Self Protected Very Low Iq High Side Driver with Analog Current Sense

NCV84140

The NCV84140 is a fully protected single channel high side driver that can be used to switch a wide variety of loads, such as bulbs, solenoids, and other actuators. The device incorporates advanced protection features such as active inrush current management, over–temperature shutdown with automatic restart and an overvoltage active clamp. A dedicated Current Sense pin provides precision analog current monitoring of the output as well as fault indication of short to V_D , short circuit to ground and OFF state open load detection. An active high Current Sense Enable pin allows all diagnostic and current sense features to be enabled.

Features

- Short Circuit Protection with Inrush Current Management
- CMOS (3 V / 5 V) Compatible Control Input
- Very Low Standby Current
- Very Low Current Sense Leakage
- Proportional Load Current Sense
- Current Sense Enable
- Off State Open Load Detection
- Output Short to V_D Detection
- Overload and Short to Ground Indication
- Thermal Shutdown with Automatic Restart
- Undervoltage Shutdown
- Integrated Clamp for Inductive Switching
- Loss of Ground and Loss of V_D Protection
- ESD Protection
- Reverse Battery Protection with External Components
- AEC-Q100 Qualified
- This is a Pb-Free Device

Typical Applications

- Switch a Variety of Resistive, Inductive and Capacitive Loads
- Can Replace Electromechanical Relays and Discrete Circuits
- Automotive / Industrial

FEATURE SUMMARY

R _{DSon} (typical) T _J = 25°C	R _{ON}	140	mΩ
Output Current Limit (typical)	I _{lim}	12	Α
OFF-state Supply Current (max)	I _{D(off)}	0.5	μΑ



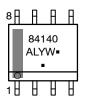
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SOIC-8 CASE 751-07 STYLE 11

MARKING DIAGRAM

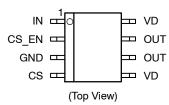


84140 = Specific Device Code A = Assembly Location

L = Wafer Lot Y = Year W = Work Week ■ Pb-Free Package

(Note: Microdot may be in either location)

PIN CONNECTIONS



ORDERING INFORMATION

Device	Package	Shipping [†]
NCV84140DR2G	SOIC-8 (Pb-Free)	2500 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

BLOCK DIAGRAM & PIN CONFIGURATION

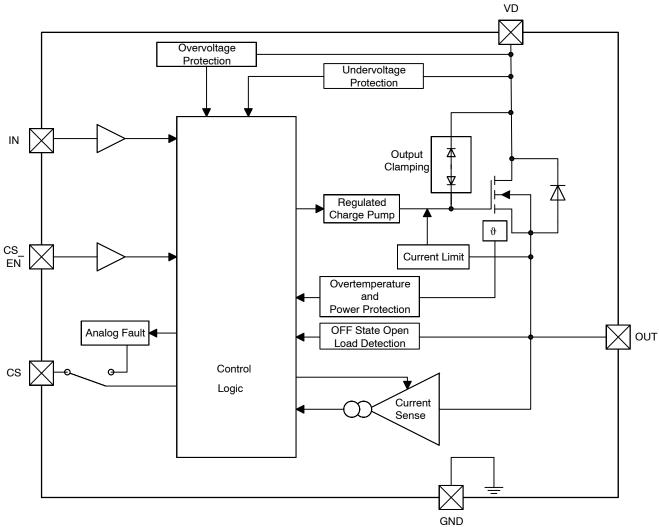


Figure 1. Block Diagram

Table 1. SO8 PACKAGE PIN DESCRIPTION

Pin#	Symbol	Description
1	IN	Logic Level Input
2	CS_EN	Current Sense Enable
3	GND	Ground
4	CS	Analog Current Sense Output
5	V_D	Supply Voltage
6	OUT	Output
7	OUT	Output
8	V_D	Supply Voltage

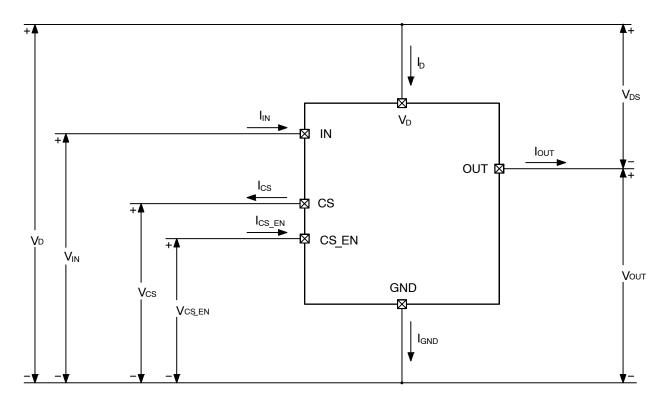


Figure 2. Voltage and Current Conventions

Table 2. Connection suggestions for unused and or unconnected pins

Connection	Input	Output	Current Sense	Current Sense Enable
Floating	Х	Х	Not Allowed	X
To Ground	Through 10 kΩ resistor	Not Allowed	Through 1 k Ω Resistor	Through 10 k Ω resistor

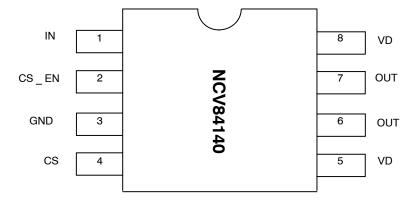


Figure 3. Pin Configuration (Top View)

ELECTRICAL SPECIFICATIONS

Table 3. MAXIMUM RATINGS

Rating	Symbol	,	Unit	
DC Supply Voltage	V _D	-0.3	38	V
Max Transient Supply Voltage (Note 1) Load Dump – Suppresses	U _S *	-	45	V
Input Voltage	V _{IN}	-10	10	V
Input Current	I _{IN}	-5	5	mA
Reverse Ground Pin Current	I _{GND}	=	-200	mA
Output Current (Note 2)	I _{OUT}	-6	Internally Limited	Α
Reverse CS Current	I _{CS}	_	-200	mA
CS Voltage	V _{CS}	V _D – 41	V _D	V
CS_EN Voltage	V _{CS_EN}	-10	10	V
CS_EN Current	I _{CS_EN}	-5	5	mA
Power Dissipation Tc = 25°C (Note 5)	P _{tot}		1.17	W
Electrostatic Discharge (Note 3) (HBM Model 100 pF / 1500 Ω) Input Current Sense Current Sense Enable Output V _D Charged Device Model CDM-AEC-Q100-011	V _{ESD}	4 4 4 4 4 750	- - - - -	DC kV kV kV kV
Single Pulse Inductive Load Switching Energy (L = 5 mH, I_L = 3.84 A, T_{Jstart} = 150°C, V_D tied to GND during inductive discharge)	E _{AS}	-	36.8	mJ
Operating Junction Temperature	TJ	-40	+150	°C
Storage Temperature	T _{storage}	-55	+150	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Load Dump Test B (with centralized load dump suppression) according to ISO16750–2 standard. Guaranteed by design. Not tested in

ESD Human Body Model tested per AEC-Q100-002 (JS-001-2017)

Field Induced Charge Device Model ESD characterization is not performed on plastic molded packages with body sizes smaller than 2 x 2 mm due to the inability of a small package body to acquire and retain enough charge to meet the minimum CDM discharge current waveform characteristic defined in JEDEC JS-002-2018

Table 4. THERMAL RESISTANCE RATINGS

Parameter	Symbol	Max. Value	Units
Thermal Resistance Junction-to-Lead (Note 4) Junction-to-Ambient (Note 4) Junction-to-Ambient (Note 5)	R _{θJL} R _{θJA} R _θ JA	29 65 106	°C/W

^{4. 645} mm² pad size, mounted on four-layer 1s2p PCB - FR4, 2 oz. Cu thickness for top layer and 1 oz. Cu thickness for inner layers (planes not electrically connected)

5. 2 cm² pad size, mounted on single-layer 1s0p PCB - FR4, 2 oz. Cu thickness

production. Passed Class C (or A, B) according to ISO16750–1.

