

# 4.2V to 18V, 2A 1ch Synchronous Buck Converter with Integrated FET

## BD9328EFJ

### General Description

The BD9328EFJ is a synchronous step-down switching regulator with built-in two low-resistance N-Channel MOSFETs. This IC can supply continuous output current of 2A over a wide input range, and provides not only fast transient response, but also easy phase compensation because of current mode control.

### Features

- Uses Low ESR Output Ceramic Capacitors
- Low Standby Current
- 380 kHz Fixed Operating Frequency
- Feedback Voltage
  - $0.9V \pm 1.5\%$  ( $T_a=25^\circ C$ )
  - $0.9V \pm 2.0\%$  ( $T_a=-25^\circ C$  to  $+85^\circ C$ )
- Under Voltage Protection
- Thermal Shutdown
- Over Current Protection

### Applications

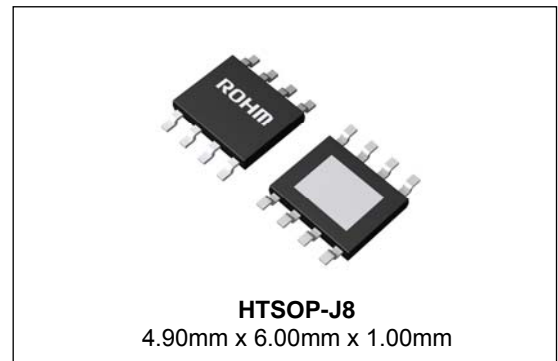
Distributed Power Systems  
Pre-Regulator for Linear Regulators

### Key Specifications

- Input Voltage Range: 4.2V to 18V
- Output Voltage Range: 0.9V to ( $V_{IN} \times 0.7$ )V
- Output Current: 2A (Max)
- Switching Frequency: 380kHz(Typ)
- Hi-Side FET ON-Resistance: 0.15Ω(Typ)
- Lo-Side FET ON-Resistance: 0.13Ω(Typ)
- Standby Current: 15μA (Typ)
- Operating Temperature Range:  $-40^\circ C$  to  $+85^\circ C$

### Package

W (Typ) D (Typ) H (Max)



### Typical Application Circuit

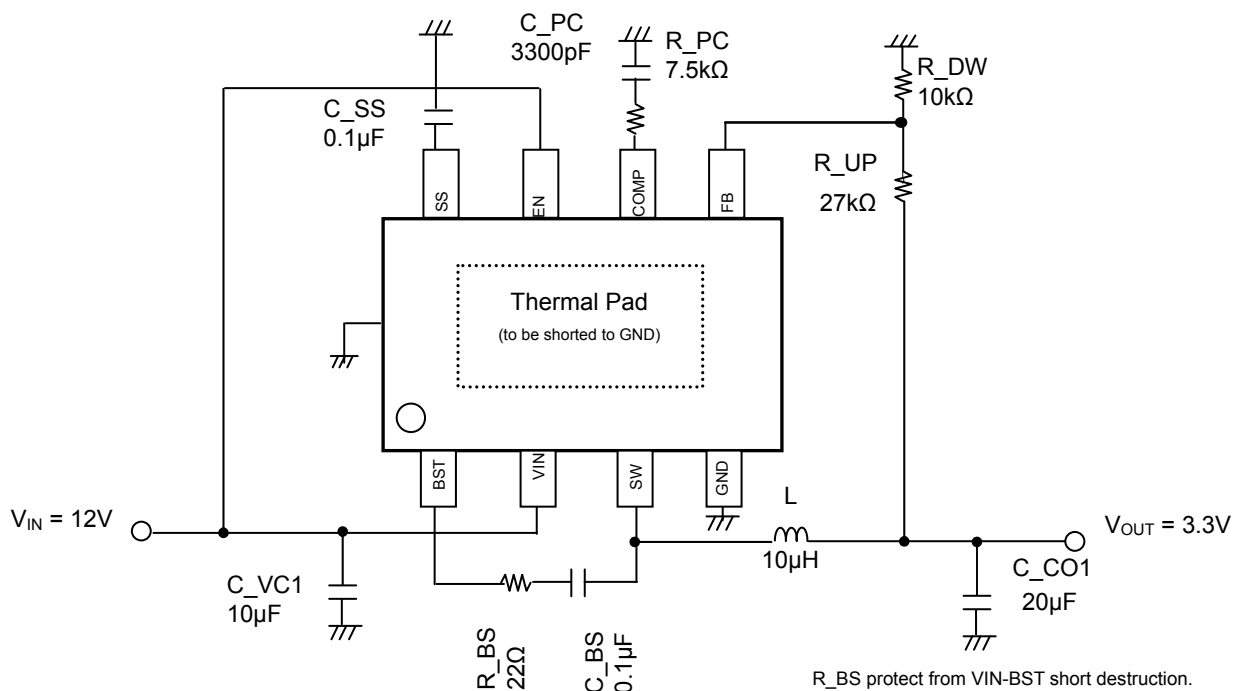


Figure 1. Typical Application Circuit

Pin Configuration

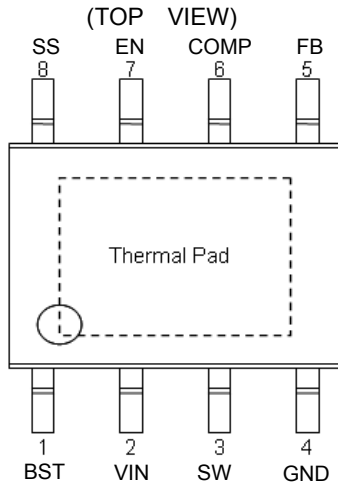


Figure 2. Pin Configuration

Block Diagram

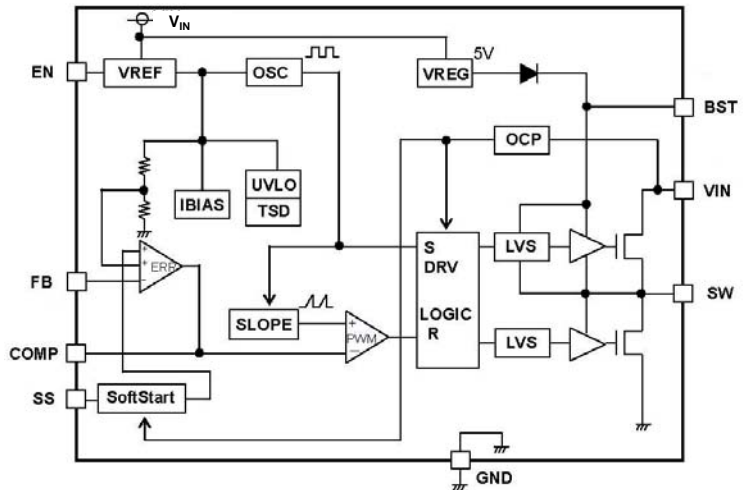


Figure 3. Block Diagram

Pin Description

Pin No.	Pin Name	Function
1	BST	High-side gate drive boost input
2	VIN	Power input
3	SW	Power switching output
4	GND	Ground
5	FB	Feedback input
6	COMP	Compensation node
7	EN	Enable input
8	SS	Soft start control input

## Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol	Rating	Unit
Supply Voltage	V <sub>IN</sub>	20	V
Switch Voltage	V <sub>SW</sub>	20	V
Power Dissipation for HTSOP-J8	Pd	3.76 (Note 1)	W
Package Thermal Resistance $\theta_{ja}$ (Note 2)	$\theta_{ja}$	29.27	°C/W
Package Thermal Resistance $\theta_{jc}$ (Note 2)	$\theta_{jc}$	3.75	°C/W
Operating Temperature Range	Topr	-40 to +85	°C
Storage Temperature Range	Tstg	-55 to +150	°C
Maximum Junction Temperature	Tjmax	150	°C
BST Voltage	V <sub>BST</sub>	V <sub>SW</sub> +7	V
EN Voltage	V <sub>EN</sub>	20	V
All Other Pins	V <sub>OTH</sub>	20	V

