## DATA SHEET

## PCA146x series <br> 32 kHz watch circuits with adaptive motor pulse

Product specification
Supersedes data of 1998 Mar 18
File under Integrated Circuits, IC16

## 32 kHz watch circuits with adaptive motor pulse

## FEATURES

- 32 kHz oscillator, amplitude regulated with excellent frequency stability
- High immunity of the oscillator to leakage currents
- Time calibration electrically programmable and reprogrammable (via EEPROM)
- A quartz crystal is the only external component required
- Very low current consumption; typically 170 nA
- Output for bipolar stepping motors of different types
- Up to $50 \%$ reduction in motor current compared with conventional circuits, by self adaption of the motor pulse width to match the required torque of the motor
- No loss of motor steps possible because of on-chip detection of the induced motor voltage
- Detector for lithium or silver-oxide battery voltage levels
- Indication for battery end-of-life
- Stop function for accurate timing
- Power-on reset for fast testing
- Various test modes for testing the mechanical parts of the watch and the IC.


## GENERAL DESCRIPTION

The PCA146x series devices are CMOS integrated circuits specially suited for battery-operated, quartz-crystal-controlled wrist-watches, with a bipolar stepping motor.

## ORDERING INFORMATION

| TYPE <br> NUMBER | PACKAGE ${ }^{(1)}$ |  |  |
| :--- | :---: | :--- | :---: |
|  | NAME | DESCRIPTION | VERSION |
| PCA1461U | - | chip in tray | - |
| PCA1461U/10 | - | chip on foil | - |
| PCA1462U | - | chip in tray | - |
| PCA1462U/7 | - | chip with bumps on tape | - |
| PCA1462U/10 | - | chip on foil | - |
| PCA1463U | - | chip in tray | - |
| PCA1463U/10 | - | chip on foil | - |
| PCA1465U/10 | - | chip on foil | - |
| PCA1465U/7 | - | chip with bumps on tape | - |
| PCA1467U/10 | - | chip on foil | - |

## Note

1. Figure 1 and Chapter "Package outline" show details of standard package, available for large orders only. Chapter "Chip dimensions and bonding pad locations" shows exact pad locations for other delivery formats.

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## PINNING

| SYMBOL | PIN | DESCRIPTION |
| :--- | :---: | :--- |
| V $_{\text {SS }}$ | 1 | ground (0 V) |
| TEST | 2 | test output |
| OSC IN | 3 | oscillator input |
| OSC OUT | 4 | oscillator output |
| V $_{\text {DD }}$ | 5 | supply voltage |
| M1 | 6 | motor 1 output |
| M2 | 7 | motor 2 output |
| RESET | 8 | reset input |

## FUNCTIONAL DESCRIPTION AND TESTING

The motor output delivers pulses of six different stages depending on the torque required to turn the motor (Figs. 3 and 4). Every motor pulse is followed by a detection phase which monitors the waveform of the induced motor voltage. When a step is missed a correction sequence will be started (Fig.2).

## Motor pulses

The circuit produces motor pulses of six different stages (stage 1 to 5, stage 8). Each stage has two independent modes: silver-oxide and lithium. The voltage level of $\mathrm{V}_{\mathrm{DD}}$ determines which mode is selected (see Section "Voltage level detector").
Stages 1 to 5 (both modes) are used in normal operation, stage 8 occurs under the following conditions:

- Correction pulse after a missing step (both modes)
- End-of-life mode
- If stage 5 is not enough to turn the motor (both modes).

In the silver-oxide mode, the ON state of the motor pulse varies between $56.25 \%$ and $100 \%$ of the duty factor $t_{D F}=977 \mu$ s depending on the stage (Fig.3). It increases in steps of $6.25 \%$ per stage.


Fig. 1 Pin configuration, PCA146xT, (PMFP8).

In the lithium mode, the ON state of the motor pulse is reduced by $18.75 \%$ of the duty factor $t_{D F}$ (Fig.4) to compensate for the increase in the voltage level.

After a RESET the circuit always starts and continues with stage 1, when all motor pulses have been executed. A failure to execute all motor pulses results in the circuit going into stage 2 , this sequence will be repeated through to stage 8.

