## LV8740V

## Monolithic Linear IC

## PWM Current Control Stepper Motor Driver

## ON Semiconductor ${ }^{\text {® }}$

http:/lonsemi.com

## Overview

The LV8740V is a 2 -channel H -bridge driver IC that can switch a stepper motor driver, which is capable of micro-step drive and supports Quarter-step excitation, and two channels of a brushed DC motor driver, which supports forward, reverse, brake, and standby of a motor. It is ideally suited for driving brushed DC motors and stepper motors used in office equipment and amusement applications.

## Function

- Single-channel PWM current control stepper motor driver (selectable with DC motor driver channel 2) incorporated.
- On resistance (upper side : $0.3 \Omega$; lower side : $0.2 \Omega$; total of upper and lower : $0.5 \Omega ; \mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{IO}=2.5 \mathrm{~A}$ )
- Excitation mode can be set to Full-step, Half-step full torque, Half-step , or Quarter-step
- Excitation step proceeds only by step signal input
- Motor current selectable in four steps
- BiCDMOS process IC
- Output short-circuit protection circuit (selectable from latch-type or auto reset-type) incorporated
- Unusual condition warning output pins
- No control supply required


## Specifications

Absolute Maximum Ratings at $\mathrm{Ta}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Supply voltage 1 | $\mathrm{V}_{\mathrm{M}}$ max | VM, VM1, VM2 | 38 | V |
| Output peak current | lo peak | $\mathrm{tw} \leq 10 \mathrm{~ms}$, duty $20 \%$, Each 1ch | 3.0 | A |
| Output current | $\mathrm{l}_{0}$ max | Each 1ch | 2.5 | A |
| Logic input voltage | $\mathrm{V}_{\mathrm{IN}}$ | ST , OE , DM , MD1/DC11, MD2/DC12, <br> FR/DC21, STP/DC22, RST , EMM , ATT1, ATT2 | -0.3 to +6.0 | V |
| MONI/EMO input voltage | $\mathrm{V}_{\mathrm{MONI}} / \mathrm{V}_{\text {EMO }}$ |  | -0.3 to +6.0 | V |
| Allowable power dissipation | Pd max | * | 3.45 | W |
| Operating temperature | Topr |  | -30 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -55 to +150 | ${ }^{\circ} \mathrm{C}$ |

* Specified circuit board : $90 \times 90 \times 1.6 \mathrm{~mm}^{3}$ : 2-Layer glass epoxy printed circuit board with back mounting.

Caution 1) Absolute maximum ratings represent the value which cannot be exceeded for any length of time.
Caution 2) Even when the device is used within the range of absolute maximum ratings, as a result of continuous usage under high temperature, high current, high voltage, or drastic temperature change, the reliability of the IC may be degraded. Please contact us for the further details

Recommended Operating Conditions at $\mathrm{Ta}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :--- | :---: | :---: |
| Supply voltage range | $\mathrm{V}_{\mathrm{M}}$ | $\mathrm{VM}, \mathrm{VM1}, \mathrm{VM2}$ | 9 to 35 | V |
| Logic input voltage | VIN | $\mathrm{ST}, \mathrm{OE}, \mathrm{DM}, \mathrm{MD1/DC11}, \mathrm{MD2/DC12}$, <br> FR/DC21, STP/DC22, RST, EMM , ATT1, <br> ATT2 | 0 to 5.5 | V |
| VREF input voltage range | VREF |  | 0 to 3.0 | V |

Electrical Characteristics at $\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{M}}=24 \mathrm{~V}, \mathrm{VREF}=1.5 \mathrm{~V}$

