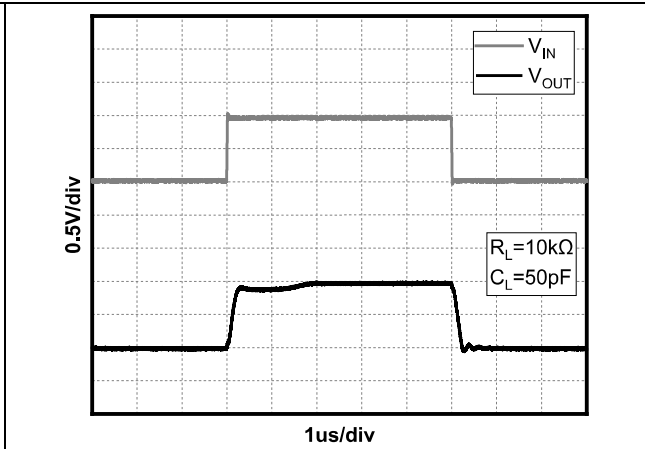


**TYPICAL CHARACTERISTICS**

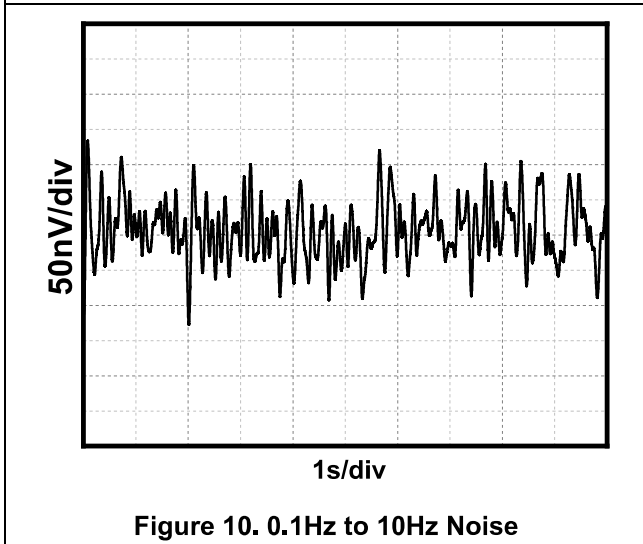
At  $T_A = 25^\circ\text{C}$ ,  $V_S = +5\text{V}$ ,  $G=+1$ ,  $V_{IN}=V_{OUT}= V_S /2$ , unless otherwise noted.



**Figure 8. Small-Signal Step Response( $V_S=2\text{V}$ )**



**Figure 9. Large-Signal Step Response( $V_S=2\text{V}$ )**



**Figure 10. 0.1Hz to 10Hz Noise**

## Detailed Description

### Overview

The AD8552ARZ-MS/AD8554ARZ-MS devices are a low noise, unity-gain stable, rail-to-rail precision operational amplifier that operate in a single-supply voltage range of 2V to 5.5V ( $\pm 1V$  to  $\pm 2.75V$ ). A high supply voltage of 6V (absolute maximum) can permanently damage the amplifier. Rail-to-rail input and output wobbles significantly increase the dynamic range, especially in low-supply applications. Good layout practices require that a 0.1 $\mu$ F capacitor be used where it is tightly threaded through the power supply pin.

### Phase Reversal Protection

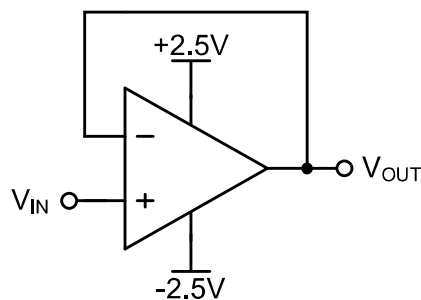
The AD8552ARZ-MS/AD8554ARZ-MS devices have internal phase-reversal protection. Many op amps exhibit phase reversal when the input is driven beyond the linear common-mode range. This condition is most often encountered in noninverting circuits when the input is driven beyond the specified common-mode voltage range, causing the output to reverse into the opposite rail. The input of the AD8552ARZ-MS/AD8554ARZ-MS prevents phase reversal with excessive commonmode voltage. Instead, the appropriate rail limits the output voltage.

## Typical Applications

### 1 Voltage Follower

As shown in Figure 11, the voltage gain is 1. With this circuit, the output voltage  $V_{OUT}$  is configured to be equal to the input voltage  $V_{IN}$ . Due to the high input impedance and low output impedance, the circuit can also stabilize the output voltage, the output voltage expression is

