# System Switching Regulator IC with Built-in FET (5V) 

## -Description

7 Channel Switching Regulator Controller for Digital Camera that contains an internal FET.
Built in the function that dim a white LED for back light with a diming set signal from a microcomputer.

## -Features

1) 1.5 V minimum input operating
2) Supplies power for the internal circuit by step-up converter(CH1).
3) CH 1 step-up converter, CH 2 cross converter, $\mathrm{CH} 3,4$ step-down converter, CH5 inverting converter for CCD, CH 6 boost converter for CCD, CH 7 boost converter for LED
4) All channels contain internal Power MOSFET and compensation. Built-In Over Voltage Protection (OVP) for $\mathrm{CH} 1,2,7$
5) Operating frequency $2.0 \mathrm{MHz}(\mathrm{CH} 3,4), 1 \mathrm{MHz}(\mathrm{CH} 1,2,5 \sim 7)$
6) Contains sequence control circuit for $\mathrm{CH} 1 \sim 4$. It is possible to select sequence $\mathrm{CH} 1 \Rightarrow \mathrm{CH} 3 \Rightarrow \mathrm{CH} 4 \Rightarrow \mathrm{CH} 2$ and $\mathrm{CH} 1 \Rightarrow \mathrm{CH} 4 \Rightarrow \mathrm{CH} 3 \Rightarrow \mathrm{CH} 2$ by SEQ_CTL pin
7) Built-In discharge switch ( $\mathrm{CH} 2,3,4$ ) and contains off sequence control circuit for $\mathrm{CH} 1 \sim 4$. $\mathrm{CH} 1,3$ turn off after $\mathrm{CH} 2,4$ output voltage discharged.
8) Built-In Short-circuit Protection (SCP)
9) CH 1 have backgate control circuit CH 6 have high side switches with soft start function.
10) Thermally enhanced UQFN036V5050 package( $5 \mathrm{~mm} \square 0.4 \mathrm{~mm}$ pitch)

## -Applications

For Digital Camera

- Absolute maximum ratings $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right)$

| Parameter | Symbol | Ratings | Unit |
| :---: | :---: | :---: | :---: |
| Maximum applied power Supply voltage | HX2BAT,VCCOUT | $-0.3 \sim 7$ | V |
| Maximum applied input voltage | VHx1~4, 56 | $-0.3 \sim 7$ | V |
|  | ( Hx56-Lx5) Voltage | $-0.3 \sim 15$ | V |
|  | VLx6 | -0.3~22 | V |
|  | VLx7 | $-0.3 \sim 30$ | V |
| Maximum Output current | IomaxHx1, Lx1 | $\pm 2.2$ | A |
|  | lomaxHx2 | $\pm 1.5$ | A |
|  | IomaxHx3 | $\pm 1.2$ | A |
|  | IomaxHx4 | $\pm 1.0$ | A |
|  | IomaxHx56 | $\pm 1.5$ | A |
|  | IomaxHS6L | +1.2 | A |
|  | IomaxLx7,8 | $\pm 1.0$ | A |
| Power Dissipation | Pd | 0.88 (*1) | W |
| Operating Temperature | Topr | $-25 \sim+85$ | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | Tstg | -55~+150 | ${ }^{\circ} \mathrm{C}$ |
| Maximum applied power Supply voltage | HX2BAT,VCCOUT | +150 | ${ }^{\circ} \mathrm{C}$ |

${ }^{*} 1$ Should be derated by $7.04 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ at $\mathrm{Ta}=25^{\circ} \mathrm{C}$ or more. When mounted on a glass epoxy PCB of $74.2 \mathrm{~mm} \times 74.2 \mathrm{~mm} \times 1.6 \mathrm{~mm}$

## －Operating condition

| Parameter | Symbol | Ratings |  |  | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min． | Typ． | Max． |  |  |
| Power supply voltage | VBAT | 1.5 | － | 5.5 | V |  |
| VREF terminal connection capacity | CVREF | 0.047 | 0.1 | 0.47 | $\mu \mathrm{F}$ |  |
| PWM7 input frequency range | fpwm | 20 | － | 100 | kHz |  |
| 【Driver】 |  |  |  |  |  |  |
| CH1 NMOS／PMOS drain current | Idpl1 | － | － | 2.1 | A |  |
| CH2 Hx2BAT input current | lin2 | － | － | 1.4 | A |  |
| CH3 PMOS drain current | Idpl3 | － | － | 1.0 | A |  |
| CH4 PMOS drain current | Idpl4 | － | － | 0.5 | A |  |
| CH5 PMOS drain current | Idpl5 | － | － | 1.4 | A |  |
| CH6 HS6L input current | IdpI6 | － | － | 1.1 | A |  |
| CH6，7 NMOS drain current | Idn16，7 | － | － | 0.9 | A |  |
| 【Output voltage setting range】 |  |  |  |  |  |  |
| CH1 |  | 4.5 | － | 5.4 | V |  |
| CH 2 |  | （※） | － | 5.4 | V | ※Use with the following range． |
| CH3 |  | 1.0 | － | 4.4 | V |  |
| CH 4 |  | 1.0 | － | 4.4 | V |  |
| CH5 |  | －9．5 | － | －1．5 | V |  |
| CH6 |  | 5.5 | － | 16 | V | ※Use with VBAT＜Vo6 |
| CH7 |  | 5.5 | － | 26 | V | ※Use with VBAT＜Vo7 |