When the motor pulses at stage 5 are not large enough to turn the motor, stage 8 is implemented for a maximum of 8 minutes with no attempt to keep current consumption low. After stage 8 has been executed the procedure is repeated from RESET.
The circuit operates for 8 minutes at a fixed stage, if every motor pulse is executed. The next 480 motor pulses are then produced at the next lower stage unless a missing step is detected. If a step is missed a correction sequence is produced and for a maximum of 8 minutes the motor pulses are increased by one stage.

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Fig. 2 Possible motor output waveform in normal operation with motor connected.

$t_{\text {OFF }}$ for stage 1 to $5=488 \mu \mathrm{~s}-$ stage $\times 61 \mu \mathrm{~s}$
ton for stage 1 to $5=488 \mu \mathrm{~s}+$ stage $\times 61 \mu \mathrm{~s}$

Fig. 3 Motor pulses in the silver-oxide mode ( $\mathrm{V}_{\mathrm{DD}}=1.55 \mathrm{~V}$ ).


Fig. 4 Motor pulses in the lithium mode ( $\mathrm{V}_{\mathrm{DD}}=2.1 \mathrm{~V}$ ).

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## Voltage level detector

The supply voltage is compared with the internal voltage reference $\mathrm{V}_{\text {LIT }}$ and $\mathrm{V}_{\text {EOL }}$ every minute. The first voltage level detection is carried out 30 ms after RESET.

When a lithium voltage level is detected ( $\mathrm{V}_{\mathrm{DD}} \geq \mathrm{V}_{\mathrm{LIT}}$ ), the circuit starts operating in the lithium mode (Fig.4).

When the detected $\mathrm{V}_{\text {DD }}$ voltage level is between $\mathrm{V}_{\text {LIT }}$ and $\mathrm{V}_{\text {EOL }}$, the circuit operates in the silver-oxide mode (Fig.3).

If the battery end-of-life is detected ( $\mathrm{V}_{\mathrm{DD}}<\mathrm{V}_{\mathrm{EOL}}$ ), the detection and stage control is switched OFF and the waveform produced is an unchopped version of the stage 8 waveform. To indicate this condition the waveform is produced in bursts of 4 pulses every 4 s .

## Detection of motor movement

After a motor pulse, the motor is short-circuited to $V_{D D}$ for 1 ms . Afterwards the energy in the motor inductor will be dissipated to measure only the current generated by the induced motor voltage. During the time $t_{\text {DI }}$ (dissipation of energy time) all switches shown in Fig. 5 are open to reduce the current as fast as possible. The current will now flow through the diodes D3 and D2, or D4 and D1. Then the first of 52 possible measurement cycles ( $\mathrm{t}_{\mathrm{MC}}$ ) starts to measure the induced current.


Fig. 5 Motor driving and detecting circuit.

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## Detection criteria

The PCA146x uses current detection in two defined parts of the detection phase to determine if the motor has moved (refer to Figs 6 and 7). The detection criteria are:

## part 1

- Minimum value of $P=1$; where $P=$ number of measured positive current polarities after $t_{D I}$.


## part 2

- Minimum value of $\mathrm{N}=2$; where $\mathrm{N}=$ number of measured positive current polarities since the first negative current polarity after part 1 was detected (see Fig.6).

If the opposite polarity is measured in one part, the internal counter is reset, so the results of all measurements in this part are ignored.

The waveform of the induced current must enable all these measurements within the time $t_{D}$ after the end of a positive motor pulse in order to be accepted as a waveform of an executed motor pulse.

If the detection criterion is satisfied earlier, a measurement cycle will not be started and the switches P1 and P2 stay closed, the motor is switched to $V_{D D}$.

Every measurement cycle ( $\mathrm{t}_{\mathrm{MC}}$ ) has 4 phases. These are detailed in Table 1.

Note that detection and pulse width control will be switched OFF when the battery voltage is below the end-of-life voltage ( $\mathrm{V}_{\mathrm{EOL}}$ ), or if stage 5 is not sufficient to turn the motor.