(Note 1) Derating is done 30.08 mW/°C when operating above Ta ≥ 25°C (Mount on 4-layer 70.0mm x 70.0mm x 1.6mm board)

(Note 2) Mount on a 4-layer 50mm x 30mm x 1.6mm application board

**Caution:** Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

## Recommended Operating Conditions (Ta= -40°C to +85°C)

Parameter	Symbol	Rating			Unit
		Min	Typ	Max	
Supply Voltage	V <sub>IN</sub>	4.2	12	18	V
SW Voltage	V <sub>SW</sub>	-0.5	-	+18	V
Output Current	I <sub>SW3</sub>	-	-	2	A
Output Voltage Range	V <sub>RANGE</sub>	0.9	-	V <sub>IN</sub> × 0.7	V

Electrical Characteristics (Unless otherwise specified V<sub>IN</sub>=12V Ta=25°C)

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
Error Amplifier Block						
FB Input Bias Current	I <sub>FB</sub>	-	0.02	2	μA	
Feedback Voltage1	V <sub>FB1</sub>	0.886	0.900	0.914	V	Voltage Follower
Feedback Voltage2	V <sub>FB2</sub>	0.882	0.900	0.918	V	Ta=-25°C to +85°C
SW Block – SW						
Hi-Side FET ON-Resistance	R <sub>ONH</sub>	-	0.15	-	Ω	I <sub>SW</sub> = -0.8A
Lo-Side FET ON-Resistance	R <sub>ONL</sub>	-	0.13	-	Ω	I <sub>SW</sub> = 0.8A
Hi/Lo-Side FET Leak Current	I <sub>LEAKN</sub>	-	0	10	μA	V <sub>IN</sub> = 18V, V <sub>SW</sub> = 0V / 18V
Switch Current Limit	I <sub>LIMIT3</sub>	3	-	-	A	
Maximum Duty Cycle	M <sub>DUTY</sub>	-	90	-	%	V <sub>FB</sub> = 0V
General						
Enable Sink Current	I <sub>EN</sub>	90	180	270	μA	V <sub>EN</sub> = 12V
Enable Threshold Voltage	V <sub>EN</sub>	1.0	1.2	1.4	V	
Under Voltage Lockout Threshold	V <sub>UVLO</sub>	3.5	3.75	4.0	V	V <sub>IN</sub> Rising
Under Voltage Lockout Hysteresis	V <sub>HYS</sub>	-	0.3	-	V	
Soft Start Current	I <sub>SS</sub>	5	10	15	μA	V <sub>SS</sub> = 0 V
Soft Start Time	t <sub>SS</sub>	-	22	-	ms	C <sub>SS</sub> = 0.1 μF
Operating Frequency	f <sub>OSC</sub>	300	380	460	kHz	
Circuit Current	I <sub>CC</sub>	-	1.2	3	mA	V <sub>FB</sub> = 1.5V, V <sub>EN</sub> = 12V
Standby Current	I <sub>QUI</sub>	-	15	27	μA	V <sub>EN</sub> = 0V

**Typical Performance Curves**

(Unless otherwise specified,  $V_{IN}=12V$   $T_a=25^{\circ}C$ )

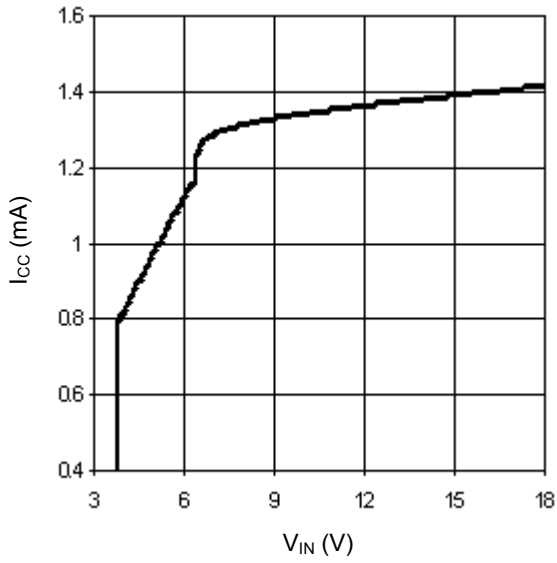


Figure 4. Circuit Current vs Input Voltage (No Switching)

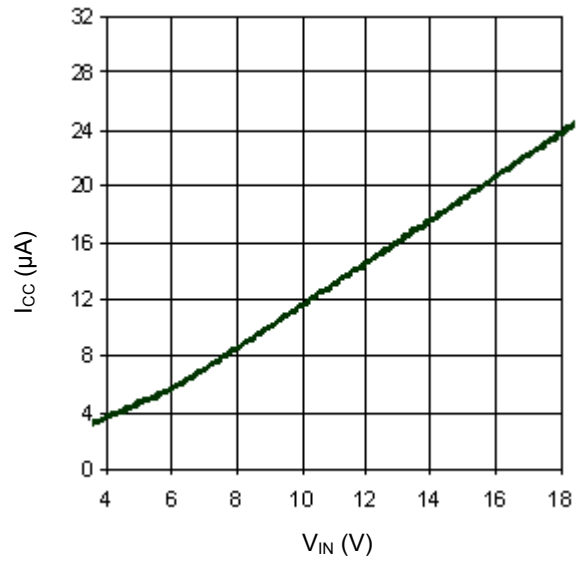


Figure 5. Standby Current vs Input Voltage (Shutdown Mode)