| Parameter |  | Symbol | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | min |  | typ | max |  |
| Standby mode current drain |  |  | IMstn | ST = "L", l(VM)+I(VM1)+I(VM2) |  | 180 | 250 | $\mu \mathrm{A}$ |
| Current drain |  | IM | $\begin{aligned} & \text { ST = "H", OE = "L", no load } \\ & \text { I(VM)+I(VM1)+I(VM2) } \end{aligned}$ |  | 3 | 5 | mA |
| VREG5 output voltage |  | Vreg5 | $l^{\prime}=-1 \mathrm{~mA}$ | 4.7 | 5.0 | 5.3 | V |
| Thermal shutdown temperature |  | TSD | Design guarantee | 150 | 180 | 210 | ${ }^{\circ} \mathrm{C}$ |
| Thermal hysteresis width |  | $\Delta \mathrm{TSD}$ | Design guarantee |  | 40 |  | ${ }^{\circ} \mathrm{C}$ |
| Motor Driver |  |  |  |  |  |  |  |
| Output on-resistance |  | Ronu | $\mathrm{I} \mathrm{O}=2.5 \mathrm{~A}$, Upper-side on resistance |  | 0.3 | 0.4 | $\Omega$ |
|  |  | Rond | $\mathrm{I} \mathrm{O}=2.5 \mathrm{~A}$, Lower-side on resistance |  | 0.2 | 0.25 | $\Omega$ |
| Output leakage current |  | Ioleak | $\mathrm{VM}=35 \mathrm{~V}$ |  |  | 50 | $\mu \mathrm{A}$ |
| Diode forward voltage |  | VD | $\mathrm{ID}=-2.5 \mathrm{~A}$ |  | 1.1 | 1.3 | V |
| ST pin input current |  | ${ }_{\text {ISTL }}$ | $\mathrm{V}_{\text {IN }}=0.8 \mathrm{~V}$ | 3 | 8 | 15 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{IsTH}^{\text {H }}$ | $\mathrm{V}_{\text {IN }}=5 \mathrm{~V}$ | 48 | 80 | 112 | $\mu \mathrm{A}$ |
| Logic pin input current (other ST pin) |  | ${ }_{\text {I }} \mathrm{NL}$ | OE , DM , MD1/DC11, MD2/DC12, <br> FR/DC21, STP/DC22, RST , EMM , ATT1, <br> ATT2, $\mathrm{V}_{\text {IN }}=0.8 \mathrm{~V}$ | 3 | 8 | 15 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{l}_{1} \mathrm{H}$ | $\mathrm{V}_{\text {IN }}=5 \mathrm{~V}$ | 30 | 50 | 70 | $\mu \mathrm{A}$ |
| Logic input voltage | High | $\mathrm{V}_{1 \times}{ }^{\text {h }}$ | ST, OE , DM , MD1/DC11, MD2/DC12, FR/DC21, STP/DC22 , RST , EMM , ATT1, ATT2 | 2.0 |  | 5.5 | V |
|  | Low | $\mathrm{V}_{\text {IN }}$ |  | 0 |  | 0.8 | V |
| Current setting comparator threshold voltage (Current step switch) | Quarter <br> step <br> resolution | Vtdac0_W | Step 0(When initialized : channel 1 comparator level) | 0.290 | 0.300 | 0.310 | V |
|  |  | Vtdac1_W | Step 1 (Initial state+1) | 0.260 | 0.270 | 0.280 | V |
|  |  | Vtdac2_W | Step 2 (Initial state+2) | 0.200 | 0.210 | 0.220 | V |
|  |  | Vtdac3_W | Step 3 (Initial state +3 ) | 0.095 | 0.105 | 0.115 | V |
|  | Half step resolution | Vtdac0_H | Step 0 (When initialized: channel 1 comparator level) | 0.290 | 0.300 | 0.310 | V |
|  |  | Vtdac2_H | Step 2 (Initial state+1) | 0.200 | 0.210 | 0.220 | V |
|  | Half step resolution (full torque) | Vtdac0_HF | Step 0 (Initial state, channel 1 comparator level) | 0.290 | 0.300 | 0.310 | V |
|  |  | Vtdac2_HF | Step 2 (Initial state+1) | 0.290 | 0.300 | 0.310 | V |
|  | Full step resolution | Vtdac2_F | Step 2 | 0.290 | 0.300 | 0.310 | V |
| Current setting comparator threshold voltage (Current attenuation rate switch) |  | Vtatt00 | ATT1=L, ATT2=L | 0.290 | 0.300 | 0.310 | V |
|  |  | Vtatt01 | ATT1 $=\mathrm{H}, \mathrm{ATT} 2=\mathrm{L}$ | 0.190 | 0.200 | 0.210 | V |
|  |  | Vtatt10 | ATT1 $=$ L, ATT2 $=\mathrm{H}$ | 0.140 | 0.150 | 0.160 | V |
|  |  | Vtatt11 | ATT1 $=\mathrm{H}, \mathrm{ATT} 2=\mathrm{H}$ | 0.090 | 0.100 | 0.110 | V |
| Chopping frequency |  | Fchop | $\mathrm{RCHOP}=20 \mathrm{k} \Omega$ | 45 | 62.5 | 75 | kHz |
| VREF pin input current |  | Iref | $\mathrm{VREF}=1.5 \mathrm{~V}$ | -0.5 |  |  | $\mu \mathrm{A}$ |
| MONI pin saturation voltage |  | Vsatmon | ${ }^{1} \mathrm{MONI}=1 \mathrm{~mA}$ |  | 50 | 100 | mV |
| Charge pump |  |  |  |  |  |  |  |
| VG output voltage |  | VG |  | 28 | 28.7 | 29.8 | V |
| Rise time |  | tONG | VG $=0.1 \mu \mathrm{~F}$, Between CP1-CP2 0.1uF $\mathrm{ST}=" \mathrm{H} " \rightarrow \mathrm{VG}=\mathrm{VM}+4 \mathrm{~V}$ |  |  | 0.5 | ms |
| Oscillator frequency |  | Fosc | $\mathrm{RCHOP}=20 \mathrm{k} \Omega$ | 90 | 125 | 150 | kHz |