Table 1 Measurement cycle

| SYMBOL | PHASE | DESCRIPTION |
| :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{M} 1}$ | 1 | During $\mathrm{t}_{\mathrm{M} 1}$ the switches P 1 and P 2 are closed in order to switch the motor to $\mathrm{V}_{\mathrm{DD}}$, so the induced current flows unaffected through the motor inductance. |
| $\mathrm{t}_{\mathrm{M} 2}$ | 2 | Measures the induced current; during a maximum time $\mathrm{t}_{\mathrm{M} 2}$ all switches are open until a change is sensed by one of the level detectors (L1, L2). The motor is short-circuited to $\mathrm{V}_{\mathrm{DD}}$. <br> Depending on the direction of the interrupted current: <br> - The current flows through diodes D3 and D2, causing the voltage at M1 to decrease in relation to M2; <br> - The current flows through diodes D4 and D1, causing the voltage at M2 to decrease in relation to M1. <br> A successfully detected current polarity is normally characterized by a short pulse of 0.5 to $10 \mu \mathrm{~s}$ with a voltage up to $\pm 2.1 \mathrm{~V}$, failed polarity detection by the maximum pulse width of $61 \mu$ s and a voltage of $\pm 0.5 \mathrm{~V}$ (see Fig.7). |
| $\mathrm{t}_{\text {M }}$ | 3 | The switches P1 and P2 remain closed for the time $\mathrm{t}_{\mathrm{M} 3}$. |
| $\mathrm{t}_{\mathrm{M} 4}$ | 4 | If the circuit detects fewer pulses than $P$ and $N$ respectively, a pulse of the time $t_{M 4}$ occurs to reduce the induced current. Therefore P2 and P1 are opened and N 1 and N 2 are closed. Otherwise P1 and P2 remain closed. |

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Fig. 6 Typical current waveform of a successfully executed motor pulse.


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## Correction sequence (see Fig.8)

If a missing step is detected, a correction sequence is produced. This consists of a small pulse ( $\mathrm{t}_{\mathrm{c}_{1}}$ ) which gives the motor a defined position and after 29.30 ms a pulse of stage $8\left(\mathrm{t}_{\mathrm{c} 2}\right)$ to turn the motor.


Fig. 8 Correction sequence after a missing motor step with motor connected.

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## Time calibration

Taking a normal quartz crystal with frequency 32768 kHz , frequency deviation ( $\Delta \mathrm{f} / \mathrm{f}$ ) of $\pm 15 \times 10^{-6}$ and $\mathrm{C}_{\mathrm{L}}=8.2 \mathrm{pF}$; the oscillator frequency is offset (by using non-symmetrical internal oscillator input and output capacitances of 10 pF and 15 pF ) such that the frequency deviation is positive-only. This positive deviation can then be compensated for to maintain time-keeping accuracy.

Once the positive frequency deviation is measured, a corresponding number ' $n$ ' (see Table 2) can be programmed into the device's EEPROM. This causes $n$ pulses of frequency 8192 Hz to be inhibited every minute of operation, which achieves the required calibration.

The programming circuit is shown in Fig.9. The required number n is programmed into EEPROM by varying $\mathrm{V}_{\mathrm{DD}}$ according to the steps shown in Fig.10, which are explained below:

1. The positive quartz frequency deviation $(\Delta f / f)$ is measured, and the corresponding values of $n$ are found according to Table 2.
2. $\mathrm{V}_{\mathrm{DD}}$ is increased to 5.1 V allowing the contents of the EEPROM to be checked from the motor pulse period $\mathrm{t}_{\mathrm{T} 3}$ at nominal frequency.

Table 2 Quartz crystal frequency deviation, n and $\mathrm{t}_{\mathrm{T} 3}$

| FREQUENCY <br> DEVIATION <br> $\Delta \mathbf{f} / \mathbf{f}$ <br> $\left(\times \mathbf{1 0}^{-6}\right)$ | NUMBER OF <br> PULSES <br> $\mathbf{( n )}$ | $\mathbf{t}_{\mathbf{T} 3}$ <br> $\mathbf{( m s )}$ |
| :---: | :---: | :---: |
| $0^{(1)}$ | 0 | $31.250^{(2)}$ |
| +2.03 | 1 | 31.372 |
| +4.06 | 2 | 31.494 |
| . | . | . |
| . | . | . |
| . | . | . |
| +127.89 | 63 | 38.936 |

