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Intelligent Power Module (IPM) 600 V, 15 A

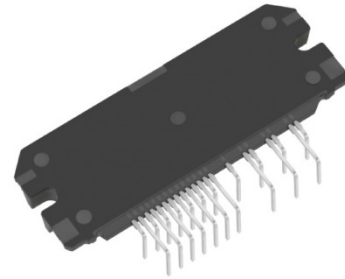
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Overview

This “Inverter IPM” is highly integrated device containing all High Voltage (HV) control from HV-DC to 3-phase outputs in a single small SIP module. Output stage uses IGBT / FRD technology and implements Under Voltage Protection (UVP) and Over Current Protection (OCP) with a Fault Detection output flag. Internal Boost diodes are provided for high side gate boost drive.

PACKAGE PICTURE



SIP29 56x21.8

Function

- Single control power supply due to Internal bootstrap circuit for high side pre-driver circuit
- All control inputs and status outputs are at low voltage levels directly compatible with microcontrollers.
- A single power supply drive is enabled through the use of bootstrap circuits for upper power supplies
- Built-in dead-time for shoot-thru protection
- Having open emitter output for low side IGBTs; individual shunt resistor per phase for OCP
- Externally accessible embedded thermistor for substrate temperature measurement
- Shutdown function ‘ITRIP’ to disable all operations of the 6 phase output stage by external input

Certification

- UL1557 (File number : E339285)

Specifications

Absolute Maximum Ratings at Tc = 25°C

| Parameter | Symbol | Remarks | Ratings | Unit |
|----------------------------|--------------------|---|-------------------------|------|
| Supply voltage | V _{CC} | V+ to U-, V-, W-, surge < 500 V *1 | 450 | V |
| Collector-emitter voltage | V _{CE} | V+ to U, V, W or U, V, W, to U-, V-, W- | 600 | V |
| Output current | I _o | V+,U-,V-,W-,U,V,W terminal current | ±15 | A |
| | | V+,U-,V-,W-,U,V,W terminal current, Tc = 100°C | ±8 | A |
| Output peak current | I _{op} | V+,U-,V-,W-,U,V,W terminal current, P.W. = 1 ms | ±30 | A |
| Pre-driver voltage | VD1, 2, 3, 4 | VB1 to U, VB2 to V, VB3 to W, VDD to VSS *2 | 20 | V |
| Input signal voltage | V _{IN} | HIN1, 2, 3, LIN1, 2, 3 | -0.3 to V _{DD} | V |
| FLTEN terminal voltage | V _{FLTEN} | FLTEN terminal | -0.3 to V _{DD} | V |
| Maximum power dissipation | P _d | IGBT per 1 channel | 35 | W |
| Junction temperature | T _j | IGBT, FRD, Pre-Driver IC | 150 | °C |
| Storage temperature | T _{stg} | | -40 to +125 | °C |
| Operating case temperature | T _c | IPM case | -40 to +100 | °C |
| Tightening torque | | A screw part *3 | 0.9 | Nm |
| Withstand voltage | V _{is} | 50 Hz sine wave AC 1 minute *4 | 2000 | VRMS |

Reference voltage is “V_{SS}” terminal voltage unless otherwise specified.

*1 : Surge voltage developed by the switching operation due to the wiring inductance between + and U- (V-, W-) terminal.

*2 : VD1 = VB1 to U, VD2 = VB2 to V, VD3 = VB3 to W, VD4 = V_{DD} to V_{SS} terminal voltage.

*3 : Flatness of the heat-sink should be less than -50µm to +100µm.

*4 : Test conditions : AC 2500 V, 1 second

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

ORDERING INFORMATION

See detailed ordering and shipping information on page 15 of this data sheet.

Electrical Characteristics at Tc = 25°C, VD1, VD2, VD3, VD4 = 15 V

| Parameter | Symbol | Conditions | Test circuit | Min | Typ | Max | Unit |
|--|------------------|--|--------------|------|------|------|------|
| Power output section | | | | | | | |
| Collector-emitter cut-off current | ICE | VCE = 600 V | Fig.1 | - | - | 100 | μA |
| Bootstrap diode reverse current | IR(BD) | VR(BD) = 600 V | | - | - | 100 | μA |
| Collector to emitter saturation voltage | VCE(SAT) | Ic = 15 A, Tj = 25°C | Fig.2 | - | 1.7 | 2.4 | V |
| | | Ic = 8 A, Tj = 100°C | | - | 1.4 | - | |
| Diode forward voltage | VF | IF = -15 A, Tj = 25°C | Fig.3 | - | 1.9 | 2.6 | V |
| | | IF = -8 A, Tj = 100°C | | - | 1.4 | - | |
| Junction to case thermal resistance | θj-c(T) | IGBT | - | - | - | 3.5 | °C/W |
| | θj-c(D) | FWD | | - | - | 5 | |
| Control (Pre-driver) section | | | | | | | |
| Pre-driver power dissipation | ID | VD1, 2, 3 = 15 V | Fig.4 | - | 0.08 | 0.4 | mA |
| | | VD4 = 15 V | | - | 1.6 | 4 | |
| High level Input voltage | Vin H | HIN1, HIN2, HIN3, LIN1, LIN2, LIN3 to VSS | - | 2.5 | - | - | V |
| Low level Input voltage | Vin L | | - | - | - | 0.8 | V |
| Logic 1 input leakage current | IIN+ | VIN = +3.3 V | - | - | 100 | 143 | μA |
| Logic 0 input leakage current | IIN- | VIN = 0 V | - | - | - | 2 | μA |
| FLTEN terminal sink current | IoSD | FAULT : ON / VFLTEN = 0.1 V | - | - | 2 | - | mA |
| FLTEN clearance delay time | FLTCLR | From time fault condition clear | - | 1.55 | 1.9 | 2.25 | ms |
| FLTEN Threshold | VEN+ | VEN rising | - | - | - | 2.5 | V |
| | VEN- | VEN falling | - | 0.8 | - | - | V |
| ITRIP threshold voltage | VITRIP | ITRIP(16) to VSS(29) | - | 0.44 | 0.49 | 0.54 | V |
| ITRIP to shutdown propagation delay | tITRIP | | - | 340 | 550 | 800 | ns |
| ITRIP blanking time | tITRIPBL | | - | 250 | 350 | - | ns |
| VCC and VBS supply undervoltage protection reset | VCCUV+ VBSUV+ | | - | 10.5 | 11.1 | 11.7 | V |
| VCC and VBS supply undervoltage protection set | VCCUV- VBSUV- | | - | 10.3 | 10.9 | 11.5 | V |
| VCC and VBS supply undervoltage hysteresis | VCCUVH VBSUVH | | - | 0.14 | 0.2 | - | V |
| Thermistor for substrate temperature Monitor | Rt | Resistance between TH(27) and VSS(29) | - | 42.3 | 47 | 51.7 | kΩ |