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| Parameter | Symbol | Conditions | Ratings |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | min | typ | max |  |
| Output short-circuit protection |  |  |  |  |  |  |
| EMO pin saturation voltage | Vsatemo | lemo $=1 \mathrm{~mA}$ |  | 50 | 100 | mV |
| CEM pin charge current | Icem | Vcem=0V | 7 | 10 | 13 | $\mu \mathrm{A}$ |
| CEM pin threshold voltage | Vtcem |  | 0.8 | 1.0 | 1.2 | V |

Pin Assignment


## Package Dimensions

unit : mm (typ)
3285B



Substrate Specifications (Substrate recommended for operation of LV8740V)

| Size | $: 90 \mathrm{~mm} \times 90 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ |
| :--- | :--- |
| Material | $:$ Glass epoxy |
| Copper wiring density | $: \mathrm{L} 1=85 \% / \mathrm{L} 2=90 \%$ |



L1: Copper wiring pattern diagram


L2 : Copper wiring pattern diagram

## Cautions

1) The data for the case with the Exposed Die-Pad substrate mounted shows the values when $90 \%$ or more of the Exposed Die-Pad is wet.
2) For the set design, employ the derating design with sufficient margin.

Stresses to be derated include the voltage, current, junction temperature, power loss, and mechanical stresses such as vibration, impact, and tension.
Accordingly, the design must ensure these stresses to be as low or small as possible.
The guideline for ordinary derating is shown below :
(1)Maximum value $80 \%$ or less for the voltage rating
(2)Maximum value $80 \%$ or less for the current rating
(3)Maximum value $80 \%$ or less for the temperature rating
3) After the set design, be sure to verify the design with the actual product.

Confirm the solder joint state and verify also the reliability of solder joint for the Exposed Die-Pad, etc.
Any void or deterioration, if observed in the solder joint of these parts, causes deteriorated thermal conduction, possibly resulting in thermal destruction of IC.