## － CH 2 output voltage setting range



Fig． 1 CH 2 output voltage setting range
Ripple voltage level of CH 2 cross converter would be big by cross talk with embedded oscillator when oscillating Duty of step down side is $50 \%$ ．Therefore please not to set oscillating duty of CH 2 with $50 \%$ ．

## -Protective functions

| Parameter | SCP | OCP | OVP | Conditions |
| :---: | :---: | :---: | :---: | :---: |
| CH1 step-up synchronous rectification | $\bigcirc$ | O | O | Stop when shorted output OVP: VCCOUTmonitor |
| CH 2 step-up voltage. | 0 | $\bigcirc$ | $\times$ | SCP:INV monitor |
| CH3 step-down synchronous rectification | $\bigcirc$ | $\bigcirc$ | $\times$ | SCP:INV monitor |
| CH4 step-up synchronous rectification | 0 | O | $\times$ | SCP:INV monitor |
| CH5 inverse Di rectification | $\bigcirc$ | O | $\times$ | SCP: Error amp output (internal node) monitor |
| CH6 step-up Di rectification | $\bigcirc$ | $\bigcirc$ | $\times$ | SCP: Error amp output (internal node) monitor |
| CH7 step-up back light | $\times$ | O | $\bigcirc$ | OVP:VO7 monitor |

- Over current protective part

| Parameter | Symbol | Limits |  |  | Unit | Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Typ. | Max. |  |  |
| CH1 LX1 OCP detecting current | IOCP1 | 2.5 | - | - | A |  |
| CH2 HX2BAT OCP detecting current | IOCP2 | 2.0 | - | - | A |  |
| CH3 HX3 OCP detecting current | IOCP3 | 1.2 | - | - | A |  |
| CH4 HX4 OCP detecting current | IOCP4 | 1.2 | - | - | A |  |
| CH5 LX5 OCP detecting current | IOCP5 | 1.8 | - | - | A |  |
| CH6 HS6L OCP detecting current | IOCP6H | 1.5 | - | - | A |  |
| CH6 LX6 OCP detecting current | IOCP6L | 1.2 | - | - | A |  |
| CH7 Lx7 OCP detecting current | IOCP7 | 1.2 | - | - | A |  |

- Recommended maximum load current

|  |  | $\begin{aligned} & \text { Vo } \\ & \text { (V) } \end{aligned}$ | Vin (V) | $\begin{array}{\|l\|l\|} \hline \text { Io_max } \\ \hline(\mathrm{mA}) \\ \hline \end{array}$ | condition |  |  | $\begin{aligned} & \text { Vo } \\ & \text { (V) } \end{aligned}$ | Vin <br> (V) | $\underset{(\mathrm{mA})}{\text { Io_max }}$ | condition |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CH1 | Boost | 5.0 | 1.8 | 400 | $\begin{aligned} & \mathrm{L}=4.3 \mu \mathrm{H} \\ & \text { (TOKO:DE4518C) } \\ & \mathrm{C}=22 \mu \mathrm{~F} \\ & \mathrm{R} 1=390 \mathrm{k} \Omega, \\ & \mathrm{R} 2=75 \mathrm{k} \Omega \end{aligned}$ | CH5 | Reversal | -6.5 | 2.5 | 100 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=10 \mu \mathrm{~F} \\ & \mathrm{R} 1=156 \mathrm{k} \Omega \\ & \mathrm{R} 2=30 \mathrm{k} \Omega \\ & \mathrm{Cc}=1000 \mathrm{pF} \end{aligned}$ |
|  |  |  | 2.5 | 750 |  |  |  |  | 3.6 | 100 |  |
|  |  |  | 3.6 | 850 |  |  |  |  | 4.2 | 100 |  |
|  |  |  | 4.2 | 850 |  |  |  |  | 5.0 | 100 |  |
| CH2 | Boost/ Stepdown | 3.2 | 1.8 | 300 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=10 \mu \mathrm{~F} \\ & \mathrm{R} 1=440 \mathrm{k} \Omega, \\ & \mathrm{R} 2=200 \mathrm{k} \Omega \\ & \mathrm{Cc}=12 \mathrm{pF} \end{aligned}$ | CH6 | Boost | 13 | 2.5 | 30 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=10 \mu \mathrm{~F} \\ & \mathrm{R} 1=360 \mathrm{k} \Omega \\ & \mathrm{R} 2=30 \mathrm{k} \Omega \end{aligned}$ |
|  |  |  | 2.5 | 600 |  |  |  |  | 3.0 | 40 |  |
|  |  |  | 3.6 | 600 |  |  |  |  | 4.2 | 50 |  |
|  |  |  | 4.2 | 600 |  |  |  |  | 5.0 | 50 |  |
| CH3 | Step down | 1.2 | 1.8 | 800 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=10 \mu \mathrm{~F} \\ & \mathrm{R} 1=300 \mathrm{k} \Omega, \\ & \mathrm{R} 2=600 \mathrm{k} \Omega \end{aligned}$ | CH7 | Boost (worth 3 light LED) | 11.4 | 1.8 | 25 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=4.7 \mu \mathrm{~F} \end{aligned}$ |
|  |  |  | 2.5 | 1000 |  |  |  |  | 2.5 | 40 |  |
|  |  |  | 3.6 | 1000 |  |  |  |  | 3.6 | 40 |  |
|  |  |  | 4.2 | 1000 |  |  |  |  | 4.2 | 40 |  |
| CH 4 | Step down | 1.8 | 2.5 | 500 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=10 \mu \mathrm{~F} \\ & \mathrm{R} 1=300 \mathrm{k} \Omega, \\ & \mathrm{R} 2=240 \mathrm{k} \Omega \end{aligned}$ | CH7 | Boost (worth 4 light LED) | 14 | 2.5 | 30 | $\begin{aligned} & \mathrm{L}=4.7 \mu \mathrm{H} \\ & \text { (TOKO:DE2815) } \\ & \mathrm{C}=4.7 \mu \mathrm{~F} \end{aligned}$ |
|  |  |  | 3.0 | 500 |  |  |  |  | 3.6 | 40 |  |
|  |  |  | 3.6 | 500 |  |  |  |  | 4.2 | 40 |  |
|  |  |  | 4.2 | 500 |  |  |  |  | 5.0 | 40 |  |

