

RAA210925

Pin-Configurable Dual 25A DC/DC Power Module with PMBus Interface

FN9352
Rev.1.00
May 31, 2019

The [RAA210925](#) is a pin-strap configurable dual 25A step-down PMBus-compliant DC/DC power supply module that integrates a digital PWM controller, synchronous MOSFETs, power inductor, and passive components. Only input and output capacitors are needed to finish the design. Because of its thermally-enhanced HDA packaging technology, the module can deliver up to 25A of continuous output current without the need for airflow or additional heat sinking. The RAA210925 simplifies configuration and control of Renesas [digital power technology](#) while offering an upgrade to full PMBus configuration through the pin-compatible ISL8274M.

Operating across an input voltage range of 4.5V to 14V, the RAA210925 offers adjustable output voltages down to 0.6V and achieves up to 95.5% conversion efficiencies. A unique ChargeMode™ control architecture provides a single clock cycle response to an output load step and can support switching frequencies up to 1067kHz. The power module integrates all power and most passive components and requires only a few external components to operate. The RAA210925 comes with a preprogrammed configuration for operating in Pin-strap mode. Output voltage, switching frequency, input UVLO, soft-start/stop delay and ramp times, tracking function, and the device SMBus address can be programmed with external pin-strap resistors. A standard PMBus interface addresses fault management, as well as real-time full telemetry and point-of-load monitoring. The RAA210925 is supported by PowerNavigator™ software, a Graphical User Interface (GUI), that can configure modules for desired solutions.

The RAA210925 is available in a low profile, compact 18mmx23mmx7.5mm fully encapsulated, thermally enhanced HDA package.

Related Literature

For a full list of related documents, visit our website:

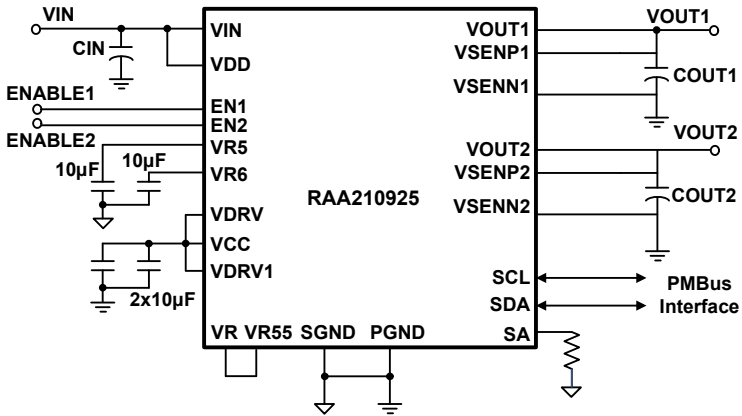
- [RAA210925](#) device page

Features

- 25A/25A dual-channel output current
 - 4.5V to 14V single rail input voltage
 - Up to 95.5% efficiency
- Programmable output voltage
 - 0.6V to 5V output voltage settings
 - ±1.2% accuracy over line/load/temperature
- ChargeMode control loop architecture
 - 296kHz to 1067kHz fixed switching frequency operations
 - No compensation required
 - Fast single clock cycle transient response
- PMBus interface and/or pin-strap mode
 - Programmable through PMBus
 - Pin-strap mode for standard settings
 - Real-time telemetry for V_{IN} , V_{OUT} , I_{OUT} , temperature, duty cycle, and f_{SW}
- Complete over/undervoltage, current, and temperature protections with fault logging
- [PowerNavigator](#) supported
- Thermally enhanced 18mmx23mmx7.5mm HDA package

Applications

- Server, telecom, storage, and datacom
- Industrial/ATE and networking equipment
- General purpose power for ASIC, FPGA, DSP, and memory



Note: This figure represents a typical implementation of the RAA210925. For PMBus operation, it is recommended to tie the enable pin (EN) to SGND.

Figure 1. Application Circuit

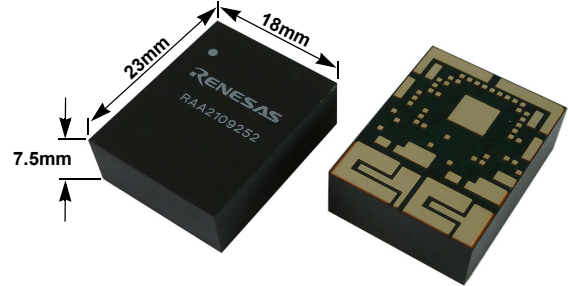


Figure 2. Small Package for High Power Density

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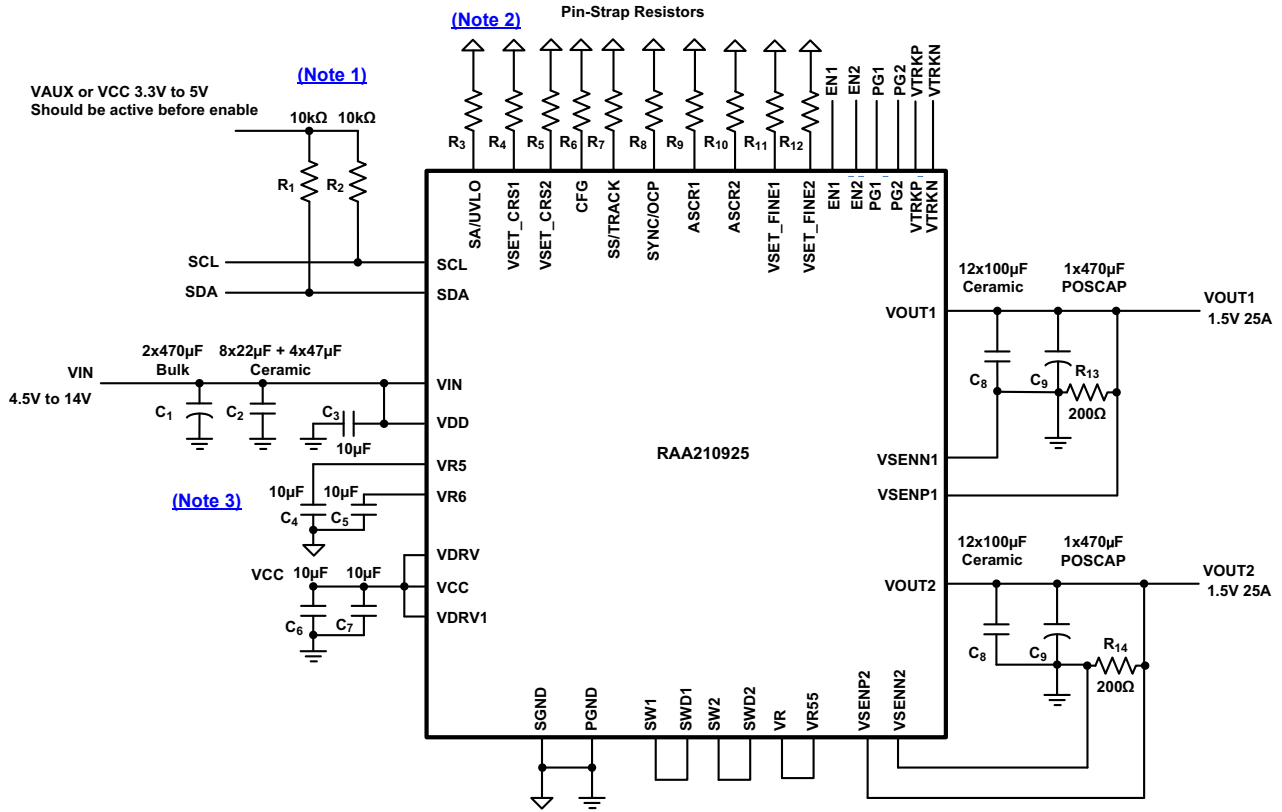
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1. Overview

1.1 Typical Application Circuit



Notes:

1. R₁ and R₂ are not required if the PMBus host already has I²C pull-up resistors.
2. R₃ through R₁₁ can be selected according to the tables for the pin-strap resistor setting in this document.
3. V₂₅, VR, and VR₅₅ do not need external capacitors. V₂₅ can be no connection.

Figure 3. RAA210925 Digital PMBus Module Dual 25A/25A Application with Pin-Strap Settings

Table 1. RAA210925 Design Guide Matrix and Output Voltage Response

| V _{IN} (V) | V _{OUT} (V) | f _{sw} (kHz) | I _{OUT} (A) | Avg OCP (A) | C _{IN} (μF) | Cout_Bulk (μF) (Note 4) | Cout_Ceramic (μF) (Note 4) | ASCR Gain (Note 6) |
|---------------------|----------------------|-----------------------|----------------------|-------------|---|-------------------------|----------------------------|--------------------|
| 12 | 5 | 1067 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 6*100 | 275 |
| 12 | 5 | 615 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 6*100 | 175 |
| 12 | 3.3 | 800 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 8*100 | 300 |
| 12 | 3.3 | 571 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 8*100 | 175 |
| 12 | 2.5 | 1067 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 9*100 | 600 |
| 12 | 2.5 | 615 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 9*100 | 275 |
| 12 | 2.5 | 471 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 9*100 | 175 |
| 12 | 1.8 | 889 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 600 |
| 12 | 1.8 | 421 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 225 |
| 12 | 1.8 | 364 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 175 |
| 12 | 1.5 | 889 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 525 |
| 12 | 1.5 | 421 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 225 |
| 12 | 1.5 | 320 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 140 |
| 12 | 1.2 | 727 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 600 |
| 12 | 1.2 | 296 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 4*470 | 12*100 | 225 |
| 12 | 1 | 615 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 450 |
| 12 | 1 | 296 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 5*470 | 12*100 | 250 |
| 12 | 0.6 | 296 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 7*470 | 12*100 | 300 |
| 5 | 2.5 | 1067 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 9*100 | 600 |
| 5 | 2.5 | 615 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 9*100 | 275 |
| 5 | 2.5 | 471 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 9*100 | 175 |
| 5 | 1.8 | 889 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 600 |
| 5 | 1.8 | 421 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 250 |
| 5 | 1.8 | 364 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 200 |
| 5 | 1.5 | 889 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 525 |
| 5 | 1.5 | 421 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 250 |
| 5 | 1.5 | 320 | 20 | 25 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 140 |
| 5 | 1.2 | 727 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 600 |

Table 1. RAA210925 Design Guide Matrix and Output Voltage Response (Continued)

| V _{IN} (V) | V _{OUT} (V) | f _{SW} (kHz) | I _{OUT} (A) | Avg OCP (A) | C _{IN} (μF) | Cout_Bulk (μF) (Note 4) | Cout_Ceramic (μF) (Note 4) | ASCR Gain (Note 6) |
|---------------------|----------------------|-----------------------|----------------------|-------------|---|-------------------------|----------------------------|--------------------|
| 5 | 1.2 | 296 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 4*470 | 12*100 | 250 |
| 5 | 1 | 615 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 1*470 | 12*100 | 450 |
| 5 | 1 | 296 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 5*470 | 12*100 | 250 |
| 5 | 0.6 | 296 | 25 | 30 | 2*470μF POS + (8*22μF + 4*47μF) ceramic | 7*470 | 12*100 | 300 |

Notes:

4. 100μF (GRM31CD80J107ME39L) ceramic and 470μF (6TPF470MAH) are selected for output capacitor in the evaluation board.
5. Peak-to-peak V_{OUT} deviation is measured under 0%-60% load transient with 15A/μs load step slew rate.
6. ASCR gain and residual was designed to achieve 50° phase margin at room temperature.
7. Frequency is selected to achieve the highest efficiency at full load as well as avoid saturation of the inductor. For instance, select 615kHz instead of 296kHz to avoid inductor saturation. Although better efficiency is obtained at 296kHz supporting 1V, 25A, higher frequency can be selected because less output capacitor meet the transient response specification.

1.2 RAA210925 Internal Block Diagram

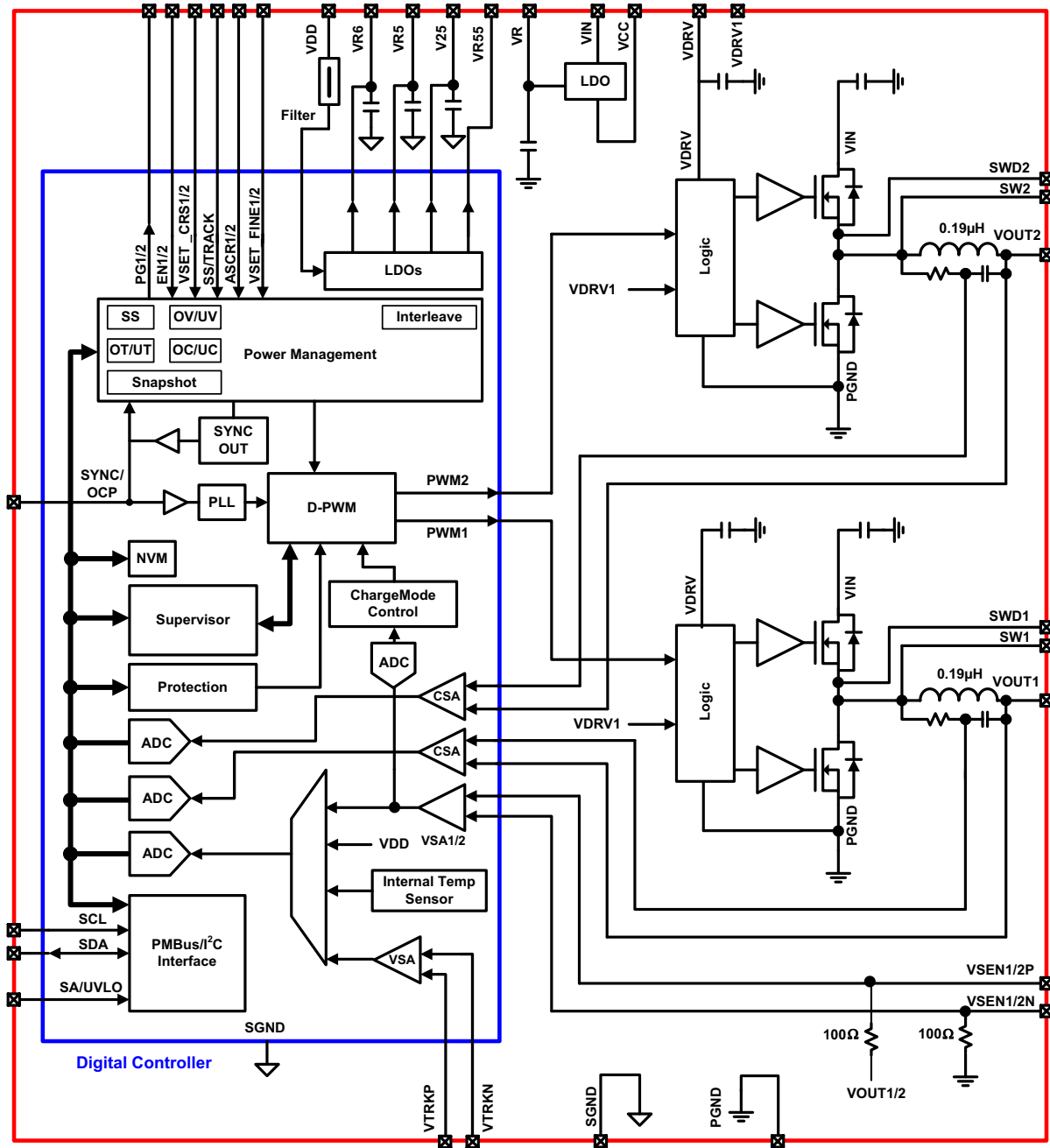


Figure 4. Internal Block Diagram

1.3 Ordering Information

| Part Number (Notes 9, 10) | Part Marking | Temp Range (°C) | Tape and Reel (Units) (Note 8) | Package (RoHS Compliant) | Pkg. Dwg. # |
|------------------------------|------------------|-----------------|-----------------------------------|-----------------------------|----------------|
| RAA2109252GLG#AG0 | RAA2109252 | -40 to +85 | - | 58 Ld 18x23 HDA Module | Y58.18x23 |
| RAA2109252GLG#HG0 | RAA2109252 | -40 to +85 | 100 | 58 Ld 18x23 HDA Module | Y58.18x23 |
| RTKA2109252H00000BU | Evaluation Board | | | | |

Notes:

8. Refer to [TB347](#) for details about reel specifications.
9. These Pb-free plastic packaged products are RoHS compliant by EU exemption 7C-I and 7A. They employ special Pb-free material sets; molding compounds/die attach materials and NiPdAu plate-e4 termination finish, which is compatible with both SnPb and Pb-free soldering operations. Pb-free products are MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
10. For Moisture Sensitivity Level (MSL), see the [RAA210925](#) device page. For more information about MSL, see [TB363](#).

Table 2. Key Differences Between Family of Parts

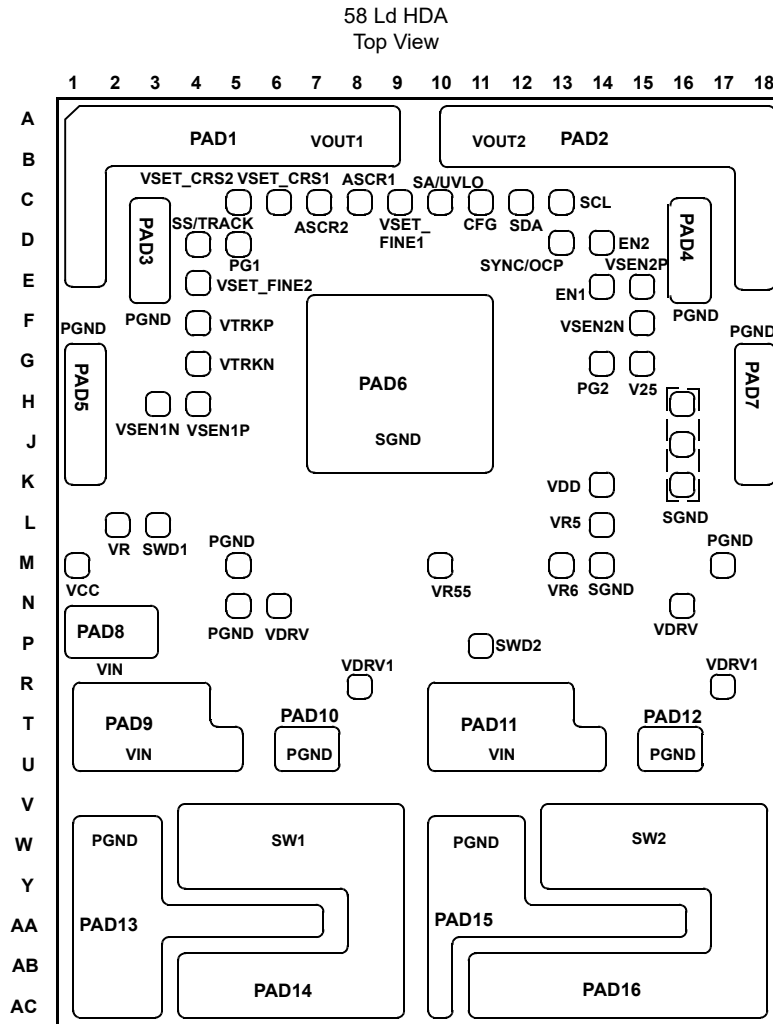
| Part Number | Description | V _{IN} Range (V) | V _{OUT} Range (V) | I _{OUT} (A) |
|-------------|---|---------------------------|----------------------------|----------------------|
| RAA210925 | 25A/25A DC/DC dual channel Power Module | 4.5 - 14 | 0.6 - 5 | 25/25 |
| RAA210825 | 25A DC/DC single channel Power Module | 4.5 - 14 | 0.6 - 5 | 25 |
| RAA210833 | 33A DC/DC single channel Power Module | 4.5 - 14 | 0.6 - 5 | 33 |
| RAA210850 | 50A DC/DC single channel Power Module | 4.5 - 14 | 0.6 - 5 | 50 |
| RAA210870 | 70A DC/DC single channel Power Module | 4.5 - 14 | 0.6 - 2.5 | 70 |

Table 3. Comparison of Simple Digital and Full Digital Parts

| | ISL8274M | RAA210925 |
|--|--|--|
| V _{IN} (V) | 4.5 - 14 | 4.5 - 14 |
| V _{OUT} (V) | 0.6 - 5 | 0.6 - 5 |
| I _{OUT} (Max) (A) | 30/30 | 25/25 |
| f _{SW} (kHz) | 296 - 1067 | 296 - 1067 |
| Digital PMBus Programmability for Configuration of Modules | All PMBus commands. NVM access to store module configuration | Configuration of modules supported by pin-strap resistors. Digital programmability supports configuration changes during run-time operation with a subset of PMBus commands. No NVM access to store module configuration |
| Power Navigator Support | Yes | Yes |
| SYNC Capability | Yes | Yes |
| Current Sharing Multi-Modules | No | No |
| DDC Pin (Inter-Device Communication) | Yes | No |

Note: For a full comparison of all the RAA210XXX and ISL827XM product offerings please visit the [simple-digital module family](#) page.

