# RENESAS

# 16-Channel High-Density T1/E1/J1 Line Interface Unit IDT82P2916

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# Interface Unit

## FEATURES

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- Integrates 16 channels T1/E1/J1 short haul line interface units for 100 Ω T1, 120 Ω E1, 110 Ω J1 twisted pair cable and 75 Ω E1 coaxial cable applications
- Per-channel configurable Line Interface options
  - · Supports various line interface options
    - Differential and Single Ended line interfaces
    - true Single Ended termination on primary and secondary side of transformer for E1 75 Ω coaxial cable applications
  - transformer-less for Differential interfaces
  - Fully integrated and software selectable receive and transmit termination
    - Option 1: Fully Internal Impedance Matching with integrated receive termination resistor
    - Option 2: Partially Internal Impedance Matching with common external resistor for improved device power dissipation
    - Option 3: External impedance Matching termination
  - Supports global configuration and per-channel configuration to T1, E1 or J1 mode

#### • Per-channel programmable features

- Provides T1/E1/J1 short haul waveform templates and userprogrammable arbitrary waveform templates
- Provides two JAs (Jitter Attenuator) for each channel of receiver and transmitter
- Supports AMI/B8ZS (for T1/J1) and AMI/HDB3 (for E1) encoding and decoding

#### Per-channel System Interface options

- Supports Single Rail, Dual Rail with clock or without clock and sliced system interface
- Integrated Clock Recovery for the transmit interface to recover transmit clock from system transmit data

#### Per-channel system and diagnostic functions

- Provides transmit driver over-current detection and protection with optional automatic high impedance of transmit interface
- Detects and generates PRBS (Pseudo Random Bit Sequence), ARB (Arbitrary Pattern) and IB (Inband Loopback) in either receive or transmit direction
- Provides defect and alarm detection in both receive and transmit directions.
  - Defects include BPV (Bipolar Violation) /CV (Code Violation) and EXZ (Excessive Zeroes)
  - Alarms include LLOS (Line LOS), SLOS (System LOS), TLOS (Transmit LOS) and AIS (Alarm Indication Signal)
- Programmable LLOS detection /clear levels. Compliant with ITU and ANSI specifications
- · Various pattern, defect and alarm reporting options

- Serial hardware LLOS reporting (LLOS, LLOS0) for all 16 channels
- Configurable per-channel hardware reporting with RMF/TMF (Receive / Transmit Multiplex Function)
- Register access to individual registers or 16-bit error counters
- Supports Analog Loopback, Digital Loopback and Remote Loopback
- Supports T1.102 line monitor
- Hitless Protection Switching (HPS) without external Relays
  - Supports 1+1 and 1:1 hitless protection switching
  - Asynchronous hardware control (OE, RIM) for fast global high impedance of receiver and transmitter (hot switching between working and backup board)
  - High impedance transmitter and receiver while powered down
  - Per-channel register control for high impedance, independent for receiver and transmitter

#### Clock Inputs and Outputs

- + Flexible master clock (N x 1.544 MHz or N x 2.048 MHz) (1  $\leq$  N  $\leq$  8, N is an integer number)
- Two selectable reference clock outputs
  - from the recovered clock of any of the 16 channels
  - from external clock input
  - from device master clock
- Integrated clock synthesizer can multiply or divide the reference clock to a wide range of frequencies: 8 KHz, 64 KHz, 2.048 MHz, 4.096 MHz, 8.192 MHz, 19.44 MHz and 32.768 MHz
- Cascading is provided to select a single reference clock from multiple devices without the need for any external logic

#### Microprocessor Interface

- · Supports Serial microprocessor interface
- Other Key Features
  - IEEE1149.1 JTAG boundary scan
  - Two general purpose I/O pins
  - 3.3 V I/O with 5 V tolerant inputs
  - 3.3 V and 1.8 V power supply
  - Package: 484-pin Fine Pitch BGA (19 mm X 19 mm)

#### Applicable Standards

- AT&T Pub 62411 Accunet T1.5 Service
- ANSI T1.102, T1.403 and T1.231
- · Bellcore TR-TSY-000009, GR-253-CORE and GR-499-CORE
- ETSI CTR12/13
- ETS 300166 and ETS 300 233
- G.703, G.735, G.736, G.742, G.772, G.775, G.783 and G.823
- 0.161
- ITU I.431 and ITU O.171

# **APPLICATIONS**

- SDH/SONET multiplexers
- Central office or PBX (Private Branch Exchange)
- Digital access cross connects
- Remote wireless modules
- Microwave transmission systems

### DESCRIPTION

The 82P2916 is a 16 channels high-density T1/E1/J1 short haul Line Interface Unit. Each channel of the 82P2916 can be independently configured. The configuration is performed through a Serial microprocessor interface.

In the receive path, through a Single Ended or Differential line interface, the received signal is processed by an adaptive Equalizer and then sent to a Slicer. Clock and data are recovered from the digital pulses output from the Slicer. After passing through an enabled or disabled Receive Jitter Attenuator, the recovered data is decoded using B8ZS/ AMI/HDB3 line code rule in Single Rail NRZ Format mode and output to the system, or output to the system without decoding in Dual Rail NRZ Format mode and Dual Rail RZ Format mode. In the transmit path, the data to be transmitted is input on TDn in Single Rail NRZ Format mode or TDPn/TDNn in Dual Rail NRZ Format mode and Dual Rail RZ Format mode, and is sampled by a transmit reference clock. The clock can be supplied externally from TCLKn or recovered from the input transmit data by an internal Clock Recovery. A selectable JA in Tx path is used to de-jitter gapped clocks. To meet T1/ E1/J1 waveform standards, five preset T1 templates, two E1 templates and one J1 template, as well as an arbitrary waveform generator are provided. The data through the Waveform Shaper, the Line Driver and the Tx Transmitter is output on TTIPn and TRINGn.

Alarms (including LOS, AIS) and defects (including BPV, EXZ) are detected in both receive line side and transmit system side. AIS alarm, PRBS, ARB and IB patterns can be generated /detected in receive / transmit direction for testing purpose. Analog Loopback, Digital Loopback and Remote Loopback are all integrated for diagnostics.

A line monitor function per T1.102 is available to provide a Non-Intrusive Monitoring of channels of other devices.

JTAG per IEEE 1149.1 is also supported by the 82P2916.

### **BLOCK DIAGRAM**

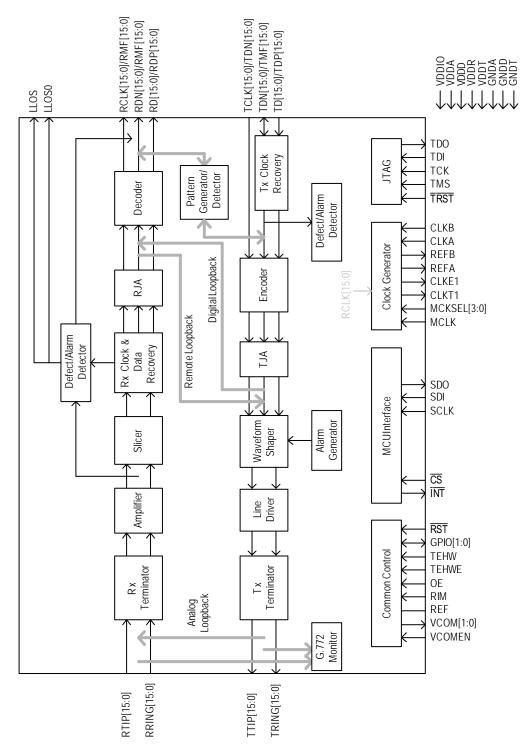


Figure-1 Functional Block Diagram

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### 1 PIN ASSIGNMENT

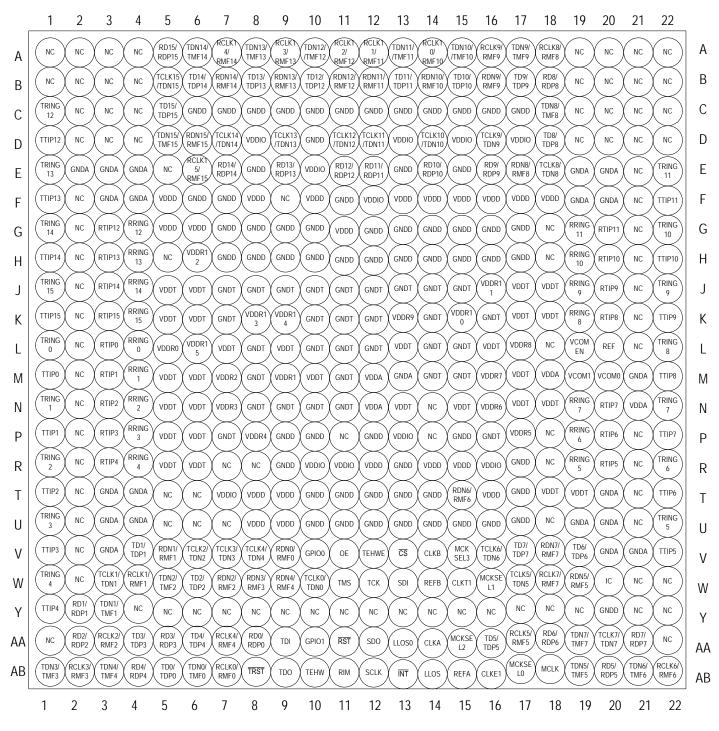


Figure-2 484-Pin FPBGA 0.8 mm pitch (Top View)

# 2 PIN DESCRIPTION

Name	1/0	Pin No. <sup>1</sup>	Description
		Line Interface	
RTIPn	Input	L3, M3, N3, P3, R3, R20, P20, N20, K20, J20, H20, G20, G3, H3, J3, K3	RTIPn / RRINGn: Receive Bipolar Tip/Ring for Channel 0 ~ 15 The receive line interface supports both Receive Differential mode and Receive Single Ender
RRINGn		L4, M4, N4, P4, R4, R19, P19, N19,	mode. In Receive Differential mode, the received signal is coupled into RTIPn and RRINGn via a 1:
(n=0~15)		K19, J19, H19, G19, G4, H4, J4, K4	In Receive Single Ended mode, RRINGn should be left open. The received signal is input of RTIPn via a 2:1 (step down) transformer or without a transformer (transformer-less). These pins will become High-Z globally or channel specific in the following conditions: • Global High-Z: • Connecting the RIM pin to low; • Loss of MCLK • During and after power-on reset, hardware reset or global software reset; • Per-channel High-Z
	Outrast		Receiver power down by writing '1' to the R_OFF bit (b5, RCF0,)
TTIPn	Output	M1, P1, T1, V1, Y1, V22, T22, P22, M22, K22, H22, F22, D1, F1, H1, K1	TTIPn / TRINGn: Transmit Bipolar Tip /Ring for Channel 0 ~ 15 The transmit line interface supports both Transmit Differential mode and Transmit Single
TRINGn		L1, N1, R1, U1, W1, U22, R22, N22,	Ended mode. In Transmit Differential mode, TTIPn outputs a positive differential pulse while TRINGn ou
(n=0~15)		L22, J22, G22, E22, C1, E1, G1, J1	<ul> <li>puts a negative differential pulse. The pulses are coupled to the line side via a 1:2 (step up transformer or without a transformer (transformer-less).</li> <li>In Transmit Single Ended mode, TRINGn should be left open (it is shorted to ground internally). The signal presented at TTIPn is output to the line side via a 1:2 (step up) transformer.</li> <li>These pins will become High-Z globally or channel specific in the following conditions:</li> <li>Global High-Z:</li> <li>Connecting the OE pin to low;</li> <li>Loss of MCLK;</li> <li>During and after power-on reset, hardware reset or global software reset;</li> <li>Per-channel High-Z</li> </ul>
			<ul> <li>Writing '0' to the OE bit (b6, TCF0,)<sup>2</sup>;</li> <li>Loss of TCLKn in Transmit Single Rail NRZ Format mode or Transmit Dual Rail NR Format mode, except that the channel is in Remote Loopback or transmit internal pattern with XCLK <sup>3</sup>;</li> <li>Transmitter power down by writing '1' to the T_OFF bit (b5, TCF0,);</li> <li>Per-channel software reset;</li> <li>The THZ_OC bit (b4, TCF0,) is set to '1' and the transmit driver over-current detected.</li> <li>Refer to Section 3.3.8 Output High-Z on TTIP and TRING for details.</li> </ul>

Note:

1. The pin number of the pins with the footnote 'n' is listed in order of channel (CH0  $\sim$  CH15).

2. The content in the brackets indicates the position and the register name of the preceding bit. After the register name, if the punctuation ',...' is followed, this bit is in a per-channel register. The addresses and details are included in Chapter 5 Programming Information.

3. XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

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Name	I/O	Pin No.	Description				
	System Interface						
RDn / RDPn (n=0~15)	Output	AA8, Y2, AA2, AA5, AB4, AB20, AA18, AA21, B18, E16, E14, E12, E11, E9, E7, A5	RDn: Receive Data for Channel 0 ~ 15 When the receive system interface is configured to Single Rail NRZ Format mode, this multi- plex pin is used as RDn. The decoded NRZ data is updated on the active edge of RCLKn. The active level on RDn is selected by the RD_INV bit (b3, RCF1,). When the receiver is powered down, RDn will be in High-Z state or low, as selected by the RHZ bit (b6, RCF0,).				
			<ul> <li>RDPn: Positive Receive Data for Channel 0 ~ 15</li> <li>When the receive system interface is configured to Dual Rail NRZ Format mode, Dual Rail RZ Format mode or Dual Rail Sliced mode, this multiplex pin is used as RDPn.</li> <li>In Receive Dual Rail NRZ Format mode, the un-decoded NRZ data is output on RDPn and RDNn and updated on the active edge of RCLKn.</li> <li>In Receive Dual Rail RZ Format mode, the un-decoded RZ data is output on RDPn and RDNn and updated on the active edge of RCLKn.</li> <li>In Receive Dual Rail Sliced mode, the raw RZ sliced data is output on RDPn and RDNn.</li> <li>For Receive Differential line interface, an active level on RDPn indicates the receipt of a positive pulse on RTIPn and a negative pulse on RRINGn; while an active level on RDNn indicates the receipt of a negative pulse on RTIPn and a positive pulse on RRINGn.</li> <li>For Receive Single Ended line interface, an active level on RDPn indicates the receipt of a positive pulse on RTIPn; while an active level on RDPn indicates the receipt of a negative pulse on RTIPn.</li> <li>The active level on RDPn and RDNn is selected by the RD_INV bit (b3, RCF1,).</li> <li>When the receiver is powered down, RDPn and RDNn will be in High-Z state or low, as selected by the RHZ bit (b6, RCF0,).</li> </ul>				
RDNn / RMFn (n=0~15)	Output	V9, V5, W7, W8, W9, W19, T15, V18, E17, B16, B14, B12, B11, B9, B7, D6	<ul> <li>RDNn: Negative Receive Data for Channel 0 ~ 15</li> <li>When the receive system interface is configured to Dual Rail NRZ Format mode, Dual Rail RZ Format mode or Dual Rail Sliced mode, this multiplex pin is used as RDNn. (Refer to the description of RDPn for details).</li> <li>RMFn: Receive Multiplex Function for Channel 0 ~ 15</li> <li>When the receive system interface is configured to Single Rail NRZ Format mode, this multiplex pin is used as RMFn.</li> <li>RMFn is configured by the RMF_DEF[2:0] bits (b7~5, RCF1,) and can indicate PRBS/ARB, LAIS, LEXZ, LBPV, LEXZ+LBPV, LLOS, output recovered clock (RCLK) or XOR output of positive and negative sliced data. Refer to Section 3.5.7.1 RMFn Indication for details. The output on RMFn is updated on the active edge of RCLKn. The active level of RMFn is always high.</li> <li>When the receiver is powered down, RMFn will be in High-Z state or low, as selected by the RHZ bit (b6, RCF0,).</li> </ul>				

Name	1/0	Pin No.	Description
RCLKn / RMFn (n=0~15)	Output	AB7, W4, AA3, AB2, AA7, AA17, AB22, W18, A18, A16, A14, A12, A11, A9, A7, E6	<ul> <li>RCLKn: Receive Clock for Channel 0 ~ 15</li> <li>When the receive system interface is configured to Single Rail NRZ Format mode, Dual Rail NRZ Format mode or Dual Rail RZ Format mode, this multiplex pin is used as RCLKn.</li> <li>RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock which is recovered from the received signal.</li> <li>The data output on RDPn/RDNn (in Receive Dual Rail NRZ Format mode, Receive Dual Rail RZ Format mode and Receive Dual Rail Sliced) is updated on the active edge of RCLKn. The active edge is selected by the RCK_ES bit (b4, RCF1,).</li> <li>In LLOS condition, RCLKn output high or XCLK, as selected by the RCKH bit (b7, RCF0,) (refer to Section 3.5.3.1 Line LOS (LLOS) for details).</li> <li>When the receiver is powered down, RCLKn will be in High-Z state or low, as selected by the RHZ bit (b6, RCF0,).</li> <li>RMFn: Receive Multiplex Function for Channel 0 ~ 15</li> <li>When the receive system interface is configured to Dual Rail Sliced mode, this multiplex pin is used as RMFn.</li> <li>(Refer to the description of RMFn of the RDNn/RMFn multiplex pin for details).</li> </ul>
LLOS	Output	AB14	LLOS: Receive Line Loss Of Signal LLOS synchronizes with the output of CLKE1 and can indicate the LLOS (Line LOS) status of all 16 channels in a serial format. When the clock output on CLKE1 is enabled, LLOS indicates the LLOS status of the 16 chan- nels in a serial format and repeats every seventeen cycles. The start filler is positioned by LLOS0. Refer to the description of LLOS0 below for details. LLOS is updated on the rising edge of CLKE1 and is always active high. When the clock output of CLKE1 is disabled, LLOS will be held in High-Z state. (Refer to Section 3.5.3.1 Line LOS (LLOS) for details.)
LLOS0	Output	AA13	LLOS0: Receive Line Loss Of Signal for Start Position LLOS0 can indicate the start position on the LLOS pin. When the clock output on CLKE1 is enabled, LLOS0 pulses high for one CLKE1 clock cycle to indicate the start position on the LLOS pin. When CLKE1 outputs 8 KHz clock, LLOS0 pulses high for one 8 KHz clock cycle (125 µs) every seventeen 8 KHz clock cycles; when CLKE1 outputs 2.048 MHz clock, LLOS0 pulses high for one 2.048 MHz clock cycle (488 ns) every seventeen 2.048 MHz clock cycles. LLOS0 is updated on the rising edge of CLKE1. When the clock output on CLKE1 is disabled, LLOS0 will be held in High-Z state. (Refer to Section 3.5.3.1 Line LOS (LLOS) for details.)

Name	1/0	Pin No.			Description	
TDn / TDPn (n=0~15)	Input	AB5, V4, W6, AA4, AA6, AA16, V19, V17, D18, B17, B15, B13, B10, B8, B6, C5	<ul> <li>TDn: Transmit Data for Channel 0 ~ 15</li> <li>When the transmit system interface is configured to Single Rail NF plex pin is used as TDn.</li> <li>TDn accepts Single Rail NRZ data. The data is sampled into the d TCLKn.</li> <li>The active level on TDn is selected by the TD_INV bit (b3, TCF1,</li> <li>TDPn: Positive Transmit Data for Channel 0 ~ 15</li> <li>When the transmit system interface is configured to Dual Rail NRZ</li> <li>RZ Format mode, this multiplex pin is used as TDPn.</li> <li>In Transmit Dual Rail NRZ Format mode, the pre-encoded NRZ</li> <li>TDNn and sampled on the active edge of TCLKn.</li> <li>In Transmit Dual Rail RZ Format mode, the pre-encoded RZ data is The line code is as follows (when the TD_INV bit (b3, TCF1,) is '</li> </ul>		the device on the active edge of CF1,). II NRZ Format mode or Dual Rail NRZ data is input on TDPn and data is input on TDPn and TDNn.	
			TDPn	TDNn	Output Pulse on TTIPn	Output Pulse on TRINGn *
			0	0	Space	Space
			0	1	Negative Pulse	Positive Pulse
			1	0	Positive Pulse	Negative Pulse
			1	1	Space	Space
			Note: * For Transmit	Single Ended line i	nterface, TRINGn should be op	pen.
					Nn is selected by the TD_II	VV bit (b3, TCF1,).
TDNn / TMFn (n=0~15)	Input / Output	AB6, Y3, W5, AB1, AB3, AB19, AB21, AA19, C18, A17, A15, A13, A10, A8, A6, D5	When the transmiplex pin is used (Refer to the dest TMFn: Transmi When the transmi Rail RZ Format TMFn is configu SAIS, TOC, TLC tion for details. The output on T TMFn is always	mit system interfa as TDNn. scription of TDPn t Multiplex Func mit system interfa mode, this multipl red by the TMF_ DS, SEXZ, SBPV, MFn is updated o high.	for details). tion for Channel 0 ~ 15 ace is configured to Single lex pin is used as TMFn. DEF[2:0] bits (b7~5, TCF1, SEXZ+SBPV, SLOS. Refe in the active edge of TCLKr	ail NRZ Format mode, this multi- Rail NRZ Format mode or Dual ) and can indicate PRBS/ARB, r to Section 3.5.7.2 TMFn Indica- n (if available). The active level of
TCLKn / TDNn (n=0~15)	Input	W10, W3, V6, V7, V8, W17, V16, AA20, E18, D16, D14, D12, D11, D9, D7, B5	When the transu Rail NRZ Forma TCLKn inputs a The data input of mit Dual Rail NR TMFn (in Transu The active edge <b>TDNn: Negative</b> When the transu pin is used as T	It mode, this multi 1.544 MHz (in T1 on TDn (in Transm Z Format mode) mit Single Rail NI is selected by the e <b>Transmit Data</b> in nit system interface	ace is configured to Single plex pin is used as TCLKn. /J1 mode) or 2.048 MHz (ir nit Single Rail NRZ Format is sampled on the active ed RZ Format mode) is update e TCK_ES bit (b4, TCF1,) for Channel 0 ~ 15 ce is configured to Dual Ra	n E1 mode) clock. mode) or TDPn/TDNn (in Trans- lge of TCLKn. The data output on ed on the active edge of TCLKn.

Name	1/0	Pin No.		Description		
			Clock			
MCLK MCKSEL[0]	Input	AB18 AB17	<ul> <li>MCLK: Master Clock Input</li> <li>MCLK provides a stable reference timing for the IDT82P2916. MCLK should be a clock with +/-32 ppm (in T1/J1 mode) or +/-50 ppm (in E1 mode) accuracy. The clock frequency of MCLK is informed to the device by MCKSEL[3:0].</li> <li>If MCLK misses (duty cycle is less than 30% for 10 µs) and then recovers, the device will be reset automatically.</li> <li>MCKSEL[3:0]: Master Clock Selection</li> </ul>			
			These four pins inform the device of the			
MCKSEL[1]		W16	MCKSEL[3:0] <sup>*</sup>	Frequency (MHz)		
MCKSEL[2]		AA15	0000	1.544		
MCKSEL[3]		V15	0001	1.544 X 2		
			0010	1.544 X 3		
			0011	1.544 X 4		
			0100	1.544 X 5		
			0101	1.544 X 6		
			0110	1.544 X 7		
			0111	1.544 X 8		
			1000	2.048		
			1001	2.048 X 2		
			1010	2.048 X 3		
			1011	2.048 X 4		
			1100	2.048 X 5		
			1101	2.048 X 6		
			1110	2.048 X 7		
			1111	2.048 X 8		
			Note: 0: GNDD 1: VDDIO			
CLKT1	Output	W15	CLKG).			
CLKE1	Output	AB16	CLKE1: 8 KHz / E1 Clock Output         The output on CLKE1 can be enabled or disabled, as determined by the CLKE1_EN bit (b3, CLKG).         When the output is enabled, CLKE1 outputs an 8 KHz or 2.048 MHz clock, as selected by the CLKE1 bit (b2, CLKG). The output is locked to MCLK.         When the output is disabled, CLKE1 is in High-Z state.			

Name	1/0	Pin No.	Description
REFA	Output	AB15	REFA: Reference Clock Output A REFA can output three kinds of clocks: a recovered clock of one of the 16 channels, an exter- nal clock input on CLKA or a free running clock. The clock frequency is programmable. Refer to Section 3.6.2 Clock Outputs on REFA/REFB for details. The output on REFA can also be disabled, as determined by the REFA_EN bit (b6, REFA). When the output is disabled, REFA is in High-Z state.
REFB	Output	W14	REFB: Reference Clock Output B REFB can output a recovered clock of one of the 16 channels, an external clock input on CLKB or a free running clock. Refer to Section 3.6.2 Clock Outputs on REFA/REFB for details. The output on REFB can also be disabled, as determined by the REFB_EN bit (b6, REFB). When the output is disabled, REFB is in High-Z state.
CLKA	Input	AA14	CLKA: External T1/E1 Clock Input A External T1/J1 (1.544 MHz) or E1 (2.048 MHz) clock is input on this pin. The CKA_T1E1 bit (b5, REFA) should be set to match the clock frequency. When not used, this pin should be connected to GNDD.
CLKB	Input	V14	CLKB: External T1/E1 Clock Input B External T1/J1 (1.544 MHz) or E1 (2.048 MHz) clock is input on this pin. The CKB_T1E1 bit (b5, REFB) should be set to match the clock frequency. When not used, this pin should be connected to GNDD.
			Common Control
VCOM[0] VCOM[1]	Output	M20 M19	VCOM: Voltage Common Mode [1:0] These pins are used only when the receive line interface is in Receive Differential mode and connected without a transformer (transformer-less). To enable these pins, the VCOMEN pin must be connected high. Refer to Figure-6 for the connection. When these pins are not used, they should be left open.
VCOMEN	Input (Pull-Down)	L19	VCOMEN: Voltage Common Mode Enable This pin should be connected high only when the receive line interface is in Receive Differen- tial mode and connected without a transformer (transformer-less). When not used, this pin should be left open.
REF	-	L20	<b>REF:</b> Reference Resistor An external resistor (10 K $\Omega$ , ±1%) is used to connect this pin to ground to provide a standard reference current for internal circuit. This resistor is required to ensure correct device operation.
RIM	Input (Pull-Down)	AB11	RIM: Receive Impedance Matching In Receive Differential mode, when RIM is low, all 16 receivers become High-Z and only exter- nal impedance matching is supported. In this case, the per-channel impedance matching con- figuration bits - the R_TERM[2:0] bits (b2-0, RCF0,) and the R120IN bit (b4, RCF0,) - are ignored. In Receive Differential mode, when RIM is high, impedance matching is configured on a per- channel basis by the R_TERM[2:0] bits (b2-0, RCF0,) and the R120IN bit (b4, RCF0,). This pin can be used to control the receive impedance state for Hitless Protection applica- tions. Refer to Section 4.4 Hitless Protection Switching (HPS) Summary for details. In Receive Single Ended mode, this pin should be left open.

Name	1/0	Pin No.	Description
OE	Input	V11	OE: Output Enable OE enables or disables all Line Drivers globally. A high level on this pin enables all Line Drivers while a low level on this pin places all Line Drivers in High-Z state and independent from related register settings. Note that the functionality of the internal circuit is not affected by OE. If this pin is not used, it should be tied to VDDIO. This pin can be used to control the transmit impedance state for Hitless protection applica- tions. Refer to Section 4.4 Hitless Protection Switching (HPS) Summary for details.
TEHWE	Input (Pull-Up)	V12	TEHWE: Hardware T1/J1 or E1 Mode Selection Enable When this pin is open, the T1/J1 or E1 operation mode is selected by TEHW globally. When this pin is low, the T1/J1 or E1 operation mode is selected by the T1E1 bit (b0, CHCF,) on a per-channel basis.
TEHW	Input (Pull-Up)	AB10	TEHW: Hardware T1/J1 or E1 Mode Selection When TEHWE is open, this pin selects the T1/J1 or E1 operation mode globally: Low - E1 mode; Open - T1/J1 mode. When TEHWE is low, the input on this pin is ignored.
GPIO[0] GPIO[1]	Output / Input	V10 AA10	GPIO: General Purpose I/O [1:0] These two pins can be defined as input pins or output pins by the DIR[1:0] bits (b1~0, GPIO) respectively. When the pins are input, their polarities are indicated by the LEVEL[1:0] bits (b3~2, GPIO) respectively. When the pins are output, their polarities are controlled by the LEVEL[1:0] bits (b3~2, GPIO) respectively.
RST	Input	AA11	<b>RST</b> : Reset (Active Low) A low pulse on this pin resets the device. This hardware reset process completes in 2 µs max- imum. Refer to Section 4.1 Reset for an overview on reset options.
			MCU Interface
INT	Output	AB13	<b>INT:</b> Interrupt Request This pin indicates interrupt requests for all unmasked interrupt sources. The output characteristics (open drain or push-pull internally) and the active level are deter- mined by the INT_PIN[1:0] bits (b3-2, GCF).
ĊS	Input	V13	CS: Chip Select (Active Low)This pin must be asserted low to enable the microprocessor interface.A transition from high to low must occur on this pin for each Read/Write operation and CSshould remain low until the operation is over.
SCLK	Input	AB12	SCLK: Shift Clock In Serial microprocessor interface, this multiplex pin is used as SCLK. SCLK inputs the shift clock for the Serial microprocessor interface. Data on SDI is sampled by the device on the rising edge of SCLK. Data on SDO is updated on the falling edge of SCLK.
SDI	Input	W13	SDI: Serial Data Input In Serial microprocessor interface, this multiplex pin is used as SDI. Address and data on this pin are serially clocked into the device on the rising edge of SCLK.
SDO	Output	AA12	SDO: Serial Data Output In Serial microprocessor interface, this multiplex pin is used as SDO. Data on this pin is serially clocked out of the device on the falling edge of SCLK.

Name	I/0	Pin No.	Description	
JTAG (per IEEE 1149.1)				
TRST	Input Pull-Down	AB8	TRST: JTAG Test Reset (Active Low)A low signal on this pin resets the JTAG test port. To ensure deterministic operation of the testlogic, TMS should be held high when the signal on TRST changes from low to high.This pin may be left unconnected when JTAG is not used.This pin has an internal pull-down resistor.	
TMS	Input Pull-up	W11	TMS: JTAG Test Mode Select The signal on this pin controls the JTAG test performance and is sampled on the rising edge of TCK. To ensure deterministic operation of the test logic, TMS should be held high when the signal on TRST changes from low to high. This pin may be left unconnected when JTAG is not used. This pin has an internal pull-up resistor.	
ТСК	Input	W12	TCK: JTAG Test Clock The clock for the JTAG test is input on this pin. TDI and TMS are sampled on the rising edge of TCK and TDO is updated on the falling edge of TCK. When TCK is idle at low state, all stored-state devices contained in the test logic shall retain their state indefinitely. This pin should be connected to GNDD when JTAG is not used.	
TDI	Input Pull-up	AA9	TDI: JTAG Test Data Input The test data is input on this pin. It is clocked into the device on the rising edge of TCK. This pin has an internal pull-up resistor. This pin may be left unconnected when JTAG is not used.	
TDO	Output	AB9	TDO: JTAG Test Data Output The test data is output on this pin. It is clocked out of the device on the falling edge of TCK. TDO is a High-Z output signal except during the process of data scanning.	
			Power & Ground	
VDDIO		D8, D13, D15, D17, E10, F12, P13, R10, R11, R16, T7	VDDIO: 3.3 V I/O Power Supply	
VDDA		N21, M12, N12, M18	VDDA: 3.3 V Analog Core Power Supply	
VDDD		F5, F8, F10, F13, F14, F15, F16, F17, F18, G5, G6, G11, R12, R14, R15, T8, T9, T16, U8, U9	VDDD: 1.8 V Digital Core Power Supply	
VDDRn (N=0~15)		H6, J16, K8, K9, K13, K15, L5, L6, L17, M7, M9, M16, N7, N16, P8, P17	VDDRn: 3.3 V Power Supply for Receiver	
VDDT		J5, J6, J17, J18, K5, K6, K17, K18, L7, L9, L13, L16, M5, M6, M10, M17, N5, N6, N13, N15, N17, N18, P5, P6, R5, R6, T18, T19	VDDT: 3.3 V Power Supply for Transmitter Driver	
GNDA		E2, E3, E4, E19, E20, F3, F4, F19, F20, M13, M21, T3, T4, T20, U3, U4, U19, U20, V3, V20, V21	GNDA: GND for Analog Core / Receiver	

Name	1/0	Pin No.	Description
GNDD		C6, C7, C8, C9, C10, C11, C12, C13, C14, C15, C16, C17, D10, E8, E13, E15, F6, F7, F11, G7, G8, G9, G10, G12, G13, G14, G15, G16, G17, H7, H8, H9, H10, H11, H12, H13, H14, H15, H16, H17, P9, P10, P12, P15, R9, R13, R17, T10, T11, T12, T13, T14, T17, U10, U11, U12, U13, U14, U15, U16, U17, Y20	GNDD: Digital GND
GNDT		J7, J8, J9, J10, J11, J12, J13, J14, J15, K7, K10, K11, K12, K14, K16, L8, L10, L11, L12, L14, L15, M8, M11, M14, M15, N8, N9, N10, N11, P7, P16	GNDT: Analog GND for Transmitter Driver
			TEST
NC	-	A1, A2, A3, A4, A19, A20, A21, A22, B1, B2, B3, B4, B19, B20, B21, B22, C2, C3, C4, C19, C20, C21, C22, D2, D3, D4, D19, D20, D21, D22, E5, E21, F2, F9, F21, G2, G18, G21, H2, H5, H18, H21, J2, J21, K2, K21, L2, L18, L21, M2, N2, N14, P2, P11, P14, P18, P21, R2, R7, R8, R18, R21, T2, T5, T6, T21, U2, U5, U6, U7, U18, U21, V2, W2, W21, W22, Y4, Y5, Y6, Y7, Y8, Y9, Y10, Y11, Y12, Y13, Y14, Y15, Y16, Y17, Y18, Y19, Y21, Y22, AA1, AA22,	NC: No Connected These pins should be left open.
			Others
IC		W20	IC: Internal Connected This pin is for IDT use only and should be connected to GNDD.

