Intelligent Power Module (IPM) 600 V, 8 A

NFAQ0860L36T

The NFAQ0860L36T is a fully-integrated inverter power stage consisting of a high-voltage driver, six IGBT's and a thermistor, suitable for driving permanent magnet synchronous motors (PMSM), brushless-DC (BLDC) motors and AC asynchronous motors. The IGBT's are configured in a 3-phase bridge with separate emitter connections for the lower legs for maximum flexibility in the choice of control algorithm. The power stage has a full range of protection functions including cross-conduction protection, external shutdown and under-voltage lockout functions. An internal comparator and reference connected to the over-current protection circuit allows the designer to set the over-current protection level.

Features

- Three-phase 8 A / 600 V IGBT Module with Integrated Drivers
- Compact 29.6 mm x 18.2 mm Dual In-Line Package
- Built-in Under Voltage Protection
- Cross-conduction Protection
- ITRIP Input to Shut Down All IGBT's
- Integrated Bootstrap Diodes and Resistors
- Thermistor for Substrate Temperature Measurement
- Shut Down Pin
- UL1557 Certification (File Number: E339285)

Typical Applications

- Industrial Pumps
- Industrial Fans
- Industrial Automation
- Home Appliances

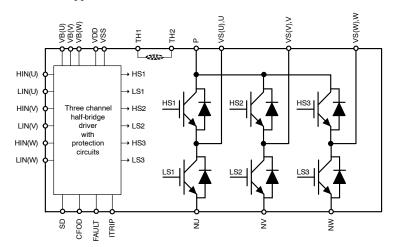


Figure 1. Function Diagram



ON Semiconductor®

www.onsemi.com



DIP38 29.6x18.2 CASE 125BS

MARKING DIAGRAM



NFAQ0860L36T = Specific Device Code

ZZZ = Assembly Lot Code

A = Assembly Location

T = Test Location

Y = Year

WW = Work Week

Device marking is on package top side

ORDERING INFORMATION

Device	Package	Shipping (Qty / Packing)
NFAQ0860L36T	DIP38 (Pb-Free)	400 / Box

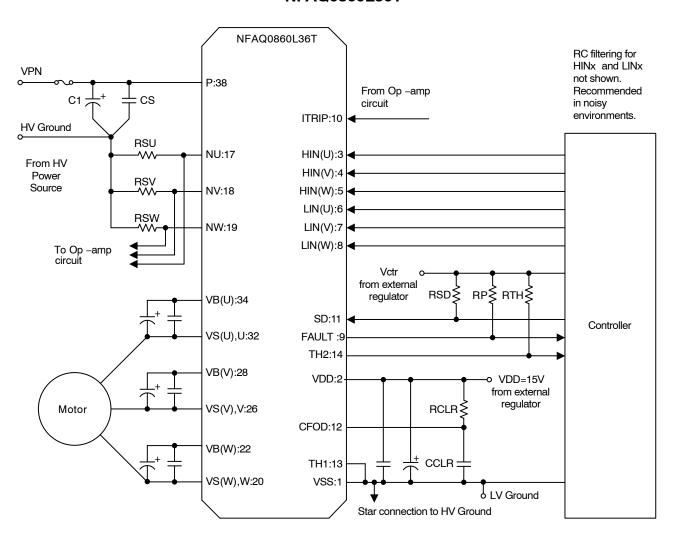


Figure 2. Application Schematic

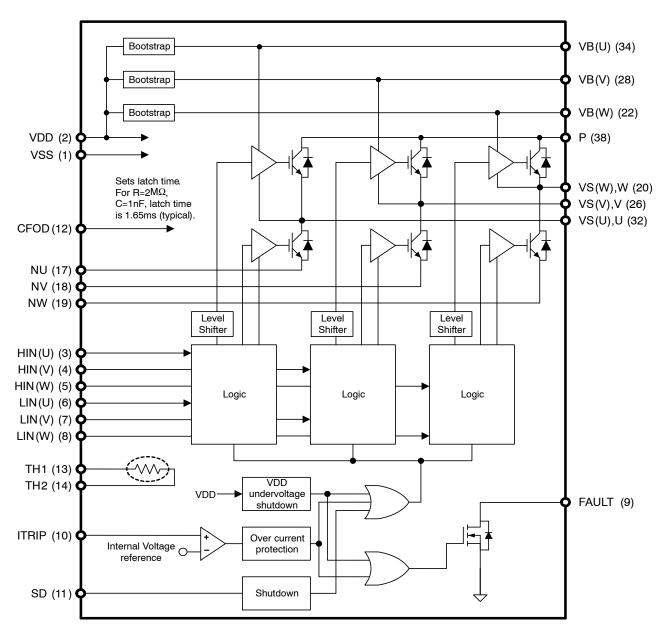


Figure 3. Simplified Block Diagram

Table 1. PIN FUNCTION DESCRIPTION

Pin	Name	Description
1	VSS	Low-Side Common Supply Ground
2	VDD	Low-Side Bias Voltage for IC and IGBTs Driving
3	HIN(U)	Signal Input for High-Side U Phase
4	HIN(V)	Signal Input for High-Side V Phase
5	HIN(W)	Signal Input for High-Side W Phase
6	LIN(U)	Signal Input for Low-Side U Phase
