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# Isolated SiC MOSFET Gate Driver Evaluation Board User's Manual

# NCP51752 EVBUM

#### Introduction

This user guide supports the evaluation board for the NCP51752. It should be used in conjunction with the NCP51752 datasheets as well as **onsemi**'s application notes and technical support team. Please visit **onsemi**'s website at <u>www.onsemi.com</u>.

This document describes the proposed solution for an isolated single channel gate driver using the NCP51752 family. This user's guide also includes information regarding operating procedures, input/output connections, an electrical schematic, printed circuit board (PCB) layout, and a bill of material (BOM) for the evaluation board.

These evaluation boards can be used to evaluate:

- NCP51752xyDR2G
- NCV51752xyDR2G

#### Description

The NCP51752 is a family of isolated single-channel gate driver with +4.5 A / -9 A source and sink peak current respectively. They are designed for fast switching to drive power MOSFETs and SiC MOSFET power switches. The NCP51752 offers short and matched propagation delays.

For improved reliability, dV/dt immunity and even faster turn-off, the NCP51752 has an embedded negative bias rail mechanism between GND2 and VEE pins.

The NCP51752 offers other important protection function such as independent under-voltage lockout for both-side driver. It's Vcc UVLO threshold with referenced to GND2 for true UVLO level regardless of voltage level between GND2 and VEE pin. The NCP51752 is available in a 4 mm SOIC-8 package and can support isolation voltage up to  $3.75 \text{ kV}_{RMS}$ .

#### **Key Features**

- Feature Options
  - V<sub>CC</sub> UVLO Referenced to GND2
  - Built-in Negative Bias between GND2 and V<sub>EE</sub> pins
- 3 V to 20 V Input Supply Voltage
- Output Supply Voltage from 6.5 V to 30 V with 6 V and 8 V for MOSFET, 12 V and 17 V for SiC MOSFET, Threshold.
- 4.5 A Peak Source, 9 A Peak Sink Output Current Capability
- Two Input Configurations with Negative 5 V Handling Capability on Input Pins
- Minimum CMTI of 200 V/ns dV/dt
- Propagation Delay Typical 36 ns with
  5 ns Max Delay Matching
- Available Package Footprint
  - Type-A : TO-3P, TO-247-3L, D-PAK, and D2PAK
  - ◆ Type-B : TO-247-4L
  - Type-C : D2PAK-7L



(A) For TO-247-3L



(B) For D2PAK Type-A



Type-B



Туре-С

Figure 1. Evaluation Board Picture

PIN CONNECTIONS



#### FUNCTIONAL BLOCK DIAGRAM



### **EVALUATION BOARD OPERATION**

This section describes how to operate the NCP51752 family evaluation board (EVB). Make external connections to the NCP51752 EVB using either the installed test–points or by installing wires into the connectors. The main connections that must be made to the EVB are the analog supply voltage, input signal, and output load and monitoring equipment.

#### Features

- Evaluation board for the NCP51752 product family in the narrow body (4 mm) SOIC-8 package.
- 3 V to 20 V Input Supply Voltage
- Output Supply Voltage from 6.5 V to 30 V with 6 V and 8 V for MOSFET, 12 V and 17 V for SiC, Threshold.
- 4.5 A and 9 A source/sink current driving capability
- TTL-compatible inputs and allowable input voltage up to V<sub>DD</sub> with for IN+, and IN- pins
- 3-position header with for IN+, and IN- pins
- Support to test with MOSFETs, and SiC MOSFETs with connection to external power stage

#### **Power and Ground**

NOTE: Connecting the all power supplies in reverse polarity (backwards) will instantly device when power is turned on and device damage can result.

The primary side of the EVB (V<sub>DD</sub>) operates from a single 3 V to 20 V power supply and connected via J6. Test point (TP8) is available for monitoring the primary–side power supply. The EVB provides connections for evaluating the output side (V<sub>CC</sub>, V<sub>EE</sub>) power supply from a minimum 6.5 V to maximum 30 V with respect to  $V_{EE}$  pin.

The negative bias (between GND2 and  $V_{EE}$ ) is generated itself from  $V_{CC}$  supply voltage.  $V_{cc}$  can be supplied at J4.

 $V_{CC}$  and  $V_{EE}$  can be monitored with respect to GND2 (TP4) via TP3 and TP5, respectively.

The V<sub>CC</sub> pin should be bypassed with a capacitor with a value of at least ten times the gate capacitance of the power device, and over 100 nF and located as close to the device as possible for the purpose of decoupling. A low ESR, ceramic surface mount capacitor is necessary. We had recommends using at least 2 capacitors; an over 100 nF ceramic surface-mount capacitor, and another a tantalum or electrolytic capacitor of few microfarads added in parallel (e.g. C6, C7, C9 and C10).