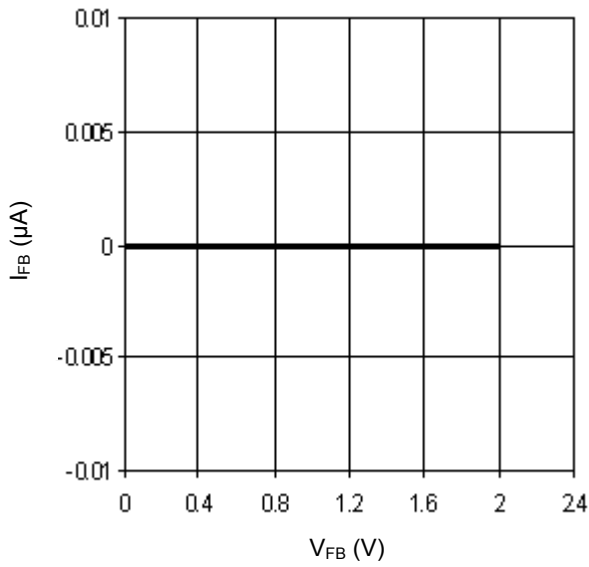


Figure 6. Input Bias Current vs Feedback Voltage

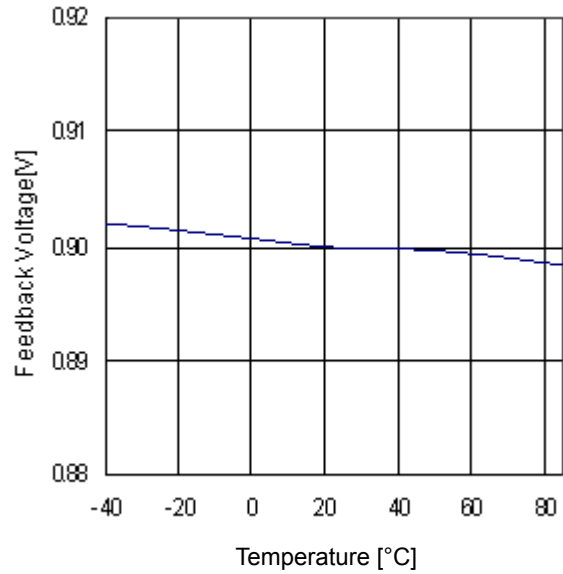


Figure 7. Feedback Voltage vs Temperature

Typical Performance Curves - continued

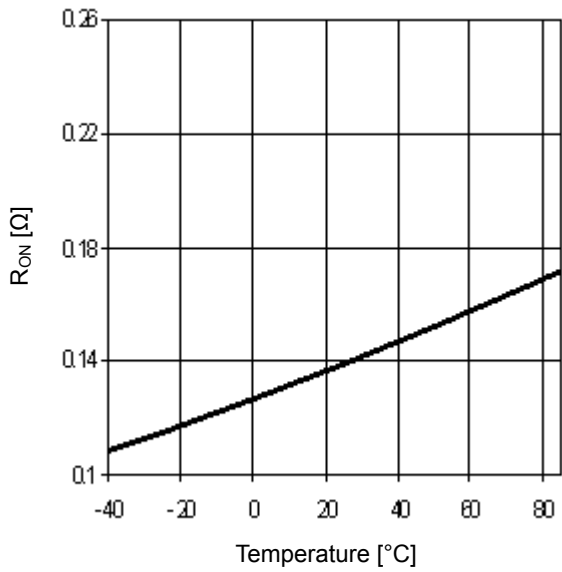


Figure 8. Hi-Side, Low-Side FET ON-Resistance vs Temperature

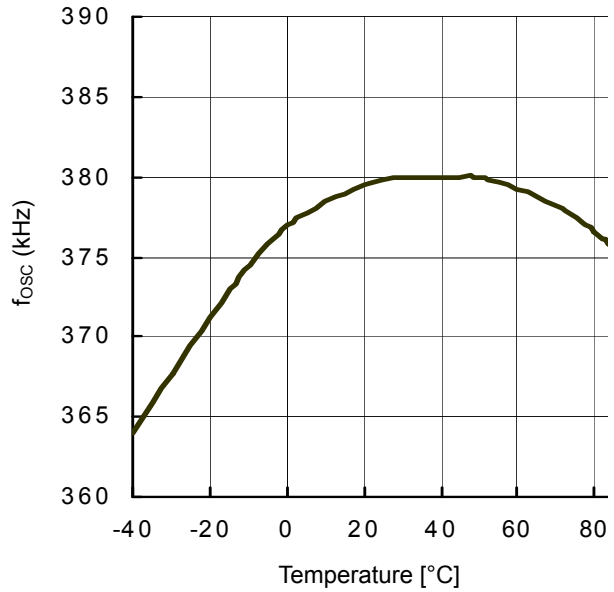


Figure 9. Operating Frequency vs Temperature

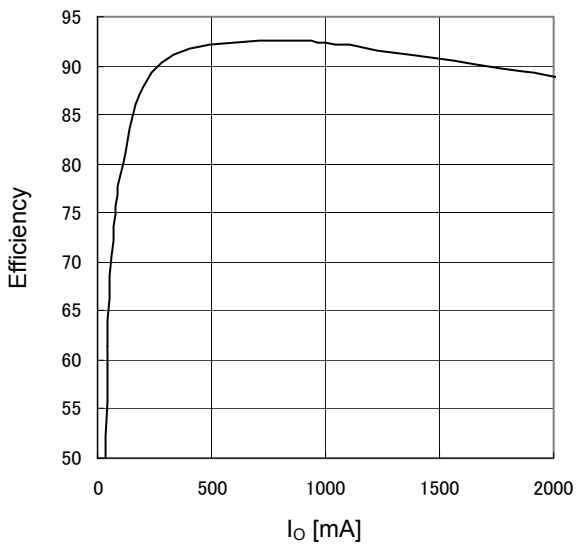


Figure 10. STEP-Down Efficiency vs I<sub>O</sub> (V<sub>IN</sub>= 12V V<sub>OUT</sub>= 3.3V L=10μH)

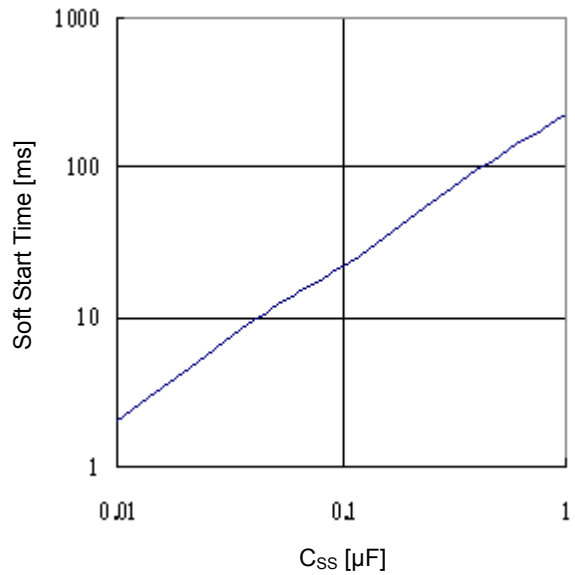


Figure 11. Soft Start Time vs Soft Start Capacitor

Typical Waveforms

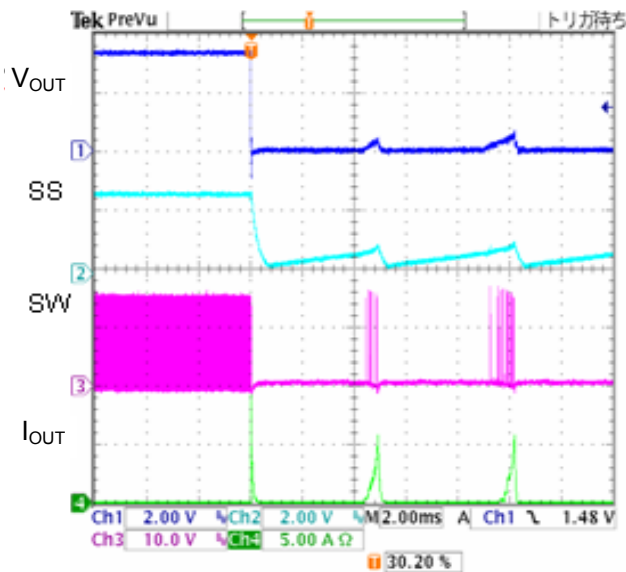


Figure 12. Over Current Protection  
(V<sub>OUT</sub> is shorted to GND)

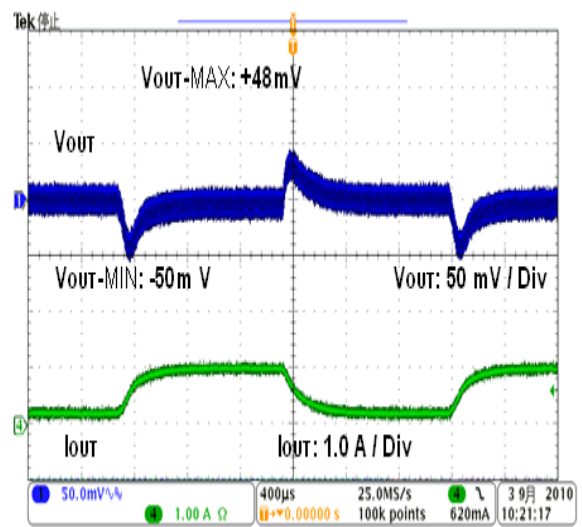


Figure 13. Transient Response  
(V<sub>IN</sub>= 12V V<sub>OUT</sub>= 3.3V L= 10μH C<sub>OUT</sub>=20μF I<sub>OUT</sub>= 0.2-1.0A )

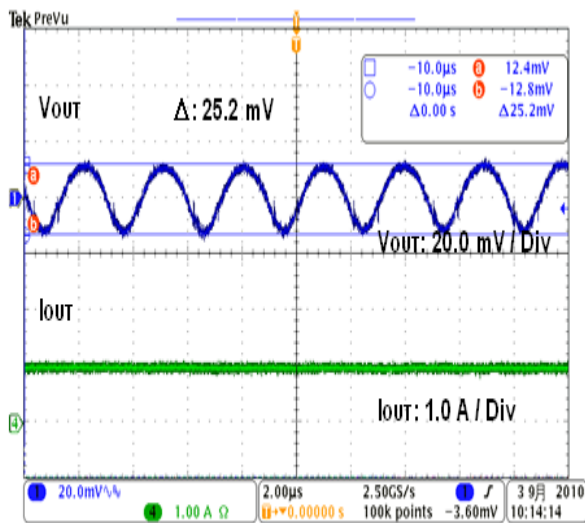


Figure 14. Output Ripple Voltage  
(V<sub>IN</sub>= 12V V<sub>OUT</sub>= 3.3V L= 10μH C<sub>OUT</sub>=20μF I<sub>OUT</sub>= 1.0A )

