PAC5232/PAC5232A Data Sheet

Power Application Controller®

Multi-Mode Power Manager™

Configurable Analog Front End™

Application Specific Power Drivers™

Arm® Cortex®-M0 Controller Core



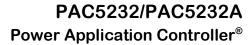




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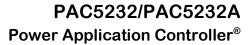
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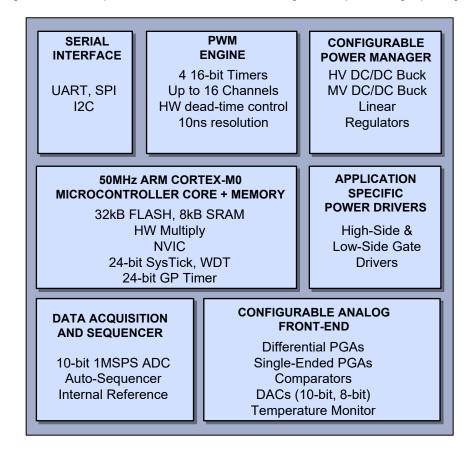


1 GENERAL DESCRIPTION

The PAC5232 is a Power Application Controller® (PAC) product that is optimized for high-speed motor control and driving. The PAC5232 integrates a 50MHz Arm® Cortex®-M0 32-bit microcontroller core with a highly-configurable Power manager, Active-Semi's proprietary and patent-pending Configurable Analog Front-EndTM and Application Specific Power DriversTM to form the most compact microcontroller-based power and motor control solution available.

The PAC5232 microcontroller features 32kB of embedded FLASH and 8kB of SRAM memory, a 1MSPS analog-to-digital converter (ADC) with programmable auto-sequencer, 3.3V/5V IO, flexible clock control system, PWM and general-purpose timers and several serial communications interfaces.

The Configurable Power Manager (CPM) provides "all-in-one" efficient power management solution for multiple types of power sources. It features a configurable high-voltage switching supply controller capable of operating a buck converter, a configurable medium-voltage switching regulator, and four linear regulated voltage supplies. The Application Specific Power Drivers (ASPD) are 180V power drivers designed for half bridge, H-bridge, 3-phase, and general purpose driving. The Configurable Analog Front End (CAFE) comprises differential programmable gain amplifiers, single-ended programmable gain amplifiers, comparators, digital-to-analog converters, and I/Os for programmable and inter-connectible signal sampling, feedback amplification, and sensor monitoring of multiple analog input signals.

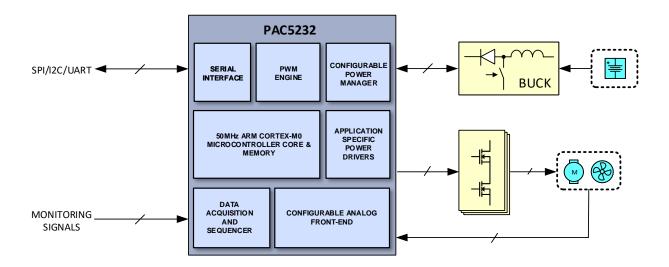


The PAC5232 is available in a 51-pin, 8x8mm TQFN package.



2 PAC® FAMILY APPLICATIONS

- Garden Tools
- Telecom Fans
- Light Electric Vehicle
- Battery Powered Motor Controllers and Drivers from 48VC 120VDC





3 PRODUCT SELECTION SUMMARY

		POW MANA					ABLE ONT E			PPLICA CIFIC DRIVE	POWER			MCI	J		APPLICATION
PART NUMBER	PIN PKG	INPUT VOLTAGE	DC/DC	DIFF-PGA	PGA	COMPARATOR	DAC	ADC CHANNEL	VBST/VSRC	POWER DRIVER	PWM CHANNEL	SPEED (MHz)	FLASH (KB)	SRAM (KB)	GPIO	SERIAL COMM	
PAC5232 PAC5232A	51L 8x8 QFN	25V- 160V	Υ	3	4	10	2	9	180V	3 LS (2A) 3 HS (2A)	6@VP 6@VCCIO	50	32	8	29	UART SPI I2C SWD	3 half-bridge 3 phase control BEMF Trapezoidal or FOC

Notes: DIFF-PGA = differential programmable gain amplifier; HS = high-side, LS = low-side, PGA = programmable gain amplifier, VSRC = Bootstrap Voltage Source

4 ORDERING INFORMATION

PART NUMBER	TEMPERATURE RANGE	PACKAGE	PINS	PACKING	
PAC5232QX PAC5232AQX	40°C to 405°C	511 0v0 OEN	51 + Exposed Pad	Tray (2600 piece)	
PAC5232AQX-T	-40°C to 105°C	51L 8x8 QFN	51 + Exposed Fad	Tape and Reel (3000 piece)	

PAC5232/PAC5232A



Power Application Controller®

5 FEATURES

5.1 Feature Overview

Configurable Power Manager

- High-voltage buck switching supply controller
 - Input Voltage: 25V 160V
 - Configurable Output Voltage: 12V or 15V
- 5V medium-voltage switching supply regulator
- 3 Linear regulators with power and hibernate management
- Power and temperature monitor, warning, fault detection

Proprietary Configurable Analog Front-End

- o 10 Analog Front-End IO pins
- 3 Differential Programmable Gain Amplifiers
- 4 Single-ended Programmable Gain Amplifiers
- Programmable Over-Current Protection
- 10 Comparators
- 2 10-bit DACs

Proprietary Application Specific Power Drivers

- 3 180V high-side gate drivers with 2A gate driving capability
- 3 low-side gate drivers with 2A gate driving capability
- Configurable propagation delay and fault protection

3.3V I/Os

 2 general-purpose I/Os with tri-state and dedicated analog input to ADC

True 5V I/Os

- 11 general-purpose I/Os with tri-state, pull-up, pull-down and dedicated I/O supply
- Configurable as true 5V or 3.3V I/Os

50MHz Arm[®] Cortex[®]-M0 32-bit Microcontroller Core

- Single-cycle 32-bit x 32-bit hardware multiplier
- Integrated sleep and deep sleep modes
- Nested Vectored Interrupt Controller (NVIC) with 20 Interrupts with 3 levels of priority
- 24-Bit SysTick Timer
- Wake-up Interrupt Controller (WIC) allowing power-saving sleep modes
- Clock-gating allowing low-power operation

Memory

- o 32kB FLASH
- o 8kB SRAM

Analog to Digital Converter (ADC)

- o 10-bit resolution
- o 1MSPS
- Dual Programmable ADC Auto-Sequencers

Flexible clock and PLL from internal 1.25% oscillator, ring oscillator, external clock or crystal

9 Timing Generators

- Four 16-bit timers with up to 16 PWM/CC blocks and 7 independent dead-time controllers
- 24-bit watchdog timer
- o 4s or 8s watchdog timer
- o 24-bit real time clock
- 24-bit SysTick timer
- Wake-up timer for sleep modes from 0.125s to 8s

SPI, I2C, and UART serial communication interfaces

SWD debug interface with interface disable function



6 ABSOLUTE MAXIMUM RATINGS

The table below shows the absolute maximum ratings for this device.

To prevent damage to the device, do not exceed these limits. Exposure to the absolute maximum rating conditions for long periods of time may affect device reliability.

The device is not guaranteed to function properly outside of the operating conditions.

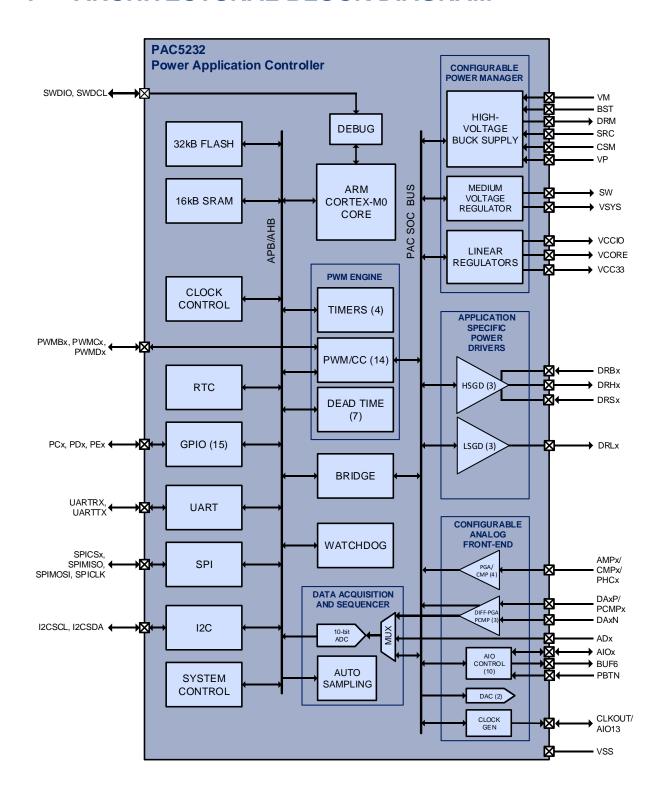
PARA	METER	VALUE	UNIT
VM to VSS	-0.3 to 160	V	
BST to VSS	-0.3 to 180	V	
BST to SRC	-0.3 to 20	V	
SRC to VSS		-10 to VM + 15	V
DRM to SRC		-0.3 to BST + 0.3	V
VP to VSS		-0.3 to 20	V
SW to VSS		-0.3 to V _P + 0.3	V
CSM to VP		-0.3 to 0.3	V
VSYS to VSS	-0.3 to 6	V	
AIO<9:7>, AIO<5:0>, VCCIO, AIO6 to VSS	-0.3 to V _{SYS} + 0.3	V	
PD <x>, PE<x> to VSS</x></x>	-0.3 to V _{CCIO} + 0.3	V	
PC <x> to VSS</x>	-0.3 to V _{CC33} + 0.3	V	
PA <x>, PB<x>, PC<x>, PD<x>, PE<x> pin in</x></x></x></x></x>	7.5	mA	
PA <x>, PB<x>, PC<x>, PD<x>, PE<x> sum o</x></x></x></x></x>	25	mA	
VCC33 to VSS		-0.3 to 4.1	V
VCORE to VSS		-0.3 to 2.5	V
DRL0, DRL1, DRL2 to VSS		-0.3 to V _P + 0.3	V
DRB3, DRB4, DRB5 to VSS		-0.3 to 180	V
DRS3, DRS4, DRS5 to VSS		-10 to VM + 15	V
DRB3 to DRS3, DRB4 to DRS4, DRB5 to DR	S5	-0.3 to 20	V
DRH3 to DRS3, DRH4 to DRS4, DRH5 to DR	-0.3 to V _{DRBx} + 0.3	V	
VSS RMS Current ¹		0.2	A _{RMS}
Operating temperature range	-40 to 105	°C	
Flootroctatic Discharge (ESD)	Human body model (JEDEC)	2	kV
Electrostatic Discharge (ESD)	Charge device model (JEDEC)	1	kV

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¹ Peak current may be 10 times higher than the RMS value for pulses shorter than 10µs.