Reference voltage is "VSS" terminal voltage unless otherwise specified.

| Parameter | Symbol | Conditions | Test circuit | Min | Typ | Max | Unit |
|------------------------------------|------------------|---|--------------|--------------|------|-----|------|
| Switching Character | | | | | | | |
| Switching time | t ON | I _o = 15 A Inductive load | Fig.5 | - | 0.45 | - | μs |
| | t OFF | | | - | 0.55 | - | |
| Turn-on switching loss | E _{on} | I _o = 15 A, V ⁺ = 300 V, V _{DD} = 15 V, L = 3.9 mH T _c = 25°C | Fig.5 | - | 410 | - | μJ |
| Turn-off switching loss | E _{off} | | | - | 390 | - | μJ |
| Total switching loss | E _{tot} | | | - | 800 | - | μJ |
| Turn-on switching loss | E _{on} | I _o = 8 A, V ⁺ = 300 V, V _{DD} = 15 V, L = 3.9 mH T _c = 100°C | Fig.5 | - | 270 | - | μJ |
| Turn-off switching loss | E _{off} | | | - | 280 | - | μJ |
| Total switching loss | E _{tot} | | | - | 550 | - | μJ |
| Diode reverse recovery energy | E _{rec} | I _o = 8 A, V ⁺ = 400 V, V _{DD} = 15 V, L=3.9mH, T _c =100°C | - | - | 12 | - | μJ |
| Diode reverse recovery time | T _{rr} | | - | - | 54 | - | ns |
| Reverse bias safe operating area | RBSOA | I _o = 30 A, V _{CE} = 450 V | - | Full square- | | | - |
| Short circuit safe operating area | SCSOA | V _{CE} = 400 V, T _c = 100°C | - | 4 | - | - | μs |
| Allowable offset voltage slew rate | dv/dt | Between U(V,W) to U-(V-,W-) | - | -50 | - | 50 | V/ns |

Reference voltage is "V_{SS}" terminal voltage unless otherwise specified.

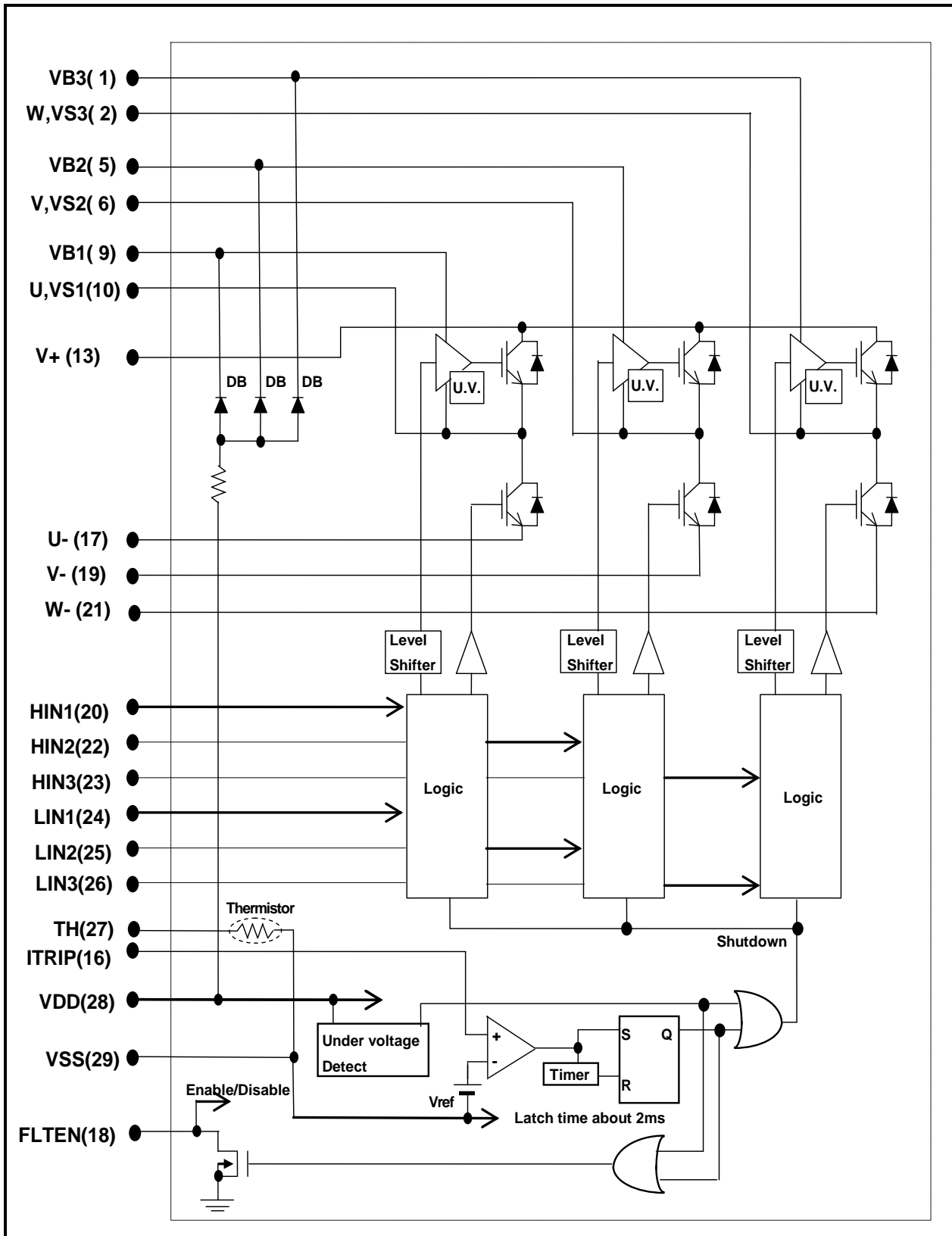
Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

