Document Number T4240 Rev. 1, 05/2016

QorlQ T4240 Data Sheet

Features

- 12 e6500 cores built on Power Architecture® technology and arranged as clusters of four e6500 cores sharing a 2 MB L2 cache
- 1.5 MB CoreNet platform cache (CPC)
- Hierarchical interconnect fabric
 - CoreNet fabric supporting coherent and noncoherent transactions with prioritization and bandwidth allocation amongst CoreNet end-points
 - 1.6 Tbps coherent read bandwidth
- Three 64-bit DDR3 SDRAM memory controllers
- DDR3 and DDR3L with ECC and interleaving support
- Data Path Acceleration Architecture (DPAA) incorporating acceleration for the following functions:
 - Packet parsing, classification, and distribution (Frame Manager 1.1)
 - Queue management for scheduling, packet sequencing, and congestion management (Queue Manager 1.1)
 - Hardware buffer management for buffer allocation and de-allocation (Buffer Manager 1.1)
 - Cryptography Acceleration (SEC 5.0)
 - RegEx Pattern Matching Acceleration (PME 2.0)
 - Decompression/Compression Acceleration (DCE 1.0)
 - DPAA chip-to-chip interconnect via RapidIO Message Manager (RMan 1.0)

T4240

- 32 SerDes lanes at up to 10 Gb/s
- Ethernet interfaces
 - Up to four 10 Gbps Ethernet MACs
 - Up to sixteen 1 Gbps Ethernet MACs
 - Combinations of 1 Gbps and 10 Gbps Ethernet MACs
 - − IEEE Std 1588TM support
- High-speed peripheral interfaces
 - Four PCI Express 2.0/3.0 controllers running at up to 8 GT/s with one controllers supporting end-point, single-root I/O virtualization (SR-IOV)
 - Two Serial RapidIO 2.0 controllers running at up to 5 Gbaud
 - Interlaken look-aside interface for TCAM connection
- Additional peripheral interfaces
 - Two Serial ATA (SATA 2.0) controllers
 - Two high-speed USB 2.0 controllers with integrated PHY
 - Enhanced secure digital host controller (SD/MMC/ eMMC)
 - Enhanced Serial peripheral interface (eSPI)
 - Four I2C controllers
 - Four 2-pin UARTs or two 4-pin DUARTs
 - Integrated flash controller supporting NAND and NOR flash
- Three 8-channel DMA engines
- 1932 FC-PBGA package, 45 mm x 45 mm, 1mm pitch

NXP reserves the right to change the production detail specifications as may be required to permit improvements in the design of its products.



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Table of Contents

1	Overv	view	
2	Pin as	signments	
	2.1	1932 ball layout diagrams 4	
	2.2	Pinout list1	0
3	Electr	ical characteristics7	3
	3.1	Overall DC electrical characteristics7	3
	3.2	Power sequencing	0
	3.3	Power-down requirements	2
	3.4	Power characteristics	3
	3.5	Power-on ramp rate	0
	3.6	Input clocks	1
	3.7	RESET initialization	6
	3.8	DDR3 and DDR3L SDRAM controller9	7
	3.9	eSPI interface1	03
	3.10	DUART interface1	06
	3.11	Ethernet interface, Ethernet management interface 1 and 2,	
		IEEE Std 1588	07
	3.12	USB interface 1	16
	3.13	Integrated flash controller1	18
	3.14	Enhanced secure digital host controller (eSDHC)1	21
	3.15	Multicore programmable interrupt controller (MPIC)1	23

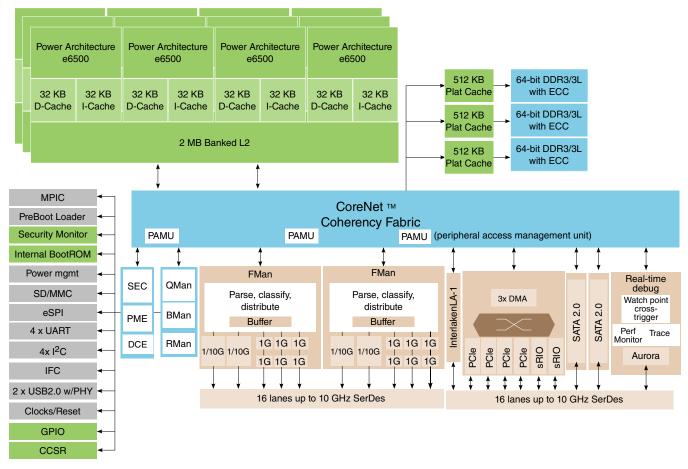
	3.16	JTAG controller124
	3.17	I2C interface
	3.18	GPIO interface
	3.19	High-speed serial interfaces (HSSI)
4	Hardv	vare design considerations188
	4.1	System clocking
	4.2	Power supply design
	4.3	Decoupling recommendations213
	4.4	SerDes block power supply decoupling recommendations213
	4.5	Connection recommendations
	4.6	Thermal
	4.7	Recommended thermal model226
	4.8	Thermal management information
5	Packa	ge information
	5.1	Package parameters for the FC-PBGA229
	5.2	Mechanical dimensions of the FC-PBGA229
6	Secur	ity fuse processor
7	Order	ing information231
	7.1	Part numbering nomenclature231
	7.2	Orderable part numbers addressed by this document232
8	Revis	ion history234

1 Overview

The T4240 QorIQ integrated multicore communications processor combines 12 dualthreaded cores built on Power Architecture® technology with high-performance data path acceleration and network and peripheral bus interfaces required for networking, telecom/ datacom, wireless infrastructure, and military/aerospace applications.

This chip can be used for combined control, data path, and application layer processing in routers, switches, gateways, and general-purpose embedded computing systems. Its high level of integration offers significant performance benefits compared to multiple discrete devices, while also simplifying board design.

This figure shows the block diagram of the chip.





2 Pin assignments

2.1 1932 ball layout diagrams

This table shows the complete view of the T4240 ball map. Figure 3, Figure 4, Figure 5, and Figure 6 show quadrant views of the ballmap.

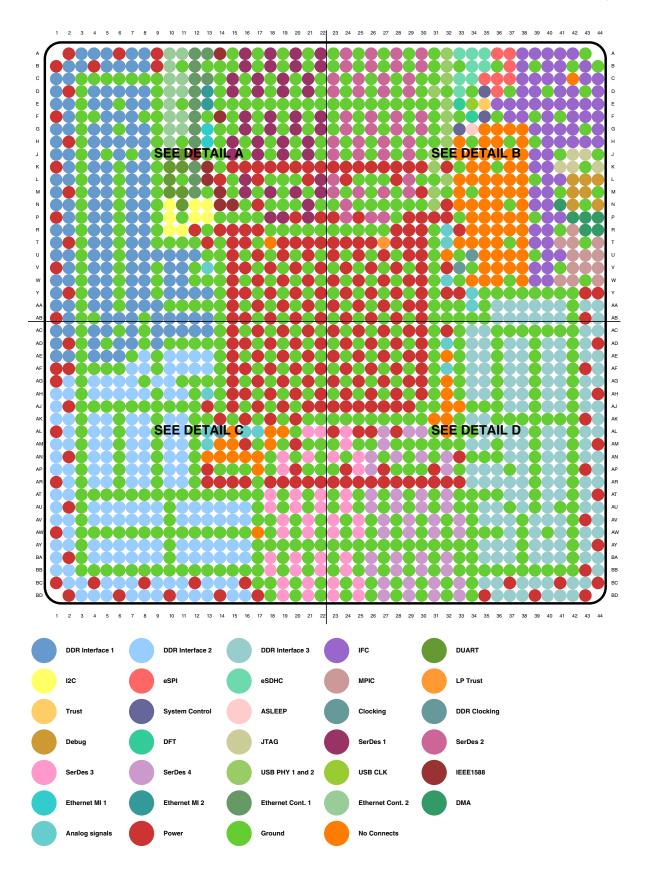


Figure 2. Complete BGA Map for the T4240

Pin assignments

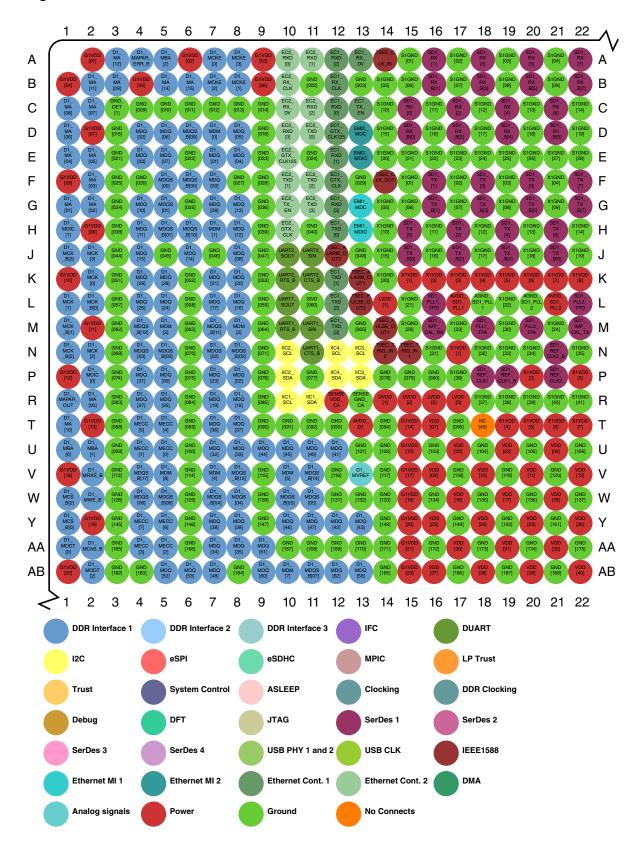


Figure 3. Detail A

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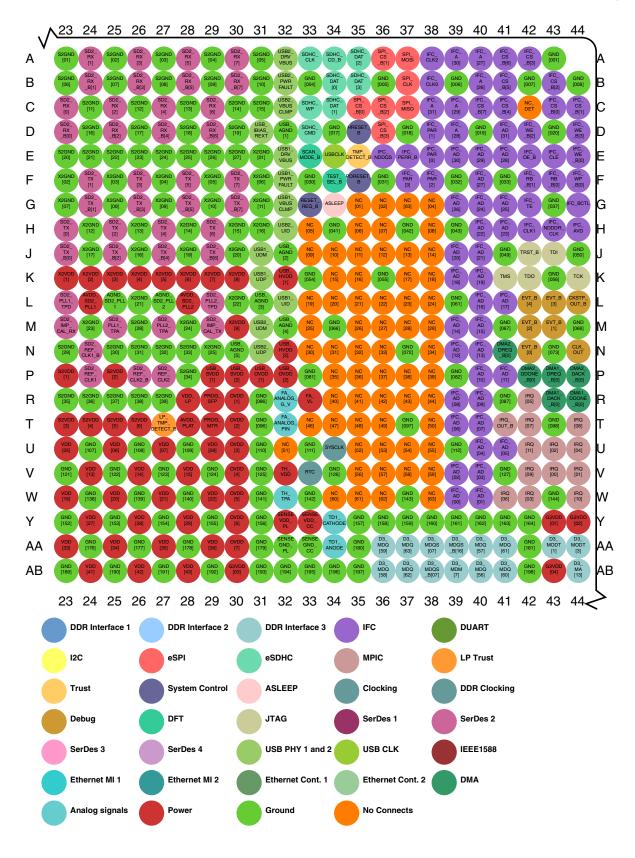


Figure 4. Detail B

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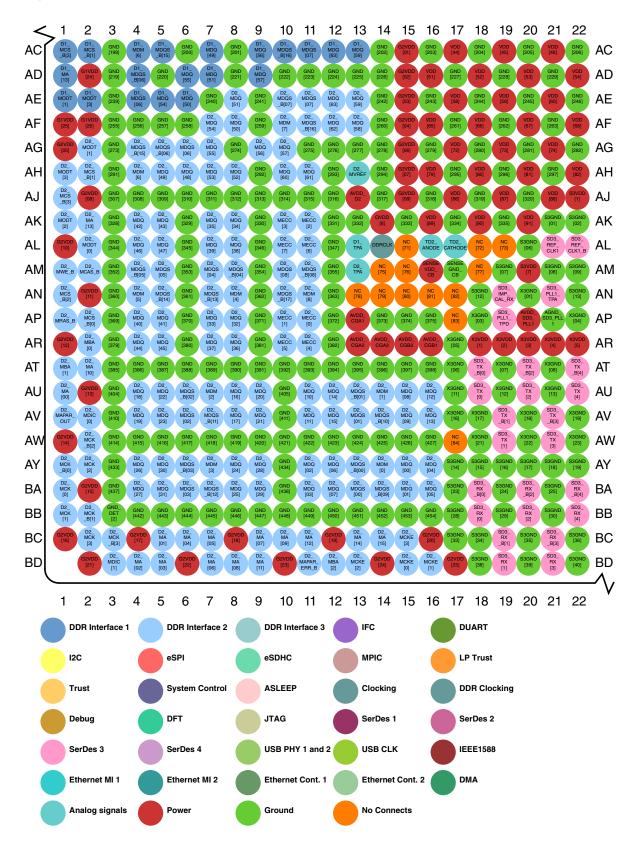


Figure 5. Detail C

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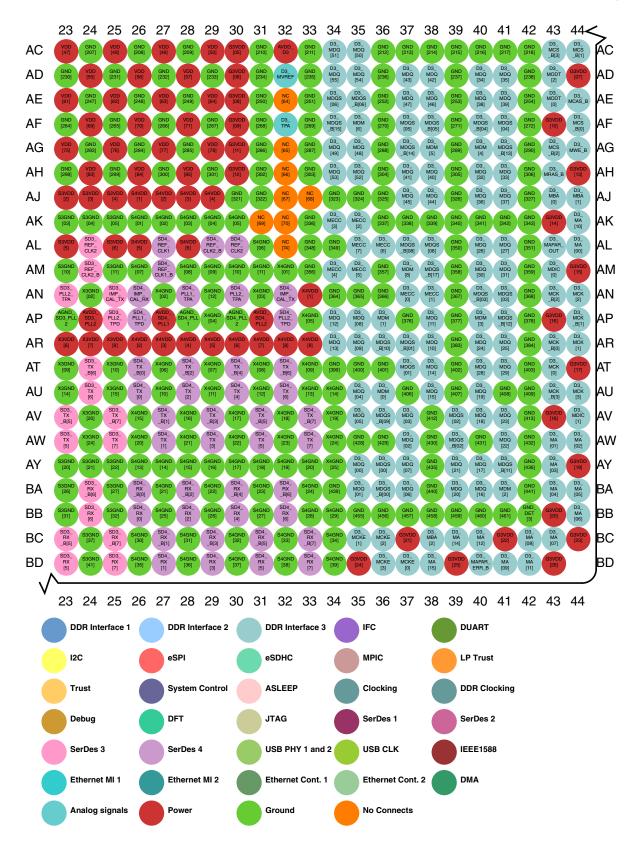


Figure 6. Detail D

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2.2 Pinout list

This table provides the pinout listing for the T4240 by bus. Primary functions are **bolded** in the table.

Signal	Signal description	Package pin number	Pin type	Power supply	Notes			
DDR SDRAM Memory Interface 1								
D1_MA00	Address	R2	0	G1V _{DD}				
D1_MA01	Address	G1	0	G1V _{DD}				
D1_MA02	Address	G2	0	G1V _{DD}				
D1_MA03	Address	F2	0	G1V _{DD}				
D1_MA04	Address	E1	0	G1V _{DD}				
D1_MA05	Address	E2	0	G1V _{DD}				
D1_MA06	Address	D1	0	G1V _{DD}				
D1_MA07	Address	C2	0	G1V _{DD}				
D1_MA08	Address	C1	0	G1V _{DD}				
D1_MA09	Address	B3	0	G1V _{DD}				
D1_MA10	Address	T1	0	G1V _{DD}				
D1_MA11	Address	B2	0	G1V _{DD}				
D1_MA12	Address	A3	0	G1V _{DD}				
D1_MA13	Address	AD1	0	G1V _{DD}				
D1_MA14	Address	B5	0	G1V _{DD}				
D1_MA15	Address	B6	0	G1V _{DD}				
D1_MAPAR_ERR_B	Address Parity Error	A4	I	G1V _{DD}	1, 18			
D1_MAPAR_OUT	Address Parity Out	R1	0	G1V _{DD}				
D1_MBA0	Bank Select	U1	0	G1V _{DD}				
D1_MBA1	Bank Select	U2	0	G1V _{DD}				
D1_MBA2	Bank Select	A5	0	G1V _{DD}				
D1_MCAS_B	Column Address Strobe	AA2	0	G1V _{DD}				
D1_MCK0	Clock	K2	0	G1V _{DD}				
D1_MCK0_B	Clock Complements	L2	0	G1V _{DD}				
D1_MCK1	Clock	L1	0	G1V _{DD}				
D1_MCK1_B	Clock Complements	M1	0	G1V _{DD}				
D1_MCK2	Clock	N2	0	G1V _{DD}				
D1_MCK2_B	Clock Complements	N1	0	G1V _{DD}				
D1_MCK3	Clock	J2	0	G1V _{DD}				
D1_MCK3_B	Clock Complements	J1	0	G1V _{DD}				
D1_MCKE0	Clock Enable	A7	0	G1V _{DD}	2			

Table 1. Pinout list by bus

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D1_MCKE1	Clock Enable	B8	0	G1V _{DD}	2
D1_MCKE2	Clock Enable	B7	0	G1V _{DD}	2
D1_MCKE3	Clock Enable	A8	0	G1V _{DD}	2
D1_MCS0_B	Chip Select	Y1	0	G1V _{DD}	
D1_MCS1_B	Chip Select	AC2	0	G1V _{DD}	
D1_MCS2_B	Chip Select	W1	0	G1V _{DD}	
D1_MCS3_B	Chip Select	AC1	0	G1V _{DD}	
D1_MDIC0	Driver Impedence Calibration	P2	IO	G1V _{DD}	3
D1_MDIC1	Driver Impedence Calibration	H1	IO	G1V _{DD}	3
D1_MDM0/D1_MDQS09	Data Mask	D7	0	G1V _{DD}	1
D1_MDM1/D1_MDQS10	Data Mask	H7	0	G1V _{DD}	1
D1_MDM2/D1_MDQS11	Data Mask	M8	0	G1V _{DD}	1
D1_MDM3/D1_MDQS12	Data Mask	M5	0	G1V _{DD}	1
D1_MDM4/D1_MDQS13	Data Mask	V7	0	G1V _{DD}	1
D1_MDM5/D1_MDQS14	Data Mask	V10	0	G1V _{DD}	1
D1_MDM6/D1_MDQS15	Data Mask	AC4	0	G1V _{DD}	1
D1_MDM7/D1_MDQS16	Data Mask	AB10	0	G1V _{DD}	1
D1_MDM8/D1_MDQS17	Data Mask	V5	0	G1V _{DD}	1
D1_MDQ00	Data	F7	IO	G1V _{DD}	
D1_MDQ01	Data	E7	IO	G1V _{DD}	
D1_MDQ02	Data	D4	IO	G1V _{DD}	
D1_MDQ03	Data	E4	IO	G1V _{DD}	
D1_MDQ04	Data	E8	10	G1V _{DD}	
D1_MDQ05	Data	D8	IO	G1V _{DD}	
D1_MDQ06	Data	D5	10	G1V _{DD}	
D1_MDQ07	Data	E5	10	G1V _{DD}	
D1_MDQ08	Data	J8	IO	G1V _{DD}	
D1_MDQ09	Data	G7	IO	G1V _{DD}	
D1_MDQ10	Data	G4	IO	G1V _{DD}	
D1_MDQ11	Data	H4	IO	G1V _{DD}	
D1_MDQ12	Data	H8	IO	G1V _{DD}	
D1_MDQ13	Data	G8	IO	G1V _{DD}	
D1_MDQ14	Data	J6	IO	G1V _{DD}	
D1_MDQ15	Data	J4	IO	G1V _{DD}	
D1_MDQ16	Data	L8	IO	G1V _{DD}	
D1_MDQ17	Data	L7	IO	G1V _{DD}	
D1_MDQ18	Data	R8	IO	G1V _{DD}	
D1_MDQ19	Data	R7	IO	G1V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
D1_MDQ20	Data	K8	10	G1V _{DD}	
D1_MDQ21	Data	K7	IO	G1V _{DD}	
D1_MDQ22	Data	P8	IO	G1V _{DD}	
D1_MDQ23	Data	P7	IO	G1V _{DD}	
D1_MDQ24	Data	L5	Ю	G1V _{DD}	
D1_MDQ25	Data	L4	Ю	G1V _{DD}	
D1_MDQ26	Data	R5	Ю	G1V _{DD}	
D1_MDQ27	Data	R4	IO	G1V _{DD}	
D1_MDQ28	Data	K5	Ю	G1V _{DD}	
D1_MDQ29	Data	K4	Ю	G1V _{DD}	
D1_MDQ30	Data	P5	IO	G1V _{DD}	
D1_MDQ31	Data	P4	IO	G1V _{DD}	
D1_MDQ32	Data	U7	10	G1V _{DD}	
D1_MDQ33	Data	U8	Ю	G1V _{DD}	
D1_MDQ34	Data	AA7	IO	G1V _{DD}	
D1_MDQ35	Data	AA8	10	G1V _{DD}	
D1_MDQ36	Data	T7	Ю	G1V _{DD}	
D1_MDQ37	Data	Т8	10	G1V _{DD}	
D1_MDQ38	Data	Y7	Ю	G1V _{DD}	
D1_MDQ39	Data	Y8	Ю	G1V _{DD}	
D1_MDQ40	Data	U11	Ю	G1V _{DD}	
D1_MDQ41	Data	U12	Ю	G1V _{DD}	
D1_MDQ42	Data	Y12	Ю	G1V _{DD}	
D1_MDQ43	Data	Y13	10	G1V _{DD}	
D1_MDQ44	Data	U9	Ю	G1V _{DD}	
D1_MDQ45	Data	U10	Ю	G1V _{DD}	
D1_MDQ46	Data	Y10	Ю	G1V _{DD}	
D1_MDQ47	Data	Y11	Ю	G1V _{DD}	
D1_MDQ48	Data	AB7	Ю	G1V _{DD}	
D1_MDQ49	Data	AC7	Ю	G1V _{DD}	
D1_MDQ50	Data	AE6	Ю	G1V _{DD}	
D1_MDQ51	Data	AD7	Ю	G1V _{DD}	
D1_MDQ52	Data	AB5	Ю	G1V _{DD}	
D1_MDQ53	Data	AB6	Ю	G1V _{DD}	
D1_MDQ54	Data	AE5	Ю	G1V _{DD}	
D1_MDQ55	Data	AD6	Ю	G1V _{DD}	
D1_MDQ56	Data	AC9	Ю	G1V _{DD}	
D1_MDQ57	Data	AD9	Ю	G1V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D1_MDQ58	Data	AB13	Ю	G1V _{DD}	
D1_MDQ59	Data	AC13	Ю	G1V _{DD}	
D1_MDQ60	Data	AB9	Ю	G1V _{DD}	
D1_MDQ61	Data	AA9	Ю	G1V _{DD}	
D1_MDQ62	Data	AB12	Ю	G1V _{DD}	
D1_MDQ63	Data	AC12	10	G1V _{DD}	
D1_MDQS00	Data Strobe	F5	10	G1V _{DD}	
D1_MDQS00_B	Data Strobe	F6	Ю	G1V _{DD}	
D1_MDQS01	Data Strobe	G5	Ю	G1V _{DD}	
D1_MDQS01_B	Data Strobe	H5	Ю	G1V _{DD}	
D1_MDQS02	Data Strobe	N7	Ю	G1V _{DD}	
D1_MDQS02_B	Data Strobe	N8	Ю	G1V _{DD}	
D1_MDQS03	Data Strobe	N4	10	G1V _{DD}	
D1_MDQS03_B	Data Strobe	N5	Ю	G1V _{DD}	
D1_MDQS04	Data Strobe	W8	Ю	G1V _{DD}	
D1_MDQS04_B	Data Strobe	W7	Ю	G1V _{DD}	
D1_MDQS05	Data Strobe	W11	10	G1V _{DD}	
D1_MDQS05_B	Data Strobe	W10	Ю	G1V _{DD}	
D1_MDQS06	Data Strobe	AE4	Ю	G1V _{DD}	
D1_MDQS06_B	Data Strobe	AD4	Ю	G1V _{DD}	
D1_MDQS07	Data Strobe	AC11	Ю	G1V _{DD}	
D1_MDQS07_B	Data Strobe	AB11	Ю	G1V _{DD}	
D1_MDQS08	Data Strobe	W4	Ю	G1V _{DD}	
D1_MDQS08_B	Data Strobe	W5	Ю	G1V _{DD}	
D1_MDQS09/D1_MDM0	Data Strobe (x4 support)	D7	Ю	G1V _{DD}	
D1_MDQS09_B	Data Strobe (x4 support)	D6	Ю	G1V _{DD}	28
D1_MDQS10/D1_MDM1	Data Strobe (x4 support)	H7	Ю	G1V _{DD}	
D1_MDQS10_B	Data Strobe (x4 support)	H6	Ю	G1V _{DD}	28
D1_MDQS11/D1_MDM2	Data Strobe (x4 support)	M8	10	G1V _{DD}	
D1_MDQS11_B	Data Strobe (x4 support)	M7	Ю	G1V _{DD}	28
D1_MDQS12/D1_MDM3	Data Strobe (x4 support)	M5	Ю	G1V _{DD}	
D1_MDQS12_B	Data Strobe (x4 support)	M4	Ю	G1V _{DD}	28
D1_MDQS13/D1_MDM4	Data Strobe (x4 support)	V7	Ю	G1V _{DD}	
D1_MDQS13_B	Data Strobe (x4 support)	V8	Ю	G1V _{DD}	28
D1_MDQS14/D1_MDM5	Data Strobe (x4 support)	V10	Ю	G1V _{DD}	
D1_MDQS14_B	Data Strobe (x4 support)	V11	10	G1V _{DD}	28
D1_MDQS15/D1_MDM6	Data Strobe (x4 support)	AC4	Ю	G1V _{DD}	
D1_MDQS15_B	Data Strobe (x4 support)	AC5	Ю	G1V _{DD}	28

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D1_MDQS16/ D1_MDM7	Data Strobe (x4 support)	AB10	10	G1V _{DD}	
D1_MDQS16_B	Data Strobe (x4 support)	AC10	10	G1V _{DD}	28
D1_MDQS17/D1_MDM8	Data Strobe (x4 support)	V5	10	G1V _{DD}	
D1_MDQS17_B	Data Strobe (x4 support)	V4	10	G1V _{DD}	28
D1_MECC0	Error Correcting Code	U5	10	G1V _{DD}	
D1_MECC1	Error Correcting Code	U4	10	G1V _{DD}	
D1_MECC2	Error Correcting Code	AA5	10	G1V _{DD}	
D1_MECC3	Error Correcting Code	AA4	10	G1V _{DD}	
D1_MECC4	Error Correcting Code	T5	10	G1V _{DD}	
D1_MECC5	Error Correcting Code	T4	10	G1V _{DD}	
D1_MECC6	Error Correcting Code	Y5	10	G1V _{DD}	
D1_MECC7	Error Correcting Code	Y4	10	G1V _{DD}	
D1_MODT0	On Die Termination	AA1	0	G1V _{DD}	2
D1_MODT1	On Die Termination	AE1	0	G1V _{DD}	2
D1_MODT2	On Die Termination	AB2	0	G1V _{DD}	2
D1_MODT3	On Die Termination	AE2	0	G1V _{DD}	2
D1_MRAS_B	Row Address Strobe	V2	0	G1V _{DD}	
D1_MWE_B	Write Enable	W2	0	G1V _{DD}	
	DDR SDRAM Memor	y Interface 2	<u>)</u>		
D2_MA00	Address	AU1	0	G2V _{DD}	
D2_MA01	Address	BC5	0	G2V _{DD}	
D2_MA02	Address	BD4	0	G2V _{DD}	
D2_MA03	Address	BD5	0	G2V _{DD}	
D2_MA04	Address	BC6	0	G2V _{DD}	
D2_MA05	Address	BC7	0	G2V _{DD}	
D2_MA06	Address	BD7	0	G2V _{DD}	
D2_MA07	Address	BC9	0	G2V _{DD}	
D2_MA08	Address	BD8	0	G2V _{DD}	
D2_MA09	Address	BC10	0	G2V _{DD}	
D2_MA10	Address	AT2	0	G2V _{DD}	
D2_MA11	Address	BD9	0	G2V _{DD}	
D2_MA12	Address	BC11	0	G2V _{DD}	
D2_MA13	Address	AK2	0	G2V _{DD}	
D2_MA14	Address	BC13	0	G2V _{DD}	
D2_MA15	Address	BC14	0	G2V _{DD}	
D2_MAPAR_ERR_B	Address Parity Error	BD11	Ι	G2V _{DD}	1, 18
D2_MAPAR_OUT	Address Parity Out	AV1	0	G2V _{DD}	
D2_MBA0	Bank Select	AR2	0	G2V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D2_MBA1	Bank Select	AT1	0	G2V _{DD}	
D2_MBA2	Bank Select	BD12	0	G2V _{DD}	
D2_MCAS_B	Column Address Strobe	AM2	0	G2V _{DD}	
D2_MCK0	Clock	BA1	0	G2V _{DD}	
D2_MCK0_B	Clock Complements	AY1	0	G2V _{DD}	
D2_MCK1	Clock	BB1	0	G2V _{DD}	
D2_MCK1_B	Clock Complements	BB2	0	G2V _{DD}	
D2_MCK2	Clock	AY2	0	G2V _{DD}	
D2_MCK2_B	Clock Complements	AW2	0	G2V _{DD}	
D2_MCK3	Clock	BC2	0	G2V _{DD}	
D2_MCK3_B	Clock Complements	BC3	0	G2V _{DD}	
D2_MCKE0	Clock Enable	BD15	0	G2V _{DD}	2
D2_MCKE1	Clock Enable	BD16	0	G2V _{DD}	2
D2_MCKE2	Clock Enable	BD13	0	G2V _{DD}	2
D2_MCKE3	Clock Enable	BC15	0	G2V _{DD}	2
D2_MCS0_B	Chip Select	AP2	0	G2V _{DD}	
D2_MCS1_B	Chip Select	AH2	0	G2V _{DD}	
D2_MCS2_B	Chip Select	AN1	0	G2V _{DD}	
D2_MCS3_B	Chip Select	AJ1	0	G2V _{DD}	
D2_MDIC0	Driver Impedence Calibration	AV2	IO	G2V _{DD}	3
D2_MDIC1	Driver Impedence Calibration	BD3	10	G2V _{DD}	3
D2_MDM0/D2_MDQS09	Data Mask	AY14	0	G2V _{DD}	1
D2_MDM1/D2_MDQS10	Data Mask	AU14	0	G2V _{DD}	1
D2_MDM2/D2_MDQS11	Data Mask	AU7	0	G2V _{DD}	1
D2_MDM3/D2_MDQS12	Data Mask	AY7	0	G2V _{DD}	1
D2_MDM4/D2_MDQS13	Data Mask	AN8	0	G2V _{DD}	1
D2_MDM5/D2_MDQS14	Data Mask	AN4	0	G2V _{DD}	1
D2_MDM6/D2_MDQS15	Data Mask	AH4	0	G2V _{DD}	1
D2_MDM7/D2_MDQS16	Data Mask	AF10	0	G2V _{DD}	1
D2_MDM8/D2_MDQS17	Data Mask	AN11	0	G2V _{DD}	1
D2_MDQ00	Data	AY15	IO	G2V _{DD}	
D2_MDQ01	Data	BA15	IO	G2V _{DD}	
D2_MDQ02	Data	AY11	IO	G2V _{DD}	
D2_MDQ03	Data	BA11	IO	G2V _{DD}	
D2_MDQ04	Data	AY16	IO	G2V _{DD}	
D2_MDQ05	Data	BA16	IO	G2V _{DD}	
D2_MDQ06	Data	AY12	IO	G2V _{DD}	
D2_MDQ07	Data	BA12	IO	G2V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
D2_MDQ08	Data	AU15	10	G2V _{DD}	
D2_MDQ09	Data	AV15	IO	G2V _{DD}	
D2_MDQ10	Data	AU11	IO	G2V _{DD}	
D2_MDQ11	Data	AV11	10	G2V _{DD}	
D2_MDQ12	Data	AU16	Ю	G2V _{DD}	
D2_MDQ13	Data	AV16	10	G2V _{DD}	
D2_MDQ14	Data	AU12	Ю	G2V _{DD}	
D2_MDQ15	Data	AV12	Ю	G2V _{DD}	
D2_MDQ16	Data	AU8	ю	G2V _{DD}	
D2_MDQ17	Data	AV8	Ю	G2V _{DD}	
D2_MDQ18	Data	AU4	IO	G2V _{DD}	
D2_MDQ19	Data	AV4	IO	G2V _{DD}	
D2_MDQ20	Data	AU9	Ю	G2V _{DD}	
D2_MDQ21	Data	AV9	IO	G2V _{DD}	
D2_MDQ22	Data	AU5	IO	G2V _{DD}	
D2_MDQ23	Data	AV5	10	G2V _{DD}	
D2_MDQ24	Data	AY8	Ю	G2V _{DD}	
D2_MDQ25	Data	BA8	10	G2V _{DD}	
D2_MDQ26	Data	AY4	10	G2V _{DD}	
D2_MDQ27	Data	BA4	Ю	G2V _{DD}	
D2_MDQ28	Data	AY9	10	G2V _{DD}	
D2_MDQ29	Data	BA9	Ю	G2V _{DD}	
D2_MDQ30	Data	AY5	Ю	G2V _{DD}	
D2_MDQ31	Data	BA5	Ю	G2V _{DD}	
D2_MDQ32	Data	AP8	10	G2V _{DD}	
D2_MDQ33	Data	AP7	Ю	G2V _{DD}	
D2_MDQ34	Data	AK8	Ю	G2V _{DD}	
D2_MDQ35	Data	AK7	Ю	G2V _{DD}	
D2_MDQ36	Data	AR8	Ю	G2V _{DD}	
D2_MDQ37	Data	AR7	Ю	G2V _{DD}	
D2_MDQ38	Data	AL8	Ю	G2V _{DD}	
D2_MDQ39	Data	AL7	Ю	G2V _{DD}	
D2_MDQ40	Data	AP4	Ю	G2V _{DD}	
D2_MDQ41	Data	AP5	Ю	G2V _{DD}	
D2_MDQ42	Data	AK4	Ю	G2V _{DD}	
D2_MDQ43	Data	AK5	Ю	G2V _{DD}	
D2_MDQ44	Data	AR4	Ю	G2V _{DD}	
D2_MDQ45	Data	AR5	Ю	G2V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D2_MDQ46	Data	AL4	IO	G2V _{DD}	
D2_MDQ47	Data	AL5	IO	G2V _{DD}	
D2_MDQ48	Data	AH6	IO	G2V _{DD}	
D2_MDQ49	Data	AH5	IO	G2V _{DD}	
D2_MDQ50	Data	AF8	IO	G2V _{DD}	
D2_MDQ51	Data	AE8	IO	G2V _{DD}	
D2_MDQ52	Data	AH8	IO	G2V _{DD}	
D2_MDQ53	Data	AH7	IO	G2V _{DD}	
D2_MDQ54	Data	AF7	10	G2V _{DD}	
D2_MDQ55	Data	AG7	IO	G2V _{DD}	
D2_MDQ56	Data	AG9	IO	G2V _{DD}	
D2_MDQ57	Data	AG10	IO	G2V _{DD}	
D2_MDQ58	Data	AF13	IO	G2V _{DD}	
D2_MDQ59	Data	AE13	IO	G2V _{DD}	
D2_MDQ60	Data	AH10	IO	G2V _{DD}	
D2_MDQ61	Data	AH11	IO	G2V _{DD}	
D2_MDQ62	Data	AF12	Ю	G2V _{DD}	
D2_MDQ63	Data	AE12	IO	G2V _{DD}	
D2_MDQS00	Data Strobe	BA13	IO	G2V _{DD}	
D2_MDQS00_B	Data Strobe	AY13	IO	G2V _{DD}	
D2_MDQS01	Data Strobe	AV13	IO	G2V _{DD}	
D2_MDQS01_B	Data Strobe	AU13	IO	G2V _{DD}	
D2_MDQS02	Data Strobe	AV6	IO	G2V _{DD}	
D2_MDQS02_B	Data Strobe	AU6	IO	G2V _{DD}	
D2_MDQS03	Data Strobe	BA6	IO	G2V _{DD}	
D2_MDQS03_B	Data Strobe	AY6	IO	G2V _{DD}	
D2_MDQS04	Data Strobe	AM7	IO	G2V _{DD}	
D2_MDQS04_B	Data Strobe	AM8	10	G2V _{DD}	
D2_MDQS05	Data Strobe	AM5	10	G2V _{DD}	
D2_MDQS05_B	Data Strobe	AM4	10	G2V _{DD}	
D2_MDQS06	Data Strobe	AG6	10	G2V _{DD}	
D2_MDQS06_B	Data Strobe	AG5	IO	G2V _{DD}	
D2_MDQS07	Data Strobe	AE11	IO	G2V _{DD}	
D2_MDQS07_B	Data Strobe	AE10	IO	G2V _{DD}	
D2_MDQS08	Data Strobe	AM10	IO	G2V _{DD}	
D2_MDQS08_B	Data Strobe	AM11	IO	G2V _{DD}	
D2_MDQS09/D2_MDM0	Data Strobe (x4 support)	AY14	IO	G2V _{DD}	
D2_MDQS09_B	Data Strobe (x4 support)	BA14	IO	G2V _{DD}	28

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D2_MDQS10/ D2_MDM1	Data Strobe (x4 support)	AU14	IO	G2V _{DD}	
D2_MDQS10_B	Data Strobe (x4 support)	AV14	IO	G2V _{DD}	28
D2_MDQS11/D2_MDM2	Data Strobe (x4 support)	AU7	IO	G2V _{DD}	
D2_MDQS11_B	Data Strobe (x4 support)	AV7	IO	G2V _{DD}	28
D2_MDQS12/D2_MDM3	Data Strobe (x4 support)	AY7	IO	G2V _{DD}	
D2_MDQS12_B	Data Strobe (x4 support)	BA7	10	G2V _{DD}	28
D2_MDQS13/ D2_MDM4	Data Strobe (x4 support)	AN8	10	G2V _{DD}	
D2_MDQS13_B	Data Strobe (x4 support)	AN7	10	G2V _{DD}	28
D2_MDQS14/ D2_MDM5	Data Strobe (x4 support)	AN4	10	G2V _{DD}	
D2_MDQS14_B	Data Strobe (x4 support)	AN5	IO	G2V _{DD}	28
D2_MDQS15/ D2_MDM6	Data Strobe (x4 support)	AH4	IO	G2V _{DD}	
D2_MDQS15_B	Data Strobe (x4 support)	AG4	IO	G2V _{DD}	28
D2_MDQS16/ D2_MDM7	Data Strobe (x4 support)	AF10	IO	G2V _{DD}	
D2_MDQS16_B	Data Strobe (x4 support)	AF11	IO	G2V _{DD}	28
D2_MDQS17/ D2_MDM8	Data Strobe (x4 support)	AN11	10	G2V _{DD}	
D2_MDQS17_B	Data Strobe (x4 support)	AN10	10	G2V _{DD}	28
D2_MECC0	Error Correcting Code	AP11	10	G2V _{DD}	
D2_MECC1	Error Correcting Code	AP10	IO	G2V _{DD}	
D2_MECC2	Error Correcting Code	AK11	10	G2V _{DD}	
D2_MECC3	Error Correcting Code	AK10	IO	G2V _{DD}	
D2_MECC4	Error Correcting Code	AR11	10	G2V _{DD}	
D2_MECC5	Error Correcting Code	AR10	IO	G2V _{DD}	
D2_MECC6	Error Correcting Code	AL11	IO	G2V _{DD}	
D2_MECC7	Error Correcting Code	AL10	IO	G2V _{DD}	
D2_MODT0	On Die Termination	AL2	0	G2V _{DD}	2
D2_MODT1	On Die Termination	AG2	0	G2V _{DD}	2
D2_MODT2	On Die Termination	AK1	0	G2V _{DD}	2
D2_MODT3	On Die Termination	AH1	0	G2V _{DD}	2
D2_MRAS_B	Row Address Strobe	AP1	0	G2V _{DD}	
D2_MWE_B	Write Enable	AM1	0	G2V _{DD}	
	DDR SDRAM Memo	ry Interface 3	3		
D3_MA00	Address	AL44	0	G3V _{DD}	
D3_MA01	Address	AW43	0	G3V _{DD}	
D3_MA02	Address	AW44	0	G3V _{DD}	
D3_MA03	Address	AY43	0	G3V _{DD}	
D3_MA04	Address	BA43	0	G3V _{DD}	
D3_MA05	Address	BA44	0	G3V _{DD}	
D3_MA06	Address	BB44	0	G3V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D3_MA07	Address	BC43	0	G3V _{DD}	
D3_MA08	Address	BC42	0	G3V _{DD}	
D3_MA09	Address	BD41	0	G3V _{DD}	
D3_MA10	Address	AK44	0	G3V _{DD}	
D3_MA11	Address	BD42	0	G3V _{DD}	
D3_MA12	Address	BC40	0	G3V _{DD}	
D3_MA13	Address	AB44	0	G3V _{DD}	
D3_MA14	Address	BC39	0	G3V _{DD}	
D3_MA15	Address	BD38	0	G3V _{DD}	
D3_MAPAR_ERR_B	Address Parity Error	BD40	I	G3V _{DD}	1, 18
D3_MAPAR_OUT	Address Parity Out	AL43	0	G3V _{DD}	
D3_MBA0	Bank Select	AJ43	0	G3V _{DD}	
D3_MBA1	Bank Select	AJ44	0	G3V _{DD}	
D3_MBA2	Bank Select	BC38	0	G3V _{DD}	
D3_MCAS_B	Column Address Strobe	AE44	0	G3V _{DD}	
D3_MCK0	Clock	AT43	0	G3V _{DD}	
D3_MCK0_B	Clock Complements	AR43	0	G3V _{DD}	
D3_MCK1	Clock	AR44	0	G3V _{DD}	
D3_MCK1_B	Clock Complements	AP44	0	G3V _{DD}	
D3_MCK2	Clock	AN44	0	G3V _{DD}	
D3_MCK2_B	Clock Complements	AN43	0	G3V _{DD}	
D3_MCK3	Clock	AU44	0	G3V _{DD}	
D3_MCK3_B	Clock Complements	AU43	0	G3V _{DD}	
D3_MCKE0	Clock Enable	BD37	0	G3V _{DD}	2
D3_MCKE1	Clock Enable	BC35	0	G3V _{DD}	2
D3_MCKE2	Clock Enable	BC36	0	G3V _{DD}	2
D3_MCKE3	Clock Enable	BD36	0	G3V _{DD}	2
D3_MCS0_B	Chip Select	AF44	0	G3V _{DD}	
D3_MCS1_B	Chip Select	AC44	0	G3V _{DD}	
D3_MCS2_B	Chip Select	AG43	0	G3V _{DD}	
D3_MCS3_B	Chip Select	AC43	0	G3V _{DD}	
D3_MDIC0	Driver Impedence Calibration	AM43	IO	G3V _{DD}	3
D3_MDIC1	Driver Impedence Calibration	AV44	IO	G3V _{DD}	3
D3_MDM0/D3_MDQS09	Data Mask	AU36	0	G3V _{DD}	1, 28
D3_MDM1/D3_MDQS10	Data Mask	AP36	0	G3V _{DD}	1, 28
D3_MDM2/D3_MDQS11	Data Mask	BA41	0	G3V _{DD}	1, 28
D3_MDM3/D3_MDQS12	Data Mask	AP40	0	G3V _{DD}	1, 28
D3_MDM4/D3_MDQS13	Data Mask	AG40	0	G3V _{DD}	1, 28

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
D3_MDM5/D3_MDQS14	Data Mask	AG38	0	G3V _{DD}	1, 28
D3_MDM6/D3_MDQS15	Data Mask	AF35	0	G3V _{DD}	1, 28
D3_MDM7/D3_MDQS16	Data Mask	AB39	0	G3V _{DD}	1, 28
D3_MDM8/D3_MDQS17	Data Mask	AM37	0	G3V _{DD}	1, 28
D3_MDQ00	Data	AY35	IO	G3V _{DD}	
D3_MDQ01	Data	BA35	IO	G3V _{DD}	
D3_MDQ02	Data	AW37	Ю	G3V _{DD}	
D3_MDQ03	Data	AV37	Ю	G3V _{DD}	
D3_MDQ04	Data	AU35	Ю	G3V _{DD}	
D3_MDQ05	Data	AV35	Ю	G3V _{DD}	
D3_MDQ06	Data	BA37	Ю	G3V _{DD}	
D3_MDQ07	Data	AY37	Ю	G3V _{DD}	
D3_MDQ08	Data	AP35	Ю	G3V _{DD}	
D3_MDQ09	Data	AR35	Ю	G3V _{DD}	
D3_MDQ10	Data	AR38	Ю	G3V _{DD}	
D3_MDQ11	Data	AP38	Ю	G3V _{DD}	
D3_MDQ12	Data	AP34	Ю	G3V _{DD}	
D3_MDQ13	Data	AR34	Ю	G3V _{DD}	
D3_MDQ14	Data	AT38	Ю	G3V _{DD}	
D3_MDQ15	Data	AU38	Ю	G3V _{DD}	
D3_MDQ16	Data	BA40	Ю	G3V _{DD}	
D3_MDQ17	Data	AY40	Ю	G3V _{DD}	
D3_MDQ18	Data	AV40	Ю	G3V _{DD}	
D3_MDQ19	Data	AU40	Ю	G3V _{DD}	
D3_MDQ20	Data	BA39	Ю	G3V _{DD}	
D3_MDQ21	Data	AY39	Ю	G3V _{DD}	
D3_MDQ22	Data	AW41	Ю	G3V _{DD}	
D3_MDQ23	Data	AV41	IO	G3V _{DD}	
D3_MDQ24	Data	AR40	IO	G3V _{DD}	
D3_MDQ25	Data	AR41	IO	G3V _{DD}	
D3_MDQ26	Data	AL40	IO	G3V _{DD}	
D3_MDQ27	Data	AL41	IO	G3V _{DD}	
D3_MDQ28	Data	AT40	IO	G3V _{DD}	
D3_MDQ29	Data	AT41	IO	G3V _{DD}	
D3_MDQ30	Data	AM40	IO	G3V _{DD}	
D3_MDQ31	Data	AM41	IO	G3V _{DD}	
D3_MDQ32	Data	AH40	IO	G3V _{DD}	
D3_MDQ33	Data	AH41	IO	G3V _{DD}	

Table 1. Pinout list by bus (continued)

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
D3_MDQ34	Data	AD40	10	G3V _{DD}	
D3_MDQ35	Data	AD41	10	G3V _{DD}	
D3_MDQ36	Data	AJ40	10	G3V _{DD}	
D3_MDQ37	Data	AJ41	10	G3V _{DD}	
D3_MDQ38	Data	AE40	10	G3V _{DD}	
D3_MDQ39	Data	AE41	10	G3V _{DD}	
D3_MDQ40	Data	AH38	10	G3V _{DD}	
D3_MDQ41	Data	AH37	10	G3V _{DD}	
D3_MDQ42	Data	AD38	10	G3V _{DD}	
D3_MDQ43	Data	AD37	10	G3V _{DD}	
D3_MDQ44	Data	AJ38	Ю	G3V _{DD}	
D3_MDQ45	Data	AJ37	10	G3V _{DD}	
D3_MDQ46	Data	AE38	10	G3V _{DD}	
D3_MDQ47	Data	AE37	10	G3V _{DD}	
D3_MDQ48	Data	AG35	10	G3V _{DD}	
D3_MDQ49	Data	AG34	10	G3V _{DD}	
D3_MDQ50	Data	AC35	10	G3V _{DD}	
D3_MDQ51	Data	AC34	10	G3V _{DD}	
D3_MDQ52	Data	AH35	10	G3V _{DD}	
D3_MDQ53	Data	AH34	10	G3V _{DD}	
D3_MDQ54	Data	AD35	Ю	G3V _{DD}	
D3_MDQ55	Data	AD34	10	G3V _{DD}	
D3_MDQ56	Data	AB40	Ю	G3V _{DD}	
D3_MDQ57	Data	AA40	10	G3V _{DD}	
D3_MDQ58	Data	AB36	Ю	G3V _{DD}	
D3_MDQ59	Data	AA36	Ю	G3V _{DD}	
D3_MDQ60	Data	AB41	10	G3V _{DD}	
D3_MDQ61	Data	AA41	Ю	G3V _{DD}	
D3_MDQ62	Data	AB37	10	G3V _{DD}	
D3_MDQ63	Data	AA37	10	G3V _{DD}	
D3_MDQS00	Data Strobe	AY36	Ю	G3V _{DD}	
D3_MDQS00_B	Data Strobe	BA36	10	G3V _{DD}	
D3_MDQS01	Data Strobe	AT37	10	G3V _{DD}	
D3_MDQS01_B	Data Strobe	AR37	Ю	G3V _{DD}	
D3_MDQS02	Data Strobe	AV39	Ю	G3V _{DD}	
D3_MDQS02_B	Data Strobe	AW39	10	G3V _{DD}	
D3_MDQS03	Data Strobe	AN41	10	G3V _{DD}	
D3_MDQS03_B	Data Strobe	AN40	10	G3V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
D3_MDQS04	Data Strobe	AF41	10	G3V _{DD}	
D3_MDQS04_B	Data Strobe	AF40	10	G3V _{DD}	
D3_MDQS05	Data Strobe	AF37	10	G3V _{DD}	
D3_MDQS05_B	Data Strobe	AF38	10	G3V _{DD}	
D3_MDQS06	Data Strobe	AE34	10	G3V _{DD}	
D3_MDQS06_B	Data Strobe	AE35	10	G3V _{DD}	
D3_MDQS07	Data Strobe	AA38	10	G3V _{DD}	
D3_MDQS07_B	Data Strobe	AB38	10	G3V _{DD}	
D3_MDQS08	Data Strobe	AL38	10	G3V _{DD}	
D3_MDQS08_B	Data Strobe	AL37	10	G3V _{DD}	
D3_MDQS09/D3_MDM0	Data Strobe (x4 support)	AU36	IO	G3V _{DD}	
D3_MDQS09_B	Data Strobe (x4 support)	AV36	IO	G3V _{DD}	28
D3_MDQS10/D3_MDM1	Data Strobe (x4 support)	AP36	10	G3V _{DD}	
D3_MDQS10_B	Data Strobe (x4 support)	AR36	IO	G3V _{DD}	28
D3_MDQS11/D3_MDM2	Data Strobe (x4 support)	BA41	10	G3V _{DD}	
D3_MDQS11_B	Data Strobe (x4 support)	AY41	10	G3V _{DD}	28
D3_MDQS12/D3_MDM3	Data Strobe (x4 support)	AP40	10	G3V _{DD}	
D3_MDQS12_B	Data Strobe (x4 support)	AP41	10	G3V _{DD}	28
D3_MDQS13/D3_MDM4	Data Strobe (x4 support)	AG40	10	G3V _{DD}	
D3_MDQS13_B	Data Strobe (x4 support)	AG41	10	G3V _{DD}	28
D3_MDQS14/ D3_MDM5	Data Strobe (x4 support)	AG38	10	G3V _{DD}	
D3_MDQS14_B	Data Strobe (x4 support)	AG37	10	G3V _{DD}	28
D3_MDQS15/ D3_MDM6	Data Strobe (x4 support)	AF35	10	G3V _{DD}	
D3_MDQS15_B	Data Strobe (x4 support)	AF34	10	G3V _{DD}	28
D3_MDQS16/ D3_MDM7	Data Strobe (x4 support)	AB39	10	G3V _{DD}	
D3_MDQS16_B	Data Strobe (x4 support)	AA39	10	G3V _{DD}	28
D3_MDQS17/ D3_MDM8	Data Strobe (x4 support)	AM37	Ю	G3V _{DD}	
D3_MDQS17_B	Data Strobe (x4 support)	AM38	10	G3V _{DD}	28
D3_MECC0	Error Correcting Code	AN37	Ю	G3V _{DD}	
D3_MECC1	Error Correcting Code	AN38	10	G3V _{DD}	
D3_MECC2	Error Correcting Code	AK35	Ю	G3V _{DD}	
D3_MECC3	Error Correcting Code	AK34	Ю	G3V _{DD}	
D3_MECC4	Error Correcting Code	AM34	Ю	G3V _{DD}	
D3_MECC5	Error Correcting Code	AM35	Ю	G3V _{DD}	
D3_MECC6	Error Correcting Code	AL36	10	G3V _{DD}	
D3_MECC7	Error Correcting Code	AL35	10	G3V _{DD}	
D3_MODT0	On Die Termination	AE43	0	G3V _{DD}	2
D3_MODT1	On Die Termination	AA43	0	G3V _{DD}	2

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
D3_MODT2	On Die Termination	AD43	0	G3V _{DD}	2
D3_MODT3	On Die Termination	AA44	0	G3V _{DD}	2
D3_MRAS_B	Row Address Strobe	AH43	0	G3V _{DD}	
D3_MWE_B	Write Enable	AG44	0	G3V _{DD}	
	Integrated Flash	Controller			
IFC_A26/GPIO2_18	IFC Address	B40	0	OV _{DD}	1
IFC_A27/GPIO2_19	IFC Address	A40	0	OV _{DD}	1
IFC_A28/GPIO2_20	IFC Address	D39	0	OV _{DD}	1
IFC_A29/GPIO2_21	IFC Address	C39	0	OV _{DD}	1
IFC_A30/GPIO2_22	IFC Address	A39	0	OV _{DD}	1
IFC_A31/GPIO2_23	IFC Address	C38	0	OV _{DD}	1
IFC_AD00/cfg_gpinput0	IFC Address/Data	W39	10	OV _{DD}	4, 21
IFC_AD01/cfg_gpinput1	IFC Address/Data	W40	10	OV _{DD}	4, 21
IFC_AD02/cfg_gpinput2	IFC Address/Data	V39	10	OV _{DD}	4, 21
IFC_AD03/cfg_gpinput3	IFC Address/Data	V40	10	OV _{DD}	4, 21
IFC_AD04/cfg_gpinput4	IFC Address/Data	U40	10	OV _{DD}	4, 21
IFC_AD05/cfg_gpinput5	IFC Address/Data	U41	10	OV _{DD}	4, 21
IFC_AD06/cfg_gpinput6	IFC Address/Data	T39	10	OV _{DD}	4, 21
IFC_AD07/cfg_gpinput7	IFC Address/Data	T40	IO	OV _{DD}	4, 21
IFC_AD08/cfg_rcw_src0	IFC Address/Data	R39	IO	OV _{DD}	4, 21
IFC_AD09/cfg_rcw_src1	IFC Address/Data	R40	IO	OV _{DD}	4, 21
IFC_AD10/cfg_rcw_src2	IFC Address/Data	P40	IO	OV _{DD}	4, 21
IFC_AD11/cfg_rcw_src3	IFC Address/Data	P41	IO	OV _{DD}	4, 21
IFC_AD12/cfg_rcw_src4	IFC Address/Data	N39	IO	OV _{DD}	4, 21
IFC_AD13/cfg_rcw_src5	IFC Address/Data	N40	IO	OV _{DD}	4, 21
IFC_AD14/cfg_rcw_src6	IFC Address/Data	M39	IO	OV _{DD}	4, 21
IFC_AD15/cfg_rcw_src7	IFC Address/Data	M40	IO	OV _{DD}	4, 21
IFC_AD16	IFC Address/Data	L40	IO	OV _{DD}	29
IFC_AD17	IFC Address/Data	L41	IO	OV _{DD}	5, 20
IFC_AD18	IFC Address/Data	K39	IO	OV _{DD}	5, 20
IFC_AD19	IFC Address/Data	K40	IO	OV _{DD}	5, 20
IFC_AD20	IFC Address/Data	J39	IO	OV _{DD}	5, 20
IFC_AD21/cfg_dram_type	IFC Address/Data	J40	IO	OV _{DD}	4, 21
IFC_AD22	IFC Address/Data	H40	IO	OV _{DD}	20
IFC_AD23	IFC Address/Data	H41	10	OV _{DD}	20
IFC_AD24	IFC Address/Data	G40	IO	OV _{DD}	20
IFC_AD25/GPIO2_25/ IFC_WP1_B	IFC Address/Data	G41	IO	OV _{DD}	20

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
IFC_AD26/GPIO2_26/ IFC_WP2_B	IFC Address/Data	G39	IO	OV _{DD}	20
IFC_AD27/GPIO2_27/ IFC_WP3_B	IFC Address/Data	F40	Ю	OV _{DD}	20
IFC_AD28/GPIO2_28	IFC Address/Data	E41	IO	OV _{DD}	20
IFC_AD29/GPIO2_29/ IFC_RB2_B	IFC Address/Data	E40	Ю	OV _{DD}	20
IFC_AD30/GPIO2_30/ IFC_RB3_B	IFC Address/Data	E39	Ю	OV _{DD}	20
IFC_AD31/GPIO2_31/ IFC_RB4_B	IFC Address/Data	D41	Ю	OV _{DD}	20
IFC_AVD	IFC Address Valid	H44	0	OV _{DD}	1, 5
IFC_BCTL	IFC Buffer Control	G44	0	OV _{DD}	1
IFC_CLE/cfg_rcw_src8	IFC Command Latch Enable	E43	0	OV _{DD}	1, 4, 25
IFC_CLK0	IFC Clock	B38	0	OV _{DD}	1
IFC_CLK1	IFC Clock	H42	0	OV _{DD}	1
IFC_CLK2	IFC Clock	A38	0	OV _{DD}	1
IFC_CS0_B	IFC Chip Select	C43	0	OV _{DD}	1, 6
IFC_CS1_B/GPIO2_10	IFC Chip Select	C44	0	OV _{DD}	1, 6
IFC_CS2_B/GPIO2_11	IFC Chip Select	B43	0	OV _{DD}	1, 6
IFC_CS3_B/GPIO2_12	IFC Chip Select	A42	0	OV _{DD}	1, 6
IFC_CS4_B/GPIO1_09	IFC Chip Select	C41	0	OV _{DD}	1, 6
IFC_CS5_B/GPIO1_10	IFC Chip Select	B41	0	OV _{DD}	1, 6
IFC_CS6_B/GPIO1_11	IFC Chip Select	A41	0	OV _{DD}	1, 6
IFC_CS7_B/GPIO1_12	IFC Chip Select	C40	0	OV _{DD}	1, 6
IFC_NDDDR_CLK	IFC NAND DDR Clock	H43	0	OV _{DD}	1
IFC_NDDQS	IFC DQS Strobe	E36	10	OV _{DD}	20
IFC_OE_B	IFC Output Enable	E42	0	OV _{DD}	1, 5
IFC_PAR0/GPIO2_13	IFC Address and Data Parity	E38	10	OV _{DD}	20
IFC_PAR1/GPIO2_14	IFC Address and Data Parity	D38	10	OV _{DD}	20
IFC_PAR2/GPIO2_16	IFC Address and Data Parity	F38	IO	OV _{DD}	20
IFC_PAR3/GPIO2_17	IFC Address and Data Parity	F37	IO	OV _{DD}	20
IFC_PERR_B/GPIO2_15	IFC Parity Error	E37	I	OV _{DD}	1, 18
IFC_RB0_B	IFC Ready / Busy CS0	F43	I	OV _{DD}	8
IFC_RB1_B	IFC Ready / Busy CS1	F42	Ι	OV _{DD}	8
IFC_RB2_B/ IFC_AD29 / GPIO2_29	IFC Ready / Busy CS 2	E40	I	OV _{DD}	1
IFC_RB3_B/ IFC_AD30 / GPIO2_30	IFC Ready / Busy CS 3	E39	I	OV _{DD}	1

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
IFC_RB4_B/ IFC_AD31 / GPIO2_31	IFC Ready / Busy CS 4	D41	I	OV _{DD}	1
IFC_TE/cfg_ifc_te	IFC External Transceiver Enable	G42	0	OV _{DD}	1, 4
IFC_WE0_B	IFC Write Enable	E44	0	OV _{DD}	1, 5
IFC_WE2_B	IFC Write Enable	D42	0	OV _{DD}	1
IFC_WE3_B	IFC Write Enable	D44	0	OV _{DD}	1
IFC_WP0_B	IFC Write Protect	F44	0	OV _{DD}	1, 5
IFC_WP1_B/ IFC_AD25 / GPIO2_25	IFC Write Protect	G41	0	OV _{DD}	1
IFC_WP2_B/ IFC_AD26 / GPIO2_26	IFC Write Protect	G39	0	OV _{DD}	1
IFC_WP3_B/ IFC_AD27 / GPIO2_27	IFC Write Protect	F40	0	OV _{DD}	1
	DUART	•			
UART1_CTS_B/GPIO1_21/ UART3_SIN	Clear To Send	N11	Ι	DV _{DD}	1
UART1_RTS_B/GPIO1_19/ UART3_SOUT	Ready to Send	M10	0	DV _{DD}	1
UART1_SIN/GPIO1_17	Receive Data	M11	Ι	DV _{DD}	1
UART1_SOUT/GPIO1_15	Transmit Data	L10	0	DV _{DD}	1
UART2_CTS_B/GPIO1_22/ UART4_SIN	Clear To Send	K11	Ι	DV _{DD}	1
UART2_RTS_B/GPIO1_20/ UART4_SOUT	Ready to Send	K10	0	DV _{DD}	1
UART2_SIN/GPIO1_18	Receive Data	J11	Ι	DV _{DD}	1
UART2_SOUT/GPIO1_16	Transmit Data	J10	0	DV _{DD}	1
UART3_SIN/ UART1_CTS_B / GPIO1_21	Receive Data	N11	Ι	DV _{DD}	1
UART3_SOUT/ UART1_RTS_B/GPIO1_19	Transmit Data	M10	0	DV _{DD}	1
UART4_SIN/ UART2_CTS_B / GPIO1_22	Receive Data	K11	Ι	DV _{DD}	1
UART4_SOUT/ UART2_RTS_B/GPIO1_20	Transmit Data	K10	0	DV _{DD}	1
	I2C				
IIC1_SCL	Serial Clock (supports PBL)	R10	IO	DV _{DD}	7, 8
IIC1_SDA	Serial Data (supports PBL)	R11	10	DV _{DD}	7, 8
IIC2_SCL	Serial Clock	N10	10	DV _{DD}	7, 8
IIC2_SDA	Serial Data	P10	Ю	DV _{DD}	7, 8
IIC3_SCL/GPIO4_00	Serial Clock	N13	10	DV _{DD}	7, 8
IIC3_SDA/GPIO4_01	Serial Data	P13	10	DV _{DD}	7, 8

 Table 1. Pinout list by bus (continued)