## Block Diagram



LV8740V
Pin Functions

| Pin No. | Pin name | Description |
| :---: | :---: | :---: |
| 1 | VG | Charge pump capacitor connection pin |
| 2 | VM | Motor power supply connection pin |
| 3 | CP2 | Charge pump capacitor connection pin |
| 4 | CP1 | Charge pump capacitor connection pin |
| 5 | VREG5 | Internal power supply capacitor connection pin |
| 6 | ATT2 | Motor holding current switching pin |
| 7 | ATT1 | Motor holding current switching pin |
| 8 | EMO | Output short-circuit state warning output pin |
| 9 | CEM | Pin to connect the output short-circuit state detection time setting capacitor |
| 10 | EMM | Over current mode switching pin |
| 11 | RCHOP | Chopping frequency setting resistor connection pin |
| 12 | MONI | Position detection monitor pin |
| 13 | RST | Reset signal input pin |
| 14 | STP/DC22 | STM STEP signal input pin/DCM2 output control input pin |
| 15 | FR/DC21 | STM forward/reverse rotation signal input pin/DCM2 output control input pin |
| 16 | MD2/DC12 | STM excitation mode switching pin/DCM1 output control input pin |
| 17 | MD1/DC11 | STM excitation mode switching pin/DCM1 output control input pin |
| 18 | DM | Drive mode (STM/DCM) switching pin |
| 19 | OE | Output enable signal input pin |
| 20 | ST | Chip enable pin |
| 21 | VREF | Constant current control reference voltage input pin |
| 22 | GND | Signal system ground |
| 23, 24 | OUT2B | Channel 2 OUTB output pin |
| 25 | PGND2 | Channel 2 Power system ground |
| 28, 29 | VM2 | Channel 2 motor power supply connection pin |
| 30, 31 | RF2 | Channel 2 current-sense resistor connection pin |
| 32, 33 | OUT2A | Channel 2 OUTA output pin |
| 34, 35 | OUT1B | Channel 1 OUTB output pin |
| 36, 37 | RF1 | Channel 1 current-sense resistor connection pin |
| 38, 39 | VM1 | Channel 1 motor power supply pin |
| 42 | PGND1 | Channel 1 Power system ground |
| 43, 44 | OUT1A | Channel 1 OUTA output pin |
| $\begin{aligned} & 26,27 \\ & 40,41 \end{aligned}$ | NC | No Connection <br> (No internal connection to the IC) |

Equivalent Circuits

| Pin No. | Pin | Equivalent Circuit |
| :---: | :---: | :---: |
| $\begin{gathered} 6 \\ 7 \\ 10 \\ 13 \\ 14 \\ 15 \\ 16 \\ 17 \\ 18 \\ 19 \end{gathered}$ | ATT2 <br> ATT1 <br> EMM <br> RST <br> STP/DC22 <br> FR/DC21 <br> MD2/DC12 <br> MD1/DC11 <br> DM <br> OE |  |
| 20 | ST |  |
| $\begin{gathered} 23,24 \\ 25 \\ 28,29 \\ 30,31 \\ 32,33 \\ 34,35 \\ 36,37 \\ 38,39 \\ 42 \\ 43,44 \end{gathered}$ | OUT2B <br> PGND2 <br> VM2 <br> RF2 <br> OUT2A <br> OUT1B <br> RF1 <br> VM1 <br> PGND1 <br> OUT1A |  |

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| Pin No. | Pin | Equivalent Circuit |
| :---: | :---: | :---: |
| $\begin{aligned} & 1 \\ & 2 \\ & 3 \\ & 4 \end{aligned}$ | VG <br> VM <br> CP2 <br> CP1 |  |
| 21 | VREF |  |
| 5 | VREG5 |  |
| $\begin{gathered} 8 \\ 12 \end{gathered}$ | EMO <br> MONI |  |

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Pin No.

## Description of operation

## 1. Input Pin Function

1-1) Chip enable function
This IC is switched between standby and operating mode by setting the ST pin. In standby mode, the IC is set to power-save mode and all logic is reset. In addition, the internal regulator circuit and charge pump circuit do not operate in standby mode.

| ST | Mode | Internal regulator | Charge pump |
| :---: | :---: | :---: | :---: |
| Low or Open | Standby mode | Standby | Standby |
| High | Operating mode | Operating | Operating |

1-2) Drive mode switching pin function
The IC drive mode is switched by setting the DM pin. In STM mode, stepper motor channel 1 can be controlled by the CLK-IN input. In DCM mode, DC motor channel 2 or stepper motor channel 1 can be controlled by parallel input. Stepper motor control using parallel input is Full-step or Half-step full torque.

