

Data Sheet May 3, 2007 FN7330.3

200MHz Slew Enhanced VFA

The EL5100, EL5101, and EL5300 represent high-speed voltage feedback amplifiers based on the current feedback amplifier architecture. This gives the typical high slew rate benefits of a CFA family along with the stability and ease of use associated with the VFA type architecture. This family is available in single, dual, and triple versions, with 200MHz, 400MHz, and 700MHz versions. This family operates on single 5V or ±5V supplies from minimum supply current. The EL5100 and EL5300 also feature an output enable function, which can be used to put the output in to a high-impedance mode. This enables the outputs of multiple amplifiers to be tied together for use in multiplexing applications.

Features

- · Pb-free plus anneal available (RoHS compliant)
- Specified for 5V or ±5V applications
- Power-down to 17μA/amplifier
- -3dB bandwidth = 200MHz
- ±0.1dB bandwidth = 20MHz
- Low supply current = 2.5mA
- Slew rate = 2200V/µs
- Low offset voltage = 4mV max
- Output current = 100mA
- A_{VOL} = 1000
- Diff gain/phase = 0.08%/0.1°

Applications

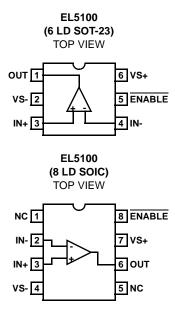
- · Video amplifiers
- PCMCIA applications
- A/D drivers
- Line drivers
- · Portable computers
- High speed communications
- · RGB applications
- Broadcast equipment
- Active filtering

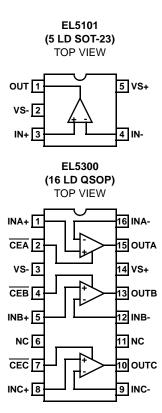
Ordering Information

PART NUMBER	PART MARKING	TAPE AND REEL	PACKAGE	PKG. DWG. #
EL5100IS	5100IS	-	8 Ld SOIC (150 mil)	MDP0027
EL5100IS-T7	5100IS	7"	8 Ld SOIC (150 mil)	MDP0027
EL5100IS-T13	5100IS	13"	8 Ld SOIC (150 mil)	MDP0027
EL5100ISZ (Note)	5100ISZ	-	8 Ld SOIC (150 mil) (Pb-free)	MDP0027
EL5100ISZ-T7 (Note)	5100ISZ	7"	8 Ld SOIC (150 mil) (Pb-free)	MDP0027
EL5100ISZ-T13 (Note)	5100ISZ	13"	8 Ld SOIC (150 mil) (Pb-free)	MDP0027
EL5100IW-T7	у	7" (3k pcs)	6 Ld SOT-23	MDP0038
EL5100IW-T7A	у	7" (250 pcs)	6 Ld SOT-23	MDP0038
EL5101IW-T7	2	7" (3k pcs)	5 Ld SOT-23	MDP0038
EL5101IW-T7A	2	7" (250 pcs)	5 Ld SOT-23	MDP0038
EL5300IU	5300IU	-	16 Ld QSOP (150 mil)	MDP0040
EL5300IU-T7	5300IU	7"	16 Ld QSOP (150 mil)	MDP0040
EL5300IU-T13	5300IU	13"	16 Ld QSOP (150 mil)	MDP0040
EL5300IUZ (Note)	5300IUZ	-	16 Ld QSOP (150 mil) (Pb-free)	MDP0040
EL5300IUZ-T7 (Note)	5300IUZ	7"	16 Ld QSOP (150 mil) (Pb-free)	MDP0040
EL5300IUZ-T13 (Note)	5300IUZ	13"	16 Ld QSOP (150 mil) (Pb-free)	MDP0040

NOTE: Intersil Pb-free plus anneal products employ special Pb-free material sets; molding compounds/die attach materials and 100% matte tin plate termination finish, which are RoHS compliant and compatible with both SnPb and Pb-free soldering operations. Intersil Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.

Pinouts





3

EL5100, EL5101, EL5300

Absolute Maximum Ratings (T_A = +25°C)

Thermal Information

Storage Temperature Range	65°C to +150°C
Ambient Operating Temperature Range	40°C to +85°C
Operating Junction Temperature	+150°C
Pb-free reflow profile	see link below
http://www.intersil.com/pbfree/Pb-FreeReflo	w.asp

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

IMPORTANT NOTE: All parameters having Min/Max specifications are guaranteed. Typical values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$

DC Electrical Specifications $V_S = \pm 5V$, GND = 0V, $T_A = +25$ °C, $V_{CM} = 0V$, $V_{OUT} = 0V$, $V_{ENABLE} = GND$ or OPEN, Unless Otherwise Specified.