1.4 Pin Configuration



1.5 Pin Descriptions

| Pin Number | Pin Name | Type | Description |
|--|------------|------|--|
| PAD1 | VOUT1 | PWR | Power supply output voltage. Channel 1 provides an output voltage from 0.6V to 5V. Refer to the "Functional Description" on page 22 to set the maximum output current from these pads. |
| PAD2 | VOUT2 | PWR | Power supply output voltage. Channel 2 provides an output voltage from 0.6V to 5V. Refer to the "Functional Description" on page 22 to set the maximum output current from these pads. |
| PAD3, PAD4, PAD5, PAD7, PAD10, PAD12, PAD13, PAD15 | PGND | PWR | Power ground. Refer to the "Layout Guide" on page 35 for the PGND pad connections and I/O capacitor placement. |
| PAD6 | SGND | PWR | Signal ground. Refer to "Layout Guide" on page 35 for the SGND pad connections. |
| PAD8, PAD9, PAD11 | VIN | PWR | Input power supply voltage to power the module. Input voltage ranges from 4.5V to 14V. |
| PAD14 | SW1 | PWR | Switching node pads for Channel 1. The SW1 pad dissipates the heat and provides good thermal performance. Refer to "Layout Guide" on page 35 for the SW1 pad connections. |
| PAD16 | SW2 | PWR | Switching node pads for Channel 2. The SW2 pad dissipates the heat and provides good thermal performance. Refer to "Layout Guide" on page 35 for the SW2 pad connections. |
| C5 | VSET_CRS2 | I | Output voltage selection pin for Channel 2. Sets the VOUT2 set point and VOUT2 max. Use VSET_FINE2 for fine-tuning. |
| C6 | VSET_CRS1 | I | Output voltage selection pin for Channel 1. Sets the VOUT1 set point and VOUT1 max. Use VSET_FINE1 for fine-tuning. |
| C7 | ASCR2 | I | ChargeMode control ASCR parameters selection pin for Channel 2. Sets ASCR gain and residual values. |
| C8 | ASCR1 | I | ChargeMode control ASCR parameters selection pin for Channel 1. Sets ASCR gain and residual values. |
| C9 | VSET_FINE1 | I | Output voltage fine-tuning. Provides increased VOUT1 resolution based on programmed VSET_CRS1 value. |
| C10 | SA/UVLO | I | Serial address selection pin. Assigns unique address for each individual device or enables certain management features. This pin also sets the UVLO level. |
| C11 | CFG | O | Clock source configuration. If the clock source is internal, set the internal FREQUENCY_SWITCH according to the SYNC pin resistor setting. If the clock source is external, the internal FREQUENCY_SWITCH is set according to the CFG pin resistor. |
| C12 | SDA | I/O | Serial data. Connect to external host and/or to other Digital-DC™ devices. A pull-up resistor is required. |
| C13 | SCL | I/O | Serial clock. Connect to external host and/or to other Digital-DC devices. A pull-up resistor is required. |
| D4 | SS/ TRACK | I | Soft-start/stop selection pin. Sets the turn on/off delay and ramp time as well as tracking configuration. |
| D5 | PG1 | O | Power-good output for Channel 1. Power-good is configured as an open-drain output. |
| D13 | SYNC/ OCP | I/O | Clock synchronization input and OCP setting pin. Sets the frequency of the internal switch clock, synchronizes to an external clock, or an output internal clock. If using external synchronization, the external clock must be active before enable. Different OCP levels can be set with this pin. |
| D14 | EN2 | I | Enable pin for Channel 2. Logic high to enable the module output. |
| E14 | EN1 | I | Enable pin for Channel 1. Logic high to enable the module output. |
| E4 | VSET_FINE2 | I/O | Output voltage fine-tuning. Provides increased VOUT2 resolution based on programmed VSET_CRS2 value. |
| E15 | VSEN2P | I | Differential output voltage sense feedback for Channel 2. Connect to positive output regulation point. |

| Pin Number | Pin Name | Type | Description |
|--------------------|----------|------|--|
| F4 | VTRKP | I | Tracking sense positive input. Tracks an external voltage source. |
| F15 | VSEN2N | I | Differential output voltage sense feedback for Channel 2. Connect to negative output regulation point. |
| G4 | VTRKN | I | Tracking sense negative input (return). |
| G14 | PG2 | O | Power-good output for Channel 2. Power-good is configured as an open-drain output. |
| G15 | V25 | PWR | Internal 2.5V reference that powers internal circuitry. No external capacitor required for this pin. Not recommended to power external circuits. |
| H3 | VSEN1N | I | Differential output voltage sense feedback for Channel 1. Connect to a negative output regulation point. |
| H4 | VSEN1P | I | Differential output voltage sense feedback for Channel 1. Connect to a positive output regulation point. |
| H16, J16, K16, M14 | SGND | PWR | Signal grounds. Use multiple vias to connect the SGND pins to the internal SGND layer. |
| K14 | VDD | PWR | Input supply voltage for controller. Connect VDD pad to VIN supply. |
| L2 | VR | PWR | Internal LDO bias pin. Tie VR to VR55 directly with a short loop trace. Not recommended to power external circuits. |
| L3 | SWD1 | PWR | Switching node driving pins for Channel 1. Directly connect to the SW1 pad with short loop wires. |
| P11 | SWD2 | PWR | Switching node driving pins for Channel 2. Directly connect to the SW2 pad with short loop wires. |
| L14 | VR5 | PWR | Internal 5V reference that powers internal circuitry. Place a 10 μ F decoupling capacitor for this pin. Maximum external loading current is 5mA. |
| M1 | VCC | PWR | Internal LDO output. Connect VCC to VDRV for internal LDO driving. |
| M5, M17, N5 | PGND | PWR | Power grounds. Using multiple vias to connect the PGND pins to the internal PGND layer. |
| M10 | VR55 | PWR | Internal 5.5V bias voltage for internal LDO use only. Tie VR55 pin directly to the VR pin. Not recommended to power external circuits. |
| M13 | VR6 | PWR | Internal 6V reference that powers internal circuitry. Place a 10 μ F decoupling capacitor for this pin. Not recommended to power external circuits. |
| N6, N16 | VDRV | PWR | Power supply for internal FET drivers. Connect a 10 μ F bypass capacitor to each of these pins. These pins can be driven by the internal LDO through the VCC pin or by the external power supply directly. Keep the driving voltage between 4.5V and 5.5V. For 5V input application, use external supply or connect this pin to VIN. |
| R8, R17 | VDRV1 | I | Bias pin of the internal FET drivers. Always tie to VDRV. |

2. Specifications

2.1 Absolute Maximum Ratings

| Parameter | Minimum | Maximum | Unit |
|---|--------------|---------|-------------|
| Input Supply Voltage, VIN Pin | -0.3 | 17 | V |
| Input Supply Voltage for Controller, VDD Pin | -0.3 | 17 | V |
| MOSFET Switch Node Voltage, SW1/2, SWD1/2 (Note 11) | -0.3 | 25 | V |
| MOSFET Driver Supply Voltage, VDRV, VDRV1 Pin | -0.3 | 6.0 | V |
| Output Voltage, VOUT1/2 Pin | -0.3 | 6.0 | V |
| Internal Reference Supply Voltage | | | |
| VR6 Pin | -0.3 | 6.6 | V |
| VR, VR5, VR55 Pin | -0.3 | 6.5 | V |
| V25 Pin | -0.3 | 3 | V |
| Logic I/O Voltage for EN1/2, PG1/2, ASCR1/2, SA/UVLO, SCL, SDA, SYNC/OCF, SS/TRACK, VSET_CRS1/2, VSET_FINE1/2 | -0.3 | 6.0 | V |
| Analog Input Voltages | | | |
| VSEN1P, VSEN2P, VTRKP | -0.3 | 6.0 | V |
| VSEN1N, VSEN2N, VTRKN | -0.3 | 0.3 | V |
| ESD Rating | Value | | Unit |
| Human Body Model (Tested per JS-001-2017) | 2 | | kV |
| Machine Model (Tested per JESD22-A115C) | 200 | | V |
| Charged Device Model (Tested per JS-002-2014) | 750 | | V |
| Latch-Up (Tested per JESD78E; Class 2, Level A) | 100 | | mA |

Note:

11. Do not apply DC voltage higher than 17V to the pins.

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

2.2 Thermal Information

| Thermal Resistance (Typical) | θ_{JA} (°C/W) | θ_{JC} (°C/W) |
|----------------------------------|----------------------|----------------------|
| 58 Ld HDA Package (Notes 12, 13) | 5.3 | 1.1 |

Notes:

12. θ_{JA} is defined by simulation in free air with the module mounted on an 8-layer evaluation board 4.7x4.8inch in size with 2oz Cu on all layers.
13. For θ_{JC} , the "case temp" location is the center of the package underside.

| Parameter | Minimum | Maximum | Unit |
|--|-------------------------------|---------|------|
| Maximum Junction Temperature (Plastic Package) | | +125 | °C |
| Storage Temperature Range | -55 | +150 | °C |
| Pb-Free Reflow Profile | see Figure 33 | | |

2.3 Recommended Operation Conditions

| Parameter | Minimum | Maximum | Unit |
|---|---------|---------|------|
| Input Supply Voltage Range, V_{IN} | 4.5 | 14 | V |
| Input Supply Voltage Range for Controller, V_{DD} | 4.5 | 14 | V |
| Output Voltage Range, V_{OUT} | 0.6 | 5 | V |
| Output Current Range, $I_{OUT(DC)}$ Per Channel (Note 16) | 0 | 25 | A |
| Operating Junction Temperature Range, T_J | -40 | +125 | °C |

2.4 Electrical Specifications

$V_{IN} = V_{DD} = 12V$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$. **Boldface limits apply across the operating temperature range, $-40^{\circ}C$ to $+85^{\circ}C$.**

| Parameter | Symbol | Test Conditions | Min (Note 14) | Typ | Max (Note 14) | Unit |
|---|-----------------------|--|---------------|-------------|---------------|-------------|
| Input and Supply Characteristics | | | | | | |
| Input Supply Current for Controller | I_{DD} | $V_{IN} = V_{DD} = 12V$, $V_{OUT} = 0V$, module not enabled | | 40 | 50 | mA |
| 6V Internal Reference Supply Voltage | V_{R6} | | 5.5 | 6.1 | 6.6 | V |
| 5V Internal Reference Supply | V_{R5} | $I_{VR5} < 5mA$ | 4.5 | 5.2 | 5.5 | V |
| 2.5V Internal Reference Supply | V_{25} | | 2.25 | 2.5 | 2.75 | V |
| Internal LDO Output Voltage | V_{CC} | | | 5.3 | | V |
| Internal LDO Output Current | I_{VCC} | $V_{IN} = V_{DD} = 12V$, V_{CC} connected to VDRV, module enabled | 50 | | | mA |
| Input Supply Voltage for Controller Read Back Resolution | $V_{DD_READ_RES}$ | | | ± 20 | | mV |
| Input Supply Voltage for Controller Read Back Total Error (Note 17) | $V_{DD_READ_ERR}$ | PMBus read | | ± 2 | | % FS |
| Output Characteristics | | | | | | |
| Output Voltage Adjustment Range | V_{OUT_RANGE} | | 0.54 | | 5.5 | V |
| Output Voltage Set-Point Resolution | V_{OUT_RES} | Configured using PMBus | | ± 0.025 | | % V_{OUT} |
| Output Voltage Set-Point Accuracy (Notes 15, 17) | V_{OUT_ACCY} | Includes line, load, and temperature ($-20^{\circ}C \leq T_A \leq +85^{\circ}C$) | -1.2 | | 1.2 | % |
| Output Voltage Read Back Resolution | $V_{OUT_READ_RES}$ | | | ± 0.15 | | % FS |
| Output Voltage Read Back Total Error (Note 17) | $V_{OUT_READ_ERR}$ | PMBus read | -2 | | 2 | % FS |
| Output Current Range (Note 16) | I_{OUT_RANGE} | Per channel | | | 25 | A |
| Output Current Read Back Total Error | $I_{OUT_READ_ERR}$ | PMBus read at max load $V_{OUT} = 1.5V$ | | ± 3 | | A |
| Soft-Start and Sequencing | | | | | | |
| Delay Time from Enable to V_{OUT} Rise | t_{ON_DELAY} | Configured using pin-strap resistor or PMBus | 2 | | 300 | ms |
| t_{ON_DELAY} Accuracy | $t_{ON_DELAY_ACCY}$ | | | ± 2 | | ms |

$V_{IN} = V_{DD} = 12V$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$. **Boldface limits apply across the operating temperature range, $-40^{\circ}C$ to $+85^{\circ}C$.** (Continued)

| Parameter | Symbol | Test Conditions | Min (Note 14) | Typ | Max (Note 14) | Unit |
|---|-------------------------|--|------------------------------------|---------------|------------------------------------|-------------|
| Output Voltage Ramp-Up Time | t_{ON_RISE} | Configured using pin-strap resistor or PMBus | 0.5 | | 120 | ms |
| Output Voltage Ramp-Up Time Accuracy | $t_{ON_RISE_ACCY}$ | | | ± 250 | | μs |
| Delay Time from Disable to V_{OUT} Fall | t_{OFF_DELAY} | Configured using pin-strap resistor or PMBus | 2 | | 300 | ms |
| t_{OFF_DELAY} Accuracy | $t_{OFF_DELAY_ACCY}$ | | | ± 2 | | ms |
| Output Voltage Fall Time | t_{OFF_FALL} | Configured using pin-strap resistor or PMBus | 0.5 | | 120 | ms |
| Output Voltage Fall Time Accuracy | $t_{ON_FALL_ACCY}$ | | | ± 250 | | μs |
| Power-Good | | | | | | |
| Power-Good Delay | V_{PG_DELAY} | | | 3 | | ms |
| Temperature Sense | | | | | | |
| Temperature Sense Range | T_{SENSE_RANGE} | | -50 | | 150 | $^{\circ}C$ |
| Internal Temperature Sensor Accuracy | INT_TEMP_{ACCY} | Tested at $+100^{\circ}C$ | -5 | | 5 | $^{\circ}C$ |
| Fault Protection | | | | | | |
| V_{DD} Undervoltage Threshold Range | $V_{DD_UVLO_RANGE}$ | Measured internally | 4.18 | | 16 | V |
| V_{DD} Undervoltage Threshold Accuracy (Note 17) | $V_{DD_UVLO_ACCY}$ | | | ± 2 | | %FS |
| V_{DD} Undervoltage Response Time | $V_{DD_UVLO_DELAY}$ | | | 10 | | μs |
| V_{OUT} Overvoltage Threshold Range | $V_{OUT_OV_RANGE}$ | Factory default | | $1.15V_{OUT}$ | | V |
| | | Configured using pin-strap resistor or PMBus | $1.05V_{OUT}$ | | V_{OUT_MAX} | V |
| V_{OUT} Undervoltage Threshold Range | $V_{OUT_UV_RANGE}$ | Factory default | | $0.85V_{OUT}$ | | V |
| | | Configured using pin-strap resistor or PMBus | 0 | | $0.95V_{OUT}$ | V |
| V_{OUT} OV/UV Threshold Accuracy (Note 15) | V_{OUT_OV/UV_ACCY} | | -2 | | 2 | % |
| V_{OUT} OV/UV Response Time | V_{OUT_OV/UV_DELAY} | | | 10 | | μs |
| Output Current Limit Set-Point Accuracy (Note 17) | I_{LIMIT_ACCY} | Tested at $I_{OUT_AVG_OC_FAULT_LIMIT} = 35A$ | | ± 10 | | % FS |
| Over-temperature Protection Threshold (Controller Junction Temperature) | $T_{JUNCTION}$ | Factory default | | 115 | | $^{\circ}C$ |
| | | Configured using PMBus | -40 | | 115 | $^{\circ}C$ |
| Thermal Protection Hysteresis | $T_{JUNCTION_HYS}$ | | | 15 | | $^{\circ}C$ |
| Oscillator and Switching Characteristics | | | | | | |
| Switching Frequency Range | f_{SW_RANGE} | | 296 | | 1067 | kHz |
| Switching Frequency Set-Point Accuracy | f_{SW_ACCY} | | -5 | | 5 | % |

$V_{IN} = V_{DD} = 12V$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$. **Boldface limits apply across the operating temperature range, $-40^{\circ}C$ to $+85^{\circ}C$.** (Continued)

| Parameter | Symbol | Test Conditions | Min (Note 14) | Typ | Max (Note 14) | Unit |
|---|-----------------------------|---|------------------------------------|-----|------------------------------------|------|
| Minimum Pulse Width Required from External SYNC Clock | EXT_SYNC _{PW} | Measured at 50% amplitude | 150 | | | ns |
| Drift Tolerance for External SYNC Clock | EXT_SYNC _{DRIFT} | External SYNC clock equal to 500kHz is not supported | -10 | | 10 | % |
| Logic Input/Output Characteristics | | | | | | |
| Bias Current at the Logic Input Pins | I _{LOGIC_BIAS} | EN1/2, PG1/2, SA/UVLO, SCL, SDA, ASCR1/2, SS/TRACK, SYNC/OCP, VSET_CRS1/2, VSET_FINE1/2 | -100 | | +100 | nA |
| Logic Input Low Threshold Voltage | V _{LOGIC_IN_LOW} | | | | 0.8 | V |
| Logic Input High Threshold Voltage | V _{LOGIC_IN_HIGH} | | 2.0 | | | V |
| Logic Output Low Threshold Voltage | V _{LOGIC_OUT_LOW} | 2mA sinking | | | 0.5 | V |
| Logic Output High Threshold Voltage | V _{LOGIC_OUT_HIGH} | 2mA sourcing | 2.25 | | | V |
| PMBus Interface Timing Characteristic | | | | | | |
| PMBus Operating Frequency | f _{SMB} | | 100 | | 400 | kHz |

Notes:

14. Compliance to datasheet limits is assured by one or more methods: Production test, characterization, and/or design. Controller is independently tested before module assembly.
15. V_{OUT} measured at the termination of the VSEN1/2P and VSEN1/2N sense points.
16. The MAX load current is determined by the thermal ["Derating Curves" on page 19](#).
17. "FS" stands for full scale of recommended maximum operation range.

3. Typical Performance Curves

3.1 Efficiency Performance

Operating condition: $T_A = +25^\circ\text{C}$, no air flow. $C_{OUT} = 1 \times 470\mu\text{F}$ POSCAP + $12 \times 100\mu\text{F}$ Ceramic. Typical values are used unless otherwise noted. The efficiency curves were measured on the evaluation board. For test conditions, refer to [Table 1 on page 6](#).