－Electrical characteristics（Unless specified， $\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{VCCOUT}=5.0 \mathrm{~V}, \mathrm{VBAT}=3 \mathrm{~V}, \mathrm{STB} 13 \sim 7=3 \mathrm{~V}, \mathrm{UPIC} 8=2.5 \mathrm{~V}$ ）

| Parameter | Symbol | Limits |  |  | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min． | Typ． | Max． |  |  |
| 【Low－voltage input malfunction prevention circuit】 |  |  |  |  |  |  |
| Detecting voltage | Vstd1 | － | 2.3 | 2.4 | V | VCCOUT monitor |
| Release voltage | Vstd2 | 2.3 | 2.5 | 2.7 | V | VCCOUT monitor |
| Hysteresis width | $\Delta \mathrm{Vstd}$ | 100 | 200 | 300 | mV |  |
| 【Short Circuit Protection】 |  |  |  |  |  |  |
| SCP detect time | Tscp | 20 | 25 | 30 | ms |  |
| Timer start threshold voltage | Vtcinv | 0.38 | 0.48 | 0.58 | V | INV monitor $\mathrm{CH} 2 \sim 4$ |
| 【Start－up Circuit】 |  |  |  |  |  |  |
| Frequency | Fstart | 150 | 300 | 600 | kHz | HX2BAT $=1.8 \mathrm{~V}$ |
| Start－up HX2BAT Voltage | Vst1 | 1.5 | － | － | V |  |
| Start－up CH Soft Start Time | Tss1 | 1.8 | 3.0 | 5.3 | msec |  |
| 【Oscillating circuit】 |  |  |  |  |  |  |
| Frequency $\mathrm{CH} 3,4$ | fosc1 | 1.6 | 2.0 | 2.4 | MHz |  |
| Frequency CH1，2，5－7 | fosc2 | 0.8 | 1.0 | 1.2 | MHz |  |
| Max duty 1（step－up） | Dmax1 | 81 | 86 | 90 | \％ |  |
| Max duty CH2 Lx21 | Dmax21 | － | － | 100 | \％ |  |
| Max dutyCH2 Lx22 | Dmax22 | 81 | 86 | 90 | \％ |  |
| Max duty 3，4（step－down） | Dmax34 | － | － | 100 | \％ |  |
| Max duty5，6，7 | Dmax567 | 81 | 86 | 90 | \％ |  |
| 【Error Amp】 |  |  |  |  |  |  |
| Input Bias current | IINV | － | 0 | 50 | nA | INV1～7，NON5＝3．0V |
| INV threshold 1 | VINV1 | 0.79 | 0.80 | 0.81 | V | CH1，3，4 |
| INV threshold 2 | VINV2 | 0.99 | 1.00 | 1.01 | V | CH2，6 |
| INV7 threshold 1 | VINV71 | 570 | 600 | 630 | mV | PWM7，Duty＝100\％ |
| INV7 threshold 2 | VINV72 | 436 | 450 | 473 | mV | PWM7，Duty＝75\％ |
| INV7 threshold 3 | VINV73 | 223 | 240 | 257 | mV | PWM7，Duty＝40\％ |
| INV7 threshold 4 | VINV74 | 15 | 30 | 45 | mV | PWM7，Duty＝5\％ |
| 【For Inverting Base Bias Voltage Vref】 |  |  |  |  |  |  |
| CH5 Output Voltage | VOUT5 | －6．072 | －6．000 | －5．928 | V | NON5，15k $\Omega, 72 \mathrm{k} \Omega$ |
| Line Regulation | DVLi | － | 4.0 | 12.5 | mV | $\mathrm{VCCOUT}=2.8 \sim 5.5 \mathrm{~V}$ |
| Output Current When Shorted | los | 0.2 | 1.0 | － | mA | Vref＝0V |
| 【Soft Start】 |  |  |  |  |  |  |
| CH2，5，6 Soft Start Time | Tss2，5，6 | 3.1 | 5.3 | 7.4 | msec |  |
| CH3，4 Soft Start Time | Tss3，4 | 1.2 | 2.1 | 3.0 | msec |  |
| CH7 Duty Restriction time | TDTC | 5.0 | 8.2 | 11.8 | msec |  |

