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[^1]
## FAN5345

## Series Boost LED Driver with Single－Wire Digital Interface

## Features

－Asynchronous Boost Converter
－Drives LEDs in Series：
－FAN5345S20X：20V Output
－FAN5345S30X：30V Output
－ 2.5 V to 5.5 V Input Voltage Range
－Single－Wire Digital Control Interface to Set LED Brightness Levels
－ 32 Linear Steps
－ 1.2 MHz Fixed Switching Frequency
－Soft－Start Capability
－Input Under－Voltage Lockout（UVLO）
－Output Over－Voltage Protection（OVP）
－Short－Circuit Detection
－Thermal Shutdown（TSD）Protection
－Small Form－Factor 6－Lead SSOT23 Package

## Applications

－Cellular Mobile Handsets
－Mobile Internet Devices
－Portable Media Players
－PDA，DSC，MP3 Players

## Description

The FAN5345 is an asynchronous constant－current LED driver that drives LEDs in series to ensure equal brightness for all the LEDs．FAN5345S20X has an output voltage of 20 V and can drive up to 5 LEDs in series．FAN5345S30X has an output voltage of 30 V and drive up to 8 LEDs in series．Optimized for small form－factor applications，the 1．2 MHz fixed switching frequency allows the use of small inductors and capacitors．

The FAN5345 uses a simple single－wire digital control interface to program the brightness levels of the LEDs in 32 linear steps by applying digital pulses．
For safety，the device features integrated over－voltage，over－ current，short－circuit detection，and thermal－shutdown protection．In addition，input under－voltage lockout protection is triggered if the battery voltage is too low．

The FAN5345 is available in a 6－lead SSOT23 package． It is＂green＂and RoHS compliant．（Please see http：／／uww．fairchildsemi．com／company／green／index．html for Fairchild＇s definition of green）．

## Ordering Information

| Part Number | Output Voltage Option | Temperature Range | Package |
| :---: | :---: | :---: | :--- |
| FAN5345S20X | 20 V | -40 to $85^{\circ} \mathrm{C}$ | 6－Lead，Super－SOT <br> TM －6，JEDEC MO－193， |
| FAN5345S30X | 30 V |  |  |

## Typical Application Diagram



Figure 1. Typical Application

Block Diagram


Figure 2. Functional Block Diagram

Pin Configuration


Figure 3. Pin Assignments Top View

Pin Definitions

| Pin \# | Name | Description |
| :---: | :---: | :--- |
| 5 | VOUT | Boost Output Voltage. Output of the boost regulator. Connect the LEDs to this pin. Connect Cout <br> (output capacitor) to GND. |
| 1 | VIN | Input Voltage. Connect to power source and decouple with Cin to GND. |
| 4 | EN | Enable Brightness Control. Program dimming levels by driving pin with digital pulses. |
| 3 | FB | Voltage Feedback. The boost regulator regulates this pin to 0.250 V to control the LED string current. <br> Tie this pin to a current setting resistor (RET) between GND and the cathode of the LED string. |
| 6 | SW | Switching Node. Tie inductor L1 from VIN to SW pin. |
| 2 | GND | Ground. Tie directly to a GND plane. |

## Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

| Symbol | Parameter |  | Min. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| VIN | VIN Pin |  | -0.3 | 6.0 | V |
| $\mathrm{V}_{\mathrm{FB}}, \mathrm{V}_{\text {EN }}$ | FB, EN Pins |  | -0.3 | $\mathrm{V}_{\text {IN }}+0.3$ | V |
| Vsw | SW Pin | FAN5345S20X | -0.3 | 22.0 | V |
|  |  | FAN5345X30X | -0.3 | 33.0 | V |
| Vout | VOUT Pin | FAN5345S20X | -0.3 | 22.0 | V |
|  |  | FAN5345X30X | -0.3 | 33.0 | V |
| ESD | Electrostatic Discharge Protection | Human Body Mo | 1.5 |  | kV |
|  |  | Charged Device | 1.5 |  |  |
| $\mathrm{T}_{J}$ | Junction Temperature |  | -40 | +150 | ${ }^{\circ} \mathrm{C}$ |
| TSTG | Storage Temperature |  | -65 | +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{L}}$ | Lead Soldering Temperature, 10 Seconds |  | , | +260 | ${ }^{\circ} \mathrm{C}$ |