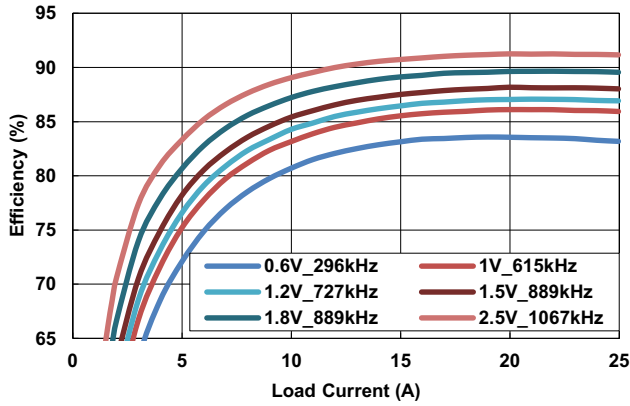


Figure 5. Single Channel Efficiency vs Output Current

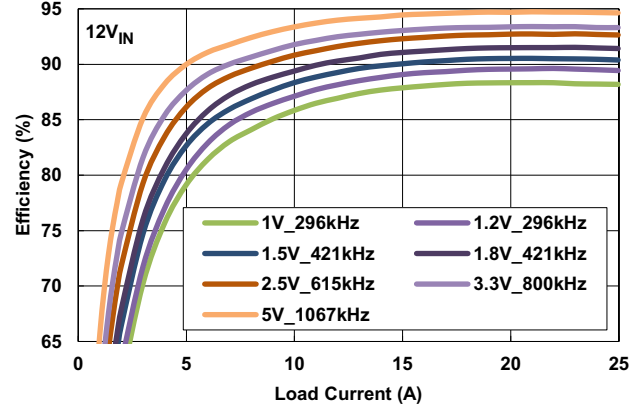


Figure 6. Single Channel Efficiency vs Output Current

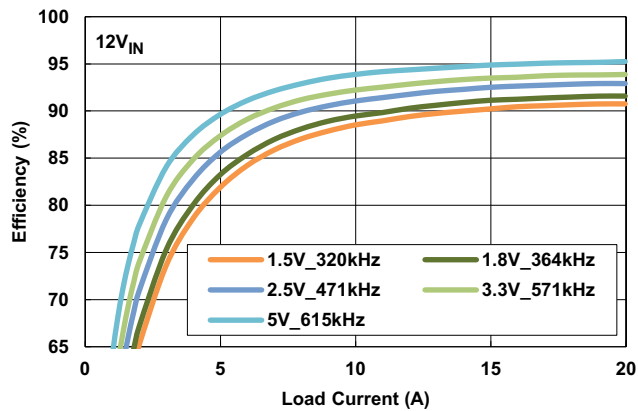


Figure 7. Single Channel Efficiency vs Output Current

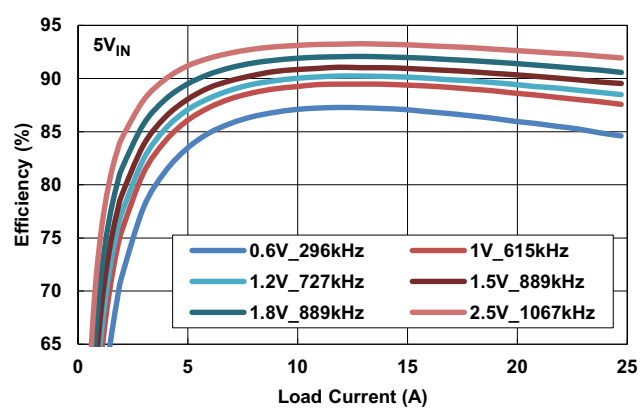


Figure 8. Single Channel Efficiency vs Output Current

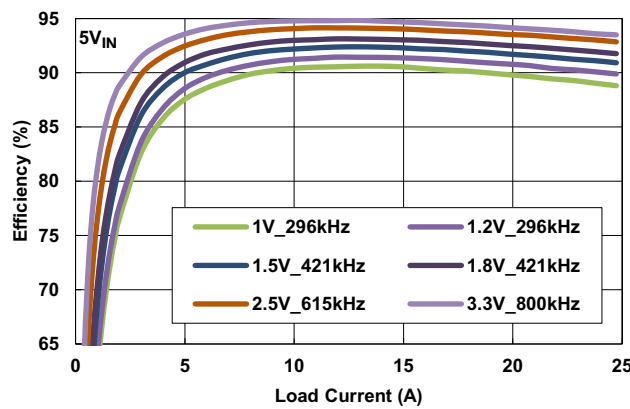


Figure 9. Single Channel Efficiency vs Output Current

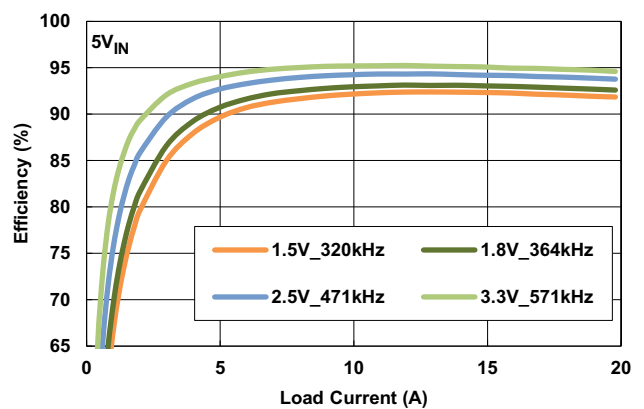


Figure 10. Single Channel Efficiency vs Output Current

3.2 Startup and Shutdown

Operating condition: $T_A = +25^\circ\text{C}$, no air flow. $C_{OUT} = 1 \times 470\mu\text{F POSCAP} + 12 \times 100\mu\text{F Ceramic}$. Typical values are used unless otherwise noted.

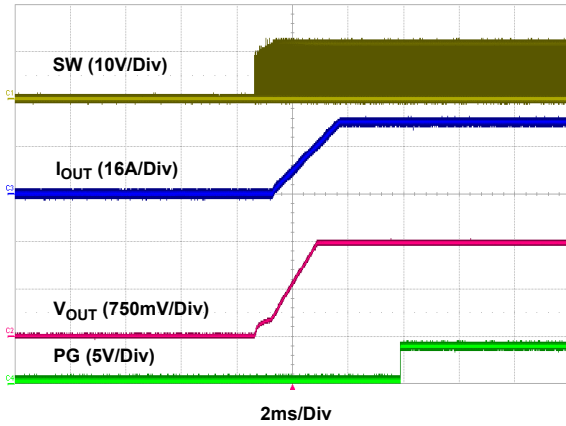


Figure 11. Single Channel Startup $12V_{IN}$, $1.5V_{OUT}$, 25A

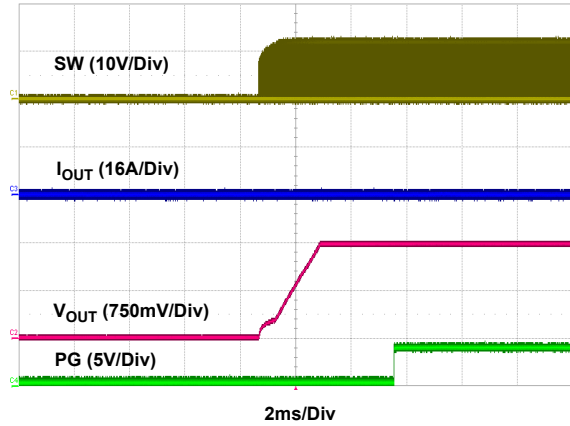


Figure 12. Single Channel Startup $12V_{IN}$, $1.5V_{OUT}$, 0A

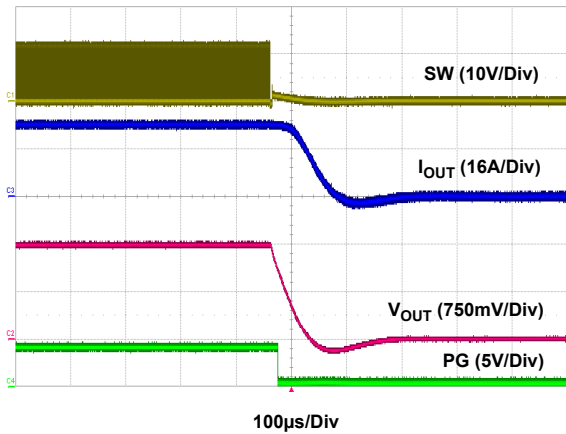


Figure 13. Single Channel Shutdown $12V_{IN}$, $1.5V_{OUT}$, 25A

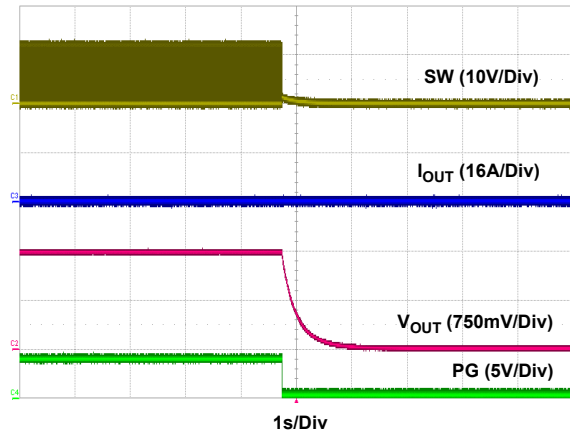


Figure 14. Single Channel Shutdown $12V_{IN}$, $1.5V_{OUT}$, 0A

3.3 Derating Curves

All of the following curves were plotted at $T_J = +125^\circ\text{C}$. The derating curves were measured on the evaluation board. For test conditions, refer to [Table 1 on page 6](#). Load current is applied per channel, two channels are operating at the same time.

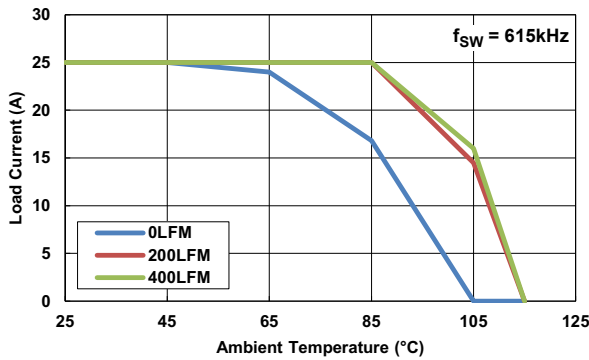


Figure 15. 12V_{IN} to 1V_{OUT}

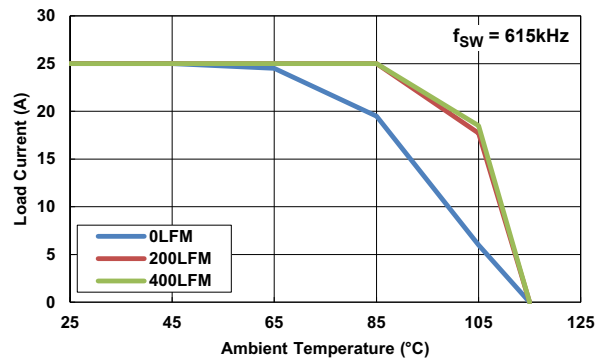


Figure 16. 5V_{IN} to 1V_{OUT}

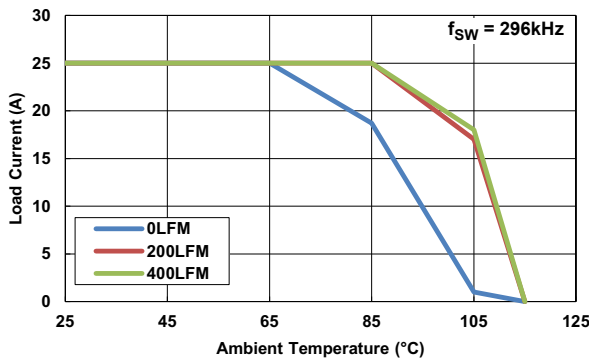


Figure 17. 12V_{IN} to 1V_{OUT}

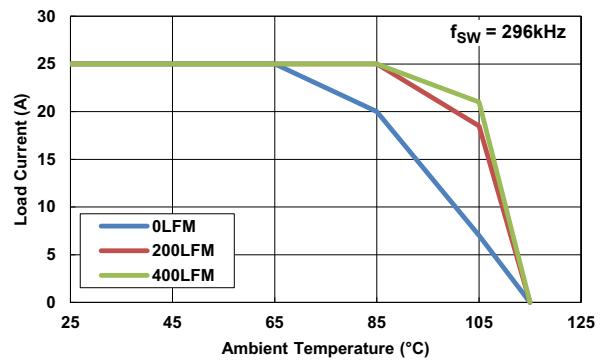


Figure 18. 5V_{IN} to 1V_{OUT}

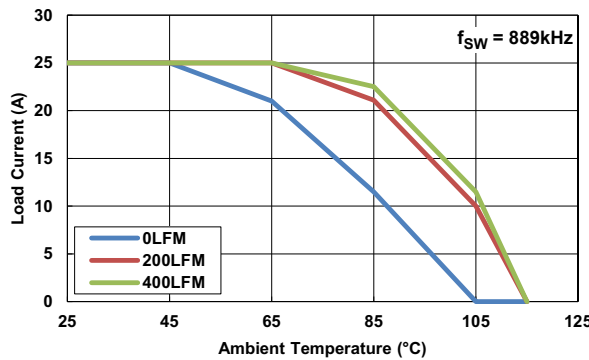


Figure 19. 12V_{IN} to 1.5V_{OUT}

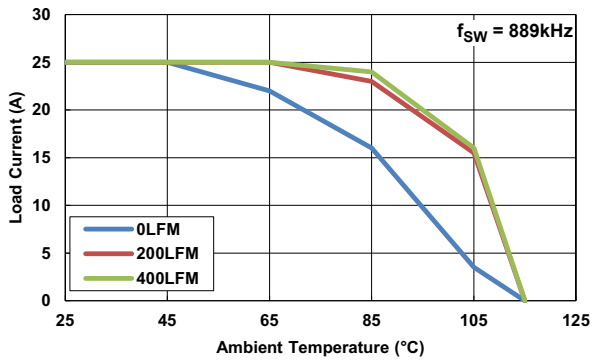


Figure 20. 5V_{IN} to 1.5V_{OUT}

All of the following curves were plotted at $T_J = +125^\circ\text{C}$. The derating curves were measured on the evaluation board. For test conditions, refer to [Table 1 on page 6](#). Load current is applied per channel, two channels are operating at the same time. **(Continued)**

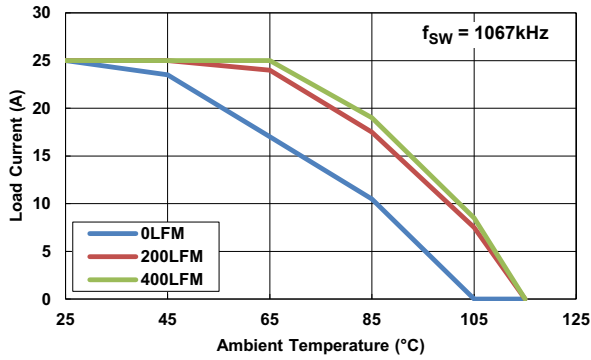


Figure 21. 12V_{IN} to 2.5V_{OUT}

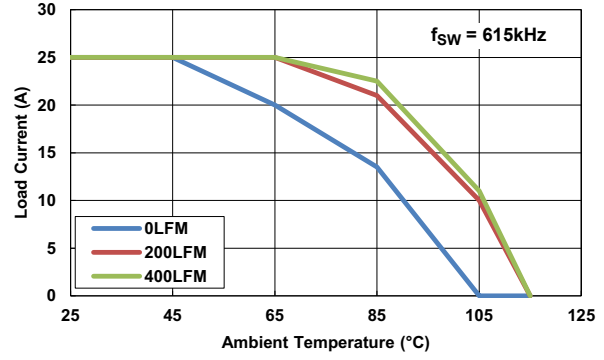


Figure 22. 5V_{IN} to 2.5V_{OUT}

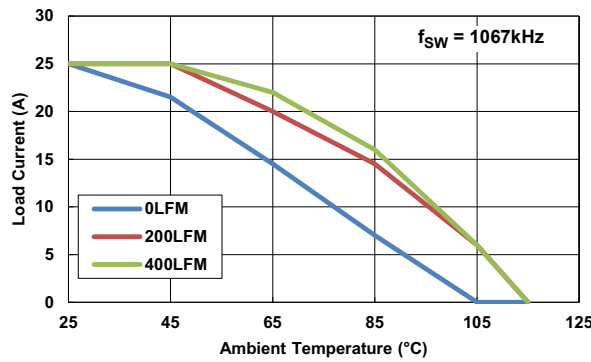


Figure 23. 12V_{IN} to 5V_{OUT}

3.4 Transient Response Performance

Operating condition: $T_A = +25^{\circ}\text{C}$, no air flow. Refer to [Table 1 on page 6](#) for output capacitor and ASCR settings. Typical values are used unless otherwise noted.

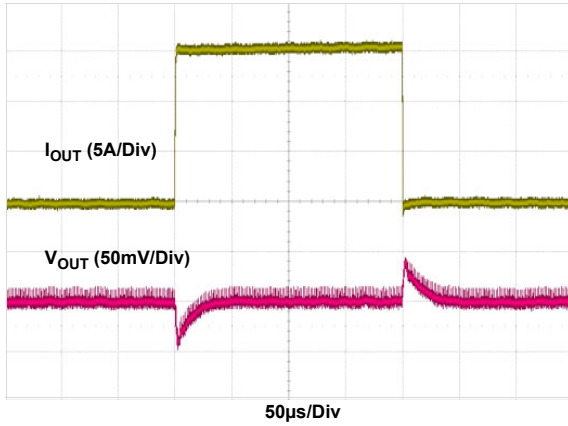


Figure 24. 0A-15A, 15A/µs, 12V_{IN}, 1V_{OUT}, 615kHz

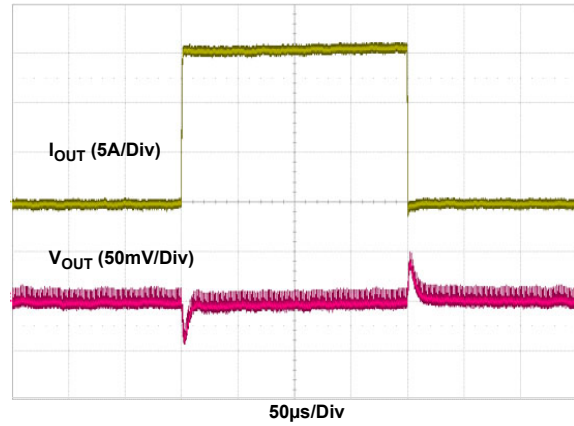


Figure 25. 0A-15A, 15A/µs, 12V_{IN}, 1.5V_{OUT}, 889kHz

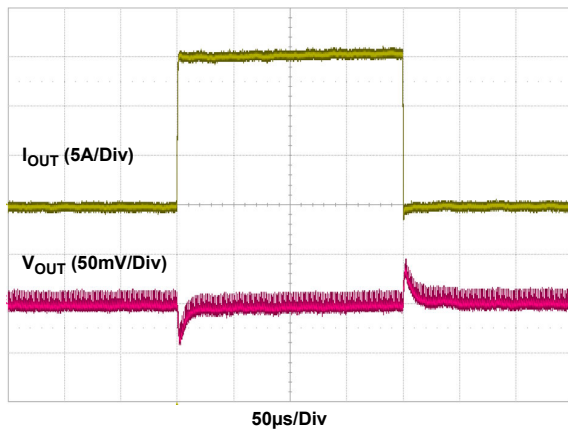


Figure 26. 0A-15A, 15A/µs, 12V_{IN}, 1.8V_{OUT}, 889kHz

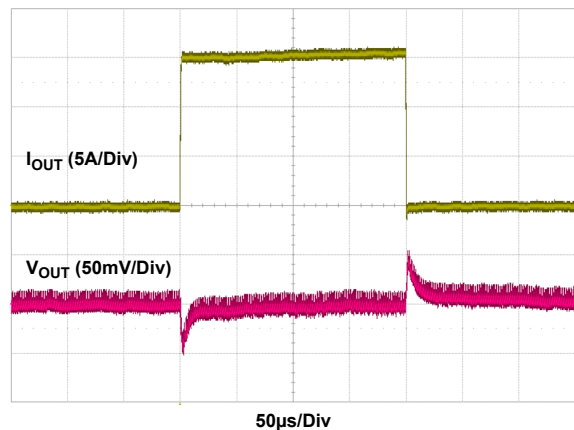


Figure 27. 0A-15A, 15A/µs, 12V_{IN}, 2.5V_{OUT}, 1067kHz

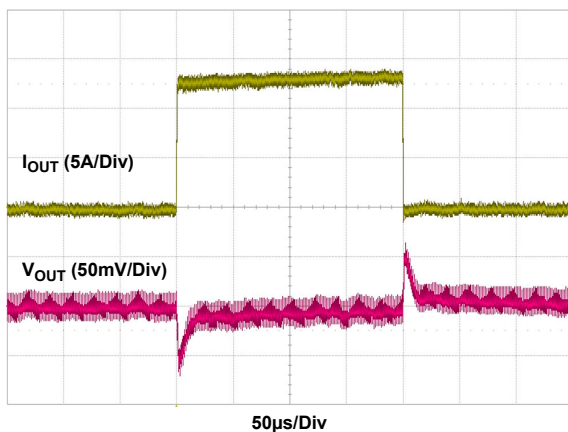


Figure 28. 0A-12.5A, 15A/µs, 12V_{IN}, 3.3V_{OUT}, 800kHz

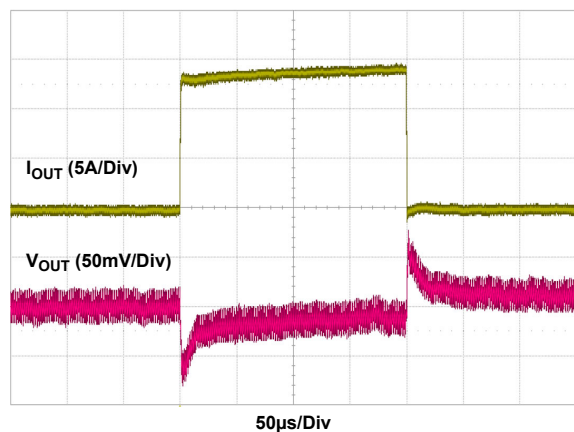


Figure 29. 0A-12.5A, 15A/µs, 12V_{IN}, 5V_{OUT}, 1067kHz

4. Functional Description

4.1 SMBus Communications

The RAA210925 provides a SMBus digital interface that enables the user to configure the module operation as well as monitor the input and output parameters. The RAA210925 can be used with any SMBus host device. In addition, the module is compatible with PMBus Power System Management Protocol Specification Parts I and II version 1.2. The RAA210925 accepts most standard PMBus commands. When configuring the device using PMBus commands, it is recommended that the enable pin is tied to SGND.

