NPT Series N-Channel IGBT with Anti-Parallel Hyperfast Diode

35 A, 1200 V

HGTG10N120BND

The HGTG10N120BND is a Non-Punch Through (NPT) IGBT design. This is a new member of the MOS gated high voltage switching IGBT family. IGBTs combine the best features of MOSFETs and bipolar transistors. This device has the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The IGBT used is the development type TA49290. The Diode used is the development type TA49189.

The IGBT is ideal for many high voltage switching applications operating at moderate frequencies where low conduction losses are essential, such as: AC and DC motor controls, power supplies and drivers for solenoids, relays and contactors.

Formerly Developmental Type TA49302.

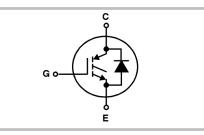
Features

- 35 A, 1200 V, $T_C = 25^{\circ}C$
- 1200 V Switching SOA Capability
- Typical Fall Time: 140 ns at $T_J = 150^{\circ}C$
- Short Circuit Rating
- Low Conduction Loss
- This is Pb-Free Device



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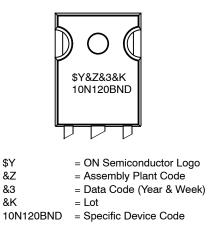
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CASE 340CK

MARKING DIAGRAMS



ORDERING INFORMATION

Part Number Package Brand HGTG10N120BND TO-247 10N120BND

NOTE: When ordering, use the entire part number.

ABSOLUTE MAXIMUM RATINGS (T_C = 25°C, Unless Otherwise Specified)

Description	Symbol	HGTG10N120BND	Units	
Collector to Emitter Voltage	BV _{CES}	1200	V	
Collector Current Continuous At $T_C = 25^{\circ}C$ At $T_C = 110^{\circ}C$	I _{C25} I _{C110}	35 17	A A	
Collector Current Pulsed (Note 1)	I _{CM}	80	А	
Gate to Emitter Voltage Continuous	V _{GES}	±20	V	
Gate to Emitter Voltage Pulsed	V _{GEM}	±30	V	
Switching Safe Operating Area at $T_J = 150^{\circ}C$ (Figure 2)	SSOA	55 A at 1200 V		
Power Dissipation Total at $T_C = 25^{\circ}C$ Power Dissipation Derating $T_C > 25^{\circ}C$	P _D	298 2.38	W W/°C	
Operating and Storage Junction Temperature Range	T _J , T _{STG}	-55 to 150	°C	
Maximum Lead Temperature for Soldering	Τ _L	260	°C	
Short Circuit Withstand Time (Note 2) at V_{GE} = 15 V	t _{SC}	8	μs	
Short Circuit Withstand Time (Note 2) at V _{GE} = 12 V	t _{SC}	15	μs	

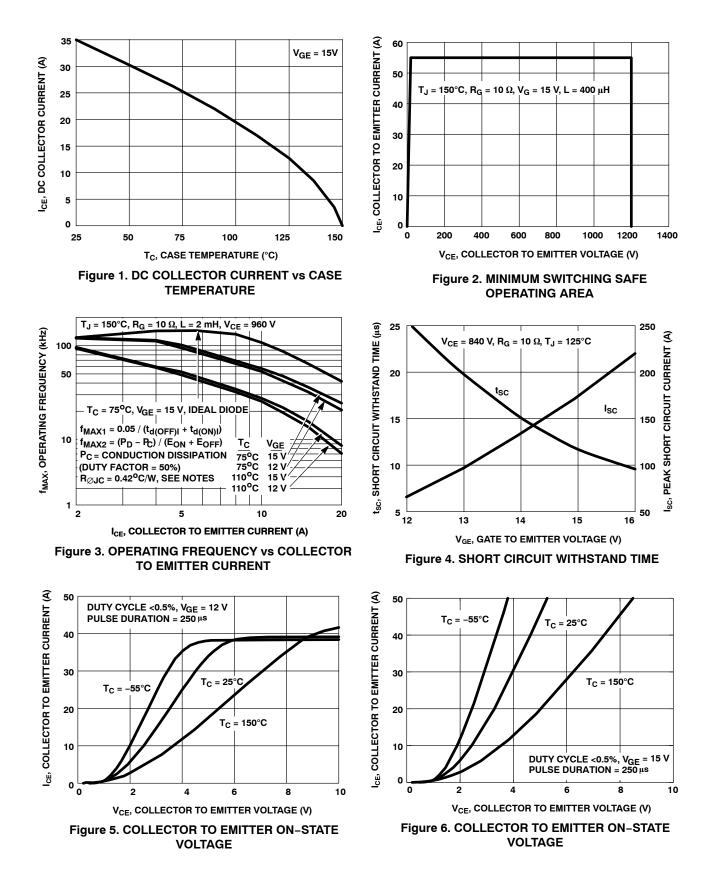
Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.
Pulse width limited by maximum junction temperature.
V_{CE(PK)} = 840 V, T_J = 125°C, R_G = 10 Ω.

