RAA306012

3-Phase Smart Gate Driver

Rev.1.00
Oct 5, 2023

1 Description

The RAA306012 is a smart gate driver IC for 3-phase brushless DC (BLDC) motor applications. It integrates three half-bridge gate drivers that are capable of driving up to three N-channel MOSFET bridges and supports bridge voltages from 6V to 65V. Each gate driver supports up to 0.64A source and 1.28A sink peak drive current with adjustable drive strength control. Adaptive and adjustable dead times are implemented to ensure robustness and flexibility. The active gate holding mechanism prevents miller effect induced cross-conduction and further enhances robustness.

The device integrates power supplies that power internal analog and logic circuitry, high-side/low-side gate drivers, and a dedicated supply for powering external microcontrollers. The device also features a low-power sleep mode that consumes only 28µA to maximize battery life in portable applications.

The driver control inputs can be configured to either 3-phase HI/LI or 3-phase PWM modes. Three accurate differential amplifiers with adjustable gain are integrated to support ground-side shunt current sensing for each bridge. The device can also support both BLDC sensor/sensorless motor drive by the integrated comparators or BEMF sense amplifier.

The device can be configured to use the SPI interface. All the parameters can be set through the SPI interface, and allows better monitoring.

Extensive protection functions include supply voltage OV/UV protection, buck regulator OV/UV/OC protection, charge pump UV protection, MOSFET V_{DS} OC protection, current sense OC protection, MOSFET V_{GS} fault, thermal warning, and thermal shutdown. Fault conditions are reported on a dedicated nFAULT pin and each status bit in the Fault Status registers.

2 Features

- Operating power supply voltage:
  - VBRIDGE: 6V to 65V (78V abs max)
  - VM: 6V to 60V (65V abs max)
- Operating ambient temperature: -40°C to 125°C
- 3-Phase gate drivers for BLDC application
  - Switching frequency range up to 200kHz
  - Peak 0.64A/1.28A source/sink current with 16 adjustable drive strength through SPI interface
  - Adaptive and adjustable dead time
- Flexible configuration for gate driver
  - 3-phase HI/LI mode and 3-phase PWM mode
  - Input control signal configuration
  - Support half-bridge, full-bridge configuration
- Fully integrated power supply architecture
  - Two VCC LDOs allow for Sleep mode low I_Q
  - 500mA buck switching regulator generates drive voltage (5V to 15V adjustable)
  - 100mA adjustable output LDO for MCU supplies
- Three accurate differential amplifiers
  - Four levels of sense gain setting
  - Supports DC offset calibration during power-up and on-the-fly
- BEMF sense amplifier for sensorless motor drive
- Three comparators for hall sensor motor drive
- Integrated protection features
  - VM over/undervoltage lockout (VM_OV/UVLO)
  - Charge pump undervoltage (VCP_UV)
  - Buck regulator fault (VDRV_OV/UV, SR_OCP)
  - MOSFET V_{DS} OCP (VDS_OCP)
  - Current sense OCP (CS_OCP)
  - MOSFET V_{GS} fault (VGS_FAULT)
  - Thermal warning/shutdown (TWARN/OTSD)
  - Fault indicator (nFAULT pin)
- 7mm×7mm 48 Ld QFN package (0.5mm pitch)

3 Applications

- Power tools and Garden tools
- Printers, Vacuum cleaners, Fans, Pumps, and Robotics

4 Overview

4.1 Typical Application Circuits

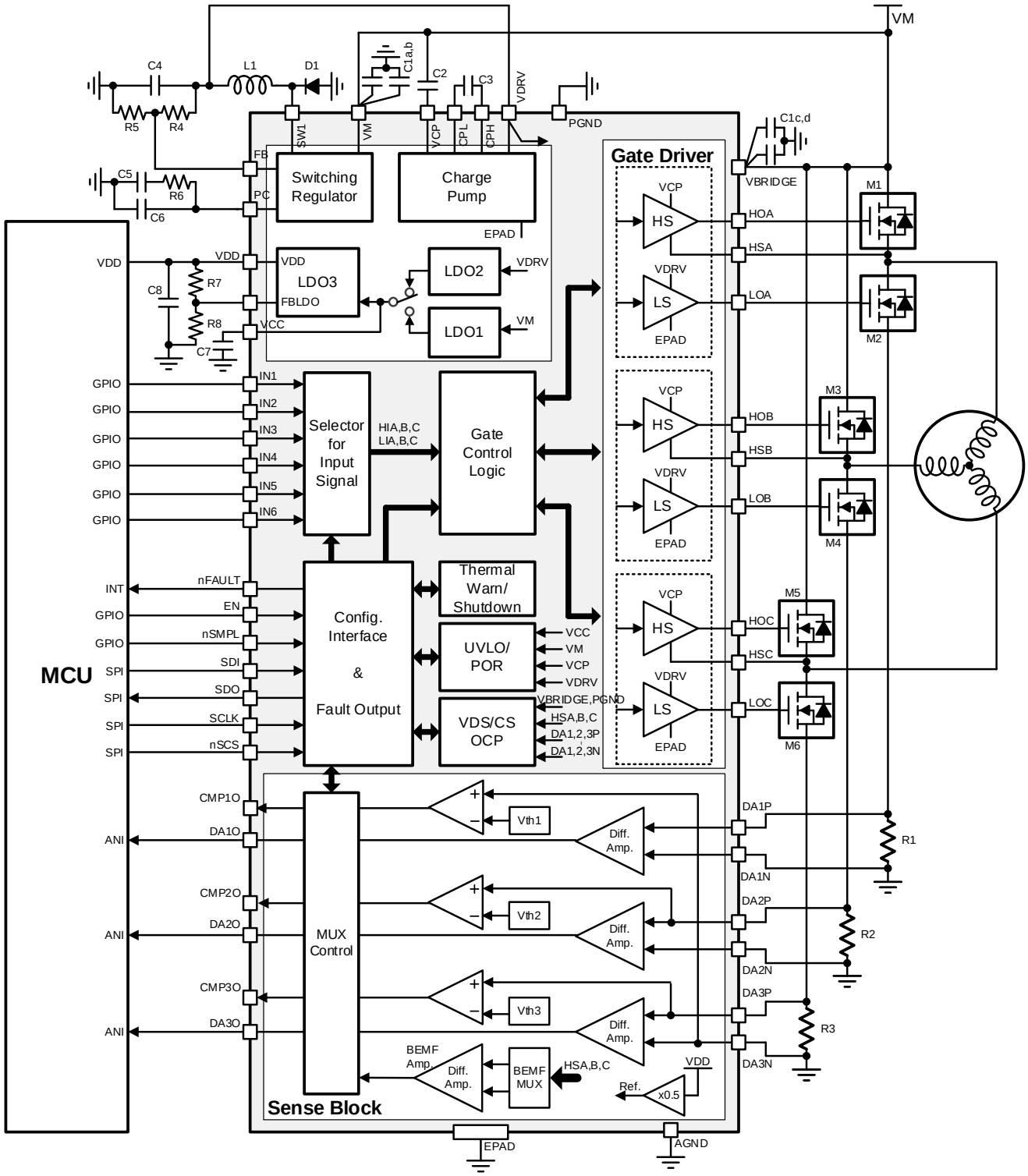


Figure 4.1-1 Simplified Block Diagram and Application – 3 Shunt Sensorless FOC Motor Drive

4.1 Typical Application Circuits (continued)

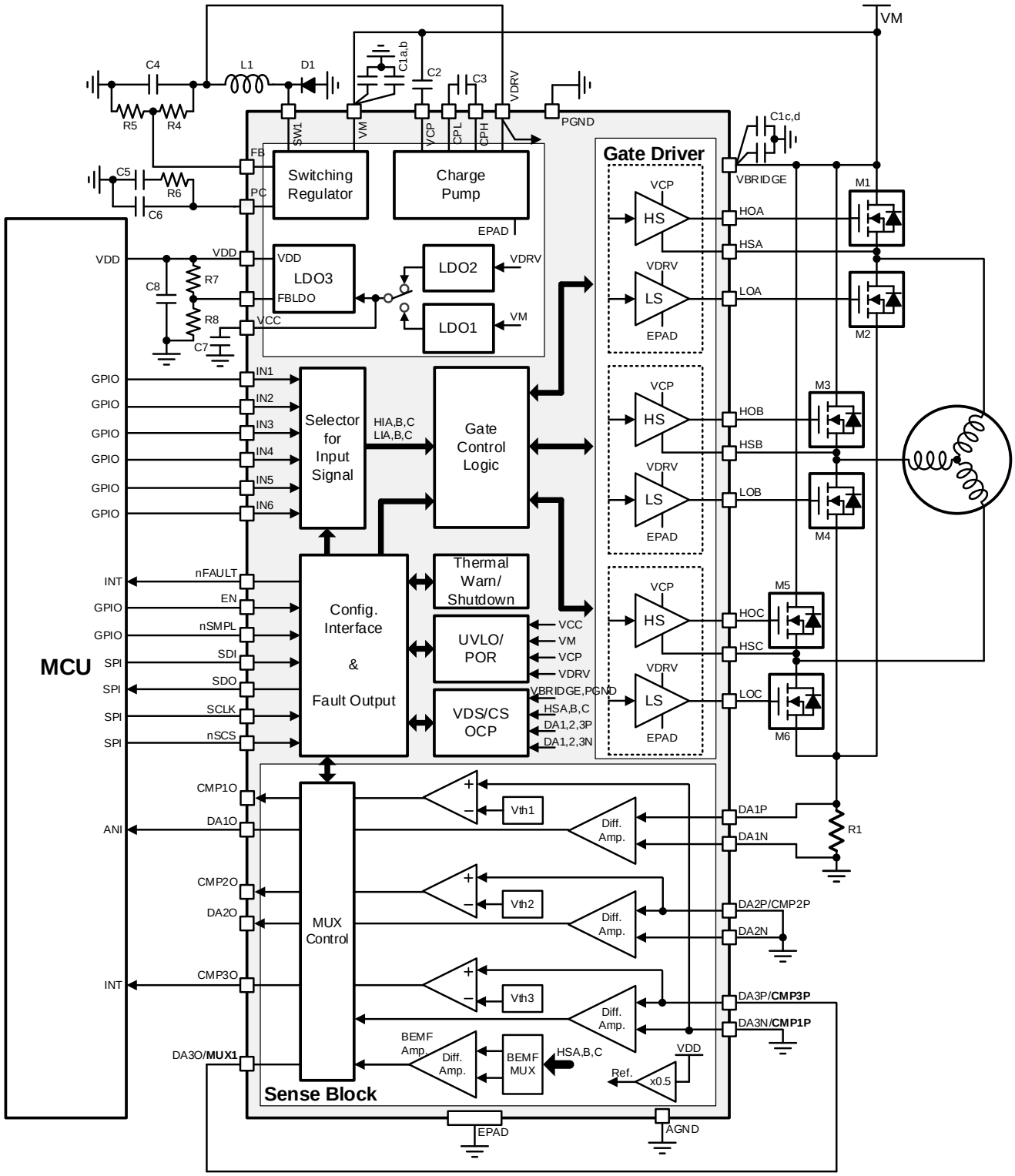


Figure 4.1-2 Simplified Block Diagram and Application – Sensorless Motor Drive by BEMF Sensing Comparator

4.1 Typical Application Circuits (continued)

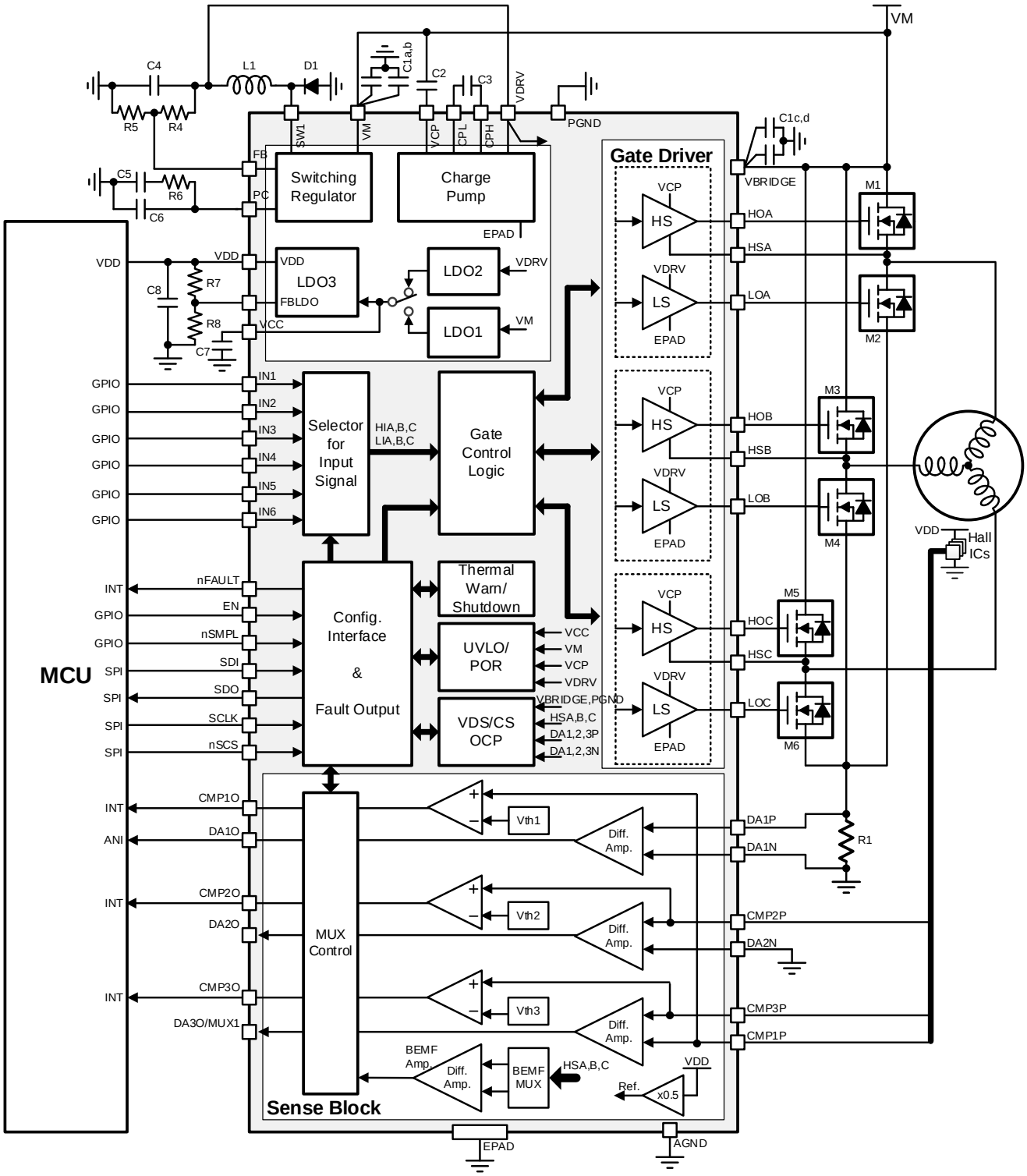


Figure 4.1-3 Simplified Block Diagram and Application – Hall Sensor Motor Drive by Using 3 Comparators

4.1 Typical Application Circuits (continued)

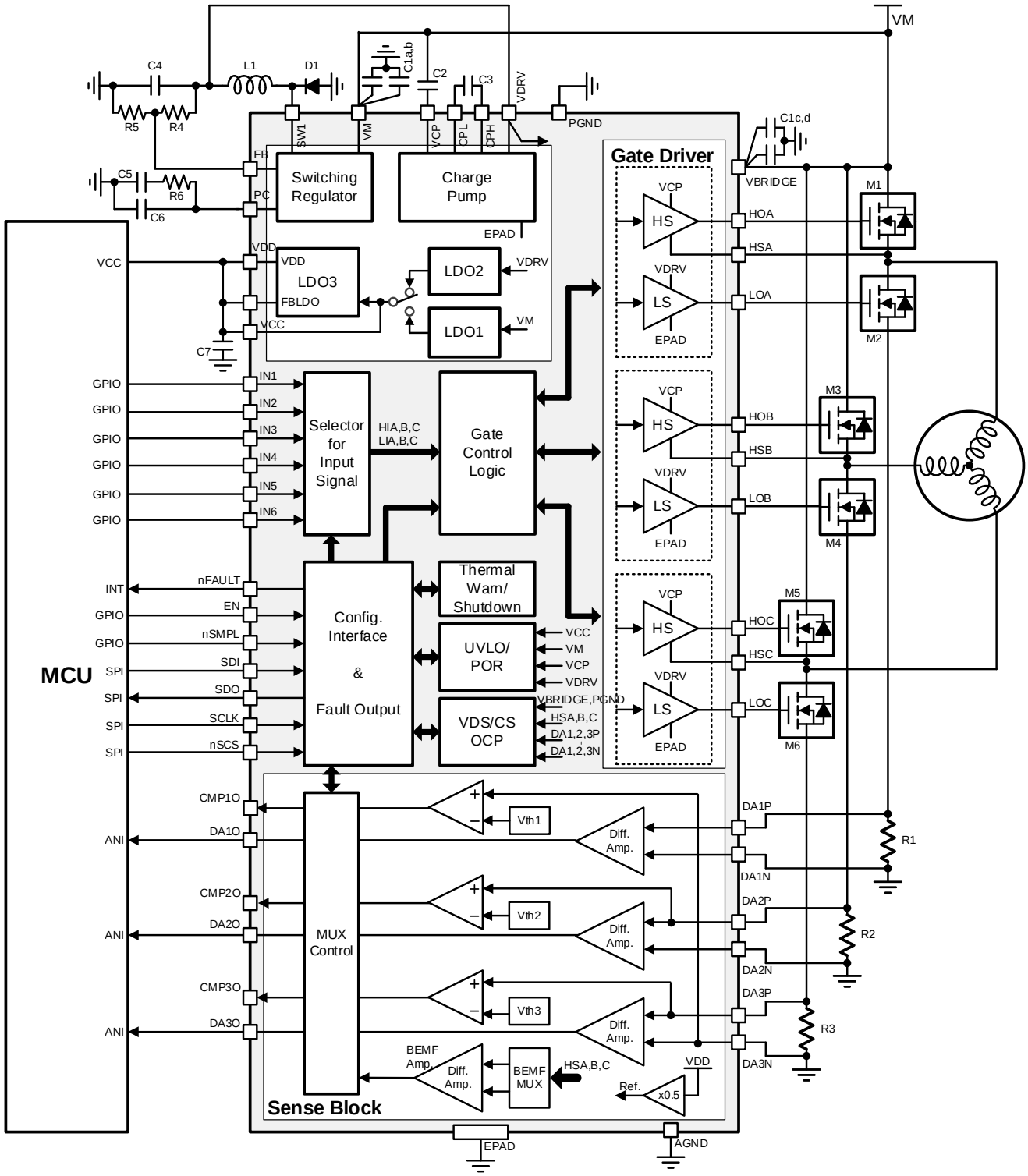


Figure 4.1-4 Simplified Block Diagram and Application – 3 Shunt Sensorless FOC Motor Drive with 5V MCU Supply

4.2 Pin Configurations

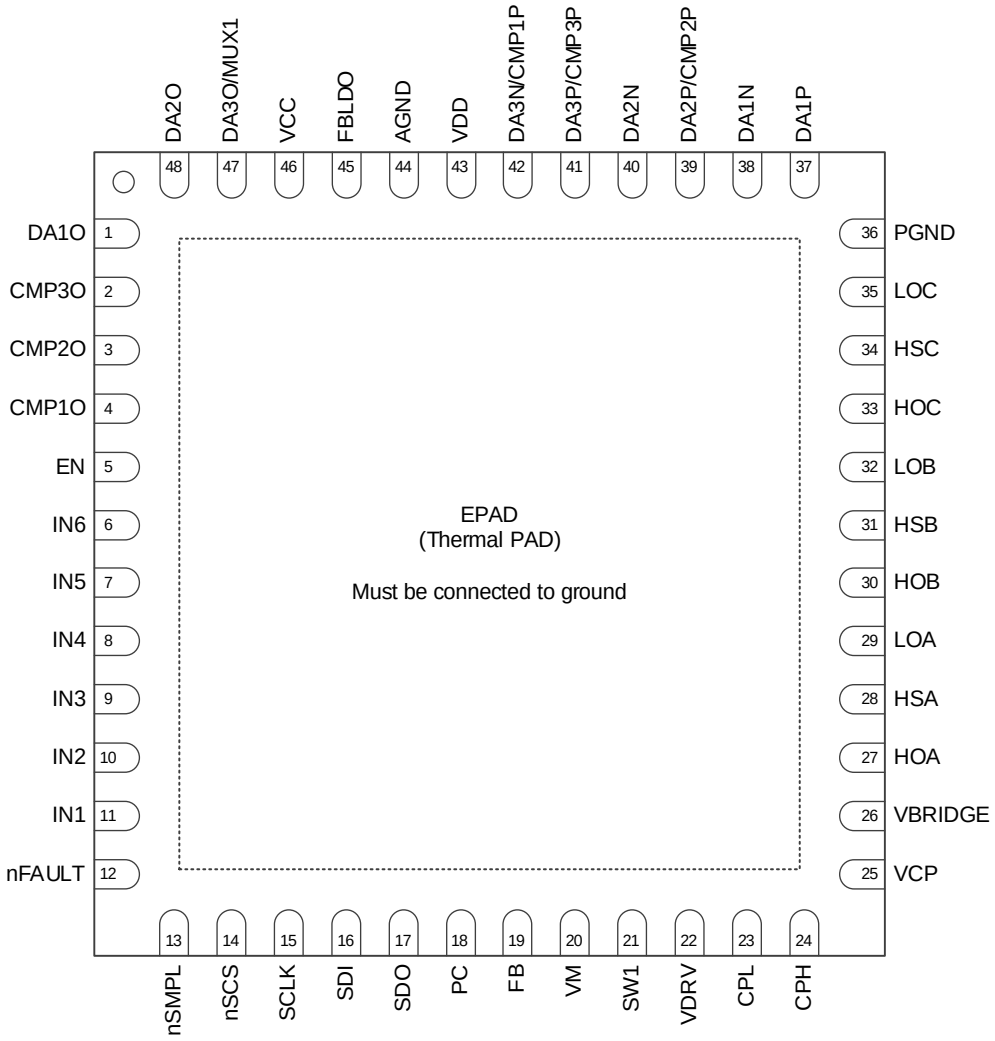


Figure 4.2-1 Pin Configuration Diagram (Top View)

4.3 Pin Descriptions

Table 4.3-1 Pin Descriptions

PIN		TYPE	I/O level	Function
Number	Name			
1	DA1O	OUT	VDD	Output of differential amplifier 1.
2	CMP3O	IN / OUT	VDD	Control input for the detect phase selection of BEMF sense amplifier. / Output of comparator 3.
3	CMP2O	IN / OUT	VDD	Control input for the detect phase selection of BEMF sense amplifier. / Output of comparator 2.
4	CMP1O	IN / OUT	VDD	Control input for the detect phase selection of BEMF sense amplifier. / Output of comparator 1.
5	EN	IN	VDD	Enable control pin for Operating Mode. When this pin is logic low, the device goes to a low-power sleep mode.
6	IN6	IN	VDD	Gate driver control input 6. Control pin of each phase gate driver is selectable by SPI.
7	IN5	IN	VDD	Gate driver control input 5. Control pin of each phase gate driver is selectable by SPI.
8	IN4	IN	VDD	Gate driver control input 4. Control pin of each phase gate driver is selectable by SPI.
9	IN3	IN	VDD	Gate driver control input 3. Control pin of each phase gate driver is selectable by SPI.
10	IN2	IN	VDD	Gate driver control input 2. Control pin of each phase gate driver is selectable by SPI.
11	IN1	IN	VDD	Gate driver control input 1. Control pin of each phase gate driver is selectable by SPI.
12	nFAULT	Open-drain OUT	VDD	Fault indicator output.
13	nSMPL	IN	VDD	Sampling control input of BEMF sense amplifier or differential amplifiers.
14	nSCS	IN	VDD	SPI chip select input.
15	SCLK	IN	VDD	SPI clock input.
16	SDI	IN	VDD	SPI data input.
17	SDO	Open-drain OUT	VDD	SPI data output.
18	PC	OUT	VCC	gm amplifier output for phase compensation of buck switching regulator.
19	FB	IN	VCC	Voltage feedback input of buck switching regulator (Ref.=0.8V).
20	VM	POWER	VM	Power supply input. Connect bypass capacitors between VM and analog ground.
21	SW1	OUT	VM	Switch node of buck switching regulator.
22	VDRV	POWER	VDRV	Output of buck switching regulator, Low-side gate driver supply. Connect to bypass capacitors between VDRV and analog ground.
23	CPL	OUT	VDRV	Charge pump low-side switch node. Connect a flying capacitor between CPH and CPL pins.
24	CPH	OUT	VCP	Charge pump high-side switch node. Connect a flying capacitor between CPH and CPL pins.

4.3 Pin Descriptions (continued)

Table 4.3-2 Pin Descriptions (continued)

PIN		TYPE	I/O level	Function
Number	Name			
25	VCP	POWER	VCP	Charge pump output. Connect a bypass capacitor between VCP and VBRIDGE pins.
26	VBRIDGE	IN	VBRIDGE	Charge pump output reference and high-side MOSFET drain sense Input. Connect a bypass capacitor between VBRIDGE pin and power ground.
27	HOA	OUT	VCP	Phase A high-side gate driver output. Connect to the high-side MOSFET gate.
28	HSA	IN	VBRIDGE	Phase A high-side source sense input. Connect to the high-side MOSFET source.
29	LOA	OUT	VDRV	Phase A low-side gate driver output. Connect to the low-side MOSFET gate.
30	HOB	OUT	VCP	Phase B high-side gate driver output. Connect to the high-side MOSFET gate.
31	HSB	IN	VBRIDGE	Phase B high-side source sense input. Connect to the high-side MOSFET source.
32	LOB	OUT	VDRV	Phase B low-side gate driver output. Connect to the low-side MOSFET gate.
33	HOC	OUT	VCP	Phase C high-side gate driver output. Connect to the high-side MOSFET gate.
34	HSC	IN	VBRIDGE	Phase C high-side source sense input. Connect to the high-side MOSFET source.
35	LOC	OUT	VDRV	Phase C low-side gate driver output. Connect to the low-side MOSFET gate.
36	PGND	GND	GND	Ground sense input of external power stage.
37	DA1P	IN	VDD	Positive input of differential amplifier 1.
38	DA1N	IN	VDD	Negative input of differential amplifier 1.
39	DA2P/CMP2P	IN	VDD	Positive input of differential amplifier 2 and positive input of comparator 2.
40	DA2N	IN	VDD	Negative input of differential amplifier 2.
41	DA3P/CMP3P	IN	VDD	Positive input of differential amplifier 3 and positive input of comparator 3.
42	DA3N/CMP1P	IN	VDD	Negative input of differential amplifier 3 and positive input of comparator 1.
43	VDD	POWER	VDD	Internal series regulator output and power supply of output buffers. Connect to a bypass capacitor between VDD and AGND.
44	AGND	GND	GND	Device analog ground.
45	FBLDO	IN	VCC	Voltage feedback input of internal series regulator (Ref.=1.2V).
46	VCC	POWER	VCC	Internal series regulator output(5V). Connect to a bypass capacitor between VCC and AGND.
47	DA3O/MUX1	OUT	VDD	Output of differential amplifier 3, BEMF sense amplifier, and multiplexer.
48	DA2O	OUT	VDD	Output of differential amplifier 2.
-	EPAD (Thermal PAD)	GND	GND	Power ground for gate driver and charge pump. Must be connected to power ground.