# **3 FUNCTIONAL DESCRIPTION**

### 3.1 T1 / E1 / J1 MODE SELECTION

The IDT82P2916 can be configured to T1/J1 mode or E1 mode globally or on a per-channel basis. The configuration is determined by the TEHWE pin, the TEHW pin and the T1E1 bit (b0, CHCF,...). Refer to Table-1 for details of the operation mode selection.

#### Table-1 Operation Mode Selection

	Global Programming		Per-Channel Programming	
TEHWE Pin	Open		Low	
TEHW Pin	Open Low		(The configuration of this pin is ignored)	
T1E1 Bit	(The configuration of this bit is ignored).		0	1
Operation Mode	T1/J1 E1		T1/J1	E1

#### 3.2 RECEIVE PATH

#### 3.2.1 R<sub>X</sub> TERMINATION

The receive line interface supports Receive Differential mode and Receive Single Ended mode, as selected by the R\_SING bit (b3, RCF0,...). In Receive Differential mode, both RTIPn and RRINGn are used to receive signal from the line side. In Receive Single Ended mode, only RTIPn is used to receive signal.

In Receive Differential mode, the line interface can be connected with T1 100  $\Omega$ , J1 110  $\Omega$  or E1 120  $\Omega$  twisted pair cable or E1 75  $\Omega$  coaxial cable. In Receiver Single Ended mode, the line interface can only be connected with 75  $\Omega$  coaxial cable.

The receive impedance matching is realized by using internal impedance matching or external impedance matching for each channel in different applications.

#### 3.2.1.1 Receive Differential Mode

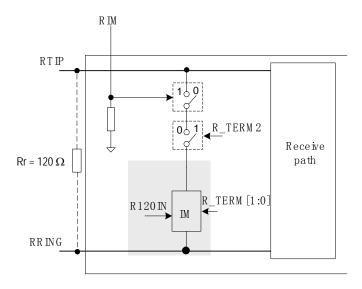
In Receive Differential mode, three kinds of impedance matching are supported: Fully Internal Impedance Matching, Partially Internal Impedance Matching and External Impedance Matching. Figure-3 shows an overview of how these Impedance Matching modes are switched.

Fully Internal Impedance Matching circuit uses an internal programmable resistor (IM) only and does not use an external resistor. This configuration saves external components and supports 1:1 Hitless Protection Switching (HPS) applications without relays. Refer to Section 4.4 Hitless Protection Switching (HPS) Summary.

Partially Internal Impedance Matching circuit consists of an internal programmable resistor (IM) and a value-fixed 120  $\Omega$  external resistor (Rr). Compared with Fully Internal Impedance Matching, this configuration provides considerable savings in power dissipation of the device. For example, In E1 120  $\Omega$  PRBS mode, the power savings would be 0.44 W. For power savings in other modes, please refer to Chapter 8

Physical And Electrical Specifications.

External Impedance Matching circuit uses an external resistor (Rr) only.



#### Figure-3 Switch between Impedance Matching Modes

To support some particular applications, such as hot-swap or Hitless Protection Switch (HPS) hot-switchover, RTIPn/RRINGn must be forced to enter high impedance state (i.e., External Impedance Matching). For hot-swap, RTIPn/RRINGn must be always held in high impedance state during /after power up; for HPS hot-switchover, RTIPn/RRINGn must enter high impedance state immediately after switchover. Though each channel can be individually configured to External Impedance Matching through register access, it is too slow for hitless switch. Therefore, a hardware pin - RIM - is provided to globally control the high impedance for all 16 receivers.

When RIM is low, only External Impedance Matching is supported for all 16 receivers and the per-channel impedance matching configuration bits - the R\_TERM[2:0] bits (b2~0, RCF0,...) and the R120IN bit (b4, RCF0,...) - are ignored.

When RIM is high, impedance matching is configured on a perchannel basis. Three kinds of impedance matching are all supported and selected by the R\_TERM[2:0] bits (b2~0, RCF0,...) and the R120IN bit (b4, RCF0,...). The R\_TERM[2] bit (b2, RCF0,...) should be set to match internal or external impedance. If the R\_TERM[2] bit (b2, RCF0,...) is '0', internal impedance matching is enabled. The R120IN bit (b4, RCF0,...) should be set to select Partially Internal Impedance Matching or Fully Internal Impedance Matching. The internal programmable resistor (IM) is determined by the R\_TERM[1:0] bits (b1~0,

Table-2	Impedance Matchin	g Value in Receive	Differential Mode

RCF0,...). If the R\_TERM[2] bit (b2, RCF0,...) is '1', external impedance matching is enabled. The configuration of the R120IN bit (b4, RCF0,...) and the R\_TERM[1:0] bits (b1~0, RCF0,...) is ignored.

A twisted pair cable can be connected with a 1:1 transformer or without a transformer (transformer-less), while a coaxial cable must be connected with a 1:1 transformer. Table 2 lists the recommended impedance matching value in different applications. Figure-4 to Figure-6 show the connection for one channel.

The transformer-less connection will offer a termination option with reduced cost and board space. However, the waveform amplitude is not standard compliant, and surge protection and common mode depression should be enhanced depending on equipment environment.

Partially Internal Impedance Matching (R120IN = 0) <sup>1</sup>		Fully Internal Impedance Matching (R120IN = 1) <sup>1, 2</sup>		External Impedance Matching	
R_TERM[2:0]	Rr	R_TERM[2:0]	Rr	R_TERM[2:0] <sup>3</sup>	Rr
000		000			100 Ω
001		001	(an an)		110 Ω
010		010	(open)		120 Ω
011	120 Ω	011		1XX	75 Ω
000					100 Ω
001		(not supp	ported)		110 Ω
010					120 Ω
-	R_TERM[2:0]         000         001         010         011         000         001	R_TERM[2:0]         Rr           000         001           010         120 Ω           000         001	R_TERM[2:0]         Rr         R_TERM[2:0]           000         000         000           001         001         001           010         120 Ω         011           000         001         (not supp	R_TERM[2:0]         Rr         R_TERM[2:0]         Rr           000         000         000         000         (open)           010         120 Ω         011         (open)         (open)           000         001         010         (open)         (open)	$\begin{array}{ c c c c c c c c c c c c c c c c c c c$

1. Partially Internal Impedance Matching and Fully Internal Impedance Matching are not supported when RIM is low.

2. Fully Internal Impedance Matching is not supported in transformer-less applications.

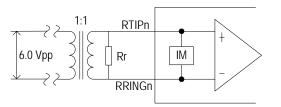
3. When RIM is low, the setting of the R\_TERM[2:0] bits is ignored.

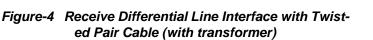
4. In transformer-less applications, the device should be protected against over-voltage. There are three important standards for over-voltage protection:

UL1950 and FCC Part 68;

Telcordia (Bellcore) GR-1089

• ITU-T K.20, K.21 and K.41





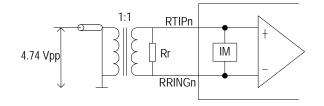
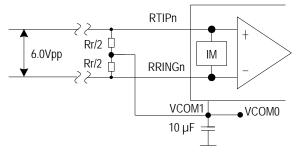


Figure-5 Receive Differential Line Interface with Coaxial Cable (with transformer)



Note: 1. Two Rr/2 resistors should be connected to VCOM[1:0] that are coupled to ground via a 10  $\mu\text{F}$  capacitor, which provide 60  $\Omega$ common mode input resistance.

2. In this mode, lightning protection should be enhanced. 3. The maximum input dynamic range of RTIP/TRING pin is -0.3 V ~3.6 V (in line monitor mode it is -0.3 V ~ 2 V)

Figure-6 Receive Differential Line Interface with Twisted Pair Cable (transformer-less, non standard compliant)

#### 3.2.1.2 Receive Single Ended Mode

Receive Single Ended mode can only be used in 75  $\Omega$  coaxial cable applications.

In Receive Single Ended mode, only External Impedance Matching is supported. External Impedance Matching circuit uses an external resistor (Rr) only. The value of the resistor is 18.75  $\Omega$  (see Figure-7 for details) when the single end is connected with a 2:1 transformer or is 75  $\Omega$  (see Figure-8 for details) when the single end is connected without a transformer.

In Receive Single Ended mode, the RIM pin should be left open and the configuration of the R\_TERM[2:0] bits (b2~0, RCF0,...) is ignored.

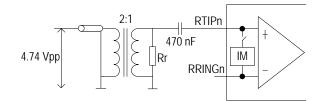
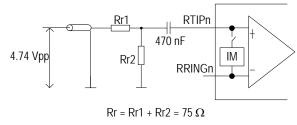


Figure-7 Receive Single Ended Line Interface with Coaxial Cable (with transformer)



Note: In this mode, port protection should be enhanced.

Figure-8 Receive Single Ended Line Interface with Coaxial Cable (transformer-less, non standard compliant)

#### 3.2.2 EQUALIZER

The equalizer compensates high frequency attenuation to enhance receive sensitivity.

#### 3.2.2.1 Line Monitor

In both T1/J1 and E1 short haul applications, the Protected Non-Intrusive Monitoring per T1.102 can be performed between two devices. The monitored channel of one device is in normal operation, and the monitoring channel of the other device taps the monitored one through a high impedance bridging circuit (refer to Figure-9 and Figure-10).

After the high resistance bridging circuit, the signal arriving at RTIPn/ RRINGn of the monitoring channel is dramatically attenuated. To compensate this bridge resistive attenuation, Monitor Gain can be used to boost the signal by 20 dB, 26 dB or 32 dB, as selected by the MG[1:0] bits (b1~0, RCF2,...). For normal operation, the Monitor Gain should be set to 0 dB, i.e., the Monitor Gain of the monitored channel should be 0 dB.

The monitoring channel can be configured to any of the External, Partially Internal or Fully Internal Impedance Matching mode. Here the external r or internal IM is used for voltage division, not for impedance matching. That is, the r (IM) and the two R make up of a resistance bridge. The resistive attenuation of this bridge is 20lg(r/(2R+r)) dB.

Note that line monitor is only available in differential line interface.

#### 3.2.2.2 Receive Sensitivity

The receive sensitivity is the minimum range of receive signal level for which the receiver recovers data error-free with -18 dB interference signal added.

For Receive Differential line interface, the receive sensitivity is -15 dB.

For Receive Single Ended line interface, the receive sensitivity is -12 dB.

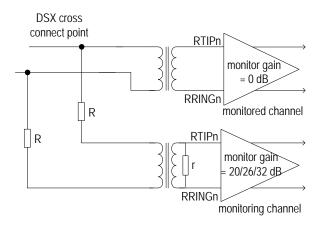


Figure-9 Receive Path Monitoring

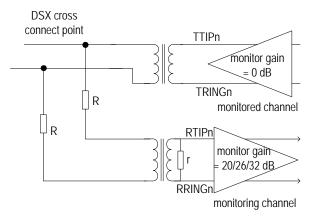


Figure-10 Transmit Path Monitoring

#### 3.2.3 SLICER

The Slicer is used to generate a standard amplitude mark or a space according to the amplitude of the input signals. The input signal is sliced at 50% of the peak value.

#### 3.2.4 R<sub>x</sub> CLOCK & DATA RECOVERY

The Rx Clock & Data Recovery is used to recover the clock signal from the received data. It is accomplished by an integrated Digital Phase Locked Loop (DPLL). The recovered clock tracks the jitter in the data output from the Slicer and keeps the phase relationship between data and clock during the absence of the incoming pulse.

Note that the IDT82P2916 also provides programmable REFA and REFB pins to output any of the 16 recovered line clocks. Refer to Section 3.6 Clock Inputs and Outputs for details.

#### 3.2.5 DECODER

The Decoder is used only when the receive system interface is in Single Rail NRZ Format mode. When the receive system interface is in other modes, the Decoder is bypassed automatically. (Refer to Section 3.2.6 Receive System Interface for the description of the receive system interface).

In T1/J1 mode, the received signal is decoded by AMI or B8ZS line code rule. In E1 mode, the received signal is decoded by AMI or HDB3 line code rule. The line code rule is selected by the R\_CODE bit (b2, RCF1,...).

#### 3.2.6 RECEIVE SYSTEM INTERFACE

The received data can be output to the system side in four modes: Single Rail NRZ Format mode, Dual Rail NRZ Format mode, Dual Rail RZ Format mode and Dual Rail Sliced mode, as selected by the R\_MD[1:0] bits (b1~0, RCF1).

If data is output on RDn in NRZ format and the recovered clock is output on RCLKn, the receive system interface is in Single Rail NRZ Format mode. In this mode, the data is decoded and updated on the active edge of RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock. The Receive Multiplex Function (RMFn) signal is updated on the active edge of RCLKn and can be selected to indicate PRBS/ARB, LAIS, LEXZ, LBPV, LEXZ + LBPV, LLOS, output recovered clock (RCLK) or XOR output of positive and negative sliced data. Refer to Section 3.5.7.1 RMFn Indication for the description of RMFn.

If data is output on RDPn and RDNn in NRZ format and the recovered clock is output on RCLKn, the receive system interface is in Dual Rail NRZ Format mode. In this mode, the data is un-decoded and updated on the active edge of RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock.

If data is output on RDPn and RDNn in RZ format and the recovered clock is output on RCLKn, the receive system interface is in Dual Rail RZ Format mode. In this mode, the data is un-decoded and updated on the active edge of RCLKn. RCLKn outputs a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock.

If data is output on RDPn and RDNn in RZ format directly after passing through the Slicer, the receive system interface is in Dual Rail Sliced mode. In this mode, the data is raw sliced and un-decoded. RMFn can be selected to indicate PRBS/ARB, LAIS, LEXZ, LBPV, LEXZ + LBPV, LLOS, output recovered clock (RCLK) or XOR output of positive and negative sliced data. Refer to Chapter 3.5.7.1 RMFn Indication for the description of RMFn.

Table-3 summarizes the multiplex pin used in different receive system interface.

Table-3	Multiplex Pin Used in Rec	eive System Interface
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Multiplex Pin Used On Receive System Interface			
RDn / RDPn	RDNn / RMFn	RCLKn / RMFn	
RDn <sup>1</sup>	RMFn <sup>2</sup>	RCLKn <sup>3</sup>	
RDPn <sup>1</sup>	RDNn <sup>1</sup>	RCLKn <sup>3</sup>	
RDPn <sup>1</sup>	RDNn <sup>1</sup>	RCLKn <sup>3</sup>	
RDPn <sup>1</sup>	RDNn <sup>1</sup>	RMFn <sup>2</sup>	
	RDn / RDPn RDn <sup>1</sup> RDPn <sup>1</sup> RDPn <sup>1</sup>	InterfaceRDn / RDPnRDNn / RMFnRDn 1RMFn 2RDPn 1RDNn 1RDPn 1RDNn 1	

1. The active level on RDn, RDPn and RDNn is selected by the RD\_INV bit (b3, RCF1,...).

2. RMFn is always active high.

3. The active edge of RCLKn is selected by the RCK\_ES bit (b4, RCF1,...).

#### 3.2.7 RECEIVER POWER DOWN

Set the R\_OFF bit (b5, RCF0,...) to '1' will power down the corresponding receiver.

In this way, the corresponding receive circuit is turned off and the RTIPn/RRINGn pins are forced to High-Z state. The pins on receive system interface (including RDn/RDPn, RDNn/RMFn, RCLKn/RMFn) will be in High-Z state if the RHZ bit (b6, RCF0,...) is '1' or in low level if the RHZ bit (b6, RCF0,...) is '0'.

After clearing the R\_OFF bit (b5, RCF0,...), it will take 1 ms for the receiver to achieve steady state, i.e., to return to the previous configuration and performance.

#### 3.3 TRANSMIT PATH

#### 3.3.1 TRANSMIT SYSTEM INTERFACE

The data from the system side is input to the device in three modes: Single Rail NRZ Format mode, Dual Rail NRZ Format mode and Dual Rail RZ Format mode, as selected by the  $T_MD[1:0]$  bits (b1~0, TCF1,...).

If data is input on TDn in NRZ format and a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock is input on TCLKn, the transmit system interface is in Single Rail NRZ Format mode. In this mode, the data is encoded and sampled on the active edge of TCLKn. TMFn is updated on the active edge of TCLKn and can be selected to indicate PRBS/ARB, SAIS, TOC, TLOS or SEXZ. Refer to Section 3.5.7.2 TMFn Indication for the description of TMFn.

If data is input on TDPn and TDNn in NRZ format and a 1.544 MHz (in T1/J1 mode) or 2.048 MHz (in E1 mode) clock is input on TCLKn, the transmit system interface is in Dual Rail NRZ Format mode. In this mode, the data is pre-encoded and sampled on the active edge of TCLKn.

If data is input on TDPn and TDNn in RZ format and no transmit clock is input, the transmit system interface is in Dual Rail RZ Format mode. In this mode, the data is pre-encoded. TMFn can be selected to indicate PRBS/ARB, SAIS, TOC, TLOS, SEXZ, SBPV, SEXZ + SBPV or SLOS. Refer to Section 3.5.7.2 TMFn Indication for the description of TMFn. The Tx Clock Recovery block is used to recover the clock signal from the data input on TDPn and TDNn. Refer to Section 3.3.2 Tx Clock Recovery.

Table-4 summarizes the multiplex pin used in different transmit system interface.

Table-4	Multiplex Pi	n Used in	Transmit	System Interface	ł
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Transmit System Interface	Multiplex Pin Used On Transmit System Interface			
incidee	TDn / TDPn	TDNn / TMFn	TCLKn / TDNn	
Single Rail NRZ Format	TDn <sup>1</sup>	TMFn <sup>2</sup>	TCLKn <sup>3</sup>	
Dual Rail NRZ Format	TDPn <sup>1</sup>	TDNn <sup>1</sup>	TCLKn <sup>3</sup>	
Dual Rail RZ Format	TDPn <sup>1</sup>	TMFn <sup>2</sup>	TDNn <sup>1</sup>	

#### Note:

1. The active level on TDn, TDPn and TDNn is selected by the TD\_INV bit (b3, TCF1,...).

2. TMFn is always active high.

3. The active edge of TCLKn is selected by the TCK\_ES bit (b4, TCF1,...). If TCLKn is missing, i.e., no transition for more than 64 T1/E1 clock cycles, the TCKLOS\_S bit (b3, STAT0,...) will be set. A transition from '0' to '1' on the TCKLOS\_S bit (b3, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the TCKLOS\_S bit (b3, STAT0,...) will set the TCKLOS\_IS bit (b3, INTS0,...) to '1', as selected by the TCKLOS\_IES bit (b3, INTES,...). When the TCKLOS\_IS bit (b3, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the TCKLOS\_IM bit (b3, INTM0,...).

#### 3.3.2 T<sub>X</sub> CLOCK RECOVERY

The Tx Clock Recovery is used only when the transmit system interface is in Dual Rail RZ Format mode. When the transmit system interface is in other modes, the Tx Clock Recovery is bypassed automatically.

The Tx Clock Recovery is used to recover the clock signal from the data input on TDPn and TDNn.

#### 3.3.3 ENCODER

The Encoder is used only when the transmit system interface is in Single Rail NRZ Format mode. When the transmit system interface is in other modes, the Encoder is bypassed automatically.

In T1/J1 mode, the data to be transmitted is encoded by AMI or B8ZS line code rule. In E1 mode, the data to be transmitted is encoded by AMI or HDB3 line code rule. The line code rule is selected by the T\_CODE bit (b2, TCF1,...).

#### 3.3.4 WAVEFORM SHAPER

The IDT82P2916 provides two ways to manipulate the pulse shape before data is transmitted:

- Preset Waveform Template;
- User-Programmable Arbitrary Waveform.

#### 3.3.4.1 Preset Waveform Template

In T1/J1 applications, the waveform template meets T1.102. The T1 template is shown in Figure-11. It is measured in the far end, as shown in Figure-12. The J1 template is measured in the near end line side.

In T1 applications, to meet the template, five preset waveform templates are provided according to five grades of cable length. The selection is made by the PULS[3:0] bits (b3~0, PULS,...). In J1 applications, the PULS[3:0] bits (b3~0, PULS,...) should be set to '0111'. Refer to Table-5 for details.

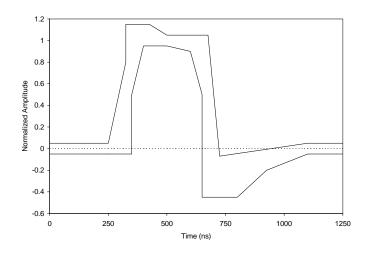


Figure-11 DSX-1 Waveform Template

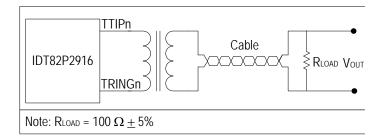


Figure-12 T1 Waveform Template Measurement Circuit

Cable Conditions	PULS[3:0]
DSX1 - 0 ~ 133 ft	0010
DSX1 - 133 ~ 266 ft	0011
DSX1 - 266 ~ 399 ft	0100
DSX1 - 399 ~ 533 ft	0101
DSX1 - 533 ~ 655 ft	0110
J1 - 0 ~ 655 ft	0111

Table-5 PULS[3:0] Setting in T1/J1 Mode

In E1 applications, the waveform template meets G.703, as shown in Figure-13. It is measured in the near end line side, as shown in Figure-14.

In E1 applications, the PULS[3:0] should be set to '0000' if differential signals (output from TTIP and TRING) are coupled to a 75  $\Omega$  coaxial cable using Internal Impedance matching mode; the PULS[3:0] should be set to '0001' for other E1 interfaces. Refer to Table-6 for details.

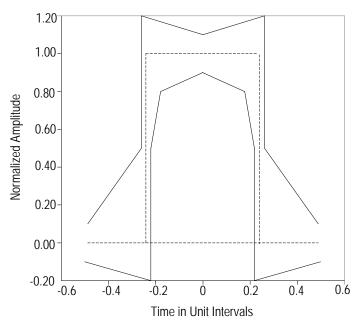
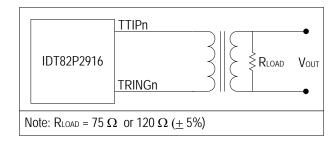


Figure-13 E1 Waveform Template



#### Figure-14 E1 Waveform Template Measurement Circuit

Table-6	PULS[3:0] Setting in E1 Mode
---------	------------------------------

Interface Conditions	PULS[3:0]
E1 75 $\Omega$ differential interface, Internal Impedance matching mode	0000
Other E1 interface	0001

After one of the preset waveform templates is selected, the preset waveform amplitude can be adjusted to get the desired waveform.

In T1 mode, the standard value of the SCAL[5:0] bits (b5~0, SCAL,...) is '110110' which is also the default value. The adjusting is made by increasing or decreasing by '1' from the standard value to scale up or down at a percentage ratio of 2% against the preset waveform amplitude.

In E1 mode, the SCAL[5:0] bits (b5~0, SCAL,...) should be set to '100001' to get the standard amplitude. The adjusting is made by increasing or decreasing by '1' from the standard value to scale up or down at a percentage ratio of 3%.

In summary, do the following step by step, the desired waveform will be got based on the preset waveform template:

- Select one preset waveform template by setting the PULS[3:0] bits (b3~0, PULS,...);
- Write '100001 to the SCAL[5:0] bits (b5~0, SCAL,...) if E1 mode is selected.
- Write the scaling value to the SCAL[5:0] bits (b5~0, SCAL,...) to scale the amplitude of the selected preset waveform template (- this step is optional).

#### 3.3.4.2 User-Programmable Arbitrary Waveform

When the PULS[3:0] bits (b3~0, PULS,...) are set to '1XXX', userprogrammable arbitrary waveform will be used in the corresponding channel.

Each waveform shape can extend up to  $1\frac{1}{4}$  UIs (Unit Interval), and is

divided into 20 sub-phases that are addressed by the SAMP[4:0] bits (b4~0, AWG0,...). The waveform amplitude of each phase is represented by a binary byte, within the range from +63 to -63, stored in the WDAT[6:0] bits (b6~0, AWG1,...) in signed magnitude form. The maximum number +63 (D) represents the maximum positive amplitude of the transmit pulse while the most negative number -63 (D) represents the maximum negative amplitude of the transmit pulse. Therefore, up to 20 bytes are used.

There are eight standard templates which are stored in a local ROM. One of them can be selected as reference and made some changes to get the desired waveform.

To do this, the first step is to choose a set of waveform value from the standard templates. The selected waveform value should be the most similar to the desired waveform shape. Table-7 to Table-14 list the sample data of each template.

Then modify the sample data to get the desired transmit waveform shape. By increasing or decreasing by '1' from the standard value in the SCAL[5:0] bits ( $b5\sim0$ , SCAL,...), the waveform amplitude will be scaled up or down.

- In summary, do the following for the write operation:
- Modify the sample data in the AWG1 register;
- Write the AWG0 register to implement the write operation, including:
  - Write the sample address to the SAMP[4:0] bits (b4~0, AWG0,...);
  - Write '0' to the RW bit (b5, AWG0,...);
  - Write '1' to the DONE bit (b6, AWG0,...).

Do the following for the read operation:

- Write the AWG0 register, including:
  - Write sample address to the SAMP[4:0] bits (b4~0, AWG0,...);
  - Write '1' to the RW bit (b5, AWG0,...);
  - Write '1' to the DONE bit (b6, AWG0,...);
- · Read the AWG1 register to get the sample data.

When the write operation is completed, write the scaling value to the SCAL[5:0] bits (b5~0, SCAL,...) to scale the amplitude of the selected standard waveform (- this step is optional).

When more than one UI is used to compose the waveform template and the waveform amplitude is not set properly, the overlap of the two consecutive waveforms will make the waveform amplitude overflow (i.e., exceed the maximum limitation). This overflow is captured by the DAC\_IS bit (b7, INTS0,...) and will be reported by the INT pin if enabled by the DAC\_IM bit (b7, INTM0,...).

Refer to application note AN-513 'User-Programmable Arbitrary Waveform for DSX1' for more details.

# RENESAS

#### Table-7 Transmit Waveform Value for T1 0 ~ 133 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	17H	27H	27H	26H	25H	25H	25H	24H	23H	4AH	4AH	49H	47H	45H	44H	43H	42H	41H	00H	00H

#### Table-8 Transmit Waveform Value for T1 133 ~ 266 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	1BH	2EH	2CH	2AH	29H	28H	27H	26H	25H	50H	4FH	4DH	4AH	48H	46H	44H	43H	42H	41H	00H

#### Table-9 Transmit Waveform Value for T1 266 ~ 399 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	1FH	34H	2FH	2CH	2BH	2AH	29H	28H	25H	57H	53H	50H	4BH	48H	46H	44H	43H	42H	41H	00H

#### Table-10 Transmit Waveform Value for T1 399 ~ 533 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	20H	3BH	35H	2FH	2EH	2DH	2CH	2AH	28H	58H	58H	53H	4CH	48H	46H	44H	43H	42H	41H	00H

#### Table-11 Transmit Waveform Value for T1 533 ~ 655 ft

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	20H	3FH	38H	33H	2FH	2EH	2DH	2CH	29H	5FH	5EH	57H	4FH	49H	47H	44H	43H	42H	41H	00H

#### Table-12 Transmit Waveform Value for E1 75 ohm

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	00H	00H	00H	0CH	30H	00H														

#### Table-13 Transmit Waveform Value for E1 120 ohm

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	00H	00H	00H	0FH	3CH	00H														

#### Table-14 Transmit Waveform Value for J1 0 $\sim 655 \mbox{ ft}$

SAMP[4:0]	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
WDAT[6:0]	17H	27H	27H	26H	25H	25H	25H	24H	23H	4AH	4AH	49H	47H	45H	44H	43H	42H	41H	00H	00H

#### 3.3.5 LINE DRIVER

The Line Driver can be set to High-Z for protection or in redundant applications.

The following two ways will set the Line Driver to High-Z:

- Setting the OE pin to low will globally set all the Line Drivers to High-Z;
- Setting the OE bit (b6, TCF0,...) to '0' will set the corresponding Line Driver to High-Z.

By these ways, the functionality of the internal circuit is not affected and TTIPn and TRINGn will enter High-Z state immediately.

#### 3.3.5.1 Transmit Over Current Protection

The Line Driver monitors the Transmit Over Current (TOC) on the line interface. When TOC is detected, the driver's output (i.e., output on TTIPn/TRINGn) is determined by the THZ\_OC bit (b4, TCF0,...). If the THZ\_OC bit (b4, TCF0,...) is '0', the driver's output current (peak to peak) is limited to 100 mA; if the THZ\_OC bit (b4, TCF0,...) is '1', the driver's output will enter High-Z. TOC is indicated by the TOC\_S bit (b4, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the TOC\_S bit (b4, STAT0,...) will set the TOC\_IS bit (b4, INTS0,...) to '1', as selected by the TOC\_IES bit (b4, INTES,...). When the TOC\_IS bit (b4, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the TOC\_IM bit (b4, INTM0,...).

TOC may be indicated by the TMFn pin. Refer to Section 3.5.7.2 TMFn Indication for details.

#### 3.3.6 T<sub>X</sub> TERMINATION

The transmit line interface supports Transmit Differential mode and Transmit Single Ended mode, as selected by the T\_SING bit (b3, TCF0,...). In Transmit Differential mode, both TTIPn and TRINGn are used to transmit signals to the line side. In Transmit Single Ended mode, only TTIPn is used to transmit signal.

The line interface can be connected with T1 100  $\Omega$ , J1 110  $\Omega$  or E1 120  $\Omega$  twisted pair cable or E1 75  $\Omega$  coaxial cable.

The transmit impedance matching is realized by using internal impedance matching or external impedance matching for each channel in different applications.

#### 3.3.6.1 Transmit Differential Mode

In Transmit Differential mode, different applications have different impedance matching. For T1/J1 applications, only Internal Impedance Matching is supported. For E1 applications, both Internal and External Impedance Matching are supported.

Internal Impedance Matching circuit uses an internal programmable resistor (IM) only.

External Impedance Matching circuit uses an external resistor (Rt) only.

A twisted pair cable can be connected with a 1:2 (step up) transformer or without a transformer (transformer-less), while a coaxial cable must be connected with a 1:2 transformer.

The T\_TERM[2:0] bits (b2~0, TCF0,...) should be set according to different cable conditions, whether a transformer is used, and what kind of Impedance Matching is selected.

Table-15 lists the recommended impedance matching value in different applications. Figure-15 to Figure-17 show the connection for one channel in different applications.

The transformer-less connection will offer a termination option with reduced cost and board space. However, the waveform amplitude is not standard compliant, and surge protection and common mode depression should be enhanced depending on equipment environment.

Table-15 Impedance Matching Value in Transmit Differential Mode

Cable Condition	Interna	I Impedance Mat	ching	Extern	al Impedance Mate	ching
Cable Condition	T_TERM[2:0]	Rt	PULS[3:0]	T_TERM[2:0]	Rt	PULS[3:0]
T1 100 $\Omega$ twisted pair (with transformer)	000		Table-5		(not supported)	1
J1 110 $\Omega$ twisted pair (with transformer)	001		0111	•	(not supported)	
E1 120 $\Omega$ twisted pair (with transformer)	010		0001	111	10 Ω	0000
E1 75 $\Omega$ coaxial (with transformer)	011	0	0000			0000
T1 100 $\Omega$ twisted pair (transformer-less)	100		Table-5		I	
J1 110 $\Omega$ twisted pair (transformer-less)	101		0111		(not supported)	
E1 120 $\Omega$ twisted pair (transformer-less)	110		0001			

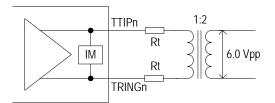


Figure-15 Transmit Differential Line Interface with Twisted Pair Cable (with Transformer)

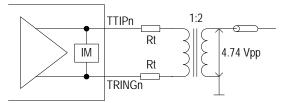
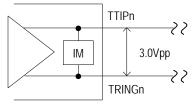


Figure-16 Transmit Differential Line Interface with Coaxial Cable (with transformer)



Note: In this mode, port protection should be enhanced.