## Recommended Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to absolute maximum ratings.

| Symbol | Parameter | Comments | Min. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| VIN | $\mathrm{V}_{\text {IN }}$ Supply Voltage | $\cdots$ - | 2.5 | 5.5 | V |
| $V_{\text {OUT }}$ | Vout Voltage ${ }^{(1)}$ | FAN5345S20X | 6.2 | 18.5 | V |
|  |  | FAN5345S30X | 6.2 | 28.5 |  |
| lout | Vout Load Current | $\square$ | 5 | 25 | mA |
| $\mathrm{T}_{\mathrm{A}}$ | Ambient Temperature |  | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{J}$ | Junction Temperature |  | -40 | +125 | ${ }^{\circ} \mathrm{C}$ |

## Note:

1. The application should guarantee that minimum and maximum duty cycle should fall between $20-85 \%$ to meet the specified range.

## Thermal Properties

Junction-to-ambient thermal resistance is a function of application and board layout. This data is measured with four-layer 2 s 2 p boards in accordance to JEDEC standard JESD51. Special attention must be paid not to exceed junction temperature $T_{J(\max )}$ at a given ambient temperature $T_{A}$.

| Symbol | Parameter | Typical | Unit |
| :---: | :---: | :---: | :---: |
| $\theta_{\text {JA6 }}$ | Junction-to-Ambient Thermal Resistance, SSOT23-6 Package | 151 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## Electrical Specifications

$\mathrm{V}_{\mathrm{IN}}=2.5 \mathrm{~V}$ to 5.5 V and $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ unless otherwise noted. Typical values are at $\mathrm{TA}=+25^{\circ} \mathrm{C}$ and $\mathrm{V}_{\mathbb{I N}}=3.6 \mathrm{~V}$.