5 Specifications

5.1 Absolute Maximum Ratings

Table 5.1-1 Absolute Maximum Ratings Note1

Item	Symbol	Minimum	Maximum	unit
VBRIDGE to GND Note2	VBRIDGEabs	-0.3	78	V
VM to GND Note2	VMabs	-0.3	65	V
Voltage difference between ground pins (AGND, PGND, EPAD)	DGNDabs	-0.3	0.3	V
VCP to GND, CPH to GND, VCP to HSx, VCP to HOx (x=A,B,C)	VCPabs	-0.3	VBRIDGE + 15 Note3	V
CPL to GND	VCPLabs	-0.3	VDRV + 0.3	V
Continuous HOx (x=A,B,C) to GND	VHOxabs	-5	VCP + 0.5	V
Transient 200ns HOx (x=A,B,C) to GND	VHOxtran	-7	VCP + 0.5	V
HOx to HSx (x=A,B,C)	VGSHxabs	-0.3	15	V
Continuous HSx (x=A,B,C) to GND	VHSxabs	-5	VBRIDGE + 5 Note4	V
Transient 200ns HSx (x=A,B,C) to GND	VHSxtran	-7	VBRIDGE + 7 Note4	V
VDRV to AGND	VDRVabs	-0.3	17	V
Continuous LOx (x=A,B,C) to GND	VLOabs	-1	VDRV + 0.5	V
SW1 to GND	VSWabs	-1	VM + 0.5	V
VCC to GND	VCCabs	-0.3	5.5	V
FB, PC, FBLDO to GND	VFBabs	-0.3	VCC + 0.3	V
DAzP (z=1,2,3) to GND, DAzN (z=1,2,3) to GND	VDAINabs	-1	VDD + 0.6	V
VDD to GND	VDDabs	-0.3	5.5	V
All other Pins	VSIGabs	-0.3	VDD + 0.3	V
Ambient temperature	TA	-40	125	°C
Junction temperature	TJ	-40	150	°C
Storage temperature	Tstg	-65	150	°C
ESD Rating	**Symbol**	**Value**		**Unit**
Human Body Model (Tested per JS-001)	HBM	+/-1000		V
Charged Device Model (Tested per JS-002)	CDM	+/-500		V

Note1: Not subject to production test, specified at Ta=25°C by design

Note2: Power supply can be applied to VBRIDGE and VM independently.

Note3: VCP pin voltage with respect to HOx and HSx pins should be limited to 86V maximum.
This will limit the maximum VCPabs, minimum VHOxabs, VHSxabs, and maximum VGSHxabs when VBRIDGE is greater than 66V.
For example, when VBRIDGE=78V, VCPabs=84V, and VHOxabs=VHSxabs=-2V, VGSHxabs should be limited to (84V - 78V)=6V.
In this example, VDRV pin voltage will be also limited to about 7V in consideration of the step-up voltage of the charge pump.

Note4: In case of VDRV<=7V, the maximum VHSxabs is limited to VBRIDGE+(VDRV-2V).
The maximum VHSxtran is also limited to VBRIDGE+VDRV.

Caution: Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter.
That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

Remark1: The GND pin of each block is the followings.
Gate driver block: EPAD, Charge pump: EPAD, Other blocks: AGND

5.2 Thermal Information

Table 5.2-1 Thermal Information

Thermal resistance (Typical)	θ_{JA}[°C/W]	Ψ_{JT}[°C/W]
7mm×7mm 48Ld QFN Package Note1,2	25.5	1.86

Note1: θ_{JA} is measured in free air with the component mounted on a high-effective thermal conductivity test board with "direct attach" features. See TB379.

Note2: For Ψ_{JT}, the "case temp" location is the center of the exposed metal pad on the package underside.

5.3 Recommended Operating Conditions

Table 5.3-1 Recommended Operating Conditions

Item	Symbol	Minimum	Maximum	unit
VBRIDGE to GND	VBRIDGEope	6	65	V
VM to GND	VMope	6	60	V
VCC to GND	VCCope	4.75	5.25	V
VDRV to GND	VDRVope Note4	5	16	V
VDD to GND	VDDope	3.135	5.25	V
External load current (VDD & VCC), EN=Lo	$I_{VDD}0$	0	50 Note3	mA
External load current (VDD & VCC), EN=Hi, LDO1=On	$I_{VDD}1\_0$	0	70 Note3	mA
External load current (VDD & VCC), EN=Hi	$I_{VDD}1\_1$	0	90 Note3	mA
EN, INz (z=1,2,3,4,5,6), nFAULT, CMPzO (z=1,2,3) SDI, SDO, SCLK, nSCS, DAzP (z=1,2,3), DAzN (z=1,2,3), CMPzP (z=1,2,3), DAzO (z=1,2,3), MUX1	VSGope	0	5.25	V
Operating ambient temperature	TA	-40	125	°C
Operating junction temperature	TJ	-40	150	°C

Note3: Power dissipation and thermal limits must be observed.
External load current is defined as the total of VCC and VDD load current.

Note4: VDRV voltage must be set so that VCC voltage doesn't deviate from the recommended operating condition.

5.4 Electrical Characteristics

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Power supply (VM)						
VM Sleep Mode current	$I_{VM}0$	EN=Lo, INz (z=1,2,3,4,5,6)=Lo, No load on VDD	-	28	44	µA
VM, VBRIDGE operating current (combined)	$I_{VM}1$	EN=Hi, INz (z=1,2,3,4,5,6)=Lo	-	2	-	mA
VDRV operating current	$I_{VDRV}1$	EN=Hi, INz (z=1,2,3,4,5,6)=Lo, VDRV=12V	-	3.8	-	mA
Sleep Mode entry time delay	tsleep	From EN=Lo to Sleep Mode	-	0.55	0.85	ms
Wake-up time delay	twake	From EN=Hi to All power rails ready, C2=2.2µF, C3=0.22µF	-	6.5	-	ms
Power supply (VCC)						
Output voltage	$V_{VCC}0\_1$	VM=60V, EN=Lo, load=0mA	-	5	-	V
	$V_{VCC}0\_2$	VM=36V, EN=Lo, load=10mA	-	5	-	V
	$V_{VCC}0\_3$	VM=6V, EN=Lo, load=50mA	4.75	5	-	V
	$V_{VCC}1\_1$	VDRV=15V, EN=Hi, load=0.1mA	-	5	-	V
	$V_{VCC}1\_2$	VDRV=12V, EN=Hi, load=40mA	4.85	5	5.15	V
	$V_{VCC}1\_3$	VDRV=8V, EN=Hi, load=100mA	-	5	-	V
Current limit	$I_{LMTCC}0$	VM=6V, EN=Lo	50	80	-	mA
	$I_{LMTCC}1\_0$	VM=36V, EN=Hi, LDO1=On	80	130	-	mA
	$I_{LMTCC}1\_1$	VDRV=12V, EN=Hi	100	160	-	mA
Power supply (VDD)						
Output voltage Note2	$V_{VDD}0\_1$	VM=60V, EN=Lo, load=0mA	-	3.3	-	V
	$V_{VDD}0\_2$	VM=36V, EN=Lo, load=10mA	-	3.3	-	V
	$V_{VDD}0\_3$	VM=6V, EN=Lo, load=50mA	3.0	3.3	-	V
	$V_{VDD}1\_1$	VDRV=15V, EN=Hi, load=0.1mA	-	3.3	-	V
	$V_{VDD}1\_2$	VDRV=12V, EN=Hi, load=40mA	3.201	3.3	3.399	V
	$V_{VDD}1\_3$	VDRV=8V, EN=Hi, load=100mA	-	3.3	-	V
Current limit Note3	$I_{LMTDD}0$	VM=6V, EN=Lo	50	80	-	mA
	$I_{LMTDD}1\_0$	VM=36V, EN=Hi, LDO1=On	80	130	-	mA
	$I_{LMTDD}1\_1$	VDRV=12V, EN=Hi	100	160	-	mA

Note2: VDD specification does not apply when VDD LDO is not used by connecting FBLDO to VCC or VDD pin.

Note3: VDD current is limited by VCC current limit. These items are not tested.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Buck switching regulator (VDRV)						
Output voltage adjustment range Note1	V_{DRV}		5	-	15	V
Reference voltage	V_{REF_SR}		0.776	0.8	0.824	V
gm amplifier gain	$gm_{\_SR}$		-	200	-	µA/V
gm amplifier output capability	I_{gmSRC_SR}		-	18	-	µA
	I_{gmSNK_SR}		-	-18	-	µA
gm amplifier maximum output voltage	V_{gmH_SR}	FB=0V	-	3.95	-	V
gm amplifier minimum output voltage	V_{gmL_SR}	FB=1V	-	0.02	-	V
Ramp offset voltage	V_{ramp_SR}		-	870	-	mV
Oscillator frequency	f_{SW_SR}		400	500	575	kHz
Soft-start time	t_{SS_SR}	Time to FB=0.8V	-	2.2	-	ms
Cycle-by-cycle current limit Note1	I_{OC1_SR}		-	1.2	-	A
Hiccup current limit threshold Note1	I_{OC2_SR}		-	1.4	-	A
Charge pump (VCP)						
Output voltage with respect to VBRIDGE	V_{VCP}_1	VM=60V, VDRV=12V, load=15mA	-	11	-	V
	V_{VCP}_2	VM=36V, VDRV=12V, load=15mA	-	11	-	V
	V_{VCP}_3	VM=6V, VDRV=6V, load=15mA	5.0	5.3	-	V
Oscillator frequency	f_{SW_CP}		-	250	-	kHz
Maximum load current Note1	I_{load_CP}		28	-	-	mA

Note1: This specification is not tested in production. Only functional test.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Logic inputs (EN)						
EN rise threshold	V_{THR_EN}		-	1.1	-	V
EN fall threshold	V_{IH_EN}		-	0.92	-	V
EN threshold hysteresis	V_{HYS_EN}		-	160	-	mV
Pulldown resistance	R_{IPD_EN}		-	100	-	kΩ
Logic inputs (INz (z=1,2,3,4,5,6), CMP1O, CMP2O, CMP3O, nSMPL, SDI, SCLK, nSCS)						
Input logic low voltage	V_{IL}		-	1.21	-	V
Input logic high voltage	V_{IH}		-	1.57	-	V
Input logic hysteresis	V_{HYS}		-	360	-	mV
Input logic low current1	$I_{IL}1$	VIN=0V, for other than nSCS	-	15	-	nA
Input logic low current2	$I_{IL}2$	VIN=0V, for nSCS	-	9	-	µA
Input logic high current1	$I_{IH}1$	VIN=VDD, for other than nSCS	-	9	-	µA
Input logic high current2	$I_{IH}2$	VIN=VDD, for nSCS	-	15	-	nA
Pulldown resistance	R_{IPD}	for other than nSCS	-	380	-	kΩ
Pullup resistance	R_{IPU}	for nSCS	-	380	-	kΩ
Logic outputs (nFault, SDO)						
Output logic low voltage	$V_{OL}1$	Io=2mA	-	0.24	-	V
High impedance output leakage	I_{OHLK}	Vo=VDD	-	50	-	nA
Logic outputs (CMP1O, CMP2O, CMP3O)						
Output logic low voltage	$V_{OL}2$	Io=1mA	-	0.14	-	V
Output logic high voltage	V_{OH}	Io=-1mA	-	V_{VDD}-0.24	-	V
Gate driver block (HOx, LOx (x=A,B,C))						
High-side gate driver Output high voltage	V_{GSHH}	Io=1mA, VDRV=12V, VM=HSx=24V, with respect to HSx	-	11.45	-	V
High-side gate driver Output low voltage	V_{GSHL}	Io=-1mA, VDRV=12V, VM=HSx=24V, with respect to HSx	-	0.05	-	V
Low-side gate driver Output high voltage	V_{GSLH}	Io=1mA, VDRV=12V, VM=HSx=24V, with respect to GND	-	11.95	-	V
Low-side gate driver Output low voltage	V_{GSLL}	Io=-1mA, VDRV=12V, VM=HSx=24V, with respect to GND	-	0.05	-	V
High-side gate clamp voltage Note1	V_{CLMP}	Io=-1mA, with respect to HSx	16.0	17.2	18.4	V
Gate pulldown resistance	R_{PD}		-	200	-	kΩ
Strong sink current Note1	I_{SNK_STG}		-	1280	-	mA
Pullup source current	I_{SRC_PU}		-	50	-	mA
Pulldown sink current	I_{SNK_PD}		-	100	-	mA

Note1: This specification is not tested in production. Only functional test.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Gate driver block (HOx (x=A,B,C))						
High-side peak source current Note1	I_{SRCH0}	ISRC_HS[3:0]=0h	-	50	-	mA
	I_{SRCH1}	ISRC_HS[3:0]=1h	-	60	-	mA
	I_{SRCH2}	ISRC_HS[3:0]=2h	-	70	-	mA
	I_{SRCH3}	ISRC_HS[3:0]=3h	-	80	-	mA
	I_{SRCH4}	ISRC_HS[3:0]=4h	-	100	-	mA
	I_{SRCH5}	ISRC_HS[3:0]=5h	-	120	-	mA
	I_{SRCH6}	ISRC_HS[3:0]=6h	-	140	-	mA
	I_{SRCH7}	ISRC_HS[3:0]=7h	-	160	-	mA
	I_{SRCH8}	ISRC_HS[3:0]=8h	-	200	-	mA
	I_{SRCH9}	ISRC_HS[3:0]=9h	-	240	-	mA
	I_{SRCH10}	ISRC_HS[3:0]=Ah	-	280	-	mA
	I_{SRCH11}	ISRC_HS[3:0]=Bh	-	320	-	mA
	I_{SRCH12}	ISRC_HS[3:0]=Ch	-	400	-	mA
	I_{SRCH13}	ISRC_HS[3:0]=Dh	-	480	-	mA
	I_{SRCH14}	ISRC_HS[3:0]=Eh	-	560	-	mA
	I_{SRCH15}	ISRC_HS[3:0]=Fh	-	640	-	mA
High-side peak sink current Note1	I_{SNKH0}	ISRC_HS[3:0]=0h	-	100	-	mA
	I_{SNKH1}	ISRC_HS[3:0]=1h	-	120	-	mA
	I_{SNKH2}	ISRC_HS[3:0]=2h	-	140	-	mA
	I_{SNKH3}	ISRC_HS[3:0]=3h	-	160	-	mA
	I_{SNKH4}	ISRC_HS[3:0]=4h	-	200	-	mA
	I_{SNKH5}	ISRC_HS[3:0]=5h	-	240	-	mA
	I_{SNKH6}	ISRC_HS[3:0]=6h	-	280	-	mA
	I_{SNKH7}	ISRC_HS[3:0]=7h	-	320	-	mA
	I_{SNKH8}	ISRC_HS[3:0]=8h	-	400	-	mA
	I_{SNKH9}	ISRC_HS[3:0]=9h	-	480	-	mA
	I_{SNKH10}	ISRC_HS[3:0]=Ah	-	560	-	mA
	I_{SNKH11}	ISRC_HS[3:0]=Bh	-	640	-	mA
	I_{SNKH12}	ISRC_HS[3:0]=Ch	-	800	-	mA
	I_{SNKH13}	ISRC_HS[3:0]=Dh	-	960	-	mA
	I_{SNKH14}	ISRC_HS[3:0]=Eh	-	1120	-	mA
	I_{SNKH15}	ISRC_HS[3:0]=Fh	-	1280	-	mA

Note1: This specification is not tested in production. Only functional test.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Gate driver block (LOx (x=A,B,C))						
Low-side peak source current Note1	I_{SRCL0}	ISRC_LS[3:0]=0h	-	50	-	mA
	I_{SRCL1}	ISRC_LS[3:0]=1h	-	60	-	mA
	I_{SRCL2}	ISRC_LS[3:0]=2h	-	70	-	mA
	I_{SRCL3}	ISRC_LS[3:0]=3h	-	80	-	mA
	I_{SRCL4}	ISRC_LS[3:0]=4h	-	100	-	mA
	I_{SRCL5}	ISRC_LS[3:0]=5h	-	120	-	mA
	I_{SRCL6}	ISRC_LS[3:0]=6h	-	140	-	mA
	I_{SRCL7}	ISRC_LS[3:0]=7h	-	160	-	mA
	I_{SRCL8}	ISRC_LS[3:0]=8h	-	200	-	mA
	I_{SRCL9}	ISRC_LS[3:0]=9h	-	240	-	mA
	I_{SRCL10}	ISRC_LS[3:0]=Ah	-	280	-	mA
	I_{SRCL11}	ISRC_LS[3:0]=Bh	-	320	-	mA
	I_{SRCL12}	ISRC_LS[3:0]=Ch	-	400	-	mA
	I_{SRCL13}	ISRC_LS[3:0]=Dh	-	480	-	mA
	I_{SRCL14}	ISRC_LS[3:0]=Eh	-	560	-	mA
	I_{SRCL15}	ISRC_LS[3:0]=Fh	-	640	-	mA
Low-side peak sink current Note1	I_{SNKL0}	ISRC_LS[3:0]=0h	-	100	-	mA
	I_{SNKL1}	ISRC_LS[3:0]=1h	-	120	-	mA
	I_{SNKL2}	ISRC_LS[3:0]=2h	-	140	-	mA
	I_{SNKL3}	ISRC_LS[3:0]=3h	-	160	-	mA
	I_{SNKL4}	ISRC_LS[3:0]=4h	-	200	-	mA
	I_{SNKL5}	ISRC_LS[3:0]=5h	-	240	-	mA
	I_{SNKL6}	ISRC_LS[3:0]=6h	-	280	-	mA
	I_{SNKL7}	ISRC_LS[3:0]=7h	-	320	-	mA
	I_{SNKL8}	ISRC_LS[3:0]=8h	-	400	-	mA
	I_{SNKL9}	ISRC_LS[3:0]=9h	-	480	-	mA
	I_{SNKL10}	ISRC_LS[3:0]=Ah	-	560	-	mA
	I_{SNKL11}	ISRC_LS[3:0]=Bh	-	640	-	mA
	I_{SNKL12}	ISRC_LS[3:0]=Ch	-	800	-	mA
	I_{SNKL13}	ISRC_LS[3:0]=Dh	-	960	-	mA
	I_{SNKL14}	ISRC_LS[3:0]=Eh	-	1120	-	mA
	I_{SNKL15}	ISRC_LS[3:0]=Fh	-	1280	-	mA

Note1: This specification is not tested in production. Only functional test.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Gate driver block (HOx, LOx (x=A,B,C))						
Adaptive dead time VGS falling threshold Note1	V_{TH_VGS}		-	1	-	V
Adaptive dead time VGS threshold hysteresis Note1	V_{HYS_VGS}		-	2	-	V
Gate drive dead time	t_{DT0}	DEAD_TIME[1:0]=0h from VTH_VGS to other-side ON	-	50	-	ns
	t_{DT1}	DEAD_TIME[1:0]=1h from VTH_VGS to other-side ON	-	100	-	ns
	t_{DT2}	DEAD_TIME[1:0]=2h from VTH_VGS to other-side ON	-	200	-	ns
	t_{DT3}	DEAD_TIME[1:0]=3h from VTH_VGS to other-side ON	-	400	-	ns
Maximum gate transition time	t_{GT0}	T_GT[1:0]=0h from Source/Sink start to Pullup/down start	-	500	-	ns
	t_{GT1}	T_GT[1:0]=1h from Source/Sink start to Pullup/down start	-	1000	-	ns
	t_{GT2}	T_GT[1:0]=2h from Source/Sink start to Pullup/down start	-	2000	-	ns
	t_{GT3}	T_GT[1:0]=3h from Source/Sink start to Pullup/down start	-	4000	-	ns
Propagation delay Note1	t_{PROP}	from INz toggle to HOx/LOx start	-	40	-	ns
Differential amplifier (DAzP, DAzN, DAzO (z=1,2,3))						
Common mode input voltage range	V_{IC_CSA}	DAzP, DAzN pin	-0.5	-	2.8	V
Differential mode input voltage range	V_{ID_CSA}	DAzP - DAzN	-0.5	-	0.5	V
Input offset voltage	V_{IO_CSA}	CAL_CONN=1	-5	-	5	mV
Output voltage linear range Note1	V_{O_CSA}		0.4	-	VDD-0.4	V
Amplifier gain	G_{CSA0}	DAz_GAIN[1:0]=0h	4.85	5	5.15	V/V
	G_{CSA1}	DAz_GAIN[1:0]=1h	9.7	10	10.3	V/V
	G_{CSA2}	DAz_GAIN[1:0]=2h	19.4	20	20.6	V/V
	G_{CSA3}	DAz_GAIN[1:0]=3h	38.8	40	41.2	V/V
Settling time to +/-1% Note1	t_{SET_CSA0}	DAz_GAIN[1:0]=0h	-	250	-	ns
	t_{SET_CSA1}	DAz_GAIN[1:0]=1h	-	300	-	ns
	t_{SET_CSA2}	DAz_GAIN[1:0]=2h	-	400	-	ns
	t_{SET_CSA3}	DAz_GAIN[1:0]=3h	-	500	-	ns

Note1: This specification is not tested in production. Only functional test.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
BEMF sense amplifier (HSA, HSB, HSC, DA3O)						
Common mode input voltage range	V_{IC_EMF}	HSA, HSB, HSC pin	-2.5	-	VM+2.5	V
Differential mode input voltage range0	V_{ID_EMF0}	HSA, HSB, HSC pin - VBRIDGE/2 BEMF_GAIN[1:0]=0h	-24	-	24	V
Differential mode input voltage range3	V_{ID_EMF3}	HSA, HSB, HSC pin - VBRIDGE/2 BEMF_GAIN[1:0]=3h	-1.2	-	1.2	V
Output offset voltage0	V_{IO_EMF0}	DA3O pin with respect to VDD/2 BEMF_GAIN[1:0]=0h	-0.2	-	0.2	V
Output offset voltage3	V_{IO_EMF3}	DA3O pin with respect to VDD/2 BEMF_GAIN[1:0]=3h	-0.3	-	0.3	V
Amplifier output voltage bias	V_{REF_EMF}		-	VDD/2	-	V
BEMF sense amplifier gain	G_{EMF0}	BEMF_GAIN[1:0]=0h	0.0475	0.05	0.0525	V/V
	G_{EMF1}	BEMF_GAIN[1:0]=1h	0.095	0.1	0.105	V/V
	G_{EMF2}	BEMF_GAIN[1:0]=2h	0.475	0.5	0.525	V/V
	G_{EMF3}	BEMF_GAIN[1:0]=3h	0.95	1	1.05	V/V
Settling time to +/-1% Note1	t_{SET_EMF0}	BEMF_GAIN[1:0]=0h	-	350	-	ns
	t_{SET_EMF1}	BEMF_GAIN[1:0]=1h	-	400	-	ns
	t_{SET_EMF2}	BEMF_GAIN[1:0]=2h	-	500	-	ns
	t_{SET_EMF3}	BEMF_GAIN[1:0]=3h	-	1000	-	ns

Note1: This specification is not tested in production. Only functional test.

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Comparator (CMPzO (z=1,2,3))						
Threshold voltage for VDD=3.3V setting	V_{TH_CMP1}	CMPz_VTH[3:0] =1h	0.173	0.206	0.221	V
	V_{TH_CMP2}	CMPz_VTH[3:0] =2h	0.380	0.413	0.433	V
	V_{TH_CMP3}	CMPz_VTH[3:0] =3h	0.589	0.619	0.650	V
	V_{TH_CMP4}	CMPz_VTH[3:0] =4h	0.784	0.825	0.866	V
	V_{TH_CMP5}	CMPz_VTH[3:0] =5h	0.980	1.031	1.083	V
	V_{TH_CMP6}	CMPz_VTH[3:0] =6h	1.176	1.238	1.299	V
	V_{TH_CMP7}	CMPz_VTH[3:0] =7h	1.372	1.444	1.516	V
	V_{TH_CMP8}	CMPz_VTH[3:0] =8h	1.568	1.650	1.733	V
	V_{TH_CMP9}	CMPz_VTH[3:0] =9h	1.763	1.856	1.949	V
	V_{TH_CMP10}	CMPz_VTH[3:0] =Ah	1.959	2.063	2.166	V
	V_{TH_CMP11}	CMPz_VTH[3:0] =Bh	2.155	2.269	2.382	V
	V_{TH_CMP12}	CMPz_VTH[3:0] =Ch	2.351	2.475	2.599	V
	V_{TH_CMP13}	CMPz_VTH[3:0] =Dh	2.547	2.681	2.815	V
	V_{TH_CMP14}	CMPz_VTH[3:0] =Eh	2.743	2.888	3.032	V
	V_{TH_CMP15}	CMPz_VTH[3:0] =Fh	2.939	3.094	3.248	V
Hysteresis	V_{HYS_CMP0}	CMPz_HYS=0	-	+/-44	-	mV
	V_{HYS_CMP1}	CMPz_HYS=1	-	0	-	mV
Comparator delay	t_{DLY_CMP}		-	-	1	µs
Fault management						
VCC power-on-reset rising	V_{CCUVR}		-	4.00	-	V
VCC power-on-reset falling	V_{CCUV}		-	3.63	-	V
VM under voltage rising0	V_{VMUVR0}	VMUV_TH=0	5.2	5.5	5.8	V
VM under voltage falling0	V_{VMUV0}	VMUV_TH=0	5.0	5.3	5.6	V
VM under voltage rising1	V_{VMUVR1}	VMUV_TH=1	7.38	7.78	8.18	V
VM under voltage falling1	V_{VMUV1}	VMUV_TH=1	7.1	7.5	7.9	V
VM over voltage rising	V_{VMOV}		-	63	-	V
VM over voltage falling	V_{VMOVR}		-	60	-	V
VDRV under voltage fault rising	V_{DRVUVR}		4.2	4.4	4.6	V
VDRV under voltage fault falling	V_{DRVUV}		4.0	4.2	4.4	V
VDRV over voltage rising	V_{DRVOV}	FB pin voltage	0.92	0.95	0.98	V
VDRV over voltage falling	V_{DRVOVR}	FB pin voltage	0.80	0.825	0.85	V
VCP under voltage fault falling0	V_{CPUV0}	with respect to VBRIDGE, CPUV_TH=0	-	0.58*VDRV	-	V
VCP under voltage fault hysteresis0	$V_{CPUVHYS0}$	CPUV_TH=0	-	0.07*VDRV	-	V
VCP under voltage fault falling1	V_{CPUV1}	with respect to VBRIDGE, CPUV_TH=1	-	0.8*VDRV	-	V
VCP under voltage fault hysteresis1	$V_{CPUVHYS1}$	CPUV_TH=1	-	0.09*VDRV	-	V

5.4 Electrical Characteristics (continued)

Note: All specifications are for Ta=25°C, VM=48V unless otherwise noted.