[^0]－Electrical characteristics $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{VCCOUT}=5.0 \mathrm{~V}, \mathrm{HX}, \mathrm{HX} 2 \mathrm{BAT}=3.6 \mathrm{~V}, \mathrm{STB} 1 \sim 6=3 \mathrm{~V}, \mathrm{PWM} 7=2.5 \mathrm{~V}\right.$ ）

| Parameter |  | Symbol | Limits |  |  | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min． | Typ． | Max． |  |  |
| 【 Output Driver】 |  |  |  |  |  |  |  |
| CH1 High side SW ON Resistance |  |  | RON1P | － | 100 | 150 | $m \Omega$ | $\mathrm{H} \times 1=5 \mathrm{~V}$ |
| CH1 Low side SW ON Resistance |  | RON1N | － | 60 | 120 | $m \Omega$ | $\mathrm{VCCOUT}=5.0 \mathrm{~V}$ |
| CH2 Lx21 High side SW ON resistance |  | RON21P | － | 120 | 180 | $m \Omega$ | $\mathrm{H} \times 2 \mathrm{BAT}=3.6 \mathrm{~V}$ |
| CH2 Lx21 Low side SW ON resistance |  | RON21N | － | 120 | 180 | $m \Omega$ | VCCOUT $=5.0 \mathrm{~V}$ |
| CH2 Lx22 High side SW ON resistance |  | RON22P | － | 120 | 180 | $m \Omega$ | VOUT2＝3．6V |
| CH2 Lx22 Low side SW ON resistance |  | RON22N | － | 100 | 150 | $m \Omega$ | VCCOUT $=5.0 \mathrm{~V}$ |
| CH3 High side SW ON Resistance |  | RON3P | － | 150 | 230 | $m \Omega$ | $\mathrm{H} \times 3=3.6 \mathrm{~V}, \mathrm{VCCOUT}=5 \mathrm{~V}$ |
| CH3 Low side SW ON Resistance |  | RON3N | － | 120 | 180 | $m \Omega$ | $\mathrm{VCCOUT}=5.0 \mathrm{~V}$ |
| CH4 High side SW ON Resistance |  | RON4P | － | 200 | 300 | $m \Omega$ | $\mathrm{H} \times 4=3.6 \mathrm{~V}, \mathrm{VCCOUT}=5 \mathrm{~V}$ |
| CH4 Low side SW ON Resistance |  | RON4N | － | 150 | 230 | $m \Omega$ | VCCOUT $=5.0 \mathrm{~V}$ |
| CH5 PMOS SW ON resistance |  | RON5P | － | 450 | 700 | $m \Omega$ | $\mathrm{H} \times 56=3.6 \mathrm{~V}$ |
| CH6，7 NMOS SW ON resistance |  | RON6，7N | － | 500 | 750 | $m \Omega$ | VCCOUT $=5.0 \mathrm{~V}$ |
| CH6 Load SW ON resistance |  | RON6P |  | 150 | 230 | $m \Omega$ | $\mathrm{H} \times 56=3.6 \mathrm{~V}$ |
| LED PIN SW ON resistance |  | RLED | － | 2.0 | 3.0 | $\Omega$ | $\mathrm{VCCOUT}=5.0 \mathrm{~V}$ |
| 【Discharge switch】 |  |  |  |  |  |  |  |
| CH2 discharge SW ON resistance |  | RDSW2 | － | 500 | 1000 | $\Omega$ | VCCOUT $=5.0 \mathrm{~V}$ |
| CH3 discharge SW | ON resistance | RDSW3 | － | 500 | 1000 | $\Omega$ | VCCOUT＝5．0V |
| CH4 discharge SW ON resistance |  | RDSW4 | － | 500 | 1000 | $\Omega$ | $\mathrm{VCCOUT}=5.0 \mathrm{~V}$ |
| 【STB1～6】 |  |  |  |  |  |  |  |
| STB Control Voltage | Active | VSTBH1 | 1.5 | － | 5.5 | V |  |
|  | Non Active | VSTBL1 | －0．3 | － | 0.3 | V |  |
| Pull Down Resistance |  | RSTB1 | 250 | 400 | 700 | $\mathrm{k} \Omega$ |  |
| 【PWM7】 |  |  |  |  |  |  |  |
| PWM7 Threshold |  | VPWM7 | 1.1 | 1.5 | 1.9 | V |  |
| Pull Down Resistance |  | RPWM7 | 250 | 400 | 700 | $\mathrm{k} \Omega$ |  |
| CH7 Delay time for shutdown |  | Toff7 | 200 | 300 | － | usec |  |
| 【Circuit Current】 |  |  |  |  |  |  |  |
| Stand－by Current | VCCOUT terminal | ISTB1 | － | － | 5 | $\mu \mathrm{A}$ |  |
|  | Hx terminal | ISTB2 | － | － | 5 | $\mu \mathrm{A}$ | Step－down Cross－converter |
|  | Lx terminal | ISTB3 | － | － | 5 | $\mu \mathrm{A}$ | Step－up |
| Circuit Current when start－up （HX2BAT current when voltage supplied for the terminal） |  | IST | － | 150 | 450 | $\mu \mathrm{A}$ | HX2BAT $=1.5 \mathrm{~V}$ |
| Circuit Current（VCCOUT current when voltage supplied for the terminal） |  | Icc2 | － | 5.0 | 9.7 | mA | $\begin{aligned} & \text { INV1~7=1.2V, } \\ & \text { NON5=-0.2V } \end{aligned}$ |

This product is not designed for normal operation with in a radioactive environment．