| DM | Drive mode | Application |
| :---: | :---: | :---: |
| Low or Open | STM mode | Stepper motor channel 1 (CLK-IN) |
| High | DCM mode | DC motor channel 2 or stepper motor channel 1 (parallel) |

2. STM mode (DM = Low or Open)

## 2-1) STEP pin function

The excitation step progresses by inputting the step signal to the STP pin.

| Input |  | Operating mode |
| :---: | :---: | :---: |
| ST | STP |  |
| Low | $*$ | Standby mode |
| High |  |  |
| High |  |  |

## 2-2) Excitation mode setting function

The excitation mode of the stepper motor can be set as follows by setting the MD1 pin and the MD2 pin.

| MD1 | MD2 | Micro-step resolution <br> (Excitation mode) |  | Initial position |  |
| :---: | :---: | :--- | :---: | :---: | :---: |
|  |  | Channel 1 | Channel 2 |  |  |
| Low | Low | Full step (2 phase excitation) | $100 \%$ | $-100 \%$ |  |
| High | Low | Half step (1-2 phase excitation) <br> full torque | $100 \%$ | $0 \%$ |  |
| Low | High | Half step (1-2 phase excitation) | $100 \%$ | $0 \%$ |  |
| High | High | Quarter step <br> (W1-2 phase excitation) | $100 \%$ | $0 \%$ |  |

This is the initial position of each excitation mode in the initial state after power-on and when the counter is reset.

## 2-3) Positional detection monitor function

The MONI position detection monitoring pin is of an open drain type.
When the excitation position is in the initial position, the MONI output is placed in the ON state.
(Refer to "2-12.Examples of current waveforms in each micro-step mode.")

## 2-4)Constant-current control reference voltage setting function

This IC does the PWM fixed current chopping control of the current of the motor by the automatic operation in setting the output current. The output current in which a fixed current is controlled by the following calculation type is set by the resistance connected between the voltage and RF-GND being input to the VREF pin.

IOUT=(VREF/5)/RF resistance
*The above-mentioned, set value is an output current of each excitation mode at $100 \%$ time.
VREF input voltage attenuation function

| ATT1 | ATT2 | Current setting reference voltage attenuation ratio |
| :---: | :---: | :---: |
| Low | Low | $100 \%$ |
| High | Low | $66.7 \%$ |
| Low | High | $50 \%$ |
| High | High | $33.3 \%$ |

The output ammeter calculation type when the attenuation function of the VREF input voltage is used is as follows.
IOUT $=(\mathrm{VREF} / 5) \times($ Attenuation ratio $) /$ RF resistance
$($ Example $)$ When VREF $=1.5 \mathrm{~V}$, setting current ratio $=100 \%[(A T T 1$, ATT2 $)=($ Low, Low $)]$ and RF resistor $=0.22 \Omega$, the following output current flows :

$$
\text { IOUT }=1.5 \mathrm{~V} / 5 \times 100 \% / 0.22 \Omega=1.36 \mathrm{~A}
$$

Under such a condition, when assuming $($ ATT1, ATT2 $)=($ High, High $)$.
IOUT $=1.36 \mathrm{~A} \times 33.3 \%=453 \mathrm{~mA}$
The power saving can be done, and attenuating the output current when the motor energizes maintenance.

## 2-5) Input Timing



TstepH/TstepL : Clock H/L pulse width (min 500ns)
Tds : Data set-up time (min 500 ns )
Tdh : Data hold time (min 500ns)

## 2-6) Blanking period

If, when exercising PWM constant-current chopping control over the motor current, the mode is switched from decay to charge, the recovery current of the parasitic diode may flow to the current sensing resistance, causing noise to be carried on the current sensing resistance pin, and this may result in erroneous detection. To prevent this erroneous detection, a blanking period is provided to prevent the noise occurring during mode switching from being received. During this period, the mode is not switched from charge to decay even if noise is carried on the current sensing resistance pin.
This IC's blanking period is fixed at about $1 \mu \mathrm{~s}$ in STM mode ( $2 \mu \mathrm{~s}$ in DCM mode).

