

## **General Description**

The OB2101H is a monolithic single-phase half-bridge gate driver IC designed for high voltage, high speed, driving power MOSFET and IGBT operating up to 650V.

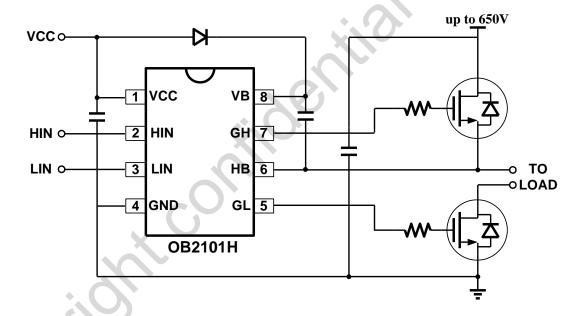
The OB2101H uses high voltage process and common mode noise canceling technique provides stable operation of high-side drivers under high dv/dt noise circumstance, and two output channels with internal deadtime to avoid cross-conduction.

The input logic level is compatible with standard 3.3V/5V. Output driver source and sink current 260mA and 530mA.

## **Features**

- Floating channel for bootstrap operation up to 650V
- Positive input logic, and 3.3V/5V input logic compatible
- Built-in low-side supply under voltage lockout (UVLO)
- Built-in high side supply under voltage lockout (UVLO)
- Built-in cross conduction prevention logic
- Built-in dead time and matched propagation delay
- Available in SOP8 package

# **Typical Application**

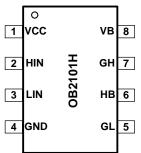




## **GENERAL INFORMATION**

## **Pin Configuration**

The pin map is shown as below for SOP8



## **Ordering Information**

	Part Number	Description
Ī	OB2101HCP	SOP8, Halogen-free, Tube
	OB2101HCPA	SOP8, Halogen-free, T&R

## **Package Dissipation Rating**

Package	RθJA (℃/W)
SOP8	90

Note: Drain Pin Connected 100mm<sup>2</sup> PCB copper clad.

**Absolute Maximum Ratings** 

Symbol	Description	Min	Max	Units
VB	High side floating supply voltage	-0.3	650	
$V_{HB}$	High side floating offset voltage	VB-25	VB+0.3	
$V_{GH}$	High side floating output voltage	V <sub>HB</sub> -0.3	VB+0.3	V
VCC	Low side and supply voltage	-0.3	20	V
$V_{GL}$	Low side gate driver output	-0.3	VCC+0.3	
V <sub>IN</sub>	Logic input voltage(HIN & LIN)	-0.3	VCC+0.3	
dV <sub>HB</sub> /dt	Allowable offset voltage transient		50	V/ns
$P_{D}$	Package power dissipation@T <sub>A</sub> ≤+25°C		0.6	W
$T_J$	Junction temerature	-40	150	°C
Ts	Storage temerature	-40	125	

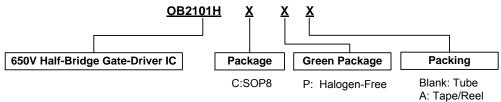
**Note:** Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.

## **Recommended Operating Conditions**

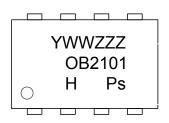
For proper operation the device should be used within the recommended conditions.

Symbol	Description	Min	Max	Units
VB	High side floating supply voltage	V <sub>HB</sub> +10	V <sub>HB</sub> +20	
$V_{HB}$	High side floating offset voltage		600	
$V_{GH}$	High side floating output voltage	$V_{HB}$	VB	\/
VCC	Low side and supply voltage	10	15	V
$V_{GL}$	Low side gate driver output	0	VCC	
$V_{IN}$	Logic input voltage(HIN & LIN)	0	VCC	
T <sub>A</sub>	Ambient temerature	-40	125	°C

## **Marking Information**







Y:Year Code WW:Week Code(01-52) ZZZ:Lot Code H:Character Code P:Green Package(Halogen-free) S:Internal Code(Optional)

## **PIN Definitions**

Symbol	Description
VCC	Low side supply voltage
HIN	Logic input for high side gate driver output(GH),in phase
LIN	Logic input for low side gate driver output(GL),in phase
GND	Low side ground
GL	Low side gate driver output
HB	High side floating supply return
GH	High side gate driver output
VB	High side floating supply

## **Dynamic Electrical Characteristics**

Setup: VCC=VB=12V, GND=HB=0V and  $T_A$ =25°C unless otherwise specified.