Parameter	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Fault management						
Thermal warning threshold Note1	T_{WARN}		-	140	-	°C
Thermal shutdown threshold Note1	T_{SD}		-	160	-	°C
Thermal hysteresis Note1	T_{HYS}		-	15	-	°C
MOSFET V_{DS} OCP threshold	V_{DSOCP}	VDS_TH[3:0]=0h	-	40	-	mV
		VDS_TH[3:0]=1h	-	60	-	mV
		VDS_TH[3:0]=2h	-	80	-	mV
		VDS_TH[3:0]=3h	-	120	-	mV
		VDS_TH[3:0]=4h	-	160	-	mV
		VDS_TH[3:0]=5h	-	200	-	mV
		VDS_TH[3:0]=6h	-	240	-	mV
		VDS_TH[3:0]=7h	-	320	-	mV
		VDS_TH[3:0]=8h	-	400	-	mV
		VDS_TH[3:0]=9h	-	480	-	mV
		VDS_TH[3:0]=Ah	-	600	-	mV
		VDS_TH[3:0]=Bh	-	720	-	mV
		VDS_TH[3:0]=Ch	-	960	-	mV
		VDS_TH[3:0]=Dh	-	1200	-	mV
		VDS_TH[3:0]=Eh	-	1600	-	mV
		VDS_TH[3:0]=Fh	-	2000	-	mV
Shunt current sense OCP threshold	V_{CSOCP}	CSOCP_TH[2:0]=0h	37	51	65	mV
		CSOCP_TH[2:0]=1h	87	105	120	mV
		CSOCP_TH[2:0]=2h	131	157	175	mV
		CSOCP_TH[2:0]=3h	179	208	234	mV
		CSOCP_TH[2:0]=4h	224	260	290	mV
		CSOCP_TH[2:0]=5h	456	516	572	mV
		CSOCP_TH[2:0]=6h	686	773	857	mV
		CSOCP_TH[2:0]=7h	917	1029	1140	mV
OCP Deglitch time	t_{DEG_OCP}	DEG_TIME[1:0]=0h	1.10	1.57	2.04	µs
		DEG_TIME[1:0]=1h	1.67	2.38	3.09	µs
		DEG_TIME[1:0]=2h	2.44	3.49	4.54	µs
		DEG_TIME[1:0]=3h	4.01	5.73	7.45	µs
OCP Retry time	t_{RETRY_OCP}	TRETRY_CSOCP, VDSOCP=0	-	4000	-	µs
		TRETRY_CSOCP, VDSOCP=1	-	70	-	µs

Note1: This specification is not tested in production. Only functional test.

5.5 SPI Timing Specification

The communication between MCU and this IC is executed by Four-wire Serial Peripheral Interface. These signals have to keep the following specifications. In some cases, the external pullup resistor for SDO pin is required depending on the SCLK period and the load capacitance including the parasitic capacitance. Please construct the suitable F/W to get the certain communication.

Table 5.5-1 SPI Timing Specification

Item	Symbol	Condition	Rated level			Unit
			MIN	TYP	MAX	
Sleep mode to SPI interface ready	t_{ready}	VCC>V_{CCUVR}, EN from low to high	-	2	-	ms
SCLK period	t_{CLK}		200	-	-	ns
SCLK high time	t_{CLKH}		100	-	-	ns
SCLK low time	t_{CLKL}		100	-	-	ns
SDI input data setup time	t_{SSDI}		40	-	-	ns
SDI input data hold time	t_{HSDI}		60	-	-	ns
SDO output data delay time	t_{dSDO}		-	-	60	ns
nSCS input setup time	t_{SnSCS}		100	-	-	ns
nSCS input hold time	t_{HnSCS}		100	-	-	ns
nSCS high time before pulling low	t_{nSCSH}		400	-	-	ns
Chip de-select off time	$t_{OFFnSCS}$		-	20	-	ns

Note: These specifications are not tested in production. Only functional test.

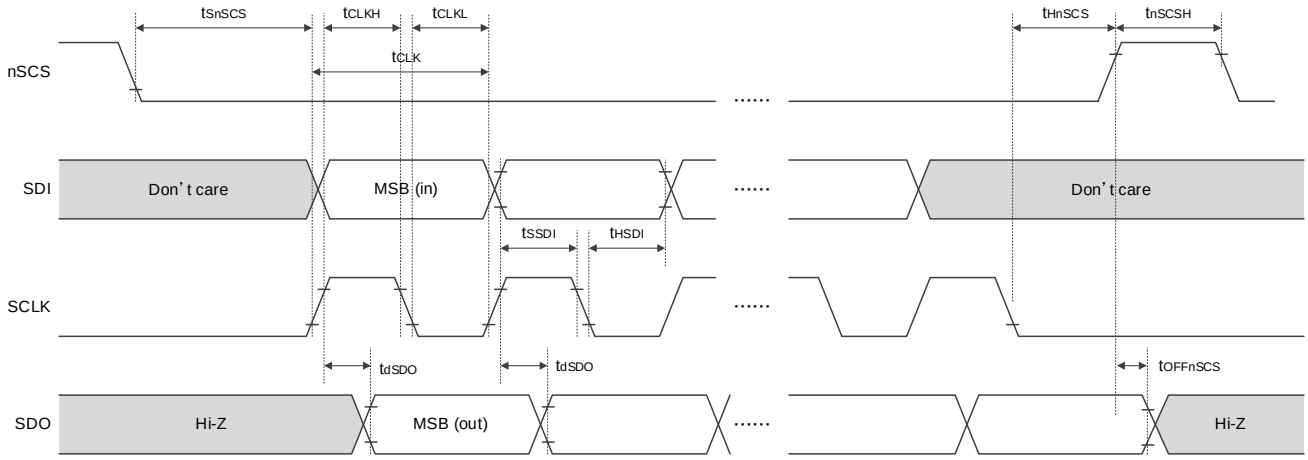


Figure 5.5-1 SPI Timing Diagram

6 Detailed Description

6.1 Power-On Sequence and Functional Modes

6.1.1 Power-On Sequence

Figure 6.1-1 shows the example of Power-On sequence. RAA306012 mode of operation and nFAULT signal work according to EN signal from MCU and supply voltages including the output of the internal regulators. Refer to section 6.1.2, 6.1.3, and 6.2.

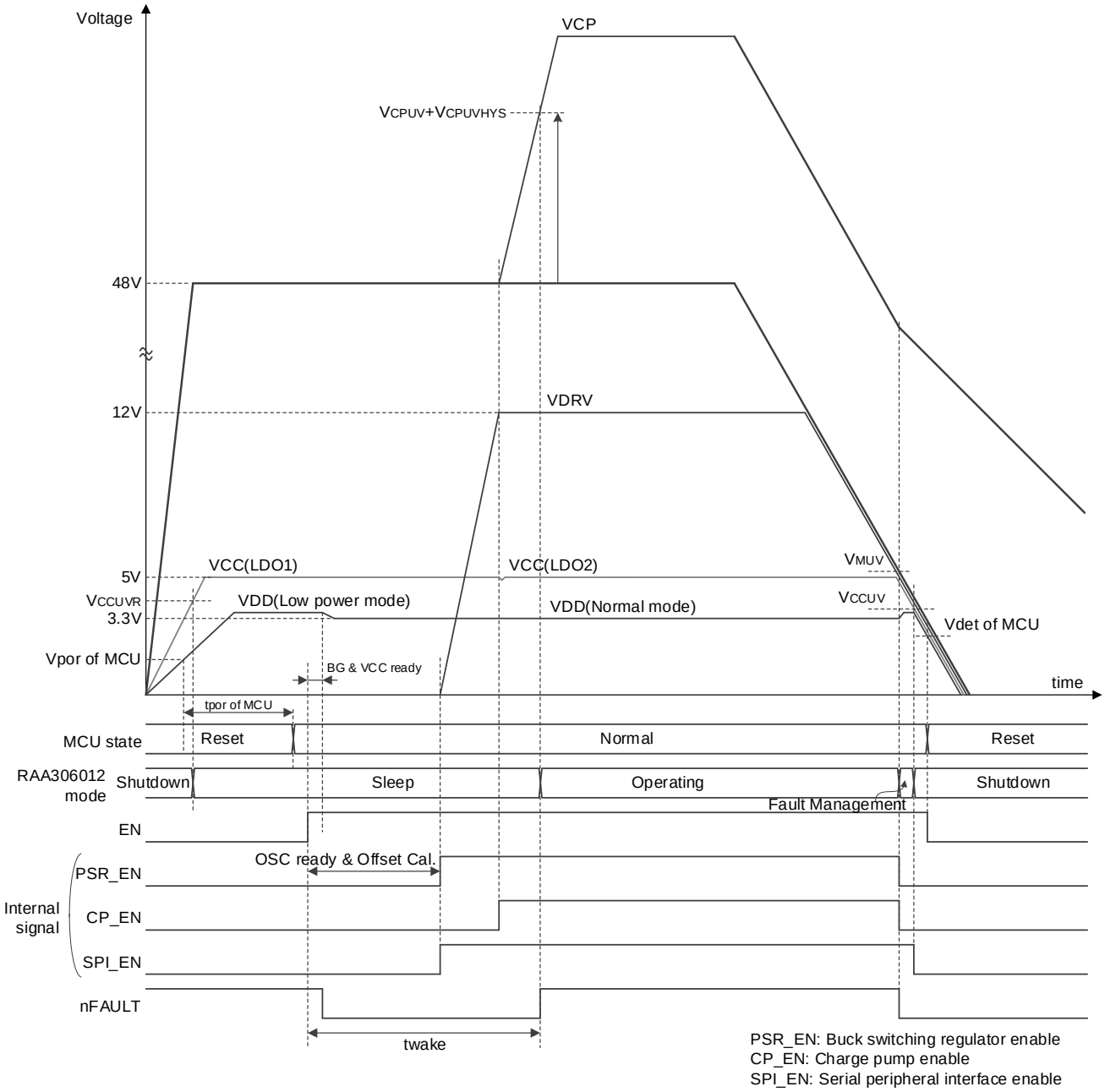


Figure 6.1-1 Power-On Sequence

6.1.2 Definitions of State of Different Modes

The device contains four modes of operation:

Shutdown Mode:

This mode represents the state where V_{CC} voltage is below the POR falling threshold (typical 3.63V). In this mode, most of internal functional blocks are disabled except for LDO1 and LDO3. LDO1 keeps powering the VCC rail until VM drops further below 1.2V. LDO3 is enabled in low power mode. Refer to section 6.3.3.

Sleep Mode:

The RAA306012 is in low-power Sleep Mode when V_{CC} is above the POR rising threshold (typical 4.0V) and EN is low. In this mode, the driver output is disabled and ignores any control input on HIx/LIx (x=A,B,C) selected from INz (z=1,2,3,4,5,6) signals. The power chain associated with the buck regulator (buck, charge pump, LDO2) is disabled. LDO1 and LDO3 are kept alive. LDO3 is enabled in low power mode. This minimizes the IC power consumption in Sleep Mode.

Operating Mode:

This mode represents the state when V_{CC} is above the POR rising threshold (typical 4.0V) and EN is pulled high. The RAA306012 is put into normal operation, high-efficiency buck regulator power chain (buck, charge pump, LDO2) and LDO3 are enabled in normal mode, and LDO1 is disabled. Driver output is enabled in response to control inputs on the HIx/LIx (x=A,B,C) selected from INz (z=1,2,3,4,5,6) signals. No occurrence of any fault is required in this mode.

Fault Management Mode:

This mode represents the state after any fault occurs. The smart gate driver reacts to fault conditions (see section 6.2 for detailed responses) and reports the fault status to the MCU using the nFAULT pin and through the SPI interface. In this mode, the functioning of blocks depends on the fault source and control settings.

6.1.3 Mode Transition

The mode transition conditions (A to J) are summarized in Table 6.1-1

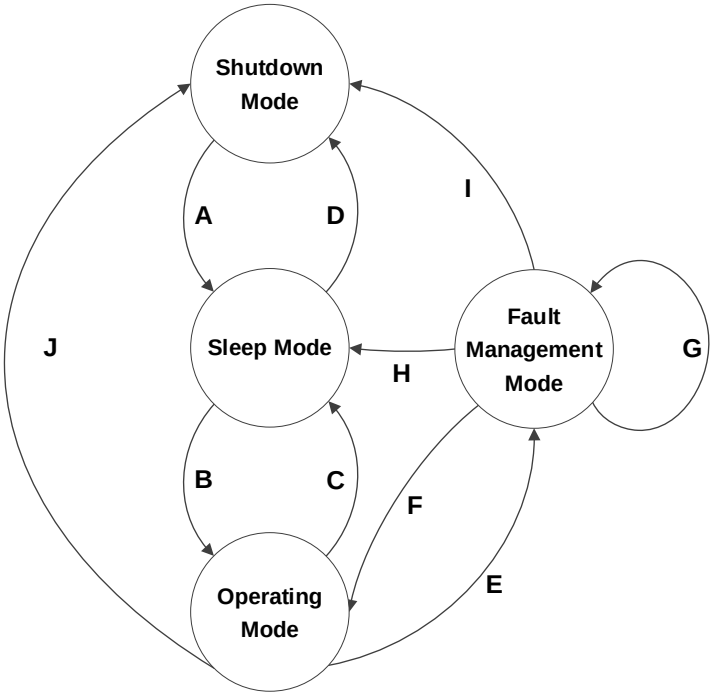


Figure 6.1-2 Mode Transition State Machine

6.1.3 Mode Transition (continued)

Table 6.1-1 Mode Transition Condition of State Machine

Transition	Exiting Mode	Entering Mode	Conditions	Actions
A	Shutdown	Sleep	VM rising, VCC > V_{CCUVR}	Disable most of the function blocks except for VCC LDO1 and VDD LDO3 Note2. Entering Sleep Mode after device start-up delay.
B	Sleep	Operating	EN goes high and stays.	Enable all function blocks except for VCC LDO1. Entering Operating Mode after Wake-up time delay "twake".
C	Operating	Sleep	EN goes low and stays. Note1	Disable most of the function blocks except for VCC LDO1 and VDD LDO3 Note2. The control registers are reset. Entering Sleep Mode after Entry time delay "tsleep".
D	Sleep	Shutdown	VM falling, and VCC < V_{CCUV}	Disable most of the function blocks except for VCC LDO1 and VDD LDO3 Note2. Entering Shutdown Mode after shutdown time delay.
E	Operating	Fault Management	Any fault condition occurs.	Respond based on fault management matrix.
F	Fault Management	Operating	Fault condition clears and operating recovers based on recovery action.	Enable all function blocks except for VCC LDO1. Entering Operating Mode after fault recovery delay.
G	Fault Management	Fault Management	Any other fault condition occurs.	Respond based on fault management matrix. Each function block disabling signal (active low) from multiple fault sources are ANDed together.
H	Fault Management	Sleep	EN goes low and stays. Note1	Disable most of the function blocks except for VCC LDO1 and VDD LDO3 Note2. The control registers are reset. Entering Sleep Mode after fault recovery delay.
I	Fault Management	Shutdown	VCC < V_{CCUV}	Disable most of the function blocks except for VCC LDO1 and VDD LDO3 Note2. The control registers are reset. Entering Shutdown Mode after shutdown time delay.
J	Operating	Shutdown	VCC < V_{CCUV}	Disable most of the function blocks except for VCC LDO1 and VDD LDO3 Note2. The control registers are reset. Entering Shutdown Mode after shutdown time delay.

Note1: EN low pulse shorter than "tsleep" maximum (0.85ms) must NOT be input to avoid the unexpected behavior.

Note2: LDO3 enters a low-power mode when the EN pin is pulled low.

6.2 Fault Management

The RAA306012 has the protect function against VM undervoltage, VM overvoltage, Charge pump undervoltage, MOSFET V_{DS} overcurrent, Current sense overcurrent, MOSFET V_{GS} fault, Thermal warning, Thermal shutdown, Buck regulator overcurrent, Buck regulator undervoltage, and Buck regulator overvoltage events. When a fault occurs, the individual fault bit is set high along with the global FAULT bit in FAULT status register. The FAULT bit is OR'ed with all the other individual status bits. The fault and recovery action of each function is shown in Table 6.2-1 and Table 6.2-2.

6.2 Fault Management (continued)

Table 6.2-1 Fault Management Matrix

Fault type	Condition	Configuration	Reports	Gate driver Input / Output	Regulator in Fault Management Mode	Recovery condition	Recovery action
VCC undervoltage (VCC_UV)	VCC < V_{CCUV} (3.63V)	-	None	Blocked / Hi-Z pulldown	LDO1, LDO3 enable, others disable	VCC > V_{CCUVR} (4.0V)	Enter Sleep Mode
VM undervoltage (VM_UV)	VM < V_{VMUV} (5.3V or 7.5V)	DIS_VMUV=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-Z pulldown	LDO1, LDO3 enable, others disable	VM > V_{VMUVR} (5.5V or 7.78V)	RA1
		DIS_VMUV=1b		Active	Keep state in original mode		RA0
VM overvoltage (VM_OV)	VM > V_{VMOV} (63V)	DIS_VMOV=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-Z pulldown	LDO1, LDO3 enable, others disable	VM < V_{VMOVR} (60V)	RA2
		DIS_VMOV=1b		Active	Keep state in original mode		RA0
VCP undervoltage (VCP_UV)	VCP < V_{CPUV} (0.58*VDRV)	DIS_VCPUV=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-Z pulldown	All reg. except LDO1 enable	VCP > V_{CPUV}+ $V_{CPUVHYS}$ (0.58*VDRV +0.07*VDRV)	RA3
		DIS_VCPUV=1b		Active	Keep state in original mode		RA0
V_{DS} overcurrent (VDS_OCP)	V_{DSX} > V_{DSOCP} (x=A,B,C)	VDSOCP_MODE=00b	nFAULT=Lo (Latched) Status bit=1b	Blocked / Depends on PDMODE	All reg. except LDO1 enable	V_{DSX} < V_{DSOCP} (x=A,B,C) & clear latch	RA4
		VDSOCP_MODE=01b				Retry after t_{RETRY_OCP}	RA5
		VDSOCP_MODE=10b		Active	Keep state in original mode	V_{DSX} < V_{DSOCP} (x=A,B,C) & clear latch	RA6
		VDSOCP_MODE=11b	None			None	No action is necessary.
Current sense overcurrent (CS_OCP)	DAzP > V_{CSOCP} (z=1,2,3)	CSOCP_MODE=00b	nFAULT=Lo (Latched) Status bit=1b	Blocked / Depends on PDMODE	All reg. except LDO1 enable	DAzP < V_{CSOCP} (z=1,2,3) & clear latch	RA4
		CSOCP_MODE=01b				Retry after tRETRY_OCP	RA5
		CSOCP_MODE=10b		Active	Keep state in original mode	DAzP < V_{CSOCP} (z=1,2,3) & clear latch	RA6
		CSOCP_MODE=11b	None			None	No action is necessary.
MOSFET V_{GS} fault (VGS_FAULT)	V_{GS} stuck > t_{GT}	DIS_VGSFLT=0b	nFAULT=Lo (Latched) Status bit=1b	Blocked / Hi-Z pulldown	All reg. except LDO1 enable	V_{GS} stuck < t_{GT} & clear latch	RA4
		DIS_VGSFLT=1b	None	Active	Keep state in original mode	None	No action is necessary.
Thermal warning (TWARN)	Tj > T_{WARN} (140°C)	TWARN_REP=0b	only Status bit=1b	Active	Keep state in original mode	Tj < T_{WARN} - T_{HYS} (125°C)	Status bit is reset, nFAULT pin is released high automatically.
		TWARN_REP=1b	nFAULT=Lo Status bit=1b				
Thermal shutdown (OTSD)	Tj > T_{SD} (160°C)	DIS_OTSD=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-Z pulldown	LDO1, LDO3 enable, others disable	Tj < T_{SD} - T_{HYS} (145°C)	RA2
		DIS_OTSD=1b		Active	Keep state in original mode		RA0
Buck reg. overcurrent limit (SR_OC1)	IL peak > I_{OC1_SR} (1.2A)	DIS_SROC=0b	None	Active	Keep state in original mode, buck reg. w/ SR_OC	None	No action is necessary.
		DIS_SROC=1b			Keep state in original mode, buck reg. w/o SR_OC		
Buck reg. overcurrent protection (SR_OCP)	IL peak > I_{OC2_SR} (1.4A)	DIS_SROC=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-Z pulldown	LDO1, LDO3 enable, buck reg. hiccup mode	Buck reg. soft-start done & IL peak < I_{OC2_SR}	RA2
		DIS_SROC=1b		Active	Keep state in original mode, buck reg. w/o SR_OC	IL peak < I_{OC2_SR}	RA0
VDRV undervoltage (VDRV_UV)	VDRV < V_{DRVUV} (4.2V)	DIS_VDRVUV=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-z pulldown	LDO1, LDO3 enable, buck reg. hiccup mode	Buck reg. soft-start done & VDRV > V_{DRVUVR} (4.4V)	RA2
		DIS_VDRVUV=1b		Active	Keep state in original mode	VDRV > V_{DRVUVR} (4.4V)	RA0
VDRV overvoltage (VDRV_OV)	FB > V_{DRVOV} (0.95V)	DIS_VDRVOV=0b	nFAULT=Lo Status bit=1b	Blocked / Hi-Z pulldown	LDO2, LDO3 enable, charge pump, buck reg. Hi-Z	FB < V_{DRVOVR} (0.825V)	RA3
		DIS_VDRVOV=1b		Active	Keep state in original mode		RA0

6.2 Fault Management (continued)

Table 6.2-2 Fault Recovery Actions

Recovery action	Case	Release/Reset timing				
		Status register reset	nFAULT release	Device mode moves to Operating Mode	Gate driver Input / Output	Regulators soft-start
RA0	Only report status	EN low pulse rising edge or CLR_FLT	Fault condition clears.	Keep state in original mode		
RA1	VM_UV	EN low pulse rising edge or CLR_FLT	Fault condition clears.			
RA2	VM_OV, OTSD, SR_OCP, VDRV_UV	EN low pulse rising edge or CLR_FLT	Fault condition clears.		EN low pulse rising edge or CLR_FLT	Fault condition clears.
RA3	VCP_UV, VDRV_OV	EN low pulse rising edge or CLR_FLT	Fault condition clears.		EN low pulse rising edge or CLR_FLT	Keep enable *Note1
RA4	VDS_OCP, CS_OCP with OCP_MODE= 00b, or VGS_FAULT	EN low pulse rising edge or CLR_FLT				Keep enable
RA5	VDS_OCP, CS_OCP with OCP_MODE= 01b	Expiration of Retry delay time				Keep enable
RA6	VDS_OCP, CS_OCP with OCP_MODE= 10b	EN low pulse rising edge or CLR_FLT		Keep state in original mode		

Note1: Charge pump and buck regulator release Hi-Z after the recovery from VDRV_OV.

6.2.1 Fault Indicator nFAULT

The nFAULT pin (open-drain configuration) is the fault indicator. It is pulled low if any of the fault conditions occur. It is pulled high when all the fault conditions are removed and all chip power rail start-ups are done. The nFAULT is latched only for MOSFET V_{DS} overcurrent, Current sense overcurrent, and MOSFET V_{GS} fault. Toggling the EN signal or setting CLR_FLT=1b in IC Control 1 register pulls the nFAULT high (if the fault is removed).

This signal notifies the MCU after any fault occurs, so the MCU can stop normal operation and enter the fault handling routine. It also informs the MCU when all fault conditions are removed and all necessary power rails are properly up, so the MCU can re-enter the normal operating routine.

6.2.2 Fault Condition Types

6.2.2.1 VCC Undervoltage (VCC_UV)

If the VCC pin voltage falls lower than the V_{CCUV} threshold at any time, the device enters Shutdown Mode, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), and most of internal function blocks are disabled except for LDO1 and LDO3. Normal operation starts again (the device enters Sleep Mode) when the VCC undervoltage condition is removed (V_{CCUVR}).

6.2.2.2 VM Undervoltage (VM_UV)

If the input supply voltage on the VM pin falls lower than the V_{VMUV} threshold at any time with DIS_VMUV=0b, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), LDO1 and LDO3 are enabled, other regulators (buck, charge pump, LDO2) are disabled, and the nFAULT pin is pulled low. The FAULT bit and VM_UV bit are also latched high in the Fault Status registers.

The nFAULT pin is released automatically and normal operation starts again (gate driver, buck, charge pump, LDO2, and LDO3 operation) when the VM undervoltage condition is removed (V_{VMUVR}). The FAULT bit and VM_UV bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

This fault detection can be disabled by setting DIS_VMUV=1b in the Fault Control registers. If the VM undervoltage condition occurs with DIS_VMUV=1b, the device keeps state in original mode but the nFAULT pin is pulled low until the VM undervoltage condition is removed (V_{VMUVR}). The FAULT bit and VM_UV bit are set high in the Fault Status registers until cleared by the CLR_FLT bit or an EN pin low pulse.

6.2.2.3 VM Overvoltage (VM_OV)

If the input supply voltage on the VM pin rises higher than the V_{VMOV} threshold at any time with DIS_VMOV=0b, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), LDO1 and LDO3 are enabled, other regulators (buck, charge pump, LDO2) are disabled, and the nFAULT pin is pulled low. The FAULT bit and VM_OV bit are also latched high in the Fault Status registers.

The nFAULT pin is released automatically and normal operation starts again (buck, charge pump, LDO2, and LDO3 operation) when the VM overvoltage condition is removed (V_{VMOVR}). The gate drivers are enabled and the FAULT bit and VM_OV bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

This fault detection can be disabled by setting DIS_VMOV=1b in the Fault Control registers. If the VM overvoltage condition occurs with DIS_VMOV=1b, the device keeps state in original mode but the nFAULT pin is pulled low until the VM overvoltage condition is removed (V_{VMOVR}). The FAULT bit and VM_OV bit are set high in the Fault Status registers until cleared by the CLR_FLT bit or an EN pin low pulse.

6.2.2.4 VCP Undervoltage (VCP_UV)

If the charge pump voltage on the VCP pin falls lower than the V_{CPUV} threshold at any time with DIS_VCPUV=0b, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), the nFAULT pin is pulled low, the buck regulator power chain (buck, charge pump, LDO2) and LDO3 keep enabled (LDO1 disabled). The FAULT bit and VCP_UV bit are also latched high in the Fault Status registers.
The nFAULT pin is released automatically when the VCP undervoltage condition is removed (V_{CPUV} + $V_{CPUVHYS}$). The gate drivers are enabled and the FAULT bit and VCP_UV bit are reset after setting the CLR_FLT bit or an EN pin low pulse.
This fault detection can be disabled by setting DIS_VCPUV=1b in the Fault Control registers. If the VCP undervoltage condition occurs with DIS_VCPUV=1b, the device keeps state in original mode but the nFAULT pin is pulled low until the VCP undervoltage condition is removed (V_{CPUV} + $V_{CPUVHYS}$). The FAULT bit and VCP_UV bit are set high in the Fault Status registers until cleared by the CLR_FLT bit or an EN pin low pulse.

6.2.2.5 MOSFET V_{DS} Overcurrent (VDS_OCP)

A MOSFET V_{DS} overcurrent is detected by monitoring the V_{DS} voltage drop across the external MOSFET r_{DS}(on). If the V_{DS} voltage across an enabled MOSFET exceeds the V_{DSOCP} threshold selected by the VDS_TH bits for longer than the t_{DEG_OCP} deglitch time selected by the DEG_TIME bits, the device judges VDS_OCP occurs and the fault actions are executed according to the VDSOCP_MODE bits in the Fault Control registers. The device has following 4 different response modes for VDS_OCP fault action.
Note: For low-side V_{DS} overcurrent, the voltage between HSx (x=A,B,C) and PGND is monitored, respectively. Therefore, this voltage includes the differential voltage across the shunt resistor.