Notes

- The pre-drive power supply low voltage protection has approximately 200 mV of hysteresis and operates as follows.
 - Upper side : The gate is turned off and will return to regular operation when recovering to the normal voltage, but the latch will continue till the input signal will turn 'low'.
 - Lower side : The gate is turned off and will automatically reset when recovering to normal voltage. It does not depend on input signal voltage.
- When assembling the IPM on the heat sink the tightening torque range is 0.6 Nm to 0.9 Nm.
- The pre-drive low voltage protection protects the device when the pre-drive supply voltage falls due to an operating malfunction.
- When use the over-current protection with external shunt resistor, please set the current protection level to be equal to or less than the rating of output peak current (I_{op}).

Module Pin-Out Description

| Pin | Name | Description |
|-----|--------|---|
| 1 | VB3 | High Side Floating Supply Voltage 3 |
| 2 | W, VS3 | Output 3 - High Side Floating Supply Offset Voltage |
| 3 | - | Without pin |
| 4 | - | Without pin |
| 5 | VB2 | High Side Floating Supply voltage 2 |
| 6 | V, VS2 | Output 2 - High Side Floating Supply Offset Voltage |
| 7 | - | Without pin |
| 8 | - | Without pin |
| 9 | VB1 | High Side Floating Supply voltage 1 |
| 10 | U, VS1 | Output 1 - High Side Floating Supply Offset Voltage |
| 11 | - | Without pin |
| 12 | - | Without pin |
| 13 | V+ | Positive Bus Input Voltage |
| 14 | - | Without pin |
| 15 | - | Without pin |
| 16 | ITRIP | Current protection pin |
| 17 | U- | Low Side Emitter Connection - Phase U |
| 18 | FLTEN | Enable input / Fault output |
| 19 | V- | Low Side Emitter Connection - Phase V |
| 20 | HIN1 | Logic Input High Side Gate Driver - Phase U |
| 21 | W- | Low Side Emitter Connection - Phase W |
| 22 | HIN2 | Logic Input High Side Gate Driver - Phase V |
| 23 | HIN3 | Logic Input High Side Gate Driver - Phase W |
| 24 | LIN1 | Logic Input Low Side Gate Driver - Phase U |
| 25 | LIN2 | Logic Input Low Side Gate Driver - Phase V |
| 26 | LIN3 | Logic Input Low Side Gate Driver - Phase W |
| 27 | TH | Thermistor output |
| 28 | VDD | +15 V Main Supply |
| 29 | VSS | Negative Main Supply |



Test Circuit

The tested phase : U+ shows the upper side of the U phase and U- shows the lower side of the U phase.

■ ICE / IR(BD)

| | | | | | | |
|---|----|----|----|----|----|----|
| | U+ | V+ | W+ | U- | V- | W- |
| M | 13 | 13 | 13 | 10 | 6 | 2 |
| N | 10 | 6 | 2 | 17 | 19 | 21 |

| | | | |
|---|-------|-------|-------|
| | U(BD) | V(BD) | W(BD) |
| M | 9 | 5 | 1 |
| N | 29 | 29 | 29 |

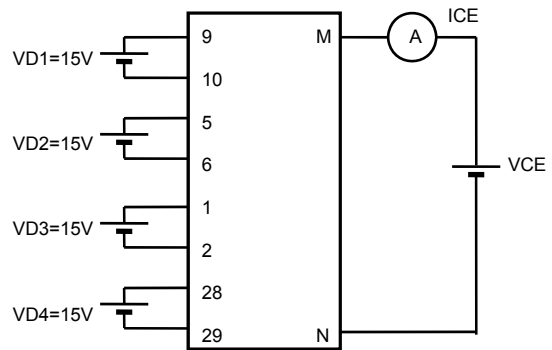


Fig.1

■ V_{CE(SAT)} (Test by pulse)

| | | | | | | |
|---|----|----|----|----|----|----|
| | U+ | V+ | W+ | U- | V- | W- |
| M | 13 | 13 | 13 | 10 | 6 | 2 |
| N | 10 | 6 | 2 | 17 | 19 | 21 |
| m | 20 | 22 | 23 | 24 | 25 | 26 |

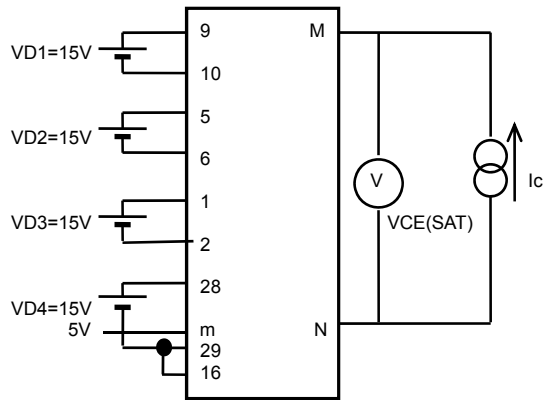


Fig.2

■ V_F (Test by pulse)

| | | | | | | |
|---|----|----|----|----|----|----|
| | U+ | V+ | W+ | U- | V- | W- |
| M | 13 | 13 | 13 | 10 | 6 | 2 |
| N | 10 | 6 | 2 | 17 | 19 | 21 |

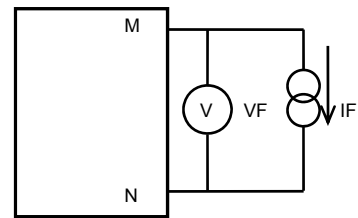


Fig.3

■ ID

| | | | | |
|---|-----|-----|-----|-----|
| | VD1 | VD2 | VD3 | VD4 |
| M | 9 | 5 | 1 | 28 |
| N | 10 | 6 | 2 | 29 |

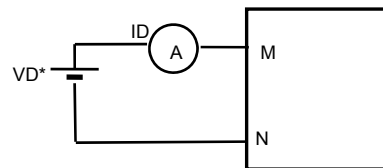


Fig.4

■ Switching time (The circuit is a representative example of the lower side U phase.)