## - Reference data (1)



Fig. 2 Start-up circuit frequency-Temp


Fig.5-1 CH134 Base voltage-Temp


Fig. 6 CH1
5.0 V voltage boost efficiency-lo


Fig. 9 CH4 1.8 V
step-down efficiency-lo


Fig. 3 Frequency CH1,2,5~7 - Temp


Fig.5-2 CH26 Base voltage - Temp


Fig. 7 CH2 3.2
Voltage boost efficiency-lo

Fig. $10 \mathrm{CH} 5-6.5 \mathrm{~V}$ inverting efficiency-lo



Fig. 4 Frequency CH3,4—Temp


Fig.5-3 CH5 Base voltage - Temp


Fig. 8 CH3
1.2 V step-down efficiency-lo


Fig. 11 CH6 13V
boost efficiency-lo

## -Reference data (2)



Fig. 12 CH7 3LED Efficiency—Input voltage


Fig. 13 IVCCOUT-VBAT (Recommended application)


Fig. 14 CH 1 start-up waveform
(VBAT=1.5V)


Fig. $17 \mathrm{CH} 2 \sim \mathrm{CH} 4$ start-up waveform (SEQ_CTL=H)


Fig. 15 CH1 start-up waveform (VBAT=3.6V)


Fig. 18 CH5 start-up waveform


Fig. $16 \mathrm{CH} 2 \sim \mathrm{CH} 4$ start-up waveform (SEQ CTL=L)


Fig. 19 CH6 start-up waveform


Fig. 20 CH 7 start-up waveform

Fig. 21 CH1~4 waveform when OFF (SEQ_CTL=H/L common)

(SEQ CTL H/L comman)


Fig. 22 CH 7 waveform when OFF

## -Block Diagram



Fig. 23 BD9355MWV Top VIEW

## -Pin description

| Pin No | Pin name | I/O | Function | note |
| :---: | :---: | :---: | :---: | :---: |
| 24 | VCCOUT | 1 | IC Power Supply Input | Part of controller Power Supply for Low side Driver |
| 23 | GND | O | Ground terminal |  |
| 10,5,15,33 | PGND1,2,34,567 | O | Ground for Internal FET |  |
| 25 | VREF5 | 0 | CH5 Reference Output |  |
| 12 | Hx1 | O | Step up output voltage terminal (Contains backgate control) |  |
| 17,7,13,1 | HX3,2BAT,4,56 | 1 | CH2-6 Pch FET Source Terminal , FET Driver Power Supply |  |
| 11,16,14,2,34,32 | Lx1, 3, 4, 5,6,7 | I/O | Terminal for Connecting Inductor |  |
| 6 | Lx21 | I/O | Terminal for Connecting Inductor For CH2 Input |  |
| 4 | Lx22 | I/O | Terminal for Connecting Inductor For CH2 Output |  |
| 3 | VOUT2 | O | CH2 DC/DC Output |  |
| 36 | HS6L | O | Output Terminal for Internal Load Switch |  |
| 21,22,19,20,27,28 | INV1,2,3,4,6,7 | 1 | Error Amp Inverted Input |  |
| 26 | NON5 | 1 | Error Amp Non-inverted input |  |
| 18,8,35 | STB1234,5,6 | 1 | ON/OFF switch <br> H : operating over 1.5 V |  |
| 30 | PWM7 | 1 | CH7 ON/OFF Control, PWM Dimming Input |  |
| 9 | SEQ_CTL | 1 | Sequence control terminal | $\begin{aligned} & \text { GND: } \mathrm{CH} 1 \rightarrow \mathrm{CH} 3 \rightarrow \mathrm{CH} 4 \rightarrow \mathrm{CH} 2 \\ & \text { VCCOUT: } \mathrm{CH} 1 \rightarrow \mathrm{CH} 3 \rightarrow \mathrm{CH} 4 \rightarrow \mathrm{CH} 2 \end{aligned}$ |
| 29 | LED | 1 | Terminal for connecting LED Cathode |  |
| 31 | VO7 | 1 | CH7 DC/DC Output |  |

## - Application circuit(1)



Fig. 24 Applied circuit diagram 1(lithium 1 cell)

## OOperation notes

- we are confident that the above applied circuit diagram should be recommended, but please thoroughly confirm its characteristics when using it. In addition, when using it with the external circuit's constant changed, please make a decision that allows a sufficient margin in light of the fluctuations of external components and ROHM's IC in terms of not only static characteristic but also transient characteristic.


## -Application circuit (2)



Fig. 25 Applied circuit diagram 2(dry battery $\times 2$ )

## OOperation notes

- we are confident that the above applied circuit diagram should be recommended, but please thoroughly confirm its characteristics when using it. In addition, when using it with the external circuit's constant changed, please make a decision that allows a sufficient margin in light of the fluctuations of external components and ROHM's IC in terms of not only static characteristic but also transient characteristic.
- Timing chart (1)

SEQ_CTL=L (GND)


SEQ_CTL=H (VCCOUT)


Fig. $26 \mathrm{CH} 1 \sim 4$ start-up sequence

- Timing chart (2)


Fig. 27 CH5,6 start-up sequence


Fig. $28 \quad$ CH7 start-up sequence

- Timing chart (3)