#### Input and Output

- 1. Connection of primary-side power supply to the V<sub>DD</sub> pin connector [J6-1, and J6-2].
- 2. Connection of secondary-side power supply to the V<sub>CC</sub> connector [J4-1, and J4-2].
- 3. Connection of non-inverting input signal (IN+) to the SIGNAL connector [J2-1, or J1].
- 4. Connection of inverting input signal (IN–) to the SIGNAL connector [J2–2, or J3].

#### **Evaluation Board Jumper Setting**

#### Table 1. EVB JUMPER SETTING

Jumper	Jumper Setting Options for Input Signals (IN+ & IN-)		
J9–IN+	Option1	Jumper not installed, IN+ signal provided by external signal and this pin is default LOW if left open	
	Option2	Jumper on J9-INP-2 and J9-INP-3 set IN+ for LOW state	Option1
	Option3	Jumper on J9-INP-2 and J9-INP-1 set IN+ for HIGH state	
J8-IN-	Option1	Jumper not installed, IN- signal provided by external signal and this pin is default HIGH if left open	
	Option2	Jumper on J8-INN-2 and J8-INN-3 set IN- low for LOW state	Option2
	Option3	Jumper on J8-INN-2 and J8-INN-1 set IN- high for HIGH state	

#### **Evaluation Board Setting before Power Up**

- 1. If the ENABLE mode is used , IN- pin (PIN3) should be connected to GND1(PIN4) through a wire-bridge between pin 2 and pin 3 of J8.
- 2. If using the DISABLE mode, should be connect IN- pin (PIN3) to V<sub>DD</sub> pin (PIN1) through a wire-bridge between pin 1 and pin 2 of J8 or this pin is default HIGH if left open.

# **Bench Setup**

The bench setup diagram includes the function generator, power supplies and oscilloscope connections.

Follow the connection procedure below and use Figure 2 as a reference.

- Make sure all the output of the function generator, power supplies are disabled before connection.
- Function generator channel–B channel (CH2) applied on IN+ (J1 or J2 pin–1) ↔ TP1 as seen in Figure 2.
- Function generator channel–A channel (CH1) applied on IN– (J3 or J2 pin–2) ↔ TP2 as seen in Figure 2.
- Power supply #1: positive node applied on J6 pin-1 (TP8), and negative node applied on J6 pin-2 (TP7).

- Power supply #2: positive node applied on J4 pin-1 (TP3), negative node connected directly to J4 pin-2 (TP5).
- Oscilloscope channel−A probes TP1 (IN+) ↔ TP7 (GND1), smaller measurement loop is preferred
- Oscilloscope channel−B probes TP9 (OUT) ↔ TP5 (VEE), smaller measurement loop is preferred.
- Oscilloscope channel−C probes TP9 (GND2) ↔ TP5 (VEE), smaller measurement loop is preferred.
- Oscilloscope channel−D probes TP9 (VCC) ↔ TP5 (VEE), smaller measurement loop is preferred.



Figure 2. Bench Setup Diagram and Configuration

#### Power–Up and Power Down Procedure

#### Power (V<sub>DD</sub> & V<sub>CC</sub>) Up Procedure

- 1. Enable power supply through pin1 of J6  $V_{DD}$  connector in primary-side
- 2. Enable power supply through pin1 of J4  $V_{CC}$  connector in secondary–side Measure the quiescent current of  $V_{CC}$  on DMM2 ranges from 0.5 mA to approximately 1.0 mA if everything is set correctly;
- 3. Measure the voltage of GND2 with respect to  $V_{EE}$  pin on DMM3 if everything is set correctly;

- Enable the function generator, two-channel outputs: channel-A (Default set to LOW or Floating for IN- pin) and channel-B for IN+ pin;
- 5. There will be:
  - a. Stable pulse output on the channel–A, channel–B, channel–C and channel–D in the oscilloscope
  - b. Scope frequency measurement is the same with function generator output;
  - c. DMM #2 read measurement results should be around 4 mA  $\pm$  1 mA under no load conditions.



CH1: IN+, CH2: OUT\_VEE, CH3: GND2\_VEE, CH4: VCC\_VEE

Figure 3. Experimental Waveform of Input to Output

Power (V<sub>DD</sub> & V<sub>CC</sub>) Down Procedure

- 1. Disable function generator
- 2. Disable power supply of  $V_{CC}$  in secondary-side
- 3. Disable power supply of  $V_{DD}$  in primary-side
- 4. Disconnect cables and probes

Figure 4 shows the NCP51752 application schematic of each evaluation board to cope with various package types.