The SMBus device address is the only parameter that must be set by the external pins.

The RAA210925 can operate without the PMBus in pin-strap mode with configurations programmed by pin-strap resistors, such as output voltage, ASCR setting, switching frequency, OCP limit, device SMBus address, input UVLO, soft-start/stop, and tracking.

4.2 Output Voltage Selection

The output voltages of both channels may be set to a voltage between 0.6V and 5V if the input voltage is higher than the desired output voltage by an amount sufficient to maintain regulation.

The VSET_CRS1/2 (V_{OUT} Coarse) and VSET_FINE1/2 (V_{OUT} Fine) pins set the output voltage. A standard 1% resistor is required. Placing a resistor between VCRS1/2 and SGND based on [Table 4](#) determines the VCRS value.

If higher resolution is desired, the VSET_FINE pin can be used to fine-tune the output voltage settings according to the following command set:

$$VOUT_COMMAND = \begin{cases} VOUT_CRS + 5mV \cdot N, & \text{if } 0.6V \leq VOUT_CRS < 1.4V \\ VOUT_CRS + 10mV \cdot N, & \text{if } 1.4V \leq VOUT_CRS < 2V \\ VOUT_CRS + 50mV \cdot N, & \text{if } 2V \leq VOUT_CRS < 3.6V \\ VOUT_CRS + 100mV \cdot N, & \text{if } 3.6V \leq VOUT_CRS < 5V \\ VOUT_CRS, & \text{if } VOUT_CRS = 5V \end{cases}$$

Use the resistors values from [Table 5 on page 23](#) to set the appropriate value of N for calculating the final output voltage.

Table 4. Output Voltage Resistor Settings

| VOUT_CRS1/2 (V) | RSET1/2 (kΩ) |
|-----------------|--------------|
| 1 | LOW |
| 1.5 | OPEN |
| 3.3 | HIGH |
| 0.6 | 10 |
| 0.675 | 11 |
| 0.7 | 12.1 |
| 0.72 | 13.3 |
| 0.75 | 14.7 |
| 0.8 | 16.2 |
| 0.85 | 17.8 |
| 0.9 | 19.6 |
| 0.93 | 21.5 |
| 0.95 | 23.7 |
| 0.98 | 26.1 |
| 1.03 | 28.7 |

Table 4. Output Voltage Resistor Settings (Continued)

| VOUT_CRS1/2 (V) | RSET1/2 (kΩ) |
|-----------------|--------------|
| 1.05 | 31.6 |
| 1.1 | 34.8 |
| 1.12 | 38.3 |
| 1.15 | 42.2 |
| 1.2 | 46.4 |
| 1.25 | 51.1 |
| 1.3 | 56.2 |
| 1.35 | 61.9 |
| 1.4 | 68.1 |
| 1.65 | 75 |
| 1.8 | 82.5 |
| 1.85 | 90.9 |
| 2 | 100 |
| 2.4 | 110 |
| 2.5 | 121 |
| 2.8 | 133 |
| 3 | 147 |
| 3.6 | 162 |
| 5 | 178 |

Table 5. VSET_FINE Settings

| N | RSET1/2 (kΩ) |
|----|--------------------------|
| 0 | 10, or OPEN |
| 1 | 11 |
| 2 | 12.1 |
| 3 | 13.3 |
| 4 | 14.7 |
| 5 | 16.2 |
| 6 | 17.8 |
| 7 | 19.6 |
| 8 | 21.5 |
| 9 | 23.7, or connect to SGND |
| 10 | 26.1 |
| 11 | 28.7 |
| 12 | 31.6 |
| 13 | 34.8 |
| 14 | 38.3 |
| 15 | 42.2 |
| 16 | 46.4 |
| 17 | 51.1 |
| 18 | 56.2 |

Table 5. VSET_FINE Settings (Continued)

| N | RSET1/2 (k Ω) |
|----|-------------------------|
| 19 | 61.9 |
| 20 | 68.1, or connect to V25 |

The output voltage may also be set to any value between 0.6V and 5V using the PMBus command VOUT_COMMAND.

By default, V_{OUT_MAX} is set to 110% of V_{OUT} set by the pin-strap resistor, which can be changed to any value up to 5.5V by the PMBus Command VOUT_MAX.

4.3 Soft-Start, Stop Delay, and Ramp Times

The RAA210925 follows an internal start-up procedure after power is applied to the VDD pin. The module requires approximately 60ms to 70ms to check for specific values stored in its internal memory and programmed by pin-strap resistors. When this process is completed, the device is ready to accept commands through the PMBus interface and the module is ready to be enabled. If the module is to be synchronized to an external clock source, the clock frequency must be stable before asserting the EN pin.

It may be necessary to set a delay from when an enable signal is received until the output voltage starts to ramp to its target value. In addition, the designer may wish to precisely set the time required for V_{OUT} to ramp to its target value after the delay period has expired. These features can be used as part of an overall inrush current management strategy or to precisely control how fast a load IC is turned on. The RAA210925 gives the system designer several options for precisely and independently controlling both the delay and ramp time periods. The soft-start delay period begins when the EN pin is asserted and ends when the delay time expires.

The soft-start delay (TON_DELAY) and ramp-up time (TON_RISE) can be set to custom values with pin-strap resistors or PMBus. When the delay time is set to 0ms, the device begins its ramp-up after the internal circuitry has initialized (approximately 2ms). When the soft-start ramp period is set to 0ms, the output ramps up as quickly as the output load capacitance and loop settings allow. It is generally recommended to set the soft-start ramp to a value greater than 2ms to prevent inadvertent fault conditions due to excessive inrush current.

Similar to the soft-start delay and ramp-up time, the delay (TOFF_DELAY) and ramp-down time (TOFF_FALL) for soft-stop/off can be set to custom values with pin-strap resistors or PMBus. In addition, the module can be configured as “immediate off” using the command ON_OFF_CONFIG, so that the internal MOSFETs are turned off immediately after the delay time expires.

The SS/TRACK pin can be used to program the soft-start/stop delay time and ramp time to some typical values as well as enable/disable the tracking function shown in [Table 6 on page 24](#).

Table 6. Soft-Start/Stop and Tracking Resistor Settings

| TON_DELAY TOFF_DELAY (ms) | | TON_RISE TOFF_FALL (ms) | | Tracking | | R (k Ω) |
|------------------------------|-----|----------------------------|-----|----------|-----|-----------------|
| Ch1 | Ch2 | Ch1 | Ch2 | Ch1 | Ch2 | |
| 5 | 5 | 2 | 2 | No | No | LOW |
| 5 | 5 | 2 | 5 | No | No | OPEN |
| 5 | 5 | 5 | 2 | No | No | HIGH |
| 5 | 5 | 5 | 5 | No | No | 10 |
| 5 | 10 | 2 | 2 | No | No | 11 |
| 5 | 10 | 2 | 5 | No | No | 12.1 |
| 5 | 10 | 5 | 2 | No | No | 13.3 |
| 5 | 10 | 5 | 5 | No | No | 14.7 |
| 10 | 5 | 2 | 2 | No | No | 16.2 |

Table 6. Soft-Start/Stop and Tracking Resistor Settings (Continued)

| TON_DELAY TOFF_DELAY (ms) | | TON_RISE TOFF_FALL (ms) | | Tracking | | R (kΩ) |
|------------------------------|-----|----------------------------|-----|------------|------------|--------|
| Ch1 | Ch2 | Ch1 | Ch2 | Ch1 | Ch2 | |
| 10 | 5 | 2 | 5 | No | No | 17.8 |
| 10 | 5 | 5 | 2 | No | No | 19.6 |
| 10 | 5 | 5 | 5 | No | No | 21.5 |
| 20 | 5 | 2 | 2 | No | No | 23.7 |
| 20 | 5 | 5 | 5 | No | No | 26.1 |
| 5 | 20 | 2 | 2 | No | No | 28.7 |
| 5 | 20 | 2 | 5 | No | No | 31.6 |
| 5 | 20 | 5 | 2 | No | No | 34.8 |
| 5 | 20 | 5 | 5 | No | No | 38.3 |
| 5 | N/A | 2 | N/A | No | Track 100% | 42.2 |
| 5 | N/A | 2 | N/A | No | Track 50% | 46.4 |
| 5 | N/A | 5 | N/A | No | Track 100% | 51.1 |
| 5 | N/A | 5 | N/A | No | Track 50% | 56.2 |
| 10 | N/A | 2 | N/A | No | Track 100% | 61.9 |
| 10 | N/A | 2 | N/A | No | Track 50% | 68.1 |
| 10 | N/A | 5 | N/A | No | Track 100% | 75 |
| 10 | N/A | 5 | N/A | No | Track 50% | 82.5 |
| N/A | 5 | N/A | 2 | Track 100% | No | 90.9 |
| N/A | 5 | N/A | 2 | Track 50% | No | 100 |
| N/A | 5 | N/A | 5 | Track 100% | No | 110 |
| N/A | 5 | N/A | 5 | Track 50% | No | 121 |
| N/A | 10 | N/A | 2 | Track 100% | No | 133 |
| N/A | 10 | N/A | 2 | Track 50% | No | 147 |
| N/A | 10 | N/A | 5 | Track 100% | No | 162 |
| N/A | 10 | N/A | 5 | Track 50% | No | 178 |

4.4 Voltage Tracking

Numerous high performance systems place stringent demands on the order in which the power supply voltages are turned on. This is particularly true when powering FPGAs, ASICs, and other advanced processor devices that require multiple supply voltages to power a single die. In most cases, the I/O interface operates at a higher voltage than the core and therefore, the core supply voltage must not exceed the I/O supply voltage according to the manufacturers' specifications.

The RAA210925 integrates a tracking scheme that allows one of its outputs (Channel 1 or Channel 2) to track a voltage that is applied to the VTRKP and VTRKN pins with no external components required. The VTRKP and VTRKN pins are analog inputs that, when the tracking mode is enabled, configure the voltage applied to the VTRKP and VTRKN pins to act as a reference for the device's output regulation.

Figure 30 illustrates the typical connection and the two tracking modes:

- **Coincident** - This mode configures the RAA210925 to ramp its output voltage at the same rate as the voltage applied to the VTRK pin until it reaches its desired output voltage. The device that is tracking another output voltage (slave) must be set to its desired steady-state output voltage.
- **Ratiometric** - This mode configures the RAA210925 to ramp its output voltage at a rate that is a percentage of the voltage applied to the VTRKP and VTRKN pins. The device that is tracking another output voltage (slave) must be set to its desired steady-state output voltage.

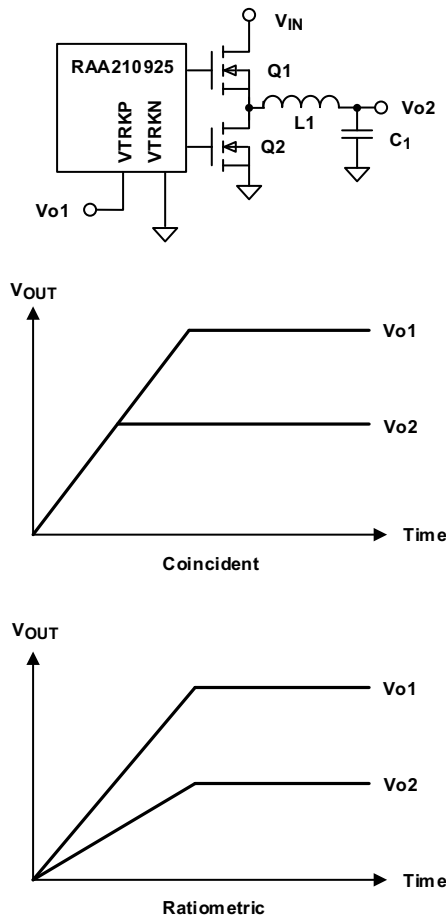


Figure 30. Tracking Modes

The master RAA210925 device in a tracking group is defined as the device that has the highest target output voltage within the group. This master device controls the ramp rate of all tracking devices and is not configured for tracking mode. The maximum tracking rise time is 1V/ms. The slave device must be enabled before the master.

Any device that is configured for tracking mode ignores its TON_DELAY and TON_RISE settings and its output takes on the turn-on/turn-off characteristics of the reference voltage present at the VTRKP and VTRKN pins.

The VOUT_COMMAND needs to be set the same as the target tracking voltage when tracking is enabled. For example, the VOUT_COMMAND of the Page1 (VOUT2 which enables the tracking) needs to set to 1V if tracking 100% is selected and a ramp of 1V is applied to VTRKP and VTRKN. The VOUT_COMMAND of Page 1 (VOUT2 which enables the tracking) needs to set to 1V if tracking 50% is selected and a ramp of 2V is applied to VTRKP and VTRKN. In Tracking mode, the minimum voltage that can be tracked is ~200mV.

4.5 Power-Good

The RAA210925 provides a Power-Good (PG) signal that indicates the output voltage is within a specified tolerance of its target level and no fault condition exists. By default, the PG pin asserts if the output is within 10% of the target voltage. This limit may be changed using the PMBus command POWER_GOOD_ON.

A PG delay period is defined as the time from when all conditions within the RAA210925 for asserting PG are met to when the PG pin is actually asserted. This feature is commonly used instead of using an external reset controller to control external digital logic. A fixed PG delay of 3ms is programmed for the RAA210925.

4.6 Switching Frequency and PLL

The device's switching frequency is set from 296kHz to 1067kHz using the pin-strap method (combined with the average OCP limit setting) as shown in [Table 7](#), or by using the PMBus command FREQUENCY_SWITCH.

Table 7. Switching Frequency and OCP Limit Resistor Setting

| Switching Frequency | | OCP Avg | |
|---------------------|-----------------------|---------|---------|
| R (kΩ) | f _{sw} (kHz) | Ch1 (A) | Ch2 (A) |
| LOW | 296 | 35 | 35 |
| OPEN | 889 | 35 | 35 |
| HIGH | 1067 | 35 | 35 |
| 10 | 296 | 30 | 35 |
| 11 | 296 | 30 | 30 |
| 12.1 | 296 | 25 | 35 |
| 13.3 | 296 | 25 | 30 |
| 14.7 | 296 | 25 | 25 |
| 16.2 | 320 | 25 | 35 |
| 17.8 | 320 | 25 | 30 |
| 19.6 | 320 | 25 | 25 |
| 21.5 | 320 | 20 | 30 |
| 23.7 | 320 | 20 | 25 |
| 26.1 | 364 | 25 | 35 |
| 28.7 | 364 | 25 | 30 |
| 31.6 | 364 | 20 | 30 |
| 34.8 | 421 | 30 | 35 |
| 38.3 | 421 | 30 | 30 |
| 42.2 | 471 | 25 | 35 |
| 46.4 | 471 | 25 | 30 |
| 51.1 | 471 | 20 | 35 |
| 56.2 | 571 | 25 | 35 |

Table 7. Switching Frequency and OCP Limit Resistor Setting (Continued)

| Switching Frequency | | OCP Avg | |
|---------------------|-----------------------|---------|---------|
| R (kΩ) | f _{sw} (kHz) | Ch1 (A) | Ch2 (A) |
| 61.9 | 571 | 25 | 30 |
| 68.1 | 571 | 20 | 35 |
| 75 | 571 | 20 | 30 |
| 82.5 | 615 | 35 | 35 |
| 90.9 | 615 | 35 | 30 |
| 100 | 615 | 30 | 30 |
| 110 | 615 | 25 | 35 |
| 121 | 615 | 25 | 30 |
| 133 | 615 | 25 | 25 |
| 147 | 727 | 35 | 35 |
| 162 | 800 | 30 | 35 |
| 178 | 800 | 30 | 30 |

The RAA210925 incorporates an internal Phase-Locked Loop (PLL) to clock the internal circuitry. The PLL can be driven by an external clock source connected to the SYNC pin. This configuration can be achieved by connecting a resistor to the CFG pin. If the clock source is set to internal, the internal frequency is set according to the SYNC pin resistor settings. If the clock source is programmed to external, the internal frequency is set according to the resistor connected to the CFG pin as shown in [Table 8](#). The external clock signal must not vary more than 10% from its initial value and should have a minimum pulse width of 150ns. For the external clock source, OCP AVG is 25A for CH1 and CH2, which is not adjustable.

Table 8. External Frequency SYNC Settings

| Clock Source | Internal FREQUENCY_SWITCH (kHz) | R _{SET} (kΩ) |
|--------------|---------------------------------|--------------------------|
| Internal | Determined by SYNC resistor | 10, or OPEN |
| External | 296 | 11 |
| External | 340 | 12.1 |
| External | 390 | 13.3 |
| External | 444 | 14.7 |
| External | 516 | 16.2, or connect to SGND |
| External | 593 | 17.8 |
| External | 696 | 19.6 |
| External | 800 | 21.5 |
| External | 941 | 23.7 |
| External | 1067 | 26.1, or connect to V25 |

4.7 Output Overcurrent Protection

The RAA210925 is protected from damage if the output is shorted to ground or if an overload condition is imposed on the output. Average output overcurrent fault threshold can be programmed by the PMBus command IOUT_AVG_OC_FAULT_LIMIT while the peak output overcurrent fault threshold can be programmed by the PMBus command IOUT_OC_FAULT_LIMIT. The default response from an average overcurrent fault is an immediate shutdown with a continuous retry of 70ms delay. A hard bound of 50A is applied to the peak overcurrent limit.

The average OCP limit can be set by the SYNC/OCP pin strap. Refer to [Table 7 on page 27](#) for more information.

4.8 Loop Compensation

The module loop response can be set by using the pin-strap method (ASCR1/2 pins) according to [Table 9](#) or through the PMBus command ASCR_CONFIG. The RAA210925 uses the ChargeMode control algorithm that responds to the output current changes within a single PWM switching cycle, achieving a smaller total output voltage variation with less output capacitance than traditional PWM controllers.

Table 9. ASCR Resistor Setting

| ASCR Gain | ASCR Residual | R (k Ω) |
|-----------|---------------|-----------------|
| 350 | 110 | LOW |
| 525 | 90 | OPEN |
| 475 | 80 | HIGH |
| 100 | 90 | 10 |
| 100 | 100 | 11 |
| 120 | 90 | 12.1 |
| 120 | 100 | 13.3 |
| 140 | 90 | 14.7 |
| 140 | 100 | 16.2 |
| 140 | 110 | 17.8 |
| 150 | 90 | 19.6 |
| 150 | 100 | 21.5 |
| 160 | 70 | 23.7 |
| 160 | 90 | 26.1 |
| 175 | 80 | 28.7 |
| 175 | 90 | 31.6 |
| 200 | 90 | 34.8 |
| 200 | 100 | 38.3 |
| 200 | 110 | 42.2 |
| 225 | 80 | 46.4 |
| 225 | 90 | 51.1 |
| 250 | 80 | 56.2 |
| 250 | 100 | 61.9 |
| 275 | 90 | 68.1 |
| 275 | 100 | 75 |
| 300 | 70 | 82.5 |
| 300 | 90 | 90.9 |
| 350 | 100 | 100 |

Table 9. ASCR Resistor Setting (Continued)

| ASCR Gain | ASCR Residual | R (kΩ) |
|-----------|---------------|--------|
| 450 | 100 | 110 |
| 450 | 110 | 121 |
| 500 | 70 | 133 |
| 500 | 90 | 147 |
| 600 | 100 | 162 |
| 600 | 110 | 178 |

4.9 SMBus Module Address Selection

Each module must have its own unique serial address to distinguish between other devices on the bus. The module address is set by connecting a resistor between pins SA/UVLO and SGND. The SA/UVLO pin also defines the input undervoltage lockout limit. The input Undervoltage Lockout (UVLO) prevents the RAA210925 from operating when the input falls below a preset threshold, indicating the input supply is out of its specified range. The UVLO threshold (V_{UVLO}) can be set between 4.18V and 16V by using the PMBus command VIN_UV_FAULT_LIMIT.