7	LIN(V)	Signal Input for Low-Side V Phase
8	LIN(W)	Signal Input for Low-Side W Phase
9	FAULT	Fault output
10	ITRIP	Input for Over Current Protection
11	SD	Shut Down Input
12	CFOD	Capacitor and Resistor for Fault Output Duration Selection
13	TH1	Thermistor Bias Voltage
14	TH2	Series Resistor for Thermistor
17	NU	Negative DC-Link Input for U Phase
18	NV	Negative DC-Link Input for V Phase
19	NW	Negative DC-Link Input for W Phase
20	VS(W), W	High-Side Bias Voltage GND for W phase IGBT Driving, Output for W Phase
22	VB(W)	High-Side Bias Voltage for W phase IGBT Driving
26	VS(V), V	High-Side Bias Voltage GND for V phase IGBT Driving, Output for V Phase
28	VB(V)	High-Side Bias Voltage for V phase IGBT Driving
32	VS(U), U	High-Side Bias Voltage GND for U phase IGBT Driving, Output for U Phase
34	VB(U)	High-Side Bias Voltage for U phase IGBT Driving
38	Р	Positive DC-Link Input

NOTE: Pins 15, 16, 21, 23, 24, 25, 27, 29, 30, 31, 33, 35, 36 and 37 are not present

Table 2. ABSOLUTE MAXIMUM RATINGS at $T_C = 25^{\circ}C$ (Note 1)

Parameter	Symbol	Conditions	Rating	Unit
Supply Voltage	VPN	P-NU,NV,NW, VPN (surge) < 500 V (Note 2)	450	V
Collector - Emitter Voltage	VCES	P-U,V,W; U-NU; V-NV; W-NW	600	V
Each IGBT Collector Current	IC	P,U,V,W,NU,NV,NW terminal current	±8	Α
		P,U,V,W,NU,NV,NW terminal current, Tc = 100°C	±4	Α
Each IGBT Collector Current (Peak)	ICp	Tc = 25°C, Under 1 ms Pulse Width	±16	Α
Corrector Dissipation	Pc	Tc = 25°C, Per One Chip	32	W
High-Side Control Bias voltage	VBS	VB(U)-VS(U), VB(V)-VS(V), (Note 3) VB(W)-VS(W)	-0.3 to +20.0	V
Control Supply Voltage	VDD	VDD-VSS	-0.3 to +20.0	V
Input Signal Voltage	VIN	HIN(U), HIN(V), HIN(W), LIN(U), LIN(V), LIN(W) – VSS	−0.3 to V _{DD}	V
FAULT Terminal Voltage	VFAULT	FAULT-VSS	–0.3 to V _{DD}	V
CFOD Terminal Voltage	VCFOD	CFOD-VSS	–0.3 to V _{DD}	V
SD Terminal Voltage	VSD	SD-VSS	–0.3 to V _{DD}	V
Current Sensing Input Voltage	VITRIP	ITRIP-VSS	-0.3 to +10.0	V
Operating Junction Temperature	Tj		150	°C
Storage Temperature	Tstg		-40 to +125	°C
Module Case Operation Temperature	Tc		-40 to +125	°C
Tightening Torque	MT	Case mounting screws	0.6	Nm
Isolation Voltage	Viso	50 Hz sine wave AC 1 minute (Note 4)	2000	Vrms