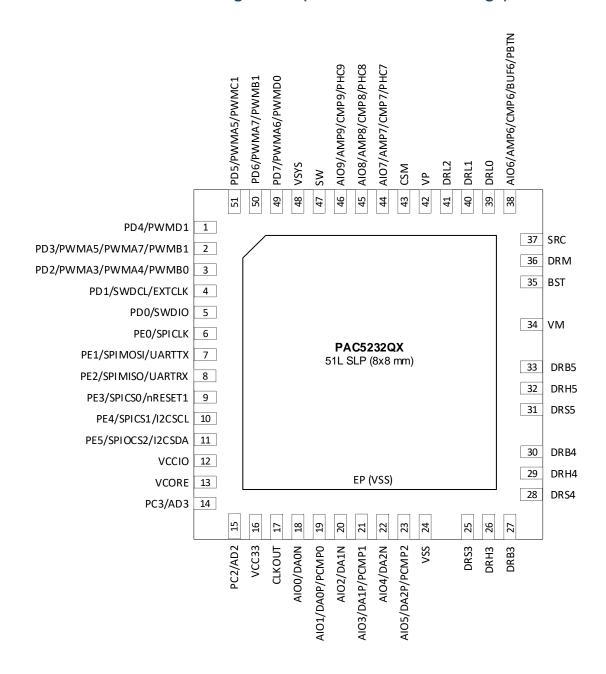
7 ARCHITECTURAL BLOCK DIAGRAM





8 PIN CONFIGURATION

8.1 PAC5232QX Pin Configuration (51L SLP 8x8 mm Package)





9 PIN DESCRIPTION

9.1 Power and Ground Pin Description

PIN NAME	PIN NUMBER	TYPE	DESCRIPTION
VCCIO	12	Power	Internally generated digital I/O 3.3V power supply. Connect a $2.2\mu F$ or higher value ceramic capacitor from V_{CCIO} to V_{SSA} .
VCORE	13	Power	Internally generated digital I/O 1.8V power supply. Connect a $2.2\mu F$ or higher value ceramic capacitor from V_{CC18} to V_{SSA} .
VCC33	16	Power	Internally generated 3.3V power supply. Connect to a 2.2µF or higher value ceramic capacitor from V _{CC33} to V _{SSA} .
VSS	24	Power	Ground.
VM	34	Power	High-Voltage Buck Regulator supply controller input. Connect a high value electrolytic capacitor in parallel with a 0.1µF ceramic capacitor from VM to VSS. This pin requires good capacitive bypass to Vss, so the ceramic capacitor must be connected with a shorter than 10mm trace from the pin.
BST	35	Power	High-Voltage Buck Regulator bootstrap input. Connect a 2.2µF or higher value ceramic capacitor from BST to SRC with a shorter than 10mm trace from the pin.
DRM	36	Power	High-Voltage Buck Regulator Switching supply driver output. Connect to the gate of the external power N-channel MOSFET.
SRC	37	Power	High-Voltage Buck Regulator Source. Connect to the source of the high-side power MOSFET of the high-voltage buck regulator.
VP	42	Power	Main power supply. Provides power to the power drivers as well as voltage feedback path for the switching supply. Connect a properly sized supply bypass capacitor in parallel with a $10\mu F$ ceramic capacitor in parallel with a $100\mu F$ aluminum capacitor from V_P to V_{SS} for voltage loop stabilization. If the switching frequency of the HV-BUCK is >= $200 kHz$, then the $100\mu F$ aluminum capacitor can be replaced with $47\mu F$, but the efficiency will be worse. This pin requires good capacitive bypassing to V_{SS} , so the ceramic capacitor must be connected with a shorter than 10mm trace from the pin.
CSM	43	Power	High-Voltage Buck Regulator Switching supply current sense input. Connect to the positive side of the current sense resistor.
SW	47	Power	Switch node for the medium-voltage buck regulator.
VSYS	48	Power	5V System power supply. Connect to a $22\mu F/6.3V$ (20%) or higher ceramic capacitor from V _{SYS} to V _{SS} .
EP (VSS)	EP	Power	Exposed pad. Must be connected to V _{ss} in a star ground configuration. Connect to a large PCB copper area for power dissipation heat sinking.

9.2 Signal Manager Pin Description

PIN NAME	PIN NUMBER	FUNCTION	TYPE	DESCRIPTION
CLKOUT	17 CLKOUT		I/O	Low-frequency clock reference output (250Hz – 2kHz), or GPIO.
AIO0	18	AIO0	I/O	Analog front end I/O 0.
AlOu		DA0N	Analog	Differential PGA 0 negative input.
AIO1	40	AIO1	I/O	Analog front end I/O 1.
AIOT	19	DA0P	Analog	Differential PGA 0 positive input.
AIO2	20	AIO2	I/O	Analog front end I/O 2.
AIO2	20	DA1N	Analog	Differential PGA 1 negative input.



AIO3	21	AIO3	I/O	Analog front end I/O 3.
AlO3	21	DA1P	Analog	Differential PGA 1 positive input.
AIO4	22	AIO4	I/O	Analog front end I/O 4.
AIO4	22	DA2N	Analog	Differential PGA 2 negative input.
AIO5	23	AIO5	I/O	Analog front end I/O 5.
AlO3	23	DA2P	Analog	Differential PGA 2 positive input.
		AIO6	I/O	Analog front end I/O 6.
		AMP6	Analog	PGA input 6.
AIO6	38	CMP6	Analog	Comparator input 6.
		BUF6	Analog	Buffer output 6.
		PBTN	Analog	Push button input.
		AIO7	I/O	Analog front end I/O 7.
AIO7	44	AMP7	Analog	PGA input 7.
AIO7		CMP7	Analog	Comparator input 7.
		PHC7	Analog	Phase comparator input 7.
		AIO8	I/O	Analog front end I/O 8.
AIO8	45	AMP8	Analog	PGA input 8.
AIOo	45	CMP8	Analog	Comparator input 8.
		PHC8	Analog	Phase comparator input 8.
		AIO9	I/O	Analog front end I/O 9.
AIO9	46	AMP9	Analog	PGA input 9.
AiO9	46	CMP9	Analog	Comparator input 9.
		PHC9	Analog	Phase comparator input 9.

9.3 Driver Manager Pin Description

PIN NAME	PIN NUMBER	TYPE	DESCRIPTION
DRS3	25	Analog	High-side gate driver source 3.
DRH3	26	Analog	High-side gate driver 3.
DRB3	27	Analog	High-side gate driver bootstrap 3. Connect a 1µF or higher value ceramic capacitor from DRB3 to DRS3 with a 10mm or shorter trace from the pin.
DRS4	28	Analog	High-side gate driver source 4.
DRH4	29	Analog	High-side gate driver 4.
DRB4	30	Analog	High-side gate driver bootstrap 4. Connect a 1µF or higher value ceramic capacitor from DRB4 to DRS4 with a 10mm or shorter trace from the pin.
DRS5	31	Analog	High-side gate driver source 5.
DRH5	32	Analog	High-side gate driver 5.
DRB5	33	Analog	High-side gate driver bootstrap 5. Connect a 1µF or higher value ceramic capacitor from DRB5 to DRS5 with a 10mm or shorter trace from the pin.
DRL0	39	Analog	Low-side gate driver 0.
DRL1	40	Analog	Low-side gate driver 1.
DRL2	41	Analog	Low-side gate driver 2.



9.4 I/O Ports Pin Description

PIN NAME	PIN NUMBER	FUNCTION	TYPE	DESCRIPTION ²
DD4		PD4	I/O	I/O port D4.
PD4	1	PWMD1	I/O	Timer D PWM/capture 1.
		PD3	I/O	I/O port D3.
DD2	0	PWMA5	I/O	Timer A PWM/capture 5.
PD3	2	PWMA7	I/O	Timer A PWM/capture 7.
		PWMB1	I/O	Timer B PWM/capture 1.
		PD2	I/O	I/O port D2.
DDO	0	PWMA3	I/O	Timer A PWM/capture 3.
PD2	3	PWMA4	I/O	Timer A PWM/capture 4.
		PWMB0	I/O	Timer B PWM/capture 0.
		PD1	I/O	I/O port D1.
PD1	4	SWDCL	I	Serial wire debug clock.
		EXTCLK	I	External clock input.
DDO	-	PD0	I/O	I/O port D0.
PD0	5	SWDIO	I/O	Serial wire debug I/O.
DEO		PE0	I/O	I/O port E0.
PE0	6	SPICLK	I/O	SPI clock.
		PE1	I/O	I/O port E1.
PE1	7	SPIMOSI	I/O	SPI master out, slave in (MOSI)
		UARTTX	0	UART transmit output.
		PE2	I/O	I/O port E2.
PE2	8	SPI MISO	I/O	SPI master in, slave out (MISO)
		UARTRX	I	UART receive input.
		PE3	I/O	I/O port E3.
PE3	9	SPICS0	0	SPI chip select 0.
		nRESET1	I	Reset input 1.
		PE4	I/O	I/O port E4.
PE4	10	SPICS1	0	SPI chip select 1.
		I2CSCL	I/O	I2C clock
		PE5	I/O	I/O port E5.
PE5	PE5 11		0	SPI chip select 2.
		I2CSDA	I/O	I2C data.
PC3	14	PC3	I/O	I/O port C3.
FUJ	14	AD3	Analog	ADC input 3.
PC2	15	PC2	I/O	I/O port C2.
1 02	10	AD2	Analog	ADC input 2.

 $^{^2}$ For a full description of all of the pin configurations for each digital I/O, see the PAC5232 User Guide for the Peripheral MUX.



		PD7	I/O	I/O port D7.
PD7	49	PWMA6	I/O	Timer A PWM/capture 6.
		PWMD0	I/O	Timer D PWM/capture 0.
		PD6	I/O	I/O port D6.
PD6	50	PWMA7	I/O	Timer A PWM/capture 7.
		PWMB1	I/O	Timer B PWM/capture 1.
		PD5	I/O	I/O port D5.
PD5	51	PWMA5	I/O	Timer A PWM/capture 5.
		PWMC1	I/O	Timer C PWM/capture 1.

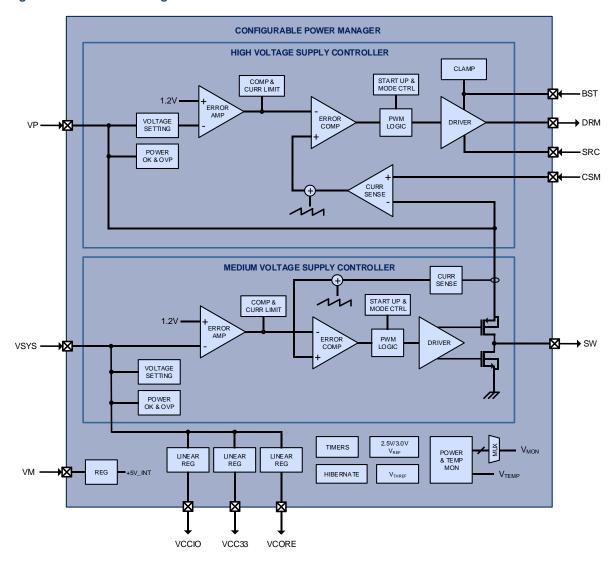


10 CONFIGURABLE POWER MANAGER (CPM)

10.1 Features

- 160V Buck DC/DC Controller (HV Buck)
 - o 25V 160V input
- 5V Switching Regulator (MV Buck)
- 3 linear regulators with power and hibernate management, including V_{REF} for ADC
- Power and temperature monitor, warning, and fault detection

Figure 10-1 CPM Block Diagram



10.2 Functional Description

The Configurable Power Manager (Figure 10-1) is optimized to efficiently provide "all-in-one" power management required by the PAC® and associated application circuitry. It incorporates a high-voltage





power supply controller that is used to convert power from a DC input source to generate a main supply output V_P. There is also an integrated medium-voltage buck DC/DC regulator to generate V_{SYS}.

Three other linear regulators provide V_{CCIO}, V_{CC33}, and V_{CORE} supplies for 3.3V I/O, 3.3V mixed signal, and 1.9V microcontroller core circuitry. The power manager also handles system functions including internal reference generation, timers, hibernate mode management, and power and temperature monitoring.

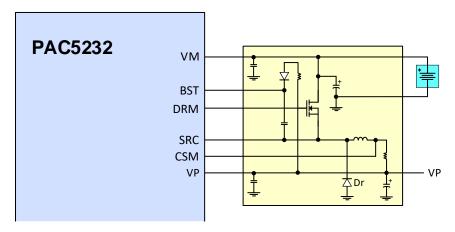
10.3 High-Voltage Supply Controller (HV-BUCK)

The PAC5232 contains a High-Voltage Supply Controller for a Buck DC/DC. This power supply is used to supply the various regulators in the PAC5232, as generating the V_P gate drive voltage for the Application Specific Driver Manager (ASPD).

The HV-BUCK controller drives an external power MOSFET for pulse-width modulation switching of an inductor or transformer for power conversion. The VM is the HV-BUCK supply controller input. The DRM output drives the gate of the N-CH MOSFET between the VM on state and Vss off state at proper duty cycle and switching frequency to ensure that the main supply voltage VP is regulated. The gate of the high-side power MOSFET is connected to the DRM pin and the source of the high-side power MOSFET is connected to SRC.

The V_P regulation voltage is initially set to 12V during start up, and can be reconfigured to be 15V by the microcontroller after initialization. When V_P is lower than the target regulation voltage, the internal feedback control circuitry causes the inductor current to increase to raise V_P . Conversely, when V_P is higher than the regulation voltage, the feedback loop control causes the inductor current to decrease to lower V_P . The feedback loop is internally stabilized. The output current capability of the switching supply is determined by the external current sense resistor. The inductor current signal is sensed differentially between the CSM pin and V_P , and has a peak current limit threshold of 0.2V.

Figure 10-2 HV-BUCK Example



The switching frequency and output voltage of the HV-BUCK can be reconfigured by the MCU. The switching frequency can be configured to be between 50kHz and 400kHz and the gate drive output voltage can be configured to either 12V or 15V to work for a range of MOSFET or IGBT based inverters.

