

-40 °C to +150 °C

3 V to 42 V

17 µA(Typ)

±2 / ±2.5 %

1.2 V to 16 V

500 mA

For Automotive 500 mA Adjustable Output

LDO Regulators

BD800M5WFPJ-C BD800M5WHFP-C BD800M5WFP2-C

General Description

The BD800M5Wxxx-C series are linear regulators designed as low current consumption products for power supplies in various automotive applications.

These products are designed for up to 45 V as an absolute maximum voltage and to operate until 500 mA for the output current with low current consumption 17 μ A (Typ). These can regulate the output with a very high accuracy (Note 1), ±2 % (BD800M5WFPJ-C, BD800M5WHFP-C) and ±2.5 % (BD800M5WFP2-C). The output voltage can be adjusted between 1.2 V and 16 V by an external resistive divider connected to the adjustment pin. These regulators are therefore an ideal for any applications requiring a direct connection to the battery and a low current consumption.

A logical "HIGH" at the EN Pin turns on the device, and in the other side, the devices are controlled to disable by a logical "LOW" input to the EN Pin.

The devices feature the integrated Over Current Protection to protect the device from a damage caused by a short-circuiting or an overload. These products also integrate Thermal Shutdown Protection to avoid the damage by overheating.

Furthermore, low ESR ceramic capacitors are sufficiently applicable for the phase compensation. (*Note 1*) The tolerance of feedback resistor is not included.

Features

- AEC-Q100 Qualified^(Note 2)
- EN Function (Output Shutdown Function)
- Over Current Protection (OCP)
- Thermal Shutdown Protection (TSD) (Note 2) Grade 1

Applications

- Power Train
- Body
- Audio System
- Navigation System

Typical Application Circuit

Components Externally Connected Capacitor: 0.1 μ F \leq C_{IN}, 1.47 μ F \leq C_{OUT} (Min) ^(Note 3) Resistor: 5 k $\Omega \leq$ R₁ \leq 200 k Ω ^(Note 4) ^(Note 5) V_{ADJ} (Typ): 0.65 V

$$R_2 = R_1 \left(\frac{V_{OUT}}{V_{ADJ}} - 1 \right)$$

(Note 3) Electrolytic, tantalum and ceramic capacitors can be used.

- (Note 4) The tolerance of feedback resistor is not included in the accuracy of output voltage.
- (Note 5) The value of a feedback resistor R_1 must be within this range.
 - R_2 value is defined by following the formula using the limitation of R_1 .
- (Note 6) If it needs better transient characteristic, insert a capacitor between the VOUT and ADJ Pins. Refer to <u>Typical Application and Layout Example</u> for the details such as equations.

Key Specifications

- Wide Temperature Range (Tj):
- Wide Operating Input Range:
- Low Current Consumption:
- Output Current Capability:
- High Output Voltage Accuracy:
- Output Voltage:

Packages

TO252-J5 HRP5 TO263-5 W(Typ) x D(Typ) x H(Max) 6.60 mm x 10.10 mm x 2.38 mm 9.395 mm x 10.540 mm x 2.005 mm 10.16 mm×15.10 mm×4.70 mm

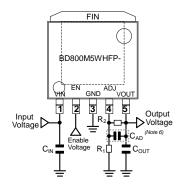




TO252-J5

TO263-5

HRP5



OProduct structure : Silicon integrated circuit OThis product has no designed protection against radioactive rays.

Contents

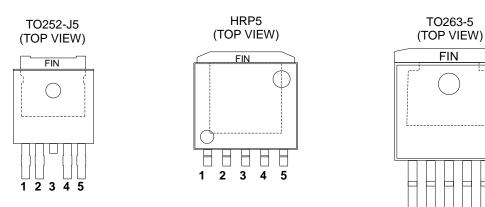
General Description	1
Features	1
Applications	
Key Specifications	
Typical Application Circuit	
Pin Configurations Pin Descriptions	
Block Diagram	
Discription of Blocks	
Absolute Maximum Ratings	
Thermal Resistance	
Operating Conditions	
Electrical Characteristics	
LDO Function	8
Enable Function	
Typical Performance Curves	
Figure 1. Output Voltage vs Input Voltage	
Figure 2. Output Voltage vs Input Voltage -Enlarged view	
Figure 3. Output Voltage vs Junction Temperature (I _{OUT} = 0.5 mA)	
Figure 4. Current Consumption + Enable Bias Current vs Input Voltage Figure 5. Current Consumption + Enable Bias Current vs Junction Temperature	
Figure 6. Current Consumption + Enable Bias Current vs Sunction remperature	
Figure 7. Output Voltage vs Output Current (Over Current Protection)	
Figure 8. Shutdown Current vs Junction Temperature $(V_{EN} = 0 V)$	
Figure 9. Dropout Voltage vs Output Current ($V_{IN} = 4.75$ V)	
Figure 10. Output Voltage vs Junction Temperature (Thermal Shutdown Protection)	
Figure 11. Output Voltage vs EN Input Voltage	11
Figure 12. EN Input Voltage vs Junction Temperature	
Figure 13. Enable Bias Current vs Junction Temperature	
Figure 14. Ripple Rejection (Vripple = 1 Vrms, I _{OUT} = 100 mA)	12
Figure 15. Line Regulation ($V_{IN} = 6 V \rightarrow 28 V$)	12
Figure 16. Load Regulation $(I_{OUT} = 0.5 \text{ mA} \rightarrow 400 \text{ mA})$	
Figure 17. Line Transient Response $(V_{IN} = 0 \ V \rightarrow 16 \ V)$	
Figure 18. Line Transient Response $(V_{IN} = 6 V \rightarrow 16 V)$	
Figure 19. Load Transient Response $(I_{OUT} = 1 \text{ mA} \leftrightarrow 500 \text{ mA})$	
Figure 20. Load Transient Response (I _{OUT} = 10 mA ↔ 500 mA)	
Figure 21. EN ON/OFF Sequence $(V_{EN} = 0 V \leftrightarrow 5 V, Tj = +25 °C)$	
Figure 22. EN ON/OFF Sequence $(V_{EN} = 0 V \leftrightarrow 5 V, Tj = +150 °C)$ Figure 23. EN ON/OFF Sequence $(V_{EN} = 0 V \leftrightarrow 5 V, Tj = -40 °C)$	17
Figure 23. EN ON/OFF Sequence (V _{EN} = 0 V \leftrightarrow 5 V, Tj = -40 °C) Figure 24. V _{IN} ON/OFF Sequence (V _{IN} = 0 V \leftrightarrow 13.5 V, Tj = +25 °C)	
Figure 24. V_{IN} ON/OFF Sequence ($V_{IN} = 0.7 \leftrightarrow 13.5 V$, $T_j = +25$ C) Figure 25. V_{IN} ON/OFF Sequence ($V_{IN} = 0.7 \leftrightarrow 13.5 V$, $T_j = +150$ °C)	19 20
Figure 26. V _{IN} ON/OFF Sequence (V _{IN} = 0 V \leftrightarrow 13.5 V, Tj = -40 °C)	
Measurement Circuit for Typical Performance Curves	
Application and Implementation	
Selection of External Components	
Input Pin Capacitor	
Output Pin Capacitor	23
Typical Application and Layout Example	
Surge Voltage Protection for Linear Regulators	
Positive Surge to the Input	
Negative Surge to the Input	
Reverse Voltage Protection for Linear Regulators	
Protection against Reverse Input/Output Voltage	
Protection against Reverse Input/Output Voltage – continued	
Protection against Input Reverse Voltage Protection against Input Reverse Voltage - continued	
Protection against Reverse Output Voltage when Output Connect to an Inductor	20
Power Dissipation	
TO252-J5	
HRP5 29	
TO263-5	30
Thermal Design	
I/O Equivalence Circuits	33
Operational Notes	
1. Reverse Connection of Power Supply	
2. Power Supply Lines	34

3.	Ground Voltage	34
4.	Ground Voltage Ground Wiring Pattern	34
5.	Recommended Operating Conditions	34
6.	Inrush Current	
7.	Thermal Consideration	34
8.	Testing on Application Boards	34
9.	Inter-pin Short and Mounting Errors	34
10.	Unused Input Pins	34
11.	Regarding the Input Pin of the IC	35
12.	Ceramic Capacitor	35
13.	Thermal Shutdown Protection Circuit(TSD)	
14.	Over Current Protection Circuit (OCP)	35
15.	Enable Pin	35
Ordering	Information	36
Marking	Diagrams	36
Physical	Dimension and Packing Information	37
) History	

1 2 3

45

Pin Configurations

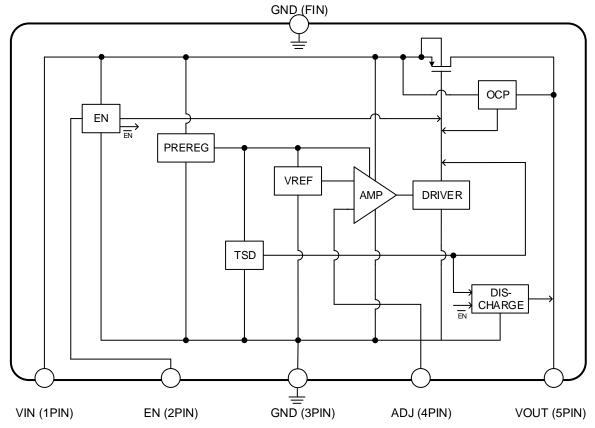


Pin Descriptions

Pin N	No. Pin Name	Function	Descriptions
1	VIN	Input Supply Voltage Pin	It is necessary to use a capacitor with a capacitance of 0.1 μ F or higher between the VIN Pin and GND. The detail of a selection is described in <u>Selection of External Components</u> . If the inductance of power supply line is high, please adjust input capacitor value.
2	EN	Control Output ON/OFF Pin	A logical "HIGH" ($V_{EN} \ge 2.0$ V) at the EN Pin enables the device and "LOW" ($V_{EN} \le 0.8$ V) at the EN Pin disables the device.
3	GND	Ground Pin	Ground
4	ADJ	Adjustment Pin For Output Voltage	Connect an external resistor to adjust output voltage.
5	VOUT	Output Voltage Pin	Connect an external resistor to adjust output voltage. It is necessary to use a capacitor with a capacitance of 1.47 μ F(Min) or higher between the VOUT Pin and GND. The detail of a selection is described in <u>Selection of External Components</u>
FIN	N GND	Ground Pin	Ground. This pin should be connected to Analog ground / Power ground or Heat Sink.