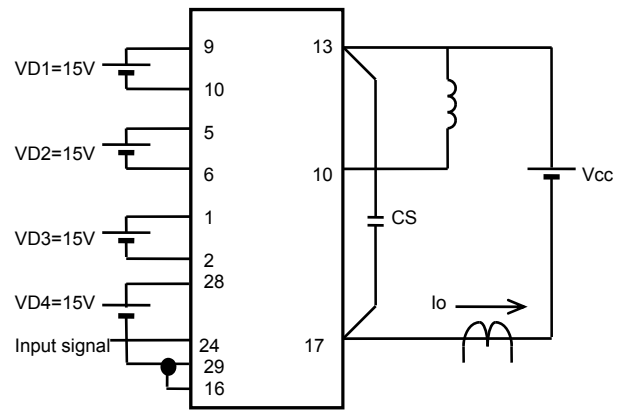
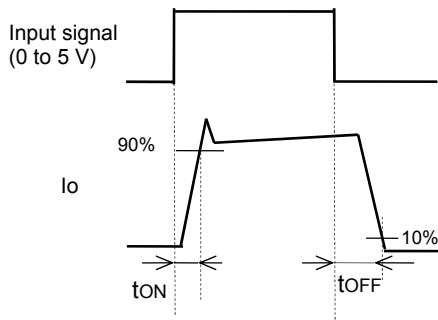


Fig.5

■ RB-SOA (The circuit is a representative example of the lower side U phase.)

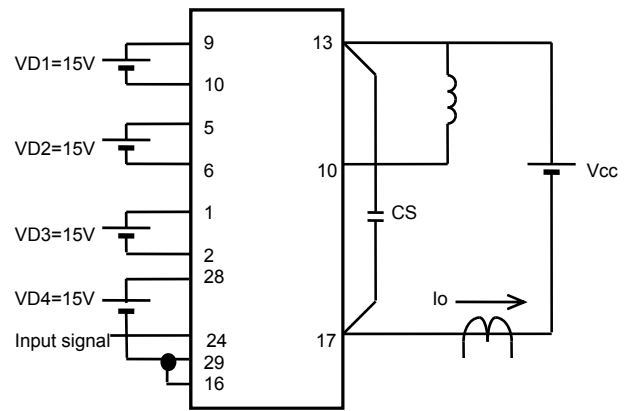
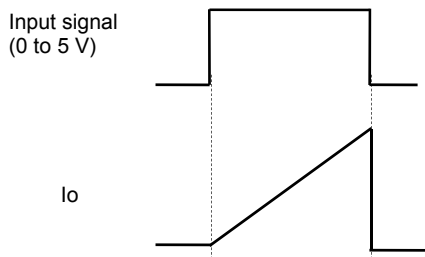


Fig.6

Input / Output Timing Chart

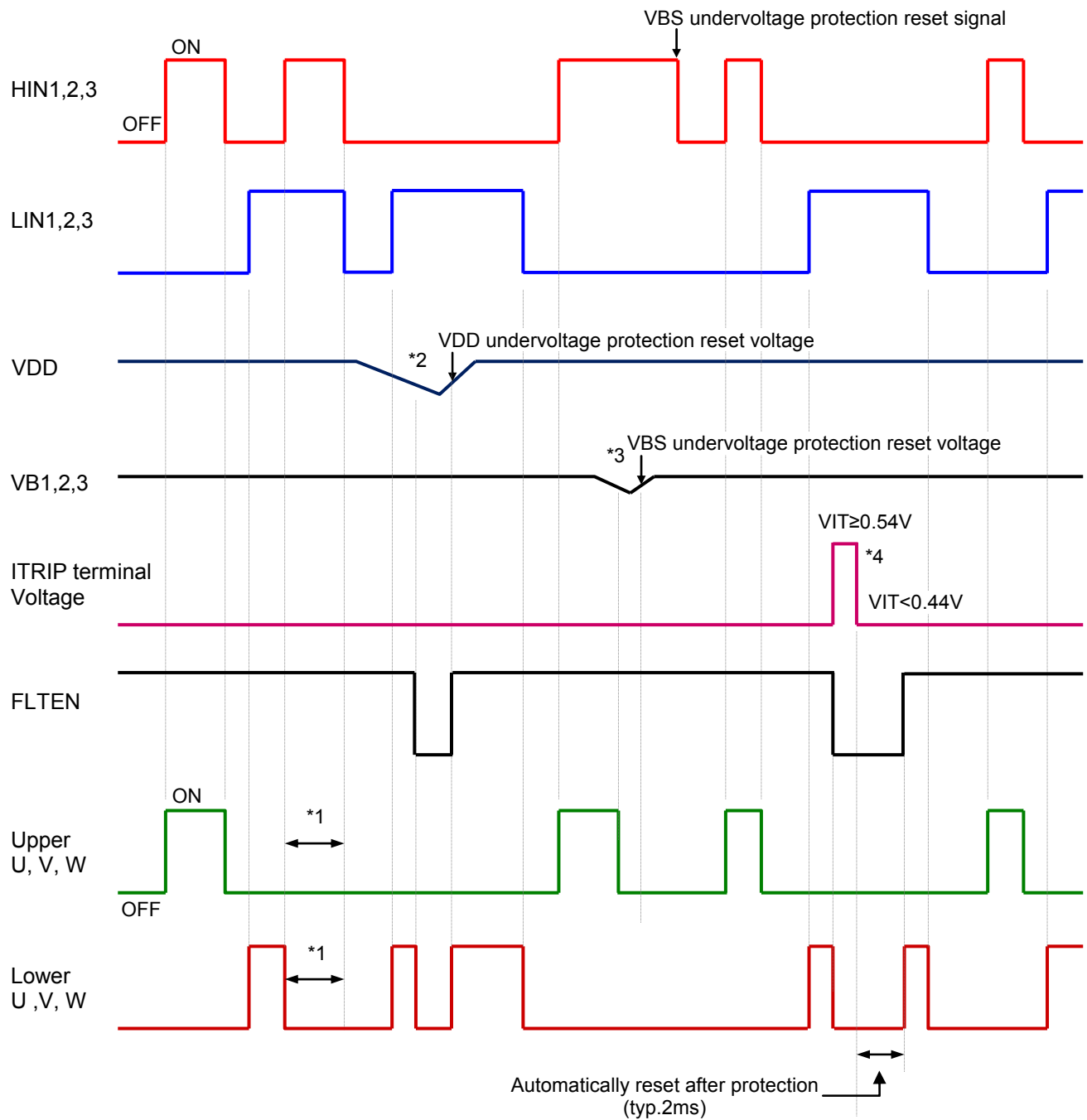
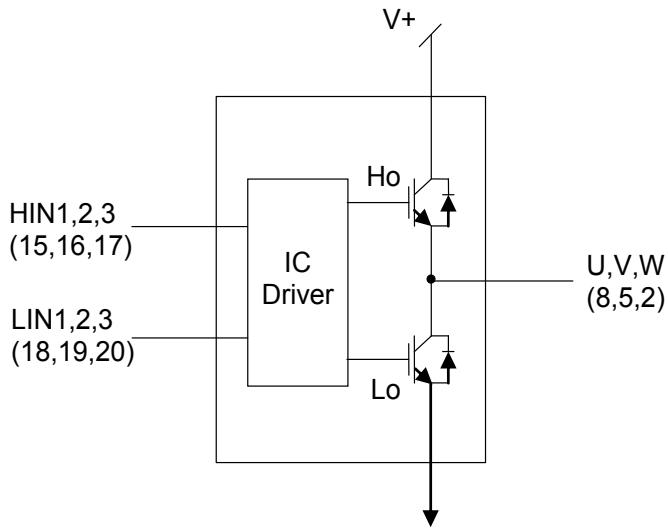