Figure-17 Transmit Differential Line Interface with Twisted Pair Cable (transformer-less, non standard compliant)

#### 3.3.6.2 Transmit Single Ended Mode

Transmit Single Ended mode can only be used in 75  $\boldsymbol{\Omega}$  coaxial cable applications.

In Transmit Single Ended mode, only Internal Impedance Matching is supported. Internal Impedance Matching circuit uses an internal programmable resistor (IM) only. The T\_TERM[2:0] bits (b2~0, TCF0,...) should be set to '011'. The output amplitude is 4.74 Vpp when PULS[3:0] is '0001' and the SCAL[5:0] bits (b5~0, SCAL...) is '100001'.<sup>1</sup>

In Single Ended mode, special care has to be taken for termination and overall setup. Refer to separate application note for details.

A 1:2 (step up) transformer should be used in application.

Figure-18 shows the connection for one channel.

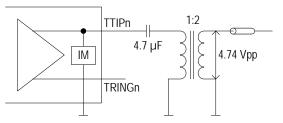


Figure-18 Transmit Single Ended Line Interface with Coaxial Cable (with transformer)

 The waveform in this mode is not standard. However, if the arbitrary waveform generator is used, the waveform could pass the template marginally.

#### 3.3.7 TRANSMITTER POWER DOWN

Set the T\_OFF bit (b5, TCF0,...) to '1' will power down the corresponding transmitter.

In this way, the corresponding transmit circuit is turned off. The pins on the transmit line interface (including TTIPn and TRINGn) will be in High-Z state. The input on the transmit system interface (including TDn, TDPn, TDNn and TCLK) is ignored. The output on the transmit system interface (i.e. TMFn) will be in High-Z state.

After clearing the T\_OFF bit (b5, TCF0,...), it will take 1 ms for the transmitter to achieve steady state, i.e., return to the previous configuration and performance.

#### 3.3.8 OUTPUT HIGH-Z ON TTIP AND TRING

TTIPn and TRINGn can be set to High-Z state globally or on a perchannel basis.

The following three conditions will set TTIPn and TRINGn to High-Z state globally:

- · Connecting the OE pin to low;
- Loss of MCLK (i.e., no transition on MCLK for more than 1 ms);
- Power on reset, hardware reset by pulling RST to low for more than 2 µs or global software reset by writing the RST register.

The following six conditions will set TTIPn and TRINGn to High-Z state on a per-channel basis:

- Writing '0' to the OE bit (b6, TCF0,...);
- Loss of TCLKn in Transmit Single Rail NRZ Format mode or Transmit Dual Rail NRZ Format mode (i.e., no transition on TCLKn for more than 64 XCLK<sup>1</sup> cycles) except that the channel is in Remote Loopback or transmit internal pattern with XCLK;
- Transmitter power down;
- Per-channel software reset by writing '1' to the CHRST bit (b1, CHCF,...);
- Setting the THZ\_OC bit (b4, TCF0,...) to '1' when transmit driver over-current is detected.
  - 1. XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

### 3.4 JITTER ATTENUATOR (RJA & TJA)

Two Jitter Attenuators are provided for each channel of receiver and transmitter. Each Jitter Attenuator can be enabled or disabled, as determined by the RJA\_EN/TJA\_EN bit (b3, RJA/TJA,...) respectively.

Each Jitter Attenuator consists of a FIFO and a DPLL, as shown in Figure-19.

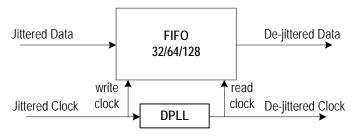


Figure-19 Jitter Attenuator

The FIFO is used as a pool to buffer the jittered input data, then the data is clocked out of the FIFO by a de-jittered clock. The depth of the FIFO can be 32 bits, 64 bits or 128 bits, as selected by the RJA\_DP[1:0]/ TJA\_DP[1:0] bits (b2~1, RJA/TJA,...). Accordingly, the typical delay produced by the Jitter Attenuator is 16 bits, 32 bits or 64 bits. The 128bit FIFO is used when large jitter tolerance is expected, while the 32-bit FIFO is used in delay sensitive applications. The DPLL is used to generate a de-jittered clock to clock out the data stored in the FIFO. The DPLL can only attenuate the incoming jitter whose frequency is above Corner Frequency (CF) by 20 dB per decade falling off. The jitter whose frequency is lower than the CF passes through the DPLL without any attenuation. In T1/J1 applications, the CF of the DPLL is 5 Hz or 1.26 Hz. In E1 applications, the CF of the DPLL is 6.77 Hz or 0.87 Hz. The CF is selected by the RJA\_BW/TJA\_BW bit (b0, RJA/TJA,...). The lower the CF is, the longer time is needed to achieve synchronization.

If the incoming data moves faster than the outgoing data, the FIFO will overflow. If the incoming data moves slower than the outgoing data, the FIFO will underflow. The overflow and underflow are both captured by the RJA\_IS/TJA\_IS bit (b5/6, INTS0,...). The occurrence of overflow or underflow will be reported by the INT pin if enabled by the RJA\_IM/TJA\_IM bit (b5/6, INTM0,...).

To avoid overflow or underflow, the JA-Limit function can be enabled by setting the RJA\_LIMT/TJA\_LIMT bit (b4, RJA/TJA,...). When the JA-Limit function is enabled, the speed of the outgoing data will be adjusted automatically if the FIFO is 2-bit close to its full or emptiness. Though the JA-Limit function can reduce the possibility of FIFO overflow and underflow, the quality of jitter attenuation is deteriorated.

The performance of the Jitter Attenuator meets ITUT I.431, G.703, G.736-739, G.823, G.824, ETSI 300011, ETSI TBR12/13, AT&T TR62411, TR43802, TR-TSY 009, TR-TSY 253 and TR-TRY 499. Refer to Section 8.12 Jitter Attenuation Characteristics for the jitter performance.

### 3.5 DIAGNOSTIC FACILITIES

The diagnostic facilities include:

- BPV (Bipolar Violation) / CV (Code Violation) detection and BPV insertion;
- EXZ (Excessive Zero) detection;
- · LOS (Loss Of Signal) detection;
- AIS (Alarm Indication Signal) detection and generation;
- Pattern generation and detection, including PRBS (Pseudo Random Bit Sequence), ARB (Arbitrary Pattern) and IB (Inband Loopback).

The above defects, alarms or patterns can be counted by an internal Error Counter, indicated by the respective interrupt bit and indicated by RMFn or TMFn.

# 3.5.1 BIPOLAR VIOLATION (BPV) / CODE VIOLATION (CV) DETECTION AND BPV INSERTION

#### 3.5.1.1 Bipolar Violation (BPV) / Code Violation (CV) Detection

BPV/CV is monitored in both the receive path and the transmit path. BPV is detected when the data is AMI coded and CV is detected when the data is B8ZS/HDB3 coded. If the transmit system interface is in Transmit Single Rail NRZ Format mode, the BPV/CV detection is disabled in the transmit path automatically.

A BPV is detected when two consecutive pulses of the same polarity are received.

A CV is detected when two consecutive BPVs of the same polarity that are not a part of the B8ZS/HDB3 zero substitution are received.

When BPV/CV is detected in the receive path, the Line Bipolar Violation LBPV\_IS bit (b4, INTS2,...) will be set and an interrupt will be reported by INT if not masked by the LBPV\_IM bit (b4, INTM2,...).

When BPV/CV is detected in the transmit path, the System Bipolar Violation SBPV\_IS bit (b5, INTS2,...) will be set and an interrupt will be reported by INT if not masked by the SBPV\_IM bit (b5, INTM2,...).

BPV/CV may be counted by an internal Error Counter or may be indicated by the RMFn or TMFn pin. Refer to Section 3.5.6 Error Counter and Section 3.5.7 Receive /Transmit Multiplex Function (RMF / TMF) Indication respectively.

#### 3.5.1.2 Bipolar Violation (BPV) Insertion

The BPV can only be inserted in the transmit path.

A BPV will be inserted on the next available mark in the data stream to be transmitted by writing a '1' to the BPV\_INS bit (b6, ERR,...). This bit will be reset once BPV insertion is done.

#### 3.5.2 EXCESSIVE ZEROES (EXZ) DETECTION

EXZ is monitored in both the receive path and the transmit path.

Different line code has different definition of the EXZ. The IDT82P2916 provides two standards of EXZ definition for each kind of line code rule. The standards are ANSI and FCC, as selected by the EXZ\_DEF bit (b7, ERR,...). Refer to Table-16 for details.

#### Table-16 EXZ Definition

Line Code	Defir	nition
Rule	ANSI (EXZ_DEF = 0)	FCC (EXZ_DEF = 1)
AMI		T1/J1 - An EXZ is detected when any string of more than 80 consecutive '0's are received. E1 - An EXZ is detected when any string of more than 15 consecutive '0's are received.
B8ZS	An EXZ is detected when any string of more than 7 consecutive '0's are received.	An EXZ is detected when any string of more than 7 consecutive '0's are received.
HDB3	An EXZ is detected when any string of more than 3 consecutive '0's are received.	
,	em interface is in Transmit Single R y to the standard of AMI.	Rail NRZ Format mode, the EXZ is

When EXZ is detected in the receive path, the LEXZ\_IS bit (b2, INTS2,...) will be set and an interrupt will be reported by  $\overline{INT}$  if not masked by the LEXZ\_IM bit (b2, INTM2,...).

When EXZ is detected in the transmit path, the SEXZ\_IS bit (b3, INTS2,...) will be set and an interrupt will be reported by  $\overline{INT}$  if not masked by the SEXZ\_IM bit (b3, INTM2,...).

EXZ may be counted by an internal Error Counter or may be indicated by the RMFn or TMFn pin. Refer to Chapter 3.5.6 Error Counter.

#### 3.5.3 LOSS OF SIGNAL (LOS) DETECTION

The IDT82P2916 detects three kinds of LOS:

- · LLOS: Line LOS, detected in the receive path;
- SLOS: System LOS, detected in the transmit system side;
- TLOS: Transmit LOS, detected in the transmit line side.

#### 3.5.3.1 Line LOS (LLOS)

Table-17 LLOS Criteria

The amplitude and density of the data received from the line side are monitored. When the amplitude of the data is less than Q Vpp for N consecutive pulse intervals, LLOS is declared. When the amplitude of the data is more than P Vpp and the average density of marks is at least 12.5% for M consecutive pulse intervals starting with a mark, LLOS is cleared. Here Q is defined by the ALOS[2:0] bits (b6~4, LOS,...). P is the sum of Q and 250 mVpp. N and M are defined by the LAC bit (b7, LOS,...). Refer to Table-17 for details.

In T1/J1 mode, LLOS detection supports ANSI T1.231 and I.431. In E1 mode, LLOS detection supports G.775 and ETSI 300233/I.431. The criteria are selected by the LAC bit (b7, LOS,...).

When LLOS is detected, the LLOS\_S bit (b0, STAT0,...) will be set. A transition from '0' to '1' on the LLOS\_S bit (b0, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the LLOS\_S bit (b0, STAT0,...) will set the LLOS\_IS bit (b0, INTS0,...) to '1', as selected by the LOS\_IES bit (b1, INTES,...). When the LLOS\_IS bit (b0, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the LLOS\_IM bit (b0, INTM0,...).

Two pins (LLOS0 and LLOS) are dedicated to LLOS indication. Whether LLOS is detected in channel 0 or not, LLOS0 is high for a CLKE1 clock cycle to indicate the start position on LLOS. LLOS indicates LLOS status of all 16 channels in a serial format and repeats every 16 cycles. Refer to Figure-20. LLOS0 and LLOS are updated on the rising edge of CLKE1. When the clock output on CLKE1 is disabled, LLOS0 and LLOS will be held in High-Z state. The output on CLKE1 is controlled by the CLKE1\_EN bit (b3, CLKG) and the CLKE1 bit (b2, CLKG). Refer to section 8.11 on page 125 for CLKE1 timing characteristics.

LLOS may be counted by an internal Error Counter or may be indicated by the RMFn pin. Refer to Section 3.5.6 Error Counter and Section 3.5.7.1 RMFn Indication respectively.

During LLOS, in Receive Single Rail NRZ Format mode, Receive Dual Rail NRZ Format mode and Receive Dual Rail RZ Format mode, RDPn/RDNn output low level. In Receive Dual Rail Sliced mode RDPn/ RDNn still output sliced data. RCLKn (if available) outputs high level or XCLK<sup>1</sup>, as selected by the RCKH bit (b7, RCF0,...).

During LLOS, if any of AIS, pattern generation in the receive path or Digital Loopback is enabled or automatic digital loopback happens, RDPn/RDNn and RCLKn output corresponding data and clock, and the setting of the RCKH bit (b7, RCF0,...) is ignored. Refer to the corresponding chapters for details.

<sup>1.</sup> XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

Operation Mode	LAC	Criteria	LLOS Declaring	LLOS Clearing
T1/J1	0	ANSI T1.231	below Q Vpp, N = 175 bits	above P Vpp, 12.5% mark density with less than 100 consecutive zeros, M = 175 bits
11/31	1	ANSI I.431	below Q Vpp, N = 1544 bits	above P Vpp, 12.5% mark density with less than 100 consecutive zeros, M = 175 bits
	0	G.775	below Q Vpp, N = 32 bits	above P Vpp, 12.5% mark density with less than 16 consecutive zeros, M = 32 bits
E1	1	ETSI 300233/ I.431	below Q Vpp, N = 2048 bits	above P Vpp, 12.5% mark density with less than 16 consecutive zeros, M = 32 bits

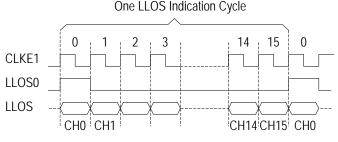


Figure-20 LLOS Indication on Pins

#### 3.5.3.2 System LOS (SLOS)

SLOS can only be detected when the transmit system interface is in Dual Rail NRZ Format mode or in Dual Rail RZ Format mode.

The amplitude and density of the data input from the transmit system side are monitored. When the input '0's are equal to or more than N consecutive pulse intervals, SLOS is declared. When the average density of marks is at least 12.5% for M consecutive pulse intervals starting with a mark, SLOS is cleared. Here N and M are defined by the LAC bit (b7, LOS,...). Refer to Table-18 for details.

In T1/J1 mode, SLOS detection supports ANSI T1.231 and I.431. In E1 mode, SLOS detection supports G.775 and ETSI 300233/I.431. The criteria are selected by the LAC bit (b7, LOS,...).

When SLOS is detected, the SLOS\_S bit (b1, STAT0,...) will be set. A transition from '0' to '1' on the SLOS\_S bit (b1, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the SLOS\_S bit (b1, STAT0,...) will set the SLOS\_IS bit (b1, INTS0,...) to '1', as selected by the LOS\_IES bit (b1, INTES,...). When the SLOS\_IS bit (b1, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the SLOS\_IM bit (b1, INTM0,...).

SLOS may be counted by an internal Error Counter or may be indicated by the TMFn pin. Refer to Section 3.5.6 Error Counter and Section 3.5.7.2 TMFn Indication respectively.

Operation Mode	LAC	Criteria	SLOS Declaring <sup>1</sup>	SLOS Clearing <sup>1</sup>
T1/J1	0	ANSI T1.231	no pulse detected for N consecutive pulse intervals, N = 175 bits	12.5% mark density with less than 100 consecutive zeros for M consecutive pulse intervals, M = 175 bits
11/51	1	ANSI I.431	no pulse detected for N consecutive pulse intervals, N = 1544 bits	12.5% mark density with less than 100 consecutive zeros for M consecutive pulse intervals, M = 175 bits
E1	0	G.775	no pulse detected for N consecutive pulse intervals, N = 32 bits	12.5% mark density with less than 16 consecutive zeros for M consecutive pulse intervals, M = 32 bits
E1	1	ETSI 300233/ I.431	no pulse detected for N consecutive pulse intervals, N = 2048 bits	12.5% mark density with less than 16 consecutive zeros for M consecutive pulse intervals, M = 32 bits

#### Table-18 SLOS Criteria

Note:

1. System input ports are schmitt-trigger inputs)

#### 3.5.3.3 Transmit LOS (TLOS)

The amplitude and density of the data output on the transmit line side are monitored. When the amplitude of the data is less than a certain voltage for a certain period, TLOS is declared. The voltage is defined by the TALOS[1:0] bits (b3-2, LOS,...). The period is defined by the TDLOS[1:0] bits (b1-0, LOS,...). When a valid pulse is detected, i.e., the amplitude is above the setting in the TALOS[1:0] bits (b3-2, LOS,...), TLOS is cleared.

When TLOS is detected, the TLOS\_S bit (b2, STAT0,...) will be set. A transition from '0' to '1' on the TLOS\_S bit (b2, STAT0,...) or any transition (from '0' to '1' or from '1' to '0') on the TLOS\_S bit (b2, STAT0,...) will set the TLOS\_IS bit (b2, INTS0,...) to '1', as selected by the TLOS\_IES bit (b2, INTES,...). When the TLOS\_IS bit (b2, INTS0,...) is '1', an interrupt will be reported by INT if not masked by the TLOS\_IM bit (b2, INTM0,...).

TLOS may be counted by an internal Error Counter or may be indicated by the TMFn pin. Refer to Section 3.5.6 Error Counter and Section 3.5.7.2 TMFn Indication respectively.

TLOS can be used to monitor the LOS in the transmit line side between two channels. The connection between the two channels is shown in Figure-21. The two channels can be of the same device or different devices on the premises that the transmit line interfaces are in the same mode and at least the output of one channel is in High-Z state. Table-19 lists each results in this case. In the left two columns, the OE bit (b6, TCF0,...) of the two channels controls the output status in the transmit line side to ensure that at least one channel is in High-Z state. The middle two columns list the internal operation status. In the right two columns, the TLOS\_S bit (b2, STAT0,...) of the two channels indicates the TLOS status in the transmit line side.

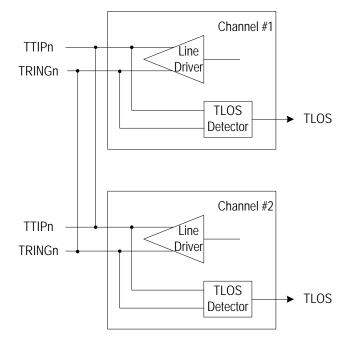


Figure-21 TLOS Detection Between Two Channels

Output Status ~ Con	trolled By the OE Bit	Internal Ope	ration Status	TLOS Status ~ Indicated By the TLOS_S E					
Channel #1	Channel #2	Channel #1	Channel #2	Channel #1	Channel #2				
Normal ~ 1	High-Z ~ 0	Normal	(don't-care)	No TLOS ~ 0	No TLOS ~ 0				
Normal ~ 1	High-Z ~ 0	Failure	Normal	TLOS Detected ~ 1 *	TLOS Detected ~ 1				
High-Z ~ 0	Normal ~ 1	(don't-care)	Normal	No TLOS ~ 0	No TLOS ~ 0				
High-Z ~ 0	Normal ~ 1	Normal	Failure	TLOS Detected ~ 1	TLOS Detected ~ 1 *				
High-Z ~ 0	High-Z ~ 0	(don't-care)	(don't-care)	TLOS Detected ~ 1	TLOS Detected ~ 1				
Note: * The TLOS_S bit (b2, STATO	Note: The TLOS_S bit (b2, STAT0,) may not be set if there is any catastrophic failure in the channel.								

#### Table-19 TLOS Detection Between Two Channels

#### 3.5.4 ALARM INDICATION SIGNAL (AIS) DETECTION AND GEN-ERATION

#### 3.5.4.1 Alarm Indication Signal (AIS) Detection

AIS is monitored in both the receive path and the transmit path.

When the mark density in the received data or in the data input from the transmit system side meets certain criteria, AIS is declared or cleared. In T1/J1 mode, the criteria are in compliance with ANSI T1.231. In E1 mode, the criteria are in compliance with ITU G.775 or ETSI 300233, as selected by the LAC bit (b7, LOS,...). Refer to Table-20 for details.

#### Table-20 AIS Criteria

	ITU G.775 for E1 (LAC = 0)	ETSI 300233 for E1 (LAC = 1)	ANSI T1.231 for T1 (LAC = 0 or 1)
AIS Declaring	Less than 3 zeros are received in each of two consecutive 512-bit data streams.	Less than 3 zeros are received in a 512-bit data stream.	Less than 9 zeros are received in a 8192-bit stream, i.e., less than 99.9% of marks in a period of 5.3 ms are received.
AIS Clearing	3 or more zeros are received in each of two consecutive 512-bit data streams.	3 or more zeros are received in a 512-bit data stream.	9 or more zeros are received in a 8192-bit data stream.

When AIS is detected in the receive path, the LAIS\_S bit (b6, STAT1,...) will be set. A transition from '0' to '1' on the LAIS\_S bit (b6, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the LAIS\_S bit (b6, STAT1,...) will set the LAIS\_IS bit (b6, INTS1,...) to '1', as selected by the AIS\_IES bit (b6, INTES,...). When the LAIS\_IS bit (b6, INTS1,...) is '1', an interrupt will be reported by INT if not masked by the LAIS\_IM bit (b6, INTM1,...).

When AIS is detected in the transmit path, the SAIS\_S bit (b7, STAT1,...) will be set. A transition from '0' to '1' on the SAIS\_S bit (b7, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the SAIS\_S bit (b7, STAT1,...) will set the SAIS\_IS bit (b7, INTS1,...) to '1', as selected by the AIS\_IES bit (b6, INTES,...). When the SAIS\_IS bit (b7, INTS1,...) is '1', an interrupt will be reported by INT if not masked by the SAIS\_IM bit (b7, INTM1,...).

AIS may be counted by an internal Error Counter or may be indicated by the RMFn or TMFn pin. Refer to Section 3.5.6 Error Counter and Section 3.5.7 Receive /Transmit Multiplex Function (RMF / TMF) Indication respectively.

#### 3.5.4.2 (Alarm Indication Signal) AIS Generation

AIS can be generated automatically in the receive path and the transmit path.

In the receive path, when the ASAIS\_LLOS bit (b2, AISG,...) is set, AIS will be generated automatically once LLOS is detected. When the ASAIS\_SLOS bit (b3, AISG,...) is set, AIS will be generated automatically once SLOS is detected. When AIS is generated, RDPn/RDNn output all '1's. RCLKn (if available) outputs XCLK.

In the transmit path, when the ALAIS\_LLOS bit (b0, AISG,...) is set, AIS will be generated automatically once LLOS is detected. When the ALAIS\_SLOS bit (b1, AISG,...) is set, AIS will be generated automatically once SLOS is detected. When AIS is generated, TTIPn/TRINGn output all '1's.

In the transmit path, the AIS transmission is controlled by the TXAIS bit (b4, AISG,...). When the TXAIS bit (b4, AISG,...) is set to '1', all '1's pattern is transmitted at TTIPn/TRINGn.

AIS generation uses XCLK<sup>1</sup> as reference clock.

If pattern (including PRBS, ARB and IB) is generated in the same direction, the priority of pattern generation is higher. The generated pattern will overwrite automatic AIS. Refer to Section 3.5.5.1 Pattern Generation for the output data and clock.

<sup>1.</sup> XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

# 3.5.5 PRBS, QRSS, ARB AND IB PATTERN GENERATION AND DETECTION

The pattern includes: Pseudo Random Bit Sequence (PRBS), Quasi-Random Signal Source (QRSS), Arbitrary Pattern (ARB) and Inband Loopback (IB).

#### 3.5.5.1 Pattern Generation

The pattern can be generated in the receive path or the transmit path, as selected by the PG\_POS bit (b3, PG,...).

The pattern to be generated is selected by the PG\_EN[1:0] bits (b5~4, PG,...).

If PRBS is selected, three kinds of PRBS patterns with maximum zero restriction according to ITU-T 0.151 and AT&T TR62411 are provided. They are:  $(2^20 - 1)$  QRSS per 0.150-4.5,  $(2^15 - 1)$  PRBS per 0.152 and  $(2^11 - 1)$  PRBS per 0.150, as selected by the PRBG\_SEL[1:0] bits (b1~0, PG,...).

If ARB is selected, the content is programmed in the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...).

If IB is selected, the IB generation is in compliance with ANSI T1.403. The length of the IB code can be 3 to 8 bits, as determined by the IBGL[1:0] bits ( $b5\sim4$ , IBL,...). The content is programmed in the IBG[7:0] bits ( $b7\sim0$ , IBG,...).

The selected pattern is transmitted repeatedly until the PG\_EN[1:0] bits (b5~4, PG,...) is set to '00'.

When pattern is generated in the receive path, the reference clock is XCLK or the recovered clock from the received signal, as selected by the PG\_CK bit (b6, PG,...). The selected reference clock is also output on RCLKn (if available).

When pattern is generated in the transmit path, the reference clock is XCLK<sup>1</sup> or the transmit clock, as selected by the PG\_CK bit (b6, PG,...). The transmit clock refers to the clock input on TCLKn (in Transmit Single Rail NRZ Format mode and in Transmit Dual Rail NRZ Format mode) or the clock recovered from the data input on TDPn and TDNn (in Transmit Dual Rail RZ Format mode).

In summary, do the followings step by step to generate pattern:

- Select the generation direction by the PG\_POS bit (b3, PG,...);
- Select the reference clock by the PG\_CK bit (b6, PG,...);
- Select the PRBS pattern by the PRBG\_SEL[1:0] bits (b1~0, PG,...) when PRBS is to be generated; program the ARB pattern in the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...) when ARB is to be generated; or set the length and the content of the IB code in the IBGL[1:0] bits (b5~4, IBL,...) and in the IBG[7:0] bits (b7~0, IBG,...) respectively when IB is to be generated;

• Set the PG\_EN[1:0] bits (b5~4, PG,...) to generate the pattern.

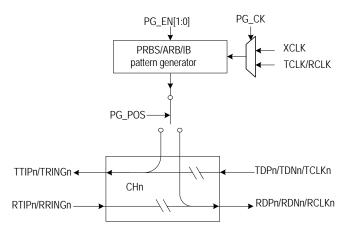
If PRBS or ARB is selected to be generated, the following two steps can be optionally implemented after the pattern is generated:

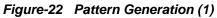
- Insert a single bit error by writing '1' to the ERR\_INS bit (b5, ERR,...);
- Invert the generated pattern by setting the PAG\_INV bit (b2, PG,...).

If pattern is generated in the receive path, the generated pattern should be encoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1) in Receive Dual Rail NRZ Format mode, Receive Dual Rail RZ Format mode and Receive Dual Rail Sliced mode. The encoding rule is selected by the R\_CODE bit (b2, RCF1,...).

If pattern is generated in the transmit path, the generated pattern should be encoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1). The encoding rule is selected by the T\_CODE bit (b2, TCF1,...).

The pattern generation is shown in Figure-22 and Figure-23.





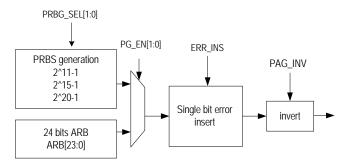


Figure-23 Pattern Generation (2)

The priority of pattern generation is higher than that of AIS generation. If they are generated in the same direction, the generated pattern will overwrite the generated AIS.

<sup>1.</sup> XCLK is derived from MCLK. It is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode.

#### 3.5.5.2 Pattern Detection

Data received from the line side or data input from the transmit system side may be extracted for pattern detection. The direction of data extraction is determined by the PD\_POS bit (b3, PD,...). One of PRBS or ARB pattern is selected for detection and IB detection is always active.

If data is extracted from the receive path, before pattern detection the data should be decoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1). The decoding rule is selected by the R\_CODE bit (b2, RCF1,...).

If data is extracted from the transmit path, before pattern detection the data should be decoded by using AMI or B8ZS (for T1/J1) / HDB3 (for E1) in Transmit Dual Rail NRZ Format mode and Transmit Dual Rail RZ Format mode. The decoding rule is selected by the T\_CODE bit (b2, TCF1,...).

#### <u>Pseudo Random Bit Sequence (PRBS) /Arbitrary Pattern (ARB)</u> <u>Detection</u>

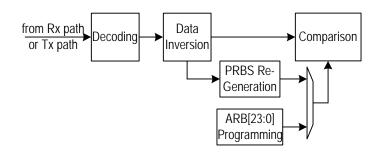
The extracted data can be optionally inverted by the PAD\_INV bit (b2, PD,...) before PRBS/ARB detection.

The extracted data is used to compare with the desired pattern. The desired pattern is re-generated from the extracted data if the desired pattern is (2^20 - 1) QRSS per 0.150-4.5, (2^15 - 1) PRBS per 0.152 or (2^11 - 1) PRBS per 0.150; or the desired pattern is programmed in the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...) if the desired pattern is ARB. The desired pattern is selected by the PAD\_SEL[1:0] bits (b1~0, PD,...).

In summary, do the followings step by step to detect PRBS/ARB:

- Select the detection direction by the PD\_POS bit (b3, PD,...);
- Set the ARB[23:0] bits (b7~0, ARBH~ARBM~ARBL,...) if the ARB pattern is desired - this step is omitted if the PRBS pattern is desired;
- Select the desired PRBS/ARB pattern by the PAD\_SEL[1:0] bits (b1~0, PD,...).

The priority of decoding, data inversion, pattern re-generation, bit programming and pattern comparison is shown in Figure-24.



#### Figure-24 PRBS / ARB Detection

During comparison, if the extracted data coincides with the re-generated PRBS pattern or the programmed ARB pattern for more than 64-bit hopping window, the pattern is synchronized and the PA\_S bit (b5, STAT1,...) will be set.

In synchronization state, if more than 6 PRBS/ARB errors are detected in a 64-bit hopping window, the pattern is out of synchronization and the PA\_S bit (b5, STAT1,...) will be cleared.

In synchronization state, each mismatched bit will generate a PRBS/ ARB error. When a PRBS/ARB error is detected during the synchronization, the ERR\_IS bit (b1, INTS2,...) will be set and an interrupt will be reported by INT if not masked by the ERR\_IM bit (b1, INTM2,...). The PRBS/ARB error may be counted by an internal Error Counter. Refer to Section 3.5.6 Error Counter.

A transition from '0' to '1' on the PA\_S bit (b5, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the PA\_S bit (b5, STAT1,...) will set the PA\_IS bit (b5, INTS1,...) to '1', as selected by the PA\_IES bit (b5, INTES,...). When the PA\_IS bit (b5, INTS1,...) is '1', an interrupt will be reported by  $\overline{INT}$  if not masked by the PA\_IM bit (b5, INTM1,...).

The PRBS/ARB synchronization status may be indicated by the RMFn or TMFn pin. Refer to Section 3.5.7 Receive /Transmit Multiplex Function (RMF / TMF) Indication.

#### Inband Loopback (IB) Detection

The IB detection is in compliance with ANSI T1.403.

The extracted data is used to compare with the target IB code. The length of the target activate/deactivate IB code can be 3 to 8 bits, as determined by the IBAL[1:0]/IBDL[1:0] bits (b3~2/b1~0, IBL,...). The content of the target activate/deactivate IB code is programmed in the IBA[7:0]/IBD[7:0] bits (b7~0, IBDA/IBDD,...). Refer to Figure-25.

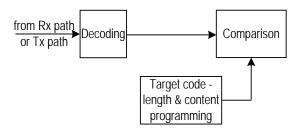


Figure-25 IB Detection

During comparison, if the extracted data coincides with the target activate/deactivate IB code with no more than  $10^{-2}$  bit error rate for a certain period, the IB code is detected. The period depends on the setting of the AUTOLP bit (b3, LOOP,...).

If the AUTOLP bit (b3, LOOP,...) is '0', Automatic Digital/Remote Loopback is disabled. In this case, when the activate IB code is detected for more than 40 ms, the IBA\_S bit (b1, STAT1,...) will be set to indicate the activate IB code detection; when the deactivate IB code is detected for more than 40 ms (T1/J1 mode) / 30 ms (E1 mode), the IBD\_S bit (b0, STAT1,...) will be set to indicate the deactivate IB code detection.

If the AUTOLP bit (b3, LOOP,...) is '1', Automatic Digital/Remote Loopback is enabled. In this case, when the activate IB code is detected for more than 5.1 seconds, the IBA\_S bit (b1, STAT1,...) will be set to indicate the activate IB code detection. The detection of the activate IB code in the receive path will activate Remote Loopback or the detection of the activate IB code in the transmit path will activate Digital Loopback (refer to Section 3.5.8.2 Remote Loopback & Section 3.5.8.3 Digital Loopback). When the deactivate IB code is detected for more than 5.1 seconds, the IBD\_S bit (b0, STAT1,...) will be set to indicate the deactivate IB code detection. The detection of the deactivate IB code in the receive path will deactivate Remote Loopback or the detection of the deactivate IB code in the transmit path will deactivate Digital Loopback (refer to Section 3.5.8.2 Remote Loopback & Section 3.5.8.3 Digital Loopback). When the transmit path will deactivate IB code in the receive path will deactivate Remote Loopback or the detection of the deactivate IB code in the transmit path will deactivate Digital Loopback (refer to Section 3.5.8.2 Remote Loopback & Section 3.5.8.3 Digital Loopback).