Block Diagram

TO252-J5 / HRP5 / TO263-5



Description of Blocks

Block Name	Function	Description of Blocks			
EN	Control Output Voltage ON/OFF	A logical "HIGH" (V _{EN} \ge 2.0 V) at the EN Pin enables the device and "LOW" (V _{EN} \le 0.8 V) at the EN Pin disables the device.			
PREREG	Internal Power Supply	Power supply for internal circuit.			
TSD	Thermal Shutdown Protection	In case maximum power dissipation is exceeded or the ambient temperature is higher than the Maximum Junction Temperature, overheating causes the chip temperature (Tj) to rise. The TSD protection circuit detects this and forces the output to turn off in order to protect the device from overheating (Typ:175 °C). When the junction temperature decreases to low, the output turns on automatically.			
VREF	Reference Voltage Generate the reference voltage.				
AMP	Error Amplifier	The error amplifier amplifies the difference between the feedback voltage of the output voltage and the reference voltage.			
DRIVER	Output MOSFET Driver	Drive the output MOSFET (Power Tr.)			
OCP	Over Current Protection	If the output current increases higher than the maximum output current, it is limited by Over Current Protection in order to protect the device from a damage caused by an over current (Typ:1400 mA). While this block is operating, the output voltage may decrease because the output current is limited. If an abnormal state is removed and the output current value returns to normal, the output voltage also returns to normal state.			
DISCHARGE	Output Discharge Function	Output pin is discharged when EN = Low input or TSD detected.(Typ:4 k Ω)			

Absolute Maximum Ratings

Parameter	Symbol	Ratings	Unit
Supply Voltage ^(Note 1)	V _{IN}	-0.3 to +45	V
EN Pin Voltage ^(Note 2)	V _{EN}	-0.3 to +45	V
Output Pin Voltage	V _{OUT}	-0.3 to +20 (≤ V _{IN} + 0.3)	V
ADJ Pin Voltage	V _{ADJ}	-0.3 to +7	V
Junction Temperature Range	Tj	-40 to +150	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Maximum Junction Temperature	Tjmax	150	°C
ESD Withstand Voltage (HBM) ^(Note 3)	$V_{ESD_{HBM}}$	±2000	V
ESD Withstand Voltage (CDM) ^(Note 4)	V_{ESD_CDM}	±750	V

Caution 1: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Caution 2: Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, design a PCB with thermal resistance and power dissipation taken into consideration by increasing board size and copper area so as not to exceed the maximum junction temperature rating.

(Note 1) Do not exceed Tjmax.

(*Note 3*) ESD susceptibility Human Body Model "HBM"; base on ANSI/ESDA/JEDEC JS001 (1.5 k Ω , 100 pF). (*Note 4*) ESD susceptibility Charged Device Model "CDM"; base on JEDEC JESD22-C101.

Thermal Resistance (Note 5)

Decemeter	Symbol	Thermal Res	Linit	
Parameter	Symbol	1s ^(Note 7)	2s2p ^(Note 8)	Unit
TO252-J5	· ·			
Junction to Ambient	θ _{JA}	120.1	24.4	°C/W
Junction to Top Characterization Parameter ^(Note 6)	Ψ_{JT}	19	3	°C/W
HRP5	· ·			
Junction to Ambient	θ _{JA}	91.3	21.4	°C/W
Junction to Top Characterization Parameter ^(Note 6)	Ψ_{JT}	8	3	°C/W
TO263-5	· ·		·	
Junction to Ambient	θ _{JA}	80.2	21.8	°C/W
Junction to Top Characterization Parameter ^(Note 6)	Ψ_{JT}	10	2	°C/W

(Note 5) Based on JESD51-2A(Still-Air). Using BD800M5WFPJ-C, BD800M5WHFP-C, BD800M5WFP2-C Chips.

(Note 6) The thermal characterization parameter to report the difference between junction temperature and the temperature at the top center of the outside surface of the component package.

(Note 7) Using a PCB board based on JESD51-3. (Note 8) Using a PCB board based on JESD51-5, 7

vote 8) Using a PCB board based (JN JESD51-5, 7.					
Layer Number of Measurement Board	Material	Board Size				
Single	FR-4	114.3 mm x 76.2 mm x	(1.57 mmt			
Тор						
Copper Pattern	Thickness					
Footprints and Traces	70 µm					
Layer Number of	Material	Board Size		Thermal \		
Measurement Board				Pitch	L	Diameter
4 Layers	FR-4	114.3 mm x 76.2 mm	x 1.6 mmt	1.20 mm	¢	0.30 mm
Тор		2 Internal Laye	ers	Botto	om	
Copper Pattern	Thickness	Copper Pattern	Thickness	Copper Pattern	n	Thickness
Footprints and Traces	70 µm	74.2 mm x 74.2 mm	35 µm	74.2 mm x 74.2 n	nm	70 µm
Note 9) This thermal via connects v	with the conner na	attorn of all lavors				

(Note 9) This thermal via connects with the copper pattern of all layers.

Operating Conditions(-40°C \leq Tj \leq +150°C)

Parameter	Symbol	Min	Max	Unit
Input Voltage ^(Note 1)	V _{IN}	$V_{OUT}(Max) + \Delta V d(Max)$	42	V
Start-Up Voltage ^(Note 2)	V _{IN Start-Up}	3	-	V
Output Voltage	V _{OUT}	1.2	16.0	V
Feedback Resistor ADJ vs GND ^(Note 3)	R ₁	5	200	kΩ
Output Control Voltage	V _{EN}	0	42	V
Output Current	Ι _{Ουτ}	0	500	mA
Input Capacitor ^(Note4)	C _{IN}	0.1	-	μF
Output Capacitor ^(Note 5)	C _{OUT}	1.47	1000	μF
Output Capacitor Equivalent Series Resistance	ESR(C _{OUT})	-	5	Ω
VOUT-ADJ Capacitor	C _{ADJ}	-	1000	pF
Operating Temperature	Та	-40	+125	°C

(Note 1) Minimum Input Voltage must be 3.3 V or more.

Please consider that the output voltage would be dropped (Dropout voltage ΔVd) by the output current. (*Note 2*) If V_{OUT} setting is 3 V or less, V_{OUT} (Min) = 90 % × V_{OUT} (Typ) when V_{IN} = 3 V, I_{OUT} = 0 mA. (*Note 3*) If it needs better transient characteristic, insert a capacitor between the VOUT and ADJ Pins. Refer to <u>Typical Application and Layout Example</u> for the details such as equations. (Note 4) If the inductance of power supply line is high, please adjust input capacitor value.

(Note 5) Set the value of the capacitor so that it does not fall below the minimum value. Take into consideration the temperature characteristics and DC device characteristics.

Electrical Characteristics

LDO Function (V_{OUT} setting = 5 V, R₁ = 120 k Ω , R₂ = 803 k Ω) Unless otherwise specified, Tj = -40 °C to +150 °C, V_{IN} = 13.5 V, V_{EN} = 5 V, I_{OUT} = 0 mA Typical values are defined at Tj = 25 °C, V_{IN} = 13.5 V

Parameter	Symbol	Limits			Unit	Conditions
Falameter	Symbol	Min	Тур	Max	Unit	Conditions
Shutdown Current	I _{SHUT}	-	1	5	μA	V _{EN} = 0 V Tj ≤ 125 °C
		-	17	34	μA	l _{OUT} ≤ 500 mA Tj ≤ 25 °C
		-	17	43	μA	I _{OUT} ≤ 500 mA Tj ≤ 85 °C
Current Consumption ^(Note 1)	I _{CC}	-	17	46	μA	l _{o∪⊤} ≤ 500 mA Tj ≤ 105 °C
		-	17	49	μA	l _{o∪⊤} ≤ 500 mA Tj ≤ 125 °C
		-	17	53	μA	I _{OUT} ≤ 500 mA Tj ≤ 150 °C
		0.637	0.650	0.663	V	TO252-J5, HRP5 package 6 V \leq V _{IN} \leq 42 V 0.5 mA \leq I _{OUT} \leq 400 mA
Reference Voltage	V _{ADJ}	0.633	0.650	0.667	V	TO263-5 package $6 V \le V_{IN} \le 42 V$ $0.5 \text{ mA} \le I_{OUT} \le 400 \text{ mA}$
5		-	0.30	0.52	V	$V_{\rm IN} = V_{\rm OUT} \times 0.95$ $I_{\rm OUT} = 300 \text{ mA}$
Dropout Voltage	ΔVd	-	0.50	0.87	V	$V_{IN} = V_{OUT} \times 0.95$ $I_{OUT} = 500 \text{ mA}$
Ripple Rejection	R.R.	60	70	-	dB	f = 120 Hz Vripple = 1 Vrms I _{OUT} = 100 mA
	Deg	-	0.02	0.40	% × V _{OUT}	6 V ≤ V _{IN} ≤ 28 V Tj ≤ 125 °C
Line Regulation	Reg.I	-	0.02	0.60	% х V _{оит}	$6 V \le V_{IN} \le 28 V$ Tj $\le 150 °C$
Lood Pogulation	Beat	-	0.02	0.40	% × V _{OUT}	0.5 mA ≤ I _{OUT} ≤ 400 mA Tj ≤ 125 °C
Load Regulation	Reg.L	-	0.02	0.60	% × V _{OUT}	0.5 mA ≤ I _{OUT} ≤ 400 mA Tj ≤ 150 °C
Overload Current Protection	I _{OUT(OCP)}	750	1400	-	mA	$6 V \le V_{IN} \le 42 V$ $V_{OUT} = 90 \% \times V_{OUT}(Typ)$
Thermal Shutdown Temperature	Tj _(TSD)	151	175	-	°C	-

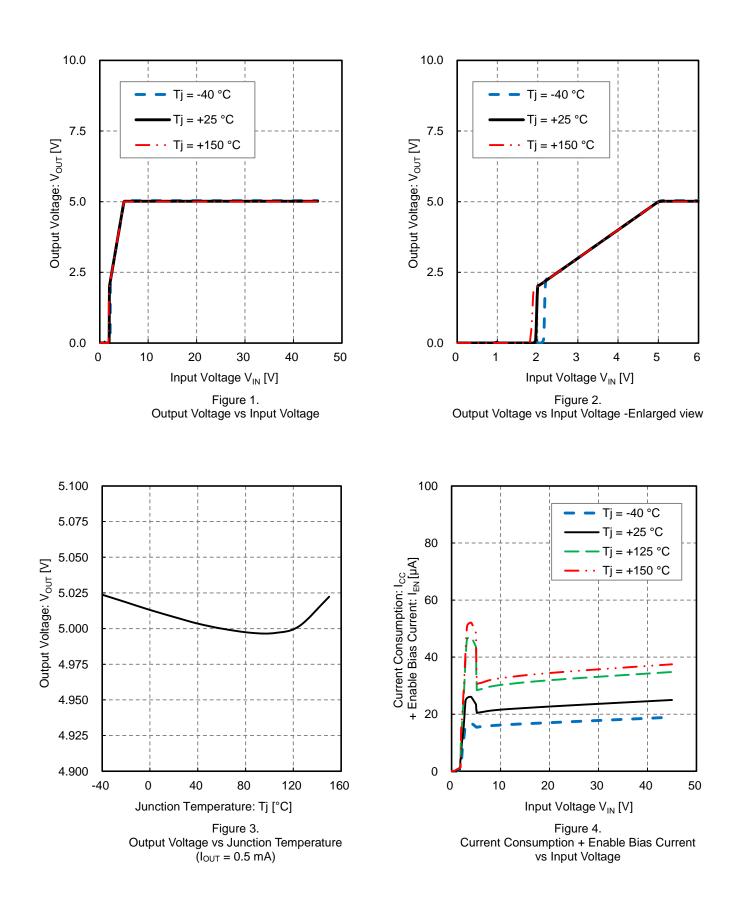
(Note 1) It does not contain the current of R_1 and R_2 .

