SMPS Series N-Channel IGBT

600 V

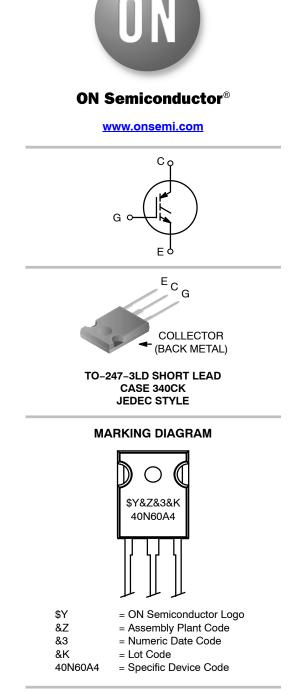
HGTG40N60A4

The HGTG40N60A4 is a MOS gated high voltage switching device combining the best features of a MOSFET and a bipolar transistor. This device has the high input impedance of a MOSFET and the low on-state conduction loss of a bipolar transistor. The much lower on-state voltage drop varies only moderately between 25°C and 150°C. This IGBT is ideal for many high voltage switching applications operating at high frequencies where low conduction losses are essential. This device has been optimized for high frequency switch mode power supplies

Formerly Developmental Type TA49347.

Features

- 100 kHz Operation at 390 V, 40 A
- 200 kHz Operation at 390 V, 20 A
- 600 V Switching SOA Capability
- Typical Fall Time 55 ns at $T_J = 125^{\circ}C$
- Low Conduction Loss
- This is a Pb–Free Device



ORDERING INFORMATION

See detailed ordering and shipping information on page 7 of this data sheet.

ABSOLUTE MAXIMUM RATINGS (T_C = 25°C unless otherwise specified)

Parameter	Symbol	HGTG40N60A4	Unit
Collector to Emitter Voltage	BV _{CES}	600	V
Collector Current Continuous At $T_C = 25^{\circ}C$ At $T_C = 110^{\circ}C$	I _{C25} I _{C110}	75 63	A A
Collector Current Pulsed (Note 1)	I _{CM}	300	А
Gate to Emitter Voltage Continuous	V _{GES}	±20	V
Gate to Emitter Voltage Pulsed	V _{GEM}	±30	V
Switching Safe Operating Area at $T_J = 150^{\circ}$ C, Figure 2	SSOA	200 A at 600 V	
Power Dissipation Total at $T_C = 25^{\circ}C$	PD	625	W
Power Dissipation Derating $T_C > 25^{\circ}C$		5	W/°C
Operating and Storage Junction Temperature Range	T _J , T _{STG}	–55 to 150	°C
Maximum Lead Temperature for Soldering	TL	260	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected. 1. Pulse width limited by maximum junction temperature.

ELECTRICAL CHARACTERISTICS (T_J = 25° C unless otherwise specified)