2. Reverse Output current has to be limited by the load to stay within absolute maximum ratings and thermal performance.

^{3.} This device series incorporates ESD protection and is tested by the following methods:

ELECTRICAL CHARACTERISTICS (7 V \leq V $_D$ \leq 28 V; -40° C \leq T $_J$ \leq 150 $^{\circ}$ C unless otherwise specified)

Table 5. POWER

				Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit	
Operating Supply Voltage	V_D		4	_	28	V	
Undervoltage Shutdown	V _{UV}		-	3.5	4	V	
Undervoltage Shutdown Hysteresis	V _{UV_hyst}		-	0.4	-	V	
On Resistance	R _{ON}	I _{OUT} = 1 A, T _J = 25°C	-	140	_	mΩ	
		I _{OUT} = 1 A, T _J = 150°C	-	-	295		
		I _{OUT} = 1 A, V _D = 4.5 V, T _J = 25°C	-	-	210		
Supply Current (Note 6)	I _D	OFF-state: $V_D = 13 \text{ V}$, $V_{IN} = V_{OUT} = 0 \text{ V}$, $T_j = 25^{\circ}\text{C}$	-	0.2	0.5	μΑ	
		OFF-state: $V_D = 13 \text{ V}$, $V_{IN} = V_{OUT} = 0 \text{ V}$, $T_j = 85^{\circ}\text{C}$	-	0.2	0.5	μΑ	
		OFF-state: V _D = 13 V, V _{IN} = V _{OUT} = 0 V, Tj = 125°C	-	-	3	μΑ	
		ON-state: $V_D = 13 \text{ V}$, $V_{\text{IN}} = 5 \text{ V}$, $I_{\text{OUT}} = 0 \text{ A}$	-	1.9	3.5	mA	
On State Ground Current	I _{GND(ON)}	V _D = 13 V, V _{CS_EN} = 5 V V _{IN} = 5 V, I _{OUT} = 1 A	-	-	6	mA	
Output Leakage Current	ΙL	V _{IN} = V _{OUT} = 0 V, V _D = 13 V, Tj = 25°C	0	-	0.5	μΑ	
		V _{IN} = V _{OUT} = 0 V, V _D = 13 V, Tj = 125°C	0	_	3		

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

6. Includes PowerMOS leakage current.

Table 6. LOGIC INPUTS (V $_D$ = 13.5 V; $-40^{\circ}C \leq T_J \leq 150^{\circ}C)$

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Input Voltage - Low	V _{IN_low}		-	-	0.9	V
Input Current – Low	I _{IN_low}	V _{IN} = 0.9 V	1	-	_	μΑ
Input Voltage - High	V_{IN_high}		2.1	-	_	V
Input Current – High	I _{IN_high}	V _{IN} = 2.1 V	-	-	10	μΑ
Input Hysteresis Voltage	V _{IN_hyst}		-	0.2	_	V
Input Clamp Voltage	V _{IN_cl}	I _{IN} = 1 mA	12	13	14	V
		I _{IN} = -1 mA	-14	-13	-12	1
CS_EN Voltage - Low	V _{CSE_low}		-	-	0.9	V
CS_EN Current - Low	I _{CSE_low}	V _{CS_EN} = 0.9 V	1	-	_	μΑ
CS_EN Voltage - High	V _{CSE_high}		2.1	-	_	V
CS_EN Current - High	I _{CSE_high}	V _{CS_EN} = 2.1 V	-	-	10	μΑ
CS_EN Hysteresis Voltage	V _{CSE_hyst}		-	0.2	_	V
CS_EN Clamp Voltage	V _{CSE_cl}	I _{CS_EN} = 1 mA	12	13	14	V
		I _{CS_EN} = -1 mA	-14	-13	-12	

Table 7. SWITCHING CHARACTERISTICS (VD = 13 V, $-40^{\circ}C \le T_{J} \le 150^{\circ}C$)

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Turn-On Delay Time	t _{d_on}	V_{IN} high to 20% V_{OUT} , R_L = 13 Ω , T_J = 25°C	5	70	120	μs
Turn-Off Delay Time	t _{d_off}	V_{IN} low to 80% V_{OUT} , R_L = 13 Ω , T_J = 25°C	5	40	100	μs
Slew Rate On	dV _{out} /dt _{on}	20% to 80% V_{OUT} , R_L = 13 Ω , T_J = 25°C	0.1	0.27	0.7	V / μs
Slew Rate Off	dV _{out} /dt _{off}	80% to 20% V_{OUT} , R_L = 13 Ω , T_J = 25°C	0.1	0.35	0.7	V / μs
Turn-On Switching Loss (Note 7)	E _{on}	R _L = 13 Ω	-	0.15	0.18	mJ
Turn-Off Switching Loss (Note 7)	E _{off}	R _L = 13 Ω	-	0.1	0.18	mJ
Differential Pulse Skew, (t _(OFF) – t _(ON)) see Figure 4 (Switching Characteristics)	t _{skew}	R _L = 13 Ω	-50	-	50	μs

^{7.} Not subjected to production testing.

Table 8. OUTPUT DIODE CHARACTERISTICS

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Forward Voltage	V_{F}	$I_{OUT} = -1 \text{ A}, T_{J} = 150^{\circ}\text{C}$	-	_	0.7	V

Table 9. PROTECTION FUNCTIONS (Note 8) (7 V \leq VD \leq 18 V, $-40^{\circ}C \leq$ TJ \leq 150°C)

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Temperature Shutdown (Note 9)	T _{SD}		150	175	200	°C
Temperature Shutdown Hysteresis (T _{SD} - T _R) (Note 9)	T _{SD_hyst}		-	7	_	°C
Reset Temperature (Note 9)	T _R		T _{RS} +1	T _{RS} +7	-	°C
Thermal Reset of Status (Note 9)	T _{RS}		135	_	-	°C
Delta T Temperature Limit (Note 9)	T _{DELTA}	$T_J = -40^{\circ}C, V_D = 13 V$	_	60	-	°C
DC Output Current Limit	I _{limH}	V _D = 13 V	8	12	16	Α
		4 V < V _D < 18 V	_	-	16	Α
Short Circuit Current Limit during Thermal Cycling (Note 9)	I _{limTCycling}	V _D = 13 V T _R < Tj < T _{TSD}	-	4	_	Α
Switch Off Output Clamp Voltage	V _{OUT_clamp}	$I_{OUT} = 0.2 \text{ A}, V_{IN} = 0 \text{ V}, L = 20 \text{ mH}$	V _D – 41	V _D - 46	V _D - 52	V
Overvoltage Protection	V _{OV}	V _{IN} = 0 V, I _D = 20 mA	41	46	52	V
Output Voltage Drop Limitation	V _{DS_ON}	I _{OUT} = 0.07 A	_	20	_	mV

^{8.} To ensure long term reliability during overload or short circuit conditions, protection and related diagnostic signals must be used together with a fitting hardware & software strategy. If the device operates under abnormal conditions, this hardware & software solution must limit the duration and number of activation cycles.