Fig. 7

Notes

- *1 : Shows the prevention of shoot-thru via control logic, however, more dead time must be added to account for switching delay externally.
- *2 : When V_{DD} decreases all gate output signals will go low and cut off all 6 IGBT outputs. When V_{DD} rises the operation will resume immediately.
- *3 : When the upper side voltage at VB1, VB2 and VB3 drops only the corresponding upper side output is turned off. The outputs return to normal operation immediately after the upper side gate voltage rises.
- *4 : When V_{ITRIP} exceeds threshold all IGBT's are turned off and normal operation resumes 2 ms (typ) after over current condition is removed.



| FLTEN | Itrip | HIN1,2,3 | LIN1,2,3 | U,V,W |
|-------|-------|----------|----------|-------|
| 1 | 0 | 1 | 0 | Vbus |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 0 | 0 | Off |
| 1 | 0 | 1 | 1 | Off |
| 1 | 1 | X | X | Off |
| 0 | X | X | X | Off |

Fig. 8

Sample Application Circuit

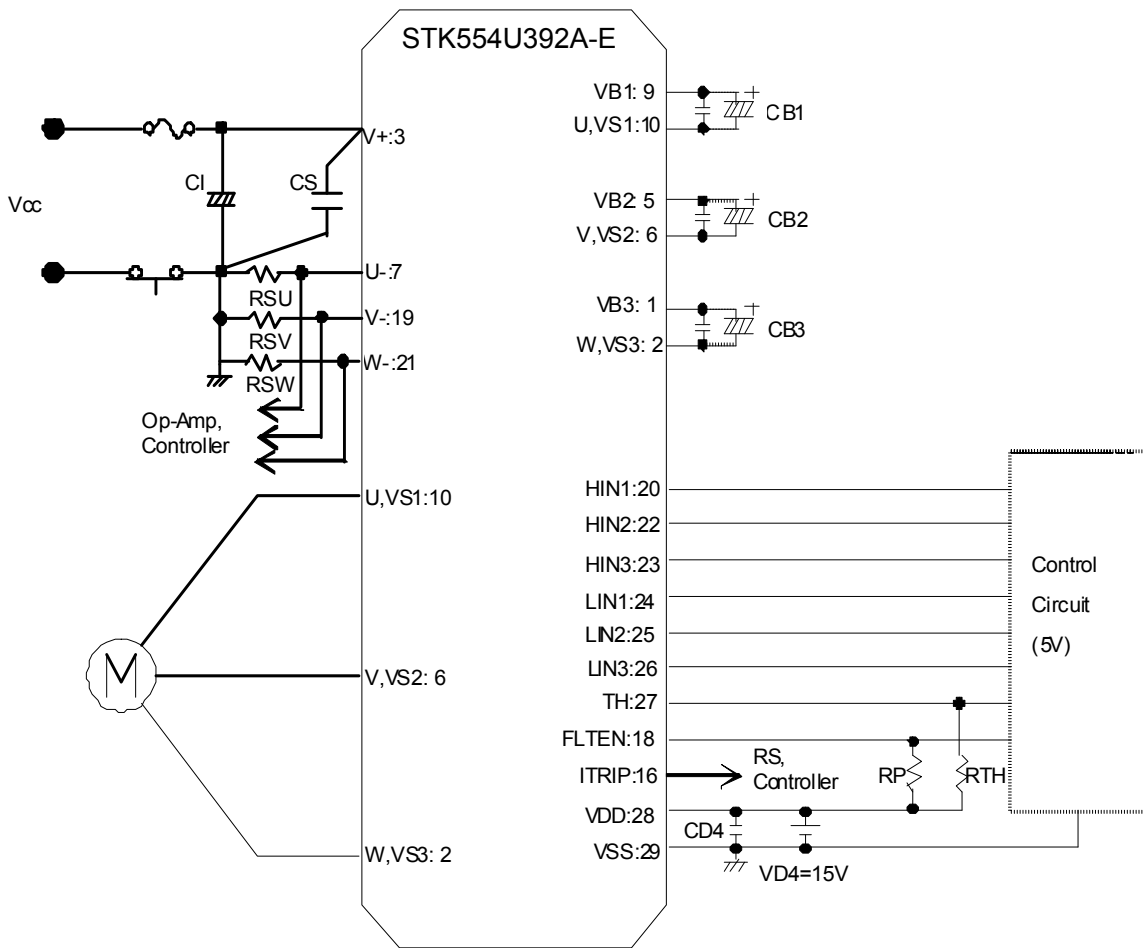


Fig. 9

Recommended Operating Condition at Tc = 25°C

| Item | Symbol | Conditions | Min | Typ | Max | Unit |
|-----------------------------|-----------------|---------------------------------------|------|-----|------|------|
| Supply voltage | V _{CC} | V+ to U-(V-,W-) | 0 | 280 | 450 | V |
| Pre-driver supply voltage | VD1, 2, 3 | VB1 to U, VB2 to V, VB3 to W | 12.5 | 15 | 17.5 | V |
| | VD4 | V _{DD} to V _{SS} *1 | 13.5 | 15 | 16.5 | |
| ON-state input voltage | VIN(ON) | HIN1, HIN2, HIN3, | 3.0 | - | 5.0 | V |
| OFF-state input voltage | VIN(OFF) | LIN1, LIN2, LIN3 | 0 | - | 0.3 | |
| PWM frequency | fPWM | | 1 | - | 20 | kHz |
| Dead time | DT | Turn-off to turn-on (external) | 0.5 | - | - | μs |
| Allowable input pulse width | PWIN | ON and OFF | 1 | - | - | μs |
| Tightening torque | | 'M3' type screw | 0.6 | - | 0.9 | Nm |

*1 : Pre-drive power supply (VD4 = 15 ±1.5 V) must have the capacity of I_o = 20 mA (DC), 0.5 A (Peak).

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

Usage Precaution

- This IPM includes internal bootstrap diode and resistor. By adding a bootstrap capacitor "CB", a high side drive voltage is generated; each phase requires an individual bootstrap capacitor. The recommended value of CB is in the range of 1 to 47 μF, however, this value needs to be verified prior to production. If selecting the capacitance more than 47 μF (±20%), connect a resistor (about 20 Ω) in series between each 3-phase upper side power supply terminals (VB1, 2, 3) and each bootstrap capacitor.