| Symbol | Parameter | Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power Supplies |  |  |  |  |  |  |
| $I_{\text {SD }}$ | Shutdown Supply Current | EN = GND |  | 0.30 | 0.90 | $\mu \mathrm{A}$ |
| $\mathrm{I}_{\text {Q(ACTIVE) }}$ | Quiescent Current at $\mathrm{I}_{\text {LOAD }}=0 \mathrm{~mA}$ | Device Not Switching, No Load |  | 300 |  | $\mu \mathrm{A}$ |
| Vuvio | Under-Voltage Lockout Threshold | $\mathrm{V}_{\text {IN }}$ Rising | 2.10 | 2.35 | 2.60 | V |
|  |  | $\mathrm{V}_{\text {IN }}$ Falling | 1.80 | 2.05 | 2.30 |  |
| Vuvhyst | Under-Voltage Lockout Hysteresis |  |  | 250 |  | mV |
| EN: Enable Pin |  |  |  |  |  |  |
| $\mathrm{V}_{\mathrm{IH}}$ | HIGH-Level Input Voltage |  | 1.2 |  |  | V |
| $\mathrm{V}_{\text {IL }}$ | LOW-Level Input Voltage |  |  |  | 0.4 | V |
| $\mathrm{R}_{\mathrm{EN}}$ | EN Pull-Down Resistance |  | 200 | 300 | 400 | k $\Omega$ |
| TLO | EN Low Time for Dimming ${ }^{(3)}$ | $\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}$; Figure 28 | 0.5 |  | 300 | $\mu \mathrm{s}$ |
| $\mathrm{THI}_{\mathrm{H}}$ | Delay Between Steps ${ }^{(3)}$ | $\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}$; Figure 28 | 0.5 | - |  | $\mu \mathrm{s}$ |
| $\mathrm{T}_{\text {SD }}$ | EN Low, Shutdown Pulse Width | $\mathrm{V}_{\mathbb{I N}}=3.6 \mathrm{~V}$; from Falling Edge of EN |  |  | 1 | ms |
| Feedback and Reference |  |  |  |  |  |  |
| $V_{\text {FB }}$ | Feedback Voltage | $\begin{aligned} & \mathrm{I}_{\mathrm{LED}}=20 \mathrm{~mA} \text { from }-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} \\ & 2.7 \mathrm{~V} \leq \mathrm{V}_{\text {IN }} \leq 5.5 \mathrm{~V} \end{aligned}$ | 230 | $250$ | 270 | mV |
| $\mathrm{I}_{\text {FB }}$ | Feedback Input Current | $V_{F B}=250 \mathrm{mV}$ |  | 0.1 | 1.0 | $\mu \mathrm{A}$ |
| Power Outputs |  |  |  |  |  |  |
| $\mathrm{R}_{\mathrm{DS}(\mathrm{ON}) \text { _Q1 }}$ | Boost Switch On Resistance | $\begin{aligned} & \mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}, \mathrm{I}_{\mathrm{SW}}=100 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{IN}}=2.5 \mathrm{~V}, \mathrm{I}_{\mathrm{SW}}=100 \mathrm{~mA} \end{aligned}$ |  | 600 |  | $\mathrm{m} \Omega$ |
| $\mathrm{I}_{\text {SW(OFF }}$ | SW Node Leakage ${ }^{(2)}$ | $\begin{aligned} & \mathrm{EN}=0, \mathrm{~V}_{\text {IN }}=\mathrm{V}_{\text {SW }}=\mathrm{V}_{\text {OUT }}=5.5 \mathrm{~V}, \mathrm{~V}_{\text {LED }}= \\ & \mathrm{OV} \end{aligned}$ |  | 0.1 | 2.0 | $\mu \mathrm{A}$ |
| ILIM-PK | Boost Switch Peak Current Limit | $\begin{aligned} & \text { FAN5345S20X: } \mathrm{V}_{\text {IN }}=3.2 \mathrm{~V} \text { to } 4.3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}} \\ & =20^{\circ} \mathrm{C} \text { to }+60^{\circ} \mathrm{C}, \mathrm{~V}_{\mathrm{F}}=3.4 \mathrm{~V}, 4 \text { LEDs } \end{aligned}$ | 200 | 300 | 400 | mA |
|  |  | FAN5345S30X | 500 | 750 | 1000 |  |
| Oscillator |  |  |  |  |  |  |
| $\mathrm{f}_{\text {sw }}$ | Boost Regulator Switching Frequency |  | 0.95 | 1.15 | 1.35 | MHz |
| Output and Protection |  |  |  |  |  |  |
| Vovp | Boost Output Over-Voltage | FAN5345S20X | 18.0 | 20.0 | 21.5 | V |
|  | Protection | FAN5345S30X | 27.5 | 30.0 | 32.5 |  |
|  | OVP Hysteresis | FAN5345S20X |  | 0.8 |  |  |
|  |  | FAN5345S30X |  | 1.0 |  |  |
| $V_{\text {tisc }}$ | V out Short-Circuit Detection Threshold | Vout Falling |  | $\mathrm{V}_{\text {IN }}-1.4$ |  | V |
| $\mathrm{V}_{\text {THSC }}$ | Vout Short-Circuit Detection Threshold | Vout Rising |  | $\mathrm{V}_{\text {IN }}-1.2$ |  | V |
| $\mathrm{D}_{\text {MAX }}$ | Maximum Boost Duty Cycle ${ }^{(3,4)}$ |  | 85 |  |  |  |
| $\mathrm{D}_{\text {MIN }}$ | Minimum Boost Duty Cycle ${ }^{(3,4)}$ |  |  |  | 20 | \% |
| T TSD | Thermal Shutdown |  |  | 150 |  | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {HYS }}$ | Thermal Shutdown Hysteresis |  |  | 35 |  | ${ }^{\circ} \mathrm{C}$ |

## Notes:

2. SW leakage current includes the leakage current of two internal switches; SW to GND and SW to Vout
3. Not tested in production; guaranteed by design.
4. Application should guarantee that minimum and maximum duty cycle fall between $20-85 \%$ to meet the specified range.