Table 10. OPEN–LOAD DETECTION (7 V \leq V $_D \leq$ 18 V, $-40^{\circ}C \leq$ $T_J \leq$ 150°C)

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Open-load Off State Detection Threshold	V _{OL}	V _{IN} = 0 V, V _{CS_EN} = 5 V	2	-	4	٧
Open-load Detection Delay at Turn Off	t _{d_OL_off}		100	350	700	μs
Off State Output Current	I _{OLOFF1}	V _{IN} = 0 V, V _{OUT} = V _{OL}	-3	-	3	μΑ
Output rising edge to CS rising edge during open load	t _{d_OL}	$V_{OUT} = 4 \text{ V}, V_{IN} = 0 \text{ V}$ $V_{CS} = 90\% \text{ of } V_{CS_High}$	-	5	30	μS

^{9.} Not subjected to production testing.

Table 11. CURRENT SENSE CHARACTERISTICS (7 V \leq V $_D$ \leq 18 V, $-40^{\circ}C$ \leq T $_J$ \leq 150°C)

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Current Sense Ratio	K ₀	I _{OUT} = 0.010 A, V _{CS} = 0.5 V, V _{CS_EN} = 5 V	260	_	800	
Current Sense Ratio	K ₁	I _{OUT} = 0.025 A, V _{CS} = 0.5 V, V _{CS_EN} = 5 V	265	490	720	
Current Sense Ratio Drift (Note 11)	$\Delta K_1 / K_1$	I _{OUT} = 0.025 A, V _{CS} = 0.5 V, V _{CS_EN} = 5 V	-25	_	25	%
Current Sense Ratio	K ₂	I _{OUT} = 0.07 A, V _{CS} = 4 V, V _{CS_EN} = 5 V	270	475	675	
Current Sense Ratio Drift (Note 11)	$\Delta K_2 / K_2$	I _{OUT} = 0.07 A, V _{CS_EN} = 5 V	-20	_	20	%
Current Sense Ratio	K ₃	I _{OUT} = 0.15 A, V _{CS} = 4V, V _{CS_EN} = 5 V	275	475	625	
Current Sense Ratio Drift (Note 11)	$\Delta K_3 / K_3$	I _{OUT} = 0.15 A, V _{CS_EN} = 5 V	-15	_	15	%
Current Sense Ratio	K ₄	I _{OUT} = 0.7 A, V _{CS} = 4 V, V _{CS_EN} = 5 V	375	475	540	
Current Sense Ratio Drift (Note 11)	$\Delta K_4 / K_4$	I _{OUT} = 0.7 A, V _{CS_EN} = 5 V	-10	_	10	%
Current Sense Ratio	K ₅	I _{OUT} = 2 A, V _{CS} = 4 V, V _{CS_EN} = 5 V	420	450	480	
Current Sense Ratio Drift (Note 11)	$\Delta K_5 / K_5$	I _{OUT} = 2 A, V _{CS_EN} = 5 V	-5	-	5	%
Current Sense Leakage Current	CS _{IIkg}	I _{OUT} = 0 A, V _{CS} = 0 V V _{CS_EN} = 5 V, V _{IN} = 0 V	-	-	1	μΑ
		I _{OUT} = 0 A, V _{CS} = 0 V V _{CS_EN} = 5 V, V _{IN} = 5 V	-	-	2	
		I _{OUT} = 1 A, V _{CS} = 0 V V _{CS_EN} = 0 V, V _{IN} = 5 V,	-	-	0.5	
CS Max Voltage	CS _{Max}	$V_D = 7 \text{ V}, V_{IN} = 5 \text{ V}, R_{CS} = 2.7 \text{ k}\Omega, \\ I_{OUT} = 2 \text{ A}, T_J = 150^{\circ}\text{C}, V_{CS_EN} = 5 \text{ V}$	5	_	7	V
Current Sense Voltage in Fault Condition (Note 10)	V _{CS_fault}	$V_D = 13 \text{ V}, V_{IN} = 0 \text{ V}, R_{CS} = 1 \text{ k}, V_{OUT} = 4 \text{ V}, V_{CS_EN} = 5 \text{ V}$	-	10	-	V
Current Sense Current in Fault Condition (Note 10)	I _{CS_fault}	$V_D = 13 \text{ V}, V_{CS} = 5 \text{ V}, V_{IN} = 0 \text{ V}, V_{OUT} = 4 \text{ V}, V_{CS_EN} = 5 \text{ V}$	7	20	30	mA
Output Saturation Current (Note 10)	I _{OUT_sat}	$V_D = 7 \text{ V}, V_{CS} = 4 \text{ V}, V_{IN} = 5 \text{ V}, $ $T_J = 150^{\circ}\text{C}, V_{CS_EN} = 5 \text{ V}$	2	-	-	Α
CS_EN High to CS High Delay Time	t _{CS_High1}	$V_{IN} = 5 \text{ V}, V_{CS_EN} = 0 \text{ to } 5 \text{ V}, \\ R_{CS} = 1 \text{ k}\Omega, R_L = 13 \Omega$	-	-	100	μs
CS_EN Low to CS Low Delay Time	t _{CS_Low1}	$V_{IN} = 5 \text{ V}, V_{CS_EN} = 5 \text{ to } 0 \text{ V}, \\ R_{CS} = 1 \text{ k}\Omega, R_L = 13 \Omega$	-	5	25	μs
V _{in} High to CS High Delay Time	t _{CS_High2}	$V_{IN} = 0 \text{ to } 5 \text{ V}, V_{CS_EN} = 5 \text{ V},$ $R_{CS} = 1 \text{ k}\Omega, R_L = 13 \Omega$	-	100	250	μS
V _{in} Low to CS Low Delay Time	t _{CS_Low2}	$V_{IN} = 5 \text{ to } 0 \text{ V, } V_{CS_EN} = 5 \text{ V,}$ $R_{CS} = 1 \text{ k}\Omega, R_L = 13 \Omega$	-	50	250	μs
Delay Time I _D Rising Edge to Rising Edge of CS	Δt_{CS_High2}	V_{IN} = 5 V, V_{CS} EN = 5 V R_{CS} = 1 k Ω , I_{CS} = 90% of I_{CS} Max	-	-	100	μs

^{10.} The following fault conditions included are: Over-temperature, Power Limitation, and OFF State Open-Load Detection. 11. Not subjected to production testing.