Latch mode (VDSOCP_MODE=00b):
After a VDS_OCP in this mode, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs) or the outputs are pulled low according to the PDMODE bit, and the nFAULT pin is pulled low. The FAULT bit, VDS_OCP bit, and corresponding VDSyx_OCP (y=H,L, x=A,B,C) bit are latched high in the Fault Status registers.
The nFAULT pin is released, normal operation starts again, and the FAULT bit, VDS_OCP bit, and VDSyx_OCP (y=H,L, x=A,B,C) bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

Automatic Retry mode (VDSOCP_MODE=01b):
After a VDS_OCP in this mode, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs) or the outputs are pulled low according to the PDMODE bit and the nFAULT pin is pulled low. The FAULT bit, VDS_OCP bit, and corresponding VDSyx_OCP (y=H,L, x=A,B,C) bit are latched high in the Fault Status registers.
The nFAULT pin is released and normal operation starts again automatically after the t_{RETRY_OCP} time passes. The FAULT bit, VDS_OCP bit, and VDSyx_OCP (y=H,L, x=A,B,C) bit keep latched until the retry starts.

Report Only mode (VDSOCP_MODE=10b):
No action takes place (the gate drivers keep active) after a VDS_OCP in this mode. The nFAULT pin is pulled low and the FAULT bit, VDS_OCP bit, and corresponding VDSyx_OCP (y=H,L, x=A,B,C) bit are latched high in the Fault Status registers.
The controller handles the VDS_OCP appropriately by controlling INz (z=1,2,3,4,5,6) signals or EN signal. The nFAULT pin is released and the FAULT bit, VDS_OCP bit, and VDSyx_OCP (y=H,L, x=A,B,C) bit are reset when the VDS_OCP condition is removed and the CLR_FLT bit or an EN pin low pulse is set.

Disable mode (VDSOCP_MODE=11b):
No action and no report take place in this mode. The gate drivers remain active, and the nFAULT pin and the fault status bits remain in original mode.

6.2.2.6 Current Sense Overcurrent (CS_OCP)

Current sense overcurrent is detected by monitoring the voltage drop across the external current sense resistor. If the differential voltage between DazP and DazN (z=1,2,3) exceeds the V_{CSOCP} threshold selected by the CSOCP_TH bits for longer than the t_{DEG_OCP} deglitch time selected by the DEG_TIME bits, the device judges CS_OCP occurs and the fault actions are executed according to the CSOCP_MODE bits. The device has following 4 different response modes for CS_OCP fault action. CSz_OCP (z=1,2,3) bits corresponding to each overcurrent can be disabled independently by setting DIS_CSzOCP (z=1,2,3) bits in the Fault Control registers to high. No action and no report take place in the case of DIS_CSzOCP (z=1,2,3)=1. If some differential amplifiers or DazP, DazN (z=1,2,3) pins are not used for shunt current sensing, corresponding DIS_CSzOCP (z=1,2,3) bits should be set to high.

Latch mode (CSOCP_MODE=00b):

After a CS_OCP in this mode, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs) or the outputs are pulled low according to the PDMODE bit and the nFAULT pin is pulled low. The FAULT bit, CS_OCP bit, and corresponding CSz_OCP (z=1,2,3) bit are latched high in the Fault Status registers.
The nFAULT pin is released, normal operation starts again, and the FAULT bit, CS_OCP bit, and corresponding CSz_OCP (z=1,2,3) bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

Automatic Retry mode (CSOCP_MODE=01b):

After a CS_OCP in this mode, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs) or the outputs are pulled low according to the PDMODE bit and the nFAULT pin is pulled low. The FAULT bit, CS_OCP bit, and corresponding CSz_OCP (z=1,2,3) bit are latched high in the Fault Status registers.
The nFAULT pin is released and normal operation starts again automatically after the t_{RETRY_OCP} time passes. The FAULT bit, CS_OCP bit, and corresponding CSz_OCP (z=1,2,3) bit keep latched until the retry starts.

Report Only mode (CSOCP_MODE=10b):

No action takes place (the gate drivers keep active) after a CS_OCP in this mode. The nFAULT pin is pulled low and the FAULT bit, CS_OCP bit, and corresponding CSz_OCP (z=1,2,3) bit are latched high in the Fault Status registers.
The controller handles the CS_OCP appropriately by controlling INz (z=1,2,3,4,5,6) signals or EN signal.
The nFAULT pin is released and the FAULT bit, CS_OCP bit, and corresponding CSz_OCP (z=1,2,3) bit are reset when the CS_OCP condition is removed and the CLR_FLT bit or an EN pin low pulse is set.

Disable mode (CSOCP_MODE=11b):

No action and no report take place in this mode. The gate drivers keep active, the nFAULT pin and the fault status bits keep state in original mode.

6.2.2.7 MOSFET V_{GS} Fault (VGS_FAULT)

MOSFET V_{GS} fault is detected by monitoring the gate-source voltage V_{GS} of the external MOSFET after the maximum gate transition time (t_{GT}). If the V_{GS} does not rise (over 3V typical) or drop (below 1V typical) by the abnormality of HOx, or LOx pins (shorted to other pins), or the inappropriate settings of ISRC_HS, ISRC_LS, and T_GT bits, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs) and the nFAULT pin is pulled low. The FAULT bit, VGS_FAULT bit, and corresponding VGSyx_FAULT (y=H,L, x=A,B,C) bit are latched high in the Fault Status registers.
The nFAULT pin is released, normal operation starts again, and the FAULT bit, VGS_FAULT bit, and VGSyx_FAULT (y=H,L, x=A,B,C) bits are reset after setting the CLR_FLT bit or an EN pin low pulse.
This fault detection can be disabled by setting DIS_VGSFLT=1b in the Fault Control registers. If the VGS_FAULT condition occurs with DIS_VGSFLT=1b, no action and no report take place. The gate drivers keep active, the nFAULT pin and the fault status bits keep state in original mode.

6.2.2.8 Thermal Warning (TWARN)

If the die temperature exceeds the trip point of the thermal warning temperature (T_{WARN}), TWARN bit is set high in the Fault Status register. The device keeps state in original mode. When the die temperature falls lower than the recovery point of the thermal warning ($T_{WARN} – T_{HYS}$), TWARN bit is cleared automatically. TWARN bit can be output to the nFAULT pin by setting TWARN_REP bit=1b through the SPI interface.

6.2.2.9 Thermal Shutdown (OTSD)

If the die temperature exceeds the trip point of the thermal shutdown temperature (T_{SD}) with DIS_OTSD=0b, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), LDO1 and LDO3 are enabled, other regulators (buck, charge pump, LDO2) are disabled, and the nFAULT pin is pulled low. The FAULT bit and OTSD bit are also latched high in the Fault Status registers.

The nFAULT pin is released automatically and normal operation starts again (buck, charge pump, LDO2, and LDO3 operation) when the die temperature falls lower than the recovery point of the thermal shutdown ($T_{SD} – T_{HYS}$). The gate drivers are enabled, and the FAULT bit and OTSD bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

The thermal shutdown can be disabled by setting DIS_OTSD=1b in the Fault Control registers. If the OTSD condition occurs with DIS_OTSD=1b, the device keeps state in original mode but the nFAULT pin is pulled low until the OTSD condition is removed ($T_{SD} – T_{HYS}$). The FAULT bit and OTSD bit are set high in the Fault Status registers until cleared by the CLR_FLT bit or an EN pin low pulse. The controller handles the OTSD appropriately by controlling INz (z=1,2,3,4,5,6) signals or EN signal to avoid the high die temperature.

6.2.2.10 Buck Regulator Overcurrent Limiting (SR_OC1)

The overcurrent function of the buck regulator protects against any overload condition and output short at worst case by monitoring the current flowing through the high-side MOSFET. The device has two overcurrent function. The overcurrent function SR_OC1 limits the high-side MOSFET peak current cycle-by-cycle. No action and no report take place by SR_OC1 event except for the buck regulator. The gate drivers keep active, the nFAULT pin and the fault status bits keep state in original mode.

6.2.2.11 Buck Regulator Overcurrent Protection (SR_OCP)

The second overcurrent function SR_OCP has the higher overcurrent threshold I_{OC2_SR} than I_{OC1_SR}. If the high-side MOSFET current reaches I_{OC2_SR}, the PWM shut off after two-cycle delay and the buck regulator enters Hiccup mode. In Hiccup mode, the PWM is disabled for a dummy cycle (63ms). After this dummy cycle, the true soft-start cycle is attempted again.

When SR_OCP occurs, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), LDO1 and LDO3 are enabled, charge pump and LDO2 are disabled, and the nFAULT pin is pulled low. The FAULT bit, SR_FAULT bit, and SR_OCP bit are also latched high in the Fault Status registers.

The nFAULT pin is released automatically when the overcurrent condition is removed (I_{OC2_SR}) and soft-start of power rails is done. The gate drivers are enabled and the FAULT bit, SR_FAULT bit, and SR_OCP bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

6.2.2.12 Buck Regulator VDRV Undervoltage (VDRV_UV)

If the VDRV pin voltage of the buck regulator output falls lower than the V_{DRVUV} threshold, the buck regulator enters Hiccup mode. The gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), LDO1 and LDO3 are enabled, charge pump and LDO2 are disabled, and the nFAULT pin is pulled low. The FAULT bit, SR_FAULT bit, and VDRV_UV bit are also latched high in the Fault Status registers.

The nFAULT pin is released automatically when the undervoltage condition is removed (V_{DRVUVR}) and soft-start of power rails is done. The gate drivers are enabled and the FAULT bit, SR_FAULT bit, and VDRV_UV bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

6.2.2.13 Buck Regulator VDRV Overvoltage (VDRV_OV)

If the VDRV pin voltage of the buck regulator output rises higher than the V_{DRVOV} threshold by FB pin voltage, the gate drivers are disabled (blocked inputs, Hi-Z pulldown outputs), LDO2 and LDO3 keep enable, charge pump and buck regulator stop switching until the fault condition is removed, and the nFAULT pin is pulled low. The FAULT bit, SR_FAULT bit, and VDRV_OV bit are also latched high in the Fault Status registers.

The nFAULT pin is released automatically and normal operation starts again (buck, charge pump) when the overvoltage condition is removed (V_{DRVOVR}). The gate drivers are enabled and the FAULT bit, SR_FAULT bit, and VDRV_OV bit are reset after setting the CLR_FLT bit or an EN pin low pulse.

6.3 Power Architecture

6.3.1 Block Diagram and On/Off Table of Power Supply Blocks

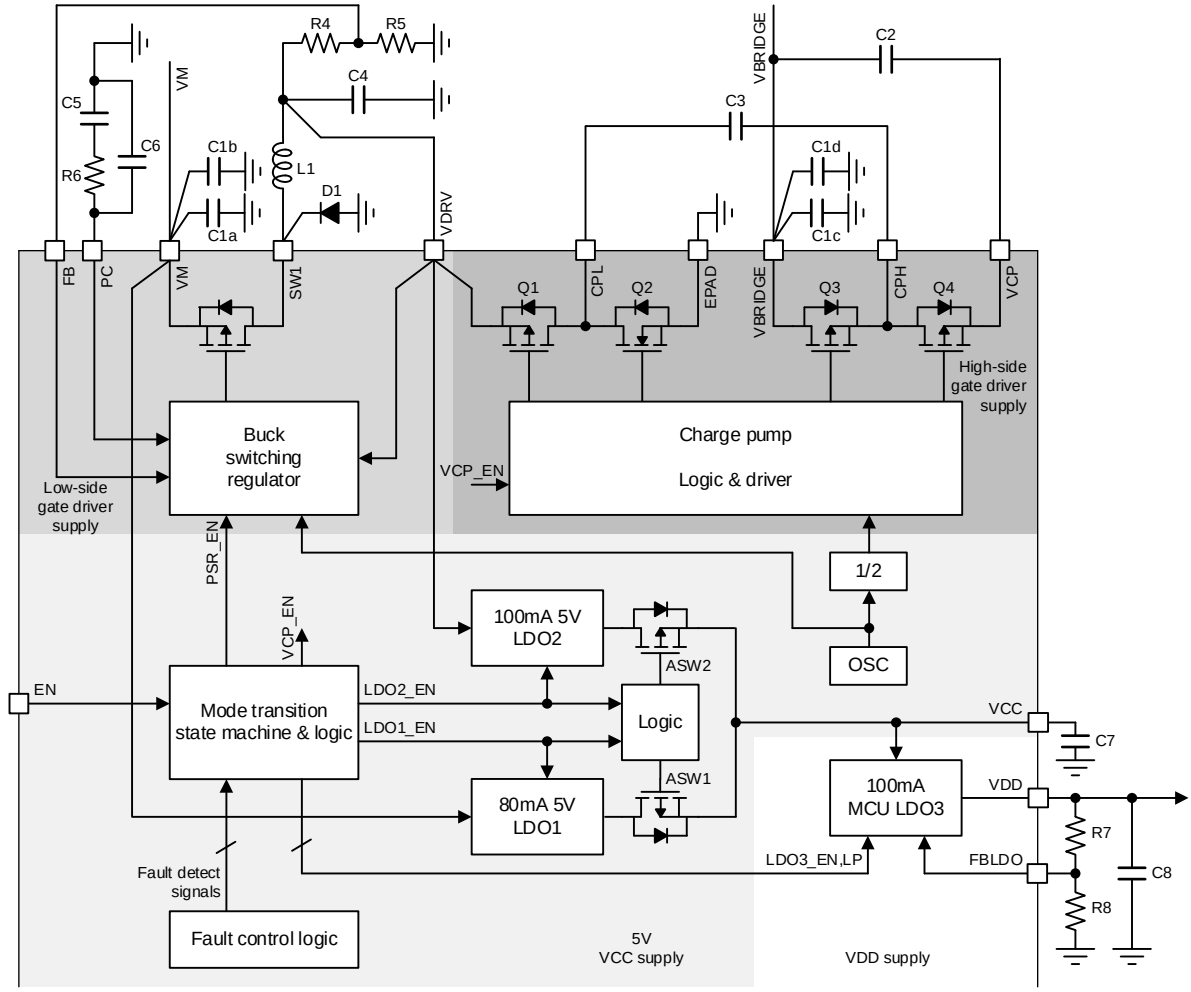


Figure 6.3-1 Block Diagram of Power Supply Blocks

Table 6.3-1 On/Off Table of Power Supply Blocks

Mode	80mA 5V LDO1	100mA MCU LDO3	100mA 5V LDO2	Buck switching regulator	Charge pump
Shutdown	On	On Note1	Off		
Sleep	On	On Note1	Off		
Operating	Off	On	On		
Fault Management	Reaction based on fault management matrix				

Note1: LDO3 enters a low-power mode when the EN pin is pulled low.

6.3.2 VCC Supply

5V VCC supply consists of two LDOs and two analog switches for LDO output selection.

5V LDO1 for EN=Lo:

This LDO is a high voltage LDO fed from VM. It is enabled to generate VCC power upon device power-up (VCC reaches above rising POR) until the buck regulator soft-start is completed. Whenever LDO1 is enabled, ASW1 is also turned on.

Note: the bandgap reference used in LDO1 is untrimmed.

5V LDO2 for EN=Hi:

This LDO is a low voltage LDO fed from the buck regulator output. It supplies VCC in Operating Mode after the buck regulator soft-start is completed. ASW2 is on when LDO2 is used.

Note: the bandgap reference used in LDO2 is trimmed, to achieve better VCC accuracy in Operating Mode.

6.3.3 VDD Supply

VDD LDO is a low voltage LDO3 fed from VCC, designed to power the MCU and peripherals in applications as required. It has a dedicated feedback pin (FBLDO) that allows for fine adjustment of output voltage within the recommended operating condition. The internal reference is 1.2V. The output voltage is tightly regulated during normal Operating Mode. When the EN pin is pulled low, LDO3 enters a low-power mode with not-so-tight regulation.

6.3.4 Low-Side Gate Driver Supply (VDRV)

Low-side gate driver supply (VDRV) is generated by a 500mA buck switching regulator fed from VM. VDRV adjustable range is from 5V to 15V. The buck regulator integrates a 1.0Ω high voltage (65V) PMOS and the corresponding gate driver. It also integrates all the control circuitry and logic to achieve peak current mode control scheme. The freewheeling diode and inductor need to be placed externally. Regulator switching frequency is 500kHz, with two-level peak current limit at 1.2A (cycle-by-cycle current limit) and 1.4A (peak OCP threshold). Regulator output is monitored and protected from OV and UV conditions. Under medium or high load conditions, the regulator runs in continuous current mode (CCM). However, under light-load conditions, it can run in discontinuous current mode (DCM) due to the nature of asynchronous rectification. Moreover, in case of high VM low VDRV operation in light load, it can run in pulse skipping mode due to the minimum on-time limitation.

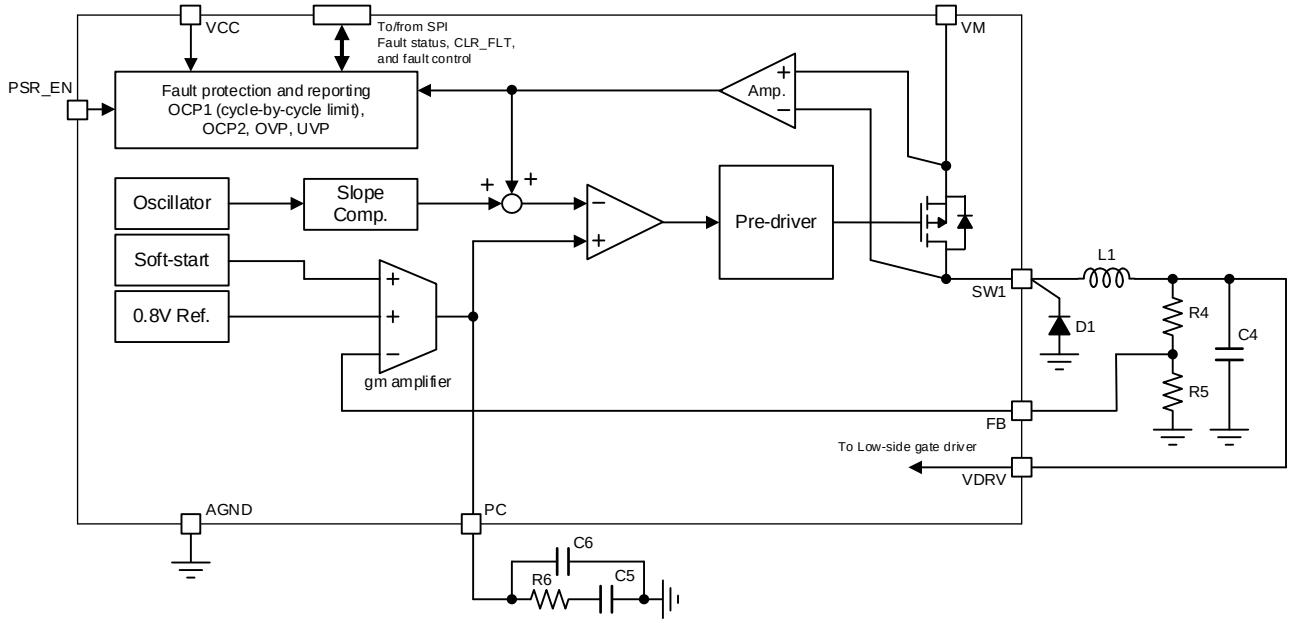


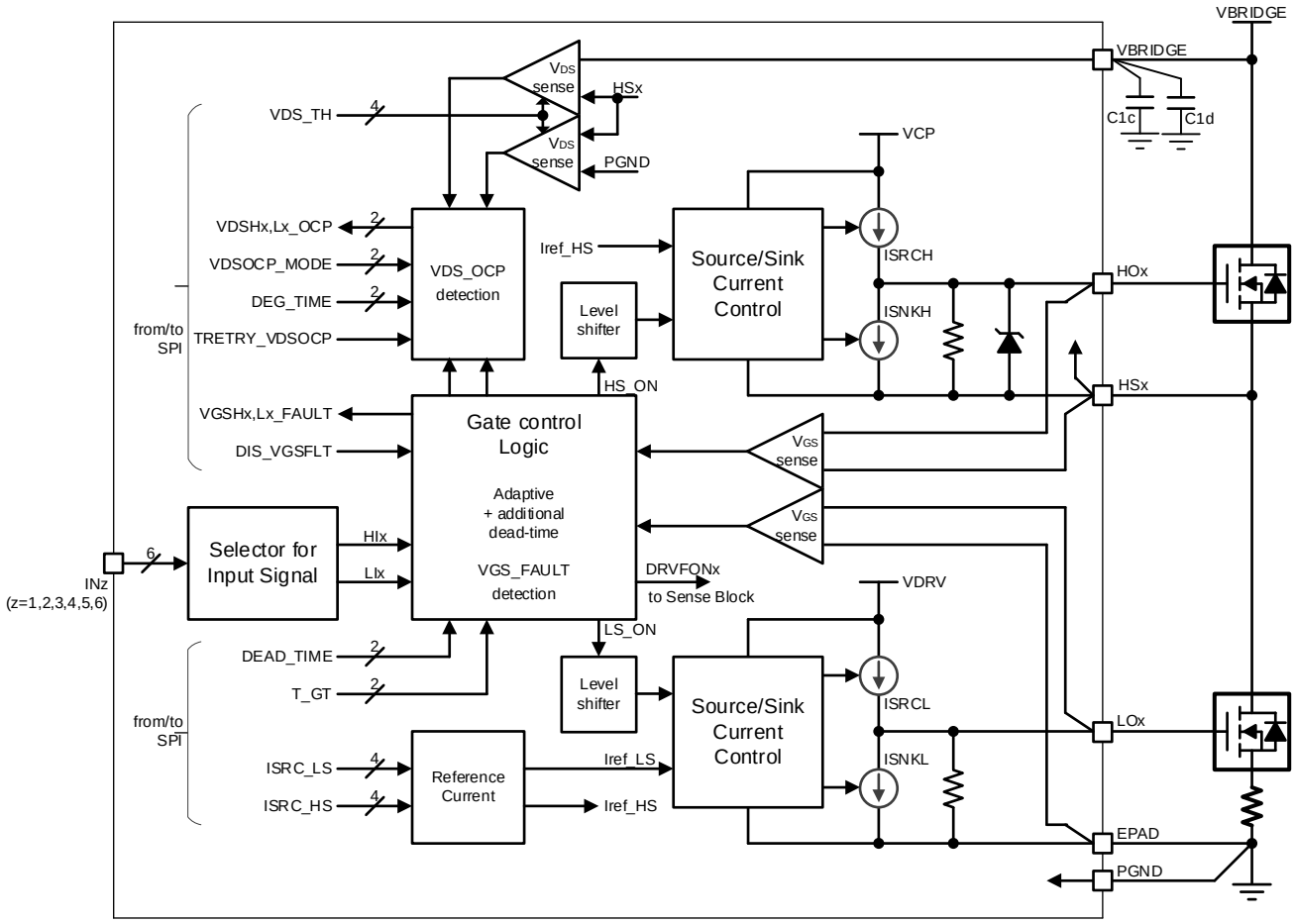
Figure 6.3-2 Buck Switching Regulator Block Diagram

6.3.5 High-Side Gate Driver Supply (VCP)

This high voltage charge pump is to generate a steady high-side gate driver supply rail at the level VBRIDGE+VDRV. The CPL pin switches between VDRV and EPAD (Ground) by complementary switches Q1 and Q2. The CPH pin switches between VBRIDGE and VCP (VBRIDGE+VDRV) by complementary switches Q3 and Q4. Q1 and Q4 are turned off and on at the same time, while Q2 and Q3 are turned on and off at the same time. In this way, during on-time of Q2 and Q3 (off-time of Q1 and Q4), the flying capacitor C3 across CPH and CPL gets charged by VBRIDGE. During on-time of Q1 and Q4 (off-time of Q2 and Q3), C2 gets charged up towards VBRIDGE+VDRV by VDRV. From its operation it can be seen that this charge pump always runs in full mode with minimal power dissipation. The complementary switches operate at 250kHz, which is 1/2 the internal oscillator frequency used by the buck regulator, and the duty cycle is 50%. The maximum loading target is 28mA. The charge pump output is monitored and undervoltage protection is implemented.

6.4 Gate Driver

6.4.1 Block Diagram



Note1: “x” in this figure stands for A,B, and C.

Figure 6.4-1 Block Diagram of Gate Driver

6.4.2 Gate Driver Control Modes

When the device is put into Operating Mode, the gate driver sets its output state based on the control signal present on HIx and LIx (x=A,B,C) signals which is selected from INz (z=1,2,3,4,5,6) signals according to HOx_SEL and LOx_SEL (x=A,B,C) bits in the Gate Driver Input Selection registers. Two gate driver control modes are available:

- Three-Phase HI/LI Mode
- Three-Phase PWM Mode

Detailed descriptions and logic truth tables are listed in the following Three-Phase HI/LI Mode and Three-Phase PWM Mode sections.

6.4.2.1 Three-Phase HI/LI Mode

This mode is enabled when PWMMODE bit=0b in the IC Control 1 register. In this mode, HIx and LIx (x=A,B,C) inputs serve as control inputs for each individual driver output, logic active high. For each phase, the HIx (x=A,B,C) input signal controls the high-side gate driver output HOx (x=A,B,C) directly, while the LIx (x=A,B,C) input signal controls the low-side gate driver output LOx (x=A,B,C) directly. See Table 6.4-1.

Table 6.4-1 Three-Phase HI/LI Mode Truth Table (x=A,B,C)

LIx	HIx	LOx	HOx - HSx	HSx
0	0	Low	Low	Hi-Z
0	1	Low	High	High
1	0	High	Low	Low
1	1	Low	Low	Hi-Z

6.4.2.2 Three-Phase PWM Mode

This mode is enabled when PWMMODE bit=1b in the IC Control 1 register. In this mode, LIx (x=A,B,C) serves as the enable (logic high)/disable (logic low) of the driver output of each bridge. HIx (x=A,B,C) serves as control input for each bridge. See Table 6.4-2.

Table 6.4-2 Three-Phase PWM Mode Truth Table (x=A,B,C)

LIx	HIx	LOx	HOx - HSx	HSx
0	0	Low	Low	Hi-Z
0	1	Low	Low	Hi-Z
1	0	High	Low	Low
1	1	Low	High	High

6.4.3 Adjustable Slew-Rate

The gate driver architecture allows for the accurate setting of the gate drive source (I_{SRC}) and sink current (I_{SNK}). It is helpful to more accurately control and adjust the slew-rate of switch node voltage, which is beneficial for radiated emission optimization, controlling the reverse recovery of the body diode and avoiding CdV/dt induced cross-conduction. For all gate driver outputs, 16 levels of sourcing/sinking current can be supported by ISRC_HS or ISRC_LS bits through the SPI interface. The configurable range is from 50mA to 640mA for sourcing and 100mA to 1280mA for sinking.

Note: The driver sink current is automatically set to be double the source current.

The maximum duration of driver peak source/sink current (maximum gate transition time (t_{GT})) can also be configured to ensure the MOSFET turns on fully. This maximum gate transition time (t_{GT}) can be configured to four levels of options (500ns, 1000ns, 2000ns, 4000ns) by T_GT bits through the SPI interface.

6.4.4 Gate Driver Robustness Enhancement

Strong sinking current to avoid CdV/dt induced cross-conduction

Additionally, within the same bridge phase, whenever one of the gate drivers is during gate transition of turning on or turning off the corresponding external MOSFET, the complementary gate driver performs a strong sinking current (I_{SNK_STG}) to avoid CdV/dt induced cross-conduction. The maximum duration of the strong sinking current is also equal to the maximum gate transition time (t_{GT}).