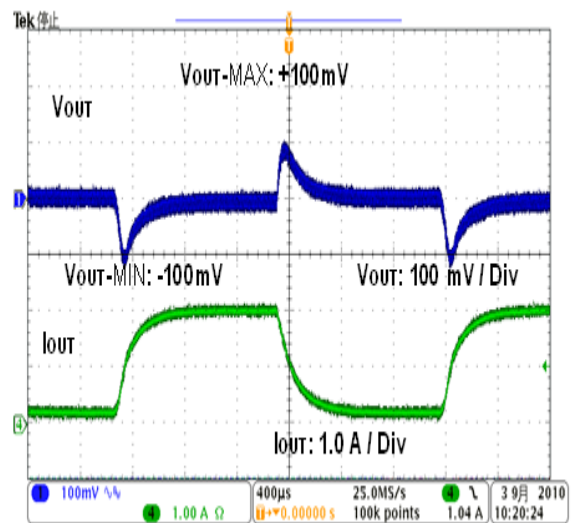


Figure 15. Transient Response  
(V<sub>IN</sub>= 12V V<sub>OUT</sub>= 3.3V L= 10μH C<sub>OUT</sub>=20μF I<sub>OUT</sub>= 0.2-2.0A)

Typical Waveforms - continued

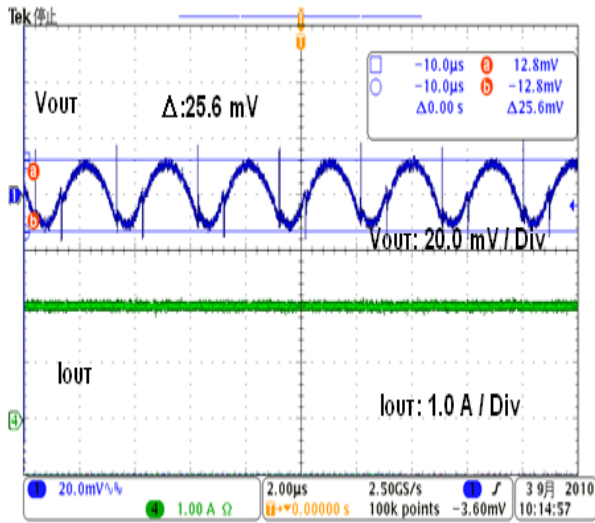


Figure 16. Output Ripple Voltage  
 ( $V_{IN} = 12\text{V}$   $V_{OUT} = 3.3\text{V}$   $L = 10\mu\text{H}$   $C_{OUT} = 20\mu\text{F}$   $I_{OUT} = 2.0\text{A}$ )

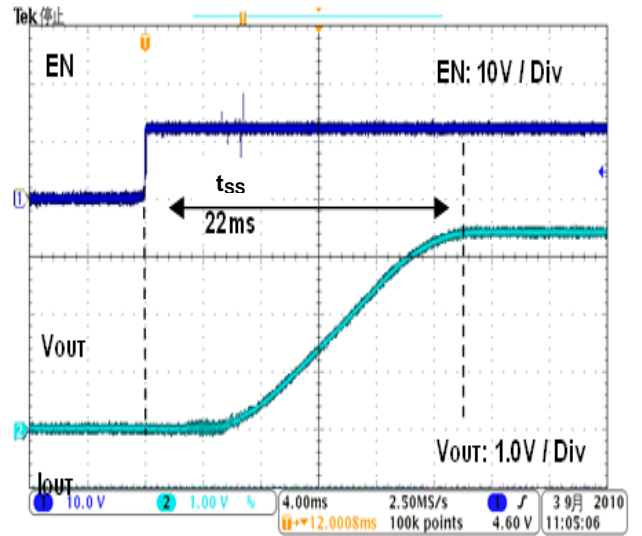
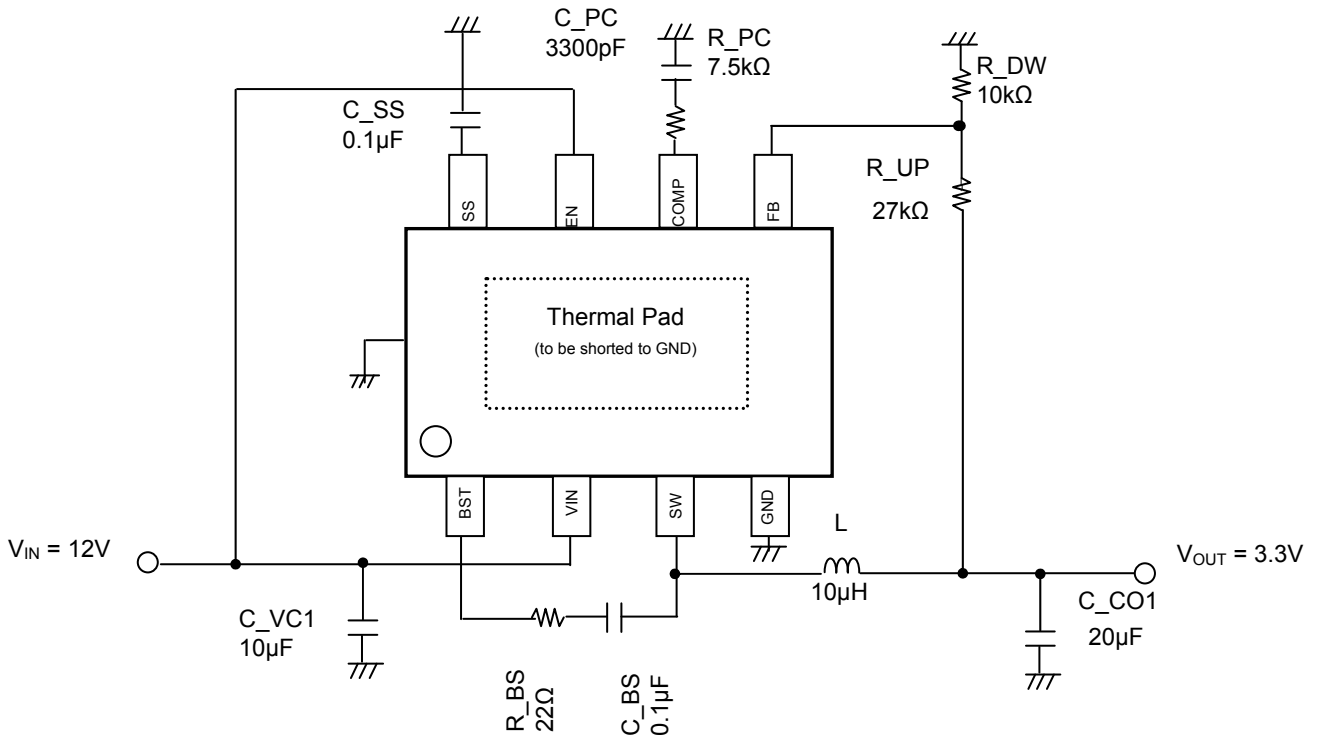


Figure 17. Start-Up Waveform  
 ( $V_{IN} = 12\text{V}$   $V_{OUT} = 3.3\text{V}$   $L = 10\mu\text{H}$   $C_{SS} = 0.1\mu\text{F}$ )

Application Information

1. Typical Application Circuit



R\_BS protect from VIN-BST short destruction.

Figure 18. Typical Application Circuit

	Symbol	Maker	Part No	
Input Capacitor	C_VC1	TDK	C3225JB1E106K	10µF/25V
Output Capacitor	C_CO1	TDK	C3216JB1C106M	10µF/16V
Inductor	L	TDK	SLF10165-100M3R8	10µH/3.8A



## 2. Block Operation

(1) VREG

This block generates a constant voltage for DC/DC boosting.

(2) VREF

This block generates an internal reference voltage of 5.1 V (Typ).