## Typical Characteristics

$$
\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{I}_{\text {LED }}=25 \mathrm{~mA}, \mathrm{~L}=10 \mu \mathrm{H}, \mathrm{C}_{\text {OUT }}=1.0 \mu \mathrm{~F}, \text { and } \mathrm{C}_{\mathrm{IN}}=10.0 \mu \mathrm{~F}
$$



Figure 4. 3 LEDs: Efficiency vs. LED Current vs. Input Voltage


Figure 6. 5 LEDs: Efficiency vs. LED Current
vs. Input Voltage


Figure 8. 7 LEDs: Efficiency vs. LED Current vs. Input Voltage


Figure 5. 4 LEDs: Efficiency vs. LED Current vs. Input Voitage


Figure 7. 6 LEDs: Efficiency vs. LED Current
vs. Input Voltage


Figure 9. 8 LEDs: Efficiency vs. LED Current vs. Input Voltage

## Typical Characteristics

$$
\mathrm{V}_{\mathrm{IN}}=3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{I}_{\text {LED }}=25 \mathrm{~mA}, \mathrm{~L}=10 \mu \mathrm{H}, \mathrm{C}_{\text {OUT }}=1.0 \mu \mathrm{~F}, \text { and } \mathrm{C}_{\mathrm{IN}}=10.0 \mu \mathrm{~F}
$$



Figure 10. Efficiency vs. Input Voltage vs. Temperature for 5 LEDs in Series


Figure 12. Delta of $V_{F B}$ Over Input Voltage and Temperature for 7 LEDs with $L=10 \mu \mathrm{H}$ and Cout $=1.0 \mu \mathrm{~F}$


Figure 14. OVP vs. Input Voltage: FAN5345S20X


Figure 11. Efficiency vs. Input Voltage vs. Temperature for 7 LEDs in Series


Figure 13. Frequency vs. Input Voltage vs. Temperature


Figure 15. OVP vs. Input Voltage: FAN5345S30X

## Typical Characteristics

$\mathrm{V}_{\text {IN }}=3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{I}_{\text {LED }}=25 \mathrm{~mA}, \mathrm{~L}=10 \mu \mathrm{H}, \mathrm{C}_{\text {OUT }}=1.0 \mu \mathrm{~F}$, and $\mathrm{C}_{\text {IN }}=10.0 \mu \mathrm{~F}$.


Figure 16. Shutdown Current vs. Input Voltage


Figure 18. Dimming Operation


Figure 20. Line Transient Response for 6 LEDs


Figure 17, Quiescent Current vs. Input Voltage


Figure 19. Line Transient Response for 5 LEDs


Figure 21. Line Transient Response for 7 LEDs

## Typical Characteristics

$\mathrm{V}_{\text {IN }}=3.6 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{I}_{\text {LED }}=25 \mathrm{~mA}, \mathrm{~L}=10 \mu \mathrm{H}, \mathrm{C}_{\text {OUT }}=1.0 \mu \mathrm{~F}$, and $\mathrm{C}_{\text {IN }}=10.0 \mu \mathrm{~F}$.