Pin assignments

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
IIC4_SCL/GPIO4_02/EVT5_B	Serial Clock	N12	IO	DV _{DD}	7, 8
IIC4_SDA/GPIO4_03/EVT6_B	Serial Data	P12	IO	DV _{DD}	7, 8
	eSPI Inter	ace		1	1
SPI_CLK	SPI Clock	B37	0	OV _{DD}	1
SPI_CS0_B/GPIO2_00/ SDHC_DAT4	SPI Chip Select	C35	0	OV _{DD}	1, 22
SPI_CS1_B/GPIO2_01/ SDHC_DAT5	SPI Chip Select	A36	0	OV _{DD}	1, 22
SPI_CS2_B/GPIO2_02/ SDHC_DAT6	SPI Chip Select	C36	0	OV _{DD}	1, 22
SPI_CS3_B/GPIO2_03/ SDHC_DAT7	SPI Chip Select	D36	0	OV _{DD}	1, 22
SPI_MISO	Master In Slave Out	C37	I	OV _{DD}	
SPI_MOSI	Master Out Slave In	A37	IO	OV _{DD}	20
	eSDHC	;			
SDHC_CD_B	Card Detection	A34	I	OV _{DD}	26
SDHC_CLK/GPIO2_09	Host to Card Clock	A33	0	OV _{DD}	1
SDHC_CMD/GPIO2_04	Command/Response	D33	Ю	OV _{DD}	22
SDHC_DAT0/GPIO2_05	Data	B34	10	OV _{DD}	22
SDHC_DAT1/GPIO2_06	Data	C34	10	OV _{DD}	22
SDHC_DAT2/GPIO2_07	Data	A35	Ю	OV _{DD}	22
SDHC_DAT3/GPIO2_08	Data	B35	10	OV _{DD}	22
SDHC_DAT4/ SPI_CS0_B / GPIO2_00	Data	C35	Ю	OV _{DD}	
SDHC_DAT5/ SPI_CS1_B / GPIO2_01	Data	A36	Ю	OV _{DD}	
SDHC_DAT6/ SPI_CS2_B / GPIO2_02	Data	C36	IO	OV _{DD}	
SDHC_DAT7/ SPI_CS3_B / GPIO2_03	Data	D36	IO	OV _{DD}	
SDHC_WP	Card Write Protection	C33	Ι	OV _{DD}	26
	Programmable Interr	upt Controlle	er		
IRQ00	External Interrupts	V43	Ι	OV _{DD}	1
IRQ01	External Interrupts	V44	Ι	OV _{DD}	1
IRQ02	External Interrupts	U43	Ι	OV _{DD}	1
IRQ03/GPIO1_23	External Interrupts	W42	Ι	OV _{DD}	1
IRQ04/GPIO1_24	External Interrupts	U44	Ι	OV _{DD}	1
IRQ05/GPIO1_25	External Interrupts	R42	Ι	OV _{DD}	1
IRQ06/GPIO1_26	External Interrupts	W41	Ι	OV _{DD}	1
IRQ07/GPIO1_27	External Interrupts	T42	Ι	OV _{DD}	1

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
IRQ08/GPIO1_28	External Interrupts	T44	Ι	OV _{DD}	1
IRQ09/GPIO1_29	External Interrupts	V42	I	OV _{DD}	1
IRQ10/GPIO1_30	External Interrupts	W44	I	OV _{DD}	1
IRQ11/GPIO1_31	External Interrupts	U42	I	OV _{DD}	1
IRQ_OUT_B/EVT9_B	Interrupt Output	T41	0	OV _{DD}	1, 6, 7
	LP Trust	t		1	
LP_TMP_DETECT_B	Low Power Tamper Detect	T27	I	V _{DD} _LP	
	Trust				
TMP_DETECT_B	Tamper Detect	E35	I	OV _{DD}	1
	System Cor	itrol			
HRESET_B	Hard Reset	D35	IO	OV _{DD}	6, 7
PORESET_B	Power On Reset	F35	I	OV _{DD}	
RESET_REQ_B	Reset Request (POR or Hard)	G33	0	OV _{DD}	1, 5
	Power Manage	ement			
ASLEEP/GPIO1_13/ cfg_xvdd_sel	Asleep	G34	0	OV _{DD}	1, 4
	Clocking]			1
RTC/GPIO1_14	Real Time Clock	V33	I	OV _{DD}	1
SYSCLK	System Clock	U34	I	OV _{DD}	
	DDR Clock	ing			1
DDRCLK	DDR Controllers Clock	AL14	I	OV _{DD}	
	Debug				
CKSTP_OUT_B	Checkstop Out	L44	0	OV _{DD}	1, 6, 7
CLK_OUT	Clock Out	N44	0	OV _{DD}	2
EVT0_B	Event 0	N42	IO	OV _{DD}	9
EVT1_B	Event 1	M43	IO	OV _{DD}	
EVT2_B	Event 2	M42	IO	OV _{DD}	
EVT3_B	Event 3	L43	IO	OV _{DD}	
EVT4_B	Event 4	L42	IO	OV _{DD}	
EVT5_B/ IIC4_SCL /GPIO4_02	Event 5	N12	IO	DV _{DD}	
EVT6_B/IIC4_SDA/GPIO4_03	Event 6	P12	IO	DV _{DD}	
EVT7_B/ DMA2_DACK0_B / GPIO4_08	Event 7	P44	Ю	OV _{DD}	
EVT8_B/ DMA2_DDONE0_B / GPIO4_09	Event 8	P42	IO	OV _{DD}	
EVT9_B/IRQ_OUT_B	Event 9	T41	IO	OV _{DD}	
	DFT				
SCAN_MODE_B	Reserved for internal use only	E33	Ι	OV _{DD}	10
TEST_SEL_B	Reserved for internal use only	F34	I	OV _{DD}	10

 Table 1. Pinout list by bus (continued)

Table 1.	Pinout list by	v bus ((continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
	JTAG				
тск	Test Clock	K44	I	OV _{DD}	
TDI	Test Data In	J43	I	OV _{DD}	9
TDO	Test Data Out	K42	0	OV _{DD}	2
TMS	Test Mode Select	K41	I	OV _{DD}	9
TRST_B	Test Reset	J42	I	OV _{DD}	9
	SerDes 1				•
SD1_IMP_CAL_RX	SerDes Receive Impedence Calibration	M16	I	S1V _{DD}	11
SD1_IMP_CAL_TX	SerDes Transmit Impedance Calibration	M22	I	X1V _{DD}	16
SD1_PLL1_TPA	Reserved for internal use only	M18	0	AVDD_SD1_PLL1	12
SD1_PLL1_TPD	Reserved for internal use only	L16	0	X1V _{DD}	12
SD1_PLL2_TPA	Reserved for internal use only	M20	0	AVDD_SD1_PLL2	12
SD1_PLL2_TPD	Reserved for internal use only	L22	0	X1V _{DD}	12
SD1_REF_CLK1	SerDes PLL 1 Reference Clock	P18	I	S1V _{DD}	
SD1_REF_CLK1_B	SerDes PLL 1 Reference Clock Complement	P19	I	S1V _{DD}	
SD1_REF_CLK2	SerDes PLL 2 Reference Clock	P21	I	S1V _{DD}	
SD1_REF_CLK2_B	SerDes PLL 2 Reference Clock Complement	N21	I	S1V _{DD}	
SD1_RX0	SerDes Receive Data (positive)	C15	I	S1V _{DD}	
SD1_RX0_B	SerDes Receive Data (negative)	D15	I	S1V _{DD}	
SD1_RX1	SerDes Receive Data (positive)	A16	I	S1V _{DD}	
SD1_RX1_B	SerDes Receive Data (negative)	B16	I	S1V _{DD}	
SD1_RX2	SerDes Receive Data (positive)	C17	I	S1V _{DD}	
SD1_RX2_B	SerDes Receive Data (negative)	D17	I	S1V _{DD}	
SD1_RX3	SerDes Receive Data (positive)	A18	I	S1V _{DD}	
SD1_RX3_B	SerDes Receive Data (negative)	B18	I	S1V _{DD}	
SD1_RX4	SerDes Receive Data (positive)	C19	I	S1V _{DD}	
SD1_RX4_B	SerDes Receive Data (negative)	D19	I	S1V _{DD}	

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
SD1_RX5	SerDes Receive Data (positive)	A20	I	S1V _{DD}	
SD1_RX5_B	SerDes Receive Data (negative)	B20	I	S1V _{DD}	
SD1_RX6	SerDes Receive Data (positive)	C21	I	S1V _{DD}	
SD1_RX6_B	SerDes Receive Data (negative)	D21	I	S1V _{DD}	
SD1_RX7	SerDes Receive Data (positive)	A22	I	S1V _{DD}	
SD1_RX7_B	SerDes Receive Data (negative)	B22	I	S1V _{DD}	
SD1_TX0	SerDes Transmit Data (positive)	H15	0	X1V _{DD}	
SD1_TX0_B	SerDes Transmit Data (negative)	J15	0	X1V _{DD}	
SD1_TX1	SerDes Transmit Data (positive)	F16	0	X1V _{DD}	
SD1_TX1_B	SerDes Transmit Data (negative)	G16	0	X1V _{DD}	
SD1_TX2	SerDes Transmit Data (positive)	H17	0	X1V _{DD}	
SD1_TX2_B	SerDes Transmit Data (negative)	J17	0	X1V _{DD}	
SD1_TX3	SerDes Transmit Data (positive)	F18	0	X1V _{DD}	
SD1_TX3_B	SerDes Transmit Data (negative)	G18	0	X1V _{DD}	
SD1_TX4	SerDes Transmit Data (positive)	H19	0	X1V _{DD}	
SD1_TX4_B	SerDes Transmit Data (negative)	J19	0	X1V _{DD}	
SD1_TX5	SerDes Transmit Data (positive)	F20	0	X1V _{DD}	
SD1_TX5_B	SerDes Transmit Data (negative)	G20	0	X1V _{DD}	
SD1_TX6	SerDes Transmit Data (positive)	H21	0	X1V _{DD}	
SD1_TX6_B	SerDes Transmit Data (negative)	J21	0	X1V _{DD}	
SD1_TX7	SerDes Transmit Data (positive)	F22	0	X1V _{DD}	
SD1_TX7_B	SerDes Transmit Data (negative)	G22	0	X1V _{DD}	

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
	SerDes 2				
SD2_IMP_CAL_RX	SerDes Receive Impedence Calibration	M23	I	S2V _{DD}	11
SD2_IMP_CAL_TX	SerDes Transmit Impedance Calibration	M29	I	X2V _{DD}	16
SD2_PLL1_TPA	Reserved for internal use only	M25	0	AVDD_SD2_PLL1	12
SD2_PLL1_TPD	Reserved for internal use only	L23	0	X2V _{DD}	12
SD2_PLL2_TPA	Reserved for internal use only	M27	0	AVDD_SD2_PLL2	12
SD2_PLL2_TPD	Reserved for internal use only	L29	0	X2V _{DD}	12
SD2_REF_CLK1	SerDes PLL 1 Reference Clock	P24	I	S2V _{DD}	
SD2_REF_CLK1_B	SerDes PLL 1 Reference Clock Complement	N24	I	S2V _{DD}	
SD2_REF_CLK2	SerDes PLL 2 Reference Clock	P27	I	S2V _{DD}	
SD2_REF_CLK2_B	SerDes PLL 2 Reference Clock Complement	P26	I	S2V _{DD}	
SD2_RX0	SerDes Receive Data (positive)	C23	I	S2V _{DD}	
SD2_RX0_B	SerDes Receive Data (negative)	D23	I	S2V _{DD}	
SD2_RX1	SerDes Receive Data (positive)	A24	I	S2V _{DD}	
SD2_RX1_B	SerDes Receive Data (negative)	B24	Ι	S2V _{DD}	
SD2_RX2	SerDes Receive Data (positive)	C25	Ι	S2V _{DD}	
SD2_RX2_B	SerDes Receive Data (negative)	D25	I	S2V _{DD}	
SD2_RX3	SerDes Receive Data (positive)	A26	I	S2V _{DD}	
SD2_RX3_B	SerDes Receive Data (negative)	B26	I	S2V _{DD}	
SD2_RX4	SerDes Receive Data (positive)	C27	I	S2V _{DD}	
SD2_RX4_B	SerDes Receive Data (negative)	D27	I	S2V _{DD}	
SD2_RX5	SerDes Receive Data (positive)	A28	I	S2V _{DD}	
SD2_RX5_B	SerDes Receive Data (negative)	B28	Ι	S2V _{DD}	
SD2_RX6	SerDes Receive Data (positive)	C29	I	S2V _{DD}	
SD2_RX6_B	SerDes Receive Data (negative)	D29	I	S2V _{DD}	

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
SD2_RX7	SerDes Receive Data (positive)	A30	I	S2V _{DD}	
SD2_RX7_B	SerDes Receive Data (negative)	B30	I	S2V _{DD}	
SD2_TX0	SerDes Transmit Data (positive)	H23	0	X2V _{DD}	
SD2_TX0_B	SerDes Transmit Data (negative)	J23	0	X2V _{DD}	
SD2_TX1	SerDes Transmit Data (positive)	F24	0	X2V _{DD}	
SD2_TX1_B	SerDes Transmit Data (negative)	G24	0	X2V _{DD}	
SD2_TX2	SerDes Transmit Data (positive)	H25	0	X2V _{DD}	
SD2_TX2_B	SerDes Transmit Data (negative)	J25	0	X2V _{DD}	
SD2_TX3	SerDes Transmit Data (positive)	F26	0	X2V _{DD}	
SD2_TX3_B	SerDes Transmit Data (negative)	G26	0	X2V _{DD}	
SD2_TX4	SerDes Transmit Data (positive)	H27	0	X2V _{DD}	
SD2_TX4_B	SerDes Transmit Data (negative)	J27	0	X2V _{DD}	
SD2_TX5	SerDes Transmit Data (positive)	F28	0	X2V _{DD}	
SD2_TX5_B	SerDes Transmit Data (negative)	G28	0	X2V _{DD}	
SD2_TX6	SerDes Transmit Data (positive)	H29	0	X2V _{DD}	
SD2_TX6_B	SerDes Transmit Data (negative)	J29	0	X2V _{DD}	
SD2_TX7	SerDes Transmit Data (positive)	F30	0	X2V _{DD}	
SD2_TX7_B	SerDes Transmit Data (negative)	G30	0	X2V _{DD}	
	SerDes	3	•		ł
SD3_IMP_CAL_RX	SerDes Receive Impedence Calibration	AN19	I	S3V _{DD}	11
SD3_IMP_CAL_TX	SerDes Transmit Impedance Calibration	AN25	I	X3V _{DD}	16
SD3_PLL1_TPA	Reserved for internal use only	AN21	0	AVDD_SD3_PLL1	12
SD3_PLL1_TPD	Reserved for internal use only	AP19	0	X3V _{DD}	12
SD3_PLL2_TPA	Reserved for internal use only	AN23	0	AVDD_SD3_PLL2	12

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
SD3_PLL2_TPD	Reserved for internal use only	AP25	0	X3V _{DD}	12
SD3_REF_CLK1	SerDes PLL 1 Reference Clock	AL21	1	S3V _{DD}	
SD3_REF_CLK1_B	SerDes PLL 1 Reference Clock Complement	AL22	I	S3V _{DD}	
SD3_REF_CLK2	SerDes PLL 2 Reference Clock	AL24	I	S3V _{DD}	
SD3_REF_CLK2_B	SerDes PLL 2 Reference Clock Complement	AM24	I	S3V _{DD}	
SD3_RX0	SerDes Receive Data (positive)	BB18	I	S3V _{DD}	
SD3_RX0_B	SerDes Receive Data (negative)	BA18	I	S3V _{DD}	
SD3_RX1	SerDes Receive Data (positive)	BD19	I	S3V _{DD}	
SD3_RX1_B	SerDes Receive Data (negative)	BC19	I	S3V _{DD}	
SD3_RX2	SerDes Receive Data (positive)	BB20	I	S3V _{DD}	
SD3_RX2_B	SerDes Receive Data (negative)	BA20	I	S3V _{DD}	
SD3_RX3	SerDes Receive Data (positive)	BD21	I	S3V _{DD}	
SD3_RX3_B	SerDes Receive Data (negative)	BC21	I	S3V _{DD}	
SD3_RX4	SerDes Receive Data (positive)	BB22	I	S3V _{DD}	
SD3_RX4_B	SerDes Receive Data (negative)	BA22	I	S3V _{DD}	
SD3_RX5	SerDes Receive Data (positive)	BD23	I	S3V _{DD}	
SD3_RX5_B	SerDes Receive Data (negative)	BC23	I	S3V _{DD}	
SD3_RX6	SerDes Receive Data (positive)	BB24	I	S3V _{DD}	
SD3_RX6_B	SerDes Receive Data (negative)	BA24	I	S3V _{DD}	
SD3_RX7	SerDes Receive Data (positive)	BD25	I	S3V _{DD}	
SD3_RX7_B	SerDes Receive Data (negative)	BC25	I	S3V _{DD}	
SD3_TX0	SerDes Transmit Data (positive)	AU18	0	X3V _{DD}	
SD3_TX0_B	SerDes Transmit Data	AT18	0	X3V _{DD}	

Table 1. Pinout list by bus (continued)

(negative)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
SD3_TX1	SerDes Transmit Data (positive)	AW19	0	X3V _{DD}	
SD3_TX1_B	SerDes Transmit Data (negative)	AV19	0	X3V _{DD}	
SD3_TX2	SerDes Transmit Data (positive)	AU20	0	X3V _{DD}	
SD3_TX2_B	SerDes Transmit Data (negative)	AT20	0	X3V _{DD}	
SD3_TX3	SerDes Transmit Data (positive)	AW21	0	X3V _{DD}	
SD3_TX3_B	SerDes Transmit Data (negative)	AV21	0	X3V _{DD}	
SD3_TX4	SerDes Transmit Data (positive)	AU22	0	X3V _{DD}	
SD3_TX4_B	SerDes Transmit Data (negative)	AT22	0	X3V _{DD}	
SD3_TX5	SerDes Transmit Data (positive)	AW23	0	X3V _{DD}	
SD3_TX5_B	SerDes Transmit Data (negative)	AV23	0	X3V _{DD}	
SD3_TX6	SerDes Transmit Data (positive)	AU24	0	X3V _{DD}	
SD3_TX6_B	SerDes Transmit Data (negative)	AT24	0	X3V _{DD}	
SD3_TX7	SerDes Transmit Data (positive)	AW25	0	X3V _{DD}	
SD3_TX7_B	SerDes Transmit Data (negative)	AV25	0	X3V _{DD}	
	SerDes 4				1
SD4_IMP_CAL_RX	SerDes Receive Impedence Calibration	AN26	I	S4V _{DD}	11
SD4_IMP_CAL_TX	SerDes Transmit Impedance Calibration	AN32	I	X4V _{DD}	16
SD4_PLL1_TPA	Reserved for internal use only	AN28	0	AVDD_SD4_PLL1	12
SD4_PLL1_TPD	Reserved for internal use only	AP26	0	X4V _{DD}	12
SD4_PLL2_TPA	Reserved for internal use only	AN30	0	AVDD_SD4_PLL2	12
SD4_PLL2_TPD	Reserved for internal use only	AP32	0	X4V _{DD}	12
SD4_REF_CLK1	SerDes PLL 1 Reference Clock	AL27	I	S4V _{DD}	
SD4_REF_CLK1_B	SerDes PLL 1 Reference Clock Complement	AM27	I	S4V _{DD}	
SD4_REF_CLK2	SerDes PLL 2 Reference Clock	AL30	I	S4V _{DD}	
SD4_REF_CLK2_B	SerDes PLL 2 Reference Clock Complement	AL29	I	S4V _{DD}	

Table 1.	Pinout list by	y bus ((continued)
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Signal	Signal description	Package pin number	Pin type	Power supply	Notes
SD4_RX0	SerDes Receive Data (positive)	BB26	I	S4V _{DD}	
SD4_RX0_B	SerDes Receive Data (negative)	BA26	I	S4V _{DD}	
SD4_RX1	SerDes Receive Data (positive)	BD27	I	S4V _{DD}	
SD4_RX1_B	SerDes Receive Data (negative)	BC27	I	S4V _{DD}	
SD4_RX2	SerDes Receive Data (positive)	BB28	I	S4V _{DD}	
SD4_RX2_B	SerDes Receive Data (negative)	BA28	I	S4V _{DD}	
SD4_RX3	SerDes Receive Data (positive)	BD29	I	S4V _{DD}	
SD4_RX3_B	SerDes Receive Data (negative)	BC29	I	S4V _{DD}	
SD4_RX4	SerDes Receive Data (positive)	BB30	I	S4V _{DD}	
SD4_RX4_B	SerDes Receive Data (negative)	BA30	I	S4V _{DD}	
SD4_RX5	SerDes Receive Data (positive)	BD31	I	S4V _{DD}	
SD4_RX5_B	SerDes Receive Data (negative)	BC31	I	S4V _{DD}	
SD4_RX6	SerDes Receive Data (positive)	BB32	I	S4V _{DD}	
SD4_RX6_B	SerDes Receive Data (negative)	BA32	I	S4V _{DD}	
SD4_RX7	SerDes Receive Data (positive)	BD33	I	S4V _{DD}	
SD4_RX7_B	SerDes Receive Data (negative)	BC33	I	S4V _{DD}	
SD4_TX0	SerDes Transmit Data (positive)	AU26	0	X4V _{DD}	
SD4_TX0_B	SerDes Transmit Data (negative)	AT26	0	X4V _{DD}	
SD4_TX1	SerDes Transmit Data (positive)	AW27	0	X4V _{DD}	
SD4_TX1_B	SerDes Transmit Data (negative)	AV27	0	X4V _{DD}	
SD4_TX2	SerDes Transmit Data (positive)	AU28	0	X4V _{DD}	
SD4_TX2_B	SerDes Transmit Data (negative)	AT28	0	X4V _{DD}	

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
SD4_TX3	SerDes Transmit Data (positive)	AW29	0	X4V _{DD}	
SD4_TX3_B	SerDes Transmit Data (negative)	AV29	0	X4V _{DD}	
SD4_TX4	SerDes Transmit Data (positive)	AU30	0	X4V _{DD}	
SD4_TX4_B	SerDes Transmit Data (negative)	AT30	0	X4V _{DD}	
SD4_TX5	SerDes Transmit Data (positive)	AW31	0	X4V _{DD}	
SD4_TX5_B	SerDes Transmit Data (negative)	AV31	0	X4V _{DD}	
SD4_TX6	SerDes Transmit Data (positive)	AU32	0	X4V _{DD}	
SD4_TX6_B	SerDes Transmit Data (negative)	AT32	0	X4V _{DD}	
SD4_TX7	SerDes Transmit Data (positive)	AW33	0	X4V _{DD}	
SD4_TX7_B	SerDes Transmit Data (negative)	AV33	0	X4V _{DD}	
	USB PHY 1	& 2			
USB1_DRVVBUS	USB PHY Digital signal - Drive VBUS	E32	0	USB_HV _{DD}	
USB1_PWRFAULT	USB PHY Digital signal - Power Fault	F32	I	USB_HV _{DD}	
USB1_UDM	USB PHY Data Minus	J31	IO	USB_HV _{DD}	
USB1_UDP	USB PHY Data Plus	K31	IO	USB_HV _{DD}	
USB1_UID	USB PHY ID Detect	L32	I	USB_OV DD	
USB1_VBUSCLMP	USB PHY VBUS	G32	I	USB_HV DD	
USB2_DRVVBUS	USB PHY Digital signal - Drive VBUS	A32	0	USB_HV _{DD}	
USB2_PWRFAULT	USB PHY Digital signal - Power Fault	B32	I	USB_HV _{DD}	
USB2_UDM	USB PHY Data Minus	M31	IO	USB_HV _{DD}	
USB2_UDP	USB PHY Data Plus	N31	IO	USB_HV _{DD}	
USB2_UID	USB PHY ID Detect	H32	I	USB_OV DD	
USB2_VBUSCLMP	USB PHY VBUS	C32	I	USB_HV _{DD}	
USB_IBIAS_REXT	USB PHY Impedance Calibration	D31	IO	-	23
	USB CLK	,			
USBCLK	USB PHY Clock In	E34	I	OV _{DD}	

Table 1. Pinout list by bus (continued)

Pin assignments

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
TSEC_1588_ALARM_OUT1/ GPIO3_03	Alarm Out 1	K13	0	LV _{DD}	1
TSEC_1588_ALARM_OUT2/ GPIO3_04	Alarm Out 2	J12	0	LV _{DD}	1
TSEC_1588_CLK_IN/ GPIO3_00	Clock In	A14	I	LV _{DD}	1
TSEC_1588_CLK_OUT / GPIO3_05	Clock Out	F14	0	LV _{DD}	1
TSEC_1588_PULSE_OUT1/ GPIO3_06	Pulse Out 1	M14	0	LV _{DD}	1
TSEC_1588_PULSE_OUT2/ GPIO3_07	Pulse Out 2	L13	0	LV _{DD}	1
TSEC_1588_TRIG_IN1/ GPIO3_01	Trigger In 1	N15	I	LV _{DD}	1
TSEC_1588_TRIG_IN2/ GPIO3_02	Trigger In 2	N14	I	LV _{DD}	1
	Ethernet Managemer	nt Interface 1			
EMI1_MDC	Management Data Clock	G13	0	LV _{DD}	
EMI1_MDIO	Management Data In/Out	H13	10	LV _{DD}	6
	Ethernet Managemer	t Interface 2	2		•
EMI2_MDC	Management Data Clock (1.2V open drain)	D13	0	OV _{DD}	7, 13
EMI2_MDIO	Management Data In/Out (1.2V open drain)	E13	IO	OV _{DD}	7, 13
	Ethernet Contr	oller 1			-
EC1_GTX_CLK/GPIO3_13	Transmit Clock Out	F12	0	LV _{DD}	1
EC1_GTX_CLK125	Reference Clock	D12	I	LV _{DD}	
EC1_RXD0/GPIO3_19	Receive Data	C12	I	LV _{DD}	1
EC1_RXD1/GPIO3_18	Receive Data	E12	I	LV _{DD}	1
EC1_RXD2/GPIO3_17	Receive Data	A12	I	LV _{DD}	1
EC1_RXD3/GPIO3_16	Receive Data	G12	I	LV _{DD}	1
EC1_RX_CLK/GPIO3_15	Receive Clock	B12	I	LV _{DD}	1
EC1_RX_DV/GPIO3_14	Receive Data Valid	A13	I	LV _{DD}	1
EC1_TXD0/GPIO3_11	Transmit Data	H12	0	LV _{DD}	1
EC1_TXD1/GPIO3_10	Transmit Data	K12	0	LV _{DD}	1
EC1_TXD2/GPIO3_09	Transmit Data	L12	0	LV _{DD}	1
EC1_TXD3/GPIO3_08	Transmit Data	M12	0	LV _{DD}	1
EC1_TX_EN/GPIO3_12	Transmit Enable	C13	0	LV _{DD}	1, 14
	Ethernet Contr	oller 2		1	
EC2_GTX_CLK/GPIO3_25	Transmit Clock Out	H10	0	LV _{DD}	1
EC2_GTX_CLK125	Reference Clock	E10	I	LV _{DD}	

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin	Pin type	Power supply	Notes
		number			
EC2_RXD0/GPIO3_31	Receive Data	A10	Ι	LV _{DD}	1
EC2_RXD1/GPIO3_30	Receive Data	A11	Ι	LV _{DD}	1
EC2_RXD2/GPIO3_29	Receive Data	C11	Ι	LV _{DD}	1
EC2_RXD3/GPIO3_28	Receive Data	D10	Ι	LV _{DD}	1
EC2_RX_CLK/GPIO3_27	Receive Clock	B10	Ι	LV _{DD}	1
EC2_RX_DV/GPIO3_26	Receive Data Valid	C10	Ι	LV _{DD}	1
EC2_TXD0/GPIO3_23	Transmit Data	D11	0	LV _{DD}	1
EC2_TXD1/GPIO3_22	Transmit Data	F10	0	LV _{DD}	1
EC2_TXD2/GPIO3_21	Transmit Data	F11	0	LV _{DD}	1
EC2_TXD3/GPIO3_20	Transmit Data	G11	0	LV _{DD}	1
EC2_TX_EN/GPIO3_24	Transmit Enable	G10	0	LV _{DD}	1, 14
	DMA		_		
DMA1_DACK0_B/GPIO4_05	DMA1 channel 0 acknowledge	R43	0	OV _{DD}	1
DMA1_DDONE0_B/GPIO4_06	DMA1 channel 0 done	R44	0	OV _{DD}	1
DMA1_DREQ0_B/GPIO4_04	DMA1 channel 0 request	P43	Ι	OV _{DD}	1
DMA2_DACK0_B/GPIO4_08/ EVT7_B	DMA2 channel 0 acknowledge	P44	0	OV _{DD}	1
DMA2_DDONE0_B/ GPIO4_09/EVT8_B	DMA2 channel 0 done	P42	0	OV _{DD}	1
DMA2_DREQ0_B/GPIO4_07	DMA2 channel 0 request	N41	Ι	OV _{DD}	1
	Analog sigr	als			·
D1_MVREF	SSTL1.35/1.5 Reference Voltage	V13	Ι	G1V _{DD} /2	
D1_TPA	Reserved for internal use only	AL13	-	-	12
D2_MVREF	SSTL1.35/1.5 Reference Voltage	AH13	Ι	G2V _{DD} /2	
D2_TPA	Reserved for internal use only	AM13	-	-	12
D3_MVREF	SSTL1.35/1.5 Reference Voltage	AD32	Ι	G3V _{DD} /2	
D3_TPA	Reserved for internal use only	AF32	-	-	12
FA_ANALOG_G_V	Reserved for internal use only	R32	-	-	15
FA_ANALOG_PIN	Reserved for internal use only	T32	-	-	15
TD1_ANODE	Thermal diode anode pin	AA34	-	Internal Diode	19, 24
TD1_CATHODE	Thermal diode cathode pin	Y34	-	Internal Diode	19, 24
TD2_ANODE	Thermal diode anode pin	AL16	-	Internal Diode	19, 24
TD2_CATHODE	Thermal diode cathode pin	AL17	-	Internal Diode	19, 24
TH_TPA	Thermal Test Point Analog	W32	-	-	12
	Power-On-Reset Co	onfiguration			
cfg_dram_type/ IFC_AD21	Power-On-Reset Configuration Signal	J40	Ι	OV _{DD}	1, 4

 Table 1. Pinout list by bus (continued)

	Table 1. Pinout list by	bus (co	lilliue	u)	
Signal	Signal description	Package pin number	Pin type	Power supply	Notes
cfg_gpinput0/IFC_AD00	Power-On-Reset Configuration Signal	W39	I	OV _{DD}	1, 4
cfg_gpinput1/IFC_AD01	Power-On-Reset Configuration Signal	W40	I	OV _{DD}	1, 4
cfg_gpinput2/IFC_AD02	Power-On-Reset Configuration Signal	V39	I	OV _{DD}	1, 4
cfg_gpinput3/IFC_AD03	Power-On-Reset Configuration Signal	V40	I	OV _{DD}	1, 4
cfg_gpinput4/IFC_AD04	Power-On-Reset Configuration Signal	U40	I	OV _{DD}	1, 4
cfg_gpinput5/IFC_AD05	Power-On-Reset Configuration Signal	U41	I	OV _{DD}	1, 4
cfg_gpinput6/IFC_AD06	Power-On-Reset Configuration Signal	T39	I	OV _{DD}	1, 4
cfg_gpinput7/IFC_AD07	Power-On-Reset Configuration Signal	T40	I	OV _{DD}	1, 4
cfg_ifc_te/IFC_TE	Power-On-Reset Configuration Signal	G42	I	OV _{DD}	1, 4
cfg_rcw_src0/ IFC_AD08	Power-On-Reset Configuration Signal	R39	I	OV _{DD}	1, 4
cfg_rcw_src1/ IFC_AD09	Power-On-Reset Configuration Signal	R40	I	OV _{DD}	1, 4
cfg_rcw_src2/IFC_AD10	Power-On-Reset Configuration Signal	P40	I	OV _{DD}	1, 4
cfg_rcw_src3/IFC_AD11	Power-On-Reset Configuration Signal	P41	I	OV _{DD}	1, 4
cfg_rcw_src4/IFC_AD12	Power-On-Reset Configuration Signal	N39	I	OV _{DD}	1, 4
cfg_rcw_src5/IFC_AD13	Power-On-Reset Configuration Signal	N40	I	OV _{DD}	1, 4
cfg_rcw_src6/IFC_AD14	Power-On-Reset Configuration Signal	M39	I	OV _{DD}	1, 4
cfg_rcw_src7/IFC_AD15	Power-On-Reset Configuration Signal	M40	I	OV _{DD}	1, 4
cfg_rcw_src8/ IFC_CLE	Power-On-Reset Configuration Signal	E43	I	OV _{DD}	1, 4
cfg_xvdd_sel/ ASLEEP / GPIO1_13	Power-On-Reset Configuration Signal	G34	I	OV _{DD}	1, 4
	General Purpose In	put/Output			
GPIO1_09/IFC_CS4_B	General Purpose Input/Output	C41	IO	OV _{DD}	
GPIO1_10/IFC_CS5_B	General Purpose Input/Output	B41	10	OV _{DD}	
GPIO1_11/IFC_CS6_B	General Purpose Input/Output	A41	IO	OV _{DD}	
GPIO1_12/IFC_CS7_B	General Purpose Input/Output	C40	IO	OV _{DD}	

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GPIO1_13/ ASLEEP / cfg_xvdd_sel	General Purpose Input/Output	G34	0	OV _{DD}	1, 4
GPIO1_14/RTC	General Purpose Input/Output	V33	IO	OV _{DD}	
GPIO1_15/UART1_SOUT	General Purpose Input/Output	L10	10	DV _{DD}	
GPIO1_16/UART2_SOUT	General Purpose Input/Output	J10	10	DV _{DD}	
GPIO1_17/ UART1_SIN	General Purpose Input/Output	M11	IO	DV _{DD}	
GPIO1_18/UART2_SIN	General Purpose Input/Output	J11	IO	DV _{DD}	
GPIO1_19/ UART1_RTS_B / UART3_SOUT	General Purpose Input/Output	M10	IO	DV _{DD}	
GPIO1_20/ UART2_RTS_B / UART4_SOUT	General Purpose Input/Output	K10	IO	DV _{DD}	
GPIO1_21/ UART1_CTS_B / UART3_SIN	General Purpose Input/Output	N11	IO	DV _{DD}	
GPIO1_22/ UART2_CTS_B / UART4_SIN	General Purpose Input/Output	K11	IO	DV _{DD}	
GPIO1_23/IRQ03	General Purpose Input/Output	W42	IO	OV _{DD}	
GPIO1_24/ IRQ04	General Purpose Input/Output	U44	IO	OV _{DD}	
GPIO1_25/ IRQ05	General Purpose Input/Output	R42	IO	OV _{DD}	
GPIO1_26/ IRQ06	General Purpose Input/Output	W41	10	OV _{DD}	
GPIO1_27/ IRQ07	General Purpose Input/Output	T42	IO	OV _{DD}	
GPIO1_28/ IRQ08	General Purpose Input/Output	T44	10	OV _{DD}	
GPIO1_29/ IRQ09	General Purpose Input/Output	V42	IO	OV _{DD}	
GPIO1_30/ IRQ10	General Purpose Input/Output	W44	IO	OV _{DD}	
GPIO1_31/ IRQ11	General Purpose Input/Output	U42	IO	OV _{DD}	
GPIO2_00/ SPI_CS0_B / SDHC_DAT4	General Purpose Input/Output	C35	IO	OV _{DD}	
GPIO2_01/ SPI_CS1_B / SDHC_DAT5	General Purpose Input/Output	A36	IO	OV _{DD}	
GPIO2_02/ SPI_CS2_B / SDHC_DAT6	General Purpose Input/Output	C36	IO	OV _{DD}	
GPIO2_03/ SPI_CS3_B / SDHC_DAT7	General Purpose Input/Output	D36	IO	OV _{DD}	
GPIO2_04/SDHC_CMD	General Purpose Input/Output	D33	10	OV _{DD}	
GPIO2_05/SDHC_DAT0	General Purpose Input/Output	B34	IO	OV _{DD}	
GPIO2_06/SDHC_DAT1	General Purpose Input/Output	C34	IO	OV _{DD}	
GPIO2_07/SDHC_DAT2	General Purpose Input/Output	A35	IO	OV _{DD}	
GPIO2_08/SDHC_DAT3	General Purpose Input/Output	B35	IO	OV _{DD}	
GPIO2_09/SDHC_CLK	General Purpose Input/Output	A33	IO	OV _{DD}	
GPIO2_10/IFC_CS1_B	General Purpose Input/Output	C44	IO	OV _{DD}	

 Table 1. Pinout list by bus (continued)

B43

General Purpose Input/Output

OV _{DD}

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GPIO2_11/IFC_CS2_B

Signal	Signal description	Package	Pin	Power supply	Notes
orginal		pin number	type		
GPIO2_12/IFC_CS3_B	General Purpose Input/Output	A42	10	OV _{DD}	
GPIO2_13/IFC_PAR0	General Purpose Input/Output	E38	10	OV _{DD}	
GPIO2_14/IFC_PAR1	General Purpose Input/Output	D38	10	OV _{DD}	
GPIO2_15/IFC_PERR_B	General Purpose Input/Output	E37	10	OV _{DD}	
GPIO2_16/IFC_PAR2	General Purpose Input/Output	F38	10	OV _{DD}	
GPIO2_17/IFC_PAR3	General Purpose Input/Output	F37	10	OV _{DD}	
GPIO2_18/IFC_A26	General Purpose Input/Output	B40	10	OV _{DD}	
GPIO2_19/IFC_A27	General Purpose Input/Output	A40	10	OV _{DD}	
GPIO2_20/IFC_A28	General Purpose Input/Output	D39	10	OV _{DD}	
GPIO2_21/IFC_A29	General Purpose Input/Output	C39	10	OV _{DD}	
GPIO2_22/IFC_A30	General Purpose Input/Output	A39	10	OV _{DD}	
GPIO2_23/IFC_A31	General Purpose Input/Output	C38	IO	OV _{DD}	
GPIO2_25/ IFC_AD25 / IFC_WP1_B	General Purpose Input/Output	G41	IO	OV _{DD}	
GPIO2_26/ IFC_AD26 / IFC_WP2_B	General Purpose Input/Output	G39	IO	OV _{DD}	
GPIO2_27/ IFC_AD27 / IFC_WP3_B	General Purpose Input/Output	F40	IO	OV _{DD}	
GPIO2_28/IFC_AD28	General Purpose Input/Output	E41	10	OV _{DD}	
GPIO2_29/ IFC_AD29 / IFC_RB2_B	General Purpose Input/Output	E40	IO	OV _{DD}	
GPIO2_30/ IFC_AD30 / IFC_RB3_B	General Purpose Input/Output	E39	IO	OV _{DD}	
GPIO2_31/ IFC_AD31 / IFC_RB4_B	General Purpose Input/Output	D41	IO	OV _{DD}	
GPIO3_00/ TSEC_1588_CLK_IN	General Purpose Input/Output	A14	IO	LV _{DD}	
GPIO3_01/ TSEC_1588_TRIG_IN1	General Purpose Input/Output	N15	Ю	LV _{DD}	
GPIO3_02/ TSEC_1588_TRIG_IN2	General Purpose Input/Output	N14	Ю	LV _{DD}	
GPIO3_03/ TSEC_1588_ALARM_OUT1	General Purpose Input/Output	K13	Ю	LV _{DD}	
GPIO3_04/ TSEC_1588_ALARM_OUT2	General Purpose Input/Output	J12	Ю	LV _{DD}	
GPIO3_05/ TSEC_1588_CLK_OUT	General Purpose Input/Output	F14	Ю	LV _{DD}	
GPIO3_06/ TSEC_1588_PULSE_OUT1	General Purpose Input/Output	M14	Ю	LV _{DD}	
GPIO3_07/ TSEC_1588_PULSE_OUT2	General Purpose Input/Output	L13	Ю	LV _{DD}	
GPIO3_08/ EC1_TXD3	General Purpose Input/Output	M12	10	LV _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
Ognai		pin number	type		Notes
GPIO3_09/EC1_TXD2	General Purpose Input/Output	L12	10	LV _{DD}	
GPIO3_10/EC1_TXD1	General Purpose Input/Output	K12	IO	LV _{DD}	
GPIO3_11/EC1_TXD0	General Purpose Input/Output	H12	IO	LV _{DD}	
GPIO3_12/EC1_TX_EN	General Purpose Input/Output	C13	IO	LV _{DD}	
GPIO3_13/EC1_GTX_CLK	General Purpose Input/Output	F12	IO	LV _{DD}	
GPIO3_14/EC1_RX_DV	General Purpose Input/Output	A13	IO	LV _{DD}	
GPIO3_15/EC1_RX_CLK	General Purpose Input/Output	B12	IO	LV _{DD}	
GPIO3_16/EC1_RXD3	General Purpose Input/Output	G12	IO	LV _{DD}	
GPIO3_17/ EC1_RXD2	General Purpose Input/Output	A12	IO	LV _{DD}	
GPIO3_18/EC1_RXD1	General Purpose Input/Output	E12	IO	LV _{DD}	
GPIO3_19/ EC1_RXD0	General Purpose Input/Output	C12	IO	LV _{DD}	
GPIO3_20/ EC2_TXD3	General Purpose Input/Output	G11	IO	LV _{DD}	
GPIO3_21/ EC2_TXD2	General Purpose Input/Output	F11	IO	LV _{DD}	
GPIO3_22/ EC2_TXD1	General Purpose Input/Output	F10	IO	LV _{DD}	
GPIO3_23/ EC2_TXD0	General Purpose Input/Output	D11	IO	LV _{DD}	
GPIO3_24/ EC2_TX_EN	General Purpose Input/Output	G10	IO	LV _{DD}	
GPIO3_25/EC2_GTX_CLK	General Purpose Input/Output	H10	IO	LV _{DD}	
GPIO3_26/ EC2_RX_DV	General Purpose Input/Output	C10	IO	LV _{DD}	
GPIO3_27/EC2_RX_CLK	General Purpose Input/Output	B10	IO	LV _{DD}	
GPIO3_28/EC2_RXD3	General Purpose Input/Output	D10	IO	LV _{DD}	
GPIO3_29/ EC2_RXD2	General Purpose Input/Output	C11	IO	LV _{DD}	
GPIO3_30/EC2_RXD1	General Purpose Input/Output	A11	IO	LV _{DD}	
GPIO3_31/EC2_RXD0	General Purpose Input/Output	A10	IO	LV _{DD}	
GPIO4_00/IIC3_SCL	General Purpose Input/Output	N13	10	DV _{DD}	
GPIO4_01/IIC3_SDA	General Purpose Input/Output	P13	10	DV _{DD}	
GPIO4_02/ IIC4_SCL /EVT5_B	General Purpose Input/Output	N12	IO	DV _{DD}	
GPIO4_03/ IIC4_SDA /EVT6_B	General Purpose Input/Output	P12	10	DV _{DD}	
GPIO4_04/DMA1_DREQ0_B	General Purpose Input/Output	P43	Ю	OV _{DD}	
GPIO4_05/ DMA1_DACK0_B	General Purpose Input/Output	R43	IO	OV _{DD}	
GPIO4_06/DMA1_DDONE0_B	General Purpose Input/Output	R44	IO	OV _{DD}	
GPIO4_07/DMA2_DREQ0_B	General Purpose Input/Output	N41	IO	OV _{DD}	
GPIO4_08/ DMA2_DACK0_B / EVT7_B	General Purpose Input/Output	P44	IO	OV _{DD}	
GPIO4_09/ DMA2_DDONE0_B/EVT8_B	General Purpose Input/Output	P42	IO	OV _{DD}	
	Power and Groun	d Signals			
GND001	GND	A43			
GND002	GND	B11			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND003	GND	B13			
GND004	GND	B33			
GND005	GND	B36			
GND006	GND	B39			
GND007	GND	B42			
GND008	GND	B44			
GND009	GND	C4			
GND010	GND	C5			
GND011	GND	C6			
GND012	GND	C7			
GND013	GND	C8			
GND014	GND	C9			
GND015	GND	D3			
GND016	GND	D9			
GND017	GND	D34			
GND018	GND	D37			
GND019	GND	D40			
GND020	GND	D43			
GND021	GND	E3			
GND022	GND	E6			
GND023	GND	E9			
GND024	GND	E11			
GND025	GND	F3			
GND026	GND	F4			
GND027	GND	F8			
GND028	GND	F9			
GND029	GND	F13			
GND030	GND	F33			
GND031	GND	F36			
GND032	GND	F39			
GND033	GND	F41			
GND034	GND	G3			
GND035	GND	G6			
GND036	GND	G9			
GND037	GND	G43			
GND038	GND	H3			
GND039	GND	H9			
GND040	GND	H11			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND041	GND	H34			
GND042	GND	H37			
GND043	GND	H39			
GND044	GND	J3			
GND045	GND	J5			
GND046	GND	J7			
GND047	GND	J9			
GND048	GND	J13			
GND049	GND	J41			
GND050	GND	J44			
GND051	GND	K3			
GND052	GND	K6			
GND053	GND	K9			
GND054	GND	K33			
GND055	GND	K36			
GND056	GND	K43			
GND057	GND	L3			
GND058	GND	L6			
GND059	GND	L9			
GND060	GND	L11			
GND061	GND	L39			
GND062	GND	M3			
GND063	GND	M6			
GND064	GND	M9			
GND065	GND	M13			
GND066	GND	M34			
GND067	GND	M41			
GND068	GND	M44			
GND069	GND	N3			
GND070	GND	N6			
GND071	GND	N9			
GND072	GND	N37			
GND073	GND	N43			
GND074	GND	P3			
GND075	GND	P6			
GND076	GND	P9			
GND077	GND	P11			
GND078	GND	P14			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND079	GND	P15			
GND080	GND	P16			
GND081	GND	P33			
GND082	GND	P39			
GND083	GND	R3			
GND084	GND	R6			
GND085	GND	R9			
GND086	GND	R31			
GND087	GND	R41			
GND088	GND	Т3			
GND089	GND	Т6			
GND090	GND	Т9			
GND091	GND	T10			
GND092	GND	T11			
GND093	GND	T12			
GND094	GND	T14			
GND095	GND	T17			
GND096	GND	T31			
GND097	GND	T37			
GND098	GND	T43			
GND099	GND	U3			
GND100	GND	U6			
GND101	GND	U13			
GND102	GND	U14			
GND103	GND	U16			
GND104	GND	U18			
GND105	GND	U20			
GND106	GND	U22			
GND107	GND	U24			
GND108	GND	U26			
GND109	GND	U28			
GND110	GND	U31			
GND111	GND	U33			
GND112	GND	U39			
GND113	GND	V3			
GND114	GND	V6			
GND115	GND	V9			
GND116	GND	V12			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND117	GND	V14			
GND118	GND	V17			
GND119	GND	V19			
GND120	GND	V10			
GND121	GND	V23			
GND122	GND	V25			
GND123	GND	V27			
GND124	GND	V29			
GND125	GND	V31			
GND126	GND	V34			
GND127	GND	V41			
GND128	GND	W3			
GND129	GND	W6			
GND130	GND	W9			
GND131	GND	W12			
GND132	GND	W13			
GND133	GND	W14			
GND134	GND	W16			
GND135	GND	W18			
GND136	GND	W20			
GND137	GND	W22			
GND138	GND	W24			
GND139	GND	W26			
GND140	GND	W28			
GND141	GND	W31			
GND142	GND	W33			
GND143	GND	W37			
GND144	GND	W43			
GND145	GND	Y3			
GND146	GND	Y6			
GND147	GND	Y9			
GND148	GND	Y14			
GND149	GND	Y17			
GND150	GND	Y19			
GND151	GND	Y21			
GND152	GND	Y23			
GND153	GND	Y25			
GND154	GND	Y27			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND155	GND	Y29			
GND156	GND	Y31			
GND157	GND	Y35			
GND158	GND	Y36			
GND159	GND	Y37			
GND160	GND	Y38			
GND161	GND	Y39			
GND162	GND	Y40			
GND163	GND	Y41			
GND164	GND	Y42			
GND165	GND	AA3			
GND166	GND	AA6			
GND167	GND	AA10			
GND168	GND	AA11			
GND169	GND	AA12			
GND170	GND	AA13			
GND171	GND	AA14			
GND172	GND	AA16			
GND173	GND	AA18			
GND174	GND	AA20			
GND175	GND	AA22			
GND176	GND	AA24			
GND177	GND	AA26			
GND178	GND	AA28			
GND179	GND	AA31			
GND180	GND	AA35			
GND181	GND	AA42			
GND182	GND	AB3			
GND183	GND	AB4			
GND184	GND	AB8			
GND185	GND	AB14			
GND186	GND	AB17			
GND187	GND	AB19			
GND188	GND	AB21			
GND189	GND	AB23			
GND190	GND	AB25			
GND191	GND	AB27			
GND192	GND	AB29			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin	Pin type	Power supply	Notes
		number	type		
GND193	GND	AB31			
GND194	GND	AB32			
GND195	GND	AB33			
GND196	GND	AB34			
GND197	GND	AB35			
GND198	GND	AB42			
GND199	GND	AC3			
GND200	GND	AC6			
GND201	GND	AC8			
GND202	GND	AC14			
GND203	GND	AC16			
GND204	GND	AC18			
GND205	GND	AC20			
GND206	GND	AC22			
GND207	GND	AC24			
GND208	GND	AC26			
GND209	GND	AC28			
GND210	GND	AC31			
GND211	GND	AC33			
GND212	GND	AC36			
GND213	GND	AC37			
GND214	GND	AC38			
GND215	GND	AC39			
GND216	GND	AC40			
GND217	GND	AC41			
GND218	GND	AC42			
GND219	GND	AD3			
GND220	GND	AD5			
GND221	GND	AD8			
GND222	GND	AD10			
GND223	GND	AD11			
GND224	GND	AD12			
GND225	GND	AD13			
GND226	GND	AD14			
GND227	GND	AD17			
GND228	GND	AD19			
GND229	GND	AD21			
GND230	GND	AD23			

Table 1. Pinout list by bus (continued)

GND231GNDAD25GND232GNDAD27GND233GNDAD20GND234GNDAD31GND235GNDAD33GND236GNDAD33GND237GNDAD39GND238GNDAD42GND240GNDAE3GND241GNDAE14GND242GNDAE18GND243GNDAE18GND244GNDAE18GND245GNDAE20GND246GNDAE28GND247GNDAE28GND249GNDAE28GND249GNDAE28GND250GNDAE31GND255GNDAE33GND256GNDAF3 <tr< th=""><th>Signal</th><th>Signal description</th><th>Package pin number</th><th>Pin type</th><th>Power supply</th><th>Notes</th></tr<>	Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND233GNDAD29GND234GNDAD31GND235GNDAD33GND236GNDAD38GND237GNDAD39GND238GNDAD42GND239GNDAE14GND240GNDAE14GND241GNDAE14GND242GNDAE16GND243GNDAE16GND244GNDAE18GND245GNDAE20GND246GNDAE28GND247GNDAE28GND248GNDAE28GND250GNDAE39GND251GNDAF39GND255GNDAF39GND256GNDAF49GND258GNDAF14<	GND231	GND				
GND234GNDAD31GND235GNDAD33GND236GNDAD36GND237GNDAD39GND238GNDAD42GND239GNDAE3GND240GNDAE1GND241GNDAE14GND242GNDAE14GND243GNDAE14GND244GNDAE18GND245GNDAE20GND246GNDAE24GND247GNDAE24GND248GNDAE28GND250GNDAE38GND251GNDAE39GND255GNDAF3GND256GNDAF4GND258GNDAF4GND258GNDAF14GND258 <td>GND232</td> <td>GND</td> <td>AD27</td> <td></td> <td></td> <td></td>	GND232	GND	AD27			
GND235GNDAD33GND236GNDAD36GND237GNDAD39GND238GNDAD29GND239GNDAE3GND240GNDAE7GND241GNDAE9GND242GNDAE16GND243GNDAE18GND244GNDAE18GND245GNDAE20GND244GNDAE21GND245GNDAE24GND247GNDAE28GND248GNDAE28GND250GNDAE31GND251GNDAE39GND255GNDAF31GND256GNDAF4GND256GNDAF4GND257GNDAF4GND263GNDAF14 <td>GND233</td> <td>GND</td> <td>AD29</td> <td></td> <td></td> <td></td>	GND233	GND	AD29			
GND236GNDAD36GND237GNDAD39GND238GNDAD42GND239GNDAE3GND240GNDAE13GND241GNDAE14GND242GNDAE14GND243GNDAE16GND244GNDAE18GND245GNDAE10AE20GND246GNDAE20GND246GNDAE20GND246GNDAE20GND246GNDAE20GND247GNDAE20GND248GNDAE24GND250GNDAE33GND251GNDAE39GND255GNDAF39GND256GNDAF4GND263GNDAF14GND263	GND234	GND	AD31			
GND237GNDAD39GND238GNDAD42GND239GNDAE3GND240GNDAE7GND241GNDAE19GND242GNDAE14GND243GNDAE18GND244GNDAE20GND245GNDAE22GND246GNDAE24GND247GNDAE24GND248GNDAE28GND249GNDAE28GND250GNDAE33GND251GNDAE39GND255GNDAF4GND256GNDAF5GND259GNDAF6GND261GNDAF14GND262GNDAF19 </td <td>GND235</td> <td>GND</td> <td>AD33</td> <td></td> <td></td> <td></td>	GND235	GND	AD33			
GND238GNDAD42GND239GNDAE3GND240GNDAE7GND241GNDAE9GND242GNDAE14GND243GNDAE16GND244GNDAE16GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE28GND249GNDAE28GND250GNDAE33GND252GNDAE39GND253GNDAE39GND254GNDAF4GND255GNDAF3GND256GNDAF9GND259GNDAF9GND260GNDAF14GND263GNDAF19GND263GNDAF21 <td< td=""><td>GND236</td><td>GND</td><td>AD36</td><td></td><td></td><td></td></td<>	GND236	GND	AD36			
GND239GNDAE3GND240GNDAE7GND241GNDAE9GND242GNDAE14GND243GNDAE16GND244GNDAE18GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE24GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE39GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF9GND259GNDAF14GND263GNDAF14GND263GNDAF14GND263<	GND237	GND	AD39			
GND240GNDAE7GND241GNDAE9GND242GNDAE14GND243GNDAE16GND244GNDAE18GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE24GND247GNDAE23GND248GNDAE24GND248GNDAE31GND250GNDAE31GND251GNDAE33GND253GNDAE43GND255GNDAF4GND258GNDAF5GND259GNDAF19GND260GNDAF19GND263GNDAF19GND264GNDAF29GND265 <td>GND238</td> <td>GND</td> <td>AD42</td> <td></td> <td></td> <td></td>	GND238	GND	AD42			
GND241GNDAE9GND242GNDAE14GND243GNDAE16GND244GNDAE18GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE28GND249GNDAE28GND250GNDAE31GND251GNDAE36GND253GNDAE39GND254GNDAF3GND255GNDAF4GND258GNDAF4GND259GNDAF14GND260GNDAF14GND263GNDAF21GND263GNDAF23	GND239	GND	AE3			
GND242GNDAE14GND243GNDAE16GND244GNDAE18GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE28GND249GNDAE28GND250GNDAE31GND251GNDAE33GND253GNDAE39GND254GNDAF3GND255GNDAF4GND256GNDAF5GND257GNDAF5GND259GNDAF14GND260GNDAF14GND261GNDAF19GND263GNDAF19GND263GNDAF19GND263GNDAF23GND263GNDAF23	GND240	GND	AE7			
GND243GNDAE16GND244GNDAE18GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE26GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE39GND254GNDAE49GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF14GND260GNDAF14GND261GNDAF19GND263GNDAF19GND264GNDAF20GND265GNDAF21GND264GNDAF29	GND241	GND	AE9			
GND244GNDAE18GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE26GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE39GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND258GNDAF5GND259GNDAF9GND260GNDAF14GND261GNDAF17GND263GNDAF21GND264GNDAF23GND265GNDAF23GND266GNDAF23GND266GNDAF23GND266GNDAF29GND266GNDAF29 <t< td=""><td>GND242</td><td>GND</td><td>AE14</td><td></td><td></td><td></td></t<>	GND242	GND	AE14			
GND245GNDAE20GND246GNDAE22GND247GNDAE24GND248GNDAE26GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE39GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND258GNDAF5GND260GNDAF14GND261GNDAF17GND263GNDAF21GND264GNDAF23GND265GNDAF25GND266GNDAF29GND266GNDAF29GND266GNDAF29GND266 </td <td>GND243</td> <td>GND</td> <td>AE16</td> <td></td> <td></td> <td></td>	GND243	GND	AE16			
GND246GNDAE22GND247GNDAE24GND248GNDAE26GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE36GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND259GNDAF6GND260GNDAF14GND261GNDAF19GND262GNDAF23GND263GNDAF23GND264GNDAF23GND265GNDAF23GND266GNDAF23GND266GNDAF23GND266GNDAF23GND266GNDAF23GND266GNDAF29GND266GNDAF29<	GND244	GND	AE18			
GND247GNDAE24GND248GNDAE26GND249GNDAAE28GND250GNDAE31GND251GNDAE33GND252GNDAE36GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF5GND258GNDAF9GND260GNDAF14GND261GNDAF21GND263GNDAF23GND265GNDAF23GND266GNDAF27GND266GNDAF29GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267<	GND245	GND	AE20			
GND248GNDAE26GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE36GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF6GND259GNDAF14GND260GNDAF14GND261GNDAF17GND263GNDAF21GND265GNDAF25GND266GNDAF25GND266GNDAF29GND266GNDAF29GND266GNDAF29GND266GNDAF29GND266 <td>GND246</td> <td>GND</td> <td>AE22</td> <td></td> <td></td> <td></td>	GND246	GND	AE22			
GND249GNDAE28GND250GNDAE31GND251GNDAE33GND252GNDAE36GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF6GND259GNDAF6GND260GNDAF14GND263GNDAF19GND264GNDAF23GND265GNDAF25GND266GNDAF27GND266GNDAF29GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267 <td>GND247</td> <td>GND</td> <td>AE24</td> <td></td> <td></td> <td></td>	GND247	GND	AE24			
GND250GNDAE31GND251GNDAE33GND252GNDAE36GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF6GND259GNDAF14GND260GNDAF14GND261GNDAF19GND263GNDAF21GND263GNDAF25GND265GNDAF25GND266GNDAF27GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267GNDAF29GND267 <td>GND248</td> <td>GND</td> <td>AE26</td> <td></td> <td></td> <td></td>	GND248	GND	AE26			
GND251GNDAE33GND252GNDAE36GND253GNDAE39GND254GNDAE42GND255GNDAF33GND256GNDAF4GND257GNDAF5GND258GNDAF6GND259GNDAF14GND260GNDAF14GND261GNDAF19GND263GNDAF21GND264GNDAF23GND265GNDAF27GND267GNDAF29	GND249	GND	AE28			
GND252 GND AE36 GND253 GND AE39 GND254 GND AE42 GND255 GND AF3 GND256 GND AF3 GND256 GND AF4 GND257 GND AF5 GND258 GND AF6 GND259 GND AF9 GND260 GND AF14 GND262 GND AF17 GND263 GND AF21 GND263 GND AF23 GND264 GND AF25 <	GND250	GND	AE31			
GND253GNDAE39GND254GNDAE42GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF6GND259GNDAF9GND260GNDAF14GND261GNDAF19GND263GNDAF21GND264GNDAF23GND265GNDAF25GND266GNDAF27GND267GNDAF29	GND251	GND	AE33			
GND254GNDAE42GND255GNDAF3GND256GNDAF4GND257GNDAF5GND258GNDAF6GND259GNDAF9GND260GNDAF14GND261GNDAF17GND263GNDAF21GND264GNDAF25GND265GNDAF27GND266GNDAF29GND267GNDAF29	GND252	GND	AE36			
GND255 GND AF3 GND256 GND AF4 GND257 GND AF5 GND257 GND AF5 GND258 GND AF6 GND259 GND AF9 GND260 GND AF14 GND260 GND AF14 GND261 GND AF17 GND262 GND AF19 GND263 GND AF21 GND264 GND AF23 GND265 GND AF27 GND266 GND AF29 GND267 GND AF29 <	GND253	GND	AE39			
GND256 GND AF4 GND257 GND AF5 GND258 GND AF6 GND259 GND AF9 GND260 GND AF14 GND261 GND AF14 GND262 GND AF17 GND263 GND AF21 GND264 GND AF23 GND265 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND254	GND	AE42			
GND257GNDAF5GND258GNDAF6GND259GNDAF9GND260GNDAF14GND261GNDAF17GND262GNDAF19GND263GNDAF21GND264GNDAF23GND265GNDAF25GND266GNDAF27GND267GNDAF29	GND255	GND	AF3			
GND258 GND AF6 GND259 GND AF9 GND260 GND AF14 GND261 GND AF14 GND262 GND AF17 GND262 GND AF19 GND263 GND AF21 GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND256	GND	AF4			
GND259 GND AF9 GND260 GND AF14 GND261 GND AF17 GND262 GND AF19 GND263 GND AF19 GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND266 GND AF27 GND267 GND AF29	GND257	GND	AF5			
GND260 GND AF14 GND261 GND AF17 GND262 GND AF19 GND263 GND AF21 GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND258	GND	AF6			
GND261 GND AF17 GND262 GND AF19 GND263 GND AF21 GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND259	GND	AF9			
GND262 GND AF19 GND263 GND AF21 GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND260	GND	AF14			
GND263 GND AF21 GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND261	GND	AF17			
GND264 GND AF23 GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND262	GND	AF19			
GND265 GND AF25 GND266 GND AF27 GND267 GND AF29	GND263	GND	AF21			
GND266 GND AF27 GND267 GND AF29	GND264	GND	AF23			
GND267 GND AF29	GND265	GND	AF25			
	GND266	GND	AF27			
GND268 GND AF31	GND267	GND	AF29			
	GND268	GND	AF31			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND269	GND	AF33			
GND270	GND	AF36			
GND271	GND	AF39			
GND272	GND	AF42			
GND273	GND	AG3			
GND274	GND	AG8			
GND275	GND	AG11			
GND276	GND	AG12			
GND277	GND	AG13			
GND278	GND	AG14			
GND279	GND	AG16			
GND280	GND	AG18			
GND281	GND	AG20			
GND282	GND	AG22			
GND283	GND	AG24			
GND284	GND	AG26			
GND285	GND	AG28			
GND286	GND	AG31			
GND287	GND	AG33			
GND288	GND	AG36			
GND289	GND	AG39			
GND290	GND	AG42			
GND291	GND	AH3			
GND292	GND	AH9			
GND293	GND	AH12			
GND294	GND	AH14			
GND295	GND	AH17			
GND296	GND	AH19			
GND297	GND	AH21			
GND298	GND	AH23			
GND299	GND	AH25			
GND300	GND	AH27			
GND301	GND	AH29			
GND302	GND	AH31			
GND303	GND	AH33			
GND304	GND	AH36			
GND305	GND	AH39			
GND306	GND	AH42			

Table 1. Pinout list by bus (continued)