(3) TSD/UVLO

TSD (Thermal shutdown)/UVLO (Under Voltage Lockout) protection block.

The TSD circuit shuts down the IC at high temperature.

The UVLO circuit shuts down the IC when the VIN voltage is low.

(4) Error Amp Block (ERR)

This block compares the reference voltage and the feedback voltage from the output. The output voltage of this block, which is connected to COMP pin, determines the switching duty cycle. At the time of startup, since the soft start is operated by the SS pin voltage, the COMP pin voltage is limited to the SS pin voltage.

(5) Oscillator Block (OSC)

This block generates the oscillating frequency.

(6) SLOPE Block

This block generates the triangular waveform with the use of the clock created by OSC. The generated triangular waveform is sent to the PWM comparator.

(7) PWM Block

The COMP pin voltage output of the error amp is compared to the SLOPE block's triangular waveform to determine the switching duty. Since the switching duty cycle is limited by the maximum duty ratio which is determined internally, 100% duty cycle cannot be achieved.

(8) DRV Block

A DC/DC driver block that accepts signal from the PWM block to drive the power FETs.

(9) OCP Block

OCP (Over Current Protection) block. The current that flows through the FETs is detected, and OCP starts when it reached 3.0A (min). After OCP detection, switching is turned off and the SS capacitor is discharged. OCP is not a "latch type" but an "auto restart".

(10) Soft Start Circuit

This circuit prevents output voltage overshoot or inrush current by making the output voltage rise gradually while restricting the current at the time of startup.

3. Selecting Application Components

(1) Output LC Filter Constant Selection (Buck Converter)

The Output LC filter is required to supply constant current to the output load. A larger inductance value at this filter results in less inductor ripple current ( $\Delta I_L$ ) and less output ripple voltage. However, inductors with large values tend to have slower load transient-response, a larger physical size, a lower saturation current, and a higher series resistance. A smaller value of inductance has almost opposite characteristics as above. So, choosing the Inductor ripple current ( $\Delta I_L$ ) between 20% to 40% of the averaged inductor current (equivalent to the output load current) is a good compromise.

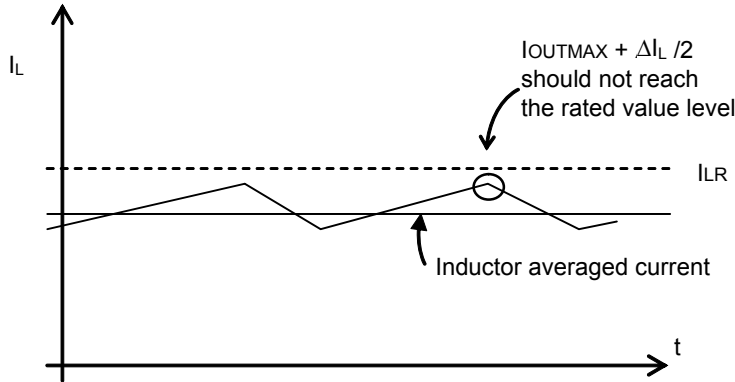


Figure 19

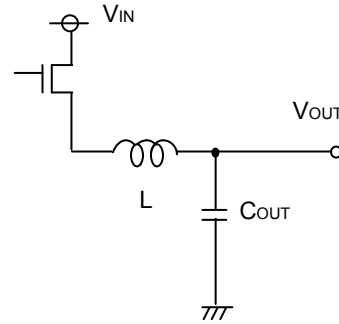


Figure 20

Setting  $\Delta I_L = 30\% \times \text{Averaged Inductor Current (2A)} = 0.6 \text{ [A]}$

$$L = V_{OUT} \times (V_{IN} - V_{OUT}) \times \frac{1}{V_{IN} \times f_{OSC} \times \Delta I_L} = 10\mu \text{ [H]}$$

Where:

$V_{IN} = 12V$ ,  $V_{OUT} = 3.3V$ ,  $f_{osc} = 380 \text{ kHz}$ ,  
 $f_{osc}$  is the switching frequency

Also the inductor should have a higher saturation current than  $I_{OUTMAX} + \Delta I_L / 2$ .

The output capacitor  $C_{OUT}$  affects the output ripple-voltage. Choose a high-value capacitor to achieve a smaller ripple-voltage that is enough to meet the application requirement.

Output ripple voltage  $\Delta V_{RPL}$  is calculated using the following equation:

$$\Delta V_{RPL} = \Delta I_L \times \left( R_{ESR} + \frac{1}{8 \times C_{OUT} \times f_{OSC}} \right) \text{ [V]}$$

where:

$R_{ESR}$  is the parasitic series resistance of the output capacitor.  
 Setting  $C_{OUT} = 20\mu F$ ,  $R_{ESR} = 10m\Omega$

$$\Delta V_{RPL} = 0.6 \times (10m + 1/(8 \times 20\mu \times 380k)) = 15.8mV$$

(2) Loop Compensation

Choosing compensation capacitor  $C_{CMP}$  and resistor  $R_{CMP}$

The current-mode buck converter has 2-poles and 1-zero system. Choosing the appropriate compensation resistor and capacitor is important to achieve a good load-transient response and good stability. An example of a DC/DC converter application bode plot is shown in Figure 22.

The compensation resistor,  $R_{CMP}$ , determines the cross over frequency  $F_{CRS}$  (the frequency at which the total DC-DC loop-gain falls to 0dB).

Setting a higher cross-over frequency achieves good response speed, but less stability. On the other hand, setting the cross-over frequency to a lower value may result to better stability, but poorer response speed.

Setting the cross-over frequency to 1/10 of the switching frequency shows good performance at most applications.

- (a) Choosing phase compensation resistor  $R_{CMP}$   
 The compensation resistor  $R_{CMP}$  can be calculated by the following formula:

$$R_{CMP} = \frac{2\pi \times V_{OUT} \times f_{CRS} \times C_{OUT}}{V_{FB} \times G_{MP} \times G_{MA}} \quad [\Omega]$$

Where:

- $V_{OUT}$  is the Output voltage
- $f_{CRS}$  is the Cross over frequency
- $C_{OUT}$  is the Output capacitor
- $V_{FB}$  is the Internal feedback voltage (**0.9V(TYP)**)
- $G_{MP}$  is the Current Sense Gain (**7.8A/V(TYP)**)
- $G_{MA}$  is the Error Amplifier Trans-conductance (**300 $\mu$ A/V(TYP)**)

Setting  $V_{OUT} = 3.3V$ ,  $f_{CRS} = 38kHz$ ,  $C_{OUT} = 20\mu F$ ;

$$R_{CMP} = \frac{2\pi \times 3.3 \times 38k \times 20\mu}{0.9 \times 7.8 \times 300\mu} = 7482.5 \approx 7.5k \quad [\Omega]$$

- (b) Choosing phase compensation capacitor  $C_{CMP}$   
 For stability of the DC/DC converter, cancellation of the phase delay that derives from output capacitor  $C_{OUT}$  and resistive load  $R_{OUT}$  is possible by inserting the phase advance.  
 The phase advance can be added by the zero on compensation resistor  $R_{CMP}$  and capacitor  $C_{CMP}$ .  
 Making  $f_z = f_{CRS} / 6$  gives a first-order estimate of  $C_{CMP}$ .

Compensation Capacitor  $C_{CMP} = \frac{1}{2\pi \times R_{CMP} \times f_z} \quad [F]$

Setting  $f_z = f_{CRS}/6 = 6.3kHz$ ;

Compensation Capacitor  $C_{CMP} = \frac{1}{2\pi \times R_{CMP} \times 6.3k} = 3.368 \times 10^{-9} \approx 3.3 \times 10^{-9} \quad [F]$

However, the best values for zero and  $f_{CRS}$  differ between applications. Decide the values accordingly after calculation using the formula above and confirmation on the actual application.

