# **PN5331B3HN**

## **Near Field Communication (NFC) controller**

Rev. 3.4 — 29 November 2017 157534

Product data sheet COMPANY PUBLIC

## 1. General description

The PN533 is a highly integrated transceiver module for contactless communication at 13.56 MHz based on the 80C51 microcontroller core. It supports 6 different operating modes:

- ISO/IEC 14443A/MIFARE Reader/Writer
- FeliCa Reader/Writer
- ISO/IEC 14443B Reader/Writer
- ISO/IEC 14443A/MIFARE Card MIFARE 1 KB or MIFARE 4 KB emulation
- FeliCa Card emulation
- ISO/IEC 18092, ECMA 340 Peer-to-Peer

The PN533 implements a demodulator and decoder for signals from ISO/IEC 14443A/MIFARE compatible cards and transponders. The PN533 handles the complete ISO/IEC 14443A framing and error detection (Parity & CRC).

The PN533 supports MIFARE 1 KB or MIFARE 4 KB emulation products. The PN533 supports contactless communication using MIFARE Higher transfer speeds up to 424 kbit/s in both directions.

The PN533 can demodulate and decode FeliCa coded signals. The PN533 handles the FeliCa framing and error detection. The PN533 supports contactless communication using FeliCa Higher transfer speeds up to 424 kbit/s in both directions.

The PN533 supports layers 2 and 3 of the ISO/IEC 14443 B Reader/Writer communication scheme, except anticollision. This must be implemented in firmware as well as upper layers.

In card emulation mode, the PN533 is able to answer to a Reader/Writer command either according to the FeliCa or ISO/IEC 14443A/MIFARE card interface scheme. The PN533 generates the load modulation signals, either from its transmitter or from the LOADMOD pin driving an external active circuit. A complete secure card functionality is only possible in combination with a secure IC using the NFC-WI/S<sup>2</sup>C interface.

Compliant to ECMA 340 and ISO/IEC 18092 NFCIP-1 Passive and Active communication modes, the PN533 offers the possibility to communicate to another NFCIP-1 compliant device, at transfer speeds up to 424 kbit/s. The PN533 handles the complete NFCIP-1 framing and error detection.

The PN533 transceiver can be connected to an external antenna for Reader/Writer or Card/PICC modes, without any additional active component.



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## **Near Field Communication (NFC) controller**

The PN533 supports the following host interfaces:

- USB 2.0 full speed interface (bus powered or non bus powered)
- High Speed UART (HSU)

PN533 has also a master I<sup>2</sup>C interface enabling the drive on an I<sup>2</sup>C peripheral (i.e. memory).

In addition, a power switch is included to supply power to a secure IC.

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## **Near Field Communication (NFC) controller**

## 2. Features and benefits

- 80C51 microcontroller core with 45056 bytes ROM and 1224 bytes RAM
- Highly integrated demodulator and decoder
- Buffered output drivers to connect an antenna with minimum number of external components
- Integrated RF level detector
- Integrated data mode detector
- Supports ISO/IEC 14443A/MIFARE
- Supports ISO/IEC 14443B (Reader/Writer mode only)
- Typical operating distance in Reader/Writer mode for communication to ISO/IEC 14443A/MIFARE, ISO/IEC 14443B or FeliCa cards up to 50 mm depending on antenna size and tuning
- Typical operating distance in NFCIP-1 mode up to 50 mm depending on antenna size, tuning and power supply
- Typical operating distance in ISO/IEC 14443A/MIFARE or FeliCa card emulation mode of approximately 100 mm depending on antenna size, tuning and external field strength
- Supports MIFARE 1 KB or MIFARE 4 KB emulation encryption in Reader/Writer mode and MIFARE higher transfer speed communication at 212 kbit/s and 424 kbit/s
- Supports contactless communication according to the FeliCa protocol at 212 kbit/s and 424 kbit/s
- Integrated RF interface for NFCIP-1 up to 424 kbit/s
- Possibility to communicate on the RF interface above 424 kbit/s using external analog components
- Supported host interfaces
  - ◆ USB 2.0 full speed interface
  - High-speed UART
- Restricted I<sup>2</sup>C master interface to control an external I2C EEPROM
- Dedicated host interrupts
- Low power modes
  - Hard-Power-down mode
  - Soft-Power-down mode
- Automatic wake-up on HSU interfaces when device is in Power-down mode
- Programmable timers
- 27.12 MHz Crystal oscillator
- On-Chip PLL to generate internally The 96 MHz for the USB interface
- Power modes
  - USB bus power mode
  - 2.5 V to 3.6 V power supply operating range in non USB bus power mode
- Power switch for external secure companion chip
- Dedicated IO ports for external device control
- Integrated antenna detector for production tests
- ECMA 373 NFC-WI interface to connect an external secure IC

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## 3. Applications

- Computing application
- Consumer applications

## 4. Quick reference data

Table 1. Quick reference data

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{BUS}$	USB Supply Voltage (USB mode)			4.02	5	5.25	V
	Supply Voltage (non USB mode)	$V_{BUS} = D_{VDD}V_{SS} = 0 V$		2.5	3.3	3.6	V
$T_{VDD}$ , $A_{VDD}$ , $D_{VDD}$	Supply Voltage	$T_{VDD} = A_{VDD} = D_{VDD}$ $V_{SS} = 0 \text{ V}$	[1]	2.5	3.3	3.6	V
$P_{VDD}$	Supply Voltage for host interface	$V_{SS} = 0 V$		1.6		3.6	V
S <sub>VDD</sub>	Supply Voltage for SAM interface	$V_{SS} = 0 V$ (SV <sub>DD</sub> Switch Enabled)		DV <sub>DD</sub> -0.1		$DV_DD$	V
I <sub>VBUS</sub>	Maximum load current (USB mode)	measured on V <sub>BUS</sub>				150	mA
	Maximum Inrush current limitation	At power up(curlimoff = 0)				100	mA
I <sub>HPD</sub>	Hard Power Down Current (Not powered from USB)	$\begin{aligned} &A_{VDD} = D_{VDD} = T_{VDD} = P_{VDD} \\ &= 3 \text{ V, RF level detector off} \end{aligned}$				10	μΑ
I <sub>SPD</sub>	Soft Power down Current(Not powered from USB)	$A_{VDD} = D_{VDD} = T_{VDD} = P_{VDD}$ = 3 V, RF level detector on				30	μΑ
I <sub>suspend</sub>	USB suspend Current	$A_{VDD} = D_{VDD} = T_{VDD} = P_{VDD}$ = 3 V, RF level detector on (without resistor on DP/DM)				250	μΑ
I <sub>DVDD</sub>	Digital Supply Current	$\begin{aligned} &A_{VDD} = D_{VDD} = T_{VDD} = P_{VDD} \\ &= 3 \text{ V, RF level detector on,} \\ &S_{VDD} \text{ switch off} \end{aligned}$	[1]		15		mA
I <sub>SVDD</sub>	S <sub>VDD</sub> Supply Current	$S_{VDD} = 3 \text{ V}, S_{VDD} \text{ switch On}$				30	mA
I <sub>AVDD</sub>	Analog Supply Current	$\begin{aligned} A_{VDD} &= D_{VDD} = T_{VDD} = P_{VDD} \\ &= 3 \text{ V, RF level detector on} \end{aligned}$			6		mA
I <sub>TVDD</sub>	Transmitter Supply Current	During RF Transmission, TVDD=3V			60	100	mA
P <sub>tot</sub>	continuous total power dissipation.	$T_{amb}$ = -30 to +85 °C				0.55	W
T <sub>amb</sub>	operating ambient temperature			-30		+85	°C

<sup>[1]</sup>  $D_{VDD}$ ,  $A_{VDD}$  and  $T_{VDD}$  shall always be at the same supply voltage.

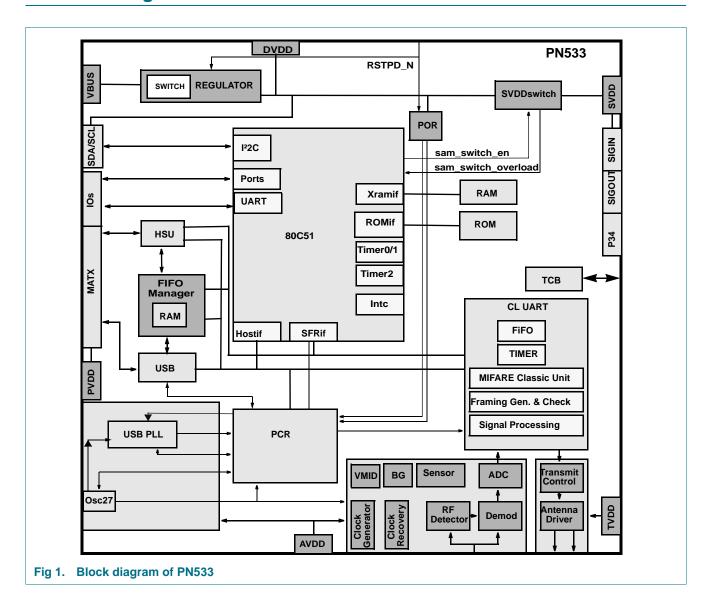
## 5. Ordering information

Table 2. Ordering information

Type number	Package					
	Name	Description	Version			
PN5331B3HN/C270[1]	HVQFN40	Heatsink Very thin Quad Flat package; 40 pins, plastic, body 6 x 6 x 0.85 mm; leadless; MSL level 2.	SOT618-1			

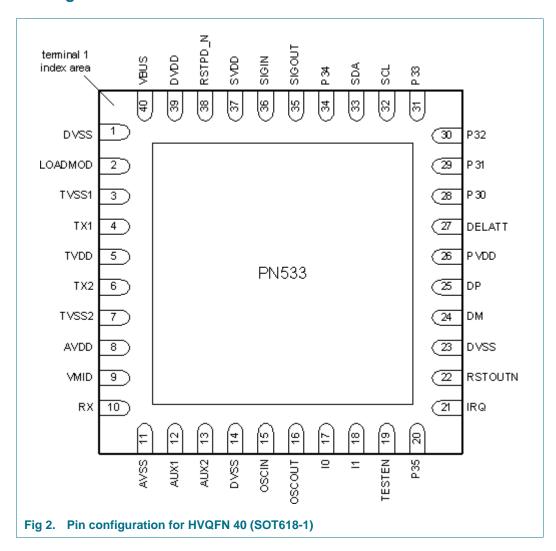
<sup>[1]</sup> Refer to Section 17.4 "Licenses"

## 6. Block diagram



## 7. Pinning information

## 7.1 Pinning



## 7.2 Pin description

Table 3. PN533 Pin description

Symbol	Pin	Type	Pad Ref Voltage	Description
DVSS	1	PWR		Digital ground
LOADMOD	2	0	DVDD	Load Modulation output provides digital signal for FeliCa and MIFARE card operating mode
TVSS1	3	PWR		Transmitter ground: supplies the output stage of TX1 and TX2
TX1	4	0	TVDD	Transmitter 1: transmits modulated 13.56 MHz energy carrier
TVDD	5	PWR		Transmitter power supply: supplies the output stage of TX1 and TX2
TX2	6	0	TVDD	Transmitter2: delivers the modulated 13.56 MHz energy carrier
TVSS2	7	PWR		Transmitter ground: supplies the output stage of TX1 and TX2
AVDD	8	PWR		Analog power supply
VMID	9	PWR	AVDD	Internal reference voltage: This pin delivers the internal reference voltage.
RX	10	I	AVDD	Receiver Input: Input pin for the reception signal, which is the load modulated 13.56 MHZ energy carrier from the antenna circuit
AVSS	11	PWR		Analog ground
AUX1	12	0	DVDD	Auxiliary output 1: This pin delivers analog and digital test signals
AUX2	13	0	DVDD	Auxiliary output 2: This pin delivers analog and digital test signals
DVSS	14	PWR		Digital Ground
OSCIN	15	I	AVDD	Crystal Oscillator Input: input to the inverting amplifier of the oscillator. This pin is also the input for an externally generated clock (fosc = 27.12 MHZ).
OSCOUT	16	0	AVDD	Crystal Oscillator output: Output of the inverting amplifier of the oscillator.
10	17	I	DVDD	Interface mode lines: selects the used host interface (refer to Table 75
I1	18	I	DVDD	"Config I0_I1 register (address 6103h) bit allocation" for details). In test mode I0 is used as test signals.
TESTEN	19	I	DVDD	Test enable pin:
				When set to 1 enable the test mode.
				When set to 0 reset the TCB and disable the access to the test mode.
P35	20	Ю	DVDD	General purpose IO signal
P70_IRQ	21	Ю	PVDD	Interrupt request: Output to signal an interrupt event to the host (Port 7 bit 0)
RSTOUT	22	0	PVDD	Output reset signal. When Low it indicates that the circuit is in reset state.
DVSS	23	PWR		Digital Ground
DM	24	Ю	PVDD	USB D- data line in USB mode or TX in HSU mode (refer to <u>Table 74 "HOST</u> interface selection" on page 47 for details).
				In test mode this signal is used as input and output test signal
DP	25	Ю	PVDD	USB D+ data line in USB mode or RX in HSU mode (refer to <u>Table 74 "HOST</u> interface selection" on page 47 for details).
				In test mode this signal is used as input and output test signal.
PVDD	26	PWR		IO pad power supply
DELATT	27	0	PVDD	Optional output for an external 1.5 KOhms resistor connection on D+.
P30	28	Ю	PVDD	General purpose IO signal. Can be configured to act either as RX line of the second serial interface UART or general purpose IO.
				In test mode this signal is used as input and output test signal.

Table 3. PN533 Pin description ... continued

Symbol	Pin	Туре	Pad Ref Voltage	Description
P31	29	Ю	PVDD	General purpose IO signal. Can be configured to act either as TX line of the second serial interface UART or general purpose IO.
				In test mode this signal is used as input and output test signal.
P32_INT0	30	Ю	PVDD	General purpose IO signal. Can also be used as an interrupt source In test mode this signal is used as input and output test signal.
P33_INT1	31	Ю	PVDD	General purpose IO signal. Can be used to generate an HZ state on the output of the selected interface for the Host communication and to enter into power down mode without resetting the internal state of PN533.
				In test mode this signal is used as input and output test signal.
SCL	32	Ю	DVDD	I <sup>2</sup> C clock line - open drain in output mode
SDA	33	Ю	DVDD	I <sup>2</sup> C data line - open drain in output mode
P34	34	Ю	SVDD	General purpose IO signal or clk signal for the SAM
SIGOUT	35	0	SVDD	Contactless communication interface output: delivers a serial data stream according to NFCIP-1 and output signal for the SAM.
				In test mode this signal is used as test signal output.
SIGIN	36	I	SVDD	Contactless communication interface input: accepts a digital, serial data stream according to NFCIP-1 and input signal from the SAM.
				In test mode this signal is used as test signal input.
SVDD	37	PWR		Output power for SAM power supply. Switched on by Firmware with an overload detection. Used as a reference voltage for SAM communication.
RSTPD_N	38	I	PVDD	Reset and Power Down: When LOW, internal current sources are switched off, the oscillator is inhibited, and the input pads are disconnected from the outside world.
				With a negative edge on this pin the internal reset phase starts.
DVDD	39	PWR		Digital Power Supply
VBUS	40	PWR		USB power supply.
-				

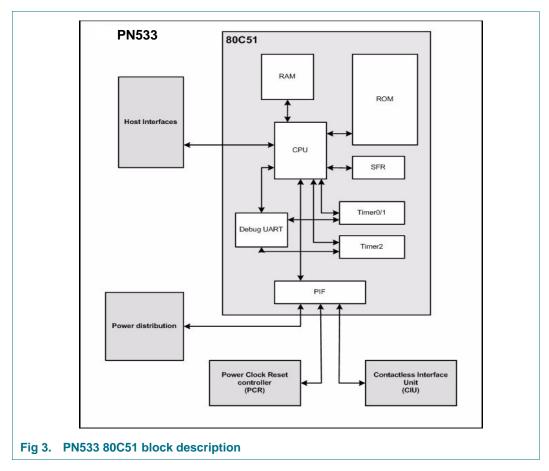
## 8. Functional description

## 8.1 80C51

The PN533 is controlled via an embedded 80C51 microcontroller core (for more details http://www.standardics.nxp.com/support/documents/microcontrollers/?scope=80C51). Its principle features are listed below:

- 6-clock cycle CPU. One machine cycle comprises 6 clock cycles or states (S1 to S6).
   An instruction needs at least one machine cycle.
- ROM interface
- RAM interface to embedded IDATA and XRAM memories (see Figure 4 on page 10)
- · Peripheral interface (PIF)
- Power control module to manage the CPU power consumption
- Clock module to control CPU clock during Shutdown and Wake-up modes
- Port module interface to configure I/O pads
- Interrupt controller
- Three timers
- Debug UART

The block diagram describes the main blocks described in this 80C51 section.



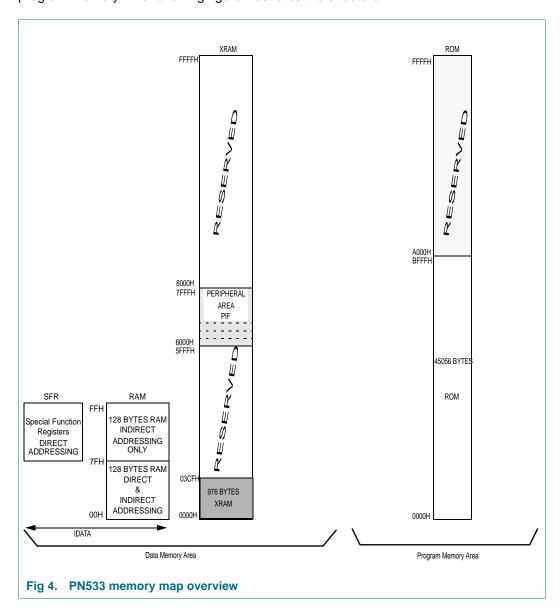
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## 8.1.1 **PN533** memory map

The memory map of PN533 is composed of 2 main memory spaces: data memory and program memory. The following figure illustrates the structure.



## 8.1.2 Data memory

Data memory is itself divided into 2 spaces:

- 384-byte IDATA with byte-wide addressing
  - 258-byte RAM
  - 128-byte SFR
- 1 bank of 64 KB extended RAM (XRAM) with 2-byte-wide addressing

#### 8.1.2.1 IDATA memory

The IDATA memory is mapped into 3 blocks, which are referred as Lower IDATA RAM, Upper IDATA RAM, and SFR. Addresses to these blocks are byte-wide, which implies an address space of only 256 bytes. However, 384 bytes can be addressed within IDATA memory through the use of direct and indirect address mechanisms.

- Direct addressing: the operand is specified by an 8-bit address field in the instruction.
- Indirect addressing: the instruction specifies a register where the address of the operand is stored.

For the range 80h to FFh, direct addressing will access the SFR space; indirect addressing accesses Upper IDATA RAM. For the range 00h to 7Fh, Lower IDATA RAM is accessed, regardless of addressing mode. This behavior is summarized in the table below:

Table 4. IDATA memory addressing

Address	Addressing mode					
	Direct	Indirect				
00h to 7Fh	Lower IDATA RAM	Lower IDATA RAM				
80h to FFh	SFRs	Upper IDATA RAM				

The SFRs and their addresses are described in the Table 5:

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Table 5. SFR map of NFC controller

0.00	Address	Bit- Byte-addressable									
Ē	Auuless	addressable		Dyte-audiessable							
				\/5.1.45	T.						
	F8h	IP1		XRAMP		P3CFGA	P3CFGB				
	F0h	В				P7CFGA	P7CFGB				
	E8h	IE1	CIU_Status2	CIU_FIFOData	CIU_FIFOLevel	CIU_WaterLevel	CIU_Control	CIU_BitFran			
	E0h	ACC									
	D8h	I <sup>2</sup> CC0N	I <sup>2</sup> CSTA	I <sup>2</sup> CDAT	I <sup>2</sup> CADR						
	D0h	PSW	CIU_Command	CIU_CommlEn	CIU_DivIEn	CIU_Commlrq	CIU_DivIrq	CIU_Erro			
	C8h	T2CON	T2MOD	RCAP2L	RCAP2H	T2L	T2H				
	C0h										
	B8h	IP0									
₽	B0h	P3									
inform	A8h	IE0			HSU_STA	HSU_CTR	HSU_PRE	HSU_CN			
ation .	A0h		FITEN	FDATA	FSIZE						
	98h	S0CON	SBUF	RWL	TWL	FIFOFS	FIFOFF	SFF			
5	90h										
<u>.</u>	88h	T01CON	T01MOD	TOL	T1L	T0H	T1H				
1	80h		SP	DPL	DPH						

#### 8.1.2.2 XRAM memory

The XRAM memory is divided into 2 memory spaces:

- 0000h to 5FFFh: reserved for addressing embedded RAM. For the PN533, only accesses between 0000h and 03C7h are valid.
- 6000h to 7FFFh: reserved for addressing embedded peripherals. This space is divided into 32 regions of 256 bytes each. Addressing can be performed using R0 or R1 and the XRAMP SFR.

The Table 6 depicts the mapping of internal peripherals into XRAM.

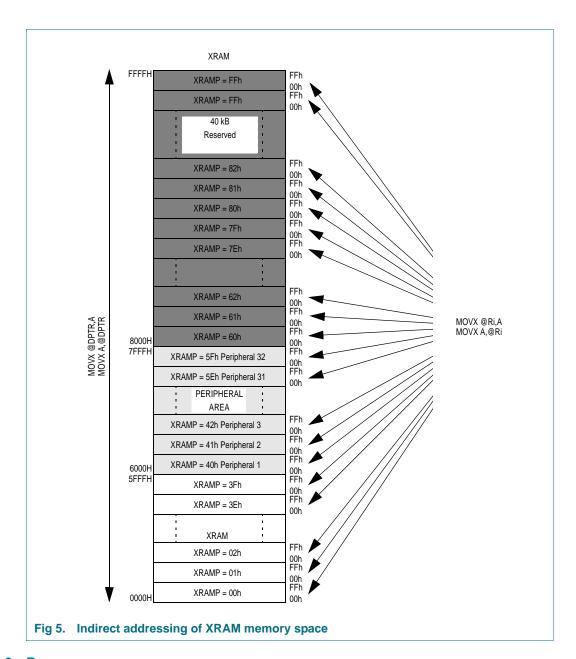
Table 6. Peripheral mapping into XRAM memory space

Base Address	End Address	Description
6000h	60FFh	Reserved.
6100h	61FFh	IOs and miscellaneous registers configuration Refer to Section 8.2 "General purpose IOs configurations" on page 37
6200h	62FFh	Power Clock and Reset controller Refer to Section 8.5.10 "PCR extension registers" on page 102
6300h	633Fh	Contactless Unit Interface Refer to Section 8.6 "Contactless Interface Unit (CIU)" on page 108
6340h	FFFFh	Reserved

XRAM is accessed via the dedicated MOVX instructions. There are two access modes:

- 16-bit data pointer (DPTR): the full XRAM address space can be accessed.
- paging mechanism: the upper address byte is stored in the SFR register XRAMP; the lower byte is stored in either R1 or R0.

The Figure 5 illustrates both mechanisms.



## 8.1.3 Program memory

PN533 program memory ranges from 0000h to AFFFh, which is physically mapped to the 44 KB ROM.

## 8.1.4 PCON module

The Power Control (PCON) module is configured using the PCON SFR register.

Table 7. PCON register (SFR: address 87h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	SMOD			-			CPU_PD	-
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	R	R	R	R/W	R/W

Table 8. Description of PCON bits

Bit	Symbol	Description
7	SMOD	Serial MODe: When set to logic level 1, the baud rate of the Debug UART is doubled
6 to 2	-	Reserved.
1	CPU_PD	Power-down: When set to logic level 1, the microcontroller goes in Power-down mode
0	Reserved	This bit should only ever contain logic level 0.

## 8.1.5 Interrupt Controller

The interrupt controller has the following features:

- 1interrupt source
- Interrupt enable registers IE0 and IE1
- Interrupt priority registers IP0 and IP1
- Wake-up from Power-down state

#### 8.1.5.1 Interrupt vectors

The mapping between interrupt sources and interrupt vectors is shown in Table 9.

Table 9. Interrupt vector

Interrupt number	Interrupt vector	Interrupt sources	Incremental priority level (conflict resolution level)
0	0003h	External P32_INT0	Highest
1	000Bh	Timer0 interrupt	
2	0013h	External P33_INT1	
3	001Bh	Timer1 interrupt	
4	0023h	Debug UART interrupt	
5	002Bh	Timer2 interrupt	
6	0033h	NFC-WI interrupt	
7	003Bh	Reserved	
8	0043h	Reserved	
9	004Bh	CIU interrupt 1	
10	0053h	CIU interrupt 0	
11	005Bh	I <sup>2</sup> C interrupt	
12	0063h	FIFO or HSU interrupts	
13	006Bh	USB interrupt	
14	0073h	General Purpose IRQ	Lowest

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## 8.1.5.2 Interrupt enable: IE0 and IE1 registers

Each interrupt source can be individually enabled or disabled by setting a bit in IE0 or IE1. In register IE0, a global interrupt enable bit can be set to logic level 0 to disable all interrupts at once.

The 2 following tables describe IE0.

Table 10. Interrupt controller IE0 register (SFR: address A8h) bit allocation

			_	-	_			
Bit	7	6	5	4	3	2	1	0
Symbol	IE0_7	IE0_6	IE0_5	IE0_4	IE0_3	IE0_2	IE0_1	IE0_0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 11. Description of IE0 bits

Bit	Symbol	Description
7	IE0_7	Global interrupt enable When set to logic 1, the interrupts can be enabled. When set to logic 0, all the interrupts are disabled.
6	IE0_6	NFC-WI counter interrupt enable When set to logic 1, NFC-WI interrupt is enabled.
5	IE0_5	Timer2 interrupt enable When set to logic 1, Timer2 interrupt is enabled.
4	IE0_4	Debug UART interrupt enable When set to logic 1, Debug UART interrupt is enabled.
3	IE0_3	Timer1 interrupt enable When set to logic 1, Timer1 interrupt is enabled
2	IE0_2	P33_INT1 interrupt enable When set to logic 1, P33_INT1 pin interrupt is enabled. The polarity of P33_INT1 can be inverted (see Table 75 on page 47).
1	IE0_1	Timer0 interrupt enable When set to logic 1, Timer0 interrupt is enabled.
0	IE0_0	P32_INT0 interrupt enable When set to logic 1, P32_INT0 pin interrupt is enabled.

The 2 following tables describe IE1.

Table 12. Interrupt controller IE1 register (SFR: address E8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	IE1_7	IE1_6	IE1_5	IE1_4	IE1_3	IE1_2	-	-
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 13. Description of IE1 bits

Bit	Symbol	Description
7	IE1_7	General purpose IRQ interrupt enable.
		When set to logic 1, enables interrupt function of P34, P35, P50_SCL and P71 according to their respective enable and level control bits. See <u>Table 19 on page 19</u> , <u>Table 138 on page 104</u> and <u>Table 144 on page 107</u> .
6	IE1_6	USB interrupt enable.
		When set to logic level 1, enables USB interrupt.
5	IE1_5	FIFO and HSU interrupt enable.
		When set to logic 1, enables FIFO interrupts, SPI interrupts, HSU interrupt.
4	IE1_4	I2C interrupt enable.
		When set to logic 1, enables I <sup>2</sup> C interrupt.
3	IE1_3	CIU interrupt 0 enable.
		When set to logic 1, enables CIU interrupt 0: CIU_IRQ_0.
2	IE1_2	CIU interrupt 1 enable.
		When set to logic 1, enables the CIU interrupt 1: CIU_IRQ_1.
1 to 0	-	Reserved. This bit must be set to logic level 0.

## 8.1.5.3 Interrupt prioritization: IP0 and IP1 registers

Each interrupt source can be individually programmed to be one of two priority levels by setting or clearing a bit in the interrupt priority registers IPO and IP1. If two interrupt requests of different priority levels are received simultaneously, the request with the high priority is serviced first. On the other hand, if the interrupts are of the same priority, precedence is resolved by comparing their respective conflict resolution levels (see

#### Table 9 on page 15

for details). The processing of a low priority interrupt can be interrupted by one with a high priority.

A RETI (Return From Interrupt) instruction jumps to the address immediately succeeding the point at which the interrupt was serviced. The instruction found at the return address will be executed, prior to servicing any pending interrupts.

The 2 following tables describe IP0.

Table 14. Interrupt controller IP0 register (SFR: address B8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	IP0_7	IP0_6	IP0_5	IP0_4	IP0_3	IP0_2	IP0_1	IP0_0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 15. Description of IP0 bits

Bit	Symbol	Description
7	IP0_7	Reserved
6	IP0_6	When set to logic level 1, NFC-WI interrupt is set to high priority.
5	IP0_5	When set to logic level 1, Timer2 interrupt is set to high priority.
4	IP0_4	When set to logic level 1, Debug UART interrupt is set to high priority.
3	IP0_3	When set to logic level 1, Timer1 interrupt is set to high priority.
2	IP0_2	When set to logic level 1, external P33_INT1 pin is set to high priority.
1	IP0_1	When set to logic level 1, Timer0 interrupt is set to high priority.
0	IP0_0	When set to logic level 1, external P32_INT0 pin is set to high priority.

The 2 following tables describe IP1.

Table 16. Interrupt controller IP1 register (SFR: address F8h) bit allocation

Bit	7	6	5	4	3	2	1 0
Symbol	IP1_7	IP1_6	IP1_5	IP1_4	IP1_3	IP1_2	-
Reset	0	0	0	0	0	0	00
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 17. Description of IP1 bits

Bit	Symbol	Description
7	IP1_7	When set to logic level 1, General Purpose IRQ interrupt is set to high priority.
6	IP1_6	When set to logic level 1, USB interrupt is set to high priority.
5	IP1_5	When set to logic level 1, combined FIFO and HSU interrupt is set to high priority.
4	IP1_4	When set to logic level 1, I <sup>2</sup> C interrupt is set to high priority.
3	IP1_3	When set to logic level 1, CIU interrupt 0 is set to high priority.
2	IP1_2	When set to logic level 1, CIU interrupt 1 is set to high priority.
1	-	Reserved. This bit must be set to logic level 0.
0	IP1_0	When set to logic level 1, interrupt number 7 is set to high priority.

## 8.1.5.4 General purpose IRQ control

The general purpose interrupts are controlled by register GPIRQ.

NOTE: this is not a standard feature of the 8051.

Table 18. GPIRQ register (address 6107h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	gpirq_ level_ P71	gpirq_ level_ DP	gpirq_ level_ P35	gpirq_ level_ P34	gpirq_ enable _P71	gpirq_ enable_ DP	gpirq_ enable_ P35	gpirq_ enable_ P34
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 19. Description of GPIRQ bits

Bit	Symbol	Description
7	-	2000.p.no
6	gpirq_level_DP	Configures the polarity of signal on P50 to generate a GPIRQ interrupt event (assuming gpirq_enable_P50 is set).
		When set to logic 0, an interrupt will be generated if P50_SCL is at logic 0.
		When set to logic 1, an interrupt will be generated if P50_SCL is at logic 1.
5	gpirq_level_P35	Configures the polarity of signal on P35 to generate a GPIRQ interrupt event (assuming gpirq_enable_P35 is set).
		When set to logic 0, an interrupt will be generated if P35 is at logic 0.
		When set to logic 1, an interrupt will be generated if P35 is at logic 1.
4	gpirq_level_P34	Configures the polarity of signal on P34 to generate a GPIRQ interrupt event (assuming gpirq_enable_P34 is set).
		When set to logic 0, an interrupt will be generated if P34 is at logic 0.
		When set to logic 1, an interrupt will be generated if P34 is at logic 1.
		<b>Remark:</b> If hide_svdd_sig of the register control_rngpower is set and gpirq_enable_P34 is also set then this bit will be asserted independently of the level on the pad P34.
3	-	
2	gpirq_enable_DP	When set to logic 1, enables pad DP to generate a GPIRQ interrupt event.
1	gpirq_enable_P35	When set to logic 1, enables pad P35 to generate a GPIRQ interrupt event.
0	gpirq_enable_P34	When set to logic 1, enables pad P34 to generate a GPIRQ interrupt event.

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#### **Near Field Communication (NFC) controller**

## 8.1.6 Timer0/1 description

Timer0/1 are general purpose timer/counters. Timer0/1 has the following functionality:

- Configurable edge or level detection interrupts
- Timer or counter operation
- 4 timer/counter modes
- · Baud rate generation for Debug UART

Timer0/1 comprises two 16-bit timer/counters: Timer0 and Timer1. Both can be configured as either a timer or an event counter.

Each of the timers can operate in one of four modes:

- Mode 0: 13-bit timer/counter
- Mode 1: 16-bit timer/counter
- Mode 2: 8-bit timer/counter with programmable preload value
- Mode 3: two individual 8-bit timer/counters (Timer0 only)

In the 'timer' function, the timer/counter is incremented every machine cycle. The count rate is 1/6 of the CPU clock frequency (CPU\_CLK).

In the 'counter' function, the timer/counter is incremented in response to a 1-to-0 transition on the input pins P34 / SIC\_CLK (Timer0) or P35 (Timer1). In this mode, the external input is sampled during state S5 of every machine cycle. If the associated pin is at logic level 1 for a machine cycle, followed by logic level 0 on the next machine cycle, the count is incremented. The new count value appears in the timer/counter in state S3 of the machine cycle following the one in which the transition was detected. The maximum count rate is 1/12 of the CPU\_CLK frequency. There are no restrictions on the duty cycle of the external input signal but to ensure that a given level is sampled at least once before it changes, it should be held for at least one full machine cycle.

The overflow output 't1\_ovf' of Timer1 can be used as a baud rate generator for the Debug UART. The Timer1 interrupt should be disabled in this case. For most applications which drive the Debug UART, Timer1 is configured for 'timer' operation and in auto-reload mode.

## 8.1.6.1 Timer0/1 registers

The Timer0/1 module contains six Special Function Registers (SFRs) which can be accessed by the CPU.

Table 20. Timer0/1 Special Function registers list

Name	Size [bytes]	Address Offset	Description	Access
T01CON	1	88h	Timer0/1 control register	R/W
T01MOD	1	89h	Timer0/1 mode register	R/W
TOL	1	8Ah	Timer0 timer/counter lower byte	R/W
T1L	1	8Bh	Timer1 timer/counter lower byte	R/W
T0H	1	8Ch	Timer0 timer/counter upper byte	R/W
T1H	1	8Dh	Timer1 timer/counter upper byte	R/W

The firmware performs a register read in state S5 and a register write in state S6. The hardware loads bits TF0 and TF1 of the register T01CON during state S2 and state S4 respectively. The hardware loads bits IE0 and IE1 of the register T01CON during state S1 and reset these bits during state S2. The registers T0L, T0H, T1L, T1H are updated by the hardware during states S1, S2, S3 and S4 respectively. At the end of a machine cycle, the firmware load has overrided the hardware load as the firmware writes in state S6.

Table 21. Timer0/1 SFR registers CPU state access

	CPU STATE									
Register	Bit	S1	S2	S3	S4	S5	S6			
T01CON	TF0		HW read			SW read	SW write			
	TF1				HW read	SW read	SW write			
	IE0/IE1	HW write	HW reset			SW read	SW write			
TOL		HW write				SW read	SW write			
TOH			HW write			SW read	SW write			
T1L				HW write		SW read	SW write			
T1H					HW write	SW read	SW write			

## 8.1.6.2 T01CON register

The register is used to control Timer0/1 and report its status.

Table 22. Timer0/1 T01CON register (SFR address 88h), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 23. Description of Timer0/1 T01CON register bits

Bit	Symbol	Description
7	TF1	Timer1 overflow.
		Set to logic level 1 by hardware on a Timer1 overflow. The flag is set to logic level 0 by the CPU after 2 machine cycles.
6	TR1	Timer1 run control.
		Set by firmware only. When set to logic level 1, Timer1 is enabled.
5	TF0	Timer0 overflow.
		Set by hardware on a Timer0 overflow. The flag is set to logic level 0 by the CPU after 2 machine cycles.
4	TR0	Timer0 run control.
		Set by firmware only. When set to logic level 1, Timer0 is enabled.
3	IE1	External Interrupt1 event.
		Set to logic level 1 by hardware when an external interrupt is detected on P33_INT1.
2	IT1	External Interrupt1 control.
		Set by firmware only. When set to logic level 1, Interrupt1 triggers on a falling edge of P33_INT1.
		When set to logic level 0, Interrupt1 triggers on a low level of P33_INT1.
1	IE0	External Interrupt0 event.
		Set to logic level 1 by hardware when an external interrupt is detected on P32_INT0.
0	IT0	External Interrupt0 control.
		Set by firmware only. When set to logic level 1, Interrupt0 triggered by a falling edge on P32_INT0.
		When set to logic level 0, Interrupt0 triggered by a low level on P32_INT0.

## 8.1.6.3 T01MOD register

This register is used to configure Timer0/1.

Table 24. Timer 0/1 T01MOD register (SFR address 89h), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	GATE1	C/T1	M11	M10	GATE0	C/T0	M01	M00
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 25. Description of T01MOD bits

Bit	Symbol	Description
7	GATE1	Timer1 gate control. Set by firmware only.
		When set to logic level 1, Timer1 is enabled only when P33_INT1 is high and bit TR1 of register T01CON is set.
		When set to logic 0, Timer1 is enabled.
6	C/T1	Timer1 timer/counter selector. Set by firmware only.
		When set to logic level 1, Timer1 is set to counter operation.
		When set to logic level 0, Timer1 is set to timer operation.
5 to 4	M[11:10]	Timer1 mode. Set by firmware only.
		<ul><li>Mode 0: M11 = 0 and M10 = 0</li></ul>
		- 8192 counter
		<ul> <li>T1L serves as a 5-bit prescaler</li> </ul>
		<ul><li>Mode 1: M11 = 0 and M10 = 1</li></ul>
		<ul> <li>16-bit timer/counter</li> </ul>
		<ul> <li>T1H and T1L are cascaded</li> </ul>
		<ul><li>Mode 2: M11 = 1 and M10 = 0</li></ul>
		<ul> <li>8-bit auto-reload timer/counter.</li> </ul>
		<ul> <li>T1H stores value to be reloaded into T1L each time T1L overflows.</li> </ul>
		<ul><li>Mode 3: M11 = 1 and M10 = 1</li></ul>
		<ul> <li>Timer1 is stopped (count frozen).</li> </ul>

Table 25. Description of T01MOD bits ...continued

Bit	Symbol	Description
3	GATE0	Timer0 gate control. Set by firmware only.
		When set to logic level 1, Timer0 is enabled only when P32_INT0 is high and bit TR0 of register T01CON is set.
		When set to logic level 0, Timer0 is enabled.
2	C/T0	Timer0 timer/counter selector. Set by firmware only.
		When set to logic level 1, Timer0 is set to counter operation.
		When set to logic level 0, Timer0 is set to timer operation.
1 to 0	M[01:00]	Timer0 mode. Set by firmware only.
		<ul> <li>Mode 0: M01 = 0 and M00 = 0</li> </ul>
		<ul> <li>8192 timer</li> </ul>
		<ul> <li>T0L acts as a 5-bit prescaler.</li> </ul>
		<ul> <li>Mode 1: M01 = 0 and M00 = 1</li> </ul>
		<ul> <li>16-bit timer/counter</li> </ul>
		<ul> <li>T0H and T0L are cascaded.</li> </ul>
		<ul> <li>Mode 2: M01 = 1 and M00 = 0</li> </ul>
		<ul> <li>8-bit auto-reload timer/counter</li> </ul>
		<ul> <li>T0H stores value to be reloaded into T0L each time T0L overflows.</li> </ul>
		<ul> <li>Mode 3: M01 = 1 and M00 = 1</li> </ul>
		<ul> <li>Timer0 split into two 8-bit timer/counters T0H and T0L</li> </ul>
		<ul> <li>T0H is controlled by the control bit of Timer1: bit TR1 of register T01CON</li> </ul>
		<ul> <li>T0L is controlled by standard Timer0 control: "{P32_INT0 OR (NOT GATE0)} AND bit TR0".</li> </ul>

## 8.1.6.4 TOL and TOH registers

These are the actual timer/counter bytes for Timer0: T0L is the lower byte; T0H is the upper byte.

Table 26. Timer0/1 T0L register (SFR address 8Ah), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T0L.7	T0L.6	T0L.5	T0L.4	T0L.3	T0L.2	T0L.1	T0L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

## Table 27. Description of T0L bits

Bit	Symbol	Description
7:0	T0L.7 to T0L.0	Timer0 timer/counter lower byte

## Table 28. Timer0/1 T0H register (SFR address 8Ch), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T0H.7	T0H.6	T0H.5	T0H.4	T0H.3	T0H.2	T0H.1	T0H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

## Table 29. Description of T0H bits

Bit	Symbol	Description
7 to 0	T0H.7 to T0H.0	Timer0 timer/counter upper byte

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#### 8.1.6.5 T1L and T1H registers

These are the actual timer/counter bytes for Timer1. T1L is the lower byte, T1H is the upper byte.

Table 30. Timer0/1 T1L register (SFR address 8Bh), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T1L.7	T1L.6	T1L.5	T1L.4	T1L.3	T1L.2	T1L.1	T1L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

#### Table 31. Description of T1L bits

Bit	Symbol	Description
7 to 0	T1L.7 to T1L.0	Timer1 timer/counter lower byte

#### Table 32. Timer0/1 T1H register (SFR address 8Dh), bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T1H.7	T1H.6	T1H.5	T1H.4	T1H.3	T1H.2	T1H.1	T1H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

#### Table 33. Description of T1H bits

Bit	Symbol	Description
7 to 0	T1H.7 to T1H.0	Timer1 timer/counter upper byte

## 8.1.6.6 Incrementer

The two 16-bit timer/counters are built around an 8-bit incrementer. The Timer0/1 are incremented in the CPU states S1 to S4; the overflow flags are set in CPU states S2 and S4.

- CPU state S1: TOL is incremented if Timer0 is set to:
  - timer operation
  - counter operation and when a 1-to-0 transition is detected on P34 / SIC CLK input.
- CPU state S2: TOH is incremented if:
  - T0L overflows. The overflow flag TF0 in register T01CON is updated.
- CPU state S3: T1L is incremented if Timer1 is set to:
  - timer operation or
  - counter operation and when a 1-to-0 transition is detected on P35 input.
- · CPU state S4: T1H is incremented if:
  - T1L overflows. The overflow flag TF1 in register T01CON is updated.

#### 8.1.6.7 Overflow detection

For both the upper and lower bytes of the Timer0/1, an overflow is detected by comparing the incremented value of the most significant bit with its previous value. An overflow occurs when this bit changes from logic level 1 to logic level 0. An overflow event in the lower byte is clocked into a flip-flop and is used in the next state as the increment enable for the upper byte. An overflow event in the upper byte will set the corresponding overflow bit in the T01CON register to logic level 1. The upper byte overflow is also clocked into a flip-flop to generate the output signals 't0\_ovf' and 't1\_ovf'.

The overflow flags TF0 and TF1, found in register T01CON, are loaded during states S2 and S4 respectively. The interrupt controller of the 80C51 scans all requests at state S2. Thus, an overflow of Timer0 or Timer1 is detected one machine cycle after it occurred. When the request is serviced, the interrupt routine sets the overflow flag to logic 0.

Execution of the interrupt routine starts on the fourth machine cycles following the timer overflow. When Timer0/1 receives the acknowledge from the CPU:

- the overflow flag TF0 in register T01CON is set to logic level 0
- two machine cycles later, the overflow flag TF1 in register T01CON is set to logic level 0

If during the same machine cycle, an overflow flag is set to logic level 0 due to a CPU acknowledge and set to logic level 1 due to an overflow, the set to logic level 1 is the strongest.

#### 8.1.7 Timer2 description

Timer2 supports a subset of the standard Timer2 found in the 8052 microcontroller. Timer2 can be configured into 2 functional modes via the T2CON and T2MOD registers:

- Mode1: Auto-reload up/down counting
- Mode2: Baud rate generation for Debug UART

Timer2 can operate either as a timer or as an event counter.

#### 8.1.7.1 Timer2 registers

Timer2 contains six Special Function Registers (SFRs) which can be accessed by the CPU.

Table 34. Timer2 SFR register List

Name	Size [bytes]	SFR address	Description	Access
T2CON	1	C8h	Timer2 control register	R/W
T2MOD	1	C9h	Timer2 mode register	R/W
RCAP2L	1	CAh	Timer2 reload lower byte	R/W
RCAP2H	1	CBh	Timer2 reload upper byte	R/W
T2L	1	CCh	Timer2 timer/counter lower byte	R/W
T2H	1	CDh	Timer2 timer/counter upper byte	R/W

Timer2 registers can be written to by either hardware or firmware. If both the hardware and firmware attempt to update the registers T2H, T2L, RCAP2H or RCAP2L during the same machine cycle, the firmware write takes precedence. A firmware write occurs in state S6 of the machine cycle.

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Each increment or decrement of Timer2 occurs in state S1 except when in baud rate generation mode and configured as a counter. In this mode, Timer2 increments on each clock cycle. When configured as a timer, Timer2 is incremented every machine cycle. Since a machine cycle consists of 6 clock periods, the count rate is 1/6 of the CPU clock frequency.

## 8.1.7.2 T2CON register

The register is used to control Timer2 and report its status.

Table 35. Timer2 T2CON register (SFR address C8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TF2	-	RCLK0	TCLK0	-	TR2	C/T2	-
Reset	0	0	0	0	0	0	0	0
Access	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 36. Description of T2CON bits

	<u> </u>	otion of 12CON bits
Bit	Symbol	Description
7	TF2	Timer2 overflow
		Set to logic level 1 by a Timer2 overflow.
		Set to logic level 0 by firmware.
		TF2 is not set when in baud rate generation mode.
6	-	Reserved.
5	RCLK0	Timer2 Debug UART Receive Clock selector. Set by firmware only.
		When set to logic level 1, Debug UART uses Timer2 overflow pulses.
		When set to logic level 0, Debug UART uses overflow pulses from another source (e.g. Timer1 in a standard configuration).
4	TCLK0	Timer2 Debug UART Transmit Clock selector. Set by firmware only.
		When set to logic level 1, Debug UART uses Timer2 overflow pulses.
		When set to logic level 0, Debug UART uses overflow pulses from another source (e.g. Timer1 in a standard configuration).
3	-	Reserved.
2	TR2	Timer2 Run control.
		Set by firmware only.
		When set to logic level 1, Timer2 is started.
		When set to logic level 0, Timer2 is stopped.
1	C/T2	Timer2 Counter/Timer selector.
		Set by firmware only.
		When set to logic level 1, Timer2 is set to counter operation.
		When set to logic level 0, Timer2 is set to timer operation.
0	-	Reserved. This bit must be set to logic level 0 by firmware.

## 8.1.7.3 T2MOD register

This Special Function Register is used to configure Timer2.

Table 37. Timer2 T2MOD register (SFR address C9h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	T2RD	-	DCEN
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R/W	R/W

Table 38. Description of TMOD bits

Bit	Symbol	Description
7 to 3	-	Reserved.
2	T2RD	Timer2 ReaD flag.
		Set by hardware and firmware.
		This bit is set to logic level 1 by hardware, if T2H is incremented between reading T2L and reading T2H. This bit is set to logic level 0, on the trailing edge of next T2L read.
		This bit is used to indicate that the 16-bit Timer 2 register is not read properly since the T2H part was incremented by hardware before it was read.
1	-	Reserved
0	DCEN	Timer2 Down Count ENable.
		Set by firmware only.
		When this bit is set, Timer2 can be configured (in auto_reload mode) as an up-counter.
		When this bit is reset, Timer2 can be configured (in auto-reload mode) as a down-counter.

#### 8.1.7.4 T2L, T2H registers

These are the actual timer/counter bytes. T2L is the lower byte, T2H the upper byte.

On the fly reading can give a wrong value since T2H can be changed after T2L is read and before T2H is read. This situation is indicated by flag T2RD in T2MOD.

These two 8-bit registers are always combined to operate as one 16-bit timer/counter.

Table 39. Timer2 T2L register (SFR address CCh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T2L.7	T2L.6	T2L.5	T2L.4	T2L.3	T2L.2	T2L.1	T2L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 40. Description of T2L bits

Bit	Symbol	Description
7 to 0	T2L.7 to T2L.0	Timer2 timer/counter lower byte

Table 41. Timer2 T2H register (SFR address CDh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	T2H.7	T2H.6	T2H.5	T2H.4	T2H.3	T2H.2	T2H.1	T2H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 42. Description of T2H bits

Bit	Symbol	Description
7 to 0	T2H.7 to T2H.0	Timer2 timer/counter upper byte

#### 8.1.7.5 RCAP2L, RCAP2H registers

These are the reload bytes. In the reload mode the T2H/T2L counters are loaded with the values found in the RCAP2H/RCAP2L registers respectively.

Table 43. Timer2 RCAP2L register (SFR address CAh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	R2L.7	R2L.6	R2L.5	R2L.4	R2L.3	R2L.2	R2L.1	R2L.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 44. Description of RCAP2L bits

Bit	Symbol	Description
7 to 0	R2L.7 to R2L.0	Timer2 lower reload byte

Table 45. Timer2 RCAP2H register (SFR address CBh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	R2H.7	R2H.6	R2H.5	R2H.4	R2H.3	R2H.2	R2H.1	R2H.0
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 46. Description of RCAP2H bits

Bit	Symbol	Description
7 to 0	R2H.7 to R2H.0	Timer2 upper reload byte

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## 8.1.8 Debug UART

The Debug UART is implemented to assist debug using UART\_RX and UART\_TX pins.

#### 8.1.8.1 Feature list

The Debug UART has the following characteristics:

- Full duplex serial port
- Receive buffer to allow reception of a second byte while the first byte is being read out by the CPU
- Four modes of operation which support 8-bit and 9-bit data transfer at various baud rates
- Supports multi-processor communication
- Baud rate can be controlled through Timer1 or Timer2 baud rate generator

#### 8.1.8.2 Debug UART functional description

The serial port has a receive buffer: a second byte can be stored while the previous one is read out of the buffer by the CPU. However, if the first byte has still not been read by the time reception of the second byte is complete, one of the bytes will be lost.

The receive and transmit data registers of the serial port are both accessed by firmware via the Special Function Register S0BUF. Writing to S0BUF loads the transmit register; reading from S0BUF accesses a physically separate receive register.

The serial port can operate in 4 modes. These modes are selected by programming bits SM0 and SM1 in S0CON:

- Mode 0:
  - Serial data are received and transmitted through UART\_RX. UART\_TX outputs the shift clock. 8 bits are transmitted/received (LSB first)
    - Baud rate: fixed at 1/6 of the frequency of the CPU clock
- Mode 1:
  - 10 bits are transmitted through UART\_TX or received through UART\_RX: a start bit (0), 8 data bits (LSB first), and a stop bit (1)
  - Receive: The received stop bit is stored into bit RB8 of register S0CON
  - Baud rate: variable (depends on overflow of Timer1 or Timer2)
- Mode 2:
  - 11 bits are transmitted through UART\_TX or received through UART\_RX: start bit
     (0), 8 data bits (LSB first), a 9th data bit, and a stop bit (1)
  - Transmit: the 9th data bit is taken from bit TB8 of S0CON. For example, the parity bit could be loaded into TB8.
  - Receive: the 9th data bit is stored into RB8 of S0CON, while the stop bit is ignored
  - Baud rate: programmable to either 1/16 or 1/32 the frequency of the CPU clock

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#### • Mode 3:

- 11 bits are transmitted through UART\_TX or received through UART\_RX: a start bit (0), 8 data bits (LSB first), a 9th data bit, and a stop bit (1). In fact, mode 3 is the same as mode 2 in all aspects except the baud rate
- Transmit: as mode 2, the 9th data bit is taken from TB8 of S0CON
- Receive: as mode 2, the 9th data bit is stored into RB8 of S0CON
- Baud rate: depends on overflows of Timer1 or Timer2

The Debug UART initiates transmission and/or reception as follows.