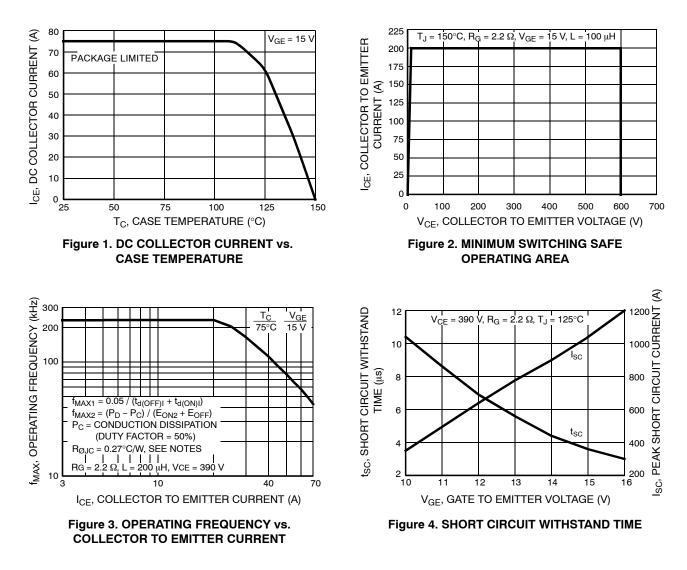
Parameter	Symbol	Test Condition		Min	Тур	Max	Unit
Collector to Emitter Breakdown Voltage	BV _{CES}	$I_{C} = 250 \ \mu A, \ V_{GE} = 0 \ V$		600	-	-	V
Emitter to Collector Breakdown Voltage	BV _{ECS}	I _C = –10 mA, V _{GE} = 0 V		20	-	-	V
Collector to Emitter Leakage Current	urrent I_{CES} $V_{CE} = BV_{CES}$ $T_J =$		$T_J = 25^{\circ}C$	-	-	250	μA
			$T_J = 125^{\circ}C$	-	-	3.0	mA
Collector to Emitter Saturation Voltage	V _{CE(SAT)}	I _C = 40 A, V _{GE} = 15 V	$T_J = 25^{\circ}C$	-	1.7	2.7	V
			T _J = 125°C	-	1.5	2.0	V
Gate to Emitter Threshold Voltage	V _{GE(TH)}	$I_{C} = 250 \ \mu A, \ V_{CE} = V_{GE}$		4.5	5.6	7	V
Gate to Emitter Leakage Current	I _{GES}	V _{GE} = ±20 V		-	-	±250	nA
Switching SOA	SSOA	$ \begin{array}{l} T_J = 150^{\circ}C, \ R_G = 2.2 \ \Omega, \ V_{GE} = 15 \ V, \\ L = 100 \ \mu H, \ V_{CE} = \ 600 \ V \end{array} $		200	-	_	A
Gate to Emitter Plateau Voltage	V _{GEP}	$I_{C} = 40 \text{ A}, V_{CE} = 0.5 \text{ BV}_{CES}$		-	8.5	-	V
On-State Gate Charge		V _{GE} = 15 V	-	350	405	nC	
		$V_{CE} = 0.5 BV_{CES}$	V _{GE} = 20 V	-	450	520	nC
Current Turn-On Delay Time	t _{d(ON)} I	IGBT and Diode at $T_J = 25^{\circ}C$, $I_{CE} = 40 \text{ A}$, $V_{CE} = 0.65 \text{ BV}_{CES}$, $V_{GE} = 15 \text{ V}$, $R_G = 2.2 \Omega$, $L = 200 \mu$ H, Test Circuit (Figure 20)		-	25	-	ns
Current Rise Time	t _{rl}			-	18	-	ns
Current Turn-Off Delay Time	t _{d(OFF)} I			-	145	-	ns
Current Fall Time	t _{fl}			-	35	-	ns
Turn-On Energy (Note 3)	E _{ON1}			-	400	-	μJ
Turn-On Energy (Note 3)	E _{ON2}			-	850	-	μJ
Turn-Off Energy (Note 2)	E _{OFF}			-	370	-	μJ

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current Turn-On Delay Time	t _{d(ON)} I	IGBT and Diode at $T_J = 125^{\circ}C$,	-	27	-	ns
Current Rise Time	t _{rl}	$\begin{array}{l} I_{CE} = 40 \text{ A,} \\ V_{CE} = 0.65 \text{ BV}_{CES}, \\ V_{GE} = 15 \text{ V,} \\ R_G = 2.2 \ \Omega, \\ L = 200 \ \mu\text{H}, \\ \text{Test Circuit (Figure 20)} \end{array}$	-	20	-	ns
Current Turn-Off Delay Time	t _{d(OFF)} I		-	185	225	ns
Current Fall Time	t _{fl}		-	55	95	ns
Turn-On Energy (Note 3)	E _{ON1}		-	400	-	μJ
Turn-On Energy (Note 3)	E _{ON2}		-	1220	1400	μJ
Turn-Off Energy (Note 2)	E _{OFF}	1	-	700	800	μJ
Thermal Resistance Junction To Case	$R_{\theta JC}$		-	-	0.2	°C/W

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

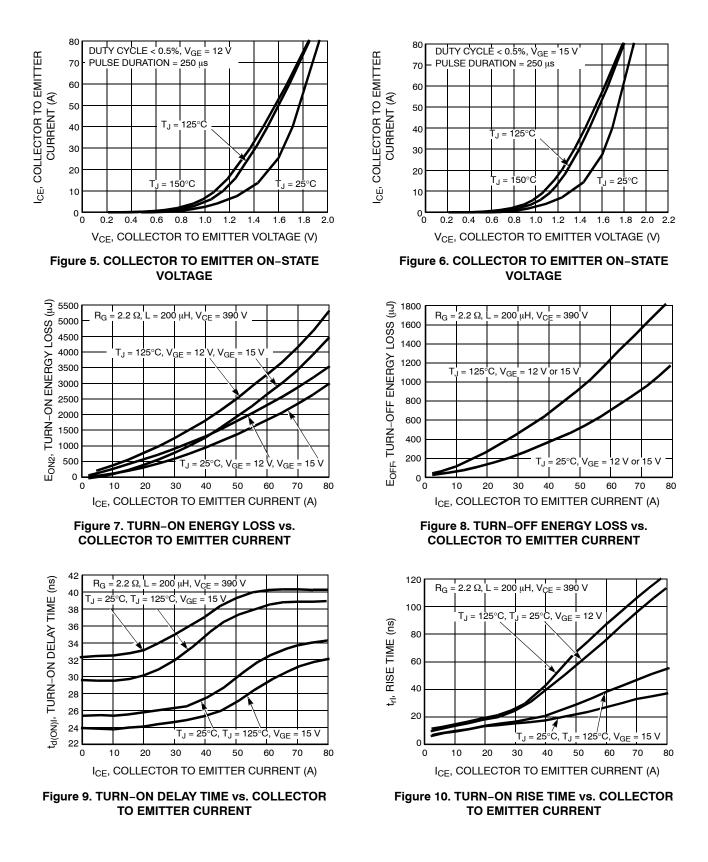
 Turn-Off Energy Loss (E_{OFF}) is defined as the integral of the instantaneous power loss starting at the trailing edge of the input pulse and ending at the point where the collector current equals zero (I_{CE} = 0 A). All devices were tested per JEDEC Standard No. 24–1 Method for Measurement of Power Device Turn-Off Switching Loss. This test method produces the true total Turn-Off Energy Loss.