- (c) The condition of the loop compensation stability  
 The stability of DC/DC converter is important. To ensure operation stability, check if the loop compensation has enough phase-margin. For the condition of loop compensation stability, the phase-delay must be less than 150 degrees at 0 dB Gain.  
 Feed-forward capacitor,  $C_{RUP}$ , boosts phase margin over a limited frequency range and is sometimes used to improve loop response.  $C_{RUP}$  will be more effective if  $R_{UP} \gg R_{UP} || R_{DW}$

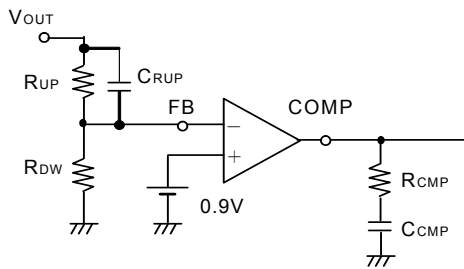


Figure 21

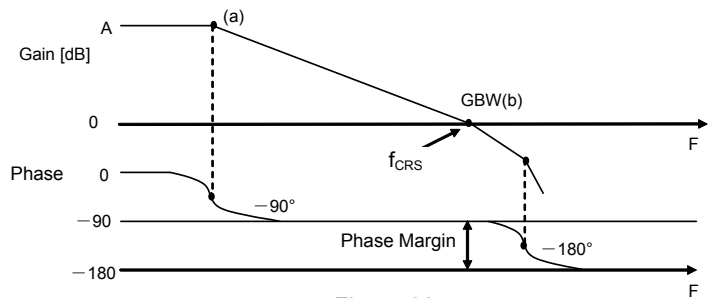


Figure 22

- (3) Design of Feedback Resistance constant  
 Set the feedback resistance as shown below.

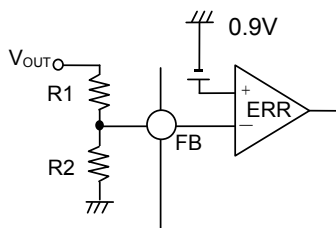


Figure 23

$$V_{OUT} = \frac{R1 + R2}{R2} \times 0.9 \quad [V]$$

4. Soft Start Function

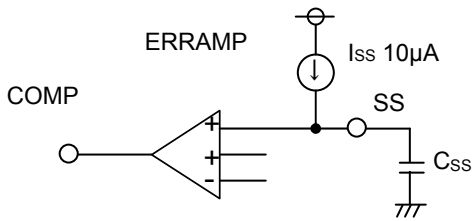


Figure 24

An adjustable soft-start function to prevent high inrush current during start-up is available.

The soft-start time is set by the external capacitor connected to SS pin.

The soft start time is given by

$$t_{SS} [s] = 2.2 \times C_{SS} / I_{SS}$$

The charge time of C<sub>SS</sub>

$$t_{SS1} [s] = 0.6 \times C_{SS} / I_{SS}$$

The SS terminal rising time

$$t_{SS2} [s] = 1.6 \times C_{SS} / I_{SS}$$

Setting C<sub>SS</sub> = 0.1µF

$$t_{SS} [s] = (0.6 + 1.6) \times 0.1\mu / 10\mu = 22 \quad [ms]$$

Please confirm the overshoot of the output voltage and inrush current in deciding the SS capacitor value.

5. EN Function

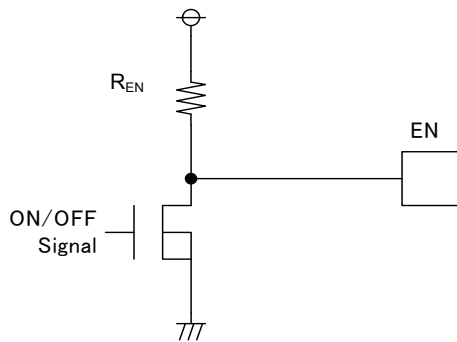
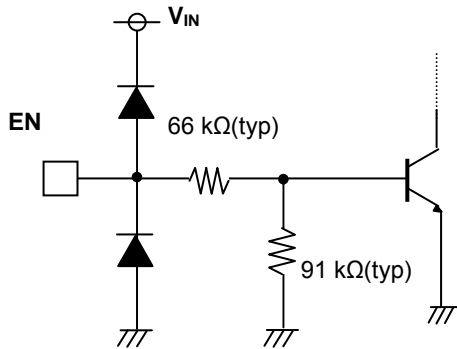


Figure 25

The EN terminal controls the IC's shut down.

Leaving EN terminal open shuts down the IC.

To start the IC, the EN terminal should be connected to V<sub>IN</sub> or to another power source.

When the EN voltage exceeds 1.2V (typ), the IC starts operating.

(Attention)

If the falling edge of EN input is too slow, output chattering occurs. This may cause large inverse current from output to input to flow and V<sub>IN</sub> voltage to increase, leading to destruction of the IC. Thus, set the fall time of EN signal within 100µs when controlling the ON/OFF operation of the IC.

This requirement is not needed when EN pin is connected with V<sub>IN</sub> and EN is not controlled.

As a recommendation, control EN with an open drain MOSFET connected as shown on Figure 25.

**6. Layout Pattern Consideration**

Two high pulsing current loops exist in the buck regulator system. The first loop, when FET is ON, starts from the input capacitors, to the VIN terminal, to the SW terminal, to the inductor, to the output capacitors, and then returns to the input capacitor through GND. The second loop, when FET is OFF, starts from the low FET, to the inductor, to the output capacitor, and then returns to the low FET through GND. To reduce the noise and improve the efficiency, please minimize these two loop areas. The input capacitor, output capacitor and the low FET should be connected to the PCB's GND plain. PCB Layout may greatly affect the thermal performance, noise and efficiency. So please take extra care when designing PCB Layout patterns.

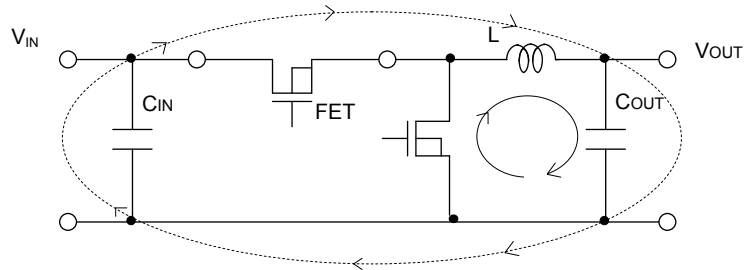


Figure 26. Current Loop in Buck Regulator System

- (1) The thermal Pad on the back side of the IC has the greatest thermal conduction into the chip. So using the GND plane as broad and wide as possible can help thermal dissipation. Adding thermal via for dissipation of heat into the different layers is also effective.
- (2) The input capacitors should be connected as close as possible to the VIN terminal.
- (3) When there is an unused area on the PCB, please arrange the copper foil plain of DC nodes, such as GND, VIN and VOUT for better heat dissipation of the IC or circumference parts.
- (4) To avoid the noise influence from AC coupling with the other lines, keep the switching lines such as SW as short as possible, and coil traces as short and as thick as possible.
- (5) Keep sensitive signal traces such as traces connected to FB and COMP away from SW pin.
- (6) The inductor and the output capacitors should be placed close to SW pin as much as possible.