2-7) Reset function

| RST | Operating mode |
| :---: | :---: |
| Low | Normal operation |
| High | Reset state |



When the RST pin is set High, the output excitation position is forced to the initial state, and the MONI output enters ON a state. When RST is set Low after that, the excitation position proceeds to the next STEP input.

2-8) Output enable function

| OE | Operating mode |
| :---: | :---: |
| High | Output OFF |
| Low | Output ON |



When the OE pin is set High, the output is forced OFF and goes to high impedance.
However, the internal logic circuits are operating, so the excitation position proceeds when the STEP signal is input to the STP pin. Therefore, when OE is returned to Low, the output level conforms to the excitation position proceeded by the STEP input.

2-9) Forward/reverse switching function

| FR | Operating mode |
| :---: | :---: |
| Low | Clockwise (CW) |
| High | Counter-clockwise (CCW) |



The internal D/A converter proceeds by one bit at the rising edge of the input STEP pulse. In addition, CW and CCW mode are switched by setting the FR pin.
In CW mode, the channel 2 current phase is delayed by $90^{\circ}$ relative to the channel 1 current. In CCW mode, the channel 2 current phase is advanced by $90^{\circ}$ relative to the channel 1 current.

## 2-10) Setting the chopping frequency

For constant-current control, chopping operation is made with the frequency determined by the external resistor (connected to the RCHOP pin).
The chopping frequency to be set with the resistance connected to the RCHOP pin (pin 11) is as shown below.


2-11) Output current vector locus (one step is normalized to 90 degrees)


Setting current ration in each micro-step mode

| STEP | Quarter-step (\%) |  | Half-step (\%) |  | Half-step full torque (\%) |  | Full-step (\%) |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Channel 1 | Channel 2 | Channel 1 | Channel 2 | Channel 1 | Channel 2 | Channel 1 | Channel 2 |
| $\theta 0$ | 0 | 100 | 0 | 100 | 0 | 100 |  |  |
| $\theta 1$ | 35 | 90 |  |  |  |  |  |  |
| $\theta 2$ | 70 | 70 | 70 | 70 | 100 | 100 | 100 | 100 |
| $\theta 3$ | 90 | 35 |  |  |  |  |  |  |
| $\theta 4$ | 100 | 0 | 100 | 0 | 100 | 0 |  |  |

2-12) Examples of current waveforms in each micro-step mode
Full step (CW mode)


Half step full torque (CW mode)


Half step (CW mode)


Quarter step (CW mode)


## 2-13) Current control operation specification

(Sine wave increasing direction)

(Sine wave decreasing direction)


In each current mode, the operation sequence is as described below :

- At rise of chopping frequency, the CHARGE mode begins.(The section in which the CHARGE mode is forced regardless of the magnitude of the coil current (ICOIL) and set current (IREF) exists for $1 / 16$ of one chopping cycle.)
- The coil current (ICOIL) and set current (IREF) are compared in this forced CHARGE section.

When (ICOIL<IREF) state exists in the forced CHARGE section ;
CHARGE mode up to ICOIL $\geq$ IREF, then followed by changeover to the SLOW DECAY mode, and finally by the FAST DECAY mode for the $1 / 16$ portion of one chopping cycle.
When (ICOIL<IREF) state does not exist in the forced CHARGE section;
The FAST DECAY mode begins. The coil current is attenuated in the FAST DECAY mode till one cycle of chopping is over.
Above operations are repeated. Normally, the SLOW (+FAST) DECAY mode continues in the sine wave increasing direction, then entering the FAST DECAY mode till the current is attenuated to the set level and followed by the SLOW DECAY mode.

## 3.DCM Mode (DM-High)

## 3-1) DCM mode output control logic

| Parallel input |  | Output |  | Mode |
| :---: | :---: | :---: | :---: | :---: |
| DC11 (21) | DC12 (22) | OUT1 (2)A | OUT1 (2) B |  |
| Low | Low | OFF | OFF | Standby |
| High | Low | High | Low | CW (Forward) |
| Low | High | Low | High | CCW (Reverse) |
| High | High | Low | Low | Brake |

## 3-2) Reset function

| RST | Operating mode | MONI |
| :---: | :---: | :---: |
| High or Low | Reset operation not performed | High output |

The reset function does not operate in DCM mode. In addition, the MONI output is High, regardless of the RST pin state.