When not using the bootstrap circuit, each upper side pre-drive power supply requires an external independent power supply.
- It is essential that wiring length between terminals in the snubber circuit be kept as short as possible to reduce the effect of surge voltages. Recommended value of "CS" is in the range of 0.1 to 10 μF.
- The "FLTEN" terminal (Pin 18) is I/O terminal; Fault output / Enable input. It is used to indicate an internal fault condition of the module and also can be used to disable the module operation.
- Inside the IPM, a thermistor used as the temperature monitor for internal substrate is connected between V_{SS} terminal and TH terminal, therefore, an external pull up resistor connected between the TH terminal and an external power supply should be used. The temperature monitor example application is as follows, please refer the Fig.10, and Fig.11 below.
- The pull-down resistor (: 33 kΩ (typ)) is connected with the inside of the signal input terminal, but please connect the pull-down resistor(about 2.2 to 3.3 kΩ) outside to decrease the influence of the noise by wiring etc.
- As protection of IPM to the unusual current by a short circuit etc., it recommends installing shunt resistors and an over-current protection circuit outside. Moreover, for safety, a fuse on V_{cc} line is recommended.
- Disconnection of terminals U, V, or W during normal motor operation will cause damage to IPM, use caution with this connection.
- The "ITRIP" terminal (Pin 16) is the input terminal to shut down. When V_{ITRIP} exceeds threshold (0.44 V to 0.54 V) all IGBT's are turned off. And normal operation resumes 2 ms (typ) after over current condition is removed. Therefore, please turn all the input signals off (Low) in case of detecting error at the "FLTEN" terminal.
- When input pulse width is less than 1 μs, an output may not react to the pulse. (Both ON signal and OFF signal)

- This data shows the example of the application circuit and does not guarantee a design as the mass production set.

The characteristic of thermistor

| Parameter | Symbol | Condition | Min | Typ. | Max | Unit |
|--|-----------|---------------------------|------|------|------|--------------------|
| Resistance | R_{25} | $T = 25^{\circ}\text{C}$ | 44.6 | 47.0 | 49.4 | $\text{k}\Omega$ |
| Resistance | R_{125} | $T = 125^{\circ}\text{C}$ | 1.28 | 1.41 | 1.53 | $\text{k}\Omega$ |
| B-Constant (25 to 50°C) | B | | 4010 | 4050 | 4091 | K |
| Temperature Range | | | -40 | | +125 | $^{\circ}\text{C}$ |