- Transmission is initiated, in modes 0, 1, 2, 3, by any instruction that uses S0BUF as destination
- Reception is initiated, in mode 0, if RI and REN in S0CON are set to logic 0 and 1 respectively
- Reception is initiated in modes 1, 2, 3 by the incoming start bit if REN in S0CON is set to a logic 1

The Debug UART contains 2 SFRs:

Table 47. Debug UART SFR register list

Name	Size [bytes]	SFR address	Description	Access
S0CON	1	0098h	Control and status register	R/W
S0BUF	1	0099h	Transmit and receive buffer	R/W

## 8.1.8.3 S0CON register

The Special Function Register SOCON is the control and status register of the Debug UART. This register contains the mode selection bits (SM2, SM1, SM0), the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

Table 48. Debug UART SOCON register (SFR: address 98h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	SM0	SM1	SM2	REN	TB8	RB8	TI	RI
Reset	0	0	0	0	0	0	0	0
Access	R/W							

Table 49. Description of S0CON bits

Bit	Symbol	Description
7 to 6	SM (0:1)	Mode selection bit 0 and 1.
		Set by firmware only. The Debug UART has 4 modes ( <u>Table 50 "Debug UART modes"</u> on page 33).
5	SM2	Multi-processor communication enable.
		Enables the multi-processor communication feature. Set by firmware only.
		In mode 2 and 3:
		if SM2 is set to logic level 1, then RI will not be activated and RB8 and S0BUF will not be loaded if the 9th data bit received is a logic 0
		if SM2 is set to logic level 0, it has no influence on the activation of RI and RB8
		In mode 1:
		if SM2 is set to logic level 1, then RI will not be activated and RB8 and S0BUF will not be loaded if no valid stop bit was received
		if SM2 is set to logic level 0, it has no influence on the activation of RI and RB8
		In mode 0, SM2 has no influence
4	REN	Serial reception enable.
		Set by firmware only.
		When set to logic level 1, enables reception.
3	TB8	Transmit data bit.
		Set by firmware only.
		In modes 2 and 3, the value of TB8 is transmitted as the 9th data bit

In modes 0 and 1, the TB8 bit is not used

Table 49. Description of S0CON bits ... continued

Bit	Symbol	Description
2	RB8	Receive data bit.
		Set by hardware and by firmware.[1]
		When set to logic level 1:
		In modes 2 or 3, the hardware stores the 9th data bit that was received in RB8
		In mode 1, the hardware stores the stop bit that was received in RB8
		In mode 0, the hardware does not change RB8.
1	TI	Transmit interrupt flag.
		TI must be set to logic level 0 by firmware.
		In modes 2 or 3, when transmitting, the hardware sets to logic level 1 the transmit interrupt flag TI at the end of the 9th bit time
		In modes 0 or 1, when transmitting, the hardware sets to logic level 1 the transmit interrupt flag TI at the end of the 8th bit time.
0	RI	Receive interrupt flag.
		RI must be set to logic level 0 by firmware.
		In modes 2 or 3, when receiving, the hardware sets to logic level 1 the receive interrupt flag 1 clock period after sampling the 9th data bit (if SM2=1 setting RI can be blocked, see bit description of SM2 above)
		In mode 1, when receiving, the hardware sets to logic level 1 the receive interrupt flag 1 clock period after sampling the stop bit [2]
		In mode 0, when receiving, the hardware sets to logic level 1 RI at the end of the CPU state 1 of the 9th machine cycle after the machine cycle where the data reception started by a write to S0CON.

<sup>[1]</sup> If SM2 is set to logic 1, loading RB8 can be blocked, see bit description of SM2 above.

**Remark:** The SOCON register supports a locking mechanism to prevent firmware read-modify-write instructions to overwrite the contents while hardware is modifying the contents of the register.

Table 50. Debug UART modes

Mode	SM0	SM1	Description	Baud rate
0	0	0	Shift register	f <sub>clk</sub> /6
1	0	1	8 bits Debug UART	Variable
2	1	0	9 bits Debug UART	f <sub>clk</sub> /64 or f <sub>clk</sub> /32
3	1	1	9 bits Debug UART	Variable

<sup>[2]</sup> If SM2 is set to logic 1, setting RI can be blocked, see bit description of SM2 above.

#### 8.1.8.4 SOBUF register

This register is implemented twice. Writing to S0BUF writes to the transmit buffer. Reading from S0BUF reads from the receive buffer. Only hardware can read from the transmit buffer and write to the receive buffer.

Table 51. Debug UART SOBUF Register (SFR: address 99h) bit allocation

	•		•	•				
Bit	7	6	5	4	3	2	1	0
Symbol		S0BUF[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 52. Description of S0BUF bits

Bit	Symbol	Description
7 to 0	S0BUF[7:0]	Writing to S0BUF writes to the transmit buffer.
		Reading from S0BUF reads from the receive buffer.

#### 8.1.8.5 Mode 0 baud rate

In mode 0, the baud rate is derived from the CPU states signals and thus:

Baud rate in mode 0 (1)

 $\frac{clk}{6}$ 

The next table lists the baud rates in Debug UART mode 0.

Table 53. Baud rates in mode 0

Conditions	Min	Тур	Max	Unit
f <sub>CLK</sub>	6.78	13.56	27.12	MHz
Baud rate	1.13	2.26	4.52	Mb/s

#### 8.1.8.6 Mode 2 baud rate

In mode 2, the baud rate depends on the value of bit SMOD from the SFR register PCON.

Baud Rate using mode 2 (2)

$$\frac{2^{SMOD}}{32} \times f_{clk}$$

The next table lists the baud rates in Debug UART mode 2.

Table 54. Baud rates in mode 2

Conditions	Min	Тур	Max	Unit
f <sub>CLK</sub>	6.78	13.56	27.12	MHz
Baud rate (SMOD=0)	212	424	847.5	kb/s
Baud rate (SMOD=1)	424	847.5	1695	kb/s

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#### 8.1.8.7 Mode 1 and 3 baud rates

In modes 1 and 3, the baud rates are determined by the rate of timer1 and timer2 overflow bits: 't1\_ovf' and 't2\_ovf'. The register bit TCLK0 from the register T2CON selects if 't1\_ovf' or 't2\_ovf' should be used as a source when transmitting. The register bit RCLK0 from the register T2CON selects if 't1\_ovf' or 't2\_ovf' should be used as a source when receiving. The timers interrupt should be disabled when used to define the Debug UART baud rates.

The data rate is also dependant on the value of the bit SMOD from the SFR register PCON.

If over1rate is the equivalent 't1\_ovf' frequency and over2rate is the equivalent 't2\_ovf' frequency then:

Baud rate in mode 1 and 3 when related to timer1 overflow

to timer1 overflow

OD

(3)

$$\frac{2^{SMOD}}{32} \cdot over1rate$$

See also Section 8.1.8.8 "Baud rates using Timer1 (Debug UART mode 1 and 3)"

Baud rate in mode 1 and 3 when related to timer2 overflow

(4)

$$\frac{1}{16}$$
 · over2rate

See also Section 8.1.8.9 "Baud rates using Timer2 (Debug UART mode 1 and 3)"

The next table shows the trigger select:

Table 55. Trigger select

RCLK0	TCLK0	SMOD	receive trigger rate	transmit trigger rate
0	-	0	over1rate/32	-
0	-	1	over1rate/16	-
1	-	-	over2rate/16	-
-	0	0	-	over1rate/32
-	0	1	-	over1rate/16
-	1	-	-	over2rate/16

## 8.1.8.8 Baud rates using Timer1 (Debug UART mode 1 and 3)

The Timer1 interrupt should be disabled in this application. The Timer1 itself can be configured for either 'timer' or 'counter' operation, and in any of its 3 running modes. In the most typical applications, it is configured for 'timer' operation, in the auto-reload mode (Timer1 mode 2: high nibble of T01MOD = 0010b). In that case the baud rate is given by the formula:

$$\frac{2^{SMOD}}{32} \times \frac{f_{clk}}{6 \times (256 - T1H)}$$

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When rewriting this formula, the value for the Timer1 reload value T1H is calculated from the desired baud rate as follows:

$$256 - \frac{2^{SMOD} \times f_{clk}}{32 \times 6 \times Baudrate}$$

One can achieve very low baud rates with Timer1 by leaving the Timer1 interrupt enabled, and configuring the timer to run as a 16-bit timer (high nibble of T01MOD = 0001b), and using the Timer1 interrupt to do a 16-bit firmware reload. Note: the frequency  $f_{\text{clk}}$  is the internal microcontroller frequency. If there is no clock divider then  $f_{\text{clk}} = f_{\text{osc}}$ .

For details on programming Timer1 to function as baud rate generator for the Debug UART see Section 8.1.6 "Timer0/1 description" on page 20.

The next table lists the maximum baud rates for using mode 2 of Timer1.

Table 56. Maximum baud rates using mode 2 of Timer1

Reload value	f <sub>CLK</sub> divided by	SMOD	Baud rate at f <sub>CLK</sub>		Unit	
			6.78	13.56	27.12	MHz
FF	96	1	70.6	141.2	282.5	kb/s

The next table shows commonly used baud rates using mode 2 of Timer1 and a CLK frequency of 27.12 MHz.

Table 57. Baud rates using mode 2 of Timer1 with f<sub>CLK</sub> = 27.12 MHz

Reload value	f <sub>CLK</sub> divided by	SMOD	Baud rate at f <sub>CLK</sub>	Unit
FC	706	0	38.4	kb/s
F9	1412	0	19.2	kb/s
F1	2825	0	9.6	kb/s
E3	5650	0	4.8	kb/s
C5	11300	0	2.4	kb/s
8A	22600	0	1.2	kb/s

### 8.1.8.9 Baud rates using Timer2 (Debug UART mode 1 and 3)

Timer2 has a programming mode to function as baud rate generator for the Debug UART. In this mode the baud rate is given by formula:

$$\frac{f_{clk}}{16 \times [65536 - (T2RCH, T2RCL)]}$$

When rewriting this formula, the value for the Timer2 reload values T2RCH/L is calculated from the desired baud rate as follows:

$$65536 - \frac{f_{clk}}{16 \times Baudrate}$$

For details on programming Timer2 to function as baud rate generator for the Debug UART (see <u>Section 8.1.7 "Timer2 description"</u>).

Note: the frequency  $f_{clk}$  is the internal microcontroller frequency. If there is no clock divider then  $f_{clk} = f_{osc}$ .

The next table lists the maximum baud rates when using Timer2.

Table 58. Maximum baud rates using Timer2

Reload value T2RCH/L	f <sub>CLK</sub> divided by	Baud rate			Unit
		6.78	13.56	27.12	MHz
FFFF	16	424	847.5	1695	kb/s

# 8.2 General purpose IOs configurations

This chapter describes the different configurations for the IO pads:

P70\_IRQ

P35

P34, alternate function SIC\_CLK

P33\_INT1

P32 INT0

P31, alternate function UART\_TX

P30, alternate function UART RX

Note that in Hard Power Down mode, these ports are disconnected from their supply rail.

For a given port x, there are three configuration registers:

PxCFGA[n]

PxCFGB[n]

Px[n]

where x is 3 or 7 and n is the bit index.

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At maximum 4 different controllable modes can be supported. These modes are defined with the following bits:

- PxCFGA[n]=0 and PxCFGB[n]=0: Open drain
- PxCFGA[n]=1 and PxCFGB[n]=0: Quasi Bidirectional (Reset mode)
- PxCFGA[n]=0 and PxCFGB[n]=1: input (High Impedance)
- PxCFGA[n]=1 and PxCFGB[n]=1: Push/pull output

Px[n] is used to write or read the port value.

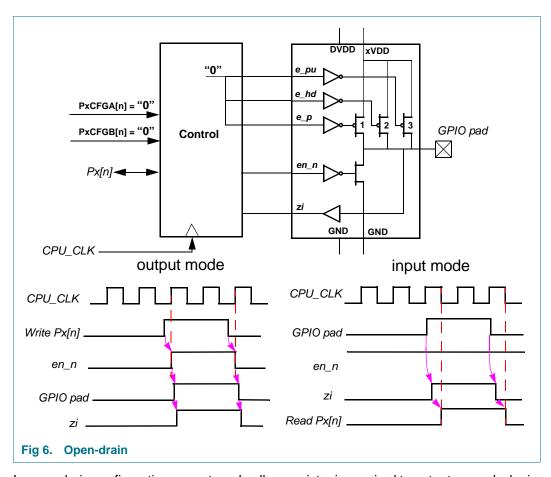
Here is the list of the registers used for these GPIO configuration

Table 59. Timer0/1 Special Function registers List

Name	Size [bytes]	SFR address	Description	Access
P3CFGA	1	FCh	Port 3 configuration	R/W
P3CFGB	1	FDh	Port 3 configuration	R/W
P3	1	B0h	Port 3 value	R/W
P7CFGA	1	F4h	Port 7configuration	R/W
P7CFGB	1	F5h	Port 7 configuration	R/W
P7	1	F7h	Port 7 value	R/W

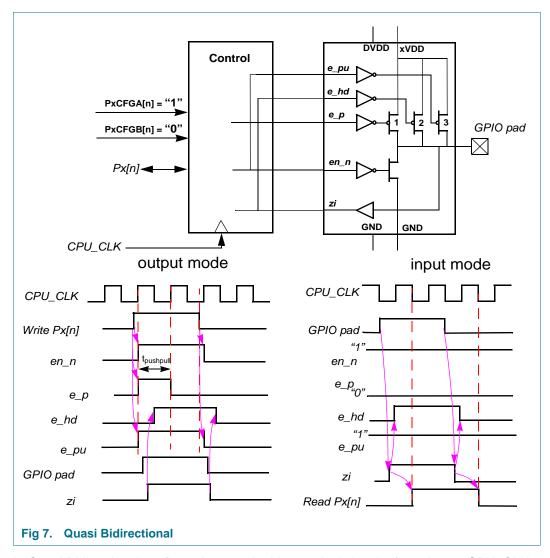
# 8.2.1 Pad configurations description

### 8.2.1.1 Open-drain



In open drain configuration, an external pull-up resistor is required to output or read a logic level 1. When writing polarity Px[n] to logic level 0, the GPIO pad is pulled down to logic level 0. When writing polarity Px[n] to logic level 1 the GPIO pad is in High Impedance.

# 8.2.1.2 Quasi Bidirectional

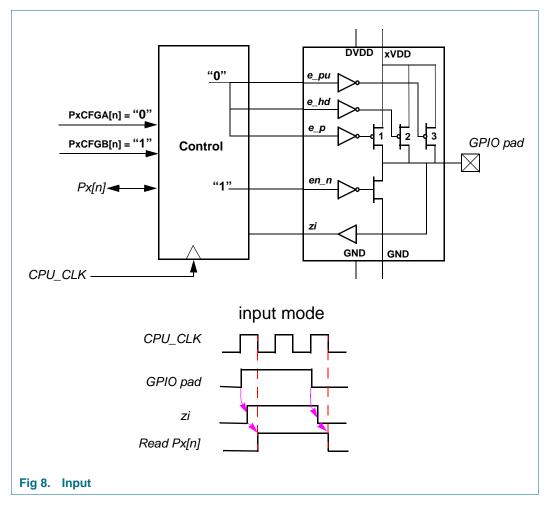


In Quasi Bidirectional configuration, e\_p is driven to logic level 1 for only one CPU\_CLK period when writing Px[n]. During the t<sub>pushpull</sub> time the pad drive a strong logic level 1 at its output. While Px[n] is logic level 1, this state will be held by the weak hold transistor (e\_hd), which implements a latch function. Because of the weaker nature of this hold transistor, the pad cell can now act as an input as well. A third very weak pull-up transistor (e\_pu) ensures that an open input is read as logic level 1.

On a transition from logic level 0 to logic level 1 externally driven on GPIO pad, when the voltage on the pad is at the supply voltage divided by 2, the pull-up (e\_hd) is activated.

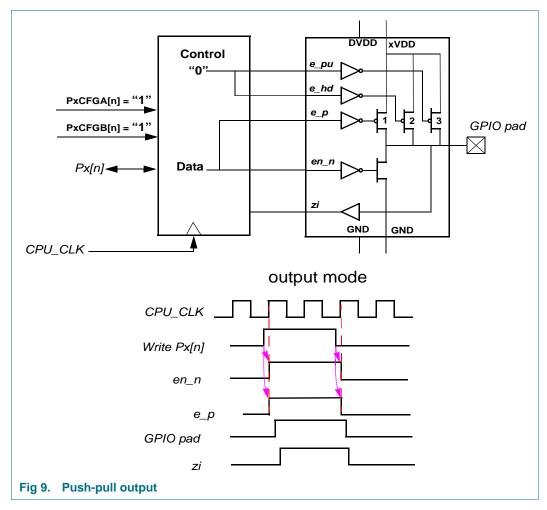
The maximum current that can be sourced by the e\_pu and e\_hd transistors is 150  $\mu A$  total at 3.6 V.

### 8.2.1.3 Input



In input configuration, no pull up or hold resistor are internally connected to the pad.

# 8.2.1.4 Push-pull output



In push-pull output, the output pin drives a strong logic level 0 or a logic level 1 continuously. It is possible to read back the pin output value.

# 8.2.2 GPIO registers description

### 8.2.2.1 P7CFGA register

Table 60. P7CFGA register (SFR: address F4h) bit allocation

		3 ( -		,				
Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	-	P7CFGA[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R	R/W

Table 61. Description of P7CFGA bits

Bit	Symbol	Description
7 to 1	-	Reserved.
0	P7CFGA[0]	In conjuction with P7CFGB[0], it configures the functional mode of P70_IRQ pin.

### 8.2.2.2 P7CFGB register

Table 62. P7CFGB register (SFR: address F5h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	-	P7CFGB[0]
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R/W

Table 63. Description of P7CFGB bits

Bit	Symbol	Description
7 to 1	-	Reserved.
0	P7CFGB[0]	In conjuction with P7CFGA[0], it configures the functional mode of P70_IRQ pin.

### 8.2.2.3 P7 register

Table 64. P7 register (SFR: address F7h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	-	P7[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R	R/W

Table 65. Description of P7 bits

Bit	Symbol	Description
7 to 1	-	Reserved.
0	P7[0]	Writing to P7[0] writes the corresponding value to the P70_IRQ pin according to the configuration mode defined by P7CFGA[0] and P7CFGB[0].  Reading from P7[0] reads the state of P70_IRQ pin.

# 8.2.2.4 P3CFGA register

Table 66. P3CFGA Register (SFR: address FCh) bit allocation

			•		,			
Bit	7	6	5	4	3	2	1	0
Symbol	-	-	P3CFGA[5]	P3CFGA[4]	P3CFGA[3]	P3CFGA[2]	P3CFGA[1]	P3CFGA[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 67. Description of P3CFGA Register bits

Bit	Symbol	Description
7:6		Reserved.
5	P3CFGA[5]	In conjuction with P3CFGB[5], it configures the functional mode of P35
4	P3CFGA[4]	In conjuction with P3CFGB[4], it configures the functional mode of P34
3	P3CFGA[3]	In conjuction with P3CFGB[3], it configures the functional mode of P33_INT1
2	P3CFGA[2]	In conjuction with P3CFGB[2], it configures the functional mode of P32_INT0
1	P3CFGA[1]	In conjuction with P3CFGB[1], it configures the functional mode of P31
0	P3CFGA[0]	In conjuction with P3CFGB[0], it configures the functional mode of P30

### 8.2.2.5 P3CFGB register

Table 68. P3CFGB register (SFR: address FDh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	P3CFGB[5]	P3CFGB[4]	P3CFGB[3]	P3CFGB[2]	P3CFGB[1]	P3CFGB[0]
Reset	0	0	0	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

Table 69. Description of P3CFGB bits

Bit	Symbol	Description
7 to 6		Reserved.
5	P3CFGB[5]	In conjuction with P3CFGA[5], it configures the functional mode of P35.
4	P3CFGB[4]	In conjuction with P3CFGA[4], it configures the functional mode of P34.
3	P3CFGB[3]	In conjuction with P3CFGA[3], it configures the functional mode of P33.
2	P3CFGB[2]	In conjuction with P3CFGA[2], it configures the functional mode of P32.
1	P3CFGB[1]	In conjuction with P3CFGA[1], it configures the functional mode of P31.
0	P3CFGB[0]	In conjuction with P3CFGA[0], it configures the functional mode of P30.

# 8.2.2.6 P3 register

Table 70. P3 register (SFR: address B0h) bit allocation

		9.010.	(0) 111 0000					
Bit	7	6	5	4	3	2	1	0
Symbol	-	-	P3[5]	P3[4]	P3[3]	P3[2]	P3[1]	P3[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 71. Description of P3 bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5	P3[5]	Writing to P3[5] writes the corresponding value to P35 pin according to the configuration mode defined by P3CFGA[5] and P3CFGB[5].
		Reading from P3[5] reads the state of P35 pin.
4	P3[4]	When P34 alternate function SIC_CLK is not used, writing to P3[4] writes the corresponding value to P34 pin according to the configuration mode defined by P3CFGA[4] and P3CFGB[4].
		Reading from P3[4] reads the state of P34 pin.
3	P3[3]	Writing to P3[3] writes the corresponding value to P33_INT1 pin according to the configuration mode defined by P3CFGA[3] and P3CFGB[3].
		Reading from P3[3] reads the state of P33_INT1 pin.
2	P3[2]	Writing to P3[2] writes the corresponding value to P32_INT0 pin according to the configuration mode defined by P3CFGA[2] and P3CFGB[2].
		Reading from P3[2] reads the state of P32_INT0 pin.
1	P3[1]	When the P31 pin alternate function UART_TX is not used, writing to P3[1] writes the corresponding value to P31 pin according to the configuration mode defined by P3CFGA[1] and P3CFGB[1].
		Reading from P3[1] reads the state of P31 pin.
0	P3[0]	When the P30 pin alternate function UART_RX is not used, writing to P3[0] writes the corresponding value to P30 pin according to the configuration mode defined by P3CFGA[0] and P3CFGB[0].
		Reading from P3[0] reads the state of P30 pin.

### 8.2.2.7 P5 register

Table 72. P5 register (SFR: address D7h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	P5[2]	P5[1]	P5[0]
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R/W	R/W

Table 73. Description of P5 bits

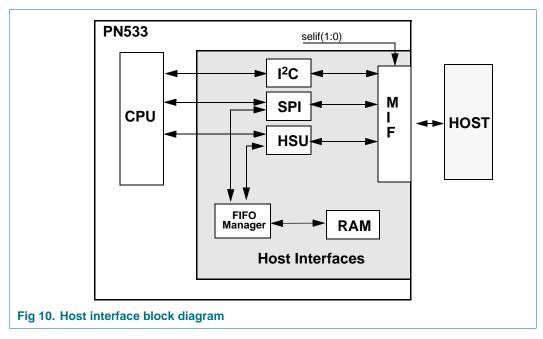
Bit	Symbol	Description
7 to 3	-	Reserved.
2	P5[2]	<b>Bit 2 of P5 register</b> is attached to USB signal "MP_ready" that indicates USB block is ready for a new transaction like write command, write data or read data.
		This bit is polled by embedded firmware that manage USB transactions.
1	P5[1]	<b>Writing to P5[1] writes</b> the corresponding value to SDA I <sup>2</sup> C pin according to the open drain configuration mode.
		Reading from P5[1] reads the state of SDA I <sup>2</sup> C pin.
0	P5[0]	<b>Writing to P5[0] writes</b> the corresponding value to SCL of the I <sup>2</sup> C pin according to the open drain configuration mode.  Reading from P5[0] reads the state of NSS I <sup>2</sup> C pin.

Remark: P5 supports only open drain mode

### 8.3 Host interfaces

PN533 must be able to support different kind of interfaces to communicate with the HOST. All the interfaces that have to be supported are exclusive.

- USB interface
- High Speed UART (HSU): supporting specific high baud rates



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# 8.3.1 MATX description

After every Power-On or Hard Reset (RSTPD\_N at low level), the PN533 also resets its interfaces and checks the current HOST interface type.

The PN533 identifies the selected HOST interface by means of the logic levels on the control pins I0 and I1 after the Reset Phase. This is done by a combination of fixed pin connections

The Power for the MATX is delivered from PVDD.

The firmware must copy the value of the pads I0 and I1 to respectively selif[0] and selif[1].

Table 74. HOST interface selection

PN533		Host Interface selected						
Pin number	Config_IO_I1 register	HSU	HSU	USB non bus powered	USB bus powered			
18	selif[1]	0	1	0	1			
17	selif[0]	0	0	1	1			
25	DP	RX	RX	DP	DP			
24	DM	TX	TX	DM	DM			

### 8.3.1.1 MATX register

The Config I0\_I1 register is used to select the host interface. It manages also the polarity of P33\_INT1.

Table 75. Config I0\_I1 register (address 6103h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	int1_pol	-	pad_I1	-	pad_I0	enselif	Selif	[1:0]
Reset	0	0	Χ	0	Χ	0	0	0
Access	R/W	R	R/W	R	R/W	R/W	R/W	R/W

Table 76. Description of Config I0\_I1 bits

Bit	Symbol	Description
7	int1_pol	When set to logic 1, the value of the P33_INT1 pin is inverted.
6	-	Reserved.
5	pad_l1	When read this bit gives the state of the I1 pin.
4	-	Reserved.
3	pad_I0	When read this bit gives the state of the I0 pin.
2	enselif	When set to logic 1, this bit indicates that the self bits are valid and that the selected interface on the MIF can drive the pins.  The firmware must copy the value of the pads I0 and I1 to respectively selif[0] and selif[1]
		When set to logic 0, the MIF cannot drive the IO lines.
1 to 0	Selif[1:0]	These bits are used by the firmware to select the host interface communication link, see <u>Table 74 on page 47</u> .

# 8.3.1.2 Pads NSS/P50\_SCL/HSU\_RX and MOSI/SDA / HSU\_TX

The I<sup>2</sup>C SDA and SCL IO ports are configured in open drain mode. A pull-up resistor is required for both pins to output or read a logical '1'.

In HSU mode, HSU\_RX is in input mode and HSU\_TX is in push-pull mode.

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### **Near Field Communication (NFC) controller**

## 8.3.2 I<sup>2</sup>C interface

It is recommended to refer the I<sup>2</sup>C standard for more information.

The I<sup>2</sup>C interface implements a Master I<sup>2</sup>C bus interface with integrated shift register, shift timing generation. I<sup>2</sup>C Standard mode (100 kHz SCLK) and Fast mode (400 kHz SCLK) are supported.

General Call +W is supported, not hardware General Call (GC +R).

The mains characteristics of the I<sup>2</sup>C module are:

- Support Master I<sup>2</sup>C bus
- Standard and Fast mode supported

The I<sup>2</sup>C module is control through 5 registers:

Table 77. I<sup>2</sup>C register list

Name	Size [bytes]	Address	Description	Access
I <sup>2</sup> CCON	1	D8h (SFR)	Control register	R/W
I <sup>2</sup> CSTA	1	D9h (SFR)	Status register	R/W
I <sup>2</sup> CDAT	1	DAh (SFR)	Data register	R/W
I <sup>2</sup> CADR	1	DBh (SFR)	Slave Address register	R/W

### 8.3.2.1 I<sup>2</sup>C functional description

The I<sup>2</sup>C interface may operate in any of the following two modes:

- Master Transmitter
- Master Receiver

Two types of data transfers are possible on the I<sup>2</sup>C bus:

- Data transfer from a Master transmitter to a Slave receiver. The first byte transmitted by the Master is the Slave address. Next follows a number of data bytes. The Slave returns an acknowledge bit after each received byte.
- Data transfer from a Slave transmitter to a Master receiver. The first byte (the Slave address) is transmitted by the Master. The Slave then returns an acknowledge bit.
   Next follows the data bytes transmitted by the Slave to the Master. The Master returns an acknowledge bit after each received byte except the last byte. At the end of the last received byte, a "not acknowledge" is returned.

In a given application, the I<sup>2</sup>C interface may operate as a Master or as a Slave.

In the Slave mode, the I<sup>2</sup>C interface hardware looks for its own Slave address and the general call address. If one of these addresses is detected, an interrupt is requested. When the PN533 microcontroller wishes to become the bus Master, the hardware waits until the bus is free before the Master mode is entered so that a possible Slave action is not interrupted. If bus arbitration is lost in the Master mode, the I<sup>2</sup>C interface switches to the Slave mode immediately and can detect its own Slave address in the same serial transfer.

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#### **Near Field Communication (NFC) controller**

#### 8.3.2.2 Master transmitter mode

As a Master, the I<sup>2</sup>C logic will generate all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the I<sup>2</sup>C bus will not be released.

I<sup>2</sup>C data are output through SDA while P50\_SCL outputs the serial clock. The first byte transmitted contains the Slave address of the receiving device (7-bit SLA) and the data direction bit. In this case the data direction bit (R/W) will be a logic '0' (W). I<sup>2</sup>C data are transmitted 8 bits at a time. After each byte is transmitted, an acknowledge bit is received. START and STOP conditions are output to indicate the beginning and the end of a serial transfer.

In the Master transmitter mode, a number of data bytes can be transmitted to the Slave receiver. Before the Master transmitter mode can be entered, I<sup>2</sup>CCON must be initialized with the ENS1 bit set to logic level 1 and the STA, STO and SI bits set to logic level 0. ENS1 must be set to logic level 1 to enable the I<sup>2</sup>C interface. If the AA bit is set to logic level 0, the I<sup>2</sup>C interface will not acknowledge its own Slave address or the general call address if they are present on the bus. This will prevent the I<sup>2</sup>C interface from entering a Slave mode.

The Master transmitter mode may now be entered by setting the STA bit. The I<sup>2</sup>C interface logic will then test the I<sup>2</sup>C bus and generate a start condition as soon as the bus becomes free. When a START condition is transmitted, the serial interrupt flag (SI) is set to logic level 1, and the status code in the status register (I<sup>2</sup>CSTA) will be 08h. This status code must be used to vector to an interrupt service routine that loads I<sup>2</sup>CDAT with the Slave address and the data direction bit (SLA+W). The SI bit in I<sup>2</sup>CCON must then be set to logic level 0 before the serial transfer can continue.

When the Slave address and the direction bit have been transmitted and an acknowledgment bit has been received, the serial interrupt flag (SI) is set to logic level 1 again, and a number of status codes in I<sup>2</sup>CSTA are possible. The appropriate action to be taken for any of the status codes is detailed in <u>Table 82 on page 55</u>. After a repeated start condition (state 10h), the I<sup>2</sup>C interface may switch to the Master receiver mode by loading I<sup>2</sup>CDAT with SLA+R.

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### **Near Field Communication (NFC) controller**

#### 8.3.2.3 Master receiver mode

As a Master, the I<sup>2</sup>C logic will generate all of the serial clock pulses and the START and STOP conditions. A transfer is ended with a STOP condition or with a repeated START condition. Since a repeated START condition is also the beginning of the next serial transfer, the I<sup>2</sup>C bus will not be released.

The first byte transmitted contains the Slave address of the transmitting device (7-bit SLA) and the data direction bit. In this case the data direction bit (R/W) will be logic level 1 (R). I<sup>2</sup>C data are received via SDA while P50\_SCL outputs the serial clock. I<sup>2</sup>C data are received 8 bits at a time. After each byte is received, an acknowledge bit is transmitted. START and STOP conditions are output to indicate the beginning and end of a serial transfer.

In the Master receiver mode, a number of data bytes are received from a Slave transmitter. The transfer is initialized as in the Master transmitter mode. When the START condition has been transmitted, the interrupt service routine must load I<sup>2</sup>CDAT with the 7-bit Slave address and the data direction bit (SLA+R). The SI bit in I<sup>2</sup>CCON must then be set to logic 0 before the serial transfer can continue.

When the Slave address and the data direction bit have been transmitted and an acknowledgment bit has been received, the serial interrupt flag (SI) is set to logic level 1 again, and a number of status codes are possible in I<sup>2</sup>CSTA. The appropriate action to be taken for each of the status codes is detailed in <u>Table 83 on page 56</u>. After a repeated start condition (state 10h), the I<sup>2</sup>C interface may switch to the Master transmitter mode by loading I<sup>2</sup>CDAT with SLA+W.

# 8.3.2.4 I<sup>2</sup>CCON register

The CPU can read from and write to this 8-bit SFR. Two bits are affected by the Serial IO (the  $I^2C$  interface) hardware: the SI bit is set to logic level 1 when a serial interrupt is requested, and the STO bit is set to logic level 0 when a STOP condition is present on the  $I^2C$  bus. The STO bit is also set to logic level 0 when ENS1 = '0'.

Table 78. I<sup>2</sup>CCON register (SFR: address D8h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	CR[2]	ENS1	STA	STO	SI	AA	CR	[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

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# **Near Field Communication (NFC) controller**

Table 79. Description of I<sup>2</sup>CCON bits

Table	ra. Desi	cription of 1-ccon bits
Bit	Symbol	Description
7	CR[2]	Serial clock frequency selection in Master mode.
		Together with CR[1:0], this bit determines the clock rate (serial clock frequency) when the I <sup>2</sup> C interface is in a Master mode. Special attention has to be made on the I <sup>2</sup> C bit frequency in case of dynamic switching of the CPU clock frequency.
6	ENS1	Serial IO enable.
		When ENS1 bit is to logic level 0, SDA and P50_SCL are in high impedance. The state of SDA and P50_SCL is ignored, the I²C interface is in the "not addressed" Slave state, and the STO bit in I²CCON is forced to logic level 0. No other bits are affected.
		When ENS1 is logic level 1, the $I^2C$ interface is enabled, assuming selif[1:0] bits are 10b (see Table 74 on page 47).
		ENS1 should not be used to temporarily release the $I^2C$ interface from the $I^2C$ bus since, when ENS1 is set to logic level 0, the $I^2C$ bus status is lost. The AA flag should be used instead.
5	STA	START control.
		When the STA bit is set to logic level 1 to enter Master mode, the I <sup>2</sup> C interface hardware checks the status of the I <sup>2</sup> C bus and generates a START condition if the bus is free. If the bus is not free, then the I <sup>2</sup> C interface waits for a STOP condition (which will free the bus) and generates a START condition after a delay of a half clock period of the internal serial clock generator.
		If STA is set to logic level 1, while the I <sup>2</sup> C interface is already in a Master mode and one or more bytes are transmitted or received, the I <sup>2</sup> C interface transmits a repeated START condition.
		STA may be set to logic level 1 at any time. This includes the case when the $I^2C$ interface is the addressed Slave.
		When the STA bit is set to logic level 0, no START condition or repeated START condition will be generated.
4	STO	STOP control.
		When the STO bit is set to logic level 1, while the I <sup>2</sup> C interface is in Master mode, a STOP condition is transmitted to the I <sup>2</sup> C bus. When the STOP condition is

When the STO bit is set to logic level 1, while the  $I^2C$  interface is in Master mode, a STOP condition is transmitted to the  $I^2C$  bus. When the STOP condition is detected on the bus, the  $I^2C$  interface hardware automatically sets STO to logic level 0.

In Slave mode, STO may be set to logic 1 to recover from an error condition. In this case, no STOP condition is transmitted to the  $I^2C$  bus. However, the  $I^2C$  interface hardware behaves as if a STOP condition has been received and switches to the defined "not addressed" Slave Receiver mode.

If the STA and STO bits are both set to logic level 1, the STOP condition is transmitted to the  $I^2C$  bus if the  $I^2C$  interface is in Master mode (in Slave mode, the  $I^2C$  interface generates an internal STOP condition which is not transmitted). The  $I^2C$  interface then transmits a START condition.

When the STO bit is set to logic level 0, no STOP condition will be generated.

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### **Near Field Communication (NFC) controller**

Table 79. Description of I<sup>2</sup>CCON bits ... continued

# Bit Symbol Description

### 3 SI Serial interrupt flag.

When SI is set to logic level 1, then if the serial interrupt from the I<sup>2</sup>C interface port is enabled, the CPU will receive an interrupt. SI is set by hardware when any one of 25 of the possible 26 states of the I<sup>2</sup>C interface are entered. The only state that does not cause SI to be set to logic level 1 is state F8h, which indicates that no relevant state information is available.

While SI is set by hardware to logic level 1, P50\_SCL is held in logic 0 when the SCL line is logic level 0, and P50\_SCL is held in high impedance when the SCL line is logic level 1.

SI must be set to logic level 0 by firmware.

When the SI flag is set to logic level 0, no serial interrupt is requested, and there is no stretching of the SCL line via P50\_SCL.

The bit IE1\_4 of register IE1 (see <u>Table 13 on page 17</u>) has also to be set to logic level 1 to enable the corresponding I<sup>2</sup>C interrupt to the CPU.

#### 2 AA Assert Acknowledge flag.

If AA is set to logic level 1, an acknowledge (low level to SDA) will be returned during the acknowledge clock pulse on the P50\_SCL line when:

- The "own Slave address" has been received.
- The general call address has been received while the general call bit (GC) in I<sup>2</sup>CADR is set.
- A data byte has been received while the I<sup>2</sup>C interface is in Master Receiver mode.
- A data byte has been received while the I<sup>2</sup>C interface is in the addressed Slave Receiver mode.

When the I<sup>2</sup>C interface is in the addressed Slave Transmitter mode, state C8h will be entered after the last serial bit is transmitted. When SI is set to logic level 0, the I<sup>2</sup>C interface leaves state C8h, enters the Not-addressed Slave Receiver mode, and the SDA line remains at logic 1. In state C8h, AA can be set to logic level 1 again for future address recognition.

When the I $^2$ C interface is in the Not-addressed Slave mode, its own Slave address and the general call address are ignored. Consequently, no acknowledge is returned, and a serial interrupt is not requested. Thus, the I $^2$ C interface can be temporarily released from the I $^2$ C bus while the bus status is monitored. While the I $^2$ C interface is released from the bus, START and STOP conditions are detected, and I $^2$ C data are shifted in. Address recognition can be resumed at any time by setting AA to logic level 1.

If AA is set to logic level 1 when the  $\rm I^2C$  own Slave address or the general call address has been partly received, the address will be recognized at the end of the byte transmission.

Table 79. Description of I<sup>2</sup>CCON bits ...continued

			•		1 10110 1110011111111111111111111111111					
Bit	Symbol	Desc	riptio	n						
1 to 0	CR[1:0]	Serial clock frequency selection in Master mode.								
		CR2	CR1	CR0	CPU_CLK division factor	2C bit frequency				
		0	0	0	10	CPU_CLK/10				
		0	0	1	20	CPU_CLK/20				
		0	1	0	30	CPU_CLK/30				
		0	1	1	40	CPU_CLK/40				
		1	0	0	80	CPU_CLK/80				
		1	0	1	120	CPU_CLK/120				
		1	1	0	160	CPU_CLK/160				
		1	1	1	(256-T1 reload value)*12 24 3072	CPU_CLK/3072 CPU_CLK/24				

#### 8.3.2.5 I<sup>2</sup>CSTA register

I<sup>2</sup>CSTA is an 8-bit read-only special function register. The three least significant bits are always at logic 0. The five most significant bits contain the status code. There are 26 possible status codes. When I<sup>2</sup>CSTA contains F8h, no relevant state information is available and no serial interrupt is requested. Reset initializes I<sup>2</sup>CSTA to F8h. All other I<sup>2</sup>CSTA values correspond to defined I<sup>2</sup>C interface states. When each of these states is entered, a serial interrupt is requested (SI = '1'), this can happen in any CPU cycle, and a valid status code will be present in I<sup>2</sup>CSTA. This status code will remain present in I<sup>2</sup>CSTA until SI is set to logic 0 by firmware.

Note that I<sup>2</sup>CSTA changes one CPU\_CLK clock cycle after SI changes, so the new status can be visible in the same machine cycle SI changes or possibly (in one out of six CPU states) the machine cycle after that. This should not be a problem since you should not read I<sup>2</sup>CSTA before either polling SI or entry of the interrupt handler (which in itself takes several machine cycles).

Table 80. I<sup>2</sup>CSTA register (SFR: address D9h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				ST[	7:0]			
Reset	1	1	1	1	1	0	0	0
Access	R	R	R	R	R	R	R	R

Table 81. Description of I<sup>2</sup>CSTA bits

Bit	Symbol	Description
7 to 0	ST[7:0]	Encoded status bit for the different functional mode. Several Status codes are returned in a certain mode (Master Transmitter, Master Receiver, Slave Transmitter, Slave Receiver) plus some miscellaneous status codes that can be returned at any time.

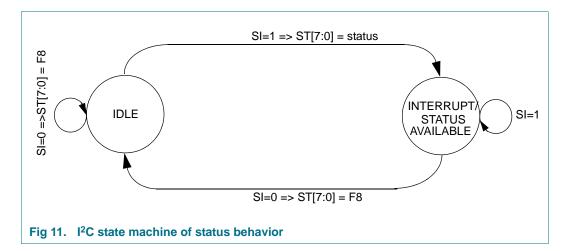


Table 82. I<sup>2</sup>C Master Transmitter Mode status code

9	Status	Status of the I <sup>2</sup> C Bus	Application f	irmwa	re Res	ponse	Next Action Taken By the I <sup>2</sup> C interface H	
	Code	and of the I <sup>2</sup> C interface Hardware	To/from I <sup>2</sup> CDAT		TO I <sup>2</sup>	CON		
	31[7.0]	I-C interface naroware		STA	STO	SI	AA	
	08h	A START condition has been transmitted	Load SLA+W	Х	0	0	Х	SLA+W will be transmitted ACK will be rece
	10h	A repeated START condition	Load SLA+W	Х	0	0	Х	As above
		has been transmitted	Load SLA+R	Х	0	0	Χ	SLA+W will be transmitted; the I <sup>2</sup> C interface will be switched to MST/(T
	18h	SLA+W has been transmitted;	Load data byte	0	0	0	Х	Data byte will be transmitted; ACK bit will b
		ACK has been received	No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
	A I		No I <sup>2</sup> CDAT action	0	1	0	Χ	STOP condition will be transmitted STO flag will be set to logic level 0
	in the second se		No I <sup>2</sup> CDAT action	1	1	0	Χ	STOP condition followed by a START cond STO flag will be set to logic level 0
	20h	SLA+W has been transmitted; NOT ACK has been received	Load data byte	0	0	0	Χ	Data byte will be transmitted ACK bit will be received
			No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
	D D D D D D D D D D D D D D D D D D D		No I <sup>2</sup> CDAT action	0	1	0	Χ	STOP condition will be transmitted STO flag will be set to logic level 0
			No I <sup>2</sup> CDAT action	1	1	0	Χ	STOP condition followed by a START cond STO flag will be set to logic level 0
9	28h	Write data byte in I <sup>2</sup> CDAT has	Load data byte	0	0	0	Х	Data byte will be transmitted; ACK bit will b
Š	8	been transmitted; ACK has been received	No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START will be transmitted
		ACK has been received	No I <sup>2</sup> CDAT action	0	1	0	Χ	STOP condition will be transmitted STO flag will be set to logic level 0
			No I <sup>2</sup> CDAT action	1	1	0	X	STOP condition followed by a START cond STO flag will be set to logic level 0

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Table 82. I<sup>2</sup>C Master Transmitter Mode status code ...continued

Status of the I <sup>2</sup> C Bus	Application f	irmwa	re Res	spons	Next Action Taken By the I <sup>2</sup> C interface H			
and of the	To/from I <sup>2</sup> CDAT		TO I <sup>2</sup> 0	CCON				
I'C interface naroware		STA	STO	SI	AA			
Write data byte in I <sup>2</sup> CDAT has	Load data byte	0	0	0	Х	Data byte will be transmitted; ACK bit will b		
been transmitted; NOT ACK has been received	No I <sup>2</sup> CDAT action	1	0	0	Χ	Repeated START will be transmitted		
	No I <sup>2</sup> CDAT action	0	1	0	X	STOP condition will be transmitted STO flag will be set to logic level 0		
	No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START cond STO flag will be set to logic level 0		
Arbitration lost in SLA+R/W or Data bytes	No I <sup>2</sup> CDAT action	0	0	0	X	I <sup>2</sup> C bus will be released; a Slave mode will		
	No I <sup>2</sup> CDAT action	1	0	0	X	A START condition will be transmitted when		
	and of the I <sup>2</sup> C interface Hardware  Write data byte in I <sup>2</sup> CDAT has been transmitted;  NOT ACK has been received  Arbitration lost in SLA+R/W or	and of the I²C interface Hardware  Write data byte in I²CDAT has been transmitted; NOT ACK has been received  Arbitration lost in SLA+R/W or Data bytes	and of the I²C interface Hardware  Write data byte in I²CDAT has been transmitted; NOT ACK has been received  Load data byte 0 No I²CDAT action 1 No I²CDAT action 0  No I²CDAT action 1  Arbitration lost in SLA+R/W or	and of the I²C interface Hardware         To/from I²CDAT         To I²C           Write data byte in I²CDAT has been transmitted; NOT ACK has been received         Load data byte 0 0 0 No I²CDAT action 1 0 No I²CDAT action 1 1 0 No I²CDAT action 1 1 1           Arbitration lost in SLA+R/W or Data bytes         No I²CDAT action 1 1 1	To/from I2CDAT   TO I2CCON	To/from I2CDAT		

Table 83. I<sup>2</sup>C Master Receiver Mode status codes

Status	Status of the I <sup>2</sup> C Bus and	Application f	irmwa	re Res	pons	е	Next Action Taken By the I <sup>2</sup> C interface F
Code ST[7:0]	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT		TO I <sup>2</sup> 0	CON		
31[7.0]			STA	STO	SI	AA	
08h	A START condition has been transmitted	Load SLA+W	Х	0	0	Х	SLA+W will be transmitted, ACK will be re
10h	A repeated START condition	Load SLA+W	Х	0	0	Х	As above
	has been transmitted	Load SL+R	X	0	0	Х	SLA+W will be transmitted; the I <sup>2</sup> C interfactor REC) mode
38h	Arbitration lost in SLA+R/W or Data bytes	No I <sup>2</sup> CDAT action	0	0	0	Х	I <sup>2</sup> C bus will be released; a Slave mode wi
		No I <sup>2</sup> CDAT action	1	0	0	Х	A START condition will be transmitted wh
40h	SLA+R has been transmitted; ACK has been received	No I <sup>2</sup> CDAT action	0	0	0	0	Data byte will be received; NOT ACK bit will be returned
		No I <sup>2</sup> CDAT action	0	0	0	1	Data byte will be received; ACK bit will be returned
48h	SLA+R has been transmitted;	No I <sup>2</sup> CDAT action	1	0	0	Х	Repeated START condition will be transm
	NOT ACK has been received	No I <sup>2</sup> CDAT action	0	1	0	Х	STOP condition will be transmitted; STO flag will be set to logic level 0
		No I <sup>2</sup> CDAT action	1	1	0	Х	STOP condition followed by a START cor STO flag will be set to logic level 0

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Table 83. I<sup>2</sup>C Master Receiver Mode status codes ...continued

3		Status of the I <sup>2</sup> C Bus and	Application f	irmwa	re Res	spons	Next Action Taken By the I <sup>2</sup> C interface F	
2	Code	the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT		TO I <sup>2</sup>	CCON		
	ST[7:0]			STA	STO	SI	AA	
	50h	Read data byte has been received; ACK has been	Read data byte or	0	0	0	0	Data byte will be received; NOT ACK bit will be returned
		returned	Read data byte	0	0	0	1	Data byte will be received; ACK bit will be returned
	58h	Read data byte has been	Read data byte	1	0	0	Х	Repeated START condition will be transmit
	received; NOT ACK has been returned	Read data byte	0	1	0	Х	STOP condition will be transmitted; STO flag will be set to logic level 0	
2			Read data byte	1	1	0	Х	STOP condition followed by a START cond STO flag will be set to logic level 0

# Table 84. I<sup>2</sup>C Miscellaneous status codes

	Status of the I <sup>2</sup> C Bus and	Application f	irmwa	re Res	pons	Next Action Taken By the I <sup>2</sup> C interface H		
Code I <sup>2</sup> CSTA		the I <sup>2</sup> C interface Hardware	To /from I <sup>2</sup> CDAT		TO I <sup>2</sup> 0	CCON		
	COIA			STA	STO	SI	AA	
	00h	Bus error	No I <sup>2</sup> CDAT action	Х	1	0	Χ	Hardware will enter the "not addressed" Sla
:	F8h	No information available	No I <sup>2</sup> CDAT action					

#### 8.3.2.6 I<sup>2</sup>CDAT register

I<sup>2</sup>CDAT contains a byte of I<sup>2</sup>C data to be transmitted or a byte which has just been received. The CPU can read from and write to this 8-bit SFR while it is not in the process of shifting a byte. This occurs when the I<sup>2</sup>C interface is in a defined state and the serial interrupt flag SI is set to logic level 1. Data in I<sup>2</sup>CDAT remains stable as long as SI is set to logic level 1. The first bit to be transmitted is the MSB (bit 7), and, after a byte has been received, the first bit of received data is located at the MSB of I<sup>2</sup>CDAT. While data is being shifted out, data on the bus is simultaneously being shifted in; I<sup>2</sup>CDAT always contains the last data byte present on the bus. Thus, in the event of lost arbitration, the transition from Master Transmitter to Slave Receiver is made with the correct data in I<sup>2</sup>CDAT.

Table 85. I<sup>2</sup>CDAT register (SFR: address DAh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		I <sup>2</sup> CDAT[7:0]							
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 86. Description of I<sup>2</sup>CDAT bits

Bit	Symbol	Description
7 to 0	I <sup>2</sup> CDAT[7:0]	<b>I2C data.</b> Eight bits to be transmitted or just received. A logic level 1 in I <sup>2</sup> CDAT corresponds to a logic level 1 on the I <sup>2</sup> C bus, and a logic level 0 corresponds to a logic level 0 on the bus. I <sup>2</sup> C data shift through I <sup>2</sup> CDAT from right to left.

I<sup>2</sup>CDAT[7:0] and the ACK flag form a 9-bit shift register which shifts in or shifts out an 8-bit byte, followed by an acknowledge bit. The ACK flag is controlled by the I<sup>2</sup>C interface hardware and cannot be accessed by the CPU. I<sup>2</sup>C data are shifted through the ACK flag into I<sup>2</sup>CDAT on the rising edges of clock pulses on P50\_SCL. When a byte has been shifted into I<sup>2</sup>CDAT, the I<sup>2</sup>C data are available in I<sup>2</sup>CDAT, and the acknowledge bit is returned by the control logic during the ninth clock pulse. I<sup>2</sup>C data are shifted out from I<sup>2</sup>CDAT via a buffer on the falling edges of clock pulses on P50\_SCL.

When the CPU writes to I<sup>2</sup>CDAT, the buffer is loaded with the contents of I<sup>2</sup>CDAT[7] which is the first bit to be transmitted to the SDA line. After nine serial clock pulses, the eight bits in I<sup>2</sup>CDAT will have been transmitted to the SDA line, and the acknowledge bit will be present in ACK. Note that the eight transmitted bits are shifted back into I<sup>2</sup>CDAT.