NumberNumberGND307GNDAJ3GND308GNDAJ4GND309GNDAJ6GND310GNDAJ6GND311GNDAJ7GND312GNDAJ8GND313GNDAJ9GND314GNDAJ10GND315GNDAJ11GND316GNDAJ14GND317GNDAJ14GND318GNDAJ16GND320GNDAJ18GND321GNDAJ30GND323GNDAJ33GND324GNDAJ36GND325GNDAJ39GND326GNDAJ39GND327GNDAJ42GND328GNDAJ39GND329GNDAK6GND320GNDAK6GND321GNDAJ42GND325GNDAJ39GND326GNDAJ39GND328GNDAK6GND331GNDAK12GND332GNDAK1	Notes
GND309 GND AJ5 GND310 GND AJ6 GND311 GND AJ7 GND312 GND AJ8 GND313 GND AJ9 GND314 GND AJ10 GND315 GND AJ11 GND316 GND AJ11 GND317 GND AJ14 GND318 GND AJ18 GND320 GND AJ20 GN321 GND AJ30 GN322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36	
GND310 GND AJ6 GND311 GND AJ7 GND312 GND AJ8 GND313 GND AJ9 GND313 GND AJ10 GND314 GND AJ10 GND315 GND AJ11 GND316 GND AJ11 GND317 GND AJ14 GND318 GND AJ18 GND320 GND AJ20 GND321 GND AJ34 GND323 GND AJ35 GND324 GND AJ34 GND325 GND AJ36 GND326 GND AJ38	
GND311 GND AJ7 GND312 GND AJ8 GND313 GND AJ9 GND314 GND AJ10 GND315 GND AJ11 GND316 GND AJ12 GND317 GND AJ14 GND318 GND AJ16 GND319 GND AJ18 GND320 GND AJ30 GND321 GND AJ30 GND322 GND AJ30 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AK3	
GND312 GND AJ8 GND313 GND AJ9 GND314 GND AJ10 GND315 GND AJ11 GND316 GND AJ12 GND317 GND AJ14 GND318 GND AJ16 GND319 GND AJ18 GND320 GND AJ30 GND321 GND AJ31 GND322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ32 GND328 GND AK3	
GND313 GND AJ9 GND314 GND AJ10 GND315 GND AJ11 GND316 GND AJ12 GND316 GND AJ14 GND317 GND AJ16 GND318 GND AJ18 GND320 GND AJ30 GND321 GND AJ31 GND322 GND AJ34 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331	
GND314GNDAJ10GND315GNDAJ11GND316GNDAJ12GND317GNDAJ14GND318GNDAJ16GND319GNDAJ18GND320GNDAJ20GND321GNDAJ30GND322GNDAJ31GND323GNDAJ36GND324GNDAJ36GND325GNDAJ39GND326GNDAJ39GND328GNDAK3GND329GNDAK3GND331GNDAK6GND331GNDAK12GND333GNDAK13GND334GNDAK15GND335GNDAK19	
GND315 GND AJ11 GND316 GND AJ12 GND317 GND AJ14 GND318 GND AJ16 GND319 GND AJ18 GND320 GND AJ20 GND321 GND AJ30 GND322 GND AJ31 GND322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ42 GND327 GND AK3 GND328 GND AK6 GND330 GND AK12 GND331 GND AK13 GND333	
GND316 GND AJ12 GND317 GND AJ14 GND318 GND AJ16 GND319 GND AJ18 GND320 GND AJ20 GND321 GND AJ30 GND322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331 GND AK13 GND333 GND AK13 GND333	
GND317 GND AJ14 GND318 GND AJ16 GND319 GND AJ18 GND320 GND AJ20 GND321 GND AJ30 GND321 GND AJ30 GND322 GND AJ31 GND323 GND AJ34 GND323 GND AJ36 GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK6 GND329 GND AK6 GND331 GND AK12 GND331 GND AK13 GND332 GND AK13 GND333	
GND318 GND AJ16 GND319 GND AJ18 GND320 GND AJ20 GND321 GND AJ30 GND322 GND AJ31 GND323 GND AJ34 GND323 GND AJ34 GND323 GND AJ35 GND324 GND AJ36 GND325 GND AJ39 GND326 GND AJ39 GND327 GND AK3 GND328 GND AK6 GND329 GND AK6 GND331 GND AK12 GND332 GND AK13 GND333 GND AK13 GND334	
GND319 GND AJ18 GND320 GND AJ20 GND321 GND AJ30 GND322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ42 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331 GND AK13 GND332 GND AK13 GND333 GND AK13 GND333 GND AK17 GND335 GND AK19	
GND320 GND AJ20 GND321 GND AJ30 GND322 GND AJ31 GND323 GND AJ34 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ42 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331 GND AK13 GND332 GND AK15 GND333 GND AK17 GND334 GND AK19	
GND321 GND AJ30 GND322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331 GND AK13 GND333 GND AK15 GND332 GND AK13 GND332 GND AK13 GND333 GND AK19 GND334 GND AK19	
GND322 GND AJ31 GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331 GND AK13 GND332 GND AK13 GND331 GND AK13 GND332 GND AK15 GND333 GND AK15 GND334 GND AK19 GND335 GND AK19	
GND323 GND AJ34 GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK12 GND331 GND AK12 GND333 GND AK13 GND333 GND AK19	
GND324 GND AJ35 GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK9 GND331 GND AK12 GND333 GND AK13 GND333 GND AK15 GND332 GND AK15 GND333 GND AK15 GND333 GND AK15 GND334 GND AK19 GND335 GND AK19	
GND325 GND AJ36 GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK9 GND331 GND AK12 GND332 GND AK13 GND332 GND AK13 GND333 GND AK15 GND333 GND AK19 GND334 GND AK19 GND335 GND AK19	
GND326 GND AJ39 GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK9 GND331 GND AK12 GND332 GND AK13 GND332 GND AK13 GND332 GND AK13 GND333 GND AK15 GND334 GND AK19 GND335 GND AK19	
GND327 GND AJ42 GND328 GND AK3 GND329 GND AK6 GND330 GND AK9 GND331 GND AK12 GND332 GND AK13 GND332 GND AK13 GND333 GND AK15 GND334 GND AK19 GND335 GND AK19	
GND328 GND AK3 GND329 GND AK6 GND330 GND AK9 GND331 GND AK12 GND332 GND AK13 GND333 GND AK15 GND333 GND AK15 GND334 GND AK19 GND335 GND AK19	
GND329 GND AK6 GND330 GND AK9 GND331 GND AK12 GND332 GND AK13 GND333 GND AK15 GND334 GND AK17 GND335 GND AK19	
GND330 GND AK9 GND331 GND AK12 GND332 GND AK13 GND333 GND AK15 GND334 GND AK17 GND335 GND AK19	
GND331 GND AK12 GND332 GND AK13 GND333 GND AK15 GND334 GND AK17 GND335 GND AK19	
GND332 GND AK13 GND333 GND AK15 GND334 GND AK17 GND335 GND AK19	
GND333 GND AK15 GND334 GND AK17 GND335 GND AK19	
GND334 GND AK17 GND335 GND AK19	
GND335 GND AK19	
GND336 GND AK33	
GND337 GND AK36	
GND338 GND AK37	
GND339 GND AK38	
GND340 GND AK39	
GND341 GND AK40	
GND342 GND AK41	
GND343 GND AK42	
GND344 GND AL3	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin	Pin type	Power supply	Notes
		number			
GND345	GND	AL6			
GND346	GND	AL9			
GND347	GND	AL12			
GND348	GND	AL33			
GND349	GND	AL34			
GND350	GND	AL39			
GND351	GND	AL42			
GND352	GND	AM3			
GND353	GND	AM6			
GND354	GND	AM9			
GND355	GND	AM12			
GND356	GND	AM33			
GND357	GND	AM36			
GND358	GND	AM39			
GND359	GND	AM42			
GND360	GND	AN3			
GND361	GND	AN6			
GND362	GND	AN9			
GND363	GND	AN12			
GND364	GND	AN34			
GND365	GND	AN35			
GND366	GND	AN36			
GND367	GND	AN39			
GND368	GND	AN42			
GND369	GND	AP3			
GND370	GND	AP6			
GND371	GND	AP9			
GND372	GND	AP12			
GND373	GND	AP14			
GND374	GND	AP15			
GND375	GND	AP16			
GND376	GND	AP37			
GND377	GND	AP39			
GND378	GND	AP42			
GND379	GND	AR3			
GND380	GND	AR6			
GND381	GND	AR9			
GND382	GND	AR12			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND383	GND	AR39			
GND384	GND	AR42			
GND385	GND	AT3			
GND386	GND	AT4			
GND387	GND	AT5			
GND388	GND	AT6			
GND389	GND	AT7			
GND390	GND	AT8			
GND391	GND	AT9			
GND392	GND	AT10			
GND393	GND	AT11			
GND394	GND	AT12			
GND395	GND	AT13			
GND396	GND	AT14			
GND397	GND	AT15			
GND398	GND	AT16			
GND399	GND	AT34			
GND400	GND	AT35			
GND401	GND	AT36			
GND402	GND	AT39			
GND403	GND	AT42			
GND404	GND	AU3			
GND405	GND	AU10			
GND406	GND	AU37			
GND407	GND	AU39			
GND408	GND	AU41			
GND409	GND	AU42			
GND410	GND	AV3			
GND411	GND	AV10			
GND412	GND	AV38			
GND413	GND	AV42			
GND414	GND	AW3			
GND415	GND	AW4			
GND416	GND	AW5			
GND417	GND	AW6			
GND418	GND	AW7			
GND419	GND	AW8			
GND420	GND	AW9			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND421	GND	AW10			
GND422	GND	AW11			
GND423	GND	AW12			
GND424	GND	AW13			
GND425	GND	AW14			
GND426	GND	AW15			
GND427	GND	AW16			
GND428	GND	AW35			
GND429	GND	AW36			
GND430	GND	AW38			
GND431	GND	AW40			
GND432	GND	AW42			
GND433	GND	AY3			
GND434	GND	AY10			
GND435	GND	AY38			
GND436	GND	AY42			
GND437	GND	BA3			
GND438	GND	BA10			
GND439	GND	BA34			
GND440	GND	BA38			
GND441	GND	BA42			
GND442	GND	BB4			
GND443	GND	BB5			
GND444	GND	BB6			
GND445	GND	BB7			
GND446	GND	BB8			
GND447	GND	BB9			
GND448	GND	BB10			
GND449	GND	BB11			
GND450	GND	BB12			
GND451	GND	BB13			
GND452	GND	BB14			
GND453	GND	BB15			
GND454	GND	BB16			
GND455	GND	BB35			
GND456	GND	BB36			
GND457	GND	BB37			
GND458	GND	BB38			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
GND459	GND	BB39			
GND460	GND	BB40			
GND461	GND	BB41			
GND_DET1	GND	C3			
GND_DET2	GND	BB3			
GND_DET3	GND	BB42			
USB_AGND1	USB PHY 1 Transceiver GND	D32			
USB_AGND2	USB PHY 1 Transceiver GND	J32			
USB_AGND3	USB PHY 1 Transceiver GND	L31			
USB_AGND4	USB PHY 1 Transceiver GND	M32			
USB_AGND5	USB PHY 1 Transceiver GND	N30			
X1GND01	Serdes 1 Transceiver GND	F15			
X1GND02	Serdes 1 Transceiver GND	F17			
X1GND03	Serdes 1 Transceiver GND	F19			
X1GND04	Serdes 1 Transceiver GND	F21			
X1GND05	Serdes 1 Transceiver GND	G14			
X1GND06	Serdes 1 Transceiver GND	G15			
X1GND07	Serdes 1 Transceiver GND	G17			
X1GND08	Serdes 1 Transceiver GND	G19			
X1GND09	Serdes 1 Transceiver GND	G21			
X1GND10	Serdes 1 Transceiver GND	H14			
X1GND11	Serdes 1 Transceiver GND	H16			
X1GND12	Serdes 1 Transceiver GND	H18			
X1GND13	Serdes 1 Transceiver GND	H20			
X1GND14	Serdes 1 Transceiver GND	H22			
X1GND15	Serdes 1 Transceiver GND	J14			
X1GND16	Serdes 1 Transceiver GND	J16			
X1GND17	Serdes 1 Transceiver GND	J18			
X1GND18	Serdes 1 Transceiver GND	J20			
X1GND19	Serdes 1 Transceiver GND	J22			
X1GND20	Serdes 1 Transceiver GND	K14			
X1GND21	Serdes 1 Transceiver GND	L15			
X1GND22	Serdes 1 Transceiver GND	L19			
X1GND23	Serdes 1 Transceiver GND	M17			
X1GND24	Serdes 1 Transceiver GND	M21			
X2GND01	Serdes 2 Transceiver GND	E31			
X2GND02	Serdes 2 Transceiver GND	F23			
X2GND03	Serdes 2 Transceiver GND	F25			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
X2GND04	Serdes 2 Transceiver GND	F27			
X2GND05	Serdes 2 Transceiver GND	F29			
X2GND06	Serdes 2 Transceiver GND	F31			
X2GND07	Serdes 2 Transceiver GND	G23			
X2GND08	Serdes 2 Transceiver GND	G25			
X2GND09	Serdes 2 Transceiver GND	G27			
X2GND10	Serdes 2 Transceiver GND	G29			
X2GND11	Serdes 2 Transceiver GND	G31			
X2GND12	Serdes 2 Transceiver GND	H24			
X2GND13	Serdes 2 Transceiver GND	H26			
X2GND14	Serdes 2 Transceiver GND	H28			
X2GND15	Serdes 2 Transceiver GND	H30			
X2GND16	Serdes 2 Transceiver GND	H31			
X2GND17	Serdes 2 Transceiver GND	J24			
X2GND18	Serdes 2 Transceiver GND	J26			
X2GND19	Serdes 2 Transceiver GND	J28			
X2GND20	Serdes 2 Transceiver GND	J30			
X2GND21	Serdes 2 Transceiver GND	L26			
X2GND22	Serdes 2 Transceiver GND	L30			
X2GND23	Serdes 2 Transceiver GND	M24			
X2GND24	Serdes 2 Transceiver GND	M28			
X2GND25	Serdes 2 Transceiver GND	N29			
X3GND01	Serdes 3 Transceiver GND	AN20			
X3GND02	Serdes 3 Transceiver GND	AN24			
X3GND03	Serdes 3 Transceiver GND	AP18			
X3GND04	Serdes 3 Transceiver GND	AP22			
X3GND05	Serdes 3 Transceiver GND	AR17			
X3GND06	Serdes 3 Transceiver GND	AT17			
X3GND07	Serdes 3 Transceiver GND	AT19			
X3GND08	Serdes 3 Transceiver GND	AT21			
X3GND09	Serdes 3 Transceiver GND	AT23			
X3GND10	Serdes 3 Transceiver GND	AT25			
X3GND11	Serdes 3 Transceiver GND	AU17			
X3GND12	Serdes 3 Transceiver GND	AU19			
X3GND13	Serdes 3 Transceiver GND	AU21			
X3GND14	Serdes 3 Transceiver GND	AU23			
X3GND15	Serdes 3 Transceiver GND	AU25			
X3GND16	Serdes 3 Transceiver GND	AV17			

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
X3GND17	Serdes 3 Transceiver GND	AV18			
X3GND18	Serdes 3 Transceiver GND	AV20			
X3GND19	Serdes 3 Transceiver GND	AV22			
X3GND20	Serdes 3 Transceiver GND	AV24			
X3GND21	Serdes 3 Transceiver GND	AW18			
X3GND22	Serdes 3 Transceiver GND	AW20			
X3GND23	Serdes 3 Transceiver GND	AW22			
X3GND24	Serdes 3 Transceiver GND	AW24			
X4GND01	Serdes 4 Transceiver GND	AM32			
X4GND02	Serdes 4 Transceiver GND	AN27			
X4GND03	Serdes 4 Transceiver GND	AN31			
X4GND04	Serdes 4 Transceiver GND	AP29			
X4GND05	Serdes 4 Transceiver GND	AP33			
X4GND06	Serdes 4 Transceiver GND	AT27			
X4GND07	Serdes 4 Transceiver GND	AT29			
X4GND08	Serdes 4 Transceiver GND	AT31			
X4GND09	Serdes 4 Transceiver GND	AT33			
X4GND10	Serdes 4 Transceiver GND	AU27			
X4GND11	Serdes 4 Transceiver GND	AU29			
X4GND12	Serdes 4 Transceiver GND	AU31			
X4GND13	Serdes 4 Transceiver GND	AU33			
X4GND14	Serdes 4 Transceiver GND	AU34			
X4GND15	Serdes 4 Transceiver GND	AV26			
X4GND16	Serdes 4 Transceiver GND	AV28			
X4GND17	Serdes 4 Transceiver GND	AV30			
X4GND18	Serdes 4 Transceiver GND	AV32			
X4GND19	Serdes 4 Transceiver GND	AV34			
X4GND20	Serdes 4 Transceiver GND	AW26			
X4GND21	Serdes 4 Transceiver GND	AW28			
X4GND22	Serdes 4 Transceiver GND	AW30			
X4GND23	Serdes 4 Transceiver GND	AW32			
X4GND24	Serdes 4 Transceiver GND	AW34			
X4GND25	Serdes 4 Transceiver GND	AY34			
S1GND01	Serdes 1 core logic GND	A15			
S1GND02	Serdes 1 core logic GND	A17			
S1GND03	Serdes 1 core logic GND	A19			
S1GND04	Serdes 1 core logic GND	A21			
S1GND05	Serdes 1 core logic GND	B14			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package	Pin	Power supply	Notes
		pin number	type		
S1GND06	Serdes 1 core logic GND	B15			
S1GND07	Serdes 1 core logic GND	B17			
S1GND08	Serdes 1 core logic GND	B19			
S1GND09	Serdes 1 core logic GND	B21			
S1GND10	Serdes 1 core logic GND	C14			
S1GND11	Serdes 1 core logic GND	C16			
S1GND12	Serdes 1 core logic GND	C18			
S1GND13	Serdes 1 core logic GND	C20			
S1GND14	Serdes 1 core logic GND	C22			
S1GND15	Serdes 1 core logic GND	D14			
S1GND16	Serdes 1 core logic GND	D16			
S1GND17	Serdes 1 core logic GND	D18			
S1GND18	Serdes 1 core logic GND	D20			
S1GND19	Serdes 1 core logic GND	D22			
S1GND20	Serdes 1 core logic GND	E14			
S1GND21	Serdes 1 core logic GND	E15			
S1GND22	Serdes 1 core logic GND	E16			
S1GND23	Serdes 1 core logic GND	E17			
S1GND24	Serdes 1 core logic GND	E18			
S1GND25	Serdes 1 core logic GND	E19			
S1GND26	Serdes 1 core logic GND	E20			
S1GND27	Serdes 1 core logic GND	E21			
S1GND28	Serdes 1 core logic GND	E22			
S1GND29	Serdes 1 core logic GND	M15			
S1GND30	Serdes 1 core logic GND	M19			
S1GND31	Serdes 1 core logic GND	N16			
S1GND32	Serdes 1 core logic GND	N18			
S1GND33	Serdes 1 core logic GND	N19			
S1GND34	Serdes 1 core logic GND	N20			
S1GND35	Serdes 1 core logic GND	N22			
S1GND36	Serdes 1 core logic GND	P17			
S1GND37	Serdes 1 core logic GND	R18			
S1GND38	Serdes 1 core logic GND	R19			
S1GND39	Serdes 1 core logic GND	R20			
S1GND40	Serdes 1 core logic GND	R21			
S1GND41	Serdes 1 core logic GND	R22			
S2GND01	Serdes 2 core logic GND	A23			
S2GND02	Serdes 2 core logic GND	A25			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
S2GND03	Serdes 2 core logic GND	A27			
S2GND04	Serdes 2 core logic GND	A29			
S2GND05	Serdes 2 core logic GND	A31			
S2GND06	Serdes 2 core logic GND	B23			
S2GND07	Serdes 2 core logic GND	B25			
S2GND08	Serdes 2 core logic GND	B27			
S2GND09	Serdes 2 core logic GND	B29			
S2GND10	Serdes 2 core logic GND	B31			
S2GND11	Serdes 2 core logic GND	C24			
S2GND12	Serdes 2 core logic GND	C26			
S2GND13	Serdes 2 core logic GND	C28			
S2GND14	Serdes 2 core logic GND	C30			
S2GND15	Serdes 2 core logic GND	C31			
S2GND16	Serdes 2 core logic GND	D24			
S2GND17	Serdes 2 core logic GND	D26			
S2GND18	Serdes 2 core logic GND	D28			
S2GND19	Serdes 2 core logic GND	D30			
S2GND20	Serdes 2 core logic GND	E23			
S2GND21	Serdes 2 core logic GND	E24			
S2GND22	Serdes 2 core logic GND	E25			
S2GND23	Serdes 2 core logic GND	E26			
S2GND24	Serdes 2 core logic GND	E27			
S2GND25	Serdes 2 core logic GND	E28			
S2GND26	Serdes 2 core logic GND	E29			
S2GND27	Serdes 2 core logic GND	E30			
S2GND28	Serdes 2 core logic GND	M26			
S2GND29	Serdes 2 core logic GND	N23			
S2GND30	Serdes 2 core logic GND	N25			
S2GND31	Serdes 2 core logic GND	N26			
S2GND32	Serdes 2 core logic GND	N27			
S2GND33	Serdes 2 core logic GND	N28			
S2GND34	Serdes 2 core logic GND	P28			
S2GND35	Serdes 2 core logic GND	R23			
S2GND36	Serdes 2 core logic GND	R24			
S2GND37	Serdes 2 core logic GND	R25			
S2GND38	Serdes 2 core logic GND	R26			
S2GND39	Serdes 2 core logic GND	R27			
S3GND01	Serdes 3 core logic GND	AK21			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
S3GND02	Serdes 3 core logic GND	AK22			
S3GND03	Serdes 3 core logic GND	AK23			
S3GND04	Serdes 3 core logic GND	AK24			
S3GND05	Serdes 3 core logic GND	AK25			
S3GND06	Serdes 3 core logic GND	AL20			
S3GND07	Serdes 3 core logic GND	AM19			
S3GND08	Serdes 3 core logic GND	AM21			
S3GND09	Serdes 3 core logic GND	AM22			
S3GND10	Serdes 3 core logic GND	AM23			
S3GND11	Serdes 3 core logic GND	AM25			
S3GND12	Serdes 3 core logic GND	AN18			
S3GND13	Serdes 3 core logic GND	AN22			
S3GND14	Serdes 3 core logic GND	AY17			
S3GND15	Serdes 3 core logic GND	AY18			
S3GND16	Serdes 3 core logic GND	AY19			
S3GND17	Serdes 3 core logic GND	AY20			
S3GND18	Serdes 3 core logic GND	AY21			
S3GND19	Serdes 3 core logic GND	AY22			
S3GND20	Serdes 3 core logic GND	AY23			
S3GND21	Serdes 3 core logic GND	AY24			
S3GND22	Serdes 3 core logic GND	AY25			
S3GND23	Serdes 3 core logic GND	BA17			
S3GND24	Serdes 3 core logic GND	BA19			
S3GND25	Serdes 3 core logic GND	BA21			
S3GND26	Serdes 3 core logic GND	BA23			
S3GND27	Serdes 3 core logic GND	BA25			
S3GND28	Serdes 3 core logic GND	BB17			
S3GND29	Serdes 3 core logic GND	BB19			
S3GND30	Serdes 3 core logic GND	BB21			
S3GND31	Serdes 3 core logic GND	BB23			
S3GND32	Serdes 3 core logic GND	BB25			
S3GND33	Serdes 3 core logic GND	BC17			
S3GND34	Serdes 3 core logic GND	BC18			
S3GND35	Serdes 3 core logic GND	BC20			
S3GND36	Serdes 3 core logic GND	BC22			
S3GND37	Serdes 3 core logic GND	BC24			
S3GND38	Serdes 3 core logic GND	BD18			
S3GND39	Serdes 3 core logic GND	BD20			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
S3GND40	Serdes 3 core logic GND	BD22			
S3GND41	Serdes 3 core logic GND	BD24			
S4GND01	Serdes 4 core logic GND	AK26			
S4GND02	Serdes 4 core logic GND	AK27			
S4GND03	Serdes 4 core logic GND	AK28			
S4GND04	Serdes 4 core logic GND	AK29			
S4GND05	Serdes 4 core logic GND	AK30			
S4GND06	Serdes 4 core logic GND	AL31			
S4GND07	Serdes 4 core logic GND	AM26			
S4GND08	Serdes 4 core logic GND	AM28			
S4GND09	Serdes 4 core logic GND	AM29			
S4GND10	Serdes 4 core logic GND	AM30			
S4GND11	Serdes 4 core logic GND	AM31			
S4GND12	Serdes 4 core logic GND	AN29			
S4GND13	Serdes 4 core logic GND	AY26			
S4GND14	Serdes 4 core logic GND	AY27			
S4GND15	Serdes 4 core logic GND	AY28			
S4GND16	Serdes 4 core logic GND	AY29			
S4GND17	Serdes 4 core logic GND	AY30			
S4GND18	Serdes 4 core logic GND	AY31			
S4GND19	Serdes 4 core logic GND	AY32			
S4GND20	Serdes 4 core logic GND	AY33			
S4GND21	Serdes 4 core logic GND	BA27			
S4GND22	Serdes 4 core logic GND	BA29			
S4GND23	Serdes 4 core logic GND	BA31			
S4GND24	Serdes 4 core logic GND	BA33			
S4GND25	Serdes 4 core logic GND	BB27			
S4GND26	Serdes 4 core logic GND	BB29			
S4GND27	Serdes 4 core logic GND	BB31			
S4GND28	Serdes 4 core logic GND	BB33			
S4GND29	Serdes 4 core logic GND	BB34			
S4GND30	Serdes 4 core logic GND	BC26			
S4GND31	Serdes 4 core logic GND	BC28			
S4GND32	Serdes 4 core logic GND	BC30			
S4GND33	Serdes 4 core logic GND	BC32			
S4GND34	Serdes 4 core logic GND	BC34			
S4GND35	Serdes 4 core logic GND	BD26			
S4GND36	Serdes 4 core logic GND	BD28			

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
S4GND37	Serdes 4 core logic GND	BD30			
S4GND38	Serdes 4 core logic GND	BD32			
S4GND39	Serdes 4 core logic GND	BD34			
AGND_SD1_PLL1	Serdes1 PLL 1 GND	L18			
AGND_SD1_PLL2	Serdes1 PLL 2 GND	L20			
AGND_SD2_PLL1	Serdes2 PLL 1 GND	L25			
AGND_SD2_PLL2	Serdes2 PLL 2 GND	L27			
AGND_SD3_PLL1	Serdes3 PLL 1 GND	AP21			
AGND_SD3_PLL2	Serdes3 PLL 2 GND	AP23			
AGND_SD4_PLL1	Serdes4 PLL 1 GND	AP28			
AGND_SD4_PLL2	Serdes4 PLL 2 GND	AP30			
SENSEGND_CA	GND Sense pin	R13			
SENSEGND_CB	GND Sense pin	AM17			
SENSEGND_CC	GND Sense pin	AA33			
SENSEGND_PL	GND Sense pin	AA32			
OVDD1	General I/O supply	R30		OV _{DD}	
OVDD2	General I/O supply	T30		OV _{DD}	
OVDD3	General I/O supply	U30		OV _{DD}	
OVDD4	General I/O supply	V30		OV _{DD}	
OVDD5	General I/O supply	W30		OV _{DD}	
OVDD6	General I/O supply	Y30		OV _{DD}	
OVDD7	General I/O supply	AA30		OV _{DD}	
OVDD8	General I/O supply	AK14		OV _{DD}	
DVDD1	UART/I2C supply	R14		DV _{DD}	
DVDD2	UART/I2C supply	R15		DV _{DD}	
G1VDD01	DDR supply for port 1	A2		G1V _{DD}	
G1VDD02	DDR supply for port 1	A6		G1V _{DD}	
G1VDD03	DDR supply for port 1	A9		G1V _{DD}	
G1VDD04	DDR supply for port 1	B1		G1V _{DD}	
G1VDD05	DDR supply for port 1	B4		G1V _{DD}	
G1VDD06	DDR supply for port 1	B9		G1V _{DD}	
G1VDD07	DDR supply for port 1	D2		G1V _{DD}	
G1VDD08	DDR supply for port 1	F1		G1V _{DD}	
G1VDD09	DDR supply for port 1	H2		G1V _{DD}	
G1VDD10	DDR supply for port 1	K1		G1V _{DD}	
G1VDD11	DDR supply for port 1	M2		G1V _{DD}	
G1VDD12	DDR supply for port 1	P1		G1V _{DD}	
G1VDD13	DDR supply for port 1	T2		G1V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
G1VDD14	DDR supply for port 1	T15		G1V _{DD}	
G1VDD15	DDR supply for port 1	U15		G1V _{DD}	
G1VDD16	DDR supply for port 1	V1		G1V _{DD}	
G1VDD17	DDR supply for port 1	V15		G1V _{DD}	
G1VDD18	DDR supply for port 1	W15		G1V _{DD}	
G1VDD19	DDR supply for port 1	Y2		G1V _{DD}	
G1VDD20	DDR supply for port 1	Y15		G1V _{DD}	
G1VDD21	DDR supply for port 1	AA15		G1V _{DD}	
G1VDD22	DDR supply for port 1	AB1		G1V _{DD}	
G1VDD23	DDR supply for port 1	AB15		G1V _{DD}	
G1VDD24	DDR supply for port 1	AD2		G1V _{DD}	
G1VDD25	DDR supply for port 1	AF1		G1V _{DD}	
G1VDD26	DDR supply for port 1	AF2		G1V _{DD}	
G2VDD01	DDR supply for port 2	AC15		G2V _{DD}	
G2VDD02	DDR supply for port 2	AD15		G2V _{DD}	
G2VDD03	DDR supply for port 2	AE15		G2V _{DD}	
G2VDD04	DDR supply for port 2	AF15		G2V _{DD}	
G2VDD05	DDR supply for port 2	AG1		G2V _{DD}	
G2VDD06	DDR supply for port 2	AG15		G2V _{DD}	
G2VDD07	DDR supply for port 2	AH15		G2V _{DD}	
G2VDD08	DDR supply for port 2	AJ2		G2V _{DD}	
G2VDD09	DDR supply for port 2	AJ15		G2V _{DD}	
G2VDD10	DDR supply for port 2	AL1		G2V _{DD}	
G2VDD11	DDR supply for port 2	AN2		G2V _{DD}	
G2VDD12	DDR supply for port 2	AR1		G2V _{DD}	
G2VDD13	DDR supply for port 2	AU2		G2V _{DD}	
G2VDD14	DDR supply for port 2	AW1		G2V _{DD}	
G2VDD15	DDR supply for port 2	BA2		G2V _{DD}	
G2VDD16	DDR supply for port 2	BC1		G2V _{DD}	
G2VDD17	DDR supply for port 2	BC4		G2V _{DD}	
G2VDD18	DDR supply for port 2	BC8		G2V _{DD}	
G2VDD19	DDR supply for port 2	BC12		G2V _{DD}	
G2VDD20	DDR supply for port 2	BC16		G2V _{DD}	
G2VDD21	DDR supply for port 2	BD2		G2V _{DD}	
G2VDD22	DDR supply for port 2	BD6		G2V _{DD}	
G2VDD23	DDR supply for port 2	BD10		G2V _{DD}	
G2VDD24	DDR supply for port 2	BD14		G2V _{DD}	
G2VDD25	DDR supply for port 2	BD17		G2V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
G3VDD01	DDR supply for port 3	Y43		G3V _{DD}	
G3VDD02	DDR supply for port 3	Y44		G3V _{DD}	
G3VDD03	DDR supply for port 3	AB30		G3V _{DD}	
G3VDD04	DDR supply for port 3	AB43		G3V _{DD}	
G3VDD05	DDR supply for port 3	AC30		G3V _{DD}	
G3VDD06	DDR supply for port 3	AD30		G3V _{DD}	
G3VDD07	DDR supply for port 3	AD44		G3V _{DD}	
G3VDD08	DDR supply for port 3	AE30		G3V _{DD}	
G3VDD09	DDR supply for port 3	AF30		G3V _{DD}	
G3VDD10	DDR supply for port 3	AF43		G3V _{DD}	
G3VDD11	DDR supply for port 3	AG30		G3V _{DD}	
G3VDD12	DDR supply for port 3	AH30		G3V _{DD}	
G3VDD13	DDR supply for port 3	AH44		G3V _{DD}	
G3VDD14	DDR supply for port 3	AK43		G3V _{DD}	
G3VDD15	DDR supply for port 3	AM44		G3V _{DD}	
G3VDD16	DDR supply for port 3	AP43		G3V _{DD}	
G3VDD17	DDR supply for port 3	AT44		G3V _{DD}	
G3VDD18	DDR supply for port 3	AV43		G3V _{DD}	
G3VDD19	DDR supply for port 3	AY44		G3V _{DD}	
G3VDD20	DDR supply for port 3	BB43		G3V _{DD}	
G3VDD21	DDR supply for port 3	BC37		G3V _{DD}	
G3VDD22	DDR supply for port 3	BC41		G3V _{DD}	
G3VDD23	DDR supply for port 3	BC44		G3V _{DD}	
G3VDD24	DDR supply for port 3	BD35		G3V _{DD}	
G3VDD25	DDR supply for port 3	BD39		G3V _{DD}	
G3VDD26	DDR supply for port 3	BD43		G3V _{DD}	
S1VDD1	SerDes1 core logic supply	N17		S1V _{DD}	
S1VDD2	SerDes1 core logic supply	P20		S1V _{DD}	
S1VDD3	SerDes1 core logic supply	P22		S1V DD	
S1VDD4	SerDes1 core logic supply	T19		S1V _{DD}	
S1VDD5	SerDes1 core logic supply	T20		S1V _{DD}	
S1VDD6	SerDes1 core logic supply	T21		S1V _{DD}	
S1VDD7	SerDes1 core logic supply	T22		S1V _{DD}	
S2VDD1	SerDes2 core logic supply	P23		S2V _{DD}	
S2VDD2	SerDes2 core logic supply	P25		S2V _{DD}	
S2VDD3	SerDes2 core logic supply	T23		S2V _{DD}	
S2VDD4	SerDes2 core logic supply	T24		S2V _{DD}	
S2VDD5	SerDes2 core logic supply	T25		S2V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
S2VDD6	SerDes2 core logic supply	T26		S2V _{DD}	
S3VDD1	SerDes3 core logic supply	AJ22		S3V _{DD}	
S3VDD2	SerDes3 core logic supply	AJ23		S3V _{DD}	
S3VDD3	SerDes3 core logic supply	AJ24		S3V _{DD}	
S3VDD4	SerDes3 core logic supply	AJ25		S3V _{DD}	
S3VDD5	SerDes3 core logic supply	AL23		S3V _{DD}	
S3VDD6	SerDes3 core logic supply	AL25		S3V _{DD}	
S3VDD7	SerDes3 core logic supply	AM20		S3V _{DD}	
S4VDD1	SerDes4 core logic supply	AJ26		S4V _{DD}	
S4VDD2	SerDes4 core logic supply	AJ27		S4V _{DD}	
S4VDD3	SerDes4 core logic supply	AJ28		S4V _{DD}	
S4VDD4	SerDes4 core logic supply	AJ29		S4V _{DD}	
S4VDD5	SerDes4 core logic supply	AL26		S4V _{DD}	
S4VDD6	SerDes4 core logic supply	AL28		S4V _{DD}	
X1VDD1	SerDes1 transceiver supply	K15		X1V _{DD}	
X1VDD2	SerDes1 transceiver supply	K16		X1V _{DD}	
X1VDD3	SerDes1 transceiver supply	K17		X1V _{DD}	
X1VDD4	SerDes1 transceiver supply	K18		X1V _{DD}	
X1VDD5	SerDes1 transceiver supply	K19		X1V _{DD}	
X1VDD6	SerDes1 transceiver supply	K20		X1V _{DD}	
X1VDD7	SerDes1 transceiver supply	K21		X1V _{DD}	
X1VDD8	SerDes1 transceiver supply	K22		X1V _{DD}	
X2VDD1	SerDes2 transceiver supply	K23		X2V DD	
X2VDD2	SerDes2 transceiver supply	K24		X2V _{DD}	
X2VDD3	SerDes2 transceiver supply	K25		X2V _{DD}	
X2VDD4	SerDes2 transceiver supply	K26		X2V DD	
X2VDD5	SerDes2 transceiver supply	K27		X2V _{DD}	
X2VDD6	SerDes2 transceiver supply	K28		X2V _{DD}	
X2VDD7	SerDes2 transceiver supply	K29		X2V DD	
X2VDD8	SerDes2 transceiver supply	K30		X2V DD	
X2VDD9	SerDes2 transceiver supply	M30		X2V _{DD}	
X3VDD1	SerDes3 transceiver supply	AR18		X3V _{DD}	
X3VDD2	SerDes3 transceiver supply	AR19		X3V _{DD}	
X3VDD3	SerDes3 transceiver supply	AR20		X3V _{DD}	
X3VDD4	SerDes3 transceiver supply	AR21		X3V _{DD}	
X3VDD5	SerDes3 transceiver supply	AR22		X3V _{DD}	
X3VDD6	SerDes3 transceiver supply	AR23		X3V _{DD}	
X3VDD7	SerDes3 transceiver supply	AR24		X3V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
X3VDD8	SerDes3 transceiver supply	AR25		X3V _{DD}	
X4VDD1	SerDes4 transceiver supply	AN33		X4V _{DD}	
X4VDD2	SerDes4 transceiver supply	AR26		X4V _{DD}	
X4VDD3	SerDes4 transceiver supply	AR27		X4V _{DD}	
X4VDD4	SerDes4 transceiver supply	AR28		X4V _{DD}	
X4VDD5	SerDes4 transceiver supply	AR29		X4V _{DD}	
X4VDD6	SerDes4 transceiver supply	AR30		X4V _{DD}	
X4VDD7	SerDes4 transceiver supply	AR31		X4V _{DD}	
X4VDD8	SerDes4 transceiver supply	AR32		X4V _{DD}	
X4VDD9	SerDes4 transceiver supply	AR33		X4V _{DD}	
LVDD1	Ethernet controller and GPIO supply	L14		LV _{DD}	
LVDD2	Ethernet controller and GPIO supply	R16		LV _{DD}	
LVDD3	Ethernet controller and GPIO supply	R17		LV _{DD}	
FA_VL	Reserved for internal use only	R33		FA_VL	15
PROG_MTR	Reserved for internal use only	T29		PROG_MTR	15
PROG_SFP	SFP Fuse Programming Override supply	R29		PROG_SFP	
TH_VDD	Thermal Monitor Unit supply	V32		TH_V _{DD}	27
VDD01	Supply for cores and platform	T16		V _{DD}	
VDD02	Supply for cores and platform	U17		V _{DD}	
VDD03	Supply for cores and platform	U19		V _{DD}	
VDD04	Supply for cores and platform	U21		V _{DD}	
VDD05	Supply for cores and platform	U23		V _{DD}	
VDD06	Supply for cores and platform	U25		V _{DD}	
VDD07	Supply for cores and platform	U27		V _{DD}	
VDD08	Supply for cores and platform	U29		V _{DD}	
VDD09	Supply for cores and platform	V16		V _{DD}	
VDD10	Supply for cores and platform	V18		V _{DD}	
VDD11	Supply for cores and platform	V20		V _{DD}	
VDD12	Supply for cores and platform	V22		V _{DD}	
VDD13	Supply for cores and platform	V24		V _{DD}	
VDD14	Supply for cores and platform	V26		V _{DD}	
VDD15	Supply for cores and platform	V28		V _{DD}	
VDD16	Supply for cores and platform	W17		V _{DD}	
VDD17	Supply for cores and platform	W19		V _{DD}	
VDD18	Supply for cores and platform	W21		V _{DD}	

 Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
VDD19	Supply for cores and platform	W23		V _{DD}	
VDD20	Supply for cores and platform	W25		V _{DD}	
VDD21	Supply for cores and platform	W27		V _{DD}	
VDD22	Supply for cores and platform	W29		V _{DD}	
VDD23	Supply for cores and platform	Y16		V _{DD}	
VDD24	Supply for cores and platform	Y18		V _{DD}	
VDD25	Supply for cores and platform	Y20		V _{DD}	
VDD26	Supply for cores and platform	Y22		V _{DD}	
VDD27	Supply for cores and platform	Y24		V _{DD}	
VDD28	Supply for cores and platform	Y26		V _{DD}	
VDD29	Supply for cores and platform	Y28		V _{DD}	
VDD30	Supply for cores and platform	AA17		V _{DD}	
VDD31	Supply for cores and platform	AA19		V _{DD}	
VDD32	Supply for cores and platform	AA21		V _{DD}	
VDD33	Supply for cores and platform	AA23		V _{DD}	
VDD34	Supply for cores and platform	AA25		V _{DD}	
VDD35	Supply for cores and platform	AA27		V _{DD}	
VDD36	Supply for cores and platform	AA29		V _{DD}	
VDD37	Supply for cores and platform	AB16		V _{DD}	
VDD38	Supply for cores and platform	AB18		V _{DD}	
VDD39	Supply for cores and platform	AB20		V _{DD}	
VDD40	Supply for cores and platform	AB22		V _{DD}	
VDD41	Supply for cores and platform	AB24		V _{DD}	
VDD42	Supply for cores and platform	AB26		V _{DD}	
VDD43	Supply for cores and platform	AB28		V _{DD}	
VDD44	Supply for cores and platform	AC17		V _{DD}	
VDD45	Supply for cores and platform	AC19		V _{DD}	
VDD46	Supply for cores and platform	AC21		V _{DD}	
VDD47	Supply for cores and platform	AC23		V _{DD}	
VDD48	Supply for cores and platform	AC25		V _{DD}	
VDD49	Supply for cores and platform	AC27		V _{DD}	
VDD50	Supply for cores and platform	AC29		V _{DD}	
VDD51	Supply for cores and platform	AD16		V _{DD}	
VDD52	Supply for cores and platform	AD18		V _{DD}	
VDD53	Supply for cores and platform	AD20		V _{DD}	
VDD54	Supply for cores and platform	AD22		V _{DD}	
VDD55	Supply for cores and platform	AD24		V _{DD}	
VDD56	Supply for cores and platform	AD26		V _{DD}	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
VDD57	Supply for cores and platform	AD28		V _{DD}	
VDD58	Supply for cores and platform	AE17		V _{DD}	
VDD59	Supply for cores and platform	AE19		V _{DD}	
VDD60	Supply for cores and platform	AE21		V _{DD}	
VDD61	Supply for cores and platform	AE23		V _{DD}	
VDD62	Supply for cores and platform	AE25		V _{DD}	
VDD63	Supply for cores and platform	AE27		V _{DD}	
VDD64	Supply for cores and platform	AE29		V _{DD}	
VDD65	Supply for cores and platform	AF16		V _{DD}	
VDD66	Supply for cores and platform	AF18		V _{DD}	
VDD67	Supply for cores and platform	AF20		V _{DD}	
VDD68	Supply for cores and platform	AF22		V _{DD}	
VDD69	Supply for cores and platform	AF24		V _{DD}	
VDD70	Supply for cores and platform	AF26		V _{DD}	
VDD71	Supply for cores and platform	AF28		V _{DD}	
VDD72	Supply for cores and platform	AG17		V _{DD}	
VDD73	Supply for cores and platform	AG19		V _{DD}	
VDD74	Supply for cores and platform	AG21		V _{DD}	
VDD75	Supply for cores and platform	AG23		V _{DD}	
VDD76	Supply for cores and platform	AG25		V _{DD}	
VDD77	Supply for cores and platform	AG27		V _{DD}	
VDD78	Supply for cores and platform	AG29		V _{DD}	
VDD79	Supply for cores and platform	AH16		V _{DD}	
VDD80	Supply for cores and platform	AH18		V _{DD}	
VDD81	Supply for cores and platform	AH20		V _{DD}	
VDD82	Supply for cores and platform	AH22		V _{DD}	
VDD83	Supply for cores and platform	AH24		V _{DD}	
VDD84	Supply for cores and platform	AH26		V _{DD}	
VDD85	Supply for cores and platform	AH28		V _{DD}	
VDD86	Supply for cores and platform	AJ17		V _{DD}	
VDD87	Supply for cores and platform	AJ19		V _{DD}	
VDD88	Supply for cores and platform	AJ21		V _{DD}	
VDD89	Supply for cores and platform	AK16		V _{DD}	
VDD90	Supply for cores and platform	AK18		V _{DD}	
VDD91	Supply for cores and platform	AK20		V _{DD}	
VDD_LP	Low Power Security Monitor supply	R28		V _{DD _} LP	
AVDD_CGA1	e6500 Cluster Group A PLL1 supply	AP13		AVDD_CGA1	

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
AVDD_CGA2	e6500 Cluster Group A PLL2 supply	AR13		AVDD_CGA2	
AVDD_CGA3	e6500 Cluster Group A PLL3 supply	AR14		AVDD_CGA3	
AVDD_CGB1	e6500 Cluster Group B PLL1 supply	AR16		AVDD_CGB1	
AVDD_CGB2	e6500 Cluster Group B PLL2 supply	AR15		AVDD_CGB2	
AVDD_PLAT	Platform PLL supply	T28		AVDD_PLAT	
AVDD_D1	DDR1 PLL supply	T13		AVDD_D1	
AVDD_D2	DDR2 PLL supply	AJ13		AVDD_D2	
AVDD_D3	DDR3 PLL supply	AC32		AVDD_D3	
AVDD_SD1_PLL1	SerDes1 PLL 1 supply	L17		AVDD_SD1_PLL1	
AVDD_SD1_PLL2	SerDes1 PLL 2 supply	L21		AVDD_SD1_PLL2	
AVDD_SD2_PLL1	SerDes2 PLL 1 supply	L24		AVDD_SD2_PLL1	
AVDD_SD2_PLL2	SerDes2 PLL 2 supply	L28		AVDD_SD2_PLL2	
AVDD_SD3_PLL1	SerDes3 PLL 1 supply	AP20		AVDD_SD3_PLL1	
AVDD_SD3_PLL2	SerDes3 PLL 2 supply	AP24		AVDD_SD3_PLL2	
AVDD_SD4_PLL1	SerDes4 PLL 1 supply	AP27		AVDD_SD4_PLL1	
AVDD_SD4_PLL2	SerDes4 PLL 2 supply	AP31		AVDD_SD4_PLL2	
SENSEVDD_CA	Vdd Sense pin for core cluster A	R12		SENSEVDD_CA	
SENSEVDD_CB	Vdd Sense pin for core cluster B	AM16		SENSEVDD_CB	
SENSEVDD_CC	Vdd Sense pin for core cluster C	Y33		SENSEVDD_CC	
SENSEVDD_PL	Vdd Sense pin for platform	Y32		SENSEVDD_PL	
USB_HVDD1	USB PHY Transceiver 3.3V Supply	K32		USB_HV _{DD}	
USB_HVDD2	USB PHY Transceiver 3.3V Supply	N32		USB_HV _{DD}	
USB_OVDD1	USB PHY Transceiver 1.8V Supply	P31		USB_OV DD	
USB_OVDD2	USB PHY Transceiver 1.8V Supply	P32		USB_OV _{DD}	
USB_SVDD1	USB PHY Analog 1.0V Supply	P29		USB_SV _{DD}	
USB_SVDD2	USB PHY Analog 1.0V Supply	P30		USB_SV DD	
	No Connection	n Pins	1		- 1
NC01	No Connection	G35			12
NC02	No Connection	G36			12
NC03	No Connection	G37			12

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin	Pin type	Power supply	Notes
		number			
NC04	No Connection	G38			12
NC05	No Connection	H33			12
NC06	No Connection	H35			12
NC07	No Connection	H36			12
NC08	No Connection	H38			12
NC09	No Connection	J33			12
NC10	No Connection	J34			12
NC11	No Connection	J35			12
NC12	No Connection	J36			12
NC13	No Connection	J37			12
NC14	No Connection	J38			12
NC15	No Connection	K34			12
NC16	No Connection	K35			12
NC17	No Connection	K37			12
NC18	No Connection	K38			12
NC19	No Connection	L33			12
NC20	No Connection	L34			12
NC21	No Connection	L35			12
NC22	No Connection	L36			12
NC23	No Connection	L37			12
NC24	No Connection	L38			12
NC25	No Connection	M33			12
NC26	No Connection	M35			12
NC27	No Connection	M36			12
NC28	No Connection	M37			12
NC29	No Connection	M38			12
NC30	No Connection	N33			12
NC31	No Connection	N34			12
NC32	No Connection	N35			12
NC33	No Connection	N36			12
NC34	No Connection	N38			12
NC35	No Connection	P34			12
NC36	No Connection	P35			12
NC37	No Connection	P36			12
NC38	No Connection	P37			12
NC39	No Connection	P38			12
NC40	No Connection	R34			12
NC41	No Connection	R35			12

Table 1. Pinout list by bus (continued)

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
NC42	No Connection	R36			12
NC43	No Connection	R37			12
NC44	No Connection	R38			12
NC45	No Connection	T18			12
NC46	No Connection	T33			12
NC47	No Connection	T34			12
NC48	No Connection	T35			12
NC49	No Connection	T36			12
NC50	No Connection	T38			12
NC51	No Connection	U32			12
NC52	No Connection	U35			12
NC53	No Connection	U36			12
NC54	No Connection	U37			12
NC55	No Connection	U38			12
NC56	No Connection	V35			12
NC57	No Connection	V36			12
NC58	No Connection	V37			12
NC59	No Connection	V38			12
NC60	No Connection	W34			12
NC61	No Connection	W35			12
NC62	No Connection	W36			12
NC63	No Connection	W38			12
NC64	No Connection	AE32			12
NC65	No Connection	AG32			12
NC66	No Connection	AH32			12
NC67	No Connection	AJ32			12
NC68	No Connection	AJ33			12
NC69	No Connection	AK31			12
NC70	No Connection	AK32			12
NC71	No Connection	AL15			12
NC72	No Connection	AL18			12
NC73	No Connection	AL19			12
NC74	No Connection	AL32			12
NC75	No Connection	AM14			12
NC76	No Connection	AM15			12
NC77	No Connection	AM18			12
NC78	No Connection	AN13			12
NC79	No Connection	AN14			12

Table 1. Pinout list by bus (continued)

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Signal	Signal description	Package pin number	Pin type	Power supply	Notes
NC80	No Connection	AN15			12
NC81	No Connection	AN16			12
NC82	No Connection	AN17			12
NC83	No Connection	AP17			12
NC84	No Connection	AW17			12
NC_DET	No Connection	C42			12

 Table 1. Pinout list by bus (continued)

1. Functionally, this pin is an output or an input, but structurally it is an I/O because it either samples configuration input during reset, is a muxed pin, or has other manufacturing test functions. This pin will therefore be described as an I/O for boundary scan.

2. This output is actively driven during reset rather than being tri-stated during reset.

3. MDIC[0] is grounded through an 237 Ω (for Rev. 1) or 187 Ω (for Rev. 2) precision 1% resistor and MDIC[1] is connected to GV _{DD} through an 237 Ω (for Rev. 1) or 187 Ω (for Rev. 2) precision 1% resistor. For either full or half driver strength calibration of DDR I/Os, use the same MDIC resistor value of 237 Ω (for Rev. 1) or 187 Ω (for Rev. 2). Memory controller register setting can be used to determine automatic calibration is done to full or half drive strength. These pins are used for automatic calibration of the DDR3/ DDR3L IOs. The MDIC[0:1] pins must be connected to 237 Ω (for Rev. 1) or 187 Ω (for Rev. 2) precision 1% resistors.

4. This pin is a reset configuration pin. It has a weak (~20 k Ω) internal pull-up P-FET that is enabled only when the processor is in its reset state. This pull-up is designed such that it can be overpowered by an external 4.7 k Ω resistor. However, if the signal is intended to be high after reset, and if there is any device on the net that might pull down the value of the net at reset, a pull-up or active driver is needed.

5. Pin must **NOT** be pulled down during power-on reset. This pin may be pulled up, driven high, or if there are any externally connected devices, left in tristate. If this pin is connected to a device that pulls down during reset, an external pull-up is required to drive this pin to a safe state during reset.

6. Recommend that a weak pull-up resistor (2 to $10 \text{ k}\Omega$) be placed on this pin to the respective power supply, or appropriate pull up resistor value for signals like HRESET_B which might require 1 k Ω .

7. This pin is an open-drain signal.

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Pin assignments

8. Recommend a pull-up resistor be placed on this pin to the respective power supply. In the I2C interface, the value of the resistor should be calculated such that maximum rise time stays under 300 ns as well as VOL be under 0.4 V at IOL = 3 mA IOL and I2C load capacitance which should not exceed 400 pF.

9. This pin has a weak (~20 k Ω) internal pull-up P-FET that is always enabled.

10. These are test signals for factory use only and must be pulled up (100 Ω to 1 k Ω) to the respective power supply for normal operation.

11. This pin requires a 200 Ω pull-up to respective power supply.

12. Do not connect. These pins should be left floating.

13. These pins must be pulled up to 1.2 V through a 180 $\Omega \pm 1\%$ resistor for MDC and a 330 $\Omega \pm 1\%$ resistor for MDIO.

14. This pin requires an external 1 k Ω pull-down resistor to prevent PHY from seeing a valid Transmit Enable before it is actively driven.

15. These pins must be pulled to ground (GND).

16. This pin requires a 698 Ω pull-up to respective power supply.

18. Recommend that a weak pull-up resistor (4.7 k Ω) be placed on this pin to the respective power supply.

19. These pins should be tied to ground if the diode is not utilized for temperature monitoring.

20. This pin requires a pull-up of 10 to 50 k Ω to its corresponding I/O supply if it is not a GPIO or not used as one.

21. This pin always needs to be either pulled up by 10 to 50 k Ω or down by 4.7 k Ω to GND, depending on the intended RCW setting to be high or low, respectively.

22. If used as SDHC signal, pull-up 10 to 100 k Ω to the respective I/O supply.

23. New board designs should leave a place holder for a series resistor and capacitor filter, which is in parallel and very close proximity to a 1%, 10 k Ω resistor pulling USB_IBIAS_REXT low. This allows the flexibility of populating them if needed to avoid board coupled noise to this pin. An SMD ceramic 100 nF low ESL in series with 100 Ω SMD resistor will do the filtration needed with slight variations that suit each board case.

24. The non-ideality factor over temperature range $85C^{\circ}$ to $105C^{\circ}$, n = 1.006 ± 0.003, with approximate error +/- 1 C^o and approximate error under +/- 3 C^o for temperature range 0 C^o to $85C^{\circ}$.

25. In GPCM mode, this pin also serves as IFC_WE1_B.

26. T4240/T4160/T4080 Rev. 2 silicon requires SDHC_CD_B and SDHC_WP signals even when eMMC/eSDHC is used.

27. TH_VDD is a quiet power domain used for the Thermal Unit. Despite being defeatured, it should be connected to a quiet recommended supply level.

28. When Dn_MDQS_B[9:17] pins are not used; terminate with 50 Ω to VTT or 100 Ω to GND. Place termination close to T4 pin when discrete x8 or x16 DRAM is used or close to the DIMM connector when signals are connected to DIMM connector to be used only by DIMMs with x8 or x16 DRAM.

29. For T4160, this pin may be left floating or pulled up. For T4080, this pin must be pulled to ground. Pull with a 4.7 k resistor.

Warning

See "Connection Recommendations" for additional details on properly connecting these pins for specific applications.

3 Electrical characteristics

This section provides the AC and DC electrical specifications for the chip. The chip is currently targeted to these specifications, some of which are independent of the I/O cell but are included for a more complete reference. These are not purely I/O buffer design specifications.

3.1 Overall DC electrical characteristics

This section describes the ratings, conditions, and other characteristics.

3.1.1 Absolute maximum ratings

This table provides the absolute maximum ratings.

	Absolute Maxir	num Ratings for Supply	Voltage Levels		
Characteristic	Symbol	Min	Мах	Unit	Notes
Core and platform supply voltage	V _{DD}	-0.3	1.08	V	9, 11

Table 2. Absolute maximum ratings¹

Table continues on the next page...

Characteristic	Symbol	Min	Мах	Unit	Notes
	Absolute Maximu	m Ratings for Storage	Temperature Condition	ıs	
∟ow Power Security Monitor supply	V _{DD_LP}	-0.3	1.08	V	11
supply voltage					
JSB PHY Analog	USB_OV _{DD}	-0.3	1.08	V	11
JSB PHY Transceiver supply voltage	USB_HV _{DD} USB_OV _{DD}	-0.3	3.63	V	11
Ethernet management nterface 2 (EMI2) I/O /oltage		-0.3	1.32	V V	7, 11
nanagement interface I (EMI1) 1588, GPIO /O voltage		-0.3	1.98		
Pad power supply for SerDes transmitter Ethernet, Ethernet	XnV _{DD}	-0.3 -0.3 -0.3	1.65 1.45 2.75	V V	8, 11
Main power supply for nternal circuitry of SerDes and pad power supply for SerDes receivers	SnV _{DD}	-0.3	1.08	V	11
DDR3L DRAM I/O voltage	GnV _{DD}	-0.3	1.42	V	11
DDR3 DRAM I/O voltage	GnV _{DD}	-0.3	1.58	V	11
		-0.3	1.98		
supply, and JTAG I/O voltage DUART, I ² C I/O voltage	DV _{DD}	-0.3	2.75	V	11
eSHDC, eSPI, DMA, MPIC, GPIO, system control and power nanagement, clocking, debug, IFC, DDRCLK	OV _{DD}	-0.3	1.98	V	11
Thermal monitor unit supply	TH_V _{DD}	-0.3	1.98	V	10, 11
Euse programming override supply	PROG_SFP	-0.3	1.98	V	11
PLL supply voltage SerDes, filtered from KnV _{DD})	AV _{DD} _SDn_PLLn	-0.3	1.65 1.48	v	
	AV _{DD} Dn	-0.3	1.65	V	11
	AV _{DD} _PLAT				
core, platform, DDR)	AV _{DD} _CGBn				

Table 2. Absolute maximum ratings¹ (continued)

Table continues on the next page...

Storage terr range	perature	T _{STG}		-55		155		°C	_
		Absol	ute Maximu	m Ratings f	or Input Sigr	hal Voltage	Levels		
Interface II	nput Signal	Syn	nbol	Min_DCV V_input	Max_DCV V_input	Min Undersho ot Voltage	Max Overshoot Voltage	Unit	Notes
DDR3 and I DRAM signa		MV _{IN}		GND	Nominal GV _{DD} x 1.05	-0.3	Nominal GV _{DD} x 1.1	V	2, 13
DDR3 and I DRAM refer		D _n _MV _{REF}		GND	Nominal GV _{DD} /2 x 1.05	-0.3	Nominal GV _{DD} /2 x 1.1	V	5
Ethernet (ex 1588, GPIO	ccept EMI2), signals	LV _{IN}		GND	Nominal LV _{DD} x 1.1	-0.3	Nominal LV _{DD} x 1.15	V	4, 5
eSHDC, eS MPIC, GPIC control and managemen debug, IFC, supply, and signals), system power nt, clocking, DDRCLK	OV _{IN}		GND	Nominal OV _{DD} x 1.1	-0.3	Nominal OV _{DD} x 1.15	V	3, 5
DUART, I ² C	signals	DV _{IN}		GND	Nominal DV _{DD} x 1.1	-0.3	Nominal DV _{DD} x 1.15	V	5, 6
SerDes signals	No internal termination selected	SV _{IN}	0.8 V maximum signal swing starting from 0.3 V	S _n GND	Nominal SnV _{DD} x 1.05	0.3	Nominal SnV _{DD} x 1.1	V	5
			0.8 V maximum signal swing starting from -0.4 V	S _n GND	Nominal SnV _{DD} x 1.05	-0.4	+0.4		
	50 Ω internal termination selected	SV _{IN}		S _n GND	+0.3	-0.4	+0.4		
USB PHY T signals	ransceiver	USB_HV _{IN}		USB_AGN D	USB_HV _{DD} + 0.3	-0.3	USB_HV _{DD} + 0.3	V	5, 12
		USB_OV _{IN}		USB_AGN D	USB_OV _D _D x 1.1	-0.3	USB_OV _D _D x 1.15	V	5, 12
Ethernet ma interface 2 s				GND	1.2 x 1.1	-0.3	1.2 x 1.15	V	—
LP Trust sig LP_TMP_D		V _{IN_LP}		GND	1.05 x V _{DD_LP}	-0.3	1.1 x V _{DD_LP}	V	_
Notes:									

Table 2. Absolute maximum ratings¹ (continued)

Table 2. Absolute maximum ratings¹

1. Functional operating conditions are given in Table 3. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.

2. **Caution:** MV_{IN} must not exceed GV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

3. Caution: OV_{IN} must not exceed OV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

4. **Caution:** LV_{IN} must not exceed LV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

5. (G,O,L,D,S, USB_H, USB_O)V_{IN} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 7. Note that the Dn_MV_{REF} maximum slew rate is restricted to 25 kv/s.

6. **Caution:** DV_{IN} must not exceed DV_{DD} by more than 0.3 V. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.

7. Ethernet MII management interface 2 pins function as open drain I/Os. The interface shall conform to 1.2 V nominal voltage levels.

8. The cfg_xvdd_sel (ASLEEP) reset configuration pin must select the correct XV_{DD} voltage.

9. Supply voltage specified at the voltage sense pin. Voltage input pins should be regulated to provide specified voltage at the sense pin. For additional information, see the "Ganged sense-line implementation example" section in the *T4240 QorlQ Integrated Processor Design Checklist* (AN4559). See also note 6 in Table 3.

10. Thermal monitoring unit is de featured on current silicon, but TH_V_{DD} should be biased always.

11. Exposing device to Absolute Maximum Ratings conditions for long periods of time may affect reliability or cause permanent damage.

12. USB Overshoot or Undershoot signal time should be under 10% of signal rise time or under 2 nSec.

13. Typical DDR interface uses ODT enabled mode. For test purposes with ODT off mode, simulation should be done first so as to make sure that the overshoot signal level at the input pin does not exceed GV_{DD} by more than 10%. The Overshoot/ Undershoot period should comply with JEDEC standards.

3.1.2 Recommended operating conditions

This table provides the recommended operating conditions for this chip.

NOTE

The values shown are the recommended operating conditions and proper device operation outside these conditions is not guaranteed.

Table 3.	Recommended	operating	conditions
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Characteri	stic	Symbol	Recommended Value	Unit	Notes
Core and platform supply voltage	At initial start-up	V _{DD}	(VID or 1.025 V) ± 30 mV	V	4, 5, 6, 7, 9
	During normal operation		VID ± 30 mV		

Table continues on the next page...

Characte	eristic	Symbol	Recommended Value	Unit	Notes
PLL supply voltage (SerDes, filtered from XnV _{DD}) Fuse programming override supply Thermal monitor unit supply eSHDC, eSPI, DMA, MPIC, GPIO, system control and power management, clocking, debug, IFC, DDRCLK supply, and JTAG I/O voltage DUART, I ² C I/O voltage DDR DRAM I/O voltage DDR JRAM I/O voltage Main power supply for internal circuitry of SerDes and pad power supply for SerDes receivers Pad power supply for SerDes transmitters Ethernet, Ethernet management interface 1 (EMI1), 1588, GPIO I/O voltage USB PHY Transceiver supply voltage USB PHY Analog supply voltage At initial start-up During normal operation Low Power Security Monitor supply	AV _{DD} _CGAn	1.8 V ± 90 mV	V	11	
		AV _{DD} _CGBn			
Ise programming override supply Isermal monitor unit supply SHDC, eSPI, DMA, MPIC, GPIO, system control and power anagement, clocking, debug, IFC, DDRCLK supply, and JTAG 0 voltage JART, I ² C I/O voltage DR DRAM I/O voltage DDR3 DDR3L ain power supply for internal circuitry of SerDes and pad power pply for SerDes receivers ad power supply for SerDes transmitters hernet, Ethernet management interface 1 (EMI1), 1588, GPIO 0 voltage hernet management interface 2 (EMI2) I/O voltage SB PHY Transceiver supply voltage SB PHY Analog supply voltage At initial start-up During normal operation w Power Security Monitor supply put voltage DDR3 and DDR3L DRAM	AV _{DD} _PLAT				
		Value AV_{DD}_CGAn $1.8 V \pm 90 mV$ V AV_{DD}_CGBn $1.8 V \pm 90 mV$ V AV_{DD}_PLAT $1.5 V \pm 75 mV or$ V AV_{DD}_Dn $1.5 V \pm 75 mV or$ V $AV_{DD}_SDn_PLLn$ $1.5 V \pm 75 mV or$ V $AV_{DD}_SDn_PLLn$ $1.5 V \pm 0 mV$ V TH_V_{DD} $1.8 V \pm 90 mV$ V AG OV_{DD} $1.8 V \pm 90 mV$ V AG OV_{DD} $1.8 V \pm 90 mV$ V AG DV_{DD} $1.5 V \pm 75 mV$ V $1.5 V \pm 75 mV$ $1.5 V \pm 75 mV$ V $1.5 V \pm 0 mV$ $1.5 V \pm 75 mV$ V Mr_DD $1.5 V \pm 0 mV$ V Mr_DD $1.8 V \pm 90 mV$ V Mr_DD_LP $1.8 V \pm 90 mV$ V Mr_DLP $1.0 V \pm 50 mV$ V Mr_DLP $1.0 V \pm 50 mV$ V Mr_DLP $1.0 V \pm 50 mV$ V Mr_DLP Mr_DR Mr_DR $Mr_$			
PLL supply voltage (SerDes, filtered	l from XnV _{DD})	AV _{DD} _SDn_PLLn	1.5 V ± 75 mV or	V	-
			1.35 V ± 67 mV		
Fuse programming override supply		PROG_SFP	1.8 V ± 90 mV	V	2
Thermal monitor unit supply		TH_V _{DD}	1.8 V ± 90 mV	V	8
PLL supply voltage (core, platform, DDR) PLL supply voltage (SerDes, filtered from XnV _{DD}) Fuse programming override supply Thermal monitor unit supply eSHDC, eSPI, DMA, MPIC, GPIO, system control and power management, clocking, debug, IFC, DDRCLK supply, and JTAG I/O voltage DUART, I ² C I/O voltage DDR DRAM I/O voltage DDR3L Main power supply for internal circuitry of SerDes and pad power supply for SerDes receivers Pad power supply for SerDes transmitters Ethernet, Ethernet management interface 1 (EMI1), 1588, GPIO I/O voltage USB PHY Transceiver supply voltage USB PHY Analog supply voltage USB PHY Analog supply voltage Input voltage DDR3 and DDR3L DRAM signals DDR3 and DDR3L DRAM reference Ethernet (except EMI2), 1588, GPIO signals eSHDC, eSPI, DMA, MPIC GPIO, system control and power management, clocking, debug, IFC, DDRCLK supply, and JTAG	OV _{DD}	1.8 V ± 90 mV	V	-	
PLL supply voltage (core, platform, DDR) PLL supply voltage (SerDes, filtered from XnV _{DD}) Fuse programming override supply Thermal monitor unit supply SHDC, eSPI, DMA, MPIC, GPIO, system control and power nanagement, clocking, debug, IFC, DDRCLK supply, and JTAGO voltage DUART, I ² C I/O voltage DDR DRAM I/O voltage DDR3L Main power supply for internal circuitry of SerDes and pad pow upply for SerDes receivers Pad power supply for SerDes transmitters Ethernet, Ethernet management interface 1 (EMI1), 1588, GPIC O voltage Iternet management interface 2 (EMI2) I/O voltage JSB PHY Transceiver supply voltage JSB PHY Analog supply voltage At initial start-up During normal operation cow Power Security Monitor supply nput voltage DDR3 and DDR3L DRAI signals DDR3 and DDR3L DRAI signals BCHO, eSPI, DMA, MP GPIO, system control an power management, clocking, debug, IFC,		DV _{DD}	2.5 V ± 125 mV	V	-
			1.8 V ± 90 mV		
DDR DRAM I/O voltage	DDR3	GnV _{DD}	1.5 V ± 75 mV	V	-
	DDR3L	_	1.35 V ± 67 mV		
	itry of SerDes and pad power	SnV _{DD}	1.0 V ± 50 mV	V	-
Pad power supply for SerDes transi	XnV _{DD}	1.5 V ± 75 mV	V	-	
			1.35 V ± 67 mV		
		LV _{DD}	2.5 V ± 125 mV	V	1
I/O voltage			1.8 V ± 90 mV		
Ethernet management interface 2 (I	EMI2) I/O voltage	-	1.2 V ± 60 mV	V	10
USB PHY Transceiver supply voltage	je	USB_HV _{DD}	3.3 V ± 165 mV	V	-
D voltage Thernet management interface 2 (EMI2) I/O voltage SB PHY Transceiver supply voltage		USB_OV _{DD}	1.8 V ± 90 mV	V	-
USB PHY Analog supply voltage	At initial start-up	USB_SV _{DD}		V	6,7,9
	During normal operation		VID ± 30 mV		
Low Power Security Monitor supply		V _{DD_LP}		_	-
Input voltage		MV _{IN}	GND to GV _{DD}		-
		Dn_MV _{REF}	GV _{DD} /2 ± 1%	V	-
		LV _{IN}	GND to LV _{DD}	V	-
	power management, clocking, debug, IFC, DDRCLK supply, and JTAG	OV _{IN}	GND to OV _{DD}	V	-
	DUART, I ² C signals	DVIN	GND to DV _{DD}	V	-
	SerDes signals	SVIN	GND to SV _{DD}	V	-

Table 3. Recommended operating conditions (continued)

Table continues on the next page...