Table 12. TRUTH TABLE

Conditions	Input	Output	CS (V _{CS_EN} = 5 V) (Note 12)
Normal Operation	L H	L H	0 Ics = I _{OUT} /K _{NOMINAL}
Overtemperature	L H	L L	0 V _{CS_fault}
Undervoltage	L H	L L	0
Overload	H H	H (no active current mgmt) Cycling (active current mgmt)	I _{CS} = I _{OUT} /K _{NOMINAL} V _{CS_fault}
Short circuit to Ground	L H	L L	0 V _{CS_fault}
OFF State Open Load	L	Н	V _{CS_fault}

^{12.} If V_{CS_EN} is low, the Current Sense output is at a high impedance, its potential depends on leakage currents and external circuitry.

WAVEFORMS AND GRAPHS

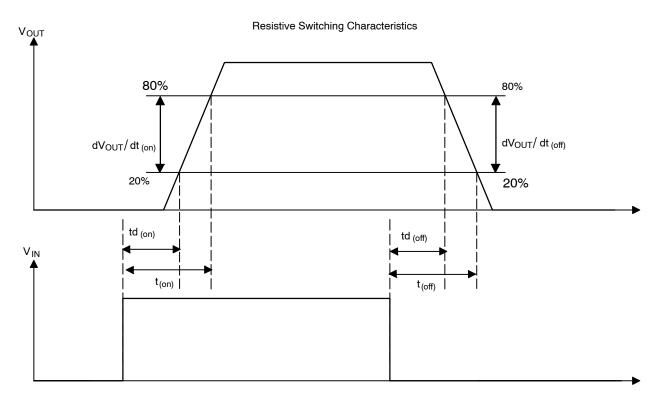


Figure 4. Switching Characteristics

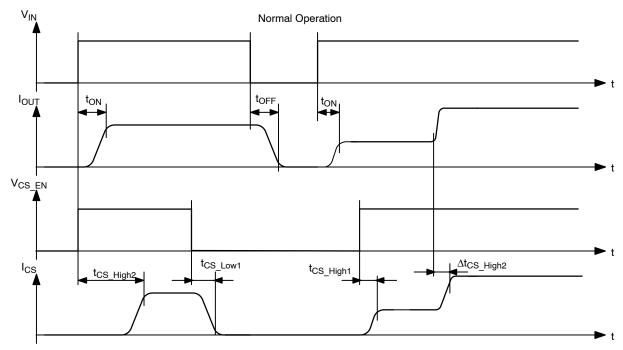


Figure 5. Normal Operation with Current Sense Timing Characteristics

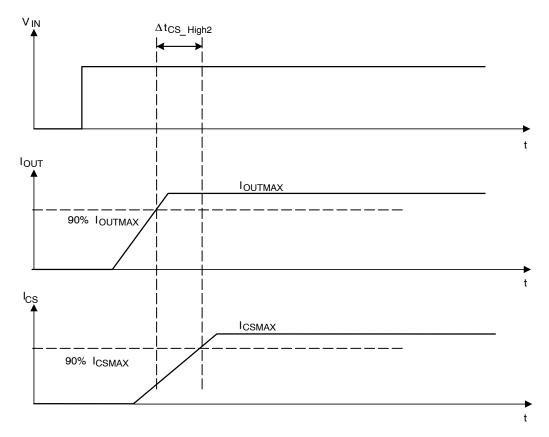


Figure 6. Delay Response from Rising Edge of I_{OUT} and Rising Edge of CS (for V_{CS_EN} = 5 V)

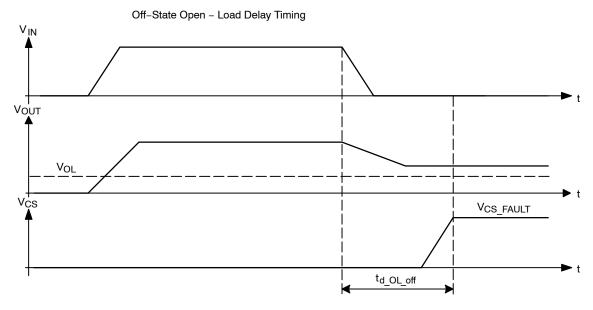


Figure 7. OFF-State Open-Load Flag Delay Timing

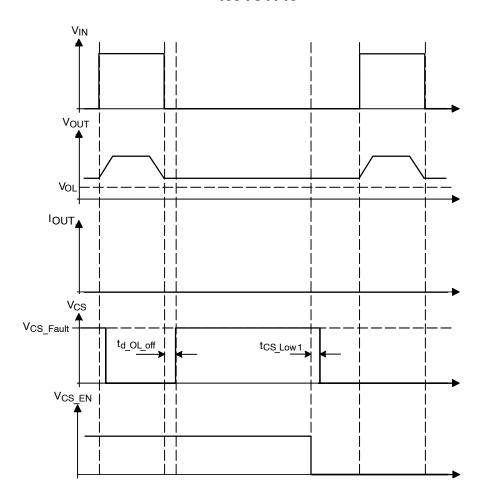


Figure 8. Off-State Open-Load with Added External Components

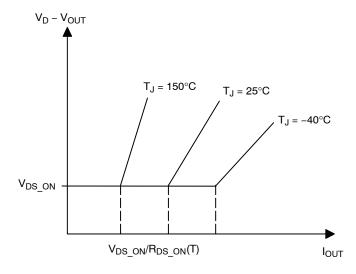
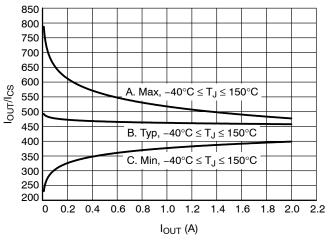