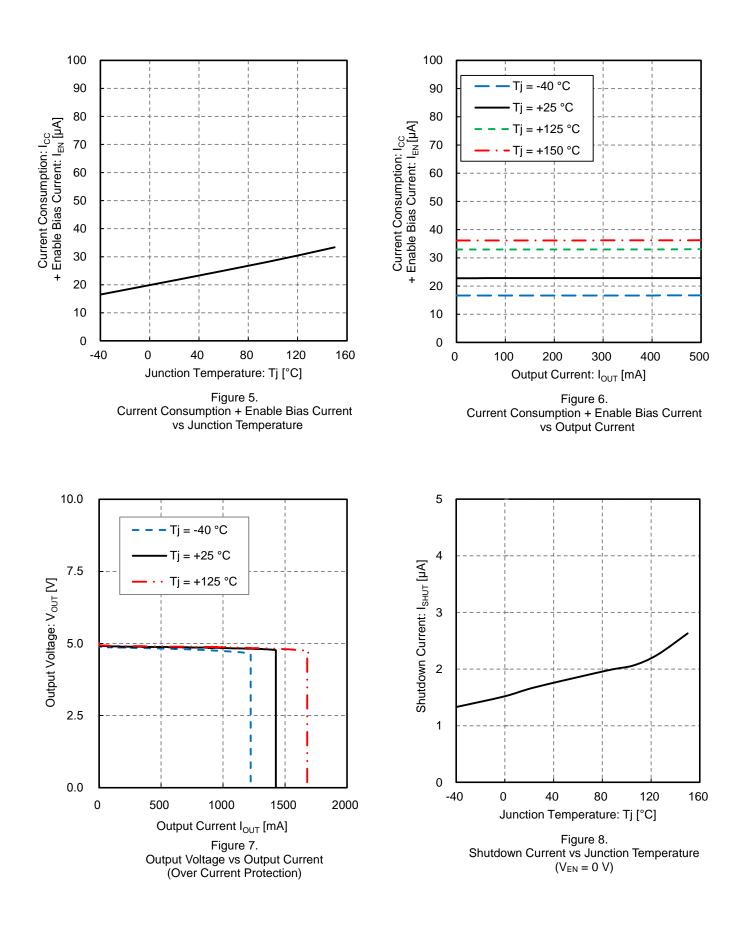
Enable Function

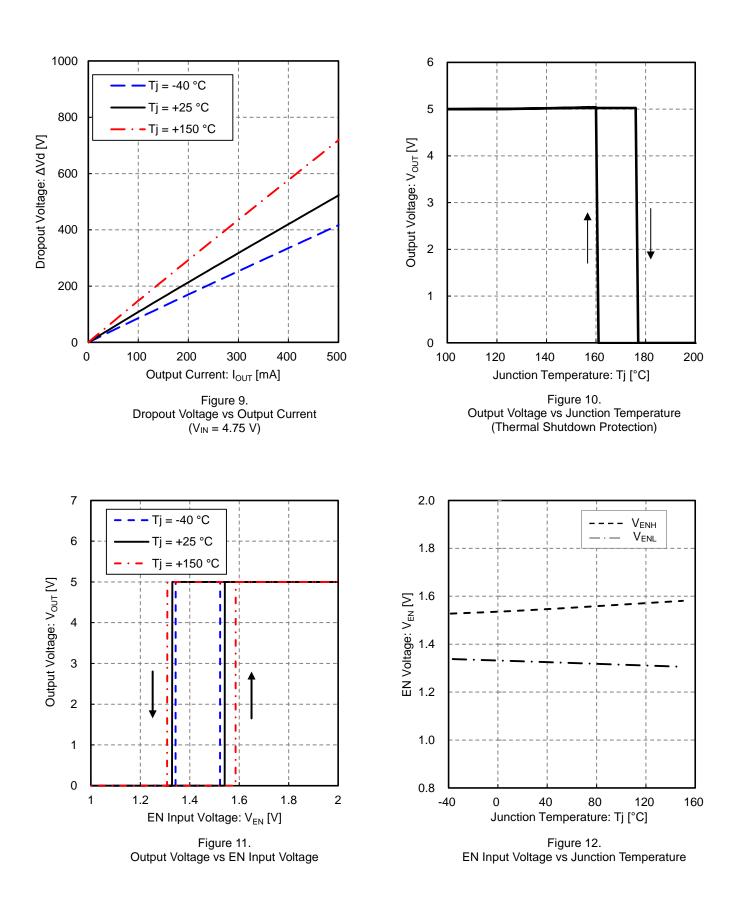
Unless otherwise specified, Tj = -40 °C to +150 °C, V_{IN} = 13.5 V, V_{EN} = 5 V, I_{OUT} = 0 mA Typical values are defined at Tj = 25 °C, V_{IN} = 13.5 V

Parameter	Symbol	Limits			Unit	Conditions
	Symbol	Min	Тур	Max	Unit	Conditions
Enable Mode Voltage	V _{ENH}	2.0	-	42.0	V	-
Disable Mode Voltage	V _{ENL}	0	-	0.8	V	-
Enable Bias Current	I _{EN}	-	4	8	μA	-