[Table 10](#) lists the available module addresses. A standard 1% resistor is required. When the UVLO threshold is hit, the module shuts down immediately. The fault needs to be cleared for the module to restart.

Table 10. SMBus Address and UVLO Resistor Setting

| PMBus Address | UVLO (V) | R (kΩ) |
|---------------|----------|--------|
| 26h | 4.2 | LOW |
| 28h | 4.5 | OPEN |
| 19h | 10.8 | 10 |
| 1Ah | 10.8 | 11 |
| 1Bh | 10.8 | 12.1 |
| 1Ch | 10.8 | 13.3 |
| 1Dh | 10.8 | 14.7 |
| 1Eh | 10.8 | 16.2 |
| 1Fh | 10.8 | 17.8 |
| 20h | 4.2 | 19.6 |
| 21h | 4.2 | 21.5 |
| 22h | 4.2 | 23.7 |
| 23h | 4.2 | 26.1 |
| 24h | 4.2 | 28.7 |
| 25h | 4.2 | 31.6 |
| 26h | 4.2 | 34.8 |
| 27h | 4.2 | 38.3 |
| 28h | 4.5 | 42.2 |
| 29h | 4.5 | 46.4 |
| 2Ah | 4.5 | 51.1 |
| 2Bh | 4.5 | 56.2 |
| 2Ch | 4.5 | 61.9 |
| 2Dh | 4.5 | 68.1 |
| 2Eh | 4.5 | 75 |

Table 10. SMBus Address and UVLO Resistor Setting (Continued)

| PMBus Address | UVLO (V) | R (kΩ) |
|---------------|----------|--------|
| 2Fh | 4.5 | 82.5 |
| 30h | 10.8 | 90.9 |
| 31h | 10.8 | 100 |
| 32h | 10.8 | 110 |
| 33h | 10.8 | 121 |
| 34h | 10.8 | 133 |
| 35h | 10.8 | 147 |
| 36h | 10.8 | 162 |
| 37h | 10.8 | 178 |

4.10 Output Overvoltage Protection

The RAA210925 offers an internal output overvoltage protection circuit that can be used to protect sensitive load circuitry from being subjected to a voltage higher than its prescribed limits. A hardware comparator compares the actual output voltage (seen at pins VSEN1/2P, VSEN1/2N) to a threshold set to 15% higher than the target output voltage (default setting). The fault threshold can be set to a desired level by the PMBus command VOUT_OV_FAULT_LIMIT. If the VSEN1/2P, VSEN1/2N voltage exceeds this threshold, the module initiates an immediate shutdown without retry.

Internal to the module, two 100Ω resistors are populated from V_{OUT} to VSEN1/2P and SGND to VSEN1/2N to protect the module from overvoltage conditions in case of open at the voltage sensing pins and differential remote sense traces due to assembly error. As long as differential remote sense traces have low resistance, V_{OUT} regulation accuracy is not compromised.

4.11 Output Prebias Protection

An output prebias condition exists when an externally applied voltage is present on a power supply's output before the power supply's control IC is enabled. Certain applications require that the converter not be allowed to sink current during start-up if a prebias condition exists at the output. The RAA210925 provides prebias protection by sampling the output voltage before initiating an output ramp.

If a prebias voltage lower than the target voltage exists after the preconfigured delay period has expired, the target voltage is set to match the existing prebias voltage and both drivers are enabled. The output voltage is then ramped to the final regulation value at the preconfigured ramp rate.

The actual time the output takes to ramp from the prebias voltage to the target voltage varies, depending on the prebias voltage. However, the total time elapsed from when the delay period expires to when the output reaches its target value matches the preconfigured ramp time (see [Figure 31 on page 32](#)).

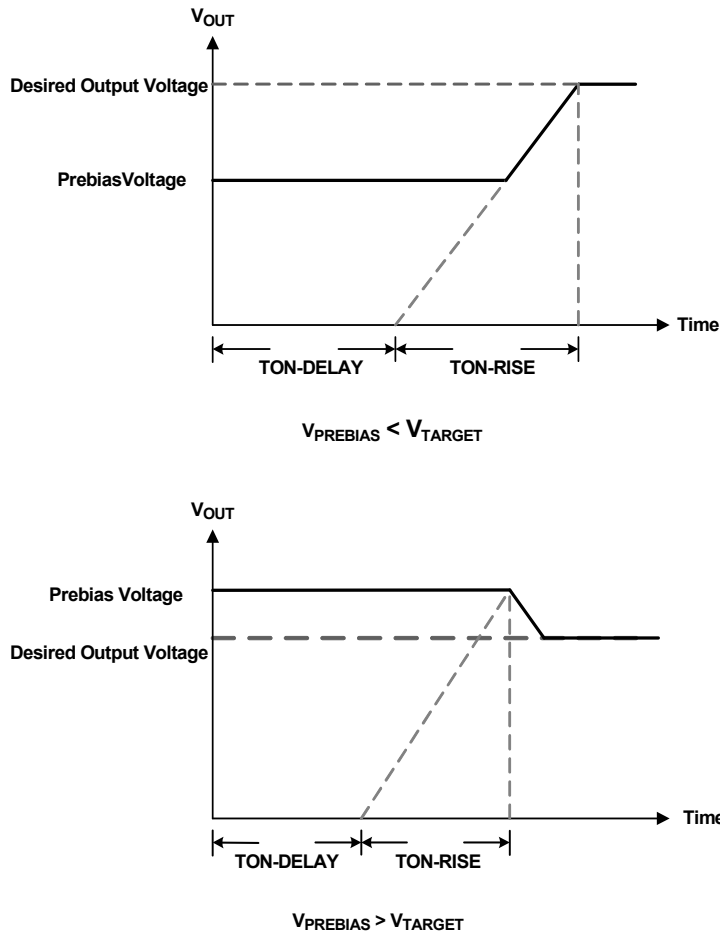


Figure 31. Output Responses to Prebias Voltages

If a prebias voltage is higher than the target voltage after the preconfigured delay period has expired, the target voltage is set to match the existing prebias voltage. Thus, both drivers are enabled with a PWM duty cycle that would ideally create the prebias voltage.

When the preconfigured soft-start ramp period has expired, the PG pin is asserted (assuming the prebias voltage is not higher than the overvoltage limit). The PWM then adjusts its duty cycle to match the original target voltage and the output ramps down to the preconfigured output voltage.

If a prebias voltage is higher than the overvoltage limit, the device does not initiate a turn-on sequence and declares an overvoltage fault condition.

4.12 Thermal Overload Protection

The RAA210925 includes a thermal sensor that continuously measures the internal temperature of the module and shuts down the controller when the temperature exceeds the preset limit. The factory default temperature limit is set to +115°C, and can be changed using the PMBus command OT_FAULT_LIMIT. Note that the temperature reading from the PMBus command is the temperature of the internal controller, which is lower than the junction temperature of the module.

The default response from an over-temperature fault is an immediate shutdown without retry.

4.13 Phase Spreading

When multiple point-of-load converters share a common DC input supply, adjust the clock phase offset of each device so that not all devices start to switch simultaneously. Setting each converter to start its switching cycle at a different point in time can dramatically reduce input capacitance requirements and efficiency losses. Because the peak current drawn from the input supply is effectively spread out over a period of time, the peak current drawn at any given moment is reduced, and the power losses proportional to the I_{RMS}^2 are reduced dramatically.

To enable phase spreading, all converters must be synchronized to the same switching clock. The phase offset of each device can also be set to any value between 0° and 360° in 22.5° increments by setting the device address appropriately as shown in [Table 11](#). This functionality can also be accessed using the PMBus command INTERLEAVE. The internal two phases of the module always maintain a phase difference of 180° . The phase offset between devices is determined from the lower four bits of the SMBus address of each interleaved device.

Table 11. INTERLEAVE

| SA | UVLO | R (k Ω) | SA in Binary | Low 4-Bits | INTERLEAVE-Ch1 | Phase Shift ($^\circ$) | Rail ID |
|-----|------|-----------------|--------------|------------|----------------|--------------------------|---------|
| 19h | 10.8 | 10 | 00011001 | 1001 | 9 | 202.5 | 25 |
| 1Ah | 10.8 | 11 | 00011010 | 1010 | 10 | 225 | 26 |
| 1Bh | 10.8 | 12.1 | 00011011 | 1011 | 11 | 247.5 | 27 |
| 1Ch | 10.8 | 13.3 | 00011100 | 1100 | 12 | 270 | 28 |
| 1Dh | 10.8 | 14.7 | 00011101 | 1101 | 13 | 292.5 | 29 |
| 1Eh | 10.8 | 16.2 | 00011110 | 1110 | 14 | 315 | 30 |
| 1Fh | 10.8 | 17.8 | 00011111 | 1111 | 15 | 337.5 | 31 |
| 20h | 4.2 | 19.6 | 00100000 | 0000 | 0 | 0 | 0 |
| 21h | 4.2 | 21.5 | 00100001 | 0001 | 1 | 22.5 | 1 |
| 22h | 4.2 | 23.7 | 00100010 | 0010 | 2 | 45 | 2 |
| 23h | 4.2 | 26.1 | 00100011 | 0011 | 3 | 67.5 | 3 |
| 24h | 4.2 | 28.7 | 00100100 | 0100 | 4 | 90 | 4 |
| 25h | 4.2 | 31.6 | 00100101 | 0101 | 5 | 112.5 | 5 |
| 26h | 4.2 | 34.8, low | 00100110 | 0110 | 6 | 135 | 6 |
| 27h | 4.2 | 38.3 | 00100111 | 0111 | 7 | 157.5 | 7 |
| 28h | 4.5 | 42.2, open | 00101000 | 1000 | 8 | 180 | 8 |
| 29h | 4.5 | 46.4 | 00101001 | 1001 | 9 | 202.5 | 9 |
| 2Ah | 4.5 | 51.1 | 00101010 | 1010 | 10 | 225 | 10 |
| 2Bh | 4.5 | 56.2 | 00101011 | 1011 | 11 | 247.5 | 11 |
| 2Ch | 4.5 | 61.9 | 00101100 | 1100 | 12 | 270 | 12 |
| 2Dh | 4.5 | 68.1 | 00101101 | 1101 | 13 | 292.5 | 13 |
| 2Eh | 4.5 | 75 | 00101110 | 1110 | 14 | 315 | 14 |
| 2Fh | 4.5 | 82.5 | 00101111 | 1111 | 15 | 337.5 | 15 |
| 30h | 10.8 | 90.9 | 00110000 | 0000 | 0 | 0 | 16 |
| 31h | 10.8 | 100 | 00110001 | 0001 | 1 | 22.5 | 17 |
| 32h | 10.8 | 110 | 00110010 | 0010 | 2 | 45 | 18 |
| 33h | 10.8 | 121 | 00110011 | 0011 | 3 | 67.5 | 19 |
| 34h | 10.8 | 133 | 00110100 | 0100 | 4 | 90 | 20 |
| 35h | 10.8 | 147 | 00110101 | 0101 | 5 | 112.5 | 21 |
| 36h | 10.8 | 162 | 00110110 | 0110 | 6 | 135 | 22 |
| 37h | 10.8 | 178 | 00110111 | 0111 | 7 | 157.5 | 23 |

4.14 Monitoring Using SMBus

The RAA210925 can monitor a wide variety of different system parameters using the PMBus commands:

- READ_VIN
- READ_VOUT
- READ_IOUT
- READ_INTERNAL_TEMP
- READ_DUTY_CYCLE
- READ_FREQUENCY

4.15 Snapshot Parameter Capture

The RAA210925 offers a special feature to capture parametric data and fault status following a fault. A detailed description is provided in the [“PMBus Commands Description” on page 41](#) under PMBus the commands SNAPSHOT and SNAPSHOT_CONTROL.

5. Layout Guide

To achieve stable operation, low losses, and good thermal performance, some layout considerations are necessary (Figure 32). Refer to the RTKA2109252H00000BU layout design.

- Establish separate SGND plane and PGND planes, then connect SGND to PGND plane on a middle layer and underneath PAD6 with a single point connection. For SGND and PGND pin connections, such as small pins H16, J16, M5 and M17..., use multiple vias for each pin to connect to inner SGND or PGND layers.
- Place enough ceramic capacitors between VIN and PGND, VOUT and PGND and bypass capacitors between VDD, VDRV and the ground plane, as close to the module as possible to minimize high frequency noise. It is very critical to place the output ceramic capacitors close to the VOUT pads and in the direction of the load current path in order to create a low impedance path for the high frequency inductor ripple current.
- Use large copper areas for power paths (VIN, PGND, VOUT) to minimize conduction loss and thermal stress. Also, use multiple vias to connect the power planes in different layers. It is recommended to enlarge PAD11 and PAD9 to place more vias on them. The ceramic capacitors CIN can be placed on the bottom layer under these two pads.
- Connect remote sensing traces to the regulation point to achieve a tight output voltage regulation and place the two traces in parallel. Route a trace from VSEN1/2N and VSEN1/2P to the point of load where the tight output voltage is desired. Avoid routing any sensitive signal traces, such as the VSENN, VSENP sensing lines near the SW pins.
- PAD14 and 16 (SW1 and SW2) are noisy pads, but they are beneficial for thermal dissipation. If the noise issue is critical for the applications, it is recommended to use only the top layer for the SW pads. For better thermal performance, use multiple vias on these pads to connect into the SW inner and bottom layers. However, use caution when placing a limited area of SW planes in any layer. The SW planes should avoid the sensing signals and should be surrounded by the PGND layer to avoid noise coupling.
- For pins SWD1 (L3) and SWD2 (P10), it is recommended to connect to the related SW1 and SW2 pads with short loop traces. The trace width should be more than 20 mils.

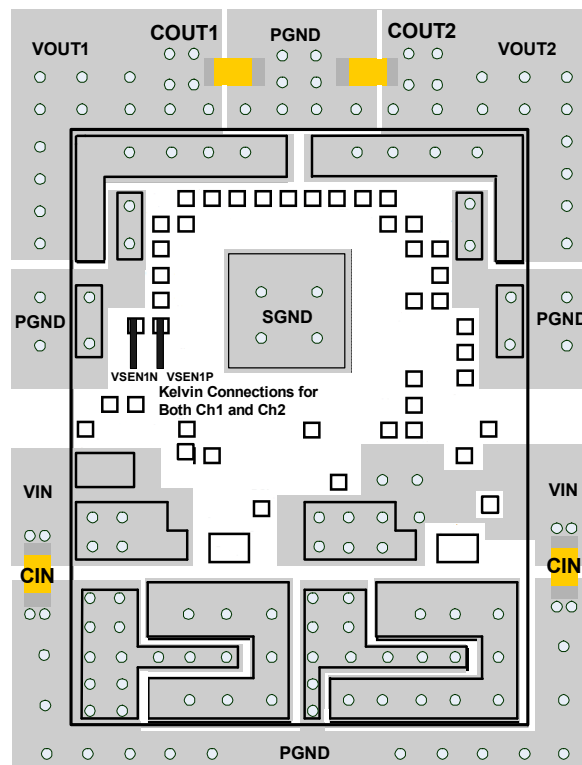


Figure 32. Recommended Layout

5.1 Thermal Considerations

Experimental power loss curves along with θ_{JA} from thermal modeling analysis can be used to evaluate the thermal consideration for the module. The derating curves are derived from the maximum power allowed while maintaining the temperature below the maximum junction temperature of +125°C. The derating curves are derived based on tests of the RTKA2109252H00000BU evaluation board, which is an 8-layer board 4.5x4inch in size with 2oz Cu on the top and bottom layers, 1oz Cu on the inner layers, and multiple via interconnects. In the actual application, other heat sources and design margins should be considered.

5.2 Package Description

The structure of the RAA210925 belongs to the High Density Array (HDA) no-lead package. This kind of package has advantages, such as good thermal and electrical conductivity, low weight, and small size. The HDA package is applicable for surface mounting technology and is being more readily used in the industry. The RAA210925 contains several types of devices, including resistors, capacitors, inductors, and control ICs. The RAA210925 is a copper leadframe based package with exposed copper thermal pads, which have good electrical and thermal conductivity. The copper leadframe and multicomponent assembly is overmolded with a polymer mold compound to protect these devices.

The package outline and typical PCB layout pattern design and typical stencil pattern design are shown on [pages 57](#) through [page 63](#). The module has a small size of 18mmx 23mmx 7.5mm.

5.3 PCB Layout Pattern Design

The bottom of RAA210925 is a leadframe footprint, which is attached to the PCB by the surface mounting process. The PCB layout pattern is shown on [pages 61](#) through [63](#). The PCB layout pattern is an array of solder mask defined PCB lands which align with the perimeters of the HDA exposed pads and I/O termination dimensions. The thermal lands on the PCB layout also feature an array of solder mask defined lands and should match 1:1 with the package exposed die pads.

5.4 Thermal Vias

A grid of 1.0mm to 1.2mm pitch thermal vias, which drops down and connects to buried copper plane(s), should be placed under the thermal land. The vias should be about 0.3mm to 0.33mm in diameter with the barrel plated to about 1.0 oz. of copper. Although adding more vias (by decreasing via pitch) improves the thermal performance, diminishing returns are seen as the number of vias is increased. Simply use as many vias as practical for the thermal land size and your board design rules allow. All vias should be capped and filled to avoid scavenging solder from the I/O solder joints and creating voids.

5.5 Stencil Pattern Design

Reflowed solder joints on the perimeter I/O lands should have about a 50 μ m to 75 μ m (2 mil to 3 mil) standoff height. The solder paste stencil design is the first step in developing optimized, reliable solder joints. Stencil aperture size to solder mask defined PCB land size ratio should typically be 1:1. The aperture width may be reduced slightly to help prevent solder bridging between adjacent I/O lands. To reduce solder paste volume on the larger thermal lands, it is recommended that an array of smaller apertures be used instead of one large aperture. It is recommended that the stencil printing area cover 50% to 80% of the PCB layout pattern. A typical solder stencil pattern is shown on [pages 59](#) through [61](#). The gap width between pads is 0.6mm. The user should consider the symmetry of the whole stencil pattern when designing its pads. A laser cut, stainless steel stencil with electropolished trapezoidal walls is recommended. Electropolishing “smooths” the aperture walls resulting in reduced surface friction and better paste release, which reduces voids. Using a Trapezoidal Section Aperture (TSA) also promotes paste release and forms a “brick like” paste deposit that assists in firm component placement. A 0.1mm to 0.15mm stencil thickness is recommended for this large pitch (1.3mm) HDA.

5.6 Reflow Parameters

Due to the low mount height of the HDA, “No-Clean” Type 3 solder paste per ANSI/J-STD-005 is recommended. A nitrogen purge is also recommended during reflow. A system board reflow profile depends on the thermal mass of the entire populated board, so it is not practical to define a specific soldering profile just for the HDA. The profile given in [Figure 33](#) is provided as a guideline, which can be customized for varying manufacturing practices and applications.