A transition from '0' to '1' on the IBA\_S/IBD\_S bit (b1/b0, STAT1,...) or any transition (from '0' to '1' or from '1' to '0') on the IBA\_S/IBD\_S bit (b1/b0, STAT1,...) will set the IBA\_IS/IBD\_IS bit (b1/b0, INTS1,...) to '1'

respectively, as selected by the IB\_IES bit (b0, INTES,...). When the IBA\_IS/IBD\_IS bit (b1/b0, INTS1,...) is '1', an interrupt will be reported on INT if not masked by the IBA\_IM/IBD\_IM bit (b1/b0, INTM1,...).

#### 3.5.6 ERROR COUNTER

An internal 16-bit Error Counter is used to count one of the following errors:

- LBPV: BPV/CV detected in the receive path (line side);
- LEXZ: EXZ detected in the receive path (line side);
- LBPV + LEXZ: BPV/CV and EXZ detected in the receive path (line side);
- SBPV: BPV/CV detected in the transmit path (system side) (disabled in Transmit Single Rail NRZ Format mode);
- SEXZ: EXZ detected in the transmit path (system side);
- SBPV + SEXZ: BPV/CV and EXZ detected in the transmit path (system side) (disabled in Transmit Single Rail NRZ Format mode);
- PRBS/ARB error.

The CNT\_SEL[2:0] bits (b4~2, ERR,...) select one of the above errors to be counted.

The Error Counter is buffered. It is updated automatically or manually, as determined by the CNT\_MD bit (b1, ERR,...).

The Error Counter is accessed by reading the ERRCH and ERRCL registers.

#### 3.5.6.1 Automatic Error Counter Updating

When the CNT\_MD bit (b1, ERR,...) is '1', the Error Counter is updated every one second automatically.

The one-second timer uses MCLK as clock reference. The expiration of each one second will set the TMOV\_IS bit (b0, INTTM) and induce an interrupt reported by  $\overline{INT}$  if not masked by the TMOV\_IM bit (b0, GCF).

When each one second expires, the Error Counter transfers the accumulated error numbers to the ERRCH and ERRCL registers and the Error Counter will be cleared to start a new round counting. The ERRCH and ERRCL registers should be read in the next second, otherwise they will be overwritten.

When the ERRCH and ERRCL registers are all '1's and there is still error to be accumulated, the registers will be overflowed. The overflow is indicated by the CNTOV\_IS bit (b0, INTS2,...) and will induce an interrupt reported by INT if not masked by the CNTOV\_IM (b0, INTM2,...).

The process of automatic Error Counter updating is illustrated in Figure-26.

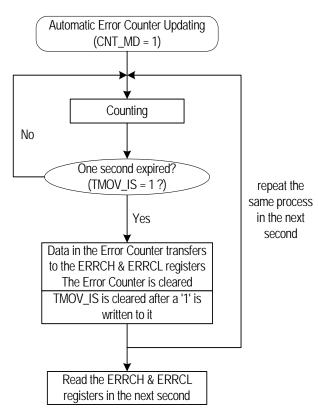


Figure-26 Automatic Error Counter Updating

#### 3.5.6.2 Manual Error Counter Updating

When the CNT\_MD bit (b1, ERR,...) is '0', the Error Counter is updated manually.

When there is a transition from '0' to '1' on the CNT\_STOP bit (b0, ERR,...), the Error Counter transfers the accumulated error numbers to the ERRCH and ERRCL registers and the Error Counter will be cleared to start a new round counting. The ERRCH and ERRCL registers should be read in the next round of error counting, otherwise they will be over-written.

When the ERRCH and ERRCL registers are all '1's and there is still error to be accumulated, the registers will be overflowed. The overflow is indicated by the CNTOV\_IS bit (b0, INTS2,...) and will induce an interrupt reported by INT if not masked by the CNTOV\_IM (b0, INTM2,...).

The process of manual Error Counter updating is illustrated in Figure-27.

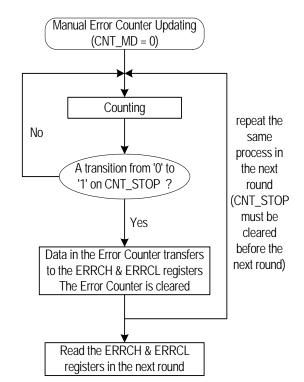


Figure-27 Manual Error Counter Updating

# 3.5.7 RECEIVE /TRANSMIT MULTIPLEX FUNCTION (RMF / TMF) INDICATION

#### 3.5.7.1 RMFn Indication

In Receive Single Rail NRZ Format mode, the RDNn/RMFn pin is used as RMFn. In Receive Dual Rail Sliced mode, the RCLKn/RMFn pin is used as RMFn. Refer to Table-3 Multiplex Pin Used in Receive System Interface for details.

#### Table-21 RMFn Indication

RMFn can indicate the status of PRBS/ARB, LAIS, LEXZ, LBPV, LEXZ + LBPV, LLOS, output recovered clock (RCLK) or XOR output of positive and negative sliced data, as selected by the RMF\_DEF[2:0] bits (b7~5, RCF1,...). Refer to Table-21 for details.

RMF_DEF[2:0]	Indication On RMF	Details
000	PRBS/ARB	RMFn is high if PRBS/ARB is detected in synchronization in the receive path. During the synchronization, RMFn goes low for a T1/E1 clock cycle if a PRBS/ARB error is detected. RMFn is low if PRBS/ARB is out of synchronization. Refer to Section 3.5.5 PRBS, QRSS, ARB and IB Pattern Generation and Detection.
001	Line Alarm Indication Signal (LAIS)	RMFn is high if AIS is detected in the receive path and low if it is cleared. This indication corresponds to the LAIS_S bit (b6, STAT1,). Refer to Section 3.5.4 Alarm Indication Signal (AIS) Detection and Generation.
010	XOR result of positive and negative sliced data	RMFn outputs XOR data of positive and negative sliced data.
011	recovered clock (RCLK)	RMFn outputs the recovered clock as RCLKn. All the description about RCLKn is applicable for RMFn.
100	Line Excessive Zeroes (LEXZ)	RMFn goes high for a T1/E1 clock cycle if an EXZ is detected in the receive path, otherwise it is low. Refer to Section 3.5.2 Excessive Zeroes (EXZ) Detection.
101	Line Bipolar Violation (LBPV)	RMFn goes high for a T1/E1 clock cycle if a BPV/CV is detected in the receive path, otherwise it is low. Refer to Section 3.5.1 Bipolar Violation (BPV) / Code Violation (CV) Detection and BPV Insertion.
110	LEXZ + LBPV	RMFn goes high for a T1/E1 clock cycle if an EXZ or a BPV/CV is detected in the receive path, otherwise it is low.
111	Line Loss of Signal (LLOS)	RMFn is high if LOS is detected in the receive path and low if it is cleared. This indication corresponds to the LLOS_S bit (b0, STAT0,). Refer to Section 3.5.3.1 Line LOS (LLOS).

#### 3.5.7.2 TMFn Indication

In Transmit Single Rail NRZ Format mode and Transmit Dual Rail RZ Format mode, the TDNn/TMFn pin is used as TMFn. Refer to Table-4 Multiplex Pin Used in Transmit System Interface for details.

TMFn can indicate the status of PRBS/ARB, SAIS, TOC, TLOS, SEXZ, SBPV, SEXZ + SBPV or SLOS, as selected by the TMF\_DEF[2:0] bits (b7~5, TCF1,...). However, the indication of SBPV, SEXZ + SBPV and SLOS is disabled automatically in Transmit Single Rail NRZ Format mode. Refer to Table-22 for details.

#### Table-22 TMFn Indication

TMF_DEF[2:0]	Indication On TMF	Details
000	PRBS/ARB	TMFn is high if PRBS/ARB is detected in synchronization in the transmit path. During the synchronization, TMFn goes low for a T1/E1 clock cycle if a PRBS/ARB error is detected. TMFn is low if PRBS/ARB is out of synchronization.
001	System Alarm Indication Signal (SAIS)	TMFn is high if AIS is detected in the transmit path and low if it is cleared. This indication corresponds to the SAIS_S bit (b7, STAT1,). Refer to Section 3.5.4 Alarm Indication Signal (AIS) Detection and Generation.
010	Transmit Over Current (TOC)	TMFn is high if transmit over current is detected and low if it is cleared. This indication corresponds to the TOC_S bit (b4, STAT0,). Refer to Section 3.3.5.1 Transmit Over Current Protection.
011	Transmit Loss of Signal (TLOS)	TMFn is high if LOS is detected in the transmit line side and low if it is cleared. This indication corresponds to the TLOS_S bit (b2, STAT0,). Refer to Section 3.5.3.3 Transmit LOS (TLOS).
100	System Excessive Zeroes (SEXZ)	TMFn goes high for a T1/E1 clock cycle if an EXZ is detected in the transmit path, otherwise it is low. Refer to Section 3.5.2 Excessive Zeroes (EXZ) Detection
101	System Bipolar Violation (SBPV) *	TMFn goes high for a T1/E1 clock cycle if a BPV/CV is detected in the transmit path, otherwise it is low. Refer to Section 3.5.1 Bipolar Violation (BPV) / Code Violation (CV) Detection and BPV Insertion.
110	System Excessive Zeroes (SEXZ) + System Bipolar Violation (SBPV) *	TMFn goes high for a T1/E1 clock cycle if an EXZ or a BPV/CV is detected in the transmit path, otherwise it is low.
111	System Loss of Signal (SLOS) *	TMFn is high if LOS is detected in the transmit system side and low if it is cleared. This indication corresponds to the SLOS_S bit (b1, STAT0,). Refer to Section 3.5.3.2 System LOS (SLOS).
Note: * In Transmit Single	Rail NRZ Format mode, the co	rresponding indication is disabled and the corresponding setting is reserved.

#### 3.5.8 LOOPBACK

There are four kinds of loopback:

- Analog Loopback
- Remote Loopback
- Digital Loopback
- Dual Loopback

Refer to Figure-1 for loopback location.

#### 3.5.8.1 Analog Loopback

Analog Loopback is enabled by the ALP bit (b0, LOOP,...). The data stream to be transmitted on the TTIPn/TRINGn pins is internally looped to the RTIPn/RRINGn pins.

In Analog Loopback mode, the data stream to be transmitted is still output to the line side, while the data stream received from the line side is covered by the Analog Loopback data.

Anytime when Analog Loopback is set, the other loopbacks (i.e., Digital Loopback and Remote Loopback) are disabled.

In Analog Loopback, the priority of the diagnostic facilities in the receive path is: pattern generation > looped data. AIS generation is disabled in both the receive path and the transmit path. Refer to Figure-28.

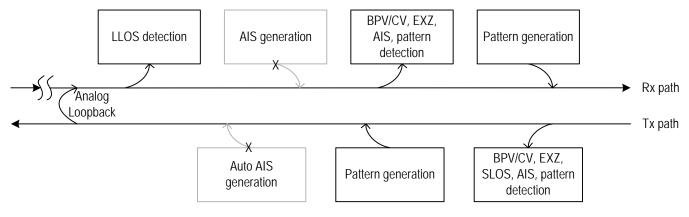


Figure-28 Priority Of Diagnostic Facilities During Analog Loopback

#### 3.5.8.2 Remote Loopback

Remote Loopback can be configured manually or automatically. Either manual Remote Loopback configuration or automatic Remote Loopback configuration will enable Remote Loopback.

Manual Remote Loopback is enabled by the RLP bit (b1, LOOP,...).

Automatic Remote Loopback is enabled when the pattern detection is assigned in the receive path (i.e., the PD\_POS bit (b3, PD,...) is '0') and the AUTOLP bit (b3, LOOP,...) is '1'. The corresponding channel will enter Remote Loopback when the activate IB code is detected in the receive path for more than 5.1 sec.; and will return from Remote Loopback when the deactivate IB code is detected in the receive path for more than 5.1 sec. Refer to section Inband Loopback (IB) Detection on page 43 for details. When automatic Remote Loopback is active, setting the AUTOLP bit (b3, LOOP,...) back to '0' will also stop automatic Remote Loopback. The setting of the PD\_POS bit (b3, PD,...) should not be changed during automatic Remote Loopback. The AUTOLP\_S bit (b7, STAT0,...) indicates the automatic Remote Loopback status.

In Remote Loopback mode, the data stream output from the RJA (if enabled) is internally looped to the Waveform Shaper. The data stream received from the line side is still output to the system side, while the data stream input from the system side is covered by the Remote Loopback data and the status on TCLKn does not affect the Remote Loopback. However, the BPV/CV, EXZ, SLOS, AIS and pattern detection in the transmit path still monitors the data stream input from the system side.

In Remote Loopback mode, the priority of the diagnostic facilities in the receive path is: pattern generation > AIS generation; the priority of the diagnostic facilities in the transmit path is: pattern generation > looped data. AIS generation is disabled in the transmit path. Refer to Figure-29.

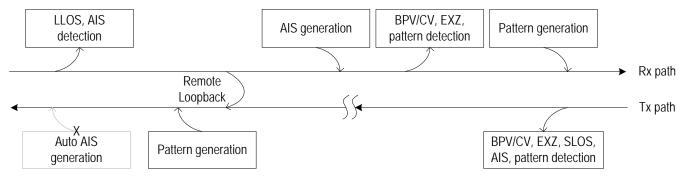


Figure-29 Priority Of Diagnostic Facilities During Manual Remote Loopback

#### 3.5.8.3 Digital Loopback

The Digital Loopback can be configured manually or automatically. Either manual Digital Loopback configuration or automatic Digital Loopback configuration will enable Digital Loopback.

Manual Digital Loopback is enabled by the DLP bit (b2, LOOP,...).

Automatic Digital Loopback is enabled when the pattern detection is assigned in the transmit path (i.e., the PD\_POS bit (b3, PD,...) is '1') and the AUTOLP bit (b3, LOOP,...) is '1'. The corresponding channel will enter Digital Loopback when the activate IB code is detected in the transmit path for more than 5.1 sec.; and will return from Digital Loopback when the deactivate IB code is detected in the transmit path for more than 5.1 sec. Refer to section Inband Loopback (IB) Detection on page 43 for details. When automatic Digital Loopback is active, setting the AUTOLP bit (b3, LOOP,...) back to '0' will also stop automatic Digital

Loopback. The setting of the PD\_POS bit (b3, PD,...) should not be changed during automatic Digital Loopback. The AUTOLP\_S bit (b7, STAT0,...) indicates the automatic Digital Loopback status.

In Digital Loopback mode, the data stream output from the TJA (if enabled) is internally looped to the Decoder (if enabled). The data stream to be transmitted is still output to the line side, while the data stream received from the line side is covered by the Digital Loopback data. However, LLOS and AIS detection in the receive path still monitors the data stream received from the line side.

In Digital Loopback mode, the priority of the diagnostic facilities in the receive path is: pattern generation > looped data; the priority of the diagnostic facilities in the transmit path is: pattern generation > looped data > AIS generation. AIS generation is disabled in the receive path.

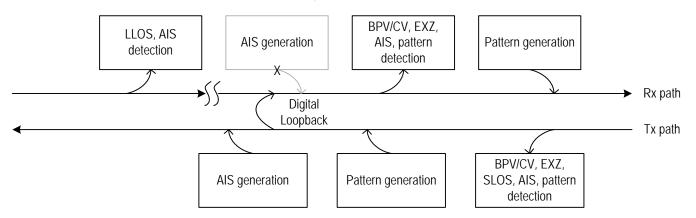


Figure-30 Priority Of Diagnostic Facilities During Digital Loopback

#### 3.5.8.4 Dual Loopback

Dual Loopback refers to the simultaneous implementation of Remote Loopback and Digital Loopback. Two kinds of combinations are supported:

- · Manual Remote Loopback + Manual Digital Loopback;
- Manual Remote Loopback + Automatic Digital Loopback.

Note that when Digital Loopback is active, automatic Remote Loopback is unavailable as the pattern detection is within the digital loop.

In Dual Loopback mode, the data stream received from the line side outputs from the RJA (if enabled), loops to the Waveform Shaper internally and does not output to the system side. The data stream to be transmitted from the system side outputs from the TJA (if enabled), loops to the Decoder (if enabled) internally and does not output to the line side. LLOS, AIS detection in the receive path monitors the data stream received from the line side. The BPV/CV, EXZ and pattern detection in the receive path monitors the digital looped data. The BPV/CV, EXZ, SLOS, AIS and pattern detection in the transmit path monitors the data stream input from the system side.

#### Manual Remote Loopback + Manual Digital Loopback

This combination of Dual Loopback is enabled when both manual Remote Loopback and manual Digital Loopback are enabled. Manual Remote Loopback is enabled by the RLP bit (b1, LOOP,...). Manual Digital Loopback is enabled by the DLP bit (b2, LOOP,...).

In this condition, the priority of the diagnostic facilities in the receive path is: pattern generation > digital looped data; the priority of the diagnostic facilities in the transmit path is: remote looped data > pattern generation. AIS generation is disabled in both the receive path and the transmit path.

Refer to Figure-31.

#### Manual Remote Loopback + Automatic Digital Loopback

This combination of Dual Loopback is enabled when both manual Remote Loopback and automatic Digital Loopback are enabled. Manual Remote Loopback is enabled by the RLP bit (b1, LOOP,...). Automatic Digital Loopback is enabled when the pattern detection is assigned in the transmit path (i.e., the PD\_POS bit (b3, PD,...) is '1') and the AUTOLP bit (b3, LOOP,...) is '1'. The corresponding channel will enter Digital Loopback when the activate IB code is detected in the transmit path for more than 5.1 sec.; and will return from Digital Loopback when the deactivate IB code is detected in the transmit path for more than 5.1 sec. Refer to section Inband Loopback (IB) Detection on page 43 for details. When automatic Digital Loopback is active, setting the AUTOLP bit (b3, LOOP,...) back to '0' will also stop automatic Digital Loopback. The setting of the PD\_POS bit (b3, PD,...) should not be changed during automatic Digital Loopback. The AUTOLP\_S bit (b7, STAT0,...) indicates the automatic Digital Loopback status.

In this condition, the priority of the diagnostic facilities in the receive path is: pattern generation > digital looped data. AIS generation in both the receive path and the transmit path, the pattern generation in the transmit path are disabled.

Refer to Figure-32.

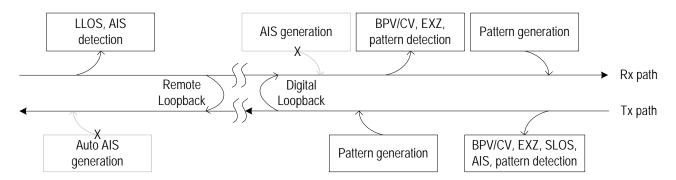


Figure-31 Priority Of Diagnostic Facilities During Manual Remote Loopback + Manual Digital Loopback

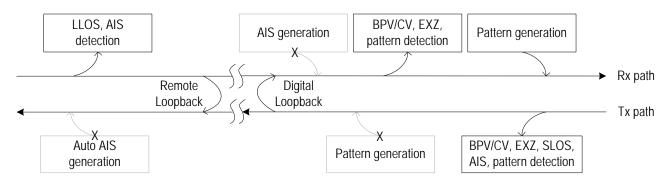


Figure-32 Priority Of Diagnostic Facilities During Manual Remote Loopback + Automatic Digital Loopback

#### 3.5.9 CHANNEL 0 MONITORING

Channel 0 can be used as a monitoring channel when not used as a regular channel. Channel 0 support G.772 Monitoring and Jitter Measurements.

#### 3.5.9.1 G.772 Monitoring

Selected by the MON[5:0] bits (b5~0, MON), any receiver or transmitter of the other 15 channels can be monitored by channel 0 (as shown in Figure-33).

When the G.772 Monitoring is implemented (the MON[5:0] bits (b5~0, MON) is not '0'), the registers of the receiver of channel 0 should be the same as those of the selected receiver /transmitter except the line interface related registers.

Once the G.772 Monitoring is implemented, the receiver of channel 0 switches to External Impedance Matching mode automatically, and the setting in the R\_TERM[2:0] bits (b2~0, RCF0,...) of channel 0 is ignored.

During the G.772 Monitoring, channel 0 processes as normal after data is received from the selected path and the operation of the monitored path is not effected.

The signal which is monitored goes through the Clock & Data Recovery of monitoring channel, (channel 0). The monitored clock can output on RCLK0. The monitored data can be observed digitally on the output pin of RCLK0, RD0/RDP0 and RDN0. LOS detector is still in use in channel 0 for the monitored signal.

In monitoring mode, channel 0 can be configured to Remote Loopback. The signal which is being monitored will output on TTIP0 and TRING0. The output signal can then be connected to a standard test equipment for non-intrusive monitoring.

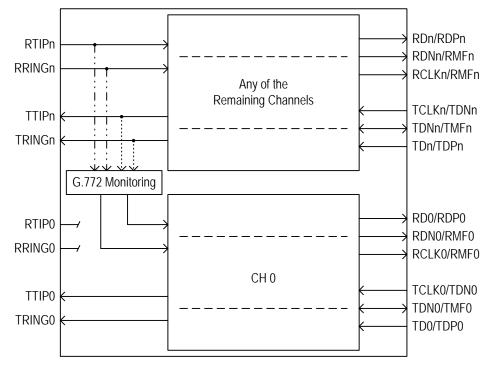


Figure-33 G.772 Monitoring

#### 3.5.9.2 Jitter Measurement (JM)

The RJA of channel 0 consists of a Jitter Measurement (JM) module. When the RJA is enabled in channel 0, the JM is used to measure the positive and negative peak value of the demodulated jitter signal of the received data stream. The bandwidth of the measured jitter is selected by the JM\_BW bit (b0, JM).

The greatest positive peak value monitored in a certain period is indicated by the JIT\_PH and JIT\_PL registers, while the greatest negative peak value monitored in the same period is indicated by the JIT\_NH and JIT\_NL registers. The relationship between the greatest positive /negative peak value and the indication in the corresponding registers is:

Positive Peak = [JIT\_PH, JIT\_PL] / 16 (UIpp);

Negative Peak = [JIT\_NH, JIT\_NL] / 16 (UIpp).

The period is determined by the JM\_MD bit (b1, JM).

When the JM\_MD bit (b1, JM) is '1', the period is one second automatically. The one-second timer uses MCLK as clock reference. The expiration of each one second will set the TMOV\_IS bit (b0, INTTM) and induce an interrupt reported by INT if not masked by the TMOV\_IM bit (b0, GCF). The TMOV\_IS bit (b0, INTTM) is cleared after a '1' is written to this bit. When each one second expires, internal buffers transfer the greatest positive/negative peak value accumulated in this one second to the JIT\_PH and JIT\_PL / JIT\_NH and JIT\_NL registers respectively and the internal buffers will be cleared to start a new round measurement. The registers should be read in the next second, otherwise they will be overwritten. Refer to Figure-34 for the process.

When the JM\_MD bit (b1, JM) is '0', the period is controlled by the JM\_STOP bit (b2, JM) manually. When there is a transition from '0' to '1' on the JM\_STOP bit (b2, JM), the internal buffers transfer the greatest positive/negative peak value accumulated in this period to the JIT\_PH and JIT\_PL / JIT\_NH and JIT\_NL registers respectively and the internal buffers will be cleared to start a new round measurement. The registers should be read in the next round of jitter measurement, otherwise they will be overwritten. Refer to Figure-35 for the process.

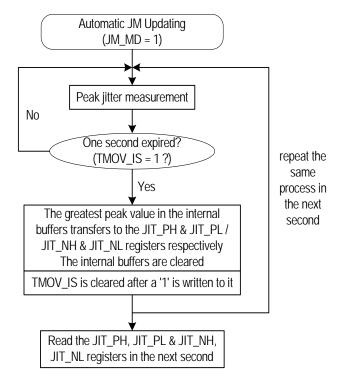


Figure-34 Automatic JM Updating

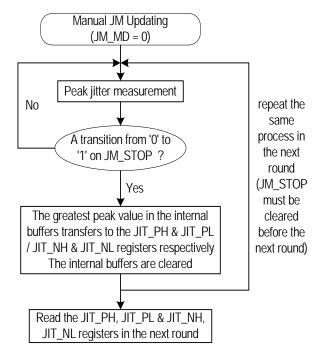


Figure-35 Manual JM Updating

# 3.6 CLOCK INPUTS AND OUTPUTS

The IDT82P2916 provides two kinds of clock outputs:

- Free running clock outputs on CLKT1 and CLKE1
- Receiver clock outputs on REFA and REFB
  - selected from any of the 16 recovered line clocks
  - driven by MCLK (free running)
  - driven by external CLKA/CLKB input

A Frequency Synthesizer is also available to scale REFA to 8 different frequencies.

The following Clock Inputs are provided:

- MCLK as programmable reference timing for the IDT82P2916.
- CLKA and CLKB as optional input clock source for REFA and REFB respectively

#### 3.6.1 FREE RUNNING CLOCK OUTPUTS ON CLKT1/CLKE1

An internal clock generator uses MCLK as reference to generate all the clocks required by internal circuits and CLKT1/CLKE1 outputs. MCLK should be a clock with +/-32 ppm (in T1/J1 mode) or +/-50 ppm (in E1 mode) accuracy. The clock frequency of MCLK is 1.544/2.048 X N MHz ( $1 \le N \le 8$ , N is an integer number), as determined by MCKSEL[3:0]. Refer to Chapter 2 Pin Description for details.

The outputs on CLKT1 and CLKE1 are free running (locking to MCLK). The output of CLKT1 is determined by the CLKT1\_EN bit (b1, CLKG) and the CLKT1 bit (b0, CLKG). Refer to Table-23. The output of CLKE1 is determined by the CLKE1\_EN bit (b3, CLKG) and the CLKE1 bit (b2, CLKG). Refer to Table-24.

#### Table-23 Clock Output on CLKT1

Contro	ol Bits	Clock Output On CLKT1
CLKT1_EN	CLKT1	olock output on other in
0	(don't-care)	High-Z
1	0	8 KHz
	1	1.544 KHz

#### Table-24 Clock Output on CLKE1

Contro	ol Bits	Clock Output On CLKE1
CLKE1_EN	CLKE1	Clock Output On CLKE1 High-Z 8 KHz
0	(don't-care)	High-Z
1	0	8 KHz
I	1	2.048 KHz

#### 3.6.2 CLOCK OUTPUTS ON REFA/REFB

The outputs on REFA and REFB can be enabled or disabled, as determined by the REFA\_EN bit (b6, REFA) and the REFB\_EN bit (b6, REFB) respectively.

When the output is disabled, REFA/REFB is in High-Z state.

When the output is enabled, the output of REFA and REFB varies in different operations. Refer to below for detailed description. Refer to Figure-36 and Figure-37 for an overview of REFA and REFB output options in normal operation.

#### 3.6.2.1 REFA/REFB in Clock Recovery Mode

In this mode (default), the clock of REFA and REFB is derived from the recovered clock of one of the 16 channels as selected by the REFA[4:0] bits (b4~0,REFA) and REFB[4:0] bits (b4~0,REFB). Determined by the FS\_BYPAS bit (b4, REFCF) a Frequency Synthesizer can be enabled for REFA (refer to Section 3.6.2.2 Frequency Synthesizer for REFA Clock Output). If the Frequency Synthesizer is disabled, REFA will output the recovered 1.544 MHz (T1) or 2.048 MHz (E1) clock depending on the line mode of the selected channel. REFB output the recovered 1.544 MHz (T1) clock depending on the line mode of the selected channel.

The recovered line clock can be output to REFA and REFB before or after it passed the receive Jitter Attenuator (RJA) selected by the JA\_BYPAS bit (b6, REFCF).

#### 3.6.2.2 Frequency Synthesizer for REFA Clock Output

For REFA a Frequency Synthesizer can be enabled or bypassed (default) as selected by FS\_BYPASS bit (b4, REFCF). The output frequency is selected by the FREQ[2:0] bits (b2~0, REFCF). Frequencies supported are 8 KHz, 64 KHz, 2.048 MHz, 4.096 MHz, 8.192 MHz, 19.44 MHz or 32.768 MHz.

#### 3.6.2.3 Free Run Mode for REFA Clock Output

REFA can also be selected to provide a free running clock locked to MCLK. To enable this mode the Frequency Synthesizer has to be enabled by setting the FS\_BYPAS bit (b4, REFCF) to '0', and the FREE bit (b3, REFCF) has to be set to '1'. REFA will provide a frequency selected by the FREQ[2:0]<sup>1</sup> bits (b2~0, REFCF) which is a free running clock locked to MCLK.

#### 3.6.2.4 REFA/REFB Driven by External CLKA/CLKB Input

In this mode, the clock of REFA and REFB is driven from an external clock input of CLKA and CLKB respectively. CLKA and CLKB are selected as an input source by setting REFA[4:0] bits (b4~0, REFA) and REFB[4:0] bits (b4~0, REFB) to any value from '11101' to '11111'.

CLKA and CLKB are an external T1/J1 (1.544 MHZ) or E1 (2.048 MHz) Clock Input. The CKA\_T1E1 bit (b5, REFA) and CKB\_T1E1 bit (b5, REFB) should be set to match the input clock frequency.

Determined by the FS\_BYPASS bit (b4, REFCF), a Frequency Synthesizer can be enabled for REFA (refer to Section 3.6.2.2 Frequency Synthesizer for REFA Clock Output). If the Frequency Synthesizer is disabled, REFA will output the 1.544 MHz (T1) or 2.048 MHz (E1) clock depending the CLKA input clock. REFB will output 1.544 MHz (T1) or 2.048 MHz (E1) depending on the CLKB input clock.

# 3.6.2.5 REFA and REFB in Loss of Signal (LOS) or Loss of Clock Condition

If the recovered clock of one of the 16 channels is selected as the clock source for REFA and REFB (refer to Section 3.6.2.1 REFA/REFB in Clock Recovery Mode) and Line LOS (LLOS) is detected in the corresponding channel, the state of output on REFA and REFB can be selected by the REFH bit (b5, REFCF). If REFH is set to '1', REFA and REFB will output a high level in case of LLOS. If REFH is set to '0' and LLOS is detected, REFA and REFB clock outputs will be locked to MCLK while the selected clock frequency will remain unchanged.

LLOS condition is set when LLOS\_S bit (b0, STAT0) is '1'. Refer to Section 3.5.3.1 Line LOS (LLOS).

Refer to Figure-38 for a detailed overview of REFA output in case of LLOS. REFB output option is only determined by the REFH bit (b5, REFCF) to be locked to MCLK or set to high level output.

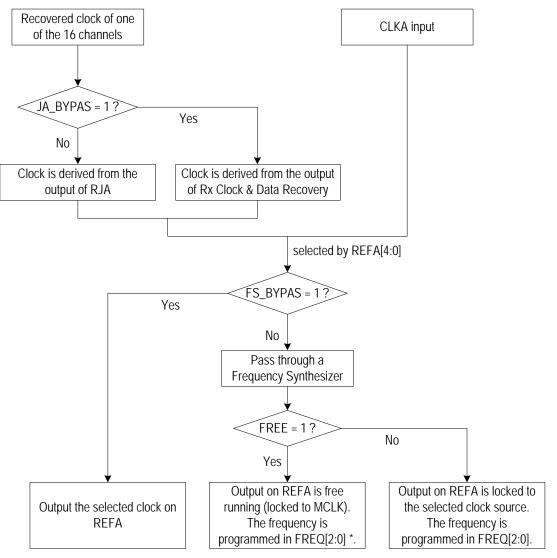
If CLKA is selected as the clock source for REFA (refer to Section 3.6.2.4 REFA/REFB Driven by External CLKA/CLKB Input) and there is no clock input on CLKA for more than 8 T1 clock cycles if T1 mode is selected (i.e. CKA\_T1E1 bit (b5, REFA) is '0') or more than 8 E1 clock cycles if E1 mode is selected (i.e. CKA\_T1E1 bit (b5, REFA) is '1'), the state of the REFA output is determined by the FS\_BYPAS bit (b4, REFCF) and the FREE bit (b3, REFCF). In case the Frequency Synthesizer is disabled (i.e. FS\_BYPAS bit (b4, REFCF) is '0'). REFA will output a high level. If the Frequency Synthesizer is enabled and the FREE bit (b3, REFCF) is set to '0', REFA will output a high level. If the Frequency Synthesizer is enabled and the FREE bit (b3, REFCF) is set to '1', REFA will be locked to MCLK.

Refer to Figure-39 for a detailed overview of REFA output in case of loss of CLKA.

If CLKB is selected as the clock source for REFB (refer to section Section 3.6.2.4 REFA/REFB Driven by External CLKA/CLKB Input) and there is no clock input on CLKB for more than 8 T1 clock cycles if T1 mode is selected (i.e. CKB\_T1E1 bit (b5, REFB) is '0') or more than 8 E1 clock cycles if E1 mode is selected (i.e. CKB\_T1E1 bit (b5, REFB) is '1'), the output on REFB is determined by the REFH bit (b5, REFCF). If REFH is set to '1', REFB will output a high level. If REFH is set to '0', the REFB clock output will be locked to MCLK.