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

1. Refer to ELECTRICAL CHARACTERISTICS, RECOMMENDED OPERATING RANGES and/or APPLICATION INFORMATION for Safe

Table 3. RECOMMENDED OPERATING RANGES

Rating	Symbol	Conditions	Min	Тур	Max	Unit
Supply Voltage	VPN	P – NU, NV, NW	0	280	450	V
High-Side Control Bias Voltage	VBS	VB(U) - VS(U), VB(V) - VS(V), VB(W) - VS(W)	13.0	15	17.5	V
Control Supply Voltage	VDD	VDD - VSS	14.0	15	16.5	V
ON-state Input Voltage	VIN(ON)	HIN(U), HIN(V), HIN(W), LIN(U), LIN(V),	3.0	-	5.0	V
OFF-state Input Voltage	VIN(OFF)	LIN(W) - VSS	0	-	0.3	V
PWM Frequency	fPWM		1	-	20	kHz
Dead Time	DT	Turn-off to Turn-on (external)	1	-	-	μs
Allowable Input Pulse Width	PWIN	ON and OFF	1	-	-	μs
Tightening Torque		'M3' Type Screw	0.4	ı	0.6	Nm

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

Operating parameters

^{2.} This surge voltage developed by the switching operation due to the wiring inductance between P and NU, NV, NW terminal.

3. VBS=VB(U)-VS(U), VB(V)-VS(V), VB(W)-VS(W)

^{4.} Test conditions : AC2500V, 1 s

Table 4. ELECTRICAL CHARACTERISTICS at T_C = 25 $^{\circ}$ C, V_{BIAS} (VBS, VDD) = 15 V unless otherwise noted.

Parameter	Test Conditions	Symbol	Min	Тур	Max	Unit
Power Output Section			I.			
Collector-Emitter Leakage Current	V _{CE} = 600 V	ICES	-	_	100	μΑ
Collector-Emitter Saturation Voltage	IN = 5 V, IC = 8 A, Tj = 25 °C	VCE(sat)	-	2.4	3.0	V
	IN = 5 V, IC = 4 A, Tj = 100 °C		_	1.9	-	V
FWDi Forward Voltage	IN = 0 V, IC = −8 A, Tj = 25 °C	VF	-	2.1	2.7	V
	IN = 0 V, IC = -4 A, Tj = 100 °C		_	1.6	-	V
Junction to Case Thermal Resistance	Inverter IGBT Part (per 1/6 Module)	Rth(j-c)Q	-	_	3.9	°C/W
	Inverter FRD Part (per 1/6 Module)	Rth(j-c)F	-	_	7.3	°C/W
Switching Character						
Switching Time	IC = 8 A, VPN = 300 V, Tj = 25°C,	t _{ON}	-	0.4	1.1	μs
	Inductive Switching	t _{OFF}	-	0.4	1.1	μs
Turn-on Switching Loss	IC = 8 A, VPN = 300 V, Tj = 25°C	E _{ON}	-	190	-	μJ
Turn-off Switching Loss		E _{OFF}	-	90	-	μJ
Total Switching Loss		E _{TOT}	-	280	-	μJ
Turn-on Switching Loss	IC = 4 A, VPN = 300 V, Tj = 100°C	E _{ON}	-	100	-	μJ
Turn-off Switching Loss		E _{OFF}	-	50	-	μJ
Total Switching Loss		E _{TOT}	-	150	-	μJ
Diode Reverse Recovery Energy	IC = 4 A, VPN = 300 V, Tj = 100°C,	E _{REC}	-	25	-	μJ
Diode Reverse Recovery Time	(di/dt set by internal driver)	t _{RR}	-	140	-	ns
Reverse Bias Safe Operating Area	IC = 16 A, V _{CE} = 450 V	RBSOA	Full Square			
Short Circuit Safe Operating Area	V _{CE} = 400 V, Tj = 100°C	SCSOA	4	-	-	μs
Driver Section						
Quiescent VBS Supply Current	VBS = 15 V, HIN = 0 V, per driver	IQBS	-	0.07	0.4	mA
Quiescent VDD Supply Current	VDD = 15 V, LIN = 0 V, VDD-VSS	IQDD	-	0.95	3.0	mA
ON Threshold Voltage	HIN(U), HIN(V), HIN(W), LIN(U), LIN(V),	VIN(ON)	-	_	2.5	V
OFF Threshold Voltage	LIN(W) – VSS	VIN(OFF)	0.8	_	-	V
Logic 1 Input Current	VIN = +3.3 V	IIN+	-	660	900	μΑ
Logic 0 Input Current	VIN = 0 V	IIN-	-	_	3	μΑ
Bootstrap ON Resistance	IB = 1 mA	RB	-	500	-	Ω
FAULT Terminal Sink Current	FAULT: ON / VFAULT = 0.1 V	IoSD	-	2	_	mA
Fault-Output Pulse Width	FAULT-VSS	tFOD	1.1	1.65	2.2	ms
CFOD Threshold	CFOD-VSS	VCFOD	-	8	_	V
Shut Down Threshold	SD-VSS	VSD+	-	_	2.5	V
		VSD-	8.0	_	_	>
ITRIP Trip Level	ITRIP-VSS	VITRIP	0.44	0.49	0.54	>
ITRIP to Shutdown Propagation Delay		tITRIP	_	1.1	-	μs
ITRIP Blanking Time		tITRIPBL	250	350	-	ns
High-Side Control Bias Voltage Under-	Reset Level	UVBSR	10.3	11.1	11.9	٧
Voltage Protection	Detection Level	UVBSD	10.1	10.9	11.7	V
	Hysteresis	UVBSH	_	0.2	_	٧
Supply Voltage Under-Voltage Protection	Reset Level	UVDDR	10.3	11.1	11.7	V
	Detection Level	UVDDD	10.1	10.9	11.5	٧
	Hysteresis	UVDDH	-	0.2	-	V
Donal of consent to the formation of the first	ad in the Electrical Characteristics for the li			lara allar		