Figure 9. Voltage Drop Limitation for V_{DS_ON}



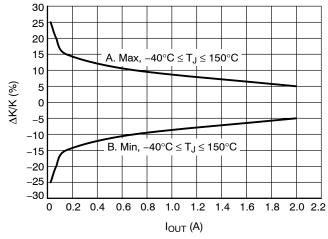


Figure 10. I_{OUT}/I_{CS} vs. I_{OUT}

Figure 11. Maximum Current Sense Ratio Drift vs. Load Current

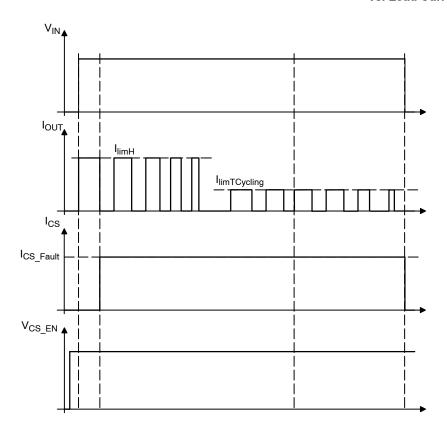


Figure 12. Short to GND or Overload

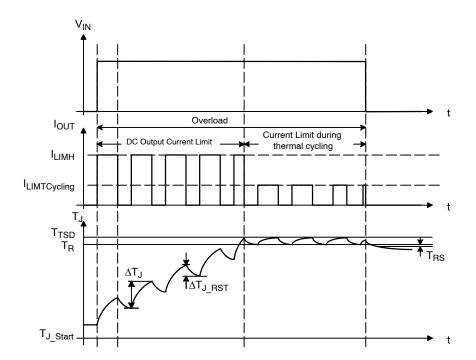


Figure 13. How T_J progresses During Short to GND or Overload

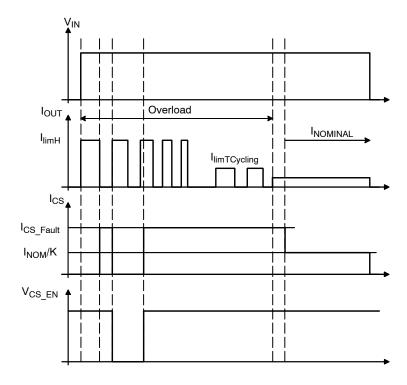


Figure 14. Discontinuous Overload or Short to GND

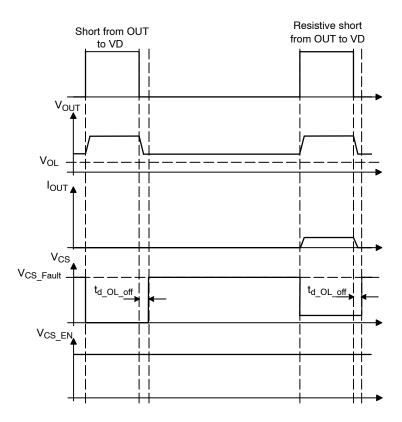


Figure 15. Short Circuit from OUT to $\ensuremath{V_D}$

TYPICAL CHARACTERISTICS

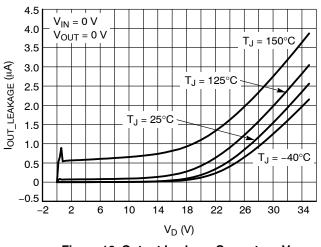


Figure 16. Output Leakage Current vs. V_D Voltage & Temperature, $V_{OUT} = 0 V$

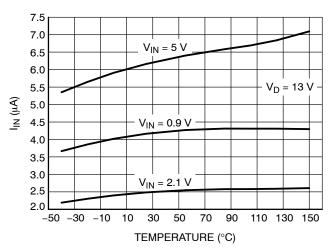


Figure 17. Input Current vs. Temperature

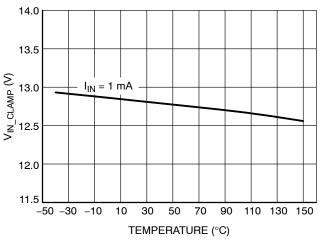


Figure 18. Input Clamp Voltage (Positive) vs. Temperature

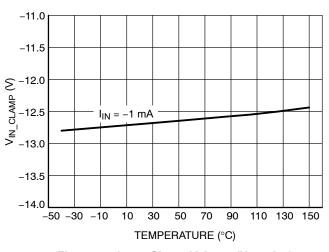


Figure 19. Input Clamp Voltage (Negative) vs. Temperature

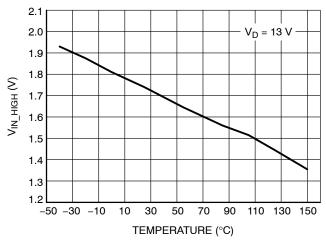


Figure 20. V_{IN} Threshold High vs. Temperature

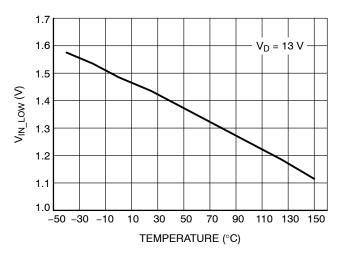
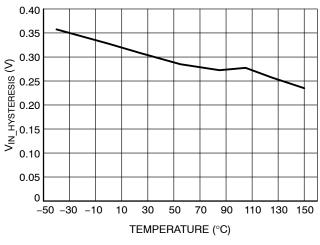


Figure 21. V_{IN} Threshold Low vs. Temperature

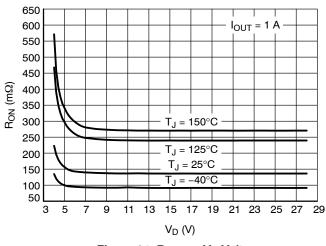
TYPICAL CHARACTERISTICS



400 $V_D = 13.5 V$ 350 $I_{OUT} = 1 A$ 300 250 Ron (mQ) 200 150 100 50 -50 -30 -10 10 30 50 70 90 110 130 150 TEMPERATURE (°C)

Figure 22. Hysteresis Input Voltage vs. Temperature

Figure 23. R_{ON} vs. Temperature



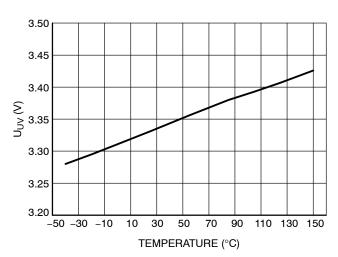
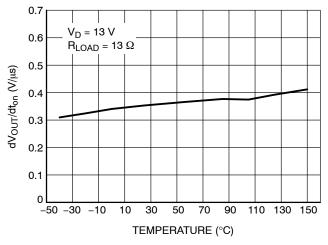