#### 8.3.2.7 I<sup>2</sup>CADR register

The CPU can read from and write to this 8-bit SFR. I<sup>2</sup>CADR is not affected by the I<sup>2</sup>C interface hardware. The content of this register is irrelevant when the I<sup>2</sup>C interface is in a Master mode. In the Slave modes, the seven most significant bits must be loaded with the microcontroller's own Slave address, and, if the least significant bit is set to logic level 1, the general call address (00h) is recognized; otherwise it is ignored.

Table 87. I<sup>2</sup>CADR register (SFR: address DBh) bit allocation

Bit	7	6	5	4	3	2	1	0		
Symbol		SA[6:0]								
Reset	0	0	0	0	0	0	0	0		
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		

Table 88. Description of I<sup>2</sup>CADR bits

Bit	Symbol	Description
7 to 1	SA[6:0]	Slave address.
		These bits correspond to the 7-bit Slave address which will be recognized on the incoming data stream from the I <sup>2</sup> C bus. When the Slave address is detected and the interface is enabled, a serial interrupt SI will be generated to the CPU.
0	GC	General call.
		When set to logic level 1, will cause the I <sup>2</sup> C logic to watch for the general call address to be transmitted on the I <sup>2</sup> C bus. If a general call address is detected and this bit is set to logic level 1, SI will be set to logic level 1.

#### 8.3.2.8 I<sup>2</sup>C\_wu\_control register

The wake up block has to be enabled before the whole chip enters in Soft-Power-down mode. The choice of the wake-up conditions is made within the register I<sup>2</sup>C\_wu\_control. Read and Write conditions can be set together.

Table 89. I<sup>2</sup>C\_wu\_control register (address 610Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	i <sup>2</sup> c_wu_en_wr	i <sup>2</sup> c_wu_en_rd	i <sup>2</sup> c_wu_en
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R/W	R/W	R/W

Table 90. Description of I<sup>2</sup>C\_wu\_control bits

Bit	Symbol	Description
7 to 3	-	Reserved.
2	i <sup>2</sup> c_wu_en_wr	When set to logic level 1, the wake-up is valid for write commands
1	i <sup>2</sup> c_wu_en_rd	When set to logic level 1, the wake-up is valid for read commands
0	i <sup>2</sup> c_wu_en	When set to logic level 1, enable the I <sup>2</sup> C wake-up conditions

### 8.3.3 FIFO manager

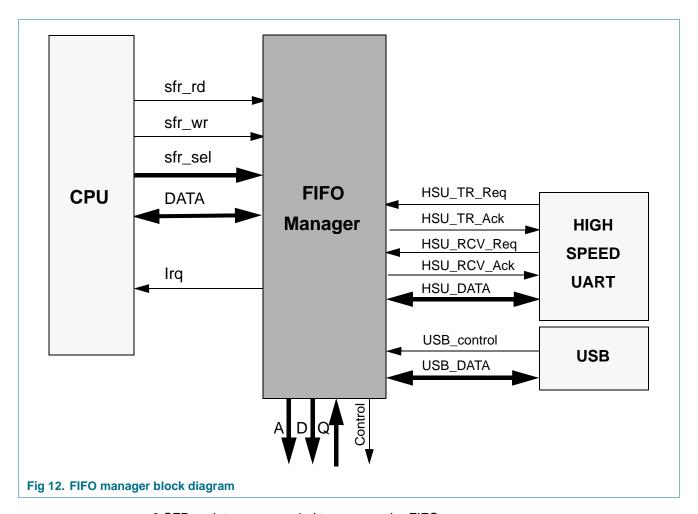
This block is designed to manage a RAM as a FIFO in order to optimize the data exchange between the CPU and the HOST.

#### 8.3.3.1 FIFO manager functional description

The RAM used for the FIFO is shared between the SPI and HSU interfaces. Indeed, these interfaces cannot be used simultaneously. The selection of the interface used is done by firmware. The FIFO manager block is the common part between the USBB and the HSU interfaces. It consists of a Data register, a Status register and also some registers to define the characteristics of the FIFO. These registers are addressed by the CPU as SFRs.

The RAM used as a FIFO is divided into two part: a receive part and a transmit part.

This block also manages the possible conflicts existing around the FIFO between the CPU and the interfaces. Indeed, a request coming from the interface (TR\_req or RCV\_req) can be simultaneous with a request to access to the data register coming from the CPU.



9 SFR registers are needed to manage the FIFO manager.

Table 91. Fifo manager SFR register list

Name	Size [bytes]	SFR Address	Description	Access
RWL	1	9Ah	FIFO Receive Waterlevel: Controls the threshold of the FIFO in reception	R/W
TWL	1	9Bh	FIFO Transmit Waterlevel: Controls the threshold of the FIFO in transmission	R/W
FIFOFS	1	9Ch	FIFO Transmit FreeSpace: Status of the number of characters which can still be loaded in the FIFO	R/W
FIFOFF	1	9Dh	FIFO Receive Fullness: Status of the number of received characters in the FIFO	R/W
SFF	1	9Eh	Global Status/Error messages	R
FIT	1	9Fh	Interrupt Source	R/W
FITEN	1	A1h	Interrupt Enable and Reset FIFO	R
FDATA	1	A2h	Data reception/transmission buffer	R/W
FSIZE	1	A3h	Control the size of the FIFO in Reception	R/W

#### 8.3.3.2 RWL register

This register defines the warning level of the Receive FIFO for the CPU. It implies a FIFO buffer overflow.

Table 92. RWL register (SFR: address 9Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				RWaterl	evel[7:0]			
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 93. Description of RWL bits

Bit	Symbol	Description
7 to 0	RWaterlevel[7:0]	Overflow threshold of the Receive FIFO to set a warning

### 8.3.3.3 TWL register

This register defines the warning level of the Transmit FIFO for the CPU. It implies a FIFO buffer underflow.

Table 94. TWL register (SFR: address 9Bh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		TWaterlevel[7:0]							
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 95. Description of TWL bits

Bit	Symbol	Description
7 to 0	TWaterlevel[7:0]	Underflow threshold of the Transmit FIFO to set a warning

### 8.3.3.4 FIFOFS register

This register indicates the number of bytes that the CPU can still load into the FIFO until the Transmit FIFO is full.

Table 96. FIFOFS register (SFR: address 9Ch) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		TransmitFreespace[7:0]							
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 97. Description of FIFOFS register bits

Bit	Symbol	Description
7 to 0	TransmitFreespace[7:0]	Freespace into the FIFO

### 8.3.3.5 FIFOFF register

This register indicates the number of bytes already received and loaded into the Receive FIFO.

Table 98. FIFOFF register (SFR: address 9Dh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		ReceiveFullness[7:0]							
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 99. Description of FIFOFF bits

Bit	Symbol	Description
7 to 0	ReceiveFullness[7:0]	Number of bytes received in the FIFO

### 8.3.3.6 SFF register

The register bits are used to allow the CPU to monitor the status of the FIFO. The primary purpose is to detect completion of data transfers.

Table 100. SFF register (SFR: address 9Eh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	FIFO_EN	-	TWLL	TFF	TFE	RWLH	RFF	RFE
Reset	0	0	1	0	1	0	0	1
Access	R/W	R	R	R	R	R	R	R

#### Table 101. Description of SFF bits

Bit	Symbol	Description
7	FIFO_EN	Fifo Enable:
		Set to logic 1 this bit enables the FIFO manager clock (CPU_CLK).
		Set to logic 0 the clock remains low.
6	-	Reserved.
5	TWLL	Transmit WaterlLevelLow:
		This bit is set to logic 1 when the number of bytes stored into the Transmit FIFO is equal or smaller than the threshold TWaterlevel.
4	TFF	Transmit FIFO Full:
		This is set to logic 1 if the transmit part of the FIFO is full. It is set to logic level 0 when a transfer is completed.
3	TFE	Transmit FIFO Empty:
		This bit indicates when the transmit part of the FIFO is empty.
		It is set to logic 0 when the CPU writes a character in the data register.
2	RWLH	Receive WaterLevel High:
		This bit is set to logic 1 when the number of bytes stored into the Receive FIFO is greater or equal to the threshold RWaterlevel.
1	RFF	Receive FIFO Full:
		This bit is set to logic 1 if the receive part of the FIFO is full. It is set to logic level 0 by reading the FDATA register.
0	RFE	Receive FIFO Empty:
		This bit indicates when the receive part of the FIFO is empty.
		Set to logic 1, when the Receive FIFO is empty.
		Set to logic level 0, when the Receive FIFO contains at least 1 byte.

# 8.3.3.7 FIT register

The FIT register contains 6 read-write bits which are logically OR-ed to generate an interrupt going to the CPU.

Table 102. FIT register (SFR: address 9Fh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Reset	-	WCOL_ IRQ	TWLL_ IRQ	TFF_ IRQ	RWLH_ IRQ	ROVR_ IRQ	RFF_ IRQ
Reset	0	0	0	0	0	0	0	0
Access	W	R	R/W	R/W	R/W	R/W	R/W	R/W

### Table 103. Description of FIT bits

Bit	Symbol	Description
7	Reset	Reset:
		Set to logic 1, Reset defines that the bits set to logic level 1 in the write command are set to logic level 0 in the register.
6	-	Reserved
5	WCOL_IRQ	Write COLlision IRQ:
		This bit is set to logic 1 when the transmitted part of the FIFO is already full (TFF is set to logic 1) and a new character is written by the CPU in the data register.
4	TWLL_IRQ	Transmit WaterILevelLow IRQ:
		This bit is set to logic 1 when the number of bytes stored into the Transmit FIFO is equal or smaller than the threshold TWaterlevel.
3	TFF_IRQ	Transmit FIFO Full IRQ:
		This is set to logic 1 if the transmitted part of the FIFO is full.
2	RWLH_IRQ	<b>Receive WaterLevel High IRQ:</b> This bit is set to logic 1 when the number of bytes stored into the Receive FIFO is greater or equal to the threshold RWaterlevel.
1	ROVR_IRQ	<b>Read OVeRrun IRQ:</b> This bit indicates that a read overrun has occured.It occurs when the receiver part of the FIFO is full and a new data transfer is completed. Then the new received data is lost and ROVR_IRQ is set.
0	RFF_IRQ	<b>Receive FIFO Full IRQ:</b> This bit is set to logic 1 if the received part of the FIFO is full.

# 8.3.3.8 FITEN register

The FITEN register enables or disables the interrupt requests to the CPU. It is also used to reset the content of the Receive and Transmit FIFO.

Table 104. FITEN register (SFR: address A1h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TFLUSH	RFLUSH	EN_ WCOL_ IRQ	EN_ TWLL_ IRQ	EN_ TFF_ IRQ	EN_ RWLH_ IRQ	EN_ ROVR_ IRQ	EN_ RFF_ IRQ
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 105. Description of FITEN bits

Bit	Symbol	Description
7	TFLUSH	When set to logic level 1, the pointer of the Transmit FIFO is reset. This bit and RFLUSH must not be set at the same time.
6	RFLUSH	When set to logic level 1, the pointer of the Receive FIFO is reset. This bit and TFLUSH must not be set at the same time but one after the other.
5	EN_WCOL_IRQ	ENable Write COLlision IRQ:
		When set to logic 1, the WCOL_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
4	EN_TWLL_IRQ	ENable Transmit WaterILevelLow IRQ:
		When set to logic 1, the TWLL_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
3	EN_TFF_IRQ	ENable Transmit FIFO Full IRQ:
		When set to logic level 1, the TFF_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
2	EN_RWLH_IRQ	ENable Receive WaterLevel High IRQ:
		When set to logic 1, the RWLH_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
1	EN_ROVR_IRQ	ENable Read OVeRrun IRQ:
		When set to logic 1, the ROVR_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
0	EN_RFF_IRQ	ENable Receive FIFO Full IRQ:
		When set to logic 1, the RFF_IRQ is enabled.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.

### 8.3.3.9 FDATA register

The FDATA register is used to provide the transmitted and received data bytes. Each data written in the data register is pushed into the Transmit FIFO. Each data read from the data register is popped from the Receive FIFO.

Table 106. FDATA register (SFR: address A2h) bit allocation

Bit	7	6	5	4	3	2	1	0
	•	J	J	-T	.417.01	_	•	J
Symbol				FDAI	A[7:0]			
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 107. Description of FDATA bits

Bit	Symbol	Description
7 to 0	FDATA[7:0]	Writing to FDATA writes to the transmit buffer.
		Reading from FDATA reads from the receive buffer.

### 8.3.3.10 FSIZE register

This register defines the size of the Receive FIFO. The maximum size is 182 bytes. The free space not used by the Receive FIFO in the RAM will be allocated to Transmit FIFO.

Table 108. FSIZE register (SFR: address A3h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		ReceiveSize[7:0]							
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 109. Description of FSIZE bits

Bit	Symbol	Description
7 to 0	ReceiveSize[7:0]	Size of the Receive FIFO

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### **Near Field Communication (NFC) controller**

# 8.3.4 High Speed UART (HSU)

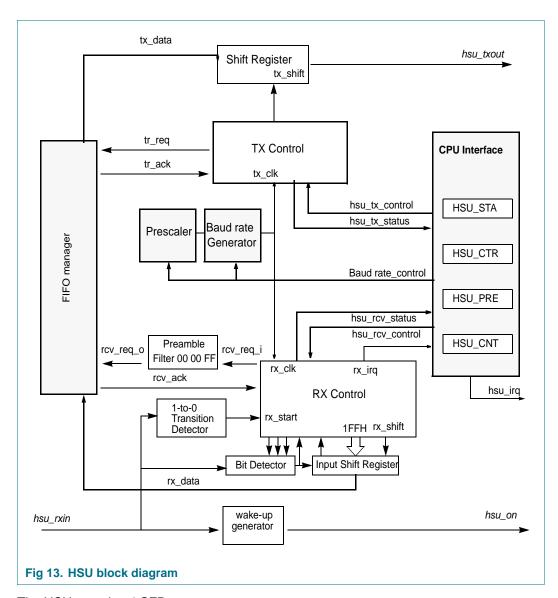
The High Speed UART (HSU) provides a high speed link to the host (up to 1.288 Mbit/s).

The HSU is a full duplex serial port. The serial port has a Receive-buffer: in conjunction with the FIFO manager, the reception of several bytes can be performed without strong CPU real time constraints. However, if the Receive FIFO still has not been read by the CPU, and the number of receive bytes is greater than the Receive FIFO size then the new incoming bytes will be lost.

The HSU receive and transmit data registers are both accessed by firmware in the FIFO manager FDATA register. Writing to FDATA loads the transmit register, reading from FDATA accesses the separate receive register.

The characteristics of the UART are the following:

- Full duplex serial port
- Receive buffer to allow reception of byte while the previous bytes are stored into the FIFO manager
- 8-bit data transfers
- Programmable baud rate generator using prescaler for transmission and reception
- Based on 27.12 MHz clock frequency
- · Dedicated protocol preamble filter
- · Wake-up generator



The HSU contains 4 SFRs:

Table 110. HSU SFR register list

Name	Size [bytes]	SFR Address	Description	Access
HSU_STA	1	ABh	HSU STAtus register	R/W
HSU_CTR	1	ACh	HSU ConTRol register	R/W
HSU_PRE	1	ADh	HSU PREscaler for baud rate generator	R/W
HSU_CNT	1	AEh	HSU CouNTer for baud rate generator	R/W

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### **Near Field Communication (NFC) controller**

#### 8.3.4.1 Mode of operation

The HSU supports only one operational mode, which has the following characteristics:

- Start bit:
  - Start bit is detected when a logic level 0 is asserted on the HSU\_RX line.
- 8 data bits
  - The data bits are sent or received LSB first.
- Stop bit:
  - During reception, the Stop bit(s) is detected when all the data bits are received and when Stop bit(s) is sampled to logic level 1. The number of Stop bits is programmable. It can be 1 or 2.
  - During Transmission, after the complete data bit transmission, a variable number of Stop bit(s) is transmitted. This number is programmable from 1 to 4.

#### 8.3.4.2 HSU Baud rate generator

To reach the high speed transfer rate, the HSU has it own baud rate generator. The baud rate generator comprises a prescaler and a counter. The prescaler is located before the counter. The purpose of the prescaler is to divide the frequency of the count signal to enlarge the range of the counter (at the cost of a lower resolution). The division factor of the prescaler is equal to 2 to the power HSU\_PRE[8:0] (Table 113 on page 71), resulting in division factors ranging from 1 (20) to 256 (28). The combination of these 2 blocks defines the bit duration and the bit sampling.

#### 8.3.4.3 HSU preamble filter

Received characters are sent to the FIFO manager after three consecutive characters have been received: 00 00 FF. When the frame is finished, and before a new frame arrives, firmware shall write a logic level 1 in the start\_frame bit of the HSU\_CTR register to re-activate the preamble filter. If firmware does not write a logic level 1 then all characters of the frame are sent to the FIFO manager (including the preamble).

#### 8.3.4.4 HSU wake-up generator

The wake-up generator is a 3-bit counter which counts on every rising edge of the HSU\_RX pin. When the counter reaches 5, the hsu\_on signal is set to logic level 1 in order to wake up the PN533. This block is useful in Soft-Power-down mode. The firmware shall reset this counter just before going in Soft-Power-down by writing a logic level 1 in the hsu\_wu\_en bit into the HSU\_CTR register.

# 8.3.4.5 HSU\_STA register

The SFR HSU\_STA is the status register of the HSU.

Table 111. HSU\_STA register (SFR: address ABh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	set_bit	-	-	disable_ preamb	irq_rx_ over_en	irq_rx_fer _ en	irq_rx_ over	irq_rx_fer
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	R/W	R/W	R/W	R/W	R/W

### Table 112. Description of HSU\_STA bits

Bit	Symbol	Description
7	set_bit	When set to logic 0 during write operation, the bits set to logic 1 in the write command are written to logic level 0 in the register.
		When set to logic 1 during write operation, the bits set to logic 1 in the write command are written to logic level 1 in the register.
6 to 5	-	Reserved
4	disable_preamb	Preamble filter disable.
		When set to logic 1, this bit disables the preamble filtering, it means that HSU_RX line transmit any received bytes to the FIFO manager.
3	irq_rx_over_en	FIFO overflow interrupt enable.
		When set to logic 1, this bit enables the interrupt generation when the bit irq_rx_over is set to logic 1.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
2	irq_rx_fer_en	Framing error interrupt enable.
		When set to logic 1, this bit enables the interrupt generation when the bit irq_rx_fer is set to logic 1.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
1	irq_rx_over	Receive FIFO overflow interrupt.
		Set to logic 1 when the FIFO manager is full (rcv_ack is set to logic level 0) and when HSU shift register is ready to send another byte to the FIFO manager.
0	irq_rx_fer	Framing error interrupt.
		Set to logic 1 when a framing error has been detected. Framing error detection is based on Stop bit sampling.
		When Stop bit is expected at logic level 1 but is sampled at logic level 0, this bit is set to logic level 1.

# 8.3.4.6 HSU\_CTR register

This register controls the configuration of the HSU.

Table 113. HSU\_CTR register (SFR: address ACh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	hsu_wu_ en	start_frame	tx_stop	bit[1:0]	rx_stopbit	tx_en	rx_en	soft_reset_n
Reset	0	0	0	0	0	0	0	1
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 114. Description of HSU\_CTR bits

Bit	Symbol	Description
7	hsu_wu_en	HSU wake-up enable.
		When set to logic 1 this bit re-activates the NSS / SCL / HSU_RX rising-edge counter. When the counter is 5 then a signal hsu_on is activated. This signal is one of the possible wake-up events from Soft-Power-down mode in the PCR block.
		The firmware shall set this bit to logic level 1 just before requesting a Soft-Power-down mode.
		The bit IE1_5 of register IE1 (see <u>Table 13 on page 17</u> ) has also to be set to logic level 1 to enable the corresponding CPU interrupt.
6	start_frame	Enables the preamble filter for next frame.
		When set to logic 1 this bit indicates that a new frame is coming. This re-activates the preamble filter (when enabled), meaning that the first "00 00 FF" characters will not be sent to the FIFO manager.
5:4	tx_stopbit[1:0]	Defines the number of stop bit during transmission.
		These 2 bits define the number of Stop bit(s) inserted at the end of the transmitted frame.
		The number of Stop bit(s) transmitted is equal to tx_stopbit +1.
3	rx_stopbit	Defines the number of stop bit during reception.
		This bit defines the number of Stop bit(s) inserted at the end of the received frame.
		The number of Stop bit(s) expected in reception is equal to rx_stopbit +1.
2	tx_en	Enables the transmission of HSU.
		When set to logic 1 this bit enables the transmission of characters.
		When set to logic 0, the transmission is disabled only after the completion of the current transmission.
1	rx_en	Enables the reception of the HSU.
		When set to logic 1 this bit enables the reception of characters.
		When set to logic 0, the reception is disabled only after the completion of the current reception.
0	soft_reset_n	HSU Reset.
		When set to logic 0, this bit disables the clock of the HSU_RX control, HSU_TX control and baud rate generator modules.

### 8.3.4.7 HSU\_PRE register

This register is used to configure the baud rate generator prescaler. The prescaler enlarges the range of the counter (at the cost of a lower resolution). The division factor of the prescaler ranges from 1  $(2^0)$  to 256  $(2^8)$ .

Table 115. HSU\_PRE register (SFR: address ADh) bit allocation

Bit	7	6	5	<u>,</u>	3	2	1	n
	•	J	J	. 7			•	U
Symbol	hsu_prescaler[7:0]							
Reset	0	0	0	1	1	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 116. Description of HSU\_PRE bits

Bit	Symbol	Description		
7 to 0	hsu_prescaler[7:0]	In conjunction with HSU_CNT, defines the HSU baud rate.		
		Baud rate = f <sub>clk</sub> / ((hsu_prescaler +1) * hsu_counter)		

### 8.3.4.8 HSU\_CNT register

This register is used to configure the baud rate generator counter.

Table 117. HSU\_CNT register (SFR: address AEh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	hsu_counter[7:0]							
Reset	0	1	1	1	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 118. Description of HSU\_CNT bits

Bit	Symbol	Description
7 to 0	hsu_counter[7:0]	In conjunction with HSU_PRE, defines the HSU baud rate.
		Baud rate = f <sub>clk</sub> / ((hsu_prescaler +1) * hsu_counter)

Here is a table of recommendation for some data rates:

Table 119. Recommendation for HSU data rates

Targeted data rate	HSU_CNT value	HSU_PRE value	Real HSU freq	Min recommended Host HSU freq	Max recommended Host HSU freq
9 600	0x71	0x18	9 516	9 326	9 706
19 200	0x9D	0x08	19 193	18 810	19 576
38 400	0x65	0x06	38 359	37 592	39 126
57 600	0x9D	0x02	57 579	56 428	58 730
115 200	0xEB	0x00	115 404	113 096	117 712
230 400	0x76	0x00	229 831	225 234	234 427
460 800	0x3B	0x00	459 661	450 467	468 854
921 600	0x1D	0x00	935 172	916 468	953 875
1 288 000	0x15	0x00	1 291 429	1 265 600	1 317 257

### **Near Field Communication (NFC) controller**

#### 8.3.5 USB

The USB module is an USB2.0 compliant device with embedded function. Special power management features such as a clock divider and clock switch are also implemented in the device.

#### 8.3.5.1 Features list

The USB module is a USB device only supporting full speed communication scheme. All embedded functions are passed to the micro controller.

The USB module includes the following features:

- SoftConnect supported
- Command GetFrameNumber supported
- Interrupt signaling to mController
- · Control EP0 endpoint of 8 bytes
- 3 interrupt endpoints of 8 bytes
- 2 bidirectional bulk endpoints of 64 bytes
- resume by host
- · remote wake up

The endpoints setup is described in the following table.

Table 120. Endpoints Setup

<b>Logical Endpoints</b>	OUT Type	IN Type	OUT Map	IN Map	OUT size	IN size	Note	
Device function description								
0	Control	Control	0.0	0.1	8	8		
1	-	Int	-	0.2		8		
2	-	Int	-	0.3		8		
3	-	Int	-	0.4		8		
4	Bulk	Bulk	0.5	0.6	64	64		

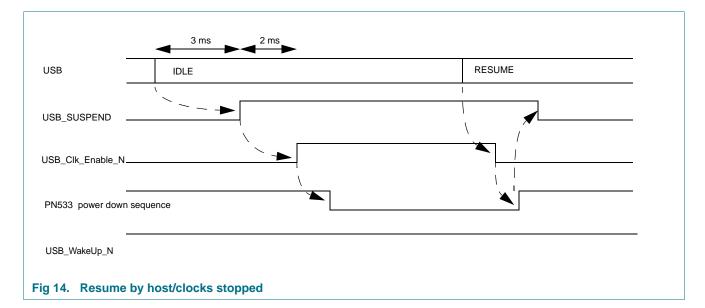
### 8.3.5.2 USB interrupt

The interrupt line of the USB module is asserted to indicates to the microcontroller that there was a transaction on one of the endpoints, or that there is new status information available.

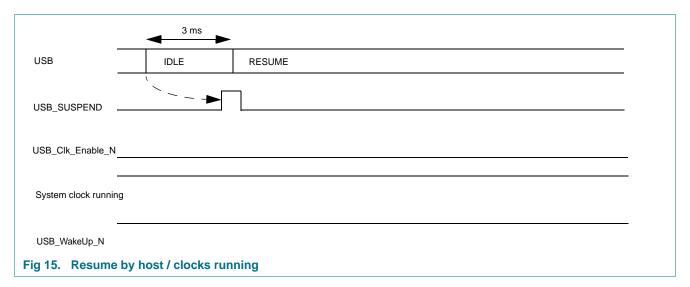
#### 8.3.5.3 Resume by host

Resume by host in suspend state: After 3 ms of no USB activity on the bus, the device goes in suspend. About 2 ms later the device will indicate that it no longer needs the clock (Clk\_Enable\_N becomes inactive).

The host can then send a resume to the device. The device needs to wake up and will require the main clock (USB\_Clk\_Enable\_N becomes active). The main clock starts running after complete PCR wake up sequence. PLL\_LOCK indicates that the main clock is running stable(a PLL can need several ms to start running at the specified frequency). The device then knows that the clock is present and can go out of suspend.



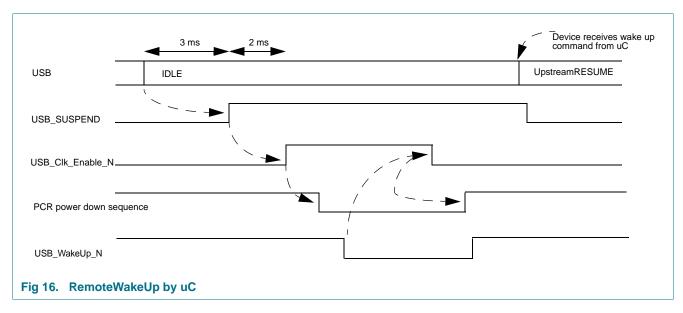
**Resume by Host before clock is disabled:** The host can decide to wake up a suspended device, before the device has switched off its main clock. The device then simply goes out of suspend.



#### 8.3.5.4 Remote wake up

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The remote wake up implemented in the PN533 platform is partly controlled by the micro controller as depicted below:



When a remote wake up is requested on an external interrupt (P32\_INT0, P33\_INT1) or a RF detector event, the Power Clock and Reset controller (PCR):

- enables the 27.12 MHZ oscillator,
- generates an interrupt to resume the CPU from Power-down mode,
- then the CPU enable the 4 MHZ oscillator, the PLL and the 48 MHZ clock.

The CPU sends a command to the USB module to perform a USB remote wake up, then the USB module exits from suspend mode and sends a resume on its upstream port.

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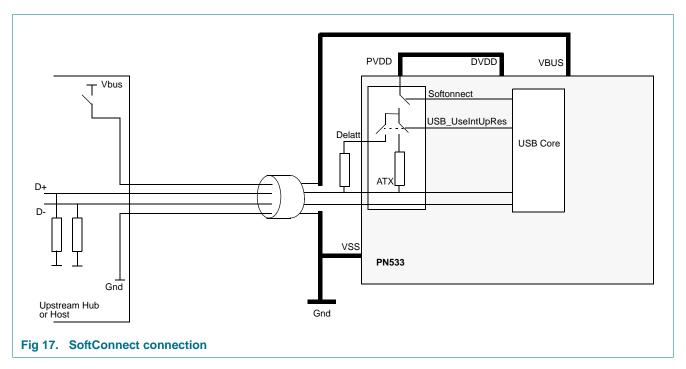
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Resume by Remote Wake up before clock is disabled: Resume by Remote Wake up before clock is disabled

If USB\_WakeUp\_N is made active before the main clock is switched off (USB\_Clk\_Enable\_N becomes inactive), the device waits until it has been suspended for 2 ms, then wakes up and sends a resume on its upstream port.

#### 8.3.5.5 Softconnect

The following figure shows how the embedded firmware Controlled Connect can be implemented.



When USB\_SoftConnect\_N is active, one of the data lines (D+) must be pulled high. This is done by using internal switch.

The internal pull up resistor or an external resistor (connected to delatt) can be used to perform the soft connection. The selection of the pull-up resistor to use is made through the USB configuration register. Refer to

#### Table 121 on page 76

Table 121. USB configuration register (6000h)

Bit	Name	Description	Reset	R/W
7 to 6	-	Reserved	XX	R/W
5	USB_UseIntUpRes	It controls switch for internal Upstream resistor when set to logic level 1; external pull up resistor is used. When set to logic level 0; internal pull up resistor is used.	0	R/W
4 to 0	-	Reserved	XXXXX	R/W

#### 8.3.5.6 USB embedded firmware view

The USB module is mapped into the XRAM memory space. It is accessible into the peripheral area on the host if internal bus. The communication between the CPU and the USB module is based on a sequence of command and data exchange.

Table 122. USB Extension memory map

Physical	Physical Address		Description	Peripheral selected	
First	Last	(Bytes)			
0x6003	0x6003	1	Write command to USB module	USB	
0x6002	0x6002	1	Write data to USB module	USB	
0x6001	0x6001	1	Read data from USB	USB	
0x6000	0x6000	1	USB configuration	USB	

**USB** Instruction set: The USB instruction set is described here after.

Table 123. USB Instruction set

Name	Recipient	Coding	Data Phase
Get Chip ID	Device	FDh	Read 2 bytes
Get ErrorCode	Device	FFh	Read 1 byte
Get Device Status	Device	FEh	Read 1 byte
Set Device Status	Device	FEh	Write 1 byte
Get Current Frame Number	Device	F5h	Read 2 bytes
Get Interrupt Register	Device	F4h	Read 2 bytes
Set Mode	Device	F3h	Write 2 bytes
Set Endpoints Enable	Device	D8h	Write 2 bytes
Set Address / Enable	Embedded Function	D0h	Write 1 byte
Get Embedded Port Status	Embedded Function	E0h	Read 1 byte
Set Embedded Port Status	Embedded Function	E0h	Write 1 byte
Select Endpoint	Function Control OUT	00h	Read 1 byte (opt)
	Function Control IN	01h	Read 1 byte (opt)
	Function Endpoint 1 IN	02h	Read 1 byte (opt)
	Function Endpoint 2 IN	03h	Read 1 byte (opt)
	Function Endpoint 3 IN	04h	Read 1 byte (opt)
	Function Endpoint 4 OUT	05h	Read 1 byte (opt)
	Function Endpoint 4 IN	06h	Read 1 byte (opt)
	Reserved	07h	-
	Reserved	08h	-
	Reserved	09h	-
Select Endpoint / Clear Interrupt	Function Control OUT	40h	Read 1 byte
	Function Control IN	41h	Read 1 byte
	Function Endpoint 1 IN	42h	Read 1 byte

Table 123. USB Instruction set ... continued

Name	Recipient	Coding	Data Phase
	Function Endpoint 2 IN	43h	Read 1 byte
	Function Endpoint 3 IN	44h	Read 1 byte
	Function Endpoint 4 OUT	45h	Read 1 byte
	Function Endpoint 4 IN	46h	Read 1 byte
	Reserved	47h	-
	Reserved	48h	-
	Reserved	49h	-
Set Endpoint Status	Function Control OUT	40h	Write 1 byte
	Function Control IN	41h	Write 1 byte
	Function Endpoint 1 IN	42h	Write 1 byte
	Function Endpoint 2 IN	43h	Write 1 byte
	Function Endpoint 3 IN	44h	Write 1 byte
	Function Endpoint 4 OUT	45h	Write 1 byte
	Function Endpoint 4 IN	46h	Write 1 byte
	Reserved	47h	-
	Reserved	48h	-
	Reserved	49h	-
Read Buffer	Selected Endpoint	F0h	Read n bytes
Write Buffer	Selected Endpoint	F0h	Write n bytes
Clear Buffer	Selected Endpoint	F2h	Read 1 byte (opt)
Validate Buffer	Selected Endpoint	FAh	none

Get VChip ID: Command: FDh

Data: Read 2 bytes

The Chip Identification is 12 bits wide. The command divides the chip Identification in bytes and returns the least significant byte first. The value of this chip ID can be determined at integration time.

The following table shown the configuration of these 2 bytes:

Table 124. Get Chip ID bytes

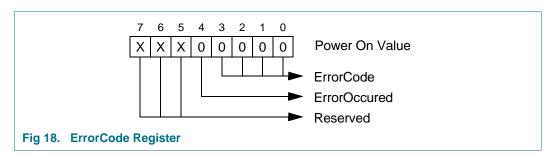
<b>Bit Position</b>	7	6	5	4	3	2	1	0	
Byte 0		DEVREV[1]							
Byte 1	0	0	0	0	DEVNAME[2]				

<sup>[1]</sup> hardware setting (8 bits) same as DEVREV, see "Device Descriptor".

<sup>[2]</sup> hardware setting (4 bits) same as DEVNAME, see "String Descriptor"

Get ErrorCode: Command: FFh

Data: Read 1 byte



Note that this is a debug command and should not be used for normal operation.

The 'Get Error Code command returns the error code of the last generated error, this command is for debugging purposes only. The 4 least significant bits form the error code. Bit 4 'Error Occurred' can be cleared by each new transfer.

The following table gives an overview of the Error Codes.

Table 125. Error codes

Error Code	Description
0000	No Error
0001	PID Encoding Error
0010	Unknown PID
0011	Unexpected Packet
0100	Error in Token CRC
0101	Error in Data CRC
0110	Time Out Error
0111	Babble
1000	Error in End of Packet
1001	Sent NAK
1010	Sent Stall
1011	Buffer Overrun Error
1100	Sent Empty Packet (ISO only)
1101	Bitstuff Error
1110	Error in Sync
1111	Wrong Toggle Bit in Data PID, ignored data

Get Device Status: Command: FEh

Data: Read 1 byte

The Get Device Status command returns the Device Status Register. Cf. the Set Device Status command.

When SuspendChange, ConnectChange and BusReset bit are set, the appropriate bit in the interrupt register is set and an interrupt is generated to the micro-controller.

The BusReset, SuspendChange and ConnectChange bit are reset by this command.

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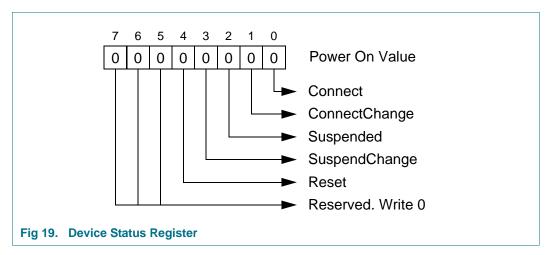
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### **Near Field Communication (NFC) controller**

Set Device Status: Command: FEh

Data: Write 1 byte

The Set Device Status command changes the Device Status Register. The value of Read Only bits is ignored.



- Connect: R/W Writing '1' will allow the device to connect its pull up
  - resistor. Writing '0' forces a disconnect. Reading returns the current connect status.
- ConnectChange: R Change of the connect status. Reading clears the
  - bit.
- Suspended: R/W This bit represents the current Suspend state. It is set to '1' when the device hasn't seen any activity on its upstream port for more than 3 ms. It is reset to '0' on any activity.

When the device is in suspend state (Suspended bit = '1') and the microcontroller writes a '0' into it, the device will generate a remote wake up. When the device is not suspended, writing a '0' has no effect. Writing a '1' into this register has never an effect.

- SuspendChange: R The Suspend Change bit is set to '1' when the Suspended bit toggles. The Suspend bit can toggle because:
  - The device goes in the suspended state
  - The device receives resume signalling on its upstream port

The Suspend Change bit is reset after the register has been read.

The Reset bit is set when the device receives a bus reset. It is cleared when read. On a bus reset the device will automatically go to the default state (unconfigured and responding to address 0).

• Reset:

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R

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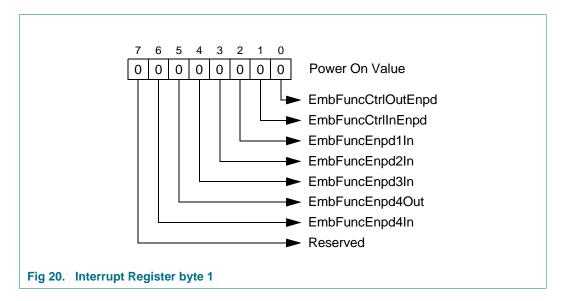
Get Current Frame Number: Command: F5h

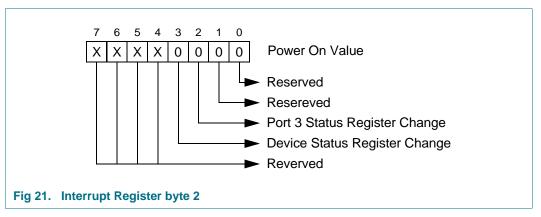
Data: Read 1 or 2 byte(s)

Data: Write 2 bytes

Get interrupt register: Command: F4h

Data: Read 2 bytes





This command indicates the origin of an interrupt. The endpoint interrupt bits (bits 0 to 9) are cleared by selecting the endpoint. The device status register change bit is reset by reading the device status change register. The Port Status Change Register change bit is reset by reading the port Status Change Register.

### **Set Mode:**

Command: F3h

Data: Write 2 bytes

0.0.1 Set Mode

Command: F3h

Data: Write 2 bytes

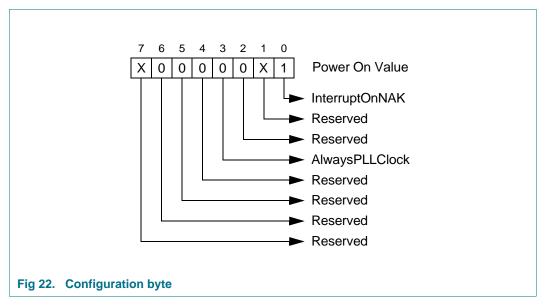
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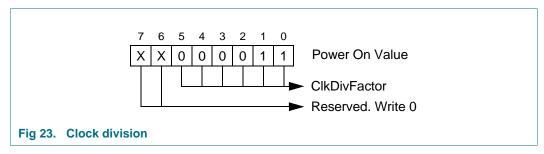
**PN5331B3HN** 

### **Configuration byte:**



- InterruptOnNAK:
- A '1' indicates that "NAKing" is reported and will generate interrupt. A '0' indicates that only successful transactions are reported.
- AlwaysPLLClock:
- indicates that the internal clocks and PLL are always running even during suspend state. A '0' indicates that the internal clock, crystal oscillator and PLL are stopped whenever not needed. To meet the strict suspend current requirement, this bit needs to be set to '0'.

#### **Clock division**



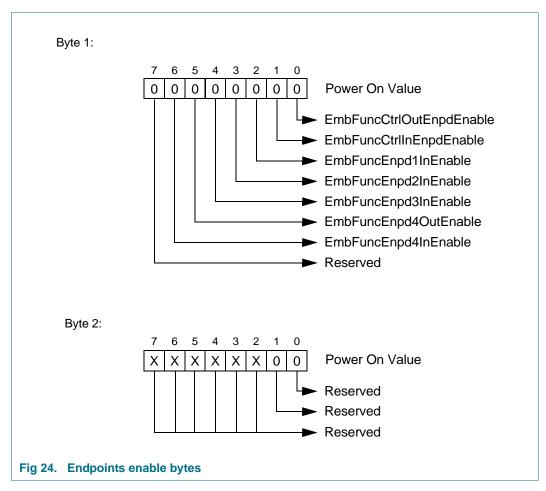
ClkDivFactor:

The value indicates clock division factor for CLOCKOUT. The output frequency is 48 MHz/(N+1) where N is the Clock Division Factor. The reset value is 3. This will produce the output frequency of 12 MHz which can then be programmed up (or down) by the user. This design ensures no glitching during frequency change. The programmed value will not be changed by a bus reset.

A '1'

Set Endpoints Enable: Command: D8h

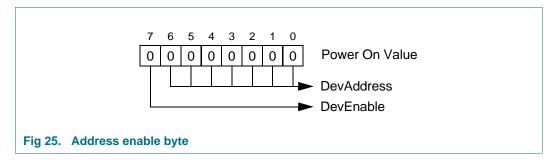
Data: Write 2 bytes



This command provides endpoint enable. The enable is defined on physical endpoint level meaning that for one endpoint the IN and OUT direction can be enabled separately.

Set Address/Enable: Command: D0h

Data: Write 1 byte



DevAddress: The value written becomes the address.

• DevEnable: A '1' enables this function

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Get Embedded Port Status: Command: E0h

Data: Read 1 byte

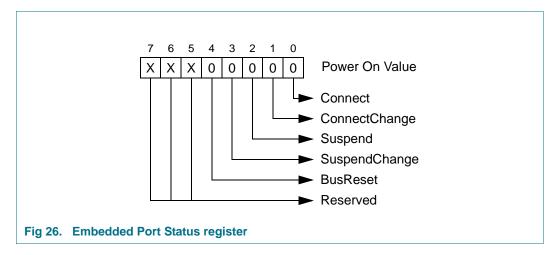
When SuspendChange and BusReset bit are set, the appropriate bit in the interrupt register is set and an interrupt is generated to the micro-controller.

The Get Embedded Port Status command returns the Embedded Port Status Register. The BusReset, SuspendChange and ConnectChange bit are reset by this command.

Set Embedded Port Status: Command: E0h

Data: Write 1 byte

The Set Embedded Port Status command changes the Embedded Port Status Register. The value of Read Only bits is ignored.



Connected:
 R/W Writing '1' will connect this embedded port. Writing

'0' will disconnect this embedded port.

ConnectChange: R Change of the connect status. Reading clears the

bit.

Suspend: R/W Embedded port is suspended. Writing '0' causes a

remote wake-up if this embedded port is suspended. Writing a '1' has no effect.

• SuspendChange: R Suspend state changed. Reading clears the bit

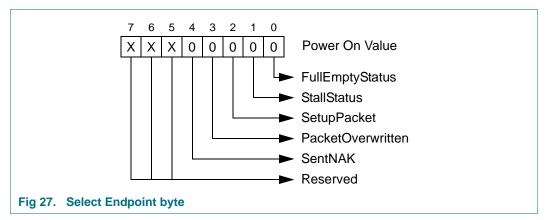
BusReset:REmbedded Port received a SetPortFeature(Reset) request. Reading clears the bit, puts the port into the enabled state and reports the end of the reset to the host.

### **Near Field Communication (NFC) controller**

Select Endpoint: Command: 00h - 09h

Data: Read 1 byte (Optional)

This command initializes an internal pointer to the start of the Selected buffer. Optionally, this command can be followed by a data read, which returns some additional info on the packet in the buffer.



- FullEmptyStatus:
- A "1" indicates the buffer of the selected endpoint is full, 
  "0" indicates an empty buffer. In case of an OUT endpoint, this bit is cleared by executing the Clear Buffer Command, if the buffer has not been over written.

In case of an IN endpoint, this bit is set by the Validate Buffer command.

- StallStatus: A "1" indicates the selected endpoint is in the stall state.
  - SetupPacket: A "1"

indicates the last received packet for the selected endpoint was a setup packet. The value of this bit is updated after each successfully received packet (i.e. an ACKED package on that particular physical endpoint).

It is cleared by doing a Select Endpoint/Clear Interrupt on this endpoint

 PacketOverwritten: '1': The previously received packet was over written by a setup packet.

The value of this bit is cleared by the 'Select Endpoint/Clear Interrupt' command.

This bit is set when a NAK is sent and the Interrupt On NAK feature is enabled. This bit is reset after the device has sent an ACK after an OUT packet or when the device has seen an ACK after sending an IN packet. This bit is only defined for the two physical control endpoints.

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#### Select Endpoint/Clear Interrupt: Command: 40h - 49h

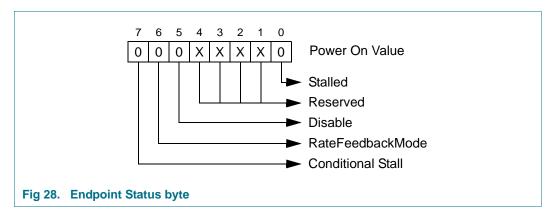
Data: Read 1 byte

Commands 40h to 49h are identical to their Select Endpoint equivalent, with the following differences:

- The command clears the associated interrupt
- The command clear the Setup and Overwritten bits in case of a control out endpoint
- The read of one byte is mandatory

Set Endpoint Status: Command: 40h - 49h

Data: Write 1 byte



Stalled: A "1" indicates the endpoint is stalled.

Disable: A "1" indicates the endpoint is disabled. After a bus-reset

each endpoint is enabled, i.e., this bit is set to '0'.

• RateFeedbackMode: '0': Interrupt endpoint in 'toggle mode'

'1': Interrupt endpoint in 'rate feedback mode'

• Conditional Stall: '1': Stall both endpoint zero endpoints, unless the

'Setup Packet' bit is set. It is only defined for control

**OUT** endpoints

Read Buffer: Command: F0h

Data: Read up to n+2 bytes

'n' is equal to the number of data bytes in the selected buffer.

This command is followed by a number of data reads, which return the contents of the selected endpoint data buffer. After each read, the internal buffer pointer is incremented by 1.

The buffer pointer is not reset to the beginning of the buffer by this command. This means that reading or writing a buffer can be interrupted by any other command (except for Select Endpoint).

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The data in the buffer are organized as follow:

Table 126. Read buffer bytes

Bit Position	7	6	5	4	3	2	1	0
Byte 0	0/1	0/1	-	-	-	-	-	-
Byte 1	- Number of Data bytes in buffer							
Byte 2		Data byte 0						
Byte n+1		Data Byte n -1						

Write Buffer: Command: F0h

Data: Write up to n+2 bytes

'n' is equal to the number of data bytes in the selected buffer.

This command is followed by a number of data writes, which load the data buffer of the selected endpoint. After each write, the internal pointer is incremented by 1. The buffer pointer is not reset to the beginning of the buffer by the Write Buffer command. This means that writing a buffer can be interrupted by any other command (except for Select Endpoint).

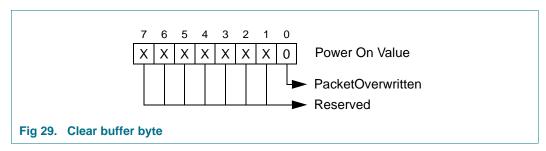
The data in the buffer are organized as follow:

Table 127. Write buffer bytes

Bit Position	7	6	5	4	3	2	1	0	
Byte 0	0/1	0/1	-	-	-	-	-	-	
Byte 1	- Number of Data bytes in buffer								
Byte 2				Data l	oyte 0				
Byte n+1		Data Byte n -1							

Clear Buffer: Command: F2h

Data: Read 1 byte (Optional)



When a packet is received completely, an internal endpoint buffer full flag is set. All subsequent packets will be refused by returning a NAK. When the microcontroller has read the data, it should free the buffer by the Clear Buffer command. When the buffer is cleared new packets will be accepted.

When bit '0' of the optional data byte is '1', the previously received packet was over written by a Setup Packet. A buffer cannot be cleared when its Packet Overwritten bit is set.

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# **Near Field Communication (NFC) controller**

Validate Buffer: Command: FAh

Data: None

When the microprocessor has written data into an IN buffer, it should set the buffer full flag by the Validate Buffer command.

This indicates that the data in the buffer is valid and can be sent to the host when the next IN token is received. A control IN buffer cannot be validated when the Packet Overwritten bit of its corresponding OUT buffer is set.

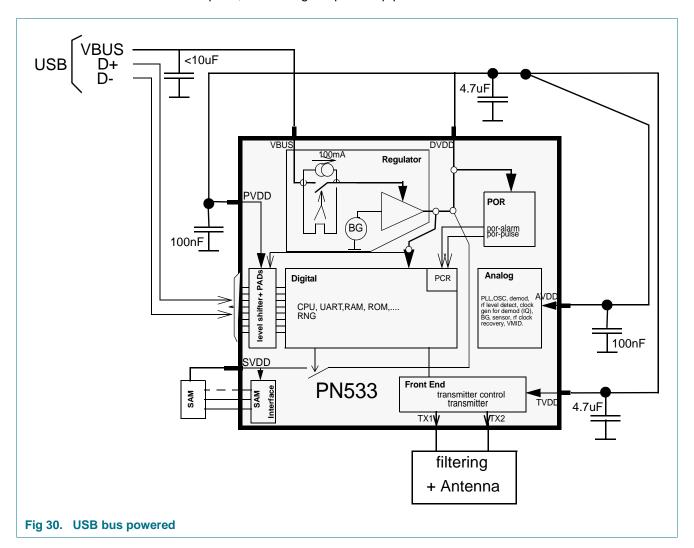
**PN5331B3HN** 

# 8.4 Power management

This chapter defines the power distribution scheme according to the different system configuration. The PN533 can be supplied by the USB connector on VBUS or directly on the VBUS, AVDD, DVDD, PVDD and TVDD. Regarding the system configuration (USB BUS powered, USB non bus powered or HOST powered), the power distribution shall be different.

### 8.4.1 USB bus powered

The power distribution is performed from the USB bus. The power delivered to the different peripherals is controlled by the PN533 chip. The <u>Figure 30 "USB bus powered"</u> depicts the system approach for the power distribution. When PN533 is supplied by the USB connector (USB powered) an internal regulator generates the supply voltage for all the parts, and during the power up phase the inrush current is limited to 100mA.



**PN5331B3HN** 

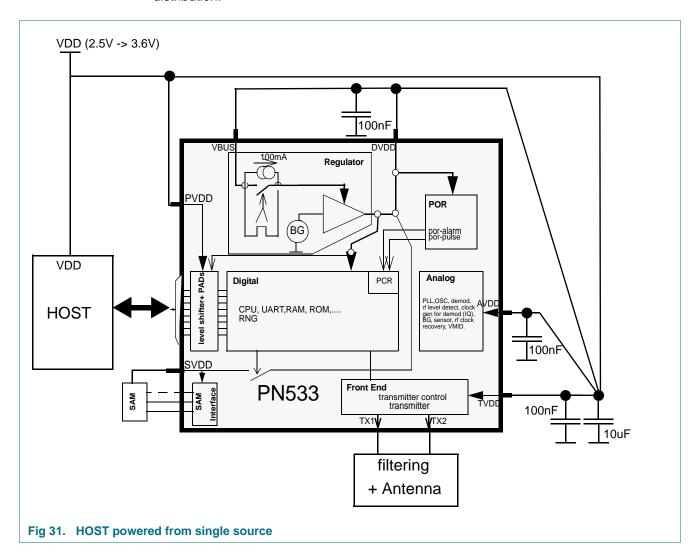
### **Near Field Communication (NFC) controller**

### 8.4.2 USB non bus powered

In this mode, the power distribution can be either HOST powered single source or HOST powered double source as described in <u>Section 8.4.3 "HOST powered (single source)"</u> and in <u>Section 8.4.4 "HOST powered (double source)"</u>. The D+ and D- IOs are referred to PVDD. The application shall take care of eventual voltage compatibility and compliancy to standards.