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

TYPICAL CHARACTERISTICS

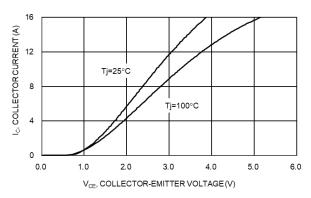


Figure 4. V_{CE} versus I_C for Different Temperatures $(V_{DD} = 15 \text{ V})$

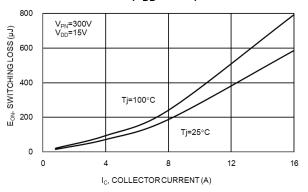


Figure 6. E_{ON} versus I_C for Different Temperatures

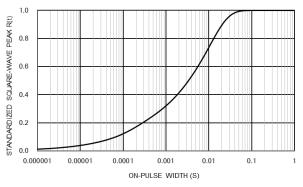


Figure 8. Thermal Impedance Plot (IGBT)

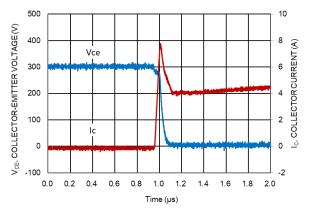


Figure 10. Turn-on Waveform Tj = 100° C, V_{CC} = 300 V

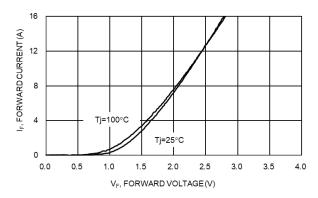


Figure 5. V_F versus I_F for Different Temperatures

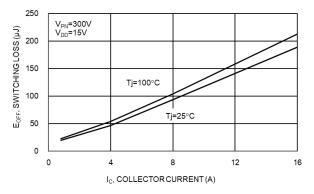


Figure 7. E_{OFF} versus I_{C} for Different Temperatures

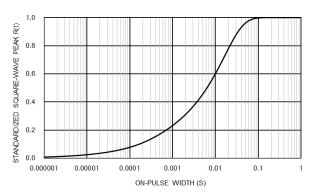


Figure 9. Thermal Impedance Plot (FRD)