(A) Schematic of Type-A for TO-3P, TO-247, and TO-252 (D-PAK), and TO-263 (D2PAK) Package



(B) Schematic of Type-B for TO-247-4L Package



(C) Schematic of Type-C for D2PAK-7L Package



# List of Test Point

Table 2 shows the test point list of NCP51752 for an evaluation board (EVB).

#### Table 2. LIST OF TEST POINT

TP	Reference	Description
TP1	INP	Non-inverting Logic Input with internal pull-down resistor to GND1.
TP2	INN	Inverting Logic Input with internal pull-up resistor to V <sub>DD</sub> .
TP3	VCC	Positive Output Supply Rail.
TP4	GND2	Gate-drive common pin. Connect this pin to the MOSFET source. $V_{CC}$ UVLO threshold referenced to GND2.
TP5	VEE	Negative output supply rail.
TP6	DRAIN	Drain of Power Device
TP7	GND1	Ground Input-side. (all signals on input-side are referenced to this pin)
TP8	VDD	Input-side Supply Voltage. It is recommended to place a bypass capacitor from $V_{DD}$ to GND1.
TP9	OUT	Gate Drive Output.
TP10	GATE	Gate of Power Device

#### **Electrical Specifications**

Table 3 shows the recommended operating conditions of NCP51752 for an evaluation board.

#### Table 3. ELECTRICAL SPECIFICATIONS

Rating	Symbol	Min.	Max.	Unit	
Power Supply Voltage – Input side	V <sub>DD</sub>	3.0	20	V	
Power Supply Voltage with respect	6-V UVLO Version	V <sub>CC</sub>	6.5	30	V
to GND2– Driver side	8-V UVLO Version		9.5	30	V
	12-V UVLO Version		13.5	30	V
	17-V UVLO Version	1	18.5	30	V
Negative Supply Voltage	VEE – GND2	-6	0	V	
Logic Input Voltage at pins IN+, and IN	V <sub>IN</sub>	0	V <sub>DD</sub>	V	
Ambient Temperature	T <sub>A</sub>	- 40	+125	°C	
Junction Temperature	TJ	-40	+125	°C	
Common Mode Transient Immunity	CMTI	200		kV/μs	

#### Bill of Material (BOM)

Table 4 shows the bill of material (BOM) of NCP51752 for an evaluation board.

Designator	Quantity	Description	Value	Footprint	Manufacturer
C1, C2	2	Capacitor, Ceramic, X7R	10 pF, 50 V	SMD 2012	Yageo
C4	1	Capacitor, Ceramic, X7R	0.22 uF, 50 V	SMD 2012	Yageo
C6	1	Capacitor, Ceramic, X7R	4.7 uF, 50 V	SMD 3216	Yageo
C7	1	Capacitor, Ceramic, X7R	10 uF, 50 V	SMD 2012	Yageo
C3 , C9, C10	3	Capacitor, Ceramic, X7R	10 uF, 50 V	SMD 3216	Yageo
C5	0	Capacitor, Ceramic, X7R	1 nF, 50 V (DNP)	SMD 2012	Yageo
C8	1	Capacitor, Ceramic, X7R	470 nF, 50 V	SMD 2012	Yageo
D1	1	Fast Recovery Diode	50V	DSN-2	
J7	1	Connector	3 Pin		CAMDEN
J2,J4, J6	3	Connector	2 Pin		CAMDEN
J1, J3	2	BNC Connector		SMB	Johnson / Cinch
J8, J9	2	Header	3 Pin	Pltch 2.54mm	JINLING
J5, J10	0	Connector	DNP		
M1		Power Switch (DNP)	For only Type–A	TO-247-3L	onsemi
			For only Type–B	TO-247-4L	onsemi
			For only Type-C	TO-2637L	onsemi
R1, R2	2	Resistors	51 Ω	SMD 2012	Rohm
R3	1	Resistors	<b>10</b> Ω	SMD 2012	Rohm
R4	1	Resistors	3.3 Ω	SMD 2012	Rohm
R5	1	Resistor	10 kΩ	SMD 2012	Rohm
U1	1	Gate driver	NCP51752CDDR2G	8 SOIC-NB	onsemi

#### Table 4. BILL OF MATERIAL

#### Input Stage

The input pins of NCP51752 is based on a TTL-compatible input-threshold logic that is independent of the V<sub>DD</sub> supply voltage for IN+, and IN- pins. The logic level compatible input provides a typically HIGH and LOW threshold of 1.63 V and 1.08 V respectively. The input signal pins impedance is 125 k $\Omega$  typically and the IN+ pin is pulled to GND1 pin and IN- pin is pulled to V<sub>DD</sub> pin as shown in Figure 5. For non-inverting input logic signal is applied to IN+ while the IN- input can be used as an enable function. If IN- is pulled HIGH, the driver output remains LOW state, regardless of the state of IN+. To enable the driver output, IN- should be tied to GND1 pin through a few ten k $\Omega$  resistor (e.g.10 k $\Omega$ ) or can be used as an active LOW enable pull down.