The Rectifier Diode (Dr) must be a low QRR diode.



10.3.1 HV-BUCK Re-start Handling

The HV-BUCK has a safety re-start mechanism that protects the device and external components in case of a DC/DC failure. This mechanism samples VM and VP when the MV-BUCK is re-started and may insert a delay before it allows the power supply to be re-started, in case of some type of short or damage with the power supply components on the PCB.

The re-start handling operates as described below.

In PAC5232, if the DC/DC has been disabled due to VM falling below $V_{UVLOF;VM}$, VM is sampled and if VM > 10V, then a 350ms delay is inserted, before sampling VM again and attempting a re-start of the DC/DC. Once the delay has expired and VM > $V_{UVLOR;VM}$, then the DC/DC will re-start. If VM falls below 8V, the DC/DC will re-start as an initial start-up without the 350ms delay.

In PAC5232A, if the DC/DC has been disabled due to VM falling below $V_{UVLOF;VM}$, VM is sampled and as soon as VM > $V_{UVLOR;VM}$, then the DC/DC will re-start. In this case, there is no delay before the re-start of the DC/DC.

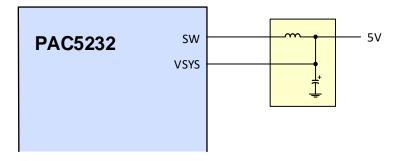
In both PAC5232 and PAC5232A, if VM is > VM UVLO falling ($V_{UVLOF;VP}$) but VP < VP UVLO falling ($V_{UVLOF;VP}$) then the DC/DC is disabled and a 350ms delay is inserted. After this delay, the DC/DC is restarted.

10.4 Medium-Voltage Buck Regulator (MV-BUCK)

The PAC5232 contains a Medium-Voltage Buck Switching Regulator that generates a 5V, 200mA supply for the device, as well as PCB functions.

The SW pin is the switch node of the Buck regulator. The Power MOSFET is integrated, so connect this pin to VSYS through an external inductor. The VSYS pin is the 5V regulator output, which should be bypassed to ground.

Figure 10-3 MV-BUCK Switching Regulator Example



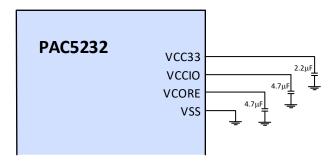
The output of VSYS is fixed at 5V and the switching frequency is 1.33MHz. This regulator supplies at least 200mA. This buck regulator offers better thermal and efficiency performance.

10.5 Linear Regulators

The CPM includes three additional linear regulators. VSYS supplies these three regulators. Once VSYS is above 4.5V, these three additional 40mA linear regulators for VCCIO, VCC33, and VCORE supplies sequentially power up.



Figure 10-4 Linear Regulators Example

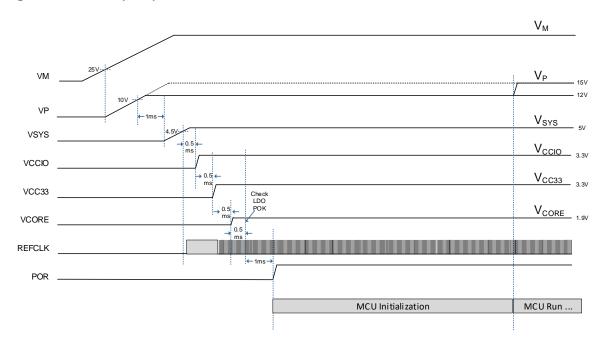


The figure above shows typical circuit connections for the linear regulators. The VCCIO regulator generates a dedicated 3.3V supply for IO. The VCC33 and VCORE regulators generate 3.3V and 1.9V, respectively. When VSYS, VCCIO, VCC33, and VCORE are all above their respective power good thresholds, and the configurable power on reset duration has expired, the microcontroller is initialized.³

10.6 Power-up Sequence

The CPM follows a typical power up sequence as in the Figure 10-5 below.

Figure 10-5 Power-Up Sequence



A typical sequence begins with motor power supply (VM) being applied and rising to 25V. When VM rises to 25V, the HV-BUCK controller is started and VP starts to rise. When VP rises over the UVLO rising threshold, then there is a 1ms delay and then the MV-BUCK is enabled. When VSYS rises to 4.5V, then

³ Note that the VCORE LDO may not have any addition load on it from the PCB. The only components connected to VCORE should be a bypass capacitor to ground.



there is a 0.5ms delay and the VCCIO LDO is enabled. Then there is a 0.5ms delay and the VCC33 LDO is enabled. Then there is a 0.5ms delay and the VCORE LDO is enabled.

There is then a 0.5ms delay and the power good threshold of all LDOs is checked. If all are OK, then there is an additional 1ms delay, then the POR signal is asserted to the MCU and it begins executing firmware.

During the firmware initialization process, the MCU may change the VP output voltage setting from the 12V default to 15V.

10.7 Hibernate Mode

The IC can go into an ultra-low power hibernate mode via the microcontroller firmware or via the optional push button (PBTN, see *Push Button* description in *Configurable Analog Front End*). In hibernate mode, only a minimal amount (typically 19 μ A at 56V) of current is used by V_M, and the CPM controller and all internal regulators are shut down to eliminate power drain from the output supplies. The system exits hibernate mode after a wake-up timer duration (configurable from 125ms to 8s or infinite) has expired or, if push button enabled, after an additional push button event has been detected. When exiting the hibernate mode, the power manager goes through the start up cycle and the microcontroller is reinitialized. Only the persistent power manager status bits (resets and faults) are retained during hibernation.

10.8 Power and Temperature Monitor

Whenever any of the V_{SYS}, V_{CCIO}, V_{CC33}, or V_{CORE} power supplies falls below their respective power good threshold voltage, a fault event is detected and the microcontroller is reset. The microcontroller stays in the reset state until V_{SYS}, V_{CCIO}, V_{CC33}, and V_{CC18} supply rails are all good again and the reset time has expired. A microcontroller reset can also be initiated by a maskable temperature fault event that occurs when the IC temperature reaches 165°C. The fault status bits are persistent during reset, and can be read by the microcontroller upon re-initialization to determine the cause of previous reset.

A power monitoring signal V_{MON} is provided onto the ADC pre-multiplexer for monitoring various internal power supplies. V_{MON} can be set to be V_{CORE}, 0.4•V_{CORE}, 0.4•V_{CCI3}, 0.4•V_{CCI0}, 0.4•V_{SYS}, 0.1•V_P, or the internal compensation voltage V_{COMP} for switching supply power monitoring.

For power and temperature warning, an IC temperature warning event at 140°C are provided as a maskable interrupt to the microcontroller. This warning allows the microcontroller to safely power down the system.

In addition to the temperature warning interrupt and fault reset, a temperature monitor signal is provided onto the ADC pre-multiplexer for IC temperature measurement.

This value has a compensation coefficient available in INFO FLASH that can be used to obatin an accurate temperature. The parameter VT300K will be stored in INFO FLASH and will indicate the compensation factor.

The temperature can then be obatined by the following formula:

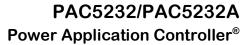
 $V_{\text{TEMP_KELVIN}} = 300 * (VM + 0.075) / (VT300K + 0.075)$



For information on on the location of this temperature coefficient, see the PAC5232 User Guide.

10.9 Voltage Reference

The reference block includes a 1.2V high-precision reference voltage used internally and for all the LDOs. There is also a high-accuracy 2.5V reference for the ADC V_{REF} on the MCU. There is also a 4-level programmable threshold voltage V_{THREF} (0.1V, 0.2V, 0.5V, and 1.25V).





10.10 Electrical Characteristics

Table 10-1 High-Voltage Buck Controller Electrical Characteristics

 $(V_M = 30V, V_P = 12V \text{ and } T_J = 25^{\circ}C \text{ unless otherwise specified})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
	V _M hibernate mode supply current	Hibernate mode, VM = 56V		19	26	μΑ
І нів;∨м	v _M nibernate mode supply current	Hibernate mode, VM = 80V		22.5		μΑ
V _{UVLOR;VM}	V _M UVLO rising		23	25	27	V
V _{UVLOF;VM}	V _M UVLO hysteresis			8		V
$V_{REF;VP}$	V _P output regulation voltage	Set to 12V	-5%	12	-5%	V
k _{POKR;VP}	V nower OK throubold	V _P rising		91		%
k pokf;vp	V _P power OK threshold	V _P falling		87		%
kovpr;vp	V _P OV protection threshold	V _P rising, blanking = 10μs		130		%
tonmin;drm	DRM minimum on time		90	200	300	ns
toffmin;drm	DRM minimum off time		390	600	1150	ns
V _{UVLOR;VP}	V _P UVLO rising			10		V
$V_{\text{UVLOF;VP}}$	V _P UVLO falling			8		V
V _{CSM;ILIM}	CSM current limit threshold		-12%	0.2	12%	V
F _{S;DRM}	Switching frequency	Frequency setting: 50kHz, 100kHz (default), 200kHz, 400kHz	-5		5	%
ISOURCE;DRM	DRM output high source current			100		mA
Isink;drm	DRM output low sink current			200	-5% 300 1150	mA
	HV-BUCK inductor value			100		μΗ
I _{DSG}	Discharge current			10		mA
V _{VM}	Motor voltage range		0		160	V
V _{SRC;VSS}	SRC to ground range		-10			V
$V_{\text{SRC;VM}}$	SRC to VM range					V
V _{BST;VSS}	BST to ground range				175	V



Table 10-2 Medium-Voltage Buck Controller Electrical Characteristics

 $(V_M = 30V, V_P = 12V \text{ and } T_J = 25^{\circ}C \text{ unless otherwise specified})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
V _{VSYS}	V _{SYS} output voltage accuracy		-3%	5	3%	V
Fsw	Switching frequency		-5%	1.33	5%	MHz
I _{VSYS;LIM}	V _{SYS} current limit		420		550	mA
1	V guitaut gurrant	V _{SYS} > 3V	200			mA
Ivsys	V _{SYS} output current	V _{SYS} < 2.5V	100			mA
V	V novem OV through and	Rising	4.25	4.5	4.75	V
V _{POK;} vsys	V _{SYS} power OK threshold	Falling		4.2		V
	V _{SYS} power OK blanking delay			10		μS
	MV-BUCK inductor value	Current rating of at least 750mA	6.8 – 20%		10 + 20%	μН
M	V 11V/10	Rising		4.5		V
Vuvlo;vsys	V _{SYS} UVLO	Falling		4.2		V
V	V OVD	Rising		5.5		V
V _{OVP} ;vsys	V _{SYS} OVP	Falling		5.2		V



Table 10-3 Linear Regulators Electrical Characteristics

 $(V_P = 12V \text{ and } T_A = -40^{\circ}\text{C to } 105^{\circ}\text{C unless otherwise specified})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Vccio	V _{CCIO} output voltage	Load = 1mA	-3%	3.3	3%	٧
V _{CCIO}	V _{CCIO} output voltage	Load = 1mA	-3%	3.3	3%	V
V _{CORE} ⁴	V _{CORE} output voltage	Load = 1mA	-3%	1.9	3%	V
I _{LIM;VCCIO}	V _{CCIO} current limit		40	65		mA
I _{LIM;VCC33}	V _{CC33} current limit		40	65		mA
I _{LIM;VCORE}	V _{CORE} current limit		40	65		mA
	LDO current fold back			50		%
t _{POK;BLANK}	Power OK blanking delay	Vccio, Vcc33, Vcore		10		μS
Rdisch	Output discharge resistance	LDO off		300		Ohm
Cvccio	VCCIO stable output capacitance		1		4.7	μF
Суссзз	VCC33 stable output capacitance		1		4.7	μF
C _{VCORE}	VCORE stable output capacitance		1		4.7	μF
V _{LDO;POK}	LDO power OK rising threshold	Hysteresis = 10%	85	90	95	%

⁴ Note that the VCORE LDO may not have any other loads. The only connection to the VCORE pin should be a bypass capacitor to ground.