Symbol	Description	Test Conditions	Min	Тур	Max	Units
GH_t <sub>on</sub>	GH turn-on propagation delay		200	310	550	
GH_t <sub>off</sub>	GH turn-off propagation delay		200	310	550	
GL_t <sub>on</sub>	GL turn-on propagation delay		200	350	550	
GL_t <sub>off</sub>	GL turn-off propagation delay		200	350	550	ns
t <sub>r</sub>	Turn-on rise time	Cload=1nF		95		
t <sub>f</sub>	Turn-off fall time	Cloau-IIIF		45		
MT	Delay matching HS & LS turn-on/off			40		



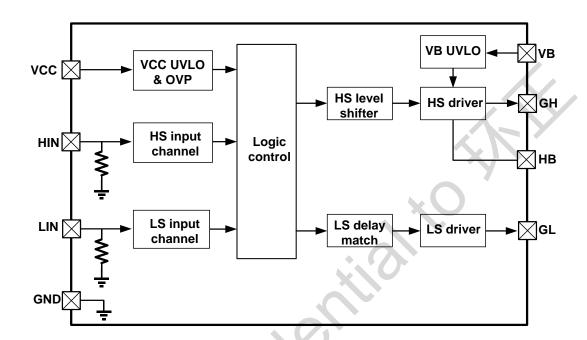
## **Static Electrical Characteristics**

Setup: VCC=VB=12V, Cload=10nF and  $T_A$ =25°C unless otherwise specified.

Symbol	Description	Test Conditions	Min	Тур	Max	Units
$V_{IH}$	Logic"1"input voltage	VCC=12V	2			V
$V_{IL}$	Logic"0"input voltage	VCC-12V			8.0	V
I <sub>IN+</sub>	Logic"1"input bias current	V <sub>IN</sub> =5V		50	150	
I <sub>IN-</sub>	Logic"0"input bias current	V <sub>IN</sub> =0V			1	
$I_{LK}$	Offset voltage leakage current	VB=HB=600V			1	uΑ
$I_{QBS}$	Quiescent VBS supply current	HIN=LIN=0V		110	200	
I <sub>QCC</sub>	Quiescent VCC supply current	HIN=LIN=0V		65	120	
V <sub>CCUV+</sub>	VCC supply under voltage positive going threshold		7.3	8.3	9.3	
V <sub>CCUV-</sub> VCC supply under voltage negative going threshold			7.8	8.8	9.8	V
$V_{BSUV+}$	V <sub>BSUV+</sub> VBS supply under voltage positive going threshold		6.1	7.1	8.1	V
$V_{BSUV}$	VBS supply under voltage negative going threshold		6.5	7.5	8.5	
I <sub>SOURCE_GH</sub>	Source current of GH driver	GH=15V, HIN=5V, with PW≤10us,	160	275		
I <sub>SINK_GH</sub>	I <sub>SINK_GH</sub> Sink current of GH driver		320	530		mA
I <sub>SOURCE_GL</sub>	I <sub>SOURCE_GL</sub> Source current of GL driver		150	260		IIIA
I <sub>SINK_GL</sub>	Sink current of GL driver	GH=0V, HIN=0V, with PW≤10us,	350	590		



## **Functional Block Diagram**





## **Typical Performance Chart**

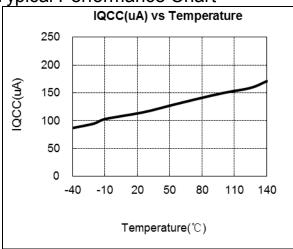
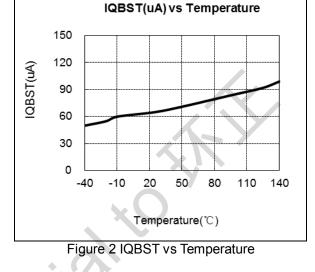