Values for two Turn-On loss conditions are shown for the convenience of the circuit designer. E_{ON1} is the turn-on loss of the IGBT only. E_{ON2} is the turn-on loss when a typical diode is used in the test circuit and the diode is at the same T_J as the IGBT. The diode type is specified in Figure 20.

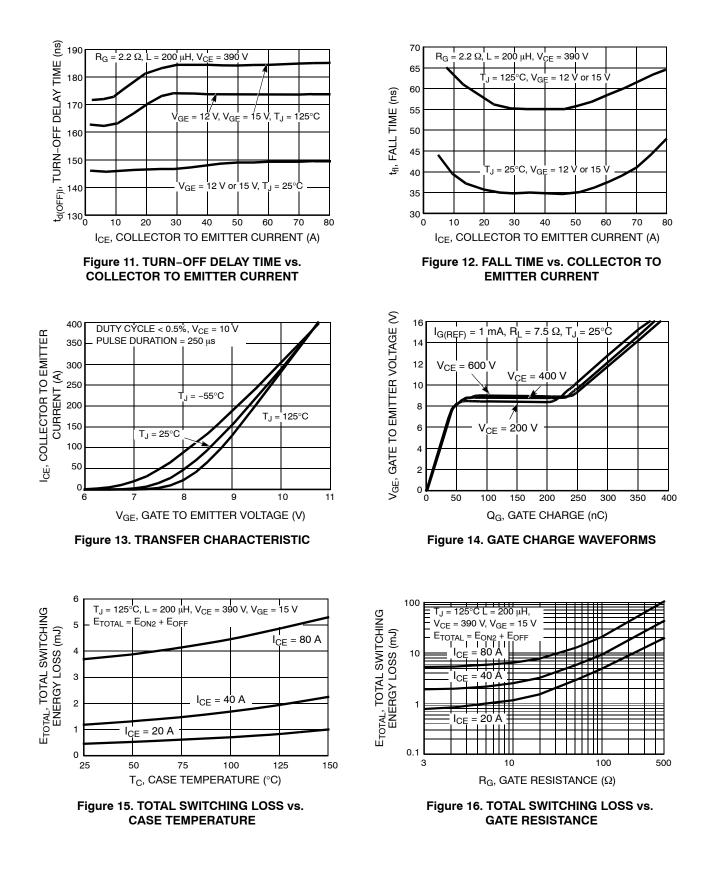


TYPICAL PERFORMANCE CURVES (unless otherwise specified)

TYPICAL PERFORMANCE CURVES (unless otherwise specified) (continued)



TYPICAL PERFORMANCE CURVES (unless otherwise specified) (continued)



TYPICAL PERFORMANCE CURVES (unless otherwise specified) (continued)

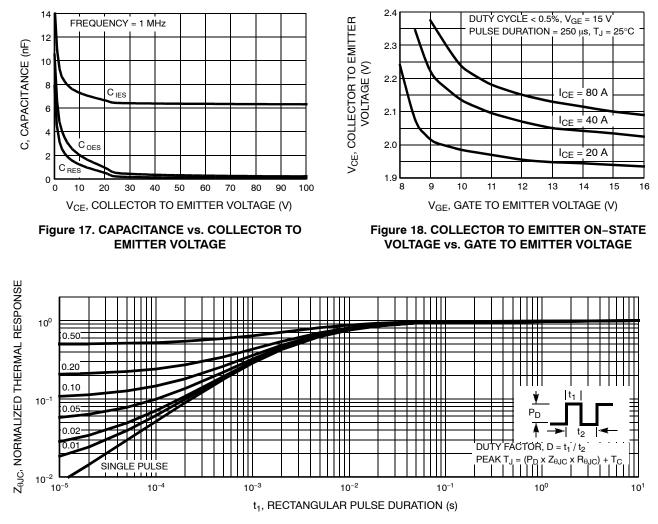
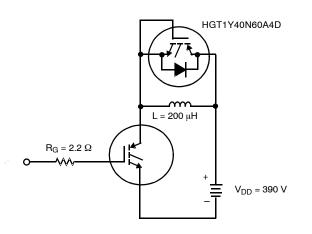