Active pullup/pulldown current to hold gate state

After the maximum gate transition time (t_{GT}), the driver actively imposes weaker current to hold the gate state. A pullup current (I_{SRC_PU}) is sourced out of the driver to maintain a high output voltage, whereas a pulldown sinking current (I_{SNK_PD}) is imposed to maintain nearly zero output voltage.

Adaptive dead time control plus configurable additional dead time

Adaptive dead time control is implemented by actively monitoring the gate of the MOSFET that is turning off first during the transition. The complementary MOSFET is allowed to start turning on only after it drops below the threshold (1V typical). In addition to adaptive dead time, you can add the extra dead time (t_{DT}) by setting DEAD_TIME bits through SPI interface.

Note: The t_{GT} value is defined as “maximum” gate transition time because in real applications, actual gate transition must be shorter than the t_{GT} setting in order to properly drive the MOSFETs. Therefore, the duration of driver peak source/sink current can get terminated before reaching the full t_{GT} duration. For example, during the turn off transition, after V_{GS} drops below the adaptive dead time threshold, peak sink current and the complementary MOSFET gate strong sinking current are terminated, which is earlier than the elapse of t_{GT}. For the turning-on transition, because there is no detection of V_{GS} reaching enough high level for the MOSFET, peak source current and complementary MOSFET gate strong sinking current sustain a full t_{GT} duration, assuming it is sufficiently long on time. Other instances involve the short on-time or off-time.

Adaptive dead time control disable function

Adaptive dead time control can be disabled by setting DIS_SADT=1b. In this case, the monitored result of the MOSFET gate voltage is ignored, and the complementary MOSFET is allowed to start turning on by the only the complementary HIx or LIx (x=A,B,C) input. After the complementary input changes to high, the extra dead time (t_{DT}) is started to avoid the shoot through current of the gate driver. After finishing the extra dead time (t_{DT}), the complementary MOSFET starts turning on. Refer to Figure 6.4-4.

Note: The inserted dead time should be shorter than the t_{GT} setting in order to properly drive the MOSFETs. The duration of driver peak source/sink current is terminated after reaching the full t_{GT} duration. Refer to Figure 6.4-5. And the inserted dead time must be longer than the MOSFET discharge time “tdchg” to avoid the shoot through current of the half-bridge MOSFET.

6.4.5 Gate Drive Timing Diagram in Three-Phase HI/LI Mode

Figure 6.4-2 shows the gate drive timing diagram if the HI/LI inserted dead time is relatively long (longer than maximum gate transition time (t_{GT}) plus extra dead time (t_{DT}) set). This case means:

- During high-side MOSFET turn-off/low-side MOSFET turn-on transition, if LIx (x=A,B,C) asserts high after the HOx-HSx (x=A,B,C) high-to-low transition time (t_{GT}) and extra dead time (t_{DT}) ends.
- During low-side MOSFET turn-off/high-side MOSFET turn-on transition, if HIx (x=A,B,C) asserts high after the LOx (x=A,B,C) high-to-low transition time (t_{GT}) and extra dead time (t_{DT}) ends.

Total effective dead time is adaptive dead time, plus extra dead time (t_{DT}) selected by DEAD_TIME bits, plus additional dead time introduced by the HIx/LIx (x=A,B,C) signals.

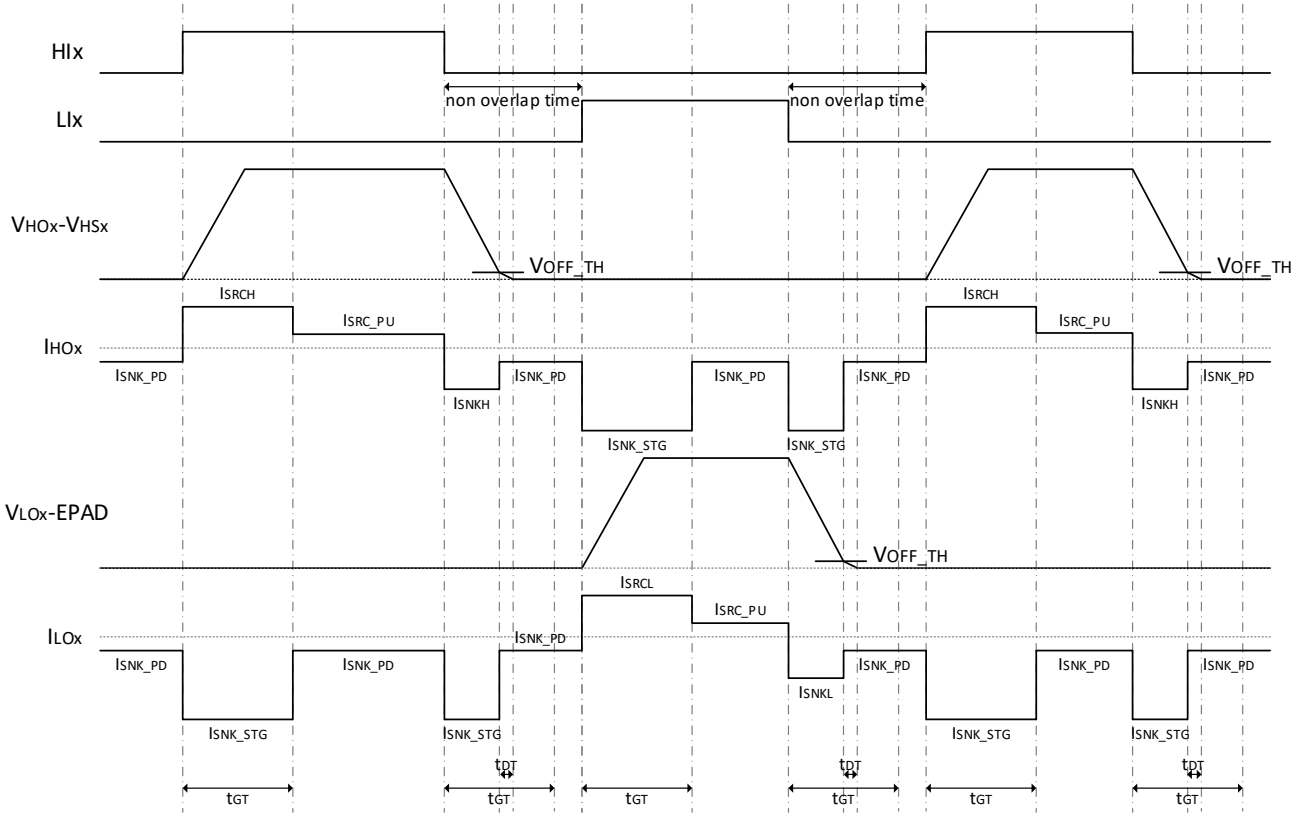


Figure 6.4-2 Gate Drive Timing Diagram in Three-Phase HI/LI Mode if the HI/LI Inserted Dead Time is Relatively Long

6.4.5 Gate Drive Timing Diagram in Three-Phase HI/LI Mode (continued)

Figure 6.4-3 shows the gate drive timing diagram if the HIx/LIx (x=A,B,C) inserted dead time is relatively short (shorter than maximum gate transition time (t_{GT}) plus extra dead time (t_{DT}) set). This case means:

- During high-side MOSFET turn-off/low-side MOSFET turn-on transition, if LIx (x=A,B,C) asserts high before the HOx-HSx (x=A,B,C) high-to-low transition time (t_{GT}) and extra dead time (t_{DT}) ends.
- During low-side MOSFET turn-off/high-side MOSFET turn-on transition, if HIx (x=A,B,C) asserts high before the LOx (x=A,B,C) high-to-low transition time (t_{GT}) and extra dead time (t_{DT}) ends.

Total effective dead time is adaptive dead time, plus extra dead time (t_{DT}) selected by DEAD_TIME bits.

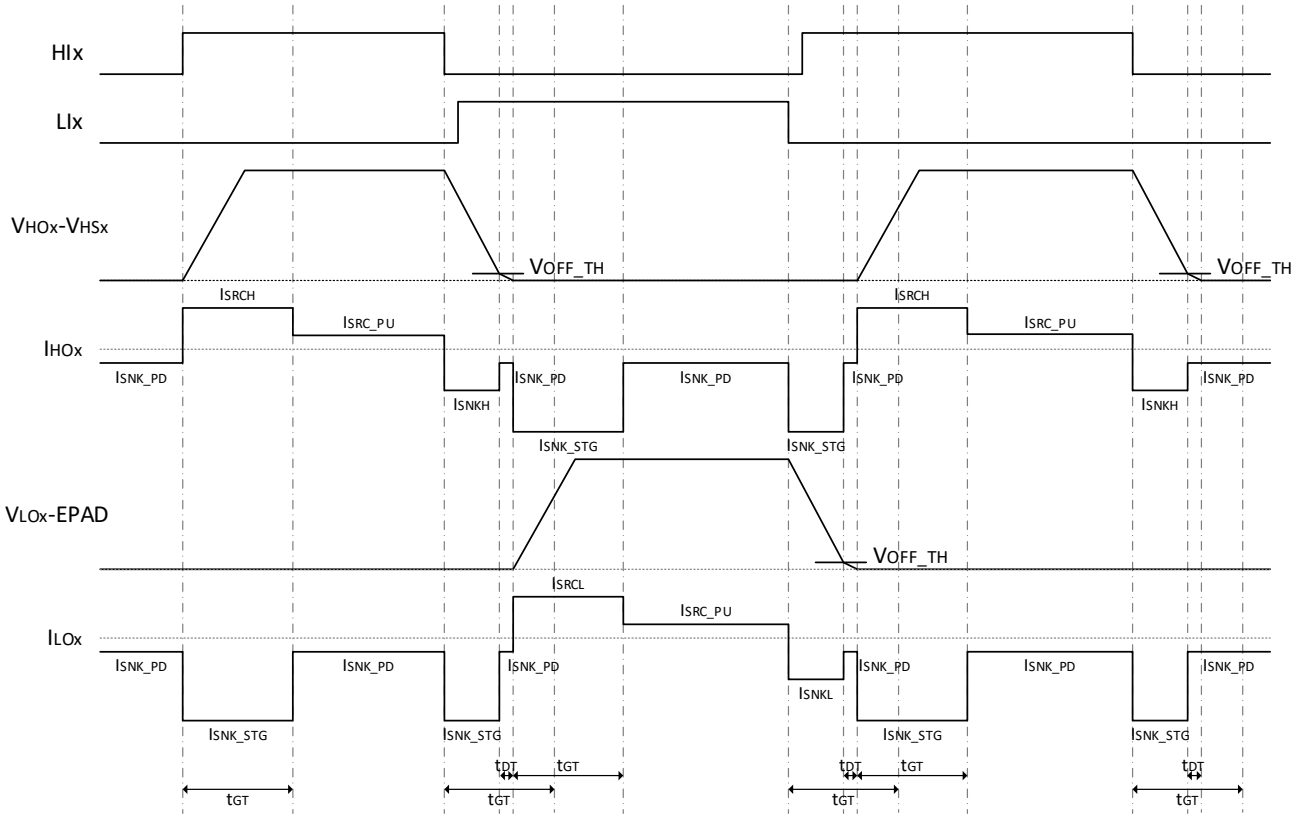


Figure 6.4-3 Gate Drive Timing Diagram in Three-Phase HI/LI Mode if the HI/LI Inserted Dead Time is Relatively Short

6.4.6 Gate Drive Timing Diagram with DIS_SADT=1b

Figure 6.4-4 shows the gate drive timing diagram with DIS_SADT=1b if the HIx/LIx (x=A,B,C) inserted dead time is shorter than maximum gate transition time (t_{GT}). This case means:

- During high-side MOSFET turn-off/low-side MOSFET turn-on transition, if LIx (x=A,B,C) asserts high after the high-side MOSFET turns off and before the HOx-HSx (x=A,B,C) high-to-low transition time (t_{GT}) ends.
- During low-side MOSFET turn-off/high-side MOSFET turn-on transition, if HIx (x=A,B,C) asserts high after the low-side MOSFET turns off and before the LOx (x=A,B,C) high-to-low transition time (t_{GT}) ends.

Total effective dead time is a non-overlap time, plus extra dead time (t_{DT}) selected by DEAD_TIME bits.

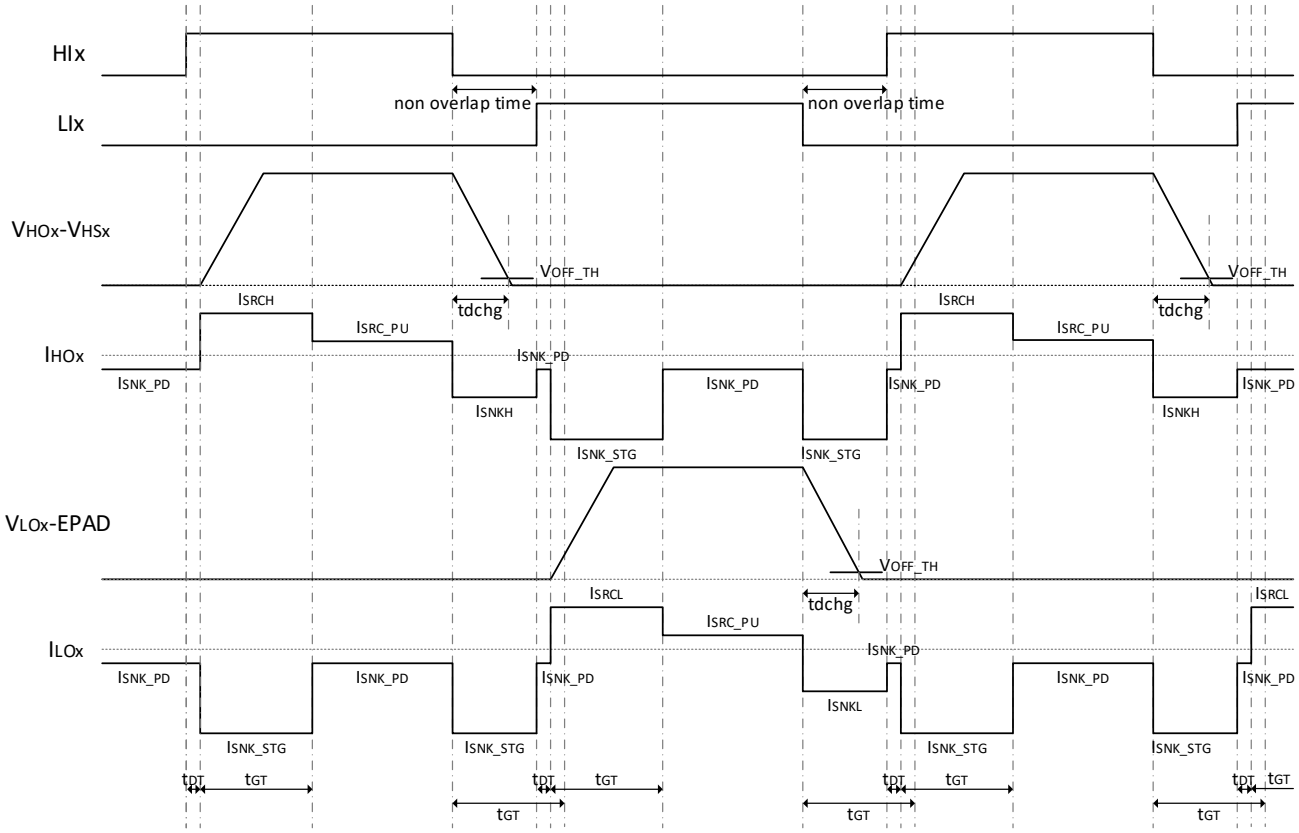


Figure 6.4-4 Gate Drive Timing Diagram with DIS_SADT=1b if the HI/LI Inserted Dead Time is Shorter than Maximum Gate Transition Time

6.4.6 Gate Drive Timing Diagram with DIS_SADT=1b (continued)

Figure 6.4-5 shows the gate drive timing diagram with DIS_SADT=1b if the HIx/LIx (x=A,B,C) inserted dead time is longer than maximum gate transition time (t_{GT}). This case means:

- During high-side MOSFET turn-off/low-side MOSFET turn-on transition, if LIx (x=A,B,C) asserts high after the HOx-HSx (x=A,B,C) high-to-low transition time (t_{GT}) ends.
- During low-side MOSFET turn-off/high-side MOSFET turn-on transition, if HIx (x=A,B,C) asserts high after the LOx (x=A,B,C) high-to-low transition time (t_{GT}) ends.

Total effective dead time is a non-overlap time, plus extra dead time (t_{DT}) selected by DEAD_TIME bits.

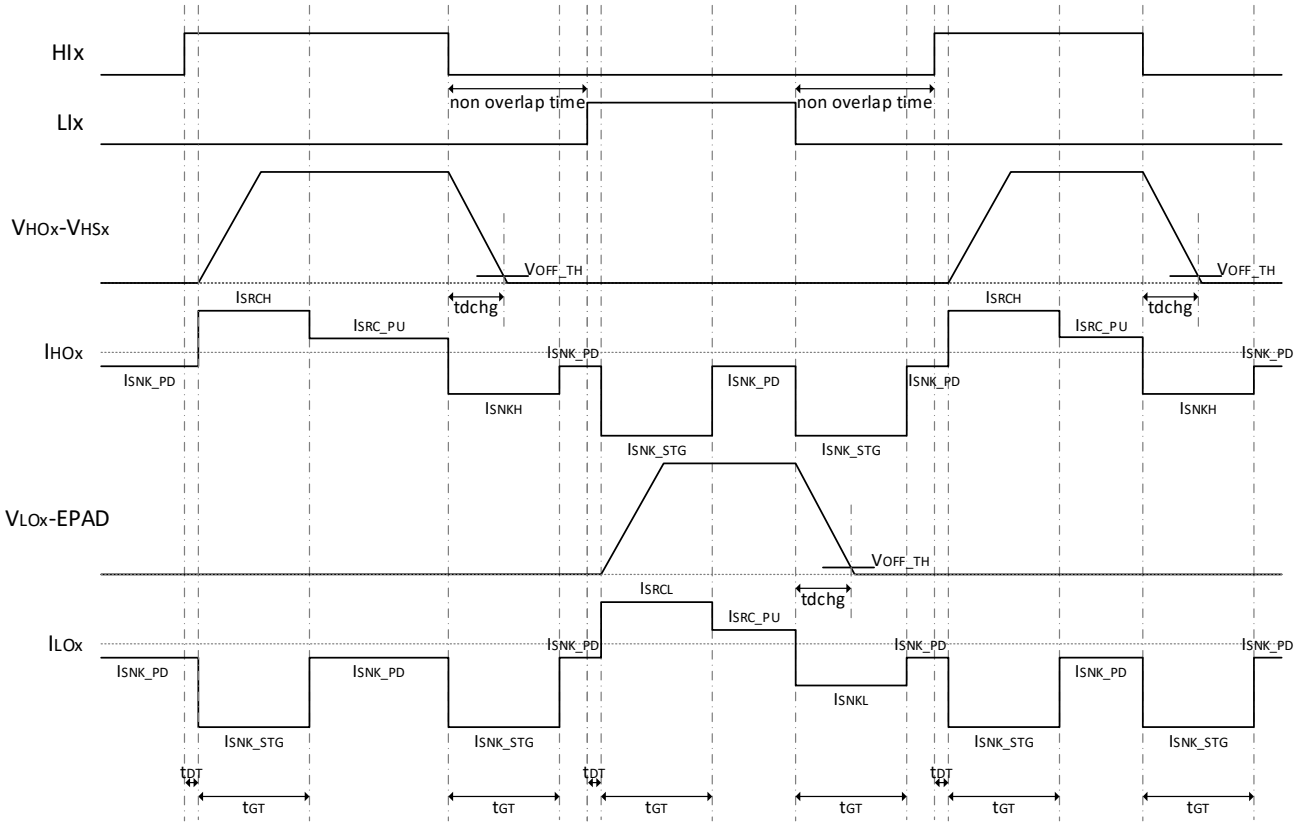


Figure 6.4-5 Gate Drive Timing Diagram with DIS_SADT=1b if the HI/LI Inserted Dead Time is Longer than Maximum Gate Transition Time

6.4.7 Gate Drive Timing Diagram in 3-Phase PWM Mode

Figure 6.4-6 shows the gate drive timing diagram when the gate driver control mode is Three-Phase PWM Mode. In this mode, the timing diagram is similar to 3-phase HI/LI Mode if the HIx/LIx (x=A,B,C) inserted dead time is zero.

Total effective dead time is adaptive dead time, plus extra dead time (t_{DT}) selected by DEAD_TIME bits.

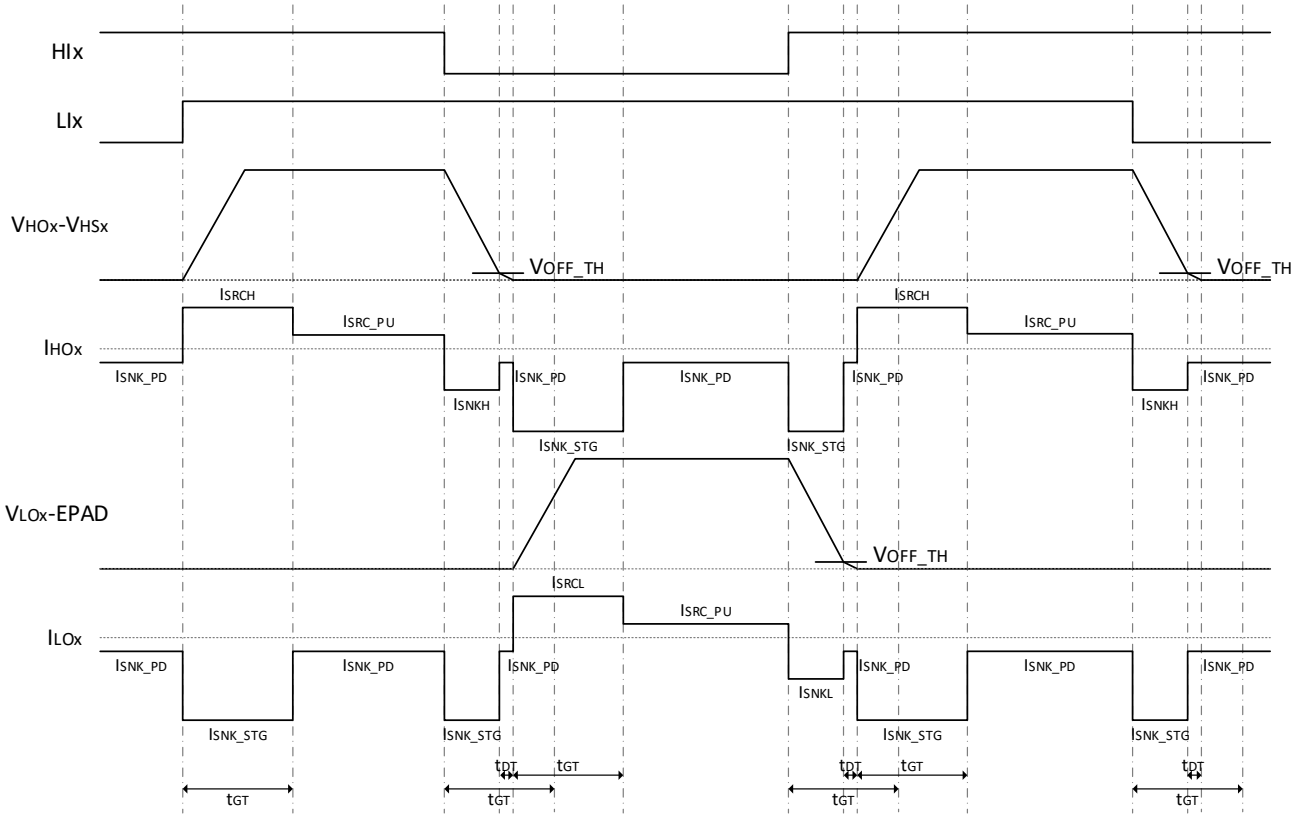


Figure 6.4-6 Gate Drive Timing Diagram in Three-Phase PWM Mode

6.5 Sense Block

6.5.1 Overview

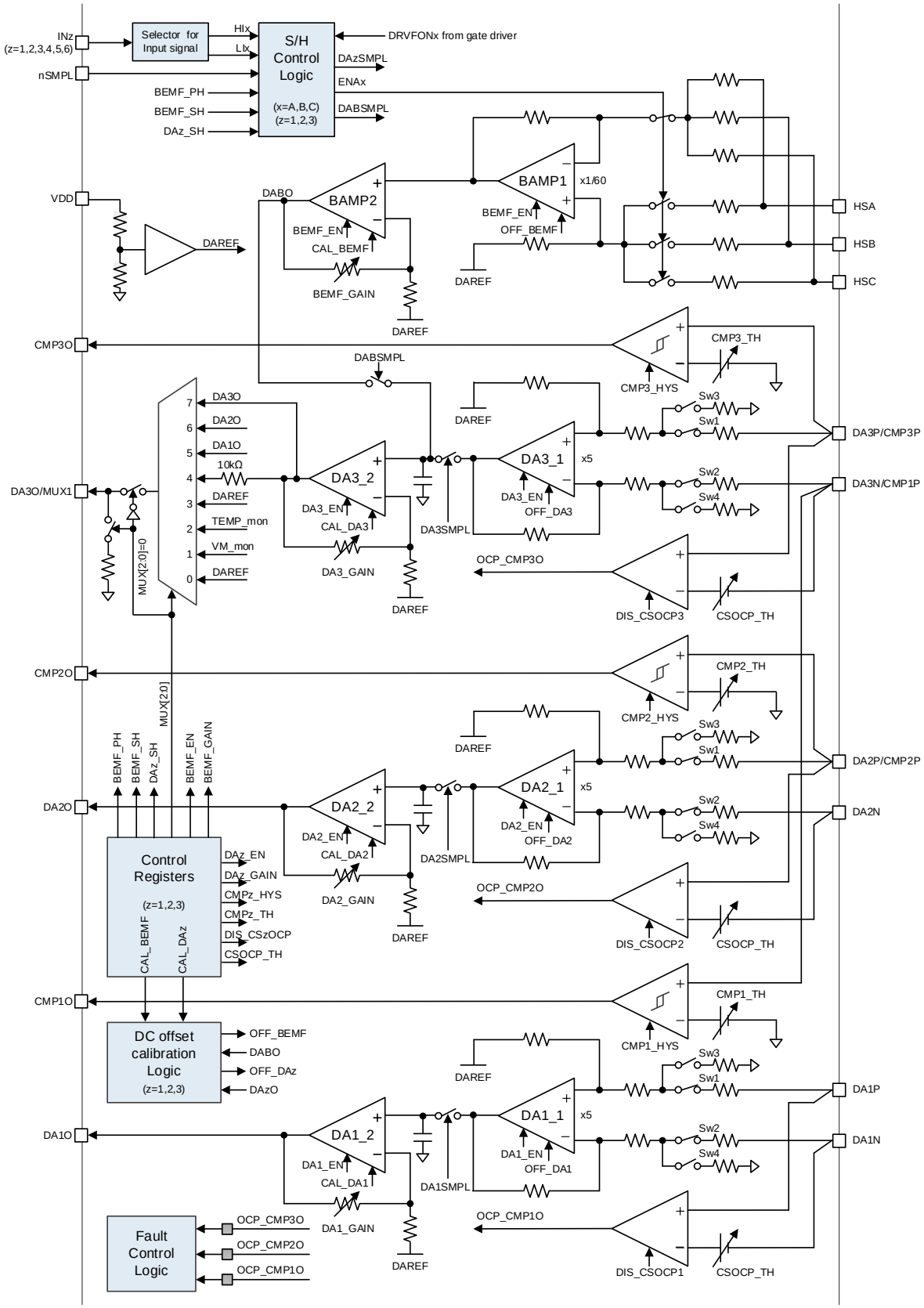


Figure 6.5-1 Block Diagram of Sense Block

6.5.1 Overview (continued)

Figure 6.5-1 shows the block diagram of sense block. The RAA306012 comprises three differential programmable gain amplifiers, three general-purpose comparators, three current sense overcurrent comparators, a BEMF sense amplifier, and an analog multiplexer. The three differential amplifiers can separately support current sensing up to three phases by using low-side shunt resistors in the external half-bridges. The general-purpose comparator has 0V to VDD input common mode range. The inputs of general-purpose comparators are DA2P, DA3P, and DA3N. The user can select these pins usage by changing the input signal only. The differential amplifiers and general-purpose comparators can be disabled individually by setting DAz_EN in the IC Control registers or CMPz_VTH (z=1,2,3) bits in the Sense Block Control registers to reduce the quiescent current. The current sense overcurrent comparator monitors the input of the differential amplifier continuously. The current sense overcurrent threshold V_{CSOCP} is selectable by the CSOCP_TH bits. Regarding BEMF (back electromotive force) sensing, two precise amplifiers form the BEMF sensing signal chain. The first differential amplifier outputs the voltage between high impedance phase and the virtual center tap. The high impedance phase is selected automatically at nSMPL falling edge, or selected by CMP1O/2O or CMP1O/3O inputs. The second programmable gain amplifier can adjust the output range suitably for the position detection by BEMF sensing. The output of BEMF sensing can be monitored through the analog multiplexer on DA3O/MUX1 pin. The analog multiplexer can output the VM voltage and die temperature by setting MUX bits in the Sense Block Control registers.