Character	istic	Symbol	Recommended Value	Unit	Notes
	USB PHY Transceiver	USB_HV _{IN}	GND to USB_HV _{DD}	V	-
	signals	USB_OV _{IN}	GND to USB_OV _{DD}	V	-
	Ethernet management interface 2 (EMI2) signals	-	GND to 1.2V	V	3
	LP Trust signal LP_TMP_DETECT_B	V _{IN_LP}	GND to V _{DD_LP}	V	-
Operating temperature range	Normal operation	T _A ,	$T_A = 0$ (min) to	°C	-
		TJ	T _J = 105(max)		
	Extended Temperature	T _A ,	T _A = -40 (min) to	°C	-
		TJ	T _J = 105(max)		
	Secure boot fuse	T _A ,	$T_A = 0$ (min) to	°C	2
	programming	TJ	T _J = 70 (max)		

 Table 3. Recommended operating conditions (continued)

1. Selecting RGMII limits to $LV_{DD} = 2.5 V$.

2. PROG_SFP must be supplied 1.8 V and the chip must operate in the specified fuse programming temperature range only during secure boot fuse programming. For all other operating conditions, PROG_SFP must be tied to GND, subject to the power sequencing constraints shown in Power sequencing.

3. Ethernet MII management interface 2 pins function as open drain I/Os. The interface conforms to 1.2 V nominal voltage levels.

4. Refer to Voltage ID (VID) controllable supply and Core and platform supply voltage filtering for additional information.

5. Supply voltage specified at the voltage sense pin. Voltage input pins should be regulated to provide specified voltage at the sense pin. For additional information, see the "Ganged sense-line implementation example" section in the *T4240 QorIQ Integrated Processor Design Checklist* (AN4559).

6. Operation at 1.1V is allowable for up to 25ms at initial power on. Alternatively the initial start-up voltage can power up straight to the VID voltage if the system has previously programmed that specific part's VID value.

7. Voltage ID (VID) operating range is between 0.975V to 1.025V. Regulator selection should be based on Vout range wider than VIDmin to VIDmax with resolution of 12.5mV or better.

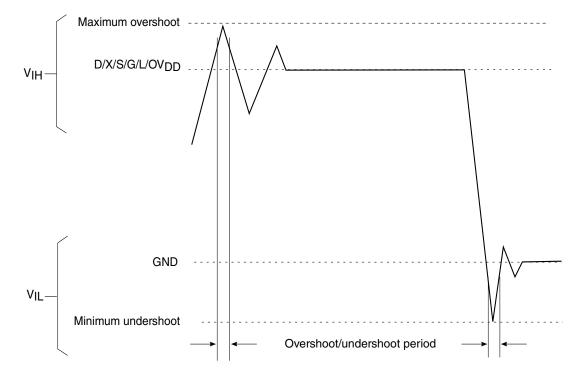
8. Keep this pin biased to the specified voltage, despite the thermal monitoring unit being de-featured.

9. If VID is known at initial start-up, set VDD=VID else if VID is not known at initial start-up, set VDD to 1.025V and change it immediately, to VDD=VID after reading VID at the beginning of booting.

10. This supply does not have a designated pin in this device because it is used only for EMI2 signals external pull-up resistor source.

11.Keep filter close to pin. Voltage and tolerance for AV_{DD} is defined at the input of the PLL supply filter and not the pin of AV_{DD} .

This figure shows the undershoot and overshoot voltages at the interfaces of the chip.



Notes:

The overshoot/undershoot period should be less than 10% of shortest possible toggling period " bit time", of the input signal or per input signal specific protocol requirement. For GPIO input signal overshoot/undershoot period, it should be less than 10% of the SYSCLK period.

Figure 7. Overshoot/Undershoot voltage for USB_OV_{IN}/USB_HV_{IN}/LV_{IN}/OV_{IN}/MV_{IN}/SV_{IN}/DV_{IN}

See Table 3 for actual recommended core voltage. Voltage to the processor interface I/Os are provided through separate sets of supply pins and must be provided at the voltages shown in Table 3. The input voltage threshold scales with respect to the associated I/O supply voltage. DV_{DD} , OV_{DD} and LV_{DD} based receivers are simple CMOS I/O circuits and satisfy appropriate LVCMOS type specifications. The DDR SDRAM interface uses differential receivers referenced by the externally supplied Dn_MV_{REF} signal (nominally set to $GV_{DD}/2$) as is appropriate for the SSTL_1.35/SSTL_1.5 electrical signaling standard. The DDR DQS receivers cannot be operated in single-ended fashion. The complement signal must be properly driven and cannot be grounded.

3.1.3 Output driver characteristics

This chip provides information on the characteristics of the output driver strengths.

NOTE

These values are preliminary estimates.

Driver type	Output impedance (Ω)	Supply voltage	Notes
DDR3 signal	18(full-strength mode)	GV _{DD} = 1.5 V	1
	27(half-strength mode)		
DDR3L signal	18(full-strength mode)	GV _{DD} = 1.35 V	1
	27(half-strength mode)		
Ethernet signals	45	LV _{DD} = 2.5 V	2
eSPI, JTAG, system control, Integrated flash controller (IFC)	45	OV _{DD} = 1.8 V	2
DUART, I ² C	45	DV _{DD} = 2.5 V	2
		DV _{DD} = 1.8 V	
1. The drive strength of the DDR3 or DDR3L interface in half-stren	igth mode is at T _j = 105 °C a	and at GV _{DD} (min).	
2. Impedance value varies by +/- 20%			

Table 4. Output drive capability

3.2 Power sequencing

For power up, the requirements are as follows:

- 1. Bring up V_{DD}, SnV_{DD}, USB_SV_{DD}, V_{DD}_LP, USB_HV_{DD}, LV_{DD}, DV_{DD}, USB_OV_{DD}, OV_{DD}, TH_V_{DD}, AV_{DD} (cores, platform, DDR), GnV_{DD}, XnV_{DD}, and AV_{DD}_SDn_PLLn. Drive PROG_SFP = GND.
 - PORESET_B input must be driven asserted and held during this step.

Power supplies in this step have no ordering requirement with respect to one another except for the USB power supplies per the following note.

NOTE

- a. USB_SV_{DD} supply must ramp before or after the USB_HV_{DD} and USB_OV_{DD} supplies have ramped. The supply set that ramp first must reach 90% of its final value before a supply from the other set can be ramped up.
- b. USB_ HV_{DD} and USB_ OV_{DD} supplies among themselves are sequence independent.
- c. USB_HV_{DD} rise time (10% to 90%) has a minimum of 100 us.
- 2. Negate PORESET_B input as long as the required assertion/hold time has been met per Table 21.
- 3. For secure boot fuse programming, use the following steps:
 - a. After negation of PORESET_B, drive PROG_SFP = 1.8 V after a required minimum delay per Table 5.

b. After fuse programming is completed, it is required to return PROG_SFP = GND before the system is power cycled (PORESET_B assertion) or powered down (V_{DD} ramp down) per the required timing specified in Table 5. See Security fuse processor, for additional details.

Warning

No activity other than that required for secure boot fuse programming is permitted while PROG_SFP is driven to any voltage above GND, including the reading of the fuse block. The reading of the fuse block may only occur while PROG_SFP = GND.

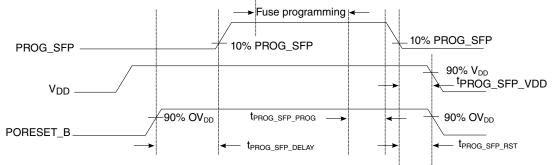
From a system standpoint, if any of the I/O power supplies ramp prior to the V_{DD} supply, there will be a brief period as the V_{DD} powers up that the I/Os associated with that I/O supply may go from being tristated to an indeterminate state (either driven to a logic one or zero), and extra current may be drawn by the device.

Only 300,000 POR cycles are permitted per lifetime of a device. Note that this value is based on design estimates and is preliminary.

All supplies must be at their stable values within 400 ms.

If using Trust Architecture Security Monitor battery backed features, then ensure that both, OVDD is ramped to recommended operational voltage, and SYSCLK is running, prior to VDD ramping up to the 0.5 Volt level. The running system clock should have a minimum frequency of 800HZ and a maximum frequency no greater than the supported maximum system clock frequency as in Table 14 table.

This figure provides the PROG_SFP timing diagram.



NOTE: PROG_SFP must be stable at 1.8 V prior to initiating fuse programming.

Figure 8. PROG_SFP timing diagram

This table provides information on the power-down and power-up sequence parameters for PROG_SFP.

Table 5. PROG_SFP timing ⁵

Driver type	Min	Max	Unit	Notes
tprog_sfp_delay	100	-	SYSCLKs	1
tprog_sfp_prog	0	-	μs	2
tprog_sfp_vdd	0	-	μs	3
tprog_sfp_rst	0	-	μs	4

1. Delay required from the deassertion of PORESET_B to driving PROG_SFP ramp up. Delay measured from PORESET_B deassertion at 90% OV_{DD} to 10% PROG_SFP ramp up.

2. Delay required from fuse programming finished to PROG_SFP ramp down start. Fuse programming must complete while PROG_SFP is stable at 1.8 V. No activity other than that required for secure boot fuse programming is permitted while PROG_SFP driven to any voltage above GND, including the reading of the fuse block. The reading of the fuse block may only occur while PROG_SFP = GND. After fuse programming is completed, it is required to return PROG_SFP = GND.

3. Delay required from PROG_SFP ramp down complete to V_{DD} ramp down start. PROG_SFP must be grounded to minimum 10% PROG_SFP before V_{DD} is at 90% V_{DD}.

4. Delay required from PROG_SFP ramp down complete to PORESET_B assertion. PROG_SFP must be grounded to minimum 10% PROG_SFP before PORESET_B assertion reaches 90% OV_{DD}.

5. Only two secure boot fuse programming events are permitted per lifetime of a device.

Warning

PROG_SFP ramp up slew rate must not exceed 25kV/s. Ramp down does not have a slew rate constraint.

3.3 Power-down requirements

The power-down cycle must complete such that power supply values are below 0.4 V before a new power-up cycle can be started.

If performing secure boot fuse programming per Power sequencing, it is required that $PROG_SFP = GND$ before the system is power cycled (PORESET_B assertion) or powered down (V_{DD} ramp down) per the required timing specified in Table 5.

NOTE

All input signals, including I/Os that are configured as inputs, driven into the chip need to monotonically increase/decrease through entire rise/fall durations.

3.4 Power characteristics

This table shows the power dissipations of the V_{DD} and $S_n V_{DD}$ supply for various operating platform clock frequencies versus the core and DDR clock frequencies when Altivec power is gated off. See the e6500 core reference manual, section 8.6.1, "Altivec Power Down - Software Controlled Entry" for details on how to place Altivec in low power state.

Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME/FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD (Core + Platform) + SVDD Power (W) ¹	VDD (Core + Platfor m) Power	SnV _{DD} power (W) ⁹	Notes
Typical	1500	667	1600	500/667	VID	1.0	65	32	29.7	2.3	2, 3
Thermal							105	42	39.7	2.3	4, 5
Maximum								50	47.7	2.3	5, 6, 7
Typical	1667	733	1866	550/733	VID	1.0	65	35	32.7	2.3	2, 3
Thermal							105	52	49.7	2.3	4, 5
Maximum								61	58.7	2.3	5, 6, 7
Typical	1800	733	1866	550/733	VID	1.0	65	38	35.7	2.3	2, 3
Thermal							105	54	51.7	2.3	4, 5
Maximum								64	61.7	2.3	5, 6, 7

Table 6. T4240 Power dissipation for rev 2 silicon with Altivec power-gated off¹

Notes:

1. Combined power of V_{DD} and $S_n V_{DD}$ with platform at power-on reset default state, all DDR controllers and all SerDes banks active. Does not include I/O power and Altivec is power-gated off.

2. Typical power assumes Dhrystone running with activity factor of 60% (on all cores) and is executing DMA on the platform with 100% activity factor.

3. Typical power based on nominal process distribution for this device.

4. Thermal power assumes Dhrystone running with activity factor of 60% (on all cores) and executing DMA on the platform at 100% activity factor.

5. Thermal and maximum power are based on worst-case process distribution for this device.

Table 6. T4240 Power dissipation for rev 2 silicon with Altivec power-gated off	Table 6.	T4240 Power dissi	pation for rev 2 silicon	with Altivec power-gated off ¹
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Power Core Plat mode freq freq (MHz) (MHz	DDR PME/FM data freq (MHz) rate (MT/s)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD (Core + Platform) + SVDD Power (W) ¹	VDD (Core + Platfor m) Power	SnV _{DD} power (W) ⁹	Notes
---	---	-------------------------------------	--------------------------	------------------------	--	--	--	-------

6. Maximum power assumes Dhrystone running with activity factor at 100% (on all cores) and is executing DMA on the platform at 115% activity factor.

7. Maximum power provided for power supply design sizing.

8. Voltage ID (VID) operating range is between 0.975 V to 1.025 V.

9. Total SnVDD Power Conditions (S1,S2,S3,S4). This represents the highest possible power at 105°C based upon worstcase voltage tolerances and data patterns. Use the equations in Table 9 for average power at 105°C.

a- SerDes1: 2 lanes @ 10.3125 G, 6 lanes @ 3.125 G.

b- SerDes2: 2 lanes @ 10.3125 G, 6 lanes @ 3.125 G.

c- SerDes3: 8 lanes @ 10.3125 G.

d- SerDes4: 4 lanes @ 10 G, 4 lanes @ 5 G.

Table 7. T4241 Power dissipation for rev 2 silicon with Altivec power-gated off¹

Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME/FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD (Core + Platform) + SVDD Power (W) ¹	VDD (Core + Platfor m) Power	SnV _{DD} power (W) ⁹	Notes
Typical	1500	667	1600	500/667	VID	1.0	65	29.1	26.8	2.3	2, 3
Thermal							105	37.5	35.2	2.3	4, 5
Maximum								45.2	42.9	2.3	5, 6, 7
Typical	1667	733	1866	550/733	VID	1.0	65	32.2	29.9	2.3	2, 3
Thermal							105	43.7	41.4	2.3	4, 5
Maximum								52.5	50.2	2.3	5, 6, 7
Typical	1800	733	1866	550/733	VID	1.0	65	34.1	31.8	2.3	2, 3
Thermal	1						105	45.1	42.8	2.3	4, 5
Maximum								54.6	52.3	2.3	5, 6, 7

Notes:

1. Combined power of V_{DD} and S_nV_{DD} with platform at power-on reset default state, all DDR controllers and all SerDes banks active. Does not include I/O power and Altivec is power-gated off.

2. Typical power assumes Dhrystone running with activity factor of 60% (on all cores) and is executing DMA on the platform with 100% activity factor.

3. Typical power based on nominal process distribution for this device.

4. Thermal power assumes Dhrystone running with activity factor of 60% (on all cores) and executing DMA on the platform at 100% activity factor.

5. Thermal and maximum power are based on worst-case process distribution for this device.

Table 7. T4241 Power dissipation for rev 2 silicon with Altivec power-gated off¹

	-										
Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME/FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD (Core + Platform) + SVDD Power (W) ¹	VDD (Core + Platfor m) Power	SnV _{DD} power (W) ⁹	Notes
6. Maximur platform at	•		-	ne running wit	h activity	factor at 10	0% (on all co	res) and is e	executing D	MA on th	ie
7. Maximur	m power	provided	for powe	r supply desig	n sizing.						
8. Voltage	ID (VID)	operating	range is	between 0.97	′5 V to 1.	025 V.					
			•	· · · ,	•		ighest possibl) for average p			ed upon w	/orst-
a- SerDes1	: 2 lanes	@ 10.31	25 G, 6 I	anes @ 3.125	G.						
b- SerDes2	2: 2 lanes	@ 10.31	25 G, 6 I	anes @ 3.125	G.						
c- SerDes3: 8 lanes @ 10.3125 G.											
d- SerDes4: 4 lanes @ 10 G, 4 lanes @ 5 G.											

This table shows the power dissipations of the V_{DD} and SnV_{DD} supplies for various operating platform clock frequencies versus the core and DDR clock frequencies when Altivec power is on.

Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME /FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD(Cor e + Platform) + SVDD (W) ¹	VDD (Core+ Platform) power	SnV _{DD} power (W) ⁹	Notes
Typical	1500	667	1600	500/667	VID	1.0	65	35	32.7	2.3	2, 3
Thermal							105	45	42.7	2.3	4, 5
Maximum								53	50.7	2.3	5, 6, 7
Typical	1667	733	1866	550/733	VID	1.0	65	38	35.7	2.3	2, 3
Thermal							105	55	52.7	2.3	4, 5
Maximum								64	61.7	2.3	5, 6, 7
Typical	1800	733	1866	550/733	VID	1.0	65	41	38.7	2.3	2, 3
Thermal	1						105	57	54.7	2.3	4, 5
Maximum								66	63.7	2.3	5, 6, 7

Table 8. T4240 Power dissipation for rev 2 silicon with Altivec enabled¹

Notes:

1. Combined power of V_{DD} and S_nV_{DD} with platform at power-on reset default state, all DDR controllers and all SerDes banks active. Does not include I/O power.

2. Typical power assumes Altivec benchmark running (on all cores) and is executing DMA on the platform with 100% activity factor.

3. Typical power based on nominal process distribution for this device.

Table 8	T4240 Power	dissination	for rev 2	silicon with	Altivec	enabled ¹
I able 0.	14240 FOWEI	uissipation		SHICOH WILH	AILIVEC	enableu

Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME /FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD(Cor e + Platform) + SVDD (W) ¹	VDD (Core+ Platform) power	SnV _{DD} power (W) ⁹	Notes
4. Thermal DMA on the	-				ing with v	vork power	activity factor	of 100% (or	n all cores)	and exec	uting
5. Thermal	and max	imum po	wer are b	ased on wors	t-case pr	ocess distril	bution for this	device.			
	•			enchmark run % activity fact	•	work powe	r activity facto	or at 100% (o	on all cores) and is	
7. Maximur	m power	provided	for powe	r supply desig	n sizing.						
8. Voltage	ID (VID)	operating	range is	between 0.97	'5 V to 1.	025 V.					
					•		ghest possibl for average p			d upon w	orst-
a- SerDes1	: 2-lanes	@ 10.31	25 G, 6-I	anes SGMII @	3.125 C	Э.					
b- SerDes2	2: 2-lanes	@ 10.31	25 G, 6-I	anes SGMII (D 3.125 C	a.					
c- SerDes3	: 8-lanes	@ 10.31	25 G.								
d- SerDes4	d- SerDes4: 4-lanes @ 10 G, 4-lanes @ 5 G.										

Table 9. T4241 Power dissipation for rev 2 silicon with Altivec enabled¹

Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME /FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD(Cor e + Platform) + SVDD (W) ¹	VDD (Core+ Platform) power	SnV _{DD} power (W) ⁹	Notes
Typical	1500	667	1600	500/667	VID	1.0	65	31.5	29.2	2.3	2, 3
Thermal							105	40.4	38.1	2.3	4, 5
Maximum								48.1	45.8	2.3	5, 6, 7
Typical	1667	733	1866	550/733	VID	1.0	65	34.9	32.6	2.3	2, 3
Thermal							105	47.1	44.8	2.3	4, 5
Maximum								55.9	53.6	2.3	5, 6, 7
Typical	1800	733	1866	550/733	VID	1.0	65	37.0	34.7	2.3	2, 3
Thermal							105	48.7	46.4	2.3	4, 5
Maximum								58.2	55.9	2.3	5, 6, 7

Notes:

1. Combined power of V_{DD} and S_nV_{DD} with platform at power-on reset default state, all DDR controllers and all SerDes banks active. Does not include I/O power.

2. Typical power assumes Altivec benchmark running (on all cores) and is executing DMA on the platform with 100% activity factor.

3. Typical power based on nominal process distribution for this device.

4. Thermal power assumes Altivec benchmark running with work power activity factor of 100% (on all cores) and executing DMA on the platform at 100% activity factor.

5. Thermal and maximum power are based on worst-case process distribution for this device.

Electrical characteristics

Table 9. T4241 Power dissipation for rev 2 silicon with Altivec enabled¹

Power mode	Core freq (MHz)	Plat freq (MHz)	DDR data rate (MT/s)	PME /FM freq (MHz)	V _{DD} ⁸ (V)	SnV _{DD} (V)	Junction temp. (°C)	VDD(Cor e + Platform) + SVDD (W) ¹	VDD (Core+ Platform) power	SnV _{DD} power (W) ⁹	Notes
	•			enchmark rur % activity fac	•	work powe	er activity facto	or at 100% (on all cores) and is	
7. Maximu	m power	provided	for powe	r supply desig	ın sizing.						
8. Voltage ID (VID) operating range is between 0.975 V to 1.025 V.											
			•	· · · ·	•		ighest possibl for average p	•		d upon w	orst-
a- SerDes	I: 2-lanes	@ 10.31	25 G, 6-I	anes SGMII (@ 3.125 (G.					
b- SerDes2	2: 2-lanes	@ 10.31	25 G, 6-I	anes SGMII (@ 3.125 (Э.					
c- SerDes3: 8-lanes @ 10.3125 G.											
d- SerDes4	1: 4-lanes	@ 10 G	, 4-lanes	@ 5 G.							

This table provides low power mode saving estimation.

Table 10. T4240/T4160/T4080 rev 2 single core, single cluster low power mode power savings, 1.0 V ^{1,2,3,7}

Mode	Temp	Core Freque ncy = 1.8 GHz	Core Frequency = 1.667 GHz	Core Frequency = 1.5 GHz	Units	Comment	Notes
PH10	65°C	0.95	0.88	0.79	Watts	Saving realized moving from PH00 to PH10 state, single core.	4
PH15	65°C	0.27	0.25	0.22	Watts	Saving realized moving from PH10 state to PH15 state, single core.	4,5
PH20	65°C	0.33	0.33	0.33	Watts	Saving realized moving from PH15 to PH20 state, single core.	4
PCL10	65°C	0.9	0.9	0.9	Watts	Saving realized moving from PH20 to PCL10 for single cluster.	6
LPM20 (T4080)	65°C	1.2	1.2	1.0	Watts	Saving realized moving from PCL10 to LPM20.	6
LPM20 (T4160)	65°C	1.2	1.2	1.0	Watts	Saving realized moving from PCL10 to LPM20.	6

Table continues on the next page...

Table 10.T4240/T4160/T4080 rev 2 single core, single cluster low power mode power
savings, 1.0 V 1,2,3,7 (continued)

Mode	Temp	Core Freque ncy = 1.8 GHz	Core Frequency = 1.667 GHz	Core Frequency = 1.5 GHz	Units	Comment	Notes
LPM20 (T4240)	65°C	1.8	1.8	1.5	Watts	Saving realized moving from PCL10 to LPM20.	6
LPM40	65°C	1.33	1.33	0.83	Watts	Saving realized moving from LPM20 to LPM40.	6
Notes:		1	1		I	I	

1. Power for V_{DD} only.

2. Typical power assumes Dhrystone running (PH00 state) with activity factor of 60%.

3. Typical power based on nominal process distribution for this device.

4. PH10, PH15, PH20 power savings with one core. Maximum savings would be N times, where N is the number of used cores.

5. Require both threads of the core to enter the same low-power mode.

6. See the e6500 reference manual and the T4240 reference manual for additional low power mode details.

7. Also applicable for lower power T4241 devices.

This table provides all the estimated I/O power supply values based on preliminary measurements.

Table 11. T4240/T4241 I/O power dissipation

I/O P	ower Supply	Used in	Parameter	Typical (mW)	Maximum (mw)	Notes	
LVCMOS	OV _{DD} 1.8 V	eSHDC, eSPI, DMA, MPIC, GPIO management, clocking, debug, IFC, DDRCLK supply, and JTAG		140		1, 3, 4, 5	
LVCMOS	LV _{DD} 1.8 V	Ethernet, Ethernet management interface 1 (EMI1), 1588, GPIO	_	122	_		
LVCMOS	LV _{DD} 2.5 V	Ethernet, Ethernet management interface 1 (EMI1), 1588, GPIO	_	198	_		
LVCMOS	DV _{DD} 1.8 V	DUART, I 2 C	—	12	—		
LVCMOS	DV _{DD} 2.5 V	DUART, I 2 C	—	17	—	1	
LVCMOS	PROG_SFP 1.8V	Fuse programming	—	200	—	1	

Table continues on the next page...

I/O Po	wer Supply	Used in	Parameter	Typical (mW)	Maximum (mw)	Notes
LVCMOS	V _{DD} LP 1 V	Low Power Security Monitor	-	8	_	
DDR I/O	GV _{DD} 1.5 V	All three DDR controllers	1866 MT/s	3500	5000	1, 2, 5
DDR I/O	GV _{DD} 1.5 V	All three DDR controllers	1600 MT/s	3100	4900	
DDR I/O	Dn_MV_REF	DDR3 and DDR3L DRAM reference	_	—	—	—
USB_PHY	USB_OV _{DD} 1.8 V	USB PHY Transceiver supply voltage	-	54		1, 5
USB_PHY	USB_HV _{DD} 3.3 V	USB PHY Transceiver supply voltage	-	59		
USB_PHY	USB_SV _{DD} 1 V	USB PHY Analog supply voltage	-	6	_	
PLL	AV _{DD} _CGAn 1.8 V AV _{DD} _CGBn 1.8 V AV _{DD} _PLAT 1.8 V	PLL of core and system	-	15 for each		1, 5
PLL_DDR	AV _{DD} _Dn 1.8 V	PLL of DDR	—	15	—	-
PLL_SerDes	AV _{DD} _SDn_PLLn 1.5 V or 1.35 V	PLL of SerDes	-	60		
SerDes, 1.35 XV _{DD} , 1.0 V SV _{DD}	Pad power supply for single SerDes module's receivers	SV _{DD}	$\label{eq:rate} \begin{array}{l} F_i = Lane \; data \\ rate \; in \; Gbps \\ N = Total \\ number \; of \; lanes \\ used \\ n_i = number \; of \\ lanes \; running \; at \\ F_i \; rate \end{array}$	$\begin{array}{l} P_{SV_{DD}} = \\ 155.047 + \\ 16.766 * N + \\ 3.287 * (Sum(n_i \\ * F_i)) \pm 15 \text{ mW} \end{array}$		6
SerDes, 1.35 XV _{DD} , 1.0 V SV _{DD}	Pad power supply for single SerDes module's transmitters	XV _{DD}	F_i = Lane data rate in Gbps N = Total number of lanes used n _i = number of lanes running at	$\begin{array}{l} P_XV_{DD} = \\ 53.256 + 50.685 \\ ^* N + 0.683 \\ (Sum(n_i * F_i)) \pm \\ 15 \ mW \end{array}$		6

Table 11. T4240/T4241 I/O power dissipation (continued)

Notes:

Table 11. T4240/T4241 I/O power dissipation

I/O Power Supply	Used in	Parameter	Typical (mW)	Maximum (mw)	Notes
1. The maximum values are dependent environmental conditions such as tempower. Expect different results dependent simulations at 105 °C junction temperations	perature, voltage and f ding on the use case. T	frequency. This is	not intended to be	e the maximum gu	
2. Typical DDR power numbers are ba of bus. These values are good for the DDR I/O power can be up to 9.6 Watts	rmal design but for sup	ply design it shou	ld be assumed 10	0% utilization of b	
3. Assuming 15 pF total capacitance I	oad.				
4. GPIOs are supported on 1.8 V and	2.5 V rails as specified	I in the hardware s	specification.		
5. The typical values are estimates an and assuming at 65° C junction tempe		ients at nominal re	ecommended volta	age for the I/O pov	ver supply
6. The total power numbers of XV_{DD} a C junction temperature. If one PLL is a how to use the formulas in estimating	used, then subtract 60	mW from the resu	llting P_ SV _{DD} . Th		
Example 1:					
On a SerDes block running SGMII at 3	3.125 Gbps on one lan	e, the SerDes typ	ical powers are ex	pected to be:	
P_SV _{DD} = 155.047 + 16.766 * 1 + 3.2	287 *(1 * 3.125) ± 15 m	W - (60 mW "beca	ause one PLL is u	sed") = 122 mW :	± 15 mW
$P_XV_{DD} = 53.256 + 50.685 * 1 + 0.685$	33 * (1 * 3.125) ± 15 m ^v	W = 106 mW ± 15	mW		
Example 2:					
On a SerDes block running PCIe at 5	Gbps on eight lanes, tl	he SerDes typical	powers are expec	ted to be:	
P_SV _{DD} = 155.047 + 16.766 * 8 + 3.2	287 * (8 * 5) ± 15 mW -	(60 mW "because	e one PLL is used	") = 361 ± 15 mW	1
$P_XV_{DD} = 53.256 + 50.685 * 8 + 0.685$	83 *(8 * 5) ± 15 mW = 4	186 mW ± 15 mW			
Example 3:					
On a single SerDes block running XFI module typical powers are expected to	•	wo lanes and SGN	/III at 3.75 G on fo	ur lanes, the singl	e SerDes
P_SV _{DD} = 155.047 + 16.766 * 6 + 3.2	287 * (2 * 10.3125 + 4 *	* 3.75) ± 15 mW =	= 373 mW ± 15 m ^v	W	
$P_XV_{DD} = 53.256 + 50.685 * 6 + 0.685$	2 * (2 * 10 2125 + 4 * *	(2.75) + 15 m M - 1	000 mM + 15 mM		

3.5 Power-on ramp rate

This section describes the AC electrical specifications for the power-on ramp rate requirements. Controlling the maximum power-on ramp rate is required to avoid excess in-rush current.

This table provides the power supply ramp rate specifications.

Table 12.	Power supply ramp rate
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Parameter	Min	Max	Unit	Notes
Required ramp rate for all voltage supplies (including $OV_{DD}/DV_{DD}/GnV_{DD}/SnV_{DD}/XnV_{DD}/LV_{DD}$, all core and platform V_{DD} supplies, Dn_MV_{REF} and all AV_{DD} supplies.)	-	25	V/ms	1, 2
Required ramp rate for PROG_SFP	-	25	V/ms	1, 2
Note:	•		•	
1. Ramp rate is specified as a linear ramp from 10 to 90%. If non-linear (for examp change from 200 to 500 mV is the most critical as this range might falsely trigger the trigger trigger the trigger the trigger trigger trigger the trigger tri			naximum ra	ite of

2. Over full recommended operating temperature range (see Table 3).

3.6 Input clocks

3.6.1 System clock (SYSCLK) and real-time clock (RTC) timing specifications

This section provides the system clock and real-time clock DC and AC timing specifications.

3.6.1.1 SYSCLK and RTC DC timing specifications

This table provides the SYSCLK and RTC DC specifications.

Table 13. SYSCLK and RTC DC electrical characteristics³

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	—	—	V	1
Input low voltage	VIL	—	—	0.6	V	1
Input capacitance (SYSCLK)	C _{IN}	—	3.3	—	pF	_
Input capacitance (RTC)	C _{IN}	—	2.6	_	pF	_
Input current ($OV_{IN} = 0 V \text{ or } OV_{IN} = OV_{DD}$)	I _{IN}	-50	—	50	μA	2
Note:						
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are	based on the res	pective min and	max OV _{IN} value	es found in Table	3.	
2. The symbol OV_{IN} , in this case, repre	esents the OV _{IN} s	symbol reference	d in Recommen	ded operating co	onditions.	
3. At recommended operating condition	ns with OV _{DD} = 1	.8 V, see Table	3.			

3.6.1.2 SYSCLK and RTC AC timing specifications

This table provides the SYSCLK AC timing specifications.

Table 14. SYSCLK AC timing specifications⁵

Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes
SYSCLK frequency	f _{SYSCLK}	66.7	—	133.3	MHz	1, 2
SYSCLK cycle time	t _{SYSCLK}	7.5	—	15	ns	1, 2
SYSCLK duty cycle	t _{KHK} / t _{SYSCLK}	40	—	60	%	2
SYSCLK slew rate	—	1	—	4	V/ns	3
SYSCLK peak period jitter	—	—	—	± 150	ps	—
SYSCLK jitter phase noise at -56 dBc	—	—	—	500	KHz	4
AC Input Swing voltage	ΔV _{AC}	0.6 x OV _{DD}	—	1 x OV _{DD}	V	6

Notes:

1. **Caution:** The relevant clock ratio settings must be chosen such that the resulting SYSCLK frequency do not exceed their respective maximum or minimum operating frequencies.

2. Measured at the rising edge and/or the falling edge at $OV_{DD}/2$.

3. Slew rate as measured from 0.35 x OV_{DD} to 0.65 x $\text{OV}_{\text{DD}}.$

4. Phase noise is calculated as FFT of TIE jitter.

5. At recommended operating conditions with $OV_{DD} = 1.8V$, see Table 3.

6. AC swing measured relative to half OV_{DD} or VIH and VIL have equal absolute offset from OV_{DD} /2, So, Swing = (VIH-VIL)/ OVDD and ΔV_{AC} = Swing x OV_{DD}

This table provides the RTC AC timing specifications.

Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes
RTC frequency	f _{RTC}		—	platform clock/16	MHz	1, 2
RTC cycle time	t _{RTC}	16/platform clock	—	—	ns	1, 2
RTC duty cycle	t _{KHK} / t _{RTC}	40	—	60	%	2
RTC slew rate	—	1	—	4	V/ns	3
RTC peak period jitter	—	—	—	± 150	ps	-
RTC jitter phase noise at -56 dBc	—	—	—	500	KHz	4
AC Input Swing voltage	ΔV _{AC}	0.6 x OV _{DD}	—	1 x OV _{DD}	V	6

Table 15. RTC AC timing specifications⁵

Notes:

1. Caution: The relevant clock ratio settings must be chosen such that it fits IEEE1588, or MPIC, or RCPM requirements.

2. Measured at the rising edge and/or the falling edge at OV_{DD}/2.

3. Slew rate as measured from 0.35 x OV_{DD} to 0.65 x OV_{DD} .

4. Phase noise is calculated as FFT of TIE jitter.

Table 15.	RTC AC timing	specifications ⁵
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Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes		
5. At recommended operating conditions with OV _{DD} = 1.8V, see Table 3.								
6. AC swing measured relative to half 0 OVDD and ΔV_{AC} = Swing x OV _{DD}	DV _{DD} or VIH and V	/IL have equal a	bsolute offset fro	om OV _{DD} /2, So,	Swing = (\	/IH-VIL)/		

3.6.2 Spread-spectrum sources

Spread-spectrum clock sources are an increasingly popular way to control electromagnetic interference emissions (EMI) by spreading the emitted noise to a wider spectrum and reducing the peak noise magnitude in order to meet industry and government requirements. These clock sources intentionally add long-term jitter to diffuse the EMI spectral content. The jitter specification given in this table considers short-term (cycle-to-cycle) jitter only. The clock generator's cycle-to-cycle output jitter should meet the chip's input cycle-to-cycle jitter requirement. Frequency modulation and spread are separate concerns; the chip is compatible with spread-spectrum sources if the recommendations listed in this table are observed.

Table 16. Spread-spectrum clock source recommendations³

Parameter	Min	Мах	Unit	Notes
Frequency modulation	-	60	kHz	-
Frequency spread	-	1.0	%	1, 2
Notes:	•		•	

1. SYSCLK frequencies that result from frequency spreading and the resulting core frequency must meet the minimum and maximum specifications given in Table 14.

2. Maximum spread-spectrum frequency may not result in exceeding any maximum operating frequency of the device.

3. At recommended operating conditions with OVDD = 1.8 V, see Table 3.

CAUTION

The processor's minimum and maximum SYSCLK and core/ platform/DDR frequencies must not be exceeded regardless of the type of clock source. Therefore, systems in which the processor is operated at its maximum rated core/platform/DDR frequency should avoid violating the stated limits by using down-spreading only.

3.6.3 Real-time clock (RTC) timing

The RTC timing input is sampled by the platform clock. The output of the sampling latch is then used as an input to the counters of the MPIC and the time base unit of the core; there is no need for jitter specification. The minimum period of the RTC signal should be greater than or equal to 16x the period of the platform clock. There is no minimum RTC frequency; RTC may be grounded if not needed.

3.6.4 Gigabit Ethernet reference clock timing

This table provides the Ethernet gigabit reference clock DC specifications.

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Input high voltage	V _{IH}	1.7	-	-	V	2
Input low voltage	V _{IL}	-	-	0.7	V	2
Input capacitance	C _{IN}	-	-	6	pF	-
Input current ($LV_{IN} = 0 V \text{ or}$ $LV_{IN} = LV_{DD}$)	I _{IN}	-50	-	50	μΑ	3
1. At recommen	ded operating cor	ditions with LV _{DD} =	= 2.5 V			·
2. The min V_{IL} a	nd max V _{IH} value	s are based on the	respective min ar	nd max LV _{IN} value	es found in Table	3.
3. The symbol L	V _{IN} , in this case, i	represents the LV _{IN}	symbol reference	ed in Recommend	ded operating con	ditions.

Table 17. ECn_GTX_CLK125 DC electrical characteristics ¹

This table provides the Ethernet gigabit reference clocks AC timing specifications.

Table 18. ECn_GTX_CLK125 AC timing specifications ¹

Parameter/Condition	Symbol	Min	Typical	Max	Unit	Notes
ECn_GTX_CLK125 frequency	t _{G125}	125 - 100 ppm	125	125 + 100 ppm	MHz	-
ECn_GTX_CLK125 cycle time	t _{G125}	-	8	-	ns	-
ECn_GTX_CLK125 rise and fall time	t _{G125R} /t _{G125F}	-	-	0.75	ns	2
$LV_{DD} = 2.5 V$						
ECn_GTX_CLK125 duty cycle	t _{G125H} /t _{G125}	47	-	53	%	3
1000Base-T for RGMII						
ECn_GTX_CLK125 jitter	-	-	-	± 150	ps	3
1. At recommended operating conditions	with $LV_{DD} = 2.5 V$	± 125 mV.	•		•	
2. Rise and fall times for ECn_GTX_CLK	125 are measured	from 0.5 and 2.0	V for LV _{DD} =	2.5 V.		

3. ECn_GTX_CLK125 is used to generate the GTX clock for the Ethernet transmitter with 2% degradation. The ECn_GTX_CLK125 duty cycle can be loosened from 47%/53% as long as the PHY device can tolerate the duty cycle generated by the GTX_CLK. See RGMII AC timing specifications for duty cycle for 10Base-T and 100Base-T reference clock.

3.6.5 DDR clock timing

This section provides the DDR clock DC and AC timing specifications.

3.6.5.1 DDR clock DC timing specifications

This table provides the DDR clock (DDRCLK) DC specifications.

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	-	V	1
Input low voltage	V _{IL}	-	-	0.6	V	1
Input capacitance	C _{IN}	-	11		pF	-
Input current (OV_{IN} = 0 V or OV_{IN} = OV_{DD})	I _{IN}	-50	-	50	μA	2
Note:			1			-
1. The min $V_{\rm IL} and \max V_{\rm IH}$ values are	based on the res	spective min and	d max OV _{IN} value	es found in Table	3.	

Table 19. DDRCLK DC electrical characteristics³

2. The symbol OV_{IN}, in this case, represents the OV_{IN} symbol referenced in Recommended operating conditions.

3. At recommended operating conditions with $OV_{DD} = 1.8$ V, see Table 3.

3.6.5.2 DDR clock AC timing specifications

This table provides the DDR clock (DDRCLK) AC timing specifications.

Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes
DDRCLK frequency	f _{DDRCLK}	66.7	-	133.3	MHz	1, 2
DDRCLK cycle time	t _{DDRCLK}	7.5	-	15	ns	1, 2
DDRCLK duty cycle	t _{KHK} / t _{DDRCLK}	40	-	60	%	2
DDRCLK slew rate	-	1	-	4	V/ns	3
DDRCLK peak period jitter	-	-	-	± 150	ps	-
DDRCLK jitter phase noise at -56 dBc	-	-	-	500	KHz	4
AC Input Swing voltage	ΔV _{AC}	0.6 x OV _{DD}	-	1 x OV _{DD}	V	6

 Table 20.
 DDRCLK AC timing specifications⁵

Notes:

1. **Caution:** The relevant clock ratio settings must be chosen such that the resulting DDRCLK frequency do not exceed their respective maximum or minimum operating frequencies.

2. Measured at the rising edge and/or the falling edge at $OV_{DD}/2$.

3. Slew rate as measured from 0.35 x OV_{DD} to 0.65 x $\text{OV}_{\text{DD}}.$

4. Phase noise is calculated as FFT of TIE jitter.

Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes	
5. At recommended operating condition	ns with OV _{DD} = 1.8	3V, see Table 3.					
6. AC swing measured relative to half OV_{DD} or VIH and VIL have equal absolute offset from $OV_{DD}/2$, So, Swing = (VIH-VIL)/ OVDD and ΔV_{AC} = Swing x OV_{DD} .							

Table 20. DDRCLK AC timing specifications⁵

3.6.6 Other input clocks

A description of the overall clocking of this device is available in the chip reference manual in the form of a clock subsystem block diagram. For information about the input clock requirements of functional modules sourced external of the chip, such as SerDes, Ethernet management, eSDHC, IFC, see the specific interface section.

3.7 **RESET** initialization

This section describes the AC electrical specifications for the RESET initialization timing requirements. This table describes the AC electrical specifications for the RESET initialization timing.

Parameter/Condition	Min	Max	Unit	Notes
Required assertion time of PORESET_B	1	-	ms	1
Required input assertion time of HRESET_B	32	-	SYSCLKs	2, 3
Maximum rise/fall time of PORESET_B signal	-	1	SYSCLK	4
Maximum rise/fall time of HRESET_B signal	-	4	SYSCLK	4
PLL input setup time with stable SYSCLK before HRESET_B negation	100	-	μs	-
Input setup time for POR configs with respect to negation of PORESET_B	4	-	SYSCLKs	2
Input hold time for all POR configs with respect to negation of PORESET_B	2	-	SYSCLKs	2
Maximum valid-to-high impedance time for actively driven POR configs with respect to negation of PORESET_B	-	5	SYSCLKs	2

 Table 21. RESET Initialization timing specifications

1. PORESET_B must be driven asserted before the core and platform power supplies are powered up.

2. SYSCLK is the primary clock input for the chip.

3. The device asserts HRESET_B as an output when PORESET_B is asserted to initiate the power-on reset process. The device releases HRESET_B sometime after PORESET_B is deasserted. The exact sequencing of HRESET_B deassertion is documented in section "Power-On Reset Sequence" in the chip reference manual.

Table 21. RESET Initialization timing specifications

Parameter/Condition	Min	Мах	Unit	Notes
4. System/board must be designed to ensure the input requirement to the guaranteed for inputs meeting this requirement by design, simulation, clable 1, notes 6 and 7, recommends a week pull up resistor for HRESE designers have to reduce the pull up resistor (min of 280 Ohms) or in a maximum rise/fall time requirement for HRESET if this pin is too loaded	haracterizatior T signal pin in ddition use bic	, or functional the range of 2k	testing. For exa K to 10K Ohms,	ample On But PCB

This table provides the PLL lock times.

Table 22.	PLL lock times
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Parameter/Condition	Min	Max	Unit	Notes
PLL lock times (Core, platform, DDR only)	-	100	μs	-

DDR3 and DDR3L SDRAM controller 3.8

This section describes the DC and AC electrical specifications for the DDR3 and DDR3L SDRAM controller interface. Note that the required $GV_{DD}(typ)$ voltage is 1.5 V when interfacing to DDR3 SDRAM and the $GV_{DD}(typ)$ voltage is 1.35 V when interfacing to DDR3L SDRAM.

NOTE

When operating at a DDR data rate of 1866 MT/s, only one dual-ranked module per memory controller is supported.

DDR3 and DDR3L SDRAM interface DC electrical 3.8.1 characteristics

This table provides the recommended operating conditions for the DDR SDRAM controller when interfacing to DDR3 SDRAM.

Table 23.	DDR3 SDRAM interface DC electrical characteristics (GV _{DD} = 1.5 V) ^{1, 7}
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Parameter	Symbol	Min	Мах	Unit	Note
I/O reference voltage	Dn_MV _{REF}	0.49 x GV _{DD}	0.51 x GV _{DD}	V	2, 3, 4
Input high voltage	V _{IH}	Dn_MV _{REF} + 0.100	GV _{DD}	V	5
Input low voltage	V _{IL}	GND	D <i>n_</i> MV _{REF} - 0.100	V	5
I/O leakage current	l _{oz}	-100	100	μA	6
Notes:	•	•		•	

Table 23. DDR3 SDRAM interface DC electrical characteristics $(GV_{DD} = 1.5 V)^{1, 7}$

Parameter	Symbol	Min	Мах	Unit	Note
1. GV_{DD} is expected to be within 50 mV of the voltage supply may or may not be from the sar	-	e supply at all times.	The DRAM's and men	nory contro	ller's
2. Dn_MV_{REF} is expected to be equal to 0.5 x 0 peak noise on Dn_MV_{REF} may not exceed the	00	66			Peak-to-
3. V _{TT} is not applied directly to the device. It is the supply to which far end signal termination is made, and it is expected to be equal to Dn_MV_{REF} with a min value of Dn_MV_{REF} - 0.04 and a max value of Dn_MV_{REF} + 0.04. V _{TT} should track variations i the DC level of Dn_MV_{REF} .					
4. The voltage regulator for Dn_MV _{REF} must m	eet the specifi	cations stated in Table	e 25.		
5. Input capacitance load for DQ, DQS, and DO	QS_B are avail	able in the IBIS mode	ls.		
6. Output leakage is measured with all outputs	disabled, 0 V	$\leq V_{OUT} \leq GV_{DD}$.			
7. For recommended operating conditions, see	Table 3.				

This table provides the recommended operating conditions for the DDR SDRAM controller when interfacing to DDR3L SDRAM.

Table 24.	DDR3L SDRAM interface DC electrical characteristics $(GV_{DD} = 1.35 V)^{1, 7}$
-----------	---

Parameter	Symbol	Min	Мах	Unit	Note
I/O reference voltage	Dn_MV _{REF}	0.49 x GV _{DD}	0.51 x GV _{DD}	V	2, 3, 4
Input high voltage	V _{IH}	Dn_MV _{REF} + 0.090	GV _{DD}	V	5
Input low voltage	V _{IL}	GND	D <i>n_</i> MV _{REF} - 0.090	V	5
I/O leakage current	I _{OZ}	-100	100	μA	6

Notes:

1. GV_{DD} is expected to be within 50 mV of the DRAM's voltage supply at all times. The DRAM's and memory controller's voltage supply may or may not be from the same source.

2. Dn_MV_{REF} is expected to be equal to 0.5 x GV_{DD} and to track GV_{DD} DC variations as measured at the receiver. Peak-to-peak noise on Dn_MV_{REF} may not exceed the Dn_MV_{REF} DC level by more than ±1% of GV_{DD} (i.e. ±13.5mV).

3. V_{TT} is not applied directly to the device. It is the supply to which far end signal termination is made, and it is expected to be equal to Dn_MV_{REF} with a min value of Dn_MV_{REF} - 0.04 and a max value of Dn_MV_{REF} + 0.04. V_{TT} should track variations in the DC level of Dn_MV_{REF} .

4. The voltage regulator for Dn_MV_{REF} must meet the specifications stated in Table 25.

5. Input capacitance load for DQ, DQS, and DQS_B are available in the IBIS models.

6. Output leakage is measured with all outputs disabled, 0 V \leq V_{OUT} \leq GV_{DD}.

7. For recommended operating conditions, see Table 3.

This table provides the current draw characteristics for Dn_MV_{REF} .

Parameter	Symbol	Min	Мах	Unit	Notes
Current draw for DDR3 SDRAM for Dn_MV _{REF}	I _{Dn_MVREF}	-	500	μA	-
Current draw for DDR3L SDRAM for Dn_MV _{REF}	I _{Dn_MVREF}	-	500	μA	-
Note:	4				•
1. For recommended operating conditions, see	Table 3.				

Table 25. Current draw characteristics for Dn_MV_{BEF}¹

3.8.2 DDR3 and DDR3L SDRAM interface AC timing specifications

This section provides the AC timing specifications for the DDR SDRAM controller interface. The DDR controller supports DDR3 and DDR3L memories. Note that the required $GV_{DD}(typ)$ voltage is 1.5 V when interfacing to DDR3 SDRAM and the required $GV_{DD}(typ)$ voltage is 1.35 V when interfacing to DDR3L SDRAM.

3.8.2.1 DDR3 and DDR3L SDRAM interface input AC timing specifications

This table provides the input AC timing specifications for the DDR controller when interfacing to DDR3 SDRAM.

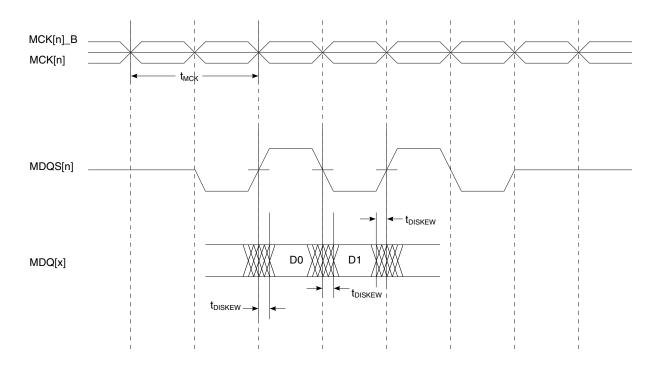
Parameter	Symbol	Min	Мах	Unit	Notes
Controller Skew for MDQS-MDQ/MECC	t _{CISKEW}			ps	1
1866 MT/s data rate		-93	93]	
1600 MT/s data rate		-112	112	1	
1333 MT/s data rate		-125	125	1	
1200 MT/s data rate		-142	142	1	
1066 MT/s data rate		-170	170	1	
Tolerated Skew for MDQS-MDQ/MECC	t _{DISKEW}			ps	2
1866 MT/s data rate		-175	175	1	
1600 MT/s data rate		-200	200	1	
1333 MT/s data rate		-250	250	1	
1200 MT/s data rate		-275	275	1	
1066 MT/s data rate		-300	300		

Table 26. DDR3 and DDR3L SDRAM interface input AC timing specifications³

1. t_{CISKEW} represents the total amount of skew consumed by the controller between MDQS[n] and any corresponding bit that is captured with MDQS[n]. This must be subtracted from the total timing budget.

2. The amount of skew that can be tolerated from MDQS to a corresponding MDQ signal is called t_{DISKEW} . This can be determined by the following equation: $t_{DISKEW} = \pm(T \div 4 - abs(t_{CISKEW}))$ where T is the clock period and $abs(t_{CISKEW})$ is the absolute value of t_{CISKEW} .

3. For recommended operating conditions, see Table 3.



This figure shows the DDR3 and DDR3L SDRAM interface input timing diagram.

Figure 9. DDR3 and DDR3L SDRAM Interface Input Timing Diagram

3.8.2.2 DDR3 and DDR3L SDRAM interface output AC timing specifications

This table contains the output AC timing targets for the DDR3 SDRAM interface.

Table 27. DDR3 and DDR3L SDRAM interface outp	put AC timing specifications ⁷
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Parameter	Symbol ¹	Min	Мах	Unit	Notes
MCK[n] cycle time	t _{MCK}	0.938	2	ns	2
ADDR/CMD output setup with respect to MCK	t _{DDKHAS}			ns	3
1866 MT/s data rate		0.410	-		
1600 MT/s data rate		0.495	-		
1333 MT/s data rate		0.606	-		
1200 MT/s data rate		0.675	-		
1066 MT/s data rate		0.744	-		
ADDR/CMD output hold with respect to MCK	t _{DDKHAX}			ns	3
1866MT/s data rate		0.390	-		
1600 MT/s data rate		0.495	-		
1333 MT/s data rate		0.606	-		
1200 MT/s data rate	1	0.675	-		
1066 MT/s data rate	1	0.744	-		

Table continues on the next page...

Table 27. DDR3 and DDR3L SDRAM interface output AC timing specifications⁷ (continued)

Parameter	Symbol ¹	Min	Мах	Unit	Notes
MCK to MDQS Skew	t _{DDKHMH}			ns	4
> 1600 MT/s data rate		-0.150	0.150		4, 6
> 1066 MT/s data rate, \leq 1600 MT/s data rate		-0.245	0.245		4, 6
MDQ/MECC/MDM output Data eye	t _{DDKXDEYE}			ns	5
1866 MT/s data rate		0.350	-		
1600 MT/s data rate		0.400	-		
1333 MT/s data rate		0.500	-		
1200 MT/s data rate		0.550	-		
1066 MT/s data rate		0.600	-		
MDQS preamble	t _{DDKHMP}	0.9 x t _{MCK}	-	ns	-
MDQS postamble	t _{DDKHME}	0.4 x t _{MCK}	0.6 x t _{MCK}	ns	-

1. The symbols used for timing specifications follow the pattern of t_{(first two letters of functional block)(signal)(state) (reference)(state) for inputs and t_{(first two letters of functional block)(reference)(state)(signal)(state)} for outputs. Output hold time can be read as DDR timing (DD) from the rising or falling edge of the reference clock (KH or KL) until the output went invalid (AX or DX). For example, t_{DDKHAS} symbolizes DDR timing (DD) for the time t_{MCK} memory clock reference (K) goes from the high (H) state until outputs (A) are setup (S) or output valid time.}

2. All MCK/MCK_B and MDQS/MDQS_B referenced measurements are made from the crossing of the two signals.

3. ADDR/CMD includes all DDR SDRAM output signals except MCK/MCK_B, MCS_B, and MDQ/MECC/MDM/MDQS.

4. Note that t_{DDKHMH} follows the symbol conventions described in note 1. For example, t_{DDKHMH} describes the DDR timing (DD) from the rising edge of the MCK[n] clock (KH) until the MDQS signal is valid (MH). t_{DDKHMH} can be modified through control of the MDQS override bits (called WR_DATA_DELAY) in the TIMING_CFG_2 register. This is typically set to the same delay as in DDR_SDRAM_CLK_CNTL[CLK_ADJUST]. The timing parameters listed in the table assume that these two parameters have been set to the same adjustment value. See the chip reference manual for a description and explanation of the timing modifications enabled by the use of these bits.

5. Available eye for data (MDQ), ECC (MECC), and data mask (MDM) outputs at the pin of the processor. Memory controller will center the strobe (MDQS) in the available data eye at the DRAM (end point) during the initialization.

6. Note that for data rates of 1200 MT/s or higher, it is required to program the start value of the DQS adjust for write leveling.

7. For recommended operating conditions, see Table 3.

NOTE

For the ADDR/CMD setup and hold specifications in Table 27, it is assumed that the clock control register is set to adjust the memory clocks by ¹/₂ applied cycle.

This figure shows the DDR3 and DDR3L SDRAM interface output timing for the MCK to MDQS skew measurement (t_{DDKHMH}).

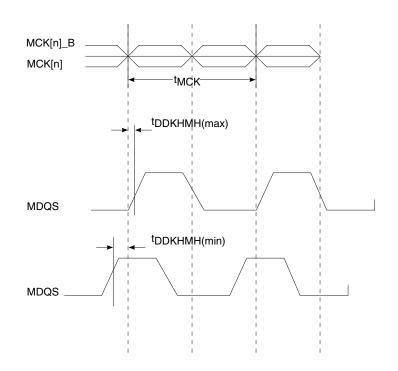


Figure 10. t_{DDKHMH} timing diagram

This figure shows the DDR3 and DDR3L SDRAM output timing diagram.

Electrical characteristics

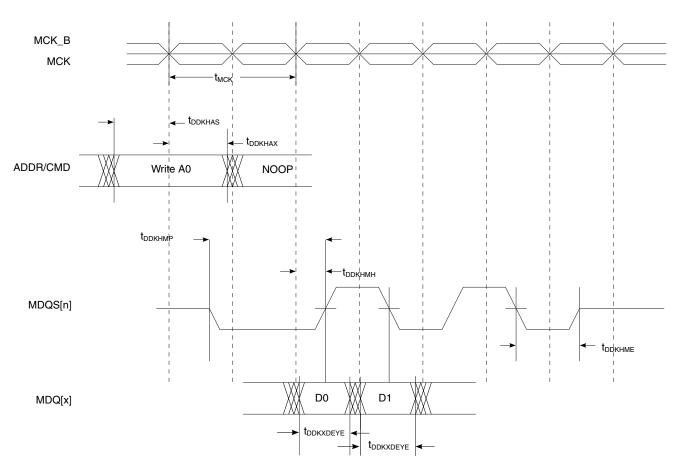


Figure 11. DDR3 and DDR3L output timing diagram

3.9 eSPI interface

This section describes the DC and AC electrical specifications for the eSPI interface.

3.9.1 eSPI DC electrical characteristics

This table provides the DC electrical characteristics for the eSPI interface operating at $OV_{DD} = 1.8 \text{ V}.$

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1
Input current ($V_{IN} = 0$ V or $V_{IN} = OV_{DD}$)	I _{IN}	-50	50	μA	2
Output high voltage	V _{OH}	1.35	-	V	-

Table 28. eSPI DC electrical characteristics (1.8 V)³

Table continues on the next page...

	i		i	i	i		
Parameter	Symbol	Min	Max	Unit	Notes		
$(OV_{DD} = min, I_{OH} = -0.5 mA)$							
Output low voltage	V _{OL}	-	0.4	V	-		
$(OV_{DD} = min, I_{OL} = 0.5 mA)$							
Notes:	•	•	•	•			
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on the respective	min and ma	x OV _{IN} value	s found in <mark>Ta</mark>	ble 3.			
2. The symbol V _{IN} , in this case, represents the OV _{IN} symbol referenced in Recommended operating conditions.							
3. For recommended operating conditions, see Table 3.	. For recommended operating conditions, see Table 3.						

 Table 28.
 eSPI DC electrical characteristics (1.8 V)³ (continued)

3.9.2 eSPI AC timing specifications

This table provides the eSPI input and output AC timing specifications.

Parameter/Condition	Symbol ²	Min	Мах	Unit	Notes
SPI_MOSI output-Master data (internal clock) hold time	t _{NIKHOX}	n1 + (t _{PLATFORM_CLK} * SPMODE[HO_ADJ])	-	ns	1, 2, 4
SPI_MOSI output-Master data (internal clock) delay	t _{NIKHOV}	-	n2 + (t _{PLATFORM_CLK} * SPMODE[HO_ADJ])	ns	1, 2, 4
SPI_CS outputs-Master data (internal clock) hold time	t _{NIKHOX2}	0	-	ns	1
SPI_CS outputs-Master data (internal clock) delay	t _{NIKHOV2}	-	6.0	ns	1
SPI inputs-Master data (internal clock) input setup time	t _{NIIVKH}	3.0	-	ns	-
SPI inputs-Master data (internal clock) input hold time	t _{NIIXKH}	0	-	ns	-
Clock-high time	t _{NIKCKH}	4	-	ns	
Clock-low time	t _{NIKCKL}	4	-	ns	-

Table 29. eSPI AC timing specifications³

Notes:

1. See the chip reference manual for details about the SPMODE register.

2. Output specifications are measured from the 50% level of the rising edge of CLKIN to the 50% level of the signal. Timings are measured at the pin.

3. The symbols used for timing specifications follow the pattern of $t_{(first two letters of functional block)(signal)(state) (reference)(state)}$ for inputs and $t_{(first two letters of functional block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{NIKHOV} symbolizes the NMSI outputs internal timing (NI) for the time t_{SPI} memory clock reference (K) goes from the high state (H) until outputs (O) are valid (V).

4. n1 and n2 values are -1.0 and 1.0 respectively.

This figure provides the AC test load for the eSPI.

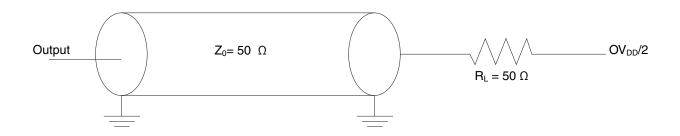
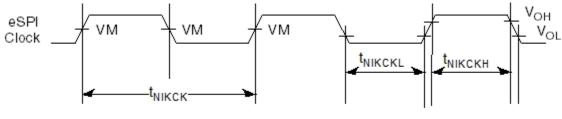


Figure 12. eSPI AC test load

This figure provides the eSPI clock output timing diagram.



VM = Midpoint Voltage (OV_{DD}/2)

Figure 13. eSPI clock output timing diagram

This figure represents the AC timing from Table 29 in master mode (internal clock). Note that although the specifications generally reference the rising edge of the clock, these AC timing diagrams also apply when the falling edge is the active edge. Also, note that the clock edge is selectable on eSPI.

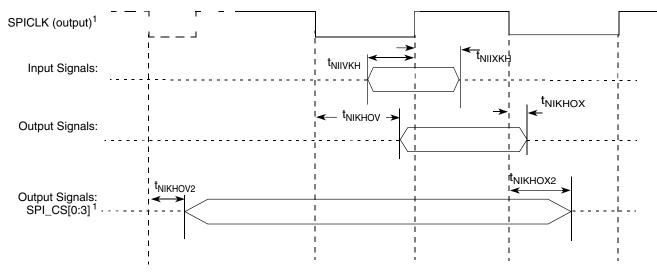


Figure 14. eSPI AC timing in master mode (internal clock) diagram

1. SPICLK appears on the interface only after CS assertion.

3.10 DUART interface

This section describes the DC and AC electrical specifications for the DUART interface.

