

LOW VOLTAGE VIDEO AMPLIFIER WITH LPF

■FEATURES

Operating Voltage

e 2.8 to 5.5V -31dB at 19MHz

SOT-23-6-1

- •6th Order LPF
- •16dB Amp. , 75 Ω Driver
- Power Save Circuit
- Bipolar Technology
- Package Outline

■APPLICATION

•Car Camera

CCTV

Car Navigation

■GENERAL DESCRIPTION

The NJM2563 is a Low Voltage Video Amplifier contained LPF circuit. Internal 75 Ω driver is easy to connect TV monitor directly.

The NJM2563 features low power and small package, and is suitable for low power design on downsizing of Car camera and CCTV.

■APPLICATION CIRCUIT



■EQUIVALENT CIRCUIT · BLOCK DIAGRAM



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■Voltage Gain Valuation

| Voltage Gain | Part No. |
|--------------|----------|
| 6.4dB | NJM2561 |
| 6.0dB | NJM2561B |
| 12.4dB | NJM2562 |
| 9.0dB | NJM2571A |

■Output DC - coupling Valuation

| Supply Voltage | Part No. |
|----------------|---------------------|
| 2.8 to 5.5V | NJM2561F1A |
| | (Screening product) |
| 2.8 to 5.5V | NJM2561B |
| 4.5 to 5.5V | NJM41031 |

■Supply Voltage Valuation

| Supply Voltage | Part No. |
|----------------|----------|
| 2.6 to 5.5V | NJM2561A |

■Operating Temperature Range Valuation

| Operating Temperature Range | Part No. |
|-----------------------------|-------------|
| -40 to 105°C | NJM2561F1-T |

■PIN CONFIGURATION



| PIN NO. | SYMBOL | DESCRIPTION |
|---------|------------|------------------------------|
| 1 | Power Save | Power Save Terminal |
| 2 | Vout | Video Signal Output Terminal |
| 3 | Vsag | SAG correction Terminal |
| 4 | Vin | Video Signal Input Terminal |
| 5 | GND | GND Terminal |
| 6 | V+ | Power Supply Terminal |

■MARK INFORMATION



■ORDERING INFORMATION

| PART NUMBER | PACKAGE OUTLINE | RoHS | HALOGEN- FREE | TERMINAL FINISH | MARKING | WEIGHT (mg) | MOQ(pcs) |
|-------------|--------------------|------|------------------|--------------------|---------|----------------|----------|
| NJM2563F1 | SOT-23-6-1 | YES | YES | Sn-2Bi | A6 | 15.0 | 3,000 |

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■ABSOLUTE MAXIMUM RATINGS

| PARAMETER | SYMBOL | RATINGS | UNIT |
|--|------------------|------------|------|
| Supply Voltage | V+ | 7.0 | V |
| Power Dissipation (Ta=25°C) ⁽⁴⁾ | PD | 410 (1) | mW |
| Operating Temperature Range | T _{opr} | -40 to 85 | °C |
| Storage Temperature Range | T _{stg} | -40 to 125 | °C |

(1) At on a board of EIA/JEDEC specification. (114.3 x 76.2 x 1.6mm 2 layers, FR-4)

■RECOMMENDED OPERATING CONDITIONS

| PARAMETER | SYMBOL | RATINGS | UNIT |
|----------------|--------|------------|------|
| Supply Voltage | V+ | 2.8 to 5.5 | V |

■POWER DISSIPATION vs. AMBIENT TEMPERATURE



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| PARAMETER | SYMBOL | TEST CONDITION | MIN. | TYP. | MAX. | UNIT | |
|------------------------------------|-----------------|--|------|------|----------------|------|--|
| Operating Current | I _{CC} | No Signal | - | 8.0 | 12.0 | mA | |
| Operating Current at Power Save | Isave | No Signal, Power Save Mode | - | 30 | 50 | μA | |
| Maximum Output Voltage Swing | Vom | f=100kHz,THD=1% | 2.2 | 2.5 | - | Vр-р | |
| Voltage Gain | Gv | Vin=100kHz, 0.3Vp-p, Input Sine Signal | 16.1 | 16.5 | 16.9 | dB | |
| Low Pass Filter | Gfy4.5M | Vin=4.5MHz/100kHz, 0.3Vp-p | -0.6 | -0.1 | 0.4 | dD | |
| Characteristic | Gfy19M | Vin=19MHz/100kHz, 0.3Vp-p | - | -31 | -21 | uВ | |
| Differential Gain | DG | Vin=0.3Vp-p, 10step Video Signal | - | 0.5 | - | % | |
| Differential Phase | DP | Vin=0.3Vp-p, 10step Video Signal | - | 0.5 | - | deg | |
| S/N Ratio | SNv | Vin=0.3Vp-p, R _L =75Ω 100% White Video Signal, 100kHz to 6MHz | - | 60 | - | dB | |
| 2nd. Distortion | Hv | Vin=0.3Vp-p, 3.58MHz, Sine Signal, R _L =75Ω | - | -50 | - | dB | |
| SW Change Voltage High Level | VthPH | Active | 1.8 | - | V ⁺ | | |
| SW Change Voltage Low Level | VthPL | Non-active | 0 | - | 0.3 | V | |