Figure 24. R_{ON} vs. V_D Voltage

Figure 25. Undervoltage Shutdown vs.
Temperature



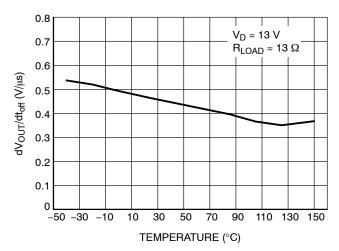


Figure 26. Slew Rate ON vs. Temperature

Figure 27. Slew Rate OFF vs. Temperature

TYPICAL CHARACTERISTICS

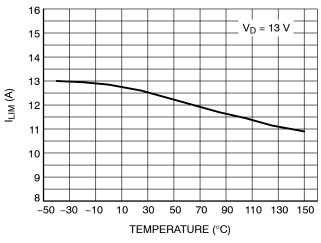


Figure 28. Current Limit vs. Temperature

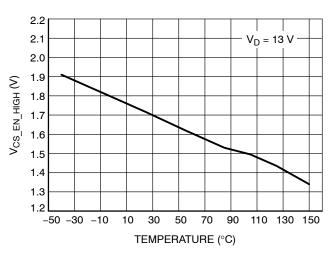


Figure 29. CS_EN Threshold High vs. Temperature

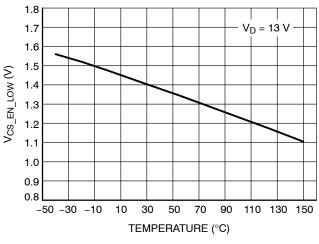


Figure 30. CS_EN Threshold Low vs. Temperature

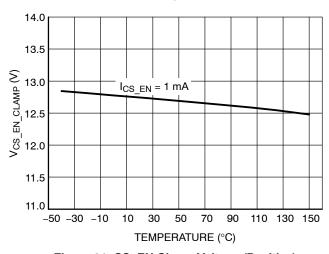


Figure 31. CS_EN Clamp Voltage (Positive) vs.
Temperature

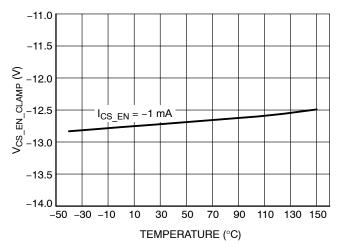


Figure 32. CS_EN Clamp Voltage (Negative) vs. Temperature

Table 13. ISO 7637-2: 2011(E) PULSE TEST RESULTS

ISO 7637-2:2011	Test Severity Levels				
Test Pulse	II	IV	Delays and Impedance	# of Pulses or Test Time	Pulse / Burst Rep. Time
1	-112	-150	2 ms, 10 Ω	500 pulses	0.5 s
2a	55	112	0.05 ms, 2 Ω	500 pulses	0.5 s
3a	-165	-220	0.1 μs, 50 Ω	1 h	100 ms
3b	112	150	0.1 μs, 50 Ω	1 h	100 ms
ISO 7637-2:2011	Test R	esults			
Test Pulse	II	IV			
1		Α			
2a	С	Е			
3a		Α			
3b	A				
Class			Function	al Status	
Α	All functions of a device perform as designed during and after exposure to disturbance.				
В	All functions of a device perform as designed during exposure. However, one or more of them can go beyond specified tolerance. All functions return automatically to within normal limits after exposure is removed. Memory functions shall remain class A.				
С	One or more functions of a device do not perform as designed during exposure but return automatically to normal operation after exposure is removed.				
D	One or more functions of a device do not perform as designed during exposure and do not return to normal operation until exposure is removed and the device is reset by simple "operator/use" action.				
E			a device do not perform as design replacing the device.	ed during and after exposure a	and cannot be returned to

APPLICATION INFORMATION

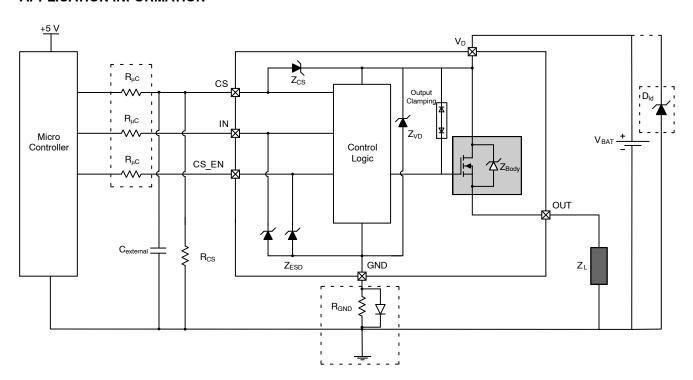


Figure 33. Application Schematic

Loss of Ground Protection

When device or ECU ground connection is lost and load is still connected to ground, the device will turn the output OFF. In loss of ground state, the output stage is held OFF independent of the state of the input. Input resistors are recommended between the device and microcontroller.

Reverse Battery Protection

Solution 1: Resistor in the GND line only (no parallel Diode)

The following calculations are true for any type of load. In the case for no diode in parallel with R_{GND} , the calculations below explain how to size the resistor.

Consider the following parameters:

 $-I_{GND}$ Maximum = 200 mA for up to $-V_D$ = 32 V.

Where $-I_{GND}$ is the DC reverse current through the GND pin and $-V_D$ is the DC reverse battery voltage.

$$- I_{\text{GND}} = \frac{- V_{\text{D}}}{R_{\text{GND}}} \tag{eq. 1}$$

Since this resistor can be used amongst multiple High–Side devices, please take note the sum of the maximum active GND currents ($I_{GND(On)max}$) for each device when sizing the resistor. Please note that if the microprocessor GND is not shared by the device GND, then R_{GND} produces a shift of ($I_{GND(On)max} \times R_{GND}$) in the input thresholds and CS output values. If the calculated power dissipation leads to too large of a resistor size or several devices have to share the same resistor, please look at the second solution for Reverse Battery Protection. Refer to Figure 35 for selecting the proper R_{GND} .

Normal Operation VIN = 5 V, Reverse Battery = 32 V

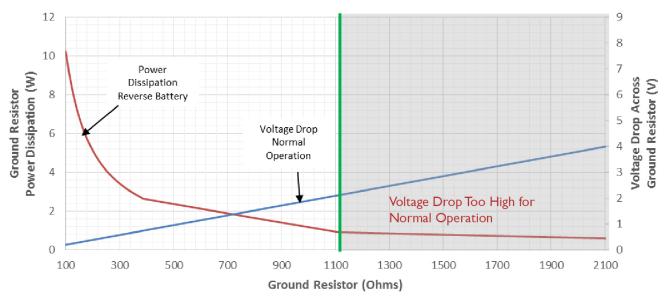


Figure 34. Reverse Battery R_{GND} Considerations

Solution 2: Diode (D_{GND}) in parallel with RGND in the ground line.

A resistor value of R_{GND} = 1 kOhm should be selected and placed in parallel to D_{GND} if the device drives an inductive load. The diode (D_{GND}) provides a ~600-700 mV shift in

the input threshold and current sense values if the micro controller ground is not common to the device ground. This shift will not vary even in the case of multiple high-side devices using the same resistor/diode network.

Undervoltage Protection

The device has two under-voltage threshold levels, V_{D_MIN} and V_{UV} . Switching function (ON/OFF) requires supply voltage to be at least V_{D_MIN} . The device features a lower supply threshold V_{UV} , above which the output can

remain in ON state. While all protection functions are guaranteed when the switch is ON, diagnostic functions are operational only within nominal supply voltage range $V_{D.}$