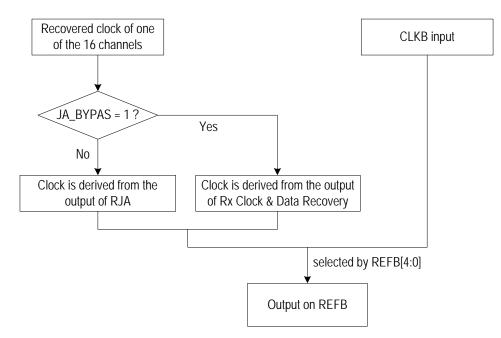
<sup>1. &#</sup>x27;000' and '011' are reserved for FREQ[2:0] in this mode.



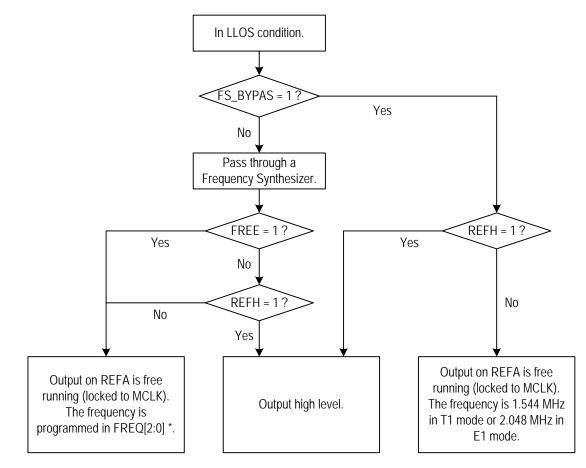


Note \*: '000' and '011' are reserved for FREQ[2:0] when REFA is free running.

#### Figure-36 REFA Output Options in Normal Operation

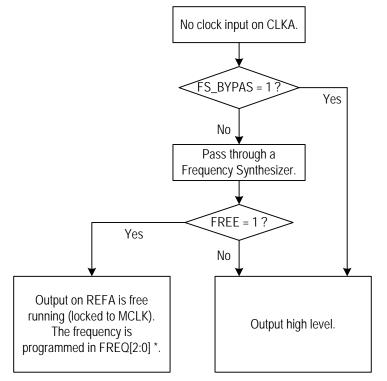






Note \*: '000' and '011' are reserved for FREQ[2:0] when REFA is free running.

## Figure-38 REFA Output in LLOS Condition (When RCLKn Is Selected)



Note \*: '000' and '011' are reserved for FREQ[2:0] when REFA is free running.

#### Figure-39 REFA Output in No CLKA Condition (When CLKA Is Selected)

#### 3.6.3 MCLK, MASTER CLOCK INPUT

MCLK provides a stable reference timing for the IDT82P2916. MCLK should be a clock with +/-32 ppm (in T1/J1 mode) or +/-50 ppm (in E1 mode) accuracy. The clock frequency of MCLK is set by pins MCKSEL[3:0] and can be N x 1.544 MHz or N x 2.048 MHz with  $1 \le N \le 8$  (N is an integer number). Refer to MCKSEL[3:0] pin description for details.

If there is a loss of MCLK (duty cycle is less than 30% for 10  $\mu$ s), the device will enter power down. In this case, both the receive and transmit circuits are turned off. The pins on the line interface will be in High-Z state. The pins on receive system interface will be in High-Z state or in low level, as selected by the RHZ bit (b6, RCF0,...). The input on the

transmit system interface is ignored and the output on the transmit system interface will be in High-Z state. Refer to Section 3.2.7 Receiver Power Down and Section 3.3.7 Transmitter Power Down for details.

If MCLK recovers after loss of MCLK the device will be reset automatically.

#### 3.6.4 XCLK, INTERNAL REFERENCE CLOCK INPUT

XCLK is derived from MCLK. For the respective channel, it is 1.544 MHz in T1/J1 mode or 2.048 MHz in E1 mode. XCLK is used as selectable reference clock for

- pattern /AIS generation
- RCLKn in LLOS
- Loss of TCLKn to determine Transmit Output High-Z.

## 3.7 INTERRUPT SUMMARY

There are altogether 20 kinds of interrupt sources as listed in Table-25. Among them, No.1 to No.19 are per-channel interrupt sources, while No. 20 is a global interrupt source.

For interrupt sources from No.1 to No.10, the occurrence of the event will cause the corresponding Status bit to be set to '1'. And selected by the Interrupt Trigger Edges Select bit, either a transition from '0' to '1' or any transition from '0' to '1' or from '1' to '0' of the Status bit will cause the Interrupt Status bit to be set to '1', which indicates the occurrence of an interrupt event.

For interrupt sources from No.11 to No.20, the occurrence of the event will cause the corresponding Interrupt Status Bit to be set to '1'.

All the interrupt can be masked by the GLB\_IM bit (b1, GCF) globally or by the corresponding interrupt mask bit individually. For all the interrupt sources, if not masked, the occurrence of the interrupt event will trigger an interrupt indicated by the INT pin. For per-channel interrupt sources, if not masked, the occurrence of the interrupt event will also cause the corresponding INT\_CHn bit (INTCH1~3) to be set '1'.

An interrupt event is cleared by writing '1' to the corresponding Interrupt Status bit. The INT\_CHn bit (INTCH1~3) will not be cleared until all the interrupts in the corresponding channel are acknowledged. The  $\overline{\rm INT}$  pin will be inactive until all the interrupts are acknowledged. Refer to Figure-40 for interrupt service flow.

No.	Interrupt Source	Status Bit	Interrupt Trigger Edges Select Bit	Interrupt Status Bit	Interrupt Mask Bit
1	TCLKn is missing.	TCKLOS_S (b3, STAT0,)	TCKLOS_IES (b3, INTES,)	TCKLOS_IS (b3, INTS0,)	TCKLOS_IM (b3, INTM0,)
2	LLOS is detected.	LLOS_S (b0, STAT0,)	LOS_IES (b1, INTES,)	LLOS_IS (b0, INTS0,)	LLOS_IM (b0, INTM0,)
3	SLOS is detected.	SLOS_S (b1, STAT0,)	LOS_IES (b1, INTES,)	SLOS_IS (b1, INTS0,)	SLOS_IM (b1, INTM0,)
4	TLOS is detected.	TLOS_S (b2, STAT0,)	TLOS_IES (b2, INTES,)	TLOS_IS (b2, INTS0,)	TLOS_IM (b2, INTM0,)
5	LAIS is detected.	LAIS_S (b6, STAT1,)	AIS_IES (b6, INTES,)	LAIS_IS (b6, INTS1,)	LAIS_IM (b6, INTM1,)
6	SAIS is detected.	SAIS_S (b7, STAT1,)	AIS_IES (b6, INTES,)	SAIS_IS (b7, INTS1,)	SAIS_IM (b7, INTM1,)
7	TOC is detected.	TOC_S (b4, STAT0,)	TOC_IES (b4, INTES,)	TOC_IS (b4, INTS0,)	TOC_IM (b4, INTM0,)
8	The PRBS/ARB pattern is detected syn- chronized.	PA_S (b5, STAT1,)	PA_IES (b5, INTES,)	PA_IS (b5, INTS1,)	PA_IM (b5, INTM1,)
9	Activate IB code is detected.	IBA_S (b1, STAT1,)	IB_IES (b0, INTES,)	IBA_IS (b1, INTS1,)	IBA_IM (b1, INTM1,)
10	Deactivate IB code is detected.	IBD_S (b0, STAT1,)	IB_IES (b0, INTES,)	IBD_IS (b0, INTS1,)	IBD_IM (b0, INTM1,)
11	The FIFO of the RJA is overflow or underflow.	-	-	RJA_IS (b5, INTS0,)	RJA_IM (b5, INTM0,)
12	The FIFO of the TJA is overflow or underflow.	-	-	TJA_IS (b6, INTS0,)	TJA_IM (b6, INTM0,)
13	Waveform amplitude is overflow.	-	-	DAC_IS (b7, INTS0,)	DAC_IM (b7, INTM0,)
14	SBPV is detected.	-	-	SBPV_IS (b5, INTS2,)	SBPV_IM (b5, INTM2,)
15	LBPV is detected.	-	-	LBPV_IS (b4, INTS2,)	LBPV_IM (b4, INTM2,)
16	SEXZ is detected.	-	-	SEXZ_IS (b3, INTS2,)	SEXZ_IM (b3, INTM2,)
17	LEXZ is detected.	-	-	LEXZ_IS (b2, INTS2,)	LEXZ_IM (b2, INTM2,)
18	PRBS/ARB error is detected.	-	-	ERR_IS (b1, INTS2,)	ERR_IM (b1, INTM2,)
19	The ERRCH and ERRCL registers are overflowed.	-	-	CNTOV_IS (b0, INTS2,)	CNTOV_IM (b0, INTM2,)
20	One second time is over.	-	-	TMOV_IS (b0, INTTM)	TMOV_IM (b0, GCF)

#### Table-25 Interrupt Summary

RENESAS

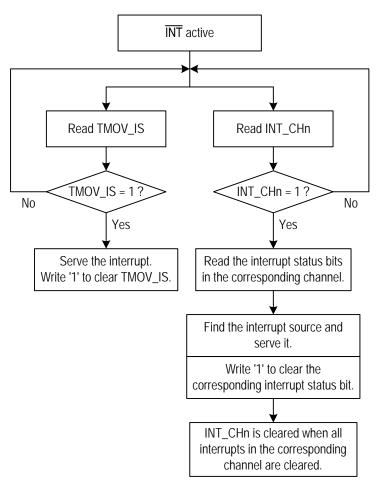


Figure-40 Interrupt Service Process

# 4 MISCELLANEOUS

## 4.1 RESET

The reset operation resets all registers, state machines as well as I/O pins to their default value or status.

The IDT82P2916 provides 4 kinds of reset:

- Power-on reset;
- · Hardware reset;
- · Global software reset;
- Per-channel software reset.

The Power-on, Hardware and Global software reset operations reset all the common blocks (including clock generator/synthesizer and microprocessor interface) and channel-related parts. The Per-channel software reset operation resets the channel-related parts. Figure-41 shows a general overview of the reset options.

During reset, all the line interface pins (i.e., TTIPn/TRINGn and RTIPn/RRINGn) are in High-Z state.

After reset, all the items listed in Table-26 are true.

#### Table-26 After Reset Effect Summary

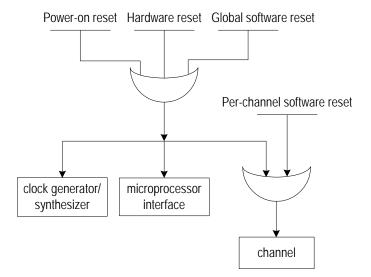


Figure-41 Reset

Effect On	Power-On Reset, Hardware Reset and Global Software Reset	Per-Channel Software Reset
TTIPn/TRINGn & RTIPn/ RRINGn	All TTIPn/TRINGn & RTIPn/RRINGn pins are in High-Z state.	Only TTIPn/TRINGn & RTIPn/RRINGn in the corresponding chan- nel are in High-Z.
Line Interface Mode	All channels are reset to T1/J1 mode.	Only the corresponding channel is reset to T1/J1 mode.
System interface	All channels are in Dual Rail NRZ Format.	Only the corresponding channel is in Dual Rail NRZ Format.
General I/O pins (i.e., D[7:0] and GPIO[1:0])	As input pins.	(No effect)
INT	Open drain output.	(No effect)
CLKT1, CLKE1, REFA, REFB	Output enable.	(No effect)
LLOS, LLOS0	Output enable.	(No effect)
TDO, SDO/ACK/RDY	High-Z.	(No effect)
state machines	All state machines are reset.	The state machines in the corresponding channel are reset.
Interrupt sources	All interrupt sources are masked.	The interrupt sources in the corresponding channel are masked.
Registers	All registers are reset to their default value.	The registers in the corresponding channel are reset to their default value except that there is no effect on the T1E1 bit.

#### 4.1.1 POWER-ON RESET

Power-on reset is initiated during power-up. When all VDD inputs (1.8V and 3.3V) reach approximately 60% of the standard value of VDD, power-on reset begins. If MCLK is applied, power-on reset will complete within 1 ms maximum; if MCLK is not applied, the device remains in reset state.

#### 4.1.2 HARDWARE RESET

Pulling the  $\overline{RST}$  pin to low will initiate hardware reset. The reset cycle should be more than 1 µs. If the  $\overline{RST}$  pin is held low continuously, the device remains in reset state.

#### 4.1.3 GLOBAL SOFTWARE RESET

Writing the RST register will initiate global software reset. Once initiated, global software reset completes in 1  $\mu s$  maximum.

#### 4.1.4 PER-CHANNEL SOFTWARE RESET

Writing a '1' to the CHRST bit (b1, CHCF,...) will initiate per-channel software reset. Once initiated, per-channel software reset completes in 1  $\mu$ s maximum and the CHRST bit (b1, CHCF,...) is self cleared.

This reset is different from other resets, for:

- It does not reset the T1E1 bit (b0, CHCF,...). That is, the operation mode of each channel is not changed;
- It does not reset the global registers, state machines and common pins (including the pins of clock generator, microprocessor interface and JTAG interface);
- It does not reset the other channels.

#### 4.2 MICROPROCESSOR INTERFACE

The microprocessor interface provides access to read and write the registers in the device. The interface consists of:

· Serial microprocessor interface;

### 4.3 POWER UP

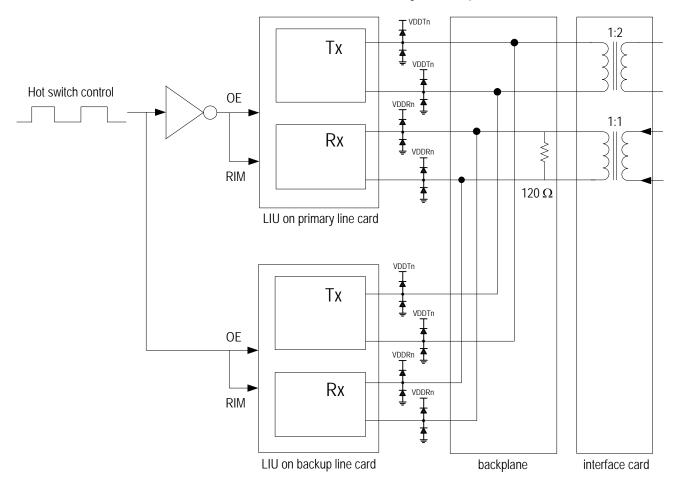
No power up sequencing for the VDD inputs (1.8 V and 3.3 V) has to be provided for the IDT82P2916. A Power-on reset will be initiated during power up. Refer to Section 4.1 Reset.

### 4.4 HITLESS PROTECTION SWITCHING (HPS) SUM-MARY

In today's telecommunication systems, ensuring no traffic loss is becoming increasingly important. To combat these problems, redundancy protection must be built into the systems carrying this traffic. There are many types of redundancy protection schemes, including 1+1 and 1:1 hardware protection without the use of external relays. Refer to Figure-42, Figure-43 and Figure-44 for different protection schemes. The IDT82P2916 provides an enhanced architecture to support both protection schemes.

IDT82P2916 highlights for HPS support:

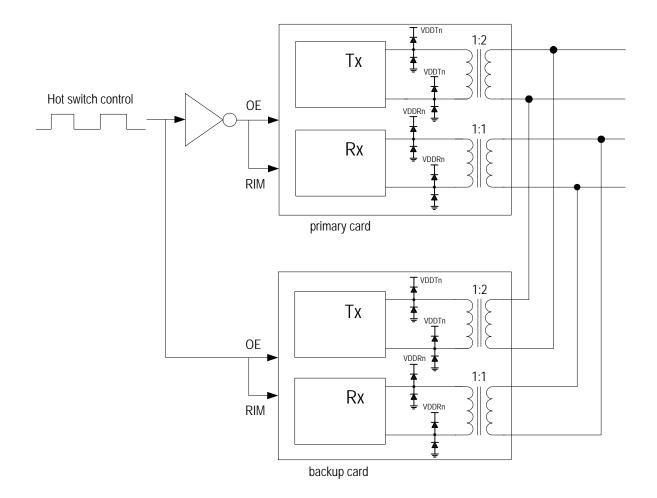
- Independent programmable receive and transmit high impedance for Tip and Ring inputs and outputs to support 1+1 and 1:1 redundancy
- Fully integrated receive termination, required to support 1:1 redundancy
- Enhanced internal architecture to guarantee High Impedance for Tip and Ring Inputs and Outputs during Power Off or Power Failure
- Asynchronous hardware control (OE, RIM) for fast global high impedance of receiver and transmitter (hot switching between working and backup board)



**Rx:** Partially Internal Impedance Matching mode. A fixed external 120  $\Omega$  resistor is placed on the backplane and provides a common termination for T1/J1/E1 applications. The R\_TERM[2:0] bits (b2-0, RCF0,...) setting is as follows: '000' for T1 100  $\Omega$  twisted pair cable, '001' for J1 110  $\Omega$  twisted pair cable, '001' for L1 120  $\Omega$  twisted pair cable and '011' for E1 75  $\Omega$  coaxial cable.

Tx: Internal Impedance Matching mode. The T\_TERM[2:0] bits (b2~0, TCF0,...) setting is as follows: '000' for T1 100 $\Omega$  twisted pair cable, '001' for J1 110  $\Omega$  twisted pair cable, '010' for E1 120  $\Omega$  twisted pair cable and '011' for E1 75  $\Omega$  coaxial cable.

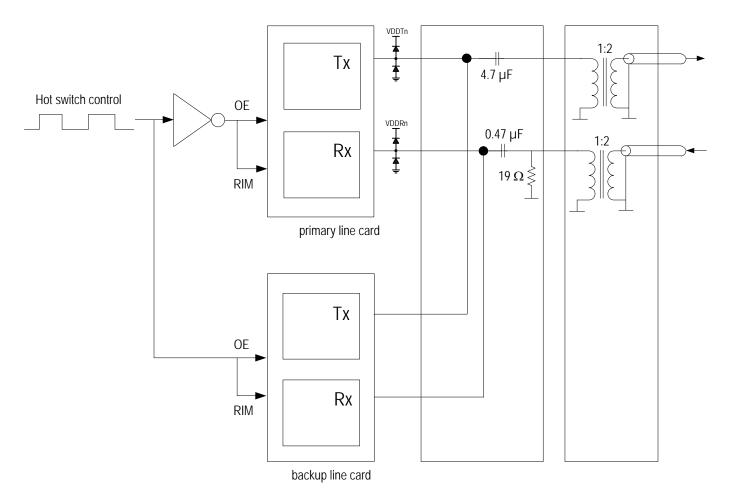




**Rx:** Fully Internal Impedance Matching mode. In this mode, there is no external resistor required. The R\_TERM[2:0] bits (b2~0, RCF0,...) setting is as follows: '000' for T1 100 $\Omega$  twisted pair cable, '001' for J1 110  $\Omega$  twisted pair cable, '010' for E1 120  $\Omega$  twisted pair cable and '011' for E1 75  $\Omega$  coaxial cable.

**Tx:** Internal Impedance Matching mode. The T\_TERM[2:0] bits (b2~0, TCF0,...) setting is as follows: '000' for T1 100  $\Omega$  twisted pair cable, '001' for J1 110  $\Omega$  twisted pair cable, '010' for E1 120  $\Omega$  twisted pair cable and '011' for E1 75  $\Omega$  coaxial cable.

#### Figure-43 1:1 HPS Scheme, Differential Interface (Individual Transformer)



**Rx:** 75  $\Omega$  External Impedance Matching mode. In this mode, there is no external resistor required. The RIM pin should be left open and the configuration of the R\_TERM[2:0] bits (b2~0, RCF0,...) is ignored.

Tx: 75 Ω Internal Impedance Matching mode. The T\_TERM[2:0] bits (b2~0, TCF0,...) should be set to '011'.

#### Figure-44 1+1 HPS Scheme, E1 75 ohm Single-Ended Interface (Shared Common Transformer)

# 5 PROGRAMMING INFORMATION

# 5.1 REGISTER MAP

### 5.1.1 GLOBAL REGISTER

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
Common	Control		I	1	1	1		1		
000	ID - Device ID Register	ID7	ID6	ID5	ID4	ID3	ID2	ID1	ID0	P 71
040	RST - Global Reset Register	RST7	RST6	RST5	RST4	RST3	RST2	RST1	RST0	P 71
080	GCF - Global Configuration Register	-	-	-	COPY	INT_PIN1	INT_PIN0	GLB_IM	TMOV_IM	P 72
	MON - G.772 Monitor Configura- tion Register	-	-	MON5	MON4	MON3	MON2	MON1	MON0	P 72
	GPIO - General Purpose I/O Pin Definition Register	-	-	-	-	LEVEL1	LEVEL0	DIR1	DIR0	P 73
Reference	e Clock Timing Option									
	CLKG - CLKT1 & CLKE1 Gener- ation Control Register	-	-	-	-	CLKE1_EN	CLKE1	CLKT1_EN	CLKT1	P 74
	REFCF - REFA/B Output Con- figuration Register	-	JA_BYPAS	REFH	FS_BYPAS	FREE	FREQ2	FREQ1	FREQ0	P 74
240	REFA - REFA Clock Sources Configuration Register	-	REFA_EN	CKA_T1E1	REFA4	REFA3	REFA2	REFA1	REFA0	P 76
280	REFB - REFB Clock Sources Configuration Register	-	REFB_EN	CKB_T1E1	REFB4	REFB3	REFB2	REFB1	REFB0	P 76
Interrupt I	Indication		L	L	•					
2C0	INTCH1 - Interrupt Requisition Source Register 1	INT_CH8	INT_CH7	INT_CH6	INT_CH5	INT_CH4	INT_CH3	INT_CH2	INT_CH1	P 77
300	INTCH2 - Interrupt Requisition Source Register 2	-	INT_CH15	INT_CH14	INT_CH13	INT_CH12	INT_CH11	INT_CH10	INT_CH9	P 77
380	INTCH3 - Interrupt Requisition Source Register 3	INT_CH0	-	-	-	-	-	-	-	P 77
	INTTM - One Second Timer Interrupt Status Register	-	-	-	-	-	-	-	TMOV_IS	P 78

#### 5.1.2 PER-CHANNEL REGISTER

Only the address of channel 1 is listed in the 'Address (Hex)' column of the following table. For the addresses of the other channels, refer to the description of each register.

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
Channel	Control			I			I			I
001	CHCF - Channel Configuration Register	-	-	-	-	-	-	CHRST	T1E1	P 78
JA Config	guration									
002	TJA - Transmit Jitter Attenuation Configuration Register	-	-	-	TJA_LIMT	TJA_EN	TJA_DP1	TJA_DP0	TJA_BW	P 79
003	RJA - Receive Jitter Attenuation Configuration Register				RJA_LIMT	RJA_EN	RJA_DP1	RJA_DP0	RJA_BW	P 80
Transmit	Path Configuration									
004	TCF0 - Transmit Configuration Register 0	-	OE	T_OFF	THZ_OC	T_SING	T_TERM2	T_TERM1	T_TERM0	P 81
005	TCF1 - Transmit Configuration Register 1	TMF_DEF2	TEM_DEF1	TMF_DEF0	TCK_ES	TD_INV	T_CODE	T_MD1	T_MD0	P 82
006	PULS - Transmit Pulse Configu- ration Register	-	-	-	-	PULS3	PULS2	PULS1	PULS0	P 83
007	SCAL - Amplitude Scaling Con- trol Register	-	-	SCAL5	SCAL4	SCAL3	SCAL2	SCAL1	SCAL0	P 84
008	AWG0 - Arbitrary Waveform Generation Control Register 0	-	DONE	RW	SAMP4	SAMP3	SAMP2	SAMP1	SAMP0	P 84
009	AWG1 - Arbitrary Waveform Generation Control Register 1	-	WDAT6	WDAT5	WDAT4	WDAT3	WDAT2	WDAT1	WDAT0	P 85
Receive I	Path Configuration									
00A	RCF0 - Receive Configuration Register 0	RCKH	RHZ	R_OFF	R120IN	R_SING	R_TERM2	R_TERM1	R_TERM0	P 86
00B	RCF1 - Receive Configuration Register 1	RMF_DEF2	RMF_DEF1	RMF_DEF0	RCK_ES	RD_INV	R_CODE	R_MD1	R_MD0	P 87
00C	RCF2 - Receive Configuration Register 2	-	-	-	-	-	-	MG1	MG0	P 88
Diagnosti	ics			L		•	L	l .	•	L
00D	LOS - LOS Configuration Regis- ter	LAC	ALOS2	ALOS1	ALOS0	TALOS1	TALOS0	TDLOS1	TDLOS0	P 89
00E	ERR - Error Detection & Inser- tion Control Register	EXZ_DEF	BPV_INS	ERR_INS	CNT_SEL2	CNT_SEL1	CNT_SEL0	CNT_MD	CNT_STOP	P 90
00F	AISG - AIS Generation Control Register	-	-	-	TXAIS	ASAIS_SL OS	ASAIS_LLO S	ALAIS_SLO S	ALAIS_LLO S	P 91
010	PG - Pattern Generation Control Register	-	PG_CK	PG_EN1	PG_EN0	PG_POS	PAG_INV	PRBG_SEL 1	PRBG_SEL 0	P 92

# RENESAS

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
	PD - Pattern Detection Control Register	-	-	-	-	PD_POS	PAD_INV	PAD_SEL1	PAD_SEL0	P 93
	ARBL - Arbitrary Pattern Gener- ation / Detection Low-Byte Reg- ister	ARB7	ARB6	ARB5	ARB4	ARB3	ARB2	ARB1	ARB0	P 94
	ARBM - Arbitrary Pattern Gen- eration / Detection Middle-Byte Register	ARB15	ARB14	ARB13	ARB12	ARB11	ARB10	ARB9	ARB8	P 94
	ARBH - Arbitrary Pattern Gener- ation / Detection High-Byte Reg- ister	ARB23	ARB22	ARB21	ARB20	ARB19	ARB18	ARB17	ARB16	P 94
	IBL - Inband Loopback Control Register	-	-	IBGL1	IBGL0	IBAL1	IBAL0	IBDL1	IBDL0	P 95
016	IBG - Inband Loopback Genera- tion Code Definition Register	IBG7	IBG6	IBG5	IBG4	IBG3	IBG2	IBG1	IBG0	P 95
017	IBDA - Inband Loopback Detec- tion Target Activate Code Defini- tion Register	IBA7	IBA6	IBA5	IBA4	IBA3	IBA2	IBA1	IBA0	P 96
018	IBDD - Inband Loopback Detec- tion Target Deactivate Code Definition Register	IBD7	IBD6	IBD5	IBD4	IBD3	IBD2	IBD1	IBD0	P 96
	LOOP - Loopback Control Reg- ister	-	-	-	-	AUTOLP	DLP	RLP	ALP	P 97
Interrupt L	Edge Selection									
	INTES - Interrupt Trigger Edges Select Register	-	AIS_IES	PA_IES	TOC_IES	TCKLOS_I ES	TLOS_IES	LOS_IES	IB_IES	P 98
Interrupt I	Mask				L					
	INTM0 - Interrupt Mask Register 0	DAC_IM	TJA_IM	RJA_IM	TOC_IM	TCKLOS_I M	TLOS_IM	SLOS_IM	LLOS_IM	P 99
	INTM1 - Interrupt Mask Register 1	SAIS_IM	LAIS_IM	PA_IM	-	-	-	IBA_IM	IBD_IM	P 100
01D	INTM2 - Interrupt Mask Register 2	-	-	SBPV_IM	LBPV_IM	SEXZ_IM	LEXZ_IM	ERR_IM	CNTOV_IM	P 101
Status Inc	dication									
01E	STAT0 - Status Register 0	AUTOLP_S	-	-	TOC_S	TCKLOS_S	TLOS_S	SLOS_S	LLOS_S	P 102
01F	STAT1 - Status Register 1	SAIS_S	LAIS_S	PA_S	-	-	-	IBA_S	IBD_S	P 103
Interrupt S	Status Indication									
	INTS0 - Interrupt Status Register 0	DAC_IS	tja_is	RJA_IS	TOC_IS	TCKLOS_I S	TLOS_IS	SLOS_IS	LLOS_IS	P 104
	INTS1 - Interrupt Status Regis- ter 1	SAIS_IS	LAIS_IS	PA_IS	-	-	-	IBA_IS	IBD_IS	P 105
	INTS2 - Interrupt Status Regis- ter 2	-	-	SBPV_IS	LBPV_IS	SEXZ_IS	LEXZ_IS	ERR_IS	CNTOV_IS	P 106

# RENESAS

Address (Hex)	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reference Page
Counter										
023	ERRCL - Error Counter Low- Byte Register	ERRC7	ERRC6	ERRC5	ERRC4	ERRC3	ERRC2	ERRC1	ERRC0	P 107
024	ERRCH - Error Counter High- Byte Register	ERRC15	ERRC14	ERRC13	ERRC12	ERRC11	ERRC10	ERRC9	ERRC8	P 107
Jitter Mea	asurement (channel 0 Only)				L	L	L		L	
7E5	JM - Jitter Measurement Config- uration For Channel 0 Register	-	-	-	-	-	JM_STOP	JM_MD	JM_BW	P 108
7E6	JIT_PL - Positive Peak Jitter Measurement Low-Byte Regis- ter	JIT_P7	JIT_P6	JIT_P5	JIT_P4	JIT_P3	JIT_P2	JIT_P1	JIT_P0	P 108
7E7	JIT_PH - Positive Peak Jitter Measurement High-Byte Regis- ter	-	-	-	-	JIT_P11	JIT_P10	JIT_P9	JIT_P8	P 108
7E8	JIT_NL - Negative Peak Jitter Measurement Low-Byte Regis- ter	JIT_N7	JIT_N6	JIT_N5	JIT_N4	JIT_N3	JIT_N2	JIT_N1	JIT_N0	P 109
7E9	JIT_NH - Negative Peak Jitter Measurement High-Byte Regis- ter	-	-	-	-	JIT_N11	JIT_N10	JIT_N9	JIT_N8	P 109

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# 5.2 REGISTER DESCRIPTION

# 5.2.1 GLOBAL REGISTER

# ID - Device ID Register

Address: 000H Type: Read Default Value: 70	)H						
7	6	5	4	3	2	1	0
ID7	ID6	ID5	ID4	ID3	ID2	ID1	ID0
Bit	Name			Descrij	ption		
7 - 0	ID[7:0]	The ID[7:0] bits are pre-s version number ('0000' is			e ID for the IDT82P2	.916. The ID[3:0] bi	ts represent the current

#### **RST - Global Reset Register**

Address: 040H Type: Write Default Value: 00	Н						
7	6	5	4	3	2	1	0
RST7	RST6	RST5	RST4	RST3	RST2	RST1	RST0
Bit	Name			Descri	otion		
7 - 0	7 - 0     RST[7:0]     Writing this register will initiate global software reset. This reset completes in 1 μs maximum.						

# RENESAS

# GCF - Global Configuration Register

7	6	5	4	3	2	1	0			
-		· ·	СОРҮ	INT_PIN1	INT_PIN0	GLB_IM	TMOV_IM			
Bit	Name			Descri	ption					
7 - 5	-	Reserved.								
4	СОРҮ	When the per-channel re ister of the other channel 0: Disable. (default) 1: Enable.	s simultaneously.		Jelermines whether tr	ne written value is d	copied to the same			
3 - 2	INT_PIN[1:0]									
1	GLB_IM	This bit is a global config 0: The per-channel interri tus bit is '1'. 1: Mask all the per-chann	upt will be generate	d when the per-chan			esponding interrupt			
0	TMOV_IM	This bit controls whether 0: Enable. 1: Mask. (default)	the interrupt is gene	erated when one seco	ond time is over. This	s one second timer	is locked to MCLK.			

# MON - G.772 Monitor Configuration Register

7	6	5	4	3	2	1	0 MON0			
-		MON5	MON4	MON3	MON2	MON1				
Bit Name Description										
7 - 6	-	Reserved.								
		transmitter or receiver to 000000: No transmitter of 000001: The receiver of of 000010: The receiver of of 00001111: The receiver of of 010000 ~ 011111: Reserv 100000: No transmitter of 100001: The transmitter of 100010: The transmitter of 00011: The transmitter of 000111: The transmitter of 100000 ~ 111111: Reserv	receiver is monitore channel 1 is monitore channel 2 is monitore hannel 15 is monitore ed. receiver is monitore of channel 1 is monit of channel 2 is monit f channel 15 is moni	ed. (default) ed. red. red. ed. tored.						

## GPIO - General Purpose I/O Pin Definition Register

	6	5	4	3	2	1	0
•	•	•	•	LEVEL1	LEVELO	DIR1	DIRO
Bit	Name			Descr	iption		
' - 4		Reserved.					
		1: Output high level. (def When the GPIO1 pin is o 0: Input low level. 1: Input high level. (defa	lefined as input, th			-	
2	LEVEL0	When the GPIO0 pin is of 0: Output low level. 1: Output high level. When the GPIO0 pin is of 0: Input low level. 1: Input high level. (defau	lefined as input, th				
1	DIR1	This bit determines whet 0: Output. 1: Input. (default)	her the GPIO1 pir	is used as output or	input.		