6.5.2 Differential Amplifiers for Current Sensing

The differential amplifier has the following functions.

Enable control:

Each differential amplifier can be disabled by DAz_EN (z=1,2,3) bit=0b according to the actual application.

Programmable gain:

The gain of differential amplifier is programmable by DAz_GAIN (z=1,2,3) bits. The gain settings are 5V/V, 10V/V, 20V/V, 40V/V.

Reference:

The output reference voltage of differential amplifier is $0.5*V_{DD}$.

Sample and hold (S/H) function:

When DAz_SH (z=1,2,3) bit is set “1”, S/H function of the differential amplifier is enabled individually. Three differential amplifier outputs are sampled during nSMPL signal=L simultaneously. This function helps 3 shunt current sensing. Refer to Figure 6.5-2.

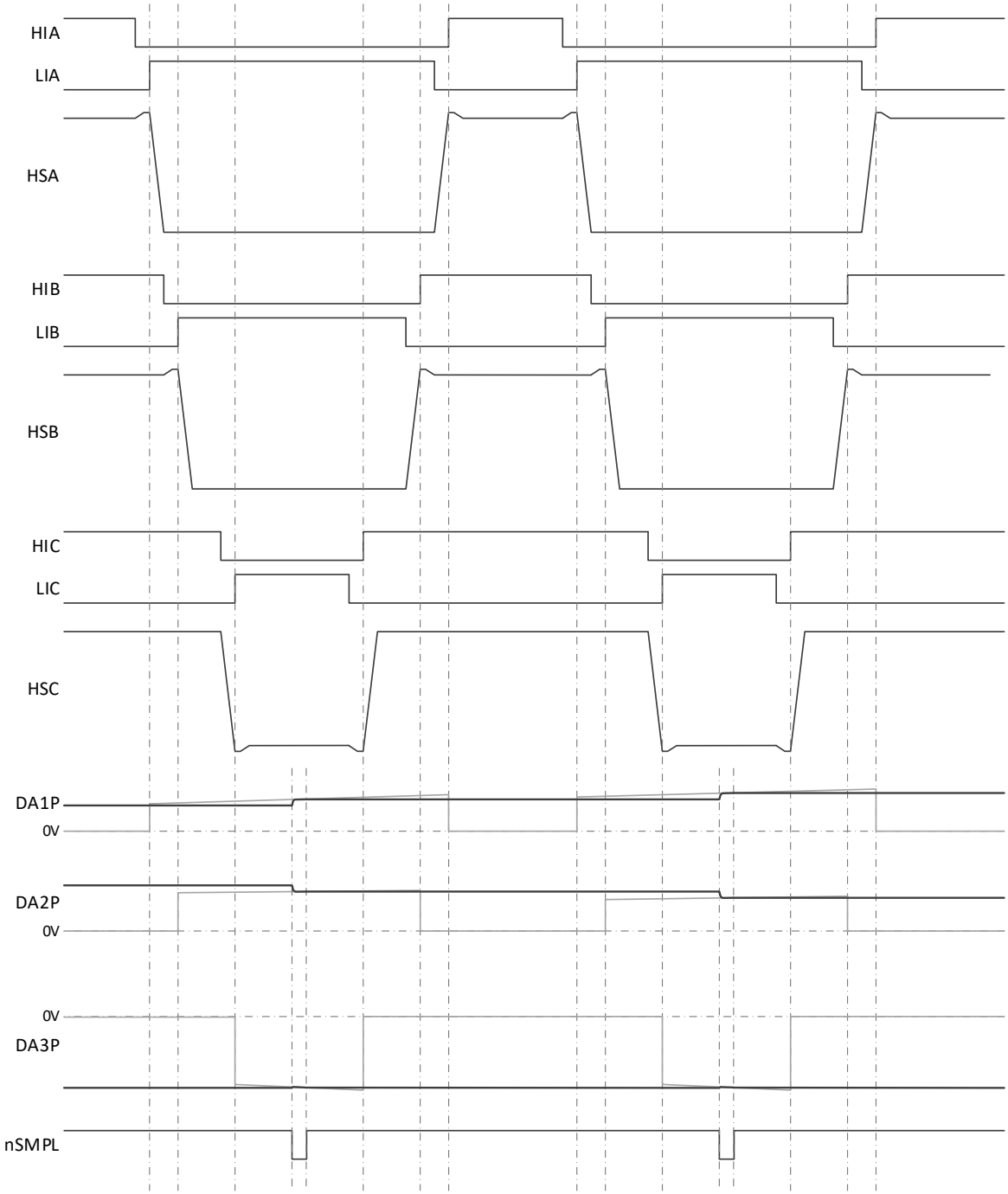
Note: The DA3O output is controlled by MUX[2:0] bits and BEMF_EN bit. When BEMF_EN bit is set to “1”, S/H switch of the differential amplifier 3 keeps turn off to avoid the conflict with BEMF sense amplifier output. When BEMF_EN bit is set to “0”, the output of the differential amplifier 3 can be monitored by MUX[2:0]=100b or 111b. Refer to section 6.5.5.

DC offset calibration:

DC offset calibration is automatically conducted upon device power up. It can also be initiated by setting CAL_DAz (z=1,2,3) bit in the Sense Block Control registers. When CAL_CONN bit is set to “0”, it is done by turning off Sw1 and Sw2, turning on Sw3 and Sw4 (connecting differential amplifier inputs to GND), and setting gain depending on DAz_GAIN (z=1,2,3) bits, and then going through an auto-zero routine to minimize amplifier input offset. When CAL_CONN bit is set to “1”, it is done by keeping Sw1 and Sw2 turn on (connecting differential amplifier inputs to the external shunt).

Note: If initiating calibration by CAL_DAz (z=1,2,3) bit, each amplifier can be calibrated individually. It takes approximately 288µs to finish calibration on each amplifier. Renesas recommends allowing 400µs per one amplifier for the calibration to complete in actual application. Although the calibration can be initiated by SPI interface even on the fly, Renesas recommends conducting the calibration when no MOSFETs are switching with all driver output pulled low to avoid any impact of noise on calibration accuracy.

6.5.2 Differential Amplifiers for Current Sensing (continued)



Note: Gray line is DAzP (z=1,2,3) signal.
Black line is the virtual signal which sampled DAzP (z=1,2,3) by nSMPL.

Figure 6.5-2 S/H Function of Current Sensing

6.5.3 BEMF Sense Amplifier

The BEMF sense amplifier has the following functions.

Enable control:

BEMF sense amplifier can be disabled by BEMF_EN=0b according to the actual application.

Programmable gain:

The gain of BEMF sense amplifier is programmable by BEMF_GAIN bits. The gain settings are 0.05V/V, 0.1V/V, 0.5V/V, 1.0V/V with DA3_GAIN=00b.

Reference:

The output reference voltage of BEMF sense amplifier is 0.5*V_{DD}.

High impedance phase selection:

In typical trapezoidal BLDC operation, only 2-phase bridges are energized at a given time. The 3rd phase is in high impedance state (both high-side and low-side MOSFETs are turned off). By sensing the differential voltage between this high impedance phase and the virtual center tap, provides you the BEMF induced in this 3rd phase stator coil, which allows you to know/estimate the rotor position relative to this 3rd phase. This device has three methods for the high impedance phase selection according to the BEMF_PH bits. When BEMF_PH bits are set to 00xb, the 3rd phase is detected by checking the state of HIx/LIx (x=A,B,C) signals in S/H control logic. The check timing is the falling edge of nSMPL signal. When BEMF_PH bits are set to 010b or 011b, the 3rd phase is selected according to CMP1O/2O or CMP1O/3O pins. When BEMF_PH bits are set to 1xxb, the 3rd phase is selected according to BEMF_PH1,0 bits directly. These functions help to achieve the position sensorless trapezoidal BLDC operation without the additional external circuits. Refer to Figure 6.5-3 and section 7.1.15.

Sample and hold (S/H) function:

BEMF sense amplifier also has S/H function. S/H capacitance and amplifier are common use with the differential amplifier 3. S/H switch becomes active only when BEMF_EN bit is set to “1”. The actual turning on timing of the S/H switch needs to occur after any possible transition is over. That requires to turn on after a reasonable delay relative to the HIx/LIx (x=A,B,C) rising edge. The delay time is realized by waiting for LS_ON (or HS_ON if it’s high-side turning on transition of HSx (x=A,B,C)) going high plus the configured gate driver transition time tGT. The actual turning off timing of the S/H switch aligns to the falling edge of the internal gate-off logic signal (which is issued shortly after HIx/LIx (x=A,B,C) falling edge). This ensures the hold value is not affected by the turning off transition.

The turn on/off timing of the S/H switch is adjustable by nSMPL signal. The S/H switch keeps turn off during nSMPL=H. If any possible transition remains after the internal S/H delay time, the adjustment of turn on timing by nSMPL signal is necessary. Refer to Figure 6.5-4.

DC offset calibration:

DC offset calibration can be initiated by setting CAL_DA3/BEMF bit with BEMF_EN=1b. The calibration phase is selectable by the combination of CAL_BCONN bit and BEMF_PH bits in the Sense Block Control registers. The amplifier gain depends on BEMF_GAIN bits. The DC offset calibration goes through an auto-zero routine to minimize amplifier input offset.

Note: If initiating calibration by CAL_DA3/BEMF bit with BEMF_EN=1b, it takes approximately 288µs to finish calibration. Renesas recommends allowing 400µs for the calibration to complete in actual application. Although the calibration can be initiated by SPI interface even on the fly, the motor has to be stop condition to avoid BEMF voltage input.

6.5.3 BEMF Sense Amplifier (continued)

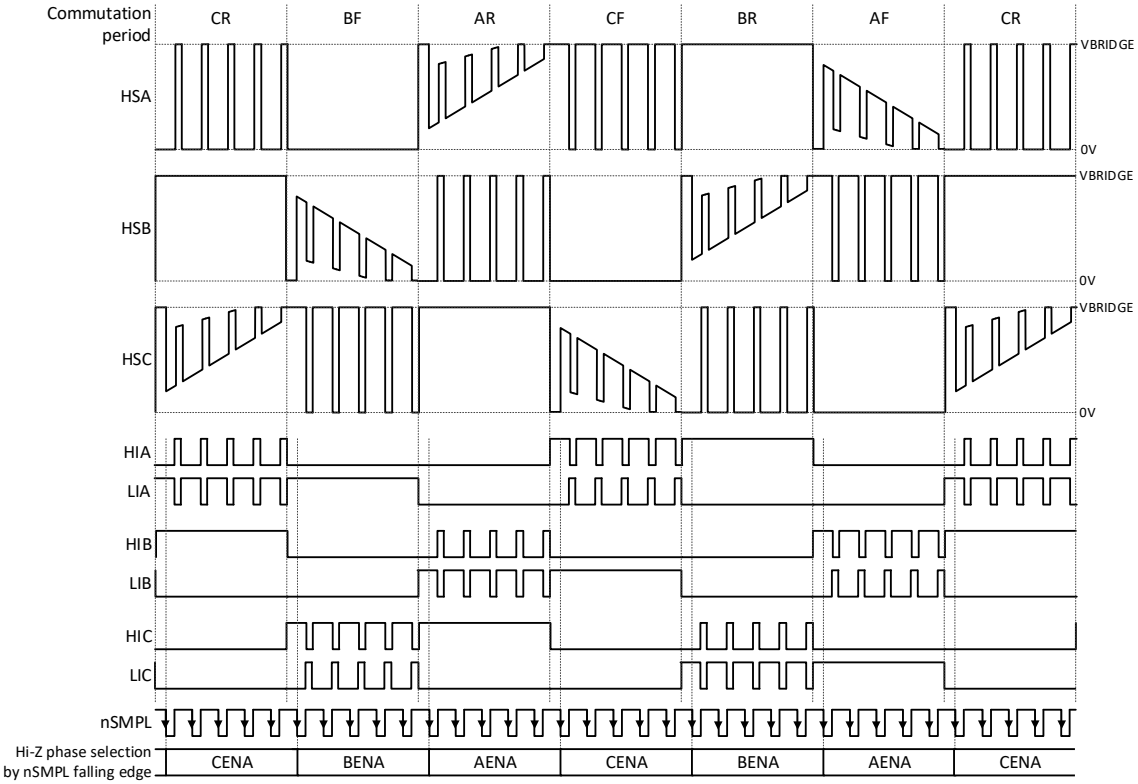


Figure 6.5-3 The Relation Between Typical Six Step Trapezoidal Drive and BEMF Detect Phase

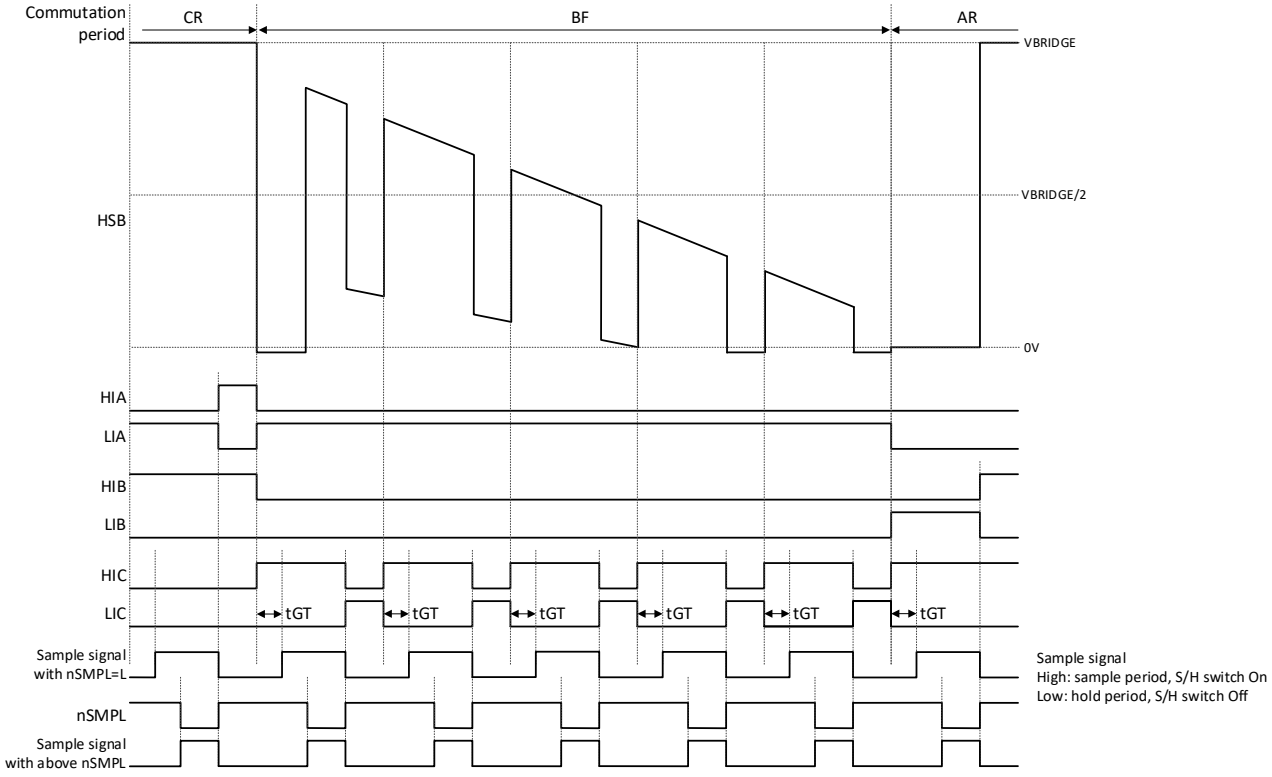


Figure 6.5-4 S/H Timing During Commutation Period BF

6.5.4 Comparators

The general-purpose comparator has the following functions.

Enable control:

Each comparator can be disabled by setting CMPz_VTH (z=1,2,3) bits to 0000b according to the actual application.

Programmable threshold voltage and hysteresis:

The threshold voltage of general-purpose comparator is programmable by CMPz_VTH (z=1,2,3) bits individually. The threshold setting including the hysteresis voltage is following equation.

- Falling: $V_{TH\_CMP} = V_{DD} / 16 \times CMPz\_VTH - 44mV \times (1 - CMPz\_HYS)$
- Rising: $V_{TH\_CMP} = V_{DD} / 16 \times CMPz\_VTH + 44mV \times (1 - CMPz\_HYS)$

Pin usage limitation:

The inputs of the general-purpose comparators are common use with the inputs of the differential amplifiers. When all of the differential amplifiers are used, the general-purpose comparators can be used for ONLY same inputs as the differential amplifiers.

6.5.5 MUX1 Output Control

DA3O/MUX1 pin has the analog multiplexer function. The following analog signals can be monitored depending on MUX bits in the Sense Block Control registers. After changing DA3O/MUX1 pin output or BEMF_EN bit, it is necessary to wait more than 4µs for the settling time of DA3O/MUX1 pin output.

- MUX=000b: GND (330kΩ pulldown)
- MUX=001b: VM monitor
- MUX=010b: Die temperature monitor
- MUX=011b: Differential amplifier reference voltage
- MUX=100b, BEMF_EN=0b: Differential amplifier 3 output with 10kΩ
- MUX=100b, BEMF_EN=1b: BEMF sense amplifier output with 10kΩ
- MUX=101b: Differential amplifier 1 output
- MUX=110b: Differential amplifier 2 output
- MUX=111b, BEMF_EN=0b: Differential amplifier 3 output without 10kΩ
- MUX=111b, BEMF_EN=1b: BEMF sense amplifier output without 10kΩ

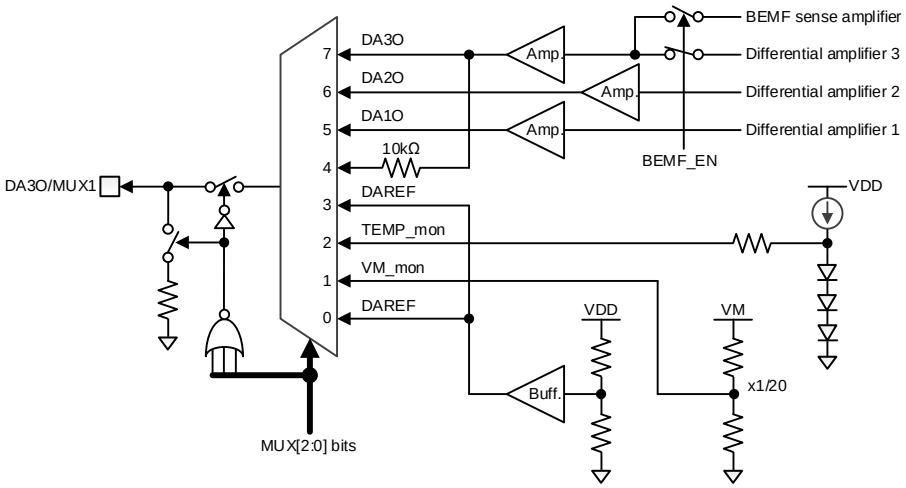


Figure 6.5-5 Block Diagram of Analog Multiplexer for DA3O/MUX1 Pin

6.5.6 VM Monitor

The output of VM attenuator can be monitored by setting 001b to MUX bits in the Sense Block Control registers. The relation between this monitored voltage (DA3O/MUX1 pin) and the VM voltage is shown in Figure 6.5-6.

The ratio of VM voltage and the monitored voltage (RVM) is 20.0 typical.

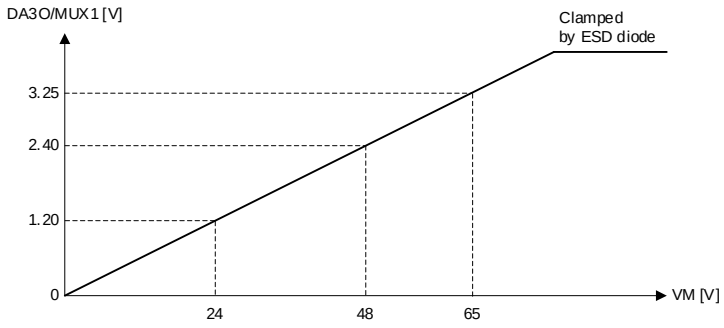


Figure 6.5-6 The Relation Between DA3O/MUX1 and VM Voltage

6.5.7 Junction Temperature Monitor

The junction temperature of the die can be monitored by setting 010b to MUX bits in the Sense Block Control 5 registers. The relation between this monitored voltage (DA3O/MUX1 pin) and the junction temperature is shown in Figure 6.5-7.

The junction temperature (Tj) is calculated by the following equation.

- Tj [°C]= 25[°C] + (2.000 – MUX1 voltage) [V] / 6.0 [mV/°C]

For example, in the case of ISENADIN=1.520[V]

- Tj [°C]= 25[°C] + (2.000 – 1.520) [V] / 6.0 [mV/°C]= 105[°C]

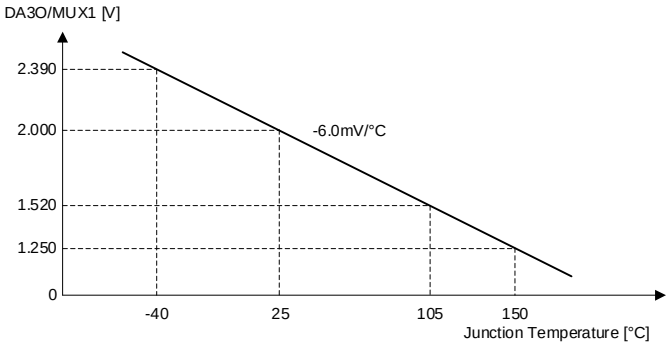


Figure 6.5-7 The Relation Between DA3O/MUX1 and the Junction Temperature

6.6 SPI Communication Format

The SPI block of this device only works in slave mode. Figure 6.6-1 shows SPI communication format of both write and read mode. If the communication format is different from Figure 6.6-1, its communication becomes invalid.

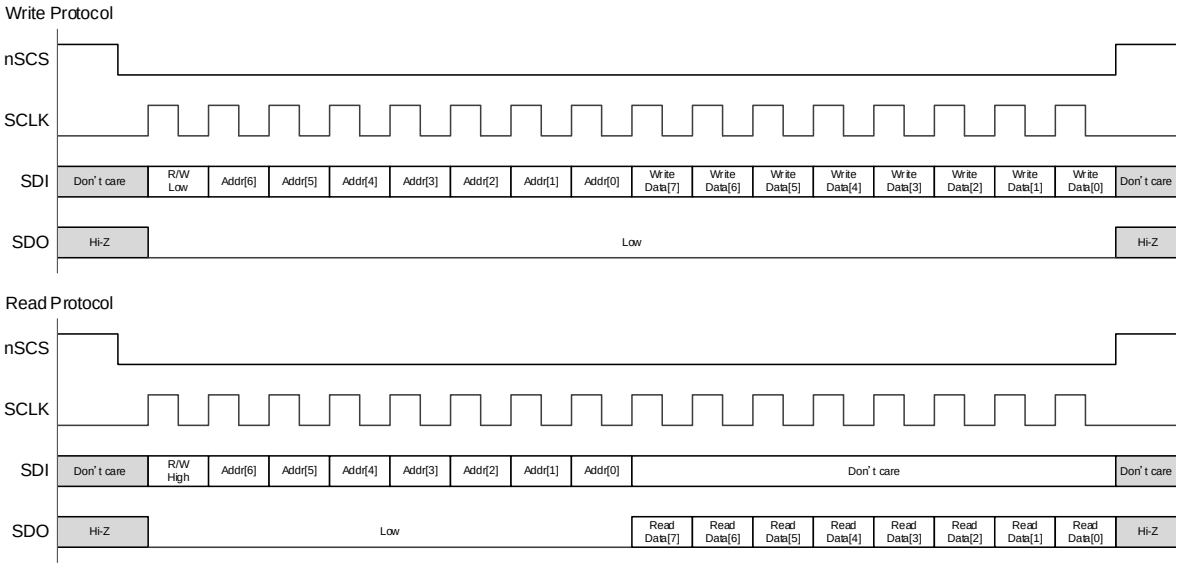


Figure 6.6-1 SPI Communication Format

7 Control Register Information

7.1 Control Register Map

Table 7-1 shows the RAA306012's control register map. The control registers are reset by entering Sleep or Shutdown Mode. Refer to Section 6.1.3 Mode Transition.