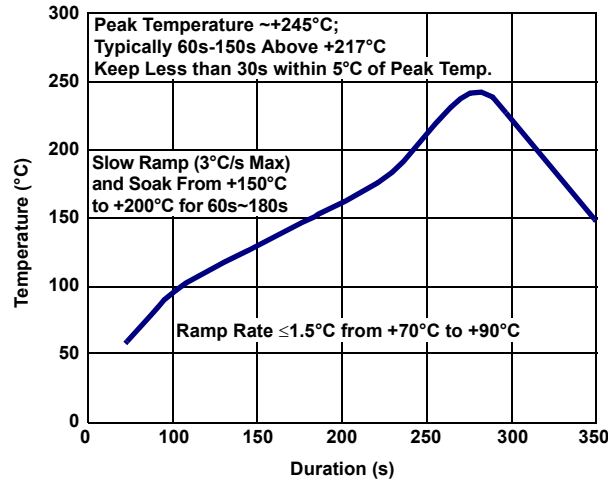


Figure 33. Typical Reflow Profile

6. PMBus Command Summary

| Command Code | Command Name | Description | Type | Data Format | Default Value | Default Setting | Page |
|--------------|---------------------|--|-----------|-------------|-----------------------|----------------------------------|--------------------|
| 00h | PAGE | Selects Controller 0, 1, or both. | R/W Byte | BIT | 00h | Channel 1 | 41 |
| 01h | OPERATION | Sets Enable and Disable modes. | R/W Byte | BIT | | | 41 |
| 02h | ON_OFF_CONFIG | Configures the EN pin and PMBus commands to turn the unit ON/OFF. | R/W Byte | BIT | 16h | Hardware Pin Enable, soft-off | 41 |
| 03h | CLEAR_FAULTS | Clears fault indications. | SEND Byte | | | | 42 |
| 21h | VOUT_COMMAND | Sets the nominal value of the output voltage. | R/W Word | L16u | | Pin-strap | 42 |
| 24h | VOUT_MAX | Sets the maximum possible value of V_{OUT} . 110% of pin-strap V_{OUT} . | R/W Word | L16u | | 1.1 x V_{OUT} Pin-strap | 42 |
| 33h | FREQUENCY_SWITCH | Sets the switching frequency. | R/W Word | L11 | | Pin-strap | 42 |
| 37h | INTERLEAVE | Configures a phase offset between devices sharing a SYNC clock. | R/W Word | BIT | 1000h(Ch1)/0000h(Ch2) | 180° phase shift between Ch1/Ch2 | 43 |
| 40h | VOUT_OV_FAULT_LIMIT | Sets the V_{OUT} overvoltage fault threshold. | R/W Word | L16u | | 1.15 x V_{OUT} Pin-strap | 43 |
| 44h | VOUT_UV_FAULT_LIMIT | Sets the V_{OUT} undervoltage fault threshold. | R/W Word | L16u | | 0.85 x V_{OUT} Pin-strap | 43 |
| 46h | IOUT_OC_FAULT_LIMIT | Sets the I_{OUT} peak overcurrent fault threshold. | R/W Word | L11 | E320h | 50A | 44 |
| 4Bh | IOUT_UC_FAULT_LIMIT | Sets the I_{OUT} valley undercurrent fault threshold. | R/W Word | L11 | E4E0h | -50A | 44 |
| 4Fh | OT_FAULT_LIMIT | Sets the over-temperature fault threshold. | R/W Word | L11 | EB98h | +115°C | 44 |
| 53h | UT_FAULT_LIMIT | Sets the under-temperature fault threshold. | R/W Word | L11 | E530h | -45°C | 45 |
| 55h | VIN_OV_FAULT_LIMIT | Sets the V_{IN} overvoltage fault threshold. | R/W Word | L11 | D3A0 | 14.5V | 45 |
| 59h | VIN_UV_FAULT_LIMIT | Sets the V_{IN} undervoltage fault threshold. | R/W Word | L11 | | Pin-strap | 45 |
| 5Eh | POWER_GOOD_ON | Sets the voltage threshold for power-good indication. | R/W Word | L16u | | 0.9 x V_{OUT} Pin-strap | 45 |
| 60h | TON_DELAY | Sets the delay time from ENABLE to start of V_{OUT} rise. | R/W Word | L11 | | Pin-strap | 46 |

| Command Code | Command Name | Description | Type | Data Format | Default Value | Default Setting | Page |
|--------------|----------------------------|---|------------|-------------|---------------|-----------------------------------|--------------------|
| 61h | TON_RISE | Sets the rise time of V_{OUT} after ENABLE and TON_DELAY. | R/W Word | L11 | | Pin-strap | 46 |
| 64h | TOFF_DELAY | Sets the delay time from DISABLE to start of V_{OUT} fall. | R/W Word | L11 | | Pin-strap | 46 |
| 65h | TOFF_FALL | Sets the fall time for V_{OUT} after DISABLE and TOFF_DELAY. | R/W Word | L11 | | Pin-strap | 46 |
| 78h | STATUS_BYTE | Returns an abbreviated status for fast reads. | Read Byte | BIT | 00h | No Faults | 47 |
| 79h | STATUS_WORD | Returns information with a summary of the units' fault condition. | Read Word | BIT | 0000h | No Faults | 48 |
| 7Ah | STATUS_VOUT | Returns the V_{OUT} specific status. | Read Byte | BIT | 00h | No Faults | 48 |
| 7Bh | STATUS_IOUT | Returns the I_{OUT} specific status. | Read Byte | BIT | 00h | No Faults | 49 |
| 7Ch | STATUS_INPUT | Returns status specific to the input. | Read Byte | BIT | 00h | No Faults | 49 |
| 7Dh | STATUS_TEMP | Returns the temperature specific status. | Read Byte | BIT | 00h | No Faults | 49 |
| 7Eh | STATUS_CML | Returns the communication, logic and memory specific status. | Read Byte | BIT | 00h | No Faults | 50 |
| 80h | STATUS_MFR_SPECIFIC | Returns the external sync clock specific status. | Read Byte | BIT | 00h | No Faults | 50 |
| 88h | READ_VIN | Returns the input voltage reading. | Read Word | L11 | | | 50 |
| 8Bh | READ_VOUT | Returns the output voltage reading. | Read Word | L16u | | | 51 |
| 8Ch | READ_IOUT | Returns the output current reading. | Read Word | L11 | | | 51 |
| 8Dh | READ_INTERNAL_TEMP | Returns the temperature reading internal to the device. | Read Word | L11 | | | 51 |
| 94h | READ_DUTY_CYCLE | Returns the duty cycle reading during the ENABLE state. | Read Word | L11 | | | 51 |
| 95h | READ_FREQUENCY | Returns the measured operating switch frequency. | Read Word | L11 | | | 51 |
| DFh | ASCR_CONFIG | Configures ASCR control loop. | R/W Block | CUS | | Pin-strap | 52 |
| E4h | DEVICE_ID | Returns the 16-byte (character) device identifier string. | Read Block | ASC | | Reads Device Version | 52 |
| E5h | MFR_IOUT_OC_FAULT_RESPONSE | Configures the I_{OUT} overcurrent fault response. | R/W Byte | BIT | B9h | Disable and Retry with 70ms delay | 52 |
| E6h | MFR_IOUT_UC_FAULT_RESPONSE | Configures the I_{OUT} undercurrent fault response. | R/W Byte | BIT | B9h | Disable and Retry with 70ms delay | 53 |

| Command Code | Command Name | Description | Type | Data Format | Default Value | Default Setting | Page |
|--------------|-------------------------|---|------------|-------------|---------------|---------------------------|--------------------|
| E7h | IOUT_AVG_OC_FAULT_LIMIT | Sets the I _{OUT} average overcurrent fault threshold. | R/W | L11 | | Set by SYNC/OCP pin-strap | 53 |
| E8h | IOUT_AVG_UC_FAULT_LIMIT | Sets the I _{OUT} average undercurrent fault threshold. | R/W | L11 | DC40h | -30A | 54 |
| EAh | SNAPSHOT | Returns 32-byte read-back of parametric and status values. | Read Block | BIT | | | 55 |
| F3h | SNAPSHOT_CONTROL | Snapshot feature control command. | R/W Byte | BIT | | | 55 |

6.1 PMBus Data Formats

- **Linear-11 (L11)** - The L11 data format uses 5-bit two's complement exponent (N) and 11-bit two's complement mantissa (Y) to represent the real world decimal value (X). The relation between the real world value (X), N, and Y is: $X = Y \cdot 2^N$



- **Linear-16 Unsigned (L16u)** - The L16u data format uses a fixed exponent (hard-coded to $N = -13h$) and a 16-bit unsigned integer mantissa (Y) to represent real world decimal value (X). The relation between the real world decimal value (X), N and Y is: $X = Y \cdot 2^{-13}$
- **Linear-16 Signed (L16s)** - The L16s data format uses a fixed exponent (hard-coded to $N = -13h$) and a 16-bit two's complement mantissa (Y) to represent real world decimal value (X). The relation between the real world decimal value (X), N and Y is: $X = Y \cdot 2^{-13}$
- **Bit Field (BIT)** - An explanation of the Bit Field for each command is provided in [“PMBus Commands Description” on page 41](#).
- **Custom (CUS)** - An explanation of the Custom data format for each command is provided in [“PMBus Commands Description” on page 41](#). A combination of Bit Field and integer are common type of Custom data format.
- **ASCII (ASC)** - A variable length string of text characters uses ASCII data format.

6.2 PMBus Use Guidelines

The PMBus is a powerful tool that allows the user to optimize circuit performance by configuring devices for their application. When configuring a device in a circuit, the device should be disabled whenever most settings are changed with PMBus commands. Some exceptions to this recommendation are OPERATION, ON_OFF_CONFIG, CLEAR_FAULTS, VOUT_COMMAND, and ASCR_CONFIG. While the device is enabled any command can be read. Many commands do not take effect until after the device has been re-enabled, hence the recommendation that commands that change device settings are written while the device is disabled.

In addition, there should be a 2ms delay between repeated READ commands sent to the same device. When sending any other command, a 5ms delay is recommended between repeated commands sent to the same device.

7. PMBus Commands Description

PAGE(00h)

Definition: Select Channel 1, Channel 2 or both channels to receive commands. All commands following this command are received and acted on by the selected channel or channels.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: 00h (Page 0)

Units: N/A

| Bits 7:4 | Bits 3:0 | Page |
|----------|----------|------|
| 0000 | 0000 | 0 |
| 0000 | 0001 | 1 |
| 1111 | 1111 | Both |

OPERATION (01h)

Definition: Sets Enable and Disable settings.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value:

Units: N/A

| Settings | Actions |
|----------|---------------|
| 00h | Immediate off |
| 40h | Soft off |
| 80h | On |

ON_OFF_CONFIG (02h)

Definition: Configures the interpretation and coordination of the OPERATION command and the ENABLE pin (EN).

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: 16h (Device starts from ENABLE pin with soft-off)

Units: N/A

| Settings | Actions |
|----------|--|
| 16h | Device starts from the ENABLE pin with soft off. |
| 17h | Device starts from the ENABLE pin with immediate off. |
| 1Ah | Device starts from the OPERATION command with soft off. |
| 1Bh | Device starts from the OPERATION command with immediate off. |

CLEAR_FAULTS (03h)

Definition: Clears all fault bits in all registers. If a fault condition still exists, the bit reasserts immediately. This command does not restart a device if it has shut down, it only clears the faults.

Data Length in Bytes: 0 Byte

Data Format: N/A

Type: Send byte

Default Value: N/A

Units: N/A

Reference: N/A

VOUT_COMMAND (21h)

Definition: Sets or reports the target output voltage. This command cannot set a value higher than VOUT_MAX.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: Pin-strap setting

Units: Volts

Range: 0V to VOUT_MAX

VOUT_MAX (24h)

Definition: Sets an upper limit on the output voltage the unit can command regardless of any other commands or combinations. The intent of this command is to provide a safeguard against a user accidentally setting the output voltage to a possibly destructive level rather than to be the primary output overprotection. Default value can be changed using PMBus.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: 1.10 x V_{OUT} pin-strap setting

Units: Volts

Range: 0V to 5.5V

FREQUENCY_SWITCH (33h)

Definition: Sets the switching frequency of the device. Initial default value is defined by a pin-strap and this value can be overridden by writing this command through the PMBus. If using an external SYNC, set this value as close as possible to the external clock value. The output must be disabled when writing this command.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin-strap setting

Units: kHz

Range: 296kHz to 1067kHz

INTERLEAVE (37h)

Definition: Configures the phase offset of a device that is sharing a common SYNC clock with other devices. The phase offset of each device can be set to any value between 0° and 360° in 22.5° increments. The internal two phases of the module always maintain a phase difference of 180°.

Data Length in Bytes: 2

Data Format: BIT

Type: R/W

Default Value: 1000h (Page0), 0000h (Page1)

Units: N/A

| Bits | Purpose | Value | Description |
|------|-------------------|---------|---|
| 15:8 | Reserved | 0 | Reserved |
| 7:4 | Group Number | 0 to 15 | Sets the group number. A value of 0 is interpreted as 16. |
| 3:0 | Position in Group | 0 to 15 | Sets position of the device's rail within the group. |

VOUT_OV_FAULT_LIMIT (40h)

Definition: Sets the V_{OUT} overvoltage fault threshold.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: $1.15 \times V_{OUT}$ pin-strap setting

Units: V

Range: 0V to V_{OUT_MAX}

VOUT_UV_FAULT_LIMIT (44h)

Definition: Sets the V_{OUT} undervoltage fault threshold. This fault is masked during ramp or when disabled.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: $0.85 \times V_{OUT}$ pin-strap setting

Units: V

Range: 0V to V_{OUT_MAX}

IOUT_OC_FAULT_LIMIT (46h)

Definition: Sets the I_{OUT} peak overcurrent fault threshold. This limit is applied to current measurement samples taken after the Current Sense Blanking Time has expired. A fault occurs after this limit is exceeded. The recommended peak OCP limit is determined by [Equation 1](#).

$$(EQ. 1) \quad IOUT_OC_FAULT_LIMIT = (IOUT_AVG_OC_FAULT_LIMIT + 0.5 \cdot IRIPPLE_{P_P}) \cdot 130\%$$

A hard bound of 50A is applied to the command value. This feature shares the OC fault bit operation (in STATUS_IOUT) and OC fault response with the IOUT_AVG_OC_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: E320h (50A)

Units: A

Range: -100A to 100A

IOUT_UC_FAULT_LIMIT (4Bh)

Definition: Sets the I_{OUT} valley undercurrent fault threshold. This limit is applied to current measurement samples taken after the Current Sense Blanking Time has expired. A fault occurs after this limit is exceeded. The recommended valley UCP limit is determined by [Equation 2](#):

$$(EQ. 2) \quad IOUT_UC_FAULT_LIMIT = (IOUT_AVG_UC_FAULT_LIMIT - 0.5 \cdot IRIPPLE_{P_P}) \cdot 130\%$$

A hard bound of -50A is applied to the command value. This feature shares the UC fault bit operation (in STATUS_IOUT) and UC fault response with IOUT_AVG_UC_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: E4E0 (-50A)

Units: A

Range: -100A to 100A

OT_FAULT_LIMIT (4Fh)

Definition: Sets the temperature at which the device should indicate an over-temperature fault. Note that the temperature must drop below the fault threshold to clear this fault.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: EB98 (+115°C)

Units: °C

Range: 0°C to +175°C

UT_FAULT_LIMIT (53h)

Definition: Sets the temperature, in degrees Celsius, of the unit where it should indicate an under-temperature fault. Note that the temperature must rise above the fault threshold to clear this fault.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: E530h (-45°C)

Units: °C

Range: -55°C to +25°C

VIN_OV_FAULT_LIMIT (55h)

Definition: Sets the V_{IN} overvoltage fault threshold.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: D3A0 (14.5V)

Units: V

Range: 0V to 16V

VIN_UV_FAULT_LIMIT (59h)

Definition: Sets the V_{IN} undervoltage fault threshold.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin-strap setting

Units: V

Range: 0V to 12V

POWER_GOOD_ON (5Eh)

Definition: Sets the voltage threshold for power-good indication. Power-good asserts after the output voltage exceeds POWER_GOOD_ON and deasserts when the output voltage is less than VOUT_UV_FAULT_LIMIT. It is recommended to set POWER_GOOD_ON higher than VOUT_UV_FAULT_LIMIT.

Data Length in Bytes: 2

Data Format: L16u

Type: R/W

Default Value: 0.9 x V_{OUT} pin-strap setting

Units: V

TON_DELAY (60h)

Definition: Sets the delay time from when the device is enabled to the start of V_{OUT} rise.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin-strap setting

Units: ms

Range: 2ms to 300ms

TON_RISE (61h)

Definition: Sets the rise time of V_{OUT} after ENABLE and TON_DELAY.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin-strap setting

Units: ms

Range: 1ms to 120ms

TOFF_DELAY (64h)

Definition: Sets the delay time from DISABLE to start of V_{OUT} fall.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin-strap setting

Units: ms

Range: 2ms to 300ms

TOFF_FALL (65h)

Definition: Sets the soft-off fall time for V_{OUT} after DISABLE and TOFF_DELAY.

Data Length in Bytes: 2

Data Format: L11

Type: R/W

Default Value: Pin-strap setting

Units: ms

Range: 0ms to 120ms

STATUS_BYTE (78h)

Definition: Returns one byte of information with a summary of the most critical faults.

Data Length in Bytes: 1

Data Format: BIT

Type: Read Only

Default Value: 00h

Units: N/A

| Bit Number | Status Bit Name | Meaning |
|------------|-------------------|--|
| 7 | BUSY | A fault was declared because the device was busy and unable to respond. |
| 6 | OFF | This bit is asserted if the unit is not providing power to the output, regardless of the reason, including simply not being enabled. |
| 5 | VOUT_OV_FAULT | An output overvoltage fault has occurred. |
| 4 | IOUT_OC_FAULT | An output overcurrent fault has occurred. |
| 3 | VIN_UV_FAULT | An input undervoltage fault has occurred. |
| 2 | TEMPERATURE | A temperature fault has occurred. |
| 1 | CML | A communications, memory, or logic fault has occurred. |
| 0 | NONE OF THE ABOVE | A fault not listed in Bits 7:1 has occurred. |

STATUS_WORD (79h)

Definition: Returns two bytes of information with a summary of the unit's fault condition. Based on the information in these bytes, the host can get more information by reading the appropriate status registers. The low byte of the STATUS_WORD is the same register as the STATUS_BYTE (78h) command.

Data Length in Bytes: 2

Data Format: BIT

Type: Read Only

Default Value: 0000h

Units: N/A

| Bit Number | Status Bit Name | Meaning |
|------------|-------------------|--|
| 15 | VOUT | An output voltage fault has occurred. |
| 14 | IOUT/POUT | An output current or output power fault has occurred. |
| 13 | INPUT | An input voltage, input current, or input power fault has occurred. |
| 12 | MFG_SPECIFIC | A manufacturer specific fault has occurred. |
| 11 | POWER_GOOD# | The POWER_GOOD signal, if present, is negated. |
| 10 | Reserved | Reserved |
| 9 | OTHER | A bit in STATUS_OTHER is set. |
| 8 | UNKNOWN | A fault type not given in Bits 15:1 of the STATUS_WORD has been detected. |
| 7 | BUSY | A fault was declared because the device was busy and unable to respond. |
| 6 | OFF | This bit is asserted if the unit is not providing power to the output, regardless of the reason, including simply not being enabled. |
| 5 | VOUT_OV_FAULT | An output overvoltage fault has occurred. |
| 4 | IOUT_OC_FAULT | An output overcurrent fault has occurred. |
| 3 | VIN_UV_FAULT | An input undervoltage fault has occurred. |
| 2 | TEMPERATURE | A temperature fault has occurred. |
| 1 | CML | A communications, memory, or logic fault has occurred. |
| 0 | NONE OF THE ABOVE | A fault not listed in Bits 7:1 has occurred. |

STATUS_VOUT (7Ah)

Definition: Returns one data byte with the status of the output voltage.

Data Length in Bytes: 1

Data Format: BIT

Type: Read Only

Default Value: 00h

Units: N/A

| Bit Number | Status Bit Name | Meaning |
|------------|-----------------|---|
| 7 | VOUT_OV_FAULT | Indicates an output overvoltage fault. |
| 6:5 | Reserved | Reserved |
| 4 | VOUT_UV_FAULT | Indicates an output undervoltage fault. |
| 3:0 | N/A | These bits are not used. |

STATUS_IOUT (7Bh)**Definition:** Returns one data byte with the status of the output current.**Data Length in Bytes:** 1**Data Format:** BIT**Type:** Read Only**Default Value:** 00h**Units:** N/A

| Bit Number | Status Bit Name | Meaning |
|------------|-----------------|--|
| 7 | IOUT_OC_FAULT | An output overcurrent fault has occurred. |
| 6:5 | Reserved | Reserved |
| 4 | IOUT_UC_FAULT | An output undercurrent fault has occurred. |
| 3:0 | N/A | These bits are not used. |

STATUS_INPUT (7Ch)**Definition:** Returns the input voltage and input current status information.**Data Length in Bytes:** 1**Data Format:** BIT**Type:** Read Only**Default Value:** 00h**Units:** N/A

| Bit Number | Status Bit Name | Meaning |
|------------|-----------------|---|
| 7 | VIN_OV_FAULT | An input overvoltage fault has occurred. |
| 6:5 | Reserved | Reserved |
| 4 | VIN_UV_FAULT | An input undervoltage fault has occurred. |
| 3:0 | N/A | These bits are not used. |

STATUS_TEMP (7Dh)**Definition:** Returns one byte of information with a summary of any temperature related faults.**Data Length in Bytes:** 1**Data Format:** BIT**Type:** Read Only**Default Value:** 00h**Units:** N/A

| Bit Number | Status Bit Name | Meaning |
|------------|-----------------|--|
| 7 | OT_FAULT | An over-temperature fault has occurred. |
| 6:5 | Reserved | Reserved |
| 4 | UT_FAULT | An under-temperature fault has occurred. |
| 3:0 | N/A | These bits are not used. |

STATUS_CML (7Eh)

Definition: Returns one byte of information with a summary of any communications, logic, and/or memory errors.