Typical Performance Curves



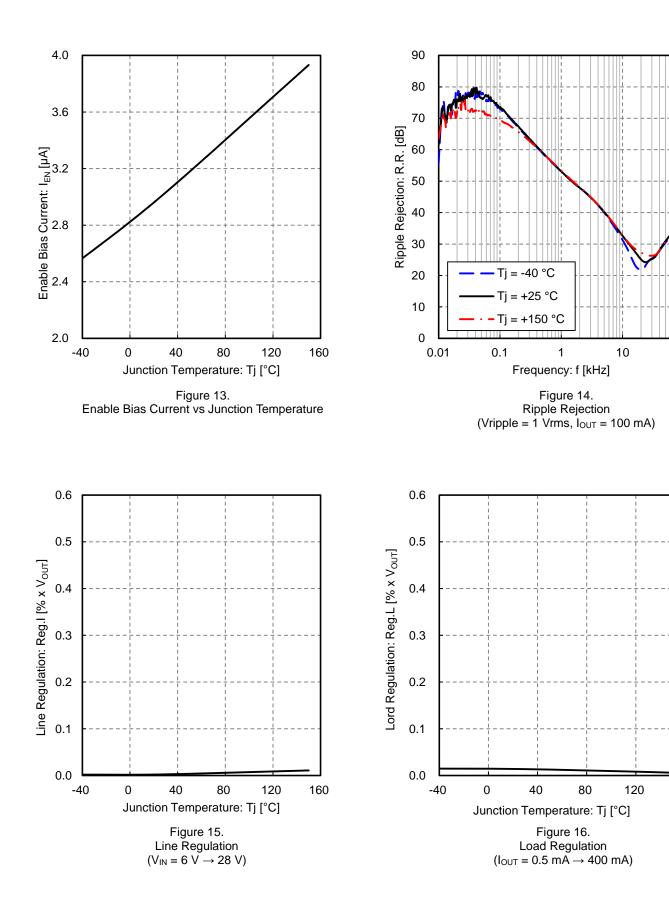


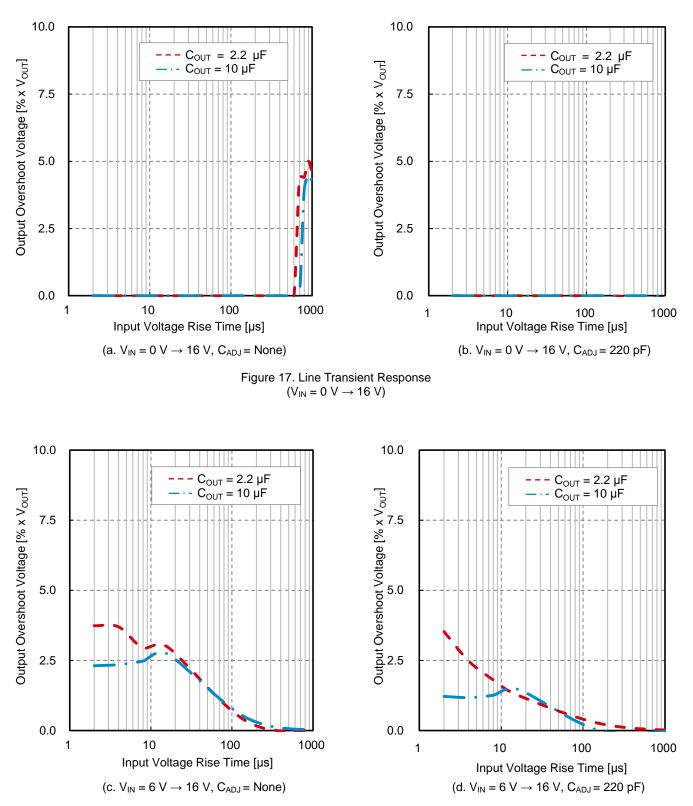


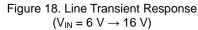
100

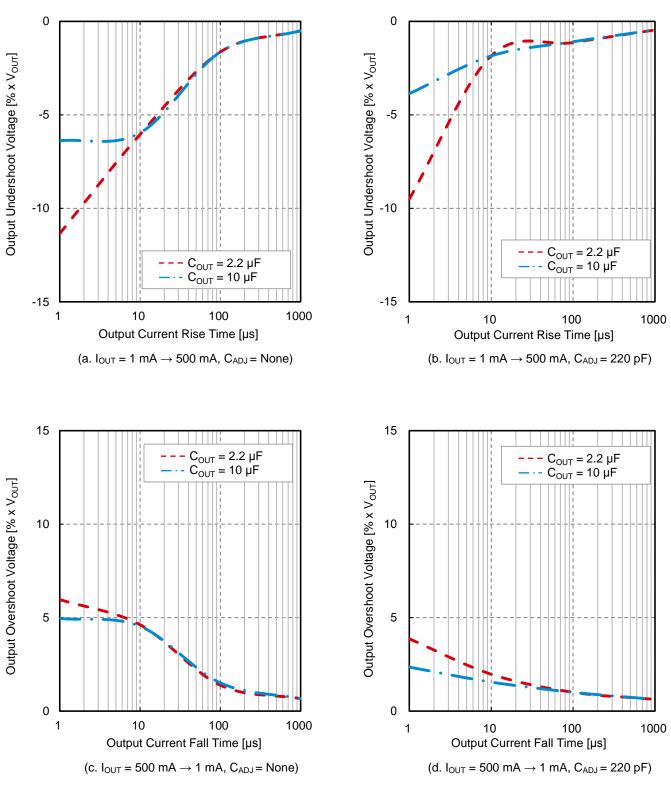
160

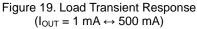
Typical Performance Curves - continued

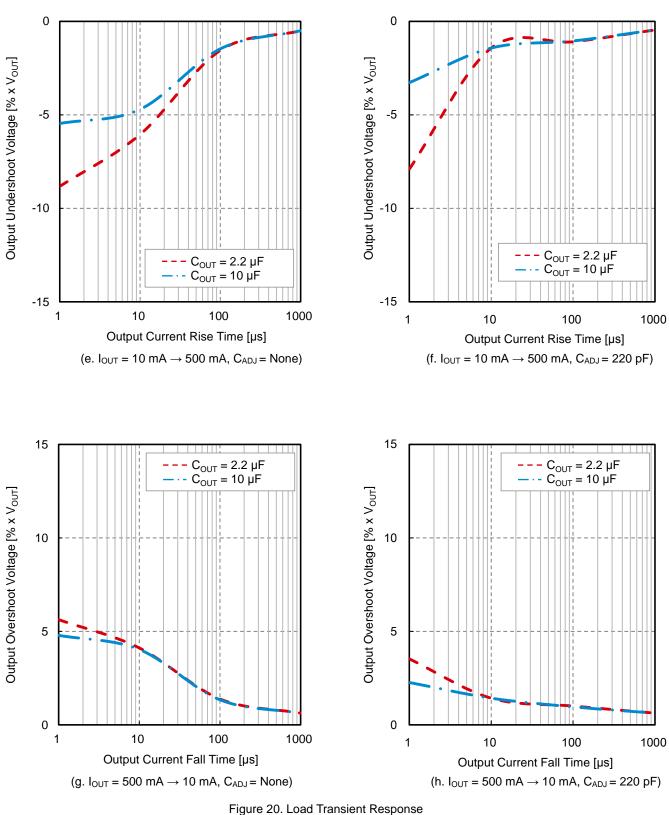












 $(I_{OUT} = 10 \text{ mA} \leftrightarrow 500 \text{ mA})$

Unless otherwise specified, $V_{IN} = 13.5 \text{ V}$, V_{OUT} setting = 5 V, $V_{EN} = 5 \text{ V}$, $R_1 = 120 \text{ k}\Omega$, $R_2 = 803 \text{ k}\Omega$, $I_{OUT} = 0 \text{ mA}$, Tj = +25 °C

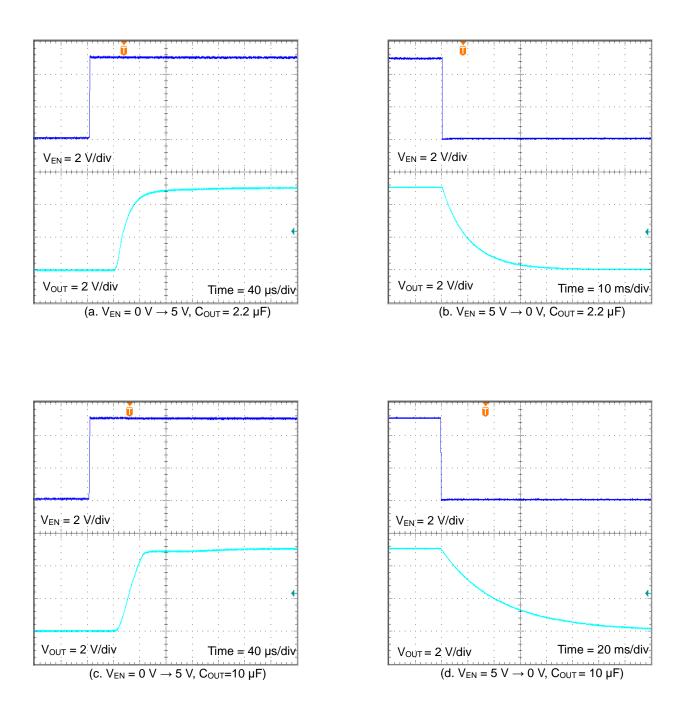


Figure 21. EN ON/OFF Sequence (V_{EN} = 0 V \leftrightarrow 5 V, Tj = +25 °C)

Unless otherwise specified, V_{IN} = 13.5 V, V_{OUT} setting = 5 V, V_{EN} = 5 V, R_1 = 120 k Ω , R_2 = 803 k Ω , I_{OUT} = 0 mA, Tj = +150 °C

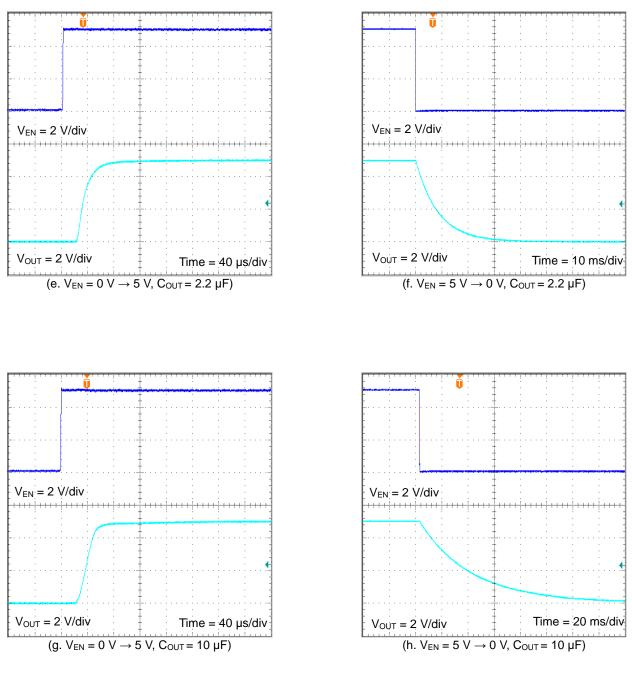


Figure 22. EN ON/OFF Sequence ($V_{EN} = 0 \ V \leftrightarrow 5 \ V$, Tj = +150 °C)

Unless otherwise specified, $V_{IN} = 13.5 \text{ V}$, V_{OUT} setting = 5 V, $V_{EN} = 5 \text{ V}$, $R_1 = 120 \text{ k}\Omega$, $R_2 = 803 \text{ k}\Omega$, $I_{OUT} = 0 \text{ mA}$, Tj = -40 °C

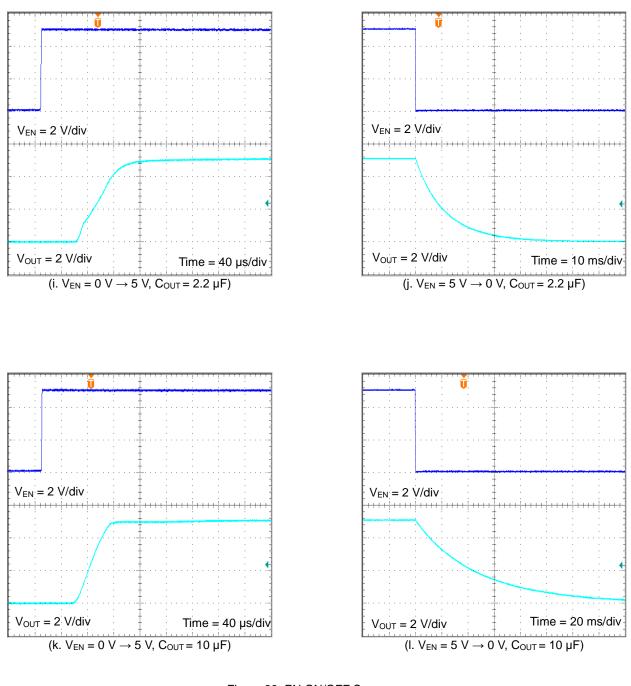


Figure 23. EN ON/OFF Sequence $(V_{EN} = 0 V \leftrightarrow 5 V, Tj = -40 °C)$

Unless otherwise specified, $V_{IN} = 13.5 \text{ V}$, V_{OUT} setting = 5 V, $V_{EN} = 5 \text{ V}$, $R_1 = 120 \text{ k}\Omega$, $R_2 = 803 \text{ k}\Omega$, $I_{OUT} = 0 \text{ mA}$, Tj = +25 °C

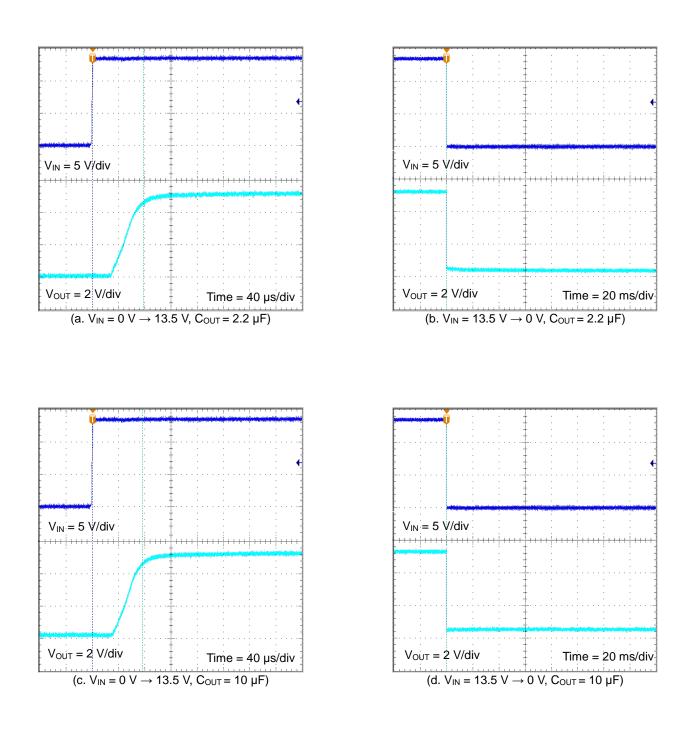


Figure 24. V_{IN} ON/OFF Sequence (V_{IN} = 0 V \leftrightarrow 13.5 V, Tj = +25 °C)

Unless otherwise specified, V_{IN} = 13.5 V, V_{OUT} setting = 5 V, V_{EN} = 5 V, R_1 = 120 k Ω , R_2 = 803 k Ω , I_{OUT} = 0 mA, Tj = +150 °C