$$V_{OUT} = V_{IN} \quad (1)$$



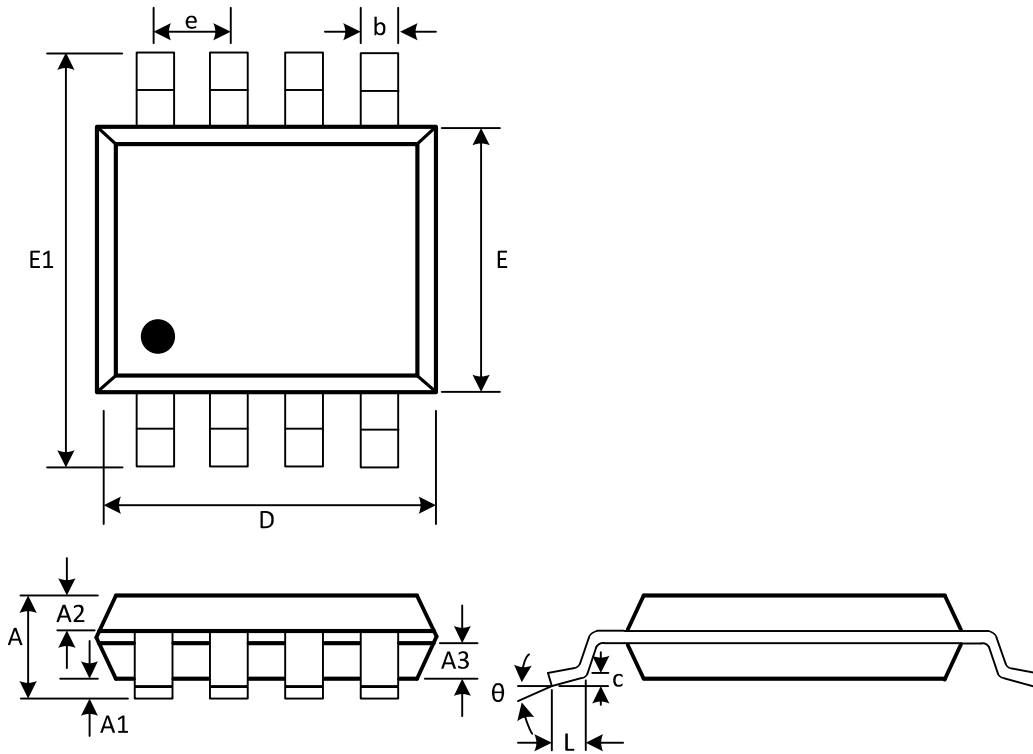
### 2 Inverting Proportional Amplifier

As shown in Figure 12, for a reverse-phase proportional amplifier, the input voltage  $V_{IN}$  is amplified by a voltage gain that depends on the ratio of  $R_1$  to  $R_2$ . The output voltage  $V_{OUT}$  is inversely with the input voltage  $V_{IN}$ . The input impedance of the circuit is equal to  $R_1$ , and the output voltage expression is

(2)

$$V_{OUT} = -\frac{R_2}{R_1} V_{IN}$$

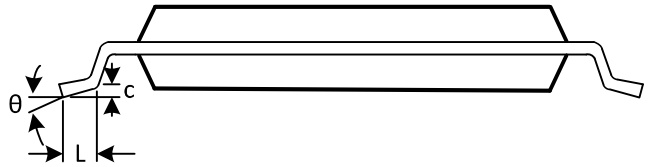
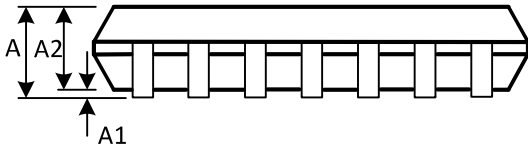
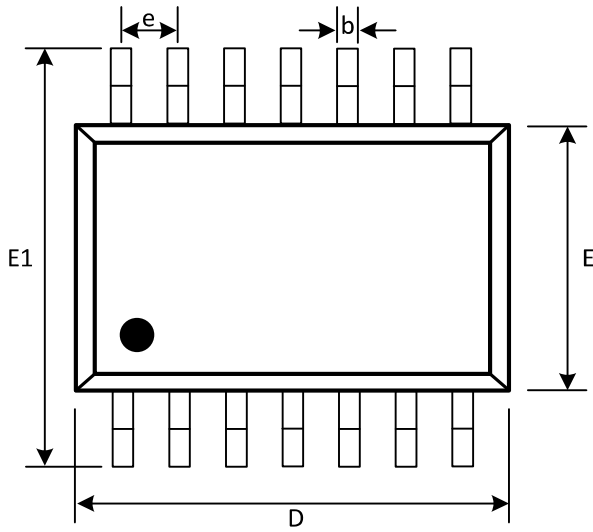
**SOP-8**



(Unit: mm)

Symbol	Min	Max
A	1.300	1.600
A1	0.050	0.200
A2	0.550	0.650
A3	0.550	0.650
b	0.356	0.456
c	0.203	0.233
D	4.800	5.000
e	1.270(BSC)	
E	3.800	4.000
E1	5.800	6.200
L	0.400	0.800
$\theta$	0°	8°

SOP-14



(Unit: mm)

Symbol	Min	Max
A	1.350	1.750
A1	0.100	0.250
A2	1.350	1.550
b	0.310	0.510
c	0.100	0.250
D	8.450	8.850
e	1.270(BSC)	
E	5.800	6.200
E1	3.800	4.000
L	0.400	1.270
$\theta$	0°	8°

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