3.10.1 DUART DC electrical characteristics

This table provides the DC electrical characteristics for the DUART interface at $DV_{DD} = 2.5 \text{ V}$.

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.7	-	V	1
Input low voltage	VIL	-	0.7	V	1
Input current ($DV_{IN} = 0 V \text{ or } DV_{IN} = DV_{DD}$)	I _{IN}	-50	50	μA	2
Output high voltage ($DV_{DD} = min, I_{OH} = -1 mA$)	V _{OH}	2.0	-	V	-
Output low voltage (DV _{DD} = min, I _{OL} = 1 mA)	V _{OL}	-	0.4	V	-
Notes:	·				
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on the r	min and max DV _{IN} re	espective val	ues found in	Table 3.	
2. The symbol DV_{IN} represents the input voltage of the transmission of the symbol DV_{IN} represents the symbol DV_{IN}	ne supply. It is refere	enced in <mark>Rec</mark>	ommended o	perating cor	ditions.
3. For recommended operating conditions, see Table	93.				

Table 30. DUART DC electrical characteristics(2.5 V)³

This table provides the DC electrical characteristics for the DUART interface at $DV_{DD} = 1.8 \text{ V}$.

Table 31.	DUART DC electrical characteristics(1.8 V) ³	
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Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1
Input current ($DV_{IN} = 0 V \text{ or } DV_{IN} = DV_{DD}$)	I _{IN}	-50	50	μA	2
Output high voltage (DV_{DD} = min, I_{OH} = -0.5 mA)	V _{OH}	1.35	-	V	-
Output low voltage (DV _{DD} = min, I_{OL} = 0.5 mA)	V _{OL}	-	0.4	V	-
Notes:					
1. The min $V_{\text{IL}} \text{and} \max V_{\text{IH}}$ values are based on the m	in and max DV _{IN} re	espective valu	ues found in	Table 3.	
2. The symbol DV_{IN} represents the input voltage of the	e supply. It is refere	enced in Reco	ommended o	perating con	ditions.

3. For recommended operating conditions, see Table 3.

3.10.2 DUART AC electrical specifications

This table provides the AC timing parameters for the DUART interface.

Table 32.	DUART	AC timing	specifications
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Parameter/Condition	Value	Unit	Notes			
Minimum baud rate	f _{PLAT} /(2 x 1,048,576)	baud	1, 3			
Maximum baud rate	f _{PLAT} /(2 x 16)	baud	1, 2			
Notes:			L			
1. f_{PLAT} refers to the internal platform clock.						
2. The actual attainable baud rate is limited b	y the latency of interrupt process	sing.				
3. The middle of a start bit is detected as the eighth sampled 0 after the 1-to-0 transition of the start bit. Subsequent bit values are sampled each 16 th sample.						

3.11 Ethernet interface, Ethernet management interface 1 and 2, IEEE Std 1588[™]

This section provides the AC and DC electrical characteristics for the Ethernet controller and the Ethernet management interfaces.

3.11.1 SGMII electrical specifications

See SGMII interface.

3.11.2 RGMII electrical specifications

This section discusses the electrical characteristics for the RGMII interface.

3.11.2.1 RGMII DC electrical characteristics

This table shows the DC electrical characteristics for the RGMII interface.

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.70	-	V	1
Input low voltage	VIL	-	0.70	V	1
Input current (LV _{IN} = 0 V or LV _{IN} = LV _{DD})	l _{IN}	-50	50	μA	2
Output high voltage (LV _{DD} = min, I _{OH} = -1.0 mA)	V _{OH}	2.00	LV _{DD} + 0.3	V	-

Table 33. RGMII DC electrical characteristics $(LV_{DD} = 2.5 V)^3$

Table continues on the next page ...

Table 33. RGMII DC electrical characteristics $(LV_{DD} = 2.5 V)^3$ (continued)

Parameter	Symbol	Min	Max	Unit	Notes
Output low voltage (LV _{DD} = min, I_{OL} = 1.0 mA)	V _{OL}	GND - 0.3	0.40	V	-
1. The min V_{IL} and max V_{IH} values are based on the respective min and max LV_{IN} values found in Table 3.					
2. The symbol LV _{IN} , in this case, represents the LV _{IN} symbol referenced in Recommended operating conditions.					
3. For recommended operating conditions, see Table 3.					

3.11.2.2 RGMII AC timing specifications

This table presents the RGMII AC timing specifications.

Parameter/Condition Symbol¹ Min Тур Max Unit Notes -750 0 1250 Data to clock output skew (at transmitter) 7.9 ps t_{SKRGT_TX} Data to clock input skew (at receiver) 1.0 2.6 ns 2.10 t_{SKRGT RX} RGMII RX_CLK Clock period duration 7.2 8.0 8.8 3 t_{RGT} ns % Duty cycle for 10BASE-T and 100BASE-TX 40 50 60 3, 4 t_{RGTH}/t_{RGT} 45 55 % Duty cycle for Gigabit t_{BGTH}/t_{BGT} 50 Rise time (20%-80%) 0.75 5,6 ns t_{RGTR} Fall time (20%-80%) 0.75 5, 6 ns t_{RGTF}

Table 34. RGMII AC timing specifications $(LV_{DD} = 2.5 V)^8$

Notes:

1. In general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII timing. Note that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).

2. This implies that PC board design will require clocks to be routed such that an additional trace delay of greater than 1.5 ns is added to the associated clock signal. Many PHY vendors already incorporate the necessary delay inside their device. If so, additional PCB delay is probably not needed.

3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.

4. Duty cycle may be stretched/shrunk during speed changes or while transitioning to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.

5. Applies to inputs and outputs.

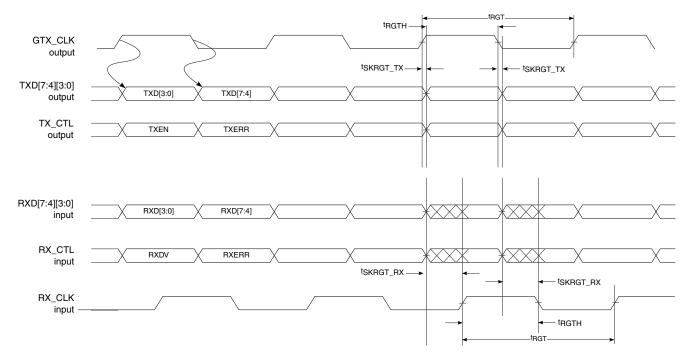
6. System/board must be designed to ensure this input requirement to the chip is achieved. Proper device operation is guaranteed for inputs meeting this requirement by design, simulation, characterization, or functional testing.

7. The frequency of ECn_RX_CLK (input) should not exceed the frequency of ECn_GTX_CLK (output) by more than 300 ppm.

8. For recommended operating conditions, see Table 3.

9. IEEE spec mandates $tSKRGT_TX = +-0.5ns$. Per erratum A-005177 we see $tSKRGT_TX$ has a wider output skew range from -0.75ns to 1.25ns which is larger than the spec asks for. If can not cope with this wide skew then use RGMII at 100 Mbps or 10 Mbps (which allows larger maximum RX skews) or terminate 1000 Mbps RGMII links with PHYs that accommodate larger RX skews or terminate to a second Rev2 device.

10. This device has better input clock to data skew tSKRGT_RX tolerance (1ns to 3.5ns) than spec (1ns to 2.6ns) requires.



This figure shows the RGMII AC timing and multiplexing diagrams.

Figure 15. RGMII AC timing and multiplexing diagrams

Warning

NXP guarantees timings generated from the MAC. Board designers must ensure delays needed at the PHY or the MAC.

3.11.3 Ethernet management interface (EMI)

This section discusses the electrical characteristics for the EMI1 and EMI2 interfaces.

Frame Manager 2's external GE MDIO configures external GE PHYs connected to EMI1 pins. Frame Manager 2's external 10GE MDIO configures external XAUI, XFI and HiGig/HiGig2 PHYs connected to EMI2 pins.

The EMI1 interface timing is compatible with IEEE Std 802.3^{TM} clause 22 and EMI2 interface timing is compatible with IEEE Std 802.3^{TM} clause 45. The External MDIO interfaces on FM1 are not available for use.

3.11.3.1 Ethernet management interface 1 DC electrical characteristics

The DC electrical characteristics for EMI1_MDIO and EMI1_MDC are provided in this section.

Table 35.	Ethernet management interface	1 DC electrical characteristics	$(LV_{DD} = 2.5 V)^3$
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Symbol	Min	Max	Unit	Notes
V _{IH}	1.70	-	V	1
V _{IL}	-	0.70	V	1
I _{IN}	-50	50	μA	2
V _{OH}	2.00	LV _{DD} + 0.3	V	-
V _{OL}	GND - 0.3	0.40	V	-
		•		
	V _{IH} V _{IL} I _{IN} V _{OH}	V _{IH} 1.70 V _{IL} - I _{IN} -50 V _{OH} 2.00	V _{IH} 1.70 - V _{IL} - 0.70 I _{IN} -50 50 V _{OH} 2.00 LV _{DD} + 0.3	V _{IH} 1.70 - V V _{IL} - 0.70 V I _{IN} -50 50 μA V _{OH} 2.00 LV _{DD} + 0.3 V

1. The min V_{IL} and max V_{IH} values are based on the respective min and max LV_{IN} values found in Table 3.

2. The symbol VIN, in this case, represents the LV_{IN} symbol referenced in Recommended operating conditions.

3. For recommended operating conditions, see Table 3.

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.60	V	1
Input current (LV _{IN} = 0V or LV _{IN} =LV _{DD})	I _{IN}	-50	50	μA	2
Output high voltage ($LV_{DD} = min, I_{OH} = -0.5 mA$)	V _{OH}	1.35	-	V	-
Output low voltage (LV _{DD} = min, I_{OL} = 0.5 mA)	V _{OL} -		0.40	V	-
Notes:	ł				T
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on th	e respective r	nin and max C	V _{IN} /QV _{IN} values foun	d in Table 3.	
2. The symbol VIN, in this case, represents the O	/ _{IN} symbol refe	erenced in Re	commended operating	g conditions.	

Table 36. DC electrical characteristics (1.8 V)

3.11.3.2 Ethernet management interface 2 DC electrical characteristics

Ethernet management interface 2 pins function as open drain I/Os. The interface conforms to 1.2 V nominal voltage levels. The DC electrical characteristics for EMI2_MDIO and EMI2_MDC are provided in this section.

Table 37.	Ethernet management interface 2 DC electrical characteristics (1.2 V)	1
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Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	0.84	-	V	-
Input low voltage	V _{IL}	-	0.36	V	-
Output low voltage (I _{OL} = 5.5 mA)	V _{OL}	-	0.2	V	-
Input capacitance	C _{IN}	-	10	pF	-
Notes:		•			
1. For recommended operating conditions,	see Table 3.				

3.11.3.3 Ethernet management interface 1 AC electrical specifications

This table provides the Ethernet management interface 1 AC timing specifications.

Parameter/	Condition	Symbol ¹	Min	Тур	Мах	Unit	Notes
MDC frequency (1/T _{MDC_CIK})	f _{MDC}	—	—	2.5	MHz	2
MDC clock pulse	width high	t _{MDCH}	160	—	—	ns	_
MDC to MDIO delay	Rev1 MDIO_CFG[EHOLD] = 0 MDIO_CFG[NEG] = 0	^t мdkhdx	(Y x t _{enet_clk}) - 3, Y = 2 x MDIO_CFG[M DIO_HOLD] + 1	_	(Y x t _{enet_cik}) + 3, Y = 2 x MDIO_CFG[M DIO_HOLD] + 1	ns	3, 4, 5
	Rev2 MDIO_CFG[NEG] = 0 MDIO_CFG[EHOLD] = 0	t _{MDKHDX}	(Y x t _{enet_clk}) - 3, Y = 2 x MDIO_CFG[M DIO_HOLD] + 1	_	(Y x t _{enet_clk}) + 3, Y = 2 x MDIO_CFG[M DIO_HOLD] + 1	ns	3, 4, 5
	Rev2 MDIO_CFG[NEG] = 0 MDIO_CFG[EHOLD] = 1	t _{MDKHDX}	(Y x t _{enet_clk}) - 3, Y = 8 x MDIO_CFG[M DIO_HOLD] +1	_	(Y x t _{enet_clk}) + 3, Y = 8 x MDIO_CFG[M DIO_HOLD] + 1	ns	3, 4, 5
Rev2 MDIO_CFG[NEG] = 1	MDIO_CFG[t _{MDKHDX}	(Y x T _{MDC_CIK}) - 3, Y = ½	_	(Y x T _{MDC_CIK}) + 3, Y = ½	ns	4
MDIO to MDC se	tup time	t _{MDDVKH}	9	—	—	ns	_
MDIO to MDC ho	ld time	t _{MDDXKH}	0	 	—	ns	—

Table 38. Ethernet management interface 1 AC timing specifications⁶

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{(first two letters of functional block)(signal)(state)(reference)(state) for inputs and <math>t_{(first two letters of functional block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time.

2. This parameter is dependent on the Ethernet clock frequency. (MDIO_CFG [MDIO_CLK_DIV] field determines the clock frequency of the MgmtClk Clock MDIO_MDC).

3. This parameter is dependent on the Ethernet clock frequency. The delay is equal to Y x Ethernet clock periods \pm 3 ns. For example, in default rev1 silicon, with an Ethernet clock of 400 MHz, the min/max delay is = (Y x t_{enet_clk}) \pm 3 ns = ((2 x 2 + 1) x 1/400 M) \pm 3 ns = 12.5 ns \pm 3 ns.

Default values for Rev 1: silicon:

- MDIO_CFG[MDIO_HOLD]= 3'b010 which selects Y = 2 x 2 + 1 = 5 tenet_clk cycles.
- MDIO_CFG[NEG] = 0, in Rev 1, NEG bit field was not visible.
- MDIO_CFG[EHOLD] = 0, in Rev 1, NEG bit field was not visible.

Default values for Rev 2 silicon:

• MDIO_CFG[MDIO_HOLD]= 3'b010, since MDIO_CFG[NEG] = 1 then $Y = \frac{1}{2}$.

Table 38. Ethernet management interface 1 AC timing specifications⁶

Parameter/Condition	Symbol ¹	Min	Тур	Мах	Unit	Notes		
 MDIO_CFG[NEG] = 1 MDIO_CFG[EHOLD] = 0 								
 For Rev 1 silicon: Y = 2 x MDIO_CFG[MDIO_HOLD] + 1 For Rev 2 silicon: If MDIO_CFG[EHOLD] = 0 and MDIO_CFG[NEG] = 0 then Y = 2 x MDIO_CFG[MDIO_HOLD] + 1 If MDIO_CFG[EHOLD] = 1 and MDIO_CFG[NEG] = 0 then Y = 8 x MDIO_CFG[MDIO_HOLD] + 1 If MDIO_CFG[NEG] = 1 then Y = ½. Thus, Y is not affected by MDIO_CFG[HOLD] and MDIO_CFG[EHOLD] when MDIO_CFG[NEG] = 1. For example, in this case, if MDC clock = 2.5 MHz, then min/max of t_{MDKHDX} delay is = Y * T_{MDC_CIK} ± 3 ns = ½ x 1/2.5 M ± 3 ns = 200 ns ± 3 ns. 								
4. t _{MDKHDX} transition:								
• For Rev 1 silcon: t _{MDKHDX} is	MDC positive edge	e to MDIO transi	tion.					
 For Rev 2 silicon: If MDIO_CFG[NEG] = 0 then t_{MDKHDX} is MDC positive edge to MDIO transition. If MDIO_CFG[NEG] = 1 then t_{MDKHDX} is MDC negative edge to MDIO transition. The default value of MDIO_CFG [MDIO_CLK_DIV] is 0 which means no MDIO clock is available. Recommender to configure this field in PBL. 								
5. t_{enet_clk} is the Ethernet clock pe	riod derived from F	rame Manager c	lock, FM clock.	t _{enet_clk} =1/FM_c	lock.			
6. For recommended operating co	onditions, see Table	93.						

3.11.3.4 Ethernet management interface 2 AC electrical characteristics

This table provides the Ethernet management interface 2 AC timing specifications.

Table 39.	Ethernet management	interface 2 AC timin	g specifications ⁶

Parameter	Condition	Symbol ¹	Min	Тур	Max	Unit	Notes
MDC frequency (1/T _{MDC_CIK})	f _{MDC}	—	—	2.5	MHz	2
MDC clock pulse	width high	t _{MDCH}	160	—	—	ns	—
MDC to MDIO delay	Rev1 MDIO_CFG[EHOLD] = 0 MDIO_CFG[NEG] = 0	t _{MDKHDX}	(Y x t _{enet_clk}) - 3, Y = 2 x MDIO_CFG[M DIO_HOLD] + 1		$(Y \times t_{enet_clk}) + 3,$ $Y = 2 \times MDIO_CFG[M]$ $DIO_HOLD] + 1$	ns	3, 4, 5
	Rev2 MDIO_CFG[NEG] = 0 MDIO_CFG[EHOLD] = 0	tмdkhdx	(Y x t _{enet_clk}) - 3, Y = 2 x MDIO_CFG[M DIO_HOLD] + 1	_	$\begin{array}{l} (Y \ x \ t_{enet_clk}) + \\ 3, \\ Y = 2 \ x \\ MDIO_CFG[M \\ DIO_HOLD] + \\ 1 \end{array}$	ns	3, 4, 5
	Rev2 MDIO_CFG[NEG] = 0	t _{MDKHDX}	(Y x t _{enet_clk}) - 3,	-	(Y x t _{enet_clk}) + 3,	ns	3, 4, 5

Table continues on the next page...

Table 39. Ethernet management interface 2 AC timing specifications⁶ (continued)

Parameter/C	Condition	Symbol ¹	Min	Тур	Max	Unit	Notes
	MDIO_CFG[EHOLD] = 1		Y = 8 x MDIO_CFG[M DIO_HOLD] + 1		Y = 8 x MDIO_CFG[M DIO_HOLD] +1		
	Rev2 MDIO_CFG[NEG] = 1	t _{MDKHDX}	(Y x T _{MDC_CIK}) - 3, Y = ½		(Y x T _{MDC_CIK}) + 3, Y = ½	ns	4
MDIO to MDC setu	up time	t _{MDDVKH}	8	—	—	ns	7
MDIO to MDC hold time		t _{MDDXKH}	0	—	_	ns	

Notes:

1. The symbols used for timing specifications follow the pattern of $t_{(first two letters of functional block)(signal)(state)(reference)(state)}$ for inputs and $t_{(first two letters of functional block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{MDKHDX} symbolizes management data timing (MD) for the time t_{MDC} from clock reference (K) high (H) until data outputs (D) are invalid (X) or data hold time. Also, t_{MDDVKH} symbolizes management data timing (MD) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{MDC} clock reference (K) going to the high (H) state or setup time.

2. This parameter is dependent on the Ethernet clock frequency (MDIO_CFG [MDIO_CLK_DIV] field determines the clock frequency of the MgmtClk Clock MDIO_MDC).

3. This parameter is dependent on the Ethernet clock frequency. The delay is equal to Y x Ethernet clock periods \pm 3 ns. For example, in default rev1 silicon, with an Ethernet clock of 400 MHz, the min/max delay is = (Y x t_{enet_clk}) = ((2 x 2 + 1) x 1/400M) \pm 3 ns = 12.5 ns \pm 3 ns.

Default values for Rev 1: silicon:

- MDIO_CFG[MDIO_HOLD] = 3'b010, which selects Y = 2 x 2 + 1 = 5 tenet_clk cycles.
- MDIO_CFG[NEG] = 0, in Rev 1, NEG bit field was not visible.
- MDIO_CFG[EHOLD] = 0, in Rev 1, NEG bit field was not visible.

Default values for Rev 2 silicon:

- MDIO_CFG[MDIO_HOLD] = 3'b010, since MDIO_CFG[NEG] = 1 then $Y = \frac{1}{2}$.
- MDIO_CFG[NEG] = 1
- MDIO_CFG[EHOLD] = 0
- For Rev 1 silicon: Y = 2 x MDIO_CFG[MDIO_HOLD] + 1

• For Rev 2 silicon:

- If MDIO_CFG[EHOLD] = 0 and MDIO_CFG[NEG] = 0 then Y = 2 x MDIO_CFG[MDIO_HOLD] + 1
- If MDIO_CFG[EHOLD] = 1 and MDIO_CFG[NEG] = 0 then Y = 8 x MDIO_CFG[MDIO_HOLD] + 1
- If MDIO_CFG[NEG] = 1 then Y = ½. Thus Y is not affected by MDIO_CFG[HOLD] and MDIO_CFG[EHOLD] when MDIO_CFG[NEG]=1. For example in this case If MDC clock = 2.5 MHz, then min/max of t_{MDKHDX} delay is = Y * T_{MDC, CIK} ± 3 ns = ½ x 1/2.5 M ± 3ns = 200 ns ±3 ns.

4. t_{MDKHDX} transition:

- For Rev 1 silcon: t_{MDKHDX} is MDC positive edge to MDIO transition.
- For Rev 2 silicon:
 - If MDIO_CFG[NEG] = 0 then t_{MDKHDX} is MDC positive edge to MDIO transition.
 - If MDIO_CFG[NEG]= 1 then t_{MDKHDX} is MDC negative edge to MDIO transition.
 - The default value of MDIO_CFG [MDIO_CLK_DIV] is 0, which means no MDIO clock is available. Recommended to configure this field in PBL.

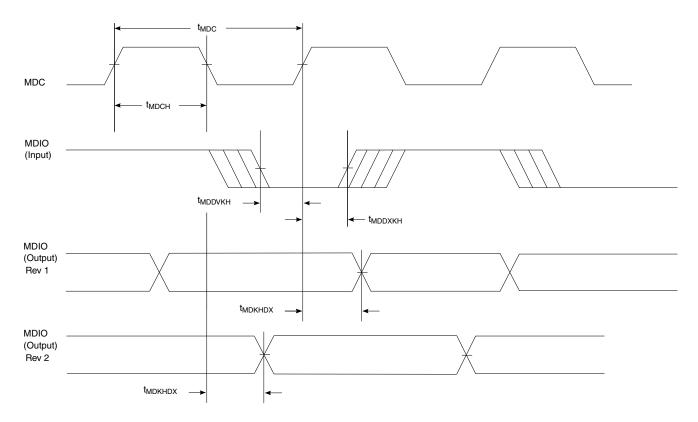
5. t_{enet clk} is the Ethernet clock period derived from Frame Manager clock (FM clock). t_{enet clk} = 1/FM_clock.

6. For recommended operating conditions, see Table 3.

Table 39. Ethernet management interface 2 AC timing specifications⁶

Parameter/Condition	Symbol ¹	Min	Тур	Мах	Unit	Notes
7. The actual setup time varies wi expected to be 68 ns measured a reduced from the default setting b associated with EMI2.	t 50% points. To er	sure setup time	is met, the EMI	2 clock frequenc	y may need t	to be

This figure shows the Ethernet management interface timing diagram.





3.11.4 IEEE 1588 electrical specifications

3.11.4.1 IEEE 1588 DC electrical characteristics

This table shows IEEE 1588 DC electrical characteristics when operating at $LV_{DD} = 2.5$ V supply.

Table 40.	IEEE 1588 DC electrical characteristics $(LV_{DD} = 2.5 V)^3$	
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Parameter	Symbol	Min	Мах	Unit	Notes
Input high voltage	V _{IH}	1.70	-	V	1

Table continues on the next page ...

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Table 40. IEEE 1588 DC electrical characteristics($LV_{DD} = 2.5 V$)³ (continued)

Parameter	Symbol	Min	Max	Unit	Notes
Input low voltage	VIL	-	0.70	V	1
Input current (LV _{IN} = 0 V or LV _{IN} = LV _{DD})	I _{IN}	-50	50	μA	2
Output high voltage (LV _{DD} = min, I _{OH} = -1.0 mA)	V _{OH}	2.00	LV _{DD} + 0.3	V	-
Output low voltage (LV _{DD} = min, I_{OL} = 1.0 mA)	V _{OL}	GND - 0.3	0.40	V	-
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on the results of the	espective min and	d max LV _{IN} valu	es found in Table	3.	•
2. The symbol $\mathrm{LV}_{\mathrm{IN}}$, in this case, represents the $\mathrm{LV}_{\mathrm{IN}}$	symbol reference	d in Recommer	nded operating co	onditions.	
3. For recommended operating conditions, see Table	3.				

This table shows IEEE 1588 DC electrical characteristics when operating at $LV_{DD} = 1.8$ V supply.

Table 41. IEEE 1588 DC electrical characteristics $(LV_{DD} = 1.8 V)^3$

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	VIL	-	0.6	V	1
Input current (LV _{IN} = 0 V or LV _{IN} = LV _{DD})	I _{IN}	-50	50	μA	2
Output high voltage (LV _{DD} = min, I _{OH} = -0.5 mA)	V _{OH}	1.35	LV _{DD} + 0.3	V	-
Output low voltage (LV _{DD} = min, I_{OL} = 0.5 mA)	V _{OL}	GND - 0.3	0.40	V	-
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on the re	spective min and	d max LV _{IN} valu	es found in Table	93.	•
2. The symbol LV_{IN} , in this case, represents the LV_{IN} s	symbol reference	d in Recommer	nded operating co	onditions.	
3. For recommended operating conditions, see Table 3	3.				

3.11.4.2 IEEE 1588 AC specifications

This table provides the IEEE 1588 AC timing specifications.

Table 42. IEEE 1588 AC timing specifications³

Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes
TSEC_1588_CLK_IN clock period	t _{T1588CLK}	6	-		ns	
TSEC_1588_CLK_IN duty cycle	t _{T1588CLKH} / t _{T1588CLK}	40	50	60	%	
TSEC_1588_CLK_IN peak-to-peak jitter	t _{T1588CLKINJ}	-	-	250	ps	-
Rise time TSEC_1588_CLK_IN (20% -80%)	t _{T1588CLKINR}	1.0	-	2.0	ns	-
Fall time TSEC_1588_CLK_IN (80% -20%)	t _{T1588CLKINF}	1.0	-	2.0	ns	-
TSEC_1588_CLK_OUT clock period	t _{T1588CLKOUT}	2 x t _{T1588CLK}	-	-	ns	2

Table continues on the next page...

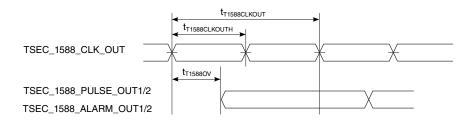
Parameter/Condition	Symbol	Min	Тур	Max	Unit	Notes
TSEC_1588_CLK_OUT duty cycle	t _{T1588CLKOTH} / t _{T1588CLKOUT}	30	50	70	%	-
TSEC_1588_PULSE_OUT1/2,	t _{T1588OV}	0.5	-	4.0	ns	-
TSEC_1588_ALARM_OUT1/2 hold time						
TSEC_1588_TRIG_IN1/2 pulse width	t _{T1588TRIGH}	2 x t _{T1588CLK}	-	-	ns	1
Notes:	1	-				
1. It needs to be at least two times the clo manual for a description of TMR_CTRL re	•	lock selected by TM	IR_CTRL[CKS	EL]. See the o	chip refere	ence

 Table 42.
 IEEE 1588 AC timing specifications³ (continued)

2. There are 3 input clock sources for 1588 i.e. TSEC_1588_CLK_IN, RTC, and MAC clock / 2 in rev1 silicon and MAC clock in rev2 silicon.

3. For recommended operating conditions, see Table 3.

This figure shows the data and command output AC timing diagram.



Note: The output delay is counted starting at the rising edge if $t_{T1588CLKOUT}$ is non-inverting. Otherwise, it is counted starting at the falling edge.

Figure 17. IEEE 1588 output AC timing

This figure shows the data and command input AC timing diagram.

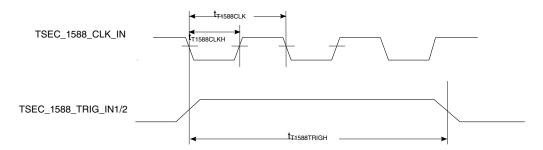


Figure 18. IEEE 1588 input AC timing

3.12 USB interface

This section provides the AC and DC electrical specifications for the USB interface.

3.12.1 USB DC electrical characteristics

This table provides the DC electrical characteristics for the USB interface at USB_HV_{DD} = 3.3 V.

Table 43. USB DC electrical characteristics (USB_HV_{DD} = 3.3 V)³

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	2.0	-	V	1, 4
Input low voltage	V _{IL}	-	0.8	V	1, 4
Input current (USB_HV _{IN} = 0 V or USB_HV _{IN} = USB_HV _{DD})	I _{IN}	-100	+100	μA	2, 4
Output high voltage (USB_HV _{DD} = min, I _{OH} = -2 mA)	V _{OH}	2.8	-	V	5
Output low voltage (USB_HV _{DD} = min, I_{OL} = 2 mA)	V _{OL}	-	0.3	V	5
Notes:		•			

1. The min V_{IL} and max V_{IH} values are based on the respective min and max USB_HV_{IN} values found in Table 3.

2. The symbol USB_HV_{IN}, in this case, represents the USB_HV_{IN} symbol referenced in Recommended operating conditions.

3. For recommended operating conditions, see Table 3.

4. These specifications only apply to the following pins: USB1_PWRFAULT, USB2_PWRFAULT, USB1_UDM (full-speed mode), USB2_UDM (full-speed mode), USB1_UDP (full-speed mode), and USB2_UDP (full-speed mode).

5. This specification only applies to USB1_DRVVBUS and USB2_DRVVBUS pins.

This table provides the DC electrical characteristics for the USBCLK at $OV_{DD} = 1.8$ V.

Table 44. USBCLK DC electrical characteristics (1.8 V)³

Parameter	Symbol	Min	Мах	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1
Input current ($V_{IN} = 0$ V or $V_{IN} = OV_{DD}$)	I _{IN}	-50	50	μΑ	2
Notes:					
1. The min V_{IL} and n	nax V _{IH} values are ba	sed on the respective	e min and max OV _{IN} v	alues found in Table	3.
2. The symbol V _{IN} , ir	n this case, represent	s the OV _{IN} symbol re	ferenced in Recomme	ended operating conc	litions.
3. For recommended	d operating conditions	s, see Table 3.			

3.12.2 USB AC timing specifications

This section describes the AC timing specifications for the on-chip USB PHY. See Chapter 7 in the *Universal Serial Bus Revision 2.0 Specification* for more information.

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Electrical characteristics

This table provides the USB clock input (USBCLK) AC timing specifications.

Symbol	Condition		Тур	Max	Unit	Notes
f _{USB_CLK_IN}	-	-	24	-	MHz	-
t _{USRF}	Measured between 10% and 90%	-	-	6	ns	2
t _{CLK_TOL}	-	-0.005	0	0.005	%	-
t _{CLK_DUTY}	Measured at rising edge and/or failing edge at $OV_{DD}/2$	40	50	60	%	-
t _{CLK_PJ}	RMS value measured with a second-order, band-pass filter of 500 kHz to 4 MHz bandwidth at 10 ⁻¹² BER	-	-	5	ps	-
	tusrf tusrf tclk_tol	fuse Providence fuse - tuse Measured between 10% and 90% tcLK_TOL - tcLK_DUTY Measured at rising edge and/or failing edge at OV _{DD} /2 tcLK_PJ RMS value measured with a second-order, band-pass filter of 500 kHz to 4 MHz	fUSB_CLK_IN t_{USRF} Measured between 10% and 90%- t_{CLK_TOL} 0.005 t_{CLK_DUTY} Measured at rising edge and/or failing edge at $OV_{DD}/2$ 40 t_{CLK_PJ} RMS value measured with a second-order, band-pass filter of 500 kHz to 4 MHz-	fUSB_CLK_IN24 t_{USRF} Measured between 10% and 90% t_{CLK_TOL} 0.0050 t_{CLK_DUTY} Measured at rising edge and/or failing edge at $OV_{DD}/2$ 4050 t_{CLK_PJ} RMS value measured with a second-order, band-pass filter of 500 kHz to 4 MHz	fUSB_CLK_IN24tUSRFMeasured between 10% and 90%6tCLK_TOL6tCLK_DUTYMeasured at rising edge and/or failing edge at $OV_{DD}/2$ 405060tCLK_PJRMS value measured with a second-order, band-pass filter of 500 kHz to 4 MHz5	$f_{USB_{CLK_IN}}$ 24-MHz t_{USRF} Measured between 10% and 90%6ns t_{CLK_TOL} 60.00500.005% t_{CLK_DUTY} Measured at rising edge and/or failing edge at $OV_{DD}/2$ 405060% t_{CLK_PJ} RMS value measured with a second-order, band-pass filter of 500 kHz to 4 MHz5ps

 Table 45.
 USBCLK AC timing specifications¹

1. For recommended operating conditions, see Table 3

2. System/board must be designed to ensure the input requirement to the device is achieved. Proper device operation is guaranteed for inputs meeting this requirement by design, simulation, characterization, or functional testing.

3.13 Integrated flash controller

This section describes the DC and AC electrical specifications for the integrated flash controller.

3.13.1 Integrated flash controller DC electrical characteristics

This table provides the DC electrical characteristics for the integrated flash controller when operating at OV_{DD} = 1.8 V.

Table 46.	. Integrated flash controller DC electrical characteristics (1.8 V) ³
-----------	--

Parameter	Symbol	Min	Мах	Unit	Note
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1
Input current	I _{IN}	-50	50	μA	2
$(V_{IN} = 0 V \text{ or } V_{IN} = OV_{DD})$					
Output high voltage	V _{OH}	1.6	-	V	-
(OV _{DD} = min, I _{OH} = -0.5 mA)					
Output low voltage	V _{OL}	-	0.32	V	-
(OV _{DD} = min, I _{OL} = 0.5 mA)					

1. The min V_{IL} and max V_{IH} values are based on the respective min and max OV_{IN} values found in Table 3.

2. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Recommended operating conditions.

Table 46. Integrated flash controller DC electrical characteristics (1.8 V)³

Parameter	Symbol	Min	Мах	Unit	Note
3. For recommended operating conditions,	see Table 3.				

3.13.2 Integrated flash controller AC timing specification

This section describes the AC timing specifications for the integrated flash controller.

3.13.2.1 Test condition

This figure provides the AC test load for the integrated flash controller.

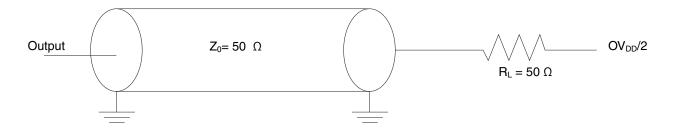


Figure 19. Integrated flash controller AC test load

3.13.2.2 Integrated flash controller AC timing specifications

All output signal timings are relative to the falling edge of any IFC_CLK. The external circuit must use the rising edge of the IFC_CLKs to latch the data.

All input timings are relative to the rising edge of IFC_CLKs.

This table describes the timing specifications of the integrated flash controller interface.

Table 47. Integrated flash controller timing specifications $(OV_{DD} = 1.8 V)^5$

Parameter/Condition	Symbol ¹	Min	Max	Unit	Notes
IFC_CLK cycle time	t _{IBK}	10	-	ns	-
IFC_CLK duty cycle	t _{IBKH} / t _{IBK}	45	55	%	-
IFC_CLK[n] skew to IFC_CLK[m]	t _{IBKSKEW}	0	±75	ps	2
Input setup	t _{IBIVKH}	4	-	ns	-
Input hold	t _{IBIXKH}	1	-	ns	-
Output delay	t _{IBKLOV}	-	1.5	ns	-
Output hold	t _{IBKLOX}	-2	-	ns	4

Table continues on the next page

Table 47. Integrated flash controller timing specifications $(OV_{DD} = 1.8 V)^5$ (continued)

Parameter/Condition	Symbol ¹	Min	Max	Unit	Notes		
IFC_CLK to output high impedance for AD	t _{IBKLOZ}	-	2	ns	3		
1. All signals are measured from OV _{DD} /2 of rising/falling edge of IFC_CLK to OV _{DD} /2 of the signal in question.							
2. Skew measured between different IFC_CLK signals at OV _{DD} /2.							
3. For purposes of active/float timing measurements, the high impedelivered through the component pin is less than or equal to the le				he total c	urrent		
4. Here the negative sign means output transit happens earlier than the falling edge of IFC_CLK.							
5. For recommended operating conditions, see Table 3.							

This figure shows the AC timing diagram.

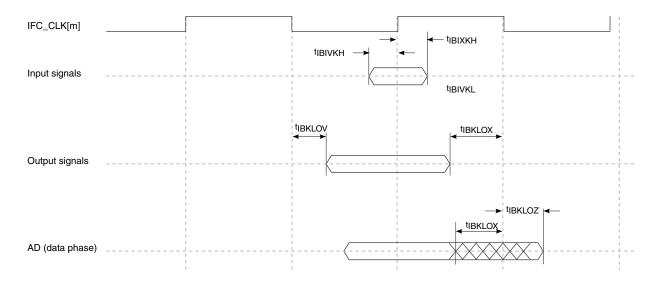


Figure 20. Integrated flash controller signals

The figure above applies to all the controllers that IFC supports.

- For input signals, the AC timing data is used directly for all controllers.
- For output signals, each type of controller provides its own unique method to control the signal timing. The final signal delay value for output signals is the programmed delay plus the AC timing delay.

This figure shows how the AC timing diagram applies to GPCM. The same principle also applies to other controllers of IFC.

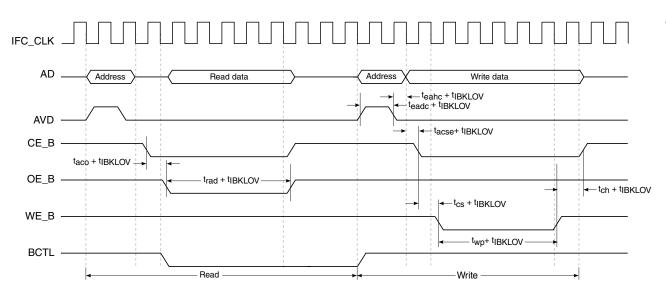


Figure 21. GPCM output timing diagram^{1, 2}

Notes for figure:

1. t_{aco} , t_{rad} , t_{eahc} , t_{eadc} , t_{acse} , t_{cs} , t_{ch} , t_{wp} are programmable. See the chip reference manual.

2. For output signals, each type of controller provides its own unique method to control the signal timing. The final signal delay value for output signals is the programmed delay plus the AC timing delay.

3.14 Enhanced secure digital host controller (eSDHC)

This section describes the DC and AC electrical specifications for the eSDHC interface.

3.14.1 eSDHC DC electrical characteristics

This table provides the DC electrical characteristics for the eSDHC interface.

Parameter	Symbol	Min	Мах	Unit	Notes
Input high voltage	V _{IH}	0.7 x OV _{DD}	-	V	1
Input low voltage	V _{IL}	-	0.3 x OV _{DD}	V	1
I/O leakage current	I _{IN} /I _{OZ}	-50	50	μA	-
Output high voltage (I_{OH} = -100 µA at OV_{DD} min)	V _{OH}	OV _{DD} - 0.2 V	-	V	-
Output low voltage (I_{OL} = 100 µA at OV_{DD} min)	V _{OL}	-	0.2	V	-

Table 48. eSDHC interface DC electrical characteristics (dual-voltage cards)³

Table continues on the next page...

Electrical characteristics

Table 48. eSDHC interface DC electrical characteristics (dual-voltage cards)³ (continued)

Parameter	Symbol	Min	Max	Unit	Notes			
Output high voltage ($I_{OH} = -100 \ \mu A$)	V _{OH}	OV _{DD} - 0.2 V	-	V	2			
Output low voltage (I _{OL} = 2 mA)	V _{OL}	-	0.3	V	2			
1. The min V_{IL} and V_{IH} values are based on the respective min and max OV_{IN} values found in Table 3.								
2. Open-drain mode is for MMC cards only.								
3. For recommended operating conditions, see Table 3.								

3.14.2 eSDHC AC timing specifications

This table provides the eSDHC AC timing specifications as defined in Figure 22.

Parameter/Condition	Symbol ¹	Min	Max	Unit	Notes
SD_CLK clock frequency:	f _{SHSCK}	0		MHz	2, 4
SD/SDIO Full-speed/high-speed mode			25/50		
MMC Full-speed/high-speed mode			20/52		
SD_CLK clock low time-Full-speed/High-speed mode	t _{SHSCKL}	10/7	-	ns	4
SD_CLK clock high time-Full-speed/High-speed mode	t _{SHSCKH}	10/7	-	ns	4
SD_CLK clock rise and fall times	t _{SHSCKR/}	-	3	ns	4
	t _{SHSCKF}				
Input setup times: SD_CMD, SD_DATx, SD_CD to SD_CLK	t _{SHSIVKH}	2.5	-	ns	3, 4, 5
Input hold times: SD_CMD, SD_DATx, SD_CD to SD_CLK	t _{SHSIXKH}	2.5	-	ns	4, 5
Output hold time: SD_CLK to SD_CMD, SD_DATx valid	t _{SHSKHOX}	-3	-	ns	4, 5
Output delay time: SD_CLK to SD_CMD, SD_DATx valid	t _{SHSKHOV}	-	3	ns	4, 5

Table 49. eSDHC AC timing specifications⁶

Notes:

1. The symbols used for timing specifications herein follow the pattern of t_{(first three letters of functional block)(signal)(state)} (reference)(state) for inputs and t_{(first three letters of functional block)(reference)(state)(signal)(state)} for outputs. For example, t_{FHSKHOV} symbolizes eSDHC high-speed mode device timing (SHS) clock reference (K) going to the high (H) state, with respect to the output (O) reaching the invalid state (X) or output hold time. Note that in general, the clock reference symbol is based on five letters representing the clock of a particular functional. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

2. In full-speed mode, the clock frequency value can be 0-25 MHz for an SD/SDIO card and 0-20 MHz for an MMC card. In high-speed mode, the clock frequency value can be 0-50 MHz for an SD/SDIO card and 0-52 MHz for an MMC card.

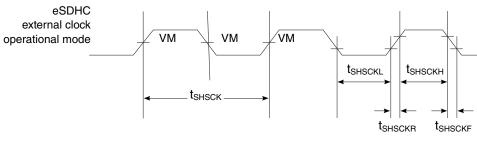
3. To satisfy setup timing, one-way board-routing delay between Host and Card, on SD_CLK, SD_CMD, and SD_DATx should not exceed 1 ns for any high speed MMC card. For any high speed or default speed mode SD card, the one way board routing delay between Host and Card, on SD_CLK, SD_CMD, and SD_DATx should not exceed 1.5ns.

4. $C_{CARD} \le 10 \text{ pF}$, (1 card), and $C_L = C_{BUS} + C_{HOST} + C_{CARD} \le 40 \text{ pF}$.

5. The parameter values apply to both full-speed and high-speed modes.

6. For recommended operating conditions, see Table 3.

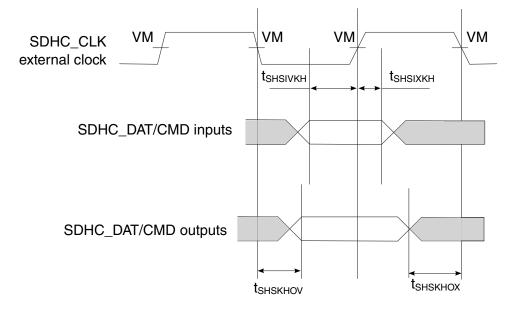
This figure provides the eSDHC clock input timing diagram.



 $VM = Midpoint voltage (OV_{DD}/2)$

Figure 22. eSDHC clock input timing diagram

This figure provides the data and command input/output timing diagram.



VM = Midpoint voltage (OV_{DD}/2)

Figure 23. eSDHC data and command input/output timing diagram referenced to clock

3.15 Multicore programmable interrupt controller (MPIC)

This section describes the DC and AC electrical specifications for the multicore programmable interrupt controller.

3.15.1 MPIC DC specifications

This figure provides the DC electrical characteristics for the MPIC interface.

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	VIL	-	0.6	V	1
Input current ($OV_{IN} = 0 V \text{ or } OV_{IN} = OV_{DD}$)	I _{IN}	-50	50	μA	2
Output high voltage (OV _{DD} = min, I _{OH} = -0.5 mA)	V _{OH}	1.35	-	V	-
Output low voltage ($OV_{DD} = min, I_{OL} = 0.5 mA$)	V _{OL}	-	0.4	V	-
Note:					
1. The min V_{IL} and max V_{IH} values are based on the	min and max C	V _{IN} respective	e values found	in Table 3.	
2. The symbol $\Omega_{\rm Max}$ in this case, represents the $\Omega_{\rm Max}$	symbol refer	anced in Table	- 3		

Table 50. MPIC DC electrical characteristics $(OV_{DD} = 1.8 \text{ V})^3$

2. The symbol OV_{IN} , in this case, represents the OV_{IN} symbol referenced in Table 3.

3. For recommended operating conditions, see Table 3.

3.15.2 MPIC AC timing specifications

This table provides the MPIC input and output AC timing specifications.

Table 51. MPIC Input AC timing specifications²

Parameter/Condition	Symbol Min		Max	Unit	Notes		
MPIC inputs-minimum pulse width t _{PIWID} 3 - SYSCLKs 1							
1. MPIC inputs and outputs are asynchron external synchronous logic. MPIC inputs a working in edge triggered mode.							
2. For recommended operating conditions	see Table 3.						

3.16 JTAG controller

This section describes the DC and AC electrical specifications for the IEEE 1149.1 (JTAG) interface.

3.16.1 JTAG DC electrical characteristics

This table provides the JTAG DC electrical characteristics.

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1

Table continues on the next page ...

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Table 52. JTAG DC electrical characteristics $(OV_{DD} = 1.8V)^3$ (continued)

Parameter	Symbol	Min	Max	Unit	Notes		
Input current (OV _{IN} = 0 V or OV _{IN} = OV _{DD})	I _{IN}	-100	50	μA	2, 4		
Output high voltage (OV_{DD} = min, I_{OH} = -0.5 mA)	V _{OH}	1.35	-	V	-		
Output low voltage ($OV_{DD} = min, I_{OL} = 0.5 mA$)	V _{OL}	-	0.4	V	-		
Notes:			•		•		
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on the	respective min	and max OV _{II}	N values found	in Table 3.			
2. The symbol V_{IN} , in this case, represents the OV _{IN} symbol found in Table 3.							
3. For recommended operating conditions, see Table	e 3.						
4. TMI, TMS, and TRST_B have internal pull-ups per	r the IEEE Std.	1149.1 speci	fication.				

3.16.2 JTAG AC timing specifications

This table provides the JTAG AC timing specifications as defined in Figure 24 through Figure 27.

Paran	neter/Condition	Symbol ¹	Min	Max	Unit	Notes
JTAG external clock fr	equency of operation	f _{JTG}	0	33.3	MHz	5
JTAG external clock cy	cle time	t _{JTG}	30	-	ns	6
JTAG external clock pulse width measured at 1.4 V		t _{JTKHKL}	15	-	ns	7
JTAG external clock rise and fall times		t _{JTGR} /t _{JTGF}	0	2	ns	8
TRST_B assert time		t _{TRST}	25	-	ns	2
Input setup times		t _{JTDVKH}	4.5	-	ns	9
Input hold times		t _{JTDXKH}	11	-	ns	10
Output valid times	Boundary-scan data	t _{JTKLDV}	-	15	ns	3, 11, 12
	TDO		-	10		
Output hold times		t _{JTKLDX}	0	-	ns	3

Table 53. JTAG AC timing specifications⁴

Notes:

1. The symbols used for timing specifications follow the pattern t_{(first two letters of functional block)(signal)(state)(reference)(state) for inputs and t_{(first two letters of functional block)(reference)(state)(signal)(state)} for outputs. For example, t_{JTDVKH} symbolizes JTAG device timing (JT) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{JTG} clock reference (K) going to the high (H) state or setup time. Also, t_{JTDXKH} symbolizes JTAG timing (JT) with respect to the time data input signals (D) reaching the invalid state (X) relative to the t_{JTG} clock reference (K) going to the high (H) state. Note that in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).}

2.TRST_B is an asynchronous level sensitive signal. The setup time is for test purposes only.

3. All outputs are measured from the midpoint voltage of the falling edge of t_{TCLK} to the midpoint of the signal in question. The output timings are measured at the pins. All output timings assume a purely resistive 50- Ω load. Time-of-flight delays must be added for trace lengths, vias, and connectors in the system.

4. For recommended operating conditions, see Table 3.

Electrical characteristics

		•			
Parameter/Condition	Symbol ¹	Min	Max	Unit	Notes
5. TCK frequency can be as high as 100MHz for interna	al debug modes		•	•	
6. If TCK = 100 MHz then t_{JTG} = 10 nsec					
7. If TCK = 100 MHz then t_{JTKHKL} = 5 nsec					
8. If TCK = 100 MHz then t _{JTGR} /t _{JTGF} =<1 nsec					
9. If TCK = 100 MHz then t_{JTDVKH} = 1.33 nsec					
10. If TCK = 100 MHz then t_{JTDXKH} = 3.3 nsec					
11. Due to value of tJTKLDV, after Update-IR or Updat	e-DR transitions	for EXTEST*	or CLAMP inst	ructions, a tra	Insition

 Table 53. JTAG AC timing specifications⁴

through the optional Run-Test-Idle state is recommended to allow for board level propagation and setup times of observation points.

12. DDR output pins when transitioning from a tristate to driving a logic 1 or 0 can require up to 24ns. Use of Run-Test Idle state is recommended after Update-IR or Update-DR TAP states.

This figure provides the AC test load for TDO and the boundary-scan outputs of the device.

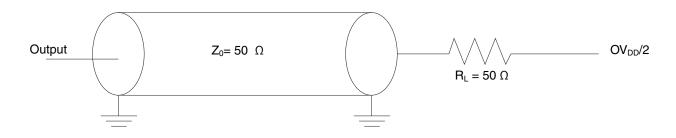
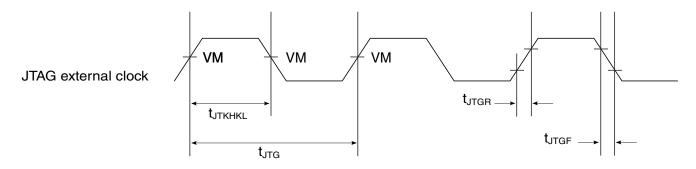


Figure 24. AC test load for the JTAG interface

This figure provides the JTAG clock input timing diagram.



VM = Midpoint voltage (OV_{DD}/2)

Figure 25. JTAG clock input timing diagram

This figure provides the TRST_B timing diagram.

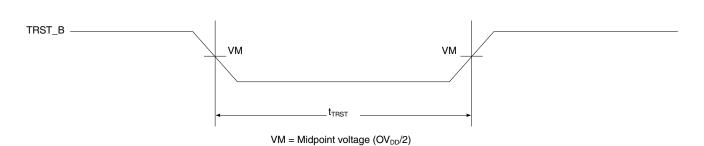


Figure 26. TRST_B timing diagram

This figure provides the boundary-scan timing diagram.

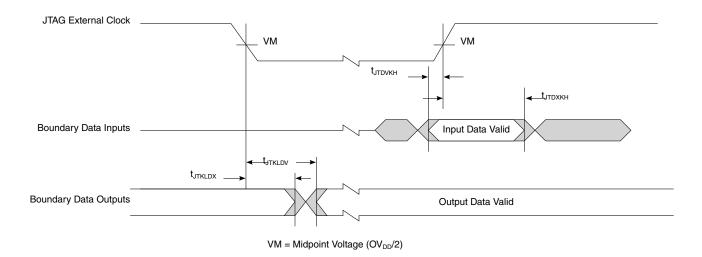


Figure 27. Boundary-scan timing diagram

3.17 I²C interface

This section describes the DC and AC electrical characteristics for the I²C interface.

3.17.1 I²C DC electrical characteristics

This table provides the DC electrical characteristics for the I^2C interfaces operating at 2.5V.

Table 54. I²C DC electrical characteristics $(DV_{DD} = 2.5V)^5$

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.7	-	V	1

Table continues on the next page...

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Electrical characteristics

Table 54.	. I ² C DC electrical characteristic	s (DV _{DD} = 2.5V)	⁵ (continued)
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Parameter	Symbol	Min	Max	Unit	Notes
Input low voltage	VIL	-	0.7	V	1
Output low voltage (DV _{DD} = min, I_{OL} = 3 mA)	V _{OL}	0	0.4	V	2
Pulse width of spikes which must be suppressed by the input filter	t _{I2KHKL}	0	50	ns	3
Leakage Input current at each I/O pin (input voltage is between 0.1 x $\rm DV_{DD}$ and 0.9 x $\rm DV_{DD}(max)$	I _{OZ}	-50	50	μA	4
Capacitance for each I/O pin	CI	-	10	pF	-
Notes:					
1. The min $V_{\text{IL}}\text{and} \max V_{\text{IH}}$ values are based on the respective min and	max DV _{IN} va	lues found	in Table 3.		
2. See the chip reference manual for information about the digital filter	used.				
3. I/O pins obstruct the SDA and SCL lines if DV_{DD} is switched off.					
4. For recommended operating conditions, see Table 3.					

This table provides the DC electrical characteristics for the I^2C interfaces operating at 1.8V.

Table 55. $I^{2}C$ DC electrical characteristics (DV_{DD} = 1.8V)⁵

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1
Output low voltage (DV _{DD} = min, I_{OL} = 3 mA)	V _{OL}	0	0.36	V	2
Pulse width of spikes which must be suppressed by the input filter	t _{I2KHKL}	0	50	ns	3
Leakage Input current each I/O pin (input voltage is between 0.1 x $\rm DV_{DD}$ and 0.9 x $\rm DV_{DD}(max)$	I _{OZ}	-50	50	μA	4
Capacitance for each I/O pin	CI	-	10	pF	-
Notos		-		-	1

Notes:

1. The min V_{IL} and max V_{IH} values are based on the respective min and max DV_{IN} values found in Table 3.

2. See the chip reference manual for information about the digital filter used.

3. I/O pins obstruct the SDA and SCL lines if DV_{DD} is switched off.

4. For recommended operating conditions, see Table 3.

3.17.2 I²C AC timing specifications

This table provides the AC timing parameters for the I²C interfaces.

Parameter/Condition	Symbol ¹	Min	Max	Unit	Notes
SCL clock frequency	f _{I2C}	0	400	kHz	2
Low period of the SCL clock	t _{I2CL}	1.3	-	μs	—
High period of the SCL clock	t _{I2CH}	0.6	-	μs	—
Setup time for a repeated START condition	t _{I2SVKH}	0.6	-	μs	—
Hold time (repeated) START condition (after this period, the first clock pulse is generated)	t _{I2SXKL}	0.6	—	μs	—
Data setup time	t _{I2DVKH}	100	-	ns	—
Data input hold time:	t _{I2DXKL}	—	-	μs	3
Data output delay time	t _{I2OVKL}	—	0.9	μs	4
Setup time for STOP condition	t _{I2PVKH}	0.6	-	μs	—
Bus free time between a STOP and START condition	t _{I2KHDX}	1.3	-	μs	—
Noise margin at the LOW level for each connected device (including hysteresis)	V _{NL}	0.1 x OV _{DD}	—	V	_
Noise margin at the HIGH level for each connected device (including hysteresis)	V _{NH}	0.2 x OV _{DD}	—	V	—
Capacitive load for each bus line	Cb	—	400	pF	—
NL . L			•		

Table 56. I²C AC timing specifications⁵

Notes:

1. The symbols used for timing specifications herein follow the pattern $t_{(first two letters of functional block)(signal)(state)(reference)(state) for inputs and <math>t_{(first two letters of functional block)(reference)(state)(signal)(state)}$ for outputs. For example, t_{I2DVKH} symbolizes I²C timing (I2) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I²C timing (I2) for the time that the data with respect to the START condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the store (K) going to the time that the data with respect to the t_{I2C} clock reference (K) reference (K) going to the time that the data with respect to the t_{I2C} clock reference (K) going to the time that the data with respect to the t_{I2C} clock reference (K) going to the time that the data with respect to the t_{I2C} clock reference (K) going to the high (H) state or setup time.

2. The requirements for I²C frequency calculation must be followed. See *Determining the I²C Frequency Divider Ratio for SCL* (AN2919).

3. As a transmitter, the chip provides a delay time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL to avoid unintended generation of a START or STOP condition. When the chip acts as the l²C bus master while transmitting, it drives both SCL and SDA. As long as the load on SCL and SDA are balanced, the chip does not generate an unintended START or STOP condition. Therefore, the 300 ns SDA output delay time is not a concern. If, under some rare condition, the 300 ns SDA output delay time is required for the chip as transmitter, see *Determining the l²C Frequency Divider Ratio for SCL* (AN2919).

4. The maximum t_{I2OVKL} has to be met only if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.

5. For recommended operating conditions, see Table 3.

This figure provides the AC test load for the I^2C .

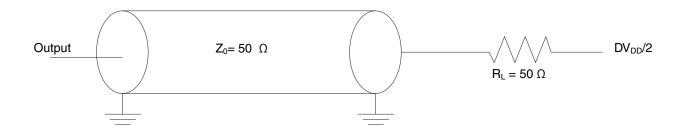


Figure 28. I²C AC test load

This figure shows the AC timing diagram for the I^2C bus.

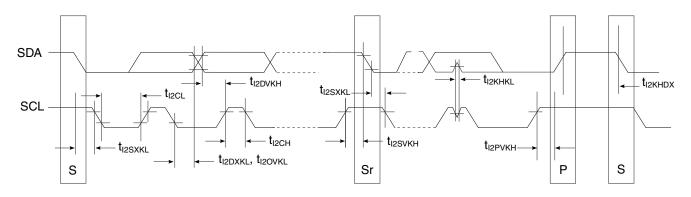


Figure 29. I²C Bus AC timing diagram

3.18 GPIO interface

This section describes the DC and AC electrical characteristics for the GPIO interface.

3.18.1 GPIO DC electrical characteristics

This table provides the DC electrical characteristics for GPIO pins operating at $LV_{DD} = 2.5 V$.

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.7	-	V	1
Input low voltage	V _{IL}	-	0.7	V	1
Input current ($V_{IN} = 0 V \text{ or } V_{IN} = LV_{DD}$)	I _{IN}	-50	50	μA	2
Output high voltage	V _{OH}	2.0	-	V	-
(LV _{DD} = min, I _{OH} = -1 mA)					

Table 57. GPIO DC electrical characteristics (2.5 V)³

Table continues on the next page...

Table 57. GPIO DC electrical characteristics (2.5 V)³ (continued)

Parameter	Symbol	Min	Max	Unit	Notes		
Output low voltage	V _{OL}	-	0.4	V	-		
$(LV_{DD} = min, I_{OL} = 1 mA)$							
1. The min V_{IL} and max V_{IH} values are based	on the respectiv	e min and max L	V _{IN} values found	in Table 3 .	•		
2. The symbol V _{IN} , in this case, represents the LV _{IN} symbol referenced in Recommended operating conditions.							
3. For recommended operating conditions, see Table 3.							

This table provides the DC electrical characteristics for GPIO pins operating at LV_{DD} or OV_{DD} = 1.8 V.

Table 58. GPIO DC electrical characteristics (1.8 V)³

Parameter	Symbol	Min	Max	Unit	Notes
Input high voltage	V _{IH}	1.25	-	V	1
Input low voltage	V _{IL}	-	0.6	V	1
Input current ($V_{IN} = 0 V \text{ or } V_{IN} = L/OV_{DD}$)	I _{IN}	-50	50	μA	2
Output high voltage	V _{OH}	1.35	-	V	-
$(L/OV_{DD} = min, I_{OH} = -0.5 mA)$					
Output low voltage	V _{OL}	-	0.4	V	-
$(L/OV_{DD} = min, I_{OL} = 0.5 mA)$					
1. The min V_{IL} and max V_{IH} values are bas	ed on the respec	tive min and ma	x L/OV _{IN} values fo	ound in Table 3.	
2. The symbol V_{IN} , in this case, represents	s the L/OV _{IN} sym	ool referenced in	Recommended o	perating condit	ions.
3. For recommended operating conditions	, see Table 3.				

This table provides the DC electrical characteristics for the LP Trust pin, LP_TMP_DETECT_B, operating at $V_{DD}LP = 1 \text{ V}$.

Table 59. LP_TMP_DETECT_B Pin DC electrical characteristics (1 V)³

Parameter	Symbol	Min	Мах	Unit	Notes
Input high voltage	V _{IH}	0.8 x V _{DD_LP}	-	V	1
Input low voltage	V _{IL}	-	0.4 x V _{DD_LP}	V	1
Input current ($V_{IN_LP} = 0 V \text{ or } V_{IN_LP} = V_{DD_LP}$)	I _{IN}	-50	50	μΑ	2
1. The min $V_{\text{IL}} \text{and} \max V_{\text{IH}}$ values are ba	sed on the respect	ive min and max	V _{DD_LP} values for	und in Table 3.	
2. The symbol V_{IN_LP} , in this case, repres	sents the V _{IN_LP} syr	mbol referenced i	n Recommended	operating condi	tions.

3. For recommended operating conditions, see Table 3.

3.18.2 GPIO AC timing specifications

This table provides the GPIO input and output AC timing specifications.

 Table 60. GPIO input AC timing specifications²

Parameter/Condition	Symbol	Min	Unit	Notes				
GPIO inputs—minimum pulse width	puts—minimum pulse width t _{PIWID} 20 ns 1							
Notes:	1							
1. GPIO inputs and outputs are asynchronous to any visible clock. GPIO outputs should be synchronized before use by any external synchronous logic. GPIO inputs are required to be valid for at least t _{PIWID} to ensure proper operation.								
2. For recommended operating conditions, see Table 3.								

This figure provides the AC test load for the GPIO.

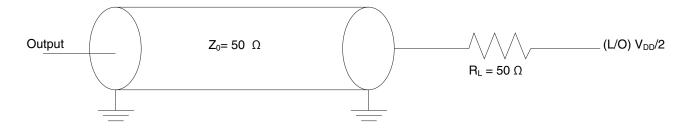


Figure 30. GPIO AC test load

3.19 High-speed serial interfaces (HSSI)

The chip features a serializer/deserializer (SerDes) interface to be used for high-speed serial interconnect applications. The SerDes interface can be used for PCI Express, SATA, Serial RapidIO, XAUI, XFI, 10GBase-KR, Aurora, Interlaken LA, HiGig/HiGig2, SGMII, 2.5x SGMII and QSGMII data transfers.

This section describes the common portion of SerDes DC electrical specifications: the DC requirement for SerDes reference clocks. The SerDes data lane's transmitter (Tx) and receiver (Rx) reference circuits are also shown.

3.19.1 Signal terms definition

The SerDes utilizes differential signaling to transfer data across the serial link. This section defines the terms that are used in the description and specification of differential signals.

This figure shows how the signals are defined. For illustration purposes only, one SerDes lane is used in the description. This figure shows the waveform for either a transmitter output (SD_TX*n* and SD_TX*n*_B) or a receiver input (SD_RX*n* and SD_RX*n*_B). Each signal swings between A volts and B volts where A > B.

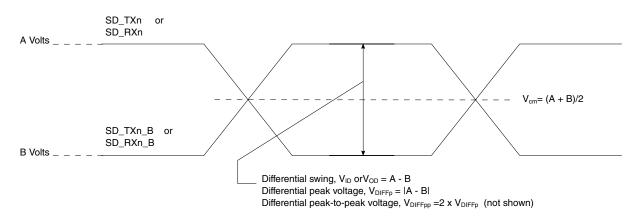


Figure 31. Differential voltage definitions for transmitter or receiver

Using this waveform, the definitions are as shown in the following list. To simplify the illustration, the definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signaling environment:

Single-Ended Swing

The transmitter output signals and the receiver input signals SD_TXn , SD_TXn_B , SD_RXn and SD_RXn_B each have a peak-to-peak swing of A - B volts. This is also referred as each signal wire's single-ended swing.

Differential Output Voltage, V_{OD} (or Differential Output Swing)

The differential output voltage (or swing) of the transmitter, V_{OD} , is defined as the difference of the two complementary output voltages: V_{SD_TXn} - $V_{SD_TXn_B}$. The V_{OD} value can be either positive or negative.

Differential Input Voltage, $V_{ID} \left(\text{or Differential Input Swing} \right)$

The differential input voltage (or swing) of the receiver, V_{ID} , is defined as the difference of the two complementary input voltages: V_{SD_RXn} - $V_{SD_RXn_B}$. The V_{ID} value can be either positive or negative.