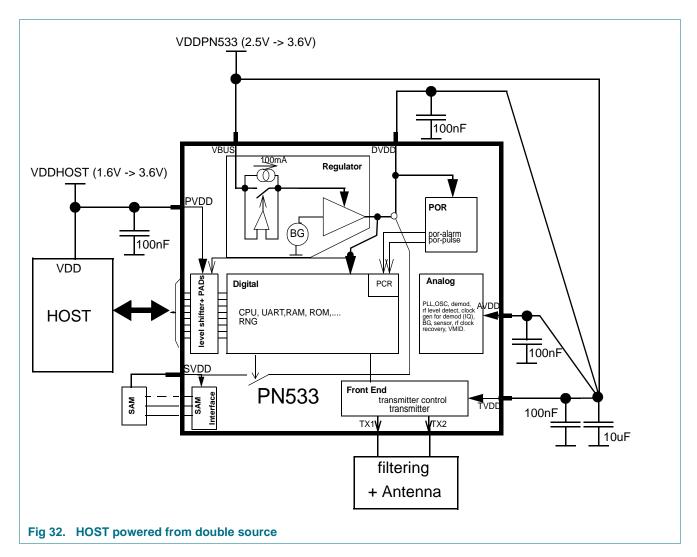
### 8.4.3 HOST powered (single source)

In that case, the power distribution is performed from a single power supply source. The <u>Figure 31 "HOST powered from single source"</u> depicts the system approach for the power distribution.



# 8.4.4 HOST powered (double source)

The power distribution is performed from the second source from the Host. The <u>Figure 32</u> "<u>HOST powered from double source</u>" depicts the system approach for the power distribution.



### 8.4.5 Low power modes

There are 2 different low power modes.

- "Hard power-down" controlled by the pin RSTPD\_N. In that case, the PN533 enters
  into the reset state and the maximum consumption depends on the connection of
  PN533 to the USB bus or not.
- "Soft power-down" controlled by a register. In that case, the PN533 enters into the idle state and the maximum consumption depends if PN533 is "USB powered" or not and if the RF detector is active or not. In that mode the PN533 can be waken up on external events.

Table 128. Current consumption in power down

Low power mode	Power supply source	Maximum current consumption
Suspend	Powered from USB	200 $\mu\text{A}$ (without resistors on D+ / D-)
Suspend with RF detector active	Powered from USB	250 $\mu\text{A}$ (without resistors on D+ / D-)
hard power-down	Not powered from USB	10 μΑ
soft power-down	Not powered from USB	25 μΑ
soft power-down with RF detector active	Not powered from USB	30 μΑ

#### 8.4.6 Power-on reset

The Power On Reset (POR) module generates the reset signals for the different parts of PN533.

The Power On Reset module is used to control the power up, power down and reset phase of PN533.

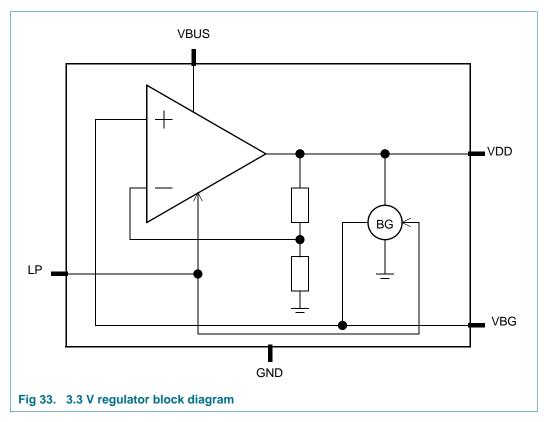
As soon as, VDD reaches Vth+Vhys, the system startup phase starts under control of the PCR.

When the RSTPD\_N is asserted, all internal current source are cut and PN533 enters reset phase.

When the power supply voltage drops below Vth, the IC goes into reset.

# 8.4.7 Regulator - short description

The regulator is used to reduce the VBUS voltage to the typical operating voltage of PN533.



The 3.3V regulator is a linear regulator with resistive feed-back. The regulator uses the Band-gap for reference voltage

#### 8.4.8 Main switch

In USB bus powered configuration, the PN533 is plugged on a USB connector. The main switch limits the inrush current to 100 mA max during the powerup. The inrush current limitation can be disabled through the bit curlimoff (<a href="Table 129">Table 129 "Control\_switch\_rng register">Touring register</a> (address 6106h) bit allocation").

#### 8.4.9 SVDD switch

The SVDD switch is used to control power to the secure IC. The switch is controlled by register Control\_switch\_rng (address 6106h). The switch is enabled with bit sic\_switch\_en. When disabled, the SVDD pin is tied to ground. A current limiter is incorporated into the switch. Current consumption exceeding 40 mA triggers the limiter and the status bit sic\_switch\_overload is set.

Register Control\_switch\_rng also controls the random generator within the Contactless Interface Unit (CIU) and the regulator current limitation.

Table 129. Control\_switch\_rng register (address 6106h) bit allocation

			•	•	•			
Bit	7	6	5	4	3	2	1	0
Symbol	-	hide_svdd _sig	sic_switch _overload	sic_switch _en	curlimoff	cpu_need _ rng	random_ dataready	-
Reset	0	1	0	0	0	0	0	1
Access	R	R/W	R	R/W	R/W	R/W	R/W	R

Table 130. Description of Control\_switch\_rng bits

Bit	Symbol	Description
7	-	Reserved.
6	hide_svdd_sig	Configures internal state of input signals SIGIN and P34 when idle. This bit can be used to avoid spikes on SIGIN and P34 when the SVDD switch is enabled or disabled.
		When set to logic 0, internal state of SIGIN and P34 are driven by pads SIGIN and P34 respectively.
		When set to logic 1, internal state of SIGIN is set to logic level 0 and internal state of P34 is set to logic 1.
5	sic_switch_overload	Indicates state of SVDD switch current limiter.
		When set to logic 0, indicates that current consumption through SVDD switch does not exceed limit (40 mA).
		When set to logic 1, the SVDD switch current limiter is activated.
4	sic_switch_en	Enables or disables power to SVDD switch.
		When set to logic 0, SVDD switch is disabled and SVDD output is tied to the ground.
		When set to logic 1, the SVDD switch is enabled and the SVDD output delivers power to secure IC and internal pads (SIGIN, SIGOUT and P34).
3	curlimoff	Configure the regulator to deliver more current than 100 mA.
		When set to 0, the 100 mA current limitations is activated.
		When set to 1, the 100 mA current limitations is desactivated.
2	cpu_need_rng	Forces random number generator into running mode.
		When set to logic 0, random number generator is under control of Contactless Interface Unit.
		When set to logic 1, random number generator is forced to run.
1	random_dataready	Indicates availability of random number.
		When set to logic 1, a new random number is available.
		Automatically set to logic 0 when register data_rng (address 6105h) is read.
0	-	Reserved.

### 8.5 Power clock and reset controller

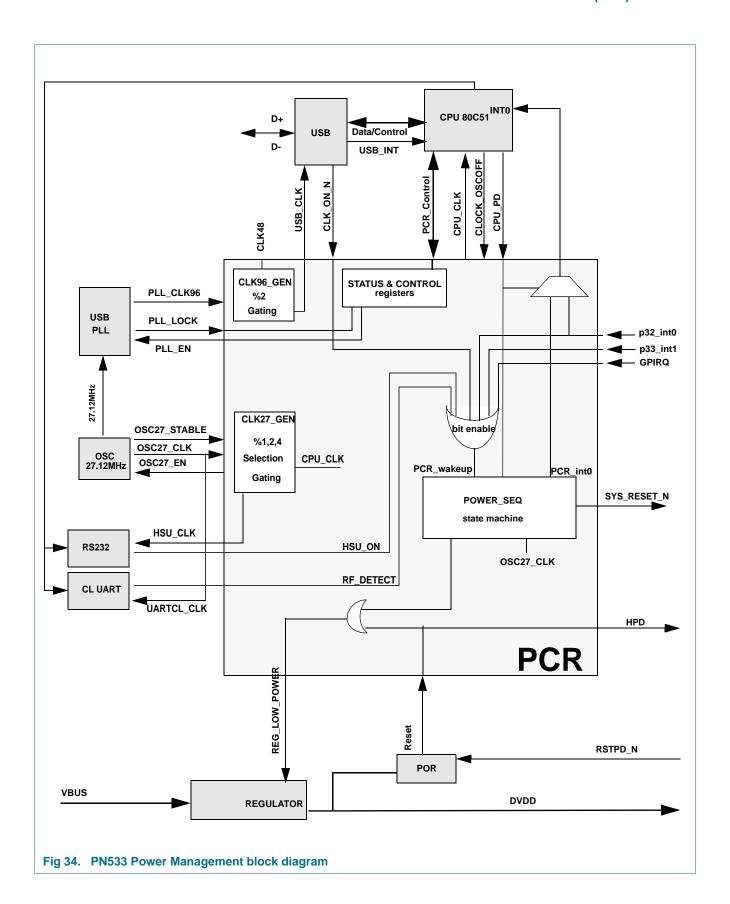
The PCR controller is the controller for the clock generation, power management and reset architecture for the PN533.

# 8.5.1 PCR in the system

This block diagram <u>Figure 34 "PN533 Power Management block diagram"</u> shows the relationship between the PCR, other embedded blocks and external signals.

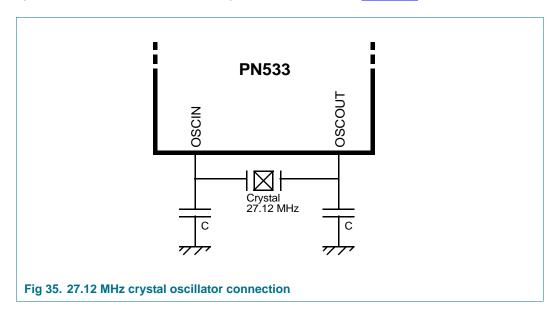
Table 131. PN533 clock source characteristics

Clock name	Frequency MHz	Tolerance	Clock source	Comments
OSC_CLK27	27.12	$\pm$ 14 kHz	OSC 27.12	Output of OSC 27
PLL_CLK96	96	± 0.25%	USB PLL	Output of USB PLL
USB_CLK	48	± 0.25%	USB PLL	
CPU_CLK	27.12/13.56/6.78		OSC 27	Default is 6.78 MHz
HSU_CLK	27.12		OSC 27	



### 8.5.2 27.12 MHz crystal oscillator

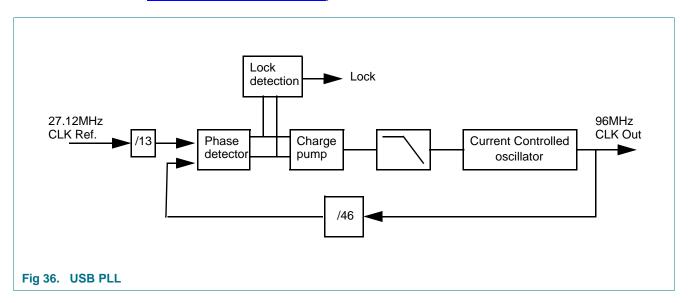
The 27.12 MHz clock applied to the PN533 is the time reference for the embedded microcontroller. Therefore stability of the clock frequency is an important factor for reliable operation. It is recommended to adopt the circuit shown in Figure 35.



### 8.5.3 PLL for USB clock generation

The 96MHz used by the USB peripheral is derived from the main 27.12MHz by mean of a semi-fractional PLL. This PLL consists of a ring oscillator running at 96MHZ, an on-chip 70KHz loop filter a divide-by-13 reference divider and a divide-by-46 feedback divider.

The PLL is controlled through several registers (see <u>Table 140 "PCR Control register</u> (address 6203h) bit allocation").



### 8.5.4 Power-up sequence

In a USB application, the embedded firmware shall start the USB PLL. The embedded firmware shall follow below sequence:

- Configure MATX in USB mode.
- Enable 96MHz PLL by setting PLL\_en bit of PCR control register (6203h) to logic level 1.
- Wait for the lock of the PLL by polling PLL\_lock bit in PCR CER register (6201h).
- When PLL is locked, enable 96 MHz clock by setting CLK\_96\_en bit of the PCR control register at logic level one.
- Enable 48 MHz USB clock that is derived from 96MHz clock by setting USB\_enable bit of PCR CER register (6201h) at logic level one.
- Set reset\_USB\_n bit of PCR control register (6203h) at logic level one.

### 8.5.5 Low power modes

There are 3 different low power modes.

- Hard-Power-down mode (HPD): controlled by the pin RSTPD\_N. The PN533 goes into reset and power consumption is at a minimum, see Section 8.5.6 "Reset modes".
- In HSU application, Soft-Power-down mode (SPD): controlled by firmware. See
   <u>Section 8.5.7 "Soft-Power-down mode (SPD)"</u> to optimize the power consumption in
  this mode.
- In USB application, Suspend mode: controlled by firmware. xxxx

Table 132. Current consumption in low power modes

Mode	Conditions	Maximum current consumption
Hard-Power-down	RSTPD_N is set to logic level 0	10 μΑ
Soft-Power-down with no RF detector	Sequence of <u>Section 8.5.7</u> is applied	<tbd></tbd>
Soft-Power-down with RF detector active	Sequence of <u>Section 8.5.7</u> is applied	30 μΑ
Suspend mode remote wake up disabled	<tbd></tbd>	<tbd></tbd>
Suspend mode with remote wake up enabled	<tbd></tbd>	<tbd></tbd>

#### **Near Field Communication (NFC) controller**

#### 8.5.6 Reset modes

The possible reset mechanisms are listed below:

Supply rail variation

When DVDD falls below Vth, the POR (Power-On-Reset) asserts an internal reset signal. The Power Sequencer disables all clocks. When DVDD rises above Vth+Vhys, the POR deasserts the internal reset signal and the Power Sequencer starts the power-up sequence. Once the PN533 is out of reset, the RSTOUT\_N pin is driven high.

Glitch on DVDD

When DVDD falls below Vth for more than 1  $\mu$ s, the POR asserts an internal reset signal. The power sequencer starts the Power-down sequence. The PN533 goes into reset and the RSTOUT\_N signal is driven low.

Hard Power-down mode (HPD)

When RSTPD\_N is set to logic level 0, the PN533 goes into Hard Power Down (HDP) mode. The PN533 goes into reset and the RSTOUT\_N signal is driven low. The power consumption is at the minimum. DVDD is tied to ground and ports are disconnected from their supply rails. The PN533 goes into reset.

# 8.5.7 Soft-Power-down mode (SPD)

In order to initiate the Soft-Power-down mode with minimal power consumption, the firmware should:

- Configure I/Os to minimize power consumption
- Shut down unused functions
  - Contactless Interface Unit with bit Power-down of SFR register D1h, see <u>Table 180</u> on page 156.
  - Disable the SVDD switch, see <u>Table 130 on page 94</u>
  - Power down the RF level detector if RF wake up is not enabled, see <u>Table 288 on</u> page 198.
- Enable relevant wake-up sources
- Disable unwanted interrupts
- Set to logic level one the CPU\_PD bit in PCON register, see Table 7 on page 15

#### **Near Field Communication (NFC) controller**

### 8.5.8 Suspend mode

After 3 ms of no USB activity on the bus, the device goes in suspend. About 2 ms later the device will indicate that it no longer needs the clock by setting to logic level one suspended bit (refer to bit "suspended" of "get device status" USB register at FEh address).

In order to initiate the Suspend mode with minimal power consumption, the firmware should:

- Poll the clock\_on bit in the CER register at 6201h address
- · Configure I/Os to minimize power consumption
- Shut down unused functions
  - Contactless Interface Unit with bit Power-down of SFR register D1h, see <u>Table 180</u> on page 156.
  - Disable the SVDD switch, see Table 130 on page 94
  - Power down the RF level detector if RF wake up is not enabled, see <u>Table 288 on</u> page 198.
- Enable relevant wake-up sources; USB wake up source is mandatory
- Disable unwanted interrupts
- Disable CLK\_96 MHz Clock, by setting to logic level zero CLK\_96\_en bit of the PCR control register at 6203h address
- Disable USB PLL by setting to logic level zero PLL\_en bit in PCR control register
- Set logic level one the CPU\_PD bit in PCON register, see <u>Table 7 on page 15</u>

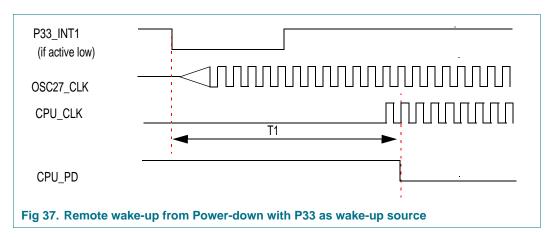
#### 8.5.9 Remote wake-up

The PN533 can be woken up from a Soft-Power-down mode or Suspend mode when an event occurs on one of the wake up sources, which has been enabled. There are seven wake-up sources:

- P32 INT0
- P33 INT1
- RF field detected
- HSU wake-up (valid when in Soft-Power-down mode)
- USB wake-up (valid when in Suspend mode)
- NFC\_WI counters
- GPIRQ: P34, P35, DP

When one of these signals is asserted, if its corresponding enable bit is set (see <u>Table 145 on page 107</u>), the Power Sequencer starts the wake-up sequence. The wake up event can only be serviced if the Power Sequencer is in the Stopped state, which means the PN533 is fully entered in Soft-Power-down mode.

<u>Figure 37</u> illustrates the wake-up mechanism, using an event on P33\_INT1 as an example. CPU\_CLK is active T1 after the falling edge of P33\_INT1 and the PN533 is ready. T1 depends on the choice of crystal oscillator and its layout. For devices such as TAS-3225A, TAS-7 or KSS2F, T1 is a maximum of 2 ms. Exit from the Power-down mode is signaled by CPU\_PD going low one clock cycle later.



When woken up, two cases are to be considered: non USB application and USB application

- In a non USB application, system is ready to operate
- In a USB application, the embedded firmware shall start the USB PLL. It shall then
  poll the PLL\_lock signal to enable the 96 MHz clocks. When the 96 MHz clock is
  enabled, the 48 MHz clock is available.

PN5331B3HN

# 8.5.10 PCR extension registers

The PCR is controlled via several registers given in Table 133:

Table 133. PCR registers

Name	Size [bytes]	Address offset	Description	Reset	R/W					
CFR	1	6200h	Clock Frequency Register	02	R/W					
CER	1	6201h	Clock Enable Register	0E	R/W					
ILR	1	6202h	Interrupt Level Register	40	R/W					
Control	1	6203h	Control	C0	R/W					
Status	1	6204h	Status	00	R					
Wakeupen	1	6205h	Wake-up Enable	00	R/W					

# 8.5.11 PCR register description

### 8.5.11.1 CFR register

The Clock Frequency Register is used to select the frequency of the CPU and its associated peripherals. The clock frequency can be changed dynamically by writing to this register at any time.

Table 134. PCR CFR register- (address 6200h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	-	-	cpu_freq[1:0]	
Reset	0	0	0	0	0	0	1	0
Access	R	R	R	R	R	R	R/W	R/W

Table 135. Description of PCR CFR bits

Bit	Symbol	Description	Description					
7 to 2	-	Reserved	Reserved					
1 to 0	cpu_frq[1:0]	Select CPU clock frequency.						
		cpu_frq[1:0]	CPU clock frequency					
		00	27.12 MHz					
		01	13.56 MHz					
		10	6.78 MHz					
		11	27.12 MHz					

# 8.5.11.2 **CER register**

The Clock Enable Register is used to enable or disable the clock of the USB and HSU interfaces (frequency is fixed at 27.12 MHz). The clock can be switched on or off at any time.

This register also contains the PLL\_lock signal that the embedded firmware should poll.

Table 136. PCR CER register (address 6201h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	clock_on	PLL_lock	hsu_enable	-	-	usb_enable
Reset	0	0	1	0	1	1	1	0
Access	R	R	R	R	R/W	R	R	R/W

Table 137. Description of PCR CER bits

	•	
Bit	Symbol	Description
7 to 6	-	Reserved.
5	clock_on	<b>USB clock_on signal</b> to poll before entering PN533 into power down.
4	PLL_lock	PLL_lock signal.
3	hsu_enable	<b>Enable HSU clock.</b> When 1, HSU is enabled. When 0, HSU is disabled.
2 to 1	-	Reserved.
0	usb_enable	Enable USB clock.

# 8.5.11.3 ILR register

The Interrupt Level Register is used to program the level of the external interrupts. Firmware can write to this register at any time.

Table 138. PCR ILR register (address 6202h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	porpulse_ latched	-	enable_pdselif	-	gpirq_level	int1_level	int0_level
Reset	0	1	0	0	0	0	0	0
Access	R	R/W	R	R/W	R	R/W	R/W	R/W

#### Table 139. Description of PCR ILR bits

Bit	Symbol	Description								
7	-	Reserved								
6	porpulse_latched	Indicates that a re	Indicates that a reset has been generated.							
		•		es that the system has been reset. The ne firmware reset sequence.						
5	-	Reserved								
4	enable_pdselif	Indicates that a re	eset has been	generated.						
		When set to logic I pins:	evel 1, P33_IN	T1 directly controls state of host interface						
			s set to logic lev selected interfa	vel 1, host interface output pins are driven ce protocol						
		<ul> <li>If P33_INT1 is into high-impe</li> </ul>	_	vel 0, host interface output pins are set						
				IT1 does not control host interface pins. cted interface protocol.						
		enable_pdselif	P33_INT1	State of host interface pins						
		0	X	Active						
		1	0	High Impedance						
		1	1	Active						
3	-	Reserved.								
2	gpirq_level	Selects gpirq inte	errupt level.							
		When set to logic	level 1, wake-ι	ip condition is true when gpirq is high.						
		When set to logic	level 0, wake-u	up condition is true when gpirq is low.						
1	int1_level	Selects P33_INT1	interrupt leve	el.						
		When set to logic I	evel 1, wake-u	p condition is true when P33_INT1 is low.						
		When set to logic high.	level 0, wake-ι	p condition is true when P33_INT1 is						
0	int0_level	Selects P32_INT0	) interrupt leve	el.						
		When set to logic high.	level 1, wake-ι	p condition is true when P32_INT0 is						

# 8.5.11.4 PCR Control register

The Control register is used to perform a firmware reset and clear wake-up conditions in the Status register.

Table 140. PCR Control register (address 6203h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	reset_USB _n	clock96 _on	PLL_en	-	clear_wakeup_ cond	soft_reset
Reset	1	1	0	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R	R/W	R/W

### Table 141. Description of PCR Control bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5	reset_USB_n	Enables a USB reset.
		When set to logic level 1, the reset for the USB block is inactive.
		When set to logic level 0, reset for the USB block is active
4	CLK96_on	Enables 96 MHz clock generation.
		When set to logic level 1, 96 MHz clocks are enabled.
		When set to logic level 0, 96 MHz clocks are disabled.
3	PLL_en	Enables the PLL.
		When set to logic level 1, PLL is enabled.
		When set to logic level 0, PLL is disabled.
1	clear_wakeup_cond	Clears value of wakeupcond in Status register.
		When set to logic level 1, wake-up conditions stored in PCR Status register are set to logic level 0. Bit is set to logic 0 automatically by hardware.
0	soft_reset	Initiates a firmware reset.
		When set to logic level 1, system goes into firmware reset mode. Bit is set to logic level 0 automatically by hardware after performing firmware reset sequence.

### 8.5.11.5 PCR Status register

The PCR Status register stores the state of the 7 wake-up events, reported within 6 flags.

Table 142. PCR Status register (address 6204h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	gpirq_wu	-	HSU_wu	CIU_wu	USB_wu	int1_wu	int0_wu
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R

An event on a given wake-up condition is flagged by a logic level 1 in the associate bit field.

Table 143. Description of PCR Status bits

when enabled
PIRQ at logic
dges on
ent).
<u>[1]</u>
pt [ <u>1]</u> .
pt.[ <u>1]</u> .
er [1]

<sup>[1]</sup> The firmware must set to logic level 0 this bit after reading it (by writing a logic 1 to bit clear\_wakeup\_cond in register PCR Control).

# 8.5.11.6 Wakeupen register

Register Wakeupen allows the selection of different wake-up events.

Table 144. PCR Wakeupen register (address 6205h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	GPIRQ_ wu_en	-	HSU_on_ en	CIU_wu_ en	clock_on_ en	int1_en	int0_en
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R	R/W	R/W

### Table 145. Description of PCR Wakeupen bits

Rit	Symbol	Description
	Oyillboi	•
7	-	Reserved.
6	GPIRQ_wu_en	General Purpose IRQ wake-up source enable.
		When set to logic 1, a GPIRQ event can wake up PN533.
5	-	Reserved.
4	HSU_on_en	HSU wake-up source enable.
		When set to logic 1, an HSU event can wake up PN533.
3	CIU_wu_en	Contactless Interface Unit wake-up source enable.
		When set to logic 1, a CIU event (RF detected or NFC-WI event) can wake up PN533.
2	clock_on_en	USB wake up source enable.
1	int1_en	P33_INT1 wake-up source enable.
		When set to logic 1, a P33_INT1 event can wake up PN533.
0	int0_en	P32_INT0 wake-up source enable.
		When set to logic 1, a P32_INT0 event can wake up PN533.
0	int0_en	-

### **Near Field Communication (NFC) controller**

# 8.6 Contactless Interface Unit (CIU)

The PN533 CIU is a modem for contactless communication at 13.56 MHz. It supports 6 different operating modes

- ISO/IEC 14443A/MIFARE Reader/Writer
- FeliCa Reader/Writer
- ISO/IEC 14443B Reader/Writer
- ISO/IEC 14443A/MIFARE KB or MIFARE 4 KB Card emulation
- FeliCa Card emulation
- ISO/IEC 18092, ECMA 340 NFCIP-1 Peer-to-Peer

The CIU implements a demodulator and decoder for signals from ISO/IEC 14443A/MIFARE compatible cards and transponders. The CIU handles the complete ISO/IEC 14443A framing and error detection (Parity & CRC).

The CIU supports MIFARE 1 KB or MIFARE 4 KB emulation products. The CIU supports contactless communication using MIFARE Higher transfer speeds up to 424 kbit/s in both directions.

The CIU can demodulate and decode FeliCa coded signals. The CIU digital part handles the FeliCa framing and error detection. The CIU supports contactless communication using FeliCa Higher transfer speeds up to 424 kbit/s in both directions.

The CIU supports layers 2 and 3 of the ISO/IEC 1444 B Reader/Writer communication scheme, except anticollision which must be implemented in firmware as well as upper layers.

In card emulation mode, the CIU is able to answer to a Reader/Writer command either according to the FeliCa or ISO/IEC 14443A/MIFARE card interface scheme. The CIU generates the load modulation signals, either from its transmitter or from the LOADMOD pin driving an external active circuit. A complete secure card functionality is only possible in combination with a secure IC using the NFC-WI/S<sup>2</sup>C interface.

Compliant to ECMA 340 and ISO/IEC 18092 NFCIP-1 Passive and Active communication modes, the CIU offers the possibility to communicate to another NFCIP-1 compliant device, at transfer speeds up to 424 kbit/s.The CIU handles the complete NFCIP-1 framing and error detection.

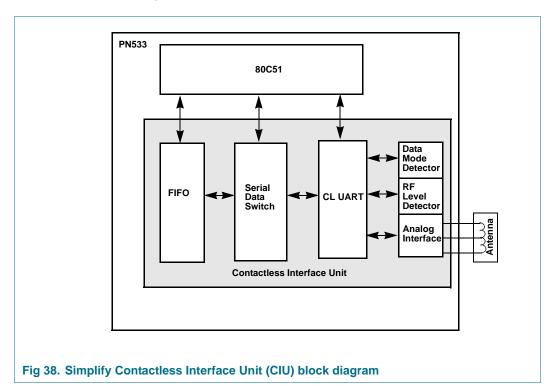
The CIU transceiver can be connected to an external antenna for Reader/Writer or Card/PICC modes, without any additional active component.

## **Near Field Communication (NFC) controller**

#### 8.6.1 Feature list

- Frequently accessed registers placed in SFR space
- Highly integrated analog circuitry to demodulate and decode received data
- Buffered transmitter drivers to minimize external components to connect an antenna.
- Integrated RF level detector
- Integrated data mode detector
- Typical operating distance of 50 mm in ISO/IEC 14443A/MIFARE or FeliCa in Reader/Writer mode depending on the antenna size, tuning and power supply
- Typical operating distance of 50 mm in NFCIP-1 mode depending on the antenna size, tuning and power supply
- Typical operating distance in ISO/IEC 14443A/MIFARE card or FeliCa card operation mode of about 100 mm depending on the antenna size, tuning and the external field strength
- Supports MIFARE 1 KB or MIFARE 4 KB emulation encryption in Reader/Writer mode
- Supports MIFARE higher data rate at 212 kbit/s and 424 kbit/s
- Supports contactless communication according to the FeliCa scheme at 212 kbit/s and 424 kbit/s
- Support of the NFC-WI/S<sup>2</sup>C interface
- 64 bytes send and receive FIFO-buffer
- Programmable timer
- CRC Co-processor
- Internal self test
- 2 interrupt sources
- Adjustable parameters to optimize the transceiver performance according to the antenna characteristics

# 8.6.2 Simplified block diagram



The Analog Interface handles the modulation and demodulation of the analog signals according to the Card emulation mode, Reader/Writer mode and NFCIP-1 mode communication scheme.

The RF level detector detects the presence of an external RF-field delivered by the antenna to the RX pin.

The data mode detector detects a ISO/IEC 14443-A MIFARE, FeliCa or NFCIP-1 mode in order to prepare the internal receiver to demodulate signals, which are sent to the PN533.

The NFC-WI/S<sup>2</sup>C interface supports communication to secure IC. It also supports digital signals for transfer speeds above 424 kbit/s.

The CL UART handles the protocol requirements for the communication schemes in co-operation with the appropriate firmware. The FIFO buffer allows a convenient data transfer from the 80C51 to the CIU and vice versa.

## 8.6.3 Reader/Writer modes

All indicated modulation indices and modes in this chapter are system parameters. This means that beside the IC settings a suitable antenna tuning is required to achieve the optimal performance.

#### 8.6.3.1 ISO/IEC 14443A Reader/Writer

The following diagram describes the communication on a physical level, the communication overview in the Table 146 describes the physical parameters.

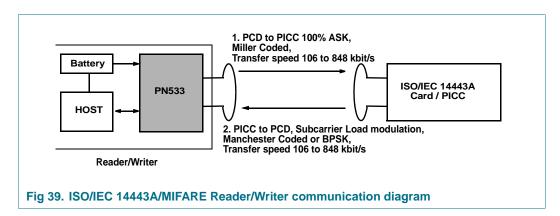
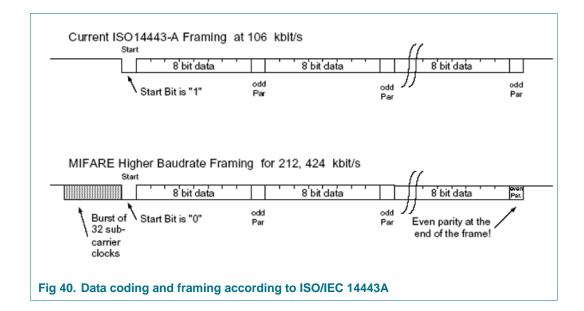


Table 146. Communication overview for ISO/IEC 14443A/MIFARE Reader/Writer

Communication scheme		ISO/IEC 14443A MIFARE	MIFARE higher baud rate				
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s	848 kbit/s		
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$	$\frac{16}{13,56MHz} \approx 1, 18\mu s$		
PN533 to	Modulation	100% ASK	>25% ASK	>25% ASK	>25% ASK		
PICC/Card	Bit coding	Modified Miller coding	Modified Miller coding	Modified Miller coding	Modified Miller coding		
PICC/Card to PN533	Modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation		
	Subcarrier frequency	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16		
	Bit coding	Manchester coding	BPSK	BPSK	BPSK		

The internal CRC co-processor calculates the CRC value according the data coding and framing defined in the ISO/IEC 14443A part 3, and handles parity generation internally according to the transfer speed.

With appropriate firmware, the PN533 can handle the complete ISO/IEC 14443A/MIFARE protocol.



#### 8.6.3.2 FeliCa Reader/Writer

The following diagram describes the communication at the physical level. <u>Table 147</u> describes the physical parameters.

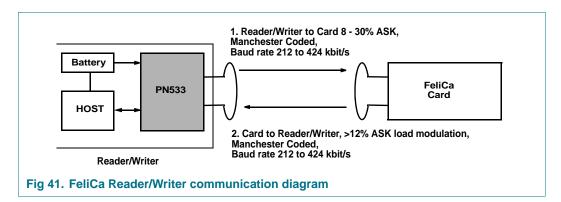


Table 147. Communication overview for FeliCa Reader/Writer

Communication	scheme	FeliCa	FeliCa higher baud rate
Baud rate		212 kbit/s	424 kbit/s
Bit length		$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$
PN533 to	Modulation	8 - 30% ASK	8 - 30% ASK
PICC/Card	Bit coding	Manchester coding	Manchester coding
PICC/Card to PN533	Modulation	>12% ASK	>12% ASK
	Bit coding	Manchester coding	Manchester coding

With appropriate firmware, the PN533 can handle the FeliCa protocol.

The FeliCa Framing and coding must comply with the following table:

Table 148. FeliCa Framing and Coding

	Preamble				SY	NC	LEN	n-D	ata	CF		
00h	00h	00h	00h	00h	00h	B2h	4Dh					

To enable the FeliCa communication a 6-byte preamble (00h, 00h, 00h, 00h, 00h, 00h) and 2-byte SYNC bytes (B2h, 4Dh) are sent to synchronize the receiver.

The following LEN byte indicates the length of the sent data bytes plus the LEN byte itself. The CRC calculation is done according to the FeliCa definitions with the MSB first.

To transmit data on the RF interface, the 80C51 has to send the LEN and data bytes to the CIU. The Preamble and SYNC bytes are generated by the CIU automatically and must not be written to the FIFO. The CIU performs internally the CRC calculation and adds the result to the frame.

The starting value for the CRC Polynomial is 2 null bytes: (00h), (00h)

Example of frame:

Table 149. FeliCa framing and coding

	Preamble						NC	LEN	2 Data	Bytes	CF	C
00	00	00	00	00	00	B2	4D	03	AB	CD	90	35

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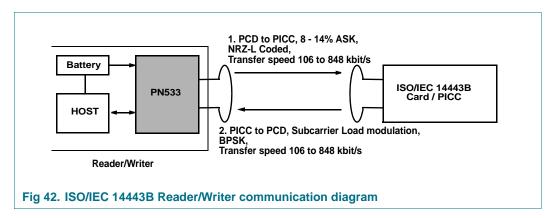
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#### 8.6.3.3 ISO/IEC 14443B Reader/Writer

The CIU supports layers 2 and 3 of the ISO/IEC 14443B Reader/Writer communication scheme, except anticollision which must be implemented in firmware as well as upper layers.

The following diagram describes the communication at the physical level. <u>Table 150</u> describes the physical parameters.



With appropriate firmware, the PN533 can handle the ISO/IEC 14443B protocol.

Table 150. Communication overview for ISO/IEC 14443B Reader/Writer

Communication	n scheme	ISO/IEC 14443B	Ту	pe B higher baud rat	е	
Baud rate Bit length		106 kbit/s	212 kbit/s	424 kbit/s	848 kbit/s	
		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,78 \mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$	$\frac{16}{13,56MHz} \approx 1,18\mu s$	
PN533 to	Modulation	8 -14% ASK	8 -14% ASK	8 -14% ASK	8 -14% ASK	
PICC/Card	Bit coding	NRZ-L	NRZ-L	NRZ-L	NRZ-L	
PICC/Card to PN533	Modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation	
	Subcarrier frequency	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16	
	Bit coding	BPSK	BPSK	BPSK	BPSK	

# 8.6.4 ISO/IEC 18092, ECMA 340 NFCIP-1 operating mode

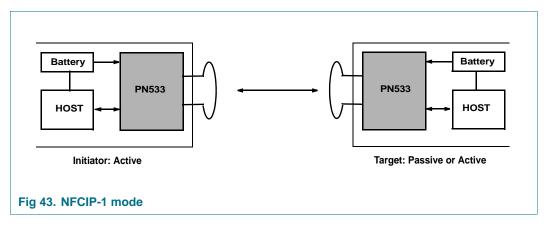
A NFCIP-1 communication takes place between 2 devices:

- Initiator: generates RF field at 13.56 MHz and starts the NFCIP-1 communication.
- Target: responds to initiator command either in a load modulation scheme in Passive Communication mode or using a self generated and self modulated RF field for Active Communication mode.

The NFCIP-1 communication differentiates between Active and Passive communication modes.

- Active Communication mode means both the initiator and the target are using their own RF field to transmit data
- Passive Communication mode means that the Target answers to an Initiator command in a load modulation scheme. The Initiator is active in terms of generating the RF field.

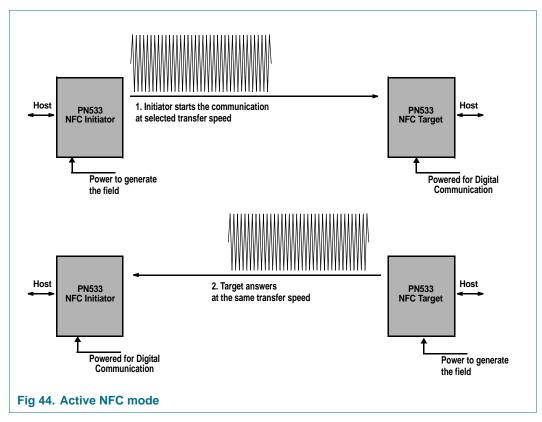
In order to fully support the NFCIP-1 standard the PN533 supports the Active and Passive Communications mode at the transfer speeds 106 kbit/s, 212 kbit/s and 424 kbit/s as defined in the NFCIP-1 standard.



With appropriate firmware, the PN533 can handle the NFCIP-1 protocol, for all communication modes and data rates, for both Initiator and Target.

#### 8.6.4.1 ACTIVE Communication mode

Active Communication Mode means both the Initiator and the Target are using their own RF field to transmit data.



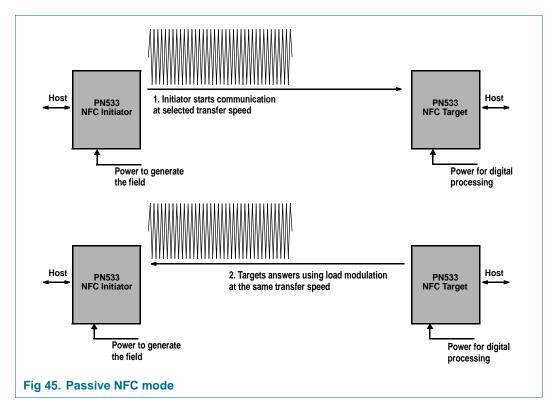
The following table gives an overview of the active communication modes:

Table 151. Communication overview for NFC Active Communication mode

Communication	scheme	ISO/IEC 18092, ECMA 340, NFCIP-1				
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s		
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$		
Initiator to Target	Modulation	100% ASK	8-30%ASK	8-30%ASK		
	Bit coding	Miller Coded	Manchester Coded	Manchester Coded		
Target to Initiator	Modulation	100% ASK	8-30%ASK	8-30%ASK		
	Bit coding	Miller Coded	Manchester Coded	Manchester Coded		

#### 8.6.4.2 PASSIVE Communication mode

Passive Communication Mode means that the target answers to an Initiator command in a load modulation scheme.



The following table gives an overview of the active communication modes:

Table 152. Communication overview for NFC Passive Communication mode

Communicatio	n scheme	ISO/IEO	C 18092, ECMA 340, N	IFCIP-1	
Baud rate		106 kbit/s 212 kbit/s 424 kbit/s			
Bit length		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	64 ≈ 4,76μs	$\frac{32}{13,56MHz} \approx 2,36\mu s$	
PN533 to	Modulation	100% ASK	100% ASK	100% ASK	
PICC/Card	Bit coding	Modified Miller coding	Modified Miller coding	Modified Miller coding	
PICC/Card to PN533	Modulation	Subcarrier load modulation	>12% ASK	>12% ASK	
	Subcarrier frequency	13.56 MHz/16	No subcarrier	No subcarrier	
	Bit coding	Manchester coding	Manchester coding	Manchester coding	

## **Near Field Communication (NFC) controller**

#### 8.6.4.3 NFCIP-1 framing and coding

The NFCIP-1 framing and coding in Active and Passive communication modes are defined in the NFCIP-1 standard: ISO/IEC 18092 or ECMA 340.

#### 8.6.4.4 NFCIP-1 protocol support

The NFCIP-1 protocol is not completely described in this document. For detailed explanation of the protocol refer to the ISO/IEC 18092 / ECMA340 NFCIP-1 standard. However the datalink layer is according to the following policy:

- Transaction includes initialization, anticollision methods and data transfer. This sequence must not be interrupted by another transaction.
- Speed should not be changed during a data transfer

In order not to disturb current infrastructure based on 13.56 MHz general rules to start NFC communication are defined in the following way:

- Per default NFCIP-1 device is in target mode, meaning its RF field is switched off.
- The RF level detector is active.
- Only if application requires the NFCIP-1 device shall switch to Initiator mode.
- Initiator shall only switch on its RF field if no external RF field is detected by RF Level detector during a time of TIDT.
- The initiator performs initialization according to the selected mode.

## **Near Field Communication (NFC) controller**

# 8.6.5 Card operating modes

The PN533 can be addressed like a FeliCa or ISO/IEC 14443A/MIFARE card. This means that the PN533 can generate an answer in a load modulation scheme according to the ISO/IEC 14443A/MIFARE or FeliCa interface description.

**Remark:** The PN533 does not support a secure storage of data. This has to be handled by a dedicated secure IC or a host. The secure IC is optional.

**Remark:** The PN533 can not be powered by the field in this mode and needs a power supply.

## 8.6.5.1 ISO/IEC 14443A/MIFARE card operating mode

With appropriate firmware, the PN533 can handle the ISO/IEC 14443A including the level 4, and the MIFARE protocols.

The following diagram describes the communication at the physical level. <u>Table 153</u> describes the physical parameters.

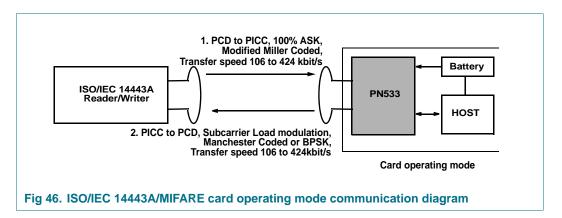


Table 153. Communication overview for ISO/IEC 14443A/MIFARE Card operating mode

Communication	scheme	ISO/IEC 14443A MIFARE	MIFARE high	er baud rate			
Baud rate		106 kbit/s	212 kbit/s 424 kbit/s				
		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,78 \mu s$	$\frac{32}{13,56MHz} \approx 2,36\mu s$			
Reader/Writer	Modulation	100% ASK	100% ASK	100% ASK			
to PN533	Bit coding	Modified Miller coding	Modified Miller coding	Modified Miller coding			

Table 153. Communication overview for ISO/IEC 14443A/MIFARE Card operating mode

Communication	n scheme	ISO/IEC 14443A MIFARE	MIFARE higher baud rate			
Baud rate		106 kbit/s	212 kbit/s	424 kbit/s		
		$\frac{128}{13,56MHz} \approx 9,44 \mu s$	$\frac{64}{13,56MHz} \approx 4,78\mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$		
PN533 to Reader/Writer	Modulation	Subcarrier load modulation	Subcarrier load modulation	Subcarrier load modulation		
	Subcarrier frequency	13.56 MHz/16	13.56 MHz/16	13.56 MHz/16		
Bit coding		Manchester coding	BPSK	BPSK		

# 8.6.5.2 FeliCa Card operating mode

With appropriate firmware, the PN533 can handle the FeliCa protocol.

The following diagram describes the communication at the physical level. <u>Table 154</u> describes the physical parameters.

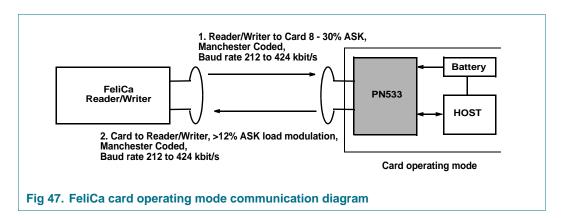


Table 154. Communication overview for FeliCa Card operating mode

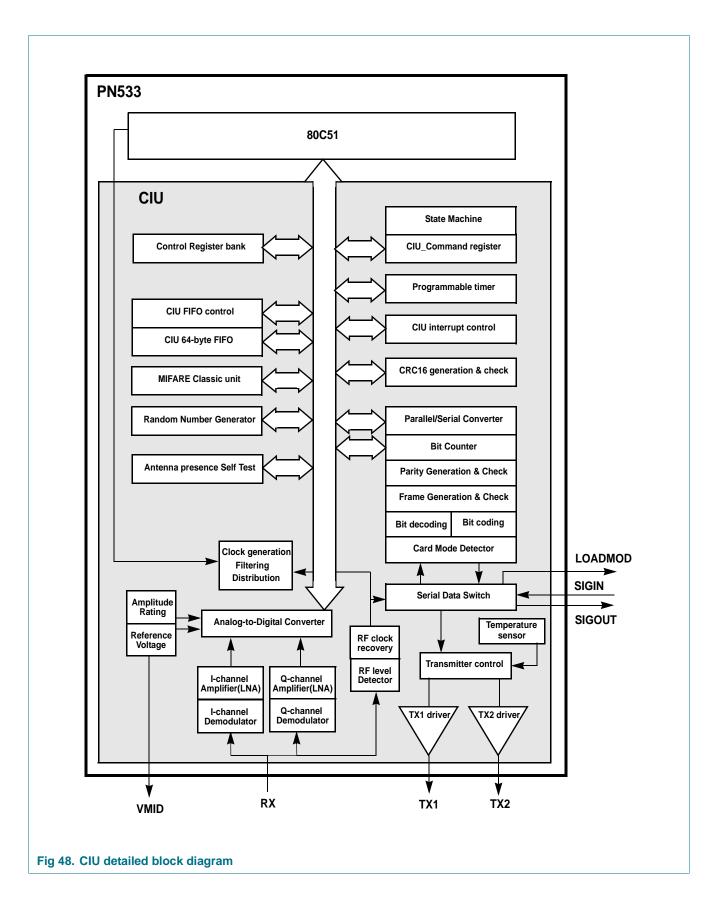
heme	FeliCa	FeliCa higher baud rate
		i choa mghei bada rate
	212 kbit/s	424 kbit/s
	$\frac{64}{13,56MHz} \approx 4,72\mu s$	$\frac{32}{13,56MHz} \approx 2,36 \mu s$
Modulation	8 - 30% ASK	8 - 30% ASK
Bit coding	Manchester coding	Manchester coding
Modulation	>12% ASK	>12% ASK
Bit coding	Manchester coding	Manchester coding
	Bit coding Modulation	$\frac{64}{13,56MHz} \approx 4,72  \mu s$ Modulation 8 - 30% ASK Bit coding Manchester coding >12% ASK

# 8.6.6 Overall CIU block diagram

The PN533 supports different contactless communication modes. The CIU supports the internal 80C51 for the different selected communication schemes such as Card Operation mode, Reader/Writer Operating mode or NFCIP-1 mode up to 424 kbit/s. The CIU generates bit- and byte-oriented framing and handles error detection according to these different contactless protocols.

Higher transfer speeds up to 3.39 Mbit/s can be handled by the digital part of the CIU. To modulate and demodulate the data an external circuit has to be connected to the communication interface pins SIGIN/SIGOUT.

**Remark:** The size and tuning of the antenna have an important impact on the achievable operating distance.



## **Near Field Communication (NFC) controller**

#### 8.6.7 Transmitter control

The signals delivered by the transmitter are on pins TX1 and pin TX2. The supply and grounds of the transmitter drivers are TVDD, TVSS1 and TVSS2.

The signals delivered are the 13.56 MHz energy carrier modulated by an envelope signal. It can be used to drive an antenna directly, using a few passive components for matching and filtering, see Section 13 "Application information" on page 227. The signals on TX1 and TX2 can be configured by the register CIU\_TxControl, see Table 213 on page 170.

The modulation index can be set by adjusting the impedance of the drivers. The impedance of the p-driver can be configured by the registers CIU\_CWGsP and CIU\_ModGsP. The impedance of the n-driver can be configured by the registers CIU\_GsNOn and CIU\_GsNOFF. Furthermore, the modulation index depends on the antenna design and tuning.

Remark: It is recommended to use a modulation index in the range of 8% for the FeliCa and NFCIP-1 communication scheme at 212 and 424 kbit/s.

The registers CIU\_TxMode and CIU\_TxAuto control the data rate and framing during the transmission and the setting of the antenna driver to support the different requirements at the different modes and transfer speeds.