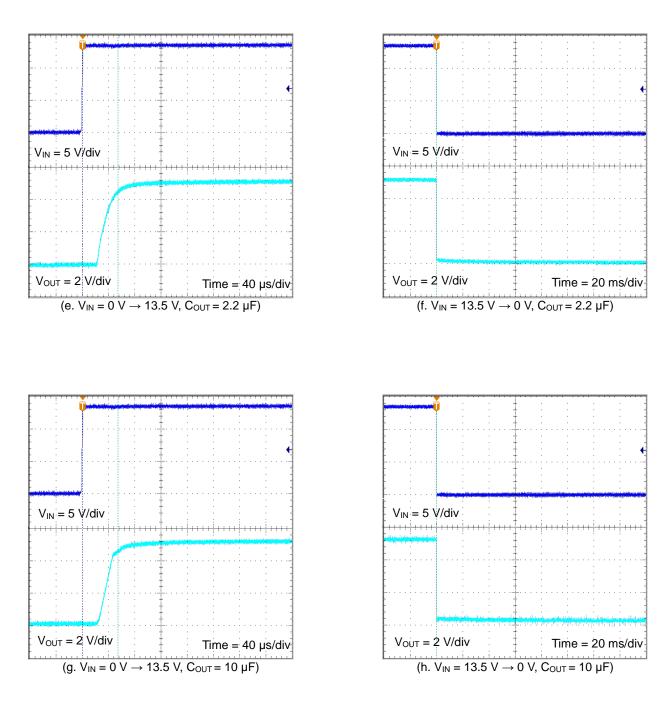


Figure 25. V_{IN} ON/OFF Sequence (V_{IN} = 0 V \leftrightarrow 13.5 V, Tj = +150 °C)

Unless otherwise specified, $V_{IN} = 13.5 \text{ V}$, V_{OUT} setting = 5 V, $V_{EN} = 5 \text{ V}$, $R_1 = 120 \text{ k}\Omega$, $R_2 = 803 \text{ k}\Omega$, $I_{OUT} = 0 \text{ mA}$, Tj = -40 °C

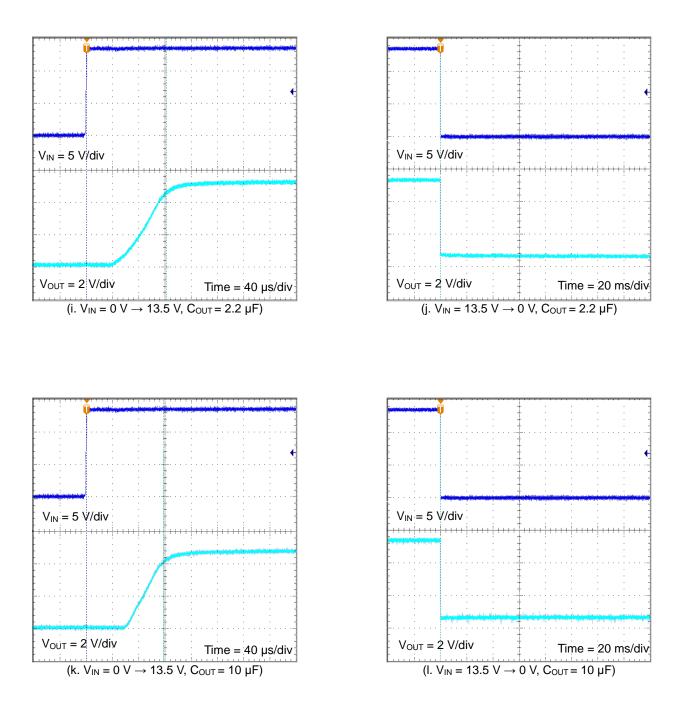
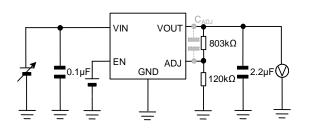
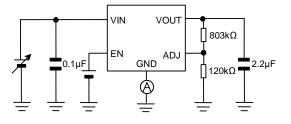


Figure 26. V_{IN} ON/OFF Sequence (V_{IN} = 0 V \leftrightarrow 13.5 V, Tj = -40 °C)

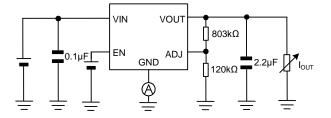
Measurement Circuit for Typical Performance Curves



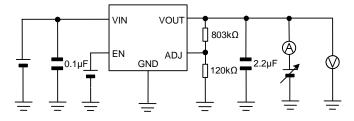
Measurement Setup for Figure 1, 2, 3, 10, 15, 17, 18



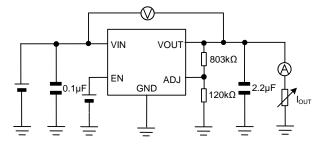
Measurement Setup for Figure 4, 5, 8, 24, 25, 26



Measurement Setup for Figure 6

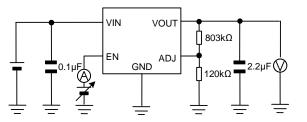


Measurement Setup for Figure 7



Measurement Setup for Figure 9

Figure 14



Measurement Setup for Figure 11, 12, 13, 21, 22, 23

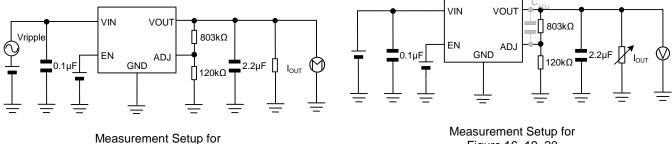


Figure 16, 19, 20

Application and Implementation

Notice: The following information is given as a reference or hint for the application and the implementation. Therefore, it does not guarantee its operation on the specific function, accuracy or external components in the application. In the application, it shall be designed with sufficient margin by enough understanding about characteristics of the external components, e.g. capacitor, and also by appropriate verification in the actual operating conditions.

Selection of External Components

Input Pin Capacitor

If the battery is placed far from the regulator or the impedance of the input-side is high, higher capacitance is required for the input capacitor in order to prevent the voltage-drop at the input line. The input capacitor and its capacitance should be selected depending on the line impedance which is between the input pin and the smoothing filter circuit of the power supply. At this time, the capacitance value setting is different each application. Generally, the capacitor with capacitance value of 0.1 μ F with good high frequency characteristic is recommended for this regulator.

Moreover, when inserting filter as ISO7637 countermeasure, insertion of ceramic capacitor from 10 nF to 470 nF (for pulses 2a) and ceramic capacitor from 100 nF to 470 nF (for pulses 3a/b) between input pin and GND pin are effective.

In addition, the consideration should be taken as the output pin capacitor, to prevent an influence to the regulator's characteristic from the deviation or the variation of the external capacitor's characteristic. All output capacitors mentioned above are recommended to have a good DC bias characteristic and a temperature characteristic (approximately ±15 %, e.g. X7R, X8R) with being satisfied high absolute maximum voltage rating based on EIA standard. These capacitors should be placed close to the input pin and mounted on the same board side of the regulator not to be influenced by implement impedance.

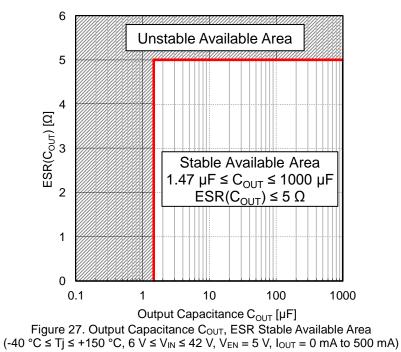
Output Pin Capacitor

The output capacitor is mandatory for the regulator in order to realize stable operation. The output capacitor with capacitance value $\ge 1.47 \ \mu\text{F}$ (Min) and ESR up to 5 Ω (Max) must be required between the output pin and the GND pin. A proper selection of appropriate both the capacitance value and ESR for the output capacitor can improve the transient behavior of the regulator and can also keep the stability with better regulation loop. The correlation of the output capacitance value and ESR is shown in the graph on the next page as the output capacitor's capacitance value and the stability region for ESR. As described in this graph, this regulator is designed to be stable with ceramic capacitors as of MLCC, with the capacitance value from 1.47 μ F to 1000 μ F and with ESR value within almost 0 Ω to 5 Ω . The frequency range of ESR can be generally considered as within about 10 kHz to 100 kHz.

Note that the provided the stable area of the capacitance value and ESR in the graph is obtained under a specific set of conditions which is based on the measurement result in single IC on our board with a resistive load. In the actual environment, the stability is affected by wire impedance on the board, input power supply impedance and also loads impedance. Therefore, please note that a careful evaluation of the actual application, the actual usage environment and the actual conditions should be done to confirm the actual stability of the system.

Generally, in the transient event which is caused by the input voltage fluctuation or the load fluctuation beyond the gain bandwidth of the regulation loop, the transient response ability of the regulator depends on the capacitance value of the output capacitor. Basically the capacitance value of $\geq 1.47 \,\mu\text{F}$ (Min) for the output capacitor is recommended as shown in the table on <u>Output Capacitance C_{OUT}, ESR Available Area</u>. Using bigger capacitance value can be expected to improve better the transient response ability in a high frequency. Various types of capacitors can be used for the output capacitor with high capacity which includes electrolytic capacitor, electro-conductive polymer capacitor and tantalum capacitor. Noted that, depending on the type of capacitors, its characteristics such as ESR ($\leq 5 \,\Omega$) absolute value range, a temperature dependency of capacitance value and increased ESR at cold temperature needs to be taken into consideration.

In addition, the same consideration should be taken as the input pin capacitor, to prevent an influence to the regulator's characteristic from the deviation or the variation of the external capacitor's characteristic. All output capacitors mentioned above are recommended to have a good DC bias characteristic and a temperature characteristic (approximately ±15 %, e.g. X7R, X8R) with being satisfied high absolute maximum voltage rating based on EIA standard. These capacitors should be placed close to the output pin and mounted on the same board side of the regulator not to be influenced by implement impedance.



Typical Application and Layout Example

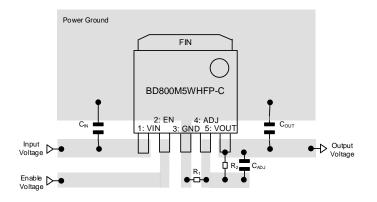


Figure 28. Typical Application and Layout Example

Parameter	Symbol	Reference Value for Application
Output Current Range	I _{OUT}	I _{OUT} ≤ 500 mA
Output Voltage Range	V _{OUT}	1.2 V to 16 V
Feedback Resistor between the ADJ and GND pins	R ₁	120 kΩ
Feedback Resistor between the ADJ and VOUT pins	R ₂	Calc. (a) $R_2 = R_1 \times (V_{OUT} / V_{ADJ} - 1) = 803 \text{ k}\Omega$ 5 V setting
ADJ Capacitor (Note 1)	C_{ADJ}	Calc. (b) $C_{ADJ} = 1 / (2\pi \times R_2 \times f_{ZERO}) = 220 \text{ pF}$
Output Capacitor	C _{OUT}	4.7 μF
Input Voltage (Note 2)	V _{IN}	13.5 V
Input Capacitor (Note 3)	CIN	0.1 µF
Enable Mode Voltage	V _{ENH}	2 V to 42V
Disable Mode Voltage	V _{ENL}	0 V to 0.8 V

(Note 1) For example, the CADJ's value is defined at 220 pF based on the calculation (b) of the above table, if it is required to improve frequency characteristics of

regulator at around f_{ZERO} = 1 kHz with the component of R₂ = 820 kΩ. (Note 2) Minimum input voltage must be 3.3 V or more. For the output voltage, please consider the voltage dropping (the minimum dropout voltage) according to the output current.