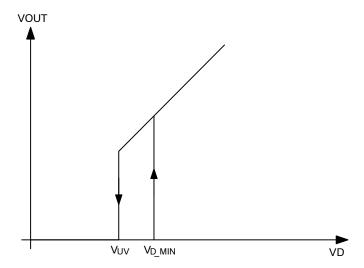


Figure 35. Undervoltage Behavior

Overvoltage Protection

The NCV84140 has two Zener diodes Z_{VD} and Z_{CS}, which provide integrated overvoltage protection. Z_{VD} protects the logic block by clamping the voltage between supply pin V_D and ground pin GND to V_{ZVD}. Z_{CS} limits voltage at current sense pin CS to $V_D - V_{ZCS}$. The output power MOSFET's output clamping diodes provide protection by clamping the voltage across the MOSFET (between V_D pin and OUT pin) to V_{CLAMP}. During overvoltage protection, current flowing through Z_{VD}, Z_{CS} and the output clamp must be limited. Load impedance Z_L limits the current in the body diode Z_{Body}. In order to limit the current in Z_{VD} a resistor, R_{GND} (150 Ω), is required in the GND path. External resistors R_{CS} and R_{SENSE} limit the current flowing through Z_{CS} and out of the CS pin into the micro-controller I/O pin. With RGND, the GND pin voltage is elevated to $V_D - V_{ZVD}$ when the supply voltage V_D rises above V_{ZVD}. ESD diodes Z_{ESD} pull up the voltage at logic pins IN, CS_EN close to the GND pin voltage $V_D - V_{ZVD}$. External resistors R_{IN}, and R_{CS} EN are required to limit the current flowing out of the logic pins into the micro-controller I/O pins. During overvoltage exposure, the device transitions into a self-protection state, with

automatic recovery after the supply voltage comes back to the normal operating range. The specified parameters as well as short circuit robustness and energy capability cannot be guaranteed during overvoltage exposure.

Overload Protection

Current limitation as well as overtemperature shutdown mechanisms are integrated into NCV84140 to provide protection from overload conditions such as bulb inrush or short to ground.

Current Limitation

In case of overload, NCV84140 limits the current in the output power MOSFET to a safe value. Due to high power dissipation during current limitation, the device's junction temperature increases rapidly. In order to protect the device, the output driver is shut down by one of the two overtemperature protection mechanisms. The output current limit is dependent on the device temperature, and will fold back once the die reaches thermal shutdown. If the input remains active during the shutdown, the output power MOSFET will automatically be re–activated after a minimum OFF time or when the junction temperature returns to a safe level.

Output Clamping with Inductive Load Switch Off

The output voltage Vout drops below GND potential when switching off inductive loads. This is because the inductance develops a negative voltage across the load in response to a decaying current. The integrated clamp of the device clamps the negative output voltage to a certain level

relative to the supply voltage V_{BAT}. During output clamping with inductive load switch off, the energy stored in the inductance is rapidly dissipated in the device resulting in high power dissipation. This is a stressful condition for the device and the maximum energy allowed for a given load inductance should not be exceeded in any application.

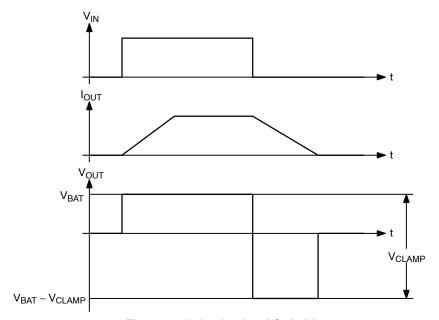


Figure 36. Inductive Load Switching

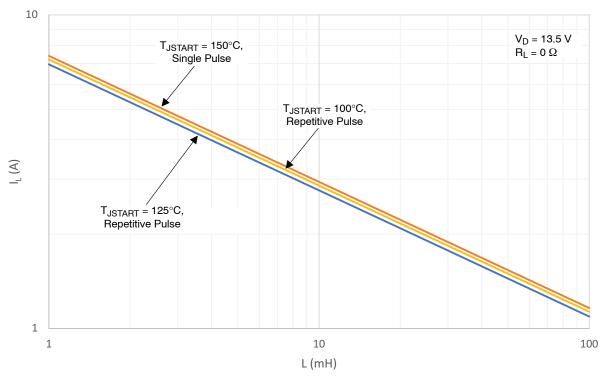


Figure 37. Maximum Switch–Off Current vs. Load Inductance, $\rm V_D$ = 13.5 V, $\rm R_L$ = 0 $\rm \Omega$

Open Load Detection in OFF State

Open load diagnosis in OFF state can be performed by activating an external resistive pull-up path (R_{PU}) to V_{BAT} . To calculate the pull-up resistance, external leakage

currents (designed pull-down resistance, humidity-induced leakage etc) as well as the open load threshold voltage $V_{\rm OL}$ have to be taken into account.

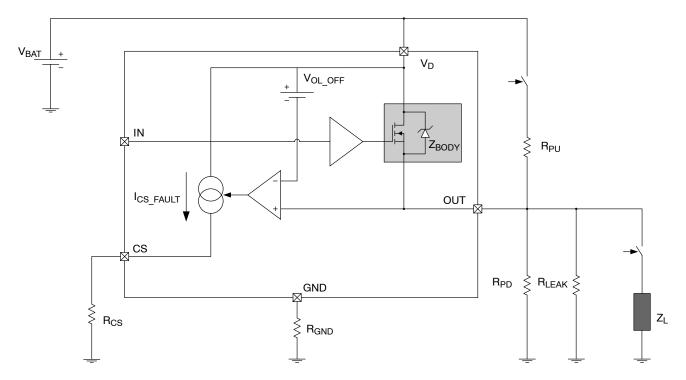


Figure 38. Open Load Detection in Off State

Current Sense in PWM Mode

When operating in PWM mode, the current sense functionality can be used, but the timing of the input signal and the response time of the current sense need to be considered. When operating in PWM mode, the following performance is to be expected. The CS_EN pin should be held high to eliminate any unnecessary delay time to the circuit. When V_{IN} switches from low to high, there will be a typical delay (t_{CS_High2}) before the current sense responds. Once this timing delay has passed, the rise time of the current

sense output (Δt_{CS_High2}) also needs to be considered. When V_{IN} switches from high to low a delay time (t_{CS_Low1}) needs to be considered. As long as these timing delays are allowed, the current sense pin can be operated in PWM mode.

EMC Performance

If better EMC performance is needed, connect a C_1 = 100 nF, C_2 = C_3 = 10 nF ceramic capacitors to the pins as close to the device as possible according to Figure 39.

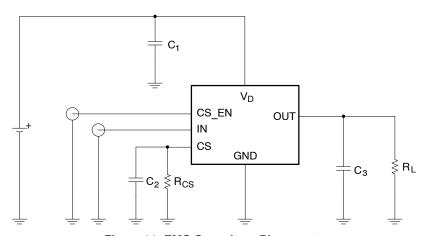


Figure 39. EMC Capacitors Placement

PACKAGE AND PCB THERMAL DATA

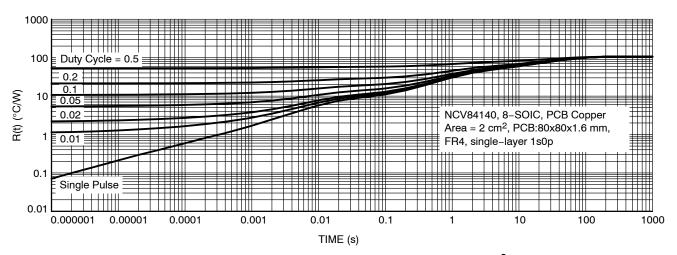


Figure 40. Junction to Ambient Transient Thermal Impedance (2 cm² Cu Area)