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
Vos	Offset Voltage		-4	1	4	mV
TCVOS	Offset Voltage Temperature Coefficient	Measured from T _{MIN} to T _{MAX}		8		μV/°C
IB	Input Bias Current	V _{IN} = 0V	-6	2	6	μΑ
los	Input Offset Current	V _{IN} = 0V	-2.5	0.5	2.5	μΑ
TCI _{OS}	Input Bias Current Temperature Coefficient	Measured from T _{MIN} to T _{MAX}		8		nA/°C
PSRR	Power Supply Rejection Ratio		70	90		dB
CMRR	Common Mode Rejection Ratio	V _{CM} from -3V to +3V	60	75		dB
CMIR	Common Mode Input Range	Guaranteed by CMRR test	-3		+3	V
R _{IN}	Input Resistance	V _{IN} = -3V to +3V	0.7	1.2		ΜΩ
C _{IN}	Input Capacitance			1		pF
I _{S,ON}	Supply Current - Enabled	Per amplifier	2.1	2.5	2.9	mA
I _{S,OFF}	Supply Current - Shut Down	V _S +, per amplifier	-5	0	5	μΑ
		V _S -, per amplifier	5	17	25	μΑ
PSOR	Power Supply Operating Range		3.3		12	V
AVOL	Open Loop Gain	R_L = 1k Ω to GND, V_{OUT} from -2.5V to +2.5V	55	60		dB
V _{OP}	Positive Output Voltage Swing	$R_L = 150\Omega$ to GND	3.2	3.4		V
		$R_L = 1k\Omega$ to GND	3.6	3.8		V
V _{ON}	Negative Output Voltage Swing	$R_L = 150\Omega$ to GND		-3.4	-3.2	V
		$R_L = 1k\Omega$ to GND		-3.8	-3.6	V
lout	Output Current	$R_L = 10\Omega$ to 0V	±60	±100		mA
V _{IH-EN}	ENABLE pin Voltage for Power Up		V _S + -4			V
V _{IL-EN}	ENABLE pin Voltage for Shut Down				V _S + -1	V
I _{EN}	Enable Pin Current	Enabled, V _{EN} = 0V	-1		1	μA
		Disabled, V _{EN} = 5V	5	17	25	μA

Closed Loop AC Electrical Specifications $V_S = \pm 5V$, $T_A = 25^{\circ}C$, $V_{ENABLE} = 0V$, $A_V = +1$, $R_F = 0\Omega$, $R_L = 150\Omega$ to GND, unless otherwise specified.

PARAMETER	DESCRIPTION	CONDITIONS	MIN	TYP	MAX	UNIT
BW	-3dB Bandwidth (V _{OUT} = 200mV _{P-P})	$V_S = \pm 5V, A_V = 1, R_F = 0\Omega$	150	200		MHz
SR	Slew Rate	$R_L = 100\Omega$, $V_{OUT} = -3V$ to $+3V$, $A_V = +2$	1500	2200	4500	V/µs
t_R, t_F	Rise Time, Fall Time	±0.1V step		2.8		ns
OS	Overshoot	±0.1V step		10		%
t _{PD}	Propagation Delay	±0.1V step		3.2		ns
t _S	0.1% Settling Time	$V_S = \pm 5V$, $R_L = 500\Omega$, $A_V = 1$, $V_{OUT} = \pm 2.5V$		20		ns
dG	Differential Gain	$A_V = 2$, $R_L = 150\Omega$, $V_{INDC} = -1$ to +1V		0.08		%
dP	Differential Phase	$A_V = 2$, $R_L = 150\Omega$, $V_{INDC} = -1$ to +1V		0.1		o
e _N	Input Noise Voltage	f = 10kHz		10		nV/√Hz
i _N	Input Noise Current	f = 10kHz		7		pA/√Hz
t _{DIS}	Disable Time			180		ns
t _{EN}	Enable Time			650		ns