(Note 3) If the inductance of power supply line is high, please adjust input capacitor value.

Surge Voltage Protection for Linear Regulators

The following shows some helpful tips to protect ICs from possible inputting surge voltage which exceeds absolute maximum ratings.

Positive Surge to the Input

If there is any potential risk that positive surges higher than absolute maximum ratings, it is applied to the input, a Zener Diode should be inserted between the VIN pin and the GND to protect the device as shown in Figure 29.

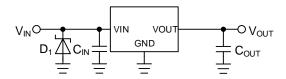


Figure 29. Surges Higher than absolute maximum ratings is Applied to the Input

Negative Surge to the Input

If there is any potential risk that negative surges below the absolute maximum ratings, (e.g.) -0.3 V, is applied to the input, a Schottky barrier diode should be inserted between the VIN and the GND to protect the device as shown in Figure 30.

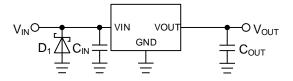


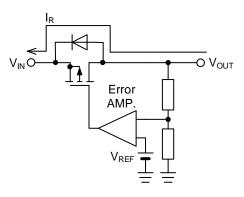
Figure 30. Surges Lower than -0.3 V is Applied to the Input

Reverse Voltage Protection for Linear Regulators

A linear regulator which is one of the integrated circuit (IC) operates normally in the condition that the input voltage is higher than the output voltage. However, it is possible to happen the abnormal situation in specific conditions which is the output voltage becomes higher than the input voltage. A reverse polarity connection between the input and the output might be occurred or a certain inductor component can also cause a polarity reverse conditions. If the countermeasure is not implemented, it may cause damage to the IC. The following shows some helpful tips to protect ICs from the reverse voltage occasion.

Protection against Reverse Input/Output Voltage

In the case that MOSFET is used for the pass transistor, a parasitic body diode between the drain-source generally exists. If the output voltage becomes higher than the input voltage and if its voltage difference exceeds V_F of the body diode, a reverse current flows from the output to the input through the body diode as shown in Figure 31. The current flows in the parasitic body diode is not limited in the protection circuit because it is the parasitic element, therefore too much reverse current may cause damage to degrade or destroy the semiconductor elements of the regulator.





Protection against Reverse Input/Output Voltage - continued

An effective solution for this problem is to implement an external bypass diode in order to prevent the reverse current flow inside the IC as shown in Figure 32. Note that the bypass diode must be turned on prior to the internal body diode of the IC. This external bypass diode should be chosen as being lower forward voltage V_F than the internal body diode. It should to be selected a diode which has a rated reverse voltage greater than the IC's input maximum voltage and also which has a rated forward current greater than the anticipated reverse current in the actual application.

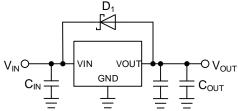


Figure 32. Bypass Diode for Reverse Current Diversion

A Schottky barrier diode which has a characteristic of low forward voltage (V_F) can meet to the requirement for the external diode to protect the IC from the reverse current. However, it also has a characteristic that the leakage (I_R) caused by the reverse voltage is bigger than other diodes. Therefore, it should be taken into the consideration to choose it because if I_R is large, it may cause increase of the current consumption, or raise of the output voltage in the light-load current condition. I_R characteristic of Schottky diode has positive temperature characteristic, which the details shall be checked with the datasheet of the products, and the careful confirmation of behavior in the actual application is mandatory.

Even in the condition when the input/output voltage is inverted, if the VIN pin is open as shown in Figure 33, or if the VIN pin becomes high-impedance condition as designed in the system, it cannot damage or degrade the parasitic element. It's because a reverse current via the pass transistor becomes extremely low. In this case, therefore, the protection external diode is not necessary.

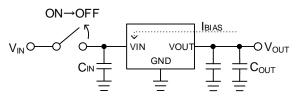


Figure 33. Open VIN

Protection against Input Reverse Voltage

When the input of the IC is connected to the power supply, accidentally if plus and minus are routed in reverse, or if there is a possibility that the input may become lower than the GND pin, it may cause to destroy the IC because a large current passes via the internal electrostatic breakdown prevention diode between the input pin and the GND pin inside the IC as shown in Figure 34.

The simplest solution to avoid this problem is to connect a Schottky barrier diode or a rectifier diode in series to the power supply line as shown in Figure 35. However, it increases a power loss calculated as $V_F \times I_{CC}$, and it also causes the voltage drop by a forward voltage V_F at the supply voltage while normal operation. Generally, since the Schottky barrier diode has lower V_F , so it contributes to rather smaller power loss than rectifier diodes. If IC has load currents, the required input current to the IC is also bigger. In this case, this external diode generates heat more, therefore select a diode with enough margin in power dissipation. On the other hand, a reverse current passes this diode in the reverse connection condition, however, it is negligible because its small amount.

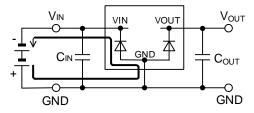


Figure 34. Current Path in Reverse Input Connection

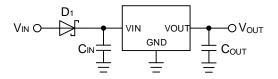


Figure 35. Protection against Reverse Polarity 1

Protection against Input Reverse Voltage - continued

Figure 36 shows a circuit in which a P-channel MOSFET is connected in series to the power. The body diode (parasitic element) is located in the drain-source junction area of the MOSFET. The drop voltage in a forward connection is calculated from the on state resistance of the MOSFET and the output current I₀. It is smaller than the drop voltage by the diode as shown in Figure 35 and results in less of a power loss. No current flows in a reverse connection where the MOSFET remains off in Figure 36.

If the gate-source voltage exceeds maximum rating of MOSFET gate-source junction with derating curve in consideration, reduce the gate-source junction voltage by connecting resistor voltage divider as shown in Figure 37.

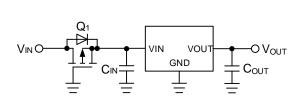


Figure 36. Protection against Reverse Polarity 2

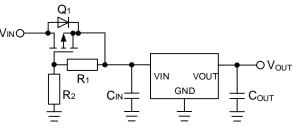


Figure 37. Protection against Reverse Polarity 3

Protection against Reverse Output Voltage when Output Connect to an Inductor

If the output load is inductive, electrical energy accumulated in the inductive load is released to the ground at the moment that the output voltage is turned off. IC integrates ESD protection diodes between the IC output and ground pins. A large current may flow in such condition finally resulting on destruction of the IC. To prevent this situation, connect a Schottky barrier diode in parallel to the integrated diodes as shown in Figure 38.

Further, if a long wire is in use for the connection between the output pin of the IC and the load, confirm that the negative voltage is not generated at the VOUT pin when the output voltage is turned off by observation of the waveform on an oscilloscope, since it is possible that the load becomes inductive. An additional diode is required for a motor load that is affected by its counter electromotive force, as it produces an electrical current in a similar way.

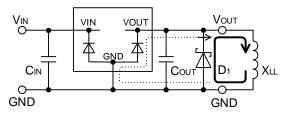


Figure 38. Current Path in Inductive Load (Output: Off)

Power Dissipation

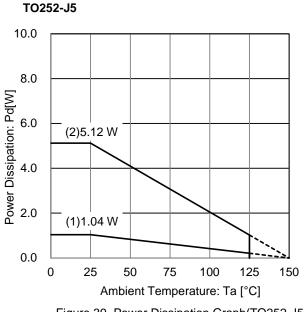


Figure 39. Power Dissipation Graph(TO252-J5)

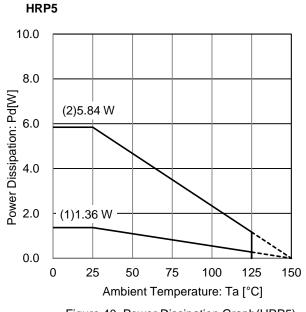


Figure 40. Power Dissipation Graph(HRP5)

(1) : 1-layer PCB (Copper foil area on the reverse side of PCB: 0 mm \times 0 mm) Board material: FR-4 Board size: 114.3 mm \times 76.2 mm \times 1.57 mmt Top copper foil: Footprint and Trace, 70 µm copper.

(2): 4-layer PCB
(Copper foil area on the reverse side of PCB: 74.2 mm × 74.2 mm) Board material: FR-4
Board size: 114.3 mm × 76.2 mm × 1.6 mmt
Top copper foil: Footprint and Traces, 70 μm copper.
2 inner layers copper foil area of PCB:
74.2 mm × 74.2 mm, 35 μm copper.
Bottom copper foil area of PCB:
74.2 mm × 74.2 mm, 70 μm copper.

 $\begin{array}{l} \mbox{Condition (1) : } \theta_{JA} = 120.1 \ ^{\circ}\mbox{C/W}, \ \Psi_{JT} \ (top \ center) = 19 \ ^{\circ}\mbox{C/W} \\ \mbox{Condition (2) : } \theta_{JA} = 24.4. \ ^{\circ}\mbox{C/W}, \ \Psi_{JT} \ (top \ center) = 3 \ ^{\circ}\mbox{C/W} \end{array}$

(1) : 1-layer PCB (Copper foil area on the reverse side of PCB: 0 mm × 0 mm) Board material: FR-4 Board size: 114.3 mm × 76.2 mm × 1.57 mmt Top copper foil: Footprint and Trace, 70 μ m copper.

(2): 4-layer PCB
(Copper foil area on the reverse side of PCB: 74.2 mm × 74.2 mm) Board material: FR-4
Board size: 114.3 mm × 76.2 mm × 1.6 mmt
Top copper foil: Footprint and Traces, 70 μm copper.
2 inner layers copper foil area of PCB:
74.2 mm × 74.2 mm, 35 μm copper.
Bottom copper foil area of PCB:
74.2 mm × 74.2 mm, 70 μm copper.

 $\begin{array}{l} \mbox{Condition (1)}: \theta_{JA} = 91.3 \ ^{\circ}\mbox{C/W}, \ \Psi_{JT} \ (top \ center) = 8 \ ^{\circ}\mbox{C/W} \\ \mbox{Condition (2)}: \theta_{JA} = 21.4 \ ^{\circ}\mbox{C/W}, \ \Psi_{JT} \ (top \ center) = 3 \ ^{\circ}\mbox{C/W} \\ \end{array}$

Power Dissipation - continued

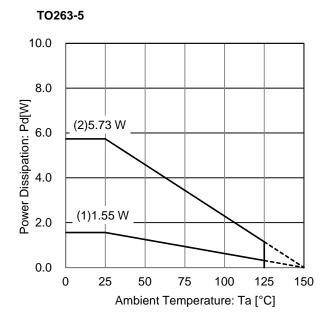


Figure 41. Power Dissipation Graph(TO263-5)

(1) : 1-layer PCB (Copper foil area on the reverse side of PCB: 0 mm \times 0 mm) Board material: FR-4 Board size: 114.3 mm \times 76.2 mm \times 1.57 mmt Top copper foil: Footprint and Trace, 70 µm copper.