10.11 Typical Performance Characteristics

(T_A = 25°C unless otherwise specified)

Figure 10-6 VDDIO LDO Voltage vs. Current

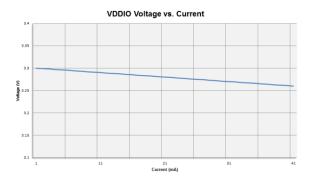


Figure 10-7 VCC33 LDO Voltage vs. Current

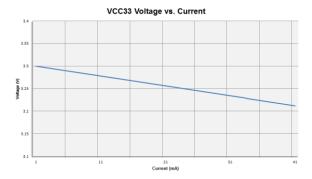
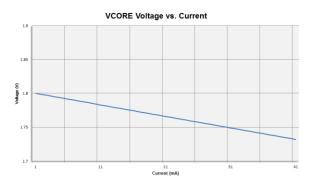


Figure 10-8 VCORE LDO Voltage vs. Current





11 CONFIGURABLE ANALOG FRONT END (CAFE)

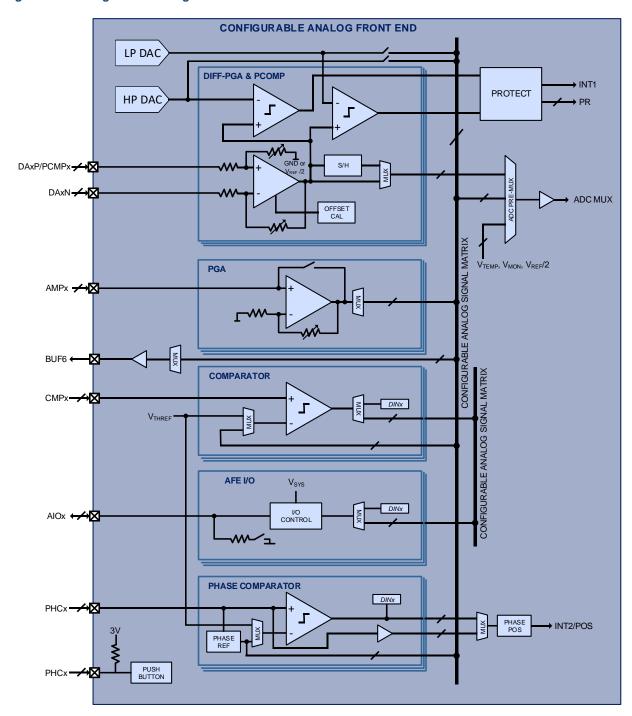
11.1 Features

- 10 Configurable Analog I/O signals
 - o Gain mode, comparator mode, I/O mode, special mode
- 3 High-Performance, Configurable Differential Amplifiers
- 4 High-Performance, Configurable Single-Ended Amplifiers
- Two high-speed comparators with protection functions
- Phase to phase, phase to center-tap modes
- Bi-directional, asymmetric configurable comparator hysteresis
- Push-button input for entering/exiting hibernate mode
- Low-frequency Clock Output for safety applications



11.2 Block Diagram

Figure 11-1 Configurable Analog Front End







11.3 Functional Description

The device includes a Configurable Analog Front End (CAFE, Figure 11-1) accessible through 8 analog and I/O pins. These pins can be configured to form flexible interconnected circuitry made up of 3 differential programmable gain amplifiers, 4 single-ended programmable gain amplifiers, 4 general purpose comparators, 3 phase comparators, 10 protection comparators, and one buffer output. These pins can also be programmed as analog feed-through pins, or as analog front end I/O pins that can function as digital inputs or digital open-drain outputs. The PAC® proprietary configurable analog signal matrix (CASM) and configurable digital signal matrix (CDSM) allow real time asynchronous analog and digital signals to be routed in flexible circuit connections for different applications. A push button function is provided for optional push button on, hibernate, and off power management function.

11.4 Differential Programmable Gain Amplifier (DA)

The DAxP and DAxN pin pair are positive and negative inputs, respectively, to a differential programmable gain amplifier. The differential gain can be programmable to be 1x, 2x, 4x, 8x, 16x, 32x, and 48x for zero ohm signal source impedance. The differential programmable gain amplifier has -0.3V to 2.5V input common mode range, and its output can be configured for routing directly to the ADC premultiplexer, or through a sample-and-hold circuit synchronized with the ADC auto-sampling mechanism. Each differential amplifier is accompanied by offset calibration circuitry, and two protection comparators for protection event monitoring. The programmable gain differential amplifier is optimized for use with signal source impedance lower than 500Ω and with matched source impedance on both positive and negative inputs for minimal offset. The effective gain is scaled by $13.5k / (13.5k + R_{SOURCE})$, where R_{SOURCE} is the matched source impedance of each input.

11.5 Single-Ended Programmable Gain Amplifier (AMP)

Each AMPx input goes to a single-ended programmable gain amplifier with signal relative to V_{SS}. The amplifier gain can be programmed to be 1x, 2x, 4x, 8x, 16x, 32x, and 48x, or as analog feed-through. The programmable gain amplifier output is routed via a multiplexer to the configurable analog signal matrix CASM.

11.6 General Purpose Comparator (CMP)

The general purpose comparator takes the CMPx input and compares it to either the programmable threshold voltage (V_{THREF}) or a signal from the configurable analog signal matrix CASM. The comparator has 0V to V_{SYS} input common mode range, and its polarity-selectable output is routed via a multiplexer to either a data input bit or the configurable digital signal matrix CDSM. Each general purpose comparator has two mask bits to prevent or allow rising or falling edge of its output to trigger second microcontroller interrupt INT2, where INT2 can be configured to active protection event PR.

11.7 Phase Comparator (PHC)

The phase comparator takes the PHCx input and compares it to either the programmable threshold voltage (V_{THREF}) or a signal from the configurable analog signal matrix CASM. The comparison signal can be set to a phase reference signal generated by averaging the PHCx input voltages. In a three-phase motor control application, the phase reference signal acts as a virtual center tap for BEMF detection. The PHCx inputs are optionally fed through to the CASM. The PHC inputs can be compared to the virtual center-tap, or phase to phase for the most efficient BEMF zero-cross detection.



The phase comparator signals can also be configured to the other two phase comparators (between AIO7, AIO8 and AIO9), to perform phase to phase comparisons.

The comparator blanking time is configurable. The blanking time configuration supports bi-directional and asymmetric configurations, which enables hysteresis for rising and falling signals.

The phase comparator has 0V to V_{SYS} input common mode range, and its polarity-selectable output is routed to a data input bit and to the phase/position multiplexer synchronized with the auto-sampling sequencers.

11.8 Protection Comparator (PCMP)

Two protection comparators are provided in association with each differential programmable gain amplifier, with outputs available to trigger protection events and accessible as read-back output bits. The HP comparator compares the amplifier output to the 10-bit HP DAC output voltage, with full scale voltage of 2.5V. The limit protection (LP) comparator compares the differential programmable gain amplifier output to the 10-bit LP DAC output voltage, with full scale voltage of 2.5V.

Each protection comparator has a mask bit to prevent or allow it to trigger the main microcontroller interrupt INT1. Each protection comparator also has one mask bit to prevent or allow it to activate protection event PR. These two protection events can be used directly by protection circuitry in the Application Specific Power Drivers (ASPD) to protect devices being driven.

11.9 Analog Output Buffer (BUF)

A subset of the signals from the configurable analog signal matrix CASM can be multiplexed to the BUF6 pin for external use. The buffer offset voltage can be minimized with the built-in swap function.

11.10 Analog Front End I/O (AIO)

The PAC5232 has 10 AlOx pins that are available. In the analog front end I/O mode, the pin can be configured to be a digital input or digital open-drain output. The AlOx input or output signal can be set to a data input or output register bit, or multiplexed to one of the signals in the configurable digital signal matrix CDSM. The signal can be set to active high (default) or active low, with V_{SYS} supply rail. Where AlO_{6,7,8,9} supports microcontroller interrupt for external signals. Each has two mask bits to prevent or allow rising or falling edge of its corresponding digital input to trigger second microcontroller interrupt INT2.

11.11 Push Button (PBTN)

The push button PBTN, when enabled, can be used by the MCU to detect a user active-low push button event and to put the system into an ultra-low-power hibernate mode. Once the system is in hibernate mode, PBTN can be used to wake up the system.

In addition, PBTN can also be used as a hardware reset for the microcontroller when it is held low for longer than 8s during normal operation. The PBTN input is active low and has a $55k\Omega$ pull-up resistor to 3V.



11.12 HP DAC and LP DAC

The 10-bit HP DAC can be used as the comparison voltage for the high-speed protection (HP) comparators, or routed for general purpose use via the AB2 signal in the CASM. The HP DAC output full scale voltage is 2.5V.

The 10-bit LP DAC can be used as the comparison voltage for the limit protection (LP) comparators, or routed for general purpose use via the AB3 signal in the CASM. The LP DAC output full scale voltage is 2.5V.

11.13 ADC Pre-Multiplexer

The ADC pre-multiplexer is a 16-to-1 multiplexer that selects between the 3 differential programmable gain amplifier outputs, AB1 through AB9, temperature monitor signal (V_{TEMP}), power monitor signal (V_{MON}), and offset calibration reference (V_{REF} / 2). The ADC pre-multiplexer can be directly controlled or automatically scanned by the auto-sampling sequencer.

When the ADC pre-multiplexer is automatically scanned, the unbuffered or sensitive signals should be masked by setting appropriate register bits.

11.14 Configurable Analog Signal Matrix (CASM)

The CASM has 9 general purpose analog signals labeled AB1 through AB9 that can be used for:

- Routing the single-ended programmable gain amplifier or analog feed-through output to AB1 through AB9
- Routing an analog signal via AB1, AB2, or AB3 to the negative input of a general purpose comparator or phase comparator
- Routing the 10-bit HP DAC output to AB2
- Routing the 10-bit LP DAC output to AB3
- Routing analog signals via AB1 through AB12 to the ADC pre-multiplexer
- Routing phase comparator feed-through signals to AB7, AB8, and AB9, and averaged voltage to AB1

11.15 Configurable Digital Signal Matrix (CDSM)

The CDSM has 7 general purpose bi-directional digital signals labeled DB1 through DB7 that can be used for:

- Routing the AlOx input to or output signals from DB1 through DB7
- Routing the general purpose comparator output signals to DB1 through DB7

11.16 Low-Frequency Clock Output

The PAC5232 has a configurable low-frequency clock output. When enabled, the device will output a 250Hz, 50% duty cycle clock to the CLKOUT pin.

The supported clock frequencies are 250Hz, 500Hz, 1kHz or 2kHz.



11.17 Cycle-by-cycle Current Limit

The PAC5232 contains hardware support for cycle by cycle current limit. The user may configure this feature to use the LPCOMP DAC as the current threshold. The CAFE will automatically perform duty cycle truncation to lower current at any time the associated phase current is greater than the setting of the LPCOMP DAC.

11.18 Temperature Protection

The PAC5232 contains an internal temperature sensor that detects temperature warnings and faults.

When the device temperature reaches the temperature warning threshold (140°C), the device sets an over-temperature warning condition. The user may configure a mask-able interrupt the MCU for this condition.

When the device temperature reaches the temperature fault threshold (165 °C), the device is shut down. There is no interrupt for this condition.

For more details on the register settings for over-temperature protection see the PAC5232 User Guide and related application notes.