Differential Peak Voltage, V_{DIFFp}

The peak value of the differential transmitter output signal or the differential receiver input signal is defined as the differential peak voltage, $V_{DIFFp} = |A - B|$ volts. Differential Peak-to-Peak, $V_{DIFFp-p}$

Since the differential output signal of the transmitter and the differential input signal of the receiver each range from A - B to -(A - B) volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak-to-peak voltage, $V_{DIFFp-p} = 2 \times V_{DIFFp} = 2 \times |(A - B)|$ volts, which is twice the differential swing in amplitude, or twice of the differential peak. For

Electrical characteristics

example, the output differential peak-to-peak voltage can also be calculated as $V_{TX-DIFFp-p} = 2 \times |V_{OD}|$.

Differential Waveform

The differential waveform is constructed by subtracting the inverting signal $(SD_TXn_B, \text{ for example})$ from the non-inverting signal $(SD_TXn, \text{ for example})$ within a differential pair. There is only one signal trace curve in a differential waveform. The voltage represented in the differential waveform is not referenced to ground. See Figure 36 as an example for differential waveform.

Common Mode Voltage, V_{cm}

The common mode voltage is equal to half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm_out} = (V_{SD_TXn} + V_{SD_TXn_B}) \div 2 = (A + B) \div 2$, which is the arithmetic mean of the two complementary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. It may be different between the receiver input and driver output circuits within the same component. It is also referred to as the DC offset on some occasions.

To illustrate these definitions using real values, consider the example of a current mode logic (CML) transmitter that has a common mode voltage of 2.25 V and outputs, TD and TD_B. If these outputs have a swing from 2.0 V to 2.5 V, the peak-to-peak voltage swing of each signal (TD or TD_B) is 500 mV p-p, which is referred to as the single-ended swing for each signal. Because the differential signaling environment is fully symmetrical in this example, the transmitter output's differential swing (V_{OD}) has the same amplitude as each signal's single-ended swing. The differential output signal ranges between 500 mV and -500 mV. In other words, V_{OD} is 500 mV in one phase and -500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage (V_{DIFFp}) is 1000 mV p-p.

3.19.2 SerDes reference clocks

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clocks inputs are SD1_REF_CLK[1:2] and SD1_REF_CLK[1:2]_B for SerDes 1, SD2_REF_CLK[1:2] and SD2_REF_CLK[1:2]_B for SerDes 2, SD3_REF_CLK[1:2] and SD3_REF_CLK[1:2]_B for SerDes 3 and SD4_REF_CLK[1:2] and SD4_REF_CLK[1:2]_B for SerDes 4.

SerDes 1-4 may be used for various combinations of the following IP blocks based on the RCW Configuration field SRDS_PRTCLn:

• SerDes 1: SGMII (1.25 and 3.125 Gbaud), QSGMII (5 Gbps only), HiGig/HiGig2 (3.125 Gbps), HiGig/HiGig2 (3.75 Gbps) or XAUI (3.125 Gb/s)

- SerDes 2: SGMII (1.25 and 3.125 Gbaud), QSGMII (5 Gbps only), XAUI (3.125 Gb/s), HiGig/HiGig2 (3.125 Gbps), HiGig/HiGig2 (3.75 Gbps), XFI (10.3125 Gb/s only) or 10GBase-KR (10.3125 Gbaud only)
- SerDes 3: PEX1/2 (2.5, 5, and 8 GT/s), SRIO1(2.5, 3.125, and 5 Gbaud) or Interlaken-LA(6.25)
- SerDes 4: PEX3/4 (2.5, 5, and 8 GT/s), SRIO2(2.5, 3.125, and 5 Gbaud), Aurora (2.5, 3.125, and 5 Gbps) or SATA1/2 (1.5 and 3.0 Gbps)

The following sections describe the SerDes reference clock requirements and provide application information.

3.19.2.1 SerDes spread-spectrum clock source recommendations

SDn_REF_CLKn/SDn_REF_CLKn_B are designed to work with spread-spectrum clock for PCI Express protocol only with the spreading specification defined in Table 61. When using spread-spectrum clocking for PCI Express, both ends of the link partners should use the same reference clock. For best results, a source without significant unintended modulation must be used.

For SATA protocol, the SerDes transmitter does not support spread-spectrum clocking. The SerDes receiver does support spread-spectrum clocking on receive, which means the SerDes receiver can receive data correctly from a SATA serial link partner using spreadspectrum clocking

The spread-spectrum clocking cannot be used if the same SerDes reference clock is shared with other non-spread-spectrum supported protocols. For example, if the spread-spectrum clocking is desired on a SerDes reference clock for PCI Express and the same reference clock is used for any other protocol such as SATA/SGMII/QSGMII/SRIO/XAUI due to the SerDes lane usage mapping option, spread-spectrum clocking cannot be used at all.

Parameter	Min	Max	Unit	Notes
Frequency modulation	30	33	kHz	-
Frequency spread	+0	-0.5	%	2
1. At recommended operating conditions. See Table 3.	•			•
2. Only down-spreading is allowed.				

Table 61. SerDes spread-spectrum clock source recommendations ¹

3.19.2.2 SerDes reference clock receiver characteristics

This figure shows a receiver reference diagram of the SerDes reference clocks.

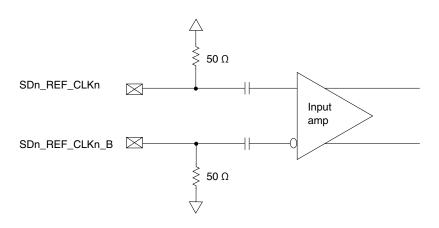


Figure 32. Receiver of SerDes reference clocks

The characteristics of the clock signals are as follows:

- The SerDes transceivers core power supply voltage requirements $(SV_{DD}n)$ are as specified in Recommended operating conditions.
- The SerDes reference clock receiver reference circuit structure is as follows:
 - The SDn_REF_CLKn and SDn_REF_CLKn_B are internally AC-coupled differential inputs as shown in Figure 32. Each differential clock input (SDn_REF_CLKn or SDn_REF_CLKn_B) has on-chip 50 Ω termination to SGNDn followed by on-chip AC-coupling.
 - The external reference clock driver must be able to drive this termination.
 - The SerDes reference clock input can be either differential or single-ended. See the differential mode and single-ended mode descriptions below for detailed requirements.
- The maximum average current requirement also determines the common mode voltage range.
 - When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8 mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA because the input is AC-coupled on-chip.
 - This current limitation sets the maximum common mode input voltage to be less than 0.4 V ($0.4 \text{ V} \div 50 \Omega = 8 \text{ mA}$) while the minimum common mode input level is 0.1 V above SGND*n*. For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0 mA to 16 mA (0-0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800 mV with the common mode voltage at 400 mV.
 - If the device driving the SD*n*_REF_CLK*n* and SD*n*_REF_CLK*n*_B inputs cannot drive 50 Ω to SGND*n* DC or the drive strength of the clock driver chip exceeds the maximum input current limitations, it must be AC-coupled off-chip.
- The input amplitude requirement is described in detail in the following sections.

3.19.2.3 DC-level requirement for SerDes reference clocks

The DC level requirement for the SerDes reference clock inputs is different depending on the signaling mode used to connect the clock driver chip and SerDes reference clock inputs, as described below:

- Differential Mode
 - The input amplitude of the differential clock must be between 400 mV and 1600 mV differential peak-to-peak (or between 200 mV and 800 mV differential peak). In other words, each signal wire of the differential pair must have a single-ended swing of less than 800 mV and greater than 200 mV. This requirement is the same for both external DC-coupled or AC-coupled connection.
 - For an external DC-coupled connection, as described in SerDes reference clock receiver characteristics, the maximum average current requirements sets the requirement for average voltage (common mode voltage) as between 100 mV and 400 mV. Figure 33 shows the SerDes reference clock input requirement for DC-coupled connection scheme.

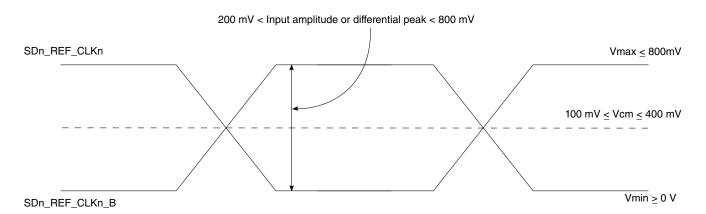


Figure 33. Differential reference clock input DC requirements (external DC-coupled)

• For an external AC-coupled connection, there is no common mode voltage requirement for the clock driver. Because the external AC-coupling capacitor blocks the DC level, the clock driver and the SerDes reference clock receiver operate in different common mode voltages. The SerDes reference clock receiver in this connection scheme has its common mode voltage set to SGND*n*. Each signal wire of the differential inputs is allowed to swing below and above the common mode voltage (SGND*n*). Figure 34 shows the SerDes reference clock input requirement for AC-coupled connection scheme.

Electrical characteristics

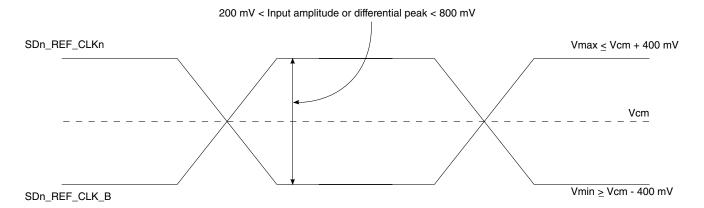


Figure 34. Differential reference clock input DC requirements (external AC-coupled)

- Single-Ended Mode
 - The reference clock can also be single-ended. The SDn_REF_CLKn input amplitude (single-ended swing) must be between 400 mV and 800 mV peak-topeak (from V_{MIN} to V_{MAX}) with SDn_REF_CLKn_B either left unconnected or tied to ground.
 - The SDn REF CLKn input average voltage must be between 200 and 400 mV. Figure 35 shows the SerDes reference clock input requirement for single-ended signaling mode.
 - To meet the input amplitude requirement, the reference clock inputs may need to be DC- or AC-coupled externally. For the best noise performance, the reference of the clock could be DC- or AC-coupled into the unused phase

(SDn REF CLKn B) through the same source impedance as the clock input (SDn_REF_CLKn) in use.

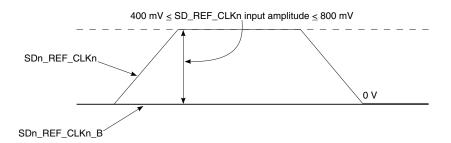


Figure 35. Single-ended reference clock input DC requirements

AC requirements for SerDes reference clocks 3.19.2.4

This table lists the AC requirements for SerDes reference clocks for protocols running at data rates up to 8 Gb/s.

This includes PCI Express (2.5, 5, 8 GT/s), SGMII (1.25 Gbaud), 2.5x SGMII (3.125 Gbaud), QSGMII (5 Gbps), Serial RapidIO (2.5, 3.125, 5 Gbaud), Aurora (2.5, 3.125, 5 Gbps), HiGig/HiGig2 (3.125 Gbps), HiGig/HiGig2 (3.75 Gbps), XAUI (3.125 Gb/s) and Interlaken-LA (6.25 Gbps) SerDes reference clocks to be guaranteed by the customer's application design.

Table 62. SDn_REF_CLKn and $SDn_REF_CLKn_B$ input clock requirements ($SnV_{DD} = 1.0 V$)

Parameter	Symbol	Min	Тур	Max	Unit	Notes
SDn_REF_CLKn/SDn_REF_CLKn_B frequency range	t _{CLK_REF}	-	100/125/156.25	-	MHz	2
SDn_REF_CLKn/SDn_REF_CLKn_B clock frequency tolerance	t _{CLK_TOL}	-300	-	300	ppm	3, 12
SDn_REF_CLKn/SDn_REF_CLKn_B clock frequency tolerance	t _{CLK_TOL}	-100	-	100	ppm	4, 12
SDn_REF_CLKn/SDn_REF_CLKn_B reference clock duty cycle	t _{CLK_DUTY}	40	50	60	%	5
SDn_REF_CLKn/ SDn_REF_CLKn_B max deterministic peak-to-peak jitter at 10 ⁻⁶ BER	t _{CLK_DJ}	-	-	42	ps	-
SDn_REF_CLKn/SDn_REF_CLKn_B total reference clock jitter at 10 ⁻⁶ BER (peak-to-peak jitter at refClk input)	t _{CLK_TJ}	-	-	86	ps	6
SDn_REF_CLKn/SDn_REF_CLKn_B 10 kHz to 1.5 MHz RMS jitter	t _{REFCLK-LF-RMS}	-	-	3	ps RMS	7
SDn_REF_CLKn/SDn_REF_CLKn_B > 1.5 MHz to Nyquist RMS jitter	tREFCLK-HF-RMS	-	-	3.1	ps RMS	7
SDn_REF_CLKn/SDn_REF_CLKn_B RMS reference clock jitter	tREFCLK-RMS-DC	-	-	1	ps RMS	8
SDn_REF_CLKn/ SDn_REF_CLKn_B rising/falling edge rate	t _{CLKRR} /t _{CLKFR}	1	-	4	V/ns	9
Differential input high voltage	-	V _{CM} +200 m V	-	-	mV	5
Differential input low voltage	-	-	-	V _{CM} -20 0 mV	mV	5
Rising edge rate (SDn_REF_CLKn) to falling edge rate (SDn_REF_CLKn) matching	Rise-Fall Matching	-	-	20	%	10, 11

1. For recommended operating conditions, see Table 3.

2. Caution: Only 100, 125 and 156.25 have been tested. In-between values do not work correctly with the rest of the system.

3. For PCI Express (2.5, 5, 8 GT/s)

4. For SGMII, 2.5x SGMII, QSGMII, sRIO, HiGig/HiGig2, XAUI, Interlaken-LA, Aurora

5. Measurement taken from differential waveform. VCM is the common mode voltage.

6. Limits from PCI Express CEM Rev 2.0

7. For PCI Express-5 GT/s, per PCI Express base specification rev 3.0

8. For PCI-Express-8 GT/s, per PCI-Express base specification rev 3.0

Table 62. SDn_REF_CLKn and $SDn_REF_CLKn_B$ input clock requirements ($SnV_{DD} = 1.0 V$)

Parameter	Symbol	Min	Тур	Max	Unit	Notes		
9. Measured from -200 mV to +200 mV on the differential waveform (derived from SD <i>n</i> _REF_CLK <i>n</i> minus SD <i>n</i> _REF_CLK <i>n</i> _B). The signal must be monotonic through the measurement region for rise and fall time. The 400 mV measurement window is centered on the differential zero crossing. See Figure 36.								
10. Measurement taken from single-ended waveform								
11. Matching applies to rising edge for SDn_REF_CLK <i>n</i> and falling edge rate for SDn_REF_CLK <i>n</i> _B. It is measured using a 200 mV window centered on the median cross point where SDn_REF_CLK <i>n</i> rising meets SDn_REF_CLK <i>n</i> _B falling. The median cross point is used to calculate the voltage thresholds that the oscilloscope uses for the edge rate calculations. The rise edge rate of SDn_REF_CLK <i>n</i> must be compared to the fall edge rate of SDn_REF_CLK <i>n</i> _B, the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 37.								
12. When 2 or more protocols share the same PLL on clock frequency tolerance must be followed.	a SerDes modu	le, the tig	htest SDn_REF_C	LKn/SDr	n_REF_C	LKn_B		

This table lists the AC requirements for SerDes reference clocks for protocols running at data rates greater than 8 Gb/s.

This includes XFI (10.3125 Gb/s) and 10GBase-KR (10.3125 GBd) SerDes reference clocks to be guaranteed by the customer's application design.

Table 63. SDn_REF_CLKn and $SDn_REF_CLKn_B$ input clock requirements ($SV_{DD}n = 1.0 V$)

Parameter	Symbol	Min	Тур	Max	Unit	Notes	
SDn_REF_CLKn/SDn_REF_CLKn_B frequency range	t _{CLK_REF}	-	156.25/ 161.1328135	-	MHz	2	
SDn_REF_CLKn/SDn_REF_CLKn_B clock frequency tolerance	t _{CLK_TOL}	-100	-	100	ppm	5	
SDn_REF_CLKn/SDn_REF_CLKn_B reference clock duty cycle	t _{CLK_DUTY}	40	50	60	%	3	
SDn_REF_CLKn/SDn_REF_CLKn_B single side band noise	@1 kHz	-	-	-85	dBC/Hz	4	
SDn_REF_CLKn/SDn_REF_CLKn_B single side band noise	@10 kHz	-	-	-108	dBC/Hz	4	
SDn_REF_CLKn/SDn_REF_CLKn_B single side band noise	@100 kH z	-	-	-128	dBC/Hz	4	
SDn_REF_CLKn/SDn_REF_CLKn_B single side band noise	@1 MHz	-	-	-138	dBC/Hz	4	
SDn_REF_CLKn/SDn_REF_CLKn_B single side band noise	@10 MHz	-	-	-138	dBC/Hz	4	
SDn_REF_CLKn/SDn_REF_CLKn_B random jitter (1.2 MHz to 15 MHz)	t _{CLK_RJ}	-	-	0.8	ps	-	
SDn_REF_CLKn/SDn_REF_CLKn_B total reference clock jitter at 10 ⁻¹² BER (1.2 MHz to 15 MHz)	t _{CLK_TJ}	-	-	11	ps	-	
SDn_REF_CLKn/SDn_REF_CLKn_B spurious noise (1.2 MHz to 15 MHz)	-	-	-	-75	dBC	-	

Table continues on the next page...

Table 63. SD n_REF_CLKn and SD $n_REF_CLKn_B$ input clock requirements (SV_{DD}n = 1.0 V) ¹ (continued)

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Differential input high voltage	-	V _{CM} +200 m V	-	-	mV	6
Differential input low voltage	-	-	-	V _{CM} -20 0 mV	mV	6
Rising edge rate (SDn_REF_CLKn) to falling edge rate (SDn_REF_CLKn) matching	Rise-Fall Matching	-	-	20	%	7, 8

1. For recommended operating conditions, see Table 3.

2. **Caution:** Only 156.25 and 161.1328135 have been tested. In-between values do not work correctly with the rest of the system.

3. Measurement taken from differential waveform.

4. Per XFP Spec. Rev 4.5, the Module Jitter Generation spec at XFI Optical Output is 10mUI (RMS) and 100 mUI (p-p). In the CDR mode the host is contributing 7 mUI (RMS) and 50 mUI (p-p) jitter.

5. When 2 or more protocols share the same PLL on a SerDes module, the tightest SDn_REF_CLKn/SDn_REF_CLKn_B clock frequency tolerance must be followed.

6. Measurement taken from differential waveform. VCM is the common mode voltage.

7. Measurement taken from single-ended waveform .

8. Matching applies to rising edge for SDn_REF_CLK*n* and falling edge rate for SDn_REF_CLK*n*_B. It is measured using a 200 mV window centered on the median cross point where SDn_REF_CLK*n* rising meets SDn_REF_CLK*n*_B falling. The median cross point is used to calculate the voltage thresholds that the oscilloscope uses for the edge rate calculations. The rise edge rate of SDn_REF_CLK*n* must be compared to the fall edge rate of SDn_REF_CLK*n*_B, the maximum allowed difference should not exceed 20% of the slowest edge rate. See Figure 37.

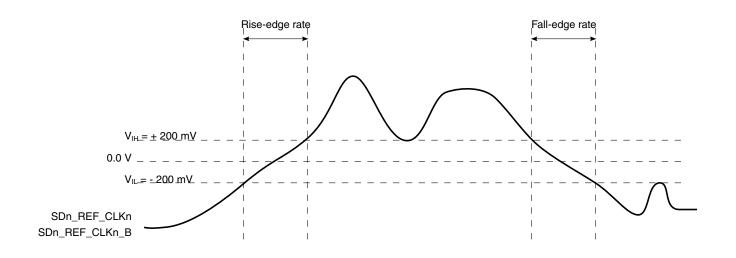


Figure 36. Differential measurement points for rise and fall time

Electrical characteristics

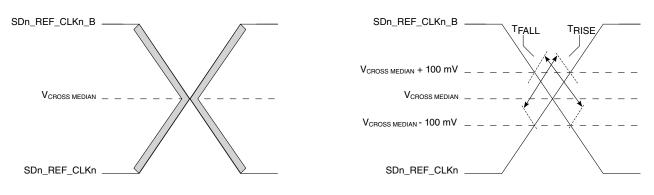


Figure 37. Single-ended measurement points for rise and fall time matching

3.19.3 SerDes transmitter and receiver reference circuits

This figure shows the reference circuits for SerDes data lane's transmitter and receiver.

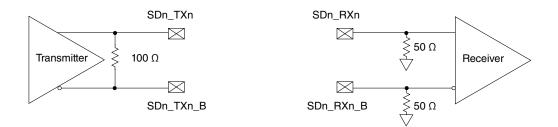


Figure 38. SerDes transmitter and receiver reference circuits

The DC and AC specification of SerDes data lanes are defined in each interface protocol section below based on the application usage:

- PCI Express
- Serial RapidIO (sRIO)
- XAUI interface
- Aurora interface
- Serial ATA (SATA) interface
- SGMII interface
- QSGMII interface
- HiGig/HiGig2 interface
- XFI interface
- Interlaken interface

Note that external AC-coupling capacitor is required for the above serial transmission protocols with the capacitor value defined in the specification of each protocol section.

3.19.4 PCI Express

This section describes the clocking dependencies, DC and AC electrical specifications for the PCI Express bus.

3.19.4.1 Clocking dependencies

The ports on the two ends of a link must transmit data at a rate that is within 600 parts per million (ppm) of each other at all times. This is specified to allow bit rate clock sources with a ± 300 ppm tolerance.

3.19.4.2 PCI Express clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

SerDes 3-4 (SD[3:4]_REF_CLK[1:2] and SD[3:4]_REF_CLK[1:2]_B) may be used for various SerDes PCI Express configurations based on the RCW Configuration field SRDS_PRTCL. PCI Express is not supported on SerDes 1 and 2.

NOTE

PCI Express operating in x8 mode is only supported at 2.5 and 5.0 GT/s.

For more information on these specifications, see SerDes reference clocks.

3.19.4.3 PCI Express DC physical layer specifications

This section contains the DC specifications for the physical layer of PCI Express on this chip.

3.19.4.3.1 PCI Express DC physical layer transmitter specifications

This section discusses the PCI Express DC physical layer transmitter specifications for 2.5 GT/s, 5 GT/s and 8 GT/s.

This table defines the PCI Express 2.0 (2.5 GT/s) DC specifications for the differential output at all transmitters. The parameters are specified at the component pins.

Table 64. PCI Express 2.0 (2.5 GT/s) differential transmitter output DC specifications $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^1$

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential peak-to-peak output voltage	V _{TX-DIFFp-p}	800	1000	1200	mV	$V_{TX-DIFFp-p} = 2 x V_{TX-D+} - V_{TX-D-} $

Table continues on the next page...

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Table 64. PCI Express 2.0 (2.5 GT/s) differential transmitter output DC specifications $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^1$ (continued)

Parameter	Symbol	Min	Typical	Max	Units	Notes	
De-emphasized differential output voltage (ratio)	V _{TX-DE-RATIO}	3.0	3.5	4.0	dB	Ratio of the $V_{TX-DIFFp-p}$ of the second and following bits after a transition divided by the $V_{TX-DIFFp-p}$ of the first bit after a transition.	
DC differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω	Transmitter DC differential mode low Impedance	
Transmitter DC impedance	Z _{TX-DC}	40	50	60	Ω	Required transmitter D+ as well as D- DC Impedance during all states	
Notes:							
1. For recommended operating conditions, see Table 3.							

This table defines the PCI Express 2.0 (5 GT/s) DC specifications for the differential output at all transmitters. The parameters are specified at the component pins.

Table 65. PCI Express 2.0 (5 GT/s) differential transmitter output DC specifications $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^1$

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential peak-to-peak output voltage	V _{TX-DIFFp-p}	800	1000	1200	mV	$V_{TX-DIFFp-p} = 2 x V_{TX-D+} - V_{TX-D-} $
Low power differential peak-to-peak output voltage	V _{TX-DIFFp-p_low}	400	500	1200	mV	$V_{TX-DIFFp-p} = 2 x V_{TX-D+} - V_{TX-D-} $
De-emphasized differential output voltage (ratio)	V _{TX-DE-RATIO-3.5dB}	3.0	3.5	4.0	dB	Ratio of the $V_{TX-DIFFp-p}$ of the second and following bits after a transition divided by the $V_{TX-DIFFp-p}$ of the first bit after a transition.
De-emphasized differential output voltage (ratio)	V _{TX-DE-RATIO-6.0dB}	5.5	6.0	6.5	dB	Ratio of the $V_{TX-DIFFp-p}$ of the second and following bits after a transition divided by the $V_{TX-DIFFp-p}$ of the first bit after a transition.
DC differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω	Transmitter DC differential mode low impedance
Transmitter DC Impedance	Z _{TX-DC}	40	50	60	Ω	Required transmitter D+ as well as D- DC impedance during all states
Impedance Notes: 1. For recommended opera			<u> </u>			Impedance during all states

This table defines the PCI Express 3.0 (8 GT/s) DC specifications for the differential output at all transmitters. The parameters are specified at the component pins.

Table 66. PCI Express 3.0 (8 GT/s) differential transmitter output DC specifications $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^3$

Parameter	Symbol	Min	Typical	Max	Units	Notes
Full swing transmitter voltage with no TX Eq	V _{TX-FS-NO-EQ}	800	-	1300	mVp-p	See Note 1.
Reduced swing transmitter voltage with no TX Eq	V _{TX-RS-NO-EQ}	400	-	1300	mV	See Note 1.
De-emphasized differential output voltage (ratio)	VTX-DE-RATIO-3.5dB	3.0	3.5	4.0	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-RATIO-6.0dB}	5.5	6.0	6.5	dB	-
Minimum swing during EIEOS for full swing	V _{TX-EIEOS-FS}	250	-	-	mVp-p	See Note 2
Minimum swing during EIEOS for reduced swing	V _{TX-EIEOS-RS}	232	-	-	mVp-p	See Note 2
DC differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω	Transmitter DC differential mode low impedance
Transmitter DC Impedance	Z _{TX-DC}	40	50	60	Ω	Required transmitter D+ as well as D- DC impedance during all states

Notes:

1. Voltage measurements for $V_{TX-FS-NO-EQ}$ and $V_{TX-RS-NO-EQ}$ are made using the 64-zeroes/64-ones pattern in the compliance pattern.

2. Voltage limits comprehend both full swing and reduced swing modes. The transmitter must reject any changes that would violate this specification. The maximum level is covered in the $V_{TX-FS-NO-EQ}$ measurement which represents the maximum peak voltage the transmitter can drive. The $V_{TX-EIEOS-FS}$ and $V_{TX-EIEOS-RS}$ voltage limits are imposed to guarantee the EIEOS threshold of 175 mV_{P-P} at the receiver pin. This parameter is measured using the actual EIEOS pattern that is part of the compliance pattern and then removing the ISI contribution of the breakout channel.

3. For recommended operating conditions, see Table 3.

3.19.4.4 PCI Express DC physical layer receiver specifications

This section discusses the PCI Express DC physical layer receiver specifications for 2.5 GT/s, 5 GT/s and 8 GT/s.

This table defines the DC specifications for the PCI Express 2.0 (2.5 GT/s) differential input at all receivers. The parameters are specified at the component pins.

Table 67. PCI Express 2.0 (2.5 GT/s) differential receiver input DC specifications (SV_{DD} = 1.0 V)⁴

Parameter	Symbol	Min	Тур	Max	Units	Notes
Differential input peak-to-peak voltage	V _{RX-DIFFp-p}	120	1000	1200		$V_{RX-DIFFp-p} = 2 \times V_{RX-D+} - V_{RX-D-} $ See Note 1.

Table continues on the next page...

Table 67. PCI Express 2.0 (2.5 GT/s) differential receiver input DC specifications (SV_{DD} = 1.0 V)⁴ (continued)

Receiver DC differential mode impedance. See Note 2 Required receiver D+ as well as D- DC Impedance (50 ± 20% tolerance). See Notes 1 and 2.
Impedance (50 ± 20% tolerance). See Notes 1 and 2.
Descriptional reserver Dynamical as D. DC.
Required receiver D+ as well as D- DC Impedance when the receiver terminations do not have power. See Note 3.
$V_{RX-IDLE-DET-DIFFp-p} = 2 \times V_{RX-D+} - V_{RX-D-}]$ Measured at the package pins of the

Notes:

1. Measured at the package pins with a test load of 50Ω to GND on each pin.

2. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.

3. The receiver DC common mode impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the receiver ground.

4. For recommended operating conditions, see Table 3.

This table defines the DC specifications for the PCI Express 2.0 (5 GT/s) differential input at all receivers. The parameters are specified at the component pins.

Table 68. PCI Express 2.0 (5 GT/s) differential receiver input DC specifications (SV_{DD} = 1.0 V)⁴

Parameter	Symbol	Min	Тур	Max	Units	Notes
Differential input peak-to-peak voltage	V _{RX-DIFFp-p}	120	1000	1200	mV	$V_{RX-DIFFp-p} = 2 \times V_{RX-D+} - V_{RX-D-} $ See Note 1.
DC differential input impedance	Z _{RX-DIFF-DC}	80	100	120	Ω	Receiver DC differential mode impedance. See Note 2
DC input impedance	Z _{RX-DC}	40	50	60	Ω	Required receiver D+ as well as D- DC Impedance ($50 \pm 20\%$ tolerance). See Notes 1 and 2.
Powered down DC input impedance	Z _{RX-HIGH-IMP-DC}	50	-	-	kΩ	Required receiver D+ as well as D- DC Impedance when the receiver terminations do not have power. See Note 3.
Electrical idle detect threshold	V _{RX-IDLE-DET-} DIFFp-p	65	-	175	mV	$V_{RX-IDLE-DET-DIFFp-p} = 2 \times V_{RX-D+} - V_{RX-D-} $ Measured at the package pins of the receiver

Table continues on the next page...

Table 68. PCI Express 2.0 (5 GT/s) differential receiver input DC specifications (SV_{DD} = 1.0 V)⁴ (continued)

Parameter	Symbol	Min	Тур	Max	Units	Notes		
Notes:								
1. Measured at the package pins with a test load of 50 Ω to GND on each pin.								
2. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.								
3. The receiver DC common mode impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the receiver ground.								
4. For recommended operating conditio	ns, see Table 3.							

This table defines the DC specifications for the PCI Express 3.0 (8 GT/s) differential input at all receivers. The parameters are specified at the component pins.

Table 69. PCI Express 3.0 (8 GT/s) differential receiver input DC specifications (SV_{DD} = 1.0 V)⁶

Parameter	Symbol	Min	Тур	Max	Units	Notes
DC differential input impedance	Z _{RX-DIFF-DC}	80	100	120	Ω	Receiver DC differential mode impedance. See Note 2
DC input impedance	Z _{RX-DC}	40	50	60	Ω	Required receiver D+ as well as D- DC Impedance $(50 \pm 20\%)$ tolerance). See Notes 1 and 2.
Powered down DC input impedance	Z _{RX-HIGH-IMP-DC}	50	-	-	kΩ	Required receiver D+ as well as D- DC Impedance when the receiver terminations do not have power. See Note 3.
Generator launch voltage	V _{RX-LAUNCH-8G}	-	800	-	mV	Measured at TP1 per PCI Express base spec. rev 3.0
Eye height (-20dB Channel)	V _{RX-SV-8G}	25	-	-	mV	Measured at TP2P per PCI Express base spec. rev 3.0. See Notes 4, 5
Eye height (-12dB Channel)	V _{RX-SV-8G}	50	-	-	mV	Measured at TP2P per PCI Express base spec. rev 3.0. See Notes 4, 5
Eye height (-3dB Channel)	V _{RX-SV-8G}	200	-	-	mV	Measured at TP2P per PCI Express base spec. rev 3.0. See Notes 4, 5
Electrical idle detect threshold	V _{RX-IDLE-DET-} DIFFp-p	65	-	175	mV	$V_{RX-IDLE-DET-DIFFp-p} = 2 \times IV_{RX-D+} - V_{RX-D-}$
						Measured at the package pins of the receiver

Notes:

1. Measured at the package pins with a test load of 50Ω to GND on each pin.

2. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM) there is a 5 ms transition time before receiver termination values must be met on all unconfigured lanes of a port.

3. The receiver DC common mode impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the receiver ground.

Table 69. PCI Express 3.0 (8 GT/s) differential receiver input DC specifications (SV_{DD} = 1.0 V)⁶

Parameter Symbol Min Typ Max Units Notes							
4. V _{RX-SV-8G} is tested at three different voltages to ensure the receiver device under test is capable of equalizing over a range of channel loss profiles. The "SV" in the parameter names refers to stressed voltage.							
5. V _{RX-SV-8G} is referenced to TP2P and is obtained after post processing data captured at TP2.							
6. For recommended operating conditions, see Table 3.							

3.19.4.5 PCI Express AC physical layer specifications

This section contains the AC specifications for the physical layer of PCI Express on this device.

3.19.4.5.1 PCI Express AC physical layer transmitter specifications

This section discusses the PCI Express AC physical layer transmitter specifications for 2.5 GT/s, 5 GT/s and 8 GT/s.

This table defines the PCI Express 2.0 (2.5 GT/s) AC specifications for the differential output at all transmitters. The parameters are specified at the component pins. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Тур	Max	Units	Notes
Unit interval	UI	399.88	400	400.12	ps	Each UI is 400 ps \pm 300 ppm. UI does not account for spread-spectrum clock dictated variations.
Minimum transmitter eye width	T _{TX-EYE}	0.75	-	-	UI	The maximum transmitter jitter can be derived as $T_{TX-MAX-JITTER} = 1 - T_{TX-EYE} =$ 0.25 UI. Does not include spread-spectrum or RefCLK jitter. Includes device random jitter at 10 ⁻¹² . See Notes 1 and 2.
Maximum time between the jitter median and maximum deviation from the median	T _{TX-EYE-MEDIAN-} to- MAX-JITTER	-	-	0.125	UI	Jitter is defined as the measurement variation of the crossing points ($V_{TX-DIFFp-p} =$ 0 V) in relation to a recovered transmitter UI. A recovered transmitter UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the transmitter UI. See Notes 1 and 2.
AC coupling capacitor	C _{TX}	75	-	200	nF	All transmitters must be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See Note 3.

Table 70. PCI Express 2.0 (2.5 GT/s) differential transmitter output AC specifications⁴

Table continues on the next page ...

Table 70. PCI Express 2.0 (2.5 GT/s) differential transmitter output AC specifications⁴ (continued)

Parameter	Symbol	Min	Тур	Max	Units	Notes
Notes:		•				•
1. Specified at the measuren consecutive transmitter UIs.	nent point into a ti	ming and	voltage	test load	d as show	wn in Figure 40 and measured over any 250
transmitter collected over any total transmitter jitter budget	y 250 consecutive collected over any median describes	transmit 250 con the poin	ter UIs. secutive It in time	The T _{TX-} e transmi	EYE-MEDI/	budget of $T_{TX-JITTER-MAX} = 0.25$ UI for the AN-to-MAX-JITTER median is less than half of the It must be noted that the median is not the per of jitter points on either side is
3. The chip's SerDes transm	itter does not have	e C _{TX} buil	t-in. An	external	AC coup	bling capacitor is required.
4. For recommended operati	ng conditions, see	Table 3.				

This table defines the PCI Express 2.0 (5 GT/s) AC specifications for the differential output at all transmitters. The parameters are specified at the component pins. The AC timing specifications do not include RefClk jitter.

Table 71.	PCI Express 2.0	(5 GT/s) differentia	al transmitter outp	out AC specifications ³
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UI	199.94	200.00	200.06	ps	Each UI is 200 ps ± 300 ppm. UI does not
-					account for spread-spectrum clock dictated variations.
T _{TX-EYE}	0.75	-	-	UI	The maximum transmitter jitter can be derived as: $T_{TX-MAX-JITTER} = 1 - T_{TX-EYE} = 0.25 UI.$ See Note 1.
T _{TX-HF-DJ-DD}	-	-	0.15	ps	-
C _{TX}	75	-	200	nF	All transmitters must be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See Note 2.
					Г _{ТХ-НF-DJ-DD} 0.15 ps С _{ТХ} 75 - 200 nF

Notes:

1. Specified at the measurement point into a timing and voltage test load as shown in Figure 40 and measured over any 250 consecutive transmitter UIs.

2. The chip's SerDes transmitter does not have C_{TX} built-in. An external AC coupling capacitor is required.

3. For recommended operating conditions, see Table 3.

This table defines the PCI Express 3.0 (8 GT/s) AC specifications for the differential output at all transmitters. The parameters are specified at the component pins. The AC timing specifications do not include RefClk jitter.

124 	4.9625	-	125.0375 31.25 12	ps ps p-p ps p-p	Each UI is 125 ps ± 300 ppm. UI does not account for spread-spectrum clock dictated variations. -
- DJ-DD -		-			-
		-	12	ps p-p	-
W-TJ -					
		-	24	ps p-p	See Note 1, 2
W-DJDD -		-	10	ps p-p	See Note 1, 2
ы -		-	18	ps p-p	See Note 2
176	6	-	265	nF	All transmitters must be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See Note 3.
			1		
		- 176	176 -	176 - 18 265	18 ps p-p

2. Measured with optimized preset value after de-embedding to transmitter pin.

3. The chip's SerDes transmitter does not have C_{TX} built-in. An external AC coupling capacitor is required.

4. For recommended operating conditions, see Table 3.

3.19.4.5.2 PCI Express AC physical layer receiver specifications

This section discusses the PCI Express AC physical layer receiver specifications for 2.5 GT/s, 5 GT/s and 8 GT/s.

This table defines the AC specifications for the PCI Express 2.0 (2.5 GT/s) differential input at all receivers. The parameters are specified at the component pins. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Тур	Max	Units	Notes
Unit Interval	UI	399.88	400.00	400.12	ps	Each UI is 400 ps \pm 300 ppm. UI does not account for spread-spectrum clock dictated variations.
Minimum receiver eye width	T _{RX-EYE}	0.4	-	-		The maximum interconnect media and transmitter jitter that can be tolerated by the receiver can be derived as $T_{RX-MAX-}$ JITTER = 1 - T_{RX-EYE} = 0.6 UI. See Notes 1 and 2.

Table 73. PCI Express 2.0 (2.5 GT/s) differential receiver input AC specifications⁴

Table continues on the next page ...

Table 73. PCI Express 2.0 (2.5 GT/s) differential receiver input AC specifications⁴ (continued)

Parameter	Symbol	Min	Тур	Max	Units	Notes
Maximum time between the jitter median and maximum deviation from the median.	T _{RX-EYE-MEDIAN-} to-MAX-JITTER	-	-	0.3	UI	Jitter is defined as the measurement variation of the crossing points ($V_{RX-DIFFp-p}$ = 0 V) in relation to a recovered transmitter UI. A recovered transmitter UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the transmitter UI. See this table notes.

Notes:

1. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 40 must be used as the receiver device when taking measurements. If the clocks to the receiver and transmitter are not derived from the same reference clock, the transmitter UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.

2. A T_{RX-EYE} = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The TRX-EYE-MEDIAN-to-MAX-JITTER specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive transmitter UIs. It must be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the receiver and transmitter are not derived from the same reference clock, the transmitter UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.

3. It is recommended that the recovered transmitter UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

4. For recommended operating conditions, see Table 3.

5. The TRX-EYE-MEDIAN-to-MAX-JITTER for common and separated reference clock architecture.

6. If spread spectrum clocking is desired, common clock receiver architecture must be used.

7. The AC specifications do not include Refclk jitter.

This table defines the AC specifications for the PCI Express 2.0 (5 GT/s) differential input at all receivers. The parameters are specified at the component pins. The AC timing specifications do not include RefClk jitter.

Table 74. PCI Express 2.0 (5 GT/s) differential receiver input AC specifications¹

Parameter	Symbol	Min	Тур	Max	Units	Notes
Unit Interval	UI	199.40	200.00	200.06	ps	1, 2
Max receiver inherent timing error	T _{RX-TJ-CC}	-	-	0.4	UI	3, 5, 6
Max receiver inherent deterministic timing error	T _{RX-DJ-DD-CC}	-	-	0.30	UI	4, 5, 6

Note:

1. Each UI is 200 ps ± 300 ppm. UI does not account for spread-spectrum clock dictated variations.

2. For recommended operating conditions, see Table 3.

3. The maximum inherent total timing error for common and separated RefClk receiver architecture.

Table 74. PCI Express 2.0 (5 GT/s) differential receiver input AC specifications¹

Parameter	Symbol	Min	Тур	Max	Units	Notes	
4. The maximum inherent deterministic tim	ing error for common an	d separated F	RefClk receive	er architecture	e.		
5. If spread spectrum clocking is desired, common clock must be used.							
6. The AC specifications do not include Re	fclk jitter.						

This table defines the AC specifications for the PCI Express 3.0 (8 GT/s) differential input at all receivers. The parameters are specified at the component pins. The AC timing specifications do not include RefClk jitter.

Table 75. PCI Express 3.0 (8 GT/s) differential receiver input AC specifications⁵

Parameter	Symbol	Min	Тур	Мах	Units	Notes
Unit Interval	UI	124.9625	125.00	125.0375	ps	Each UI is 125 ps \pm 300 ppm. UI does not account for spread- spectrum clock dictated variations. See Note 1.
Eye Width at TP2P	T _{RX-SV-8G}	0.3	-	0.35	UI	See Note 1
Differential mode interference	V _{RX-SV-DIFF-8G}	14	-	-	mV	Frequency = 2.1GHz. See Note 2.
Sinusoidal Jitter at 100 MHz	T _{RX-SV-SJ-8G}	-	-	0.1	UI p-p	Fixed at 100 MHz. See Note 3.
Random Jitter	T _{RX-SV-RJ-8G}	-	-	2.0	ps RMS	Random jitter spectrally flat before filtering. See Note 4.

Note:

1. T_{RX-SV-8G} is referenced to TP2P and obtained after post processing data captured at TP2. T_{RX-SV-8G} includes the effects of applying the behavioral receiver model and receiver behavioral equalization.

2. V_{RX-SV-DIFF-8G} voltage may need to be adjusted over a wide range for the different loss calibration channels.

3. The sinusoidal jitter in the total jitter tolerance may have any amplitude and frequency as shown in Figure 39.

4. Random jitter (Rj) is applied over the following range: The low frequency limit may be between 1.5 and 10 MHz, and the upper limit is 1.0 GHz. See Figure 39 for details. Rj may be adjusted to meet the 0.3 UI value for $T_{RX-SV-8G}$.

5. For recommended operating conditions, see Table 3.

Electrical characteristics

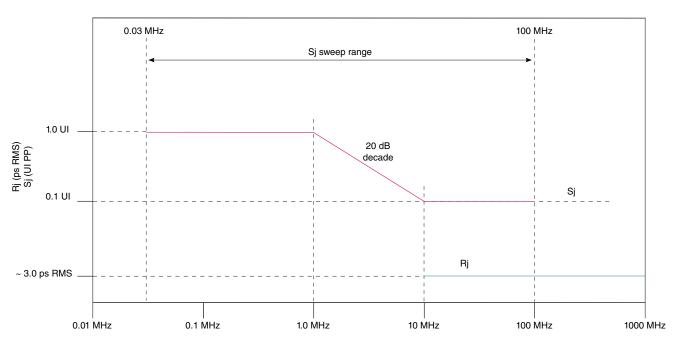


Figure 39. Swept sinusoidal jitter mask

3.19.4.6 Test and measurement load

The AC timing and voltage parameters must be verified at the measurement point. The package pins of the device must be connected to the test/measurement load within 0.2 inches of that load, as shown in the following figure.

NOTE

The allowance of the measurement point to be within 0.2 inches of the package pins is meant to acknowledge that package/ board routing may benefit from D+ and D- not being exactly matched in length at the package pin boundary. If the vendor does not explicitly state where the measurement point is located, the measurement point is assumed to be the D+ and Dpackage pins.

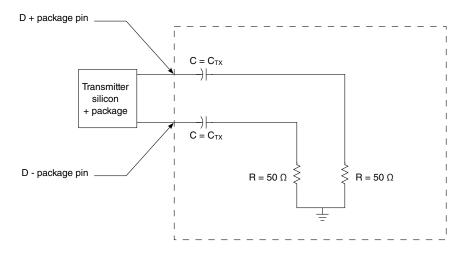


Figure 40. Test/measurement load

3.19.5 Serial RapidIO (sRIO)

This section describes the DC and AC electrical specifications for the serial RapidIO interface of the LP-Serial physical layer. The electrical specifications cover both single and multiple-lane links. Two transmitters (short run and long run) and a single receiver are specified for each of three baud rates: 2.50, 3.125 and 5 GBaud.

Two transmitter specifications allow for solutions ranging from simple board-to-board interconnect to driving two connectors across a backplane. A single receiver specification is given that accepts signals from both the short run and long run transmitter specifications.

The short run transmitter must be used mainly for chip-to-chip connections on either the same printed circuit board or across a single connector. This covers the case where connections are made to a mezzanine (daughter) card. The minimum swings of the short run specification reduce the overall power used by the transceivers.

The long run transmitter specifications use larger voltage swings that are capable of driving signals across backplanes. This allows a user to drive signals across two connectors and a backplane.

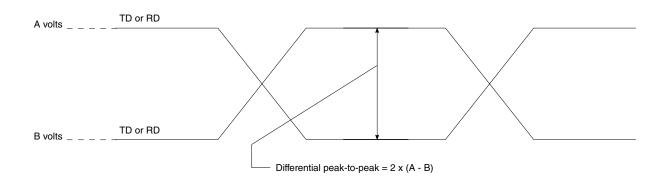
All unit intervals are specified with a tolerance of ± 100 ppm. The worst case frequency difference between any transmit and receive clock is 200 ppm.

To ensure interoperability between drivers and receivers of different vendors and technologies, AC coupling at the receiver input must be used.

3.19.5.1 Signal definitions

This section defines the terms used in the description and specification of the differential signals used by the LP-Serial links. The following figure shows how the signals are defined. The figures show waveforms for either a transmitter output (TD and TD_B) or a receiver input (RD and RD_B). Each signal swings between A volts and B volts where A > B. Using these waveforms, the definitions are as follows:

- The transmitter output signals and the receiver input signals-TD, TD_B, RD, and RD_B-each have a peak-to-peak swing of A B volts.
- The differential output signal of the transmitter, V_{OD} , is defined as V_{TD} V_{TD_B}
- The differential input signal of the receiver, V_{ID} , is defined as V_{RD} V_{RD_B}
- The differential output signal of the transmitter and the differential input signal of the receiver each range from A B to -(A B) volts
- The peak value of the differential transmitter output signal and the differential receiver input signal is A B volts.
- The peak-to-peak value of the differential transmitter output signal and the differential receiver input signal is 2 x (A B) volts.





To illustrate these definitions using real values, consider the case of a CML (current mode logic) transmitter that has a common mode voltage of 2.25 V, and each of its outputs TD and TD_B, has a swing that goes between 2.5 V and 2.0 V. Using these values, the peak-to-peak voltage swing of the signals TD and TD_B is 500 mV p-p. The differential output signal ranges between 500 mV and -500 mV. The peak differential voltage is 500 mV. The peak-to-peak differential voltage is 1000 mV p-p.

3.19.5.2 Equalization

With the use of high-speed serial links, the interconnect media causes degradation of the signal at the receiver and produces effects such as inter-symbol interference (ISI) or datadependent jitter. This loss can be large enough to degrade the eye opening at the receiver

Electrical characteristics

beyond what is allowed in the specification. To negate a portion of these effects, equalization can be used. The most common equalization techniques that can be used are as follows:

- Pre-emphasis on the transmitter
- A passive high-pass filter network placed at the receiver, often referred to as passive equalization.
- The use of active circuits in the receiver, often referred to as adaptive equalization.

3.19.5.3 Serial RapidIO clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

SerDes 3 and SerDes 4 (SD[3:4]_REF_CLK[1:2] and SD[3:4]_REF_CLK[1:2]_B) may be used for various SerDes serial RapidIO configurations based on the RCW Configuration field SRDS_PRTCL. Serial RapidIO is not supported on SerDes 1 and 2. The ref clock frequency tolerance spec is ±100ppm.

For more information on these specifications, see SerDes reference clocks.

3.19.5.4 DC requirements for serial RapidIO

This section explains the DC requirements for the serial RapidIO interface.

3.19.5.4.1 DC serial RapidIO transmitter specifications

This table defines the transmitter DC specifications for serial RapidIO operating at 2.5 and 3.125 GBaud.

Table 76.	Serial RapidIO transmitter	DC specifications-2.5 GBaud, 3.125 GBaud ²
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Parameter	Symbol	Min	Тур	Max	Unit	Notes
Long-run differential output voltage	V _{DIFFPP}	800	-	1600	mV p-p	-
Short-run differential output voltage	V _{DIFFPP}	500	-	1000	mV p-p	-
DC Differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω	Transmitter DC differential impedance
Notes:						
1. Voltage relative to COMMON of either signal comprising a differential pair						
2. For recommended operating conditi	ons, see <mark>Ta</mark> t	ole 3.				

This table defines the transmitter DC specifications for serial RapidIO operating at 5 GBaud.

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Long-run differential output voltage	V _{DIFF}	800	-	1200	mV	-
Short-run differential output voltage	V _{DIFF}	400	-	750	mV	-
Long-run de-emphasized differential output voltage (ratio)	V _{TX-DE-RATIO-3.5dB}	3	3.5	4	dB	-
Long-run de-emphasized differential output voltage (ratio)	V _{TX-DE-RATIO-6.0dB}	5.5	6.0	6.5	dB	-
Differential resistance	T _{RD}	80	100	120	Ω	-
Notes:				•	•	
1. For recommended operating conditions, see Table 3.						

Table 77. Serial RapidIO transmitter DC specifications-5 GBaud¹

3.19.5.4.2 DC serial RapidIO receiver specifications

LP-Serial receiver electrical and timing specifications are stated in the text and tables of this section.

Receiver input impedance results in a differential return loss better than 10 dB and a common mode return loss better than 6 dB from 100 MHz to (0.8) x (Baud Frequency). This includes contributions from on-chip circuitry, the chip package, and any off-chip components related to the receiver. AC coupling components are included in this requirement. The reference impedance for return loss measurements is $100-\Omega$ resistive for differential return loss and $25-\Omega$ resistive for common mode.

This table defines the receiver DC specifications for serial RapidIO operating at 2.5 and 3.125 GBaud.

 Table 78.
 Serial RapidIO receiver DC specifications-2.5 GBaud, 3.125 GBaud²

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Differential input voltage	V _{IN}	200	-	1600	mV p-p	1
DC differential receiver input impedance	Z _{RX-DIFF-DC}	80	100	120	Ω	Receiver DC differential impedance
Notes:		•				•
1. Measured at the receiver						
2. For recommended operating conditions, s	see Table 3.					

This table defines the receiver DC specifications for serial RapidIO operating at 5 GBaud.

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Long-run differential input voltage	V _{DIFF}	-	-	1200	mV	1
Short-run differential input voltage	V _{DIFF}	125	-	1200	mV	1
Differential resistance	R _{RD}	80	-	120	Ω	-
Notes:						
1. Measured at the receiver.						
2. For recommended operating conditions,	see Table 3.					

 Table 79.
 Serial RapidIO receiver DC specifications-5 GBaud²

3.19.5.5 AC requirements for serial RapidIO

This section explains the AC requirements for the serial RapidIO interface.

3.19.5.5.1 AC requirements for serial RapidIO transmitter

This table defines the transmitter AC specifications for the serial RapidIO operating at 2.5 and 3.125 GBaud. The AC timing specifications do not include RefClk jitter.

Table 80. Serial RapidIO transmitter, 2.5 GBaud and 3.125 GBaud, AC timing specifications¹

Parameter	Symbol	Min	Typical	Max	Unit		
Deterministic jitter	J _D	-	-	0.17	UI p-p		
Total jitter	J _T	-	-	0.35	UI p-p		
Unit Interval: 2.5 GBaud	UI	400 - 100ppm	400	400 + 100ppm	ps		
Unit Interval: 3.125 GBaud	UI	320 - 100ppm	320	320 + 100ppm	ps		
Notes:	•	•	•	•			
1. For recommended operating conditions, see Table 3.							

This table defines the transmitter AC specifications for the serial RapidIO operating at 5 GBaud, short range. The AC timing specifications do not include RefClk jitter.

 Table 81. Serial RapidIO transmitter, 5 GBaud, AC timing specifications¹

Parameter	Symbol	Min	Typical	Max	Unit		
Baud rate	T _{BAUD}	5.000 - 100ppm	5.000	5.000 + 100ppm	GBaud		
Uncorrelated high probability jitter	T _{UHPJ}	-	-	0.155	UI p-p		
Total jitter	TJ	-	-	0.30	UI p-p		
Notes:							
1. For recommended operating conditions, see	Table 3.						

This table defines the receiver AC specifications for serial RapidIO operating at 2.5 and 3.125 GBaud. The AC timing specifications do not include RefClk jitter.

Table 82. Serial RapidIO receiver, 2.5 GBaud and 3.125 GBaud, AC timing specifications³

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Deterministic jitter tolerance	J _D	-	-	0.37	UI p-p	1
Combined deterministic and random jitter tolerance	J _{DR}	-	-	0.55	UI p-p	1
Total jitter tolerance ²	J _T	-	-	0.65	UI p-p	1
Bit error rate	BER	-	-	10 ⁻¹²	-	-
Unit Interval: 2.5 GBaud	UI	400 - 100ppm	400	400 + 100ppm	ps	-
Unit Interval: 3.125 GBaud	UI	320 - 100ppm	320	320 + 100ppm	ps	-
Notes:	-	-		•	,	1

1. Measured at receiver

2. Total jitter is composed of three components: deterministic jitter, random jitter and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of Figure 42. The sinusoidal jitter component is included to ensure margin for low-frequency jitter, wander, noise, crosstalk, and other variable system effects.

3. For recommended operating conditions, see Table 3.

This figure shows the single-frequency sinusoidal jitter limits for 2.5 GBaud and 3.125 GBaud rates.



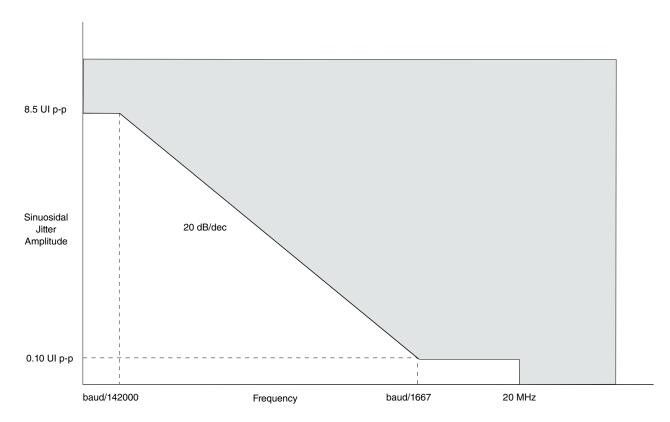


Figure 42. Single-frequency sinusoidal jitter limits, substitute the baud parameter in this figure by either 2.5G or 3.125G.

This table defines the receiver AC specifications for serial RapidIO operating at 5 GBaud. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Receiver baud rate	R _{BAUD}	5.000 - 100ppm	5.000	5.000 + 100ppm	Gbaud	-
Long-run Gaussian jitter	R _{GJ}	-	-	0.2	UI p-p	2
Long-Run Uncorrelated bounded high probability jitter	R _{UHPJ}	-	-	0.12	UI p-p	2, 3
Long-run correlated bounded high probability jitter	R _{CBHPJ}	-	-	0.63	UI p-p	2, 4
Short-run correlated bounded high probability jitter	R _{CBHPJ}	-	-	0.30	UI p-p	2, 4
Long-run bounded high probability jitter	R _{BHPJ}	-	-	0.75	UI p-p	3, 4
Short-run bounded high probability jitter	R _{BHPJ}	-	-	0.45	UI p-p	3, 4
Sinusoidal jitter, maximum	R _{SJ-max}	-	-	5.00	UI p-p	-
Sinusoidal jitter, high frequency	R _{SJ-hf}	-	-	0.05	UI p-p	-
Long-run total jitter (does not include sinusoidal jitter)	R _{Tj}	-	-	0.95	UI p-p	3, 4
Short-run total jitter (does not include sinusoidal jitter)	R _{Tj}	-	-	0.60	UI p-p	3, 4

 Table 83.
 Serial RapidIO receiver, 5G Baud, AC timing specifications¹

Table continues on the next page...

Table 83. Serial RapidIO receiver, 5G Baud, AC timing specifications¹ (continued)

Parameter	Symbol Min Typical Max Unit No								
Notes:									
1. For recommended operating conditions, see Table 3.									
2. The AC specifications do not include Refclk jitter.									
 The jitter (R_{UHPJ}) is Bounded High Probability Jitter and is commonly caused by crosstalk coupling and can have periodic and bounded PRBS jitter subcomponents. 									
4. The jitter (R_{CBHPJ}) and amplitude have to be	e correlated	, for example by a	PCB trace.						

This figure shows the single-frequency sinusoidal jitter limits for 5 GBaud rate.

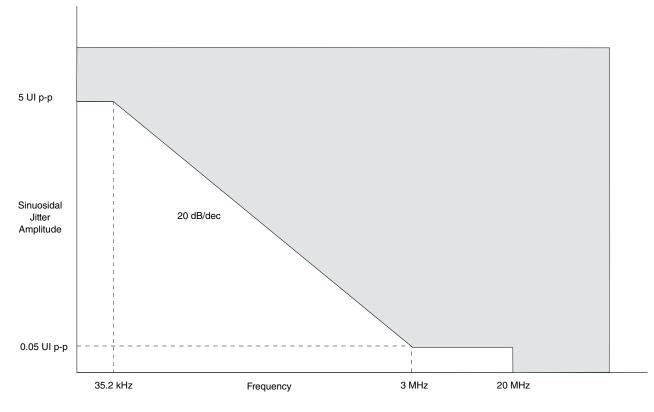


Figure 43. Single-frequency sinusoidal jitter limits

3.19.6 XAUI interface

This section describes the DC and AC electrical specifications for the XAUI bus.

3.19.6.1 XAUI DC electrical characteristics

This section discusses the XAUI DC electrical characteristics for the clocking signals, transmitter, and receiver.

3.19.6.1.1 DC requirements for XAUI SDn_REF_CLKn and SDn_REF_CLKn_B

Only SerDes 1-2 (SD[1:2]_REF_CLK[1:2] and SD[1:2]_REF_CLK[1:2]_B) may be used for various SerDes XAUI configurations based on the RCW Configuration field SRDS_PRTCL. The ref clock frequency tolerance spec is ±100ppm.

For more information on these specifications, see SerDes reference clocks.

3.19.6.1.2 XAUI transmitter DC electrical characteristics

This table defines the XAUI transmitter DC electrical characteristics.

Table 84. XAUI transmitter DC electrical characteristics $(XV_{DD} = 1.35V \text{ or } 1.5V)^1$

Parameter	Symbol	Min	Typical	Max	Unit	Notes	
Differential output voltage	V _{DIFFPP}	800	1000	1600	mV p-p	-	
DC Differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω	3	
1. For recommended operating conditio	ns, see Table	3.		•			
2. Absolute output voltage limit	2. Absolute output voltage limit						
3. Transmitter DC differential impedance							

3.19.6.1.3 XAUI receiver DC electrical characteristics

This table defines the XAUI receiver DC electrical characteristics.

Table 85. XAUI receiver DC timing specifications $(SV_{DD} = 1.0 V)^1$

Parameter	Symbol	Min	Typical	Max	Unit	Notes	
Differential input voltage	V _{IN}	200	-	1600	mV p-p	2	
DC Differential receiver input impedance	Z _{RX-DIFF-DC}	80	100	120	Ω	3	
1. For recommended operating condition	ons, see Table	3.	1	1		1	
2. Measured at the receiver.							
3. Receiver DC differential impedance	3. Receiver DC differential impedance						

3.19.6.2 XAUI AC timing specifications

This section explains the AC requirements for the XAUI interface.

3.19.6.2.1 XAUI transmitter AC timing specifications

This table defines the XAUI transmitter AC timing specifications. RefClk jitter is not included.

Parameter	Symbol	Min	Typical	Max	Unit		
Deterministic jitter	J _D	-	-	0.17	UI p-p		
Total jitter	J _T	-	-	0.35	UI p-p		
Unit Interval: 3.125 Gb/s	UI	320 - 100 ppm	320	320 + 100 ppm	ps		
1. For recommended operating conditions, see Table 3.							

 Table 86.
 XAUI transmitter AC timing specifications ¹

3.19.6.2.2 XAUI receiver AC timing specifications

This table defines the receiver AC specifications for XAUI. RefClk jitter is not included.

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Deterministic jitter tolerance	J _D	-	-	0.37	UI p-p	1
Combined deterministic and random jitter tolerance	J _{DR}	-	-	0.55	UI p-p	1
Total jitter tolerance	J _T	-	-	0.65	UI р-р	1, 2
Bit error rate	BER	-	-	10 ⁻¹²	-	-
Unit Interval: 3.125 Gb/s	UI	320 - 100 ppm	320	320 + 100 ppm	ps	-
Nataa		•		•		•

Table 87. XAUI receiver AC timing specifications³

Notes:

1. Measured at receiver.

2. Total jitter is composed of three components: deterministic jitter, random jitter, and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of Figure 43. The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk, and other variable system effects.

3. For recommended operating conditions, see Table 3.

3.19.7 Aurora interface

This section describes the Aurora clocking requirements and its DC and AC electrical characteristics.

3.19.7.1 Aurora clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

Only SerDes 4 (SD4_REF_CLK[1:2] and SD4_REF_CLK[1:2]_B) may be used for SerDes Aurora configurations based on the RCW Configuration field SRDS_PRTCL. The ref clock frequency tolerance spec is ±100ppm. Aurora is not supported on SerDes 1-3.

For more information on these specifications, see SerDes reference clocks.

3.19.7.2 Aurora DC electrical characteristics

This section describes the DC electrical characteristics for the Aurora interface.

3.19.7.2.1 Aurora transmitter DC electrical characteristics

This table defines the Aurora transmitter DC electrical characteristics.

Table 88. Aurora transmitter DC electrical characteristics ($XV_{DD} = 1.35$ V or 1.5 V)¹

Parameter	Symbol	Min	Typical	Max	Unit		
Differential output voltage	V _{DIFFPP}	800	1000	1600	mV p-p		
DC Differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω		
1. For recommended operating conditions, see Table 3.							

3.19.7.2.2 Aurora receiver DC electrical characteristics

This table defines the Aurora receiver DC electrical characteristics for the Aurora interface.

Table 89.	Aurora receiver D	C electrical	characteristics	$(SV_{DD} = 1.0V)^{1}$
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Parameter	Symbol	Min	Typical	Max	Unit	Notes	
Differential input voltage	V _{IN}	200	-	1600	mV p-p	2	
DC Differential receiver impedance	Z _{RX-DIFF-DC}	80	100	120	Ω	3	
Notes:							
1. For recommended operating condition	ons, see Table	93.					
2. Measured at receiver							
3. DC Differential receiver impedance							

3.19.7.3 Aurora AC timing specifications

This section describes the AC timing specifications for Aurora.

3.19.7.3.1 Aurora transmitter AC timing specifications

This table defines the Aurora transmitter AC timing specifications. RefClk jitter is not included.

Parameter	Symbol	Min	Typical	Max	Unit
Deterministic jitter	J _D	-	-	0.17	UI p-p
Total jitter	J _T	-	-	0.35	UI p-p
Unit interval: 2.5 Gbps	UI	400 - 100 ppm	400	400 + 100 ppm	ps
Unit interval: 3.125 Gbps	UI	320 - 100 ppm	320	320 + 100 ppm	ps
Unit interval: 5.0 Gbps	UI	200 - 100 ppm	200	200 + 100 ppm	ps
Notes:		- I		-	
1. For recommended operating condition	s, see Table 3.				

Table 90. Aurora transmitter AC timing specifications¹

3.19.7.3.2 Aurora receiver AC timing specifications

This table defines the Aurora receiver AC timing specifications. RefClk jitter is not included.

