

2.7V to 5.5V

# 2.7V to 5.5V Input, 6.0A Integrated MOSFET Single Synchronous Buck DC/DC Converter

## BD9B600MUV

#### **General Description**

BD9B600MUV is a synchronous buck switching regulator with built-in low on-resistance power MOSFETs. This IC, which is capable of providing current up to 6A, features fast transient response by employing constant on-time control system. It offers high oscillating frequency at low inductance. With its original constant on-time control method which operates low consumption at light load, this product is ideal for equipment and devices that demand minimal standby power consumption.

#### **Features**

- Synchronous Single DC/DC Converter
- Constant on-time control suitable to Deep-SLLM
- **Over Current Protection**
- Short Circuit Protection
- **Thermal Shutdown Protection**
- Under Voltage Lockout Protection
- Adjustable Soft Start
- Power Good Output
- VQFN016V3030 Package (backside heat dissipation)

#### **Applications**

- Step-down Power Supply for DSPs, FPGAs, Microprocessors, etc.
- Laptop PCs/Tablet PCs/Servers
- LCD TVs
- Storage Devices (HDDs/SSDs)
- Printers, OA Equipment
- **Entertainment Devices**
- Distributed Power Supply, Secondary Power Supply

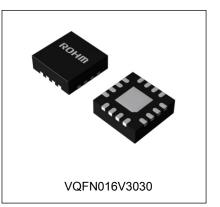
## **Typical Application Circuit**

#### **Key Specifications**

- Input Voltage Range:
- Output Voltage Range:
- 0.8 V to VPVIN x 0.8 V Maximum Operating Current: 6A (Max)
- Switching Frequency: 2MHz/1MHz (Typ)
- High-Side MOSFET ON Resistance: 30mΩ (Typ)
- Low-Side MOSFET ON Resistance: 30mΩ (Typ)
- Standby Current: 0µA (Typ)

#### Package VQFN016V3030

W (Typ) x D (Typ) x H (Max) 3.00 mm x 3.00 mm x 1.00 mm



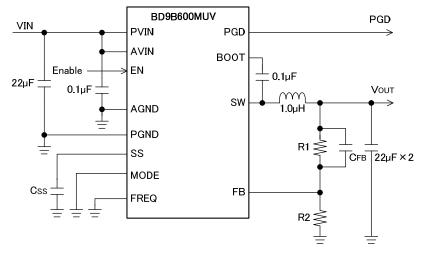
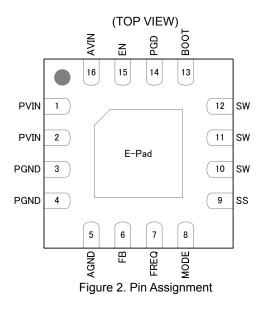


Figure 1. Application Circuit

OProduct structure : Silicon monolithic integrated circuit OThis product has no designed protection against radioactive rays

## **Pin Configuration**



## **Pin Descriptions**

Pin No.	Pin Name	Function
1, 2	PVIN	Power supply terminals for the switching regulator. These terminals supply power to the output stage of the switching regulator. Connecting a 22µF ceramic capacitor is recommended.
3, 4	PGND	Ground terminals for the output stage of the switching regulator.
5	AGND	Ground terminal for the control circuit.
6	FB	An inverting input node for the error amplifier and main comparator. See page 22 for how to calculate the resistance of the output voltage setting.
7	FREQ	Terminal for setting switching frequency. Connecting this terminal to ground makes switching to operate constant on-time corresponding to 2.0MHz. Connecting this terminal to AVIN makes switching to operate constant on-time corresponding to 1.0MHz. Please fix this terminal to AVIN or ground in operation.
8	MODE	Terminal for setting switching control mode. Connecting this terminal to AVIN forces the device to operate in the fixed frequency PWM mode. Connecting this terminal to ground enables the Deep-SLLM control and the mode is automatically switched between the Deep-SLLM control and fixed frequency PWM mode. Please fix this terminal to AVIN or ground in operation.
9	SS	Terminal for setting the soft start time. The rise time of the output voltage can be specified by connecting a capacitor to this terminal. See page 23 for how to calculate the capacitance.
10, 11, 12	SW	Switch nodes. These terminals are connected to the source of the High-Side MOSFET and drain of the Low-Side MOSFET. Connect a bootstrap capacitor of 0.1 $\mu$ F between these terminals and BOOT terminal. In addition, connect an inductor of 0.47 $\mu$ H to 1 $\mu$ H (FREQ=L), 1 $\mu$ H to 1.5 $\mu$ H (FREQ=H) considering the direct current superimposition characteristic.
13	BOOT	Terminal for bootstrap. Connect a bootstrap capacitor of 0.1 $\mu$ F between this terminal and SW terminals. The voltage of this terminal is the gate drive voltage of the High-Side MOSFET.
14	PGD	A "Power Good" terminal, an open drain output. Use of pull up resistor is needed. See page 17 for how to specify the resistance. When the FB terminal voltage reaches more than 80% of 0.8 V, the internal Nch MOSFET turns off and the output turns High.
15	EN	Enable terminal. Turning this terminal signal Low (0.8V or lower) forces the device to enter the shutdown mode. Turning this terminal signal High (2.0V or higher) enables the device. This terminal must be terminated.
16	AVIN	Terminal for supplying power to the control circuit of the switching regulator. Connecting a 0.1µF ceramic capacitor is recommended. This terminal must be connected to PVIN.
-	E-Pad	A backside heat dissipation exposed pad. Connecting to the internal PCB ground plane by using multiple vias provides excellent heat dissipation characteristics.