Fig. $29 \quad \mathrm{CH} 1 \sim 4$ OFF sequence


Fig. $30 \mathrm{CH} 5,6$ OFF sequence


Fig. 31 CH7 OFF sequence

## - CH 7 dimming function



Fig.. 32 CH 7 block diagram

## - CH7 operation

The output duty control signal for soft start starts rising by connecting terminals LED and INV7 when inputting any Duty for PWM7. And threshold voltage of erroramp being proportional to PWM7 Duty is supplied by PWM7 input signal after start up , then INV7 output voltage being proportional to PWM7 Duty is supplied as the result of negative feedback of DCDC converter. DTC7 rises up slower comparatively with oscillating frequency by fixed degree incline. Oscillating duty is restricted by DTC7 signal which is inputted to PWM comparator therefore input rush current is prevented to occur even output voltage of erroramp at start up time rises up rapidly. The time from start up to reaching set current of LED is depend on input voltage, a number of LED, PWM7 duty, resistor to set the current of LED. The time to reach set current of LED will be long when input voltage is low, a number of LED is big, set output current is big because of high duty under that condition. When you input L voltage into PWM7 pin during over $500 \mu \mathrm{sec}$ typ, Switch between LED and INV7 and switching turn off. CH7 heve Over voltage protection(OVP).When VO7 pin is over 28Vtyp,OVP stop CH7 function..OVP latch CH 7 function and reset dy All $\mathrm{STB}=\mathrm{L}$.

- Attention of CH7 start-up

In case CH 7 start up, Please turn on STB1234 before $\mathrm{CH} 7 . \mathrm{CH} 7$ can not start before $\mathrm{CH} 1,2,3,4$

## - Recommended method of setting at the time of INV7 output voltage setting.

If INV7 output setting value is made larger than previous setting value during all intervals but soft start interval (at the time of starting up), it is recommended that the value of voltage is increased step by step with the smallest possible width of step after fully evaluating the restriction at the soft side that controls rush current and switching and the vision of brightness etc. in terms of set application.

## OPWM7 Duty INV7 Voltage Value

INV7 is output voltage that proportionate to PWM7 input PWM pulse DUTY and control LED current by external resister for setting (between INV7 and GND ). LED current is decided by NOTE1 formula.
(Note 1) LED current = INV7 voltage / resistance R for LED current setting

| PWM7 InputDUTY[\%] | INV7 Output Voltage [mV] |
| :---: | :---: |
| 5 | 30 |
| 10 | 60 |
| 15 | 90 |
| 20 | 120 |
| 25 | 150 |
| 30 | 180 |
| 35 | 210 |
| 40 | 240 |
| 45 | 270 |
| 50 | 300 |
| 55 | 330 |
| 60 | 360 |
| 65 | 390 |
| 70 | 420 |
| 75 | 450 |
| 80 | 480 |
| 85 | 510 |
| 90 | 540 |
| 95 | 570 |
| 100 | 600 |

Fig. 33 PWM7 DUTY - INV7 Terminal Voltage

## -Block explanation

1. SCP, Timer Latch

It is a timer latch type of short-circuit protection circuit.
For $\mathrm{CH} 1,2,6 \sim 8$, the error AMP output voltage is monitored, and detected when the feedback voltage deviates from control, for $\mathrm{CH} 3 \sim 5$, it is detected when the voltage of INV terminal becomes lower than $60 \%$, and in 25 ms the latch circuit operates and the outputs of all the channels are fixed at OFF.
In order to reset the latch circuit, please turn off all the STB terminals before turning them on once again or turning power supply on once again.
2. U.V.L.O

It is a circuit to prevent malfunction at low voltage.
It is to prevent malfunction of internal circuit at the time of rising or dropping to a lower value of power supply voltage. If the voltage of VCCOUT terminal becomes lower than 2.3 V , then the output of each DC/DC converter is reset to OFF, and SCP's timer latch \& soft start circuit are reset. When control is deviated from, the operation of CH1 at the time of start-up will be explained in START UP OSC mentioned later.
3. Voltage Reference (VREF5)

For the reference voltage circuit of CH 5 inversion CH , the output voltage is 1.25 V and outputted from VREF5 terminal (25pin). According this voltage and the output voltage of CH 5 , the dividing resistance (resistor) is set and then the output voltage is set.
If STB5 terminal is made to be H level at the time of start-up, then increase gradually the voltage up to 1.25 V . The inversion output of CH 5 follows this voltage and performs the soft start. $0.1 \mu \mathrm{~F}$ is recommended as the external capacitor.
4. OSC

It is an oscillation circuit the frequency of which is fixed by a built-in CR.
The operating frequencies of $\mathrm{CH} 3,4$ are set at 2 MHz , and the operating frequencies of $\mathrm{CH} 1,2,5$ are set at 1 MHz .
5. ERRAMP 1~7

It is an error amplifier to detect output signal and output PWM control signal. The reference voltages of ERRAMP (Error Amplifier) of $\mathrm{CH} 1,3,4$ are internally set at 0.8 V , and the reference voltages of ERRAMP (Error Amplifier) of $\mathrm{CH} 2,6$ are set at 1.0 V . The reference voltage of CH 5 is set at GND potential, and for CH7's ERRAMP7, the maximum value of the reference voltage is set at 0.6 V . In addition, each CH incorporates a built-in element for phase compensation.
6. ERRCOMP, Start Up OSC

It is a comparator to detect the output voltage and control the start circuit, and also an oscillator that is turned ON/OFF by this comparator and starts operating from 1.5 V . The frequency of this oscillator is about 300 kHz fixed internally. This oscillator stops operating if VCC terminal becomes more than 2.5 V or the soft start time is exceeded.
7. Current mode control block