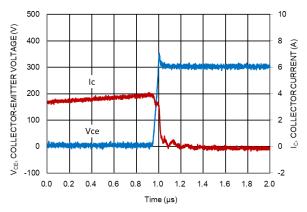
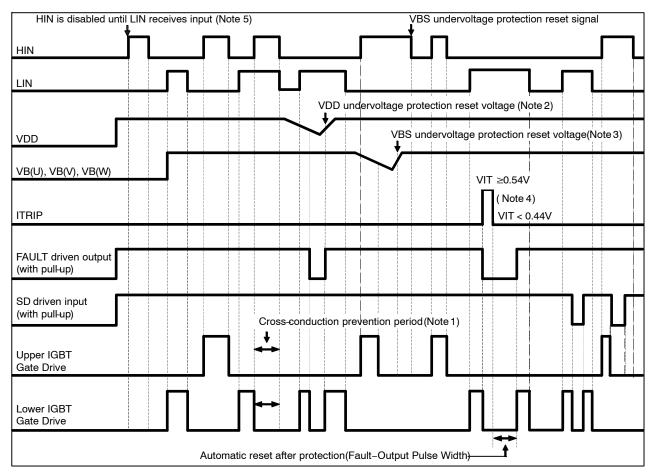


Figure 11. Turn-off Waveform Tj = 100° C, V_{CC} = 300 V

APPLICATIONS INFORMATION

Input / Output Timing Chart



NOTES:

- 1. This section of the timing diagram shows the effect of cross-conduction prevention.
- 2. This section of the timing diagram shows that when the voltage on VDD decreases sufficiently all gate output signals will go low, switching off all six IGBTs. When the voltage on VDD rises sufficiently, normal operation will resume.
- 3. This section shows that when the bootstrap voltage on VB(U) (VB(V), VB(W)) drops, the corresponding high side output U (V, W) is switched off. When the voltage on VB(U) (VB(V), VB(W)) rises sufficiently, normal operation will resume.
- 4. This section shows that when the voltage on ITRIP exceeds the threshold, all IGBTs are turned off. Normal operation resumes later after the over-current condition is removed.
- 5. After VDD has risen above the threshold to enable normal operation, the driver waits to receive an input signal on the LIN input before enabling the driver for the HIN signal.

Figure 12. Input / Output Timing Chart

Table 5. INPUT / OUTPUT LOGIC TABLE

INPUT				OUTPUT				
HIN	LIN	ITRIP	SD	High side IGBT	Low side IGBT	U,V,W	FAULT	
Н	L	L	Н	ON (Note 5)	OFF	Р	OFF	
L	Н	L	Н	OFF	ON	NU,NV,NW	OFF	
L	L	L	Н	OFF	OFF	High Impedance	OFF	
Н	Н	L	Н	OFF	OFF	High Impedance	OFF	
Х	Х	Н	Х	OFF	OFF	High Impedance	ON	
Х	Х	L	L	OFF	OFF	High Impedance	OFF	

Table 6. THERMISTOR CHARACTERISTICS

Parameter	Symbol	Condition	Min	Тур	Max	Unit
Resistance	R ₂₅	Tth = 25°C	99	100	101	kΩ
	R ₁₀₀	Tth = 100°C	5.18	5.38	5.60	kΩ
B-Constant (25 to 50°C)	В		4208	4250	4293	K
Temperature Range			-40	-	+125	°C