- Non-inverting input IN+ controls the driver output while inverting input IN- is set to LOW
- Inverting input IN- controls the driver output while non-inverting input IN+ is set to HIGH.

And we recommends an RC network is to be added on the PWM input pins, IN+ and IN-, for reducing the impact of system noise and ground bounce, for example, 51  $\Omega$  (R1, and R2) with 10 pF (C1, and C2) is an acceptable choice as

shown in Figure 5. IN+, and IN– signal can be monitored via TP1, and TP2, respectively.



Figure 5. Recommended Input Circuit

#### **Output Stage**

The output stage is able to sink/source typically around +4.5 A/-9.0 A at 25°C for the NCP51752.

The output voltage swing between  $V_{CC}$  and  $V_{EE}$  provides rail–to–rail operation with respect to GND2 pin when used

the GND2 pin connect to the emitter or source of power device as shown in Figure 6.

The power device is turned off with a negative voltage on the gate with respect to the source pin. This configuration prevents the power device from unintentionally turning on because of current induced from the Miller effect.

The EVB comes populated with a 1–nF load (C5) on the output side. The OUT and GATE can be monitored directly via TP9 and TP10, respectively.

The Type–A EVB allows for evaluation of the device with an MOSFET load in either of the standard TO–3P,

TO-247-3L, and TO-252 (D-PAK), and TO-263 (D2PAK) footprints. The Type-B EVB allows for evaluation of the device with an MOSFET load in the standard TO-247-4L.

The Type–C EVB allows for evaluation of the device with an MOSFET load in the standard D2PAK–7L.

During evaluation with an SiC MOSFET load, the pre-installed capacitive load (C5) can be disconnected from the output. The EVB provides an additional connection (J5) for applying an external power supply to the MOSFET Drain. The EVB is not intended for high voltage testing and the voltage applied to J5 should be limited to  $50 V_{DC}$ .



Figure 6. Schematic of Output Stage

#### **Functional Mode Table**

Table 5 shows the functional modes for the NCP51752 assuming  $V_{DD}$  and  $V_{CC}$  are in the recommended ranges for an evaluation board.

#### **Table 5. FUNCTIONAL MODES**

INF	PUT	GATE DRIVE OUTPUT		
IN+	IN-	OUT		
LOW	Х	Low		
Х	HIGH	Low		
HIGH	LOW	High		

NOTE: X = Don't care

#### PERFORMANCE OF EVALUATION BOARD

This section describes application guidance and operation of the NCP51752 for an evaluation board (EVB) include key functions.

#### **Protection Function**

The NCP51752 provides important protection functions such as independent under-voltage lockout for gate driver. Figure 7 shows an overall input to output timing diagram. Under-Voltage Lockout protection on the primary- and secondary-sides power supplies events in the **CASE-A**, **B** and **C** and the gate driver output (OUT) is immediately turn-off when an inverting input signal (IN-) is HIGH regardness non-inverting input (IN+) signal state in the **CASE-D**. The negative bias control circuit is enabled after the V<sub>CC</sub> power-up delay time, t<sub>VPOR to OUT</sub>, during initial V<sub>CC</sub> start-up or after POR event.



Figure 7. Overall Operating Waveforms Definitions

Figure 8 shows an experimental result of enable function that the inverting input pin (IN–) voltage goes to HIGH state in normal operation, the gate driver output is turned–off

immediately regardless non-inverting input (IN+) pin signal state.



Figure 8. Experimental Waveforms of Enable Function

#### Under-Voltage Lockout Protection V<sub>DD</sub>

The NCP51752 provides the Under-Voltage Lockout (UVLO) protection function for  $V_{DD}$  in primary-side.

As test result, the  $V_{DD}$  UVLO turn–on and off threshold voltages are around 2.8 V and 2.63 V respectively as shown in Figure 9.