Typical Performance Curves

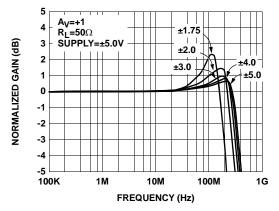


FIGURE 1. GAIN vs FREQUENCY FOR VARIOUS C_L

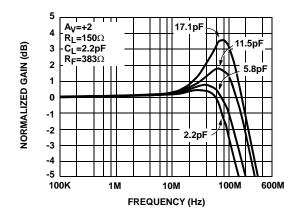


FIGURE 3. GAIN vs FREQUENCY FOR VARIOUS CIN-

5

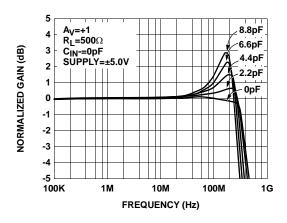


FIGURE 2. GAIN vs FREQUENCY FOR VARIOUS C_{L}

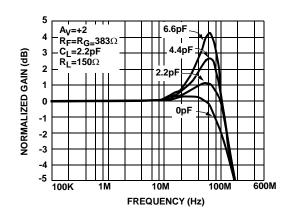


FIGURE 4. GAIN vs FREQUENCY FOR VARIOUS CIN-

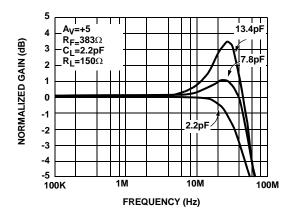


FIGURE 5. GAIN vs FREQUENCY FOR VARIOUS C_{IN} (-)