## Notes

1. Increments of $2.03 \times 10^{-6} /$ step.
2. Increments of $122 \mu \mathrm{~s} /$ step.
3. $\mathrm{V}_{\mathrm{DD}}$ is decreased to 2.5 V during a motor pulse to initialize a storing sequence.
4. The first $\mathrm{V}_{\mathrm{DD}}$ pulse to 5.1 V erases the contents of EEPROM.
5. When the EEPROM is erased a logic 1 is at the TEST pin.
6. $\mathrm{V}_{\mathrm{DD}}$ is increased to 5.1 V to read the data by pulsing $\mathrm{V}_{\mathrm{DD}} \mathrm{n}$ times to 4.5 V . After the n edge, $\mathrm{V}_{\mathrm{DD}}$ is decreased to 2.5 V .
7. $\mathrm{V}_{\mathrm{DD}}$ is increased to 5.1 V to store n bits in the EEPROM.
8. $\mathrm{V}_{\mathrm{DD}}$ is decreased to 2.5 V to terminate the storing sequence and to return to operating mode.
9. $V_{D D}$ is increased to 5.1 V to check writing from the motor pulse period $\mathrm{t}_{\mathrm{T} 3}$.
10. $\mathrm{V}_{\mathrm{DD}}$ is decreased to the operation voltage between two motor pulses to return to operating mode. (Decreasing $\mathrm{V}_{\mathrm{DD}}$ during the motor pulse would restart the programming mode).
The time calibration can be reprogrammed up to 100 times.


Fig. 9 Circuit for programming the time calibration.

(1) Rise and fall time should be greater than $400 \mu \mathrm{~s} / \mathrm{V}$ for immediately correct checking.

Fig. $10 \mathrm{~V}_{\mathrm{DD}}$ for programming.

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## Power-on reset

For correct operation of the Power-on reset the rise time of $\mathrm{V}_{\mathrm{DD}}$ from 0 V to 2.1 V should be less than 0.1 ms .
All resetable flip-flops are reset. Additionally the polarity of the first motor pulse is positive: $\mathrm{V}_{\mathrm{M} 1}-\mathrm{V}_{\mathrm{M} 2} \geq 0 \mathrm{~V}$.

## Customer testing

An output frequency of 32 Hz is provided at RESET (pin 8) to be used for exact frequency measurement. Every minute a jitter occurs as a result of the inhibition, which occurs 90 to 150 ms after disconnecting the RESET from $V_{D D}$.

Connecting the RESET to $\mathrm{V}_{\mathrm{DD}}$ stops the motor pulses leaving them in a 3-state mode and sets the motor pulse width for the next available motor pulse to stage 1 in the silver-oxide mode. A 32 Hz signal without jitter is produced at the TEST pin.
Debounce time RESET = 14.7 to 123.2 ms .
Connecting RESET to $\mathrm{V}_{\text {SS }}$ activates Tests 1 and 2 and disables the inhibition.

Test 1, $\mathrm{V}_{\mathrm{DD}}>\mathrm{V}_{\mathrm{EOL}}$. Normal function takes place except that the motor pulse period is $\mathrm{t}_{\mathrm{T} 1}=125 \mathrm{~ms}$ instead of $\mathrm{t}_{\mathrm{T}}$, and the motor pulse stage is reduced every second instead of every 8 minutes. At TEST a speeded-up 8 minute signal is available.
Test 2, $\mathrm{V}_{\mathrm{DD}}<\mathrm{V}_{\mathrm{EOL}}$. Motor pulses of stage 8 are produced, with a time period of $\mathrm{t}_{\mathrm{T} 2}=31.25 \mathrm{~ms}$.

Test and reset modes are terminated by disconnecting the RESET pin.

Test 3, $\mathrm{V}_{\mathrm{DD}}>5.1 \mathrm{~V}$. Motor pulses of stage 8 are produced, with a time period of $\mathrm{t}_{\mathrm{T} 3}=31.25 \mathrm{~ms}$ and $n \times 122 \mu$ s to check the contents of the EEPROM. At TEST a speeded-up cycle for motor pulse period signal $t_{\top}$ is available at 1024 times its normal frequency. Decreasing $\mathrm{V}_{\mathrm{DD}}$ voltage level to lower than 2.5 V between two motor pulses returns the circuit to normal operating conditions.