## CLKG - CLKT1 & CLKE1 Generation Control Register

7	6	5	4	3	2	1	0			
•	•	•		CLKE1_EN	CLKE1	CLKT1_EN	CLKT1			
Bit	Name			Descrip	tion					
7 - 4	-	Reserved.								
3	CLKE1_EN	0: The output is disabled. C	This bit controls whether the output on the CLKE1 pin is enabled. 0: The output is disabled. CLKE1 is in High-Z state. 1: The output is enabled. The frequency of CLKE1 is determined by the CLKE1 bit (b2, CLKG). (default)							
2	CLKE1	This bit is valid only when t 0: 8 KHz. 1: 2.048 MHz. (default)	he CLKE1_EN bit	(b3, CLKG) is '1'. Thi	is bit selects the clo	ock frequency output c	on the CLKE1 p			
1	CLKT1_EN	This bit controls whether th 0: The output is disabled. C 1: The output is enabled. T	LKT1 is in High-Z	state.	y the CLKT1 bit (b	0, CLKG). (default)				
0	CLKT1	This bit is valid only when t 0: 8 KHz.					on the CLKT1 pi			

## REFCF - REFA/B Output Configuration Register

-       JA_BYPAS       REFH       FS_BYPAS       FREE       FREQ2       FREQ1         Bit       Name       Description       D	
7       Reserved.         6       JA_BYPAS         This bit is valid only when the clock source for REFA or REFB is the recovered clock of one of the 16 channels in ing receiver. This bit determines whether the selected recovered clock passes through the RJA.         0: The selected recovered clock is derived from the output of RJA. (default)         1: The selected recovered clock does not pass through the RJA and is derived from the output of Rx Clock & Da         5       REFH         This bit is valid only when the selected clock source is lost. This bit controls the output on REFA/REFB.         For REFA, this bit, together with the FS_BYPAS bit (b4, REFCF) and the FREE bit (b3, REFCF), controls the when the selected clock source is the recovered clock of one of the 16 channels; this bit is ignored when th source is CLKA. Refer to the related table in the description of the FREE bit (b3, REFCF).	-
6       JA_BYPAS       This bit is valid only when the clock source for REFA or REFB is the recovered clock of one of the 16 channels in ing receiver. This bit determines whether the selected recovered clock passes through the RJA.         0: The selected recovered clock is derived from the output of RJA. (default)       1: The selected recovered clock does not pass through the RJA and is derived from the output of RX Clock & Da         5       REFH       This bit is valid only when the selected clock source is lost. This bit controls the output on REFA/REFB.         6       For REFA, this bit, together with the FS_BYPAS bit (b4, REFCF) and the FREE bit (b3, REFCF), controls the when the selected clock source is the recovered clock of one of the 16 channels; this bit is ignored when th source is CLKA. Refer to the related table in the description of the FREE bit (b3, REFCF).	
ing receiver. This bit determines whether the selected recovered clock passes through the RJA.       0: The selected recovered clock is derived from the output of RJA. (default)         1: The selected recovered clock does not pass through the RJA and is derived from the output of Rx Clock & Da         5       REFH         This bit is valid only when the selected clock source is lost. This bit controls the output on REFA/REFB.         For REFA, this bit, together with the FS_BYPAS bit (b4, REFCF) and the FREE bit (b3, REFCF), controls the when the selected clock source is the recovered clock of one of the 16 channels; this bit is ignored when th source is CLKA. Refer to the related table in the description of the FREE bit (b3, REFCF).	
For REFA, this bit, together with the FS_BYPAS bit (b4, REFCF) and the FREE bit (b3, REFCF), controls the when the selected clock source is the recovered clock of one of the 16 channels; this bit is ignored when th source is CLKA. Refer to the related table in the description of the FREE bit (b3, REFCF).	
For REFB: 0: Output free running clock. The frequency is 1.544 MHz if the selected clock source was T1 clock or 2.048 MH clock source was E1 clock. 1: Output high level. (default)	

3	li C F 1 V	n normal operation Coutput the clo REFCF). (defau Coutput free ru	tion: bock which is it) inning clock ted clock so	locked to the which is locke urce is lost, th	selected clo ed to MCLK a	ck source a	nd the fr Jency is	nternal Frequency Synthesizer equency is programmed in the FREQ[2:0] bits (b2~0 programmed in the FREQ[2:0] bits (b2~0, REFCF). PAS bit (b4, REFCF) and the REFH bit (b5, REFCF)
		Selecte Sou	d Clock Irce	FS_BYPA S	FREE	REFH		Output On REFA
					0	(don't-	High I	evel.
		CL	KA	0	1	care)		running clock, whose frequency is programmed FREQ[2:0] bits (b2~0, REFCF).
				1	(don'i	i-care)	High l	evel.
					0	0		unning clock, whose frequency is programmed FREQ[2:0] bits (b2~0, REFCF).
		Deserve	d ala ali af	0	Ī	1	High I	evel.
		one of the			1	(don't- care)		unning clock, whose frequency is programmed FREQ[2:0] bits (b2~0, REFCF).
					(don't-	0		unning clock, whose frequency is 1.544 MHz in ode or 2.048 MHz in E1 mode.
					care)	1	High I	evel.
2 - 0	FREQ[2:0] T	hese bits are v	alid only wh	en the Freque	ncy Synthes	izer on REF	A is enat	oled. These bits determine the output clock frequency
		FREQ[2:0]		ut when FS_B ncy Synthesiz refer		LKn or CLK		Output when FS_BYPAS=0 and FREE=1 (the Frequency Synthesizer is free running)
		000		z if the selecte Hz if the selec				-
		001			8 kHz			8 kHz
		010			64 kHz			64 kHz
		011		F	Reserved			-
		100		4.	096 MHz			4.096 MHz
		101		8.	192 MHz			8.192 MHz
		110		19	9.44 MHz			19.44 MHz
		111		32	.768 MHz			32.768 MHz

## REFA - REFA Clock Sources Configuration Register

7	6	5	1	0							
•	REFA_E	REFA2	REFA1	REFA0							
Bit	Name		Description								
7	-	Reserved.	served.								
6	REFA_EN	0: The output is disabled. I	nis bit controls whether the output on the REFA pin is enabled. The output is disabled. REFA is in High-Z state. The output is enabled. (default)								
5	CKA_T1E1	This bit defines the input c 0: Input T1 clock. (default) 1: Input E1 clock.	his bit defines the input clock frequency on the CLKA pin. Input T1 clock. (default)								
4 - 0	REFA[4:0]	These bits select the clock 00000: Recovered clock of 00001: Recovered clock of 00010: Recovered clock of  01111: Recovered clock of 10000 ~ 11111: The input of	channel 0. channel 1. (defaul channel 2. channel 15.	t)							

## REFB - REFB Clock Sources Configuration Register

7	6	5	4	3	2	1	0		
•	REFB_EN	REFB1	REFB0						
Bit	Name			Descrip	otion				
7	- Res	Reserved.							
6	0: T	This bit controls whether the output on the REFB pin is enabled. 0: The output is disabled. REFB is in High-Z state. 1: The output is enabled. (default)							
5	0: Ir	s bit defines the input c nput T1 clock. (default) nput E1 clock.	ock frequency on th	ne CLKB pin.					
4 - 0	000	se bits select the clock 00: Recovered clock of 01: Recovered clock of 10: Recovered clock of	channel 0. channel 1. (default	)					

#### INTCH1 - Interrupt Requisition Source Register 1

Address: 2C0H Type: Read Default Value: 0	0Н						
7	6	5	4	3	2	1	0
INT_CH8	B INT_CH	7 INT_CH6	INT_CH5	INT_CH4	INT_CH3	INT_CH2	INT_CH1
Bit	Name			Descri	ption		
7 - 0	INT_CH[8:1]	These bits indicate wheth channel 8 to 1 respectivel 0: No interrupt is generate 1: At least one interrupt is	y. ed or all the interrupt	ts are cleared in the	corresponding chan		[8:1] bits correspond to

#### INTCH2 - Interrupt Requisition Source Register 2

Address: 300H Type: Read Default Value: 0	)0H							
7	6		5	4	3	2	1	0
·	INT_CH	15 INT_	CH14 I	NT_CH13	INT_CH12	INT_CH11	INT_CH10	INT_CH9
Bit	Name				Descrip	otion		
7	-	Reserved.						
6 - 0	INT_CH[15:9]	channel 16 to 9 0: No interrupt	respectively. is generated or a	all the interru	rrupt generated in the pts are cleared in the corresponding channe	corresponding chan	-	6:9] bits correspond to

## INTCH3 - Interrupt Requisition Source Register 3

Address: 380 Type: Read Default Value							
7	6	5	4	3	2	1	0
INT_C	H0 -						·
Bit	Name			Descri	ption		
7	INT_CH0	This bit indicates whether 0: No interrupt is generat 1: At least one interrupt is	ed or all the interrupt	s are cleared in cha			
6 - 0	-	Reserved.					

## INTTM - One Second Timer Interrupt Status Register

dress: 3C0H be: Read / W fault Value: 0	rite						
7	6	5	4	3	2	1	0
-	•		•			•	TMOV_IS
Bit	Name			Descrip	tion		
7 - 1	-	Reserved.					
0	TMOV_IS	0: No one second time	hen the TMOV_IM bit (bo e over interrupt is genera ver interrupt is generated	ted; or a '1' is writte	n to this bit. (defaul		second time over.

#### 5.2.2 PER-CHANNEL REGISTER

## CHCF - Channel Configuration Register

	241H, 281H, 2C1F <i>(CH0)</i> ite	I, 101H, 141H, 181H, 1C1 I, 301H, 341H, 381H, <i>(CH</i>					
7	6	5	4	3	2	1	0
-	•	•		•	-	CHRST	T1E1
Bit	Name			Descr	iption		
7 - 2	-	Reserved.					
1	CHRST	Writing a '1' to this bit wil mum. This bit is self cleared.	l initiate per-channe	el software reset. On	ice initiated, per-ch	annel software reset o	completes in 1 µs maxi-
0	T1E1	This bit is valid only wher 0: T1/J1. (default) 1: E1. This bit can not be reset	·		T1/J1 or E1 opera	tion mode.	

## TJA - Transmit Jitter Attenuation Configuration Register

7	6									
	0	5	4	3	2	1	0			
-	•	•	TJA_LIMT	TJA_EN	TJA_DP1	TJA_DP0	TJA_BW			
Bit	Name	1		Descrij	otion					
7 - 5	-	Reserved.		2000.1						
4	TJA_LIMT	0: Disable. (default)	: Enable. The speed of the TJA outgoing data will be adjusted automatically if the FIFO in the TJA is 2-bit close to its full or e							
3	TJA_EN	This bit controls whethe 0: Disable. (default) 1: Enable.	r the TJA is enabled to	) use.						
2 - 1	TJA_DP[1:0]	These bits select the de 00: 128-bit. (default) 01: 64-bit. 1X: 32-bit.	pth of the TJA FIFO.							

## RJA - Receive Jitter Attenuation Configuration Register

	243H, 283H, 2C3⊦ <i>(CH0)</i> ite	I, 103H, 143H, 183 I, 303H, 343H, 383I	н, 1С3н, <i>(СН1~СН8)</i> ң, <i>(СН9~СН15)</i>						
7	6	5	4	3	2	1	0		
· ·	•	•	RJA_LIMT	RJA_EN	RJA_DP1	RJA_DP0	RJA_BW		
Bit	Name			Descr	iption				
7 - 5	-	Reserved.							
4	RJA_LIMT	0: Disable. (defau	Enable. The speed of the RJA outgoing data will be adjusted automatically if the FIFO in the RJA is 2-bit close to its full						
3	RJA_EN	This bit controls w 0: Disable. (defaul 1: Enable.	hether the RJA is enable t)	ed to use.					
2 - 1	RJA_DP[1:0]	These bits select t 00: 128-bit. (defau 01: 64-bit. 1X: 32-bit.	he depth of the RJA FIF It)	0.					
0	RJA_BW	0: 5 Hz (in T1/J1 r	e Corner Frequency for t node) / 6.77 Hz (in E1 m I1 mode) / 0.87 Hz (in E1	ode). (default)					

## TCF0 - Transmit Configuration Register 0

7	6	5	4	3	2	1	0	
	OE	T_OFF	T_TERM1	T_TERM0				
Bit	Name			Descrip	ption			
7	-	Reserved.						
6	OE	This bit determines the output of the Line Driver, i.e., the output on the TTIPn and TRINGn pins. 0: High-Z. (default) 1: Normal operation. This bit determines whether the transmitter is powered down. 0: Normal operation. (default) 1: Power down.						
5	T_OFF							
4	THZ_OC	This bit determines the output of the Line Driver, i.e., the output on the TTIPn and TRINGn pins when TOC is detected. 0: The output current is limited to 100 mAp-p. (default) 1: The output current is limited to 100 mAp-p within the first 1 ms after the TOC is detected and then the output is in High- when the TOC is detected for more than 1 ms.						
3	T_SING							
2 - 0	T_TERM[2:0]	These bits select the imp 000: The 100 $\Omega$ internal 001: The 110 $\Omega$ internal 010: The 120 $\Omega$ internal 011: The 75 $\Omega$ internal in 100: The 100 $\Omega$ internal	impedance matching mpedance matching impedance matching npedance matching i impedance matching	is selected for T1 1 is selected for J1 11 j is selected for E1 1 s selected for E1 75 j is selected for T1 1	00 $\Omega$ twisted pair ca 10 $\Omega$ twisted pair ca 20 $\Omega$ twisted pair ca $\Omega$ coaxial cable (wi 00 $\Omega$ twisted pair ca	ble (with transforme ole (with transformer ble (with transforme th transformer). ble (transformer-les:	). r). s).	
		101: The 110 $\Omega$ internal i 110: The 120 $\Omega$ internal i 111: The external impeda	mpedance matching					

## TCF1 - Transmit Configuration Register 1

	<i>(CH0)</i> rite	I, 305H, 345H, 385H, <i>(CH9</i>									
7	6	5	4	3	2	1	0				
TMF_DEF	TMF_DE	F1 TMF_DEF0	TCK_ES	TD_INV	T_CODE	T_MD1	T_MD0				
Bit	Name		Description								
7 - 5	TMF_DEF[2:0]       These bits are valid only in Transmit Dual Rail RZ Format mode and Transmit Single Rail NRZ Format mode. They de indication on the TMFn pin.         000:       PRBS/ARB indication when the PRBS/ARB detection is switched to the transmit path. Or reserved when the l detection is switched to the receive path. (default)         001:       SAIS indication.         010:       TOC indication.         011:       TLOS indication.         011:       TLOS indication.         011:       SEXZ indication.         101:       SEXZ indication in Transmit Dual Rail RZ Format mode. Reserved in Transmit Single Rail NRZ Format mode.         110:       SEXZ + SBPV indication in Transmit Dual Rail RZ Format mode. Reserved in Transmit Single Rail NRZ Format mode.         111:       SLOS indication in Transmit Dual Rail RZ Format mode. Reserved in Transmit Single Rail NRZ Format mode.						d when the PRBS// nat mode. IRZ Format mode.				
4	TCK_ES	This bit selects the active 0: Falling edge. (default) 1: Rising edge.	edge of the TCLKn	pin.							
3	TD_INV	This bit determines the ac 0: Active high. (default) 1: Active low.			pins.						
2	T_CODE	This bit selects the line co 0: B8ZS (in T1/J1 mode) / 1: AMI.	HDB3 (in E1 mode	). (default)							
1 - 0	T_MD[1:0]	These bits determines the 00: Transmit Single Rail N or 2.048 MHz (in E1 mode 01: Transmit Dual Rail NF T1/J1 mode) or 2.048 MH 10: Transmit Dual Rail RZ 11: Reserved.	RZ Format system e) clock is input on T RZ Format system ir z (in E1 mode) cloc	interface. The data is CLKn. hterface. The data is k is input on TCLKn.	s input on TDPn and (default)	TDNn in NRZ form					

## PULS - Transmit Pulse Configuration Register

7	6	5	2	1	3	2	1	0
	·	-	-	· Pl	JLS3	PULS2	PULS1	PULS0
Bit	Name				Descriptior	1		
7 - 4	-	Reserved.						
3 - 0	PULS[3:0]	These bits select of waveform.	one of the eight p	preset waveform te	emplates for sho	ort haul applicati	on or enable user-pro	grammable art
		PULS[3:0]	Operation Mode	Transmit Clock	Cable	Impedance	Cable Range	Cable Loss
		0000	E1	2.048 MHz	E	1 75 Ω	-	0 ~ 12 dB
		0001	E1	2.048 MHz	E	Ι 120 Ω	-	0 ~ 12 dB
		0010 (default)	DSX1	1.544 MHz		100 Ω	0 ~ 133 ft	0 ~ 0.6 dB
		0011	DSX1	1.544 MHz		100 Ω	133 ~ 266 ft	0.6 ~ 1.2 dE
		0100	DSX1	1.544 MHz		100 Ω	266 ~ 399 ft	1.2 ~ 1.8 dE
		0101	DSX1	1.544 MHz		100 Ω	399 ~ 533 ft	1.8 ~ 2.4 dE
		0110	DSX1	1.544 MHz		100 Ω	533 ~ 655 ft	2.4 ~ 3.0 dE
		0111	J1	1.544 MHz		110 Ω	-	0 ~ 12 dB
					ser-programmal			1

## SCAL - Amplitude Scaling Control Register

	247H, 287H, 2C7H <i>(CH0)</i> ite	I, 107H, 147H, 187H, 1C7F I, 307H, 347H, 387H, <i>(СН</i> 9					
7	6	5	4	3	2	1	0
•	•	SCAL5	SCAL4	SCAL3	SCAL2	SCAL1	SCALO
Bit	Name			Descrip	otion		
7 - 6 5 - 0	- SCAL[5:0]	Reserved. These bits specify a scalir In T1/J1 mode, the standar standard value will result i In E1 mode, the standard standard value will result i <b>Note:</b> The default value f should be written to these	ard value is '110110 n 2% scaling up or Value is '100001' f n 3% scaling up or or SCAL[5:0] is '110	' for the waveform and down against the waveform among for the waveform among down against the waveform and 0110' which is the T	mplitude. If necessa aveform amplitude. T aplitude. If necessar aveform amplitude. T	ry, increasing or de The scale range is fr y, increasing or dec The scale range is fr	rom +20% to -100%. creasing by '1' from the rom +100% to -100%.

## AWG0 - Arbitrary Waveform Generation Control Register 0

	248H, 288H, 2C8H, 3 <i>(CH0)</i> ite	108H, 148H, 188H, 1C8 308H, 348H, 388H, <i>(CH</i>						
7	6	5	4	3	2	1	0	
·	DONE	RW	SAMP4	SAMP3	SAMP2	SAMP1	SAMP0	
Bit	Name			Descrip	otion			
7	- F	Reserved.						
6	tı C 1	This bit is valid only when the user-programmable arbitrary waveform is enabled (i.e., the PULS[3:0] bits (b3~0, PULS,) ar to '1XXX'). This bit determines whether to enable the data writing/reading from RAM. 0: Disable. (default) 1: Enable.						
5	tı C	Fhis bit is valid only wher o '1XXX'). This bit detern ): Write data to RAM. (de 1: Read data from RAM.	mines read/write dire		orm is enabled (i.e.,	the PULS[3:0] bits (	b3~0, PULS,) are set	
4 - 0	a C C  1 1	These bits are valid only are set to '1XXX'). These 00000: The RAM sample 00011: The RAM sample 00010: The RAM sample  10001: The RAM sample 10010: The RAM sample 10011 ~ 11111: The RAM	bits specify the RAN address is 0. (defau address is 1. address is 2. address is 17. address is 18.	A sample address. lt)	waveform is enable	d (i.e., the PULS[3:0	0] bits (b3~0, PULS,)	

## AWG1 - Arbitrary Waveform Generation Control Register 1

	249H, 289H, 2C9H <i>(CH0)</i> ite	I, 109H, 149H, 189H, 1C9 I, 309H, 349H, 389H, <i>(CH</i>					
7	6	5	4	3	2	1	0
· ·	WDAT	5 WDAT5	WDAT4	WDAT3	WDAT2	WDAT1	WDAT0
Bit	Name			Descri	ption		
7	-	Reserved.					
6 - 0	WDAT[6:0]	are set to '1XXX').	template sample dat	a to be stored in R	AM which address		0] bits (b3~0, PULS,) SAMP[4:0] bits (b4~0,

## RCF0 - Receive Configuration Register 0

7	6	5	4	3	2	1	0		
RCKH	RHZ	R_OFF	R120IN	R_SING	R_TERM2	R_TERM1	R_TERM0		
Bit	Name			Descri	otion				
7	RCKH	RCKH       This bit determines the output on RCLKn when LLOS is detected. This bit is valid only when LLOS is detected and the pattern generation is disabled in the receive path.         0: XCLK. (default)       1: High level.							
6	RHZ		: High-Z. (default)						
5	R_OFF	This bit determines whether the receiver is powered down. 0: Normal operation. (default) 1: Power down.							
4	R120IN	<ul> <li>This bit is valid only when the receive line interface is in Receive Differential mode and per-channel internal impedance configuration is enabled. This bit selects the internal impedance matching mode.</li> <li>0: Partially Internal Impedance Matching mode. An internal programmable resistor (IM) and a value-fixed external res are used. (default)</li> <li>1: Fully Internal Impedance Matching mode. Only an internal programmable resistor (IM) is used.</li> </ul>							
3	R_SING		e interface. Both RT						
2 - 0	R_TERM[2:0]	<ul> <li>0: Receive Differential line interface. Both RTIPn and RRINGn are used to receive signal from the line side. (default)</li> <li>1: Receive Single Ended line interface. Only RTIPn is used to receive signal. RRINGn should be left open.</li> <li>These bits are valid only when impedance matching is configured on a per-channel basis. These bits select the im matching mode of the receive path to match the cable impedance.</li> <li>In Receive Differential mode:</li> <li>000: The 100 Ω internal impedance matching is selected for T1 100 Ω twisted pair cable.</li> <li>001: The 110 Ω internal impedance matching is selected for E1 120 Ω twisted pair cable.</li> <li>010: The 120 Ω internal impedance matching is selected for E1 75 Ω coaxial cable.</li> <li>11: The 75 Ω internal impedance matching is selected for T1 100 Ω, J1 110 Ω twisted pair cable.</li> <li>11: The 75 Ω internal impedance matching is selected for E1 75 Ω coaxial cable.</li> <li>11: The 75 Ω internal impedance matching is selected for E1 120 Ω twisted pair cable.</li> <li>11: The 75 Ω internal impedance matching is selected for E1 120 Ω twisted pair cable.</li> <li>11: The 75 Ω internal impedance matching is selected for E1 120 Ω twisted pair cable.</li> <li>11: The 75 Ω internal impedance matching is selected for E1 75 Ω coaxial cable.</li> <li>11: The 75 Ω internal impedance matching is selected for E1 120 Ω twisted pair cable and E1 75 Ω coaxia lable.</li> </ul>							

## RCF1 - Receive Configuration Register 1

20BH	, 24BH, 28BH, 2CB   <i>(CH0)</i> /rite	H, 10BH, 14BH, 18BH, 1CE H, 30BH, 34BH, 38BH, <i>(CH</i>								
7	6	5	4	3	2	1	0			
RMF_DE	F2 RMF_DE	F1 RMF_DEF0	RCK_ES	RD_INV	R_CODE	R_MD1	R_MD0			
Bit	Name		Description							
7 - 5	RMF_DEF[2:0]	These bits are valid only in put on the RMFn pin. 000: PRBS/ARB indicatio detection is switched to th 001: LAIS indication. 010: XOR data of positive 011: Recovered clock (RC 100: LEXZ indication. 101: LBPV indication. 110: LEXZ + LBPV indication.	n when the PRBS/. e transmit path. (de and negative sliced LK).	ARB detection is sw fault)			5			
4	RCK_ES	This bit selects the active 0: Rising edge. (default) 1: Falling edge.	edge of the RCLKn	pin.						
3	RD_INV	This bit determines the ac 0: Active high. (default) 1: Active low.	tive level on the RD	Pn and RDNn pins.						
2	R_CODE	This bit selects the line co 0: B8ZS (in T1/J1 mode) / 1: AMI.								
1 - 0	R_MD[1:0]	These bits determines the 00: Receive Single Rail N mode) or 2.048 MHz (in E 01: Receive Dual Rail NR T1/J1 mode) or 2.048 MH 10: Receive Dual Rail RZ J1 mode) or 2.048 MHz (in 11: Receive Dual Rail Slic the Slicer.	IRZ Format system 1 mode) recovered 2 Format system in z (in E1 mode) reco Format system inte n E1 mode) recover	n interface. The data clock is output on R terface. The data is o overed clock is output rface. The data is output red clock is output or	CLKn. output on RDPn and it on RCLKn. (defau itput on RDPn and F n RCLKn.	l RDNn in NRZ forr lt) RDNn in RZ format	nat and a 1.544 MHz and a 1.544 MHz (in <sup>-</sup>			

## RCF2 - Receive Configuration Register 2

	24CH, 28CH, 2CC <i>(CH0)</i> ite	H, 10CH, 14CH, 18CH, 10 H, 30CH, 34CH, 38CH, <i>(C</i>					
7	6	5	4	3	2	1	0
·	•	-	-	-	-	MG1	MG0
Bit	Name			Descrip	otion		
7 - 2	-	Reserved.					
1 - 0		These bits select the Mor 00: 0 dB. (default) 01: 20 dB. 10: 26 dB. 11: 32 dB.	itor Gain.				

## LOS - LOS Configuration Register

7	6		5	4	3	2	1	0
LAC	ALOS	52 A	LOS1	ALOS0	TALOS1	TALOS0	TDLOS1	TDLOSO
Bit	Name				Des	cription		
7	LAC	0: T1.231 (in <sup>-</sup>	T1/J1 mode	5, SLOS and AIS c e) / G.775 (in E1 m / ETSI 300233 & I.	ode). (default)			
6 - 4	ALOS[2:0]	vals, LLOS is	declared. T )] settings f	he consecutive pu	lise intervals (N) and mode and Line M	litude of the data is less re determined by the LA onitor mode are differer	C bit (b7, LOS,).	·
				ALOS[2:0]	Q (Vpp)	vs. 6.0 Vpp (dB)	vs. 4.74 Vpp (dB)	
				000	0.5	21.58	19.54	
			(	001 (default)	0.7	18.66	16.61	
				010	0.9	16.48	14.43	
				011	1.2	13.98	11.93	
				100	1.4	12.64	10.59	
				101	1.6	11.48	9.43	
				110	1.8	10.46	8.41	
				111	2.0	9.54	7.49	
			ALO	S[2:0] Setting ir	n Line Monitor N	lode		
				ALOS[2:0]	Q (Vpp)	vs. 6.0 Vpp (dB)	vs. 4.74 Vpp (dB)	
				000	1.0	15.56	13.52	
			(	001 (default)	1.4	12.64	10.59	
				010	1.8	10.46	8.41	
				011	2.2	8.71	6.67	
				1xx		reserved.		
3 - 2	TALOS[1:0]		e period is d ed. Il line interf efault) ded line int	determined by the ace:		e of the data is less thar ), LOS,). When the ar		

1 - 0	TDLOS[1:0]	These bits select the period. When the amplitude of the data is less than a certain voltage for the period, TLOS is declared. The
		voltage is determined by the TALOS bits (b3~2, LOS,).
		00: 16-pulse.
		01: 32-pulse. (default)
		1X: 64-pulse.

## ERR - Error Detection & Insertion Control Register

20EH,	, 24EH, 28EH, 2CEI <i>(CH0)</i> ′rite	H, 10EH, 14EH, 18EH, 1C H, 30EH, 34EH, 38EH, <i>(Cl</i>					
7	6	5	4	3	2	1	0
EXZ_DE	F BPV_IN	S ERR_INS	CNT_MD	CNT_STOP			
Bit	Name			Descri	ption		
7	EXZ_DEF	This bit selects the EXZ d 0: ANSI. (default) 1: FCC.	efinition standard.				
6	BPV_INS	This bit controls whether the writing '1' to this bit will in This bit is cleared once the theorem on the theorem on the theorem once the theorem once the theorem on the theorem once the theorem on theorem on t	sert a BPV on the n	ext available mark ir		be transmitted.	
5	ERR_INS	This bit controls whether the transition from '0' to '1' This bit is cleared once the transition from '0' to '1' This bit is cleared once the transition of transition of the transition of the transition of transition	on this bit will insert	a single bit error to			
4 - 2	CNT_SEL[2:0]	These bits select what kir 000: Disable. (default) 001: LBPV. 010: LEXZ. 011: LBPV + LEXZ. 100: SBPV. 101: SEXZ. 110: SBPV + SEXZ. 111: PRBS/ARB error.	d of error to be cou	nted by the internal	Error Counter.		
1	CNT_MD	This bit determines wheth 0: Manually by setting the 1: Every-one second auto	CNT_STOP bit (b0		updated automaticall	y or manually.	
0	CNT_STOP	This bit is valid only when A transition from '0' to '1' This bit must be cleared b	on this bit updates t	he ERRCH & ERRC	L registers.		

## AISG - AIS Generation Control Register

7	6	5	4	3	2	1	0
	•		TXAIS	ASAIS_SLOS	ASAIS_LLOS	ALAIS_SLOS	ALAIS_LLOS
Bit	Name			Descri	ption		
7 - 5	-	Reserved.					
4	TXAIS	This bit controls the trans 0: Disable. (default) 1: Transmit all ones patte In remote loopback, this	ern at TTIPn/TRING	·			
3	ASAIS_SLOS	This bit controls the AIS 0: Disable. (default) 1: Enable.	generation in the re	eceive path once SLO	S is detected.		
2	ASAIS_LLOS	This bit controls the AIS 0: Disable. (default) 1: Enable.	generation in the re	ceive path once LLO	S is detected.		
1	ALAIS_SLOS	This bit controls the AIS 0: Disable. (default) 1: Enable.	generation in the tra	ansmit path once SLC	OS is detected.		
0	ALAIS_LLOS	This bit controls the AIS 0: Disable. (default)	generation in the tra	ansmit path once LLC	OS is detected.		

## PG - Pattern Generation Control Register

7	6	5	4	3	2	1	0
-	PG_CK	PG_EN1	PG_EN0	PG_POS	PAG_INV	PRBG_SEL1	PRBG_SEL
Bit	Name			Descrip	otion		
7	-	Reserved.		-			
5 - 4	PG_CK	This bit selects the referent When the pattern is gener 0: XCLK. (default) 1: Recovered clock from t When the pattern is gener 0: XCLK. (default) 1: Transmit clock, i.e., the mat mode) or the clock re These bits select the patter	ated in the receiver he received signal. ated in the transmit clock input on TCL covered from the da	path: path: Kn (in Transmit Sing ta input on TDPn an	jle Rail NRZ Format	mode and in Transi	
5-4	FG_LN[1.0]	00: Disable. (default) 01: PRBS. 10: ARB. 11: IB.	an to be generated.				
3	PG_POS	This bit selects the patterr 0: Transmit path. (default) 1: Receive path.		ARB & IB) generation	n direction.		
				ed PRBS/ARB patte			

## PD - Pattern Detection Control Register

: Read / W ult Value: 0								
7	6	5	4	3	2	1	0	
-	· ·	•	•	PD_POS	PAD_INV	PAD_SEL1	PAD_SEL0	
Bit	Name			Descrip	otion			
7 - 4	-	Reserved.	Reserved.					
3	PD_POS	This bit selects the patter 0: Receive path. (default) 1: Transmit path.		ARB & IB) detection	direction.			
2	PAD_INV	This bit controls whether 0: Normal. (default) 1: Invert.	to invert the data bef	ore PRBS/ARB dete	ection.			
1 - 0	PAD_SEL[1:0]	These bits select the des 00: 2 <sup>20</sup> - 1 QRSS. 01: 2 <sup>15</sup> - 1 PRBS. 10: 2 <sup>11</sup> - 1 PRBS. 11: ARB. (default)	red PRBS/ARB patt	ern to be detected.				

#### ARBL - Arbitrary Pattern Generation / Detection Low-Byte Register

	252H, 292H, 2D2H <i>(CH0)</i> te	, 112H, 152H, 192H, 1D2 , 312H, 352H, 392H, <i>(CH</i>	•				
7	6	5	4	3	2	1	0
ARB7	ARB6	ARB5	ARB4	ARB3	ARB2	ARB1	ARB0
Bit	Name			Descrip	tion		
7 - 0	ARB[7:0]	These bits, together with be generated or detected				r detected. The ARE	323 bit is the first bit to

#### ARBM - Arbitrary Pattern Generation / Detection Middle-Byte Register

	253H, 293H, 2D3H <i>(CH0)</i> te	, 113H, 153H, 193H, 1D3I , 313H, 353H, 393H, <i>(CH</i>					
7	6	5	4	3	2	1	0
ARB15	ARB14	ARB13	ARB12	ARB11	ARB10	ARB9	ARB8
Bit	Name			Descri	ption		
7 - 0	ARB[15:8]	(Refer to the description	of the ARBL register	.)			