Data Length in Bytes: 1

Data Format: BIT

Type: Read Only

Default Value: 00h

Units: N/A

| Bit Number | Meaning |
|------------|---|
| 7 | Invalid or unsupported PMBus command was received. |
| 6 | The PMBus command was sent with invalid or unsupported data. |
| 5 | Packet error was detected in the PMBus command. |
| 4 | Memory/logic fault has occurred. |
| 3:2 | Reserved |
| 1 | A PMBus command tried to write to a read-only or protected command, or a communication fault other than the ones listed in this table has occurred. |
| 0 | Not used |

STATUS_MFR_SPECIFIC (80h)

Definition: Returns one byte of information providing the status of the clock synchronization faults.

Data Length in Bytes: 1

Data Format: BIT

Type: Read only

Default value: 00h

Units: N/A

| Bit Number | Field Name | Meaning |
|------------|---------------------------------|--|
| 7:4 | Reserved | Reserved |
| 3 | External Switching Period Fault | Loss of external clock synchronization has occurred. |
| 2:0 | Reserved | Reserved |

READ_VIN (88h)

Definition: Returns the input voltage reading.

Data Length in Bytes: 2

Data Format: L11

Type: Read Only

Units: V

READ_VOUT (8Bh)

Definition: Returns the output voltage reading.

Data Length in Bytes: 2

Data Format: L16u

Type: Read Only

Units: V

READ_IOUT (8Ch)

Definition: Returns the output current reading.

Data Length in Bytes: 2

Data Format: L11

Type: Read Only

Default Value: N/A

Units: A

READ_INTERNAL_TEMP (8Dh)

Definition: Returns the controller junction temperature reading from the internal temperature sensor.

Data Length in Bytes: 2

Data Format: L11

Type: Read Only

Units: °C

READ_DUTY_CYCLE (94h)

Definition: Reports the actual duty cycle of the converter during the enable state.

Data Length in Bytes: 2

Data Format: L11

Type: Read only

Units: %

READ_FREQUENCY (95h)

Definition: Reports the actual switching frequency of the converter during the enable state.

Data Length in Bytes: 2

Data Format: L11

Type: Read only

Units: kHz

ASCR_CONFIG (DFh)

Definition: Allows user configuration of ASCR settings. ASCR gain is analogous to bandwidth, ASCR residual is analogous to damping. To improve load transient response performance, increase ASCR gain. To lower transient response overshoot, increase ASCR residual. Increasing ASCR gain can result in increased PWM jitter and should be evaluated in the application circuit. Excessive ASCR gain can lead to excessive output voltage ripple. Increasing ASCR residual to improve transient response damping can result in slower recovery times, but does not affect the peak output voltage deviation. Typical ASCR gain settings range from 50 to 1000, and ASCR residual settings range from 10 to 100.

Data Length in Bytes: 4

Data Format: CUS

Type: R/W

Default Value: Pin-strap setting

| Bit | Purpose | Data Format | Value | Description |
|-------|-----------------------|-------------|----------|---------------|
| 31:25 | Unused | | 0000000h | Unused |
| 24 | Reserved | | | Reserved |
| 23:16 | ASCR Residual Setting | Integer | | ASCR residual |
| 15:0 | ASCR Gain Setting | Integer | | ASCR gain |

DEVICE_ID (E4h)

Definition: Returns the 16-byte (character) device identifier string.

Data Length in Bytes: 16

Data Format: ASC

Type: Block Read

Default Value: Part number/Die revision/Firmware revision

MFR_IOUT_OC_FAULT_RESPONSE (E5h)

Definition: Configures the I_{OUT} overcurrent fault response as defined by the following table. The command format is the same as the PMBus standard fault responses except that it sets the overcurrent status bit in STATUS_IOUT.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: B9h (Disable and continuous retry with 70ms delay)

Units: N/A

| Field Name | Actions |
|------------|--|
| 80h | Disable with no retry. |
| B9h | Disable and continuous retry with 70ms delay |

MFR_IOUT_UC_FAULT_RESPONSE (E6h)

Definition: Configures the I_{OUT} undercurrent fault response as defined by the following table. The command format is the same as the PMBus standard fault responses except that it sets the undercurrent status bit in STATUS_IOUT.

Data Length in Bytes: 1

Data Format: BIT

Type: R/W

Default Value: B9h (Disable and continuous retry with 70ms delay)

Units: N/A

| Field Name | Actions |
|------------|--|
| 80h | Disable with no retry. |
| B9h | Disable and continuous retry with 70ms delay |

IOUT_AVG_OC_FAULT_LIMIT (E7h)

Definition: Sets the I_{OUT} average overcurrent fault threshold. For down-slope sensing, this corresponds to the average of all the current samples taken during the (1-D) time interval, excluding the current sense blanking time (which occurs at the beginning of the 1-D interval). For up-slope sensing, this corresponds to the average of all the current samples taken during the D time interval, excluding the current sense blanking time (which occurs at the beginning of the D interval). This feature shares the OC fault bit operation (in STATUS_IOUT) and OC fault response with IOUT_OC_FAULT_LIMIT.

Paged or Global: Paged

Data Length in Bytes: 2

Data Format: Linear-11

Type: R/W

Protectable: Yes

Default Value: SYNC/OCP pin-strap setting

Units: A

Range: -100A to 100A

IOUT_AVG_UC_FAULT_LIMIT (E8h)

Definition: Sets the I_{OUT} average undercurrent fault threshold. For down-slope sensing, this corresponds to the average of all the current samples taken during the (1-D) time interval, excluding the current sense blanking time (which occurs at the beginning of the 1-D interval). For up-slope sensing, this corresponds to the average of all the current samples taken during the D time interval, excluding the current sense blanking time (which occurs at the beginning of the D interval). This feature shares the UC fault bit operation (in STATUS_IOUT) and UC fault response with IOUT_UC_FAULT_LIMIT.

Paged or Global: Paged

Data Length in Bytes: 2

Data Format: Linear-11

Type: R/W

Protectable: Yes

Default Value: DC40h (-30A)

Units: A

Range: -100A to 100A

SNAPSHOT (EAh)

Definition: A 32-byte read-back of parametric and status values. It allows monitoring and status data to be stored to flash following a fault condition. In case of a fault, last updated values are stored to the flash memory. When the SNAPSHOT STATUS bit is set stored, the device no longer automatically captures parametric and status values following fault until stored data are erased. Use the SNAPSHOT_CONTROL command to erase store data and clear the status bit before next ramp up. Data erased is not allowed when module is enabled.

Data Length in Bytes: 32

Data Format: Bit field

Type: Block Read

| Byte Number | Value | PMBus Command | Format |
|-------------|--|---------------------------|--------|
| 31:23 | Reserved | Reserved | 00h |
| 22 | Flash Memory Status Byte FF - Not Stored 00 - Stored | NVM Status Byte | BIT |
| 21 | Manufacturer Specific Status Byte | STATUS_MFR_SPECIFIC (80h) | Byte |
| 20 | CML Status Byte | STATUS_CML (7Eh) | Byte |
| 19 | Temperature Status Byte | STATUS_TEMPERATURE (7Dh) | Byte |
| 18 | Input Status Byte | STATUS_INPUT (7Ch) | Byte |
| 17 | I _{OUT} Status Byte | STATUS_IOUT (7Bh) | Byte |
| 16 | V _{OUT} Status Byte | STATUS_VOUT (7Ah) | Byte |
| 15:14 | Switching Frequency | READ_FREQUENCY (95h) | L11 |
| 13:12 | Reserved | Reserved | 00h |
| 11:10 | Internal Temperature | READ_INTERNAL_TEMP (8Dh) | L11 |
| 9:8 | Duty Cycle | READ_DUTY_CYCLE (94h) | L11 |
| 7:6 | Reserved | Reserved | L11 |
| 5:4 | Output Current | READ_IOUT (8Ch) | L11 |
| 3:2 | Output Voltage | READ_VOUT (8Bh) | L16u |
| 1:0 | Input Voltage | READ_VIN (88h) | L11 |

SNAPSHOT_CONTROL (F3h)

Definition: Writing a 01h causes the device to copy the current Snapshot values from NVRAM to the 32-byte Snapshot command parameter. Writing a 02h causes the device to write the current Snapshot values to NVRAM. Writing a 03h erases all Snapshot values from NVRAM. Write (02h) and Erase (03h) can only be used when the device is disabled. All other values are ignored.

Data Length in Bytes: 1

Data Format: Bit field

Type: R/W byte

| Value | Description |
|-------|---|
| 01h | Read Snapshot values from NV RAM |
| 02h | Write Snapshot values to NV RAM |
| 03h | Erase Snapshot values stored in NV RAM. |

8. Revision History

8.1 Firmware

Table 12. RAA210925 Nomenclature Guide

| Firmware Revision Code | Change Description | Note |
|------------------------|--------------------|-----------------------------|
| RAA210925-0-G0100 | Initial Release | Recommended for new designs |

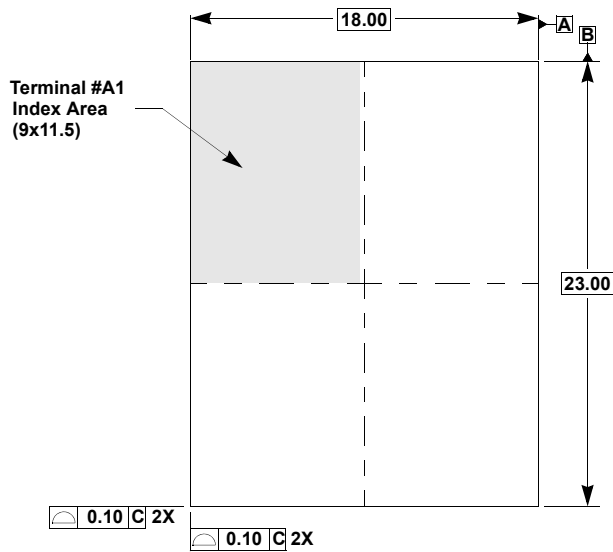
8.2 Datasheet

| Rev. | Date | Description |
|------|--------------|--|
| 1.00 | May 31, 2019 | Pin Description table on page 11 and 12 - updated Description for D5 and G14. Page 22, SMBus Communications 1st sentence: Changed PMBus to SMBus page 41 OPERATION (01h) table, 80h removed Nominal from Actions column. |
| 0.00 | Sep 12, 2018 | Initial release |

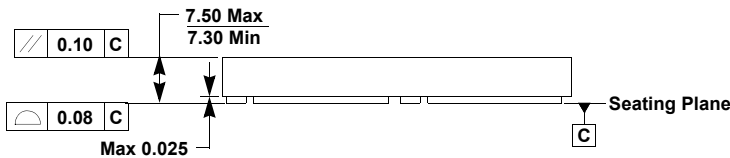
9. Package Outline Drawing

Y58.18x23
 58 I/O 18mmx23mmx7.5mm Custom HDA Module
 Rev 4, 4/18

For the most recent package outline



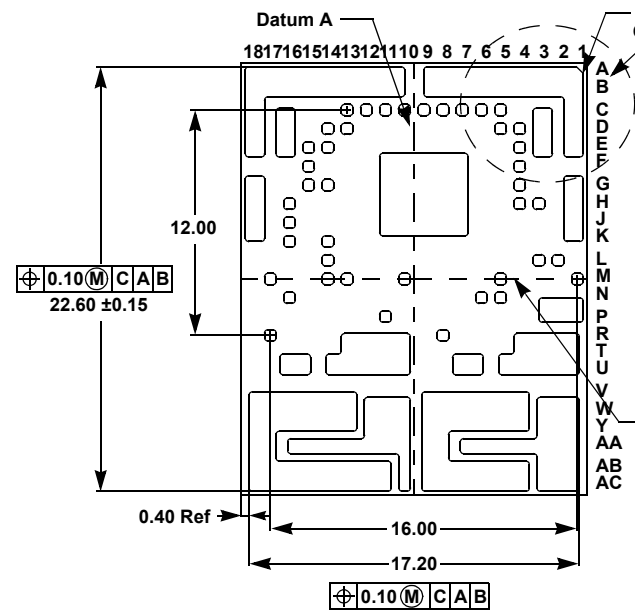
Top View



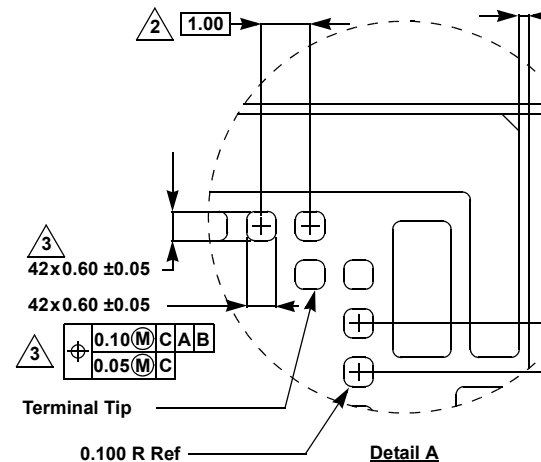
Side View

Notes:

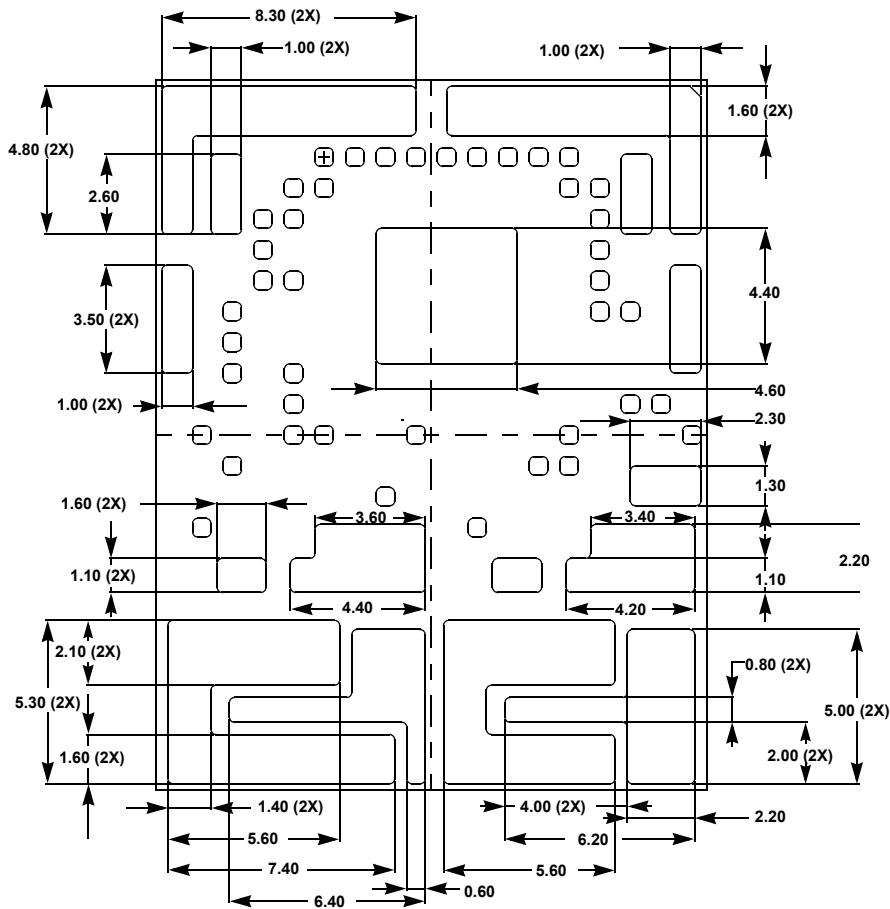
1. All dimensions are in millimeters.
2. Represents the basic land grid pitch.
3. These 42 I/Os are centered in a fixed row and column matrix at 1.0mm pitch BSC.
4. Dimensioning and tolerancing per ASME Y14.5-2009.
5. Tolerance for exposed PAD edge location dimension on page 3 is ± 0.1 mm.