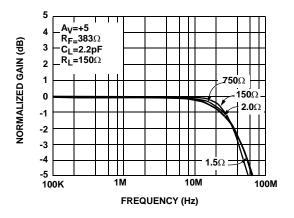


FIGURE 7. GAIN vs FREQUENCY FOR VARIOUS RL

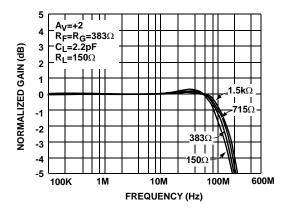


FIGURE 9. GAIN vs FREQUENCY FOR VARIOUS $R_{\rm L}$

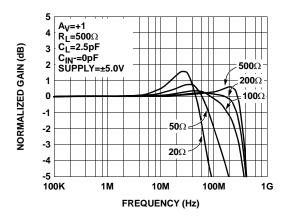


FIGURE 6. GAIN vs FREQUENCY FOR VARIOUS RL

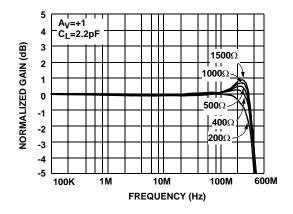


FIGURE 8. GAIN vs FREQUENCY FOR VARIOUS RL

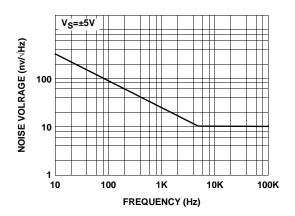


FIGURE 10. EQUIVALENT INPUT VOLTAGE NOISE vs FREQUENCY

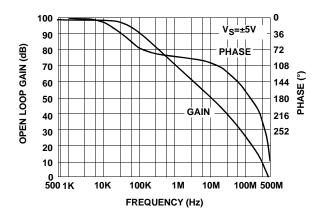


FIGURE 11. OPEN LOOP GAIN AND PHASE vs FREQUENCY

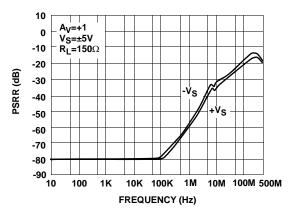


FIGURE 13. PSRR vs FREQUENCY

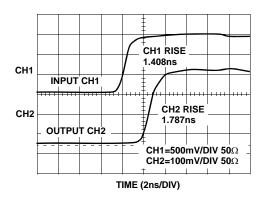


FIGURE 15. LARGE SIGNAL RISE TIME

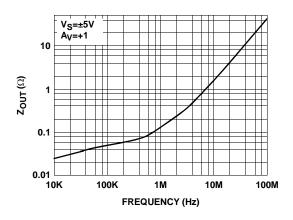


FIGURE 12. Z_{OUT} vs FREQUENCY

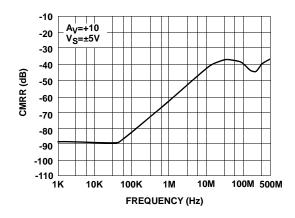


FIGURE 14. CMRR vs FREQUENCY

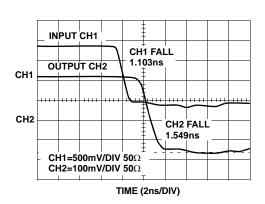


FIGURE 16. LARGE SIGNAL FALL TIME

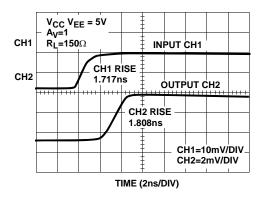


FIGURE 17. SMALL SIGNAL RISE TIME

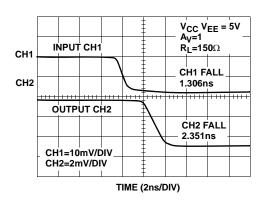


FIGURE 19. SMALL SIGNAL FALL TIME

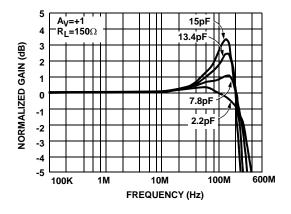


FIGURE 21. GAIN vs FREQUENCY FOR VARIOUS CL

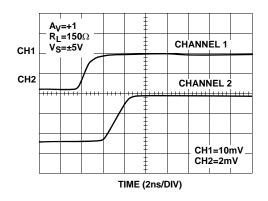


FIGURE 18. SMALL SIGNAL RISE TIME

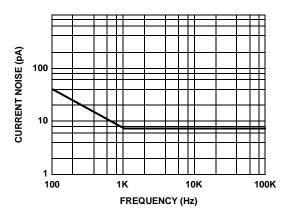


FIGURE 20. CURRENT NOISE

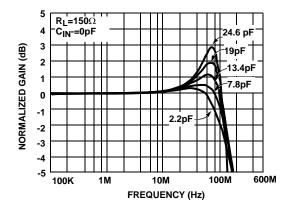


FIGURE 22. GAIN vs FREQUENCY FOR VARIOUS CL

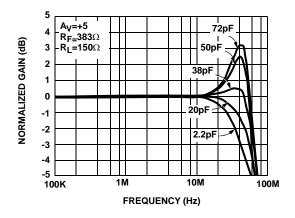


FIGURE 23. GAIN vs FREQUENCY FOR VARIOUS CL

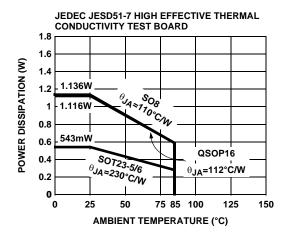


FIGURE 25. PACKAGE POWER DISSIPATION VS AMBIENT TEMPERATURE

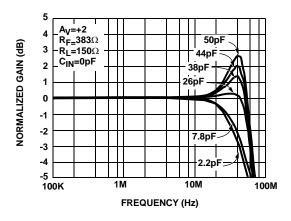


FIGURE 24. GAIN vs FREQUENCY FOR VARIOUS CL

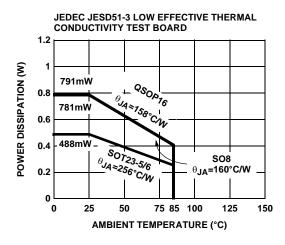


FIGURE 26. PACKAGE POWER DISSIPATION VS AMBIENT TEMPERATURE

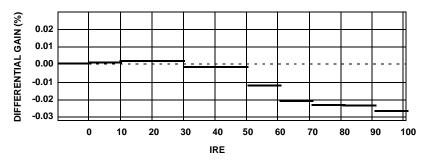


FIGURE 27. DIFFERENTIAL GAIN (%)

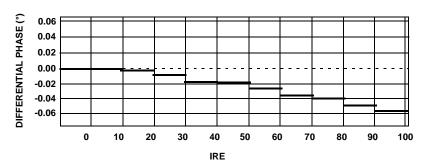


FIGURE 28. DIFFERENTIAL PHASE (°)