■ELECTRICAL CHARACTERISTICS (V⁺=3.0V,R_L=150Ω,Ta=25°C)

CONTROL TERMINAL

| PARAMETER | STATUS | NOTE | | |
|------------|--------|--------------------------|--|--|
| Power Save | Н | Power Save: OFF (Active) | | |
| | L | Power Save: ON (Mute) | | |
| | OPEN | Power Save: ON (Mute) | | |

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∎TEST CIRCUIT



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■APPLICATION CIRCUIT (1) Standard circuit









(1) Standard circuit

This circuit is for a portable equipment of small mounting space. The SAG correction reduces output coupling capacitor values. However, this circuit may cause to SAG deterioration, and lose synchronization by luminance fluctuation. Adjust the C1 value, checking the waveform containing a lot of low frequency components like a bounce waveform (Worst condition waveform of SAG). Change the capacitor of C1 into a large value to improve SAG.

(2) SAG correction unused circuit

We recommend this circuit when there is no space limitation. Connect the coupling capacitor after connecting the Vout pin and Vsag pin. The recommended value is 470µF or more.

(3) Two-line driving circuit

This circuit drives two-line of 150Ω. However, it may cause to lose synchronization by an input signal of large APL change (100% white signals more than 1Vp-p). Confirm the large APL change waveform (100% white signals more than 1Vp-p) and evaluate sufficiently.

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TERMINAL DESCRIPTION

| PIN.No. | SYMBOL | EQUIVALENT CIRCUIT | DC VOLTAGE |
|---------|------------|---|------------|
| 1 | Power Save | | - |
| 2 | Vout | Vout | 0.33V |
| 3 | Vsag | Vsag 750Ω 750Ω 750Ω 750Ω 770Ω 770Ω | - |
| 4 | Vin | $Vin \underbrace{\begin{array}{c} \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\$ | 1.1V |
| 5 | GND | - | - |
| 6 | V+ | - | - |

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■APPLICATION

SAG correction circuit

SAG correction circuit is a circuit to correct for low-frequency attenuation by high-pass filter consisting of the output coupling capacitance and load resistance. Low-frequency attenuation raises the sag in the vertical period of the video signal.

Capacitor for Vsag (Csag) is connected to the negative feedback of the amplifier. This Csag increase the low frequency gain to correct for the attenuation of low frequency gain.



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SAG correction circuit generates a low frequency component signal amplified to Vout terminal.

Changes of the luminance signal will be low-frequency components, if you want to output a large signal luminance changes. Therefore, generate correction signal of change of a luminance signal to Vout pin.

At this time, signal is over the dynamic range of Vout pin. This may cause a lack of sync signal, and waveform distortion. Please see diagram below (green waveform), if you want to output large changes of a signal luminance, such as 100% white video signal and black signal. Thus, output signal exceed dynamic range of Vout pin and may be the signal lack.



< Countermeasure for waveform distortion >

1. Please using small value the Sag compensation capacitor (VSAG).

It can ensure the dynamic range by using small value the capacitor (VSAG). It because of low-frequency variation of Vout pin is smaller. However, the output (VOUT) must be use large capacitor for this reason sag characteristics become exacerbated.

2. Please do not use the sag correction circuit.

Signal can output within dynamic range for reason it does not change the DC level of the output terminal. However, the output (VOUT) must be use large capacitor for this reason sag characteristics become exacerbated.

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< Dual drive at using SAG correction circuit >

Using sag correction circuit at dual drive circuit is below. Dual drives are less load resistance. Thus, the cut-off frequency of HPF that is composed of the output capacitor and load resistance will be small. Therefore, the sag characteristics deteriorate. Please size up to the output capacitor (Vout) for not to deteriorate the sag characteristics.