## AVAILABLE TYPES

Refer to Chapters "Ordering information" and "Functional description and testing".

|  |  |  | SPECIFICATIONS |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SHORT TYPE NUMBER | DELIVERY <br> FORMAT ${ }^{(1)}$ | $\begin{aligned} & \text { PERIOD } \\ & \mathbf{t}_{\mathbf{T}} \\ & (\mathbf{s}) \end{aligned}$ | PULSE <br> WIDTH <br> $t_{p}$ (ms) | DRIVE <br> (\%) | DETECTION CRITERION | EEPROM | $\begin{array}{\|l} \text { BATTERY } \\ \text { EOL } \\ \text { DETECTION } \end{array}$ | REMARKS |
| 1461 | U; U/10 | 1 | 7.8 | $\begin{gathered} \text { max. } 100 \\ 81 \end{gathered}$ | $\begin{aligned} & P=1 \\ & N=2 \end{aligned}$ | yes | yes | 1.5 V and 2.1 V Lithium |
| 1462 | $\begin{gathered} \hline \mathrm{U} ; \mathrm{U} / 7 ; \\ \mathrm{U} / 10 \end{gathered}$ | 1 | 5.8 | $\begin{gathered} \operatorname{max.} 100 \\ 81 \end{gathered}$ | $\begin{aligned} & P=1 \\ & N=2 \end{aligned}$ | yes | yes | 1.5 V and 2.1 V Lithium |
| 1463 | U; U/10 | 1 | 3.9 | $\begin{gathered} \text { max. } 100 \\ 81 \end{gathered}$ | $\begin{aligned} & P=1 \\ & N=2 \end{aligned}$ | yes | yes | 1.5 V and 2.1 V Lithium |
| 1465 | U/10; U/7 | 1 | 5.8 | max. 100 | $\begin{aligned} & P=1 \\ & N=2 \end{aligned}$ | yes | no | 1.5 V |
| 1467 | U/10 | 1 | 7.8 | max. 100 | $\begin{aligned} & P=1 \\ & N=2 \end{aligned}$ | yes | no | 1.5 V |

## Note

1. $U=$ Chip in tray; $U / 7=$ chip with bumps on tape; $U / 10=$ chip on foil.

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## LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

| SYMBOL | PARAMETER | CONDITIONS | MIN. | MAX. | UNIT |
| :--- | :--- | :--- | :--- | :--- | :--- |
| $\mathrm{V}_{\mathrm{DD}}$ | supply voltage | $\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V}$; note 1 | -1.8 | +6 | V |
| $\mathrm{~V}_{\mathrm{I}}$ | all input voltages |  | $\mathrm{V}_{\mathrm{SS}}$ | $\mathrm{V}_{\mathrm{DD}}$ | V |
|  | output short-circuit duration |  | indefinite |  |  |
| $\mathrm{T}_{\text {amb }}$ | operating ambient temperature |  | -10 | +60 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {stg }}$ | storage temperature |  | -30 | +100 | ${ }^{\circ} \mathrm{C}$ |

## Note

1. Connecting the battery with reversed polarity does not destroy the circuit, but in this condition a large current flows, which will rapidly discharge the battery.

## HANDLING

Inputs and outputs are protected against electrostatic discharges in normal handling. However, to be totally safe, it is advisable to take handling precautions appropriate to handling MOS devices. Advice can be found in "Data Handbook IC16, General, Handling MOS Devices".

CHARACTERISTICS
$\mathrm{V}_{\mathrm{DD}}=1.55 \mathrm{~V} ; \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{f}_{\mathrm{osc}}=32.768 \mathrm{kHz} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}$; crystal: $\mathrm{R}_{\mathrm{S}}=20 \mathrm{k} \Omega ; \mathrm{C}_{1}=2$ to $3 \mathrm{fF} ; \mathrm{C}_{\mathrm{L}}=8$ to 10 pF ; $\mathrm{C}_{0}=1$ to 3 pF ; unless otherwise specified.
Immunity against parasitic impedance $=20 \mathrm{M} \Omega$ between adjacent pins.