Application Information

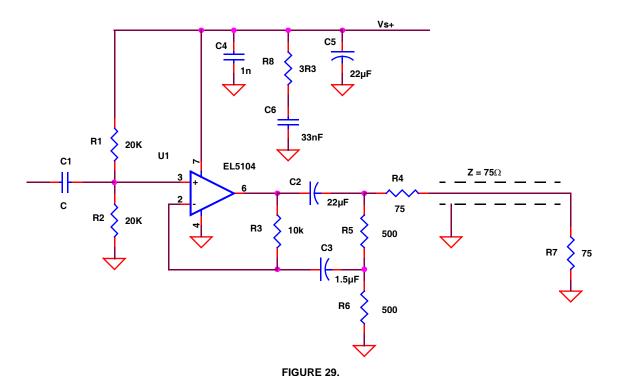
Video Amplifier with Reduced Size Output Capacitance

If you have a video line driver $Z = 75\Omega$, the DC decoupling capacitor could be relatively large.

$$C = \frac{1}{2\pi \times R \times f} =$$

$$f = 10Hz, R = Z = 75\Omega, C = 132\mu F$$

By using the circuit below, C could be reduced to $C2 = 22\mu F$.



10 0 -5 -10 -15 Conditions/comments: (1) C1 = 1µF Vs = +10V -20 (2) C1 = 0.47µF Vs = +10V -25 -30 -35 -40 -45 1.00E+00 1.00E+02 1.00E+04 1.00E+06 1.00E+08 1.00E+03 1.00E+05 1.00E+07 1.00E+09 FREQUENCY (Hz)

FIGURE 30. VIDEO-

11

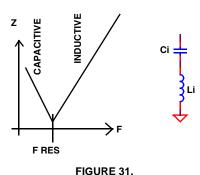
The test result is shown on Figure 30.

By selecting a different value for C1, we could reduce the effect, created by C3 R3 and get flat response from 16Hz with an 1/5 value, price and size output capacitor. There is another, very important issue by using high bandwidth amplifiers.

In the past when the bandwidth of the operational amplifier ended at a few hundred kHz even at few MHz, the power-supply bypass was not a very critical issue, since a $0.1\mu F$ capacitor "did the job", but today's amplifiers could have bandwidth, what used to be reserved for microwave circuits not to long time ago.

Therefore that high bandwidth amplifiers require the same respect what we reserve for microwave circuits. Particularly the power supply bypass and the pcb-layout could very heavily influence the performance of a modern high bandwidth amplifiers. It could happen above a few MHz, but it will happen above 100MHz, that the capacitor will behave like an inductor.

The reason for that is the very small but not zero value serial inductance of the capacitor.



The capacitor will behave as a capacitor up to its resonance frequency, above the resonance frequency it will behave as an inductor.

Just 1nHy inductance serial with 1nF capacitance will have serial resonance at:

$$F = \frac{1}{2\pi\sqrt{L \times C}}$$

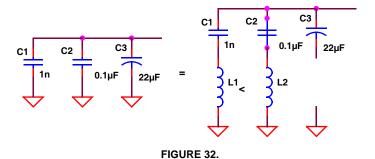
$$C = 1nF, L = 1nHy, F = 159 MHz$$

And an other 1nHy is very easy to get together with the inductance of traces on the pcb, and therefore you could encounter resonances from ca 50MHz and above anywhere. So if the amplifier has a bandwidth of a few hundred MHz, the proper power supply by-pass could become a serious if not difficult task.

Intuitively, you would use capacitors value $0.1\mu F$ parallel with a few μF tantalum, and to cure the effect of it's serial resonance put a smaller one parallel to it.

The result will surprise to you, because you will get even something worse than without the small capacitor.

What is happening there? Just look what we get:



Above its serial resonance C2* the ideal capacitance of C2 is a short, the Tantalum capacitor for high frequencies is not effective, the left over is C1 capacitor and L1 + L2 inductors, we get a parallel tank circuit, which is at it's resonance a high impedance path and do not carry any high frequency current, it does not work as bypass at all!

The impedance of a parallel tank circuit at resonance is dependent from it's Q. High Q high impedance.

The Q of a parallel tank circuit could be reduced by bypassing it with a resistor, or adding a resistor in serial to one of the reactive components. Since the bypassing would short the DC supply we do have to go to add resistor in serial to the reactive component, we will ad a resistor serial with the inductor. (See Figure 33.)