CH1, 3~7 adopt the PWM method based on current mode.
For a current- mode DC/DC converter, FET at the main side of synchronous rectification is turned on when detecting the clock edge, and turned off by detecting the peak current by means of the current comparator.
8. Cross Control

DUTY controller for CH2 cross converter. It have PWM comparator that compare 1MHz SLOPE and ERROR AMP output and logic circuit for control 4 FET ON/OFF switching.. LX21 MAX ON DUTY is $100 \%$, LX21 MAX ON DUTY is $86 \%$.
9. Back gate Control

PchFET backgate selector controller in CH 1 .
PchFET have body Di between backgate and source, drain ordinary. This circuit intercept CH1 step up output voltage by cutting body Di line at STB OFF and control soft start .CH1 softstart output voltage from OV like a slope.
10. Nch DRIVER, Pch DRIVER

Internal Nch, Pch FET driver CMOS inverter type output circuit.
11. Load SW

It is a circuit, mounted in CH6, to control the Load SW. Hx56 terminal (1pin) is input terminal, and the HS6L terminals (36pin) are output terminals.
This control circuit can prevent the rush current at the time of SW ON because the soft start starts functioning at the time of start-up. In addition, this Load SW is provided with OCP function to prevent the IC from damage. Ensure that the IC is used within Load SW's rated current when used normally.
12. ON/OFF LOGIC

It is the voltage applied to STB terminal and can control the ON/OFF of CH1~CH6.
If the voltage more than 1.5 V is applied, then it becomes ON , but if open or OV is applied, then it becomes off, furthermore, it all the channels are turned off, then the whole IC will be in standby state. In addition, STB1, 2, 3, 4~STB6 terminals contain respectively a built-in pull-down resistor of about 400k $\Omega$.
PWM7 is the input terminal of the start signal and the light control signal of CH 7 . It becomes high if the voltage more than 2.1 V is applied and becomes Low if the voltage less than 0.4 V is applied. In addition, PWM7 terminal contains a built-in pull-down resistor of about $400 \mathrm{k} \Omega$.

## 13. SOFT START

It is a circuit to apply the soft start to the output voltage of DC/DC converter and prevent the rush current at the start-up. Soft start time varies with the channels.
a. CH1 • . . . reaches the target voltage in 3.0 ms .
b. $\mathrm{CH} 3,4$ • • reach the target voltage in 2.1 ms .
c. $\mathrm{CH} 2,5,6$ • • reach the target voltage in 5.3 ms .
d. $\mathrm{CH} 7 \cdot \cdots \cdot$ reach the target voltage in 8.2 ms .
14. Brightness controller

CH 7 have LED brightness controller.INV7 is output voltage that proportionate to PWM7 input DUTY and control LED current by external resister for setting (between INV7 and GND ).
15. OVP COMP7

In CH7, When LED is OPEN, INV7 become L and output voltage increase suddenly. If this condition continues, Lx7 voltage increase and exceed break down voltage. CH7 heve Over voltage protection (OVP). When VO7 pin is inputted over 28Vtyp, OVP stop CH7 function..OVP latch CH7 function and reset dy All STB=L

## -Setting method of IC peripheral components

(1) Design of feedback resistor constant



Reference voltage is
connected to GND inside conne
IC


VOUT7


$$
\begin{align*}
& \mathrm{CH} 1,3, \text { 4output voltage } \\
& \mathrm{V}=\frac{(\mathrm{R} 1+\mathrm{R} 2)}{\mathrm{R} 2} \times 0.8[\mathrm{~V}] \cdots \tag{1}
\end{align*}
$$

$$
\mathrm{V} 0=-\frac{\mathrm{R} 2}{\mathrm{R} 1} \times 1.25[\mathrm{~V}] \cdots(2)
$$

## CH2,6 output voltage

$$
\begin{equation*}
\mathrm{V}_{\mathrm{o}}=\frac{(\mathrm{R} 1+\mathrm{R} 2)}{\mathrm{R} 2} \times 1.0[\mathrm{~V}] \cdots(3 \tag{4}
\end{equation*}
$$

CH5 output voltage

CH 7 output voltage

$$
\mathrm{Io}=\frac{\mathrm{INV7}}{\mathrm{R} 3}[\mathrm{~A}] \cdots
$$

Fig. 34 Feedback resistor setting method
(a) $\mathrm{CH} 1,3,4$ setting

The reference voltage of $\mathrm{CH} 1,3,4$ ERROR AMP is 0.8 V . Please refer to Formula (1) in Fig. 33 for determining the output voltage.
This IC incorporates built-in phase compensation. Please refer to Applied Circuit Diagram for setting the values of R1 \& R2 and ensure that the setting values of R1 \& R2 are of the order of several hundred $k \Omega$.
(b) CH5 setting

The reference voltage of CH5's ERROR AMP is connected to GND inside the IC. Therefore, a high-accuracy regulator can be configured if setting by the feedback resistance between the outputs of VREF and CH5 as shown in Fig.33. Please refer to Formula (2) in Fig. 33 for determining the output voltage. R1 is recommended as more than $20 \mathrm{k} \Omega$ because the current capacity of VREF5 is about $100 \mu \mathrm{~A}$.
(c) $\mathrm{CH} 2,6$ setting

The reference voltage of CH7's ERROR AMP is 1.0V. Please refer to Formula (3) in Fig. 33 for determining the output voltage.
(d) CH 7 setting LED current is decided by Fig. 33 (4) formula. Please decide R3 value for LED current range.
(2) Points for attention in terms of PCB layout of base-plate

OFor a switching regulator, in principle a large current transiently flows through the route of power supply - coil output capacitor. Ensure that the wiring impedance is lowered as much as possible by making the pattern as wide as possible and the layout as short as possible.
Olnterference of power supply noise with feedback terminals (INV1~7,NON5) may cause the output voltage to oscillate. Ensure that the power supply noise's interference is avoided by making the wiring between feedback resistor and feedback terminal as short as possible.