11.19 Electrical Characteristics

Table 11-1 Differential Programmable Gain Amplifier (DA) Electrical Characteristics (AIO<5:0>)

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
V _{ICMR;DA}	Input common mode range		-0.3		2.5	V
$V_{OLR;DA}$	Output linear range		0.1		V _{SYS} – 0.1	V
V _{SHR;DA}	Sample and hold range		0.1		3.5	
I _{CC;DA}	Operating supply current	Each enabled amplifier		150		μΑ
V _{OS;DA}	Input offset voltage	Gain = 8x	-8		8	mV
k _{CMRR;DA}	Common mode rejection ratio		50	80		dB
	Slew rate	Gain = 8x	10			V/μs
R _{INDIF;DA}	Differential input impedance			27		kΩ
t _{ST;DA}	Settling time	To 1% of final value			360	ns
		Gain = 1x	 			
		Gain = 2x		2		
		Gain = 4x		4		
$A_{VZI;DA}$	Differential amplifier gain (zero ohm source impedance)	Gain = 8x, $V_{DAxP}=V_{DAxN}=0V$, $T_A = 25^{\circ}C$	-2		2	%
		Gain = 16x		16	2.5 V _{SYS} - 0.1 3.5 150 8 80 27 360 1 2 4	
		Gain = 32x		32		
		Gain = 48x		48		



Table 11-2 Single-Ended Programmable Gain Amplifier (AMP) Electrical Characteristics (AIO<9:6>)

(V_{SYS} = 5V, V_{CCIO} = 3.3V and T_A = -40°C to 105°C unless otherwise specified.)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
VICMR;AMP	Input common mode range		0		Vsys	V
V _{OLR;AMP}	Output linear range		0.1		V _{SYS} - 0.1	V
I _{CC;AMP}	Operating supply current	Each enabled amplifier		80	120	μΑ
V _{OS;AMP}	Input offset voltage	Gain = 8x	-10		10	mV
	Slew rate	Gain = 1x	10			V/μs
tst;amp	Settling time	To 1% of final value			360	ns
		Gain = 1x		1		
		Gain = 2x		2		
		Gain = 4x		4		
A _{V;AMP}	Amplifier gain	Gain = 8x, V _{AMPx} =125mV, T _A = 25°C	-2		2	%
		Gain = 16x		16		
		Gain = 32x		32		
		Gain = 48x		48		
t _{ST;AMP}	Settling time	To 1% of final value			350	ns

Table 11-3 General Purpose Comparator (CMP) Electrical Characteristics (AIO<9:6>)

 $(T_A = -40$ °C to 105°C unless otherwise specified.)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
VICMR;CMP	Input common mode range		0		Vsys	V
I _{CC;CMP}	Operating supply current	Each enabled comparator		35		μΑ
V _{OS;CMP}	Input offset voltage		-10		10	mV
V _{HYS;CMP}	Hysteresis			22		mV
t _{DEL;CMP}	Comparator delay				1	μS
tdelmode;cmp	Mode change blanking delay			10		μs

Table 11-4 Phase Comparator (PHC) Electrical Characteristics (AIO<9:6>)

($V_{SYS} = 5V$, $V_{CC33} = 3.3V$ and $T_A = -40$ °C to 105°C unless otherwise specified.)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
V _{ICMR;PHC}	Input common mode range					V
I _{CC;PHC}	Operating supply current	Each enabled comparator		35		μΑ
Vos;PHC	Input offset voltage		-10		10	mV
V _{HYS;PHC}	Hysteresis			23		mV
t _{DEL;PHC}	Comparator delay	10mV difference input			1	μs



Table 11-5 Special Mode Electrical Characteristics (AIO<9:7>)

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
VICMR;SPEC	Input common mode range		0		Vsys	V
I _{CC;SPEC}	Operating supply current	Each enabled comparator		80	120	μА
	Comparator Hysteresis,	AIO<9:7>HYS = 00b (0mV)		0		mV
		AIO<9:7>HYS = 01b (6mV)	4	6	8	mV
	HYSMODE = 0	AIO<9:7>HYS = 10b (12mV)	9	12	15	mV
V		AIO<9:7>HYS = 11b (24mV)	18	24	30	mV
VHSYS;SPEC		AIO<9:7>HYS = 00b (0mV)		0		mV
	Comparator Hysteresis,	AIO<9:7>HYS = 01b (24mV)	18	24	30	mV
	HYSMODE = 1	AIO<9:7>HYS = 10b (48mV)	36	48	60	mV
		AIO<9:7>HYS = 11b (96mV)	72	96	120	mV

Table 11-6 Special Mode Electrical Characteristics (AIO6)

(T_A = -40°C to 105°C unless otherwise specified.)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
VICMR;SPEC	Input common mode range		0		Vsys	V
I _{CC;SPEC6}	Operating supply current			60		μΑ
VINOFF;SPEC6	Input offset voltage		-20		20	mV
I _{OUT;SPEC6}	Output current			2		mA

Table 11-7 Analog Front End (AIO) Electrical Characteristics (AIO<9:0>)

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
VAIO	Pin voltage range		0		5	V
V _{IH;AIO}	High-level input voltage		2.2			V
V _{IL;AIO}	Low-level input voltage				0.8	V
R _{PD;AIO}	Pull-down resistance	Input mode		1		МΩ
V _{OL;AIO}	Low-level output voltage	I _{AIOx} =7mA, open-drain output mode			0.3	V
I _{OL;AIO}	Low-level output sink current	V _{AlOx} = 0.4V, open-drain output mode	6	14		mA
I _{LK;AIO}	High-level output leakage current	V _{AlOx} = 5V, open-drain output mode		0	10	μА



Table 11-8 Push Button (PBTN) Electrical Characteristics (AIO6)

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
V _{I;PBTN}	Input voltage range		0		5	V
V _{IH;PBTN}	High-level input voltage		2.2			V
V _{IL;PBTN}	Low-level input voltage				0.8	V
R _{PU;PBTN}	Pull-up resistance	To 3V, push-button input mode		50		kΩ

Table 11-9 HP DAC and LP DAC Electrical Characteristics

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
V	DAC reference voltage	TA = 25°C	-0.5%	2.5	0.5%	V
V DACREF	V _{DACREF} DAC reference voltage	TA = -40°C to 105°C	-0.9%	2.5	0.9%	V
	HP 10-bit DAN INL		-2		2	LSB
	HP 10-bit DAC DNL		-1		1	LSB
	LP 10-bit DAC INL		-2		2	LSB
	LP 10-bit DAC DNL		-1		1	LSB

Table 11-10 Low-Frequency Clock Output (CLKOUT)

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
FCLKOUT	Low-speed clock output frequency	CLKOUT 250Hz		250		Hz
Г	_ Low-speed clock output frequency	T _A = 25°C	-10		10	%
FCLKOUT;ERR	error	T _A = -40°C - 105°C	-15		15	%

Table 11-11 Temperature Protection

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
T _{WARN}	Temperature warning threshold			140		°C
T _{WARN;HYS}	Temperature warning hysteresis			10		°C
T _{WARN;BLANK}	Temperature warning blanking			10		μs
T _{FAULT}	Temperature fault threshold			165		°C
T _{FAULT;HYS}	Temperature fault hysteresis			10		°C
T _{FAULT} ;BLANK	Temperature fault blanking			10		μs



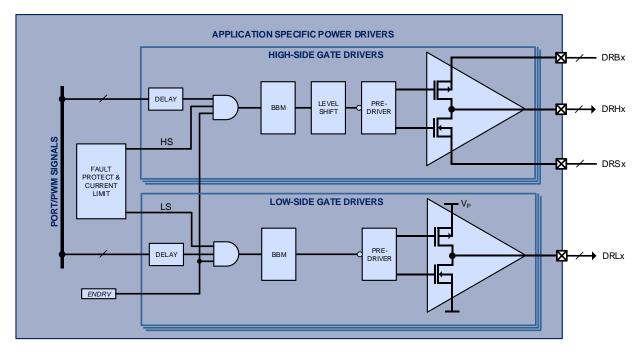
12 APPLICATION SPECIFIC POWER DRIVERS (ASPD)

12.1 Features

- 3 low-side and 3 high-side gate drivers
- 2A sink/source gate driving capability
- Configurable propagation delays
- Fast fault protection
- Cycle-by-cycle current limit function
- Configurable driver break-before-make (BBM) safety function

12.2 Block Diagram

Figure 12-1 Application Specific Power Drivers



12.3 Functional Description

The Application Specific Power Drivers (ASPD, Figure 12-1) module handles power driving for power and motor control applications. The ASPD contains three low-side gate drivers (DRLx), three high-side gate drivers (DRHx). Each gate driver can drive an external MOSFET or IGBT switch in response to high-speed control signals from the microcontroller ports, and a pair of high-side and low-side gate drivers can form a half-bridge driver.

Figure 12-2 below shows typical gate driver connections and Table 12-1 shows the ASPD available resources. The ASPD gate drivers support up to a 180V source supply.



Figure 12-2 Typical Gate Driver Connections

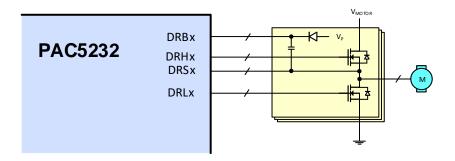


Table 12-1 Power Driver Resources by Part Numbers

PART	LOW-SI	IDE GATE DRIVER	HIGH-SIDE GATE DRIVER			
NUMBER	DRLx	SOURCE/SINK CURRENT	DRHx	SOURCE/BOOTSTRAP SOURCE/SINK CURREN		
PAC5232	3	2A/2A	3	165V/180V	2A/2A	

The ASPD includes built-in configurable fault protection for the internal gate drivers.

12.4 Low-Side Gate Driver

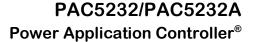
The DRLx low-side gate driver drives the gate of an external MOSFET or IGBT switch between the low-level power ground rail and high-level V_P supply rail. The DRLx output pin has sink and source output current capability of 2A. Each low-side gate driver is controlled by a microcontroller port signal with 4 configurable levels of propagation delay.

12.5 High-Side Gate Driver

The DRHx high-side gate driver drives the gate of an external MOSFET or IGBT switch between its low-level DRSx driver source rail and its high-level DRBx bootstrap rail. The DRSx pin can go up to 160V steady state (VM + 15V maximum). The DRHx output pin has sink and source output current capability of 2A.

The DRBx bootstrap pin can have a maximum operating voltage of 15V relative to the DRSx pin, and up to 175V steady state. The DRSx pin can have a maximum operating voltage of 10V relative to the VM pin. The DRSx pin is designed to tolerate momentary switching negative spikes down to -10V without affecting the DRHx output state. Each high-side gate driver is controlled by a microcontroller port signal with 4 configurable levels of propagation delay.

For bootstrapped high-side operation, connect an appropriate capacitor between DRBx and DRSx and a properly rated bootstrap diode from V_P to DRBx. To operate the DRHx output as a low-side gate driver, connect its DRBx pin to V_P and its DRSx pin to V_{SS}.





12.6 Power Drivers Control

All power drivers are initially disabled from power-on-reset. To enable the power drivers, the MCU must first enable the ASPD block. The gate drivers are controlled by the microcontroller ports and/or PWM signals according to Table 12-2, with configurable delays as shown in Table 12-3 Power Driver Propagation Delay.

Refer to the PAC5232 User Guide and PAC® application notes and user guide for additional information on power drivers control programming.

Table 12-2 Microcontroller Port and PWM to Power Driver Mapping

PART NUMBER	DRL0	DRL1	DRL2	DRH3	DRH4	DRH5
PAC5232	PWMA0	PWMA1	PWMA2	PWMA3/ PWMA4/ PWMB0	PWMA5/ PWMC0	PWMA6/ PWMD0

Table 12-3 Power Driver Propagation Delay

DRLx	DRHx
135ns	155ns

12.7 Gate Driver Fault Protection

The ASPD incorporates a configurable fault protection mechanism using protection signal from the Configurable Analog Front End (CAFE), designated as protection event 1 (PR) signal. The DRL0/DRL1/DRL2 drivers are designated as low-side group 1. The DRH3/DRH4/DRH5 gate drivers are designated as high-side group 1. The PR signal from the CAFE can be used to disable low-side group 1, high-side group 1, or both depending on the PR mask bit settings.



12.8 Electrical Characteristics

Table 12-4 Gate Driver Electrical Characteristics

 $(V_P = 12V, \text{ and } T_A = -40^{\circ}\text{C to } 105^{\circ}\text{C unless otherwise specified.}$

	A = -40°C to 105°C unless otherwise specif					
SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Low-Side Gate	Drivers (DRLx pins)					
$V_{\text{OH;DRL}}$	High-level output voltage	I _{DRLx} = -50mA	V _P -0.3			V
$V_{OL;DRL}$	Low-level output voltage	I _{DRLx} = 50mA			0.3	V
I _{OHPK;DRL}	Output high source current	10μs pulse		2		Α
I _{OLPK;DRL}	Output low sink current	10μs pulse		2		А
High-Side Gate	e Drivers (DRHx, DRBx and DRSx pins)					
V _{DRS}	Level-shift driver source voltage range		-10		VM + 10	V
VDRB	Poetstrop pin voltage range	Relative to V _{DRS}	10		20	V
V DRB	Bootstrap pin voltage range	Relative to VSS			175	V
Vuvlo:drb	Bootstrap UVLO threshold	V _{DRBx} rising	8.5			V
V UVLO;DRB	Bootstrap OVLO trireshold	Hysteresis		1		V
I _{BS;DRB}	Bootstrap supply current	Current from DRBx to DRSx		28		μА
I _{OS;DRB}	Offset supply current	Current from DRBx to ground		10		μА
$V_{\text{OH;DRH}}$	High-Level output voltage	$I_{DRHx} = -50mA$	V _{DRBx} - 0.3			V
$V_{\text{OL;DRH}}$	Low-level output voltage	I _{DRHx} = 50mA			V _{DRSx} +0.3	V
I _{OHPK;DRH}	Output high source current	10μs pulse		2		Α
I _{OLPK;DRL}	Output low sink current	10μs pulse		2		Α
High-Side and	Low-Side Gate Driver Propagation Dela	у				
		Delay setting 00b	-50%	Delay +	0 50%	ns
		Delay setting 01b	-50%	Delay +	50 50%	ns
t PD	Propagation Delay⁵	Delay setting 10b	-50%	Delay - 100	+ 50%	ns
		Delay setting 11b	-50%	Delay - 200	+ 50%	ns