Parameter	Symbol	Test (Min	Тур	Max	Units	
Collector to Emitter Breakdown Voltage	BV _{CES}	I _C = 250 μA, V _{GE} = 0 V		1200	-	-	V
Collector to Emitter Leakage Current	I _{CES}	V _{CE} = 1200 V	$T_{C} = 25^{\circ}C$	-	-	250	μΑ
			T _C = 125°C	-	170	-	μA
			T _C = 150°C	-	-	2.5	mA
Collector to Emitter Saturation Voltage	V _{CE(SAT)}	I _C = 10 A, V _{GE} = 15 V	$T_C = 25^{\circ}C$	-	2.45	2.7	V
			T _C = 150°C	-	3.7	4.2	V
Gate to Emitter Threshold Voltage	V _{GE(TH)}	I _C = 90 μA, V _{CE} = V _{GE}		6.0	6.8	-	V
Gate to Emitter Leakage Current	I _{GES}	V _{GE} = ±20 V		-	-	±250	nA
Switching SOA	SSOA	$ \begin{array}{l} T_J = 150^\circ C, \ R_G = 10 \ \Omega, \ V_{GE} = 15 \ V, \\ L = 400 \ \mu H, \ V_{CE(PK)} = 1200 \ V \end{array} $		55	_	_	A
Gate to Emitter Plateau Voltage	V _{GEP}	I _C = 10 A, V _{CE} = 600 V		-	10.4	-	V
On-State Gate Charge	Q _{G(ON)}	I _C = 10 A, V _{CE} = 600 V	V _{GE} = 15 V	-	100	120	nC
			V _{GE} = 20 V	-	130	150	nC
Current Turn-On Delay Time	t _{d(ON)} I	IGBT and Diode at $T_J = 25^{\circ}C$, $I_{CE} = 10 A$, $V_{CE} = 960 V$, $V_{GE} = 15 V$, $R_G = 10 \Omega$, L = 2 mH, Test Circuit (Figure 20)		-	23	26	ns
Current Rise Time	t _{rl}			-	11	15	ns
Current Turn-Off Delay Time	t _{d(OFF)} I			-	165	210	ns
Current Fall Time	t _{fl}			-	100	140	ns
Turn–On Energy	E _{ON}			-	0.85	1.05	mJ
Turn–Off Energy (Note 3)	E _{OFF}			-	0.8	1.0	mJ
Current Turn-On Delay Time	t _{d(ON)} I	IGBT and Diode at $T_J = 150^{\circ}C$, $I_{CE} = 10 A$, $V_{CE} = 960 V$, $V_{GE} = 15 V$, $R_G = 10 \Omega$, L = 2 mH, Test Circuit (Figure 20)		-	21	25	ns
Current Rise Time	t _{rl}			-	11	15	ns
Current Turn-Off Delay Time	t _{d(OFF)} I			-	190	250	ns
Current Fall Time	t _{fl}			-	140	200	ns
Turn–On Energy	E _{ON}			_	1.75	2.3	mJ
Turn–Off Energy (Note 3)	E _{OFF}			-	1.1	1.4	mJ
Diode Forward Voltage	V _{EC}	I _{EC} = 10 A		-	2.55	3.2	V
Diode Reverse Recovery Time	t _{rr}	I_{EC} = 10 A, dI _{EC} /dt = 200 A/µs		-	57	70	ns
		I_{EC} = 1 A, dI _{EC} /dt = 200 A/µs		-	32	40	ns
Thermal Resistance Junction To Case	$R_{\theta JC}$	IGBT		-	-	0.42	°C/W
		Diode		-	-	1.25	°C/W