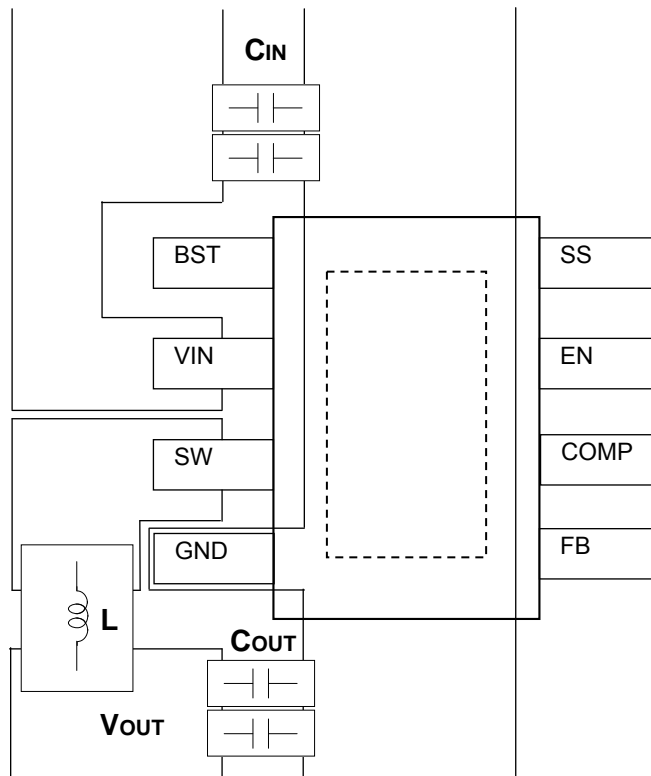


Figure 27. An example of PCB Layout Pattern

I/O Equivalence Circuit

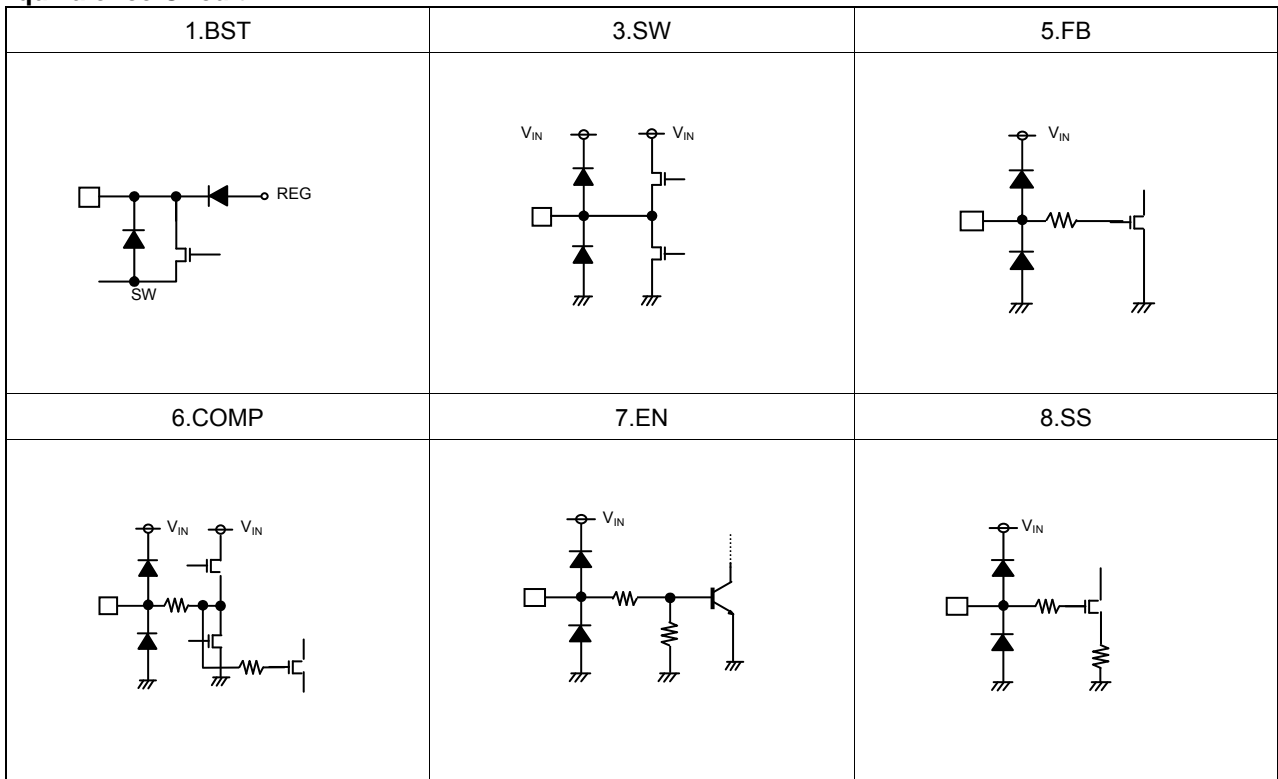
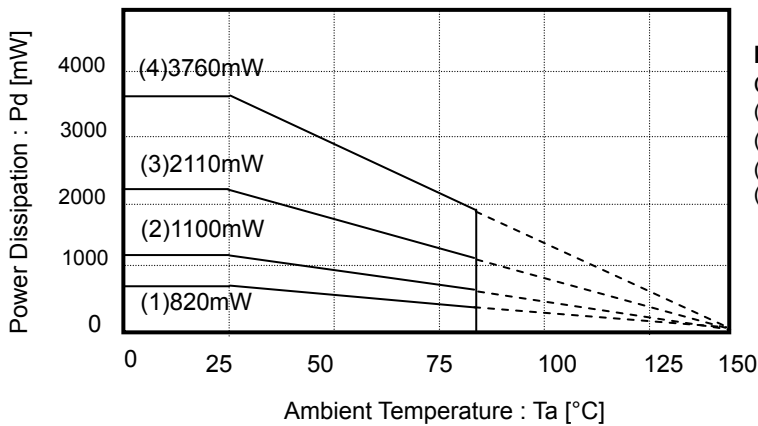


Figure 28. I/O Equivalence Circuit

Power Dissipation



HTSOP-J8 Package

- On 70mm x 70mm x 1.6 mm glass epoxy PCB
- (1) 1-layer board (Backside copper foil area 0 mm x 0 mm)
  - (2) 2-layer board (Backside copper foil area 15 mm x 15 mm)
  - (3) 2-layer board (Backside copper foil area 70 mm x 70 mm)
  - (4) 4-layer board (Backside copper foil area 70 mm x 70 mm)

## Operational Notes

### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

### 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

### 7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

### 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

### 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

### 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

Operational Notes – continued

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.  
 When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

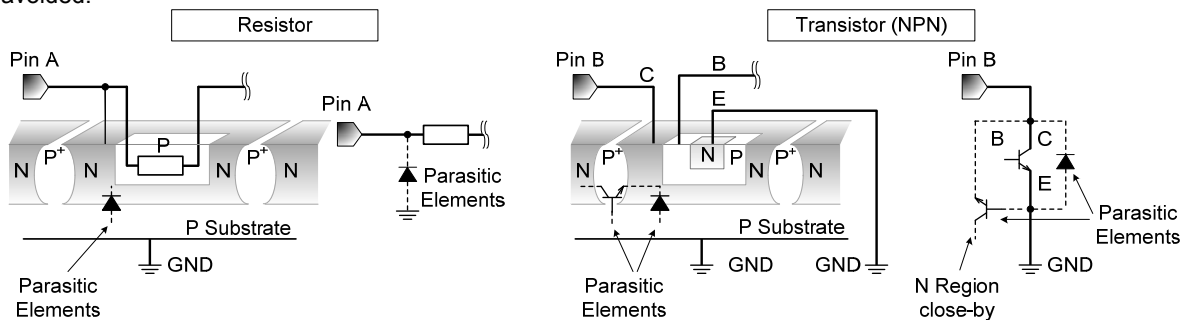


Figure 29. Example of monolithic IC structure

13. Thermal Shutdown Circuit(TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

14. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

15. EN control speed

Chattering happens if standing lowering speed is slow when standing of EN pin is lowered. The reverse current in which the input side and the pressure operation are done from the output side is generated when chattering operates with the output voltage remained, and there is a case to destruction. Please set to stand within 100us when you control ON/OFF by the EN signal.