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{DD} 1}$ | supply voltage | $\mathrm{T}_{\text {amb }}=-10$ to $+60^{\circ} \mathrm{C}$ | 1.2 | 1.55 | 2.5 | V |
| $\Delta \mathrm{V}_{\mathrm{DD}}$ | supply voltage variation | transient within 1.2 V and 2.5 V | - | - | 0.25 | V |
| $\mathrm{V}_{\mathrm{DD} 2}$ | supply voltage | programming | 5.0 | 5.1 | 5.2 | V |
| $\Delta \mathrm{V}_{\text {DDP }}$ | supply voltage pulse variation | programming | 0.55 | 0.6 | 0.65 | V |
| $\mathrm{I}_{\mathrm{DD} 1}$ | supply current | between motor pulses | - | 170 | 260 | nA |
| $\mathrm{I}_{\text {DD2 }}$ | supply current | $\mathrm{V}_{\mathrm{DD}}=2.1 \mathrm{~V}$ | - | 190 | 300 | nA |
| $\mathrm{I}_{\mathrm{DD} 3}$ | supply current | stop mode; pin 8 connected to $V_{D D}$ | - | 180 | 280 | nA |
| $\mathrm{I}_{\text {D } 4}$ | supply current | $\mathrm{V}_{\mathrm{DD}}=2.1 \mathrm{~V}$ | - | 220 | 360 | nA |
| IDD5 | supply current | $\mathrm{T}_{\text {amb }}=-10$ to $+60^{\circ} \mathrm{C}$ | - | - | 600 | nA |
| Motor output |  |  |  |  |  |  |
| $\mathrm{V}_{\text {sat }}$ | saturation voltage $\Sigma(\mathrm{P}+\mathrm{N})$ | $\mathrm{R}_{\mathrm{M}}=2 \mathrm{k} \Omega ; \mathrm{T}_{\mathrm{amb}}=-10$ to $+60^{\circ} \mathrm{C}$ | - | 150 | 200 | mV |
| $\mathrm{Z}_{\mathrm{o} \text { (sc) }}$ | output short-circuit impedance | between motor pulses $I_{\text {transistor }}<1 \mathrm{~mA}$ | - | 200 | 300 | $\Omega$ |

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| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Oscillator |  |  |  |  |  |  |
| $V_{\text {OSC ST }}$ | starting voltage |  | 1.2 | - | - | V |
| $\mathrm{gm}_{\mathrm{m}}$ | transconductance | $\mathrm{V}_{\mathrm{i}(\mathrm{p}-\mathrm{p})} \leq 50 \mathrm{mV}$ | 6 | 15 | - | $\mu \mathrm{S}$ |
| $\mathrm{t}_{\text {osc }}$ | start-up time |  | - | 1 | - | S |
| $\Delta \mathrm{f} / \mathrm{f}$ | frequency stability | $\Delta \mathrm{V}_{\mathrm{DD}}=100 \mathrm{mV}$ | - | $0.05 \times 10^{-6}$ | $0.3 \times 10^{-6}$ |  |
| $\mathrm{C}_{i}$ | input capacitance |  | 8 | 10 | 12 | pF |
| $\mathrm{C}_{0}$ | output capacitance |  | 12 | 15 | 18 | pF |
| Voltage level detector |  |  |  |  |  |  |
| $\mathrm{V}_{\text {LIT }}$ | threshold voltage |  | 1.62 | 1.80 | 1.98 | V |
| $\mathrm{V}_{\text {EOL }}$ | threshold voltage |  | 1.30 | 1.38 | 1.46 | V |
| $\Delta V_{\text {EOL }}$ | hysteresis of threshold |  | - | 10 | - | mV |
| TC ${ }_{\text {EOL }}$ | temperature coefficient |  | - | -1 | - | $\mathrm{mV} / \mathrm{K}$ |
| Reset input |  |  |  |  |  |  |
| $\mathrm{f}_{0}$ | output frequency |  | - | 32 | - | Hz |
| $\Delta \mathrm{V}_{0}$ | output voltage swing | $\mathrm{R}=1 \mathrm{M} \Omega ; \mathrm{C}=10 \mathrm{pF}$ | 1.4 | - | - | V |
| $\mathrm{t}_{\text {edge }}$ | edge time | $\mathrm{R}=1 \mathrm{M} \Omega$; $\mathrm{C}=10 \mathrm{pF}$ | - | 1 | - | $\mu \mathrm{s}$ |
| $\mathrm{l}_{\mathrm{im}}$ | peak input current | note 1 | - | 320 | - | nA |
| $\mathrm{I}_{\mathrm{i}}(\mathrm{av})$ | average input current |  | - | 10 | - | nA |