 Table 91. Aurora receiver AC timing specifications³

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Deterministic jitter tolerance	J _D	-	-	0.37	UI р-р	1
Combined deterministic and random jitter tolerance	J _{DR}	-	-	0.55	UI p-p	1
Total jitter tolerance	J _T	-	-	0.65	UI p-p	1, 2
Bit error rate	BER	-	-	10 ⁻¹²	-	-
Unit Interval: 2.5 Gbps	UI	400 - 100 ppm	400	400 + 100 ppm	ps	-
Unit Interval: 3.125 Gbps	UI	320 - 100 ppm	320	320 + 100 ppm	ps	-
Unit Interval: 5.0 Gbps	UI	200 - 100 ppm	200	200 + 100 ppm	ps	-

Notes:

1. Measured at receiver

2. Total jitter is composed of three components: deterministic jitter, random jitter, and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of Figure 42. The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk and other variable system effects.

3. For recommended operating conditions, see Table 3.

3.19.8 Serial ATA (SATA) interface

This section describes the DC and AC electrical specifications for the serial ATA (SATA) interface.

3.19.8.1 SATA DC electrical characteristics

This section describes the DC electrical characteristics for SATA.

3.19.8.1.1 SATA DC transmitter output characteristics

This table provides the differential transmitter output DC characteristics for the SATA interface at Gen1i/1m or 1.5 Gbps transmission.

Table 92. Gen1i/1m 1.5G transmitter DC specifications $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^3$

Parameter	Symbol	Min	Тур	Max	Units	Notes		
Tx differential output voltage	V _{SATA_TXDIFF}	400	500	600	mV p-p	1		
Tx differential pair impedance	Z _{SATA_TXDIFFIM}	85	100	115	Ω	2		
Notes:			•		-			
1. Terminated by 50 Ω load								
2. DC impedance								
3. For recommended operating conditions, see Table 3.								

This table provides the differential transmitter output DC characteristics for the SATA interface at Gen2i/2m or 3.0 Gbps transmission.

Table 93. Gen 2i/2m 3G transmitter DC specifications $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^2$

Parameter	Symbol	Min	Тур	Max	Units	Notes
Transmitter differential output voltage	V _{SATA_TXDIFF}	400	-	700	mV p-p	1
Transmitter differential pair impedance	Z _{SATA_TXDIFFIM}	85	100	115	Ω	-
Notes:	•	•				
1. Terminated by 50 Ω load.						
2. For recommended operating conditions	s, see Table 3.					

3.19.8.1.2 SATA DC receiver input characteristics

This table provides the Gen1i/1m or 1.5 Gbps differential receiver input DC characteristics for the SATA interface.

Parameter	Symbol	Min	Typical	Max	Units	Notes			
Differential input voltage	V _{SATA_RXDIFF}	240	500	600	mV p-p	1			
Differential receiver input impedance	Z _{SATA_RXSEIM}	85	100	115	Ω	2			
OOB signal detection threshold	V _{SATA_OOB}	50	120	240	mV p-p	-			
Notes:									
1. Voltage relative to common of eithe	r signal comprisin	g a differential pa	air						
2. DC impedance									
3. For recommended operating condit	3. For recommended operating conditions, see Table 3.								

Table 94. Gen1i/1m 1.5 G receiver input DC specifications $(SV_{DD} = 1.0 V)^3$

This table provides the Gen2i/2m or 3 Gbps differential receiver input DC characteristics for the SATA interface.

Table 95. Gen2i/2m 3 G receiver input DC specifications $(SV_{DD} = 1.0 V)^3$

Parameter	Symbol	Min	Typical	Max	Units	Notes			
Differential input voltage	V _{SATA_RXDIFF}	240	-	750	mV p-p	1			
Differential receiver input impedance	Z _{SATA_RXSEIM}	85	100	115	Ω	2			
OOB signal detection threshold	V _{SATA_OOB}	75	120	240	mV p-p	2			
Notes:									
1. Voltage relative to common of either	signal comprising	a differential	pair						
2. DC impedance									
3. For recommended operating conditions, see Table 3.									

3.19.8.2 SATA AC timing specifications

This section discusses the SATA AC timing specifications.

3.19.8.2.1 AC requirements for SATA REF_CLK

The AC requirements for the SATA reference clock listed in this table are to be guaranteed by the customer's application design. SATA does not support TX Spread Spectrum Clock as it is an optional requirement in protocol. However T4 SATA supports RX spread spectrum data as it is required in the SATA standard that all SATA Receivers handle spread spectrum. SerDes can receive spread spectrum without affecting other protocols since this doesn't affect SerDes PLL.

Electrical characteristics

Parameter	Symbol	Min	Тур	Max	Unit	Notes
SDn_REF_CLKn/SDn_REF_CLKn_B frequency range	tCLK_REF	-	100/125	-	MHz	1
SDn_REF_CLKn/SDn_REF_CLKn_B clock frequency tolerance	t _{CLK_TOL}	-350	-	+350	ppm	-
SDn_REF_CLKn/SDn_REF_CLKn_B reference clock duty cycle	tCLK_DUTY	40	50	60	%	5
SDn_REF_CLKn/SDn_REF_CLKn_B cycle-to- cycle clock jitter (period jitter)	t _{CLK_CJ}	-	-	100	ps	2
SDn_REF_CLKn/SDn_REF_CLKn_B total reference clock jitter, phase jitter (peak-to-peak)	t _{CLK_PJ}	-50	-	+50	ps	2, 3, 4
Notes:		•				

Table 96. SATA reference clock input requirements⁶

1. Caution: Only 100 and 125MHz have been tested. In-between values do not work correctly with the rest of the system.

2. At RefClk input

3. In a frequency band from 150 kHz to 15 MHz at BER of 10⁻¹²

4. Total peak-to-peak deterministic jitter must be less than or equal to 50 ps.

5. Measurement taken from differential waveform

6. For recommended operating conditions, see Table 3.

3.19.8.3 AC transmitter output characteristics

This table provides the differential transmitter output AC characteristics for the SATA interface at Gen1i/1m or 1.5 Gbps transmission. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Тур	Max	Units	Notes
Channel speed	t _{CH_SPEED}	-	1.5	-	Gbps	-
Unit Interval	T _{UI}	666.4333	666.6667	670.2333	ps	-
Total jitter data-data 5 UI	U _{SATA_TXTJ5UI}	-	-	0.355	UI p-p	1
Total jitter, data-data 250 UI	U _{SATA_TXTJ250UI}	-	-	0.47	UI p-p	1
Deterministic jitter, data-data 5 UI	U _{SATA_TXDJ5UI}	-	-	0.175	UI p-p	1
Deterministic jitter, data-data 250 UI	U _{SATA_TXDJ250UI}	-	-	0.22	UI p-p	1
Notes:			·		•	•
1. Measured at transmitter output pins	peak to peak phase	variation, rand	om data pattern	1		
2. For recommended operating condit	ions, see Table 3.					

Table 97. Gen1i/1m 1.5 G transmitter AC specifications²

This table provides the differential transmitter output AC characteristics for the SATA interface at Gen2i/2m or 3.0 Gbps transmission. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Тур	Max	Units	Notes
Channel speed	t _{CH_SPEED}	-	3.0	-	Gbps	-
Unit Interval	T _{UI}	333.2167	333.3333	335.1167	ps	-
Total jitter $f_{C3dB} = f_{BAUD} \div 500$	U _{SATA_TXTJfB/500}	-	-	0.37	UI p-p	1
Total jitter f _{C3dB} = f _{BAUD} ÷ 1667	U _{SATA_TXTJfB/1667}	-	-	0.55	UI p-p	1
Deterministic jitter, $f_{C3dB} = f_{BAUD} \div 500$	U _{SATA_TXDJfB/500}	-	-	0.19	UI p-p	1
Deterministic jitter, f _{C3dB} = f _{BAUD} ÷ 1667	U _{SATA_TXDJfB/1667}	-	-	0.35	UI p-p	1
Notes:	•		·		•	
1. Measured at transmitter output pins	peak-to-peak phase	variation, rando	om data pattern			
2. For recommended operating condition	ons, see Table 3.					

Table 98. Gen 2i/2m 3 G transmitter AC specifications²

3.19.8.4 AC differential receiver input characteristics

This table provides the Gen1i/1m or 1.5 Gbps differential receiver input AC characteristics for the SATA interface. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Typical	Мах	Units	Notes
Unit Interval	T _{UI}	666.4333	666.6667	670.2333	ps	-
Total jitter data-data 5 UI	U _{SATA_RXTJ5UI}	-	-	0.43	UI p-p	1
Total jitter, data-data 250 UI	U _{SATA_RXTJ250UI}	-	-	0.60	UI p-p	1
Deterministic jitter, data-data 5 UI	U _{SATA_RXDJ5UI}	-	-	0.25	UI p-p	1
Deterministic jitter, data-data 250 UI	U _{SATA_RXDJ250UI}	-	-	0.35	UI p-p	1
Notes:			·	·	·	
1. Measured at receiver.						
2. For recommended operating condit	ions, see Table 3.					

Table 99. Gen 1i/1m 1.5G receiver AC specifications²

This table provides the differential receiver input AC characteristics for the SATA interface at Gen2i/2m or 3.0 Gbps transmission. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit Interval	T _{UI}	333.2167	333.3333	335.1167	ps	-
Total jitter $f_{C3dB} = f_{BAUD} \div 500$	U _{SATA_RXTJfB/500}	-	-	0.60	UI p-p	1
Total jitter $f_{C3dB} = f_{BAUD} \div 1667$	U _{SATA_RXTJfB/1667}	-	-	0.65	UI p-p	1
Deterministic jitter, $f_{C3dB} = f_{BAUD} \div 500$	U _{SATA_RXDJfB/500}	-	-	0.42	UI p-p	1

Table 100. Gen 2i/2m 3G receiver AC specifications²

Table continues on the next page ...

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Electrical characteristics

Parameter	Symbol	Min	Typical	Мах	Units	Notes		
Deterministic jitter, $f_{C3dB} = f_{BAUD} \div 1667$	U _{SATA_RXDJfB/1667}	-	-	0.35	UI p-p	1		
Notes:								
1. Measured at receiver								
2. For recommended operating conditions, see Table 3.								

Table 100. Gen 2i/2m 3G receiver AC specifications² (continued)

3.19.9 SGMII interface

Each SGMII port features a 4-wire AC-coupled serial link from the SerDes interface of the chip, as shown in Figure 44, where C_{TX} is the external (on board) AC-coupled capacitor. Each SerDes transmitter differential pair features 100- Ω output impedance. Each input of the SerDes receiver differential pair features 50- Ω on-die termination to XGND*n*. The reference circuit of the SerDes transmitter and receiver is shown in Figure 38.

3.19.9.1 SGMII clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

When operating in SGMII mode, the EC*n*_GTX_CLK125 clock is not required for this port. Instead, a SerDes reference clock is required on SD[1:2]_REF_CLK[1:2] and SD[1:2]_REF_CLK[1:2]_Bpins. SerDes 1-2 may be used for SerDes SGMII configurations based on the RCW Configuration field SRDS_PRTCL.

For more information on these specifications, see SerDes reference clocks.

3.19.9.2 SGMII DC electrical characteristics

This section discusses the electrical characteristics for the SGMII interface.

3.19.9.2.1 SGMII and SGMII 2.5x transmit DC specifications

This table describes the SGMII SerDes transmitter AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs $(SDn_TXn \text{ and } SDn_TXn_B)$ as shown in Figure 45.

Table 101. SGMII DC transmitter electrical characteristics $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^4$

Parameter	Symb ol	Min	Тур	Мах	Unit	Notes
Output high voltage	V _{OH}	-	-	1.5 x V _{OD} -max	mV	1

Table continues on the next page ...

Electrical characteristics

Table 101. SGMII DC transmitter electrical characteristics $(XV_{DD} = 1.35 \text{ V or } 1.5 \text{ V})^4$ (continued)

Notes
ECR0 [AMP_RED] =
ECR0 [AMP_RED] = efault)
ECR0 [AMP_RED] =
ECR0 [AMP_RED] =

2. $|V_{OD}| = |V_{SD_TXn} - V_{SD_TXn_B}|$. $|V_{OD}|$ is also referred to as output differential peak voltage. $V_{TX-DIFFp-p} = 2 \times |V_{OD}|$.

3. The $|V_{OD}|$ value shown in the Typ column is based on the condition of XVDD_SRDSn-Typ = 1.35 V or 1.5 V, no common mode offset variation. SerDes transmitter is terminated with 100- Ω differential load between SDn_TXn and SDn_TXn_B.

4. For recommended operating conditions, see Table 3.

This figure shows an example of a 4-wire AC-coupled SGMII serial link connection.

Electrical characteristics

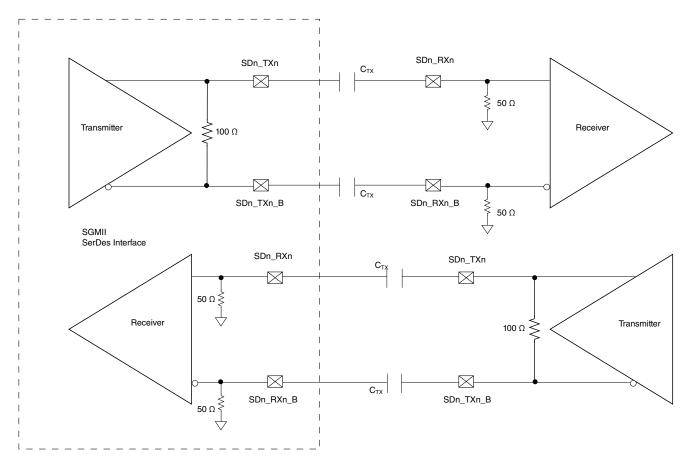


Figure 44. 4-wire AC-coupled SGMII serial link connection example

This figure shows the SGMII transmitter DC measurement circuit.

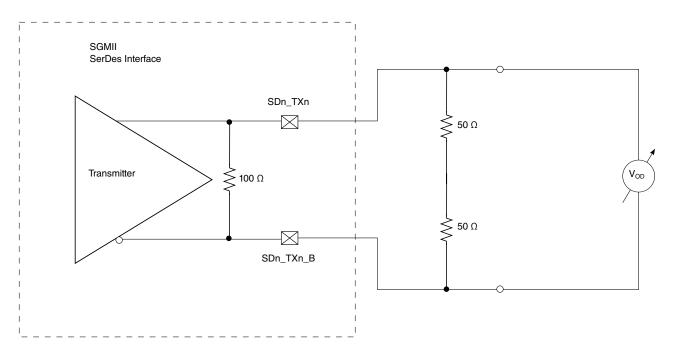


Figure 45. SGMII transmitter DC measurement circuit

This table defines the SGMII 2.5x transmitter DC electrical characteristics for 3.125 GBaud.

Table 102. SGMII 2.5x transmitter DC electrical characteristics (XV_{DD} = 1.35 V or 1.5 V)¹

Parameter	Symbo I	Min	Typical	Мах	Unit	Notes
Output differential voltage	V _{OD}	400	-	600	mV	SRDSxLNmTECR0 [AMP_RED] = 6b000000
Output impedance (differential)	R _O	80	100	120	Ω	-
Notes:						
1. For recommended opera	ting condi	tions, se	e Table 3.			

3.19.9.2.2 SGMII and SGMII 2.5x DC receiver electrical characteristics

This table lists the SGMII DC receiver electrical characteristics. Source synchronous clocking is not supported. Clock is recovered from the data.

Table 103. SGMII DC receiver electrical characteristics $(SV_{DD} = 1.0V)^4$

Parameter		Symbol	Min	Тур	Max	Unit	Notes
DC input voltage range		-	N/A			-	1
Input differential voltage	SRDSxLNmGCR1 [REIDL_TH] = 001	V _{RX_DIFFp-p}	100	-	1200	mV	2

Table continues on the next page...

Para	meter	Symbol	Min	Тур	Max	Unit	Notes
	SRDSxLNmGCR1 [REIDL_TH] = 100		175	-			
Loss of signal threshold	SRDSxLNmGCR1 [REIDL_TH] = 001	V _{LOS}	30	-	100	mV	3
	SRDSxLNmGCR1 [REIDL_TH] = 100		65	-	175		
Receiver differential input ir	npedance	Z _{RX_DIFF}	80	-	120	Ω	-
Notes:							
1. Input must be externally	AC coupled.						

Table 103. SGMII DC receiver electrical characteristics $(SV_{DD} = 1.0V)^4$ (continued)

2. V_{BX DIFFp-p} is also referred to as peak-to-peak input differential voltage.

3. The concept of this parameter is equivalent to the electrical idle detect threshold parameter in PCI Express. See PCI Express DC physical layer receiver specifications, and PCI Express AC physical layer receiver specifications, for further explanation.

4. For recommended operating conditions, see Table 3.

This table defines the SGMII 2.5x receiver DC electrical characteristics for 3.125 GBaud.

Table 104. SGMII 2.5x receiver DC timing specifications $(SV_{DD} = 1.0V)^1$

Parameter	Symbol	Min	Typical	Max	Unit	Notes		
Input differential voltage	V _{RX_DIFFp-p}	200	-	1200	mV	-		
Loss of signal threshold	V _{LOS}	75	-	200	mV	-		
Receiver differential input impedance	Z _{RX_DIFF}	80	-	120	Ω	-		
Notes:								
1. For recommended operating conditions, see Table 3.								

3.19.9.3 SGMII AC timing specifications

This section discusses the AC timing specifications for the SGMII interface.

3.19.9.3.1 SGMII and SGMII 2.5x transmit AC timing specifications

This table provides the SGMII and SGMII 2.5x transmit AC timing specifications. A source synchronous clock is not supported. The AC timing specifications do not include RefClk jitter.

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Unit Interval: 1.25 GBaud (SGMII)	UI	800 - 100 ppm	800	800 + 100 ppm	ps	1
Unit Interval: 3.125 GBaud (2.5x SGMII))	UI	320 - 100 ppm	320	320 + 100 ppm	ps	1

Table 105. SGMII transmit AC timing specifications⁴

Table continues on the next page...

Parameter	Symbol	Min	Тур	Max	Unit	Notes	
Deterministic jitter	JD	-	-	0.17	UI p-p	-	
Total jitter	JT	-	-	0.35	UI p-p	2	
AC coupling capacitor	C _{TX}	10	-	200	nF	3	
Notes:	I.	1		1		1	
1. Each UI is 800 ps \pm 100 ppm or 320 ps	± 100 ppm.						
2. See Figure 42 for single frequency sinu	isoidal jitter m	easurements.					
3. The external AC coupling capacitor is r	equired. It is r	ecommended the	at it be pla	ced near the devi	ice transmitte	r outputs.	
4. For recommended operating conditions, see Table 3.							

 Table 105.
 SGMII transmit AC timing specifications⁴ (continued)

3.19.9.3.2 SGMII AC measurement details

Transmitter and receiver AC characteristics are measured at the transmitter outputs $(SDn_TXn \text{ and } SDn_TXn_B)$ or at the receiver inputs $(SDn_RXn \text{ and } SDn_RXn_B)$ respectively, as depicted in this figure.

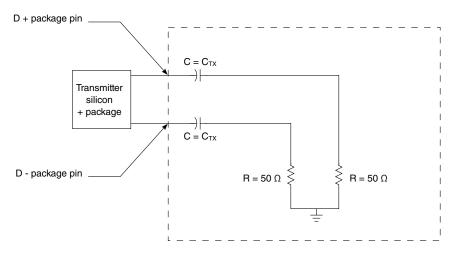


Figure 46. SGMII AC test/measurement load

3.19.9.3.3 SGMII and SGMII 2.5x receiver AC timing Specification

This table provides the SGMII and SGMII 2.5x receiver AC timing specifications. The AC timing specifications do not include RefClk jitter. Source synchronous clocking is not supported. Clock is recovered from the data.

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Deterministic jitter tolerance	J _D	-	-	0.37	UI p-p	1
Combined deterministic and random jitter tolerance	J _{DR}	-	-	0.55	UI p-p	1
Total jitter tolerance	J _T	-	-	0.65	UI p-p	1, 2
Bit error ratio	BER	-	-	10 ⁻¹²	-	-
Unit Interval: 1.25 GBaud (SGMII)	UI	800 - 100 ppm	800	800 + 100 ppm	ps	1
Unit Interval: 3.125 GBaud (2.5x SGMII])	UI	320 - 100 ppm	320	320 + 100 ppm	ps	1
Notes:			ł	1	1	1

Table 106. SGMII Receive AC timing specifications³

1. Measured at receiver

2. Total jitter is composed of three components: deterministic jitter, random jitter, and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of Figure 42. The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk and other variable system effects.

3. For recommended operating conditions, see Table 3.

The sinusoidal jitter in the total jitter tolerance may have any amplitude and frequency in the unshaded region of Figure 43.

3.19.10 QSGMII interface

This section describes the QSGMII clocking and its DC and AC electrical characteristics.

3.19.10.1 QSGMII clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

The ref clock frequency tolerance spec is ± 100 ppm. For more information on these specifications, see SerDes reference clocks.

3.19.10.2 QSGMII DC electrical characteristics

This section discusses the electrical characteristics for the SGMII interface.

3.19.10.2.1 QSGMII transmitter DC specifications

This table describes the QSGMII SerDes transmitter AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs $(SDn_TXn \text{ and } SDn_TXn_B)$.

Table 107. QSGMII DC transmitter electrical characteristics $(XV_{DD} = 1.35V \text{ or } 1.5V)^1$

Parameter	Symbol	Min	Тур	Max	Unit	Notes			
Output differential voltage	V _{DIFF}	400	-	900	mV	-			
Differential resistance	T _{RD}	80	100	120	Ω	-			
Notes:									
1. For recommended operating conditions, see Table 3.									

3.19.10.2.2 **QSGMII DC** receiver electrical characteristics

This table defines the QSGMII receiver DC electrical characteristics.

```
Table 108. QSGMII receiver DC timing specifications (SV_{DD} = 1.0V)^1
```

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Input differential voltage	V _{DIFF}	100	-	900	mV	-
Differential resistance	R _{RDIN}	80	100	120	Ω	-
Notes:			•	-		
1. For recommended operating conditions, see Table 3.						

3.19.10.3 QSGMII AC timing specifications

This section discusses the AC timing specifications for the QSGMII interface.

3.19.10.3.1 QSGMII transmit AC timing specifications

This table provides the QSGMII transmitter AC timing specifications.

Table 109. QSGMII transmit AC timing specifications¹

Parameter	Symbol	Min Typ Max		Max	Unit	Notes	
Transmitter baud rate	T _{BAUD}	5.000 - 100 ppm	5.000	5.000 + 100 ppm	Gbps	-	
Uncorrelated high probability jitter	T _{UHPJ}	-	-	0.15	UI p-p	-	
Total jitter tolerance	J _T	-	-	0.30	UI p-p	-	
Notes:	·					•	
1. For recommended operating conditions, see Table 3.							

3.19.10.3.2 QSGMII receiver AC timing Specification

This table provides the QSGMII receiver AC timing specifications.

Parameter	Symbol	Min	Тур	Max	Unit	Notes
Receiver baud rate	R _{BAUD}	5.000 - 100 ppm	5.000	5.000 + 100 ppm	Gbps	-
Uncorrelated bounded high probability jitter	R _{DJ}	-	-	0.15	UI p-p	-
Correlated bounded high probability jitter	R _{CBHPJ}	-	-	0.30	UI p-p	1
Bounded high probability jitter	R _{BHPJ}	-	-	0.45	UI p-p	-
Sinusoidal jitter, maximum	R _{SJ-max}	-	-	5.00	UI p-p	-
Sinusoidal jitter, high frequency	R _{SJ-hf}	-	-	0.05	UI p-p	-
Total jitter (does not include sinusoidal jitter)	R _{Tj}	-	-	0.60	UI p-p	-
Notes:		•			•	
1. The jitter (R_{CBHPJ}) and amplitude have to be σ	correlated, for	example, by a PC	B trace.			
2. For recommended operating conditions, see	Table 3.					

Table 110. QSGMII receive AC timing specifications²

The sinusoidal jitter may have any amplitude and frequency in the unshaded region of this figure.

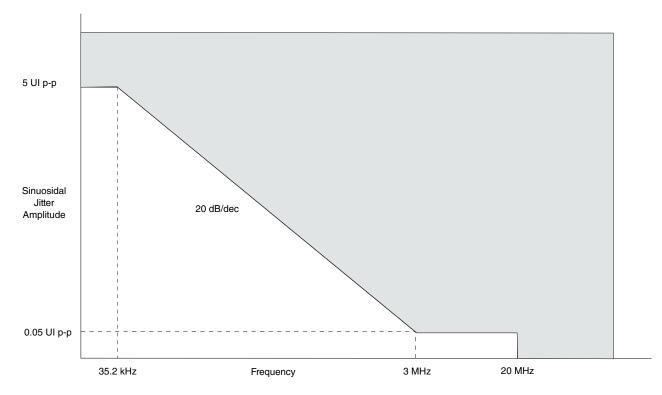


Figure 47. QSGMII single-frequency sinusoidal jitter limits

3.19.11 HiGig/HiGig2 interface

This section describes the HiGig/HiGig2 clocking requirements and its DC and AC electrical characteristics.

3.19.11.1 HiGig/HiGig2 clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

Only SerDes 1 and 2 (SD[1:2]_REF_CLK[1:2] and SD[1:2]_REF_CLK[1:2]_B) may be used for SerDes HiGig/HiGig2 configurations based on the RCW Configuration field SRDS_PRTCL. The ref clock frequency tolerance spec is ±100ppm.

For more information on these specifications, see SerDes reference clocks.

3.19.11.2 HiGig/HiGig2 DC electrical characteristics

This section describes the DC electrical characteristics for HiGig/HiGig2.

3.19.11.2.1 HiGig/HiGig2 transmitter DC electrical characteristics

This table defines the HiGig/HiGig2 transmitter DC electrical characteristics.

Table 111. HiGig/HiGig2 transmitter DC electrical characteristics (XV_{DD} = 1.35V or 1.5V)²

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Differential output voltage	V _{DIFFPP}	800	1000	1600	mV p-p	-
DC Differential transmitter impedance	Z _{TX-DIFF-DC}	80	100	120	Ω	Transmitter DC differential impedance
Notes:	•					•
1. Absolute output voltage limit						
2. For recommended operating condition	s, see Table 3	3.				

3.19.11.2.2 HiGig/HiGig2 receiver DC electrical characteristics

This table defines the HiGig/HiGig2 receiver DC electrical characteristics.

Table 112. HiGig/HiGig2 receiver DC electrical characteristics $(SV_{DD} = 1.0V)^2$

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Differential input voltage	V _{IN}	200	-	1600	mV p-p	1
DC Differential receiver impedance	Z _{RX-DIFF-DC}	80	100	120	Ω	DC Differential receiver impedance
1. Measured at receiver			•			
2. For recommended operating condit	ions, see Table	e 3.				

3.19.11.3 HiGig/HiGig2 AC timing specifications

This section describes the AC timing specifications for HiGig/HiGig2.

3.19.11.3.1 HiGig/HiGig2 transmitter AC timing specifications

This table defines the HiGig/HiGig2 transmitter AC timing specifications. RefClk jitter is not included.

Parameter	Symbol	Min	Typical	Мах	Unit
Deterministic jitter	J _D	-	-	0.17	UI p-p
Total jitter	J _T	-	-	0.35	UI p-p
Unit Interval: 3.125 Gbps (HiGig/HiGig2)	UI	320 - 100 ppm	320	320 + 100 ppm	ps
Unit Interval: 3.75 Gbps (HiGig/HiGig2)	UI	266.66 - 100 ppm	266.66	266.66 + 100 ppm	ps
Notes:					
1. For recommended operating conditions	, see Table 3.				

Table 113. HiGig/HiGig2 transmitter AC timing specifications¹

3.19.11.3.2 HiGig/HiGig2 receiver AC timing specifications

This table defines the HiGig/HiGig2 receiver AC timing specifications. RefClk jitter is not included.

Table 114. HiGig/HiGig2 receiver AC timing specifications³

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Deterministic jitter tolerance	J _D	-	-	0.37	UI p-p	1
Combined deterministic and random jitter tolerance	J _{DR}	-	-	0.55	UI p-p	1
Total jitter tolerance	J _T	-	-	0.65	UI p-p	1, 2
Unit Interval: 3.125 Gbps (HiGig/ HiGig2)	UI	320 - 100ppm	320	320 + 100ppm	ps	-
Unit Interval: 3.75 Gbps (HiGig/HiGig2)	UI	266.66 - 100ppm	266.66	266.66 + 100ppm	ps	-

1. Measured at receiver

2. Total jitter is composed of three components: deterministic jitter, random jitter, and single frequency sinusoidal jitter. The sinusoidal jitter may have any amplitude and frequency in the unshaded region of Figure 43. The sinusoidal jitter component is included to ensure margin for low frequency jitter, wander, noise, crosstalk and other variable system effects.

3. For recommended operating conditions, see Table 3.

3.19.12 XFI interface

This section describes the XFI clocking requirements and its DC and AC electrical characteristics.

3.19.12.1 XFI clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

Only SerDes 2 (SD2_REF_CLK[1:2] and SD2_REF_CLK[1:2]_B) may be used for SerDes XFI configurations based on the RCW Configuration field SRDS_PRTCL.

The ref clock frequency tolerance spec is ± 100 ppm. For more information on these specifications, see SerDes reference clocks.

3.19.12.2 XFI DC electrical characteristics

This section describes the DC electrical characteristics for XFI.

3.19.12.2.1 XFI transmitter DC electrical characteristics

This table defines the XFI transmitter DC electrical characteristics.

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Output differential voltage	V _{TX-DIFF}	360	-	770	mV	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-1.14dB	0.6	1.1	1.6	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-3.5dB	3	3.5	4	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-4.66dB	4.1	4.6	5.1	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-6.0dB	5.5	6.0	6.5	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-9.5dB	9	9.5	10	dB	-
Differential resistance	T _{RD}	80	100	120	Ω	-
Notes:	,	-1		1		1
1. For recommended operating condi	tions, see Table	3.				

Table 115. XFI transmitter DC electrical characteristics (XV_{DD} = 1.35V or 1.5V)¹

3.19.12.2.2 XFI receiver DC electrical characteristics

This table defines the XFI receiver DC electrical characteristics.

Parameter	Symbol	Min	Typical	Max	Unit	Notes	
Input differential voltage	V _{RX-DIFF}	110	-	1050	mV	1	
Differential resistance	R _{RD}	80	100	120	Ω	-	
1. Measured at receiver							
2. For recommended operating conditions, see Table 3.							

Table 116. XFI receiver DC electrical characteristics $(SV_{DD} = 1.0V)^2$

3.19.12.3 XFI AC timing specifications

This section describes the AC timing specifications for XFI.

3.19.12.3.1 XFI transmitter AC timing specifications

This table defines the XFI transmitter AC timing specifications. RefClk jitter is not included.

 Table 117. XFI transmitter AC timing specifications¹

Parameter	Symbol	Min	Typical	Max	Unit
Transmitter baud rate	T _{BAUD}	10.3125 - 100ppm	10.3125	10.3125 + 100ppm	Gb/s
Unit Interval	UI	-	96.96	-	ps
Deterministic jitter	DJ	-	-	0.155	UI p-p
Total jitter	TJ	-	-	0.30	UI p-p
Notes:		1		•	
1. For recommended operating condition	tions, see Table 3.				

3.19.12.3.2 XFI receiver AC timing specifications

This table defines the XFI receiver AC timing specifications. RefClk jitter is not included.

Table 118. XFI receiver AC timing specifications³

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Receiver baud rate	R _{BAUD}	10.3125 - 100ppm	10.3125	10.3125 + 100ppm	Gb/s	-
Unit Interval	UI	-	96.96	-	ps	-
Total non-EQJ jitter	T _{NON-EQJ}	-	-	0.45	UI p-p	1
Total jitter tolerance	TJ	-	-	0.65	UI p-p	1, 2

Table continues on the next page...

Table 118. XFI receiver AC timing specifications³ (continued)

Parameter	Symbol	Min	Typical Max Unit No					
1. The total jitter (T _J) consists of Random Jitter (R _J), Duty Cycle Distortion (DCD), Periodic Jitter (P _J), and Inter symbol Interference (ISI). Non-EQJ jitter can include duty cycle distortion (DCD), random jitter (R _J), and periodic jitter (P _J). Non-EQJ jitter is uncorrelated to the primary data stream with exception of the DCD and so cannot be equalized by the receiver under test. It can exhibit a wide spectrum. Non - EQJ = T_J - ISI = R_J + DCD + P_J								
2. The XFI channel has a loss budget of 9.6 dB @5.5GHz. The channel loss including connector @ 5.5GHz is 6dB. The channel crosstalk and reflection margin is 3.6dB. Manual tuning of TX Equalization and amplitude will be required for performance optimization.								
3. For recommended operating condition	tions, see Tat	ole 3.						

This figure shows the sinusoidal jitter tolerance of XFI receiver.

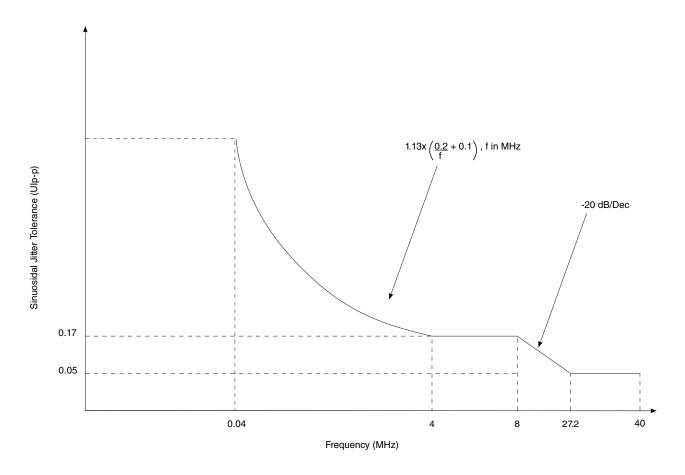


Figure 48. XFI host receiver input sinusoidal jitter tolerance

3.19.13 10GBase-KR interface

This section describes the 10GBase-KR clocking requirements and its DC and AC electrical characteristics.

3.19.13.1 10GBase-KR clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

Only SerDes 2 (SD2_REF_CLK[1:2] and SD2_REF_CLK[1:2]_B) may be used for SerDes 10GBase-KR configurations based on the RCW Configuration field SRDS_PRTCL. The ref clock frequency tolerance spec is ±100ppm.

For more information on these specifications, see SerDes reference clocks .

3.19.13.2 10GBase-KR DC electrical characteristics

This section describes the DC electrical characteristics for 10GBase-KR.

3.19.13.2.1 10GBase-KR transmitter DC electrical characteristics

This table defines the 10GBase-KR transmitter DC electrical characteristics.

Table 119. 10GBaseKR transmitter DC electrical characteristics (XV_{DD} = 1.35V or 1.5V)¹

Parameter	Symbol	Min	Typical	Мах	Unit	Notes
Output differential voltage	V _{TX-DIFF}	800	-	1200	mV	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-1.14dB	0.6	1.1	1.6	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-3.5dB	3	3.5	4	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-4.66dB	4.1	4.6	5.1	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-6.0dB	5.5	6.0	6.5	dB	-
De-emphasized differential output voltage (ratio)	V _{TX-DE-} RATIO-9.5dB	9	9.5	10	dB	-
Differential resistance	T _{RD}	80	100	120	Ω	-

3.19.13.2.2 10GBase-KR receiver DC electrical characteristics

This table defines the 10GBase-KR receiver DC electrical characteristics.

Electrical characteristics

Table 120. 10GBase-KR receiver DC electrical characteristics $(XV_{DD} = 1.35V \text{ or } 1.5V)^1$

Parameter	Symbol	Min	Typical	Мах	Unit	Notes	
Input differential voltage	V _{RX-DIFF}	-	-	1200	mV	-	
Differential resistance	R _{RD}	80	-	120	Ω	-	
1. For recommended operating conditions, see Table 3.							

3.19.13.3 10GBase-KR AC timing specifications

This section describes the AC timing specifications for 10GBase-KR.

3.19.13.3.1 10GBase-KR transmitter AC timing specifications

This table defines the 10GBase-KR transmitter AC timing specifications. RefClk jitter is not included.

 Table 121. 10GBase-KR transmitter AC timing specifications¹

Parameter	Symbol	Min	Typical	Мах	Unit			
Transmitter baud rate	T _{BAUD}	10.3125 - 100 ppm	10.3125	10.3125 + 100 ppm	GBd			
Deterministic jitter	DJ	-	-	0.155	UI p-p			
Total jitter	TJ	-	-	0.30	UI p-p			
1. For recommended operating conditions, see Table 3.								

3.19.13.3.2 10GBase-KR receiver AC timing specifications

This table defines the 10GBase-KR receiver AC timing specifications. RefClk jitter is not included.

Table 122. 10GBase-KR receiver AC timing specifications^{4,3}

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Receiver baud rate	R _{BAUD}	10.3125 - 100 ppm	10.3125	10.3125 + 100 ppm	GBd	-
Random jitter	RJ	-	-	0.130	UI p-p	1
Sinusodial jitter, maximum	S _{J-max}	-	-	0.115	UI p-p	1
Duty cycle distortion	D _{CD}	-	-	0.035	UI p-p	1
Total jitter	TJ	-	-	1.0	UI p-p	1,2

1. The AC specifications do not include Refclk jitter.

2. The Total applied Jitter Tj = ISI + Rj + DCD + Sj-max where ISI is jitter due to frequency dependent loss.

3. TX Equalization and amplitude tuning is through software for performance optimization, as in Freescale provided SDKs.

4. For recommended operating conditions, see Table 3.

3.19.14 Interlaken interface

This section describes the Interlaken clocking requirements and its DC and AC electrical characteristics.

3.19.14.1 Interlaken clocking requirements for SDn_REF_CLKn and SDn_REF_CLKn_B

Only SerDes 3 (SD3_REF_CLK[1:2] and SD3_REF_CLK[1:2]_B) may be used for SerDes Interlaken-LA configurations based on the RCW Configuration field SRDS_PRTCL. The ref clock frequency tolerance spec is ±100ppm.

For more information on these specifications, see SerDes reference clocks.

3.19.14.2 Interlaken-short reach DC electrical characteristics

This section describes the DC electrical characteristics for Interlaken-short reach.

3.19.14.2.1 Interlaken-short reach transmitter DC electrical characteristics

This table defines the Interlaken-short reach transmitter DC electrical characteristics.

Table 123. Interlaken-short reach transmitter DC electrical characteristics $(XV_{DD} = 1.35V \text{ or } 1.5V)^1$

Parameter	Symbol	Min	Typical	Max	Unit	Notes
Output differential voltage	V _{DIFF}	400	-	750	mV	-
Differential resistance	T _{RD}	80	100	120	Ω	-
Notes:			•	•		•
1. For recommended operating co	onditions, see Table 3	3.				

3.19.14.2.2 Interlaken-short reach receiver DC electrical characteristics

This table defines the Interlaken-short reach receiver DC electrical characteristics.

Table 124. Interlaken-short reach receiver DC electrical characteristics $(SV_{DD} = 1.0V)^{1}$

Parameter	Symbol	Min	Typical	Max	Unit	Notes		
Input differential voltage	V _{DIFF}	125	-	1200	mV	-		
Differential resistance	R _{RDIN}	80	100	120	Ω	-		
Notes:								
1. For recommended operating condition	1. For recommended operating conditions, see Table 3.							

3.19.14.3 Interlaken-short reach AC timing specifications

This section describes the AC timing specifications for Interlaken-short reach.

3.19.14.3.1 Interlaken-short reach transmitter AC timing specifications

This table defines the Interlaken-short reach transmitter AC timing specifications. RefClk jitter is not included.

Table 125. Interlaken-short reach transmitter AC timing specifications¹

Parameter	Symbol	Min	Typical	Max	Unit			
Transmitter baud rate	T _{BAUD}	3.125 - 100ppm	3.125	3.125 + 100ppm	Gbps			
Transmitter baud rate	T _{BAUD}	6.25 - 100 ppm	6.25	6.25 + 100 ppm	Gbps			
Uncorrelated high probability jitter	T _{UHPJ}	-	-	0.155	UI p-p			
Total jitter tolerance	TJ	-	-	0.30	UI p-p			
Notes:								
1. For recommended operating conditions, see Table 3.								

3.19.14.3.2 Interlaken-short reach receiver AC timing specifications

This table defines the Interlaken-short reach receiver AC timing specifications. RefClk jitter is not included.

R _{BAUD}	3.125 - 100 ppm	3.125	3.125 + 100	Gbps	1
	ppm			Copps	-
			ppm		
R _{BAUD}	6.25 - 100 ppm	6.25	6.25 + 100 ppm	Gbps	-
R _{UBHPJ}	-	-	0.15	UI p-p	-
R _{CBHPJ}	-	-	0.30	UI p-p	1
R _{BHPJ}	-	-	0.45	UI p-p	-
R _{SJ-max}	-	-	5.00	UI p-p	-
R _{SJ-hf}	-	-	0.05	UI p-p	-
R _{Tj}	-	-	0.60	UI p-p	-
,	br example, by a	PCB trace.		1	4
	R _{UBHPJ} R _{CBHPJ} R _{BHPJ} R _{SJ-max} R _{SJ-hf} R _{Tj} correlated, fo	Р.152 - Р.0ВНРЈ - Р.СВНРЈ - Р.Б.НРЈ - Р.Б.J.max - Р.S.J.max - Р.S.J.max - Р.S.J.max - Р.S.J.max - Р.S.J.max - Р.S.J.max -	AubhPJ- R_{UBHPJ} - R_{CBHPJ} - R_{BHPJ} - R_{SJ-max} - R_{SJ-hf} - R_{Tj} <tr< td=""><td>RubhPJ - - 0.15 RCBHPJ - - 0.30 RBHPJ - - 0.45 RSJ-max - - 5.00 RSJ-hf - 0.05 RTj - - 0.60 correlated, for example, by a PCB trace.</td><td>RUBHPJ - - 0.15 UI p-p R_{CBHPJ} - - 0.30 UI p-p R_{BHPJ} - - 0.45 UI p-p R_{SJ-max} - - 5.00 UI p-p R_{SJ-hf} - - 0.60 UI p-p R_{Tj} - - 0.60 UI p-p</td></tr<>	RubhPJ - - 0.15 RCBHPJ - - 0.30 RBHPJ - - 0.45 RSJ-max - - 5.00 RSJ-hf - 0.05 RTj - - 0.60 correlated, for example, by a PCB trace.	RUBHPJ - - 0.15 UI p-p R _{CBHPJ} - - 0.30 UI p-p R _{BHPJ} - - 0.45 UI p-p R _{SJ-max} - - 5.00 UI p-p R _{SJ-hf} - - 0.60 UI p-p R _{Tj} - - 0.60 UI p-p

Table 126. Interlaken-short reach receiver AC timing specifications²

2. For recommended operating conditions, see Table 3.

The sinusoidal jitter in the total jitter tolerance may have any amplitude and frequency in the unshaded region of this figure.

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

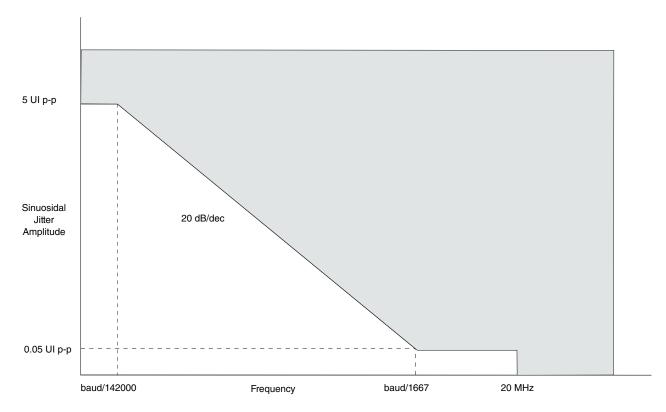


Figure 49. Single-frequency sinusoidal jitter limits

4.1 System clocking

This section describes the PLL configuration of the chip.

4.1.1 PLL characteristics

Characteristics of the chip's PLLs include the following:

- There are five selectable core cluster PLLs which generate a clock for each core cluster from the externally supplied SYSCLK input.
 - Core cluster 1 (cores 0-3) can select from cluster group A PLL 1, 2 or 3 (CGA1, 2, 3 PLL)
 - Core cluster 2 (cores 4-7) can select from cluster group A PLL 1, 2 or 3 (CGA1, 2, 3 PLL), not applicable to T4080 parts

- Core cluster 3 (cores 8-11) can select from cluster group B PLL 1 or 2 (CGB1, 2 PLL)
- The frequency ratio between each of the core cluster PLLs and SYSCLK is selected using the configuration bits as described in Core cluster to SYSCLK PLL ratio. The frequency for each core cluster 1-3 is selected using the configuration bits as described in Table 131 and Table 132.
- The platform PLL generates the platform clock from the externally supplied SYSCLK input. The frequency ratio between the platform and SYSCLK is selected using the platform PLL ratio configuration bits as described in Platform to SYSCLK PLL ratio.
- Cluster group A generates an asynchronous clock for PME from cluster group A PLL 1 or cluster group A PLL 2. Described in Frame Manager (FMn) clock select.
- Cluster group B generates an asynchronous clock for FMan 1 and FMan 2 from the platform PLL, cluster group B PLL 1, or cluster group B PLL 2. Described in Frame Manager (FMn) clock select.
- The DDR block PLL generates an asynchronous DDR clock from the externally supplied DDRCLK input. The frequency ratio is selected using the Memory Controller Complex PLL multiplier/ratio configuration bits as described in DDR controller PLL ratios.
- Each of the four SerDes blocks has 2 PLLs which generate a core clock from their respective externally supplied SD*n*_REF_CLK*n*/SD*n*_REF_CLK*n*_B inputs. The frequency ratio is selected using the SerDes PLL RCW configuration bits as described in SerDes PLL ratio.

4.1.2 Clock ranges

This table provides the clocking specifications for the processor core, platform, memory, and integrated flash controller.

Characteristic Maximum processor core frequency						Unit	Notes	
	1500 MHz		1667 MHz		1800 MHz		1	
	Min	Max	Min	Max	Min	Max	1	
Core cluster group PLL frequency	1000	1500	1000	1667	1000	1800	MHz	1, 2
Core cluster frequency	See note 2	1500	See note 2	1667	See note 2	1800	MHz	2
Platform clock frequency	400	667	400	733	400	733	MHz	1, 7
Memory bus clock frequency	533	800	533	933.333	533	933.333	MHz	1, 3, 4
IFC clock frequency	—	100	_	100	_	100	MHz	5

 Table 127. Processor, platform, and memory clocking specifications

Table continues on the next page...

Characteristic	Maximum processor core frequency							Notes
	1500 MHz		1667 MHz		1800 MHz			
	Min	Мах	Min	Мах	Min	Мах		
PME	See note 6	500	See note 6	550	See note 6	550	MHz	6
FMn	450/667	667	450/667	733	450/667	733	MHz	8

Table 127. Processor, platform, and memory clocking specifications (continued)

1. **Caution:**The platform clock to SYSCLK ratio and core to SYSCLK ratio settings must be chosen such that the resulting SYSCLK frequency, core frequency, and platform clock frequency do not exceed their respective maximum or minimum operating frequencies.

2. The core cluster can run at cluster group PLL/1, PLL/2, or PLL/4. For the PLL/1 case, the minimum frequency is 1000 MHz. With a minimum cluster group PLL frequency of 1000 MHz, this results in a minimum allowable core cluster frequency of 500 MHz for PLL/2. For the PLL/4 case, the minimum allowable core cluster frequency is platform clock frequency / 2. For the case of the minimum platform frequency = 400 MHz, the minimum core cluster frequency is 200 MHz.

3. The memory bus clock speed is half the DDR3/DDR3L data rate. DDR3/3L memory bus clock frequency is limited to min = 533 MHz.

4. The memory bus clock speed is dictated by its own PLL.

5. The integrated flash controller (IFC) clock speed on IFC_CLK[0:2] is determined by the IFC module input clock (platform clock / 2) divided by the IFC ratio programmed in CCR[CLKDIV]. See the chip reference manual for more information.

6. The PME minimum frequency is Platform Frequency / 2. For the case of the minimum platform frequency = 400 MHz, the minimum PME frequency is 200 MHz.

7. The minimum platform frequency should meet the requirements in Minimum platform frequency requirements for highspeed interfaces. For SRIO proper operations the FMAN minimum frequency has to be equal to 528 MHz.

8. If all MACs operate using RGMII or SGMII at 1.25 G, then the minimum required FMAN frequency is 450 MHz. Also, If any MAC operates at a higher rate then the minimum FMAN is 667 MHZ.

4.1.2.1 DDR clock ranges

The DDR memory controller can run only in asynchronous mode, where the memory bus is clocked with the clock provided on the DDRCLK input pin, which has its own dedicated PLL.

This table provides the clocking specifications for the memory bus.

Table 128.	Memory bus	clocking	specifications
------------	------------	----------	----------------

Characteristic	Min	Max	Unit	Notes						
Memory bus clock frequency	533	933.3333	MHz	1, 2, 3						
Notes:	Notes:									
1. Caution: The platform clock to SYSCLK ratio and core to platform clock ratio settings must be chosen such that the resulting SYSCLK frequency, core frequency, and platform frequency do not exceed their respective maximum or minimum operating frequencies. See Platform to SYSCLK PLL ratio, and Core cluster to SYSCLK PLL ratio, and DDR controller PLL ratios, for ratio settings.										
2. The memory bus clock refers to the chip's memory controllers' Dn_MCK[0:3] and Dn_MCK[0:3]_B output clocks, running at half of the DDR data rate.										
3. The memory bus clock speed is dictated by its	s own PLL. See <mark>DDR</mark>	3. The memory bus clock speed is dictated by its own PLL. See DDR controller PLL ratios.								

4.1.3 Platform to SYSCLK PLL ratio

This table lists the allowed platform clock to SYSCLK ratios.

Because the DDR operates asynchronously, the memory-bus clock-frequency is decoupled from the platform bus frequency.

For all valid platform frequencies supported on this chip, set the RCW Configuration field SYS_PLL_CFG = 0b00.

Binary Value of SYS_PLL_RAT	Platform:SYSCLK Ratio
0_0011	3:1
0_0100	4:1
0_0101	5:1
0_0110	6:1
0_0111	7:1
0_1000	8:1
0_1001	9:1
0_1010	10:1
0_1011	11:1
0_1100	12:1
0_1101	13:1
0_1110	14:1
0_1111	15:1
1_0000	16:1
All Others	Reserved

Table 129. Platform to SYSCLK PLL ratios

4.1.4 Core cluster to SYSCLK PLL ratio

The clock ratio between SYSCLK and each of the core cluster PLLs is determined by the binary value of the RCW Configuration field CGm_PLLn_RAT . This table describes the supported ratios. For all valid core cluster frequencies supported on this chip, set the RCW Configuration field $CGn_PLL_CFG = 0b00$.

This table lists the supported asynchronous core cluster to SYSCLK ratios.

Binary value of CGm_PLLn_RAT	Core cluster:SYSCLK Ratio
00_1000	8:1
00_1001	9:1
00_1010	10:1
00_1011	11:1
00_1100	12:1
00_1101	13:1
00_1110	14:1
00_1111	15:1
01_0000	16:1
01_0010	18:1
01_0100	20:1
01_0110	22:1
01_1001	25:1
01_1010	26:1
01_1011	27:1
All others	Reserved

Table 130.	Core cluster PLL to SYSCLK ratios
------------	-----------------------------------

4.1.5 Core complex PLL select

The clock frequency of each core cluster is determined by the binary value of the RCW Configuration field Cn_PLL_SEL . These tables describe the selections available to each core cluster, where each individual core cluster can select a frequency from their respective tables.

NOTE

There is a restriction that requires that the frequency provided to the e6500 core cluster after any dividers must always be greater than half of the platform frequency. Special care must be used when selecting the /2 or /4 outputs of a cluster PLL in which this restriction is observed.

Binary Value of Cn_PLL_SEL for n = 1-2	Core cluster ratio
0000	CGA PLL1/1
0001	CGA PLL1/2
0010	CGA PLL1/4
0100	CGA PLL2/1

Table 131. Core cluster [1-2] PLL select

Table continues on the next page ...

Binary Value of Cn_PLL_SEL for n = 1-2	Core cluster ratio
0101	CGA PLL2/2
0110	CGA PLL2/4
1000	CGA PLL3/1
1001	CGA PLL3/2
1010	CGA PLL3/4
All Others	Reserved

Table 131. Core cluster [1-2] PLL select (continued)

Table 132. Core cluster [3] PLL select

Binary Value of Cn_PLL_SEL for $n = 3$	Core cluster ratio
0000	CGB PLL1/1
0001	CGB PLL1/2
0010	CGB PLL1/4
0100	CGB PLL2/1
0101	CGB PLL2 2
0110	CGB PLL2/4
All Others	Reserved

4.1.6 DDR controller PLL ratios

The three DDR memory controller complexes operate asynchronous to the platform. All DDR controllers operate at the same frequency configuration.

In asynchronous DDR mode, the DDR data rate to DDRCLK ratios supported are listed in the following table. This ratio is determined by the binary value of the RCW Configuration field MEM_PLL_RAT (bits 10-15).

The RCW Configuration field MEM_PLL_CFG (bits 8-9) must be set to MEM_PLL_CFG = 0b00 for all valid DDR PLL reference clock frequencies supported on this chip.

	value of LL_RAT		values of LL_RAT	to DDRCLK			IT/s)		
Rev 1 silicon	Rev 2 silicon	Rev 1 silicon	Rev 2 silicon	Ratio value	-		frequency values that give ata values at Ratio value		
					66.6667 MHz	100 MHz	125 MHz	133.333 MHz	
00_1010	00_0101	10	5	10				1333.333	

 Table 133. DDR data Rate to DDRCLK ratios¹

Table continues on the next page...

-	value of LL_RAT		values of LL_RAT	DDR data rate to DDRCLK	Resulting DDR data-rate (MT/s)		IT/s)	
Rev 1 silicon	Rev 2 silicon	Rev 1 silicon	Rev 2 silicon	Ratio value	Examples of DDRCLK frequency values that give typical DDR data rata values at Ratio value		-	
					66.6667 MHz	100 MHz	125 MHz	133.333 MHz
00_1100	00_0110	12	6	12			1500	1600
00_1110	00_0111	14	7	14			1750	1866.667
01_0000	00_1000	16	8	16	1066.667	1600		
01_0010	00_1001	18	9	18		1800		
01_0100	00_1010	20	10	20	1333.333			
All Others	All Others	Reserved	Reserved	Reserved				
	Notes:							

Table 133. DDR data Rate to DDRCLK ratios¹ (continued)

1. This table shows examples of standard DDR data rate resulted from multiplying the MEM_PLL_RAT by some common DDRCLK frequencies like 66.66MHZ, 100MHz, or 133MHZ. Customers can supply of course different DDRCLK frequency from the common ones presented in this table and thus they have to pick up the correct MEM_PLL_RAT value that will give them a common DDR data rate and always use a value for Rev2 silicon that is half of what is supposed to be given in Rev1 or simply in rev2 MEM_PLL_RAT = 0.5 * DDR data rate/DDRCLK.

4.1.7 SerDes PLL ratio

The clock ratio between each of the three SerDes PLLs and their respective externally supplied SD*n*_REF_CLK*n*/SD*n*_REF_CLK*n*_B inputs is determined by a set of RCW Configuration fields-SRDS_PRTCL_S*n*, SRDS_PLL_REF_CLK_SEL_S*n*, and SRDS_DIV_*_S*n*-as shown in this table.

SerDes protocol (given lane)	Valid reference clock frequency	Legal setting for SRDS_PRTCL_S <i>n</i>	Legal setting for SRDS_PLL_RE F_CLK_SEL_S <i>n</i>	Legal setting for SRDS_DIV_*_Sn	Notes
	High-sp	eed serial and debug inter	faces		
PCI Express 2.5 GT/s	100 MHz	Any PCIe	0b0: 100 MHz	2b10: 2.5 G	1
(doesn't negotiate upwards)	125 MHz		0b1: 125 MHz		1
PCI Express 5 GT/s	100 MHz	Any PCIe	0b0: 100 MHz	2b01: 5.0 G	1
(can negotiate up to 5 GT/s)	125 MHz		0b1: 125 MHz		1
PCI Express 8 GT/s	100 MHz	Any PCIe	0b0: 100 MHz	2b00: 8.0 G	1
(can negotiate up to 8 GT/s)	125 MHz		0b1: 125 MHz		1
Serial RapidIO 2.5 Gbaud	100 MHz	SRIO @ 2.5/5 Gbaud	0b0: 100 MHz	0b1: 2.5 G	-
	125 MHz		0b1: 125 MHz	1	-
Serial RapidIO 3.125 Gbaud	125 MHz	SRIO @ 3.125 Gbaud	0b0: 125 MHz	Don't care	-

 Table 134.
 Valid SerDes RCW encodings and reference clocks

Table continues on the next page...

Table 134.	Valid SerDes RCW	encodings and reference clocks	(continued)

SerDes protocol (given lane)	Valid reference clock frequency	Legal setting for SRDS_PRTCL_S <i>n</i>	Legal setting for SRDS_PLL_RE F_CLK_SEL_S <i>n</i>	Legal setting for SRDS_DIV_*_Sn	Notes
	156.25 MHz		0b1: 156.25 MHz		-
Serial RapidIO 5 Gbaud	100 MHz	SRIO @ 2.5/5 Gbaud	0b0: 100 MHz	0b0: 5.0 G	-
	125 MHz		0b1: 125 MHz		-
Interlaken Lookaside (6.25	125 MHz	Interlaken LA @ 6.25	0b0: 125 MHz	Don't care	-
Gbps)	156.25 MHz	Gbps	0b1: 156.25 MHz		-
Interlaken Lookaside (10.3125	156.25 MHz	Interlaken LA @ 10.3125	0b0: 156.25 MHz	Don't care	-
Gbps)	161.1328125 MHz	Gbps	0b1: 161.1328125 MHz		-
SATA (1.5 or 3 Gbps)	100 MHz	Any SATA	0b0: 100 MHz	Don't care	2
	125 MHz		0b1: 125 MHz		
Debug (2.5 Gbps)	100 MHz	Aurora @ 2.5/5 Gbps	0b0: 100 MHz	0b1: 2.5 G	-
	125 MHz		0b1: 125 MHz		-
Debug (3.125 Gbps)	125 MHz	Aurora @ 3.125 Gbps	0b0: 125 MHz	Don't Care	-
	156.25 MHz		0b1: 156.25 MHz		-
Debug (5 Gbps)	100 MHz	Aurora @ 2.5/5 Gbps	0b0: 100 MHz	0b0: 5.0 G	-
	125 MHz		0b1: 125 MHz		-
		Networking interfaces			
SGMII (1.25 Gbaud)	100 MHz	SGMII @ 1.25 Gbaud	0b0: 100 MHz	Don't care	-
	125 MHz		0b1: 125 MHz		-
2.5x SGMII (3.125 Gbaud)	125 MHz	SGMII @ 3.125 Gbaud	0b0: 125 MHz	Don't care	-
	156.25 MHz		0b1: 156.25 MHz		-
QSGMII (5.0 Gbps)	100 MHz	Any QSGMII	0b0: 100 MHz	0b0: 5.0 G	-
	125 MHz		0b1: 125 MHz		-
XAUI (3.125 Gb/s)	125 MHz	XAUI @ 3.125 Gb/s	0b0: 125 MHz	Don't care	-
	156.25 MHz		0b1: 156.25 MHz		-
HiGig or HiGig2 (3.125 Gbps)	125 MHz	HiGig @ 3.125 Gbps	0b0: 125 MHz	Don't care	-
	156.25 MHz		0b1: 156.25 MHz		-
HiGig or HiGig2 (3.75 Gbps)	125 MHz	HiGig @ 3.75 Gbps	0b0: 125 MHz	Don't care	-
	156.25 MHz		0b1: 156.25 MHz		-
XFI (10.3125 Gbps)	156.25 MHz	XFI @ 10.3125 Gbps	0b0: 156.25 MHz	Don't care	-
10GBase-KR (10.3125 GBd)	156.25 MHz	10GBase-KR @ 10.3125 GBd	0b0: 156.25 MHz	Don't care	-

1. A spread-spectrum reference clock is permitted for PCI Express. However, if any other high-speed interfaces such as sRIO, Interlaken, SATA, SGMII, SGMII 2.5x, QSGMII, XAUI, XFI, 10GBase-KR, HiGig/HiGig2 or Aurora are used concurrently on the same SerDes bank, spread-spectrum clocking is not permitted.

2. SerDes lanes configured as SATA initially operate at 3.0 Gbps. 1.5 Gbps operation may later be enabled through the SATA IP itself. It is possible for software to set each SATA at different rates.

4.1.8 Frame Manager (FMn) clock select

The following tables describe the clocking options that may be applied to each FM. The clock selection is determined by the binary value of the RCW Clocking Configuration fields HWA_CGB_M1_CLK_SEL and HWA_CGB_M2_CLK_SEL.

Binary value of HWA_CGB_Mn_CLK_SEL	Frame Manager (FM1) clock select ¹	Frame Manager (FM2) clock select ¹
000b, 001b	Reserved	Reserved
010b	Cluster group B PLL 1/2	Cluster group B PLL 2/2
011b	Cluster group B PLL 1/3	Cluster group B PLL 2/3
100b	Cluster group B PLL 1/4	Cluster group B PLL 2/4
101b	Platform clock frequency/1	Platform clock frequency/1
110b	Cluster group B PLL 2/2	Cluster group B PLL 1/2
111b	Reserved	Cluster group B PLL 1/3
1. For max frequency, see Table 12		

Table 135. Frame Manager (FMn) clock select

4.1.9 Pattern Matching Engine (PME) clock select

The PME can be synchronous with or asynchronous to the platform, depending on configuration.

This table describes the clocking options that may be applied to the PME. The clock selection is determined by the binary value of the RCW Clocking Configuration field HWA_CGA_M1_CLK_SEL.

Binary Value of HWA_CGA_M1_CLK_SEL	PME Frequency ¹			
000b	Platform clock frequency/2 (synchronous mode)			
001b	Reserved			
010b	Cluster group A PLL 1/2 (Asynchronous mode)			
011b	Cluster group A PLL 1/3 (Asynchronous mode)			
100b	Cluster group A PLL 1/4 (Asynchronous mode)			
101b	Reserved			
110b	Cluster group A PLL 2/2 (Asynchronous mode)			
111b	Cluster group A PLL 2/3 (Asynchronous mode)			
Note:				
1. For asynchronous mode, max frequency, see Table 127.				

Table 136. Pattern Matching Engine clock select

4.1.10 Frequency options

This section discusses interface frequency options.

4.1.10.1 SYSCLK and core cluster frequency options

This table shows the expected frequency options for SYSCLK and core cluster frequencies.

Core cluster: SYSCLK Ratio ²		SYSCLK (MHz) ²	
	66.67	100.00	133.33
		Core cluster Frequency	(MHz) ¹
8:1			1067
9:1			1200
10:1		1000	1333
11:1		1100	1467
12:1		1200	1600
13:1			
14:1		1400	
15:1	1000	1500	
16:1	1067	1600	
18:1	1200	1800	
20:1	1333		
22:1	1467		
25:1	1667		
26:1			
27:1	1800		
Notes:	1		
1. Core cluster frequency values are	e shown rounded up to th	e nearest whole number (deci	mal place accuracy removed)
2. Example values.			

Table 137. SYSCLK and core cluster frequency options

4.1.10.2 SYSCLK and platform frequency options

This table shows the expected frequency options for SYSCLK and platform frequencies.

Table 138.	SYSCLK and	platform freq	uency options
------------	------------	---------------	---------------

Platform: SYSCLK Ratio ³	SYSCLK (MHz) ³			
	66.67	100.00	133.33	
		Platform Frequency (N	/Hz) ¹	
4:1		400	533	
5:1			667	
6:1	400 ²	600		
7:1		700		
8:1	533			
9:1	600			
10:1	667			
11:1	733			
12:1				
Notes:				
1. Platform frequency values are sl	hown rounded up to the ne	arest whole number (decimal	place accuracy removed)	
2. A minimum platform clock freque	ency requirement is 528MI	Iz if x4 SRIO is used.		