(2): 4-layer PCB

(Copper foil area on the reverse side of PCB: 74.2 mm x 74.2 mm)
Board material: FR-4
Board size: 114.3 mm x 76.2 mm x 1.6 mmt
Top copper foil: Footprint and Traces, 70 μm copper.
2 inner layers copper foil area of PCB:
74.2 mm x 74.2 mm, 35 μm copper.
Bottom copper foil area of PCB:
74.2 mm x 74.2 mm, 70 μm copper.

 $\begin{array}{l} \mbox{Condition (1)}: \theta_{JA} = 80.2 \ ^{\circ}\mbox{C/W}, \ \Psi_{JT} \ (\mbox{top center}) = 10 \ ^{\circ}\mbox{C/W} \\ \mbox{Condition (2)}: \theta_{JA} = 21.8 \ ^{\circ}\mbox{C/W}, \ \Psi_{JT} \ (\mbox{top center}) = 2 \ ^{\circ}\mbox{C/W} \end{array}$

Thermal Design

This product exposes a frame on the back side of the package for thermal efficiency improvement. The power consumption of the IC is decided by the dropout voltage condition, the load current and the current consumption. Refer to power dissipation curves illustrated in Figure 39 to 41 when using the IC in an environment of Ta ≥ 25 °C. Even if the ambient temperature Ta is at 25°C, chip junction temperature (Tj) can be very high depending on the input voltage and the load current. Consider the design to be Tj \le Tjmax = 150 °C in whole operating temperature range.

Should by any condition the maximum junction temperature Tjmax = 150 °C rating be exceeded by the temperature increase of the chip, it may result in deterioration of the properties of the chip. The thermal impedance in this specification is based on recommended PCB and measurement condition by JEDEC standard. Therefore, need to be careful because it might be different from the actual use condition. Verify the application and allow sufficient margins in the thermal design by the following method to calculate the junction temperature Tj. Tj can be calculated by either of the two following methods.

1. The following method is used to calculate the junction temperature Tj with ambient temperature Ta.

 $Tj = Ta + P_C \times \theta_{IA}$ [°C]

Where:

- *Tj* is the Junction Temperature
- *Ta* is the Ambient Temperature
- $P_{\mathcal{C}}$ is the Power Consumption
- $\theta_{I\!A}$ is the Thermal Resistance (Junction to Ambient)

2. The following method is also used to calculate the junction temperature Tj with top center of case's (mold) temperature T_T.

$$Tj = T_T + P_C \times \Psi_{IT}$$
 [°C]

Where:

- Tj is the Junction Temperature
- T_T is the Top Center of Case's (mold) Temperature
- $P_{\mathcal{C}}$ is the Power consumption
- Ψ_{JT} is the Thermal Resistance (Junction to Top Center of Case)

3. The following method is used to calculate the power consumption Pc (W).

$$Pc = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{CC}$$
[W]

Where:

 $P_{\mathcal{C}}$ is the Power Consumption

 V_{IN} is the Input Voltage

- V_{OUT} is the Output Voltage
- I_{OUT} is the Load Current
- I_{CC} is the Current Consumption

Calculation Example (TO263-5)

If V_{IN} = 13.5 V, V_{OUT} = 5.0 V, I_{OUT} = 200 mA, I_{CC} = 17 μ A, the power consumption Pc can be calculated as follows:

$$P_{C} = (V_{IN} - V_{OUT}) \times I_{OUT} + V_{IN} \times I_{CC} = (13.5 V - 5.0 V) \times 200 mA + 13.5 V \times 17 \mu A \approx 1.7 W$$

At the maximum ambient temperature Tamax = 85 °C, the thermal impedance (Junction to Ambient) θ_{JA} = 21.8 °C/W (4-layer PCB)

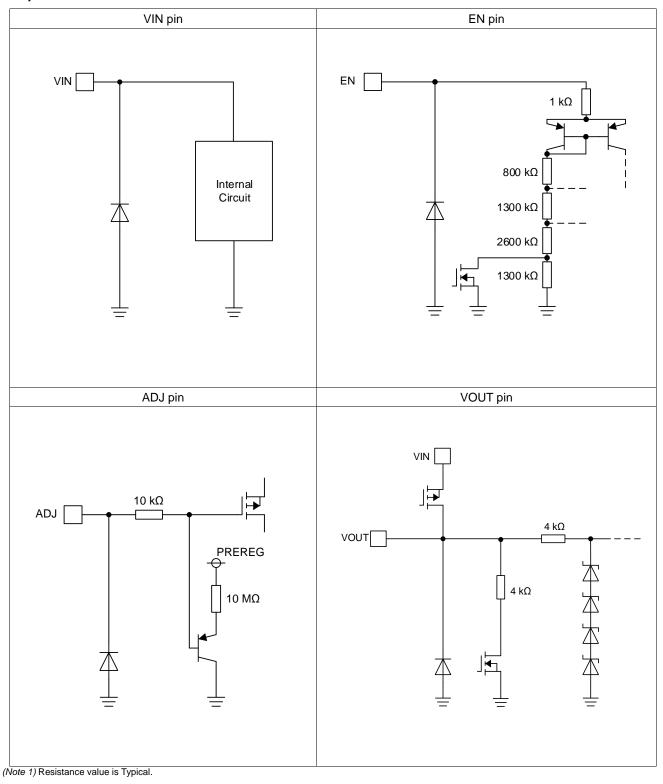
 $Tj = Tamax + P_C \times \theta_{JA}$ = 85 °C + 1.7 W × 21.8 °C/W \approx 122.1 °C

When operating the IC, the top center of case's (mold) temperature $T_T = 100$ °C, $\Psi_{JT} = 10$ °C/W (1-layer PCB)

$$Tj = T_T + P_C \times \Psi_{JT} = 100 \,^{\circ}C + 1.7 \, W \times 10 \,^{\circ}C/W = 117.0 \,^{\circ}C$$

If it is difficult to ensure the margin by the calculations above, it is recommended to expand the copper foil area of the board, increasing the layer and thermal via between thermal land pad for optimum thermal performance.

I/O Equivalence Circuits^(Note 1)



Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Recommended Operating Conditions

The function and operation of the IC are guaranteed within the range specified by the recommended operating conditions. The characteristic values are guaranteed only under the conditions of each item specified by the electrical characteristics.

6. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

7. Thermal Consideration

The power dissipation under actual operating conditions should be taken into consideration and a sufficient margin should be allowed in the thermal design. On the reverse side of the package this product has an exposed heat pad for improving the heat dissipation. The amount of heat generation depends on the voltage difference between the input and output, load current, and bias current. Therefore, when actually using the chip, ensure that the generated heat does not exceed the Pd rating. If Junction temperature is over Tjmax (=150 °C), IC characteristics may be worse due to rising chip temperature. Heat resistance in specification is measurement under PCB condition and environment recommended in JEDEC. Ensure that heat resistance in specification is different from actual environment.

8. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

9. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

10. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

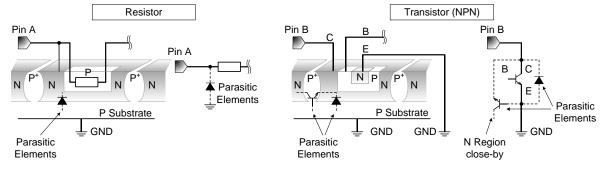
Operational Notes – continued

11. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.



12. Ceramic Capacitor

When using a ceramic capacitor, determine a capacitance value considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

13. Thermal Shutdown Protection Circuit(TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's maximum junction temperature rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF power output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

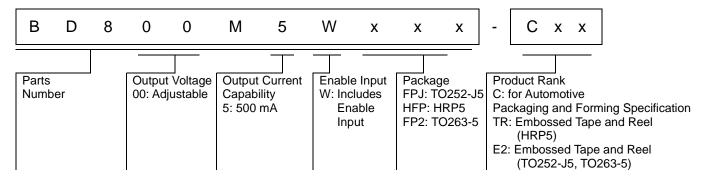
14. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

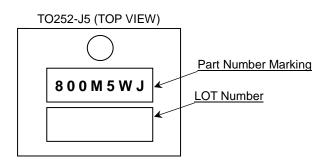
15. Enable Pin

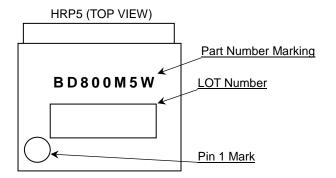
The EN pin is for controlling ON/OFF the output voltage. Do not make voltage level of chip enable keep floating level, or between V_{ENH} and V_{ENL} . Otherwise, the output voltage would be unstable or indefinite.