16. About output voltage when EN terminal on

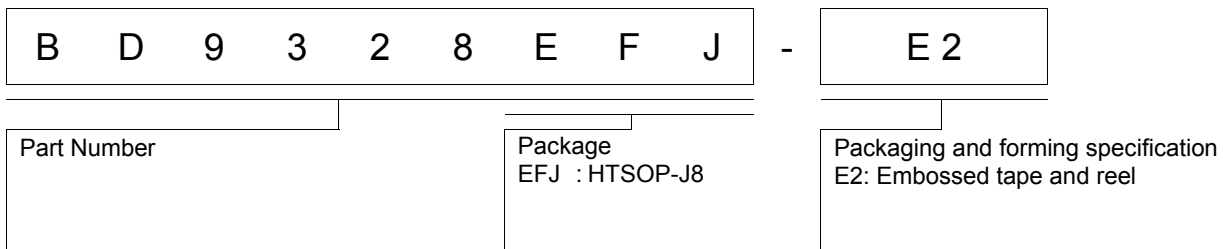
When restarting by EN terminal, BD9328EFJ starts from 0V. When an electric charge is left in an output capacitance at this time, electric current discharge from an output capacitance is performed. When many electric charges are left in an output capacitance, this electrical current discharge becomes big, and BD9328EFJ sometimes comes to destruction. Therefore please do the discharge control to follow conditions of output voltage when EN terminal on.

In case of output capacitor value is less than 100 μF,: Please set the output voltage less than 2.0V when EN terminal on.  
 In case of output capacitor value is more than 100 μF,: Please set the output voltage based on the next formula.

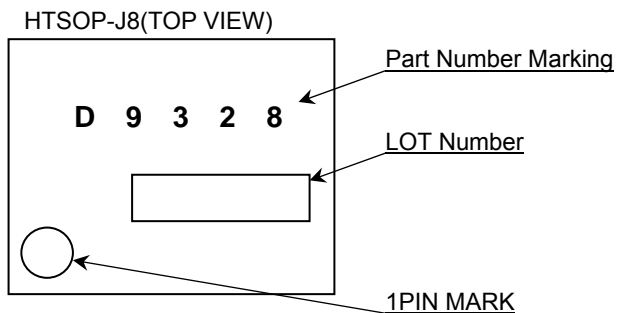
$$(\text{Output voltage when EN terminal on [V]}) < 9.15 \times (\text{Output capacitance } [\mu\text{F}])^{(-0.33)}$$



Ordering Information

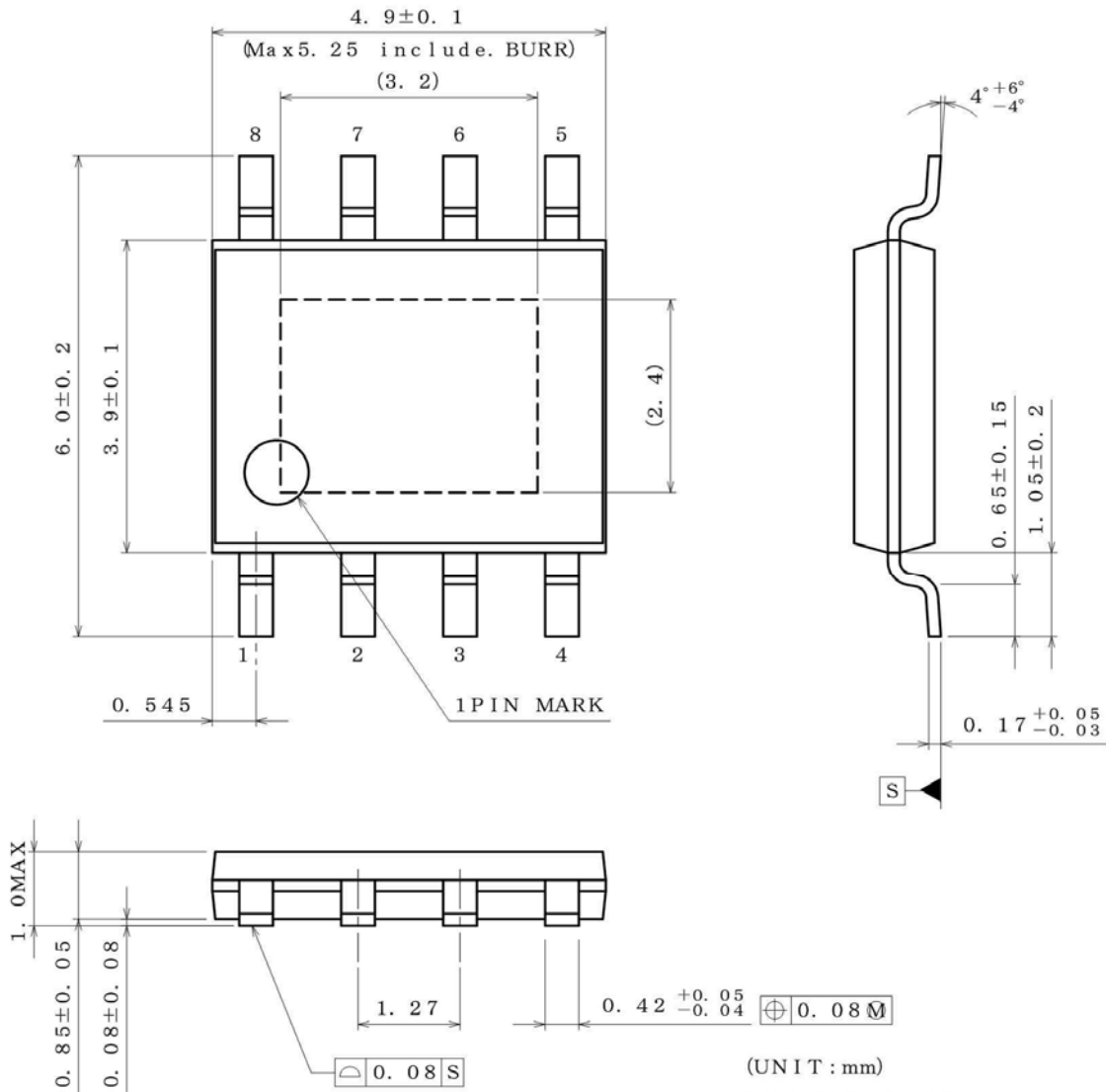


Marking Diagram

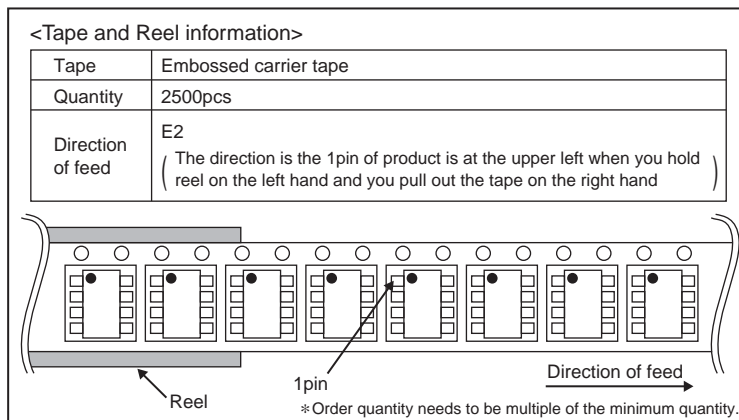


Physical Dimension Tape and Reel Information

Package Name	HTSOP-J8
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(UNIT : mm)  
 PKG : HTSOP-J8  
 Drawing No. EX169-5002-2



## Revision History

Date	Revision	Changes
11.Apr.2012	001	New Release
02.Sep.2014	002	Applied the ROHM Standard Style and improved understandability.
16.Feb.2015	003	Add "16.about output voltage when EN terminal on" in Operational Notes

# Notice

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(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA
CLASS III	CLASS III	CLASS II b	CLASS III
CLASS IV		CLASS III	

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  - Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
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  - Use of the Products in places subject to dew condensation
- The Products are not subject to radiation-proof design.
- Please verify and confirm characteristics of the final or mounted products in using the Products.
- In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- Confirm that operation temperature is within the specified range described in the product specification.
- ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

## Precaution for Mounting / Circuit board design

- When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

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### Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of Ionizer, friction prevention and temperature / humidity control).

### Precaution for Storage / Transportation

1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
  - [a] the Products are exposed to sea winds or corrosive gases, including Cl<sub>2</sub>, H<sub>2</sub>S, NH<sub>3</sub>, SO<sub>2</sub>, and NO<sub>2</sub>
  - [b] the temperature or humidity exceeds those recommended by ROHM
  - [c] the Products are exposed to direct sunshine or condensation
  - [d] the Products are exposed to high Electrostatic
2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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