## Block Diagram

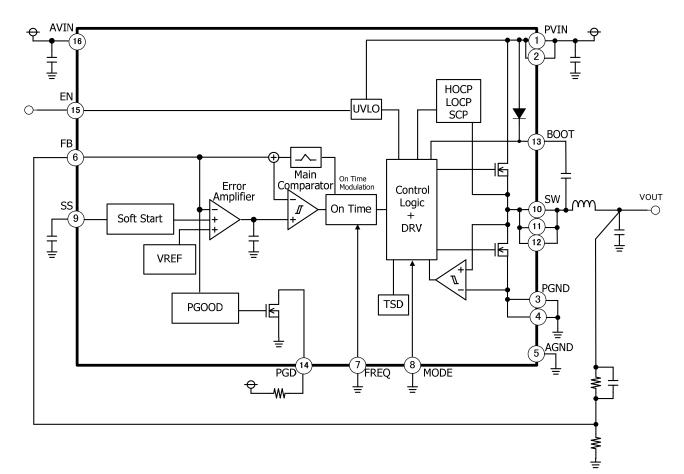


Figure 3. Block Diagram

## **Description of Blocks**

VREF

The VREF block generates the internal reference voltage.

UVLO

The UVLO block is for Under Voltage lockout protection. It will shut down the IC when VIN falls to 2.45 V (Typ) or lower. The threshold voltage has a hysteresis of 100mV (Typ).

TSD

The TSD block is for thermal protection. The thermal protection circuit shuts down the device when the internal temperature of IC rises to  $175^{\circ}C$  (Typ) or higher. Thermal protection circuit resets when the temperature falls. The circuit has a hysteresis of  $25^{\circ}C$  (Typ).

Soft Start

The Soft Start circuit slows down the rise of output voltage during start-up and controls the current, which allows the prevention of output voltage overshoot and inrush current. A built-in soft start function is provided and a soft start is initiated in 1msec (Typ) when the SS terminal is open.

- Control Logic + DRV This block is a DC/DC driver. A signal from On Time is applied to drive the MOSFETs.
- PGOOD

When the FB terminal voltage reaches more than 80% of 0.8 V, the Nch MOSFET of the built-in open drain output turns off and the output turns High.

HOCP/LOCP/SCP

After soft start is completed and in condition where output voltage is below 70% (Typ) of voltage setting, it counts the number of times of which current flowing in High side FET or Low side FET reaches over current limit. When 512 times is counted it stops operation for 1m sec (Typ) and re-operates. Counting is reset when output voltage is above 80% (Typ) of voltage setting or when EN, UVLO, SCP function is re-operated.

Error Amplifier

Adjusts Main Comparator input to make internal reference voltage equal to FB terminal voltage.

Main Comparator

Main comparator compares Error Amplifier output and FB terminal voltage. When FB terminal voltage becomes low it outputs High and reports to the On Time block that the output voltage has dropped below control voltage.

On Time

This is a block which creates On Time. Requested On Time is created when Main Comparator output becomes High. On Time is adjusted to restrict frequency change even with I/O voltage change.

## Absolute Maximum Ratings (Ta = 25°C)

Parameter	Symbol	Rating	Unit
Supply Voltage	VPVIN, VAVIN	-0.3 to +7	V
EN Terminal Voltage	VEN	-0.3 to +7	V
MODE Terminal Voltage	VMODE	-0.3 to +7	V
FREQ Terminal Voltage	VFREQ	-0.3 to +7	V
PGD Terminal Voltage	Vpgd	-0.3 to +7	V
Voltage from GND to BOOT	VBOOT	-0.3 to +14	V
Voltage from SW to BOOT	⊿Vвоот	-0.3 to +7	V
FB Terminal Voltage	Vfb	-0.3 to +7	V
SW Terminal Voltage	Vsw	-0.3 to VPVIN + 0.3	V
Output Current	Ιουτ	6.5	А
Allowable Power Dissipation <sup>(Note 1)</sup>	Pd	2.66	W
Operating Temperature Range	Topr	-40 to 85	°C
Storage Temperature Range	Tstg	-55 to 150	°C

(Note 1) VQFN016V3030: Derate by 21.3mW when operating above 25°C

PCB size: 70mm x 70mm x 1.6mm when mounted on a 4-layer glass epoxy board (copper foil area: 70mm x 70mm)

Copper foil thickness: Front side and reverse side 70µm be used, 2nd and 3rd 35µm be used.