#### ARBH - Arbitrary Pattern Generation / Detection High-Byte Register

	254H, 294H, 2D4H <i>(CH0)</i> te	, 114H, 154H, 194H, 1D4H , 314H, 354H, 394H, <i>(CH</i> 9	•				
7	6	5	4	3	2	1	0
ARB23	ARB22	ARB21	ARB20	ARB19	ARB18	ARB17	ARB16
Bit	Name			Descrij	otion		
7 - 0	ARB[23:16]	(Refer to the description of	of the ARBL register.	)			

#### IBL - Inband Loopback Control Register

/	6	5	5 4	3	2	1	0			
	-	IBGL1	IBGL0	IBAL1	IBAL0	IBDL1	IBDL0			
Bit	Name			Descrip	otion					
7 - 6	-	Reserved.								
5 - 4	IBGL[1:0]	These bits define the leng 00: 5-bit long in the IBG[4 01: 6-bit long in the IBG[5 10: 7-bit long in the IBG[6 11: 8-bit long in the IBG[7	:0] bits (b4~0, IBG, :0] bits (b5~0, IBG, :0] bits (b6~0, IBG,	.). (default) .). .).	Immed in the IBG[7:	0] bits (b7~0, IBG,)				
3 - 2	IBAL[1:0]	00: 5-bit long in the IBA[4 01: 6-bit long in the IBA[5 10: 7-bit long in the IBA[6	<ul> <li>is used to be to be in the IBG[7:0] bits (b7-0, IBC,).</li> <li>is bits define the length of the valid target activate IB detection code programmed in the IBA[7:0] bits (b7-0, IBDA,).</li> <li>is 5-bit long in the IBA[4:0] bits (b4-0, IBDA,). (default)</li> <li>is 6-bit long in the IBA[5:0] bits (b5-0, IBDA,).</li> <li>is 7-bit long in the IBA[6:0] bits (b6-0, IBDA,).</li> <li>is 8-bit long in the IBA[7:0] bits (b7-0, IBDA,).</li> </ul>							

## IBG - Inband Loopback Generation Code Definition Register

	256H, 296H, 2D6H <i>(CH0)</i> te	, 116H, 156H, 196H, 1D6l , 316H, 356H, 396H, <i>(CH</i>					
7	6	5	4	3	2	1	0
IBG7	IBG6	IBG5	IBG4	IBG3	IBG2	IBG1	IBG0
Bit	Name			Descript	tion		
7 - 0	IBG[7:0]	The IBG[X:0] bits define IBG0 bit is the last bit to b					

## IBDA - Inband Loopback Detection Target Activate Code Definition Register

	257H, 297H, 2D7H <i>(CH0)</i> ite	, 117H, 157H, 197H, 1D7 , 317H, 357H, 397H, <i>(CH</i>	• •				
7	6	5	4	3	2	1	0
IBA7	IBA6	IBA5	IBA4	IBA3	IBA2	IBA1	IBA0
Bit	Name			Descrip	otion		
7 - 0		The IBA[X:0] bits define IBL,). The IBA0 bit is th		0	ection code. The 'X'	is determined by th	ne IBAL[1:0] bits (b3~2,

#### IBDD - Inband Loopback Detection Target Deactivate Code Definition Register

	258H, 298H, 2D8H <i>(CH0)</i> ite	I, 118H, 158H, 198H, 1D8 I, 318H, 358H, 398H, <i>(CH</i>					
7	6	5	4	3	2	1	0
IBD7	IBD6	IBD5	IBD4	IBD3	IBD2	IBD1	IBD0
Bit	Name			Descrip	otion		
7 - 0	IBD[7:0]	The IBD[X:0] bits define to IBL,). The IBD0 bit is the IBD0			ection code. The 'X	' is determined by th	ne IBDL[1:0] bits (b1~0,

## LOOP - Loopback Control Register

7	6	5	4	3	2	1	0
	· ·	•	•	AUTOLP	DLP	RLP	ALP
Bit	Name			Descript	tion		
7 - 4	-	Reserved.					
3	AUTOLP	This bit determines whet 0: Automatic Digital/Rem 1: Automatic Digital/Rem vate IB code is detected the deactivate IB code is	note Loopback is disa note Loopback is ena in the transmit/recei	ibled. (default) bled. The correspond ve path for more than	ing channel will ente 5.1 sec.; and will re		
2	DLP	This bit controls whether 0: Disable. (default) 1: Enable.	Digital Loopback is e	enabled.			
1	RLP	This bit controls whether 0: Disable. (default) 1: Enable.	Remote Loopback is	s enabled.			
0	ALP	This bit controls whether 0: Disable. (default)	Analog Loopback is	enabled.			

## INTES - Interrupt Trigger Edges Select Register

7	6	5	4	3	2	1	0		
-	AIS_IE	S PA_IES	TOC_IES	TCKLOS_IES	TLOS_IES	LOS_IES	IB_IES		
Bit	Name			tion					
7	-	Reserved.							
6	AIS_IES	This bit selects the transi 0: A transition from '0' t INTS1,) / the SAIS_IS 1: Any transition from '0 LAIS_IS bit (b6, INTS1,	o '1' on the LAIS_S bit (b7, INTS1,) to ' to '1' or from '1' to	5 bit (b6, STAT1,) '1' respectively. (defa o '0' on the LAIS_S b	/ the SAIS_S bit (but) out) bit (b6, STAT1,) /	o7, STAT1,) will se			
5	PA_IES	0: A transition from '0' to	This bit selects the transition edge of the PA_S bit (b5, STAT1,). b: A transition from '0' to '1' on the PA_S bit (b5, STAT1,) will set the PA_IS bit (b5, INTS1,) to '1'. (default) : Any transition from '0' to '1' or from '1' to '0' on the PA_S bit (b5, STAT1,) will set the PA_IS bit (b5, INTS1,) to '1'.						
4	TOC_IES	0: A transition from '0' to	This bit selects the transition edge of the TOC_S bit (b4, STAT0,). D: A transition from '0' to '1' on the TOC_S bit (b4, STAT0,) will set the TOC_IS bit (b4, INTS0,) to '1'. (default) 1: Any transition from '0' to '1' or from '1' to '0' on the TOC_S bit (b4, STAT0,) will set the TOC_IS bit (b4, INTS0,) to						
3	TCKLOS_IES	This bit selects the transi 0: A transition from '0' to 1: Any transition from '0' '1'.	'1' on the TCKLOS_	S bit (b3, STAT0,)	will set the TCKLOS				
2	TLOS_IES	This bit selects the transi 0: A transition from '0' to 1: Any transition from '0'	'1' on the TLOS_S b	oit (b2, STAT0,) will	set the TLOS_IS b				
1	LOS_IES	This bit selects the transi 0: A transition from '0' to INTS0,) / the SLOS_IS 1: Any transition from '0' LLOS_IS bit (b0, INTS0,	b '1' on the LLOS_S bit (b1, INTS0,) to to '1' or from '1' to	5 bit (b0, STAT0,) / '1' respectively. (def '0' on the LLOS_S b	the SLOS_S bit (b ault) iit (b0, STAT0,) /	o1, STAT0,) will s€			
0	IB_IES	This bit selects the transi 0: A transition from '0' to the IBD_IS bit (b0, INTS 1: Any transition from '0' bit (b1, INTS1,) / the IB	tion edge of the IBA '1' on the IBA_S bit I,) to '1' respective to '1' or from '1' to '0	_S bit (b1, STAT1,) (b1, STAT1,) / the I ly. (default) )' on the IBA_S bit (b	and the IBD_S bit BD_S bit (b0, STAT 1, STAT1,) / the I	1,) will set the IBA			

## INTM0 - Interrupt Mask Register 0

ılt Value: F	FH								
7	6	5	4	3	2	1	0		
DAC_IM	TJA_IN	RJA_IM TOC_IM TCKLOS_IM TLOS_IM SLOS_IM LLO							
Bit	Name			Descrip	tion				
7	DAC_IM	0: Interrupt is enabled.	is bit is the waveform amplitude overflow interrupt mask. Interrupt is enabled. Interrupt is masked. (default)						
6	TJA_IM	0: Interrupt is enabled.	his bit is the TJA FIFO overflow and underflow interrupt mask. : Interrupt is enabled. : Interrupt is masked. (default)						
5	RJA_IM	0: Interrupt is enabled.	his bit is the RJA FIFO overflow and underflow interrupt mask.						
4	TOC_IM	This bit is the Line Driver 0: Interrupt is enabled. 1: Interrupt is masked. (d	·	Χ.					
3	TCKLOS_IM	This bit is the TCLKn mis 0: Interrupt is enabled. 1: Interrupt is masked. (d	5						
2	TLOS_IM	0: Interrupt is enabled.	his bit is the TLOS interrupt mask.						
1	SLOS_IM	This bit is the SLOS inter 0: Interrupt is enabled. 1: Interrupt is masked. (d	•						
0	LLOS_IM	This bit is the LLOS inter 0: Interrupt is enabled. 1: Interrupt is masked. (d	rupt mask.						

## INTM1 - Interrupt Mask Register 1

IBA_IM	0						
	IBD_IM						
his bit is the SAIS interrupt mask. Interrupt is enabled. Interrupt is masked. (default)							
This bit is the LAIS interrupt mask. 0: Interrupt is enabled. 1: Interrupt is masked. (default)							

## INTM2 - Interrupt Mask Register 2

	<i>(CH0)</i> Irite	DH, 31DH, 35DH, 39DH, (	СН9~СН15)						
7	6	5	4	3	2	1	0		
	· ·	SBPV_IM	LBPV_IM	LEXZ_IM	ERR_IM	CNTOV_IM			
Bit	Name			Descrij	ption				
7 - 6	-	Reserved.	eserved.						
5	SBPV_IM	0: Interrupt is enabled.	his bit is the SBPV interrupt mask. Interrupt is enabled. Interrupt is masked. (default)						
4	LBPV_IM	This bit is the LBPV inte 0: Interrupt is enabled. 1: Interrupt is masked. (	his bit is the LBPV interrupt mask. : Interrupt is enabled.						
3	SEXZ_IM	This bit is the SEXZ inte 0: Interrupt is enabled. 1: Interrupt is masked. (							
2	LEXZ_IM	This bit is the LEXZ inte 0: Interrupt is enabled. 1: Interrupt is masked. (							
1	ERR_IM	This bit is the PRBS/AR 0: Interrupt is enabled. 1: Interrupt is masked. (		х. -					
0	CNTOV_IM	This bit is the ERRCH a 0: Interrupt is enabled. 1: Interrupt is masked. (	-	overflow interrupt ma	ask.				

## STAT0 - Status Register 0

	25eh, 29eh, 2De <i>(Cho)</i>	H, 11EH, 15EH, 19EH, 1DE H, 31EH, 35EH, 39EH, <i>(CH</i>							
7	6	5	4	3	2	1	0		
AUTOLP_S	5 -	•	TOC_S	TCKLOS_S	TLOS_S	SLOS_S	LLOS_S		
Bit	Name			Descrip	tion				
7	AUTOLP_S	This bit indicates the autor 0: Out of automatic Digital 1: In automatic Digital/Rer	Remote Loopback						
6 - 5	-	Reserved.							
4	TOC_S		nis bit indicates the TOC status. No TOC is detected. (default) TOC is detected.						
3	TCKLOS_S	This bit indicates the TCLI 0: TCLKn is not missing. ( 1: TCLKn is missing.							
2	TLOS_S	This bit indicates the TLO: 0: No TLOS is detected. ( 1: TLOS is detected.							
1	SLOS_S	This bit indicates the SLO 0: No SLOS is detected. ( 1: SLOS is detected.							
0	LLOS_S	This bit indicates the LLOS 0: No LLOS is detected. (o 1: LLOS is detected.							

## STAT1 - Status Register 1

7DFH ( e: Read ault Value: 00										
7	6	5	4	3	2	1	0			
SAIS_S	LAIS_	S PA_S	IBA_S	IBD_S						
Bit	Name			Descrip	otion					
7	SAIS_S		This bit indicates the SAIS status. ): No SAIS is detected. (default) 1: SAIS is detected.							
6	LAIS_S		This bit indicates the LAIS status. 0: No LAIS is detected. (default) 1: LAIS is detected							
5	PA_S	This bit indicates the PRI 0: The PRBS/ARB patter 1: The PRBS/ARB patter	n is out of synchroniz	zation. (default)						
4 - 2	-	Reserved.								
1	IBA_S	This bit indicates the acti 0: No activate IB code is 1: Activate IB code is det more than 5.1 sec. when	detected. (default) ected for more than		FOLP bit (b3, LOOF	P,) is '0' or activate I	B code is detected			
0	IBD_S	This bit indicates the dea 0: No deactivate IB code 1: Deactivate IB code is of is '0' or deactivate IB cod	is detected. (default) letected for more tha	n 40 ms (in T1/J1 ma			OLP bit (b3, LOOF			

## INTS0 - Interrupt Status Register 0

7	6	5	4	3	2	1	0		
DAC_IS	TJA_IS	S RJA_IS	TOC_IS	TCKLOS_IS	TLOS_IS	SLOS_IS	LLOS_IS		
Bit	Name			Descrip	tion				
7	DAC_IS	This bit indicates the interrupt status of the waveform amplitude overflow. 0: No waveform amplitude overflow interrupt is generated; or a '1' is written to this bit. (default) 1: Waveform amplitude overflow interrupt is generated and is reported by the INT pin.							
6	TJA_IS	This bit indicates the inter 0: No TJA FIFO overflow 1: TJA FIFO overflow or u	rupt status of the T. or underflow interru	JA FIFO overflow or upt is generated; or a	Inderflow. 1' is written to this	bit. (default)			
5	RJA_IS	This bit indicates the interrupt status of the RJA FIFO overflow or underflow. 0: No RJA FIFO overflow or underflow interrupt is generated; or a '1' is written to this bit. (default) 1: RJA FIFO overflow or underflow interrupt is generated and is reported by the INT pin.							
4	TOC_IS	This bit indicates the interrupt status of the Line Driver TOC. 0: No TOC interrupt is generated; or a '1' is written to this bit. (default) 1: TOC interrupt is generated and is reported by the INT pin. When the TOC_IES bit (b4, INTES,) is '0', a transition from '0' '1' on the TOC_S bit (b4, STAT0,) set this bit to '1'; when the TOC_IES bit (b4, INTES,) is '1', any transition (from '0' to '1 from '1' to '0') on the TOC_S bit (b4, STAT0,) set this bit to '1'.							
3	TCKLOS_IS	This bit indicates the inter 0: No TCLKn missing inter 1: TCLKn missing interrup tion from '0' to '1' on the T sition (from '0' to '1' or fro	rrupt is generated; ot is generated and CKLOS_S bit (b3, S	or a '1' is written to th is reported by the INT STAT0,) set this bit t	pin. When the TCH o '1'; when the TCK	LOS_IES bit (b3, IN			
2	TLOS_IS	This bit indicates the inter 0: No TLOS interrupt is g 1: TLOS interrupt is gene to '1' on the TLOS_S bit '1' or from '1' to '0') on the	rupt status of TLOS enerated; or a '1' is rated and is reporte (b2, STAT0,) set tl e TLOS_S bit (b2, S	5. written to this bit. (de d by the INT pin. Wh his bit to '1'; when the TATO,) set this bit to	fault) en the TLOS_IES t e TLOS_IES bit (b2	bit (b2, INTES,) is			
1	SLOS_IS	<ul> <li>'1' or from '1' to '0') on the TLOS_S bit (b2, STAT0,) set this bit to '1'.</li> <li>This bit indicates the interrupt status of the SLOS.</li> <li>0: No SLOS interrupt is generated; or a '1' is written to this bit. (default)</li> <li>1: SLOS interrupt is generated and is reported by the INT pin. When the LOS_IES bit (b1, INTES,) is '0', a transition from '1' on the SLOS_S bit (b1, STAT0,) set this bit to '1'; when the LOS_IES bit (b1, INTES,) is '1', any transition (from '0' to from '1' to '0') on the SLOS_S bit (b1 STAT0,) set this bit to '1'.</li> </ul>							
0	LLOS_IS	from '1' to '0') on the SLOS_S bit (b1, STAT0,) set this bit to '1'.							

## INTS1 - Interrupt Status Register 1

7	6	5	4	3	2	1	0		
SAIS_IS	LAIS_I	S PA_IS	· .			IBA_IS	IBD_IS		
Bit	Name			Descrip	otion				
7	SAIS_IS	This bit indicates the interr 0: No SAIS interrupt is ger 1: SAIS interrupt is genera on the SAIS_S bit (b7, ST/ '1' to '0') on the SAIS_S bi	herated; or a '1' is wr ted and is reported b AT1,) set this bit to	ritten to this bit. (def by the INT pin. Whe '1'; when the AIS_I	n the AIS_IES bit (I				
6	LAIS_IS	This bit indicates the interrupt status of the LAIS. 0: No LAIS interrupt is generated; or a '1' is written to this bit. (default) 1: LAIS interrupt is generated and is reported by the INT pin. When the AIS_IES bit (b6, INTES,) is '0', a transition from '0' to on the LAIS_S bit (b6, STAT1,) set this bit to '1'; when the AIS_IES bit (b6, INTES,) is '1', any transition (from '0' to '1' or fi '1' to '0') on the LAIS_S bit (b6, STAT1,) set this bit to '1'.							
5	PA_IS	This bit indicates the interr 0: No PRBS/ARB pattern s 1: PRBS/ARB pattern sync is '0', a transition from '0' transition (from '0' to '1' or	synchronization inter chronization interrup to '1' on the PA_S b	rupt is generated; c t is generated and i bit (b5, STAT1,) se	or a '1' is written to s reported by the $\overline{I}$ et this bit to '1'; wh	NT pin. When the PA en the PA_IES bit (b!			
4 - 2	-	Reserved.		<u> </u>					
1	IBA_IS	This bit indicates the interr 0: No activate IB code inte 1: Activate IB code interru from '0' to '1' on the IBA_S '1' or from '1' to '0') on the	rrupt is generated; o pt is generated and 5 bit (b1, STAT1,) s	or a '1' is written to t is reported by the set this bit to '1'; whe	INT pin. When the en the IB_IES bit (b				
0	IBD_IS	This bit indicates the interr 0: No deactivate IB code in 1: Deactivate IB code inter from '0' to '1' on the IBD_S	nterrupt is generated rupt is generated an	d; or a '1' is written to nd is reported by the	e INT pin. When the				

## INTS2 - Interrupt Status Register 2

	ł <i>(CH0)</i> Vrite	Н, 322Н, 362Н, 3A2Н, <i>(CH</i> 9	~~CHT3)							
7	6	5	4	3	2	1	0			
-	•	SBPV_IS LBPV_IS SEXZ_IS LEXZ_IS ERR_IS								
Bit	Name			Descrip	ption					
7 - 6	-	Reserved.	Reserved.							
5	SBPV_IS	0: No SBPV interrupt is ge	This bit indicates the interrupt status of the SBPV. 0: No SBPV interrupt is generated; or a '1' is written to this bit. (default) 1: SBPV interrupt is generated and is reported by the INT pin.							
4	LBPV_IS	0: No LBPV interrupt is ge	This bit indicates the interrupt status of the LBPV. 0: No LBPV interrupt is generated; or a '1' is written to this bit. (default) 1: LBPV interrupt is generated and is reported by the INT pin.							
3	SEXZ_IS	This bit indicates the inter 0: No SEXZ interrupt is ge 1: SEXZ interrupt is gener	enerated; or a '1' is v	vritten to this bit. (de	efault)					
2	LEXZ_IS	This bit indicates the inter 0: No LEXZ interrupt is ge 1: LEXZ interrupt is gener	nerated; or a '1' is v	ritten to this bit. (de	efault)					
1	ERR_IS	This bit indicates the inter 0: No PRBS/ARB error int 1: PRBS/ARB error intern	errupt is generated;	or a '1' is written to						
0	CNTOV_IS	This bit indicates the inter 0: No ERRCH or ERRCL 1: ERRCH and ERRCL re	register overflow inte	errupt is generated;	or a '1' is written to th					

#### ERRCL - Error Counter Low-Byte Register

	263H, 2A3H, 2E3H <i>(CH0)</i>	, 123H, 163H, 1A3H, 1E3 , 323H, 363H, 3A3H, <i>(CH</i>	• •				
7	6	5	4	3	2	1	0
ERRC7	ERRC6	ERRC5	ERRC4	ERRC3	ERRC2	ERRC1	ERRC0
Bit	Name			Descrip	otion		
7 - 0		These bits, together with updated automatically or error counting; otherwise	manually, as determ	nined by the CNT_N			,

#### ERRCH - Error Counter High-Byte Register

	264H, 2A4H, 2E4H <i>CH0)</i>	, 124Н, 164Н, 1A4Н, 1E4ŀ , 324Н, 364Н, 3A4Н, <i>(СН</i> 9					
7	6	5	4	3	2	1	0
ERRC15	ERRC1	4 ERRC13	ERRC12	ERRC11	ERRC10	ERRC9	ERRC8
Bit	Name			Descrip	otion		
7 - 0	ERRC[15:8]	(Refer to the description o	f the ERRCL registe	er.)			

#### JM - Jitter Measurement Configuration For Channel 0 Register

/pe: Read / Wr efault Value: 00										
7	6	5	4	3	2	1	0			
-	·	•	JM_STOP JM_MD							
Bit	Name		Description							
7 - 3	-	Reserved.	Reserved.							
2	JM_STOP	This bit is valid only when A transition from '0' to '1' This bit must be cleared	on this bit updates th	he JIT_PH, JIT_PL	_ and JIT_NH, JIT_NL	. registers.				
1	JM_MD	This bit selects the jitter r 0: The period is determin 1: The period is one seco	ed manually by settin		bit (b2, JM). (default)					
0	JM_BW	This bit selects the band 0: 10 Hz ~ 40 KHz (in T1 1: 8 KHz ~ 40 KHz (in T1	/J1 mode) / 20 Hz ~	100 KHz (in E1 ma						

#### JIT\_PL - Positive Peak Jitter Measurement Low-Byte Register

Type:	ss: 7E6H Read It Value: 00	Н											
	7	6		5	4	3	2	1	0				
	JIT_P7	JIT_P6		JIT_P5	JIT_P4	JIT_P3	JIT_P2	JIT_P1	JIT_P0				
	Bit Name			Description									
7	7 - 0	JIT_P[7:0]	These bits, together with the JIT_P[11:8] bits, reflect the greatest positive peak value of the demodulated jitter signal which is measured by channel 0. They are updated automatically or manually, as determined by the JM_MD bit (b1, JM). They should be read in the next round of jitter measurement; otherwise, they will be overwritten. The relationship between the greatest positive peak value and the indication in these bits is: Positive Peak = [JIT_PH, JIT_PL] / 16 (UIpp)										

#### JIT\_PH - Positive Peak Jitter Measurement High-Byte Register

Address: 7E7H Type: Read Default Value: 0											
7	6	5	4	3	2	1	0				
-		•		JIT_P	11 JIT_P10	JIT_P9	JIT_P8				
Bit	Name		Description								
7 - 4	-	Reserved.									
3 - 0	JIT_P[11:8]	(Refer to the description of the JIT_PL register.)									

### JIT\_NL - Negative Peak Jitter Measurement Low-Byte Register

Address: 7E8H Type: Read Default Value: 0	0Н						
7	6	5	4	3	2	1	0
JIT_N7	JIT_N6	JIT_N5	JIT_N4	JIT_N3	JIT_N2	JIT_N1	JIT_N0
Bit	Name			Descri	ption		
7 - 0	JIT_N[7:0]	These bits, together with measured by channel 0. read in the next round of The relationship between Negative Peak = [JIT_NH	They are updated a jitter measurement the greatest nega	automatically or manu t; otherwise, they will tive peak value and th	ually, as determined be overwritten.	by the JM_MD bit (b	

### JIT\_NH - Negative Peak Jitter Measurement High-Byte Register

Type:	ss: 7E9H Read t Value: 00	ЭH												
	7	6		5		4		3		2		1	0	
	-					-		JIT_N11	Γ	JIT_N10	Ι	JIT_N9	JIT_N8	
	Bit	Name						Des	criptio	on				
7	7 - 4	-	Reserved.											
3	3 - 0	JIT_N[11:8]	(Refer to the	descript	ion of the	JIT_NL re	egister.)							

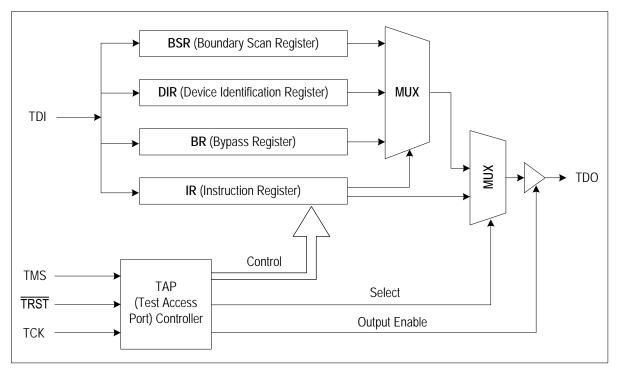
## 6 JTAG

The IDT82P2916 supports the digital Boundary Scan Specification as described in the IEEE 1149.1 standards.

The boundary scan architecture consists of data and instruction registers plus a Test Access Port (TAP) controller. The control of the TAP is achieved through signals applied to the Test Mode Select (TMS) and Test Clock (TCK) input pins. Data is shifted into the registers via the Test

Data Input (TDI) pin, and shifted out of the registers via the Test Data Output (TDO) pin. Both TDI and TDO are clocked at a rate determined by TCK.

The JTAG boundary scan registers include BSR (Boundary Scan Register), DIR (Device Identification Register), BR (Bypass Register) and IR (Instruction Register). These will be described in the following pages. Refer to Figure-45 for architecture.



### Figure-45 JTAG Architecture

### 6.1 JTAG INSTRUCTION REGISTER (IR)

The IR with instruction decode block is used to select the test to be executed or the data register to be accessed or both.

The instructions include: EXTEST, SAMPLE/PRELOAD, IDCODE, BYPASS, CLAMP and HIGHZ.

### 6.2 JTAG DATA REGISTER

### 6.2.1 DEVICE IDENTIFICATION REGISTER (IDR)

The IDR can be set to define the Version, the Part Number, the Manufacturer Identity and a fixed bit.

### 6.2.2 BYPASS REGISTER (BYP)

The BYP consists of a single bit. It can provide a serial path between the TDI input and the TDO output. Bypassing the BYR will reduce test access times.

### 6.2.3 BOUNDARY SCAN REGISTER (BSR)

The bidirectional ports interface to 2 boundary scan cells:

- In cell: The input cell is observable only.
- Out cell: The output cell is controllable and observable.

### 6.3 TEST ACCESS PORT (TAP) CONTROLLER

The TAP controller is a 16-state synchronous state machine. The states include: Test Logic Reset, Run-Test/Idle, Select-DR-Scan, Capture-DR, Shift-DR, Exit1-DR, Pause-DR, Exit2-DR, Update-DR, Select-IR-Scan, Capture-IR, Shift-IR, Exit1-IR, Pause-IR, Exit2-IR and Update-IR.

Figure-46 shows the state diagram. Note that the figure contains two main branches to access either the data or instruction registers. The value shown next to each state transition in this figure states the value present at TMS at each rising edge of TCK.

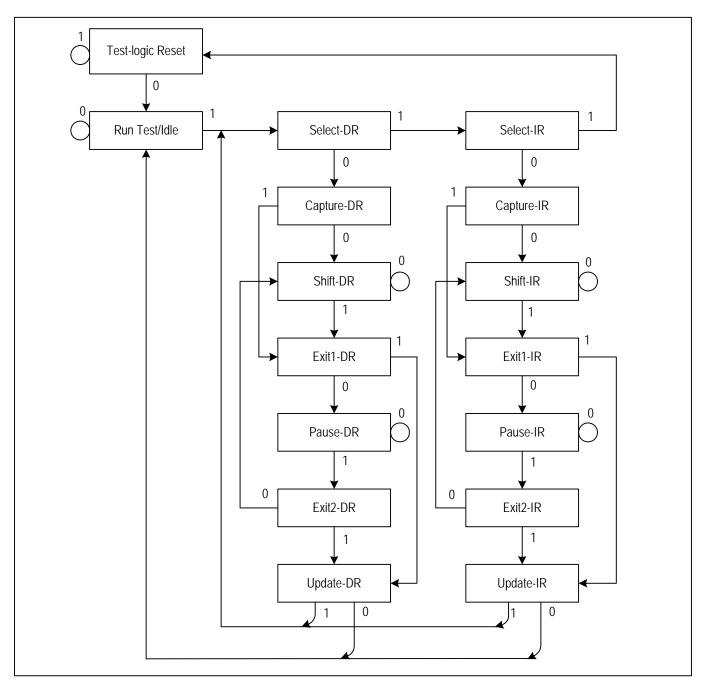


Figure-46 JTAG State Diagram

## 7 THERMAL MANAGEMENT

The device is designed to operate over the industry temperature range -40°C ~ +85°C. To ensure the functionality and reliability of the device, the maximum junction temperature,  $T_{jmax}$ , should not exceed 125°C. In some applications, the device will consume more power and a thermal solution should be provided to ensure the junction temperature  $T_j$  does not exceed  $T_{jmax}$ . Below is a table listing thermal data for the IDT82P2916.

Package	θ <sub>JC</sub> (°C/W) <sup>1</sup>	$\theta_{JB}$ (°C/W) <sup>2</sup>	$\theta_{JA}$ (°C/W) <sup>3</sup>	Airflow (m/s)
			23.7	0
			18.7	1
484-pin BF	4.2	5.3	17.0	2
			16.1	3
			15.5	4
			15.0	5

Note:

1. Junction-to-Case Thermal Resistance

2. Junction-to-Board Thermal Resistance

3. Junction-to-Ambient Thermal Resistance

### 7.1 JUNCTION TEMPERATURE

Junction temperature  $T_j$  is the temperature of package typically at the geographical center of the chip where the device's electrical circuits are. It can be calculated as follows:

Equation 1:  $T_i = T_A + P^* \theta_{JA}$ 

Where:

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance of the package

T<sub>i</sub> = Junction Temperature

T<sub>A</sub> = Ambient Temperature

P = Device Power Consumption

For the IDT82P2916, the above values are:

 $\theta_{JA} = 23.7 \text{ °C/W}$  (when airflow rate is 0 m/s. See the above table )

*T<sub>jmax</sub> = 125* °C

 $T_{A} = -40 \,^{\circ}\text{C} \sim 85 \,^{\circ}\text{C}$ 

*P* = *Refer to Section 8.3 Device Power Consumption and Dissipation* (*Typical*) 1

### 7.2 EXAMPLE OF JUNCTION TEMPERATURE CAL-CULATION

Assume:

$$T_A = 85 \ ^\circ C$$

 $\theta_{JA}$  = 23.7 °C/W (airflow: 0 m/s)

P = 1.46 W (E1 120  $\Omega$ , 100% ones, External Impedance matching)

The junction temperature T<sub>i</sub> can be calculated as follows:

 $T_i = T_A + P^* \theta_{,IA} = 85 \ ^{\circ}C + 1.46 \ W \ X \ 23.7 \ ^{\circ}C/W = 119.6 \ ^{\circ}C$ 

The junction temperature of *119.6* °C is below the maximum junction temperature of 125 °C, so no extra heat enhancement is required.

In some operation environments, the calculated junction temperature might exceed the maximum junction temperature of 125 °C and an external thermal solution such as a heatsink is required.

### 7.3 HEATSINK EVALUATION

A heatsink is expanding the surface area of the device to which it is attached.  $\theta_{JA}$  is now a combination of device case and heatsink thermal resistance, as the heat flowing from the die junction to ambient goes through the package and the heatsink.  $\theta_{JA}$  can be calculated as follows:

Equation 2: 
$$\theta_{JA} = \theta_{JC} + \theta_{HA}$$

Where:

 $\theta_{JC}$  = Junction-to-Case (heatsink) Thermal Resistance  $\theta_{HA}$  = Heatsink-to-Ambient Thermal Resistance

For the IDT82P2916,  $\theta_{IC}$  is 4.2 °C/W.

 $\theta_{HA}$  determines which heatsink can be selected to ensure the junction temperature does not exceed  $T_{jmax}$ . According to Equation 1 and 2, the heatsink-to-ambient thermal resistance  $\theta_{HA}$  can be calculated as follows:

Equation 3:  $\theta_{HA} = (T_j - T_A) / P - \theta_{JC}$ 

Assume:

$$T_{j} = 125 \ ^{\circ}C \ (T_{jmax})$$

$$T_{A} = 85 \ ^{\circ}C$$

$$P = 2.72 \ W \ (E1 \ 75 \ \Omega, \ 100\% \ ones, \ Fully \ Internal \ Impedance \ matching)$$

$$\theta_{1C} = 4.2 \ ^{\circ}C/W$$

The Heatsink-to-Ambient thermal resistance  $\theta_{\text{HA}}$  can be calculated as follows:

 $\theta_{HA} = (125 \ ^{\circ}C - 85 \ ^{\circ}C) / 2.72 \ W - 4.2 \ ^{\circ}C/W = 10.51 \ ^{\circ}C/W$ 

That is, if a heatsink whose heatsink-to-ambient thermal resistance  $\theta_{HA}$  is below or equal to  $10.51\ ^\circ\text{C/W}$  is used in such operation environment, the junction temperature will not exceed the maximum junction temperature.