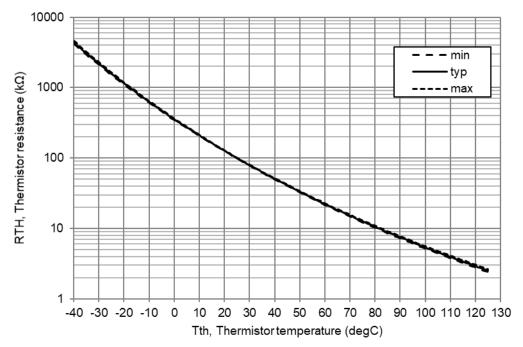


Figure 13. Thermistor Resistance versus Thermistor Temperature

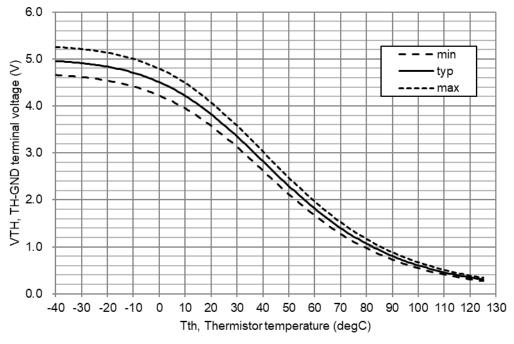


Figure 14. Thermistor Voltage versus Thermistor Temperature Conditions: RTH = 39 k Ω , Pull-up Voltage 5.0 V (see Figure 2)

FAULT Pin

The FAULT output is an open drain output requiring a pull–up resistor. If the pull–up voltage is 5 V, use a pull–up resistor with a value of 6.8 k Ω or higher. If the pull–up voltage is 15 V, use a pull–up resistor with a value of 20 k Ω or higher. The FAULT output is triggered if there is a VDD undervoltage or an overcurrent condition.

Under-voltage Protection

If VDD goes below the VDD supply under-voltage lockout falling threshold, the FAULT output is switched on. The FAULT output stays on until VDD rises above the VDD supply under-voltage lockout rising threshold. After VDD has risen above the threshold to enable normal operation, the driver waits to receive an input signal on the LIN input before enabling the driver for the HIN signal.

Overcurrent Protection

An over–current condition is detected if the voltage on the ITRIP pin is larger than the reference voltage. There is a blanking time of typically 350 ns to improve noise immunity. After a shutdown propagation delay of typically 1.1 μ s, the FAULT output is switched on. The FAULT output is held on for a time determined by the resistor and capacitor connected to the CFOD pin. If RCLR = 2 M Ω and CCLR = 1 nF, the FAULT output is switched on for 1.65 ms (typ.) because the FAULT pin goes back to high impedance when CFOD is higher than 8 V (typ.).

The over–current protection threshold should be set to be equal or lower to 2 times the module rated current (Io).

An additional fuse is recommended to protect against system level or abnormal over-current fault conditions.

Capacitors on High Voltage and VDD Supplies

Both the high voltage and VDD supplies require an electrolytic capacitor and an additional high frequency capacitor. The recommended value of the high frequency capacitor is between 100 nF and 10 μ F.

SD Pin

The SD terminal pin is used to enable or shut down the built-in driver. If the voltage on the SD pin rises above the VSD+ voltage, the output drivers are enabled. If the voltage on the SD pin falls below the VSD- voltage, the drivers are disabled.

Minimum Input Pulse Width

When input pulse width is less than 1 μ s, an output may not react to the pulse. (Both ON signal and OFF signal)

Calculation of Bootstrap Capacitor Value

The bootstrap capacitor value CB is calculated using the following approach. The following parameters influence the choice of bootstrap capacitor:

- VBS: Bootstrap power supply. 15 V is recommended.
- QG: Total gate charge of IGBT at VBS = 15 V.
 8 nC
- UVLO: Falling threshold for UVLO. Specified as 12 V.
- IDMAX: High side drive power dissipation. Specified as 0.4 mA
- TONMAX: Maximum ON pulse width of high side IGBT.

Capacitance calculation formula:

$$CB = (QG + IDMAX * TONMAX) / (VBS - UVLO)$$

CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 μ F, however, the value needs to be verified prior to production. When not using the bootstrap circuit, each high side driver power supply requires an external independent power supply.

The internal bootstrap circuit uses a MOSFET. The turn on time of this MOSFET is synchronized with the turn on of the low side IGBT. The bootstrap capacitor is charged by turning on the low side IGBT.

If the low side IGBT is held on for a long period of time (more than one second for example), the bootstrap voltage on the high side MOSFET will slowly discharge.