CH1: IN+, CH2: OUT\_VEE, CH3: VDD



#### Under-Voltage Lockout Protection V<sub>CC</sub>

The NCP51752 provides the Under–Voltage Lockout (UVLO) protection function of gate drive output in secondary–side. As test result of NCP51752DA variant (For example, VCC UVLO Threshold = 17 V and voltage between GND2 and VEE = 2 V), the V<sub>CC</sub> UVLO turn–on

and off threshold voltages are around 19.6 V and 19.1 V with respect to VEE respectively as shown in Figure 10.

Therefore, the  $V_{CC}$  true UVLO threshold turn-on and off threshold voltages are 17.6 V and 17.1 V with respect to GND2 respectively due to GND2\_VEE bias voltage is 2 V.



CH1: IN+, CH2: OUT\_VEE, CH3: VCC\_VEE

Figure 10. Experimental Waveforms of  $V_{CC}$  Under-Voltage Lockout Protection

#### Negative Bias Control Function

The NCP51752 provides a simple way to generate negative bias between GND2 and VEE pin in the gate drive loop. This negative bias is very useful in case of PCB layout and/or package leads generating high ringing in power transistor Vgs.

The NCP51752 offers different options to generate 2 V, 3 V, 4 V and 5 V between GND2 and VEE pins to accommodate all configurations and types of power transistors (SiC MOSFETs).

Figure 11 shows the timing chart of the negative bias control function.



Figure 11. Timing Chart of Negative Bias Control

Figure 12 shows the experimental result of negative bias control when supplying the  $V_{CC}$  of the NCP51752CD variant (For example, VCC UVLO Threshold = 12 V and voltage between GND2 and VEE = 5 V).

The C<sub>GND2</sub> capacitor(C8) rapidly charged to 80% of target threshold after VCC power up delay time ( $t_{VPOR to}$  <sub>OUT</sub>) from V<sub>CC</sub> = 4.7 V (POR) and then slowly charged up to 100% of target threshold.



CH1: IN+, CH2: OUT\_VEE, CH3: VCC\_VEE, CH4:GND2\_VEE

Figure 12. Experimental Waveforms of Negative Bias Control

#### **Output Driving Current Capability**

The experimental result of source and sink peak currents driving capability around 5.0 A and 9.95 A respectively at  $V_{CC} = 20$  V and room temperature as shown in Figure 13 (a).

And source and sink peak currents driving capability around 6.0 A and 10.5 A respectively at  $V_{CC}$  = 30 V and room temperature as shown in Figure 13 (b).



(a) At V<sub>DD</sub> = 5 V, and V<sub>CC</sub> = 20 V

(b) At V<sub>DD</sub> = 5 V, and V<sub>CC</sub> = 30 V

CH1: IN+, CH2: OUT\_VEE, CH3: OUT Current



#### ESD Structure

Figure 14 shows the multiple diodes related to an ESD protection components of NCP51752. This illustrates the absolute maximum rating for the device.



Figure 14. ESD Structure

#### **Printed Circuit Board**

Figure 15 shows the photograph of the NCP51752 evaluation board for the Type-A. This EVB allows for



(A) For TO-247-3L Package

evaluation of the device with an MOSFET load in either of the standard TO-3P, TO-247, TO-252 (D-PAK), and TO-263 (D2PAK) footprints.



(B) For D2PAK Package

#### Figure 15. Evaluation Board Picture of Type-A (Top View)

Figure 16 shows the printed circuit board layout of NCP51752 evaluation board for the Type–A.



(C) Bottom View

Figure 16. Printed Circuit Board of Type-A

Figure 17 shows the photograph of the NCP51752 evaluation board for the Type–B. This EVB allows for

evaluation of the device with an MOSFET load in the standard TO-247-4L footprint.



Figure 17. Evaluation Board Picture of Type-B (Top View)

Figure 18 shows the printed circuit board layout of NCP51752 evaluation board for the Type–B.



(C) Bottom View

Figure 18. Printed Circuit Board of Type-B

Figure 19 shows the photograph of the NCP51752 evaluation board board for the Type–C. This EVB allows

for evaluation of the device with an MOSFET load in the standard D2PAK-7L footprint.



Figure 19. Evaluation Board Picture of Type-C (Top View)

Figure 18 shows the printed circuit board layout of NCP51752 evaluation board for the Type–B.



Figure 20. Printed Circuit Board of Type-C

#### **Related Product Information**

- [1] Datasheet of NCP51752 available on onsemi website
- [2] Datasheet of NCV51752 available on onsemi website
- [3] **onsemi** <u>AND90180/D</u>, "Practical Design Guidelines on the Usage of an Isolated Gate Driver"

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