## Note

1. Duty factor is $1: 32$ and $\operatorname{RESET}=\mathrm{V}_{\mathrm{DD}}$ or $\mathrm{V}_{\mathrm{SS}}$.

## 32 kHz watch circuits with adaptive motor pulse

TIMING PARAMETERS

| SYMBOL | PARAMETER | SECTION | VALUE | OPTION | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{t}_{\mathrm{T}}$ | cycle for motor pulse (note 1) | motor pulse (Figs 2, 3 and 4) | 1 | 5, 10, 12 or 20 | s |
| $\mathrm{t}_{\mathrm{p}}$ | motor pulse width |  | 7.81 | 3.9 or 5.9 | ms |
| $\mathrm{t}_{\mathrm{DF}}$ | duty factor |  | 977 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{ONL}}$ | last duty factor on |  | 61 to 305 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{v}}$ | voltage detection cycle | level mode | 60 | - | s |
| $\mathrm{t}_{\text {SON }}$ | duty factor on | silver-oxide mode (Fig.3) | 550 to 794 | - | $\mu \mathrm{s}$ |
| tsoff | duty factor off |  | 427 to 183 | - | $\mu \mathrm{S}$ |
| $\mathrm{t}_{\text {SONF }}$ | first duty factor on |  | 488 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {AOFF }}$ | additional duty factor off | lithium mode (Fig.4) | 183 | - | $\mu \mathrm{s}$ |
| tLon | duty factor on |  | 305 to 611 | - | $\mu \mathrm{s}$ |
| tLOFF | duty factor off |  | 672 to 366 | - | $\mu \mathrm{s}$ |
| tLONF | first duty factor on |  | 244 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{E}}$ | EOL sequence | end-of-life mode | 4 | - | s |
| $\mathrm{t}_{\mathrm{E} 1}$ | motor pulse width |  | $t_{p}$ | - | ms |
| $\mathrm{t}_{\mathrm{E} 2}$ | time between pulses |  | 31.25 | - | ms |
| $\mathrm{t}_{\mathrm{D}}$ | detection sequence | detection (Fig.7) | 4.3 to 28.3 | - | ms |
| $\mathrm{t}_{\mathrm{DS}}$ | short-circuited motor |  | 977 | - | $\mu \mathrm{S}$ |
| $\mathrm{t}_{\mathrm{DI}}$ | dissipation of energy |  | 977 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\text {MC }}$ | measurement cycle |  | 488 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{M} 1}$ | phase 1 |  | 244 | - | $\mu \mathrm{S}$ |
| $\mathrm{t}_{\mathrm{M} 2}$ | phase 2 (measure window) |  | 61 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{M} 3}$ | phase 3 |  | 122 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{M} 4}$ | phase 4 |  | 61 | - | $\mu \mathrm{S}$ |
| P | positive current polarities |  | 1 | $\mathrm{P}<\mathrm{N}$ |  |
| N | negative current polarities |  | 2 | 2 to 6 |  |
| $\mathrm{t}_{\mathrm{C}}$ | correction sequence | correction sequence (Fig.8) | $\mathrm{tp}_{\mathrm{p}}+30.27$ | - | ms |
| $\mathrm{t}_{\mathrm{C} 1}$ | small pulse width |  | 977 | - | $\mu \mathrm{s}$ |
| $\mathrm{t}_{\mathrm{C} 2}$ | large pulse width |  | $\mathrm{t}_{\mathrm{p}}$ | - | ms |
| $\begin{aligned} & \mathrm{t}_{\mathrm{T} 1} \\ & \mathrm{t}_{\mathrm{T} 2} \\ & \mathrm{t}_{\mathrm{T} 3} \end{aligned}$ | cycles for motor-pulses in: <br> test 1 <br> test 2 <br> test 3 | testing <br> Fig. 10 | $\begin{array}{\|l\|} \hline 125 \\ 31.25 \\ 31.25 \text { to } 39 \end{array}$ | - | ms <br> ms <br> ms |
| $\mathrm{t}_{\text {DEB }}$ | debounce time for RESET $=V_{D D}$ |  | 14.7 to 123.2 | - | ms |

## Note

1. No option available when EOL indication is required.

## 32 kHz watch circuits with adaptive motor pulse

## CHIP DIMENSIONS AND BONDING PAD LOCATIONS



Chip area: $2.91 \mathrm{~mm}^{2}$.
Bonding pad dimensions: $110 \mu \mathrm{~m} \times 110 \mu \mathrm{~m}$.
Chip thickness: $200 \pm 25 \mu \mathrm{~m}$, with bumps: $270 \pm 25 \mu \mathrm{~m}$.
Fig. 11 Bonding pad locations, PCA146xU series; 8 terminals.