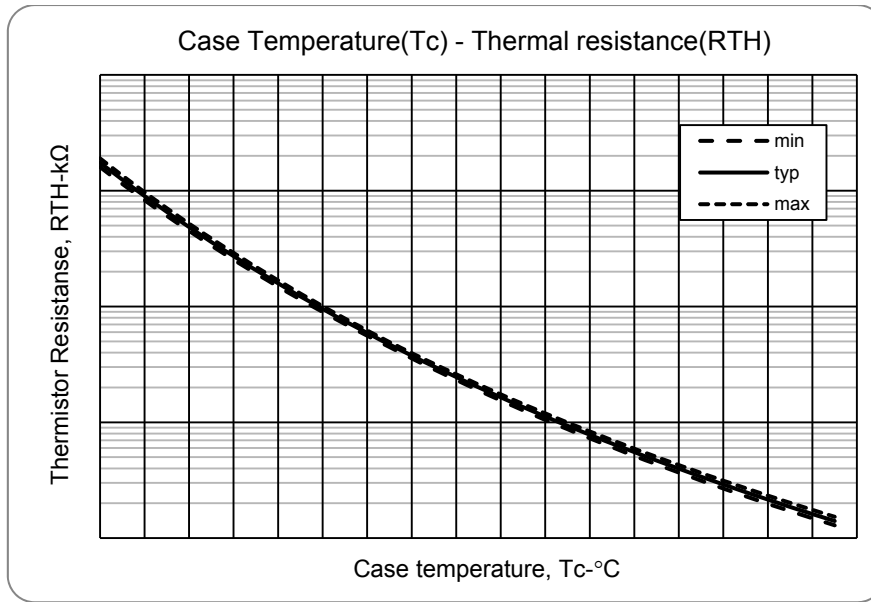


Fig. 10 Variation of thermistor resistance with temperature

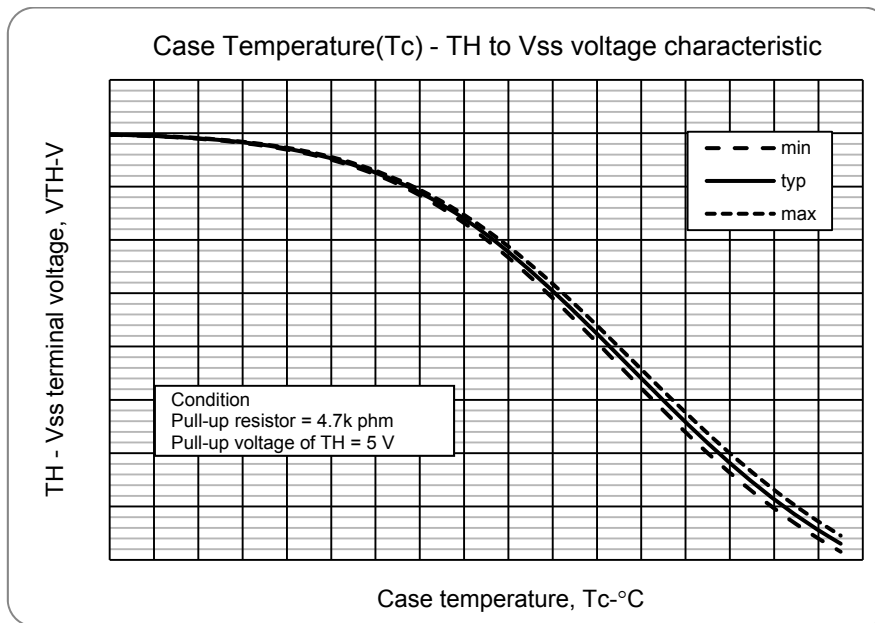


Fig. 11 Variation of temperature sense voltage with thermistor temperature

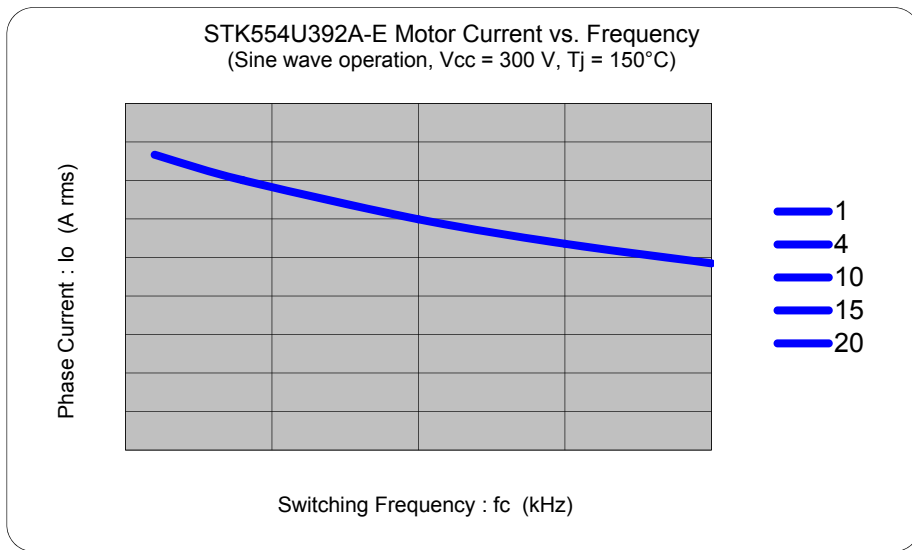


Fig. 12 Maximum sinusoidal phase current as function of switching frequency
At $T_c = 100^\circ\text{C}$, $V_{CC} = 300\text{ V}$

Switching waveform

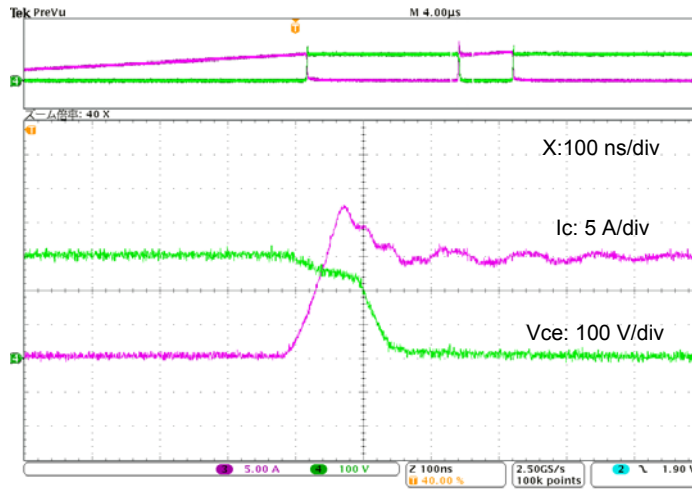


Fig. 13 IGBT Turn-on. Typical turn-on waveform at $T_c = 100^\circ\text{C}$, $V_{CC} = 300\text{ V}$, $I_c = 15\text{ A}$