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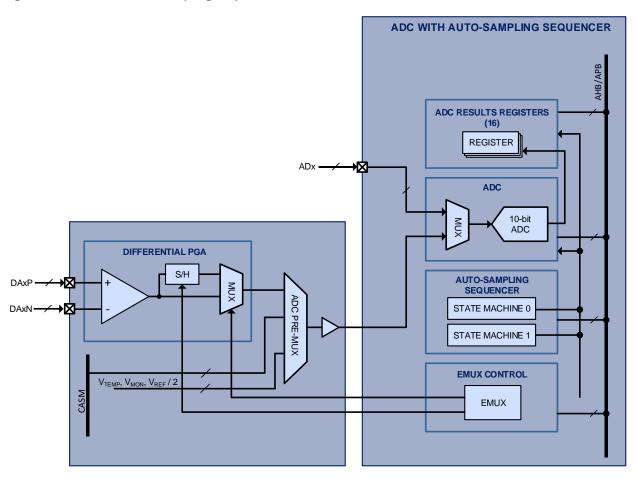
⁵ Delay from Table 12-3 Power Driver Propagation Delay



13 ADC WITH AUTO-SAMPLING SEQUENCER

13.1 ADC Block Diagram

Figure 13-1 ADC with Auto-Sampling Sequencer



13.2 Functional Description

13.2.1 ADC

The analog-to-digital converter (ADC) is a 10-bit successive approximation register (SAR) ADC with 1 µs conversion time and up to 1MSPS capability. The ADC input clock has a user-configurable divider from /1 to /8 of the system clock. The integrated analog multiplexer allows selection from up to 6 direct ADx inputs, and from up to 10 analog inputs signals in the Configurable Analog Front End (CAFE), including up to 3 differential input pairs. The ADC can be configured for repeating or non-repeating conversions and can interrupt the microcontroller when a conversion is finished.



13.2.2 Auto-Sampling Sequencer

Two independent and flexible auto-sampling sequencer state machines allow signal sampling using the ADC without interaction from MCU. Each auto-sampling sequencer state machine can be programmed to take and store up to 8 samples each in the ADC result registers from different analog inputs, is able to control the ADC MUX and ADC Pre-mux as well as the precise timing of the S/H in the Configurable Analog Front-End (CAFE). The sampling start of the auto-sampling sequencer can be precisely triggered using timers A, B, C, or D or any of their associated PWM edges (high-to-low or low-to-high). It also supports manual start or a ping-pong-scheme, where one auto-sampling sequencer state machine triggers the other when it finishes sampling.

The auto-sampling sequencer can interrupt the microcontroller when either conversion sequence is finished.

13.2.3 EMUX Control

A dedicated low latency interface controllable by the auto-sampling sequencer or register control allows changing the ADC pre-multiplexer and asserting/de-asserting the S/H circuit in the Configurable Analog Front-End (CAFE), allowing back to back conversions of multiple analog inputs without MCU interaction.

For more information on the ADC and Auto-Sequencer, see the PAC5232 User Guide.



13.3 Electrical Characteristics

Table 13-1 ADC and Sequencer Electrical Characteristics

 $(V_{CCIO} = 5V \text{ and } T_A = -40^{\circ}\text{C to } 105^{\circ}\text{C unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
ADC						
f _{ADCCLK}	ADC conversion clock input				16	MHz
fadcconv	ADC conversion time	fadcclk = 16MHz			1	μs
	ADC resolution			10		bits
	ADC effective resolution		9.2			bits
	ADC differential non-linearity (DNL)			±0.5		LSB
	ADC integral non-linearity (INL)			±1		LSB
	ADC offset error			0.6		%FS
	ADC gain error			0.12		%FS
REFERENCE	VOLTAGE					
V _{REFADC}	ADC reference voltage input	V _{REF} = 2.5V		2.5		V
	ADC reference voltage input error	T _A = 25°C	-0.5		0.5	%
SAMPLE AN	D HOLD					
tadcsh	ADC sample and hold time	fadcclk = 16MHz		188		μs
C _{ADCIC}	ADC input capacitance	ADC MUX input		1.3		pF
INPUT VOLT	AGE RANGE					
V _{ADCIN}	ADC input voltage range	ADC MUX input	0		V _{REFADC}	V
EMUX CLOC	K SPEED		•			•
f _{EMUXCLK}	EMUX engine clock input				50	MHz



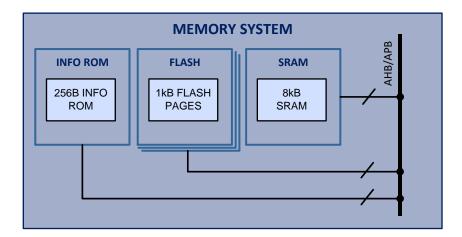
14 MEMORY SYSTEM

14.1 Features

- 32kB Embedded FLASH
 - o 100,000 program/erase cycles
 - o 10 years data retention
- 256B INFO Embedded FLASH
 - Device ID, calibration data
 - 0
- 8kB SRAM
 - 50MHz read/write
 - Data storage or code execution
- SWD code protection

14.2 Memory System Block Diagram

Figure 14-1 Memory System



14.3 Functional Description

The PAC5232 has multiple banks of embedded FLASH memory, SRAM memory as well as peripheral control registers that are program-accessible in a flat memory map.

14.4 Program FLASH

The PAC5232 Memory Controller provides access to 32 1kB pages of main program FLASH for a total of 32kB of FLASH through the system AHB bus. Each page may be individually erased or written while the MCU is executing instructions from SRAM.

14.5 INFO FLASH

The PAC5232 Memory Controller contains a 256B read-only INFO memory. This memory contains device-specific information such as the device ID and calibration data for the device.

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14.6 **SRAM**

The PAC5232 Memory Controller provides access to the 8kB SRAM for non-persistent data storage. The SRAM memory supports word (4B), half-word (2B) and byte addresses with aligned access.

The PAC5232 Memory Controller can read or write data from RAM up to 50MHz. This can be a benefit for time-critical applications. This memory can also be used for program execution when modifying the contents of FLASH.

For more information on the PAC5232 Memory Controller, see the PAC5232 User Guide.

14.7 SWD Protection

The PAC5232 allows the user to blow a fuse that prohibits use of the SWD serial interface.

This will prevent unauthorized users from accessing the contents of the device after customer production.

14.8 Electrical Characteristics

Table 14-1 Memory System Electrical Characteristics

 $(V_{CCIO} = 5V, T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Embedded FLASH						
tread;flash	FLASH read time		40			ns
twrite;flash	FLASH write time		20			μS
tperase;flash	FLASH page erase time				10	ms
N _{PERASE;FLASH}	FLASH program/erase cycles			100k		cycles
t _{DR;FLASH}	FLASH data retention		10			years
SRAM						
tacc;sram	SRAM access time	Word (32-bits), aligned	20			ns



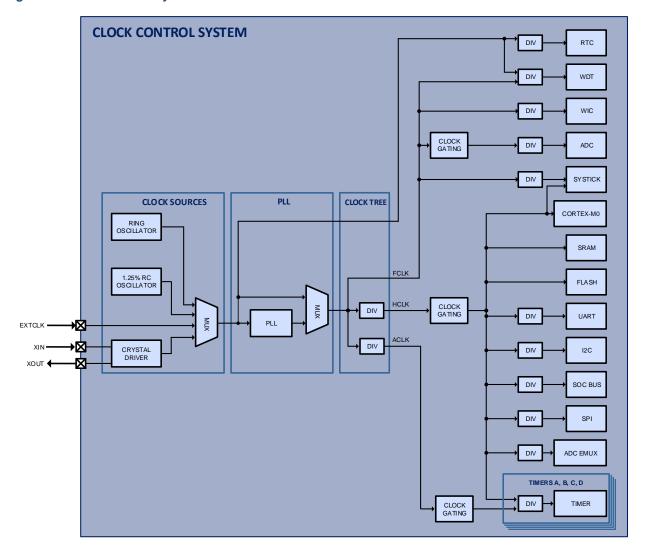
15 SYSTEM AND CLOCK CONTROL

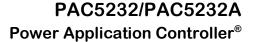
15.1 Features

- Ring oscillator with 7.5MHz, 9.6MHz, 13.8MHz, and 25.7MHz settings
- High accuracy 1.25% trimmed 4MHz RC oscillator
- External clock input up to 40MHz
- PLL with 1MHz to 25 MHz input, and 3.5MHz to 100MHz output
- /1 to /8 clock divider for HCLK
- /1 to /128 clock divider for ACLK

15.2 Block Diagram

Figure 15-1 Clock Control System







15.3 Functional Description

15.3.1 Free-Running Clock (FRCLK)

The free running clock (FRCLK) is generated from one of the 4 clock sources: ring oscillator, trimmed RC oscillator, crystal driver or external clock input. The FRCLK is used for the real-time clock (RTC), watchdog timer (WDT), input to the PLL, or FCLK source to clock the system in low power and sleep mode.

15.3.2 Auxiliary Clock (ACLK)

The auxiliary clock (ACLK) is derived from FCLK with a /1, /2, to /128 divider, and supplies the timer and dead-time blocks. It can be clocked faster or slower than HCLK and can go as high as 100MHz.

15.3.3 Clock Gating

The clock tree supports clock gating in deep-sleep mode for the timer block, ADC, SPI interface, I²C interface, UART interface, memory subsystem and the Arm[®] Cortex[®]-M0 itself.

15.3.4 Ring Oscillator (ROSC)

The integrated ring oscillator provides 4 different clocks with 7.5MHz, 9.6MHz, 13.8MHz, and 25.7MHz settings. After reset, the clock tree always defaults to this clock input with the lowest frequency setting.

15.3.5 Trimmed 4MHz RC Oscillator

The 1.25% trimmed 4MHz RC oscillator provides an accurate clock suitable for many applications. It is also used to derive the clock for the Multi-Mode Power Manager.

15.3.6 External Clock Input

The clock tree can be supplied with an external clock up to 40MHz.

15.3.7 PLL

The integrated PLL input clock is supplied by the FRCLK with an input frequency range of 1MHz to 25MHz. The PLL output frequency is adjustable from 3.5MHz to 100MHz.



15.4 Electrical Characteristics

Table 15-1 CCS Electrical Characteristics

 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	unless otherwise specified.) PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
Clock Tree (FRC	CLK, FCLK, ACLK, and HCLK)					
f _{FRCLK}	Free-running clock frequency				50	MHz
f _{FCLK}	Fast clock frequency				100	MHz
facik	Auxiliary clock frequency	After divider			100	MHz
f _{HCLK}	System clock frequency	After divider			50	MHz
Internal Oscillat	ors					
		Frequency setting = 11b		7.5		
£	Diamaga: Hatau fua maga	Frequency setting = 10b		9.6		- MHz
f _{ROSC}	Ring oscillator frequency	Frequency setting = 01b		13.8		
		Frequency setting = 00b		25.7		
,	Trimmed RC oscillator	T _A = 25°C	3.96	4	4.05	MHz
ITRIM	f _{TRIM} frequency	T _A = -40°C to 105°C	3.92	4	4.08	MHz
	Trimmed RC oscillator clock iitter	T _A = -40°C to 105°C		0.5		%
External Clock I	nput (EXTCLK)		•			
fextclk	External Clock Input Frequency				40	MHz
thigh;extclk	External Clock High Time		10			ns
t _{LOW;EXTCLK}	External Clock Low Time		10			ns
V _{IH;EXTCLK}	External Clock Input high- level input voltage		2.1			V
VIL;EXTCLK	External Clock Input low- level input voltage				0.825	V
PLL	· · · · · · · · · · · · · · · · · · ·					
f _{INPLL}	PLL input frequency range		2		25	MHz
foutpll	PLL output frequency range		3.5		100	MHz
	PLL setting time			0.5		ms
	DLL paried litter	RMS		30		ps
	PLL period jitter	Peak to peak		±150		ps



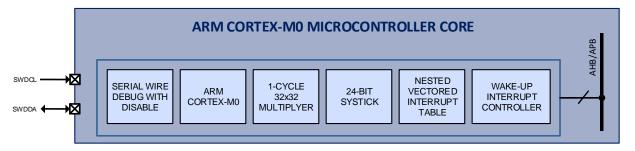
16 ARM CORTEX®-M0 MCU CORE

16.1 Features

- Arm[®] Cortex[®]-M0 core
- Fast single-cycle 32-bit x 32-bit multiplier
- 24-bit SysTick timer
- Up to 50MHz operation
- Serial wire debug (SWD), with 4 break-point and 2 watch-point unit comparators
- Nested vectored interrupt controller (NVIC) with 25 external interrupts
- Wake-up interrupt controller (WIC) with GPIO, real-time clock (RTC) and watchdog timer (WDT) interrupts enabled
- Sleep and deep-sleep mode with clock gating

16.2 Block Diagram

Figure 16-1 Arm® Cortex®-M0 Microcontroller Core



16.3 Functional Description

The Arm® Cortex®-M0 microcontroller core is configured for little endian operation and includes the fast single-cycle 32-bit multiplier and 24-bit SysTick timer and can operate at a frequency of up to 50MHz.