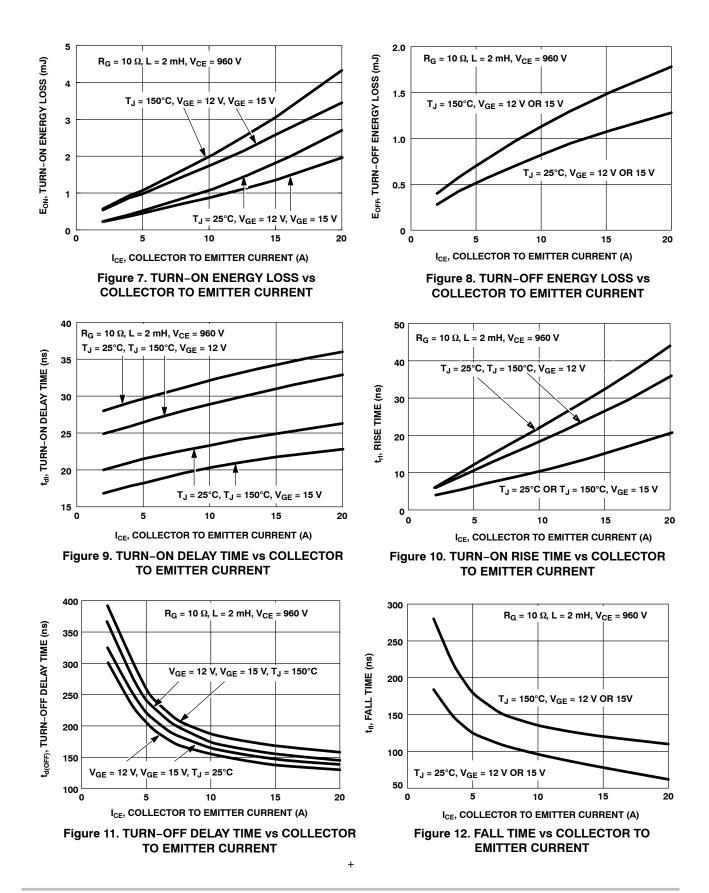
Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product

Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I_{CE} = 0A). All devices were tested per JEDEC Standard No. 24–1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

TYPICAL PERFORMANCE CHARACTERISTICS



TYPICAL PERFORMANCE CHARACTERISTICS (continued)



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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