Figure 22. Startup Waveform for Switch Voltage, Inductor Current, $\mathrm{V}_{\mathrm{FB}}$, and EN for 5 LEDs


Figure 24. Startup Waveform for Switch Voltage, Inductor Current, $\mathrm{V}_{\mathrm{FB}}$, and EN for 6 LEDS


Figure 26. Startup Waveform for Switch Voltage, Inductor Current, $\mathrm{V}_{\mathrm{FB}}$, and EN for 7 LEDs


Figure 23. Steady-State Waveform for Vout, Switch Voltage, and Inductor Current for 5 LEDs


Figure 25. Steady-State Waveform for Vout, Switch Voltage, and Inductor Current for 6 LEDs


Figure 27. Steady-State Waveform for Vout, Switch Voltage, and Inductor Current for 7 LEDs

## Circuit Description

## Overview

The FAN5345 is an inductive current-mode boost serial LED driver that achieves LED current regulation by maintaining 0.250 V across the $\mathrm{R}_{\text {SET }}$ resistor. The current through the LED string (lied) is therefore given by:

$$
\begin{equation*}
I_{L E D}=\frac{0.250}{R_{S E T}} \tag{1}
\end{equation*}
$$

The voltage $\mathrm{V}_{\text {Out }}$ is determined by the sum of the forward voltages across each LED, plus the voltage across $\mathrm{R}_{\mathrm{SET}}$, which is always 250 mV .

## UVLO and Soft-Start

If EN has been LOW for more than 1 ms , the IC may initiate a "cold start" soft-start cycle when EN rises, provided $\mathrm{V}_{\mathrm{IN}}$ is above the UVLO threshold.

## Driving Eight LEDs in Series

FAN5345S30X can drive 8 LEDs in series, but the minimum input voltage $\left(\mathrm{V}_{\mathrm{IN}}\right)$ must be greater than or equal to 2.9 V while the forward voltage of the white LED should be less than or equal to 3.2 V and the maximum LED current cannot exceed 20 mA in order to maintain stable operation.

## Digital Interface

The FAN5345 implements a single-wire digital interface to program the LED brightness to one of thirty-two (32) levels spaced in linear steps. With this single-wire solution, the FAN5345 does not require the system processor to constantly supply a signal to drive the LEDs.

## Digital Dimming Control

The FAN5345 starts driving the LEDs at the maximum brightness level. After startup, the control logic is ready to accept programming pulses to decrease the brightness level by the number of positive edges applied to the EN pin. Figure 28. Digital Pulse-Dimming Control Diagram shows the digital pulse dimming control. The dimming control function has no effect before soft-start finishes. The soft-start takes about 2 ms .

## Over-Current and Short-Circuit Detection

The boost regulator employs a cycle-by-cycle peak inductor current limit of 300 mA (typical) and 750 mA (typical) for FAN5345S20X and FAN5345S30X respectively.

## Over-Voltage / Open-Circuit Protection

If the LED string is an open circuit, FB remains at $O V$ and the output voltage continues to increase in the absence of an over-voltage protection (OVP) circuit. The FAN5345S20X OVP circuit disables the boost regulator when Vout exceeds 20.0 V and continues to keep the regulator off until $V_{\text {out }}$ drops below 19.0V. For FAN5345S30X, the OVP is 30.0 V and it turns back on when $\mathrm{V}_{\text {OUT }}$ is below 29.0 V .

## Thermal Shutdown

When the die temperature exceeds $150^{\circ} \mathrm{C}$, a reset occurs and remains in effect untii the die cools to $115^{\circ} \mathrm{C}$; at which time, the circuit is allowed to begin the soft-start sequence.

## Application Information

The reference schematic diagram is shown in Figure 29. FAN5345 is able to drive up to eight LEDs with input voltage equal or greater than $2.9 \mathrm{~V}\left(\mathrm{~V}_{\mathrm{IN}} \geq 2.9 \mathrm{~V}\right)$. However, the number of LEDs that can be used depends on forward voltage. It is recommended that the forward voltage $\left(\mathrm{V}_{\mathrm{F}}\right)$ of
the white LEDs be no greater than 3.2 V and the maximum LED current is 20 mA . FAN5345 can be also used as a boost convertor by connect the $\mathrm{V}_{\text {out }}$ point to the load directly. The return trace of the load should also return to GND through a sense resistor (R1).