In the following tables, these abbreviations are used:

- RF: 13.56 MHz clock derived from 27.12 MHz quartz divided by 2
- RF\_n: inverted 13.56 MHz clock
- GsPMos: Conductance of the transmitter PMOS
- GsNMos: Conductance of the transmitter NMOS
- CWGsP: PMOS conductance value for Continuous Wave (see <u>Table 250 on page 187</u>)
- ModGsP: refers to ModGsP[5:0], PMOS conductance value for Modulation (see <u>Table 251 on page 187</u>)
- CWGsNOn: refers to CWGsP[5:0], NMOS conductance value for Continuous Wave (see <u>Table 248 on page 186</u>)
- ModGsNOn: NMOS conductance value for Modulation when generating RF field (see Table 248 on page 186)
- CWGsNOff: NMOS conductance value for Continuous Wave when no RF is generated by the PN533 itself (see Table 240 on page 182)
- ModGsNOff: NMOS conductance value for modulation when load Modulation (see Table 240 on page 182)

**Remark:** If only 1 driver is switched on, the values for ModGsNOn and CWGsNOn are used for both drivers.

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Table 155.	Settings	for TX1

D E E	TX1 RFEn	Force 100ASK	InvTx1 RFON	InvTx1 RFOFF	Envelope	TX1	GsPMos	GsNMos	Remarks										
0 X X	Χ	0	0	0		ModGsNOff	If TX1RFEN is set to logic level 0, the pin TX1 i												
					1	0		CWGsNOff	on InvTx1RFOFF. The bit Force 100ASK has n the transconductance value.										
				1 0 1 ModGsP	the transconductance value.														
					1	1	CWGsP												
	1	0 0	0	0	0 X	X	0	RF	ModGsP	ModGsNON	If TX1RFEN is set to logic level 1, the RF phase								
			1 2		1	RF	CWGsP	CWGsNON	InvTx1RFON. The bit Force100ASK has effect;										
				X	0	RF_n	ModGsP	ModGsNON	logic level 0, TX1 is pulled to ground.										
					1	RF_n	CWGsP	CWGsNON											
<u>.</u>		1	0	X	0	0		ModGsNON											
or and a second					1	RF	CWGsP	CWGsNON											
)		1	1	1	1	1	1	1	1	1	1	1	1	Χ	0	0		ModGsNON	
					1	RF_n	CWGsP	CWGsNON											
5																			

#### Table 156. Settings for TX2

Table	oo. Settii	ngs for 12	<b>X</b> 2						
TX2 RFEn	Force 100ASK	TX2CW		InvTx2 RFOFF	Envelope	TX2	GsPMos	GsNMos	Remarks
0 X 0	0	Χ	0	0	0		ModGsNOff	If Tx2RFEn is set to logic 0, the pin TX2	
					1	0		CWGsNOff	the InvTx2RFOFF bit. The bit ForceASI Envelope modulates the transconducta
				1	0	1	ModGsP		
					1	1	CWGsP		
		1	X	0	0	0		CWGsNOff	When Tx2CW bit is set, the transconduc
					1	0		CWGsNOff	CWGsP or CWGsNOff
				1	0	1	CWGsP		-
				1	1	CWGsP			
	TX2 RFEn	TX2 Force RFEn 100ASK	TX2 Force TX2CW RFEn 100ASK	RFEn 100ASK RFON 0 X 0 X	TX2         Force RFEn 100ASK         TX2CW RFON RFOFF         InVTx2 RFON RFOFF           0         X         0         X         0	TX2 RFEN         Force 100ASK         TX2CW RFON         InVTx2 RFOFF         Envelope RFOF           0         X         0         X         0         0         1           1         1         0         1         0         1           1         X         0         0         0         1           1         X         0         0         1         1	TX2 RFC RFE RFE PRIOR         FORCE RFON RFOR RFOR RFOR RFOR RFOR RFOR RFOR	TX2 RFCN         Force 100ASK         TX2CW RFON RFOF         InvTx2 RFOF         Envelope RFOE         TX2 GSPMos           0         X         0         0         0         0           1         0         1         0         0           1         0         1         0         0           1         0         0         0         0           1         0         0         0         0           1         0         1         0         0           1         0         1         0         0	TX2 RFOR RFON RFON RFON         Envelope RFON RFOR RFOR RFOR RFOR RFOR RFOR RFOR

Table 156. Settings for TX2 ...continued

TX2 RFEn	Force 100ASK	TX2CW	InVTx2 RFON	InvTx2 RFOFF	Envelope	TX2	GsPMos	GsNMos	Remarks	
1 0	0	0	0	X	0	RF	ModGsP	ModGsNOn	•	
					1	RF	CWGsP	CWGsNOn	level 0, the phase of TX2 is depending of set to logic level 1, the transconductance	
			1	X	0	RF_n	ModGsP	ModGsNOn	CWGsNOn, independent of Envelope.	
					1	RF_n	CWGsP	CWGsNOn	- -	
		1	0	X	Χ	RF	CWGsP	CWGsNOn		
			1	X	Χ	RF_n	CWGsP	CWGsNOn		
	1	0	0	X	0	0		ModGsNOn	If TX2RFEn is set to logic level 1 and TX	
						1	RF	CWGsP	CWGsNOn	Force100ASK has effect; when Envelope pulled to ground.
			1	X	0	0		ModGsNOn	- pulled to ground.	
					1	RF_n	CWGsP	CWGsNOn		
		1	0	X	Χ	RF	CWGsP	CWGsNOn		
			1	X	Χ	RF_n	CWGsP	CWGsNOn		
	RFEn	<b>RFEn 100ASK</b> 1 0	RFEn 100ASK  1 0 0  1 1  1 0	RFEN 100ASK         RFON           1         0         0           1         0         1           1         0         0           1         0         0	RFEN 100ASK         RFON RFOFF           1         0         0         X           1         0         X           1         0         X           1         0         X           1         X           1         X           1         X	RFEN         100ASK         RFON         RFOFF           1         0         X         0           1         1         X         0           1         0         X         X           1         0         X         X           1         0         X         0           1         1         X         0           1         0         X         X	RFEN 100ASK         RFON RFOFF           1         0         X         0         RF           1         0         X         0         RF_n           1         0         X         X         RF           1         0         X         X         RF_n           1         0         X         0         0           1         0         0         0         1           1         X         0         0         0           1         X         0         0         0           1         RF_n         1         RF_n         1	RFEN 100ASK         RFON RFOFF           1         0         RF ModGsP           1         RF CWGsP           1         X         0         RF_n ModGsP           1         RF_n CWGsP           1         0         X         X         RF CWGsP           1         0         X         X         RF_n CWGsP           1         0         X         0         0           1         0         0         0         0           1         X         0         0         0           1         X         0         0         0           1         RF_n CWGsP         1         RF_n CWGsP         0	RFEN         RFON         RFOFF           1         0         RF         ModGsP         ModGsNOn           1         0         RF         CWGsP         CWGsNOn           1         0         RF         ModGsP         ModGsNOn           1         0         X         X         RF         CWGsP         CWGsNOn           1         0         X         X         RF         CWGsP         CWGsNOn           1         0         X         0         0         ModGsNOn           1         0         0         ModGsNOn         1         RF         CWGsP         CWGsNOn           1         X         0         0         ModGsNOn         1         RF         CWGsP         CWGsNOn           1         X         X         X         RF         CWGsP         CWGsNOn	

#### 8.6.8 RF level detector

The RF level detector is integrated to fulfill NFCIP-1 protocol requirements (e.g. RF collision avoidance).

Furthermore the RF level detector can be used to wake up the PN533 and to generate an interrupt.

The sensitivity of the RF level detector is adjustable in a 4-bit range using the bits RFLevel in register CIU\_RFCfg (see <u>Table 246 on page 185</u>). The sensitivity itself depends on the antenna configuration and tuning.

Possible sensitivity levels at the RX pin are listed below:

Table 157. Setting of the RF level detector

VRx typic CIU Power-down		CIU_RFCfg setting	CIU_RFCfg setting with additional amplifier
2	1.9	1111b	
1.35	1.3	1110b	
0.95	0.9	1101b	
0.6	0.57	1100b	
0.41	0.40	1011b	
0.28	0.27	1010b	
0.17	0.17	1001b	
0.12	0.12	1000b	1xxx1111b
0.085	-	0111b <sup>[1]</sup>	1xxx1110b
0.055	-	0110b <sup>[1]</sup>	1xxx1101b
0.040	-	0101b <sup>[1]</sup>	1xxx1100b
-	-	0100b[1]	1xxx1011b[1]
-	-	0011b <sup>[1]</sup>	1xxx1010b[1]
-	-	0010b[1]	1xxx1001b[1]
-	-	0001b[1]	1xxx1000b[1]
-	-	0000b[1]	1xxx0111b[1]

<sup>[1]</sup> Due to noise, it is recommended not to use this setting to avoid misleading results.

To increase the sensitivity of the RF level detector an amplifier can be activated by setting the bit RFLevelAmp in register CIU RFCfg to logic level 1 (see Table 246 on page 185).

**Remark:** With typical antenna, lower sensitivity levels without the additional amplifier set (below 1000b) can provoke misleading results because of intrinsic noise in the environment.

**Remark:** For the same reasons than above, it is recommended to use the RFLevelAmp only with upper RF level settings (above 1001b).

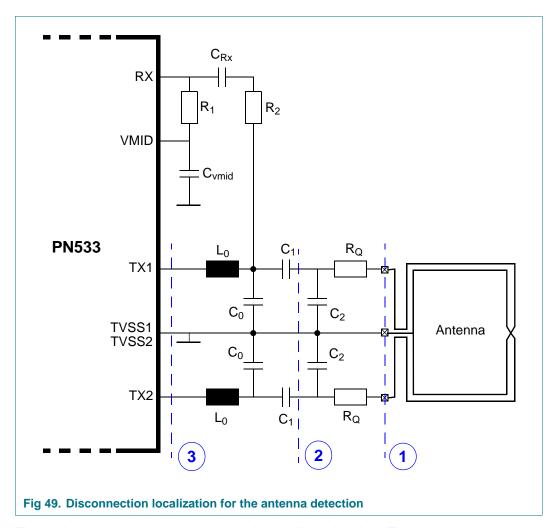
**Remark:** During the CIU Power-down mode the additional amplifier of the RF level detector is automatically switched off to ensure that the power consumption is minimal.

### 8.6.9 Antenna presence self test

The goal of the Antenna Presence Self Test is to facilitate at assembly phase the detection of the absence of the antenna and/or antenna matching components. Such a detection is done by mean of measuring the current consumption.

#### 8.6.9.1 Principle

The principle is explained with typical antenna tuning and matching components.



The testing operation can be managed via a dedicated register <u>Table 159 on page 128</u> and requires the transmitter to be activated. When activated by asserting bit 0, the detector will monitor the current consumption through the internal low dropout voltage regulator. Any violation to the current limits will be reported via bits 7 and 6 of the register.

Several levels of detection can be programmed through the register to offer a large panel of compatibility to different type of antennas. The high current threshold can be programmed from 40 mA to 150 mA with 15 mA steps (total current consumption of the IC). The low current threshold can be programmed from 5mA to 35 mA with 10 mA step (total current consumption of the IC).

There is no dedicated pin for the output of the detector. The result of the detection is to be read out from the antenna test register.

PN5331B3HN

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# **Near Field Communication (NFC) controller**

- Cases 1 and 2: If the antenna and/or the tuning network are not connected, the TVDD current is higher than the nominal one. The antenna detector detects this higher consumption and the andet\_up bit in andet\_control register is set to high
- Case 3: If the EMC filter is not correctly connected, the current within TVDD is lower than the nominal one. The antenna detector detects this lower consumption and the andet\_bot bit in andet\_control register is set to high.

To have this functionality working properly it is needed to have the transmitter generating some RF in the antenna.

# 8.6.9.2 Antenna presence detector register

Table 158. andet\_control register (address 610Ch) bit allocation

			•					
Bit	7	6	5	4	3	2	1	0
Symbol	andet_bot	andet_up	andet_	ithl[1:0]	an	det_ithh[2	::0]	andet_en
Reset	0	0	0	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

Table 159. Description of andet\_control bits

Bit	Symbol	Description
7	andet_bot	A too low power consumption has been detected
6	andet_up	A too high power consumption has been detected
5 to 4	andet_ithl[1:0	Set the low current consumption threshold to be detected
		Define the overcurrent threshold
		00: 6 mA
		01: 18 mA
		10: 29 mA
		11: 40 mA
3 to 1	andet_ithh[2:0]	Set the high current consumption threshold to be detected
		000: 38 mA
		001: 50 mA
		010: 64 mA
		011: 76 mA
		100: 89 mA
		101: 102 mA
		110: 115 mA
		111: 127 mA
0	andet_en	Enable the detection of the antenna presence detector functionality.

# 8.6.10 Random generator

The random generator is used to generate various random number needed for the NFCIP-1 protocol, as well as for MIFARE security.

It can also be used for test purpose, by generating random data through the field.

Table 160. Data\_rng register (address 6105h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		data_rng							
Reset	Χ	Х	Х	Х	Х	Х	Х	X	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 161. Description of Data\_rng bits

Bit	Symbol	Description
7 to 0	data_rng	Random number data register.

The Control\_switch\_rng register can also be used to control the behaviour of the SVDD switch.

Table 162. Control\_switch\_rng register (address 6106h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	hide_svdd_ sig	sic_switch_ overload	sic_switch_ en	-	cpu_need_ rng	random_ dataready	-
Reset	0	1	0	0	0	0	0	1
Access	R	R/W	R	R/W	R	R/W	R/W	R

Table 163. Description of Control\_switch\_rng bits

Bit	Symbol	Description
7	-	Reserved.
6	hide_svdd_sig	Configure the internal state of SIGIN and P34 in an idle state.
		This bit can be used to avoid spikes on SIGIN and P34 when the SVDD switch becomes enabled or disabled.
		When set to logic 0, the internal state of SIGIN and P34 signals are driven by respectively the pads SIGIN and P34.
		When set to logic 1, the internal state of SIGIN is fixed to 0 and the internal state of P34 is fixed set to logic 1.
5	sic_switch_overload	State of the current limitation of the SVDD switch.
		When set to logic 0, it indicates that the current consumption into the SVDD switch does no exceed the limit.
		When set to logic 1, the current limitation of the SVDD switch is activated by the switch.
4	sic_switch_en	Enable of the SVDD switch.
		When set to logic 0, the SVDD switch is disabled and the SVDD output power is tied to the ground.
		When not to logic 1, the SVDD quitable analysis and the SVDD quitaut
		When set to logic 1, the SVDD switch is enabled and the SVDD output deliver power to the secure IC and to the internal pads (SIGIN, SIGOUT and P34).

Table 163. Description of Control\_switch\_rng bits ...continued

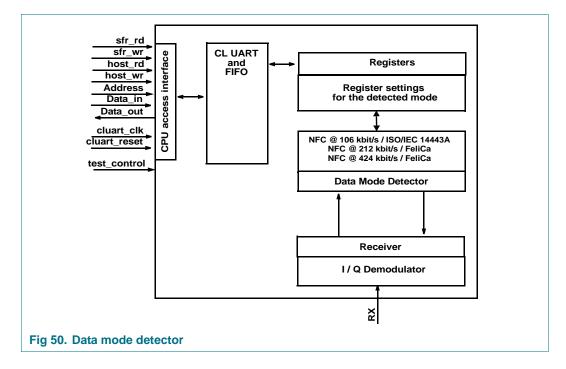
Bit	Symbol	Description
2	cpu_need_rng	Force the random number generator in running mode.
		When set to logic 0, the random number generator is under control of the CIU.
		When set to logic 1, the random number generator is forced to run.
1	random_dataready	Indicates availability of random number.
		When set to logic 1, it indicates that a new random number is available.
		It is automatically set to logic 0 when the register data_rng is read.
0	-	Reserved.

#### 8.6.11 Data mode detector

The data mode detector is able to detect received signals according to the ISO/IEC 14443A/MIFARE, FeliCa or NFCIP-1 schemes and the standard baud rates for 106 kbit/s, 212 kbit/s and 424 kbit/s in order to prepare the internal receiver in a fast and convenient way for further data processing.

The data mode detector can only be activated by the AutoColl command (see <u>Section 8.6.20.12 "AutoColl command" on page 147</u>). The mode detector is reset, when no external RF field is detected by the RF level detector.

The data mode detector could be switched off during the Autocoll command by setting the bit ModeDetOff in the register Mode to logic level 1 (see <u>Table 208 on page 167</u>).



#### 8.6.12 Serial data switch

Two main blocks are implemented in the CIU. A digital block comprising state machines, coder and decoder logic and an analog block with the modulator and antenna drivers, receiver and amplifier. The Serial Data Switch is the interface between these two blocks.

The Serial Data Switch can route the interfacing signals to the pins SIGIN and SIGOUT.

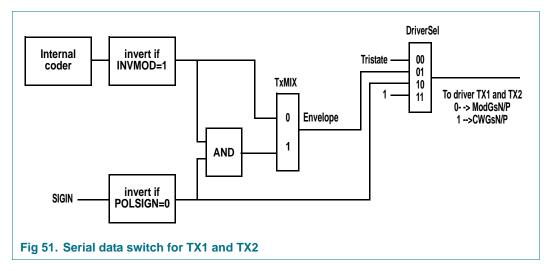
SIGOUT and SIGIN are mainly used to enable the NFC-WI/S<sup>2</sup>C interface in the secure IC to emulate card functionality with the PN533. SIGIN is capable of processing a digital signal on transfer speeds above 424 kbit/s. SIGOUT pin can also provide a digital signal that can be used with an additional external circuit to generate transfer speeds at 106 kbit/s, 212 kbit/s, 424 kbit/s and above.

Load modulation is usually performed internally by the CIU, via TX1 and TX2. However, it is possible to use LOADMOD to drive an external circuitry performing load modulation at the antenna (see optional circuitry of Figure 64 on page 227).

The Serial Data Switch is controlled by the registers CIU\_TxSel (see <u>Table 218 on page 172</u>) and CIU\_RxSel (see <u>Table 220 on page 173</u>).

#### 8.6.12.1 Serial data switch for driver and loadmod

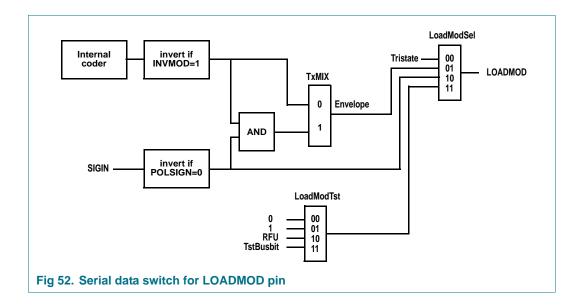
The following figure shows the serial data switch for pins TX1 and TX2.



SIGIN is in general only used for secure IC communication. If TxMix is set to logic 1 (see <u>Table 218 on page 172</u>), the driver pins are simultaneously controlled by SIGIN and the internal coder.

The following figure shows the serial data switch for the LOADMOD pin.

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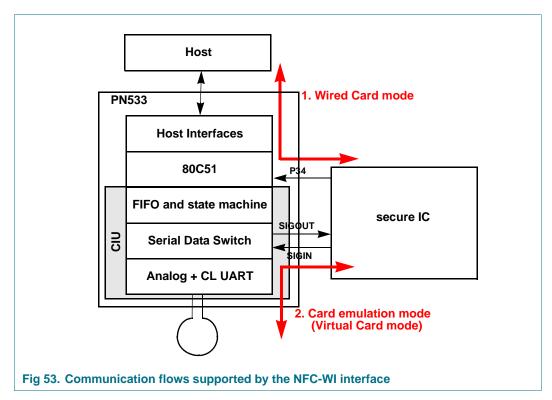


# 8.6.13 NFC-WI/S<sup>2</sup>C interface support

The NFC-WI/S²C provides the possibility to directly connect a secure IC to the PN533 in order to act as a contactless smart card IC via the PN533. The interfacing signals can be routed to the pins SIGIN and SIGOUT. SIGIN can receive either a digital FeliCa or digital ISO/IEC 14443A signal sent by the secure IC. The SIGOUT pin can provide a digital signal and a clock to communicate to the secure IC. A secure IC can be a smart card IC provided by NXP Semiconductors.

The PN533 generates the supply SVDD to the secure IC. The pins SIGIN and SIGOUT are referred to this supply, as well as pin P34 / SIC\_CLK, which can be used as an extra pin for the connection to a secure IC.

The following figure outlines the supported communication flows via the PN533 to the secure core IC.



Configured in the Wired Card mode the host controller can directly communicate to the secure IC via SIGIN/SIGOUT. In this mode the PN533 generates the RF clock and performs the communication on the SIGOUT line. To enable the Wired Card mode the clock has to be derived by the internal oscillator of the PN533 (see bits sic\_clock\_sel in Table 266 on page 191.)

Configured in Card emulation mode the secure IC can act as contactless smart card IC via the PN533. In this mode the signal on the SIGOUT line is provided by the RF field of the external Reader/Writer. To enable the Virtual Card mode the clock derived by the external RF field has to be used.

The configuration of the NFC-WI/S<sup>2</sup>C interface differs for the FeliCa and MIFARE scheme as outlined in the following chapters.

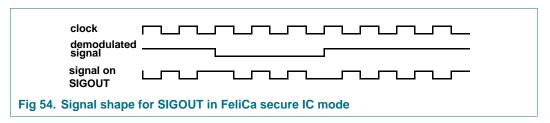
# 8.6.13.1 Signal shape for FeliCa NFC-WI/S<sup>2</sup>C interface support

The FeliCa secure IC is connected to the PN533 via the pins SIGOUT and SIGIN.

The signal at SIGOUT contains the information of the 13.56 MHz clock and the digitized demodulated signal. The clock and the demodulated signal are combined by using the logical function exclusive OR; XOR.

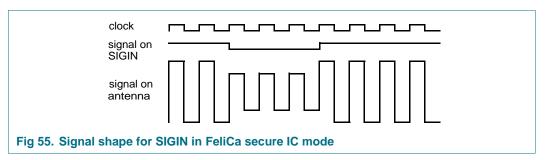
To ensure that this signal is free of spikes, the demodulated signal is digitally filtered first. The time delay for the digital filtering is in the range of one bit length. The demodulated signal changes only at a positive edge of the clock.

The register CIU\_TxSel (see Table 218 on page 172) controls the setting at SIGOUT



The response from the FeliCa secure IC is transferred from SIGIN directly to the antenna driver. The modulation is done according to the register setting of the antenna drivers.

The clock is switched to P34 / SIC\_CLK (see sic\_clk\_p34\_en bit in <u>Table 178 on page 155</u>).



**Remark:** The signal on antenna is shown in principle only. This signal is sinusoidal. The clock for SIGIN is the same as the clock for SIGOUT.

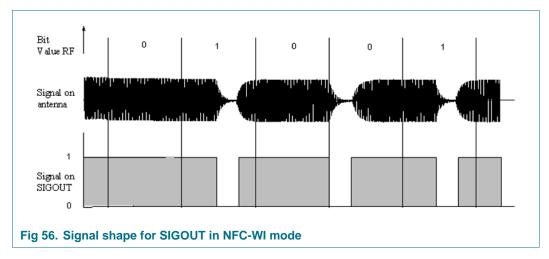
# 8.6.13.2 Signal shape for ISO/IEC14443A and MIFARE NFC-WI/S<sup>2</sup>C support

The secure IC, e.g. the SmartMX is connected to the PN533 via the pins SIGOUT, SIGIN and P34 / SIC\_CLK.

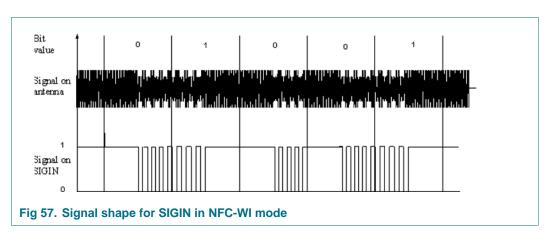
The signal at SIGOUT is a digital 13.56 MHz Miller coded signal between PVSS and SVDD. It is either derived from the external 13.56 MHz carrier signal when in Virtual Card Mode or internally generated when in Wired Card mode.

The register  $CIU\_TxSel$  controls the setting at SIGOUT.

Note: The clock settings for the Wired Card mode and the Virtual Card mode differ. Refer to the description of the bit SicClockSel in register CIU\_TestSel1.



The signal at SIGIN is a digital Manchester coded signal compliant with ISO/IEC 14443A with a subcarrier frequency of 847.5 kHz generated by the secure IC.



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#### 8.6.13.3 NFC-WI/S<sup>2</sup>C initiator mode

The PN533 includes 2 counters of 127 and 31, with digital filtering, to enable activation from the secure IC (ACT\_REQ\_Si), or the command to go from data to command mode (ESC\_REQ).

Table 164. NFC\_WI\_control register (address 610Eh) bit allocation

			_	•	•			
Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	-	nfc_wi_status	-	nfc_wi_en_ act_req_im	nfc_wi_en_ clk
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R/W	R/W	R/W

Table 165. Description of NFC\_WI\_control bits

	•	
Bit	Symbol	Description
7 to 4	-	Reserved.
3	nfc_wi_status	Indicates a NFC-WI counter has reached its limit.
		Set to logic 1, when the counter has reached its limit. It can also be used as an interrupt for the 80C51 if the IE0_6 bit is set to logic level 1 (see <u>Table 10 on page 16</u> ).
2	-	Reserved.
1	nfc_wi_en_act_req_im	<b>Selection of the NFC-WI counter</b> . This bit is used to select the 31 or 127 counter.
		When set to logic 0, the 31 counter is selected.
		When set to logic 1, the 127 counter is selected.
0	nfc_wi_en_clk	Enable the NFC-WI counters on SIGIN.
		When set to logic level 1, the counters can run and count the clock cycles within 2 and 12 MHz.

# 8.6.14 Hardware support for FeliCa and NFC polling

#### 8.6.14.1 Polling sequence functionality for initiator

- 1. Timer: The CIU has a timer, which can be programmed to generate an interrupt at the end of each timeslot, or if required at the end of the last timeslot only.
- 2. The receiver can be configured to receive frames continuously. The receiver is ready to receive immediately after the last frame has been transmitted. This mode is activated by setting to logic level 1 the bit RxMultiple in the register CIU\_RxMode. It has to be set to logic level 0 by firmware.
- The CIU adds one byte at the end of every received frame, before it is transferred into the FIFO buffer. This byte indicates whether the received frame is correct (see register Err). The first byte of each frame contains the length byte of the frame.
- 4. The length of one frame is 18 or 20 bytes (+1 byte error Info). The size of the FIFO is 64 bytes. This means 3 frames can be stored in the FIFO at the same time. If more than 3 frames are expected, the 80C51 has to read out data from the FIFO, before the FIFO is filled completely. In the case that the FIFO overflows, data is lost. (See error flag BufferOvfl).

## **Near Field Communication (NFC) controller**

#### 8.6.14.2 Polling sequence functionality for target

- 1. The 80C51 has to configure the CIU with the correct polling response parameters for the Polling command.
- To activate the automatic polling in target mode, the AutoColl Command has to be activated.
- 3. The CIU receives the polling command send out by an initiator and answers with the polling response. The timeslot is selected automatically (The timeslot itself is randomly generated, but in the range 0 to TSN, which is defined by the polling command). The CIU compares the system code, stored in byte 17 and 18 of the Config Command with the system code received with the polling command by an initiator. If the system code is equal, the CIU answers according to the configured polling response. The system code FF(hex) acts as a wildcard for the system code bytes (i.e. a target of a system code 1234(hex) answers to the polling command with one of the following system codes 1234(hex), 12FF(hex), FF34 (hex) or FFFF(hex)). If the system code does not match no answer is sent back by the PN533. If a valid command, which is not a Polling command, is received by the CIU, no answer is sent back and the command AutoColl is stopped. The received frame is stored in the FIFO.

### 8.6.14.3 Additional hardware support for FeliCa and NFC

Additionally to the polling sequence support for the FeliCa mode, the PN533 supports the check of the LEN-byte.

The received LEN-byte is checked by the registers CIU\_FeINFC1 and CIU\_FeINFC2:

DataLenMin in register CIU\_FeINFC1 defines the minimum length of the accepted frame length. This register is 6 bits long. Each value represents a length of 4.

DataLenMax in register CIU\_FeINFC2 defines the maximum length of the accepted frame. This register is 6 bits long. Each value represents a length of 4. If set to logic 0 this limit is switched off. If the length is not in the supposed area, the packed is not transferred to the FIFO and receiving is kept active.

#### Example 1:

- DataLenMin = 4
  - The length shall be greater or equal 16.
- DataLenMax = 5
  - The length shall be smaller than 20. Valid area: 16, 17, 18, 19

#### Example 2:

- DataLenMin = 9
  - The length shall be greater or equal 36.
- DataLenMax = 0
  - The length shall be smaller than 256. Valid area: 36 to 255

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## 8.6.15 CRC co-processor

The CRC preset value of the CRC co-processor can be configured to 0000h, 6363h, A671h or FFFFh depending of the bits CRCPreset in the register Mode. This is only valid when using CalcCRC command (see <a href="Section 8.6.20.7">Section 8.6.20.7</a> "CalcCRC command" on page 145).

During a communication, the preset value of the CRC coprocessor is set according to the bits CIU\_RxMode and CIU\_TxMode.

The CRC polynomial for the 16-bit CRC is fixed to  $x^{16} + x^{12} + x^5 + 1$ .

The CRC co-processor is configurable to handle the different MSB and LSB requirements for the different protocols. The bit MSBFirst in the register CIU\_Mode indicates that the data will be loaded with MSB first

The registers CRCResult-Hi and CRCResult-Lo indicate the result of the CRC calculation.

#### 8.6.16 FIFO buffer

An 64\*8 bits FIFO buffer is implemented in the CIU. It buffers the input and output data stream between the 80C51 and the internal state machine of the CIU. Thus, it is possible to handle data streams with lengths of up to 64 bytes without taking timing constraints into account.

#### 8.6.16.1 Accessing the FIFO buffer

The FIFO-buffer input and output data bus is connected to the register CIU\_FIFOData. Writing to this register stores one byte in the FIFO-buffer and increments the internal FIFO-buffer write-pointer. Reading from this register shows the FIFO-buffer contents stored at the FIFO-buffer read-pointer and decrements the FIFO-buffer read-pointer. The distance between the write- and read-pointer can be obtained by reading the register CIU\_FIFOLevel.

When the 80C51 starts a command, the CIU may, while the command is in progress, access the FIFO-buffer according to that command. Physically only one FIFO-buffer is implemented, which can be used in input- and output direction. Therefore the 80C51 has to take care, not to access the FIFO-buffer in an unintended way.

#### 8.6.16.2 Controlling the FIFO buffer

Besides writing to and reading from the FIFO-buffer, the FIFO-buffer pointers might be reset by setting the bit FlushBuffer in the register CIU\_FIFOLevel. Consequently, the FIFOLevel[6:0] bits are set to logic level 0, the bit BufferOvfl in the register CIU\_Error is set to logic level 0, the actually stored bytes are not accessible anymore and the FIFO-buffer can be filled with another 64 bytes again.

#### 8.6.16.3 Status information about the FIFO buffer

The 80C51may obtain the following data about the FIFO-buffers status:

- Number of bytes already stored in the FIFO-buffer: FIFOLevel[6:0] in register CIU FIFOLevel
- Warning, that the FIFO-buffer is quite full: HiAlert in register CIU\_Status1
- Warning, that the FIFO-buffer is quite empty: LoAlert in register CIU\_Status1
- Indication, that bytes were written to the FIFO-buffer although it was already full: BufferOvfl in register CIU Error.

BufferOvfl can be set to logic level 0 only by setting to logic level 1 bit FlushBuffer in the register CIU\_FIFOLevel.

The CIU can generate an interrupt signal

- If LoAlertIEn in register CIU\_CommIEn is set to logic 1, it will set to logic level 1 CIU\_IRQ\_0 in the register CIU\_Status1, when LoAlert in the same register changes to logic level 1.
- If HiAlertIEN in register CIU\_CommIEn is set to logic 1, it will set to logic level 1 CIU\_IRQ\_0 in the register CIU\_Status1, when HiAlert in the same register changes to logic level 1.

The flag HiAlert is set to logic 1 if only WaterLevel[5:0] bits (as set in register CIU\_WaterLevel) or less can be stored in the FIFO-buffer. It is generated by the following equation:

$$HiAlert = (64 - FIFOLenght) \le WaterLevel$$

The flag LoAlert is set to logic 1 if WaterLevel[5:0] bits (as set in register CIU\_WaterLevel) or less are actually stored in the FIFO-buffer. It is generated by the following equation:

 $LoAlert = FIFOLenght \leq WaterLevel$ 

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#### 8.6.17 CIU timer

A timer unit is implemented in the CIU: CIU\_timer. The 80C51 use CIU\_timer to manage timing relevant tasks for contactless communication. CIU\_timer may be used in one of the following configurations:

- Timeout-Counter
- Watch-Dog Counter
- Stop Watch
- Programmable One-Shot
- Periodical Trigger

CIU\_timer can be used to measure the time interval between two events or to indicate that a specific event occurred after a specific time. CIU\_timer can be triggered by events which will be explained in the following, but it does not itself influence any internal event (e.g. A timeout during data reception does not influence the reception process automatically). Furthermore, several timer related bits are set and these bits can be used to generate an interrupt.

CIU\_timer has a input clock of 6.78 MHz (derived from the 27.12 MHz quartz). CIU\_timer consists of 2 stages: 1 prescaler and 1 counter.

The prescaler is a 12 bits counter. The reload value for the prescaler can be defined between 0 and 4095 in register CIU\_TMode and CIU\_TPrescaler. This decimal value is called TPrescaler.

The reload value TReloadVal for the counter is defined with 16 bits in a range from 0 to 65535 in the registers CIU\_TReloadVal\_Lo and CIU\_TReloadVal\_Hi.

The current value of CIU\_timer is indicated by the registers CIU\_TCounterVal\_lo and CIU\_TCounterVal\_hi.

If the counter reaches 0 an interrupt will be generated automatically indicated by setting the TimerIRq flag in the register CommonIRq. If enabled, it will set to logic level 1 CIU\_IRQ\_1 in the register CIU\_Status1. TimerIRq flag can be set to logic level 1 or to logic level 0 by the 80C51. Depending on the configuration, CIU\_timer will stop at 0 or restart with the value of the registers CIU\_TReloadVal\_Lo and CIU\_TReloadVal\_Hi.

Status of CIU\_timer is indicated by the bit TRunning in the register CIU\_Status1.

CIU\_timer can be manually started by TStartNow in register Control or manually stopped by TStopNow in register Control.

Furthermore CIU\_timer can be activated automatically by setting the bit TAuto in the register CIU\_TMode to fulfill dedicated protocol requirements automatically.

The time delay of a timer stage is the reload value +1.

Maximum time:

TPrescaler = 4095, TReloadVal = 65535 => 4096\*65536/6.78 MHz = 39.59 s

Example:

To indicate 100 ms it is required to count 678 clock cycles. This means the value for TPrescaler has to be set to TPrescaler = 677. The timer has now an input clock of 100 us. The timer can count up to 65535 timeslots of 100 ms.

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## **Near Field Communication (NFC) controller**

## 8.6.18 Interrupt request system

The CIU indicates certain events by setting interrupt bits in the register CIU\_Status1 and, in addition it will set to logic level 1 CIU\_interrupt1 or CIU\_IRQ\_0. If this interrupt is enabled (see <u>Table 12 on page 17</u>) the 80C51 will be interrupted. This allows the implementation of efficient interrupt-driven firmware.

### 8.6.18.1 Interrupt sources

The following table shows the integrated interrupt flags, the corresponding source and the condition for its activation.

The interrupt flag TimerIRq in the register CIU\_CommIrq indicates an interrupt set by the timer unit. The setting is done when the timer decrements from logic level 1 down to logic level 0.

The TxIRq bit in the register CIU\_Commlrq indicates that the transmitter has finished. If the state changes from sending data to transmitting the end of frame pattern, the transmitter unit sets automatically the interrupt bit to logic level 1.

The CRC coprocessor sets the flag CRCIRq in the register CIU\_DivIrq after having processed all data from the FIFO buffer. This is indicated by the flag CRCReady set to logic level 1.

The RxIRq flag in the register CIU\_Commlrq indicates an interrupt when the end of the received data is detected.

The flag IdleIRq in the register CIU\_Commlrq is set to level 1 if a command finishes and the content of the CIU\_Command register changes to idle.

The flag HiAlertIRq in the register CIU\_CommIrq is set to level 1 if the HiAlert bit is set to logic 1, that means the Contactless FIFO buffer has reached the level indicated by the bits WaterLevel[5:0].

The flag LoAlertIRq in the register CIU\_CommIrq is set to logic level 1 if the LoAlert bit is set to logic level 1, that means the Contactless FIFO buffer has reached the level indicated by the bits WaterLevel[5:0].

The flag RFOnIRq in the register CIU\_DivIrq is set to logic level 1, when the RF level detector detects an external RF field.

The flag RFOffIRq in the register CIU\_DivIrq is set to logic level 1, when a present external RF field is switched off.

The flag ErrIRq in the register CIU\_CommIrq indicates an error detected by the CIU during sending or receiving. This is indicated by any bit set to logic 1 in register CIU\_Error.

The flag ModelRq in the register CIU\_DivIrq indicates that the data mode detector has detected the current mode.

These flags are summarized with 2 interrupt bits within the register CIU Status1:

- the high priority interrupt sources are summarized with CIU\_IRQ\_0.
- the low priority interrupt sources are summarized with CIU\_IRQ\_1.

See the register Table 191 on page 161.

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Table 166. High priority interrupt sources (CIU\_IRQ\_0)

Interrupt Flag	Interrupt source	Set automatically, WHEN
TxIRq	Transmitter	a transmitted data stream ends
RxIRq	Receiver	a received data stream ends
HiAlertIRq	FIFO-buffer	the FIFO-buffer is getting full
LoAlertIRq	FIFO-buffer	the FIFO-buffer is getting empty

#### Table 167. Low priority interrupt sources (CIU\_IRQ\_1)

Interrupt Flag	Interrupt source	Set automatically, WHEN
TimerIRq	Timer Unit	the timer counts from 1 to 0
CRCIRq	CRC-Coprocessor	all data from the FIFO buffer have been processed
IdleIRq	CIU_Command Register	a command execution finishes
RFOnIRq	RF Level Detector	an external RF field is detected
RFOffIRq	RF Level Detector	a present external RF field is switched off
ErrlRq	CIU	an error is detected
ModelRq	data mode detector	the mode has been detected

#### 8.6.19 CIU Power Reduction Modes

#### 8.6.19.1 Hard-Power-down

A Hard-Power-down is enabled when RSTPD\_N is low. None of the CIU blocks are running, even the RF level detector.

## 8.6.19.2 CIU Power-down

The CIU Power-down mode is entered immediately by setting the Power-down bit in the register CIU\_Command. All CIU blocks are switched off, except the 27.12 MHz oscillator and the RF level detector.

All registers and the FIFO will keep the content during CIU Power-down.

If the bit AutoWakeUp in the register CIU\_TxAuto is set and an external RF field is detected, the CIU Power-down mode is left automatically.

After setting bit Power-down to logic level 0 in the register CIU\_Command, it needs 1024 clocks cycle until the CIU Power-down mode is left indicated by the Power-down bit itself. Setting it to logic 0 does not immediately set it to logic level 0. It is automatically set to logic level 0 by the CIU when the CIU Power-down mode is left.

When in CIU Power-down mode and DriverSel[1:0] is no set to 00b (see <u>Table 218 on page 172</u>), to ensure a minimum impedance at the transmitter outputs, the CWGsNOn[3], CWGsNOff[3], ModGsNOn[3], ModGsNOff[3], CWGsP[5], ModGsP[5] bits are set to logic level 1, but it is not readable in the registers.

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#### 8.6.19.3 Transmitter Power-down

The Transmitter Power-down mode switches off the internal antenna drivers to turn off the RF field by setting the bits Tx1RFEn and Tx2RFEn in the register CIU\_TxControl to logic level 0. The receiver is still switched on, meaning the CIU can be accessed by a second NFC device as a NFCIP-1 target.

**Note:** In case the bit InitialRFOn has been set to logic 1, when the drivers were already switched on, it is needed either to set InitialRFOn to logic level 0, before setting the bits Tx1RFEn and Tx2RFEn in the register CIU\_TxControl to logic 0, or to set also the bits Tx1RFAutoEn and Tx2RFAutoEn in the register CIU TxAuto to logic level 0.

#### 8.6.20 CIU command set

#### 8.6.20.1 General description

The CIU behaviour is determined by an internal state machine capable to perform a certain set of commands. Writing the according command code to the CIU\_Command register starts the commands.

Arguments and/or data necessary to process a command are mainly exchanged via the FIFO buffer.

#### 8.6.20.2 General behaviour

- Each command, that needs a data stream (or data byte stream) as input will immediately process the data it finds in the FIFO buffer. An exception to this rule is the Transceive command. Using this command the transmission is started with the StartSend bit in CIU\_BitFraming register.
- Each command that needs a certain number of arguments will start processing only when it has received the correct number of arguments via the FIFO buffer.
- The FIFO buffer is not cleared automatically at command start. Therefore, it is also
  possible to write the command arguments and/or the data bytes into the FIFO buffer
  and start the command afterwards.
- Each command may be interrupted by the 80C51 by writing a new command code into the CIU\_Command register e.g.: the Idle command.

# 8.6.20.3 Commands overview

Table 168. Command overview

Command	Command code	Action
Idle	0000	No action; cancels current command execution.
Config	0001	Configures the CIU for FeliCa, MIFARE and NFCIP-1 communication.
Generate RandomID	0010	Generates 10-byte random ID number
CalcCRC	0011	Activates the CRC co-processor or perform self-test.
Transmit	0100	Transmits data from the FIFO buffer.
NoCmdChange	0111	No command change. This command can be used to modify different bits in the CIU_Command register without touching the command. e.g. Power-down bit.
Receive	1000	Activates the receiver circuitry.
SelfTest	1001	Activates the self-test. Not described in this chapter.
Transceive	1100	If bit Initiator in the register CIU_Control is set to logic 1:
		Transmits data from FIFO buffer to the antenna and activates automatically the receiver after transmission is finished.
		If bit Initiator in the register CIU_Control is set to logic 0:
		Receives data from antenna and activates automatically the transmitter after reception.
AutoColl	1101	Handles FeliCa polling (Card operating mode only) and MIFARE anticollision (Card operating mode only)
MFAuthent	1110	Performs the MIFARE 1 KB or MIFARE 4 KB emulation authentication in MIFARE Reader/Writer mode only.
Soft Reset	1111	Resets the CIU.

#### 8.6.20.4 Idle command

The CIU is in idle mode. This command is also used to terminate the actual command.

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#### 8.6.20.5 Config command

To configure the automatic MIFARE Anticollision, FeliCa Polling and NFCID3, the data used for these transactions have to be stored internally. All the following data have to be written to the FIFO in this order:

- SENS\_RES (2 bytes): in order byte0, byte1
- NFCID1 (3 bytes): in order byte0, byte1, byte 2; the first NFCID1 byte if fixed to 08h and the check byte is calculated automatically
- SEL\_RES (1 byte)
- Polling response (2 bytes (shall be 01h, FEh)+ 6 bytes NFCID2 + 8 bytes Pad + 2 bytes system code)
- NFCID3 (1 byte)

In total 25 bytes which are transferred into an internal buffer with the Config command.

The complete NFCID3 is 10 bytes long and consist of the 3 NFCID1 bytes, the 6 NFCID2 bytes and the NFCID3 byte listed above.

To read out this configuration (after it has been loaded), the command Config with an empty FIFO buffer has to be started. In this case the 25 bytes are transferred from the internal buffer to the FIFO.

The CIU has to be configured after each power up, before using the automatic Anticollision/Polling function (AutoColl command). During a Hard-Power-down (RSTPD\_N set to logic level 0) this configuration remains unchanged.

This command terminates automatically when finished and the active command is Idle.

#### 8.6.20.6 Generate RandomID command

This command generates a 10-byte random number stored in the internal 25 bytes buffer and overwrites the 10 NFCID3 bytes. This random number might be used for fast generation of all necessary ID bytes for the automatic Anticollision / Polling function.

Note: To configure the CIU, Config command has to be used first.

This command terminates automatically when finished and the active command is Idle.

#### 8.6.20.7 CalcCRC command

The content of the FIFO is transferred to the CRC co-processor and a CRC calculation is started. The result is stored in the CRCResult register. The CRC calculation is not limited to a dedicated number of bytes. The calculation is not stopped when the FIFO gets empty during the data stream. The next byte written to the FIFO is added to the calculation.

The preset value of the CRC is defined by the CRCPreset bits of the register CIU\_Mode, and the chosen value is loaded to the CRC co-processor when the command is started.

This command has to be terminated by firmware by writing any command to the CIU\_Command register e.g. the Idle command.

If SelfTest in register CIU\_AutoTest is set to logic 1, the CRC co-processor is in Self Test mode and performs a digital self-test. The result of the self-test is written in the FIFO.

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#### 8.6.20.8 Transmit command

The content of the FIFO is transmitted immediately after starting the command. Before transmitting FIFO content, all relevant register settings have to be set to transmit data in the selected mode.

This command terminates automatically when the FIFO gets empty and the active command is Idle. It can be terminated by any other command written to the CIU\_Command register.

#### 8.6.20.9 NoCmdChange command

This command does not influence any ongoing command in the CIU\_Command register. It can be used to manipulate any bit except the command bits in the CIU\_Command register, e.g. the bits RcvOff or Power-down.

#### 8.6.20.10 Receive command

The CIU activates the receiver path and waits for any data stream to be received. The correct settings for the expected mode have to be set before starting this command.

This command terminates automatically when the receive data stream ends and the active command is Idle. This is indicated either by the end of frame pattern or by the length byte depending on the selected framing and speed.

**Note:** If the bit RxMultiple in the register CIU\_RxMode is set to logic 1, the Receive command does not terminate automatically. It has to be terminated by setting any other command in the CIU\_Command register.

#### 8.6.20.11 Transceive command

This circular command repeats transmitting data from the FIFO and receiving data from the RF field continuously. If the bit Initiator in the register CIU\_Control is set to logic 1, it indicates that the first action is transmitting and after having finished transmission the receiver is activated to receive data. If the bit Initiator in the CIU\_Control register is set to logic 0, the first action is receiving and after having received a data stream, the transmitter is activated to transmit data. In the second configuration the PN533 first acts as a receiver and if a data stream is received it switches to the Transmit mode.

Table 169. Transceive command scenario

Communication step	Initiator =1	Initiator=0
1	Send	Receive
2	Receive	Send
3	Send	Receive
4	Receive	Send

Each transmission process has to be started with setting bit StartSend in the register CIU\_BitFraming. This command has to be cleared by firmware by writing any command to the CIU\_Command register e.g. the command idle.

**Note:** If the bit RxMultiple in register CIU\_RxMode is set, this command will never leave the receiving state, because the receiving will not be cancelled automatically.

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#### 8.6.20.12 AutoColl command

This command automatically handles the MIFARE activation and the FeliCa polling in the Card Operation mode. The bit Initiator in the CIU\_Control register has to be set to logic level 0 for correct operation. During this command, Mode Detector is active if not deactivated by setting the bit ModeDetOff in the CIU\_Mode register. After Mode Detector detects a mode, the mode dependent registers are set according to the received data. In case of no external RF field this command resets the internal state machine and returns to the initial state but it will not be terminated.

When the Autocoll command terminates the Transceive command gets active.

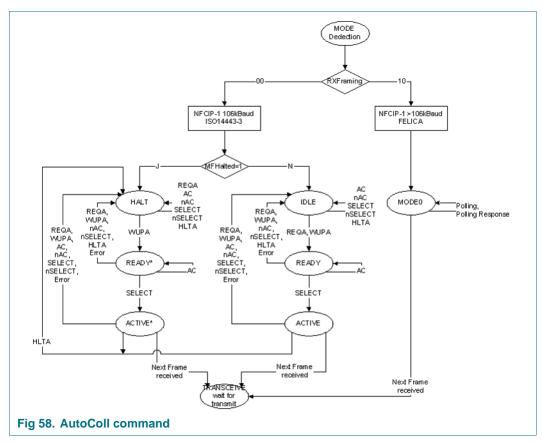
#### **During Autocoll command:**

- The CIU interrupt bits, except RfOnIRq, RfOffIRq and SIGINActIRq (see <u>Table 188 on page 159</u>), are not supported. Only the last received frame will serve the CIU interrupts.
- During ISO/IEC 14443A activation, TxCRCEn and RxCRCEn bits are defined by the AutoColl command. The changes cannot be observed at the CIU\_TxMode and CIU\_RxMode registers. When the Transceive command is active, the value of the bits is relevant.
- During Felica activation (polling), TxCRCEn and RxCRCEn bits are always relevant and are not overruled by the Autocoll command. Their value must be set to logic level 1 according the FeliCa protocol.

Note: Pay attention, that the FIFO will also receive the two CRC check bytes of the last command, even if they are already checked and correct, and if the state machine (Anticollision and Select routine) has not been executed, and 106 kbit is detected.

This command can be cleared by firmware by writing any other command to the CIU\_Command register, e.g. the Idle command. Writing the same content again to the CIU\_Command register resets the state machine.

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• NFCIP-1 106 kbps passive communication mode:

The MIFARE anticollision is finished and the command changes automatically to Transceive. The FIFO contains the ATR\_REQ frame including the start byte F0h. The bit TargetActivated in the register CIU Status2 is set to logic 1

• NFCIP-1 212 and 424 kbps passive communication mode:

The FeliCa polling command is finished and the command has automatically changed to Transceive. The FIFO contains the ATR\_REQ frame. The bit TargetActivated in the register CIU\_Status2 is set to logic 1.

• NFCIP-1 106, 212 and 424 kbps active communication mode:

This command is changing automatically to Transceive. The FIFO contains the ATR\_REQ frame. The bit TargetActivated in the register CIU\_Status2 is set to logic 0. For 106 kbps only, the first byte in the FIFO indicates the start byte F0h and the CRC is added into the FIFO.

ISO/IEC 14443A/MIFARE (Card Operating mode):

The MIFARE anticollision is finished and the command has automatically changed to Transceive. The FIFO contains the first command after the Select. The bit TargetActivated in the register CIU\_Status2 is set to logic 1.

FeliCa (Card Operating mode):

The FeliCa polling command is finished and the command has automatically changed to Transceive. The FIFO contains the command after the Polling in the FeliCa protocol. The bit TargetActivated in the register CIU\_Status2 is set to logic 1.

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#### 8.6.20.13 MFAuthent command

This command handles the MIFARE authentication in Reader/Writer mode to enable a secure communication to any MIFARE 1 KB and MIFARE 4 KB emulation card. The following data shall be written to the FIFO before the command can be activated:

- Authentication command code (60h for key A, 61h for key B)
- Block address
- Sector key byte 0
- Sector key byte 1
- Sector key byte 2
- · Sector key byte 3
- Sector key byte 4
- Sector key byte 5
- Card serial number byte 0
- Card serial number byte 1
- Card serial number byte 2
- Card serial number byte 3

In total 12 bytes shall be written to the FIFO.

Note: When the MFAuthent command is active, any FIFO access is blocked. Anyhow if there is an access to the FIFO, the bit WrErr in the register CIU\_Error is set to logic level 1.

This command terminates automatically when the MIFARE 1KB or MIFARE 4 KB emulation card is authenticated. The bit MFCrypto1On in the register CIU\_Status2 is set to logic level 1.

This command does not terminate automatically when the card does not answer, therefore CIU timer should be initialized to automatic mode. In this case, beside the bit IdleIRq, the bit TimerIRq can be used as termination criteria. During authentication processing, the bits RxIRq and TxIRq of CIU\_CommIrq register are blocked.

The Crypto1On bit is only valid after termination of the MFAuthent command (either after processing the authentication or after writing the Idle command in the register CIU\_Command).

In case there is an error during the MIFARE authentication, the ProtocolErr bit in the CIU\_Error register is set to logic 1 and the Crypto1On bit in CIU\_Status2 register is set to logic 0.

#### 8.6.20.14 SoftReset command

This command performs a reset of the CIU. The configuration data of the internal buffer remains unchanged. All registers are set to the reset values.

When SoftReset is finished, the active command switches to Idle.

#### 8.6.21 CIU tests signals

#### 8.6.21.1 CIU self-test

The CIU has the capability to perform a self-test. To start the self-test the following procedure has to be performed:

- 1. Perform a SoftReset.
- 2. Clear the internal buffer by writing 25 bytes of 00h and perform the Config command.
- 3. Enable the self-test by writing the value 09h to the register CIU\_AutoTest.
- 4. Write 00h to the FIFO.
- 5. Start the self-test with the CalcCRC command.
- 6. The self-test will be performed.
- 7. When the self-test is finished, the FIFO is contains the following bytes:
  - Correct answer for VersionReg equal to 80h:

```
0x00, 0xaa, 0xe3, 0x29, 0x0c, 0x10, 0x29, 0x6b 0x76, 0x8d, 0xaf, 0x4b, 0xa2, 0xda, 0x76, 0x99 0xc7, 0x5e, 0x24, 0x69, 0xd2, 0xba, 0xfa, 0xbc 0x3e, 0xda, 0x96, 0xb5, 0xf5, 0x94, 0xb0, 0x3a 0x4e, 0xc3, 0x9d, 0x94, 0x76, 0x4c, 0xea, 0x5e 0x38, 0x10, 0x8f, 0x2d, 0x21, 0x4b, 0x52, 0xbf 0xfb, 0xf4, 0x19, 0x94, 0x82, 0x5a, 0x72, 0x9d
```

0xba, 0x0d, 0x1f, 0x17, 0x56, 0x22, 0xb9, 0x08

#### 8.6.21.2 CIU test bus

The test bus is implemented for production test purposes. The following configuration can be used to improve the design of a system using the PN533. The test bus allows to route internal signals to output pins.