Figure 1 IQCC vs Temperature



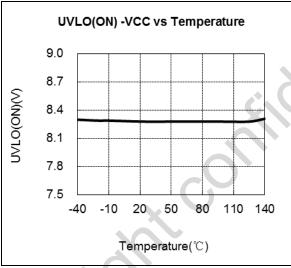


Figure 3 VCC UVLO(ON) vs Temperature

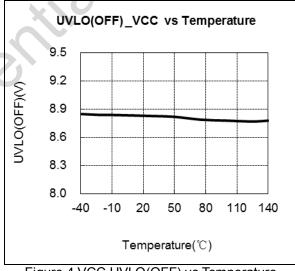


Figure 4 VCC UVLO(OFF) vs Temperature

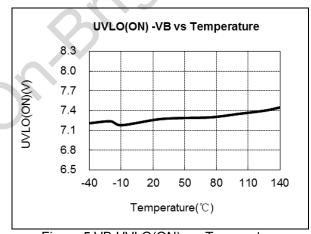


Figure 5 VB UVLO(ON) vs Temperature

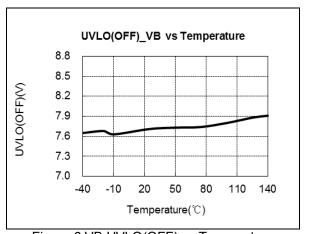


Figure 6 VB UVLO(OFF) vs Temperature



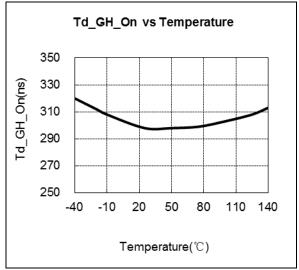


Figure 7 GH t<sub>ON</sub> vs Temperature

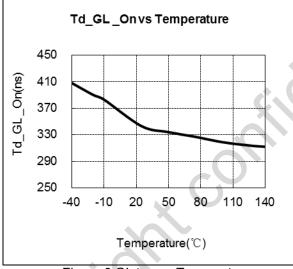


Figure 9 GL ton vs Temperature

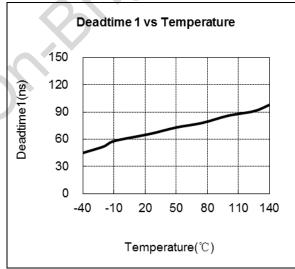


Figure 11 GLf to GHr Deadtime vs Temperature

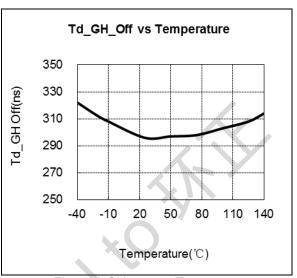


Figure 8 GH t<sub>OFF</sub> vs Temperature

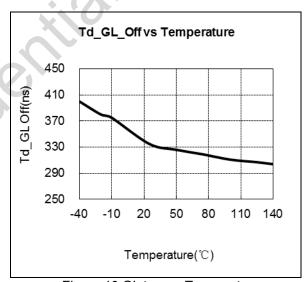


Figure 10 GL t<sub>OFF</sub> vs Temperature

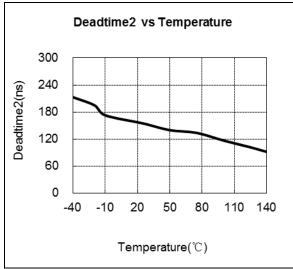


Figure 12 GHf to GLr Deadtime vs Temperature



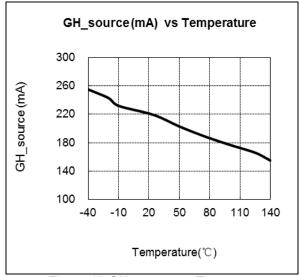


Figure 13 GH\_source vs Temperature

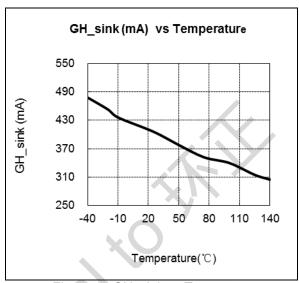


Figure 14 GH\_sink vs Temperature

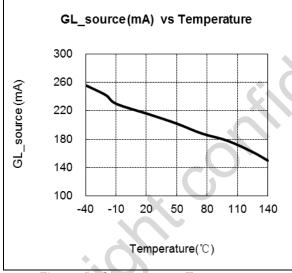


Figure 15 GL\_source vs Temperature

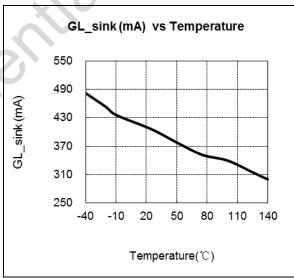
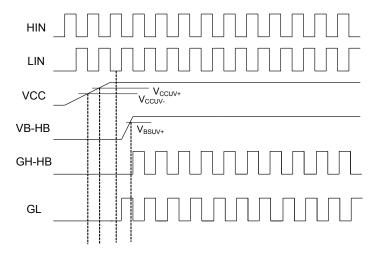


Figure 16 GL\_sink vs Temperature

# **Timing Diagram**





## **Operational Description**

## **Power Supply**

When VCC voltage increases above 8.8V (typical), OB2101H starts to work, open the gate drivers. When VCC voltage drops below 8.3V (typical), OB2101H shuts down the gate drivers. OB2101H will not resume working until the VCC voltage increases above 8.8V (typical).