The microcontroller nested vectored interrupt controller (NVIC) supports 25 external interrupts for the device's peripherals and sub-systems. For low-latency interrupt processing, the NVIC also supports interrupt tail-chaining. The wake-up interrupt controller (WIC) is able to wake up the device from low-power modes using any GPIO interrupt, as well as from the RTC or WDT. The Arm® Cortex®-M0 supports both sleep and deep-sleep low-power modes. The deep-sleep mode supports clock gating to limit standby power even further.

Firmware debug support includes 4 break-point and 2 watch-point unit comparators using the serial wire debug (SWD) protocol. The serial wire debug mechanism can be disabled to prevent device access to the firmware in the field.

For more information on the detailed operation of the Microcontroller Core in the PAC5232, see the PAC5232 User Guide.



16.4 Electrical Characteristics

Table 16-1 MCU and Clock Control System Electrical Characteristics

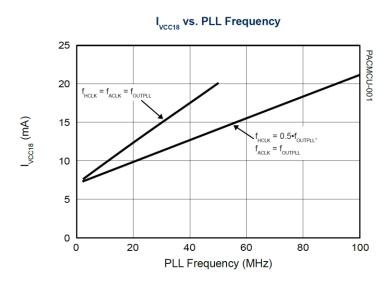
 $(T_A = -40^{\circ}C \text{ to } 105^{\circ}C \text{ unless otherwise specified.})$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
fhclk	Microcontroller Clock	HCLK			50	MHz
		f _{FRCLK} = f _{HCLK} = f _{ACLK} = ROSC 11b, PLL disabled, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		3.4		
		f _{FRCLK} = f _{HCLK} = f _{ACLK} = ROSC 10b, PLL disabled, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		4		
		f _{FRCLK} = f _{HCLK} = f _{ACLK} = ROSC 01b, PLL disabled, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		5.3		
lop;vsys	V _{SYS} operating supply current	f _{FRCLK} = f _{HCLK} = f _{ACLK} = ROSC 00b, PLL disabled, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		9		mA
		f _{FRCLK} = f _{HCLK} = f _{ACLK} = CLKREF, PLL disabled, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		2.3		
		f _{FRCLK} = f _{HCLK} = f _{ACLK} = 10MHz XTAL, PLL disabled, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		4.5		
		f _{FRCLK} = 4MHz CLKREF, f _{HCLK} = 50MHz, f _{ACLK} = f ₀ UTPLL = 100MHz, CPU halt; other clock sources, ADC, timers, and serial interfaces disabled		23.3		

16.5 Typical Performance Characteristics

(Vsys = Vccio = 5V, T_A = -40°C to 105°C unless otherwise specified.)

Figure 16-2 MCU Performance Characteristics





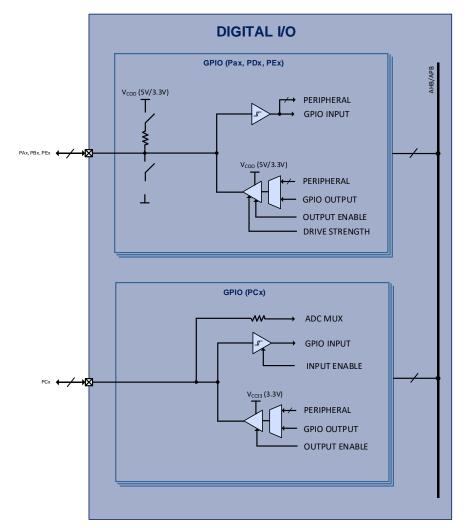
17 IO CONTROLLER

17.1 Features

- 5V-compliant I/O PAx, PDx, PEx
- 3.3V-compliant I/O PCx
- Configurable drive strength on PAx, PDx, PEx
- Configurable pull-up or pull-down on PAx, PDx, PEx

17.2 Block Diagram

Figure 17-1 IO Controller Block Diagram





17.3 Functional Description

The PAC® can support up to 4 ports with 8 I/Os each from PAx, PCx, PDx, and PEx, in addition to the I/Os on the analog front end. All PAx, PCx, PDx, and PEx ports have interrupt capability with configurable interrupt edge.

PAx, PDx, and PEx I/Os use Vccio as the I/O supply voltage that is set to 3.3V. To drive these IOs t 5V, the user may short VCCIO and VSYS on the PCB, then 5V will be the supply for these IOs.

The drive current can be configured as 8mA or 16mA. They also support weak pull-up and pull-down to save external components.

PCx uses V_{CC33} as its I/O supply voltage. The drive current is fixed to 8mA. PC0 to PC5 are also associated with analog inputs AD0 to AD5 to the ADC.

17.4 GPIO Current Injection

Under normal operation, there should not be current injected into the GPIOs on the device due to the GPIO voltage below ground or above the GPIO supply⁶. Current will be injected into the GPIO when the GPIO pin voltage is less than -0.3V or when greater than GPIO supply + 0.3V.

In order provide a robust solution when this situation occurs, the PAC52XX family of products allows a small amount of injected current into the GPIO pins, to avoid excessive leakage or device damage.

For information on the GPIO current injection thresholds, see the absolute maximum parameters for this device.

Sustained operation with the GPIO pin voltage greater than the GPIO supply or when the GPIO pin voltage is less than -0.3V may result in reduced lifetime of the device. GPIO current injection should only be a temporary condition.

_

⁶ VCC33 is the supply for any PC GPIO pin and VCCIO is the supply for any other GPIO pins.



17.5 Electrical Characteristics

Table 17-1 IO Controller Electrical Characteristics

 $(V_{SYS} = V_{CCIO} = 5V$, and $T_A = -40^{\circ}C$ to $105^{\circ}C$ unless otherwise specified.)

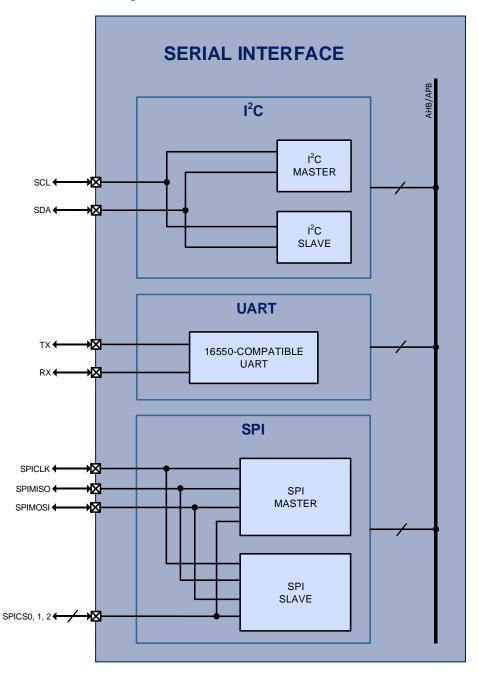
SYMBOL	PARAMETER	COND	CONDITIONS		TYP	MAX	UNIT
PAx, PDx, Pl	Ex (5V Operation)						
V _{IH}	High-level input voltage			3			V
VIL	Low-level input voltage					0.8	V
loL	Low-level output sink current	Vol. = 0.4V	Drive Strength = 0b	7			mA
IOL	(Limited by I _{VSYS})	VOL = 0.4V	Drive Strength = 1b	15			
I	High-level output source current	V _{OH} = 2.4V	Drive Strength = 0b			-7	mΛ
Іон	(Limited by I _{VSYS})	VOH = 2.4V	Drive Strength = 1b			-15	mA 5
R _{PU}	Weak pull-up resistance	V _{CCIO} = 3.3V, pull-up enabl	V _{CCIO} = 3.3V, pull-up enabled		74	104	kΩ
R_{PD}	Weak pull-down resistance		V _{CCIO} = 3.3V, pull-down enabled		84	121	kΩ
I _{IL}	Input leakage current			-10		10	μА
PCx (3.3V O	peration)						
VIH	High-level input voltage	V _{CC33} = 3.3V		2			V
VIL	Low-level input voltage	Vcc33 = 3.3V				0.8	V
loL	Low-level output sink current	V _{CC33} = 3.3V, V _{OL} = 0.4V		7			mA
I _{OH}	High-level output source current	$V_{CC33} = 3.3V, V_{OH} = 2.4V$				-7	mA
I _{IL}	Input leakage current	T _A = 125°C		-10	0	10	μΑ



18 SERIAL INTERFACE

18.1 Block Diagram

Figure 18-1 Serial Interface Block Diagram







18.2 Functional Description

The PAC5232 has up to three serial interfaces: I2C, UART and SPI.

18.3 I²C Controller

The I₂C controller is a configurable peripheral that can support various modes of operation:

- I²C master operation
 - Normal mode (100kHz), fast mode (400kHz), or fast mode plus (1MHz)
 - Single and multi-master
 - Synchronization (multi-master)
 - Arbitration (multi-master)
 - o 7-bit or 10-bit slave addressing
- I²C slave operation
 - o Normal mode (100kHz), fast mode (400kHz), or fast mode plus (1MHz)
 - Clock stretching
 - o 7-bit or 10-bit slave addressing

The I₂C peripheral may operate either by polling, or can be configured to be interrupt driven for both receive and transmit data.

18.4 UART Controller

The UART peripheral is a configurable peripheral that can support various features and modes of operation:

- Programmable clock selection
- National Instruments PC16550D compatible
- 16-deep transmit and receive FIFO and fractional clock divisor
- Up to 3.125Mbps communication speed (with HCLK = 50MHz)

The UART peripheral may operate either by polling, or can be configured to be interrupt driven for both receive and transmit data.

18.5 SPI Controller

The device contains an SPI controller that can each be used in either master or slave operation, with the following features:

- SPI master operation
 - Control of up to three different SPI slaves
 - Operation up to 25MHz
 - Flexible multiple transmit mode for variable-size SPI data with user-defined chip-select behavior
 - Chip select "shaping" through programmable additional delay for chip-select setup, hold and wait time
 - o for back-to-back transfers



- SPI master or slave operation
 - Supports clock phase and polarity control
 - o Data transmission/reception can be on 8-, 16-, 24- or 32-bit boundary
 - Selectable data bit ordering (LSB or MSB first)
 - o Programmable chip select polarity
 - Selectable "auto-retransmit" mode

The SPI peripheral may operate either by polling, or can be configured to be interrupt driven for both receive and transmit data.