The Observe\_testbus register is used to enable this functionality.

Table 170. Observe\_testbus register (address 6104h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	-	-	-	-	-	-	-	observe_ciu	
Reset	0	0	0	0	0	0	0	0	
Access	R	R	R	R	R	R	R	R/W	

Table 171. Description of Observe\_testbus bits

Bit	Symbol	Description
7 to 1	-	Reserved.
0	observe_ciu	Configure the pads P3x (P30 to P35), RSTOUT_N and P70_IRQ to observe internal CIU data bus.
		When set to logic 1, the pads are configured in output mode and show the internal data bus D0 to D6 of the CIU. P70_IRQ is the 13.56 MHz digital clock of CIU (generated from field or crystal).

The test bus signals are selected by accessing TestBusSel in register CIU\_TestSel2.

Table 172. TstBusBitSel set to 07h

Test bus bit	Test signal	Comments				
D6	sdata	shows the actual received data value.				
D5	scoll	shows if in the actual bit a collision has been detected (106 kbit/s only)				
D4	svalid	shows if sdata and scoll are valid				
D3	sover	shows that the receiver has detected a stop bit (ISO/IEC 14443A/MIFARE mode only)				
D2	RCV_reset	shows if the receiver is reset				
D1	RFon filtered	shows the value of the internal RF level detector				
D0	Envelope	shows the output of the internal coder				

Table 173. TstBusBitSel set to 0Dh

Test bus bit	Test signal	Comments
D6	clkstable	shows if the oscillator delivers a stable signal
D5	clk27/8	shows the output signal of the oscillator divided by 8
D4	clk27rf/8	shows the clk27rf signal divided by 8
D3	clk13/4	shows the clk13rf divided by 4
D2	clk27	shows the output signal of the oscillator
D1	clk27rf	shows the RF clock multiplied by 2
D0	clk13rf	shows the RF clock of 13.56 MHz

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### 8.6.21.3 Test signals at pin AUX

Each signal can be switched to pin AUX1 or AUX2 by setting SelAux1 or SelAux2 in the register CIU\_AnalogTest. See <u>Table 280 on page 195</u>

#### 8.6.21.4 PRBS

Enables the Pseudo Random Bit Stream of 9-bit or 15-bit length sequence, PRBS9 or PRBS15, according to ITU-TO150. To start the transmission of the defined datastream, Transmit command has to be activated. The preamble/Sync byte/start bit/parity bit are generated automatically depending on the selected mode.

**Note:** All relevant registers to transmit data have to be configured before entering PRBS mode according ITU-TO150.

# 8.6.22 CIU memory map

The registers of the CIU are either map into the SFR or into the XRAM memory space.

Table 174. Contactless Interface Unit SFR memory map

ADDR	Byte size	Register name	Description
D1h	1	CIU_Command	Starts and stops the command execution
D2h	1	CIU_CommlEn	Control bits to enable and disable the passing of interrupt requests
D3h	1	CIU_DivlEn	Control bits to enable and disable the passing of interrupt requests
D4h	1	CIU_Commlrq	Contains common Interrupt Request flags
D5h	1	CIU_DivIrq	Contains diverse Interrupt Request flags
D6h	1	CIU_Error	Error flags showing the error status of the last command executed
DFh	1	CIU_Status1	Contains status flags of the CRC, Interrupt Request System and FIFO buffer
E9h	1	CIU_Status2	Contain status flags of the Receiver, Transmitter and Data Mode Detector
EAh	1	CIU_FIFOData	in- and output of 64 bytes FIFO buffer
EBh	1	CIU_FIFOLevel	Indicates the number of bytes stored in the FIFO
ECh	1	CIU_WaterLevel	Defines the thresholds for FIFO under- and overflow warning
EDh	1	CIU_Control	Contains miscellaneous Control bits
EEh	1	CIU_BitFraming	Adjustments for bit oriented frames
EFh	1	CIU_Coll	Bit position of the first bit collision detected on the RF-interface

Table 175. Contactless Interface Unit extension memory map

ADDR	Byte size	Register name	Description
6301h	1	CIU_Mode	Defines general modes for transmitting and receiving
6302h	1	CIU_TxMode	Defines the transmission data rate and framing during transmission
6303h	1	CIU_RxMode	Defines the transmission data rate and framing during receiving
6304h	1	CIU_TxControl	Controls the logical behaviour of the antenna driver pins TX1 and TX2
6305h	1	CIU_TxAuto	Controls the settings of the antenna driver
6306h	1	CIU_TxSel	Selects the internal sources for the antenna driver
	1	CIU_RxSel	Selects internal receiver settings
6308h	1	CIU_RxThreshold	Selects thresholds for the bit decoder
6309h	1	CIU_Demod	Defines demodulator settings
630Ah	1	CIU_FeINFC1	Defines the length of the valid range for the received frame
630Bh	1	CIU_FeINFC2	Defines the length of the valid range for the received frame
630Ch	1	CIU_MifNFC	Controls the communication in ISO/IEC 14443/MIFARE and NFC target mode at 106 kbit/s
630Dh	1	CIU_ManualRCV	Allows manual fine tuning of the internal receiver

Table 175. Contactless Interface Unit extension memory map ...continued

ADDR		Register name	Description
	size	<b>U</b>	
630Eh	1	CIU_TypeB	Configure the ISO/IEC 14443 type B
630Fh	1	-	Reserved
6310h	1	-	Reserved
6311h	1	CIU_CRCResultMSB	Shows the actual MSB values of the CRC calculation
6312h	1	CIU_CRCResultLSB	Shows the actual LSB values of the CRC calculation
6313h	1	CIU_GsNOFF	Selects the conductance of the antenna driver pins TX1 and TX2 for load modulation when own RF field is switched OFF
6314h	1	CIU_ModWidth	Controls the setting of the width of the Miller pause
6315h	1	CIU_TxBitPhase	Bit synchronization at 106 kbit/s
6316h	1	CIU_RFCfg	Configures the receiver gain and RF level
6317h	1	CIU_GsNOn	Selects the conductance of the antenna driver pins TX1 and TX2 for modulation, when own RF field is switched ON
6318h	1	CIU_CWGsP	Selects the conductance of the antenna driver pins TX1 and TX2 when not in modulation phase
6319h	1	CIU_ModGsP	Selects the conductance of the antenna driver pins TX1 and TX2 when in modulation phase
631Ah	1	CIU_TMode	Defines settings for the internal timer
631Bh	1	CIU_TPrescaler	
631Ch	1	CIU_TReloadVal_hi	Describes the 16-bit long timer reload value (Higher 8 bits)
631Dh	1	CIU_TReloadVal_lo	Describes the 16-bit long timer reload value (Lower 8 bits)
631Eh	1	CIU_TCounterVal_hi	Describes the 16-bit long timer actual value (Higher 8 bits)
631Fh	1	CIU_TCounterVal_lo	Describes the 16-bit long timer actual value (Lower 8 bits)
6320h	1	-	Reserved
6321h	1	CIU_TestSel1	General test signals configuration
6322h	1	CIU_TestSel2	General test signals configuration and PRBS control
6323h	1	CIU_TestPinEn	Enables test signals output on pins.
6324h	1	CIU_TestPinValue	Defines the values for the 8-bit parallel bus when it is used as I/O bus
6325h	1	CIU_TestBus	Shows the status of the internal test bus
6326h	1	CIU_AutoTest	Controls the digital self-test
6327h	1	CIU_Version	Shows the CIU version
6328h	1	CIU_AnalogTest	Controls the pins AUX1 and AUX2
6329h	1	CIU_TestDAC1	Defines the test value for the TestDAC1
632Ah	1	CIU_TestDAC2	Defines the test value for the TestDAC2
632Bh	1	CIU_TestADC	Show the actual value of ADC I and Q
632Ch	1	-	Reserved for tests
632Dh	1	-	Reserved for tests
632Eh	1	-	Reserved for tests
632Fh	1	CIU_RFlevelDet	Power down of the RF level detector
6330h	1	CIU_SIC_CLK_en	Enables the use of secure IC clock on P34 / SIC_CLK.

# 8.6.23 CIU register description

#### 8.6.23.1 CIU register bit behaviour

Depending of the functionality of a register, the access condition to the bits can vary. The following table describes the access conditions:

Table 176. Behavior of register bits

Abbreviation	Behavior	Description
R/W	Read and Write	These bits can be written and read by the 80C51. Since they are used only for control means, there content is not influenced by internal state machines, e.g. CIU_CommIEn may be written and read by the CPU. It will also be read by internal state machines, but never changed by them.
DY	DYnamic	These bits can be written and read by the 80C51. Nevertheless, they may also be written automatically by CIU internal state machines, e.g. the commands in the CIU_Command register change their values automatically after their execution.
R	Read only	These registers hold flags, which value is determined by CIU internal states only, e.g. the CRCReady register can not be written from external but shows CIU internal states.
W	Write only	These registers are used for control means only. They may be written by the 80C51 but can not be read. Reading these registers returns always logic level 0.
Reserved		These registers are not implemented or reserved for NXP testing use.

### 8.6.23.2 CIU\_SIC\_CLK\_en register (6330h)

Enables the use of P34 / SIC\_CLK as secure IC clock.

Table 177. CIU\_SIC\_CLK\_en register (address 6330h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	sic_clk_p34_en	-	-	-	Errorbusbitenable	Errorbusbitsel[2:0		I[2:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	R	R/W	R/W	R/W	R/W

Table 178. Description of CIU\_SIC\_CLK\_en bits

Bit	Symbol	Description	Description					
7	sic_clk_p34_en	Set to logic 1,	Set to logic 1, this bit configures P34 / SIC_CLK to be used as secure C clock: SIC_CLK.					
		Set to logic 0,	P34 / SIC_CLK is in normal mode: P34.					
6 to 4	-	Reserved	eserved					
3	Errorbusbitenable	Set to logic 1, enable the error source selected by Errorbusbitsel on AUX pads according to SelAux1 and SelAux2 bits (code 1010b).						
2 to 0	Errorbusbitsel[2:0]	Define the error source on ErrorBusBit:						
		Value	Description					
		000	selects ProtocollErr on test bus					
		001	selects ParityErr on test bus					
		010	selects CRCErr on test bus					
		011	selects CollErr on test bus					
		100	selects BufferOvfl on test bus					
		101	selects RFErr on test bus					
		110	selects TempErr on test bus					
		111	selects WrErr on test bus					
		_	•					

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## 8.6.23.3 CIU\_Command register (D1h or 6331h)

Starts and stops the command execution.

Table 179. CIU\_Command register (address D1h or 6331h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	RcvOff	Power-down		Comi	mand	
Reset	0	0	1	0	0	0	0	0
Access	R	R	R/W	DY	DY	DY	DY	DY

### Table 180. Description of CIU\_Command bits

Bit	Symbol	Description
7 to 6	-	Reserved
5	RcvOff	Set to logic 1, the analog part of the receiver is switched off.
4	Power-down	Set to logic 1, the CIU Power-down mode is entered. This means, internal current consuming blocks of the contactless analog module are switched off, except for the RF level detector.
		Set to logic 0, the PN533 starts the wake up procedure. During this procedure this bit still shows a logic level 1. A logic level 0 indicates that the PN533 is ready for operations; see <a href="Section 8.6.19.2">Section 8.6.19.2</a> "CIU Power-down" on page 142.
		<b>Note:</b> The Power-down bit can not be set, when the SoftReset command has been activated.
3 to 0	Command	Activates a command according the Command Code.
		Reading this register shows, which command is actually executed. See Section 8.6.20 "CIU command set" on page 143.

### 8.6.23.4 CIU\_CommlEn register (D2h or 6332h)

Control bits to enable and disable the passing of interrupt requests.

Table 181. CIU\_CommlEn register (address D2h or 6332h) bit allocation

Bit	it 7 6 5		4	4 3		1	0	
Symbol	-	TxIEn	RXIEn	SiginAct IEn	ModelEn	CRCIEn	RfOnIEn	RfOfflEn
Reset	0	0	0	0	0	0	0	0
Access	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 182. Description of CIU\_CommlEn bits

Bit	Symbol	Description
7	-	Reserved.
6	TxIEn	When set to logic level 1, allows the transmitter interrupt request (indicated by bit TxIRq) to be propagated to CIU_IRQ_1.
5	RxIEn	When set to logic level 1, allows the receiver interrupt request (indicated by bit RxIRq) to be propagated to CIU_IRQ_1.
4	IdleIEn	When set to logic level 1, allows the idle interrupt request (indicated by bit IdleIRq) to be propagated to CIU_IRQ_0.
3	HiAlertIEn	When set to logic level 1, allows the high alert interrupt request (indicated by bit HiAlertIRq) to be propagated to CIU_IRQ_1.
2	LoAlertIEn	When set to logic level 1, allows the low alert interrupt request (indicated by bit LoAlertIRq) to be propagated to CIU_IRQ_1.
1	ErrlEn	When set to logic level 1, allows the error interrupt request (indicated by bit ErrIRq) to be propagated to CIU_IRQ_0.
0	TimerIEn	When set to logic level 1, allows the timer interrupt request (indicated by bit TimerIRq) to be propagated to CIU_IRQ_0.

## 8.6.23.5 CIU\_DivIEn register (D3h or 6333h)

Controls bits to enable and disable the passing of interrupt requests.

Table 183. CIU\_DivIEn register (address D3h or 6333h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	SiginAct IEn	ModelEn	CRCIEn	RfOnlEn	RfOfflEn
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R/W	R/W	R/W	R/W	R/W

Table 184. Description of CIU\_DivIEn bits

Bit	Symbol	Description
7 to 5	-	Reserved.
4	SiginAct IEn	Allows the SIGIN active interrupt request to be propagated to CIU_IRQ_0.
3	ModelEn	When set to logic level 1, allows the mode interrupt request (indicated by bit ModelRq) to be propagated to CIU_IRQ_0.
2	CRCIEn	When set to logic level 1, allows the CRC interrupt request (indicated by bit CRCIRq) to be propagated to CIU_IRQ_0.
1	RfOnlEn	When set to logic level 1, allows the RF field on interrupt request (indicated by bit RfOnIRq) to be propagated to CIU_IRQ_0.
0	RfOfflEn	When set to logic level 1, allows the RF field off interrupt request (indicated by bit RfOffIRq) to be propagated to CIU_IRQ_0.

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## 8.6.23.6 CIU\_Commlrq register (D4h or 6334h)

Contains common CIU interrupt request flags.

Table 185. CIU\_Commlrq register (address D4h or 6334h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Set1	TxIRq	RxIRq	IdleIrq	HiAltertIRq	LoAlertIRq	ErrlRq	TimerIRq
Reset	0	0	0	1	0	1	0	0
Access	W	DY	DY	DY	DY	DY	DY	DY

#### Table 186. Description of CIU\_CommIRQ bits

Bit	Symbol	Description							
7	Set1	When set to logic 0 during write operation, the bit set to logic level 1 in the write command are written to logic level 0 in the register.							
		When set to logic 1 during write operation, the bit set to logic level 1 in the write command are written to logic level 1 in the register.							
6	TxIRq	Set to logic 1, immediately after the last bit of the transmitted data was sent out.							
5	RxIRq	Set to logic 1 when the receiver detects the end of a valid datastream.							
		If the RxNoErr bit in CIU_RxMode register is set to logic 1, RxIRQ is only set to logic 1 when data bytes are available in the FIFO.							
4	IdleIrq	Set to logic 1, when a command terminates by itself e.g. when the CIU_Command register changes its value from any command to the Idle command.							
		If an unknown command is started, the CIU_Command register changes its value to the Idle command and the IdleIRq bit is set.							
		Starting the Idle Command by the 80C51 does not set IdleIRq bit.							
3	HiAltertIRq	Set to logic 1, when HiAlert bit in CIU_Status1 register is set to logic level 1. In opposition to HiAlert, HiAlertIRq stores this event and can only be reset by Set1 bit.							
2	LoAlertIRq	Set to logic 1, when LoAlert bit in CIU_Status1 register is set. In opposition to LoAlert, LoAlertIRq stores this event and can only be reset by Set1 bit.							
1	ErrlRq	Set to logic 1, if any error flag in the CIU_Error register is set.							
0	TimerIRq	Set to logic 1, when the timer decrements the TimerValue register to zero.							

<sup>[1]</sup> Remark: All bits in the register CIU\_Commlrq shall be set to logic level 0 by firmware.

# 8.6.23.7 CIU\_DivIrq register (D5h or 6335h)

Contains miscellaneous interrupt request flags.

Table 187. CIU\_DivIrq register (address D5h or 6335h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Set2	-	-	SiginActIrq	ModelRq	CRCIRq	RfOnIRq	RfOffIRq
Reset	0	0	0	Χ	0	0	X	X
Access	W	R	R	DY	DY	DY	DY	DY

#### Table 188. Description of CIU\_DivIrq bits

Bit	Symbol	Description
7	Set2	When set to logic 0 during write operation, the bit set to logic level 1 in the write command are written to logic level 0 in the register.
		When set to logic 1 during write operation, the bit set to logic level 1 in the write command are written to logic level 1 in the register.
6 to 5	-	Reserved.
4	SiginActIrq	Set to logic level 1, when SIGIN is active. See <u>Section 8.6.13 "NFC-WI/S<sup>2</sup>C interface support" on page 133</u> . This interrupt is set when either a rising or falling edge is detected on SIGIN.
3	ModelRq	Set to logic 1, when the mode has been detected by the Data Mode Detector.
		<b>Note:</b> The Data Mode Detector can only be activated by the AutoColl command and is terminated automatically having the detected the communication mode.
		<b>Note:</b> The Data Mode Detector is automatically restarted after each RF reset.
2	CRCIRq	Set to logic 1, when the CRC command is active and all data are processed.
1	RfOnIRq	Set to logic 1, when an external RF field is detected.
0	RfOffIRq	Set to logic 1, when an present external RF field is switched off.

## 8.6.23.8 CIU\_Error register (D6h or 6336h)

Error flags showing the error status of the last command executed.

Table 189. CIU\_Error register (address D6h or 6336h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	WrErr	TempErr	RFErr	BufferOvfl	CollErr	CRCErr	ParityErr	ProtocollErr
Reset	0	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R

#### Table 190. Description of CIU\_Error bits

Bit	Symbol	Description
7	WrErr	Set to logic 1, when data is written into the FIFO by the 80C51 during the AutoColl command or MFAuthent command or if data is written into the FIFO by the 80C51 during the time between sending the last bit on the RF interface and receiving the last bit on the RF interface.
6	TempErr[1]	Set to logic 1, if the internal temperature sensor detects overheating. In this case the antenna drivers are switched off automatically.
5	RFErr	Set to logic 1, if in active communication mode the counterpart does not switch on the RF field in time as defined in NFCIP-1 standard.
		<b>Note:</b> RFErr is only used in active communication mode. The bit RxFraming or the bit TxFraming has to be set to 01h to enable this functionality.
4	BufferOvfl	Set to logic 1, if the 80C51 or if the internal state machine (e.g. receiver) tries to write data into the FIFO buffer although the FIFO buffer is already full.
3	CollErr	Set to logic 1, if a bit-collision is detected. It is set to logic level 0 automatically at receiver start phase. This flag is only valid during the bitwise anticollision at 106 kbit/s. During communication schemes at 212 and 424 kbit/s this flag is always set to logic level 0.
2	CRCErr	Set to logic 1, if RxCRCEn in CIU_RxMode register is set to logic level 1 and the CRC calculation fails. It is set to logic level 0 automatically at receiver start-up phase.
1	ParityErr	Set to logic 1, if the parity check has failed. It is set to logic level 0 automatically at receiver start-up phase. Only valid for ISO/IEC 14443A/MIFARE or NFCIP-1 communication at 106 kbit/s.
0	ProtocollErr	Set to logic 1, if one out of the following cases occurs:
		<ul> <li>Set to logic 1 if the SOF is incorrect. It is set to logic level 0 automatically at receiver start-up phase. The bit is only valid for 106 kbit in Active and Passive Communication mode.</li> </ul>
		<ul> <li>If bit DetectSync in CIU_Mode register is set to logic 1 during FeliCa communication or Active Communication with transfer speeds higher than 106 kbit, ProtocolErr is set to logic 1 in case of a byte length violation.</li> </ul>
		<ul> <li>During the AutoColl command, ProtocolErr is set to logic 1, if the Initiator bit in CIU_Control register is set to logic 1.</li> </ul>
		<ul> <li>During the MFAuthent Command, ProtocolErr is set to logic 1, if the number of bytes received in one data stream is incorrect.</li> </ul>
		<ul> <li>Set to logic 1, if the Miller Decoder detects 2 pauses below the minimum time according to the ISO/IEC 14443A definitions.</li> </ul>

<sup>[1]</sup> Command execution will clear all error flags except for bit TempErr. A setting by firmware is impossible.

## 8.6.23.9 CIU\_Status1 register (DFh or 6337h)

Contains status flags of the CRC, Interrupt Request System and FIFO buffer.

Table 191. CIU\_Status1 register (address DFh or 6337h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	CIU_IRQ_1	CRCOk	CRCReady	CIU_IRQ_0	TRunning	RFOn	HiAlert	LoAlert
Reset	0	0	1	0	0	Χ	0	1
Access	R	R	R	R	R	R	R	R

#### Table 192. Description of CIU\_Status1 bits

Bit	Symbol	Description
7	CIU_IRQ_1	This bit shows, if any CIU_IRQ_1 source requests attention (with respect to the setting of the interrupt enable flags, see CIU_CommIEn and CIU_DivIEn registers).
6	CRCOk	Set to logic 1, the CRC result is zero. For data transmission and reception the bit CRCOk is undefined (use CRCErr in CIU_Error register). CRCok indicates the status of the CRC coprocessor, during calculation the value changes to logic level 0, when the calculation is done correctly, the value changes to logic level 1.
5	CRCReady	Set to logic 1, when the CRC calculation has finished. This bit is only valid for the CRC co-processor calculation using the CalcCRC command.
4	CIU_IRQ_0	This bit shows, if any CIU_IRQ_0 source requests attention (with respect to the setting of the interrupt enable flags, see CIU_CommIEn and CIU_DivIEn registers).
3	TRunning	Set to logic 1, the CIU_timer is running, e.g. the CIU_timer will decrement the CIU_TCounterVal_lo with the next timer clock.
		<b>Note:</b> In the gated mode TRunning is set to logic 1, when the CIU_timer is enabled by the register bits. This bit is not influenced by the gated signal.
2	RFOn	Set to logic 1, if an external RF field is detected. This bit does not store the state of the RF field.
1	HiAlert	Set to logic 1, when the number of bytes stored in the FIFO buffer fulfils the following equation:
		$HiAlert = (64 - FIFOLength) \le WaterLevel$
		Example:
		$FIFOLenght = 60, WaterLevel = 4 \rightarrow HiAlert = 1$
		$FIFOLenght = 59$ , $WaterLevel = 4 \rightarrow HiAlert = 0$
0	LoAlert	Set to logic 1, when the number of bytes stored in the FIFO buffer fulfills the following equation:
		$LoAlert = FIFOLength \le WaterLevel$
		Example:
		FIFOLenght = 4, WaterLevel = $4 \rightarrow LoAlert = 1$
		$FIFOLenght = 5$ , $WaterLevel = 4 \rightarrow LoAlert = 0$

## 8.6.23.10 CIU\_Status2 register (E9h or 6338h)

Contain status flags of the receiver, transmitter and Data Mode Detector.

Table 193. CIU\_Status2 register (address E9h or 6338h) bit allocation

Bit	3it 7		5	4	3	2	1	0
Symbol	TempSensClear	-	RFFreqOK	TgActivated	MFCrypto1On	ModemState[2		e[2:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R	R	DY	DY	R	R	R

#### Table 194. Description of CIU\_Status2 bits

Bit	Symbol	Descri	ption					
7	TempSensClear		Set to logic 1, this bit clears the temperature error, if the temperature is selow the alarm limit of 125 $^{\circ}$ C.					
6	-	Reserv	Reserved					
5	RFFreqOK	Indicate 13.56 N	es if the frequency detected at the RX pin is in the range of MHz.					
			ogic 1, if the frequency at the RX pin is in the range 12 MHz < frequency < 15 MHz.					
			The value of RFFreqOK is not defined if the external RF ncy is in the range of 9 to 12 MHz or in the range of 15 to z.					
4	TgActivated		ogic 1 if the Select command is received correctly or if the command was answered.					
			This bit can only be set during the AutoColl command in Passive unication mode or Card operating modes.					
		Note: 7	This bit is set to logic level 0 automatically by switching off the d.					
3	MFCrypto1On		ogic level 1, MIFARE Crypto1 unit is switched on and therefore a communication with the card is encrypted.					
		This bit can only be set to logic 1 by a successful execution of the MFAuthent command. This is only valid in Reader/Writer mode for MIFARE 1 KB or MIFARE 4 KB emulation cards.						
		This bit	t shall be set to logic level 0 by firmware.					
2 to 0	ModemState[2:0]	Modem machin	nState shows the state of the transmitter and receiver state nes.					
		Value	Description					
		000	Idle					
		001	Wait for StartSend in CIU_BitFraming register					
		010	TxWait: Wait until RF field is present, if TxWaitRF is set to logic 1. The minimum time for TxWait is defined by the TxWait register.					
		011	Transmitting					
		100	RxWait: Wait until RF field is present, if the bit RxWaitRF is set to logic 1. The minimum time for RxWait is defined by the RxWait register					
		101	Wait for data					
		110	Receiving					

### 8.6.23.11 CIU\_FIFOData register (EAh or 6339h)

In- and output of 64 byte FIFO buffer.

Table 195. CIU\_FIFOData register (address EAh or 6339h) bit allocation

Bit	7	6	5	4	3	2	1	0			
Symbol		FIFOData[7:0]									
Reset	Х	X	Х	Х	X	X	Х	Х			
Access	DY	DY	DY	DY	DY	DY	DY	DY			

#### Table 196. Description of CIU\_FIFOData bits

Bit	Symbol	Description
7 to 0	FIFOData[7:0]	Data input and output port for the internal 64 bytes FIFO buffer. The FIFO buffer acts as parallel in/parallel out converter for all data stream in- and outputs

## 8.6.23.12 CIU\_FIFOLevel register (EBh or 633Ah)

Indicates the number of bytes stored in the FIFO.

Table 197. CIU\_FIFOLevel register (address EBh or 633Ah) bit allocation

Bit	7	6	5	4	3	2	1	0			
Symbol	FlushBuffer		FIFOLevel[6:0]								
Reset	0	0	0	0	0	0	0	0			
Access	W	R	R	R	R	R	R	R			

Table 198. Description of CIU\_FIFOLevel bits

Bit	Symbol	Description
7	FlushBuffer	Set to logic 1, this bit clears the internal FIFO-buffer's read- and write-pointer and the bit BufferOvfl in the CIU_Error register immediately. Reading this bit will always return logic level 0.
6 to 0	FIFOLevel[6:0]	Indicates the number of bytes stored in the FIFO buffer. Writing to the CIU_FIFOData Register increments, reading decrements FIFOLevel.

### 8.6.23.13 CIU\_WaterLevel register (ECh or 633Bh)

Defines the thresholds for FIFO under- and overflow warning.

Table 199. CIU\_WaterLevel register (address ECh or 633Bh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	-	-	WaterLevel[6:0]						
Reset	0	0	0	0	1	0	0	0	
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W	

#### Table 200. Description of CIU\_WaterLevel bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	WaterLevel[5:0]	This register defines a threshold to indicate a FIFO buffer over- or underflow to the 80C51:
		The HiAlert bit in CIU_Status1 register is set to logic 1, if the remaining number of bytes in the FIFO-buffer space is equal or less than the defined WaterLevel[5:0] bits.
		The LoAlert bit in CIU_Status1 register is set to logic 1, if equal or less than WaterLevel[5:0] bits are in the FIFO.
		<b>Remark:</b> For the calculation of the HiAlert and LoAlert see <u>Table 192</u> on page 161.

## 8.6.23.14 CIU\_Control register (EDh or 633Ch)

Contains miscellaneous control bits.

Table 201. CIU\_Control register (address EDh or 633Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TStopNow	TStartNow	WrNFCIP-1IDtoFIFO	Initiator	-	RxLastBits[2:0]		[2:0]
Reset	0	0	0	0	0	0	0	0
Access	W	W	DY	R/W	R	R	R	R

Table 202. Description of CIU\_Control bits

Bit	Symbol	Description
7	TStopNow	Set to logic 1, the timer stops immediately.
		Reading this bit will always return logic level 0.
6	TStartNow	Set to logic 1, the timer starts immediately.
		Reading this bit will always return logic level 0.
5	WrNFCIP-1IDtoFIFO	Set to logic 1, the internal stored NFCID3 (10 bytes) is copied into the FIFO.
		Afterwards the bit is set to logic level 0 automatically.
4	Initiator	Set to logic 1, the PN533 acts as Initiator or Reader/Writer, otherwise it acts as Target. or a Card.
3	-	Reserved.
2 to 0	RxLastBits[2:0]	Shows the number of valid bits in the last received byte. If set to 000b, the whole byte is valid.

# 8.6.23.15 CIU\_BitFraming register (EEh or 633Dh)

Adjustments for bit oriented frames.

Table 203. CIU\_BitFraming register (address EEh or 633Dh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	StartSend	F	RxAlign[2:0	)]	-	Tx	LastBits[2	:0]
Reset	0	0	0	0	0	0	0	0
Access	W	R/W	R/W	R/W	R	R/W	R/W	R/W

#### Table 204. Description of CIU\_BitFraming bits

Bit	Symbol	Description
7	StartSend	Set to logic 1, the transmission of data starts.
		This bit is only valid in combination with the Transceive command.
6 to 4	RxAlign[2:0]	Used for reception of bit oriented frames:
		RxAlign[2:0] defines the bit position for the first received bit to be stored in the FIFO. Further received bits are stored in the following bit positions
		Example:
		RxAlign[2:0] = 0: The LSB of the received bit is stored at bit 0, the second received bit is stored at bit position 1.
		RxAlign[2:0] = 1: The LSB of the received bit is stored at bit 1, the second received bit is stored at bit position 2
		RxAlign[2:0] = 7: The LSB of the received bit is stored at bit 7, the second received bit is stored in the following byte at bit position 0.
		These bits shall only be used for bitwise anticollision at 106 kbit/s in Passive Communication or Reader/Writer mode. In all other modes it shall be set to logic level 0.
3	-	Reserved.
2 to 0	TxLastBits[2:0]	Used for transmission of bit oriented frames: TxLastBits defines the number of bits of the last byte that shall be transmitted. A 000b indicates that all bits of the last byte shall be transmitted.

# 8.6.23.16 CIU\_Coll register (EFh or 633Eh)

Defines the first bit collision detected on the RF interface.

Table 205. CIU\_Coll register (address EFh or 633Eh) bit allocation

Bit	it 7		5	4 3		2	1	0
Symbol	ValuesAfterColl	-	CollPosNotValid	CollPos				
Reset	1	0	1	Χ	Χ	X	Χ	X
Access	R/W	R	R	R	R	R	R	R

#### Table 206. Description of CIU\_Coll bits

Bit	Symbol	Description
7	ValuesAfterColl	If this bit is set to logic 0, all receiving bits will be cleared after a collision.
		This bit shall only be used during bitwise anticollision at 106 kbit/s, otherwise it shall be set to logic 1.
6	-	Reserved
5	CollPosNotValid	Set to logic 1, if no Collision is detected or the Position of the collision is out of range of the CollPos[4:0] bits.
		This bit shall only be interpreted in Passive Communication mode at 106 kbit/s or ISO/IEC 14443A/MIFARE Reader/Writer mode.
4 to 0	CollPos	These bits show the bit position of the first detected collision in a received frame, only data bits are interpreted.
		Example:
		00h indicates a bit collision in the 32 <sup>nd</sup> bit.
		01h indicates a bit collision in the 1st bit
		08h indicates a bit collision in the 8 <sup>th</sup> bit
		This bit shall only be interpreted in Passive Communication mode at 106 kbit/s or ISO/IEC 14443A/MIFARE Reader/Writer mode if CollPosNotValid is set to logic 0.

## 8.6.23.17 CIU\_Mode register (6301h)

Defines general modes for transmitting and receiving.

Table 207. CIU\_Mode register (address 6301h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	MSBFirst	DetectSync	TXWaitRF	RxWaitRF	PolSigin	ModeDet Off	CRCF [1	Preset :0]
Reset	0	0	1	1	1	0	1	1
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

### Table 208. Description of CIU\_Mode bits

Bit	Symbol	Description						
7	MSBFirst	The bit order	the CRC co-processor calculates the CRC with MSB first. in the registers CRCResultMSB and the sultLSB is reversed.					
		Note: During	lote: During RF communication this bit is ignored.					
6	DetectSync		1, the CIU waits for the F0h byte before the receiver is F0h byte is added as a Sync-byte for transmission.					
		This bit is only protocol.	/ valid for 106 kbit/s during NFCIP-1 data exchange					
		In all other mo	odes it shall be set to logic 0.					
5	TXWaitRF	NFCIP-1 can	the transmitter in Reader/Writer or Initiator mode for only be started, if an own RF field is generated (i.e. l/or Tx2RFen is set to logic level 1).					
4	RxWaitRF		the counter for RxWait starts only, if an external RF field is arget mode for NFCIP-1 or in Card Operating mode					
3	PolSigin	PolSigin defines the polarity of the SIGIN pin.						
		Set to logic 1, the polarity of SIGIN pin is active high.						
		Set to logic 0	the polarity of SIGIN pin is active low.					
		Note: The inte	ernal envelope signal is coded active low.					
		Note: Changi	ng this bit will generate a SiginActIrq event.					
2	ModeDetOff	Set to logic 1,	the internal Data Mode Detector is switched off.					
		Note: The Da command.	ta Mode Detector is only active during the AutoColl					
1 to 0	CRCPreset[1:0]	Defines the posterior command.	reset value for the CRC co-processor for the CalCRC					
			any communication, the preset values is selected according to the mode definition in the CIU_RxMode and registers.					
		Value	Description					
		00	00 00					
		01	63 63					
		10	A6 71					
		11	FF FF					

## 8.6.23.18 CIU\_TxMode register (6302h)

Defines the transmission data rate and framing during transmission.

Table 209. CIU\_TxMode register (address 6302h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TxCRCEn	Tx	TxSpeed[2:0]		InvMod	TxMix	TxFraming[1:0]	
Reset	0	0	0	0	0	0	0	0
Access	R/W	DY	DY	DY	R/W	R/W	DY	DY

Table 210. Description of CIU\_TxMode bits

Bit	Symbol	Description	
7	TxCRCEn	Set to logic 1, transmission.	this bit enables the CRC generation during data
		Note: This bit	shall only set to logic 0 at 106 kbit/s.
6 to 4	TxSpeed[2:0]	Defines bit ra	te while data transmission.
		Value	Description
		000	106 kbit/s
		001	212 kbit/s
		010	424 kbit/s
		011	848 kbit/s
		100	1696 kbit/s
		101	Reserved
		Reserved	
		to the bit codi	coding for transfer speeds above 424 kbit/s is equivalent ng of the Active Communication mode of the 424 kbit/s of 8092 / ECMA340.
3	InvMod	Set to logic 1	the modulation for transmitting data is inverted.
2	TxMix	•	the signal at SIGIN is mixed with the internal coder. See 2 "Serial data switch" on page 131.
1 to 0	TxFraming[1:0]	Defines the fr	aming used for data transmission.
		Value	Description
		00	ISO/IEC 14443A/MIFARE and Passive Communication mode 106 kbit/s
		01	Active Communication mode
		10	FeliCa and Passive Communication mode at 212 kbit/s and 424 kbit/s

## 8.6.23.19 CIU\_RxMode register (6303h)

Defines the reception data rate and framing during receiving.

Table 211. CIU\_RxMode register (address 6303h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	RXCRCEn	Rx	RxSpeed[2:0]		RxNoErr	RxMultiple	RxFraming[1:0]	
Reset	0	0	0	0	0	0	0	0
Access	R/W	DY	DY	DY	R/W	R/W	DY	DY

Table 212. Description of CIU\_RxMode bits

Table 2	212. Description	of CIU_RxMo	ode bits							
Bit	Symbol	Description								
7	RxCRCEn		this bit enables the CRC calculation during reception. The II not be written within the CIU FIFO.							
		Note: This bit	lote: This bit shall only set to logic level 0 at 106 kbit/s.							
6 to 4	RxSpeed[2:0]	Defines the bi	t rate while data receiving.							
		Ų i	art of the CIU handles only transfer speeds up to 424 kbit/s digital part of the CIU handles the higher transfer speeds as							
		Value	Description							
		000	106 kbit/s							
		001	212 kbit/s							
		010	424 kbit/s							
		011	848 kbit/s							
		100	1696 kbit/s							
		101	Reserved							
		110 - 111	Reserved							
		the bit coding	coding for transfer speeds above 424 kbit/s is equivalent to of the active communication mode of the 424 kbit/s of the of 8092 / ECMA340.							
3	RxNoErr	_	, a not valid received data stream (less than 4 bits received) d. The receiver will remain active.							
2	RxMultiple	Set to logic 0,	the receiver is deactivated after receiving a data frame.							
		is only valid fo set this bit, the automatically.	it is possible to receive more than one data frame. This bit r 212 and 424 kbit/s to handle the Polling command. Having e receive and transceive commands will not end In this case the multiple receiving can only be deactivated Idle command to the CIU_Command register or clearing this 51.							
			I, at the end of a received data stream an error byte is FIFO. The error byte is a copy of the CIU_Error register.							
1 to 0	RxFraming[1:0]	Defines the ex	spected framing for data reception.							
		Value	Description							
		00	ISO/IEC 14443A/MIFARE and Passive Communication mode 106 kbit/s							
		01	Active communication mode							
		10	FeliCa and Passive Communication mode at 212 kbit/s and 424 kbit/s							
		11	ISO/IEC 14443B							

## 8.6.23.20 CIU\_TxControl register (6304h)

Controls the logical behavior of the antenna driver pins TX1 and TX2. See also <u>Table 155</u> on page 124 and <u>Table 156</u> on page 124.

Table 213. CIU\_TxControl register (address 6304h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	InvTx2 RFon	InvTx1 RFon	InvTx2 RFoff	InvTx1 RFoff			Tx2 RFEn	Tx1 RFEn
Reset	1	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	W	R/W	R/W

### Table 214. Description of CIU\_TxControl bits

		-
Bit	Symbol	Description
7	InvTx2RFon	Set to logic1 and Tx2RFEn set to logic level 1, TX2 output signal is inverted.
6	InvTx1RFon	Set to logic1 and Tx1RFEn set to logic level 1, TX1 output signal is inverted.
5	InvTx2RFoff	Set to logic1 and Tx2RFEn set to logic level 0, TX2 output signal is inverted.
4	InvTx1RFoff	Set to logic1 and Tx1RFEn set to logic level 0, TX1 output signal is inverted.
3	Tx2CW	Set to logic 1, the output signal on pin TX2 will deliver continuously the un-modulated 13.56 MHz energy carrier.  Set to logic 0, Tx2CW is enabled to modulate of the 13.56 MHz energy carrier.
		det to logic of 1720V is enabled to modulate of the 13.30 Wi12 energy carrier.
2	CheckRF	Set to logic 1, Tx2RFEn and Tx1RFEn can not be set if an external RF field is detected.
		Only valid when using in combination with Tx2RFAutoEn and TX1RFAutoEn bits in CIU_TxAuto register.
1	Tx2RFEn	Set to logic 1, the output signal on pin TX2 will deliver the 13.56 MHz energy carrier modulated by the transmission data.
0	Tx1RFEn	Set to logic 1, the output signal on pin TX1 will deliver the 13.56 MHz energy carrier modulated by the transmission data.

## 8.6.23.21 **CIU\_TxAuto** register (6305h)

Controls the setting of the antenna driver.

Table 215. CIU\_TxAuto register (address 6305h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Auto RFOFF	Force 100ASK	AutoWakeUp	-	CAOn	InitialRFOn	Tx2 RFAutoEn	Tx1 RFAutoEn
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R	R/W	W	R/W	R/W

#### Table 216. Description of CIU TxAuto bits

	•	
Bit	Symbol	Description
7	AutoRFOff	Set to logic 1, own RF field is switched off after the last data bit has been transmitted as defined in the NFCIP-1 standard.
6	Force100ASK	Set to logic 1, Force100ASK forces a 100% ASK modulation independent of the setting in CIU_ModGsP register.
5	AutoWakeUp	Set to logic 1, the PN533 in CIU Power-down mode can be woken up by the RF level detector.
4		Reserved
3	CAOn	Set to logic 1, the collision avoidance is activated and internally the value n is set in accordance to the ISO/IEC 18092 / ECMA340 NFCIP-1 standards.
2	InitialRFOn	Set to logic 1, the initial RF collision avoidance is performed and the bit InitialRFOn is set to logic level 0 automatically, if the RF is switched ON.
		<b>Note:</b> The driver(s) which should be switched on, have to enabled by Tx2RFAutoEn and/or Tx1RFAutoEn bits.
		<b>Note:</b> If the own RF field is already ON when the bit InitialRFOn is set, it is not set to logic level 0.
1	Tx2RFAutoEn	Set to logic 1, RF is switched on at TX2 (i.e. $Tx2RFEn$ is set to logic level 1) after the external RF field is switched off according to the time TADT. If the InitialRFOn and $Tx2RFAutoEn$ bits are set to logic 1, RF is switched on at TX2 if no external RF field is detected during the time $T_{IDT}$ .
		<b>Note:</b> The times $T_{ADT}$ and $T_{IDT}$ are in accordance to the ISO/IEC 18092/ ECMA340 NFCIP-1 standards.
0	Tx1RFAutoEn	Set to logic 1, RF is switched on at TX1(i.e. Tx1RFEn is set to logic level 1) after the external RF field is switched off according to the time TADT. If the InitialRFOn and Tx1RFAutoEn bits are set to logic 1, RF is switched on at TX1 if no external RF field is detected during the time T <sub>IDT</sub> .
		<b>Note:</b> The times $T_{ADT}$ and $T_{IDT}$ are in accordance to the ISO/IEC 18092/ ECMA340 NFCIP-1 standards.

## 8.6.23.22 CIU\_TxSel register (6306h)

Selects the sources for the analogue transmitter part

Table 217. CIU\_TxSel register (address 6306h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	LoadModSel[1:0]		DriverSel[1:0]		SigOutSel[3:0]			
Reset	0	0	0	1	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 218. Description of CIU\_TxSel bits

Bit	Symbol	Description	
7 to 6	LoadModSel[1:0]	Selects the	signal to be output on LOADMOD
		Value	Description
		00	Tristate
		01	Modulation signal (envelope) from the internal coder
		10	Modulation signal (envelope) from SIGIN
		11	Test signal defined by LoadModtest in register CIU_TestSel1
5 to 4	DriverSel[1:0]	Selects the s	signals to be output on Tx1 and Tx2.
		Value	Description
		00	Tristate
		01	Modulation signal (envelope) from the internal coder
		10	Modulation signal (envelope) from SIGIN
		11	HIGH
			IIGH level depends on the setting of N/InvTx1RFOff and InvTx2RFON/InvTx2RFOff.
3 to 0	SigOutSel[3:0]	Select the si	gnal to be output on SIGOUT
		0000	Tristate
		0001	Low
		0010	High
		0011	Test bus signal as defined by TestBusBitSel in CIU_TestSel1.
		0100	Modulation signal (envelope) from the internal coder
		0101	Serial data stream to be transmitted
		0110	Output signal of the receiver circuit (card modulation signal regenerated and delayed). This signal is used as data output signal for secure IC interface connection using 3 lines.
			<b>Note:</b> To have a valid signal the CIU has to be set to the receiving mode by either the Transceive or Receive command. The RxMultiple bit can be used to keep the CIU in receiving mode.
			<b>Note:</b> Do not use this setting in ISO/IEC 14443A/MIFARE mode. Data collisions will not be transmitted on SIGOUT when using Manchester coding.
		0111	Serial data stream received.
			<b>Note:</b> Do not use this setting in ISO/IEC 14443A/MIFARE mode. Miller coding parameters as the bitlength can vary

Table 218. Description of CIU\_TxSel bits ...continued

Bit	Symbol	Description	1
		1000-1011	FeliCa secure IC modulation
			1000 RX*
			1001 TX
			1010 Demodulator comparator output
			1011 Reserved
			<b>Note:</b> * To have a valid signal the CIU has to be set to the receiving mode by either the Transceive or Receive commands. The bit RxMultiple can be used to keep the CIU in receiving mode
		1000-1011	MIFARE secure IC modulation
			1100 RX* with RF carrier
			1101 TX with RF carrier
			1110 RX with RF carrier unfiltered
			1111 RX envelope unfiltered
			<b>Note:</b> * To have a valid signal the CIU has to be set to the receiving mode by either the Transceive or Receive commands. The bit RxMultiple can be used to keep the CIU in receiving mode

### 8.6.23.23 CIU\_RxSel register (6307h)

Selects internal receiver settings.

Table 219. CIU\_RxSel register (address 6307h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	UartSel[1:0]		RxWait[5:0]						
Reset	1	0	0	0	0	1	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Table 220. Description of CIU\_RxSel bits

Bit	Symbol	Description	
7 to 6	UartSel[1:0]	Selects the input of	of the digital part (CL UART) of the CIU
		Value	Description
		00	Constant Low
		01	Envelope signal at SIGIN
		10	Modulation signal from the internal analog part
		11	Modulation signal from SIGIN pin. Only valid for transfer speeds above 424 kbit/s
5 to 0	RxWait[5:0]		ssion, the activation of the receiver is delayed for RxWait this 'frame guard time' any signal at pin Rx is ignored.
		(e.g. Transceive, at the mode of the Communication metransmitted data s	ignored by the Receive command. All other commands Autocoll, MFAuthent) use this parameter. Depending on IU, the counter starts differently. In Passive node the counters starts with the last modulation of the tream. In Active Communication mode the counter starts the external RF field is switched on.

### 8.6.23.24 CIU\_RxThreshold register (6308h)

Selects thresholds for the bit decoder.

Table 221. CIU\_RxThreshold register (address 6308h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		MinLe	/el[3:0]		-	Collevel[2:0]		
Reset	1	0	0	0	0	1	0	0
Access	R/W	R/W	R/W	R/W	R	R/W	R/W	R/W

#### Table 222. Description of CIU\_RxThreshold bits

Bit	Symbol	Description
7 to 4	MinLevel[3:0]	Defines the minimum signal strength at the decoder input that shall be accepted. If the signal strength is below this level, it is not evaluated.
3	-	Reserved
2 to 0	Collevel[2:0]	Defines the minimum signal strength at the decoder input that has to be reached by the weaker half-bit of the Manchester-coded signal to generate a bit-collision relatively to the amplitude of the stronger half-bit.

### 8.6.23.25 CIU\_Demod register (6309h)

Defines demodulator settings.

Table 223. CIU\_Demod register (address 6309h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	AddIO	Q[1:0]	FixIQ	-	TauR	cv[1:0]	TauSy	nc[1:0]
Reset	0	1	0	0	1	1	0	1
Access	R/W	R/W	R/W	R	R/W	R/W	R/W	R/W

Table 224. Description of CIU\_Demod bits

Bit	Symbol	Description						
7 to 6	AddIQ[1:0]	Defines the u	se of I and Q channel during reception.					
		Note: FixIQ h	has to be set to logic 0 to enable the following settings.					
		Value	Description					
		00	Select the stronger channel					
		01	Select the stronger and freeze the selected during communication					
		10	Combines the I and Q channel					
		11	RFU					
5	FixIQ	If set to logic I channel.	1 and AddIQ[0] is set to logic level 0, the reception is fixed to					
		If set to logic Q channel.	1 and AddIQ[0] is set to logic level 1, the reception is fixed to					
4	-	Reserved						
3 to 2	TauRcv[1:0]	Changes time	e-constant of internal PLL during data receiving.					
		Note: If set to	000h, the PLL is frozen during data receiving.					
1 to 0	TauSync[1:0]	Changes time	e-constant of internal PLL during burst (out of data reception)					

# 8.6.23.26 CIU\_FeINFC1 register (630Ah)

Defines the length of the FeliCa Sync bytes and the minimum length of the received frame.

Table 225. CIU\_FeINFC1 register (address 630Ah) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	FelSync	Len[1:0]	DataLenMin[5:0]						
Reset	0	0	0	0	0	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

#### Table 226. Description of CIU\_FeINFC1 bits

Bit	Symbol	Description				
7 to 6	FelSyncLen[1:0]]	Defines the length of the Sync bytes.				
		Value	Description			
		00	B2 4D			
		01	00 B2 4D			
		10	00 00 B2 4D			
		11	00 00 00 B2 4D			
5 to 0	DataLenMin[5:0]	These bits def	fine the minimum length of the accepted frame length.			
		Data	LenMin×4≤DataPacketLenght			
		CIU_Mode reg	er is ignored at 106 kbit/s if the DetectSync bit in gister is set to logic 0. If a received frame is shorter as the enMin value, the frame will be ignored.			

# 8.6.23.27 CIU\_FeINFC2 register (630Bh)

Defines the maximum length of the received frame.

Table 227. CIU\_FeINFC2 register (address 630Bh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	WaitForSelected	ShortTimeSlot			DataLen	Max[5:0]		
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 228. Description of CIU\_FeINFC2 bits

Bit	Symbol	Description				
7	WaitForSelected	Set to logic 1, the AutoColl command is automatically ended only when:				
		<ol> <li>A valid command has been received after performing a valid Select procedure according to ISO/IEC 14443A.</li> </ol>				
		<ol><li>A valid command has been received after performing a valid Polling procedure according to the FeliCa specification.</li></ol>				
		Note: If this bit is set, no Active Communication is possible.				
		<b>Note:</b> Setting this bit reduces the 80C51 interaction in case of a communication to another device in the same RF field during Passive Communication mode.				
6	ShortTimeSlot	Defines the time slot length for Active Communication mode at 424 kbit/s.				
		Set to logic 1 a short time slot is used (half of the timeslot at 212 kbit/s).				
		Set to logic 0 a long timeslot is used (equal to the timeslot for 212 kbit/s).				
5 to 0	DataLenMax[5:0]	These bits define the maximum length of the accepted frame length:				
		$DataLenMax \times 4 \ge DataPacketLenght$				
		<b>Note:</b> If set to logic 0 the maximum data length is 256 bytes.				
		This parameter is ignored at 106 kbit/s if the bit DetectSync in register CIU_Mode is set to logic 0.				
		If a received frame is larger as the defined DataLenMax value, the frame will be ignored.				

## 8.6.23.28 CIU\_MifNFC register (630Ch)

Defines ISO/IEC 14443A/MIFARE/NFC specific settings in target or card operating mode.