## 3-3) Output enable function

| OE | Operating mode |
| :---: | :---: |
| High | Output OFF |
| Low | Output ON |

When the OE pin is set High, the output is forced OFF and goes to high impedance. When the OE pin is set Low, output conforms to the control logic.

## 3-4) Current limit control time chart

When the current of the motor reaches up to the limit current by setting the current limit, this IC does the short brake control by the automatic operation so that the current should not increase more than it.


Moreover, the voltage impressed to the terminal VREF can be switched to the setting of four stages by the state of two input of ATT1 and ATT2.
VREF input voltage attenuation function

| ATT1 | ATT2 | Current setting reference voltage |
| :---: | :---: | :---: |
| Low | Low | $100 \%$ |
| High | Low | $66.7 \%$ |
| Low | High | $50 \%$ |
| High | High | $33.3 \%$ |

The output ammeter calculation type when the attenuation function of the VREF input voltage is used is as follows.
IOUT $=(\mathrm{VREF} / 5) \times($ Attenuation ratio $) /$ RF resistance
(Example) When VREF $=1.5 \mathrm{~V}$, setting current ratio $=100 \%[(\mathrm{ATT} 1, \mathrm{ATT} 2)=($ Low, Low $)]$ and RF resistor $=0.22 \Omega$, the following output current flows :

$$
\text { IOUT }=1.5 \mathrm{~V} / 5 \times 100 \% / 0.22 \Omega=1.36 \mathrm{~A}
$$

Under such a condition, when assuming $($ ATT1, ATT2 $)=($ High, High $)$.

$$
\mathrm{IOUT}=1.36 \mathrm{~A} \times 33.3 \%=453 \mathrm{~mA}
$$

3-5) Examples of current waveform in each micro-step mode when stepper motor parallel input control Full step (CW mode)


Half step full torque (CW mode)

DC11

DC21

DC12

DC22

11

I2


## 4.Output short-circuit protection circuit

This output short protection circuit that makes the output a standby mode to prevent the thing that IC destroys when the output is short-circuited by a voltage short and the earth fault, etc., and turns on the warning output to IC is built into.

## 4-1) Output short-circuit protection mode switching function

Output short-circuit protection mode of IC can be switched by the setting of EMM pin.

| EMM | State |
| :---: | :---: |
| Low or Open | Latch method |
| High | Auto reset method |

## 4-2) Latch method

In the latch mode, the output is turned off when the output current exceeds the detection current, and the state is maintained.
The output short protection circuit starts operating so that IC may detect a short output. When the short-circuit is the consecutive between internal timers $(\approx 4 \mu \mathrm{~s})$, the output where the short-circuit is first detected is turned off. Even if the following time (Tcem) of the timer latch is exceeded, the output is turned ON again, and afterwards, when the short-circuit is detected, all the outputs of correspondence ch side are still switched to the standby mode, and the state is maintained. This state is released by making it to $\mathrm{ST}=$ ="L".


## 4-3) Automatic return method

In the automatic return mode, the output wave type changes into the switching wave type when the output current exceeds the detection current.
The short-circuit detection circuit operates when a short output is detected as well as the latch method. The output is switched to the standby mode when the operation of the short-circuit detection circuit exceeds the following time (Tcem) of the timer latch, and it returns to the turning on mode again after 2 ms (TYP). At this time, the above-mentioned switching mode is repeated when is still in the over current mode until the over current mode is made clear.

## 4-4) Abnormal state warning output pin

When IC operates the protection circuit detecting abnormality, the EMO pin has been installed as a terminal that outputs this abnormality to CPU side. This pin is an open drain output, and if abnormality is detected, the EMO output becomes ( $\mathrm{EMO}=$ "L") of ON.
EMO pin enters on a state in the following.