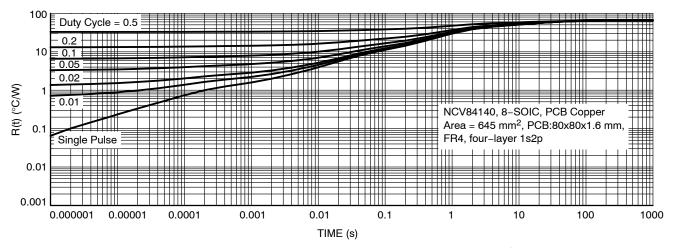
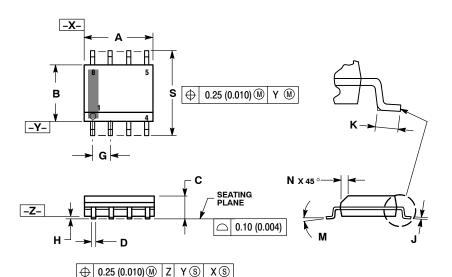


Figure 41. Junction to Ambient Transient Thermal Impedance (645 mm² Cu Area)



SOIC-8 NB CASE 751-07 **ISSUE AK**

DATE 16 FEB 2011



- NOTES:
 1. DIMENSIONING AND TOLERANCING PER
- ANSI Y14.5M, 1982.
 CONTROLLING DIMENSION: MILLIMETER.
- DIMENSION A AND B DO NOT INCLUDE MOLD PROTRUSION.
- MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE
- DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.
- 751-01 THRU 751-06 ARE OBSOLETE. NEW STANDARD IS 751-07.

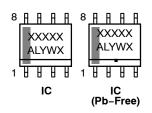
	MILLIMETERS		INCHES		
DIM	MIN	MAX	MIN	MAX	
Α	4.80	5.00	0.189	0.197	
В	3.80	4.00	0.150	0.157	
С	1.35	1.75	0.053	0.069	
D	0.33	0.51	0.013	0.020	
G	1.27 BSC		0.050 BSC		
Н	0.10	0.25	0.004	0.010	
J	0.19	0.25	0.007	0.010	
K	0.40	1.27	0.016	0.050	
М	0 °	8 °	0 °	8 °	
N	0.25	0.50	0.010	0.020	
S	5.80	6.20	0.228	0.244	

SOLDERING FOOTPRINT*



^{*}For additional information on our Pb-Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

GENERIC MARKING DIAGRAM*



XXXXX = Specific Device Code = Assembly Location = Wafer Lot

= Year = Work Week = Pb-Free Package XXXXXX AYWW AYWW H \mathbb{H} Discrete **Discrete** (Pb-Free)

XXXXXX = Specific Device Code = Assembly Location Α

= Year ww = Work Week

= Pb-Free Package

*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot "•", may or may not be present. Some products may not follow the Generic Marking.

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STYLE 9: PIN 1. EMITTER, COMMON 2. COLLECTOR, DIE #1 3. COLLECTOR, DIE #2 4. EMITTER, COMMON 5. EMITTER, COMMON 6. BASE, DIE #2 7. BASE, DIE #1 8. EMITTER, COMMON	STYLE 10: PIN 1. GROUND 2. BIAS 1 3. OUTPUT 4. GROUND 5. GROUND 6. BIAS 2 7. INPUT 8. GROUND STYLE 14: PIN 1. N-SOURCE 2. N-GATE 3. P-SOURCE	STYLE 11: PIN 1. SOURCE 1 2. GATE 1 3. SOURCE 2 4. GATE 2 5. DRAIN 2 6. DRAIN 2 7. DRAIN 1 8. DRAIN 1	STYLE 12: PIN 1. SOURCE 2. SOURCE 3. SOURCE 4. GATE 5. DRAIN 6. DRAIN 7. DRAIN 8. DRAIN
STYLE 13: PIN 1. N.C. 2. SOURCE 3. SOURCE 4. GATE 5. DRAIN 6. DRAIN 7. DRAIN 8. DRAIN	STYLE 14: PIN 1. N-SOURCE 2. N-GATE 3. P-SOURCE 4. P-GATE 5. P-DRAIN 6. P-DRAIN 7. N-DRAIN 8. N-DRAIN	8. DHAIN 1 STYLE 15: PIN 1. ANODE 1 2. ANODE 1 3. ANODE 1 4. ANODE 1 5. CATHODE, COMMON 6. CATHODE, COMMON 7. CATHODE, COMMON 8. CATHODE, COMMON	STYLE 16: PIN 1. EMITTER, DIE #1 2. BASE, DIE #1 3. EMITTER, DIE #2 4. BASE, DIE #2 5. COLLECTOR, DIE #2 6. COLLECTOR, DIE #2 7. COLLECTOR, DIE #1 8. COLLECTOR, DIE #1
STYLE 17: PIN 1. VCC 2. V2OUT 3. V1OUT 4. TXE 5. RXE 6. VEE 7. GND 8. ACC	STYLE 18: PIN 1. ANODE 2. ANODE 3. SOURCE 4. GATE 5. DRAIN 6. DRAIN 7. CATHODE 8. CATHODE	STYLE 19: PIN 1. SOURCE 1 2. GATE 1 3. SOURCE 2 4. GATE 2 5. DRAIN 2 6. MIRROR 2 7. DRAIN 1 8. MIRROR 1	STYLE 20: PIN 1. SOURCE (N) 2. GATE (N) 3. SOURCE (P) 4. GATE (P) 5. DRAIN 6. DRAIN 7. DRAIN 8. DRAIN
6. VEE 7. GND 8. ACC STYLE 21: PIN 1. CATHODE 1 2. CATHODE 2 3. CATHODE 3 4. CATHODE 4 5. CATHODE 5 6. COMMON ANODE 7. COMMON ANODE 8. CATHODE 6	STYLE 22: PIN 1. I/O LINE 1 2. COMMON CATHODE/VCC 3. COMMON CATHODE/VCC 4. I/O LINE 3 5. COMMON ANODE/GND 6. I/O LINE 4 7. I/O LINE 5 8. COMMON ANODE/GND	STYLE 23: PIN 1. LINE 1 IN 2. COMMON ANODE/GND 3. COMMON ANODE/GND 4. LINE 2 IN 5. LINE 2 OUT 6. COMMON ANODE/GND 7. COMMON ANODE/GND 8. LINE 1 OUT	0 COLLECTOR/ANODE
STYLE 25: PIN 1. VIN 2. N/C 3. REXT 4. GND 5. IOUT 6. IOUT 7. IOUT 8. IOUT	STYLE 26: PIN 1. GND 2. dv/dt 3. ENABLE 4. ILIMIT 5. SOURCE 6. SOURCE 7. SOURCE 8. VCC	STYLE 27: PIN 1. ILIMIT 2. OVLO 3. UVLO 4. INPUT+ 5. SOURCE 6. SOURCE 7. SOURCE 8. DRAIN	STYLE 28: PIN 1. SW_TO_GND 2. DASIC_OFF 3. DASIC_SW_DET 4. GND 5. V_MON 6. VBULK 7. VBULK 8. VIN
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