# 8 PHYSICAL AND ELECTRICAL SPECIFICATIONS

### 8.1 ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Min	Мах	Unit
VDDD	Digital Core Power Supply	-0.5	2.2	V
VDDA	Analog Core Power Supply	-0.5	4.6	V
VDDIO	I/O Power Supply	-0.5	4.6	V
VDDT0~15	Power Supply for Transmitter Driver	-0.5	4.6	V
VDDR0~15	Power Supply for Receiver	-0.5	4.6	V
	Input Voltage, Any Digital Pin	GND - 0.5	6	V
V <sub>in</sub>	Input Voltage, Any RTIP and RRING pin <sup>1</sup>	GND - 0.5	VDDR + 0.5	V
	ESD Voltage, Any Pin <sup>2</sup>	2000		V
	Transient Latch-up Current, Any Pin		100	mA
l <sub>in</sub>	Input Current, Any Digital Pin <sup>3</sup>	-10	10	mA
	DC Input Current, Any Analog Pin <sup>3</sup>		±100	mA
Pd	Maximum Power Dissipation in Package		1.68 <sup>4</sup>	W
Tj	Junction Temperature		125	°C
T <sub>s</sub>	Storage Temperature	-65	+150	°C

Note:

1. Reference to ground.

2. Human body model.

3. Constant input current.

4. If device power consumption exceeds this value, a heatsink or a fan must be used. Refer to Chapter 7 Thermal Management.

Caution:

Exceeding the above values may cause permanent damage. Functional operation under these conditions is not implied. Exposure to absolute maximum rating conditions for extended period may affect device reliability.

### 8.2 RECOMMENDED OPERATING CONDITIONS

Symbol	Parameter	Min	Тур.	Мах	Unit	
T <sub>op</sub>	Operating Temperature Range	-40		85 <sup>1</sup>	°C	
VDDIO	Digital I/O Power Supply	3.13	3.3	3.47	V	
VDDA	Analog Core Power Supply	3.13	3.3	3.47	V	
VDDD	Digital Core Power Supply	1.71	1.8	1.89	V	
VDDT	Power Supply for Transmitter Driver	3.13	3.3	3.47	V	
VDDR	Power Supply for Receiver	3.13	3.3	3.47	V	
V <sub>IL</sub>	Input Low Voltage	-0.5		0.8	V	
V <sub>IH</sub>	Input High Voltage	2.0		VDDIO+0.5	V	

1. An external thermal solution such as heatsink may be required depending on the mode of operation. Refer to Chapter 7 Thermal Management.

### 8.3 DEVICE POWER CONSUMPTION AND DISSIPATION (TYPICAL)<sup>1</sup>

		Total Consumption (W)				vice Power Dis mal Considera		Per-Channel Power Down Saving (mW) <sup>2</sup>			
Mode	Parameter	1.8 V	3.3 V	Total	Fully Internal R120IN=1 <sup>3</sup>	Internal Internal		Fully Internal R120IN=1 <sup>3</sup>	Partially Internal R120IN=0 <sup>4</sup>	External <sup>5</sup>	
E1/120 Ω	PRBS	0.18	1.73	1.91	1.91	1.48	1.20	80	60	40	
	100% ones	0.18	2.31	2.49	2.49	1.87	1.46	130	90	70	
E1/75 Ω	PRBS	0.18	1.86	2.04	2.04	1.77	0.98	90	60	50	
	100% ones	0.18	2.54	2.72	2.72	2.33	1.32	150	120	80	
T1/100 Ω	QRSS	0.14	1.93	2.07	2.07	1.66	1.57	100	80	60	
	100% ones	0.14	2.79	2.93	2.93	2.32	2.24	160	120	90	
J1/110 Ω	QRSS	0.14	1.87	2.01	2.01	1.59	1.56	100	70	60	
	100% ones	0.14	2.64	2.78	2.78	2.17	2.17	150	110	90	

Note:

1. Test conditions: VDDx (typical) at 25 °C operating temperature (ambient).

2. The R\_OFF bit (b5, RCF0,...) and T\_OFF bit (b5, TCF0,...) are set to '1' to enable per-channel power down.

3. The transmitter is in Internal Impedance Matching mode and the receiver is in Fully Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '1'. And the T\_TERM[2:0] bits (b2-0, TCF0,...) and R\_TERM[2:0] bits (b2-0, RCF0,...) are set according to different cable conditions.

4. The transmitter is in Internal Impedance Matching mode and the receiver is in Partially Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '0'. And the T\_TERM[2:0] bits (b2~0, TCF0,...) and R\_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

5. For E1 mode, both the transmitter and the receiver are in External Impedance Matching mode. That is, the T\_TERM[2:0] bits (b2-0, TCF0,...) are set to '111' and the R\_TERM[2:0] bits (b2-0, RCF0,...) are set to '1xx'. For T1/J1 mode, as the transmitter External Impedance Matching mode is not supported, the transmitter is in Internal Impedance Matching mode and the receiver is in Internal Impedance Matching mode. That is, the T\_TERM[2:0] bits (b2-0, TCF0,...) are set to '1xx'. For T1/J1 mode, as the T\_TERM[2:0] bits (b2-0, TCF0,...) are set according to different cable conditions and the R\_TERM[2:0] bits (b2-0, RCF0,...) are set to '1xx'.

## 8.4 DEVICE POWER CONSUMPTION AND DISSIPATION (MAXIMUM)<sup>1</sup>

Mada		Т	otal Consumption (	W)	Total Device Power Dissipation (for Thermal Consideration, W)			
Mode	Parameter	1.89 V	3.47 V	Total	Fully Internal R120IN=1 <sup>2</sup>	Partially Internal R120IN=0 <sup>3</sup>	External <sup>4</sup>	
E1/120 Ω	PRBS	0.22	1.82	2.04	2.04	1.68	1.40	
	100% ones	0.22	2.43	2.65	2.65	2.08	1.67	
E1/75 Ω	PRBS	0.22	1.95	2.17	2.17	1.98	0.98	
	100% ones	0.22	2.67	2.89	2.89	2.53	1.37	
T1/100 Ω	QRSS	0.17	2.03	2.20	2.20	1.84	1.65	
	100% ones	0.17	2.93	3.10	3.10	2.53	2.34	
J1/110 Ω	QRSS	0.17	1.96	2.13	2.13	1.78	1.63	
	100% ones	0.17	2.78	2.95	2.95	2.38	2.28	

Note:

1. Test conditions: VDDx (maximum) at 85 °C operating temperature (ambient).

2. The transmitter is in Internal Impedance Matching mode and the receiver is in Fully Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '1'. And the T\_TERM[2:0] bits (b2~0, TCF0,...) and R\_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

3. The transmitter is in Internal Impedance Matching mode and the receiver is in Partially Internal Impedance Matching mode. That is, the R120IN bit (b4, RCF0,...) is set to '0'. And the T\_TERM[2:0] bits (b2~0, TCF0,...) and R\_TERM[2:0] bits (b2~0, RCF0,...) are set according to different cable conditions.

4. For E1 mode, both the transmitter and the receiver are in External Impedance Matching mode. That is, the T\_TERM[2:0] bits (b2-0, TCF0,...) are set to '111' and the R\_TERM[2:0] bits (b2-0, RCF0,...) are set to '1xx'. For T1/J1 mode, as the transmitter External Impedance Matching mode is not supported, the transmitter is in Internal Impedance Matching mode and the receiver is in Internal Impedance Matching mode. That is, the T\_TERM[2:0] bits (b2-0, TCF0,...) are set according to different cable conditions and the R\_TERM[2:0] bits (b2-0, RCF0,...) are set to '1xx'.

### 8.5 D.C. CHARACTERISTICS

@ TA = -40 to +85 °C, VDDIO = 3.3 V  $\pm$  5%, VDDD = 1.8 V  $\pm$  5%

Symbol	Parameter	Min	Тур.	Max	Unit	Test Conditions
V <sub>OL</sub>	Output Low Voltage			0.40	V	VDDIO = 3.13 V, I <sub>OL</sub> = 4 mA, 8 mA
V <sub>OH</sub>	Output High Voltage	2.4		VDDIO	V	VDDIO = 3.13 V, I <sub>OH</sub> = 4 mA, 8 mA
$V_{T+}$	Schmitt Trigger Input Low to High Threshold	1.47			V	
V <sub>T-</sub>	Schmitt Trigger Input High to Low Threshold			0.89	V	
R <sub>pu</sub>	Internal Pull-up /Pull-down Resistor	50	70	115	KΩ	
IIL	Input Low Current	-1	0	+1	μA	V <sub>IL</sub> = GNDD
I <sub>IH</sub>	Input High Current	-1	0	+1	μA	V <sub>IH</sub> = VDDIO
C <sub>in</sub>	Input Digital Pin Capacitance			10	pF	
C <sub>out</sub>	Output Load Capacitance			50	pF	
C <sub>out</sub>	Output Load Capacitance (bus pins)			100	pF	
I <sub>ZL</sub>	Leakage Current of Digital Output in High-Z mode	-10		10	μA	GNDIO < V <sub>O</sub> < VDDIO
Z <sub>OH</sub>	Output High-Z on TTIPn, TRINGn pins	10			KΩ	

### 8.6 E1 RECEIVER ELECTRICAL CHARACTERISTICS

Paran	neter	Min	Тур.	Мах	Unit	Test Conditions
Receiver Sensitivity tial mode with Cable			15		dB	with Nominal Pulse Amplitude of 3.0 V for 120 $\Omega$
	Receiver Sensitivity of Receive Single Ended mode with Cable Loss @ 1024 kHz		12		dB	and 2.37 V for 75 $\Omega$ termination, adding -18 dB interference signal.
Signal to Noise Interference Margin		-16			dB	@cable loss 0-6 dB
Analog LOS Level (Normal Mode)	ALOS[2:0] 000 001 (default) 010 011 100 101 110 111		0.5 0.7 0.9 1.2 1.4 1.6 1.8 2.0		V <sub>pp</sub>	In Differential mode, measured between RTIP and RRING pins. In Singled Ended mode, measured between RTIP and GNDA pins Refer to Table-17 for LLOS Criteria Declare and Clear.
	LOS hysteresis		0.25			
Analog LOS Level (Line Monitor Mode)	ALOS[2:0] 000 001 (default) 010 011 1xx (reserved)		1.0 1.4 1.8 2.2		V <sub>pp</sub>	Measured on the line with the monitor gain set by the MG[1:0] bits (b1~0, RCF2,) equal to the resistive attenuation. Refer to Table-17 for LLOS Criteria Declare and Clear.
	LOS hysteresis		0.41			
Allowable Consecutiv G.775 I.431 / ETSI300233			32 2048			
LOS Reset		12.5			% ones	G.775, ETSI 300233
Receive Intrinsic Jitte	r			0.05	U.I.	JA disabled; wide band
Input Jitter Tolerance: 1 Hz ~ 20 Hz 20 Hz ~ 2.4 KHz 18 KHz ~ 100 KHz		80 10 0.6 (Refer to Figure- 50)			U.I. U.I. U.I.	G.823, with 6 dB Cable Attenuation
Receiver Differential	nput Impedance		2.6		KΩ	
Receiver Common ance to GND	Mode Input Imped-		1.6		KΩ	@1024 KHz; Rx port is high-Z



Parameter	Min	Тур.	Мах	Unit	Test Conditions
Receiver Single Ended mode Input Impedance to GND		3.1		KΩ	The RRINGn pins are open.
Receive Return Loss: 51 KHz ~ 102 KHz 102 KHz ~ 2.048 MHz 2.048 MHz ~ 3.072 MHz	12 18 14			dB dB dB	G.703
Receive Path Delay: Single Rail Dual Rail NRZ Dual Rail RZ			6.6 1.8 1.5	U.I. U.I. U.I.	JA Disabled

### 8.7 T1/J1 RECEIVER ELECTRICAL CHARACTERISTICS

Param	neter	Min	Тур.	Мах	Unit	Test Conditions
Receiver Sensitivity of tial mode with Cable L			15		dB	with Nominal Pulse Amplitude of 3.0 V for 100 $\Omega$ ter-
Receiver Sensitivity Ended mode with Cab			12		dB	<ul> <li>mination, adding -18 dB interference signal.</li> </ul>
Signal to Noise Interfe	erence Margin	-16			dB	
Analog LOS Level (Normal Mode)	ALOS[2:0] 000 001 (default) 010 011 100 101 110 111 LOS hysteresis		0.5 0.7 0.9 1.2 1.4 1.6 1.8 2.0 0.25		V <sub>pp</sub>	In Differential mode, measured between RTIP and RRING pins. In Singled Ended mode, measured between RTIP and GNDA pins Refer to Table-17 for LLOS Criteria Declare and Clear.
Analog LOS Level (Line Monitor Mode)	ALOS[2:0] 000 001 (default) 010 011 1xx (reserved)		1.0 1.4 1.8 2.2		V <sub>pp</sub>	Measured on the line with the monitor gain set by the MG[1:0] bits (b1~0, RCF2,) equal to the resistive attenuation. Refer to Table-17 for LLOS Criteria Declare and Clear.
	LOS hysteresis		0.41			
Allowable Consecutive T1.231 - 1993 I.431	e Zeros before LOS:		175 1544			
LOS Reset		12.5			% ones	G.775, ETSI 300233
Receive Intrinsic Jitter	-			0.05	U.I.	JA disabled; Wide band
Input Jitter Tolerance: 0.1 Hz ~ 1 Hz 4.9 Hz ~ 300 Hz 10 KHz ~ 100 KHz		138.0 28.0 0.4			U.I. U.I. U.I.	AT&T62411
Receiver Differential II	nput Impedance		3.1		KΩ	
Receiver Common M ance to GND	Mode Input Imped-		2.2		ΚΩ	— @772 KHz; Rx port is high-Z
Receiver Single Ended mode Input Impedance to GND			4		ΚΩ	The RRINGn pins are open.
Receive Return Loss: 39 KHz ~ 77 KHz 77 KHz ~ 1.544 MH 1.544 MHz ~ 2.316	łz	20 20 20			dB dB dB	G.703
Receive Path Delay: Single Rail Dual Rail NRZ Dual Rail RZ			6.5 2.5 1.4		U.I. U.I. U.I.	JA Disabled

### 8.8 E1 TRANSMITTER ELECTRICAL CHARACTERISTICS

Parameter		Min	Тур.	Max	Unit	Test Conditions
Output Pulse Amplitude: E1, 75 $\Omega$ load E1, 120 $\Omega$ load		2.14 2.7	2.37 3.0	2.60 3.3	V V	Differential Line Inter- face mode
Zero (Space) Level: E1, 75 $\Omega$ load E1, 120 $\Omega$ load		-0.237 -0.3		+0.237 0.3	V V	Differential Line Inter- face mode
Transmit Amplitude Variation with Supp	-1		+1	%		
Difference between Pulse Sequences f (T1.102)	for 17 consecutive pulses			200	mV	
Output Pulse Width at 50% of Nominal	Amplitude	232	244	256	ns	
Ratio of the Amplitudes of Positive an Center of the Pulse Interval (G.703)	d Negative Pulses at the	0.95		1.05		
Ratio of the Width of Positive and Nega of the Pulse Interval (G.703)	ative Pulses at the Center	0.95		1.05		
Transmit Analog LOS Level (TALOS) (Differential line interface)	TALOS[1:0] 00 01 (default) 10 11		1.2 0.9 0.6 0.4		V <sub>p</sub>	Measured on the TTIP and TRING pins.
	TALOS hysteresis		0.08			
Transmit Analog LOS Level (TALOS) (Single Ended line interface)	TALOS[1:0] 00 01 (default) 10 11 TALOS hysteresis		0.61 0.48 0.32 0.24 0.04		V <sub>p</sub>	Measured on the TTIP pin.
Transmit Return Loss (G.703): 51 KHz ~ 102 KHz 102 KHz ~ 2.048 MHz 2.048 MHz ~ 3.072 MHz		8 14 10	0.04		dB dB dB	Internal Impedance Matching
Intrinsic Transmit Jitter 20 Hz ~ 100 KHz				0.050	U.I.	TCLK is jitter free
Transmit Path Delay: Single Rail Dual Rail NRZ Dual Rail RZ			8.5 4.5 4.4		U.I. U.I. U.I.	JA is disabled
Line Short Circuit Current			100		mAp	Measured on pin

### 8.9 T1/J1 TRANSMITTER ELECTRICAL CHARACTERISTICS

Parameter		Min	Тур.	Max	Unit	Test Conditions
Output Pulse Amplitude		2.4	3.0	3.6	V	Differential Line Interface mode
Zero (Space) Level		-0.15		0.15	V	
Transmit Amplitude Variation with Supp	ly	-1		+1	%	
Difference between Pulse Sequences fr (T1.102)	or 17 consecutive pulses			200	mV	
Output Pulse Width at 50% of Nominal	Amplitude	338	350	362	ns	
Pulse Width Variation at the Half Ampli	tude (T1.102)			20	ns	
Imbalance between Positive and Neg (T1.102)	gative Pulses Amplitude	0.95		1.05		
Output Power Levels (T1.102-1993): @772 KHz @1544 KHz (Referenced to Power a	t 772 KHz)	12.6 -29		17.9	dBm dBm	
Transmit Analog LOS Level (TALOS) (Differential line interface)	TALOS[1:0] 00 01 (default) 10 11 TALOS hysteresis		1.2 0.9 0.6 0.4 0.08		Vp	Measured on the TTIP and TRING pins.
Transmit Analog LOS Level (TALOS) (Single Ended line interface)	TALOS Insteresis TALOS [1:0] 00 01 (default) 10 11 TALOS hysteresis		0.61 0.48 0.32 0.24 0.04		Vp	Measured on the TTIP pin.
Transmit Return Loss (G.703): 39 KHz ~ 77 KHz 77 KHz ~ 1.544 MHz 1.544 MHz ~ 2.316 MHz		8 14 10			dB dB dB	Internal Impedance Matching
Intrinsic Transmit Jitter: 10 Hz ~ 8 KHz 8 KHz ~ 40 KHz 10 Hz ~ 40 KHz Wide Band				0.020 0.025 0.025 0.050	U.I.p-p U.I.p-p U.I.p-p U.I.p-p	TCLK is jitter free
Transmit Path Delay (JA is disabled): Single Rail Dual Rail NRZ Dual Rail RZ			8.2 4.1 4.3		U.I. U.I. U.I.	JA is disabled
Line Short Circuit Current			100		mAp	Measure on pin

### 8.10 TRANSMITTER AND RECEIVER TIMING CHARACTERISTICS

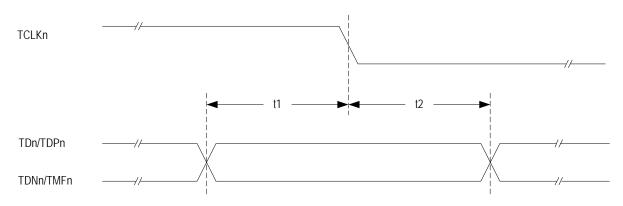
Symbol	Parameter	Min	Тур.	Max	Unit
	MCLK Frequency: E1 T1/J1		2.048 X n 1.544 X n (n = 1 ~ 8)		MHz MHz
	MCLK Tolerance	-100		100	ppm
	MCLK Duty Cycle	30		70	%
ransmit Path		·			
	TCLK Frequency: E1 T1/J1		2.048 1.544		MHz MHz
	TCLK Tolerance	-50		+50	ppm
	TCLK Duty Cycle	10		90	%
t1	Transmit Data Setup Time	40			ns
t2	Transmit Data Hold Time	40			ns
	Delay Time of OE low to Driver High-Z			1	μs
eceive Path			1 1		I
	Clock Recovery Capture Range <sup>1</sup> : E1 T1/J1		+80 / -80 +180 / -180		ppm ppm
	RCLK Duty Cycle <sup>2</sup>	40	50	60	%
t4	RCLK Pulse Width <sup>2</sup> : E1 T1/J1	457 607	488 648	519 689	ns ns
t5	RCLK Pulse Width Low Time: E1 T1/J1	203 259	244 324	285 389	ns ns
t6	RCLK Pulse Width High Time: E1 T1/J1	203 259	244 324	285 389	ns ns
	Rise/Fall Time <sup>3</sup>	20			ns
t7	Receive Data Setup Time: E1 T1/J1	200 200	244 324		ns ns
t8	Receive Data Hold Time: E1 T1/J1	200 200	244 324		ns ns

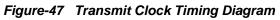
Note:

1. Relative to nominal frequency, MCLK = +100 or -100 ppm.

2. RCLK duty cycle width will vary depending on extent of the received pulse jitter displacement. Maximum and minimum RCLK duty cycles are for worst case jitter conditions (0.2 UI displacement for E1 per ITU G.823).

**3.** For all digital outputs.  $C_{load} = 15 \text{ pF}.$ 





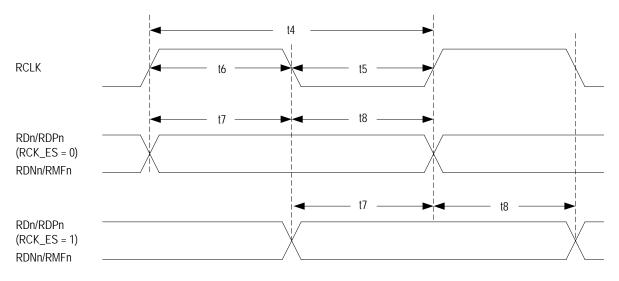


Figure-48 Receive Clock Timing Diagram

### 8.11 CLKE1 TIMING CHARACTERISTICS

Symbol	Parameter	Min	Тур.	Max	Unit
CLKE1 outputs 2.04	48 MHz clock			I	1
t1	CLKE1 Pulse Width		488		ns
t2	CLKE1 Pulse Width High Time	232	244	256	ns
t3	CLKE1 Pulse Width Low Time	232	244	256	ns
t4	LLOS Data Setup Time	217	244	271	ns
t5	LLOS Data Hold Time	217	244	271	ns
CLKE1 outputs 8kH	lz clock				1
t1	CLKE1 Pulse Width		125		μs
t2	CLKE1 Pulse Width High Time	62.4	62.5	62.6	μs
t3	CLKE1 Pulse Width Low Time	62.4	62.5	62.6	μs
t4	LLOS Data Setup Time	62.38	62.5	62.62	μs
t5	LLOS Data Hold Time	62.38	62.5	62.62	μs

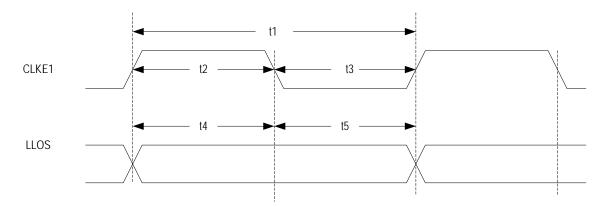


Figure-49 CLKE1 Clock Timing Diagram

### 8.12 JITTER ATTENUATION CHARACTERISTICS

Para	ameter	Min	Тур.	Max	Unit
Jitter Transfer Function Corner (-3 dB) Fi	requency:				
E1, 32/64/128-bit FIFO	JA_BW = 0 JA_BW = 1 JA_BW = 0		6.63 0.87 5		Hz Hz Hz
	JA_BW = 1		1.28		Hz
Jitter Attenuator:					
E1 (G.736) T1/J1 (AT&T pub.62411)	@ 3 Hz @ 40 Hz @ 400 Hz @ 100 KHz @ 1 Hz @ 20 Hz @ 1 KHz @ 1.4 KHz	-0.5 -0.5 +19.5 +19.5 0 0 +33.3 40			dB dB dB dB dB dB dB dB dB dB
Jitter Attenuator Latency Delay: 32-bit FIFO 64-bit FIFO 128-bit FIFO	@ 70 KHz	40	16 32 64		U.I. U.I. U.I.
Input Jitter Tolerance before FIFO O Underflow: 32-bit FIFO 64-bit FIFO 128-bit FIFO	verflow or		28 56 120		U.I. U.I. U.I.

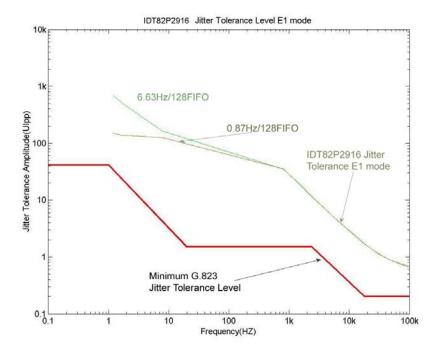


Figure-50 E1 Jitter Tolerance Performance

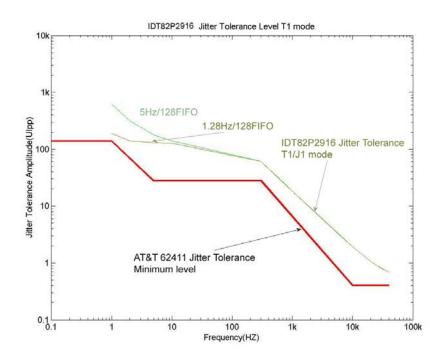


Figure-51 T1/J1 Jitter Tolerance Performance

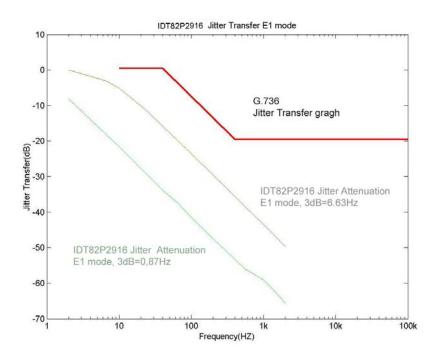


Figure-52 E1 Jitter Transfer Performance

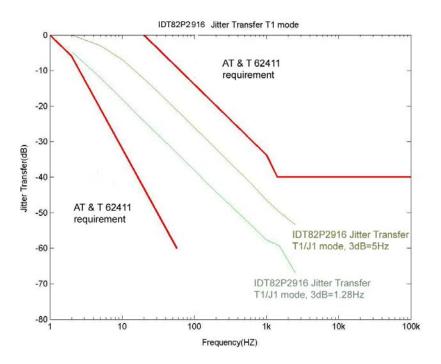


Figure-53 T1/J1 Jitter Transfer Performance

### 8.13 MICROPROCESSOR INTERFACE TIMING

### 8.13.1 SERIAL MICROPROCESSOR INTERFACE

A falling transition on  $\overline{CS}$  indicates the start of a read/write operation, and a rising transition indicates the end of the operation. After  $\overline{CS}$  is set to low, a 5-bit instruction on SDI is input to the device on the rising edge of SCLK. If the MSB is '1', it is a read operation. If the MSB is '0', it is a write operation. Following the instruction, an 11-bit address is clocked in on SDI to specify the register. If the device is in a read operation, the data read from the specified register is output on SDO on the falling edge of SCLK (refer to Figure-54). If the device is in a write operation, the data written to the specified register is input on SDI following the address byte (refer to Figure-55).

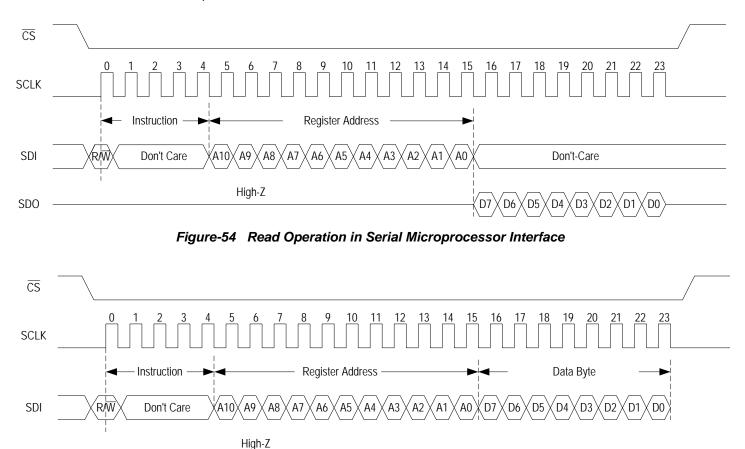


Figure-55 Write Operation in Serial Microprocessor Interface

SDO

# RENESAS

Symbol	Description	Min.	Max.	Units
f <sub>OP</sub>	SCLK Frequency		2.0	MHz
t <sub>CSH</sub>	Minimum CS High Time	100		ns
t <sub>CSS</sub>	CS Setup Time	50		ns
t <sub>CSD</sub>	CS Hold Time	100		ns
t <sub>CLD</sub>	Clock Disable Time	50		ns
t <sub>CLH</sub>	Clock High Time	205		ns
t <sub>CLL</sub>	Clock Low Time	205		ns
t <sub>DIS</sub>	Data Setup Time	50		ns
t <sub>DIH</sub>	Data Hold Time	150		ns
t <sub>PD</sub>	Output Delay		150	ns
t <sub>DF</sub>	Output Disable Time		50	ns

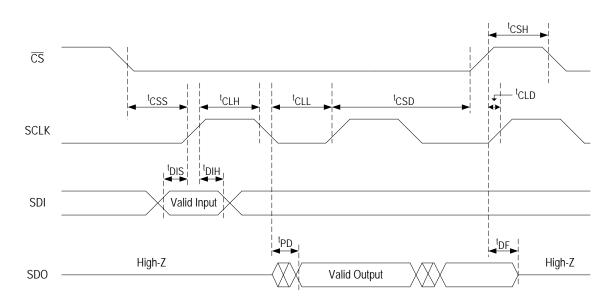
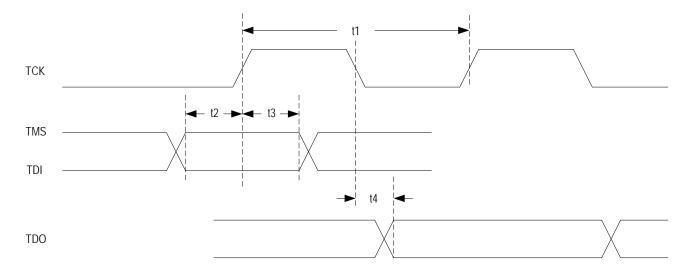


Figure-56 Timing Diagram

### 8.14 JTAG TIMING CHARACTERISTICS

Symbol	Parameter	Min	Тур.	Мах	Unit
t1	TCK Period	100			ns
t2	TMS to TCK Setup Time; TDI to TCK Setup Time	25			ns
t3	TCK to TMS Hold Time; TCK to TDI Hold Time	25			ns
t4	TCK to TDO Delay Time			50	ns





# RENESAS

# Glossary

AIS	_	Alarm Indication Signal
AMI	_	Alternate Mark Inversion
ARB	_	Arbitrary Pattern
B8ZS	_	Binary 8 Zero Substitution
BPV	_	Bipolar Violation
CF	_	Corner Frequency
CV	_	Code Violation
DPLL	_	Digital Phase Locked Loop
EXZ	_	Excessive Zeroes
FIFO	_	First In First Out
HDB3	_	High Density Bipolar 3
HPS	_	Hitless Protection Switching
IB	_	Inband Loopback
LAIS	_	Line Alarm Indication Signal
LBPV	_	Line Bipolar Violation
LEXZ	_	Line Excessive Zeroes
LLOS	_	Line Loss of Signal
LOS	_	Loss Of Signal
NRZ	_	Non-Return to Zero
PBX	_	Private Branch Exchange
PRBS	_	Pseudo Random Bit Sequence
QRSS	_	Quasi-Random Signal Source
RJA	_	Receive Jitter Attenuator
RZ	_	Return to Zero
SAIS	_	System Alarm Indication Signal
SBPV	_	System Bipolar Violation
SDH	_	Synchronous Digital Hierarchy

# RENESAS

SEXZ	_	System Excessive Zeroes
SLOS	_	System LOS
SONET	_	Synchronous Optical Network
TEPBGA	_	Thermally Enhanced Plastic Ball Grid Array
ALT	_	Transmit Jitter Attenuator
TLOS	_	Transmit Loss of Signal
ТОС	_	Transmit Over Current

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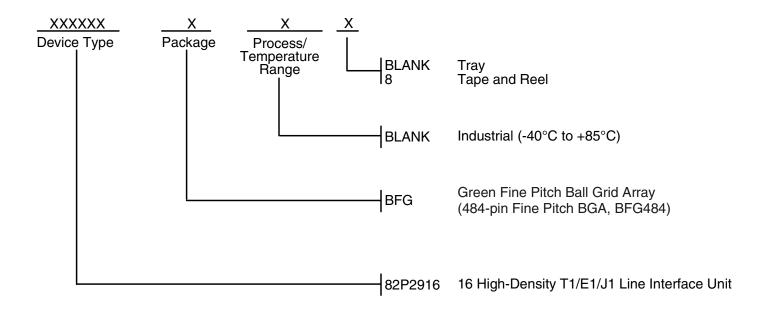
### Т

T1 / E1 / J1 mode selection	
Transmit Over Current (TOC)	

### W

waveform template	28
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## **Data Sheet Document History**

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