Table 7-1 Control Register Map

Address	Register Name	Symbol	Access Type	Initial value	7	6	5	4	3	2	1	0
0x00	Fault Status 0	FLTSTS0	R	00h	FAULT	SR_FAULT	OV_UVLO	VDS_OCP	VGS_FAULT	CS_OCP	OTSD	TWARN
0x01	Fault Status 1	FLTSTS1	R	00h	VDRV_UV	VDRV_OV	SR_OCP	VCP_UV	VM_UV	VM_OV	N/A	N/A
0x02	Fault Status 2	FLTSTS2	R	00h	VDSHA_OCP	VDSLA_OCP	VGSHA_FAULT	VGSLA_FAULT	VDSHB_OCP	VDSLB_OCP	VGSHB_FAULT	VGSLB_FAULT
0x03	Fault Status 3	FLTSTS3	R	00h	VDSHC_OCP	VDSLC_OCP	VGSHC_FAULT	VGSLC_FAULT	N/A	CS1_OCP	CS2_OCP	CS3_OCP
0x04	Fault Control 1	FLTCTL1	R/W	00h	DIS_VDRVUV	DIS_VDRVOV	DIS_SROC	DIS_VCPUV	DIS_VMUV	DIS_VMOV	DIS_OTSD	TWARN_REP
0x05	Fault Control 2	FLTCTL2	R/W	07h	CSOCP_MODE1	CSOCP_MODE0	VDSOCP_MODE1	VDSOCP_MODE0	DIS_VGSFLT	DIS_CS1OCP	DIS_CS2OCP	DIS_CS3OCP
0x06	IC Control 1	ICCTL1	R/W	35h	CLR_FLT	WRITE_LOCK2	WRITE_LOCK1	WRITE_LOCK0	PWMMODE	CSOCP_TH2	CSOCP_TH1	CSOCP_TH0
0x07	IC Control 2	ICCTL2	R/W	50h	DEAD_TIME1	DEAD_TIME0	T_GT1	T_GT0	BEMF_EN	DA1_EN	DA2_EN	DA3_EN
0x08	Gate Driver Control	GDCTL	R/W	FFh	ISRC_HS3	ISRC_HS2	ISRC_HS1	ISRC_HS0	ISRC_LS3	ISRC_LS2	ISRC_LS1	ISRC_LS0
0x09	Over Current Protection Control	OCPCTL	R/W	00h	VDS_TH3	VDS_TH2	VDS_TH1	VDS_TH0	TRETRY_CSOCP	TRETRY_VDSOCP	DEG_TIME1	DEG_TIME0
0x0A	Phase-A Gate Driver Input Selection	GDSELA	R/W	14h	CMP1_HYS	HOA_SEL2	HOA_SEL1	HOA_SEL0	VMUV_TH	LOA_SEL2	LOA_SEL1	LOA_SEL0
0x0B	Phase-B Gate Driver Input Selection	GDSELB	R/W	25h	CMP2_HYS	HOB_SEL2	HOB_SEL1	HOB_SEL0	PDMODE	LOB_SEL2	LOB_SEL1	LOB_SEL0
0x0C	Phase-C Gate Driver Input Selection	GDSELC	R/W	36h	CMP3_HYS	HOC_SEL2	HOC_SEL1	HOC_SEL0	CPUV_TH	LOC_SEL2	LOC_SEL1	LOC_SEL0
0x0D	Sense Block Control 1	SNSCTL1	R/W	AAh	BEMF_GAIN1	BEMF_GAIN0	DA1_GAIN1	DA1_GAIN0	DA2_GAIN1	DA2_GAIN0	DA3_GAIN1	DA3_GAIN0
0x0E	Sense Block Control 2	SNSCTL2	R/W	00h	CAL_BCONN	BEMF_PH2	BEMF_PH1	BEMF_PH0	BEMF_SH	DA1_SH	DA2_SH	DA3_SH
0x0F	Sense Block Control 3	SNSCTL3	R/W	88h	CMP1_VTH3	CMP1_VTH2	CMP1_VTH1	CMP1_VTH0	CMP2_VTH3	CMP2_VTH2	CMP2_VTH1	CMP2_VTH0
0x10	Sense Block Control 4	SNSCTL4	R/W	80h	CMP3_VTH3	CMP3_VTH2	CMP3_VTH1	CMP3_VTH0	CAL_CONN	CAL_DA1	CAL_DA2	CAL_DA3/BEMF
0x11	Sense Block Control 5	SNSCTL5	R/W	00h	DIS_SADT	RESERVED11_6	CTL6_UNLOCK	RESERVED11_4	RESERVED11_3	MUX2	MUX1	MUX0
0x12	Sense Block Control 6	SNSCTL6	R/W	40h	RESERVED12_7	BEMF_OFFSET	RESERVED12_5	RESERVED12_4	RESERVED12_3	RESERVED12_2	RESERVED12_1	GD_AOR

7.1.1 Fault Status 0 Register: FLTSTS0 (Address=0x00) [Default=0x00]

Figure 7.1-1 and Table 7.1-1 show the details of Fault Status 0 register.

Figure 7.1-1 Fault Status 0 Register FLTSTS0

7	6	5	4	3	2	1	0
FAULT	SR_FAULT	OV_UVLO	VDS_OCP	VGS_FAULT	CS_OCP	OTSD	TWARN
R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b

Table 7.1-1 Fault Status 0 Register FLTSTS0 Descriptions

Bit	Field	Type	Default	Description
7	FAULT	R	0b	Logic OR of all Fault Status bits
6	SR_FAULT	R	0b	Logic OR of the Fault Status bits for buck switching regulator: VDRV_UV, VDRV_OV, SR_OCP
5	OV_UVLO	R	0b	Logic OR of the Fault Status bits for undervoltage and overvoltage: VCP_UV, VM_UV, VM_OV
4	VDS_OCP	R	0b	Logic OR of the Fault Status bits for V_{DS} overcurrent: VDSHx_OCP, VDSLx_OCP
3	VGS_FAULT	R	0b	Logic OR of the Fault Status bits for V_{GS} fault : VGSHx_FAULT, VGSLx_FAULT
2	CS_OCP	R	0b	Logic OR of the Fault Status bits for current sense overcurrent: CS1_OCP, CS2_OCP, CS3_OCP
1	OTSD	R	0b	Indicator of thermal shutdown
0	TWARN	R	0b	Indicator of thermal warning

Note1: Fault status registers are reset by writing 1b to CLR_FLT, or recovery low pulse (>tsleep: 0.85ms) on EN pin.

7.1.2 Fault Status 1 Register: FLTSTS1 (Address=0x01) [Default=0x00]

Figure 7.1-2 and Table 7.1-2 show the details of Fault Status 1 register.

Figure 7.1-2 Fault Status 1 Register FLTSTS1

7	6	5	4	3	2	1	0
VDRV_UV	VDRV_OV	SR_OCP	VCP_UV	VM_UV	VM_OV	N/A	N/A
R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b

Table 7.1-2 Fault Status 1 Register FLTSTS1 Descriptions

Bit	Field	Type	Default	Description
7	VDRV_UV	R	0b	Indicator of VDRV undervoltage (V_{DRVUV})
6	VDRV_OV	R	0b	Indicator of VDRV overvoltage (V_{DRVOV})
5	SR_OCP	R	0b	Indicator of buck switching regulator overcurrent (I_{OC2_SR})
4	VCP_UV	R	0b	Indicator of VCP undervoltage (V_{CPUV})
3	VM_UV	R	0b	Indicator of VM undervoltage (V_{VMUV})
2	VM_OV	R	0b	Indicator of VM overvoltage (V_{VMOV})
1	N/A	R	0b	Not assigned
0	N/A	R	0b	Not assigned

Note1: Fault status registers are reset by writing 1b to CLR_FLT, or recovery low pulse (>tsleep: 0.85ms) on EN pin.

7.1.3 Fault Status 2 Register: FLTSTS2 (Address=0x02) [Default=0x00]

Figure 7.1-3 and Table 7.1-3 show the details of Fault Status 2 register.

Figure 7.1-3 Fault Status 2 Register FLTSTS2

7	6	5	4	3	2	1	0
VDSHA_OCP	VDSLA_OCP	VGSHA_FAULT	VGSLA_FAULT	VDSHB_OCP	VDSLB_OCP	VGSHB_FAULT	VGSLB_FAULT
R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b

Table 7.1-3 Fault Status 2 Register FLTSTS2 Descriptions

Bit	Field	Type	Default	Description
7	VDSHA_OCP	R	0b	Indicator of V_{DS} overcurrent on Phase-A high-side MOSFET
6	VDSLA_OCP	R	0b	Indicator of V_{DS} overcurrent on Phase-A low-side MOSFET
5	VGSHA_FAULT	R	0b	Indicator of V_{GS} fault on Phase-A high-side MOSFET
4	VGSLA_FAULT	R	0b	Indicator of V_{GS} fault on Phase-A low-side MOSFET
3	VDSHB_OCP	R	0b	Indicator of V_{DS} overcurrent on Phase-B high-side MOSFET
2	VDSLB_OCP	R	0b	Indicator of V_{DS} overcurrent on Phase-B low-side MOSFET
1	VGSHB_FAULT	R	0b	Indicator of V_{GS} fault on Phase-B high-side MOSFET
0	VGSLB_FAULT	R	0b	Indicator of V_{GS} fault on Phase-B low-side MOSFET

Note1: Fault status registers are reset by writing 1b to CLR_FLT, or recovery low pulse (>tsleep: 0.85ms) on EN pin.

7.1.4 Fault Status 3 Register: FLTSTS3 (Address=0x03) [Default=0x00]

Figure 7.1-4 and Table 7.1-4 show the details of Fault Status 3 register.

Figure 7.1-4 Fault Status 3 Register FLTSTS3

7	6	5	4	3	2	1	0
VDSHC_OCP	VDSLC_OCP	VGSHC_FAULT	VGSLC_FAULT	N/A	CS1_OCP	CS2_OCP	CS3_OCP
R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b	R: 0b

Table 7.1-4 Fault Status 3 Register FLTSTS3 Descriptions

Bit	Field	Type	Default	Description
7	VDSHC_OCP	R	0b	Indicator of V_{DS} overcurrent on Phase-C high-side MOSFET
6	VDSLC_OCP	R	0b	Indicator of V_{DS} overcurrent on Phase-C low-side MOSFET
5	VGSHC_FAULT	R	0b	Indicator of V_{GS} fault on Phase-C high-side MOSFET
4	VGSLC_FAULT	R	0b	Indicator of V_{GS} fault on Phase-C low-side MOSFET
3	N/A	R	0b	Not assigned
2	CS1_OCP	R	0b	Indicator of current sense overcurrent by DA1P, DA1N inputs
1	CS2_OCP	R	0b	Indicator of current sense overcurrent by DA2P, DA2N inputs
0	CS3_OCP	R	0b	Indicator of current sense overcurrent by DA3P, DA3N inputs

Note1: Fault status registers are reset by writing 1b to CLR_FLT, or recovery low pulse (>tsleep: 0.85ms) on EN pin.

7.1.5 Fault Control 1 Register: FLTCTL1 (Address=0x04) [Default=0x00]

Figure 7.1-5 and Table 7.1-5 show the details of Fault Control 1 register.

Figure 7.1-5 Fault Control 1 Register FLTCTL1

7	6	5	4	3	2	1	0
DIS_VDRVUV	DIS_VDRVOV	DIS_SROC	DIS_VCPUV	DIS_VMUV	DIS_VMOV	DIS_OTSD	TWARN_REP
R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-5 Fault Control 1 Register FLTCTL1 Descriptions

Bit	Field	Type	Default	Description
7	DIS_VDRVUV	R/W	0b	Write 1b to report status only for VDRV undervoltage (V_{DRVUV}) detection
6	DIS_VDRVOV	R/W	0b	Write 1b to report status only for VDRV overvoltage (V_{DRVOV}) detection
5	DIS_SROC	R/W	0b	Write 1b to report status only for buck switching regulator overcurrent (I_{OC2_SR}) protection
4	DIS_VCPUV	R/W	0b	Write 1b to report status only for VCP undervoltage (V_{CPUV}) detection
3	DIS_VMUV	R/W	0b	Write 1b to report status only for VM undervoltage (V_{VMUV}) detection
2	DIS_VMOV	R/W	0b	Write 1b to report status only for VM overvoltage fault (V_{VMOV}) detection
1	DIS_OTSD	R/W	0b	Write 1b to report status only for thermal shutdown
0	TWARN_REP	R/W	0b	0b: Thermal warning is reported on only TWARN bit. 1b: Thermal warning is reported on nFAULT pin, FAULT bit and TWARN bit.

7.1.6 Fault Control 2 Register: FLTCTL2 (Address=0x05) [Default=0x07]

Figure 7.1-6 and Table 7.1-6 show the details of Fault Control 2 register.

Figure 7.1-6 Fault Control 2 Register FLTCTL2

7	6	5	4	3	2	1	0
CSOCP_MODE1	CSOCP_MODE0	VDSOCP_MODE1	VDSOCP_MODE0	DIS_VGSFLT	DIS_CS1OCP	DIS_CS2OCP	DIS_CS3OCP
R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 1b	R/W: 1b	R/W: 1b

Table 7.1-6 Fault Control 2 Register FLTCTL2 Descriptions

Bit	Field	Type	Default	Description
7	CSOCP_MODE1	R/W	0b	Response mode for current sense overcurrent 00b: Latch upon current sense overcurrent [Note2] 01b: Automatic retry upon current sense overcurrent 10b: Report on nFAULT pin, FAULT, CS_OCP and CS1/2/3_OCP bits only. No action takes place. 11b: Disable. No report and no action takes place.
6	CSOCP_MODE0	R/W	0b	
5	VDSOCP_MODE1	R/W	0b	Response mode for V_{DS} overcurrent 00b: Latch upon V_{DS} overcurrent [Note2] 01b: Automatic retry upon V_{DS} overcurrent 10b: Report on nFAULT pin, FAULT, VDS_OCP, VDSHx_OCP and VDSLx_OCP bits only. No action takes place. 11b: Disable, No report and no action takes place
4	VDSOCP_MODE0	R/W	0b	
3	DIS_VGSFLT	R/W	0b	Write 1b to disable V_{GS} fault detection
2	DIS_CS1OCP	R/W	1b	Write 1b to disable current sense overcurrent by DA1P, DA1N inputs
1	DIS_CS2OCP	R/W	1b	Write 1b to disable current sense overcurrent by DA2P, DA2N inputs
0	DIS_CS3OCP	R/W	1b	Write 1b to disable current sense overcurrent by DA3P, DA3N inputs

Note2: Latch is recovered by writing 1b to CLR_FLT, or recovery low pulse (>tsleep: 0.85ms) on EN pin.

7.1.7 IC Control 1 Register: ICCTL1 (Address=0x06) [Default=0x35]

Figure 7.1-7 and Table 7.1-7 show the details of IC Control1 register.

Figure 7.1-7 IC Control 1 Register ICCTL1

7	6	5	4	3	2	1	0
CLR_FLT	WRITE_LOCK2	WRITE_LOCK1	WRITE_LOCK0	PWMMODE	CSOCP_TH2	CSOCP_TH1	CSOCP_TH0
R/W: 0b	R/W: 0b	R/W: 1b	R/W: 1b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 1b

Table 7.1-7 IC Control 1 Register ICCTL1 Descriptions

Bit	Field	Type	Default	Description
7	CLR_FLT	R/W	0b	Write 1b to clear the all flagged fault status bits. This bit is reset to 0b automatically.
6	WRITE_LOCK2	R/W	0b	Write 110b to ignore all further register write except WRITE_LOCK[2:0]. Write 011b to unlock to allow register write. Writing other values takes no effect.
5	WRITE_LOCK1	R/W	1b	
4	WRITE_LOCK0	R/W	1b	
3	PWMMODE	R/W	0b	0b: 3-Phase HI/LI mode, 1b: 3-Phase PWM mode
2	CSOCP_TH2	R/W	1b	Threshold voltage setting of current sense overcurrent by DAzP, DAzN (z=1,2,3) inputs 000b: 51mV, 001b: 105mV, 010b: 157mV, 011b: 208mV, 100b: 260mV, 101b: 516mV, 110b: 773mV, 111b: 1029mV
1	CSOCP_TH1	R/W	0b	
0	CSOCP_TH0	R/W	1b	

7.1.8 IC Control 2 Register: ICCTL2 (Address=0x07) [Default=0x50]

Figure 7.1-8 and Table 7.1-8 show the details of IC Control 2 register.

Figure 7.1-8 IC Control 2 Register ICCTL2

7	6	5	4	3	2	1	0
DEAD_TIME1	DEAD_TIME0	T_GT1	T_GT0	BEMF_EN	DA1_EN	DA2_EN	DA3_EN
R/W: 0b	R/W: 1b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-8 IC Control 2 Register ICCTL2 Descriptions

Bit	Field	Type	Default	Description
7	DEAD_TIME1	R/W	0b	Dead time from V_{GS} voltage monitor output to start timing of another-side turn on 00b: 50ns, 01b: 100ns, 10b: 200ns, 11b: 400ns
6	DEAD_TIME0	R/W	1b	
5	T_GT1	R/W	0b	Maximum gate transition time 00b: 500ns, 01b: 1000ns, 10b: 2000ns, 11b: 4000ns
4	T_GT0	R/W	1b	
3	BEMF_EN	R/W	0b	Write 1b to enable BEMF sense amplifier.
2	DA1_EN	R/W	0b	Write 1b to enable differential amplifier 1.
1	DA2_EN	R/W	0b	Write 1b to enable differential amplifier 2.
0	DA3_EN	R/W	0b	Write 1b to enable differential amplifier 3.

7.1.9 Gate Drive Control Register: GDCTL (Address=0x08) [Default=0xFF]

Figure 7.1-9 and Table 7.1-9 show the details of Gate Drive Control register.

Figure 7.1-9 Gate Drive Control Register GDCTL

7	6	5	4	3	2	1	0
ISRC_HS3	ISRC_HS2	ISRC_HS1	ISRC_HS0	ISRC_LS3	ISRC_LS2	ISRC_LS1	ISRC_LS0
R/W: 1b	R/W: 1b	R/W: 1b	R/W: 1b	R/W: 1b	R/W: 1b	R/W: 1b	R/W: 1b

Table 7.1-9 Gate Drive Control Register GDCTL Descriptions

Bit	Field	Type	Default	Description
7	ISRC_HS3	R/W	1b	High-side gate driver output source current. Sink current is 2*(source current). 0000b: 50mA, 0001b: 60mA, 0010b: 70mA, 0011b: 80mA, 0100b: 100mA, 0101b: 120mA, 0110b: 140mA, 0111b: 160mA, 1000b: 200mA, 1001b: 240mA, 1010b: 280mA, 1011b: 320mA, 1100b: 400mA, 1101b: 480mA, 1110b: 560mA, 1111b: 640mA
6	ISRC_HS2	R/W	1b	
5	ISRC_HS1	R/W	1b	
4	ISRC_HS0	R/W	1b	
3	ISRC_LS3	R/W	1b	Low-side gate driver output source current. Sink current is 2*(source current). 0000b: 50mA, 0001b: 60mA, 0010b: 70mA, 0011b: 80mA, 0100b: 100mA, 0101b: 120mA, 0110b: 140mA, 0111b: 160mA, 1000b: 200mA, 1001b: 240mA, 1010b: 280mA, 1011b: 320mA, 1100b: 400mA, 1101b: 480mA, 1110b: 560mA, 1111b: 640mA
2	ISRC_LS2	R/W	1b	
1	ISRC_LS1	R/W	1b	
0	ISRC_LS0	R/W	1b	

7.1.10 Overcurrent Protection Control Register: OCPCTL (Address=0x09) [Default=0x00]

Figure 7.1-10 and Table 7.1-10 show the details of Overcurrent Protection Control register.

Figure 7.1-10 Overcurrent Protection Control Register OCPCTL

7	6	5	4	3	2	1	0
VDS_TH3	VDS_TH2	VDS_TH1	VDS_TH0	TRETRY_CSOCP	TRETRY_VDSOCP	DEG_TIME1	DEG_TIME0
R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-10 Overcurrent Protection Control Register OCPCTL Descriptions

Bit	Field	Type	Default	Description
7	VDS_TH3	R/W	0b	Threshold voltage setting of V_{DS} overcurrent fault 0000b: 40mV, 0001b: 60mV, 0010b: 80mV, 0011b: 120mV, 0100b: 160mV, 0101b: 200mV, 0110b: 240mV, 0111b: 320mV, 1000b: 400mV, 1001b: 480mV, 1010b: 600mV, 1011b: 720mV, 1100b: 960mV, 1101b: 1200mV, 1110b: 1600mV, 1111b: 2000mV
6	VDS_TH2	R/W	0b	
5	VDS_TH1	R/W	0b	
4	VDS_TH0	R/W	0b	
3	TRETRY_CSOCP	R/W	0b	Retry time for current sense overcurrent fault with CSOCP_MODE=01b, 0b: 4000µs, 1b: 70µs
2	TRETRY_VDSOCP	R/W	0b	Retry time for V_{DS} overcurrent fault with VDSOCP_MODE=01b, 0b: 4000µs, 1b: 70µs
1	DEG_TIME1	R/W	0b	Deglitch time for both current sense and V_{DS} overcurrent fault 00b: 1.57µs, 01b: 2.38µs, 10b: 3.49µs, 11b: 5.73µs
0	DEG_TIME0	R/W	0b	

7.1.11 Phase-A Gate Driver Input Selection Register: GDSELA (Address=0x0A) [Default=0x14]

Figure 7.1-11 and Table 7.1-11 show Phase-A Gate Driver Input Selection register.

Figure 7.1-11 Phase-A Gate Driver Input Selection Register GDSELA

7	6	5	4	3	2	1	0
CMP1_HYS	HOA_SEL2	HOA_SEL1	HOA_SEL0	VMUV_TH	LOA_SEL2	LOA_SEL1	LOA_SEL0
R/W: 0b	R/W: 0b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 0b

Table 7.1-11 Phase-A Gate Driver Input Selection Register GDSELA Descriptions

Bit	Field	Type	Default	Description
7	CMP1_HYS	R/W	0b	Comparator 1 hysteresis setting, 0b: +/-44mV, 1b: 0mV
6	HOA_SEL2	R/W	0b	Input selection for Phase-A high-side gate driver Note3 000b: Lo fix, 001b: IN1, 010b: IN2, 011b: IN3, 100b: IN4, 101b: IN5, 110b: IN6, 111b: Hi-Z
5	HOA_SEL1	R/W	0b	
4	HOA_SEL0	R/W	1b	
3	VMUV_TH	R/W	0b	VM under voltage threshold setting, 0b: 5.3V, 1b: 7.5V
2	LOA_SEL2	R/W	1b	Input selection for Phase-A low-side gate driver Note3 000b: Lo fix, 001b: IN1, 010b: IN2, 011b: IN3, 100b: IN4, 101b: IN5, 110b: IN6, 111b: Hi-Z
1	LOA_SEL1	R/W	0b	
0	LOA_SEL0	R/W	0b	

Note3: When HOx_SEL or LOx_SEL bits set to 111b, the source/sink current of gate driver becomes off (Hi-Z).

7.1.12 Phase-B Gate Driver Input Selection Register: GDSELB (Address=0x0B) [Default=0x25]

Figure 7.1-12 and Table 7.1-12 show Phase-B Gate Driver Input Selection register.

Figure 7.1-12 Phase-B Gate Driver Input Selection Register GDSELB

7	6	5	4	3	2	1	0
CMP2_HYS	HOB_SEL2	HOB_SEL1	HOB_SEL0	PDMODE	LOB_SEL2	LOB_SEL1	LOB_SEL0
R/W: 0b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 1b

Table 7.1-12 Phase-B Gate Driver Input Selection Register GDSELB Descriptions

Bit	Field	Type	Default	Description
7	CMP2_HYS	R/W	0b	Comparator 2 hysteresis setting, 0b: +/-44mV, 1b: 0mV
6	HOB_SEL2	R/W	0b	Input selection for Phase-B high-side gate driver Note3 000b: Lo fix, 001b: IN1, 010b: IN2, 011b: IN3, 100b: IN4, 101b: IN5, 110b: IN6, 111b: Hi-Z
5	HOB_SEL1	R/W	1b	
4	HOB_SEL0	R/W	0b	
3	PDMODE	R/W	0b	Gate driver pulldown mode after VDS_OCP, CS_OCP, 0b: Hi-Z pulldown, 1b: driver output low
2	LOB_SEL2	R/W	1b	Input selection for Phase-B low-side gate driver Note3 000b: Lo fix, 001b: IN1, 010b: IN2, 011b: IN3, 100b: IN4, 101b: IN5, 110b: IN6, 111b: Hi-Z
1	LOB_SEL1	R/W	0b	
0	LOB_SEL0	R/W	1b	

Note3: When HOx_SEL or LOx_SEL bits set to 111b, the source/sink current of gate driver becomes off (Hi-Z).

7.1.13 Phase-C Gate Driver Input Selection Register: GDSELC (Address=0x0C) [Default=0x36]

Figure 7.1-13 and Table 7.1-13 show Phase-C Gate Driver Input Selection register.

Figure 7.1-13 Phase-C Gate Driver Input Selection Register GDSELC

7	6	5	4	3	2	1	0
CMP3_HYS	HOC_SEL2	HOC_SEL1	HOC_SEL0	CPUV_TH	LOC_SEL2	LOC_SEL1	LOC_SEL0
R/W: 0b	R/W: 0b	R/W: 1b	R/W: 1b	R/W: 0b	R/W: 1b	R/W: 1b	R/W: 0b

Table 7.1-13 Phase-C Gate Driver Input Selection Register GDSELC Descriptions

Bit	Field	Type	Default	Description
7	CMP3_HYS	R/W	0b	Comparator 3 hysteresis setting, 0b: +/-44mV, 1b: 0mV
6	HOC_SEL2	R/W	0b	Input selection for Phase-C high-side gate driver Note3 000b: Lo fix, 001b: IN1, 010b: IN2, 011b: IN3, 100b: IN4, 101b: IN5, 110b: IN6, 111b: Hi-Z
5	HOC_SEL1	R/W	1b	
4	HOC_SEL0	R/W	1b	
3	CPUV_TH	R/W	0b	VCP under voltage threshold setting, 0b: 0.58*VDRV, 1b: 0.8*VDRV
2	LOC_SEL2	R/W	1b	Input selection for Phase-C low-side gate driver Note3 000b: Lo fix, 001b: IN1, 010b: IN2, 011b: IN3, 100b: IN4, 101b: IN5, 110b: IN6, 111b: Hi-Z
1	LOC_SEL1	R/W	1b	
0	LOC_SEL0	R/W	0b	

Note3: When HOx_SEL or LOx_SEL bits set to 111b, the source/sink current of gate driver becomes off (Hi-Z).

7.1.14 Sense Block Control 1 Register: SNSCTL1 (Address=0x0D) [Default=0xAA]

Figure 7.1-14 and Table 7.1-14 show Sense Block Control 1 register.

Figure 7.1-14 Sense Block Control 1 Register SNSCTL1

7	6	5	4	3	2	1	0
BEMF_GAIN1	BEMF_GAIN0	DA1_GAIN1	DA1_GAIN0	DA2_GAIN1	DA2_GAIN0	DA3_GAIN1	DA3_GAIN0
R/W: 1b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 1b	R/W: 0b

Table 7.1-14 Sense Block Control 1 Register SNSCTL1 Descriptions

Bit	Field	Type	Default	Description
7	BEMF_GAIN1	R/W	1b	Gain setting of BEMF sense amplifier with DA3_GAIN=00b 00b: 0.05V/V, 01b: 0.1V/V, 10b: 0.5V/V, 11b: 1.0V/V
6	BEMF_GAIN0	R/W	0b	
5	DA1_GAIN1	R/W	1b	Gain setting of differential amplifier 1 00b: 5V/V, 01b: 10V/V, 10b: 20V/V, 11b: 40V/V
4	DA1_GAIN0	R/W	0b	
3	DA2_GAIN1	R/W	1b	Gain setting of differential amplifier 2 00b: 5V/V, 01b: 10V/V, 10b: 20V/V, 11b: 40V/V
2	DA2_GAIN0	R/W	0b	
1	DA3_GAIN1	R/W	1b	Gain setting of differential amplifier 3 00b: 5V/V, 01b: 10V/V, 10b: 20V/V, 11b: 40V/V
0	DA3_GAIN0	R/W	0b	

7.1.15 Sense Block Control 2 Register: SNSCTL2 (Address=0x0E) [Default=0x00]

Figure 7.1-15 and Table 7.1-15 show Sense Block Control 2 register.