Figure 29. Reference Application Schematic Diagram

## Component Placement and PCB Recommendations



Figure 30. Reference PCB Layout
FAN5345 switches at 1.2 MHz to boost the output voltage. Component placement and PCB layout need to be carefully taken into consideration to ensure stable output and to
prevent generation of noise. Figure 30 is the FAN5345 a portion of the evaluation board layout. The critical layout elements are: the $\mathrm{L} 1, \mathrm{C}_{\mathbb{I N}}, \mathrm{C}_{\mathbb{I N}}$ return trace, $\mathrm{C}_{\text {out, }}$, and the Cout return trace.

## Input Capacitor and Return Trace

The input capacitor is the first priority in a switching buck or boost regulator layout. A stable input source ( $\mathrm{V}_{\mathrm{IN}}$ ) enables a switching regulator to deliver its best performance. During the regulator's operation, it is switching at a high frequency, which makes the load of $\mathrm{C}_{\mathrm{IN}}$ change dynamically to make the input source vary at the same switching frequency as the regulator. To ensure a stable input source, $\mathrm{C}_{\mathrm{IN}}$ needs to hold enough energy to minimize the variation at the input pin of the regulator. For $\mathrm{C}_{\mathrm{IN}}$ to have a fast response of charge / discharge, the trace from $\mathrm{C}_{\mathrm{IN}}$ to the input pin of the regulator and the return trace from GND of the regulator to $\mathrm{C}_{\mathrm{IN}}$ should be as short and wide as possible to minimize trace resistance, inductance, and capacitance. During operation, the current flow from $\mathrm{C}_{\mathbb{I}}$ through the regulator to the load and back to $\mathrm{C}_{\mathrm{IN}}$ contains high-frequency variation due to switching. Trace resistance reduces the overall efficiency due to $I^{2} R$ loss. Even a small trace inductance could effectively yield ground variation to add noise on Vout. The input capacitor should be placed close to the VIN and GND pins of the regulator and traces should be as short as possible. Avoid routing the return trace through different layers because vias have strong inductance effect at high frequencies. If routing to other PCB layers is unavoidable, place vias next to the VIN and GND pins of the regulator to minimize the trace distance.

## Output Capacitor and Return Trace

The output capacitor serves the same purpose as the input capacitor, but also maintains a stable output voltage. As explained above, the current travels to the load and back to the Cout GND terminal. Cout should be placed close to the VOUT pin. The traces of Cout to L1, VOUT, and return from load to Cout should be as short and wide as possible to minimize trace resistance and inductance. To minimize noise coupling to load, a small-value capacitor can be placed between VOUT and Cout to route high-frequency noise back to GND before it gets to the load.

## Inductor

Inductor (L1) should be placed as close to the regulator as possible to minimize trace resistance and inductance for the reasons explained above.

## Sense Resistor

The sense resistor provides a feedback signal for the regulator to control output voltage. A long trace from the sense resistor to the FB pin couples noise into the FB pin. If noise is coupled into the FB pin, it causes unstable operation of the switching regulator, which affects application performance. The return trace from the sense resistor to the FB pin should be short and away from any fast-switching signal traces. The ground plane under the return trace is necessary. If the ground plan under the return trace is noisy, but not the same ground plane as the regulator; the noise could be coupled into the FB pin through PCB parasitic capacitance, yielding noisy output.
In Figure $30 ; C_{\text {IN }}$, Cout, and L1 are all placed next to the regulator. All traces are on the same layer to minimize trace resistance and inductance. Total PCB area, not including the sense resistor, is $67.2 \mathrm{~mm}^{2}(7.47 \mathrm{~mm} \times 8.99 \mathrm{~mm})$.

Table 1. Recommended External Components


## Physical Dimensions



Figure 31. 6-Lead, SuperSOT ${ }^{\text {TM }}$-6, JEDEC MO-193, 1.6mm Wide

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