3.Example values.

4.1.10.3 DDRCLK and DDR data rate frequency options

This table shows the expected frequency options for DDRCLK and DDR data rate frequencies.

Table 139. DDRCLK and DDR data rate frequency options

DDR data rate: DDRCLK	DDRCLK (MHz) ²					
Ratio ²	66.67	100.00	125.00	133.33		
	DDR Data Rate (MT/s) ¹					
10:1				1333		
12:1			1500	1600		
14:1			1750	1866		
16:1	1067	1600				
18:1		1800				
20:1	1333					
Notes:						
1. DDR data rate values are	shown rounded up to	o the nearest whole numb	ber (decimal place accur	acy removed).		

2. Example values.

4.1.10.4 SYSCLK and FMan frequency options

These table shows the expected frequency options for SYSCLK and FMan frequencies.

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Table 140. SYSCLK and FMan frequency options (clocked by CGB PLLn / 2)

Core cluster: SYSCLK Ratio ³		SYSCLK (MHz) ³	
	66.67	100.00	133.33
		FMan Frequency (MH	z) ^{1, 2}
8:1			533
9:1			600
10:1		500	667
11:1		550	733
12:1		600	
13:1			
14:1		700	
15:1	500	750	
16:1	533		
18:1	600		
20:1	667		
22:1	733		
25:1			
26:1			
27:1			
Notes:			
1. FMan frequency values are show	n rounded up to the near	est whole number (decimal pla	ace accuracy removed).
2. For min frequency, see Table 12	7.		
3 Example values			

3. Example values.

Table 141. SYSCLK and FMan frequency options (clocked by CGB PLLn / 3)

Core cluster: SYSCLK Ratio ³		SYSCLK (MHz) ³	
	66.67	100.00	133.33
		FMan Frequency (MHz))1, 2
8:1			
9:1			
10:1			
11:1			489
12:1			533
13:1			578
14:1		467	
15:1		500	
16:1		533	
18:1		600	
20:1			
22:1	489		

Table continues on the next page...

Table 141. SYSCLK and FMan frequency options (clocked by CGB PLLn / 3) (continued)

		SYSCLK (MHz) ³		
66.67	100.00	133.33		
FMan Frequency (MHz) ^{1, 2}				
556				
578				
600				
rounded up to the nearest	whole number (decimal place ac	curacy removed)		
5	578 500	56 578 578 578 578 578 578 578 578 578 578		

2. For min frequency, see Table 127

3. Example values.

Table 142. SYSCLK and FMan frequency options (clocked by CGB PLL1 / 4)

Core cluster: SYSCLK Ratio ³		SYSCLK (MHz) ³		
	66.67	100.00	133.33	
		FMan Frequency (MHz) ^{1, 2}		
8:1				
9:1				
10:1				
11:1				
12:1				
13:1				
14:1				
15:1				
16:1				
18:1		450		
20:1				
22:1				
25:1				
26:1				
27:1	450			
Notes:	1			

Notes:

1. FMan frequency values are shown rounded up to the nearest whole number (decimal place accuracy removed).

2. For min frequency, see Table 127.

3. Example values.

Table 143. SYSCLK and FMan frequency options (clocked by platform frequency/1)

Platform: SYSCLK Ratio ³	SYSCLK (MHz) ³		
	66.67	100.00	133.33
	FMan Frequency (MHz) ^{1, 2}		
4:1			533
5:1			667
6:1		600	
7:1		700	
8:1	533		
9:1	600		
10:1	667		
11:1	733		
12:1			
Notes:			
1. FMan frequency values are show	vn rounded up to the near	est whole number (decimal pla	ace accuracy removed).
2. For min frequency, see Table 127.			

3. Example values.

4.1.10.5 SYSCLK and PME frequency options

These table shows the expected frequency options for SYSCLK and PME frequencies.

Table 144. SYSCLK and PME frequency options (clocked by CGA PLLn / 2)

Core cluster: SYSCLK Ratio ²		SYSCLK (MHz) ²	
	66.67	100.00	133.33
	PME Frequency (MHz) ¹		
8:1			533
9:1			600
10:1		500	
11:1		550	
12:1		600	
13:1			
14:1			
15:1	500		
16:1	533		
18:1	600		
20:1			
22:1			
25:1			
26:1			

Table continues on the next page...

Table 144. SYSCLK and PME frequency options (clocked by CGA PLLn / 2) (continued)

Core cluster: SYSCLK Ratio ²	SYSCLK (MHz) ²		
	66.67	100.00	133.33
		PME Frequency (MHz) ¹	
27:1			
Notes:			
1. PME frequency values are shown rounded up to the nearest whole number (decimal place accuracy removed).			
2. Example values.			

Table 145. SYSCLK and PME frequency options (clocked by CGA PLLn / 3)

Core cluster: SYSCLK Ratio ²		SYSCLK (MHz) ²	
	66.67	100.00	133.33
	PME Frequency (MHz) ¹		
8:1			
9:1			400
10:1			444
11:1			489
12:1		400	533
13:1			578
14:1		467	
15:1		500	
16:1		533	
18:1	400	600	
20:1	444		
22:1	489		
25:1	556		
26:1	578		
27:1	600		
Notes:			
1. PME frequency values are shown	n rounded up to the neare	st whole number (decimal plac	ce accuracy removed).
2 Example values			

2. Example values.

Table 146. SYSCLK and PME frequency options (clocked by platform frequency/2)

Platform: SYSCLK Ratio ²	SYSCLK (MHz) ²		
	66.67 100.00 133.33		
		PME Frequency (MHz) ¹	
4:1		200	267
5:1			334

Table continues on the next page ...

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

Table 146. SYSCLK and PME frequency options (clocked by platform frequency/2) (continued)

Platform: SYSCLK Ratio ²		SYSCLK (MHz) ²	
	66.67	100.00	133.33
	PME Frequency (MHz) ¹		
6:1	200	300	400
7:1		350	
8:1	267	400	
9:1	300		
10:1	334		
11:1	367		
12:1	400		
Notes:			
1. PME frequency values are show	n rounded up to the neare	est whole number (decimal plac	ce accuracy removed).
2. Example values.			

4.1.10.6 Minimum platform frequency requirements for high-speed interfaces

The platform clock frequency must be considered for proper operation of high-speed interfaces as described below.

For proper PCI Express operation, the platform clock frequency must be greater than or equal to:

527 MHz x (PCI Express link width)

16

Figure 50. Gen 1 PEX minimum platform frequency

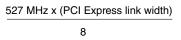


Figure 51. Gen 2 PEX minimum platform frequency

527 MHz x (PCI Express link width)

Figure 52. Gen 3 PEX minimum platform frequency

See section "Link Width," in the chip reference manual for PCI Express interface width details. Note that "PCI Express link width" in the above equation refers to the negotiated link width as the result of PCI Express link training, which may or may not be the same

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

as the link width POR selection. It refers to the widest port in use, not the combined width of the number ports in use. For instance, if two x4 PCIe Gen3 ports are in use, 527MHz platform frequency is needed to support by using Gen 3 equation (527 x 4/4, not 527 x 4 x 2/4).

4.2 Power supply design

4.2.1 Voltage ID (VID) controllable supply

To guarantee performance and power specifications, a specific method of selecting the optimum voltage-level must be implemented when the chip is used. As part of the chip's boot process, software must read the VID efuse values stored in the Fuse Status register (FUSESR) and then configure the external voltage regulator based on this information. This method requires a point of load voltage regulator for each chip.

NOTE

During the power-on reset process, the fuse values are read and stored in the FUSESR. It is expected that the chip's boot code reads the FUSESR value very early in the boot sequence and updates the regulator accordingly.

The default voltage regulator setting that is safe for the system to boot is the recommended operating V_{DD} at initial start-up of 1.025 V. It is highly recommended to select a regulator with a Vout range of at least 0.9 V to 1.1 V, with a resolution of 12.5 mV or better, when implementing a VID solution. If the VID for a specific part is already known at initial start-up, it is acceptable to program the voltage regulator to the VID value. The device does not require an initial voltage of 1.025V at start-up.

The table below lists the valid VID efuse values that will be programmed at the factory for this chip.

Binary value of DA_V / DA_ALT_V	V _{DD} voltage
00001	0.9875 V
00010	0.9750 V
10000	1.0000 V
10001	1.0125 V
10010	1.0250 V
All other values	See the complete list in the Fuse Status Register (DCFG_CCSR_FUSESR) section of the chip reference manual.

Table 147. Fuse Status Register (DCFG_CCSR_FUSESR)

If DA_ALT_V is not all zeros, then software should read DA_ALT_V for the VID value and not the DA_V. For additional information on VID, please see the chip reference manual.

4.2.1.1 Options for system design

There are several widely-accepted options available to the system designer for obtaining the benefits of a VID solution. The most common option is to use the VID solution to drive a system's controllable voltage-regulators through a sideband interface such as a simple parallel bus or PMBus interface. PMbus is similar to I²C but with extensions to improve robustness and address shortcomings of I²C; the PMBus specification can be found at www.pmbus.org. The simple parallel bus is supported by the chip through GPIO pins and the PMBus interface is supported by an I²C interface. Other VID solutions may be to access an FPGA/ASIC or separate power management chip through the IFC, SPI, or other chip-specific interface, where the other device then manages the voltage regulator. The method chosen for implementing the chip-specific voltage in the system is decided by the user.

4.2.1.1.1 Example 1: Regulators supporting parallel bus configuration

In this example, a user builds a VID solution using controllable regulators with a parallel bus. In this implementation, the user chooses to utilize any subset of the available GPIO pins on the chip except those noted below.

NOTE

GPIO pins that are muxed on an interface used by the application for loading RCW information are not available for VID use.

It is recommended that all GPIO pins used for VID are located in the same 32-bit GPIO IP block so that all bits can be accessed with a single read or write.

The general procedure for setting the core voltage regulator to the desired operating voltage is as follows:

- 1. The GPIO pins are released to high-impedance at POR. Because GPIO pins default to being inputs, they do not begin automatically driving after POR, and only work as outputs under software control.
- 2. The board is responsible for a default voltage regulator setting that is "safe" for the system to boot. To achieve this, the user puts pull-up and/or pull-down resistors on the GPIO pins as needed for that specific system. For the case where the regulator's

QorIQ T4240 Data Sheet, Rev. 1, 05/2016

interface operates at a different voltage than OV_{DD} , the chip's GPIO module can be operated in an open drain configuration.

- 3. There is no direct connection between the Fuse Status Register (FUSESR) and the chip's pins. As part of the chip's boot process, software must read the efuse values stored in the FUSESR and then configure the voltage regulator based on this information. The software determines the proper value for the parallel interface and writes it to the GPIO block data (GPDAT) register. It then changes the GPIO direction (GPDIR) register from input to output to drive the new value on the device pins, thus overriding the board configuration default value. Note that some regulators may require a series of writes so that the voltage is slowly stepped from its old to its new value.
- 4. When the voltage has stabilized, software adjusts the operating frequencies as desired.

Upon completion of configuration, some regulators may have a write-protect pin to prevent undesired data changes after configuration is complete. A single GPIO pin on the chip could be allocated for this task if desired.

4.2.1.1.2 Example 2: Regulators supporting PMBus configuration

In this example, a user builds a VID solution using controllable regulators with a PMBus interface. For the case where the regulator's interface operates at a different voltage than DV_{DD} , the chip's I²C module can be operated in an open-drain configuration.

In this implementation, the user chooses to utilize any I²C interface available on the chip. These regulators have a means for setting a safe, default, operating value either through strapping pins or through a default, non-volatile store.

NOTE

If I^2C1 controller is selected, it is important that its calling address is different than the 7-bit value of 0x50h used by the pre-boot loader (PBL) for RCW and pre-boot initialization.

The general procedure for setting the core voltage regulator to the desired operating voltage is as follows:

- 1. The board is responsible for configuring a safe default value for the controllable regulator either through dedicated pins or its non-volatile store.
- 2. As part of the chip's boot process, software must read the efuse values stored in the FUSESR register and then configure the voltage regulator based on this information. The software decides on a new configuration and sends this value across the I²C interface connected to the regulator's PMBus interface. Note that some regulators may require a series of writes so that the voltage is slowly stepped from its old to its new value.

3. When the voltage has stabilized, software adjusts the operating frequencies as desired.

Upon completion of configuration, some regulators may have a write-protect pin to prevent undesired data changes after configuration is complete. A single GPIO pin on the chip could be allocated for this task, if desired.

4.2.1.1.3 Example 3: Regulators supporting FPGA/ASIC or separate power management device configuration

In this example, a user builds a VID solution using controllable regulators that are managed by a FPGA/ASIC or a separate power-management device. In this implementation, the user chooses to utilize the IFC, eSPI or any other available chip interface to connect to the power-management device.

The general procedure for setting the core voltage regulator to the desired operating voltage is as follows:

- 1. The board is responsible for configuring a safe default value for the controllable regulator either through dedicated pins or its non-volatile store.
- 2. As part of the chip's boot process, software must read the efuse values stored in the FUSESR and then configure the voltage regulator based on this information. The software decides on a new configuration and sends this value across the IFC, eSPI, or any other interface that is used to connect to the FPGA/ASIC or separate power-management device that manages the regulator. Note that some regulators may require a series of writes so that the voltage is slowly stepped from its old to its new value.
- 3. When the voltage has stabilized, software adjusts the operating frequencies as desired.

Upon completion of configuration, some regulators may have a write-protect pin to prevent undesired data changes after configuration is complete. A single GPIO pin on the chip could be allocated for this task, if desired.

4.2.2 Core and platform supply voltage filtering

The V_{DD} supply is normally derived from a high current capacity linear or switching power supply which can regulate its output voltage very accurately despite changes in current demand from the chip within the regulator's relatively low bandwidth. Several bulk decoupling capacitors must be distributed around the PCB to supply transient current demand above the bandwidth of the voltage regulator.

These bulk capacitors should have a low ESR (equivalent series resistance) rating to ensure the quick response time necessary. They should also be connected to the power and ground planes through two vias to minimize inductance. However, customers should work directly with their power regulator vendor for best values and types of bulk capacitors.

As a guideline for customers and their power regulator vendors, NXP recommends that these bulk capacitors be chosen to maintain the positive transient power surges to less than VID+50 mV (except that a positive transient of up to +100 mV can be tolerated for less than 1 us, negative transient undershoot should comply with specification of VID-30mV) for current steps of up to 20A for 12 cores, 15A for 8 cores and 10A for 4 cores with a slew rate of 12 A/us.

These bulk decoupling capacitors will ideally supply a stable voltage for current transients into the megahertz range. Above that, see Decoupling recommendations for further decoupling recommendations.

4.2.3 PLL power supply filtering

Each of the PLLs described in System clocking is provided with power through independent power supply pins (AV_{DD} _PLAT, A_{VDD} _CGAn, A_{VDD} _CGBn and AV_{DD} _Dn and AV_{DD} _SDn_PLLn). AV_{DD} _PLAT, A_{VDD} _CGAn, A_{VDD} _CGBn and AV_{DD} _Dn voltages must be derived directly from a 1.8 V voltage source through a low frequency filter scheme. AV_{DD} _SDn_PLLn voltages must be derived directly from the XnV_{DD} source through a low frequency filter scheme. The recommended solution for PLL filtering is to provide independent filter circuits per PLL power supply, as illustrated in Figure 53, one for each of the AV_{DD} pins. By providing independent filters to each PLL, the opportunity to cause noise injection from one PLL to the other is reduced. This circuit is intended to filter noise in the PLL's resonant frequency range from a 500 kHz to 10 MHz range.

Each circuit should be placed as close as possible to the specific AV_{DD} pin being supplied to minimize noise coupled from nearby circuits. It should be possible to route directly from the capacitors to the AV_{DD} pin, which is on the periphery of the footprint, without the inductance of vias.

This figure shows the PLL power supply filter circuit.

Where:

- $R = 5 \Omega \pm 5\%$
- C1 = 10 μ F ± 10%, 0603, X5R, with ESL ≤ 0.5 nH
- C2 = 1.0 μ F ± 10%, 0402, X5R, with ESL ≤ 0.5 nH

NOTE

A higher capacitance value for C2 may be used to improve the filter as long as the other C2 parameters do not change (0402 body, X5R, ESL \leq 0.5 nH).

NOTE

Keep filter close to pin. Voltage and tolerance for AV_{DD} is defined at the input of the PLL supply filter and not the pin of AV_{DD} .

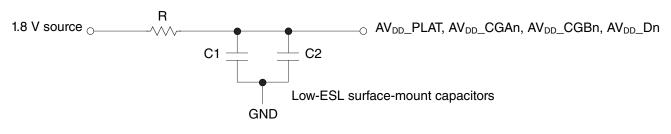


Figure 53. PLL power supply filter circuit

The AV_{DD}_SDn_PLLn signals provide power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following Figure 54. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD}_SDn_PLLn balls to ensure it filters out as much noise as possible. The ground connection should be near the AV_{DD}_SDn_PLLn balls. The 0.003- μ F capacitors closest to the balls, followed by a 4.7- μ F and 47- μ F capacitor, and finally the 0.33 Ω resistor to the board supply plane. The capacitors are connected from AV_{DD}_SDn_PLLn to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All traces should be kept short, wide, and direct.

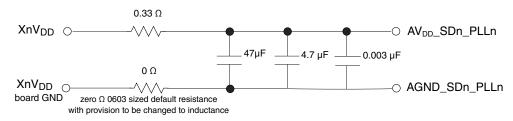


Figure 54. SerDes PLL power supply filter circuit

Note the following:

- AV_{DD} _SD*n*_PLL*n* should be a filtered version of XnV_{DD} .
- Signals on the SerDes interface are fed from the XnV_{DD} power plane.
- Voltage for AV_{DD}_SDn_PLLn is defined at the PLL supply filter and not the pin of AV_{DD}_SDn_PLLn.

QorlQ T4240 Data Sheet, Rev. 1, 05/2016

- A 47- μ F 0805 XR5 or XR7, 4.7- μ F 0603, and 0.003- μ F 0402 capacitor are recommended. The size and material type are important. A 0.33- $\Omega \pm 1\%$ resistor is recommended.
- There needs to be dedicated analog ground, AGND_SD*n*_PLL*n* for each AV_{DD}_SD*n*_PLL*n* pin up to the physical local of the filters themselves.

4.2.4 SnV_{DD} power supply filtering

 SnV_{DD} must be supplied by a decicated linear regulator.

An example solution for SnV_{DD} filtering, where SnV_{DD} is sourced from a linear regulator, is illustrated in Figure 55. The component values in this example filter are system dependent and are still under characterization, component values may need adjustment based on the system or environment noise.

Where:

- C1 = 0.003 μ F ± 10%, X5R, with ESL ≤ 0.5 nH
- C2 and C3 = 2.2 μ F ± 10%, X5R, with ESL ≤ 0.5 nH
- F1 and F2 are 0603 sized Ferrite SMD, like the Murata part BLM18PG121SH1. Its maximum DC resistance is 0.05Ω , or 0.025Ω for the parallel resultant, and each has about a 120 Ω +/- 25% of AC impedance at 100 MHz, which is half valued for the parallel resultant, with individual maximum DC current carrying capacity of 2Amps.
- Bulk and decoupling capacitors are added, as needed, per power supply design.

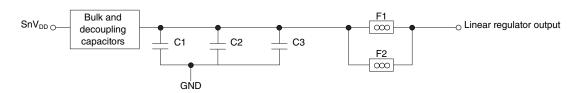


Figure 55. SV_{DD} power supply filter circuit

Note the following:

- Please refer to Power-on ramp rate, for maximum SnV_{DD} power-up ramp rate.
- There needs to be enough output capacitance or a soft start feature to assure ramp rate requirement is met.
- The ferrite beads should be placed in parallel to reduce voltage droop.
- Besides a linear regulator, a low noise dedicated switching regulator can also be used. 10 mVp-p, 50kHz 500MHz is the noise goal.

4.2.5 XnV_{DD} power supply filtering

 XnV_{DD} may be supplied by a linear regulator or sourced by a filtered GnV_{DD}. Systems may design in both options to allow flexibility to address system noise dependencies. However, for initial system bring-up, the linear regulator option is highly recommended.

An example solution for XnV_{DD} filtering, where XnV_{DD} is sourced from a linear regulator, is illustrated in Figure 56. The component values in this example filter are system dependent and are still under characterization, component values may need adjustment based on the system or environment noise.

Where:

- C1 = 0.003 μ F ± 10%, X5R, with ESL ≤ 0.5 nH
- C2 and C3 = 2.2 μ F ± 10%, X5R, with ESL ≤ 0.5 nH
- F1 and F2 are 0603 sized Ferrite SMD, like the Murata part BLM18PG121SH1. Its maximum DC resistance is 0.05Ω , or 0.025Ω for the parallel resultant, and each has about a 120+-25% Ω of AC impedance at 100 MHz, which is half valued for the parallel resultant, with individual maximum DC current carrying capacity of 2Amps.
- Bulk and decoupling capacitors are added, as needed, per power supply design.

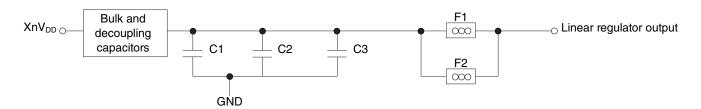


Figure 56. XnV_{DD} power supply filter circuit

Note the following:

- See Power-on ramp rate for maximum XnV_{DD} power-up ramp rate.
- There needs to be enough output capacitance or a soft-start feature to assure ramp rate requirement is met.
- The ferrite beads should be placed in parallel to reduce voltage droop.
- Besides a linear regulator, a low-noise, dedicated switching regulator can be used. 10 mVp-p, 50 kHz 500 MHz is the noise goal.

4.2.6 USB_HV_{DD} and USB_OV_{DD} power supply filtering

USB_HV_{DD} and USB_OV_{DD} must be sourced by a filtered 3.3 V and 1.8 V voltage source using a star connection. An example solution for USB_HV_{DD} and USB_OV_{DD} filtering, where USB_HV_{DD} and USB_OV_{DD} are sourced from a 3.3 V and 1.8 V voltage

source, is illustrated in the following figure. The component values in this example filter is system dependent and are still under characterization, component values may need adjustment based on the system or environment noise.

Where:

- C1 = 0.003 μ F ± 10%, X5R, with ESL ≤ 0.5 nH
- C2 and C3 = 2.2 μ F ± 10%, X5R, with ESL ≤ 0.5 nH
- F1 is an 0603 sized Ferrite SMD, like the Murata part BLM18PG121SH1. Its maximum DC resistance is 0.05Ω and it has about a 120+-25% Ω of AC impedance at 100 MHz with maximum DC current carrying capacity of 2Amps.
- Bulk and decoupling capacitors are added, as needed, per power supply design.

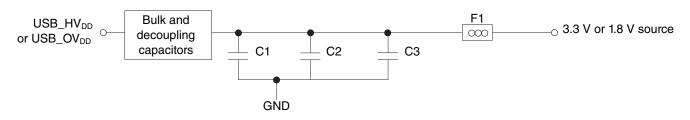


Figure 57. USB_HV_{DD} and USB_OV_{DD} power supply filter circuit

4.2.7 USB_SV_{DD} power supply filtering

 USB_SV_{DD} must be sourced by a filtered V_{DD} using a star connection. An example solution for USB_SV_{DD} filtering, where USB_SV_{DD} is sourced from V_{DD} , is illustrated in the following figure. The component values in this example filter is system dependent and are still under characterization, component values may need adjustment based on the system or environment noise.

Where:

- C1 = 2.2 μ F ± 20%, X5R, with Low ESL (for example, Panasonic ECJ0EB0J225M)
- F1 is an 0603 sized Ferrite SMD, like the Murata part BLM18PG121SH1. Its maximum DC resistance is 0.05Ω and it has about a 120+-25% Ω of AC impedance at 100 MHz with maximum DC current carrying capacity of 2Amps.
- Bulk and decoupling capacitors are added, as needed, per power supply design.

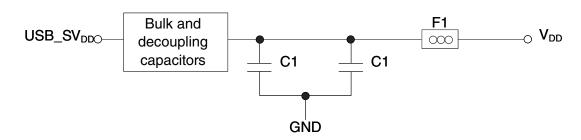


Figure 58. USB_SV_{DD} power supply filter circuit

4.3 Decoupling recommendations

Due to large address and data buses, and high operating frequencies, the device can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the chip system, and the chip itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD} , OV_{DD} , DV_{DD} , GnV_{DD} , and LV_{DD} pin of the device. These decoupling capacitors should receive their power from separate V_{DD} , OV_{DD} , DV_{DD} , GnV_{DD} , LV_{DD} , and GND power planes in the PCB, utilizing short traces to minimize inductance. Capacitors may be placed directly under the device using a standard escape pattern. Others may surround the part.

These capacitors should have a value of 0.1 μ F. Only ceramic SMT (surface mount technology) capacitors should be used to minimize lead inductance, preferably 0402 or 0603 sizes.

As presented in Core and platform supply voltage filtering, it is recommended that there be several bulk storage capacitors distributed around the PCB, feeding the V_{DD} and other planes (for example, OV_{DD} , DV_{DD} , GnV_{DD} , and LV_{DD}), to enable quick recharging of the smaller chip capacitors.

4.4 SerDes block power supply decoupling recommendations

The SerDes block requires a clean, tightly regulated source of power (SnV_{DD}) and XnV_{DD} to ensure low jitter on transmit and reliable recovery of data in the receiver. An appropriate decoupling scheme is outlined below.

NOTE

Only SMT capacitors should be used to minimize inductance. Connections from all capacitors to power and ground should be done with multiple vias to further reduce inductance.

QorlQ T4240 Data Sheet, Rev. 1, 05/2016

- 1. The board should have at least 1 x 0.1-uF SMT ceramic chip capacitor placed as close as possible to each supply ball of the device. Where the board has blind vias, these capacitors should be placed directly below the chip supply and ground connections. Where the board does not have blind vias, these capacitors should be placed in a ring around the device as close to the supply and ground connections as possible.
- 2. Between the device and any SerDes voltage regulator there should be a lower bulk capacitor for example a 10-uF, low ESR SMT tantalum or ceramic and a higher bulk capacitor for example a 100uF 300-uF low ESR SMT tantalum or ceramic capacitor.

4.5 **Connection recommendations**

The following is a list of connection recommendations:

- To ensure reliable operation, it is highly recommended to connect unused inputs to an appropriate signal level. Unless otherwise noted in this document, all unused active low and open drain I/O inputs should be pulled up to V_{DD} , OV_{DD} , DV_{DD} , GnV_{DD} , and LV_{DD} as required. All unused active high inputs should be connected to GND. All NC (no-connect) signals must remain unconnected. Power and ground connections must be made to all external V_{DD} , OV_{DD} , DV_{DD} , LV_{DD} and GND pins of the device.
- The TEST_SEL_B pin must be pulled to OV_{DD} through a 100-ohm to 1k-ohm resistor.
- The chip has temperature diodes that can be used to monitor its temperature by using some external temperature monitoring devices (such as Analog Devices, ADT7481ATM). For more information, see AN4787. The following are the specifications of the chip temperature diodes:
 - Operating range: 10-230 µA
 - Non-ideality factor over temperature range $85C^{\circ}$ to $105C^{\circ}$, n = 1.006 ± 0.003, with approximate error +/- 1 C^o and error under +/- 3 C^o for temperature range 0 C^o to $85C^{\circ}$.

4.5.1 Legacy JTAG configuration signals

Correct operation of the JTAG interface requires configuration of a group of system control pins as demonstrated in Figure 60. Care must be taken to ensure that these pins are maintained at a valid deasserted state under normal operating conditions as most have asynchronous behavior and spurious assertion will give unpredictable results.

Boundary-scan testing is enabled through the JTAG interface signals. The TRST_B signal is optional in the IEEE Std 1149.1 specification, but it is provided on all processors built on Power Architecture technology. The device requires TRST_B to be asserted during power-on reset flow to ensure that the JTAG boundary logic does not interfere with normal chip operation. While the TAP controller can be forced to the reset state using only the TCK and TMS signals, generally systems assert TRST_B during the power-on reset flow. Simply tying TRST_B to PORESET_B is not practical because the JTAG interface is also used for accessing the common on-chip processor (COP), which implements the debug interface to the chip.

The COP function of these processors allow a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert PORESET_B or TRST_B in order to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 60 allows the COP port to independently assert PORESET_B or TRST_B, while ensuring that the target can drive PORESET_B as well.

The COP interface has a standard header, shown in Figure 59, for connection to the target system, and is based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

The COP header adds many benefits such as breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features. An inexpensive option can be to leave the COP header unpopulated until needed.

There is no standardized way to number the COP header; so emulator vendors have issued many different pin numbering schemes. Some COP headers are numbered top-tobottom then left-to-right, while others use left-to-right then top-to-bottom. Still others number the pins counter-clockwise from pin 1 (as with an IC). Regardless of the numbering scheme, the signal placement recommended in Figure 59 is common to all known emulators.

4.5.1.1 Termination of unused signals

If the JTAG interface and COP header will not be used, NXP recommends the following connections:

• TRST_B should be tied to PORESET_B through a $0 k\Omega$ isolation resistor so that it is asserted when the system reset signal (PORESET_B) is asserted, ensuring that the

JTAG scan chain is initialized during the power-on reset flow. NXP recommends that the COP header be designed into the system as shown in Figure 60. If this is not possible, the isolation resistor will allow future access to TRST_B in case a JTAG interface may need to be wired onto the system in future debug situations.

• No pull-up/pull-down is required for TDI, TMS or TDO.

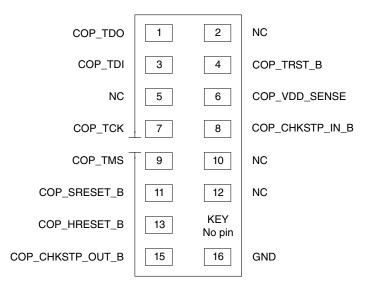
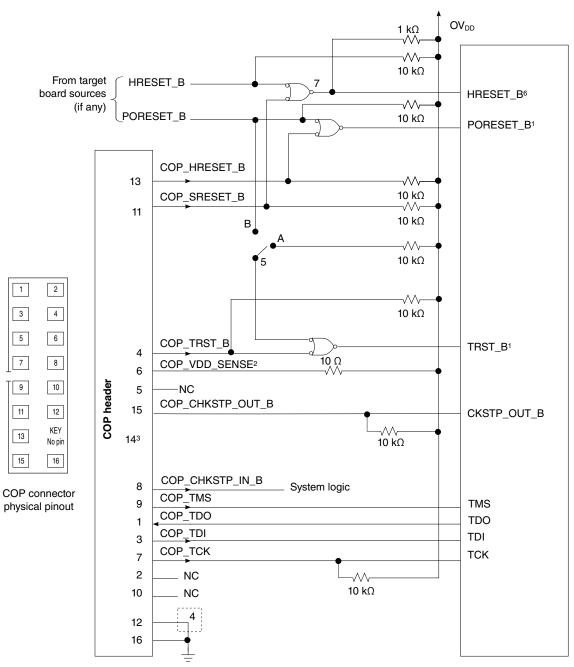


Figure 59. Legacy COP Connector Physical Pinout



Notes:

- 1. The COP port and target board should be able to independently assert PORESET_B and TRST_B to the processor in order to fully control the processor as shown here.
- 2. Populate this with a 10 Ω resistor for short-circuit/current-limiting protection.
- 3. The KEY location (pin 14) is not physically present on the COP header.
- 4. Although pin 12 is defined as a no-connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.
- 5. This switch is included as a precaution for BSDL testing. The switch should be closed to position A during BSDL testing to avoid accidentally asserting the TRST_B line. If BSDL testing is not being performed, this switch should be closed to position B.
- 6. Asserting HRESET_B causes a hard reset on the device
- 7. This is an open-drain output gate.

Figure 60. Legacy JTAG Interface Connection

4.5.2 Aurora configuration signals

Correct operation of the Aurora interface requires configuration of a group of system control pins as demonstrated in the figures below. Care must be taken to ensure that these pins are maintained at a valid deasserted state under normal operating conditions as most have asynchronous behavior and spurious assertion will give unpredictable results.

NXP recommends that the Aurora 34 pin duplex connector be designed into the system as shown in Figure 63 or the 70 pin duplex connector be designed into the system as shown in Figure 64.

If the Aurora interface will not be used, NXP recommends the legacy COP header be designed into the system as described in Termination of unused signals .

Hardware design considerations

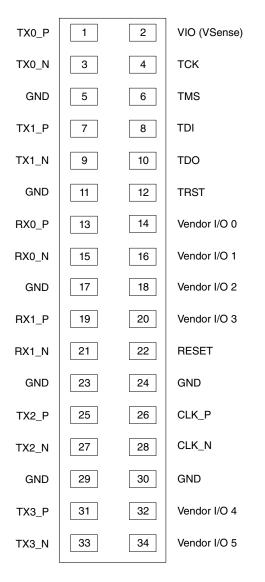
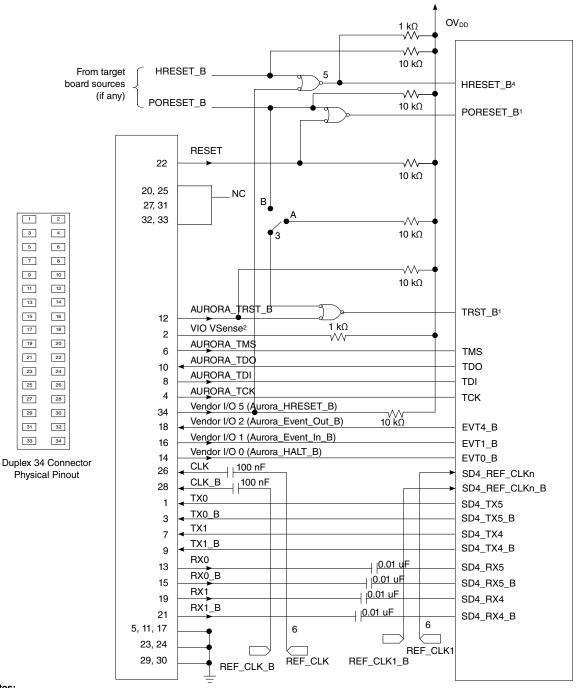


Figure 61. Aurora 34 pin connector duplex pinout

Hardware design considerations

TX0_P	1	2	VIO (VSense)
TX0_N	3	4	тск
GND	5	6	тмз
TX1_P	7	8	TDI
TX1_N	9	10	TDO
GND	11	12	TRST
RX0_P	13	14	Vendor I/O 0
RX0_N	15	16	Vendor I/O 1
GND	17	18	Vendor I/O 2
RX1_P	19	20	Vendor I/O 3
RX1_N	21	22	RESET
GND	23	24	GND
TX2_P	25	26	CLK_P
TX2_N	27	28	CLK_N
GND	29	30	GND
TX3_P	31	32	Vendor I/O 4
TX3_N	33	34	Vendor I/O 5
GND	35	36	GND
RX2_P	37	38	N/C
RX2_N	39	40	N/C
GND	41	42	GND
RX3_P	43	44	N/C
RX3_N	45	46	N/C
GND	47	48	GND
TX4_P	49	50	N/C
TX4_N	51	52	N/C
GND	53	54	GND
TX5_P	55	56	N/C
TX5_N	57	58	N/C
GND	59	60	GND
TX6_P	61	62	N/C
TX6_N	63	64	N/C
GND	65	66	GND
TX7_P	67	68	N/C
TX7_N	69	70	N/C

Figure 62. Aurora 70 pin connector duplex pinout



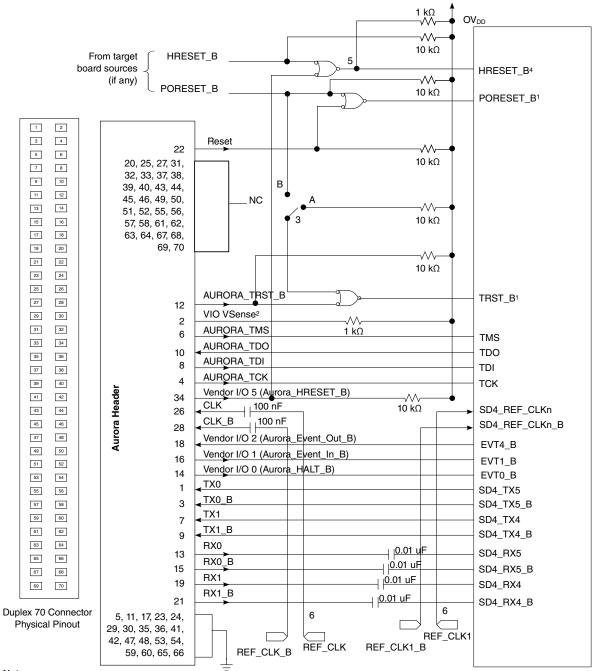
Notes:

1. The Aurora port and target board should be able to independently assert PORESET_B and TRST_B to the processor in order to fully control the processor as shown here.

- 2. Populate this with a 1 k Ω resistor for short-circuit/current-limiting protection.
- 3. This switch is included as a precaution for BSDL testing. The switch should be closed to position A during BSDL testing to avoid accidentally asserting the TRST_B line. If BSDL testing is not being performed, this switch should be closed to position B.
- 4. Asserting HRESET_B causes a hard reset on the device
- 5. This is an open-drain output gate.
- 6. REF_CLK/REF_CLK_B and REF_CLK1/REFCLK1_B are buffered clocks from the same common source.

Figure 63. Aurora 34 pin connector duplex interface connection

Hardware design considerations



Notes:

1. The Aurora port and target board should be able to independently assert PORESET_B and TRST_B to the processor in order to fully control the processor as shown here.

- 2. Populate this with a 1 k Ω resistor for short-circuit/current-limiting protection.
- 3. This switch is included as a precaution for BSDL testing. The switch should be closed to position A during BSDL testing to avoid accidentally asserting the TRST_B line. If BSDL testing is not being performed, this switch should be closed to position B.
- 4. Asserting HRESET_B causes a hard reset on the device
- 5. This is an open-drain output gate.

6. REF_CLK/REF_CLK_B and REF_CLK1/REFCLK1_B are buffered clocks from the same common source.

Figure 64. Aurora 70 pin connector duplex interface connection

4.5.3 Guidelines for high-speed interface termination

4.5.3.1 SerDes interface entirely unused

If the high-speed SerDes interface is not used at all, the unused pin should be terminated as described in this section.

Note that SnV_{DD} , XnV_{DD} and $AVDD_SDn_PLLn$ must remain powered.

For AVDD_SD*n*_PLL*n*, it must be connected to XnV_{DD} through a zero ohm resistor (instead of filter circuit shown in Figure 54).

The following pins must be left unconnected:

- SD*n*_TX[7:0]
- SD*n*_TX[7:0]_B
- SDn_IMP_CAL_RX
- SD*n*_IMP_CAL_TX

The following pins must be connected to SnGND:

- SD*n*_REF_CLK1, SDn_REF_CLK2
- SD*n*_REF_CLK1_B, SDn_REF_CLK2_B

It is recommended for the following pins to be connected to SnGND:

- SD*n*_RX[7:0]
- SD*n*_RX[7:0]_B

It is possible to independently disable each SerDes module by disabling all PLLs associated with it.

SerDes n = 1:4 is disabled as follows:

- SRDS_PLL_PD_S*n* = 2'b11 (both PLLs configured as powered down, all data lanes selected by the protocols defined in SRDS_PRTCL_S*n* associated to the PLLs are powered down as well)
- SRDS_PLL_REF_CLK_SEL_S*n* = 2'b00
- SRDS_PRTCL_S*n* = 2 (no other values permitted when both PLLs are powered down

4.5.3.2 SerDes interface partly unused

If only part of the high speed SerDes interface pins are used, the remaining high-speed serial I/O pins should be terminated as described in this section.

Note that both SnV_{DD} and XnV_{DD} must remain powered.

If any of the PLLs are un-used, the corresponding AVDD_SD*n*_PLL*n* must be connected to XnV_{DD} through a zero ohm resistor (instead of filter circuit shown in Figure 54).

The following unused pins must be left unconnected:

- SDn_TX[*n*]
- SDn_TX[*n*]_B

The following unused pins must be connected to SnGND:

- SD1_REF_CLK[1:2], SD1_REF_CLK[1:2]_B (If entire SerDes 1 unused)
- SD2_REF_CLK[1:2], SD2_REF_CLK[1:2]_B (If entire SerDes 2 unused)
- SD3_REF_CLK[1:2], SD3_REF_CLK[1:2]_B (If entire SerDes 3 unused)
- SD4_REF_CLK[1:2], SD4_REF_CLK[1:2]_B (If entire SerDes 4 unused)

It is recommended for the following unused pins to be connected to SnGND:

- SDn_RX[*n*]
- $SDn_RX[n]_B$

In the RCW configuration field SRDS_PLL_PD_S*n*, the respective bits for each unused PLL must be set to power it down. A module is disabled when both its PLLs are turned off.

Unused lanes must be powered down through the SRDSx Lane m General Control Register 0 (SRDSxLNmGCR0) as follows:

- SRDSxLNmGCR0[RRST] = 0
- SRDSxLNmGCR0[TRST] = 0
- SRDSxLNmGCR0[RX_PD] = 1
- SRDSxLNmGCR0[TX_PD] = 1

Note that in the case where the SerDes pins are connected to slots, it is acceptable to have these pins unterminated when unused.

4.5.4 USB controller connections

This section details the hardware connections required for the USB controllers.

4.5.4.1 USB divider network

This figure shows the required divider network for the VBUS interface for the chip. Additional requirements for the external components are:

- Both resistors require 1% accuracy and a current capability of up to 1 mA. They must both have the same temperature coefficient and accuracy.
- The zener diode must have a value of 5 V-5.25 V.
- The 0.6 V diode requires an $I_F = 10 \text{ mA}$, $I_R < 500 \text{ nA}$ and $V_{F(Max)} = 0.8 \text{ V}$. If the USB PHY does not support OTG mode, this diode can be removed from the schematic or made a DNP component.

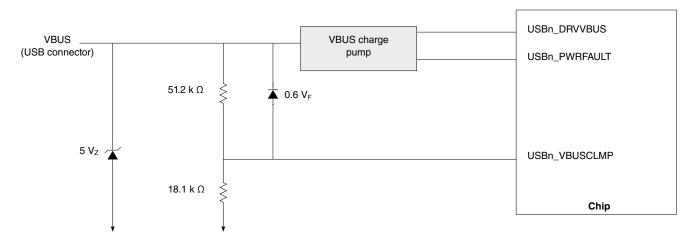


Figure 65. Divider network at VBUS

4.6 Thermal

This table shows the thermal characteristics for the chip. Note that these numbers are based on design estimates and are preliminary.

Table 148.	Package ther	mal characteristics ⁶
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Rating	Board	Symbol	Value	Unit	Notes
Junction to ambient, natural convection	Single-layer board (1s)	$R_{\Theta JA}$	11	°C/W	1, 2
Junction to ambient, natural convection	Four-layer board (2s2p)	R _{OJA}	9	°C/W	1, 3
Junction to ambient (at 200 ft./min.)	Single-layer board (1s)	R _{OJMA}	8	°C/W	1, 2
Junction to ambient (at 200 ft./min.)	Four-layer board (2s2p)	R _{OJMA}	6	°C/W	1, 2
Junction to board	-	R _{OJB}	3	°C/W	3
Junction to case top	-	R _{OJCtop}	0.3	°C/W	4
Junction to lid top	-	R _{ØJClid}	0.11	°C/W	5

Table continues on the next page...

Table 148. Package thermal characteristics⁶ (continued)

Rating	Board	Symbol	Value	Unit	Notes		
1. Junction temperature is a function of die size, on-chip (board) temperature, ambient temperature, air flow, pow resistance.		-		-			
2. Per JEDEC JESD51-3 and JESD51-6 with the board (JESD51-9) horizontal.							
3. Thermal resistance between the die and the printed-circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.							
4. Junction-to-case-top at the top of the package determined using MIL-STD 883 Method 1012.1. The cold plate temperature is used for the case temperature. Reported value includes the thermal resistance of the interface layer.							
5. Junction-to-lid-top thermal resistance determined usir cold plate, the lid top temperature is used here for the re thermal resistance of the interface layer between the pa	eference case temperature						

6. See Thermal management information, for additional details.

4.7 Recommended thermal model

Information about Flotherm models of the package or thermal data not available in this document can be obtained from your local NXP sales office.

4.8 Thermal management information

This section provides thermal management information for the flip-chip, plastic-ball, grid array (FC-PBGA) package for air-cooled applications. Proper thermal control design is primarily dependent on the system-level design-the heat sink, airflow, and thermal interface material.

The recommended attachment method to the heat sink is illustrated in Figure 66. The heat sink should be attached to the printed-circuit board with the spring force centered over the die. This spring force should not exceed 60 pounds force (270 Newtons).

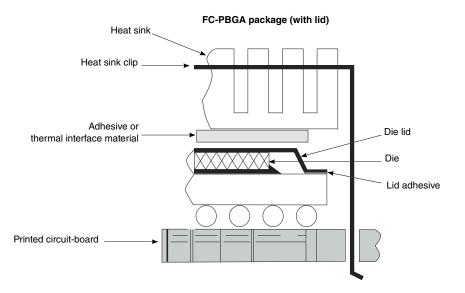


Figure 66. Package exploded, cross-sectional view-FC-PBGA (with lid)

The system board designer can choose between several types of heat sinks to place on the device. There are several commercially-available thermal interfaces to choose from in the industry. Ultimately, the final selection of an appropriate heat sink depends on many factors, such as thermal performance at a given air velocity, spatial volume, mass, attachment method, assembly, and cost.

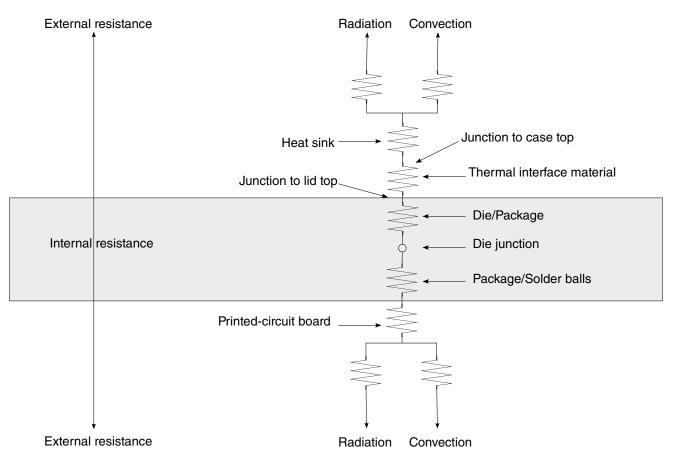
4.8.1 Internal package conduction resistance

For the package, the intrinsic internal conduction thermal resistance paths are as follows:

- The die junction-to-case thermal resistance
- The die junction-to-lid-top thermal resistance
- The die junction-to-board thermal resistance

This figure depicts the primary heat transfer path for a package with an attached heat sink mounted to a printed-circuit board.

Hardware design considerations



(Note the internal versus external package resistance)

Figure 67. Package with heat sink mounted to a printed-circuit board

The heat sink removes most of the heat from the device. Heat generated on the active side of the chip is conducted through the silicon and through the heat sink attach material (or thermal interface material), and finally to the heat sink. The junction-to-case thermal resistance is low enough that the heat sink attach material and heat sink thermal resistance are the dominant terms.

4.8.2 Thermal interface materials

A thermal interface material is required at the package-to-heat sink interface to minimize the thermal contact resistance. The performance of thermal interface materials improves with increasing contact pressure; this performance characteristic chart is generally provided by the thermal interface vendor. The recommended method of mounting heat sinks on the package is by means of a spring clip attachment to the printed-circuit board (see Figure 66).

The system board designer can choose among several types of commercially-available thermal interface materials.

5 Package information

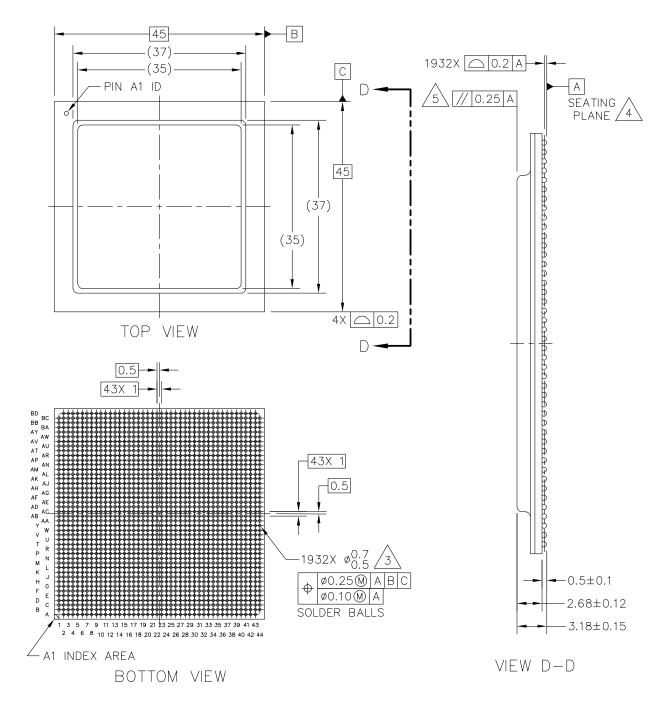
5.1 Package parameters for the FC-PBGA

The package parameters are as provided in the following list. The package type is 45 mm x 45 mm, 1932 flip-chip, plastic-ball, grid array (FC-PBGA).

- Package outline 45 mm x 45 mm
- Interconnects 1932
- Ball Pitch 1.0 mm
- Ball Diameter (typical) 0.60 mm
- Solder Balls 96.5% Sn, 3% Ag, 0.5% Cu
- Module height (typical) 3.03 mm to 3.33 mm (maximum)

5.2 Mechanical dimensions of the FC-PBGA

This figure shows the mechanical dimensions and bottom surface nomenclature of the chip.



© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.	MECHANICAL OUTL	LINE	PRINT VERSION NO	T TO SCALE
TITLE: FCPBGA, 1932 I	/O, [DOCUMEN	NT NO: 98ASA00384D	REV: A
45 X 45 X 3.18	PKG, c	CASE NU	IMBER: 2231–02	12 APR 2012
1 MM PITCH, WITH	H LID s	STANDAR	D: NON-JEDEC	

Figure 68. Mechanical dimensions of the FC-PBGA with full lid QorlQ T4240 Data Sheet, Rev. 1, 05/2016

NOTES:

- 1. All dimensions are in millimeters.
- 2. Dimensions and tolerances per ASME Y14.5M-1994.
- 3. Maximum solder ball diameter measured parallel to datum A.
- 4. Datum A, the seating plane, is determined by the spherical crowns of the solder balls.
- 5. Parallelism measurement shall exclude any effect of mark on top surface of package.

6 Security fuse processor

This chip implements the QorIQ platform's Trust Architecture, supporting capabilities such as secure boot. Use of the Trust Architecture features is dependent on programming fuses in the Security Fuse Processor (SFP). The details of the Trust Architecture and SFP can be found in the chip reference manual.

To program SFP fuses, the user is required to supply 1.8 V to the PROG_SFP pin per Power sequencing. PROG_SFP should only be powered for the duration of the fuse programming cycle, with a per device limit of two fuse programming cycles. All other times PROG_SFP should be connected to GND. The sequencing requirements for raising and lowering PROG_SFP are shown in Figure 8. To ensure device reliability, fuse programming must be performed within the recommended fuse programming temperature range per Table 3.

NOTE

Users not implementing the QorIQ platform's Trust Architecture features should connect PROG_SFP to GND.

7 Ordering information

Contact your local NXP sales office or regional marketing team for order information.

7.1 Part numbering nomenclature

This table provides the NXP QorIQ platform part numbering nomenclature.

Ordering information

pt or t	n	nn	n	x	t	е	n	с	d	r
Generation	Platform	Number of virtual cores	Derivative	Qual status	Temperature range	Encryption	Package type	CPU speed	DDR data rate	Die revision
PT = 28 nm (Prototype) T = 28 nm (Production)	4	24 = 24 virtual cores $16 = 16$ virtual cores $08 = 8$ virtual cores	0 = Standard power 1 = Low power	P = Prototype N = Qualified to industrial tier	S = Standard temp X = Extended temp	E = SEC present N = SEC not present	7 = FC- PBGA C4/C5 Pb-free	P = 1500 MHz Q = 1667 MHz T = 1800 MHz	Q = 1600 MT/s T= 1866 MT/s Z= TBD	A = Rev 1.0 B = Rev 2.0

 Table 149.
 Part numbering nomenclature

7.2 Orderable part numbers addressed by this document

This table provides the NXP orderable part numbers addressed by this document for the chip.

Part number	pt or t	n	nn	n	x	t	е	n	с	d	r
(Freq-Core/Freq- Platform/Freq-DDR (MT/s))											
T4240NSE7PQB	T=28 nm	4	24=24 virtual cores	0	N=Qualifie d	S=Std temp	E=SEC present	7	P=1500 M Hz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4240NSE7QTB	T=28 nm	4	24=24 virtual cores	0	N=Qualifie d	S=Std temp	E=SEC present	7	Q=1667 M Hz CPU	T=1866 M T/s DDR	B = Rev 2.0
T4240NSN7PQB	T=28 nm	4	24=24 virtual cores	0	N=Qualifie d	S=Std temp	N=No SEC present	7	P=1500 M Hz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4240NSN7QTB	T=28 nm	4	24=24 virtual cores	0	N=Qualifie d	S=Std temp	N=No SEC present	7	Q=1667 M Hz CPU	T=1866 M T/s DDR	B = Rev 2.0

Table 150. Orderable part numbers addressed by this document

Table continues on the next page...

Table 150.	Orderable part numbers addressed by this document (continued)	
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Part number	pt or t	n	nn	n	x	t	е	n	с	d	r
(Freq-Core/Freq- Platform/Freq-DDR (MT/s))											
T4240NSE7TTB	T = 28 nm	4	24=24 virtual cores	0	N=Qualifie d	S=Std temp	E= SEC present	7	T=1800 MHz CPU	T=1866 MT/S DDR	B = Rev 2.0
T4240NSN7TTB	T = 28 nm	4	24=24 virtual cores	0	N=Qualifie d	S=Std temp	N = No SEC present	7	T=1800 MHz CPU	T=1866 MT/S DDR	B = Rev 2.0
T4240NXE7PQB	T=28 nm	4	24=24 virtual cores	0	N=Qualifie d	X=extende d temp	E=SEC present	7	P=1500 M Hz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4240NXN7PQB	T=28 nm	4	24=24 virtual cores	0	N=Qualifie d	X=extende d temp	N=No SEC present	7	P=1500 M Hz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4241NSE7PQB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	S= Std temp	E=SEC present	7	P=1500 MHz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4241NSE7QTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	S= Std temp	E=SEC present	7	Q=1667 M Hz CPU	T=1866 MT/s DDR	B = Rev 2.0
T4241NSE7TTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	S= Std temp	E=SEC present	7	T=1800 MHz CPU	T=1866 MT/s DDR	B = Rev 2.0
T4241NSN7PQB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	S= Std temp	N = SEC not present	7	P=1500 MHz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4241NSN7QTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	S= Std temp	N = SEC not present	7	Q=1667 M Hz CPU	T=1866 MT/s DDR	B = Rev 2.0
T4241NSN7TTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	S= Std temp	N = SEC not present	7	T=1800 MHz CPU	T=1866 MT/s DDR	B = Rev 2.0
T4241NXE7PQB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	X=extende d temp	E=SEC present	7	P=1500 MHz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4241NXE7QTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	X=extende d temp	E=SEC present	7	Q=1667 M Hz CPU	T=1866 MT/s DDR	B = Rev 2.0
T4241NXE7TTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	X=extende d temp	E=SEC present	7	T=1800 MHz CPU	T=1866 MT/s DDR	B = Rev 2.0
T4241NXN7PQB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	X=extende d temp	N = SEC not present	7	P= 1500MHz CPU	Q=1600 M T/s DDR	B = Rev 2.0
T4241NXN7QTB	T=28 nm	4	24=24 virtual cores	1	N=Qualifie d	X=extende d temp	N = SEC not present	7	Q=1667 M Hz CPU	T=1866 MT/s DDR	B = Rev 2.0

Table continues on the next page...

Part number	pt or t	n	nn	n	x	t	е	n	с	d	r
(Freq-Core/Freq- Platform/Freq-DDR (MT/s))											
T4241NXN7TTB	T=28 nm		24=24 virtual cores	1	N=Qualifie d	X=extende d temp	N = SEC not present	7	T=1800 MHz CPU	T=1866 MT/s DDR	B = Rev 2.0

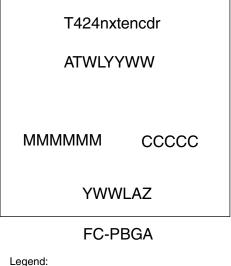
Notes:

1. The 1866 MT/s DDR rate is associated with 733 MHz platform frequency in high speed parts while the 1600 MT/s DDR rate is associated with 667 MHz platform clock frequency in lower speed parts in the part number encoding.

2. The T4241/T4161/T4081 are identical to T4240/T4160/T4080 except they consume less power. See the power requirements in Power characteristics.

7.2.1 Part marking

Parts are marked as in the example shown in this figure.



T424nxtencdr is the orderable part number. ATWLYYWW is the test traceability code. MMMMMM is the mask number. CCCCC is the country code. YWWLAZ is the assembly traceability code.

Figure 69. Part marking for T424n FC-PBGA chip

8 Revision history

This table summarizes revisions to this document.

Table 151.	Revision history
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Revision	Date	Description
1	05/2016	Updated the document title to conform with new naming requirements.
		Rebranded to NXP company name.
		 Removed all references to and sections for 10.3215G Interlaken.
		 Updated the speed units throughout the document.
		In the pinout list table:
		Modified notes 6, 8, 23, and 24.
		Added note 29.
		Added note 28 to the D3_MDMn pins.
		Changed the note reference on IFC_AD16 to 29.
		 Updated note 29 so the pull down resistance is 4.7 K instead of 10 to 50 kΩ.
		 In Table 2 : Changed format to group "Supply Voltage Levels," "Storage Temperature Conditions,"
		and "Signal Voltage Levels."
		 Reduced Maximum VDD, SnVDD supply voltage level from 1.1 V to 1.08 V.
		 Reduced Maximum VDD, SIVDD supply voltage level from 1.17 v to 1.06 v. Reduced the max GnVDD I/O voltage levels from 1.65 to 1.58 (DDR3) and from 1.45 to
		1.42 (DDR3L).
		 Increased the storage temperature range max value from 150 to 155.
		 Added "Min_DCV V_input," "Max_DCV V_input," and "Max Overshoot Voltage"
		columns for Signal Voltage level signals.
		In the SerDes signals, added additional rows for "No internal termination selected" and
		"50 ohm internal termination selected".
		 Added the LP Trust signal LP_TMP_DETECT_B.
		 Renamed "USBn_V_{IN}_3P3" and "USBn_V_{IN}_1P8" in note 5 and stressed the max slew
		rate of Dn_MVREF to 25 kv/s.
		 Updated note 9 to include "See also note 6 in Table 3".
		 Updated note 10 to include required biasing.
		Added notes 11, 12, and 13.
		• In Table 3, added table note 11 and added the LP Trust signal.
		Updated Figure 7.
		 In Power sequencing, added a paragraph for V_{DD}_LP special power sequencing. Also relaxed
		power lines stability time from 75 ms to 400 ms.
		In Table 6 : Added 1 8 GHz newer numbers
		Added 1.8 GHz power numbers.Updated note 9.
		In Table 8 :
		Added 1.8 GHz power numbers.
		Changed the 1667 MHz freq DDR data rate from 1867 to 1866.
		Updated note 9.
		Added Table 7 and Table 9 for the T4241 low-power device.
		In Table 10 :
		Added the 1.8 GHz frequency.
		 Added the T4240/T4160 and T4080 LPM20 data.
		 Updated power numbers so they are relevant to 65°C.
		 Added a note saying that these numbers are good for the T4241 device.
		• In Table 11 :
		 Reduced the 1600 MT/s GV_{DD} typical and max values from 3150 to 3100 and 4920 to
		4900, respectively.
		 Improved the PLL_SerDes typical value from 40 to 60. Added a family family (a part Q) (a part of Q) (b part of Q) (b part of Q) (b part of Q) (c part of Q) (c
		 Added a formula for XV_{DD} and SV_{DD} typical power estimation rather than have multiple
		rows showing different SerDes configuration power.
		 Removed the note: "Maximum DDR power numbers are based on one 2-rank DIMM with 100% utilization " (provinuely note 5) and renumbered the notes.
		with 100% utilization," (previously note 5) and renumbered the notes.
		Updated note 6 and added example.Added low power devices to the table title.
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Table continues on the next page...

Revision	Date	Description
		In Table 13, included RTC clock DC specifications and updated the input capacitance data for
		both SYSCLK and RTC clock pins.
		 In Table 14, changed the maximum SYSCLK AC swing value from "TBD" to "1 x OV_{DD}" and updated note 6.
		 In "SYSCLK and RTC AC timing specifications," added Table 15, "RTC AC timing
		specification."
		In Real-time clock (RTC) timing, removed the 50% duty cycle requirement from the RTC
		period minimum.
		In Table 19, changed the DDRCLK input pin capacitance typical value from 7 to 11 and
		removed the max value.
		 In Table 20 : Changed the minimum DDRCLK cycle time from 5 ns to 7.5 ns.
		 Changed the maximum DDRCLK AC swing value from "TBD" to "1 x OV_{DD}".
		Updated note 6.
		 In Table 21, relaxed HRESET_B signal rise/fall time to 4 SYSCLK cycles.
		In Table 29, updated the SPI_MOSI hold time min and SPI_MOSI delay max formulas.
		Updated the t _{MDKHDX} delay equation presentation in Table 38 and Table 39. (When
		MDIO_CFG[NEG] = 0 then Y = 0.5 and t_{MDKHDX} is Y x $T_{MDC_CIK} \pm 3$ ns.)
		 In Table 42 : Updated the TSEC_1588_CLK_IN clock period min value and removed the max value.
		 Updated the TSEC_1588_CLK_OUT clock period min value.
		 Added "hold time" to TSEC_1588_ALARM_OUT1/2.
		 Changed the TSEC_1588_TRIG_IN1/2 pulse width min value.
		 Removed notes 4 and 5 and updated notes 1 and 2.
		Updated all note references.
		• In Table 46, changed VOH/VOL min and max from (0.8 x OVDD, 0.4) to (1.6 V, 0.32 V).
		 In Table 55, changed IOL at 1.8 V from 1 mA to 3 mA. In "GPIO DC electrical characteristics," added Table 59, "LP_TMP_DETECT_B pin DC
		electrical characteristics."
		 In Table 76, Table 84, and Table 111, removed the absolute output voltage limits (min -0.4 V,
		max 2.30 V).
		• In Table 105, changed the figure reference in note 2 from Figure 43 to Figure 42.
		In Table 122, added note 2 and 3 and updated all note references.
		 In Table 127, added the 1800 MHz data columns and added note 8 to describe why FMAN might have two different minimum for successing.
		 might have two different minimum frequencies. In Table 133, changed the 133.333 MHz DDRCLK frequency example value to 1866.667.
		 In Table 137, added the 1800 MHz Core cluster: SYSCLK Ratio options.
		 In Minimum platform frequency requirements for high-speed interfaces, removed the SRIO
		equation that shows minimum platform frequency.
		• In Table 147, changed the VDD voltage note for "All other values". After the table, added
		paragraph for if DA_ALT_V is not all zeroes.
		In PLL power supply filtering, updated the second NOTE.
		 Updated Figure 53 and Figure 54. In Connection recommendations, added the expected temperature error to the non-ideality
		factor temperature range.
		 Updated Mechanical dimensions of the FC-PBGA to include package parameters.
		• In Table 149 :
		Added 16 and 08 cores to column nn.
		 Changed column n to "0 = Standard power; 1 = Low power".
		• Added symbol "T = 1800 MHz" to column C.
		In Orderable part numbers addressed by this document, added two new orderable 1.8 G parts and added T4241 part numbers
		and added T4241 part numbers.Updated Part marking to include low power numbers.
-		
0	07/2014	Initial release

 Table 151.
 Revision history (continued)

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