< Dual drive at not using SAG correction circuit >

We recommended two-example dual drive circuit with not use sag correction circuit. Please change the configuration to be used according to the situation. Please configure to meet the following conditions. Then you can adjust the characteristics of each configuration.





< Using SAG correction circuit >



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Input signal: bounce signal (IRE0%, IRE100%, 30Hz), resistance=75Ω, Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal

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< Not using SAG correction circuit > Input_signal: bounce signal (IRE0%, IRE100%, 30Hz), resistance=150Ω, Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal



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Input signal: White100% to Black, resistance150Ω, Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal

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Input signal: White100% to Black, resistance=75Ω, Waveform: yellow: input signal, green: Vout signal, purple: Vout1signal

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Clamp circuit

1. Operation of Sync-tip-clamp

Input circuit will be explained. Sync-tip clamp circuit (below the clamp circuit) operates to keep a sync tip of the minimum potential of the video signal. Clamp circuit is a circuit of the capacitor charging and discharging of the external input Cin. It is charged to the capacitor to the external input Cin at sync tip of the video signal. Therefore, the potential of the sync tip is fixed.

And it is discharged charge by capacitor Cin at period other than the video signal sync tip. This is due to a small discharge current to the IC.

In this way, this clamp circuit is fixed sync tip of video signal to a constant potential from charging of Cin and discharging of Cin at every one horizontal period of the video signal.

The minute current be discharged an electrical charge from the input capacitor at the period other than the sync tip of video signals. Decrease of voltage on discharge is dependent on the size of the input capacitor Cin.

If you decrease the value of the input capacitor, will cause distortion, called the H sag. Therefore, the input capacitor recommend on more than 0.1 uF.



clamp potential clamp potential clamp potential clamp potential clamp potential charge period discharge period charge period charge period charge period discharge period charge period charge period discharge period charge period charge period charge period discharge period charge period charge period discharge period charge period charge period charge period discharge period charge period charge period charge period discharge period charge period

< Waveform of input terminal >

2. Input impedance

The input impedance of the clamp circuit is different at the capacitor discharge period and the charge period.

The input impedance of the charging period is a few k Ω . On the other hand, the input impedance of the discharge period is several M Ω . Because is a small discharge-current through to the IC.

Thus the input impedance will vary depending on the operating state of the clamp circuit.

3. Impedance of signal source

Source impedance to the input terminal, please lower than 200Ω . A high source impedance, the signal may be distorted. If so, please to connect a buffer for impedance conversion.

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Ver.6.1



■PACKAGE OUTLINE

SOT-23-6-1(MTP6-1)





■SOLDER FOOT PRINT

| PKG | b | 1 | С | e1 | e | |
|------------|------|------|-----------|------|------|---|
| SOT-23-6-1 | 0.70 | 1.00 | 1.90 | 2.40 | 0.95 | [|
| | | | UNIT : mm | | | |



Note : These solder foot print dimensions are just examples. When designing PCB, please estimate the pattern carefully.

UNIT : mm



■PACKING SPECIFICATION

General Description

NJRC delivers ICs in 4 methods, plastic tube container, two kinds of Taping, tray and vinyl bag packing. Except adhesive tape treated anti electrostatic and contain carbon are using as the ESD (Electrostatic Discharge Damage) protection.

SOT-23(MTP) Emboss Taping (TE1)

| Symbol | SOT-23-6-1 | Remark |
|----------------|------------|---------------------|
| A | 3.3±0.1 | Bottom size |
| В | 3.2±0.1 | Bottom size |
| Do | 1.55 | |
| D ₁ | 1.05 | |
| E | 1.75±0.1 | |
| F | 3.5±0.05 | |
| Po | 4.0±0.1 | |
| P ₁ | 4.0±0.1 | |
| P ₂ | 2.0±0.05 | |
| Т | 0.25±0.05 | |
| T ₂ | 1.57 | |
| W | 8.0±0.3 | |
| W1 | 5.5 | Thickness 0.1MAX |

| Symbol | SOT-23-6-1 | |
|----------------|------------|-----------|
| A | Ø180±1 | |
| В | Ø 60±1 | |
| C | Ø 13±0.2 | |
| D | Ø 21±0.8 | |
| E | 2±0.5 | |
| W | 9±0.5 | |
| W ₁ | 1.2±0.2 | Unit : mm |
| Contents | 3,000pcs | |



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Label

Label



■RECOMMENDED MOUNTING METHOD

* Recommended reflow soldering procedure



The temperature indicates at the surface of mold package.

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