Figure 7.1-15 Sense Block Control 2 Register SNSCTL2

7	6	5	4	3	2	1	0
CAL_BCONN	BEMF_PH2	BEMF_PH1	BEMF_PH0	BEMF_SH	DA1_SH	DA2_SH	DA3_SH
R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-15 Sense Block Control 2 Register SNSCTL2 Descriptions

Bit	Field	Type	Default	Description
7	CAL_BCONN	R/W	0b	Input selection of BEMF sense amplifier during BEMF offset calibration 0b: The amplifier inputs are connected to the reference voltage of BEMF sense amplifier (DAREF) 1b: The amplifier inputs are connected to the phase selected by BEMF_PH bits
6	BEMF_PH2	R/W	0b	Detect phase selection of BEMF sense amplifier Note4 00xb: Select automatically from the input signals of the gate driver at every nSMPL falling edge 010b: Select by CMP1O and CMP2O pins (CMP1O, CMP2O)= (0,0): No selection, (0,1): Phase-A, (1,0): Phase-B, (1,1): Phase-C 011b: Select by CMP1O and CMP3O pins (CMP1O, CMP3O)= (0,0): No selection, (0,1): Phase-A, (1,0): Phase-B, (1,1): Phase-C 100b: No selection, 101b: Phase-A, 110b: Phase-B, 111b: Phase-C
5	BEMF_PH1	R/W	0b	
4	BEMF_PH0	R/W	0b	
3	BEMF_SH	R/W	0b	S/H control setting of BEMF sense amplifier Note4 0b: keep sampling, 1b: sampling during nSMPL signal=Lo & PWM ON after tGT
2	DA1_SH	R/W	0b	S/H control setting of differential amplifier 1 Note4 0b: keep sampling, 1b: sampling during nSMPL signal=Lo
1	DA2_SH	R/W	0b	S/H control setting of differential amplifier 2 Note4 0b: keep sampling, 1b: sampling during nSMPL signal=Lo
0	DA3_SH	R/W	0b	S/H control setting of differential amplifier 3 Note4 0b: keep sampling, 1b: sampling during nSMPL signal=Lo

Note4: Refer to the detail description of Differential Amplifiers for Current Sensing and BEMF sense amplifier

7.1.16 Sense Block Control 3 Register: SNSCTL3 (Address=0x0F) [Default=0x88]

Figure 7.1-16 and Table 7.1-16 show Sense Block Control 3 register.

Figure 7.1-16 Sense Block Control 3 Register SNSCTL3

7	6	5	4	3	2	1	0
CMP1_VTH3	CMP1_VTH2	CMP1_VTH1	CMP1_VTH0	CMP2_VTH3	CMP2_VTH2	CMP2_VTH1	CMP2_VTH0
R/W: 1b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 1b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-16 Sense Block Control 3 Register SNSCTL3 Descriptions

Bit	Field	Type	Default	Description
7	CMP1_VTH3	R/W	1b	Threshold voltage setting of Comparator 1 0000b: Disable, 0001b to 1111b: Threshold voltage= VDD /16 x CMP1_VTH
6	CMP1_VTH2	R/W	0b	
5	CMP1_VTH1	R/W	0b	
4	CMP1_VTH0	R/W	0b	
3	CMP2_VTH3	R/W	1b	Threshold voltage setting of Comparator 2 0000b: Disable, 0001b to 1111b: Threshold voltage= VDD /16 x CMP2_VTH
2	CMP2_VTH2	R/W	0b	
1	CMP2_VTH1	R/W	0b	
0	CMP2_VTH0	R/W	0b	

7.1.17 Sense Block Control 4 Register: SNSCTL4 (Address=0x10) [Default=0x80]

Figure 7.1-17 and Table 7.1-17 show Sense Block Control 4 register.

Figure 7.1-17 Sense Block Control 4 Register SNSCTL4

7	6	5	4	3	2	1	0
CMP3_VTH3	CMP3_VTH2	CMP3_VTH1	CMP3_VTH0	CAL_CONN	CAL_DA1	CAL_DA2	CAL_DA3/BEMF
R/W: 1b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-17 Sense Block Control 4 Register SNSCTL4 Descriptions

Bit	Field	Type	Default	Description
7	CMP3_VTH3	R/W	1b	Threshold voltage setting of Comparator 3 0000b: Disable, 0001b to 1111b: Threshold voltage= VDD /16 x CMP3_VTH
6	CMP3_VTH2	R/W	0b	
5	CMP3_VTH1	R/W	0b	
4	CMP3_VTH0	R/W	0b	
3	CAL_CONN	R/W	0b	Input selection of differential amplifier during DC offset calibration 0b: The amplifier inputs are connected to GND. 1b: The amplifier inputs are connected to the external shunt.
2	CAL_DA1	R/W	0b	Write 1b to enable DC offset calibration for differential amplifier 1. This bit is automatically reset to 0 after calibration is done.
1	CAL_DA2	R/W	0b	Write 1b to enable DC offset calibration for differential amplifier 2. This bit is automatically reset to 0 after calibration is done.
0	CAL_DA3/BEMF	R/W	0b	Write 1b to this bit to enable DC offset calibration for differential amplifier 3 if BEMF sensing is disabled (BEMF_EN=0b). Write 1b to this bit to enable DC offset calibration for BEMF sensing amplifiers if BEMF sensing is enabled (BEMF_EN=1b). This bit automatically resets to 0 after calibration is done

7.1.18 Sense Block Control 5 Register: SNSCTL5 (Address=0x11) [Default=0x00]

Figure 7.1-18 and Table 7.1-18 show Sense Block Control 5 register.

Figure 7.1-18 Sense Block Control 5 Register SNSCTL5

7	6	5	4	3	2	1	0
DIS_SADT	RESERVED11_6	CTL6_UNLOCK	RESERVED11_4	RESERVED11_3	MUX2	MUX1	MUX0
R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-18 Sense Block Control 5 Register SNSCTL5 Descriptions

Bit	Field	Type	Default	Description
7	DIS_SADT	R/W	0b	Write 1b to disable the adaptive dead time control function
6	RESERVED11_6	R/W	0b	Reserved. The write value should be 0b.
5	CTL6_UNLOCK	R/W	0b	Write 0b to ignore SNSCTL6 register write. Write 1b to unlock to allow SNSCTL6 register write.
4	RESERVED11_4	R/W	0b	Reserved. The write value should be 0b.
3	RESERVED11_3	R/W	0b	Reserved. The write value should be 0b.
2	MUX2	R/W	0b	Output selection of DA3O/MUX1 pin 000b: GND (pulldown: 330kΩ) 001b: VM monitor 010b: TEMP monitor 011b: Differential amplifier reference voltage 101b: Differential amplifier 1 output 110b: Differential amplifier 2 output In case of BEMF_EN=0b, 100b: Differential amplifier 3 output w/ 10kΩ 111b: Differential amplifier 3 output w/o 10kΩ In case of BEMF_EN=1b, 100b: BEMF sense amplifier output w/ 10kΩ 111b: BEMF sense amplifier output w/o 10kΩ
1	MUX1	R/W	0b	
0	MUX0	R/W	0b	

7.1.19 Sense Block Control 6 Register: SNSCTL6 (Address=0x12) [Default=0x40]

Figure 7.1-19 and Table 7.1-19 show Sense Block Control 6 register. CTL6_UNLOCK=1b is necessary to allow SNSCTL6 register write. After writing SNSCTL6 register, CTL6_UNLOCK should be set to 0b.

Figure 7.1-19 Sense Block Control 6 Register SNSCTL6

7	6	5	4	3	2	1	0
RESERVED12_7	BEMF_OFFSET	RESERVED12_5	RESERVED12_4	RESERVED12_3	RESERVED12_2	RESERVED12_1	GD_AOR
R/W: 0b	R/W: 1b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b	R/W: 0b

Table 7.1-19 Sense Block Control 6 Register SNSCTL6 Descriptions

Bit	Field	Type	Default	Description
7	RESERVED12_7	R/W	0b	Reserved. The write value should be 0b.
6	BEMF_OFFSET	R/W	1b	Data selection of BEMF sense amplifier DC offset 0b: calibration data, 1b: trimming data by shipping test This bit automatically sets to 0 after DC offset calibration for BEMF sense amplifier is done.
5	RESERVED12_5	R/W	0b	Reserved. The write value should be 0b.
4	RESERVED12_4	R/W	0b	Reserved. The write value should be 0b.
3	RESERVED12_3	R/W	0b	Reserved. The write value should be 0b.
2	RESERVED12_2	R/W	0b	Reserved. The write value should be 0b.
1	RESERVED12_1	R/W	0b	Reserved. The write value should be 0b.
0	GD_AOR	R/W	0b	Write 1b to enable the active override mode of the gate driver logic.

8 External Circuit

The external circuit in the case of 3 shunt sensorless FOC is shown as Figure 8-1. The recommended value of each external component is shown in Table 8-1.

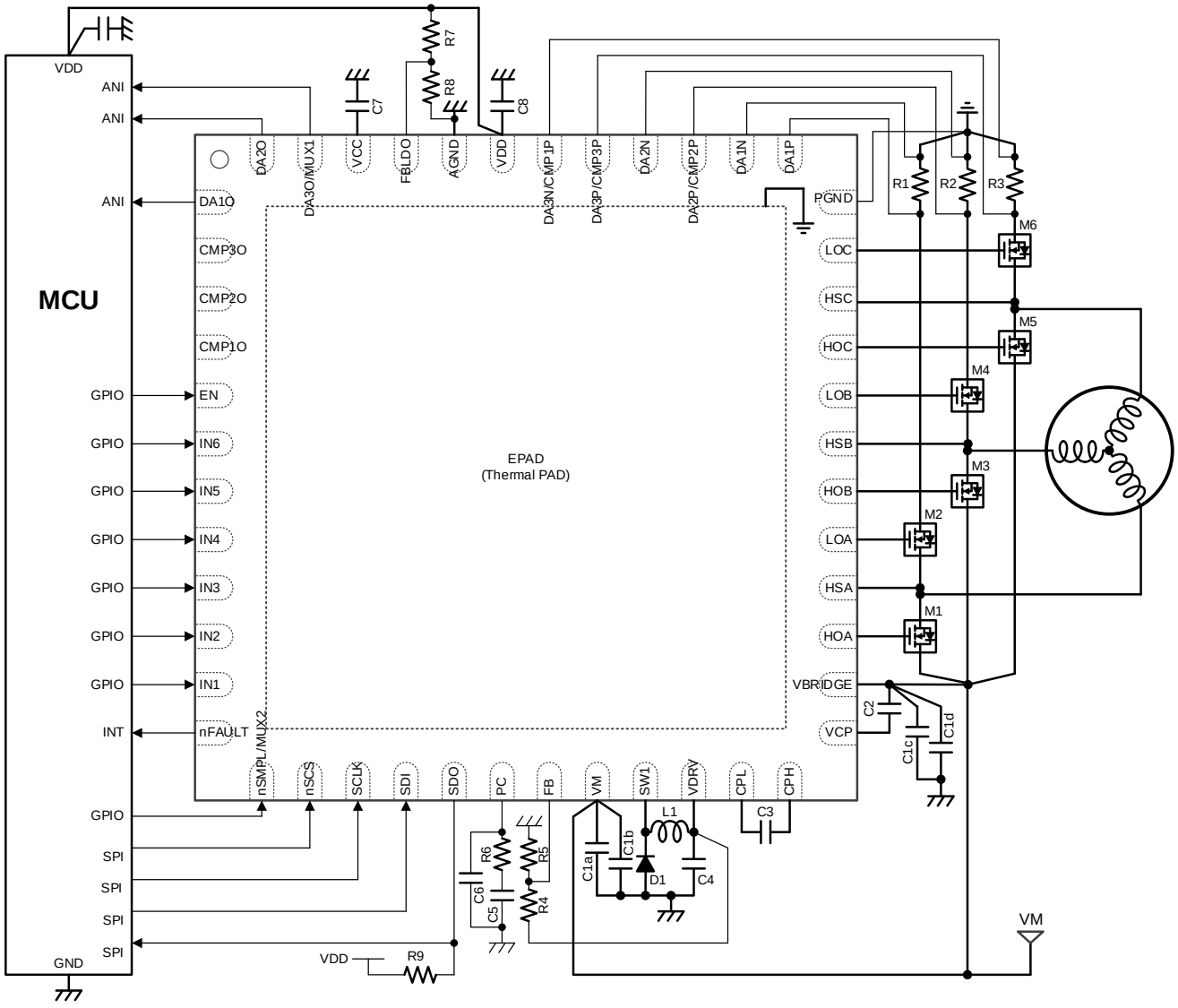


Figure 8-1 External Circuit Example – 3 Shunt Sensorless FOC

8 External Circuit (continued)

Table 8-1 The External Component List

Part No.	Recommended value	Ratings	Purpose	Notes
R1 to R3	Depend on application	Depend on application	Shunt resistance for current sense	
R4	48.7kΩ	-	Bleeder resistance for VDRV output voltage setting	1
R5	3.48kΩ	-	Bleeder resistance for VDRV output voltage setting	1
R6	60.4kΩ	-	Phase compensation resistance for the switching regulator	5
R7	160kΩ	-	Bleeder resistance for VDD output voltage setting	2
R8	91kΩ	-	Bleeder resistance for VDD output voltage setting	2
R9	DNP	-	Pullup resistance for SDO pin. Pullup function of MCU I/O can be used.	3
C1a	3x 4.7µF	100V	Bypass capacitance for VM terminal	6
C1b	0.1µF	100V	Bypass capacitance for VM terminal	6
C1c	4.7µF	100V	Bypass capacitance for VBRIDGE terminal	6
C1d	0.1µF	100V	Bypass capacitance for VBRIDGE terminal	6
C2	2.2µF	25V	Bypass capacitance for VCP terminal	4
C3	0.22µF	100V	Pumping capacitance for the charge pump	4
C4	10µF	25V	Output capacitance for the switching regulator, VDRV terminal	5
C5	2200pF	10V	Phase compensation capacitance for the switching regulator	5
C6	DNP	10V	Phase compensation capacitance for the switching regulator	
C7	22µF	10V	Output capacitance for the linear regulator, VCC5V terminal	
C8	22µF	10V	Output capacitance for the linear regulator, VDD terminal	
M1 to M6	Depend on application	Depend on application	Power MOSFET for the motor drive	
L1	22µH or 33µH	>2A	Coil for the switching regulator	
D1	0.6V	100V, >2A	Schottky rectifier diode for the switching regulator	
MCU	RX13T or RL78/G1F	3.3 or 5V operation	Microcontroller for the motor drive control	

Please refer to the application note for the details to select the parts.

Note1: VDRV output voltage is 12V with these resistors.

Note2: VDD output voltage is 3.310V with these resistors.

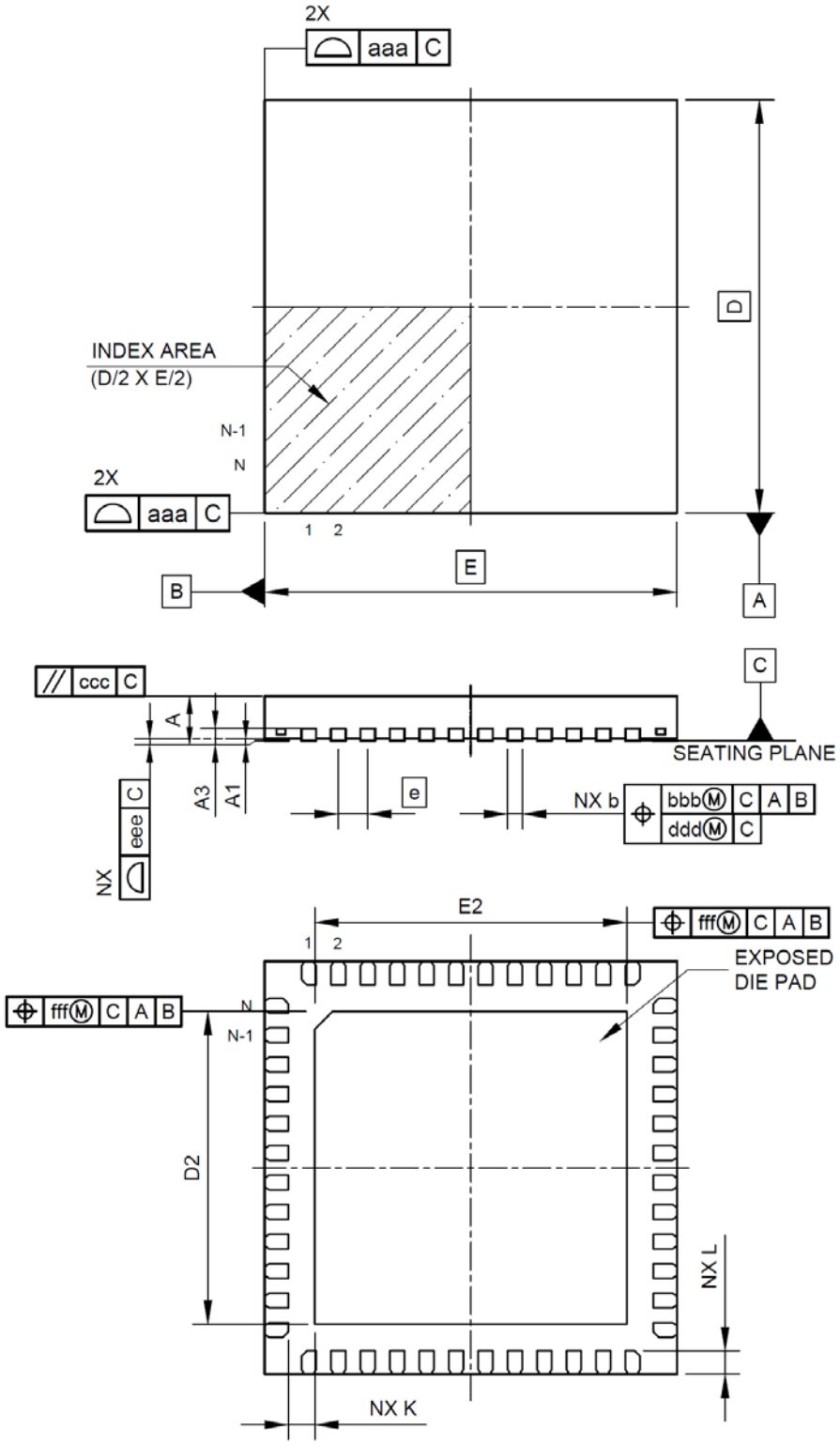
Note3: In some cases, the external pullup resistor for SDO pin is required depending on the SCLK period and the load capacitance including the parasitic capacitance.

Note4: Please consider the effective capacitance. The smaller C3 causes the larger voltage drop of VCP. The smaller C2 causes the larger voltage ripple of VCP.

Note5: Please select the suitable value of R6 and C5 depending on C4 effective capacitance.

Note6: The suitable capacitance depends on the constraints of the application and characteristic.

9 Package Specification



Symbols	Dimension in Millimeters		
	Min.	Typ.	Max.
A	-	-	0.80
A1	0.00	0.02	0.05
A3	0.203 REF.		
b	0.20	0.25	0.30
D	-	7.00	-
E	-	7.00	-
e	-	0.50	-
N	48		
L	0.30	0.40	0.50
K	0.20	-	-
D2	5.20	5.30	5.40
E2	5.20	5.30	5.40
aaa	0.15		
bbb	0.10		
ccc	0.10		
ddd	0.05		
eee	0.08		
fff	0.10		

10 Revision History

Rev.	Date	Summary of Revised	Object Page
1.00	5-Oct-23	Initial release	All

Notice

1. Descriptions of circuits, software and other related information in this document are provided only to illustrate the operation of semiconductor products and application examples. You are fully responsible for the incorporation or any other use of the circuits, software, and information in the design of your product or system. Renesas Electronics disclaims any and all liability for any losses and damages incurred by you or third parties arising from the use of these circuits, software, or information.
2. Renesas Electronics hereby expressly disclaims any warranties against and liability for infringement or any other claims involving patents, copyrights, or other intellectual property rights of third parties, by or arising from the use of Renesas Electronics products or technical information described in this document, including but not limited to, the product data, drawings, charts, programs, algorithms, and application examples.
3. No license, express, implied or otherwise, is granted hereby under any patents, copyrights or other intellectual property rights of Renesas Electronics or others.
4. You shall be responsible for determining what licenses are required from any third parties, and obtaining such licenses for the lawful import, export, manufacture, sales, utilization, distribution or other disposal of any products incorporating Renesas Electronics products, if required.
5. You shall not alter, modify, copy, or reverse engineer any Renesas Electronics product, whether in whole or in part. Renesas Electronics disclaims any and all liability for any losses or damages incurred by you or third parties arising from such alteration, modification, copying or reverse engineering.
6. Renesas Electronics products are classified according to the following two quality grades: “Standard” and “High Quality”. The intended applications for each Renesas Electronics product depends on the product’s quality grade, as indicated below.
   "Standard": Computers; office equipment; communications equipment; test and measurement equipment; audio and visual equipment; home electronic appliances; machine tools; personal electronic equipment; industrial robots; etc.
   "High Quality": Transportation equipment (automobiles, trains, ships, etc.); traffic control (traffic lights); large-scale communication equipment; key financial terminal systems; safety control equipment; etc.

   Unless expressly designated as a high reliability product or a product for harsh environments in a Renesas Electronics data sheet or other Renesas Electronics document, Renesas Electronics products are not intended or authorized for use in products or systems that may pose a direct threat to human life or bodily injury (artificial life support devices or systems; surgical implantations; etc.), or may cause serious property damage (space system; undersea repeaters; nuclear power control systems; aircraft control systems; key plant systems; military equipment; etc.). Renesas Electronics disclaims any and all liability for any damages or losses incurred by you or any third parties arising from the use of any Renesas Electronics product that is inconsistent with any Renesas Electronics data sheet, user’s manual or other Renesas Electronics document.
7. No semiconductor product is absolutely secure. Notwithstanding any security measures or features that may be implemented in Renesas Electronics hardware or software products, Renesas Electronics shall have absolutely no liability arising out of any vulnerability or security breach, including but not limited to any unauthorized access to or use of a Renesas Electronics product or a system that uses a Renesas Electronics product. RENESAS ELECTRONICS DOES NOT WARRANT OR GUARANTEE THAT RENESAS ELECTRONICS PRODUCTS, OR ANY SYSTEMS CREATED USING RENESAS ELECTRONICS PRODUCTS WILL BE INVULNERABLE OR FREE FROM CORRUPTION, ATTACK, VIRUSES, INTERFERENCE, HACKING, DATA LOSS OR THEFT, OR OTHER SECURITY INTRUSION (“Vulnerability Issues”). RENESAS ELECTRONICS DISCLAIMS ANY AND ALL RESPONSIBILITY OR LIABILITY ARISING FROM OR RELATED TO ANY VULNERABILITY ISSUES. FURTHERMORE, TO THE EXTENT PERMITTED BY APPLICABLE LAW, RENESAS ELECTRONICS DISCLAIMS ANY AND ALL WARRANTIES, EXPRESS OR IMPLIED, WITH RESPECT TO THIS DOCUMENT AND ANY RELATED OR ACCOMPANYING SOFTWARE OR HARDWARE, INCLUDING BUT NOT LIMITED TO THE IMPLIED WARRANTIES OF MERCHANTABILITY, OR FITNESS FOR A PARTICULAR PURPOSE.
8. When using Renesas Electronics products, refer to the latest product information (data sheets, user’s manuals, application notes, “General Notes for Handling and Using Semiconductor Devices” in the reliability handbook, etc.), and ensure that usage conditions are within the ranges specified by Renesas Electronics with respect to maximum ratings, operating power supply voltage range, heat dissipation characteristics, installation, etc. Renesas Electronics disclaims any and all liability for any malfunctions, failure or accident arising out of the use of Renesas Electronics products outside of such specified ranges.
9. Although Renesas Electronics endeavors to improve the quality and reliability of Renesas Electronics products, semiconductor products have specific characteristics, such as the occurrence of failure at a certain rate and malfunctions under certain use conditions. Unless designated as a high reliability product or a product for harsh environments in a Renesas Electronics data sheet or other Renesas Electronics document, Renesas Electronics products are not subject to radiation resistance design. You are responsible for implementing safety measures to guard against the possibility of bodily injury, injury or damage caused by fire, and/or danger to the public in the event of a failure or malfunction of Renesas Electronics products, such as safety design for hardware and software, including but not limited to redundancy, fire control and malfunction prevention, appropriate treatment for aging degradation or any other appropriate measures. Because the evaluation of microcomputer software alone is very difficult and impractical, you are responsible for evaluating the safety of the final products or systems manufactured by you.
10. Please contact a Renesas Electronics sales office for details as to environmental matters such as the environmental compatibility of each Renesas Electronics product. You are responsible for carefully and sufficiently investigating applicable laws and regulations that regulate the inclusion or use of controlled substances, including without limitation, the EU RoHS Directive, and using Renesas Electronics products in compliance with all these applicable laws and regulations. Renesas Electronics disclaims any and all liability for damages or losses occurring as a result of your noncompliance with applicable laws and regulations.
11. Renesas Electronics products and technologies shall not be used for or incorporated into any products or systems whose manufacture, use, or sale is prohibited under any applicable domestic or foreign laws or regulations. You shall comply with any applicable export control laws and regulations promulgated and administered by the governments of any countries asserting jurisdiction over the parties or transactions.
12. It is the responsibility of the buyer or distributor of Renesas Electronics products, or any other party who distributes, disposes of, or otherwise sells or transfers the product to a third party, to notify such third party in advance of the contents and conditions set forth in this document.
13. This document shall not be reprinted, reproduced or duplicated in any form, in whole or in part, without prior written consent of Renesas Electronics.
14. Please contact a Renesas Electronics sales office if you have any questions regarding the information contained in this document or Renesas Electronics products.

(Note1) “Renesas Electronics” as used in this document means Renesas Electronics Corporation and also includes its directly or indirectly controlled subsidiaries.
(Note2) “Renesas Electronics product(s)” means any product developed or manufactured by or for Renesas Electronics.

(Rev.5.0-1 October 2020)

Corporate Headquarters

TOYOSU FORESIA, 3-2-24 Toyosu,
Koto-ku, Tokyo 135-0061, Japan
www.renesas.com

Trademarks

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.

Contact information

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales office, please visit:
www.renesas.com/contact/.

X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for Gate Drivers ***category:***

Click to view products by Renesas ***manufacturer:***

Other Similar products are found below :

56956 57.404.7355.5 LT4936 57.904.0755.0 0131700000 LTP70N06 LVP640 5J0-1000LG-SIL LY2-US-AC240 LY3-UA-DC24 LZNQ2-US-DC12 LZP40N10 60100564 60249-1-CUT-TAPE 0134220000 6035 60713816 61161-90 6131-204-23149P 6131-205-17149P 6131-209-15149P 6131-218-17149P 6131-220-21149P 6131-260-2358P 6131-265-11149P CS1HCPU63 6150-5001 CSB4 CSK-38-60006 CSK-38-60008 621A 622-4053LF 6273 M40N08MA-H M55155/29XH06 64-807 65-1930-6 CV500ISB02 M83723/88Y1407N CWD012-2 CWD03-3 CX3225SB16934D0PPSC2 CX5032GB10000D0PPS02 687-772NF1 70.140.1653 70.200.0653.0 703001B01F060 70-3601 706006D02F0601 706210