## -PIN equivalent circuit

INV1~INV6
(Error amplifier's inversion input)


STB1234
(operating when $\mathrm{CH} 1 \sim 4$ ON/OFF switch is High)


```
VREF5
(CH5 Standard voltage output)
```



NON5
(Error amplifier's non-inversion input)


STB5,6
(operating when $\mathrm{CH} 5,6$ ON/OFF switch is High)


## PWM7

(CH7 start signal, LED modulated light signal input)


Hx1,2BAT,3,4, VOUT2 (Pch FET source terminal) Lx1,21,22,3,4(Nch,Pch FET drain terminal)

HS6H (PMOS high side SW input terminal) HS6L (OMOS high side SW input terminal)


PGND567
INV7(CH7 Error amplifier's inversion input)
LED(LED cathode connection terminal)


Hx1,2BAT(Pch FET FET source terminal) Lx1,21(Nch,Pch FET drain terminal) PGND1,2 (output stage earthing terminal)


Fig. 35 PIN equivalent circuit

## - Notes for use

1) Absolute Maximum Ratings

Although the quality of this product has been tightly controlled, deterioration or even destruction may occur if the absolute maximum ratings, such as for applied pressure and operational temperature range, are exceeded. Furthermore, we are unable to assume short or open mode destruction conditions. If special modes which exceed the absolute maximum ratings are expected, physical safely precautions such as fuses should be considered.
2) GND Potential

The potential of the GND pin should be at the minimum potential during all operation status. In addition, please try to do not become electric potential below GND for the terminal other than NON5 including the transient phenomenon in practice. Please do not go down below 0.3 V for the NON5 terminal with transient phenomenon and the like when you use.
3) Heat Design

Heat design should consider tolerance dissipation (Pd) during actual use and margins which should be set with plenty of room.
4) Short-circuiting Between Terminals and Incorrect Mounting When attaching to the printed substrate, pay special attention to the direction and proper placement of the IC. If the IC is attached incorrectly, it may be destroyed. Destruction can also occur when there is a short, which can be caused by foreign objects entering between outputs or an output and the power GND.
5) Operation in Strong Magnetic Fields

Exercise caution when operating in strong magnet fields, as errors can occur.
6) About common impedance

Please do sufficient consideration for the wiring of power source and GND with the measures such as lowering common impedance, making ripple as small as possible (making the wiring as thick and short as possible, dropping ripple from L.C) and the like.
7) Heat Protection Circuit (TSD circuit)

This IC has a built-in Temperature Protection Circuit (TSD circuit). The temperature protection circuit (TSD circuit) is only to cut off the IC from thermal runaway, and has not been designed to protect or guarantee the IC. Therefore, the user should not plan to activate this circuit with continued operation in mind.
8) Rush current at the time of power supply injection.

Because there are times when rush current flows instantaneously in internal logical uncertain state at the time of power source turning on with CMOS IC, please pay attention to the power source coupling capacity, the width of GND pattern wiring and power source, and the reel.
9) IC Terminal Input

This IC is a monolithic IC, and between each element there is a $P+$ isolation and $P$ substrate for element separation. There is a P-N junction formed between this P-layer and each element's N-layer, which makes up various parasitic elements. For example, when resistance and transistor are connected with a terminal as in fig.35:

OWhen GND>(terminal A) at the resistance, or GND>(terminal B) at the transistor (NPN), the P-N junction operates as a parasitic diode.
OAlso, when GND>(terminal B) at the transistor, a parasitic NPN transistor operates by the N-layer of other elements close to the aforementioned parasitic diode.
With the IC's configuration, the production of parasitic elements by the relationships of the electrical potentials is inevitable. The operation of the parasitic elements can also interfere with the circuit operation, leading to malfunction and even destruction. Therefore, uses which cause the parasitic elements to operate, such as applying voltage to the input terminal which is lower than the GND(P-substrate), should be avoided.


Fig.. 36 Simple Structure of Bipolar IC (Sample)

## -Ordering part number



UQFN036V5050


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|  |  | CLASSIII |  |

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NCV1077CSTBT3G XCL207A123CR-G MPM54304GMN-0002 MPM54304GMN-0004 MPM54304GMN-0003
XDPE132G5CG000XUMA1 MP8757GL-P MIC23356YFT-TR LD8116CGL HG2269M/TR OB2269 XD3526 U6215A U6215B U6620S
LTC3412IFE LT1425IS MAX25203BATJA/VY+ MAX77874CEWM + XC9236D08CER-G ISL95338IRTZ MP3416GJ-P BD9S201NUX-
CE2 MP5461GC-Z MPQ4415AGQB-Z MPQ4590GS-Z MAX38640BENT18+T MAX77511AEWB+ MAX20406AFOD/VY+


[^0]:    （＊）Recommend resistor value over $20 \mathrm{k} \Omega$ between VREF5 to NON5，because VREF5 current is under $100 \mu \mathrm{~A}$ ． This product is not designed for normal operation with in a radioactive environment．