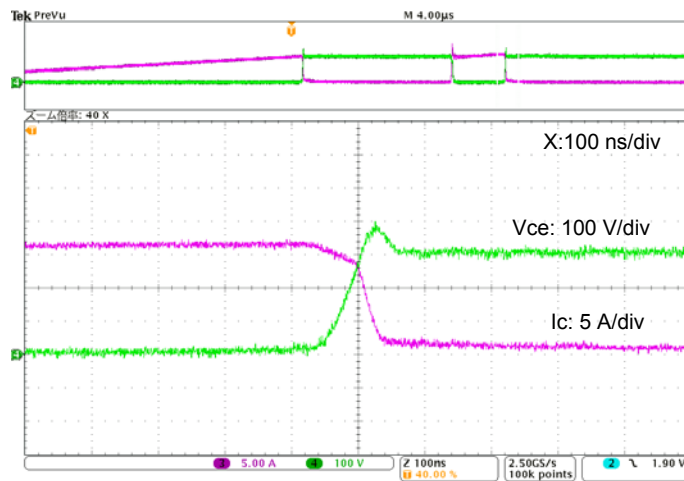


Fig. 14 IGBT Turn-off. Typical turn-off waveform $T_c = 100^\circ\text{C}$, $V_{CC} = 300\text{ V}$, $I_c = 15\text{ A}$

CB capacitor value calculation for bootstrap circuit

Calculate condition

| Item | Symbol | Value | Unit |
|--|---------|-------|------|
| Upper side power supply | VBS | 15 | V |
| Total gate charge of output power IGBT at 15 V | Qg | 132 | nC |
| Upper side power supply low voltage protection | UVLO | 12 | V |
| Upper side power dissipation | IDmax | 400 | μA |
| ON time required for CB voltage to fall from 15V to UVLO | Ton-max | - | s |

Capacitance calculation formula

CB must not be discharged below to the upper limit of the UVLO - the maximum allowable on-time (Ton-max) of the upper side is calculated as follows:

$$\begin{aligned} VBS \times CB - Qg - IDmax \times Ton-max &= UVLO \times CB \\ CB &= (Qg + IDmax \times Ton-max) / (VBS - UVLO) \end{aligned}$$

The relationship between Ton-max and CB becomes as follows. CB is recommended to be approximately 3 times the value calculated above. The recommended value of CB is in the range of 1 to 47 μF, however, the value needs to be verified prior to production.

Cb vs Tonmax

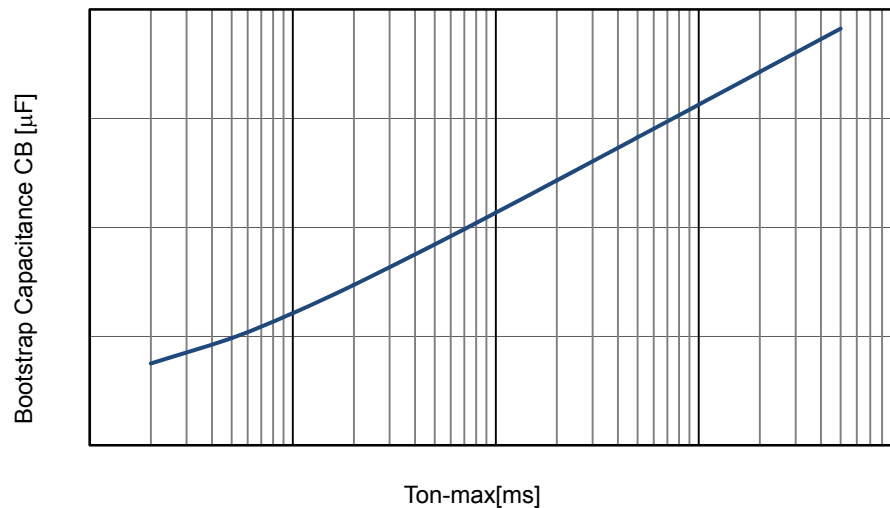
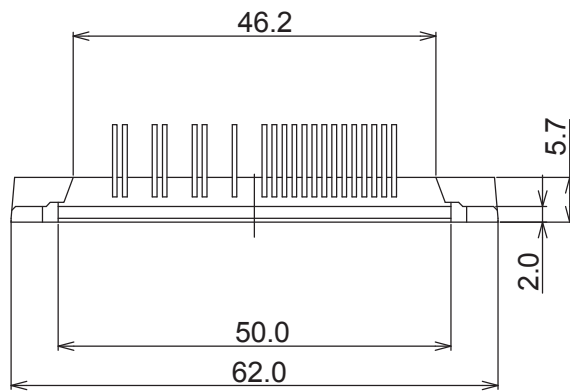
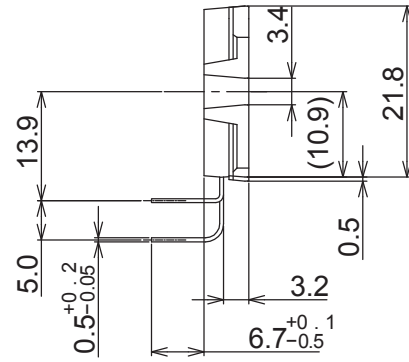
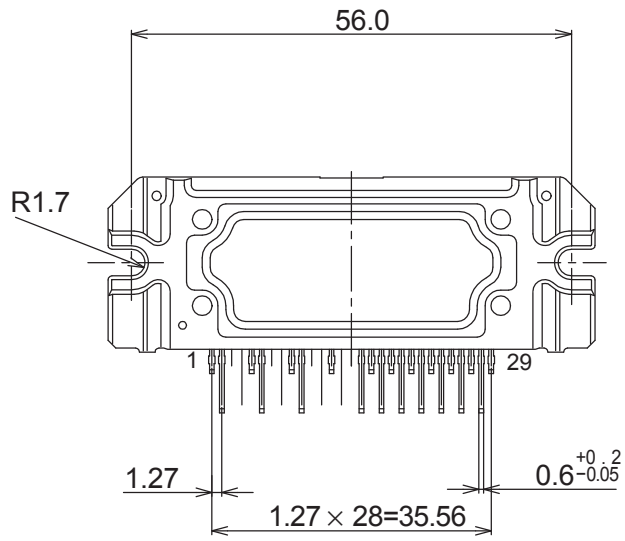


Fig. 15 Ton-max vs CB characteristic

PACKAGE DIMENSIONS
unit : mm

SIP29 56x21.8
CASE 127BW
ISSUE O

missing pin : 3,4,7,8,11,12,14,15



ORDERING INFORMATION

| Device | Package | Shipping (Qty / Packing) |
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