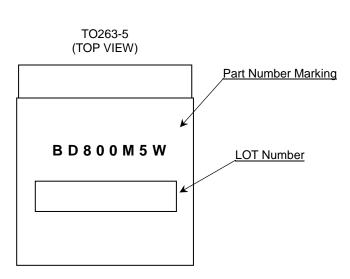
Ordering Information

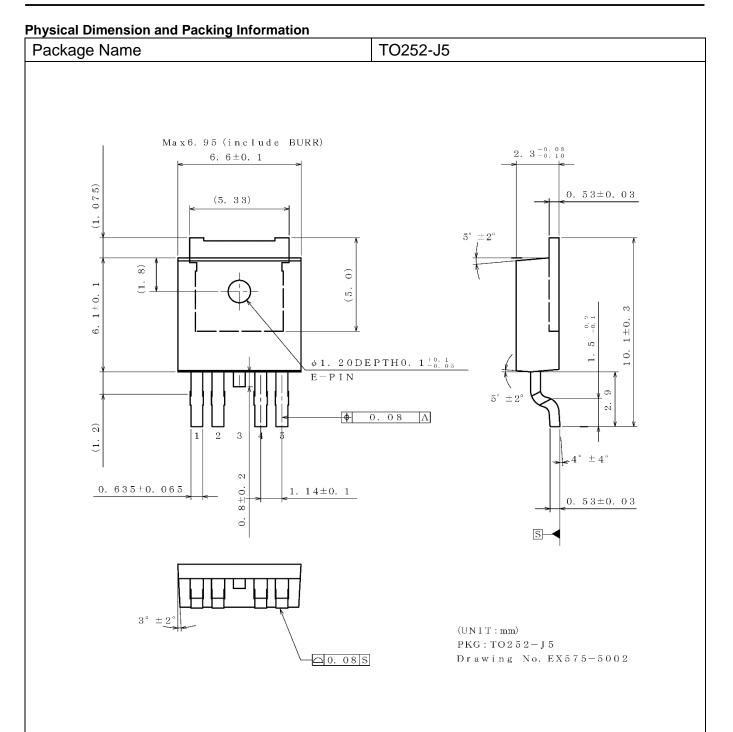


Marking Diagrams



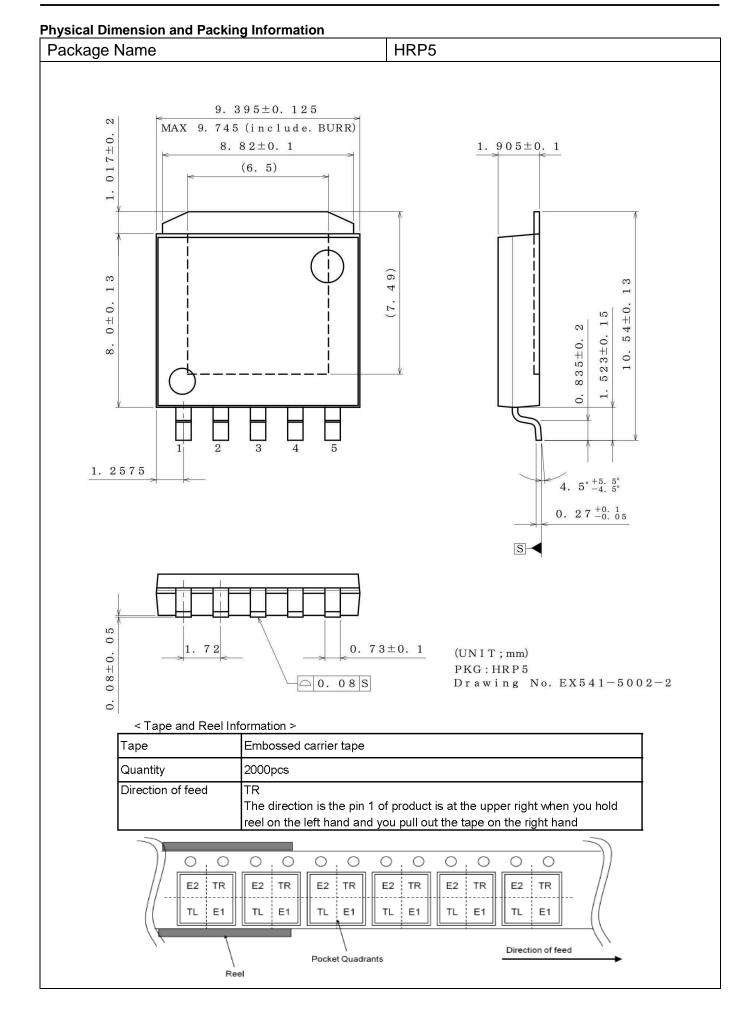


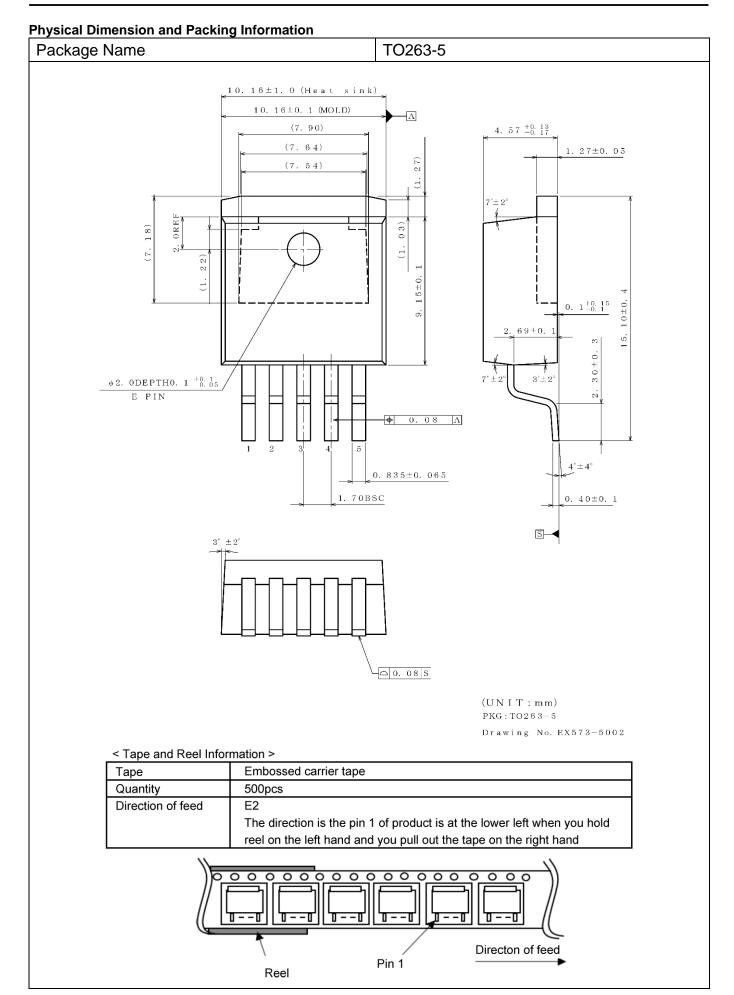




< Tape and Reel Information >

Tape	Embossed carrier tape with dry pack
Quantity	2000pcs
Direction of feed	E2
	The direction is the pin 1 of product is at the lower left
	when you hold reel on the left hand and you pull out the tape on the right hand
	Pin 1 Directon of feed





Revision History

Date	Revision	Changes
29.Mar.2019	001	New Release
02.Mar.2020	002	Add TO252-J5 Package. Add Output Voltage Accuracy ±2.5 % Add DISCHARGE block in Block Diagram and change OCP and PREREG. Change Y-axis from I_{CC} to $I_{CC}+I_{EN}$ in Figure 4 and 6. Change from monochrome to color for Figure 17 to 20. Change Measurement Circuit for Figure 6. Change Figure in Typical Application and Layout Example. Delete Item of Output Capacitor ESR for stability in Typical Application and Layout Example. Change resistance of VOUT Pin from 40 k Ω to 4 k Ω in I/O Equivalence Circuits.

Notice

Precaution on using ROHM Products

 If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment (Note 1), aircraft/spacecraft, nuclear power controllers, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

JAPAN	USA	EU	CHINA
CLASSII	CLASSⅢ	CLASS II b	CLASSI
CLASSⅣ		CLASSII	

2. ROHM designs and manufactures its Products subject to strict quality control system. However, semiconductor products can fail or malfunction at a certain rate. Please be sure to implement, at your own responsibilities, adequate safety measures including but not limited to fail-safe design against the physical injury, damage to any property, which a failure or malfunction of our Products may cause. The following are examples of safety measures:

[a] Installation of protection circuits or other protective devices to improve system safety

[b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure

- 3. Our Products are not designed under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc, prior to use, must be necessary:
 - [a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
 - [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
 - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (Exclude cases where no-clean type fluxes is used. However, recommend sufficiently about the residue.); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse, is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

Precautions Regarding Application Examples and External Circuits

- 1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
- 2. You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
 - [a] the Products are exposed to sea winds or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

Precaution for Product Label

A two-dimensional barcode printed on ROHM Products label is for ROHM's internal use only.

Precaution for Disposition

When disposing Products please dispose them properly using an authorized industry waste company.

Precaution for Foreign Exchange and Foreign Trade act

Since concerned goods might be fallen under listed items of export control prescribed by Foreign exchange and Foreign trade act, please consult with ROHM in case of export.

Precaution Regarding Intellectual Property Rights

- 1. All information and data including but not limited to application example contained in this document is for reference only. ROHM does not warrant that foregoing information or data will not infringe any intellectual property rights or any other rights of any third party regarding such information or data.
- 2. ROHM shall not have any obligations where the claims, actions or demands arising from the combination of the Products with other articles such as components, circuits, systems or external equipment (including software).
- 3. No license, expressly or implied, is granted hereby under any intellectual property rights or other rights of ROHM or any third parties with respect to the Products or the information contained in this document. Provided, however, that ROHM will not assert its intellectual property rights or other rights against you or your customers to the extent necessary to manufacture or sell products containing the Products, subject to the terms and conditions herein.

Other Precaution

- 1. This document may not be reprinted or reproduced, in whole or in part, without prior written consent of ROHM.
- 2. The Products may not be disassembled, converted, modified, reproduced or otherwise changed without prior written consent of ROHM.
- 3. In no event shall you use in any way whatsoever the Products and the related technical information contained in the Products or this document for any military purposes, including but not limited to, the development of mass-destruction weapons.
- 4. The proper names of companies or products described in this document are trademarks or registered trademarks of ROHM, its affiliated companies or third parties.

General Precaution

- 1. Before you use our Products, you are requested to carefully read this document and fully understand its contents. ROHM shall not be in any way responsible or liable for failure, malfunction or accident arising from the use of any ROHM's Products against warning, caution or note contained in this document.
- 2. All information contained in this document is current as of the issuing date and subject to change without any prior notice. Before purchasing or using ROHM's Products, please confirm the latest information with a ROHM sales representative.
- 3. The information contained in this document is provided on an "as is" basis and ROHM does not warrant that all information contained in this document is accurate and/or error-free. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties resulting from inaccuracy or errors of or concerning such information.

X-ON Electronics

Largest Supplier of Electrical and Electronic Components

Click to view similar products for LDO Voltage Regulators category:

Click to view products by ROHM manufacturer:

Other Similar products are found below :

AP7363-SP-13 NCV8664CST33T3G L79M05TL-E AP7362-HA-7 PT7M8202B12TA5EX TCR3DF185,LM(CT TLF4949EJ NCP4687DH15T1G NCV8703MX30TCG LP2951CN NCV4269CPD50R2G AP7315-25W5-7 NCV47411PAAJR2G AP2111H-1.2TRG1 ZLDO1117QK50TC AZ1117ID-ADJTRG1 NCV4263-2CPD50R2G NCP706ABMX300TAG NCP114BMX075TCG MC33269T-3.5G TLE4471GXT AP7315-33SA-7 NCV4266-2CST33T3G NCP715SQ15T2G NCV8623MN-50R2G NCV563SQ18T1G NCV8664CDT33RKG NCV4299CD250R2G NCP715MX30TBG NCV8702MX25TCG TLE7270-2E NCV562SQ25T1G AP2213D-3.3TRG1 AP2202K-2.6TRE1 NCV8170BMX300TCG NCV8152MX300180TCG NCP700CMT45TBG AP7315-33W5-7 NCP154MX180300TAG AP2113AMTR-G1 NJW4104U2-33A-TE1 MP2013AGG-5-P NCV8775CDT50RKG NJM2878F3-45-TE1 S-19214B00A-V5T2U7 S-19214B50A-V5T2U7 S-19213B50A-V5T2U7 S-19214BC0A-E8T1U7*1 S-19213B00A-V5T2U7 S-19213B33A-V5T2U7