- When a voltage short, the earth fault or the load is short-circuited and the output short-circuit protection circuit operates, the output pin
- When the junction temperature of IC rises, and the overheating protection circuit operates


## 4-5) Timer latch time (Tcem)

The time to output OFF when an output short-circuit occurs can be set by the capacitor connected between the CEM pin and GND. The capacitor (Ccem) value can be determined as follows :
Timer latch : Tcem
Tcem $\approx \mathrm{C} \times \mathrm{V} / \mathrm{I}[\mathrm{sec}]$
V $:$ Threshold voltage of comparator TYP 1 V
I $:$ CEM charge current TYP $10 \mu \mathrm{~A}$

## 5.Thermal shutdown function

The thermal shutdown circuit is included, and the output is turned off when junction temperature Tj exceeds $180^{\circ} \mathrm{C}$ and the abnormal state warning output is turned on at the same time.
When the temperature falls hysteresis level, output is driven again (automatic restoration)
The thermal shutdown circuit doesn't guarantee protection of the set and the destruction prevention of IC, because it works at the temperature that is higher than rating $\left(\operatorname{Tjmax}=150^{\circ} \mathrm{C}\right)$ of the junction temperature

$$
\begin{aligned}
\mathrm{TTSD} & =180^{\circ} \mathrm{C}(\text { typ }) \\
\triangle \mathrm{TSD} & =40^{\circ} \mathrm{C}(\text { typ })
\end{aligned}
$$

## 6.Charge Pump Circuit

When the ST pin is set High, the charge pump circuit operates and the VG pin voltage is boosted from the VM voltage to the VM + VREG5 voltage. If the VG pin voltage is not boosted to VM+4V or more, the output pin cannot be turned on. Therefore it is recommended that the drive of motor is started after the time has passed tONG or more.


VG Pin Voltage Schematic View

## Application Circuits

1. Stepper motor driver application circuit example(DM="L")


Each constant setting type in the example of the above-mentioned circuit is as follows.
When setting current ratio $=100 \%, \mathrm{VREF}=1.5 \mathrm{~V}$, the following output current flows :

$$
\begin{aligned}
\text { IOUT } & =\text { VREF } / 5 / \text { RF resistance } \\
& =1.5 \mathrm{~V} / 5 \times 100 \% / 0.22 \Omega=1.36 \mathrm{~A}
\end{aligned}
$$

Chopping frequency setting.
$62.5 \mathrm{kHz}(\mathrm{RCHOP}=20 \mathrm{k} \Omega)$
Time of timer latch when output is short-circuited
Tcem $=$ Ccem $*$ Vtcem/Icem

$$
=100 \mathrm{pF} * 1 \mathrm{~V} / 10 \mu \mathrm{~A}=10 \mu \mathrm{~s}
$$

2. DC motor driver application circuit example


Each constant setting type in the example of the above-mentioned circuit is as follows.
When setting current LIMIT $=100 \%$, VREF $=1.5 \mathrm{~V}$, the following output current flows :

$$
\begin{aligned}
\text { Ilimit } & =\mathrm{VREF} / 5 / \mathrm{RF} \text { resistance } \\
& =1.5 \mathrm{~V} / 5 \times 100 \% / 0.22 \Omega=1.36 \mathrm{~A}
\end{aligned}
$$

Chopping frequency setting.
$62.5 \mathrm{kHz}(\mathrm{RCHOP}=20 \mathrm{k} \Omega)$
Time of timer latch when output is short-circuited

$$
\begin{aligned}
\text { Tcem } & =\text { Ccem } * \text { Vtcem/Icem } \\
& =100 \mathrm{pF} * 1 \mathrm{~V} / 10 \mu \mathrm{~A}=10 \mu \mathrm{~s}
\end{aligned}
$$

ORDERING INFORMATION

| Device | Package | Shipping (Qty / Packing) |
| :---: | :---: | :---: |
| LV8740V-TLM-E | SSOP44J (275mil) <br> (Pb-Free) | $2000 /$ Tape \& Reel |
| LV8740V-MPB-E | SSOP44J (275mil) <br> (Pb-Free) | $30 /$ Fan-Fold |

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