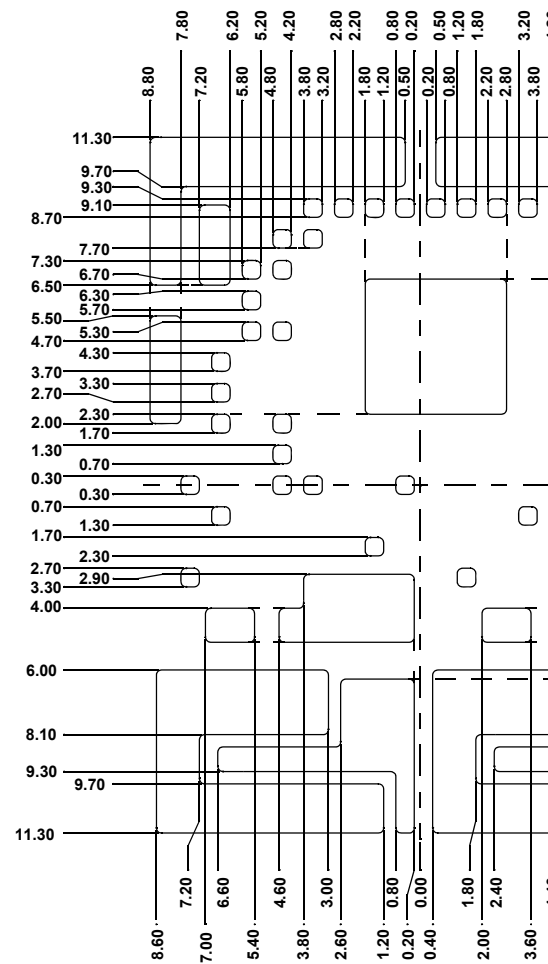
Bottom View



Detail A

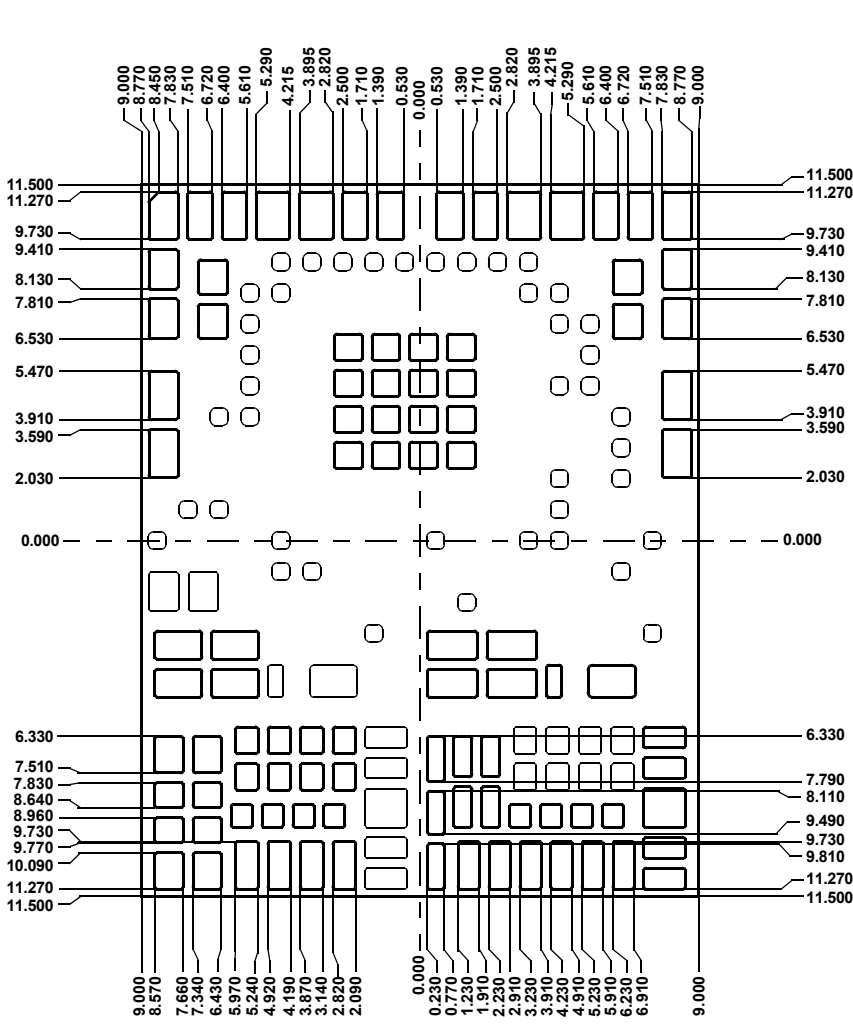


Size Details for the 16 Exposed Pads

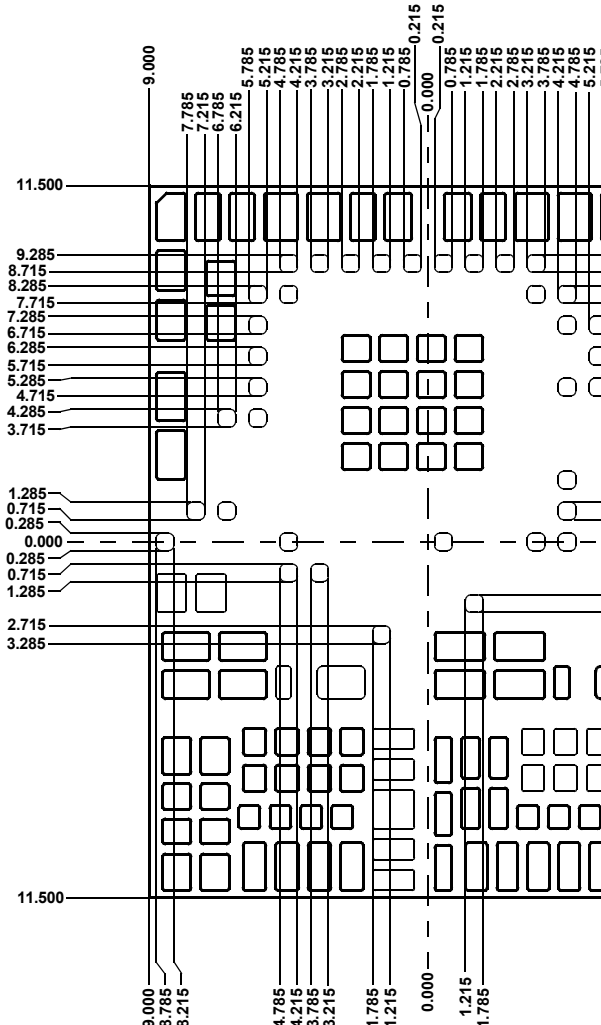


Terminal and Pad Edge Details

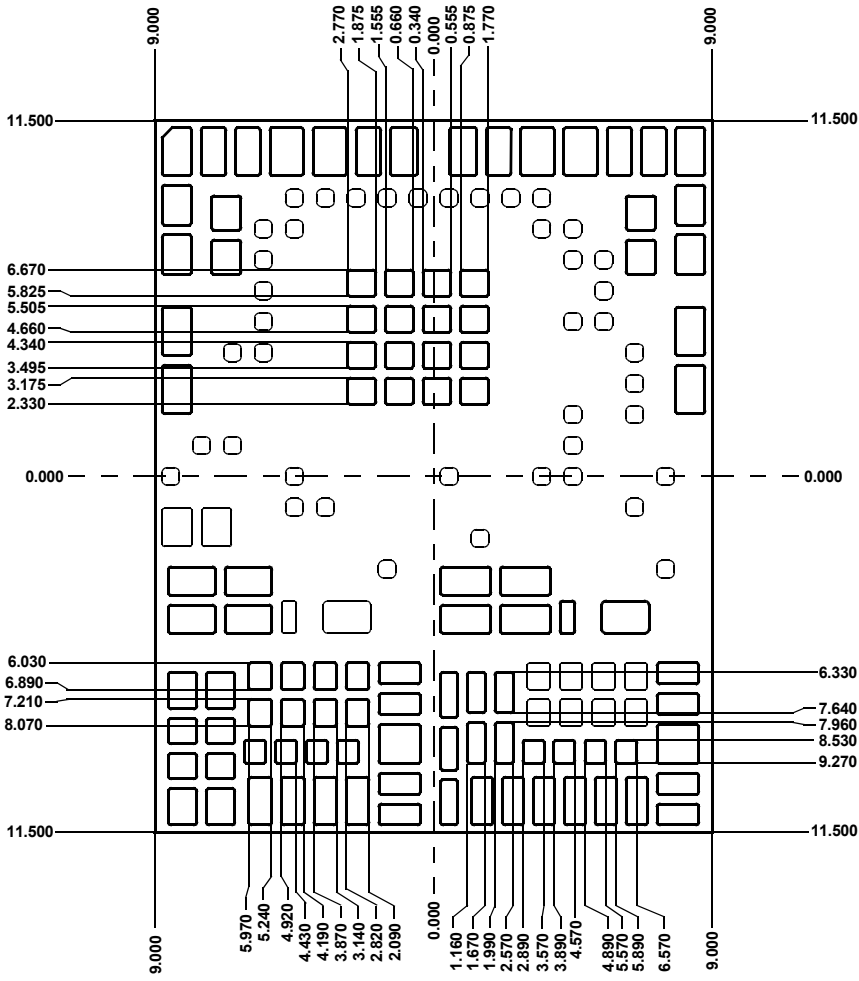
Bottom View



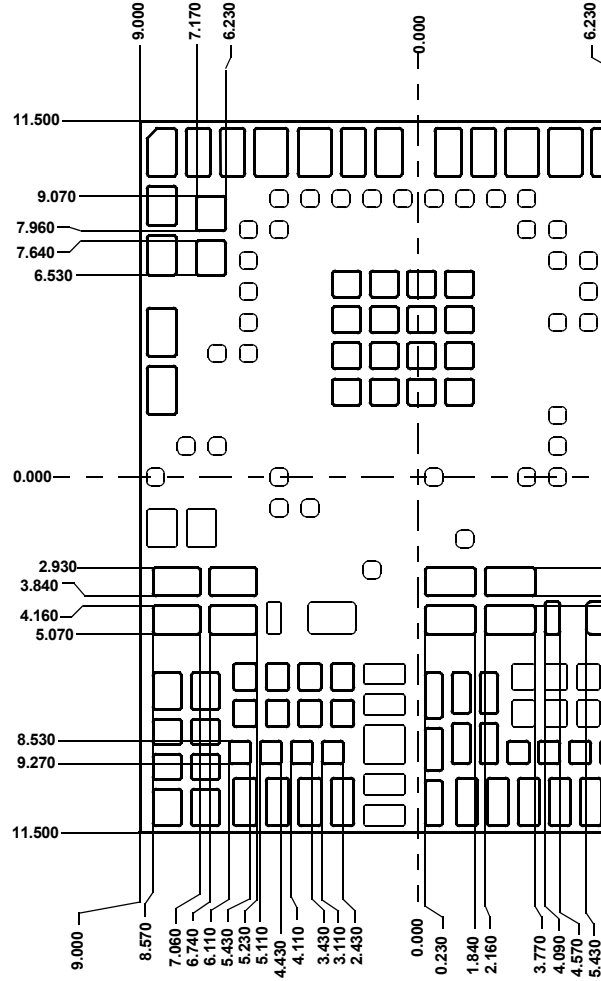
Stencil Opening Edge Position - 1



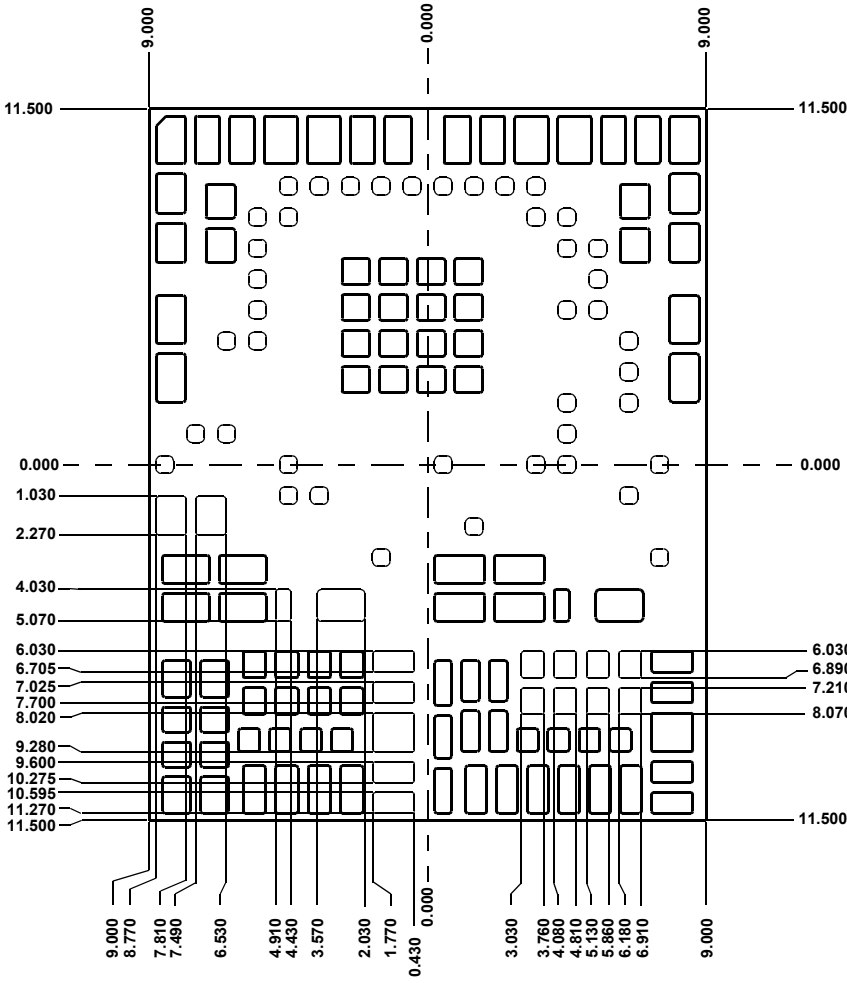
Stencil Opening Edge Position - 2



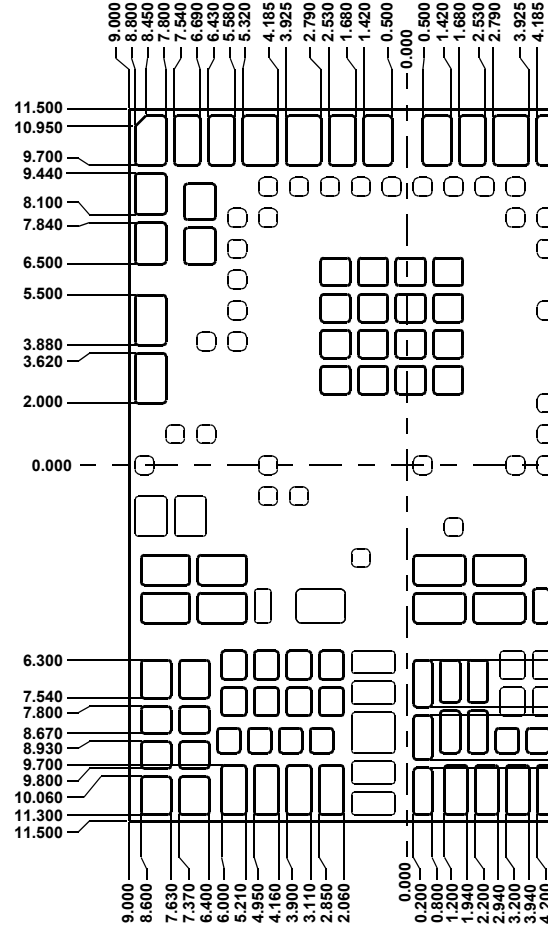
Stencil Opening Edge Position - 3



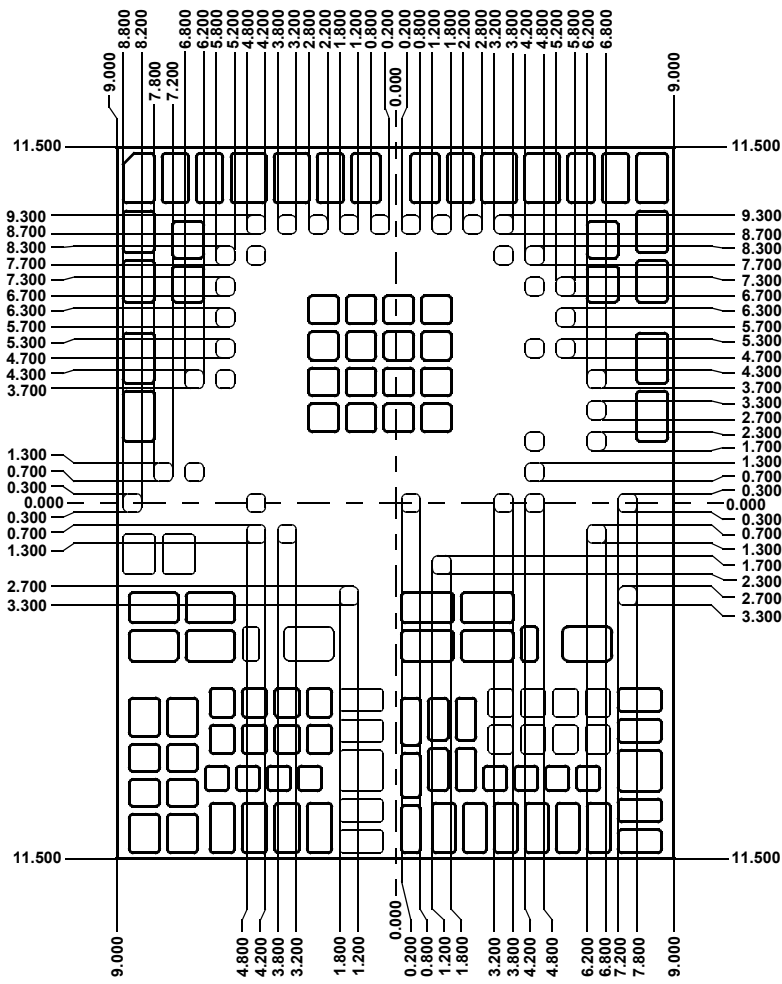
Stencil Opening Edge Position - 4



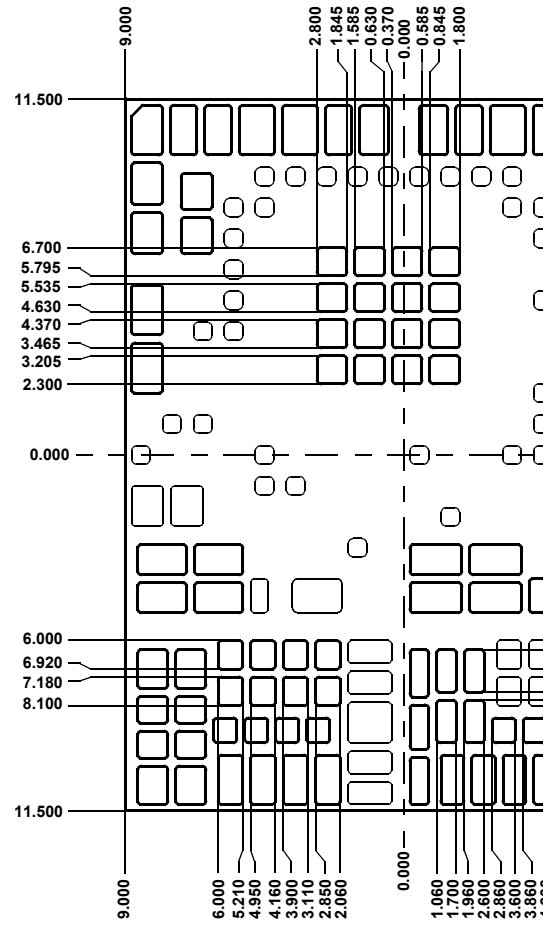
Stencil Opening Edge Position - 5



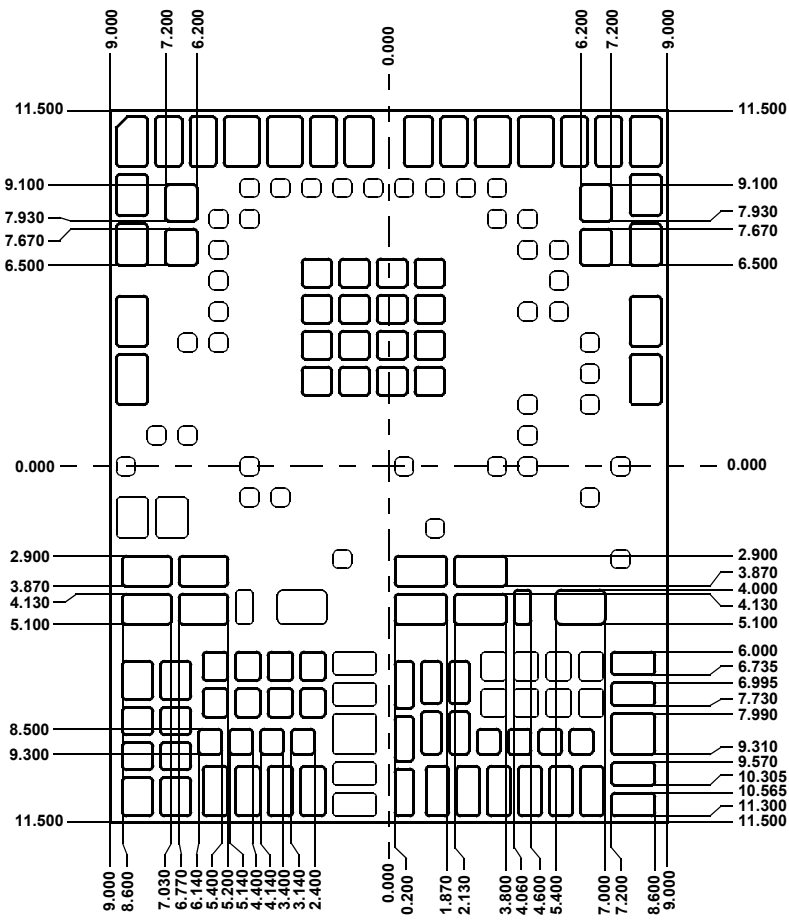
PCB Land Pattern - 1 (for Reference)



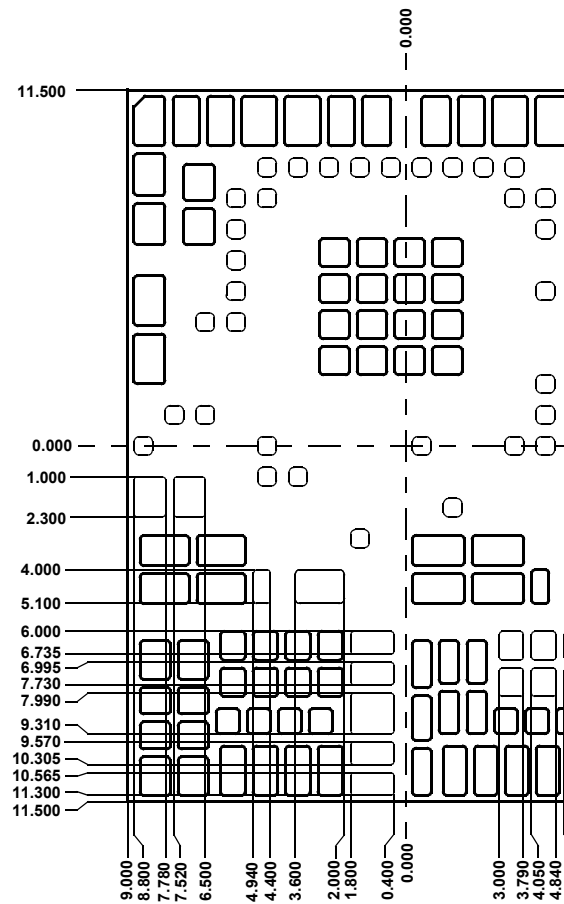
PCB Land Pattern - 2 (for Reference)



PCB Land Pattern - 3 (for Reference)



PCB Land Pattern - 4 (for Reference)



PCB Land Pattern - 5 (for Reference)

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(Rev.4.0-1 November 2017)

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