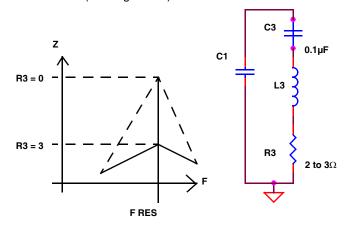


FIGURE 33.

The final power supply bypass circuit will look:

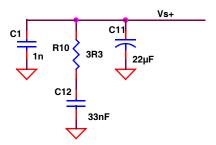
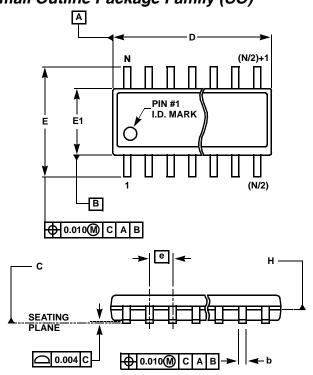
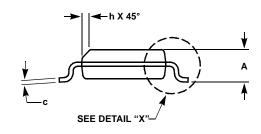
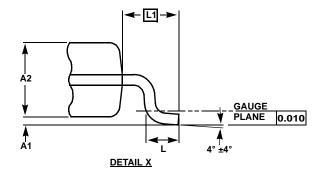


FIGURE 34.

Small Outline Package Family (SO)







MDP0027

SMALL OUTLINE PACKAGE FAMILY (SO)

	INCHES								
SYMBOL	SO-8	SO-14	SO16 (0.150")	SO16 (0.300") (SOL-16)	SO20 (SOL-20)	SO24 (SOL-24)	SO28 (SOL-28)	TOLERANCE	NOTES
Α	0.068	0.068	0.068	0.104	0.104	0.104	0.104	MAX	-
A1	0.006	0.006	0.006	0.007	0.007	0.007	0.007	±0.003	-
A2	0.057	0.057	0.057	0.092	0.092	0.092	0.092	±0.002	-
b	0.017	0.017	0.017	0.017	0.017	0.017	0.017	±0.003	-
С	0.009	0.009	0.009	0.011	0.011	0.011	0.011	±0.001	-
D	0.193	0.341	0.390	0.406	0.504	0.606	0.704	±0.004	1, 3
Е	0.236	0.236	0.236	0.406	0.406	0.406	0.406	±0.008	-
E1	0.154	0.154	0.154	0.295	0.295	0.295	0.295	±0.004	2, 3
е	0.050	0.050	0.050	0.050	0.050	0.050	0.050	Basic	÷
L	0.025	0.025	0.025	0.030	0.030	0.030	0.030	±0.009	÷
L1	0.041	0.041	0.041	0.056	0.056	0.056	0.056	Basic	=
h	0.013	0.013	0.013	0.020	0.020	0.020	0.020	Reference	=
Ν	8	14	16	16	20	24	28	Reference	=

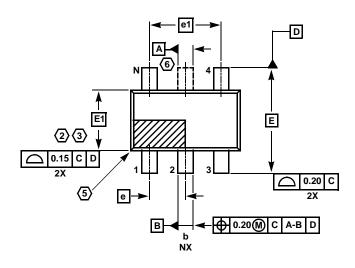
NOTES

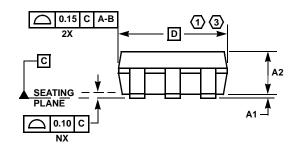
Rev. M 2/07

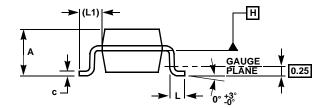
- 1. Plastic or metal protrusions of 0.006" maximum per side are not included.
- 2. Plastic interlead protrusions of 0.010" maximum per side are not included.
- 3. Dimensions "D" and "E1" are measured at Datum Plane "H".
- 4. Dimensioning and tolerancing per ASME Y14.5M-1994

intersil

SOT-23 Package Family







14

MDP0038

SOT-23 PACKAGE FAMILY

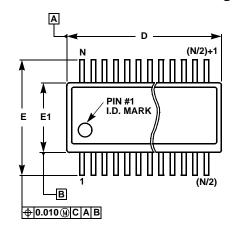
	MILLIM			
SYMBOL	SOT23-5	SOT23-6	TOLERANCE	
Α	1.45	1.45	MAX	
A1	0.10	0.10	±0.05	
A2	1.14	1.14	±0.15	
b	0.40	0.40	±0.05	
С	0.14	0.14	±0.06	
D	2.90	2.90	Basic	
Е	2.80	2.80	Basic	
E1	1.60	1.60	Basic	
е	0.95	0.95	Basic	
e1	1.90	1.90	Basic	
L	0.45	0.45	±0.10	
L1	0.60	0.60	Reference	
N	5	6	Reference	