Table 229. CIU\_MifNFC register (address 630Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	SensMiller[2:0]		TauMiller[1:0]		MFHalted	TxWa	it[1:0]	
Reset	0	1	1	0	0	0	1	0
Access	R/W	R/W	R/W	R/W	R/W	DY	R/W	R/W

#### Table 230. Description of CIU\_MifNFC bits

Bit	Symbol	Description
7 to 5	SensMiller[2:0]	This bit defines the sensitivity of the Miller decoder.
4 to 3	TauMiller[1:0]	This bit defines the time constant of the Miller decoder.
2	MFHalted	Set to logic 1, this bit indicates that the CIU is set to HALT mode in Card Operating mode at 106 kbit/s. This bit is either set by the 80C51 or by the internal state machine and indicates that only the code 52h is accepted as a Request command.
		This bit is automatically set to logic level 0 by RF reset.
1 to 0	TxWait[1:0]	In combination with TxBitPhase[6:0] in CIU_TxBitPhase register, defines the additional response time for the target at 106 kbit/s in Passive Communication mode and during the AutoColl command. See CIU_TxBitPhase register.

## 8.6.23.29 CIU\_ManualRCV register (630Dh)

Allows manual fine tuning of the internal receiver.

**IMPORTANT NOTE:** For standard application it is not recommended to change this register settings.

Table 231. CIU\_ManualRCV register (address 630Dh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	FastFilt MF_SO	Delay MF_SO	ParityDisable	LargeBWPLL	ManualHPCF	HPCI	F[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 232. Description of CIU\_ManualRCV bits

Bit	Symbol	Description
7	-	Reserved
6	FastFiltMF_SO	If this bit is set to logic 1, the internal filter for the Miller-Delay circuit is set to Fast-Mode
		<b>Note:</b> This bit should only be set to logic 1, if the Miller pauses length expected are less than 400 ns. At 106 kbit/s, the Miller pauses duration is around 3 $\mu$ s.
5	DelayMF_SO	If this bit is set to logic 1, when SigoutSel=1100b (register 6306h), the Signal at SIGOUT-pin is delayed according the delay defined by TxBitPhase[6:0] (register 6315h) and TxWait bits (register 630Ch).
		<b>Note:</b> In ISO/IEC 14443A/MIFARE Card MIFARE 1 KB or MIFARE 4 KB emulation (Virtual Card) mode (DriverSel = 10b and SigoutSel=1110b), the Signal at SIGIN must then be 128 /fc faster compared to the ISO/IEC 14443A restrictions on the RF-Field for the Frame Delay Time.
		<b>Note:</b> This delay shall only be activated for setting bits SigOutSel to (1110b) or (1111b) in register CIU_TxSel.
		If this bit is set to logic 0, the SIGOUT-pin delay is not adjustable.
		<b>Note:</b> In ISO/IEC 14443A/MIFARE card 1 KB or MIFARE 4 KB emulation (Virtual Card) mode (DriverSel = 10b and SigoutSel=1110b), the ISO/IEC 14443A restrictions on the RF-Field for the Frame Delay Time should be adjusted on the secure IC side
4	ParityDisable	If this bit is set to logic 1, the generation of the Parity bit for transmission and the parity check for receiving is switched off. The received parity bit is handled like a data bit.

Table 232. Description of CIU\_ManualRCV bits ...continued

Bit	Symbol	Description	1				
3	LargeBWPLL	Set to logic 1, the bandwidth of the internal PLL for clock recovery is extended.					
			e bandwidth is extended, the PLL filtering effect is weaker formance of the communication may be affected.				
2	ManualHPCF	Set to logic 0, the HPCF[1:0] bits are ignored and the HPCF[1:0] settings are adapted automatically to the receiving mode.					
1 to 0	HPCF[1:0]	Selects the internal rece	High Pass Corner Frequency (HPCF) of the filter in the eiver chain				
		Value	Description				
		00	For signals with frequency spectrum down to 106 kHz				
		01	For signals with frequency spectrum down to 212 kHz				
		10	For signals with frequency spectrum down to 424 kHz				
		11	For signals with frequency spectrum down to 848 kHz				

## 8.6.23.30 **CIU\_TypeB** register (630Eh)

Selects the specific settings for the ISO/IEC 14443B

Table 233. CIU\_TypeB register (address 630Eh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	Rx SOFReq	Rx EOFReq	-	EOFSOF Width	NoTx SOF	NoTx EOF	TxEG	T[1:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R	R/W	R/W	R/W	R/W	R/W

### Table 234. Description of CIU\_TypeB bits

Bit	Symbol	Description						
7	RxSOFReq		If this bit is set to logic 1, the SOF is required. A datastream starting without SOF is ignored.					
			to logic level 0, a datastream with and without SOF is SOF will be removed and not written into the FIFO.					
6	RxEOFReq	without EOF v	to logic 1, the EOF is required. A datastream ending vill generate a protocol error: ProtocollErr in the CIU_Error eset to logic level 1.					
			to logic level 0, a datastream with and without EOF is EOF will be removed and not written into the FIFO.					
5	-	Reserved.						
4	EOFSOFWidth	If this bit is set to logic 1, the SOF and EOF will have the maximum length defined in the ISO/IEC 14443B.						
			to logic level 0, the SOF and EOF will have the minimum in the ISO/IEC 14443B.					
3	NoTxSOF	If this bit is set	to logic 1, the generation of the SOF is suppressed.					
2	NoTxEOF	If this bit is set	to logic 1, the generation of the EOF is suppressed.					
1 to 0	TxEGT[1:0]	These bits def	ine the length of the EGT, as defined in the 3B					
		Value	Description					
		00	0 bit					
		01	1 bit					
		10	2 bits					
		11	3 bits					

#### 8.6.23.31 CIU\_CRCResultMSB register (6311h)

Shows the actual MSB values of the CRC calculation.

Note: The CRC is split into two 8-bit registers. See also the CIU\_CRCResultLSB register.

Note: Setting the bit MSBFirst in CIU\_Mode register reverses the bit order, the byte order is not changed

Table 235. CIU\_CRCResultMSB register (address 6311h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				CRCResu	ltMSB[7:0]			
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R	R

Table 236. Description of CIU\_CRCResultMSB bits

Bit	Symbol	Description
7 to 0	CRCResultMSB[7:0]	This register shows the actual value of the most significant byte of the CRC calculation. It is valid only if CRCReady bit in CIU_Status1 register is set to logic 1.

#### 8.6.23.32 CIU\_CRCResultLSB register (6312h)

Shows the actual LSB values of the CRC calculation.

Note: The CRC is split into two 8-bit registers. See also the CIU\_CRCResultMSB register.

Note: Setting the bit MSBFirst in CIU\_Mode register reverses the bit order, the byte order is not changed

Table 237. CIU\_CRCResultLSB register (address 6312h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				CRCResu	ltLSB[7:0]			
Reset	1	1	1	1	1	1	1	1
Access	R	R	R	R	R	R	R	R

Table 238. Description of CIU\_CRCResultLSB bits

Bit	Symbol	Description
7 to 0	CRCResultLSB[7:0]	This register shows the actual value of the most significant byte of the CRC register. It is valid only if CRCReady bit in CIU_Status1 register is set to logic 1.

#### 8.6.23.33 CIU\_GsNOff register (6313h)

Selects the conductance for the N-driver of the antenna driver pins TX1 and TX2 when there is no RF generated by the PN533.

Table 239. CIU\_GsNOff register (address 6313h) bit allocation

Bit	7	6	5	4	3	2	1	0		
Symbol		CWGsNOff[3:0]				ModGsNOff[3:0]				
Reset	1	0	0	0	1	0	0	0		
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		

Table 240. Description of CIU\_GsNOff bits

Bit	Symbol	Description
7 to 4	CWGsNOff[3:0]	The value of this register defines the conductance of the output N-driver during the time of no modulation and when there is no RF generated by the PN533 (neither Tx1RFEn nor Tx2RFEn is set to logic level 1).
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, CWGsNOff[3] is set to logic level 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if no RF is generated by the driver, otherwise the value CWGsNOn in the CIU_GsNOn register is used.
3 to 0	ModGsNOff[3:0]	The value of this register defines the conductance of the output N-driver for the time of modulation and when there is no RF generated by the PN533 (neither Tx1RFEn nor Tx2RFEn is set to logic level 1).
		This may be used to regulate the modulation index when doing load modulation.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down, if DriverSel[1:0] is not equal to 01b, ModGsNOff[3] is set to logic level 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if no RF is generated by the driver, otherwise the value ModGsNOn in the CIU_GsNOn register is used.

#### 8.6.23.34 CIU\_ModWidth register (6314h)

Controls the setting of the modulation width.

Table 241. CIU\_ModWidth register (address 6314h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol				ModWid	dth[7:0]			
Reset	0	0	1	0	0	1	1	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 242. Description of CIU\_ModWidth bits

Bit	Symbol	Description
7 to 0	ModWidth[7:0]	These bits define the width of the Miller modulation as initiator in Active and Passive Communication mode as multiples of the carrier frequency (ModWidth+1 / fc). The maximum value is half the bit period.
		Acting as a target in Passive Communication mode at 106 kbit/s or in Card Operating mode for ISO/IEC 14443A/MIFARE these bits are used to change the duty cycle of the subcarrier frequency.
		Number of cycles with low value: NCLV = (Modwidth modulo 8)+1
		Number of cycles with high value: NCHV = 16 - NCLV

#### 8.6.23.35 CIU\_TxBitPhase register (6315h)

Adjust the bit phase at 106 kbit/s during transmission.

Table 243. CIU\_TxBitPhase register (address 6315h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	RcvClkChange			Txl	BitPhase[6	6:0]		
Reset	1	0	0	0	0	1	1	1
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 244. Description of CIU\_TxBitPhase bits

Bit	Symbol	Description
7	RcvClkChange	Set to logic 1, the demodulator's clock is derivated from the external RF field.
6 to 0 TxBitPhase[6:0]		TXBitPhase[6:0] in addition with TxWait bits (register 63 0Ch), define a delay to adjust the bit synchronization during Passive Communication mode at 106 kbit/s and in ISO/IEC 14443A/MIFARE Reader/Writer mode. TxBitphase[6:0] are representing a delay in number of carrier frequency clock cycles.
		<b>Note:</b> The ranges to be used for TxWait[1:0] and TxBitPhase[6:0] are between:
		TXWait=01b and TxBitPhase = 1Bh (equivalent to an added delay of 20 clock cycles) and TXWait=01b and TxBitPhase = 7Fh (equivalent to an added delay of 120 clock cycles)
		TxWait=10b and TxBitPhase = 00h (equivalent to an added delay of 121 clock cycles) and TxWait=10b and TxBitPhase = 0Fh (equivalent to an added delay of 136 clock cycles)
		Note: The delay can vary depending of antenna circuits.
		<b>Note:</b> When DriverSel = 01b (the transmitter modulation input is coming from the internal coder), this delay is added to the waiting period before transmitting data in all communication modes.
		<b>Note:</b> When SigoutSel=1110b (CIU_TxSel register), and DelayMF_SO =1b (CIU_ManualRCV register), this delay is added on SIGOUT.
		<b>Note:</b> If the Signal at SIGIN is 128/fc faster compared to the ISO/IEC 14443A restrictions on the RF-Field for the Frame Delay Time, this delay is made so that if the FDT is correct when DriverSel = 01b, the same values of TxWait[1:0] and TxBitPhase[6:0] are also correct for this configuration when DriverSel = 10b (the transmitter modulation input is coming from SIGIN).

#### CIU\_RFCfg register (6316h) 8.6.23.36

Configures the receiver gain and RF level detector sensitivity.

Table 245. CIU\_RFCfg register (address 6316h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	RFLevelAmp	F	RxGain[2:0]			RFLevel[3:0]			
Reset	0	1	0	0	1	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Bit	Symbol	Descrip tion	
7	RFLevelAmp	J	c 1, this bit activates the RF level detector's amplifier, see 6.8 "RF level detector" on page 126.
6 to 4	RxGain[2:0]	This regist	ter defines the receivers signal voltage gain factor:
		Value	Description
		000	18 dB
		001	23 dB
		010	18 dB
		011	23 dB
		100	33 dB
		101	38 dB
		110	43 dB
		111	48 dB
3 to 0	RFLevel[3:0]		e sensitivity of the RF level detector, for description see Section level detector on page 126.

#### 8.6.23.37 **CIU\_GsNOn register (6317h)**

Selects the conductance for the N-driver of the antenna driver pins TX1 and TX2 when generating RF.

Table 247. CIU\_GsNOn register (address 6317h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	CWGsNOn[3:0]				ModGsNOn[3:0]				
Reset	1	0	0	0	1	0	0	0	
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

#### Table 248. Description of CIU\_GsNOn bits

Bit	Symbol	Description
7 to 4	CWGsNOn[3:0]	The value of this register defines the conductance of the output N-driver. during times of no modulation and when the PN533 generates the RF field.
		This may be used to regulate the output power and subsequently current consumption and operating distance.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, CWGsNOn[3] is set to logic level 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if RF is generated by the driver (either Tx1RFEn or Tx2RFEn is set to logic level 1), otherwise the value CWGsNOFF in the register CIU_GsNOff is used.
3 to 0	ModGsNOn[3:0]	The value of this register defines the conductance of the output N-driver for the time of modulation and when the PN533 generates the RF field.
		This may be used to regulate the modulation index.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, ModGsNOn[3] is set to logic level 1. This is not readable in the register.
		<b>Note:</b> The value of the register is only used if RF is generated by the driver (either Tx1RFEn or Tx2RFEn is set to logic level 1), otherwise the value ModGsNOff in the register CIU_GsNOFF is used.

#### 8.6.23.38 CIU\_CWGsP register (6318h)

Defines the conductance of the P-driver.

Table 249. CIU\_CWGsP register (address 6318h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-			CWGs	P[5:0]		
Reset	0	0	1	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 250. Description of CIU\_CWGsP bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	CWGsP[5:0]	The value of this register defines the conductance of the output P-driver, during times of no modulation.
		This may be used to regulate the output power and subsequently current consumption and operating distance.
		Note: The conductance value is binary weighted.
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, CWGsP[5] is set to logic level 1. This is not readable in the register.

#### 8.6.23.39 CIU\_ModGsP register (6319h)

Defines the driver P-output conductance for the time of modulation.

Table 251. CIU\_ModGsP register (address 6319h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-			ModGs	sP[5:0]		
Reset	0	0	1	0	0	0	0	0
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

Table 252. Description of CIU\_ModGsP bits

Bit	Symbol	Description				
7 to 6	-	Reserved.				
5 to 0	ModGsP[5:0]	The value of this register defines the conductance of the output P-driver for the time of modulation.				
		This may be used to regulate the modulation index.				
		Note: The conductance value is binary weighted.				
		<b>Note:</b> During CIU Power-down mode, if DriverSel[1:0] is not equal to 01b, ModGsP[5] is set to logic level 1. This is not readable in the register.				
		<b>Note:</b> If Force100ASK in CIU_TxAuto register is set to logic level 1, the ModGsP[5:0] setting has no effect.				

#### 8.6.23.40 CIU\_TMode register (631Ah)

Defines settings for the internal timer.

Table 253. CIU\_TMode register (address 631Ah) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TAuto	TGate	ed[1:0]	TAutoRestart		TPrescal	er_Hi[3:0]	
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 254. Description of CIU\_TMode bits

Bit	Symbol	Description					
7	TAuto	Set to logic 1, the timer starts automatically at the end of the transmission in all communication modes at all speed, or when InitialRFOn (in CIU_TxAuto) is set to logic 1 and the external RF field is switched on. The timer stops immediately after receiving the first data bit if RxMultiple in the CIU_RxMode register is set to logic level 0.					
			s set to logic 1, the timer never stops. In this case the stopped by setting the bit TStopNow in register to 1.				
		Set to logic 0 protocol.	indicates, that the timer is not influenced by the				
6 to 5	TGated[1:0]	The internal t	mer is running in gated mode.				
			gated mode, the bit TRunning is logic level 1 when the ed by the register bits. This bit does not influence the				
		Value	Description				
		00	No gated mode				
		01	Gated by SIGIN				
		10	Gated by AUX1				
		11	Reserved				
4	TAutoRestart	•	the timer automatically restart its count-down from e defined within when reaches zero.				
		Set to logic 0 to logic 1.	the timer decrements to zero and the bit TimerIRq is set				
3 to 0	TPrescaler_Hi[3:0]	Defines highe	er 4 bits for the TPrescaler.				
		The following	formula is used to calculate $f_{\text{Timer}}$ :				
			$f_{Timer} = 6.78MHz/T_{PreScaler}$				
		For detailed of	lescription see Section 8.6.17 "CIU_timer" on page 140.				
		Note: TPreScaler is defined with TPreScaler_Hi[3:0] in this register and TPreScaler_LO[7:0] in CIU_TPrescaler.					

#### 8.6.23.41 CIU\_TPrescaler register (631Bh)

Define the LSB of the Timer-Prescaler.

Table 255. CIU\_TPrescaler register (address 631Bh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TPrescaler_LO[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 256. Description of CIU TPrescaler bits

Bit	Symbol	Description				
7 to 0 TPrescaler_LO[7		Defines lower 8 bits for TPrescaler. The following formula is used to calculate $f_{\text{Timer}}$				
		$f_{Timer} = 6.78MHz/T_{PreScaler}$				
		For detailed description see Section 8.6.17 "CIU_timer" on page 140.				
		Note: The TPreScaler time is defined with TPreScaler_Hi[3:0] in CIU_TMode and TPreScaler_LO[7:0] in this register.				

#### 8.6.23.42 CIU\_TReload\_hi register (631Ch)

Defines the MSB of the 16-bit long timer reload value.

Table 257. CIU\_TReloadVal\_hi register (address 631Ch) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TReloadVal_Hi[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 258. Description of CIU\_TReloadVal\_hi bits

Bit	Symbol	Description
7 to 0	TReloadVal_Hi[7:0]	Defines the higher 8 bits for the TReloadValue.
		With a start event the timer loads with the TReloadValue. Changing this register affects the timer only with the next start event.
		<b>Note:</b> The reload value is defined with TReloadVal_Hi[7:0] in this register and TReloadVal_Lo[7:0] in CIU_TReloadVal_lo

#### 8.6.23.43 CIU\_TReloadVal\_lo register (631Dh)

Defines the LSB of the 16 bit long timer reload value.

Table 259. CIU\_TReload\_lo register (address 631Dh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TReloadVal_Lo[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 260. Description of CIU\_TReload\_lo bits

Bit	Symbol	Description
7 to 0	TReloadVal_Lo[7:0]	Defines the lower 8 bits for the TReloadValue.
		With a start event the timer loads with the TReloadValue. Changing this register affects the timer only with the next start event.
		<b>Note:</b> The reload value is defined with TReloadVal_Lo[7:0] in this register and TReloadVal_Hi[7:0] in CIU_TReload_Hi.

#### 8.6.23.44 CIU\_TCounterVal\_hi register (631Eh)

Defines the MSB byte of the current value of the timer.

Table 261. CIU\_TCounterVal\_hi register (address 631Eh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		TCounterVal_Hi[7:0]							
Reset	х	х	х	х	х	х	х	х	
Access	R	R	R	R	R	R	R	R	

Table 262. Description of CIU\_TCounterVal\_hi bits

Bit	Symbol	Description
7 to 0	TCounterVal_Hi[7:0]	MSB of the current value of the timer (Higher 8 bits).

#### 8.6.23.45 Register CIU\_TCounterVal\_lo (631Fh)

Defines the LSB byte of the current value of the timer.

Table 263. CIU\_TCounterVal\_lo register (address 631Fh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TCounterVal_LO[7:0]						
Reset	х	х	х	х	х	х	х	х
Access	R	R	R	R	R	R	R	R

Table 264. Description of CIU\_TCounterVal\_lo bits

Bit	Symbol	Description
7 to 0	TCounterVal_LO[7:0]	LSB of the current value of the timer (Lower 8 bits).

#### 8.6.23.46 CIU\_TestSel1 register (6321h)

General test signal configuration.

Table 265. CIU\_TestSel1 register (address 6321h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	LoadMo	dTst[1:0]	:0] SICclksel[1:0]		SICCIkD1	1 TstBusBitSel[2:0]		[2:0]
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 266. Description of CIU\_TestSel1 bits

Bit	Symbol	Description	Description			
7 to 6	LoadModTst[1:0]	Defines the te	est signal for the LOADMOD pin			
		<b>Note:</b> The bits LoadModSel in register CIU_TxSel has to be logic 1 to enable LoadModTst:				
		Value	Description			
		00	Low			
		01	High			
		10	RFU			
		11	TstBusBit as defined by the TestBusBitSel bit of this register			
5 to 4	SICclksel[1:0]	Defines the so	ource for the 13.56 MHz secure IC clock			
		Value	Description			
		00	GND - secure IC clock is switched off			
		01	Clock derivated by the internal oscillator			
		10	Internal CIU clock			
		11	Clock derivated from the RF Field			
3	SICCIkD1	•	the secure IC clock is delivered to P31 / UART_TX if the bit is set to logic 1.			
2 to 0	TstBusBitSel(2:0]	Select the TstBusBit from the test bus.				

#### 8.6.23.47 CIU\_TestSel2 register (6322h)

General test signal configuration and PRBS control.

Table 267. CIU\_TestSel2 register (address 6322h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	TstBusFlip	PRBS9	PRBS15		Tst	BusSel[4	l:0]	
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 268. Description of CIU\_TestSel2 bits

	<u> </u>	
Bit	Symbol	Description
7	TstBusFlip	If set to logic 1, the internal test bus(D6-D0) is mapped to the external test bus pins by the following order: D4,D3, D2,D6,D5, D0, D1. See Section 8.6.21.2 "CIU test bus" on page 151.
6	PRBS9	Starts and enables the PRBS9 sequence according ITU-TO150.
		<b>Note:</b> All relevant register to transmit data have to be configured before entering PRBS9 mode.
		<b>Note:</b> The data transmission of the defined sequence is started by the Transmit command.
5	PRBS15	Starts and enables the PRBS15 sequence according ITU-TO150.
		<b>Note:</b> All relevant register to transmit data have to be configured before entering PRBS15 mode.
		<b>Note:</b> The data transmission of the defined sequence is started by the Transmit command.
4 to 0	TstBusSel[4:0]	Selects the test bus source. See Section 8.6.21.2 "CIU test bus" on page 151.

#### 8.6.23.48 CIU\_TestPinEn register (6323h)

Enable the output drivers for the test pins.

Table 269. CIU\_TestPinEn register (address 6323h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		TestPinEn[7:0]						
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 270. Description of CIU\_TestPinEn bits

Bit	Symbol	Description
7 to 0	TestPinEn[7:0]	Each of the bit enables the output driver for an internal test pin: P70_IRQ (MSB), RSTOUT_N, P35, P34 / SIC_CLK, P33_INT1, P32_INT0, P31 / UART_TX, P30 / UART_RX (LSB). DataEn[7] enables P70_IRQ, DataEn[0] enables P30 / UART_RX.
		<b>Note:</b> The data transmission of the defined sequence is started by the Transmit command.

#### 8.6.23.49 CIU\_TestPinValue register (6324h)

Defines the values for the 7 bit test bus signals to be I/O on P70\_IRQ, RSTOUT\_N, P35, P34 / SIC\_CLK, P33\_INT1, P32\_INT0, P31 / UART\_TX and P30 / UART\_RX pins.

Table 271. CIU\_TestPinValue register (address 6324h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	useio		TestPinValue[6:0]					
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Table 272. Description of CIU\_TestPinValue bits

Bit	Symbol	Description
7	useio	Set to logic 1, this bit enables the I/O functionality for the internal test bus on the pins P70_IRQ (MSB), RSTOUT, P35, P34 / SIC_CLK, P33_INT1, P32_INT0, P31 / UART_TX, P30 / UART_RX (LSB)
		<b>Note:</b> Before using P34 / SIC_CLK as a test output, the SVDD switch should be closed. See register address 6106h.
6 to 0	TestPinValue[6:0]	UseIO set to logic 1, Read or write the value of the test bus.  UseIO set to logic 0, Read 000_0000. No write.

#### 8.6.23.50 CIU\_TestBus register (6325h)

Shows the status of the internal test bus.

Table 273. CIU\_TestBus register (address 6325h) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol		TestBus[7:0]							
Reset	х	х	х	х	х	х	х	х	
Access	R	R	R	R	R	R	R	R	

Table 274. Description of CIU\_TestBus bits

Bit	Symbol	Description
7 to 0	TestBus[7:0]	Shows the status of the internal test bus. The test bus is selected by the register CIU_TestSel2.
		See Section 8.6.21.2 "CIU test bus" on page 151.

#### 8.6.23.51 CIU\_AutoTest register (6326h)

Controls the digital self-test.

Table 275. CIU\_AutoTest register (address 6326h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	AmpRcv	-	-	SelfTest[3:0]			
Reset	0	1	0	0	0	0	0	0
Access	R	R	R	R	R/W	R/W	R/W	R/W

Table 276. Description of CIU\_AutoTest bits

Bit	Symbol	Description
7	-	Reserved.
6	AmpRcv	Set to logic 1 the internal signal processing in the receiver chain is performed non-linear. This increases the operating distance in communication modes at 106 kbit/s.
		<b>Note:</b> Due to non linearity the effects of MinLevel and CollLevel in CIU_RxThreshold register are as well non linear.
5 to 4	-	Reserved
3 to 0	SelfTest[3:0]	Enables the digital Self Test. The self-test can be started by the Selftest command in the CIU_Command register. The self-test is enabled by 1001.
		<b>Note:</b> For default operation the self-test has to be disabled (0000).

#### 8.6.23.52 CIU\_Version register (6327h)

Shows the version of the CIU.

Table 277. CIU\_Version register (address 6327h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol		Produ	ct[3:0]		Version[3:0]			
Reset	1	0	0	0	0	0	0	0
Access	R	R	R	R	R	R	R	R

Table 278. Description of CIU\_Version bits

Bit	Symbol	Description
7 to 4	Product	Product 1000 (PN533)
3 to 0	Version	Version 0000

#### 8.6.23.53 CIU\_AnalogTest register (6328h)

Controls the pins AUX1 and AUX2.

Table 279. CIU\_AnalogTest register (address 6328h) bit allocation

Bit	7	6	5	4	3	2	1	0		
Symbol		AnalogSelAux1[3:0]				AnalogSelAux2[3:0]				
Reset	0	0	0	0	0	0	0	0		
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		

#### Table 280. Description of CIU\_AnalogTest bits

Bit	Symbol	Description
7 to 4	AnalogSelAux1[3:0]	Controls the AUX1 pin. Note: All test signals are described in Section 8.6.21.3 "Test signals at pin AUX" on page 152.
		0000 Tristate
		0001 DAC output: register CIU_TestDAC1[1]
		0010 DAC output: test signal corr1[1]
		0011 DAC output: test signal corr2[1]
		0100 DAC output: test signal MinLevel[1]
		0101 DAC output: ADC_I[1]
		0110 DAC output: ADC_Q[1]
		0111 DAC output: ADC_I combined with ADC_Q[1]
		1000 Test signal for production test
		1001 secure IC clock
		1010 ErrorBusBit as described in <u>Table 178 on page 155</u>
		1011 Low
		1100 TxActive
		At 106 kbit/s: High during Start bit, Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Preamble, Sync,
		Data bits and CRC
		1101 RxActive
		At 106 kbit/s: High during Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Data bits and CRC
		1110 Subcarrier detected
		At 106 kbit/s: not applicable At 212 kbit/s and 424 kbit/s: High during last part of preamble, Sync, Data bits and CRC.
		1111 Test bus bit as defined by the TstBusBitSel in <u>Table 266 on page 191</u>

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Table 280. Description of CIU\_AnalogTest bits ...continued

Bit	Symbol	Desci	ription
3 to 0	AnalogSelAux2[3:0]		ols the AUX2 pin. Note: All test signals are described in on 8.6.21.3 "Test signals at pin AUX" on page 152.
		0000	Tristate
		0001	DAC output: register CIU_TestDAC2[2]
		0010	DAC output: test signal corr1[2]
		0011	DAC output: test signal corr2[2]
		0100	DAC output: test signal MinLevel[2]
		0101	DAC output: ADC_I[2]
		0110	DAC output: ADC_Q[2]
		0111	DAC output: ADC_I combined with ADC_Q[2]
		1000	Test signal for production test
		1001	secure IC clock
		1010	ErrorBusBit as described in Table 178 on page 155
		1011	Low
		1100	TxActive
			At 106 kbit/s: High during Start bit, Data bits, Parity and CRC
			At 212 kbit/s and 424 kbit/s: High during Preamble, Sync, Data bits and CRC
		1101	RxActive
			At 106 kbit/s: High during Data bits, Parity and CRC At 212 kbit/s and 424 kbit/s: High during Data bits and CRC
		1110	Subcarrier detected
			At 106 kbit/s: not applicable At 212 kbit/s and 424 kbit/s: High during last part of preamble, Sync, Data bits and CRC.
		1111	Test bus bit as defined by the TstBusBitSel in <u>Table 265 on page 191</u>

<sup>[1]</sup> Current output. The use of 1  $\,\mathrm{k}\Omega$  pull down resistor on AUX1 is recommended.

<sup>[2]</sup> Current output. The use of 1  $\,\mathrm{k}\Omega$  pull down resistor on AUX2 is recommended.

#### 8.6.23.54 CIU\_TestDAC1 register (6329h)

Defines the test value for TestDAC1.

Table 281. CIU\_TestDAC1 register (address 6329h) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	TestDAC1[5:0]					
Reset	0	0	Х	X	X	X	Х	X
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 282. Description of CIU\_TestDAC1 bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	TestDAC1[5:0]	Defines the test value for TestDAC1. The output of the DAC1 can be switched to AUX1 by setting AnalogSelAux1 to 0001 in the CIU_AnalogTest register.

#### 8.6.23.55 CIU\_TestDAC2 register (632Ah)

Defines the test value for TestDAC2.

Table 283. CIU\_TestDAC2 register (address 632Ah) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	-	-	TestDAC2[6:0]						
Reset	0	0	Х	Х	Х	Х	Х	X	
Access	R	R	R/W	R/W	R/W	R/W	R/W	R/W	

Table 284. Description of CIU\_TestDAC2 bits

Bit	Symbol	Description
7 to 6	-	Reserved.
5 to 0	TestDAC2[6:0]	Defines the test value for TestDAC2. The output of the DAC2 can be switched to AUX2 by setting AnalogSelAux2 to 0001 in the CIU_AnalogTest register.

#### 8.6.23.56 CIU\_TestADC register (632Bh)

Shows the actual value of ADC I and Q channel.

Table 285. CIU\_TestADC register (address 632Bh) bit allocation

Bit	7	6	5	4	3	2	1	0	
Symbol	ADC_I[3:0]			ADC_Q[3:0]					
Reset	Х	Х	Х	Х	Х	Х	Χ	X	
Access	R	R	R	R	R	R	R	R	

Table 286. Description of CIU\_TestADC bits

Bit	Symbol	Description
7 to 4	ADC_I[3:0]	Shows the actual value of ADC I channel.
3 to 0	ADC_Q[3:0]	Shows the actual value of ADC Q channel.

## 8.6.23.57 CIU\_RFlevelDet register (632Fh)

Power down of the RF level detector.

Table 287. CIU\_RFlevelDet register (address 632Fh) bit allocation

Bit	7	6	5	4	3	2	1	0
Symbol	-	-	-	pd_rflvldet	-	-	-	-
Reset	0	0	0	0	0	0	0	0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### Table 288. Description of CIU\_RFlevelDet bits

Bit	Symbol	Description
7 to 5	-	Reserved. These bits must be set to logic level 0.
4	pd_rfleveldet	Power down of the RF level detector.
		When set to logic 1, the RF level detector is in power down mode.
3 to 0	-	Reserved. These bits must be set to logic level 0.

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# 8.7 Registers map

# 8.7.1 Standard registers

#### Table 289. Standard registers mapping

Register address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bi
6000h to 6102h	Reserved							
6103h	Config I0_I1	int1_pol		pad_I1		pad_I0	enselif	
6104h	Observe_testbus			1			1	
6105h	Data_rng					data_rng		
6106h	Control_switch_rng		hide_svdd_sig	sic_switch_overload	sic_switch_en		cpu_need_rng	random_c
6107h	GPIRQ	gpirq_level_P71	gpirq_level_P50	gpirq_level_P35	gpirq_level_P34	gpirq_enable_P71	gpirq_enable_P50	gpirq_ena
6108h					Reserved			
6109h	LDO	Rese	erved	overcurrent_status	sel_over	current[1:0]	enoffset	soft_high
610Ah	i <sup>2</sup> c_wu_control						i <sup>2</sup> c_wu_en_wr	i <sup>2</sup> c_wu
610Bh					Reserved			
610Ch	Andet_control	andet_bot	andet_up	andet_ith	nI[1:0]		andet_ithh[2:0]	
610Dh	Reserved							
610Eh	NFC_WI_control					nfc_wi_status		nfc_wi_en_
610Fh					Reserved			
6200h	PCR CFR							
6201h	PCR CER					hsu_enable		
6202h	PCR ILR		porpulse_latched		enable_pdselif		gpirq_level	int1_
6203h	PCR Control			Rese	erved			clear_wak
6204h	PCR Status	i <sup>2</sup> c_wu	gpirq_wu	SPI_wu	HSU_wu	CIU_wu		int1
6205h	PCR Wakeupen	i <sup>2</sup> c_wu_en	GPIRQ_wu_en	SPI_on_en	HSU_on_en	CIU_wu_en		int1
6206h to 6300h					Reserved			1
6301h	CIU_Mode	MSBFirst	DetectSync	TXWaitRF	RxWaitRF	PolSigin	ModeDetOff	
6302h	CIU_TxMode	TxCRCEn		TxSpeed[2:0]		InvMod	TxMix	
6303h	CIU_RxMode	RXCRCEn		RxSpeed[2:0]		RxNoErr	RxMultiple	
6304h	CIU_TxControl	InvTx2RFon	InvTx1RFon	InvTx2RFoff	InvTx1RFoff	Tx2CW	CheckRF	Tx2F
6305h	CIU_TxAuto	AutoRFOFF	Force100ASK	AutoWakeUp		CAOn	InitialRFOn	Tx2RF
6306h	CIU_TxSel	LoadMo	dSel[1:0]	DriverSe	·l[1:0]		Sig	OutSel[3:0]
6307h	CIU_RxSel					xWait[5:0]		
6308h	CIU_RxThreshold		MinL	evel[3:0]				Colle
6309h	CIU_Demod	Addio	Q[1:0]	FixIQ		TauRo	cv[1:0]	

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PN5	Table 28	89. Standard regi	sters mapping	continued	continued						
PN5331B3HN	Register address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bi		
.	630Ah	CIU_FeINFC1	FelSync	Len[1:0]			DataLenMin[5:0]				
	630Bh	CIU_FeINFC2	WaitForSelected	ShortTimeSlot			Data	aLenMax[5:0]			
	630Ch	CIU_MifNFC		SensMiller[2:0]		TauMi	iller[1:0]	MFHalted			
	630Dh	CIU_ManualRCV		FastFiltMF_SO	DelayMF_SO	ParityDisable	LargeBWPLL	ManualHPCF			
	630Eh	CIU_TypeB	RxSOFReq	RxEOFReq	Reserved	EOFSOFWidth	NoTxSOF	NoTxEOF			
	630Fh to 6310h										
	6311h	CIU_CRCResultMSB				CRC	ResultMSB[7:0]				
	6312h	CIU_CRCResultLSB				CRC	ResultLSB[7:0]				
	6313h	CIU_GsNOff		CWGsNOff[3:0] ModGsNOff[3:0]							
	6314h	CIU_ModWidth				M	odWidth[7:0]				
All info	6315h	CIU_TxBitPhase	RcvClkChange	TxBitPhase[6:0]							
All information provided in this document is subject to legal disclaimers	6316h	CIU_RFCfg	RFLevelAmp		RxGain[2:0]	-	RFLevel[3:				
on pro	6317h	CIU_GsNOn		CWGs1	CWGsNOn[3:0]			Mode	GsNOn[3:0]		
vided	6318h	CIU_CWGsP	Rese	erved CWGsP[5:0]			WGsP[5:0]				
in this	6319h	CIU_ModGsP	Rese	erved		ModGsP[5:0]					
docu	631Ah	CIU_TMode	TAuto	TGate	TGated[1:0] TAutoRestart TPrescaler_H				caler_Hi[3:0]		
ment	631Bh	CIU_TPrescaler				TPre	escaler_LO[7:0]				
is sub	631Ch	CIU_TReloadVal_Hi				TRe	loadVal_Hi[7:0]				
ect to	631Dh	CIU_TReloadVal_Lo				TRel	loadVal_Lo[7:0]				
legal	631Eh	CIU_TCounterVal_hi				TCou	unterVal_Hi[7:0]				
disclai	631Fh	CIU_TCounterVal_lo				TCou	unterVal_LO[7:0]				
mers.	6320h					Reserved					
	6321h	CIU_TestSel1	LoadMod	dTst[1:0]	SICclkse	el[1:0]	SICCIkD1		TstBusE		
	6322h	CIU_TestSel2	TstBusFlip	PRBS9	PRBS15			TstBusSel[4:0	0]		
	6323h	CIU_TestPinEn				Te	estPinEn[7:0]				
	6324h	CIU_TestPinValue	useio				TestPinValue[6	3:0]			
	6325h	CIU_TestBus				Т	TestBus[7:0]				
	6326h	CIU_AutoTest	Reserved	AmpRcv	Reser	rved		Se	elfTest[3:0]		
0	6327h	CIU_Version		Pro	oduct			1	Version		
NXP	6328h	CIU_AnalogTest		AnalogSe	elAux1[3:0]			Analoç	gSelAux2[3:0		
B.V. 2	6329h	CIU_TestDAC1	Rese	rved			Tes	stDAC1[5:0]			
017. /	632Ah	CIU_TestDAC2	Rese	erved			Tes	stDAC2[5:0]			
\ll righ	632Bh	CIU_TestADC		ADC_I[3:0]				ADC_Q[3:0]	]		
© NXP B.V. 2017. All rights reserved.	632Ch to 632Eh			Reserved							

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Table 289. Standard regi	sters mappingcontin	ued
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Register address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bi	
632Fh	CIU_RFlevelDet				pd_rfleveldet			I	
6330h	SIC_CLK	sic_clk_p34_en				Errorbusbitenable		Errorbus	
6331h	CIU_Command			RcvOff	Power-down		C	ommand	
6332h	CIU_CommlEn	Reserved	TxIEn	RXIEn	IdleIEn	HiAlertIEn	LoAlertIEn	Errl	
6333h	CIU_DivIEn				SiginAct IEn	ModelEn	CRCIEn	RfOr	
6334h	CIU_Commlrq	Set1	TxIRq	RxIRq	IdleIrq	HiAltertIRq	LoAlertIRq	Errl	
6335h	CIU_DivIrq	Set2		1	SiginActIrq	ModelRq	CRCIRq	RfOr	
6336h	CIU_Error	WrErr	TempErr	RFErr	BufferOvfl	CollErr	CRCErr	Parit	
6337h	CIU_Status1	CIU_IRQ_1	CRCOk	CRCReady	CIU_IRQ_0	TRunning	RFOn	HiA	
6338h	CIU_Status2	TempSensClear		RFFreqOK	TgActivated	MFCrypto1On		Modem	
6339h	CIU_FIFOData			1	F	FIFOLevel[6:0]			
633Ah	CIU_FIFOLevel	FlushBuffer							
633Bh	CIU_WaterLevel				Wate	WaterLevel[5:0]			
633Ch	CIU_Control	TStopNow	TStartNow	WrNFCIP-1IDtoFIF O	Initiator			RxLas	
633Dh	CIU_BitFraming	StartSend		RxAlign[2:0]				TxLas	
633Eh	CIU_Coll	ValuesAfterColl	Reserved	CollPosNotValid			CollPos		
633Fh to FFFFh					Reserved				

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# 8.7.2 SFR registers

# Table 290. SFR registers mapping

SFR address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2			
80h					Reserved					
81h[1]	SP Stack Pointer					SP[7:0]				
82h[1]	DPL Data Pointer Low				[	DPL[7:0]				
83h[1]	DPH Data Pointer High				D	PLH7:0]				
84h to 86h					Reserved					
87h	PCON	SMOD			Reserved			CF		
88h	T01CON	TF1	TR1	TF0	TR0	IE1	IT1			
89h	T01MOD	GATE1	C/T1	M11	M10	GATE0	C/T0			
8Ah	TOL	T0L.7	T0L.6	T0L.5	T0L.4	T0L.3	T0L.2	Т		
8Bh	T1L	T1L.7	T1L.6	T1L.5	T1L.4	T1L.3	T1L.2	Т		
8Ch	ТОН	T0H.7	T0H.6	T0H.5	T0H.4	T0H.3	T0H.2	Т		
8Dh	T1H	T1H.7	T1H.6	T1H.5	T1H.4	T1H.3	T1H.2	Т		
8Eh to 97h		1	Reserved							
98h	SOCON	SM0	SM1	SM2	REN	TB8	RB8			
99h	SOBUF				Sc	BUF[7:0]		'		
9Ah	RWL				RWa	terlevel[7:0]				
9Bh	TWL				TWa	terlevel[7:0]				
9Ch	FIFOFS				Transmi	tFreespace[7:0]				
9Dh	FIFOFF				Receiv	eFullness[7:0]				
9Eh	SFF	FIFO_EN	Reserved	TWLL	TFF	TFE	RWLH			
9Fh	FIT	Reset	Reserved	WCOL_IRQ	TWLL_IRQ	TFF_IRQ	RWLH_IRQ	RO\		
A0h	Reserved							'		
A1h	FITEN	TFLUSH	RFLUSH	EN_WCOL_IRQ	EN_TWLL_IRQ	EN_TFF_ IRQ	EN_RWLH_IRQ	EN_R		
A2h	FDATA				F	DATA[7:0]		'		
A3h	FSIZE				Rece	eiveSize[7:0]				
A4h to A7h					Reserved					
A8h	IE0	IE0_7	IE0_6	IE0_5	IE0_4	IE0_3	IE0_2	II		
A9h	SPIcontrol	Reserv	ed	Enable	Reserved	СРНА	CPOL			
AAh	SPIstatus		Reserved TR_FE RCV_OVR							
ABh	HSU_STA	set_bit	Re	eserved	disable_preamb	amb irq_rx_over_en irq_rx_fer_en				
ACh	HSU_CTR	hsu_wu_en	start_frame	tx_stop	obit[1:0]	rx_stopbit	tx_en	r.		
ADh	HSU_PRE		1	-	hsu_p	rescaler[7:0]		1		
AEh	HSU_CNT				hsu_	counter[7:0]				

Table 290. SFR registers mapping ...continued

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SFR address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	
AFh					Reserved			
B0h	P3	Reserve	ed	P3[5]	P3[4]	P3[3]	P3[2]	ı
B1h to B7h					Reserved			
B8h	IP0	IP0_7	IP0_6	IP0_5	IP0_4	IP0_3	IP0_2	H
B9h to C7h		+	+	-	Reserved	-	-	
C8h	T2CON	TF2	Reserved	RCLK0	TCLK0	Reserved	TR2	(
C9h	T2MOD			Reserved			T2RD	
CAh	RCAP2L	R2L.7	R2L.6	R2L.5	R2L.4	R2L.3	R2L.2	R
CBh	RCAP2H	R2H.7	R2H.6	R2H.5	R2H.4	R2H.3	R2H.2	R
CCh	T2L	T2L.7	T2L.6	T2L.5	T2L.4	T2L.3	T2L.2	Т
CDh	T2H	T2H.7	T2H.6	T2H.5	T2H.4	T2H.3	T2H.2	Т
CEh to CFh					Reserved			
D0h[1]	PSW Program Status Word	PSW[7:0]						
D1h	CIU_Command	Reserve	ed	RcvOff	Power-down		Con	mmand
D2h	CIU_CommlEn		TxIEn	RXIEn	IdleIEn	HiAlertlEn	LoAlertIEn	Е
D3h	CIU_DivIEn		Reserved		SiginAct IEn	ModelEn	CRCIEn	Rf
D4h	CIU_Commlrq	Set1	TxIRq	RxIRq	IdleIrq	HiAltertIRq	LoAlertIRq	Е
D5h	CIU_DivIrq	Set2	Re	eserved	SiginActIrq	ModelRq	CRCIRq	Rf0
D6h	CIU_Error	WrErr	TempErr	RFErr	BufferOvfl	CollErr	CRCErr	Pa
D7h	Reserved							
D8h	I <sup>2</sup> CCON	CR[2]	ENS1	STA	STO	SI	AA	
D9h	I <sup>2</sup> CSTA					ST[7:0]		
DAh	I <sup>2</sup> CDAT				J <sup>2</sup> (	CDAT[7:0]		
DBh	I <sup>2</sup> CADR				SA[6:0]			
DCh to DEh					Reserved			
DFh	CIU_Status1	CIU_IRQ_1	CRCOk	CRCReady	CIU_IRQ_0	TRunning	RFOn	Н
E0h[1]	ACC Accumulator	ACC[7:0]		· 				
E1h to E7h	Reserved							
E8h	IE1	IE1_7	Reserved	IE1_5	IE1_4	IE1_3	IE1_2	Re
E9h	CIU_Status2	TempSensClear	Reserved	RFFreqOK	TgActivated	MFCrypto1On		Modem
EAh	CIU_FIFOData				FIF	OData[7:0]		
EBh	CIU_FIFOLevel	FlushBuffer				FIFOLevel[6:	.0]	
ECh	CIU_WaterLevel					Water	rLevel[5:0]	
EDh	CIU_Control	TStopNow	TStartNow	WrNFCIP-1ID to FIFO	Initiator	Reserved		RxLas
EEh	CIU_BitFraming	StartSend		RxAlign[2:0]		Reserved		TxLast

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Table 290. SFR registers mapping ...continued

	•							
SFR address	Register name	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	
EFh	CIU_Coll	ValuesAfterColl		CollPosNotValid			CollPos	
F0h[1]	B register					B[7:0]		
F1h to F3h					Reserved			
F4h	P7FGA						P7CFGA[2]	P70
F5h	P7FGB						P7CFGB[2]	P70
F6h	Reserved							
F7h	P7						P7[2]	ı
F8h	IP1	IP1_7		IP1_5	IP1_4	IP1_3	IP1_2	
F9h					Reserved		'	
FAh[1]	XRAMP						XRAMP[4:0]	
FBh	Reserved							
FCh	P3FGA			P3CFGA[5]	P3CFGA[4]	P3CFGA[3]	P3CFGA[2]	P30
FDh	P3FGB			P3CFGB[5]	P3CFGB[4]	P3CFGB[3]	P3CFGB[2]	P30
FEh to FFh		1		1	Reserved	1		

<sup>[1]</sup> This register is not described in this document as it is a standard 80C51 register.

# 9. Limiting values

Table 291. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions	Min	Max	Unit
$\begin{array}{l} PV_{DD}, \\ SV_{DD}, \\ TV_{DD}, \\ AV_{DD}, \\ DV_{DD} \end{array}$	Supply Voltages		-0.5	4	V
$V_{BUS}$	USB Supply Voltage		-0.5	5.5	V
$P_{tot}$	Total power dissipation			608	mW
I <sub>SVDD</sub>	Maximum current in SVDD switch			30	mΑ
$V_{INRX}$	Dynamic RX Input voltage Range	input signal at 13.56 MHz	-0.7	AVDD+1	V
V <sub>INTX1</sub> , V <sub>INTX2</sub>	DynamicTX1 input Voltage Range	input signal at 13.56 MHz	-1.2	TVDD+1.3	V
I <sub>TX1</sub> , I <sub>TX2</sub>	Maximum current in transmitter TX1	input signal at 13.56 MHz	-300	300	mA
$V_{ESD}$	Electrostatic discharge voltage				
$V_{ESDH}$	ESD Susceptibility (Human Body model)	EIA/JESD22-A114-E		± 2.0	kV
$V_{\text{ESDM}}$	ESD Susceptibility (Machine model)	EIA/JESD22-A115-A		200	V
$V_{\text{ESDC}}$	ESD Susceptibility (Charge Device model)	EIA/JESD22-C101-C		± 1.0	kV
T <sub>stg</sub>	Storage temperature		-55	150	°C
Tj	Junction temperature		-40	100	°C

# 10. Recommended operating conditions

**Table 292. Operating conditions** 

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$T_{amb}$	Ambient Temperature			-30	+25	+85	°C
$V_{BUS}$	USB Supply Voltage	$V_{SS} = 0 V$	[1]	4.02	5	5.25	V
	Supply Voltage (Non USB mode)	$VBUS = D_{VDD}$ $V_{SS} = 0 V$		2.5	3.3	3.6	V
$TV_DD, \ AV_DD, \ DV_DD$	Supply Voltages	$TV_{DD} = AV_{DD} = DV_{DD}$ $V_{SS} = 0 V$	[2][3]	2.5	3.3	3.6	V
$PV_{DD}$	Supply Voltage for host interface	$V_{SS} = 0 V$	[3]	1.6	1.8 to 3.3	3.6	V

<sup>[1]</sup>  $V_{SS}$  represents  $DV_{SS}$ ,  $TV_{SS1}$ ,  $TV_{SS2}$ ,  $AV_{SS}$ .

#### 11. Thermal characteristics

Table 293. Thermal characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R <sub>thj-a</sub>	thermal resistance from junction to ambient	in free air with exposed pad soldered on a 4 layer Jedec PCB-0.5		37	41.1	K/W

<sup>[2]</sup>  $AV_{SS}$ ,  $DV_{DD}$  and  $TV_{DD}$  shall always be on the same voltage level.

<sup>[3]</sup> Supply voltage of  $AV_{SS}$ ,  $DV_{DD}$  and  $TV_{DD}$  below 3 V reduces the performance (e.g. the achievable operating distance).

#### 12. Characteristics

Unless otherwise specified, the limits are given for the full operating conditions. The typical value is given for 25°C,  $DV_{DD}$  = 3.4 V and  $PV_{DD}$  = 3 V in non USB bus power mode,  $V_{BUS}$  = 5 V in USB power mode.

Timings are only given from characterization results.

#### 12.1 Power management characteristics

#### 12.1.1 Current consumption characteristics

Table 294. Current consumption characteristics

Tubic 254.	our chi consumption char	actoristics					
Symbol	Parameter	Conditions		Min	Тур	Max	Unit
I <sub>HPD</sub>	Hard-Power-down current (Not powered from USB)	$AV_{DD} = DV_{DD} = TV_{DD} = PV_{DD} = 3 V$ , RF level detector off	<u>[5]</u>		1.3	10	μА
I <sub>SPD</sub>	Soft-Power-down current (Not powered from USB)	$AV_{DD} = DV_{DD} = TV_{DD} = PV_{DD} = 3 V,$ $SV_{DD} = 0 V, RF level$ detector on	<u>[5]</u>		9	30	μΑ
I <sub>suspend</sub>	USB suspend current	$V_{BUS} = 5 \text{ V},$ $AV_{DD} = DV_{DD} = TV_{DD} =$ $PV_{DD} = 3 \text{ V}, SV_{DD} = 3 \text{ V},$ PVDD = 3  V RF level detector on (without resistor on $\pm$ D)	<u>[5]</u>		120	250	μΑ
$I_{DVDD}$	Digital supply current	$AV_{DD} = DV_{DD} = TV_{DD} = PV_{DD} = 3 V,$ RF level detector on			12		
I <sub>AVDD</sub>	Analog supply current	$AV_{DD} = DV_{DD} = TV_{DD} = PV_{DD} = 3 V,$ RF level detector on			3	6	mA
I <sub>AVDDrcvoff</sub>		$AV_{DD} = DV_{DD} = TV_{DD} = PV_{DD} = 3 V,$ RF level detector off			1.5	5	mA
I <sub>PVDD</sub>	Pad supply current		[2]			30	mA
I <sub>SVDD</sub>	Output supply current	sam_switch_en set to 1	<u>[3]</u>			30	mΑ
I <sub>TVDD</sub>	Transmitter supply current	Continuous wave, $TV_{DD} = 3 V$	[1][4]		60	100	mA

<sup>[1]</sup> ITVDD depends on TV<sub>DD</sub> and the external circuitry connected to Tx1 and Tx2.