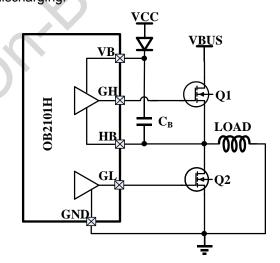
#### **Low-Side Driver**

The low-side driver is designed to drive a ground referenced N-channel MOSFET. Its low Rdson allows the external MOSFET to be turned on and off quickly. When a low-side driver is on, VCC voltage is applied to the gate of the external MOSFET.

## **High-Side Driver**

The high-side driver is designed to drive a floating N-channel MOSFET, whose source terminal is referenced to the HB pin. A low-power, high-speed, level-shifting circuit isolates the low-side referenced circuitry from the high-side referenced driver. Power to the high-side driver and UVLO circuit is supplied by the bootstrap circuit.

The bootstrap circuit consists of the diode and capacitor ( $C_B$ ). In a typical application, the HB pin is at ground potential when the low-side MOSFET is on. The bootstrap diode allows capacitor  $C_B$  to be charged up to VCC-V<sub>F</sub> during this time (where V<sub>F</sub> is the forward voltage drop of the bootstrap diode). After the low-side MOSFET is turned off and the GH pin turns on, the voltage across capacitor  $C_B$  is applied to the gate of the upper external MOSFET. As the upper MOSFET turns on, voltage on the HB pin rises with the source of the high-side MOSFET until it reaches  $V_{BUS}$ . As the VB and HB pin rise, the bootstrap diode is reverse biased preventing capacitor  $C_B$  from discharging.

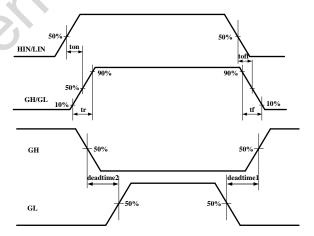


## **Propagation Delay and Dead-time**

Propagation delay and dead-time are important considerations, as shown below. BLDC controllers use two switching MOSFETs operating complementarily. These MOSFETs must not be on at the same time or a short circuit will occur, causing high peak currents and higher power dissipation in the MOSFETs.

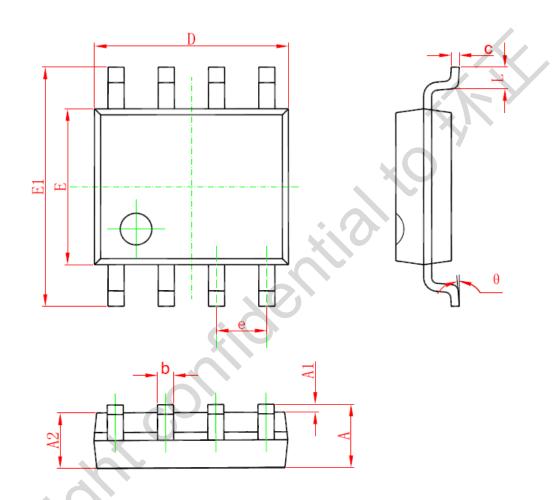
Make sure the input signal pulse width is greater than the minimum specified pulse width. An input signal that is less than the minimum pulse width results in no output pulse or an output pulse whose width is significantly less than the input.

The maximum duty cycle (ratio of high side ontime to switching period) is controlled by the minimum pulse width of the low side and by the time required for the  $C_{\rm B}$  capacitor to charge during the off-time. Adequate time must be allowed for  $C_{\rm B}$  capacitor to charge up before the high-side driver is turned on.





# PACKAGE MECHANICAL DATA SOP8 PACKAGE OUTLINE DIMENSIONS



Cumbal	Dimensions In Millimeters		Dimensions In Inches		
Symbol	Min	Max	Min	Min	
Α	1.350	1.750	0.053	0.069	
A1	0.050	0.250	0.002	0.010	
A2	1.250	1.650	0.049	0.065	
b	0.310	0.510	0.012	0.020	
С	0.100	0.250	0.004	0.010	
D	4.700	5.150	0.185	0.203	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
е	1.270	(BSC)	0.050	(BSC)	
L	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	



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