Table 3 Bonding pad locations (dimensions in $\mu \mathrm{m}$ )
All $x / y$ coordinates are referenced to bottom left pad ( $V_{D D}$ ), see Fig.11.

| PAD | $\mathbf{x}$ | $\mathbf{y}$ |
| :--- | :---: | :---: |
| $\mathrm{V}_{\text {SS }}$ | 1290 | 1100 |
| TEST | 940 | 1100 |
| OSC IN | 481 | 1100 |
| OSC OUT | -102 | 1100 |
| V $_{\text {DD }}$ | 0 | 0 |
| M1 | 578 | 0 |
| M2 | 930 | 0 |
| RESET | 1290 | 0 |
| chip corner (max. value) | -497.5 | -170 |

## 32 kHz watch circuits with adaptive motor pulse

## APPLICATION INFORMATION


(1) Quartz crystal case should be connected to $V_{D D}$. Stray capacitance and leakage resistance from RESET, M1 or M2 to OSC IN should be less than 0.5 pF or larger than $20 \mathrm{M} \Omega$.
(2) Motor, probe and stray capacitance from $M 2$ or $M 1$ to $V_{D D}$ or $V_{S S}$ should be less than $C_{M}=80 \mathrm{pF}$ for correct operation of the detection circuit. Driving the motor at its minimum energy, probe and stray capacitance must be avoided.

Fig. 12 Typical application circuit diagram.

## 32 kHz watch circuits with adaptive motor pulse

PCA146x series

## PACKAGE OUTLINE

PMFP8: plastic micro flat package; 8 leads (straight)
SOT144-1


DIMENSIONS (mm are the original dimensions)

| UNIT | $\mathbf{A}_{\mathbf{2}}$ | $\mathbf{b}$ | $\mathbf{c}$ | $\mathbf{D}^{(\mathbf{1})}$ | $\mathbf{E}^{(\mathbf{1})}$ | $\mathbf{e}$ | $\mathbf{H}_{\mathbf{E}}$ | $\mathbf{L}$ | $\mathbf{m}$ <br> $\mathbf{m a x}$. | $\mathbf{n}$ <br> $\mathbf{m a x}$. | $\mathbf{Q}_{\mathbf{1}}$ | $\mathbf{Q}_{\mathbf{2}}$ | $\mathbf{t}$ | $\mathbf{w}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| mm | 0.90 | 0.40 | 0.19 | 3.1 | 3.1 | 0.80 | 4.6 | 0.75 | 0.26 | 0.3 | 0.40 | 0.40 | 0.95 | 0.1 |

Note

1. Plastic or metal protrusions of 0.15 mm maximum per side are not included.

| OUTLINE <br> VERSION | REFERENCES |  |  |  | EUROPEAN <br> PROJECTION | ISSUE DATE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | IEC | JEDEC | EIAJ |  |  |  |

# 32 kHz watch circuits with adaptive motor pulse 

## SOLDERING

## Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "IC Package Databook" (order code 9398652 90011).

## Reflow soldering

Reflow soldering techniques are suitable for all SO packages.
Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.
Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to $250^{\circ} \mathrm{C}$.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at $45^{\circ} \mathrm{C}$.

## Wave soldering

Wave soldering techniques can be used for all SO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is $260^{\circ} \mathrm{C}$, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than $150^{\circ} \mathrm{C}$ within 6 seconds. Typical dwell time is 4 seconds at $250^{\circ} \mathrm{C}$.
A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

## Repairing soldered joints

Fix the component by first soldering two diagonallyopposite end leads. Use only a low voltage soldering iron (less than 24 V ) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to $300^{\circ} \mathrm{C}$. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and $320^{\circ} \mathrm{C}$.

# 32 kHz watch circuits with adaptive motor pulse 

## DEFINITIONS

| Data sheet status |  |
| :--- | :--- |
| Objective specification | This data sheet contains target or goal specifications for product development. |
| Preliminary specification | This data sheet contains preliminary data; supplementary data may be published later. |
| Product specification | This data sheet contains final product specifications. |
| Limiting values |  |
| Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or <br> more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation <br> of the device at these or at any other conditions above those given in the Characteristics sections of the specification <br> is not implied. Exposure to limiting values for extended periods may affect device reliability. |  |
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## 32 kHz watch circuits with adaptive motor pulse <br> PCA146x series

## NOTES

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PCA146x series

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