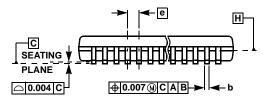
Rev. F 2/07

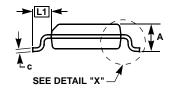
NOTES:

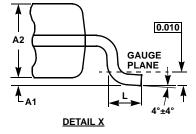
- 1. Plastic or metal protrusions of 0.25mm maximum per side are not included.
- Plastic interlead protrusions of 0.25mm maximum per side are not included.
- 3. This dimension is measured at Datum Plane "H".
- 4. Dimensioning and tolerancing per ASME Y14.5M-1994.
- Index area Pin #1 I.D. will be located within the indicated zone (SOT23-6 only).
- 6. SOT23-5 version has no center lead (shown as a dashed line).

Quarter Size Outline Plastic Packages Family (QSOP)









MDP0040

QUARTER SIZE OUTLINE PLASTIC PACKAGES FAMILY

	INCHES				
SYMBOL	QSOP16	QSOP24	QSOP28	TOLERANCE	NOTES
Α	0.068	0.068	0.068	Max.	-
A1	0.006	0.006	0.006	±0.002	-
A2	0.056	0.056	0.056	±0.004	-
b	0.010	0.010	0.010	±0.002	-
С	0.008	0.008	0.008	±0.001	-
D	0.193	0.341	0.390	±0.004	1, 3
E	0.236	0.236	0.236	±0.008	-
E1	0.154	0.154	0.154	±0.004	2, 3
е	0.025	0.025	0.025	Basic	-
L	0.025	0.025	0.025	±0.009	-
L1	0.041	0.041	0.041	Basic	-
N	16	24	28	Reference	-

Rev. F 2/07

NOTES:

- Plastic or metal protrusions of 0.006" maximum per side are not included.
- Plastic interlead protrusions of 0.010" maximum per side are not included.
- 3. Dimensions "D" and "E1" are measured at Datum Plane "H".
- 4. Dimensioning and tolerancing per ASME Y14.5M-1994.

All Intersil U.S. products are manufactured, assembled and tested utilizing ISO9000 quality systems. Intersil Corporation's quality certifications can be viewed at www.intersil.com/design/quality

Intersil products are sold by description only. Intersil Corporation reserves the right to make changes in circuit design, software and/or specifications at any time without notice. Accordingly, the reader is cautioned to verify that data sheets are current before placing orders. Information furnished by Intersil is believed to be accurate and reliable. However, no responsibility is assumed by Intersil or its subsidiaries for its use; nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Intersil or its subsidiaries.

For information regarding Intersil Corporation and its products, see www.intersil.com

X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Renesas manufacturer:

Other Similar products are found below:

EL4511CUZ-T7 PYB15-Q24-S5-H-U PQA30-D24-S24-DH PQA30-D48-S12-TH PYB30-Q24-T312-H-U PYB15-Q24-S5-H-T PYB15-Q24-S12-H-T V7815-500-SMT PYB20-Q48-S12-H-T PQZ6-Q24-S15-D PYB20-Q48-S5-H-T PYB20-Q24-S12-H-T VLED15-120-350 VGS-75-12 PYB15-Q24-S12-H-U HI3-574AJN-5Z ISL6561IRZ ICL3238ECAZ R5F100GFAFB#V0 VGS-50-15 VGS-50-24 HFA1100IBZ ISL6525CBZ VGS-25-24 VGS-50-5 VGS-100-12 ICL7663SACBAZA ISL6117CBZA ISL6443IRZ ISL12027IBZ HIP4081AIB ISL84514IBZ ISL84521IVZ ISL12027IB27AZ ISL6741IBZ X9116WS8IZ M30620FCAFP#U3 ETSA120500UD-P5P-SZ PDQ2-D24-S12-S PDS1-S12-D12-M PDS1-S12-D15-M PYB15-Q24-S12-T PYB20-Q48-S12 R0K33062PS000BE R0K505220S000BE R0K561664S000BE R0K570865S000BE HC55185AIMZ R7S721001VCBGAC0 EPSA050250UB-P5P-EJ