18.6 Dynamic Characteristics

Table 18-1 Serial Interface Dynamic Characteristics

(\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	nd T 10°C to	105°C unloce	otherwise specified.)
VSYS = VCCIO = JV.a	11u 1 A = -40 C 10	100 0 0111600	Offici Mise specified.)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
I ² C						
		Standard mode (100kHz)	2.8			MHz
f _{I2CCLK}	I ² C input clock frequency	Full-speed mode (400kHz)	2.8			MHz
		Fast mode (1MHz)	6.14			MHz
UART						
f _{UARTCLK}	USART input clock frequency				f _{HCLK} /16	MHz
	UART baud rate	f _{HCLK} = 50MHz			3.125	Mbps
SPI						
f _{SPICLK} USART input clock frequency	LICART in a charle for any and	Master mode			f _{HCLK} /2	MHz
	OSAINT IIIPUL GIOCK TEQUETICS	Slave mode			f _{HCLK} /2	MHz



Table 18-2 I²C Dynamic Characteristics

 $(V_{SYS} = V_{CCIO} = 5V, \text{ and } T_A = -40^{\circ}\text{C} \text{ to } 105^{\circ}\text{C} \text{ unless otherwise specified.})$

SYMBOL	and T _A = -40°C to 105°C unless otherwise sp PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
		Standard mode	0		100	kHz
f _{SCL}	SCL clock frequency	Full-speed mode	0		400	kHz
		Fast mode	0		1000	kHz
		Standard mode	4.7			μs
t _{LOW}	SCL clock low	Full-speed mode	1.3			μs
		Fast mode	0.5			μs
		Standard mode	4.0			μs
t _{HIGH}	SCL clock high	Full-speed mode	0.6			μs
		Fast mode	0.26			μs
		Standard mode	4.0			μs
t _{HD;STA}	Hold time for a repeated START condition	Full-speed mode	0.6			μs
		Fast mode	0.26			μs
		Standard mode	4.7			μs
t _{SU;STA}	Set-up time for a repeated START condition	Full-speed mode	0.6			μs
		Fast mode	0.26			μs
		Standard mode	0		3.45	μs
t _{HD;DAT}	Data hold time	Full-speed mode	0		0.9	μs
		Fast mode	0			μs
		Standard mode	250			ns
t _{SU;DAT}	Data setup time	Full-speed mode	100			ns
		Fast mode	50			ns
		Standard mode	4.0			μs
t _{su;sto}	Set-up time for STOP condition	Full-speed mode	0.6			μs
		Fast mode	0.26			μs
		Standard mode	4.7			μs
t _{BUF}	Bus free time between a STOP and START condition	Full-speed mode	1.3			μs
	and on an osmanion	Fast mode	0.5			μs
		Standard mode			1000	ns
t _r	Rise time for SDA and SCL	Full-speed mode	20		300	ns
		Fast mode			120	ns
		Standard mode			300	ns
t _f	Fall time for SDA and SCL	Full-speed mode			300	ns
		Fast mode			120	ns
Сь	Capacitive load for each bus line	Standard mode, full-speed mode			400	pF
υ	Capacitive load for odoll but lillo	Fast mode			550	pF



Figure 18-2 I²C Timing Diagram

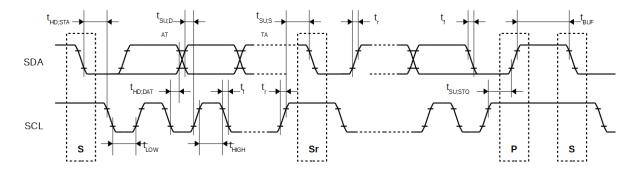


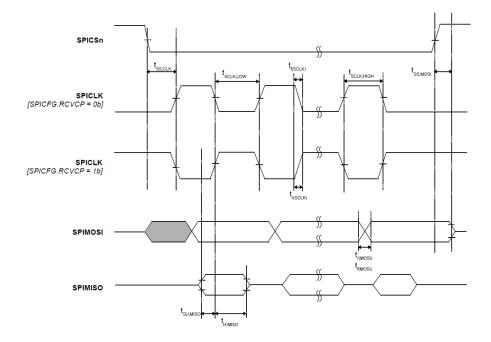


Table 18-3 SPI Dynamic Characteristics

($V_{SYS} = V_{CCIO} = 5V$, and $T_A = -40^{\circ}C$ to $105^{\circ}C$ unless otherwise specified.)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT
tsclk;HIGH	SPICLK Input High Time	SPICLK = 25MHz	30			ns
t _{SCLK;LOW}	SPICLK Input low Time		30			ns
tss;sclk	SPICSn to SPICLK Time		120			ns
t _{SS;MOSI}	SPICSn to SPIMISO High-impedance time		10		50	ns
$t_{r(SCLK)}$	SPICLK Rise Time			10	25	ns
t _{f(SCLK)}	SPICLK Fall Time			10	25	ns
$t_{r(MOSI)}$	SPIMOSI Rise Time			10	25	ns
t _{f(MOSI)}	SPIMOSI Fall Time			10	25	ns
t _{r(MISO)}	SPIMISO Rise Time			10	25	ns
t _{f(MISO)}	SPIMISO Fall Time			10	25	ns
t _{SU;MISO}	SPIMISO Setup Time		20			ns
t _{H;MISO}	SPIMISO Hold Time		20			ns

Figure 18-3 SPI Timing Diagram





19 TIMERS

19.1 Block Diagram

Figure 19-1 PWM Timers Block Diagram

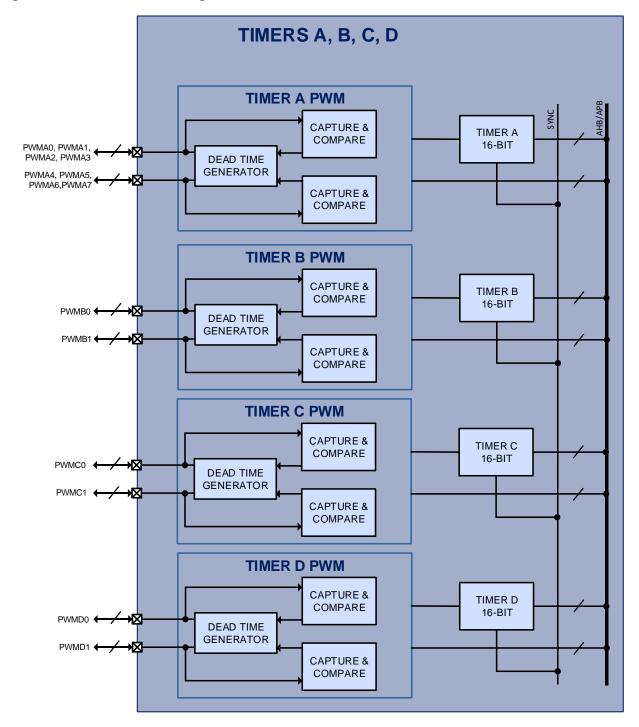




Figure 19-2 SOC Bus Watchdog and Wake-Up Timer

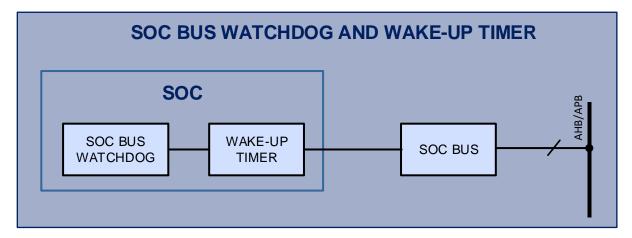
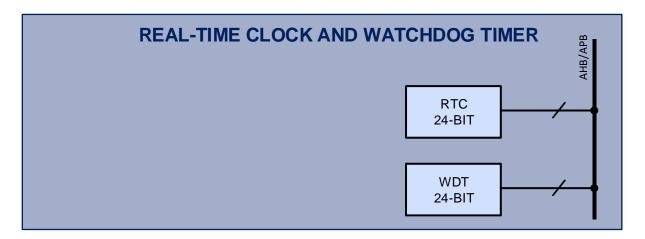


Figure 19-3 Real-Time Clock and Watchdog Timer



19.2 Functional Description

The device includes 9 timers: timer A, timer B, timer C, timer D, watchdog timer 1 (WDT), watchdog timer 2, wake-up timer, real-time clock (RTC), and SysTick timer. The device supports up to 14 different PWM signals and has up to 7 dead-time controllers. Timers A, B, C and D can be concatenated to synchronize to a single clock and start/stop signal for applications that require a synchronized timer period between timers.

19.2.1 Timer A

Timer A is a general purpose 16-bit timer with 8 PWM/capture and compare units. It has 4 pairs of PWM signals going into 4 dead-time controllers. Timer A can be concatenated with timers B, C, and D to synchronize the PWM/capture and compare units. It can use either ACLK or HCLK as clock input with an additional clock divider from /1 to /128.



19.2.2 Timer B

Timer B is a general purpose 16-bit timer with 2 PWM/capture and compare units. It has one pair of PWM signals going into one dead-time controller, as well as 2 additional compare units that can be used for additional system time bases for interrupts. Timer B can be concatenated with timers A, C, and D to synchronize the PWM/capture and compare units. It can use either ACLK or HCLK as clock input with an additional clock divider from /1 to /128.

19.2.3 Timer C

Timer C is a general purpose 16-bit timer with 2 PWM/capture and compare units. It has one pair of PWM signals going into one dead-time controller. Timer C can be concatenated with timers A, B, and D to synchronize the PWM/capture and compare units. It can use either ACLK or HCLK as clock input with an additional clock divider from /1 to /128.

19.2.4 Timer D

Timer D is a general purpose 16-bit timer with 2 PWM/capture and compare units. It has one pair of PWM signals going into one dead-time controller. Timer D can be concatenated with timers A, B, and C to synchronize the PWM/capture and compare units. It can use either ACLK or HCLK as clock input with an additional clock divider from /1 to /128.

19.2.5 Watchdog Timer

The 24-bit watchdog timer (WDT) can be used for long time period measurements or periodic wake up from sleep mode.

The watchdog timer can be used as a system watchdog, or as an interval timer, or both. The watchdog timer can use either FRCLK or FCLK as clock input with an additional clock divider from /2 to /65536.

19.2.6 SOC Bus Watchdog Timer

The watchdog timer 2 is used to monitor internal SOC Bus communication. It will trigger device reset if there is no SOC Bus communication to the AFE for 4s or 8s.

19.2.7 Wake-Up Timer

The wake-up timer can be used for very low power hibernate and sleep modes to wake up the micro controller periodically.

It can be configured to be 125ms, 250ms, 500ms, 1s, 2s, 4, or 8s.

19.2.8 Real-Time Clock

The 24-bit real-time clock (RTC) can be used for time measurements when an accurate clock source is used. This timer can also be used for periodic wake up from sleep mode. The RTC uses FRCLK as clock input with an additional clock divider from /2 to /65536.



20 THERMAL CHARACTERISTICS

Table 20-1 Thermal Characteristics

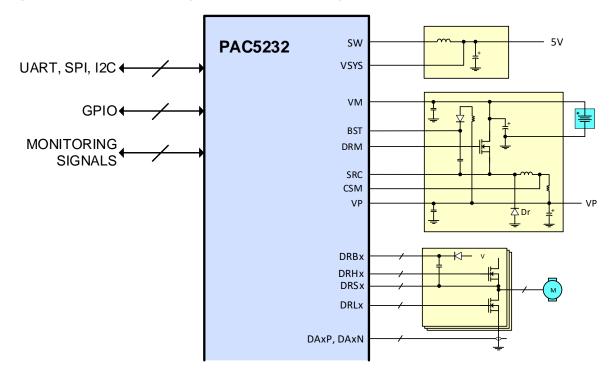
PARAMETER	VALUE	UNIT
Operating ambient temperature range	-40 to 105	°C
Operating junction temperature range	-40 to 125	°C
Storage temperature range	-55 to 150	°C
Lead temperature (Soldering, 10 seconds)	300	°C
Junction-to-case thermal resistance (⊕JC)	2.897	°C/W
Junction-to-ambient thermal resistance (Θ_{JA})	23.36	°C/W



21 APPLICATION EXAMPLES

The following simplified diagram shows an example of a single-motor, low-voltage application using the PAC5232 device.

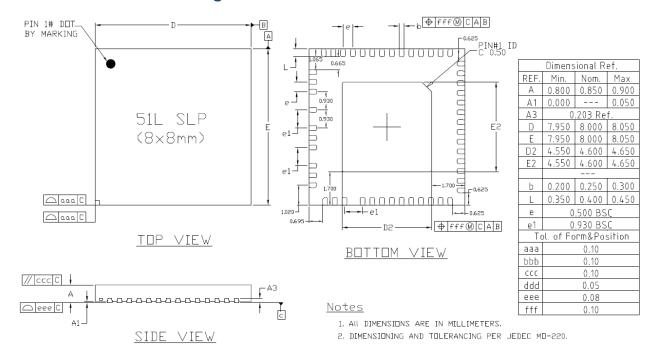
Figure 21-1 3-Phase Motor Using PAC5232 (Simplified Diagram)





22 PACKAGE OUTLINE AND DIMENSIONS

22.1 51L 8x8 QFN Package Outline and Dimensions







Contact Information

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FSB50250AB FNA41060 MSVBTC50E MSVCPM3-54-12 MSVCPM3-63-12 MSVCPM4-63-12 FSB50550AB L99MC6TR

LC898301XA-MH LV8413GP-TE-L-E MSVGW45-14-3 MSVGW54-14-4 IRSM515-025DA4 LB11851FA-BH STIPN2M50-H

BH67172NUX-GE2 DMPUC-R2 AIS-D-42MA FNB81560T3 NCV7535DBR2G NCP81274MNTXG LB11685AV-W-AH

IM240S6Y1BAKMA1 FSB50450BL FSB50450BSL FSB50550BSL IKCM15L60GD DA7280-00FV2 IM240M6Z1BALMA1 FSB50550BL

MP9517GJS-P TLE9877QXA40XUMA3 BA6868FM-E2 HT8835ARDZ IGCM04F60GA