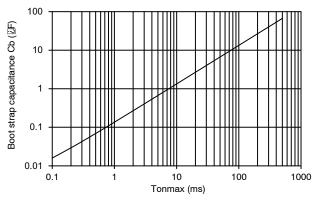


Figure 15. Bootstrap Capacitance versus Tonmax

TEST CIRCUITS

• ICES

	U+	V+	W+	U–	V-	W-
Α	38	38	38	32	26	20
В	32	26	20	17	18	19

U+, V+, W+ : High side phase U-, V-, W- : Low side phase

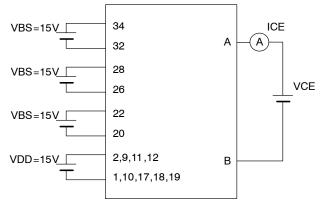


Figure 16. Test Circuit for I_{CE}

• VCE(sat) (Test by pulse)

	U+	V+	W+	U–	V-	W-
Α	38	38	38	32	26	20
В	32	26	20	17	18	19
С	3	4	5	6	7	8

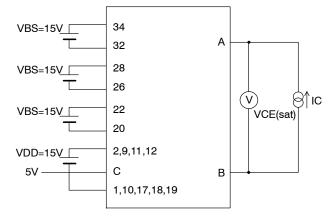


Figure 17. Test Circuit for V_{CE(SAT)}

• VF (Test by pulse)

	U+	V+	W+	U-	V-	W-
Α	38	38	38	32	26	20
В	32	26	20	17	18	19

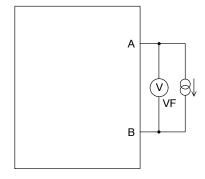


Figure 18. Test Circuit for V_{F}

• RB (Test by pulse)

	U+	V+	W+
Α	2	2	2
В	34	28	22
С	6	7	8

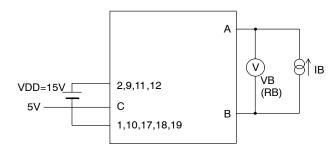


Figure 19. Test Circuit for RB

• IQBS, IQDD

	VBS U+	VBS V+	VBS W+	V_{DD}
Α	34	28	22	2
В	32	26	20	1

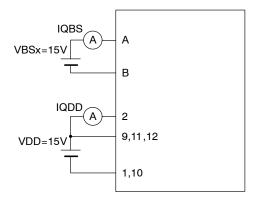


Figure 20. Test Circuit for ID

• Switching Time (The circuit is a representative example of the Inverter Low side U phase.)

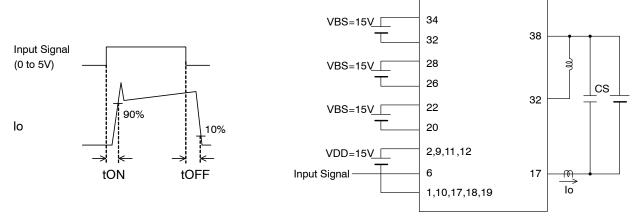
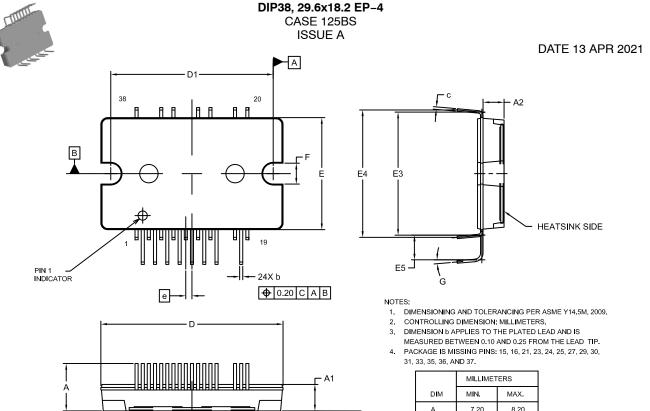


Figure 21. Test Circuit for Switching Time



_	<u> </u>		_	_	Α	7.20	8.20
			C	SEATING PLANE	A1	4.00	4.60
			Ш	PLANE	A2	3.10	3.70
1 N N	f				b	0.40	0.60
	ءااء				С	0.35	0.60
					D	29.10	30.10
		1 1			D1	26.30	26.50
					D2	19.20	20.20
		= 			E	17.70	18.70
		- [2 =1			E2	14.90	15.90
					E3	19.50	20.50
					E4	20.70 REF	
11 11		<u> </u>			E5	3.50	4.50
ט ט	الاس				e	1.00 BSC	
)2					F	2.90	3.90
	-1				G	4°	6°

GENERIC MARKING DIAGRAM*

- D

XXXX = Specific Device Code

ZZZ = Lot ID

AT = Assembly & Test Location

Y = Year

WW = Work Week

*This information is generic. Please refer to device data sheet for actual part marking. Pb-Free indicator, "G" or microdot "•", may or may not be present. Some products may not follow the Generic Marking.