Typical value using a complementary driver configuration and an antenna matched to 40  $\Omega$  between TX1 and TX2 at 13.56 MHz.

<sup>[2]</sup> IPV<sub>DD</sub> depends on the overall load at the digital pins.

<sup>[3]</sup>  $ISV_{DD}$  depends on the overall load on  $SV_{DD}$  pad.

<sup>[4]</sup> During operation with a typical circuitry the overall current is below 100 mA.

<sup>[5]</sup> ISPD and IHPD are the total currents over all supplies.

#### 12.1.2 Voltage regulator characteristics

Table 295. Voltage regulator characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{BUS}$	USB Supply voltage (USB mode)	$V_{SS} = 0 V$		4.02	5	5.25	V
$D_{VDD}$	Supply voltage after Inrush current limitation (USB mode)	From $I_{DVDD} = 0$ to $I_{DVDD} = 150$ mA		2.95	3.3	3.6	V
$IV_{BUS}$	Maximum load current (USB mode)	measured on $V_{\text{BUS}}$				150	mA
	Maximum Inrush current limitation	At power-up (curlimofff = 0)				100	mA
$V_{th1}$	Reset threshold on DV <sub>DD</sub> falling			1.90	2.15	2.40	V
Vhys	Vth1 hysteresis			35	60	85	mV
Cdec	DVDD decoupling capacitor		[2]	8	10		μF

<sup>[1]</sup> The internal regulator is only enabled when the USB interface is selected by I0 and I1.

### 12.2 Antenna presence self test thresholds

The following values are guaranteed by design. Only functional is done in production for cases andet\_ithl[1:0]=10b and for andet\_ithl[2:0]=011b.

Table 296. Antenna presence detection lower levels characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>AndetH</sub>	I <sub>DVDD</sub> lower current threshold for	andet_ithl[1:0] = $00b$		5		mΑ
	antenna presence detection	andet_ithl[1:0] = 01b		15		mΑ
		andet_ithl[1:0] = 10b		25		mΑ
		andet_ithl[1:0] = 11b		35		mΑ

Table 297. Antenna Presence Detection Upper Levels characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>AndetH</sub>	I <sub>DVDD</sub> upper current threshold for	andet_ithh[2:0] = 000b		45		mΑ
	antenna presence detection	andet_ithh[2:0] = 001b		60		mΑ
		andet_ithh[2:0] = 010b		75		mΑ
		andet_ithh[2:0] = 011b		90		mΑ
		andet_ithh[2:0] = 100b		105		mΑ
		andet_ithh[2:0] = 101b		120		mΑ
		andet_ithh[2:0] = 110b		135		mΑ
		andet_ithh[2:0] = 111b		150		mA

<sup>[2]</sup> For more details on Implementation of decoupling capacitor refer to Figure 30, Figure 31 and Figure 32 in "Power distribution section".

#### 12.3 Typical 27.12 MHz Crystal requirements

Table 298. Crystal requirements

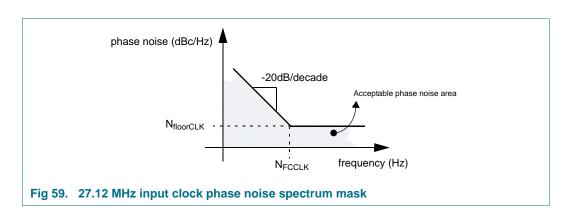
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$f_{XTAL}$	XTAL frequency		27.107	27.12	27.133	MHz
ESR	Equivalent series resistance				100	Ω
$C_{LOAD}$	Load capacitance			10		pF
$P_{XTAL}$	Drive level		100			μW

## 12.4 Pin characteristics for 27.12 MHz XTAL Oscillator (OSCIN, OSCOUT)

Table 299. Pin characteristics for 27.12 MHz XTAL Oscillator (OSCIN, OSCOUT)

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
I <sub>LeakOSCIN</sub>	Input Leakage current	RSTPD_N = 0 V		-1		1	mA
V <sub>IHOSCIN</sub>	High level Input voltage			$0.7 \times AV_{DD}$		$AV_{DD}$	V
V <sub>ILOSCIN</sub>	Low level Input voltage			0		$0.3 \times AV_{DD}$	V
D <sub>COSCIN</sub>	DC input voltage				0.65		V
C <sub>inOSCIN</sub>	OSCIN Input Capacitance	$AV_{DD} = 2.8 \text{ V},$ VDC = 0.65  V, VAC = 1  Vpp			2		pF
V <sub>OHOSCOUT</sub>	High level output voltage				1.1		V
V <sub>OLOSCOUT</sub>	Low level output voltage				0.2		V
C <sub>inOSCOUT</sub>	Input Capacitance				2		pF
f <sub>OSCIN</sub>	Clock Frequency			-500 ppm	27.12	+500 ppm	MHZ
D <sub>FEC</sub>	Duty Cycle of Clock Frequency			40	50	60	%
$N_{floorCLK}$	Clock phase noise floor		[1]			-140	dBc/Hz
N <sub>FCCLK</sub>	Clock phase noise corner	NfloorCLK = -140dBc/Hz; -20dB/decade slope	[1]			50	kHz

<sup>[1]</sup> NFCLK and NCCLK define the mask for maximum acceptable phase noise of the clock signal at the OSCIN, OSCOUT inputs. See Figure 59 "27.12 MHz input clock phase noise spectrum mask".



PN5331B3HN

#### 12.5 RSTPD\_N input pin characteristics

Table 300. RSTPD\_N input pin characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{IH}$	High level input voltage		PV <sub>DD</sub> -0.4		$PV_DD$	V
V <sub>IL</sub>	Low level input voltage		0		0.4	V
I <sub>IH</sub>	High level input current	$V_I = PV_{DD}$	-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$	-1		1	μΑ
C <sub>in</sub>	Input capacitance			2.5		pF

#### 12.6 Input pin characteristics for I0, I1 and TESTEN

Table 301. Input pin characteristics for I0, I1 and TESTEN

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{IH}$	High level input voltage				$DV_DD$	V
V <sub>IL</sub>	Low level input voltage		[2] 0		$0.3 \times \text{DV}_\text{DD}$	V
I <sub>IH</sub>	High level input current I0 and I1	$V_I = DV_{DD}$	<u>⅓</u> -1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$	-1		1	μΑ
$C_{in}$	Input capacitance			2.5		pF

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $DV_{DD}$  - 0.4 V.

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V.

<sup>[3]</sup> TESTEN should never be set to high level in the application. It is used for production test purpose only. It is recommended to connect TESTEN to ground although there is a pull-down included.

## 12.7 RSTOUT\_N output pin characteristics

Table 302. RSTOUT\_N output pin characteristics

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{OH}$	High level output voltage	$PV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$		$0.7 \times PV_{DD}$		$PV_{DD}$	V
		$PV_{DD} = 1.8 \text{ V},$ $I_{OH} = -2 \text{ mA}$	[2]	$0.7 \times PV_{DD}$		$PV_{DD}$	V
V <sub>OL</sub>	Low level output voltage	$PV_{DD} = 3 V$ , $I_{OL} = 4 \text{ mA}$		0		$0.3 \times PV_{DD}$	V
		$PV_{DD} = 1.8 \text{ V},$ $I_{OL} = 2 \text{ mA}$	[2]	0		$0.3 \times PV_{DD}$	V
I <sub>OH</sub>	High level output current	$PV_{DD} = 3 \text{ V},$ $V_{OH} = 0.8 \times PV_{DD}$	[3]	-4			mA
		$PV_{DD} = 1.8 \text{ V},$ $V_{OH} = 0.7 \times PV_{DD}$		-2			mA
I <sub>OL</sub>	Low level output current	$PV_{DD} = 3 V$ , $V_{OL} = 0.2 \times PV_{DD}$	[3]	4			mΑ
		$PV_{DD} = 1.8 \text{ V},$ $V_{OL} = 0.3 \times PV_{DD}$		2			mA
C <sub>out</sub>	Load capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{aligned} \text{PV}_{\text{DD}} &= 3 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.8 \times \text{PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.2 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$	[1]			13.5	ns
		$\begin{aligned} & \text{PV}_{\text{DD}} = 1.8 \text{ V,} \\ & \text{V}_{\text{OH}} = 0.7 \times \text{PV}_{\text{DD}}, \\ & \text{V}_{\text{OL}} = 0.3 \times \text{PV}_{\text{DD}}, \\ & \text{C}_{\text{out}} = 30 \text{ pF} \end{aligned}$				10.8	ns

<sup>[1]</sup>  $I_{OH}$  and  $I_{OL}$  give the output drive capability from which the rise and fall times may be calculated as a function of the load capacitance.

<sup>[2]</sup> Data at PVDD= 1.8V are only given from characterization results.

<sup>[3]</sup> The  $I_{OH}$  and  $I_{OL}$  give the output driving capability and allow to calculate directly the rise and fall time as function of the load capacitance

# 12.8 Input/output characteristics for pin P70\_IRQ

Table 303. Input/output pin characteristics for pin P70\_IRQ

High level input voltage						
		[1]	$0.7 \times PV_{DD}$		$PV_{DD}$	V
Low level input voltage		[2]	0		$0.3 \times PV_{DD}$	V
Push-pull mode high level output voltage	$PV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$		$0.7 \times PV_{DD}$		$PV_{DD}$	V
	$PV_{DD} = 1.8 \text{ V},$ $I_{OH} = -2 \text{ mA}$	[3]	$0.7 \times PV_{DD}$		$PV_{DD}$	V
Push-pull mode low level output voltage	$PV_{DD} = 3 V$ , $I_{OL} = 4 \text{ mA}$		0		$0.3 \times PV_{DD}$	V
	$PV_{DD} = 1.8 \text{ V},$ $I_{OL} = 2 \text{ mA}$	[3]	0		$0.3 \times PV_{DD}$	V
Input mode high level input current	$V_I = DV_{DD}$		-1		1	μΑ
Input mode low level input current	V <sub>I</sub> = 0 V		-1		1	μΑ
High level output current	$PV_{DD} = 3 V$ , $V_{OH} = 0.8 \times PV_{DD}$	[4]	-4			mA
Low level output current	$PV_{DD} = 3 V$ , $V_{OL} = 0.2 \times PV_{DD}$	[4]	4			mΑ
Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
Input capacitance				2.5		pF
Load capacitance					30	pF
Rise and fall times	$\begin{aligned} \text{PV}_{\text{DD}} &= 3 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.8 \times \text{PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.2 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$				13.5	ns
	$\begin{aligned} \text{PV}_{\text{DD}} &= 1.8 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.7 \times \text{PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.3 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$				10.8	ns
	Push-pull mode low level output voltage  Input mode high level input current Input mode low level input current High level output current Low level output current Input leakage current Input capacitance Load capacitance	level output voltage	level output voltage $I_{OH} = -4 \text{ mA}$ $PV_{DD} = 1.8 \text{ V},  [3]$ $I_{OH} = -2 \text{ mA}$ $PV_{DD} = 3 \text{ V},  [3]$ $I_{OL} = 4 \text{ mA}$ $PV_{DD} = 1.8 \text{ V},  [3]$ $I_{DL} = 4 \text{ mA}$ $PV_{DD} = 1.8 \text{ V},  [3]$ $I_{DL} = 2 \text{ mA}$ $Input mode high level input current$ $Input mode low level input current$ $Input mode low level input current$ $High level output  PV_{DD} = 3 \text{ V},  [4]$ $V_{OH} = 0.8 \times PV_{DD}$ $Input leakage current$ $Input capacitance$ $Input capacitance$ $Input capacitance$ $Input capacitance$ $Input leakage current$ $Input capacitance$ $Input capacitance$ $Input capacitance$ $Input leakage current PV_{DD} = 3 \text{ V},  [4]$ $V_{OH} = 0.2 \times PV_{DD}$ $V_{OH} = 0.8 \times PV_{DD},  V_{OH} = 0.8 \times PV_{DD},  V_{OH} = 0.2 \times PV_{DD},  V_{OH} = 0.2 \times PV_{DD},  V_{OH} = 0.7 \times PV_{DD},  V_{OH} = 0.3 \times PV_{DD}$		level output voltage $ \begin{array}{l} I_{OH} = \text{-4 mA} \\ \hline PV_{DD} = 1.8 \text{ V,} \\ I_{OH} = \text{-2 mA} \\ \hline \\ PV_{DD} = 3 \text{ V,} \\ I_{OL} = 4 \text{ mA} \\ \hline \\ PV_{DD} = 1.8 \text{ V,} \\ I_{OL} = 2 \text{ mA} \\ \hline \\ Input mode high level input current \\ \hline \\ Input mode low level input current \\ \hline \\ Input mode low level input current \\ \hline \\ Input mode low level output current \\ \hline \\ U_I = DV_{DD} \\ \hline \\ Input evel output current \\ \hline \\ V_{OH} = 3 \text{ V,} \\ V_{OH} = 0.8 \times PV_{DD} \\ \hline \\ Input leakage current \\ \hline \\ Input capacitance \\ \hline \\ Input capacitance \\ \hline \\ Rise and fall times \\ \hline \\ PV_{DD} = 3 \text{ V,} \\ V_{OH} = 0.4 \text{ V} \\ V_{OH} = 0.4 \text{ V} \\ V_{OH} = 0.4 \text{ V} \\ V_{OH} = 0.8 \times PV_{DD}, \\ V_{OH} = 0.8 \times PV_{DD}, \\ V_{OH} = 0.8 \times PV_{DD}, \\ V_{OH} = 0.2 \times PV_{DD}, \\ V_{OH} = 0.3 \times PV_{DD}, \\ V_{OH} = 0.7 \times PV_{DD}, \\ V_{OH} = 0.7 \times PV_{DD}, \\ V_{OH} = 0.3 \times PV_{DD}, \\$	level output voltage $ \frac{I_{OH} = -4 \text{ mA}}{PV_{DD} = 1.8 \text{ V,}} \qquad \frac{[3]}{I_{OH} = -2 \text{ mA}} = 0 $ PV <sub>DD</sub> PV <sub>D</sub>

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $PV_{DD}$  - 0.4 V.

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V.

<sup>[3]</sup> Data at PV<sub>DD</sub> = 1.8 V are only given from characterization results.

<sup>[4]</sup> The I<sub>OH</sub> and I<sub>OL</sub> give the output driving capability and allow to calculate directly the rise and fall time as function of the load capacitance.

# 12.9 Input/output pin characteristics for P30 / UART\_RX, P31 / UART\_TX, P32 INT0, P33 INT1

Table 304. Input/output pin characteristics for P30 / UART\_RX, P31 / UART\_TX, P32\_INT0, P33\_INT1

Parameter	Conditions		Min	Тур	Max	Unit
High level input voltage		[1]	$0.7 \times PV_{DD}$		$PV_{DD}$	V
Low level input voltage		[2]	0		$0.3 \times PV_{DD}$	V
Push-pull mode high level output voltage	$PV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$		PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
	$PV_{DD} = 1.8 \text{ V},$ $I_{OH} = -2 \text{ mA}$	[3]	PV <sub>DD</sub> - 0.4		$PV_{DD}$	V
Push-pull mode low level output voltage	$PV_{DD} = 3 V$ , $I_{OL} = 4 \text{ mA}$		0		0.4	V
	$PV_{DD} = 1.8 \text{ V},$ $I_{OL} = 2 \text{ mA}$	[3]	0		0.4	V
Input mode high level input current	$V_I = PV_{DD}$		-1		1	μΑ
Input mode low level input current	$V_I = 0 V$		-1		1	μΑ
High level output current	$PV_{DD} = 3 V,$ $V_{OH} = 0.8 \times PV_{DD}$	[4]	-4			mA
Low level output current	$PV_{DD} = 3 V$ , $V_{OL} = 0.2 \times PV_{DD}$	[4]	4			mA
Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
Input capacitance				2.5		pF
Load capacitance					30	pF
Rise and fall times	$\begin{aligned} \text{PV}_{\text{DD}} &= 3 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.8 \times \text{PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.2 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$				13.5	ns
	$\begin{aligned} \text{PV}_{\text{DD}} &= 1.8 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.7 \text{ PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.3 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$				10.8	ns
	High level input voltage Low level input voltage Push-pull mode high level output voltage  Push-pull mode low level output voltage  Input mode high level input current Input mode low level input current  High level output current  Low level output current Input leakage current Input capacitance Load capacitance		$ \begin{array}{c c} \mbox{High level input voltage} & \mbox{ \begin{tabular}{ c c c c } \hline Low level input voltage \\ \hline \mbox{Push-pull mode high level output voltage} & \mbox{PV}_{DD} = 3 \ V, \\ \mbox{I}_{OH} = -4 \ mA \\ \hline \mbox{PV}_{DD} = 1.8 \ V, \\ \mbox{I}_{OL} = 2 \ mA \\ \hline \mbox{PV}_{DD} = 3 \ V, \\ \mbox{I}_{OL} = 4 \ mA \\ \hline \mbox{PV}_{DD} = 1.8 \ V, \\ \mbox{I}_{OL} = 2 \ mA \\ \hline \mbox{Input mode high level input current}} & \mbox{V}_{I} = \mbox{PV}_{DD} \\ \hline \mbox{Input mode low level input current} & \mbox{V}_{I} = \mbox{PV}_{DD} \\ \hline \mbox{High level output current}} & \mbox{PV}_{DD} = 3 \ V, \\ \mbox{V}_{OH} = 0.8 \times \mbox{PV}_{DD} \\ \hline \mbox{Input leakage current}} & \mbox{RSTPD}_{N} = 0.4 \ V \\ \hline \mbox{Input capacitance} \\ \hline \mbox{Load capacitance}} & \mbox{PV}_{DD} = 3 \ V, \\ \mbox{V}_{OH} = 0.8 \times \mbox{PV}_{DD}, \\ \mbox{V}_{OL} = 0.2 \times \mbox{PV}_{DD}, \\ \mbox{V}_{OL} = 0.2 \times \mbox{PV}_{DD}, \\ \mbox{V}_{OL} = 0.2 \times \mbox{PV}_{DD}, \\ \mbox{V}_{OL} = 0.7 \ \mbox{PV}_{DD}, \\ \mbox{V}_{OH} = 0.7 \ \mbox{PV}_{DD}, \\ \mbox{V}_{OL} = 0.3 \times \mb$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	$\begin{array}{llllllllllllllllllllllllllllllllllll$	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $PV_{DD}$  - 0.4 V.

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V

<sup>[3]</sup> Data at  $PV_{DD} = 1.8 \text{ V}$  are only given from characterization results.

<sup>[4]</sup> The I<sub>OH</sub> and I<sub>OL</sub> give the output driving capability and allow to calculate directly the rise and fall time as function of the load capacitance.

## 12.10 Input/output pin characteristics for P35

Table 305. Input/output pin characteristics for P35

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level Input voltage		[1]	$0.7 \times \text{DV}_\text{DD}$		$DV_DD$	V
$V_{IL}$	Low level Input voltage		[2]	0		$0.3 \times \text{DV}_\text{DD}$	V
V <sub>OH</sub>	High level output voltage	$DV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$		DV <sub>DD</sub> - 0.4		$DV_DD$	V
V <sub>OL</sub>	Low level output voltage	$DV_{DD} = 3 V$ , $I_{OL} = 4 \text{ mA}$		0		0.4	V
I <sub>IH</sub>	High level input current	$V_I = DV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_1 = 0 V$		-1		1	μΑ
I <sub>OH</sub>	High level output current	$DV_{DD} = 3 V$ , $V_{OH} = 0.8 \times PV_{DD}$	[4]	-4			mA
I <sub>OL</sub>	Low level output current	$DV_{DD} = 3 V$ , $V_{OL} = 0.2 \times PV_{DD}$	[4]	4			mA
I <sub>Leak</sub>	Input leakage current	RSTPD_N = 0.4 V		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{aligned} &DV_{DD} = 3 \ V, \\ &V_{OH} = DV_{DD} - 0.4, \\ &V_{OL} = 0.4, \\ &C_{ou\ t} = 30\ pF \end{aligned}$				16.5	nS

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $DV_{DD}$  - 0.4 V.

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4  $\rm V.$ 

<sup>[3]</sup> The  $I_{OH}$  and  $I_{OL}$  give the output driving capability and allow to calculate directly the rise and fall time as function of the load capacitance.

# 12.11 Input/output pin characteristics for DP and DM

Table 306. Input/output pin characteristics for DP and DM for USB interface

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level input voltage	$PV_{DD} = 3.3 V$		2		3.6	V
$V_{IL}$	Low level input voltage		[1]	0		8.0	V
V <sub>OH</sub>	High level output voltage	$PV_{DD} = 3.3 \text{ V},$ $R_{PD} = 1.5 \Omega \text{ to V}_{SS}$		2.8		$PV_{DD}$	V
$V_{OL}$	Low level output voltage	$PV_{DD} = 3.3 \text{ V},$ $R_{PD} = 1.5 \Omega \text{ to } PV_{DD}$		0		0.3	V
I <sub>OH</sub>	High level output current	$PV_{DD} = 3.3 \text{ V},$ $V_{OH} = 0.8 \times PV_{DD}$	[2]	-4			mA
		$PV_{DD} = 1.8 \text{ V},$ $V_{OH} = 0.7 \times PV_{DD}$		-2			mA
I <sub>OL</sub>	Low level output current	$PV_{DD} = 3.3 \text{ V},$ $V_{OL} = 0.2 \times PV_{DD}$	[2]	4			mA
		$PV_{DD} = 1.8 \text{ V},$ $V_{OL} = 0.3 \times PV_{DD}$		2			mA
I <sub>IH</sub>	High level input current	$V_I = PV_{DD}$				1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$				1	μΑ
I <sub>Leak</sub>	Input leakage current	RSTPD_N = 0 V		-1		1	μΑ
C <sub>in</sub>	Input capacitance				2.5	3.5	pF
Z <sub>inp</sub>	Input Impedance exclusive of pullup/pulldown (for low/full speed)			300			kΩ
$Z_{DRV}$	Driver output resistance			28		44	Ω
T <sub>FDRATE</sub>	Full-speed Data rate			11.97		12.03	Mb/s
T <sub>DJ1</sub>	source Jitter total (including			-3.5		3.5	ns
T <sub>DJ2</sub>	frequency tolerance) to next transition for paired transition			-4		4	ns
T <sub>FDEOP</sub>	Source Jitter for differential transition to SE0 transition			-2		5	ns
T <sub>JR1</sub>	Receiver Jitter: to next			-18.5		18.5	ns
T <sub>JR2</sub>	transition for paired transitions			-9		9	ns
$T_{FEOPT}$	Source SE0 interval of EOP			160		175	ns
$T_{FEOPR}$	Receiver SE0 interval of EOP			82			ns
T <sub>FST</sub>	Width of SE0 interval during differential transition					14	ns

<sup>[1]</sup> The value doesn't guaranty the power down consumptions. To reach the specified power down consumptions, the limit is 0.4 V.

<sup>[2]</sup> The I<sub>OH</sub> and I<sub>OL</sub> give the output driving capability and allow to calculate directly the rise and fall time as function of the load capacitance.

Table 307. USB DP/DM differential receiver input levels

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{DI}$	differential input sensitivity	-	0.2	-	-	V
$V_{CM}$	differential common mode voltage	-	8.0	-	2.5	V

#### Table 308. USB DP/DM driver characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>FR</sub>	rise time	CL = 50  pF; 10% to 90% of $[V_{OH} - V_{OL}]$	4	-	20	ns
t <sub>FF</sub>	fall time	CL = 50  pF; 10% to 90% of $[V_{OH} - V_{OL}]$	4	-	20	ns
F <sub>RFM</sub>	differential rise/fall time matching $(t_{FR}/_{tFF})$	excluding the first transition from Idle state	90	-	111.1	%
V <sub>CRS</sub>	output signal crossover voltage	excluding the first transition from Idle state	1.3	-	2.0	V

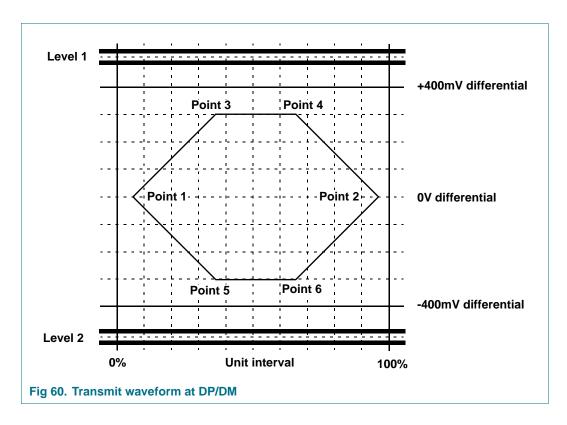


Table 309. Input Pin characteristics for DP for HSU interface

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level Input voltage		[1]	$0.7 \times PV_{DD}$		$PV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage		[2]	0		$0.3 \times PV_{DD}$	V
I <sub>IH</sub>	High level input current	$V_i = PV_{DD}$				1	mA
I <sub>IL</sub>	Low level input current	$V_i = 0 V$				1	mA
I <sub>Leak</sub>	Input Leakage current	RSTPD_N = 0 V		-1		1	mA
C <sub>in</sub>	Input Capacitance				2.5	3.5	pF

<sup>[1]</sup> The value doesn't guaranty the power down consumptions. To reach the specified power down consumptions, the limit is  $PV_{DD}$  - 0.4 V.

Table 310. Output Pin characteristics for DM for HSU interface

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{OH}$	High level output voltage	$PV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$		PV <sub>DD</sub> -0.4	ļ	$PV_{DD}$	V
		$PV_{DD} = 1.8 \text{ V},$ $I_{OH} = -2 \text{ mA}$		PV <sub>DD</sub> -0.4	ļ	$PV_{DD}$	V
$V_{OL}$	Low level output voltage	$PV_{DD} = 3 V$ , $I_{OL} = -4 \text{ mA}$		0		0.4	V
		$PV_{DD} = 1.8 \text{ V},$ $I_{OL} = -2 \text{ mA}$		0		0.4	V
I <sub>OH</sub>	High level output current	$PV_{DD} = 3 V$ , $V_{OH} = 0.8 \times PV_{DD}$	[1]	-4			mA
		$PV_{DD} = 1.8 \text{ V},$ $V_{OH} = 0.7 \times PV_{DD}$		-2			mA
I <sub>OL</sub>	Low level output current	$PV_{DD} = 3.3 \text{ V},$ $V_{OL} = 0.2 \times PV_{DD}$	[1]	4			mA
		$PV_{DD} = 1.8 \text{ V},$ $V_{OL} = 0.3 \times PV_{DD}$		2			mA
I <sub>Leak</sub>	Input leakage current	RSTPD_N = 0 V		-1		1	mA
C <sub>out</sub>	Load Capacitance					30	pF
trise,fall	Rise and fall time	$\begin{aligned} \text{PV}_{\text{DD}} &= 3 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.8 \times \text{PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.2 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$	[1]			13.5	nS
		$\begin{aligned} \text{PV}_{\text{DD}} &= 1.8 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.7 \times \text{PV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.3 \times \text{PV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{aligned}$				10.8	nS

<sup>[1]</sup> The I<sub>OH</sub> and I<sub>OL</sub> give the output driving capability and allow to calculate directly the rise and fall time as function of the load capacitance

<sup>[2]</sup> The value doesn't guaranty the power down consumptions. To reach the specified power down consumptions, the limit is 0.4 V.

### 12.12 Input pin characteristics for SCL

Table 311. Input/output drain output pin characteristics for SCL I<sup>2</sup>C interface

	<u> </u>						
Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level Input voltage		[1]	$0.7 \times PV_{DD}$		$DV_DD$	V
$V_{IL}$	Low level Input voltage		[2]	0		$0.3 \times DV_{DD} \\$	V
$V_{OL}$	Low level output voltage	$DV_{DD} = 3 V$ , $I_{OL} = -4 \text{ mA}$		0		0.3	V
I <sub>IH</sub>	High level input current	$V_I = DV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	RSTPD_N = 0.4 V		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
t <sub>r</sub>	Rise time SCL		[3]	20		300	ns
t <sub>f</sub>	Fall time SCL		[3]	20		300	ns

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $DV_{DD}$  - 0.4 V.

### 12.13 Input/output pin characteristics for SDA

Table 312. Input/output pin characteristics for SDA I<sup>2</sup>C interface

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level Input voltage		[1]	$0.7 \times PV_{DD}$		$DV_DD$	V
$V_{IL}$	Low level Input voltage		[2]	0		$0.3 \times DV_{DD} \\$	V
$V_{OL}$	Low level output voltage	$DV_{DD} = 3 V$ , $I_{OL} = -4 \text{ mA}$		0		0.3	V
I <sub>IH</sub>	High level input current	$V_{I} = DV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF
C <sub>out</sub>	Load Capacitance					30	pF
t <sub>r</sub>	Rise time SDA		[3]	20		300	ns
t <sub>f</sub>	Fall time SDA		[3]	20		300	ns

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $\mathrm{DV}_{\mathrm{DD}}$  - 0.4 V.

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V.

<sup>[3]</sup> The PN533 has a slope control according to the I<sup>2</sup>C specification for the Fast mode. The slope control is always present and not dependant of the I<sup>2</sup>C speed.

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V.

<sup>[3]</sup> The PN533 has a slope control according to the I<sup>2</sup>C specification for the Fast mode. The slope control is always present and not dependant of the I<sup>2</sup>C speed.

### 12.14 Output pin characteristics for Delatt

Table 313. Output pin characteristics for Delatt

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{OH}$	High level Output voltage		<u>[1]</u>	$0.7 \times \text{SV}_\text{DD}$		$SV_{DD}$	V
V <sub>IL</sub>	Low level Input voltage			0		$0.3 \times PV_{DD}$	V
I <sub>IH</sub>	Input mode high level input current	$V_{I} = SV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Input mode low level input current	$V_I = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μА
C <sub>in</sub>	Input Capacitance				2.5		pF

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $PV_{DD}$  - 0.4 V.

### 12.15 Input pin characteristics for SIGIN

Table 314. Input/output pin characteristics for SIGIN

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level Input voltage		[1]	$0.7 \times \text{SV}_{DD}$		$SV_{DD}$	V
$V_{IL}$	Low level Input voltage		[2]	0		$0.3 \times \text{SV}_{DD}$	V
I <sub>IH</sub>	High level input current	$V_I = SV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Low level input current	$V_I = 0 V$		-1		1	μΑ
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		pF

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $SV_{DD}$  - 0.4 V.

### 12.16 Output pin characteristics for SIGOUT

Table 315. Output pin characteristics for SIGOUT

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{OH}$	High level output voltage	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OH}$ = -4 mA	SV <sub>DD</sub> - 0.4		$SV_{DD}$	V
$V_{OL}$	Low level output voltage	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OL}$ = +4 mA	0		0.4	V
I <sub>OH</sub>	High level output current	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OH}$ = -4 mA	-0.4			mA
I <sub>OL</sub>	Low level output current	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OL}$ = +4 mA	4			mA
I <sub>Leak</sub>	Input leakage current	$RSTPD_N = 0.4 V$	-1		1	μΑ
C <sub>in</sub>	Input Capacitance			2.5		pF
C <sub>out</sub>	Load Capacitance				30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{split} \text{SV}_{\text{DD}} &= 3 \text{ V,} \\ \text{V}_{\text{OH}} &= 0.8 \times \text{SV}_{\text{DD}}, \\ \text{V}_{\text{OL}} &= 0.2 \times \text{SV}_{\text{DD}}, \\ \text{C}_{\text{out}} &= 30 \text{ pF} \end{split}$			9	ns

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V.

### 12.17 Input/output pin characteristics for P34

Table 316. Input/output pin characteristics for P34

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
$V_{IH}$	High level input voltage		[1]	$0.7 \times \text{SV}_\text{DD}$		$SV_{DD}$	V
V <sub>IL</sub>	Low level input voltage		[2]	0		$0.3 \times SV_{DD}$	V
V <sub>OH</sub>	Push-pull mode high level output voltage	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OH}$ = -4 mA		SV <sub>DD</sub> - 0.4		SV <sub>DD</sub>	V
V <sub>OL</sub>	Push_pull mode low level output voltage	$DV_{DD} - 0.1 < SV_{DD}$ $< DV_{DD}$ $I_{OH} = +4 \text{ mA}$		0		0.4	V
I <sub>IH</sub>	Input mode high level input current	$V_I = SV_{DD}$		-1		1	μΑ
I <sub>IL</sub>	Input mode low level input current	V <sub>I</sub> = 0 V		-1		1	μΑ
I <sub>OH</sub>	High level output voltage	$DV_{DD}$ - 0.1 < $SV_{DD}$ < $DV_{DD}$ $I_{OH}$ = -4 mA		-0.4			V
I <sub>OL</sub>	Low level output voltage	$DV_{DD} - 0.1 < SV_{DD}$ $< DV_{DD}$ $I_{OL} = +4 \text{ mA}$		4			V
I <sub>Leak</sub>	Input leakage current	RSTPD_N = 0.4 V		-1		1	μΑ
C <sub>in</sub>	Input Capacitance				2.5		рF
C <sub>out</sub>	Load Capacitance					30	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{split} & DV_{DD} = 0.1 < DV_{DD} \\ & V_{OH} = 0.8 \times SV_{DD}, \\ & V_{OL} = 0.2 \times SV_{DD}, \\ & C_{out} = 30 \text{ pF} \end{split}$	[3]		13.5		ns

<sup>[1]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is  $SV_{DD}$  - 0.4 V.

### 12.18 Output pin characteristics for LOADMOD

Table 317. Output pin characteristics for LOADMOD

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{OH}$	High level output voltage	$DV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$	DV <sub>DD</sub> - 0.4		$DV_DD$	V
$V_{OL}$	Low level output voltage	$DV_{DD} = 3 V$ , $I_{OL} = 4 \text{ mA}$	0		0.4	V
$C_{out}$	Load Capacitance				10	pF
t <sub>rise,fall</sub>	Rise and fall times	$\begin{aligned} & DV_{DD} = 3 \; V, \\ & V_{OH} = 0.8 \times DV_{DD}, \\ & V_{OL} \times = 0.2 \times DV_{DD}, \\ & C_{out} = 10 \; pF \end{aligned}$			4.5	ns

<sup>[2]</sup> To minimize power consumption when in Soft-Power-down mode, the limit is 0.4 V.

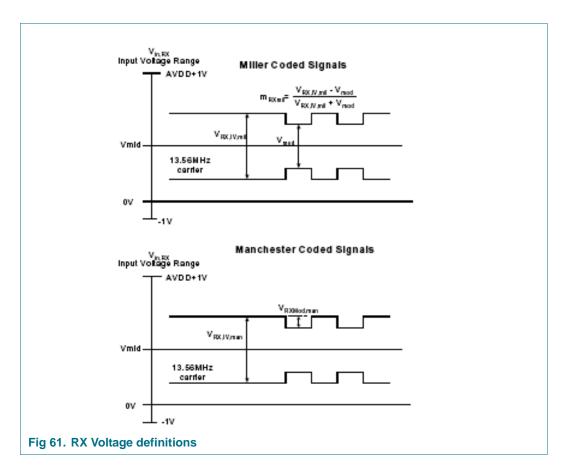
<sup>[3]</sup>  $I_{OH}$  and  $I_{OL}$  specify the output drive capability from which the rise and fall times may be calculated as a function of the load capacitance.

# 12.19 Input pin characteristics for RX

Table 318. Input pin characteristics for RX

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{INRX}$	Dynamic Input voltage Range	signal frequency at 13.56 MHz	-0.7		AV <sub>DD</sub> +1	V
C <sub>inrx</sub>	RX Input Capacitance		6	10	14	pF
R <sub>inrx</sub>	RX Input Series resistance	AV <sub>DD</sub> = 3 V, Receiver active, VRX = 1 Vpp, 1.5 V DC offset	315	350	385	Ω
$V_{RX,MinIV,Mill}$	Minimum Dynamic Input voltage, Miller coded	106 kbit/s		150	500	mVpp
V <sub>RX,MinIV,Man</sub>	Minimum Dynamic Input voltage, Manchester Coded	212 and 424 kbit/s		100	200	mVpp
$V_{RX,MaxIV,Mill}$	Maximum Dynamic Input voltage, Miller coded	106 kbit/s	$AV_{DD}$			Vpp
V <sub>RX,MaxIV,Man</sub>	Maximum Dynamic Input voltage, Manchester Coded	212 and 424 kbit/s	$AV_DD$			Vpp
$V_{mRX,Mill}$	Minimum Modulation index, Miller coded	106 kbit/s VRX = 1.5 Vpp, SensMiller = 3		33		%
V <sub>RXMod,Man</sub>	Minimum modulation voltage	RxGain = 6 and 7	<u>[1]</u>		6	mV
V <sub>RXMod,Man</sub>	Minimum modulation voltage	RxGain = 4 and 5	[1]		18	mV
$V_{RXMod,Man}$	Minimum modulation voltage	RxGain = 0 to 3	[1]		120	mV

<sup>[1]</sup> The minimum modulation voltage is valid for all modulation schemes except Miller coded signals.



### 12.20 Output pin characteristics for AUX1/AUX2

Table 319. Output pin characteristics for AUX1/AUX2

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{OH}$	High level output voltage	$DV_{DD} = 3 V$ , $I_{OH} = -4 \text{ mA}$	DV <sub>DD</sub> - 0.4		$DV_DD$	V
$V_{OL}$	Low level output voltage	$DV_{DD} = 3 V$ , $I_{OL} = 4 \text{ mA}$	$DV_SS$		DV <sub>SS</sub> +0.4	V
I <sub>OH</sub>	High level output current	$DV_{DD} = 3 V,$ $V_{OH} = DV_{DD} -0.3$	-4			mA
I <sub>OL</sub>	Low level output current	$DV_{DD} = 3 \text{ V, } V_{OL} = DV_{DD} - 0.3$	4			mA
I <sub>Leak</sub>	Input leakage current	RSTPD_N = 0 V	-1		1	μΑ
C <sub>in</sub>	Input Capacitance			2.5		pF
$C_{out}$	Load Capacitance				15	pF

# 12.21 Output pin characteristics for TX1/TX2

Table 320. Output pin characteristics for TX1/TX2

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>OH</sub> , C32, 3 V	High level output voltage	$TV_{DD} = 3 \text{ V}$ and $I_{TX} = 32 \text{ mA}$ , $CWGsN = Fh$			150	mV
V <sub>OH</sub> , C80, 3 V	High level output voltage	$TV_{DD} = 3 \text{ V}$ and $I_{TX} = 80 \text{ mA}$ , $CWGsN = Fh$			400	mV
V <sub>OL</sub> , C32, 3 V	Low level output voltage	$TV_{DD} = 2.5V$ and $I_{TX} = 32$ mA, $CWGsN = Fh$			240	mV
V <sub>OL</sub> , C80, 3 V	Low level output voltage	$TV_{DD}$ = 2.5 V and $I_{TX}$ = 80 mA, $CWGsN$ = Fh			640	mV

Table 321. Output resistance for TX1/TX2

$3 \text{ V, V}_{TX} = \text{TV}_{DD} - \text{CWGsP} = 01\text{h}$ $3 \text{ V, V}_{TX} = \text{TV}_{DD} - \text{CWGSP} = 0.00$	133 67	180	251	Ω
,,	67			
CWGsP = 02h	07	90	125	Ω
, ,,,	34	46	62	Ω
, 17	17	23	31	Ω
, IX DD	8.5	12	15.5	Ω
, 17	4.7	6	7.8	Ω
==	2.3	3	4.4	Ω
	34	46	62	Ω
, ,,,	17	23	31	Ω
	8.5	12	15.5	Ω
. 17. 55	4.7	6	7.8	Ω
, ,,,	2.3	3	4.4	Ω
	3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsP = 04h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsP = 08h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsP = 10h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsP = 20h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsP = 3Fh 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsP = 3Fh 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsN = 10h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsN = 20h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsN = 20h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGsN = 80h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGSN = 80h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGSN = 80h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - , CWGSN = F0h	3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 , CWGsP = 04h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 17 , CWGsP = 08h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 8.5 , CWGsP = 10h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 4.7 , CWGsP = 20h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 2.3 , CWGsP = 3Fh 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 , CWGsN = 10h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 17 , CWGsN = 20h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 4.7 , CWGsN = 40h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 4.7 , CWGsN = 40h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 4.7 , CWGsN = 80h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 4.7 , CWGsN = 80h 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 2.3	3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 46 , CWGsP = 04h 17 23 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 17 23 , CWGsP = 08h 12 12 12 12 12 12 12 12 12 12 12 12 12	3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 46 62 CWGsP = 04h 77 23 31 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 17 23 31 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 8.5 12 15.5 CWGsP = 10h 78 7.8 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 4.7 6 7.8 CWGsP = 20h 78 78 46 62 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 46 62 CWGsP = 3Fh 79 31 31 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 46 62 CWGsN = 10h 79 31 31 3 V, V <sub>TX</sub> = TV <sub>DD</sub> - 34 46 62 CWGsN = 20h 30 70 70 70 70 70 70 70 70 70 70 70 70 70

### 12.22 Timing for Reset

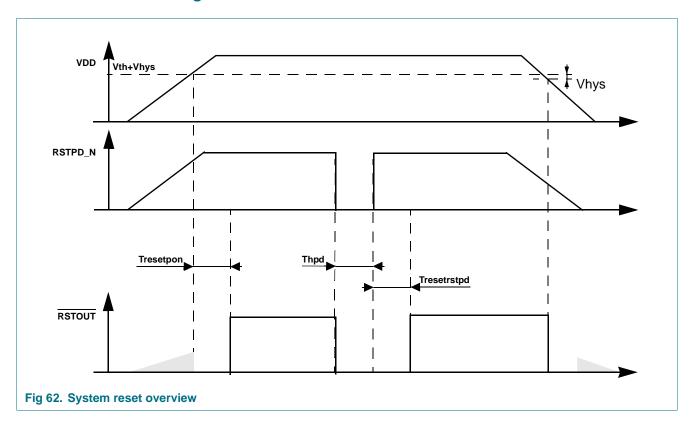


Table 322. Reset duration time

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
T <sub>resetpon</sub>	Reset time at power on		[1] 0.1	0.4	2	ms
$T_{HPD}$	Hard Power-down time	User dependent	<u>[2]</u> 20			ns
T <sub>resetRSTPD_N</sub>	Reset time when RSTPD_N is released		<u>11</u> 0.1	0.4	2	ms

<sup>[1]</sup> Depends on the 27.12 MHz crystal oscillator startup time.

<sup>[2]</sup> In the case  $T_{HPD}$  pulse is shorter than 20 ns, it may happened that the IC would be partially resetted.

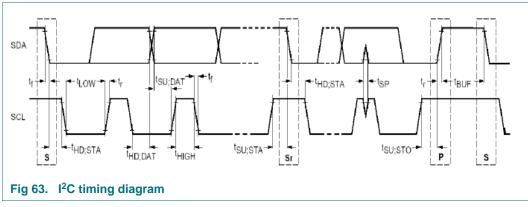
# 12.23 Timing for the I<sup>2</sup>C interface

Table 323. I<sup>2</sup>C timing specification

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
f <sub>SCL</sub>	SCL clock frequency			0		400	kHz
t <sub>HD; STA</sub>	Hold time (repeated) START condition.	After this period, the first clock pulse is generated		600			ns
t <sub>SU; STA</sub>	Set-up time for a repeated START condition			600			ns
t <sub>SU; STO</sub>	Set-up time for STOP condition			600			ns
t <sub>LOW</sub>	LOW period of the P50_SCL clock			1300			ns
t <sub>HIGH</sub>	HIGH period of the P50_SCL clock			600			ns
t <sub>HD; DAT</sub>	Data hold time			0		900	ns
t <sub>SU; DAT</sub>	Data set-up time			100			ns
t <sub>r</sub>	Rise time P50_SCL and SDA		[1]	20		300	ns
t <sub>f</sub>	Fall time P50_SCL and SDA		[1]	20		300	ns
t <sub>BUF</sub>	Bus free time between a STOP and START condition			1.3			ms
tStrWuSpd	Stretching time on P50_SCL when woken-up on its own address		[2]			1	ms
t <sub>HDSDA</sub>	Internal hold time for SDA			330		590	ns
t <sub>HDSDA</sub>	Internal hold time for SDA in SPD mode		[3]		270		ns

<sup>[1]</sup> The PN533 has a slope control according to the I<sup>2</sup>C specification for the Fast mode. The slope control is always present and not dependant of the I<sup>2</sup>C speed.

<sup>[3]</sup> The PN533 has an internal hold time of around 270ns for the SDA signal to bridge the undefined region of the falling edge of P50\_SCL.



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<sup>[2] 27.12</sup> MHz quartz starts in less than 800 μs. For example, quartz like TAS-3225A, TAS-7 or KSS2F with appropriate layout.

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### **Near Field Communication (NFC) controller**

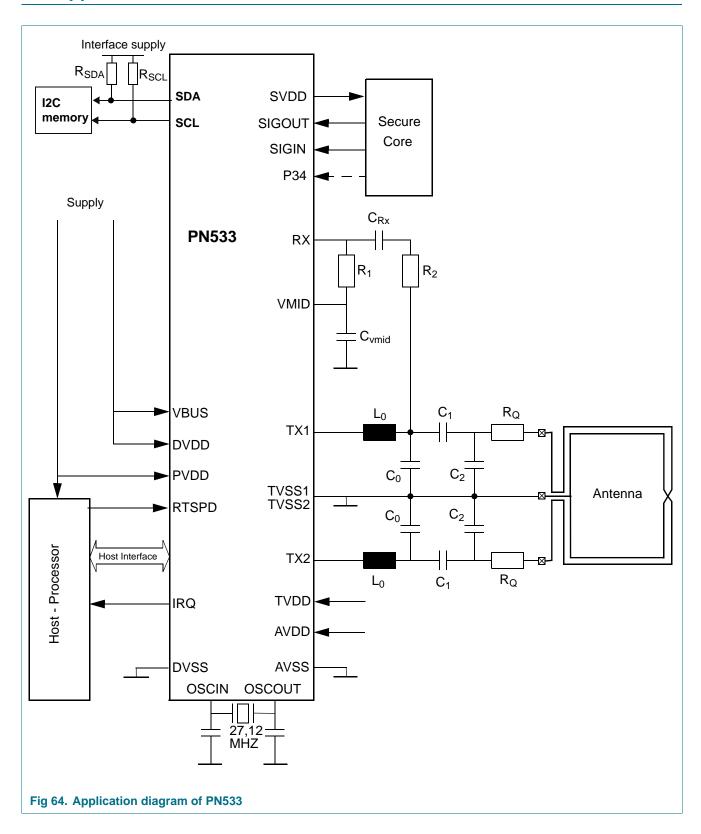
### 12.24 Temperature sensor

Table 324. Temperature sensor characteristics

Symbol	Parameter	Conditions	Mi	n Typ	Max	Unit
T <sub>sens</sub>	CIU Temperature sensor threshold		<u>[1]</u> 10	125	140	°C

<sup>[1]</sup> Temperature sensor embedded in PN533 does not aim to monitor the temperature. It helps to prevent catastrophic failure avoiding destruction of the IC. In any case the application must ensure by other means that ambient temperature do not exceed 85 °C as specified in operating condition table.

# 13. Application information



# 14. Package outline

HVQFN40: plastic thermal enhanced very thin quad flat package; no leads; 40 terminals; body  $6 \times 6 \times 0.85 \text{ mm}$ 

SOT618-1

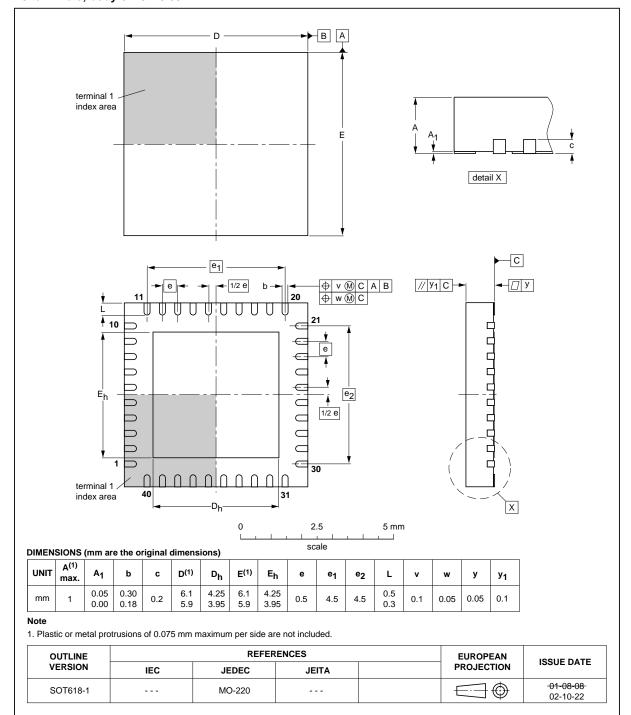


Fig 65. Package outline HVQFN40 (SOT618-1)

This package is MSL level 2.

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# 15. Abbreviations

### **Table 325. Abbreviations**

Acronym	Description
ASK	Amplitude Shift keying
CIU	Contactless interface Unit
HPD	Hard Power Down (see Section 8.5.6 on page 99)
HZ	High Impedance
Initiator	Generates RF field at 13.56 MHz and starts the NFCIP-1 communication.
Load modulation Index	The load modulation index is defined as the card's voltage ratio (Vmax - Vmin) / (Vmax + Vmin) measured at the card's coil.
Modulation Index	The modulation index is defined as the voltage ratio (Vmax - Vmin) / (Vmax + Vmin).
MSL	Moisture Sensitivity Level
PCD	Proximity Coupling Device. Definition for a Card Reader/ Writer according to the ISO/IEC 14443 Specification
PCR	Power Clock Reset controller
PICC	Proximity Cards. Definition for a contactless Smart Card according to the ISO/IEC 14443 specification
SAM	Secure Access Module
SIC	Secure Integrated Circuit (can be a Smart Card IC, a Secure Access Module (SAM),)
SPI	Serial Parallel Interface
SPD	Soft Power Down mode (see Section 8.5.7 on page 99)
Target	Responds to initiator command either using load modulation scheme (RF field generated by Initiator) or using modulation of self-generated RF field (no RF field generated by initiator during target answer).

# 16. Revision history

### Table 326. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes
PN5331B3HN v. 3.3	20171129	Product data sheet	-	PN5331B3HN v. 3.3
Modiifcations:	<ul> <li>Security statu</li> </ul>	is changed into Company	Pubic, no content cha	nge
PN5331B3HN v. 3.3	20120927	Product data sheet	-	157531
Modiifcations:	Section 17.4	"Licenses": updated		
157531	20081210	Product data sheet		Rev. 3.0
Modiifcations:	111 and <u>Table</u> 114:			FARE Reader/Writer" on page 143B Reader/Writer" on page
157530	1 July 2008	Product data sheet		Rev. 1.0
157510	10 April 2008  Initial version	Objective data sheet		

### 17. Legal information

### 17.1 Data sheet status

Document status[1][2]	Product status[3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

- [1] Please consult the most recently issued document before initiating or completing a design.
- [2] The term 'short data sheet' is explained in section "Definitions"
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