Figure 19. NORMALIZED TRANSIENT THERMAL RESPONSE, JUNCTION TO CASE



TEST CIRCUIT AND WAVEFORMS

Figure 20. INDUCTIVE SWITCHING TEST CIRCUIT

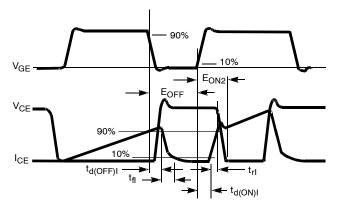


Figure 21. SWITCHING TEST WAVEFORMS

HANDLING PRECAUTIONS FOR IGBTs

Insulated Gate Bipolar Transistors are susceptible to gateinsulation damage by the electrostatic discharge of energy through the devices. When handling these devices, care should be exercised to assure that the static charge built in the handler's body capacitance is not discharged through the device. With proper handling and application procedures, however, IGBTs are currently being extensively used in production by numerous equipment manufacturers in military, industrial and consumer applications, with virtually no damage problems due to electrostatic discharge. IGBTs can be handled safely if the following basic precautions are taken:

- Prior to assembly into a circuit, all leads should be kept shorted together either by the use of metal shorting springs or by the insertion into conductive material such as "ECCOSORBD[™] LD26" or equivalent.
- 2. When devices are removed by hand from their carriers, the hand being used should be grounded by any suitable means for example, with a metallic wristband.
- 3. Tips of soldering irons should be grounded.
- 4. Devices should never be inserted into or removed from circuits with power on.
- 5. *Gate Voltage Rating* Never exceed the gate–voltage rating of V_{GEM} . Exceeding the rated V_{GE} can result in permanent damage to the oxide layer in the gate region.
- 6. Gate Termination The gates of these devices are essentially capacitors. Circuits that leave the gate open– circuited or floating should be avoided. These conditions can result in turn–on of the device due to voltage buildup on the input capacitor due to leakage currents or pickup.

7. *Gate Protection* - These devices do not have an internal monolithic Zener diode from gate to emitter. If gate protection is required an external Zener is recommended.

OPERATING FREQUENCY INFORMATION

Operating frequency information for a typical device (Figure 3) is presented as a guide for estimating device performance for a specific application. Other typical frequency vs collector current (I_{CE}) plots are possible using the information shown for a typical unit in Figures 6, 7, 8, 9 and 11. The operating frequency plot (Figure 3) of a typical device shows f_{MAX1} or f_{MAX2} ; whichever is smaller at each point. The information is based on measurements of a typical device and is bounded by the maximum rated junction temperature.

 f_{MAX1} is defined by $f_{MAX1} = 0.05 / (t_{d(OFF)I} + t_{d(ON)I})$. Deadtime (the denominator) has been arbitrarily held to 10% of the on–state time for a 50% duty factor. Other definitions are possible. $t_{d(OFF)I}$ and $t_{d(ON)I}$ are defined in Figure 21. Device turn–off delay can establish an additional frequency limiting condition for an application other than T_{JM} . $t_{d(OFF)I}$ is important when controlling output ripple under a lightly loaded condition.

 f_{MAX2} is defined by $f_{MAX2} = (P_D - P_C) / (E_{OFF} + E_{ON2})$. The allowable dissipation (P_D) is defined by $P_D = (T_{JM} - T_C) / R_{\theta JC}$. The sum of device switching and conduction losses must not exceed P_D. A 50% duty factor was used (Figure 21) and the conduction losses (P_C) are approximated by $P_C = (V_{CE} \ x \ I_{CE}) / 2$.

 E_{ON2} and E_{OFF} are defined in the switching waveforms shown in Figure 25. E_{ON2} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn–on and E_{OFF} is the integral of the instantaneous power loss ($I_{CE} \times V_{CE}$) during turn–off. All tail losses are included in the calculation for E_{OFF} ; i.e., the collector current equals zero ($I_{CE} = 0$).

ORDERING INFORMATION

Part Number	Package	Brand	Shipping
HGTG40N60A4	TO-247	40N60A4	450 Units / Tube

NOTE: When ordering, use the entire part number.

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