DOCUMENT NUMBER:	98AON92315G	Electronic versions are uncontrolled except when accessed directly from the Document Repositor Printed versions are uncontrolled except when stamped "CONTROLLED COPY" in red.		
DESCRIPTION:	DIP38, 29.6x18.2 EP-4		PAGE 1 OF 1	

ON Semiconductor and are trademarks of Semiconductor Components Industries, LLC dba ON Semiconductor or its subsidiaries in the United States and/or other countries. ON Semiconductor reserves the right to make changes without further notice to any products herein. ON Semiconductor makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does ON Semiconductor assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. ON Semiconductor does not convey any license under its patent rights nor the rights of others.

onsemi, ONSEMI, and other names, marks, and brands are registered and/or common law trademarks of Semiconductor Components Industries, LLC dba "onsemi" or its affiliates and/or subsidiaries in the United States and/or other countries. onsemi owns the rights to a number of patents, trademarks, copyrights, trade secrets, and other intellectual property. A listing of onsemi's product/patent coverage may be accessed at www.onsemi.com/site/pdf/Patent-Marking.pdf. Onsemi reserves the right to make changes at any time to any products or information herein, without notice. The information herein is provided "as-is" and onsemi makes no warranty, representation or guarantee regarding the accuracy of the information, product features, availability, functionality, or suitability of its products for any particular purpose, nor does onsemi assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. Buyer is responsible for its products and applications using onsemi products, including compliance with all laws, regulations and safety requirements or standards, regardless of any support or applications provided by onsemi. "Typical" parameters which may be provided in onsemi data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. onsemi does not convey any license under any of its intellectual property rights nor the rights of others. onsemi products are not designed, intended, or authorized for use as a critical component in life support systems or any EDA class 3 medical devices or medical devices with a same or similar classification in a foreign jurisdiction or any devices intended for implantation in the human body. Should Buyer pu

PUBLICATION ORDERING INFORMATION

LITERATURE FULFILLMENT:
Email Requests to: orderlit@onsemi.com

onsemi Website: www.onsemi.com

TECHNICAL SUPPORT North American Technical Support: Voice Mail: 1 800-282-9855 Toll Free USA/Canada Phone: 011 421 33 790 2910

Europe, Middle East and Africa Technical Support:

Phone: 00421 33 790 2910

For additional information, please contact your local Sales Representative

X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Motor/Motion/Ignition Controllers & Drivers category:

Click to view products by ON Semiconductor manufacturer:

Other Similar products are found below:

FSB50550TB2 FSBF15CH60BTH MSVCPM2-63-12 MSVGW45-14-2 MSVGW54-14-3 MSVGW54-14-5 NTE7043 LA6565VR-TLM-E LB11650-E LB1837M-TLM-E LB1845DAZ-XE LC898300XA-MH SS30-TE-L-E 26700 LV8281VR-TLM-H BA5839FP-E2 IRAM236-1067A LA6584JA-AH LB11847L-E NCV70501DW002R2G AH293-PL-B STK672-630CN-E TND315S-TL-2H FNA23060 FSB50250AB FNA41060 MSVB54 MSVBTC50E MSVCPM3-54-12 MSVCPM3-63-12 MSVCPM4-63-12 MSVTA120 FSB50550AB NCV70501DW002G LC898301XA-MH LV8413GP-TE-L-E MSVGW45-14-3 MSVGW45-14-4 MSVGW45-14-5 MSVGW54-14-4 STK984-091A-E MP6519GQ-Z LB11651-E IRSM515-025DA4 LV8127T-TLM-H MC33812EKR2 NCP81382MNTXG TDA21801 LB11851FA-BH NCV70627DQ001R2G