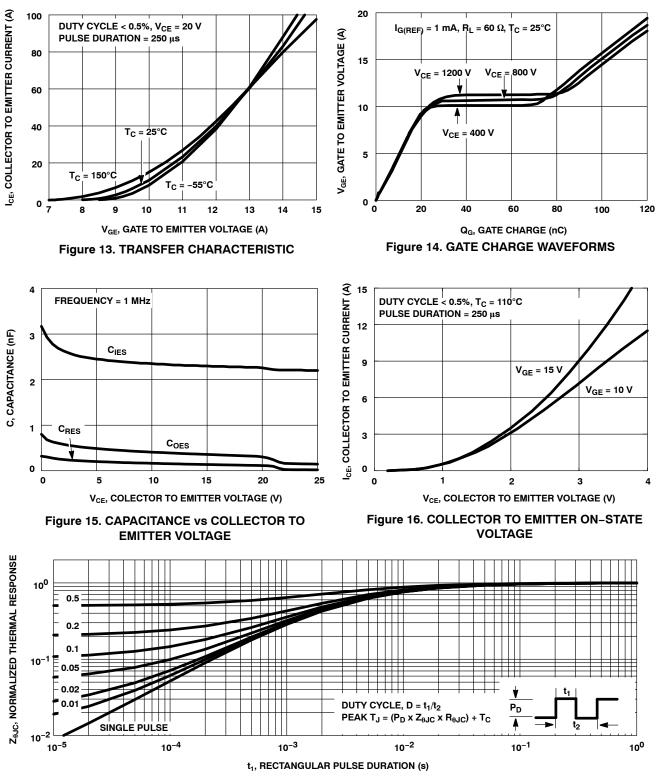
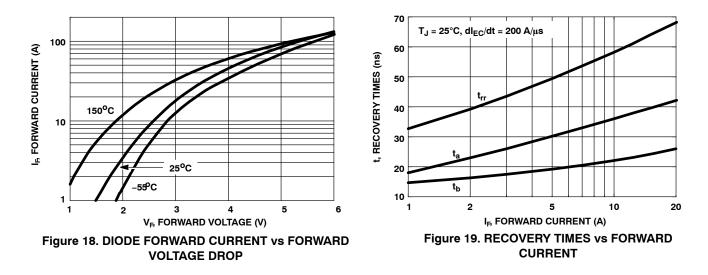


Figure 17. NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE

TYPICAL PERFORMANCE CHARACTERISTICS (continued)





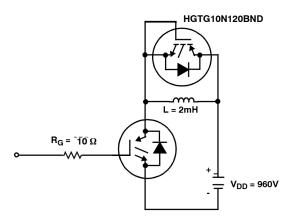
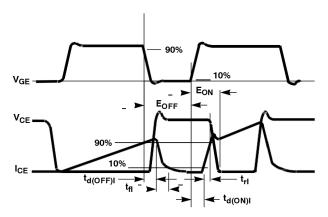


Figure 20. INDUCTIVE SWITCHING TEST CIRCUIT





HANDLING PRECAUTIONS FOR IGBTS

Insulated Gate Bipolar Transistors are susceptible to gate-insulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD[™] LD26" or equivalent
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means for example, with a metallic wristband
- 3. Tips of soldering irons should be grounded
- 4. Devices should never be inserted into or removed from circuits with power on
- 5. Gate Voltage Rating Never exceed the gate–voltage rating of V_{GEM} . Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region
- 6. Gate Termination The gates of these devices are essentially capacitors. Circuits that leave the gate open– circuited or floating should be avoided. These conditions can result in turn–on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup
- Gate Protection These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended

OPERATING FREQUENCY INFORMATION

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information s11hown for a typical unit in Figures 5, 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows f_{MAX1} or f_{MAX2} ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 f_{MAX1} is defined by $f_{MAX1} = 0.05/(t_{d(OFF)I} + t_{d(ON)I})$. Deadtime (the denominator) has been arbitrarily held to 10% of the on-state time for a 50% duty factor. Other definitions are possible. $t_{d(OFF)I}$ and $t_{d(ON)I}$ are defined in Figure 21. Device turn-off delay can establish an additional frequency limiting condition for an application other than T_{JM} . $t_{d(OFF)I}$ is important when controlling output ripple under a lightly loaded condition.

 f_{MAX2} is defined by $f_{MAX2} = (P_D - P_C)/(E_{OFF} + E_{ON})$. The allowable dissipation (P_D) is defined by $P_D = (T_{JM} - T_C)/R_{\theta JC}$. The sum of device switching and conduction losses must not exceed P_D . A 50% duty factor was used (Figure 3) and the conduction losses (P_C) are approximated by

$$P_{C} = (V_{CE} \times I_{CE})/2 \qquad (eq. 1)$$

 E_{ON} and E_{OFF} are defined in the switching waveforms shown in Figure 21. E_{ON} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn–on and E_{OFF} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn–off. All tail losses are included in the calculation for E_{OFF} ; i.e., the collector current equals zero ($I_{CE} = 0$).

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