**Caution:** Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

## Recommended Operating Conditions (Ta= -40°C to +85°C)

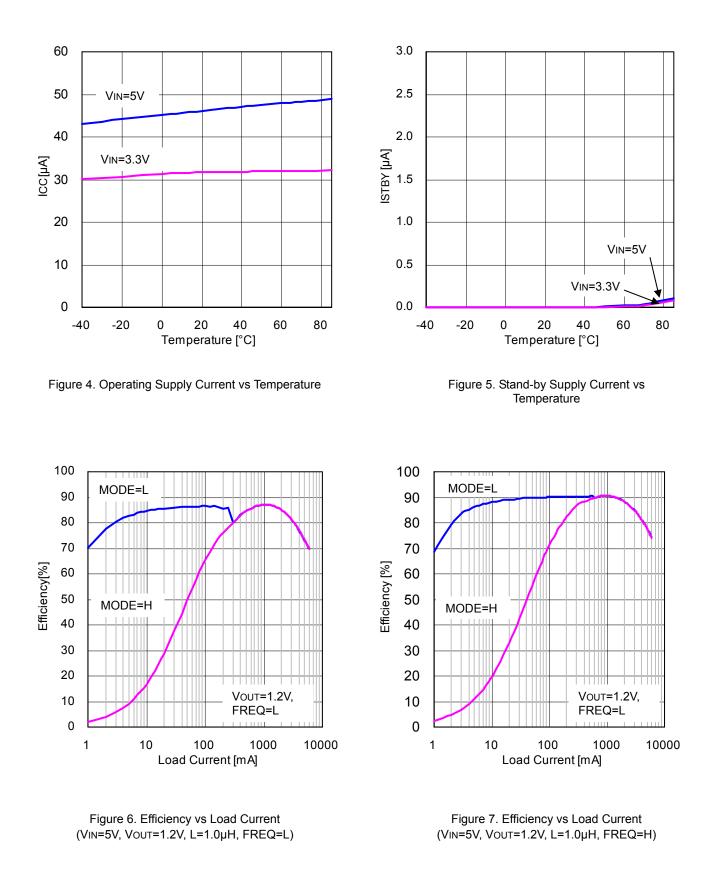
Parameter	Symbol	Min	Тур	Max	Unit
Supply Voltage	Vpvin, Vavin	2.7	-	5.5	V
Output Current (Note 2)	Ιουτ	-	-	6	А
Output Voltage Range	VRANGE	0.8	-	VPVIN × 0.8	V

(Note 2) Pd, ASO should not be exceeded

## Electrical Characteristics (Unless otherwise specified Ta=25°C, V<sub>AVIN</sub> = V<sub>PVIN</sub> = 5V, V<sub>EN</sub> = 5V, V<sub>MODE</sub> = GND)

Parameter	Symbol	Min	Тур	Max	Unit	Conditions
AVIN pin	1	11			l.	
Standby Supply Current	Istb	-	0	10	μA	EN=GND
Operating Supply Current	Icc	-	45	80	μA	FREQ=AVIN, IOUT=0mA Non switching
UVLO Detection Threshold	VUVLO1	2.35	2.45	2.55	V	Vin falling
UVLO Release Threshold	VUVLO2	2.425	2.55	2.7	V	Vin rising
UVLO Hysteresis	VUVLOHYS	50	100	200	mV	
Enable						
EN Input High Level Voltage	VENH	2.0	-	-	V	
EN Input Low Level Voltage	VENL	-	-	0.8	V	
EN Input Current	IEN	-	5	10	μA	EN=5V
Reference Voltage, Error Amplifier						
FB Terminal Voltage	Vfb	0.792	0.8	0.808	V	
FB Input Bias Current	IFB	-	-	1	μA	FB=0.8V
Internal Soft Start Time	Tss	0.5	1.0	2.0	ms	With internal constant
Soft Start Terminal Current	lss	0.5	1.0	2.0	μA	
Control						
FREQ Input High Level Voltage	Vfrqh	VAVIN-0.3	-	-	V	
FREQ Input Low Level Voltage	VFRQL	-	-	0.3	V	
MODE Input High Level Voltage	VMODEH	VAVIN-0.3	-	-	V	
MODE Input Low Level Voltage	VMODEL	-	-	0.3	V	
On time1	ONT1	96	120	144	ns	Vout=1.2V, FREQ=GND
On time2	ONT2	192	240	288	ns	VOUT=1.2V, FREQ=AVIN
Power Good						
Power Good Rising Threshold	Vpgdh	75	80	85	%	FB rising, VPGDH=FB/VFBx100
Power Good Falling Threshold	Vpgdl	65	70	75	%	FB falling, VPGDL=FB/VFBx100
Output Leakage Current	Ilkpgd	-	0	5	μA	PGD=5V
Power Good On Resistance	Rpgd	-	100	200	Ω	
Power Good Low Level Voltage	$PGDV_L$	-	0.1	0.2	V	I <sub>PGD</sub> =1mA
SW						
High Side FET On Resistance	Ronh	-	30	60	mΩ	BOOT - SW = 5 V
Low Side FET On Resistance	RONL	-	30	60	mΩ	
High Side Output Leakage Current	RILH	-	0	10	μA	No switching
Low Side Output Leakage Current	RILL	-	0	10	μA	No switching

## **Typical Performance Curves**



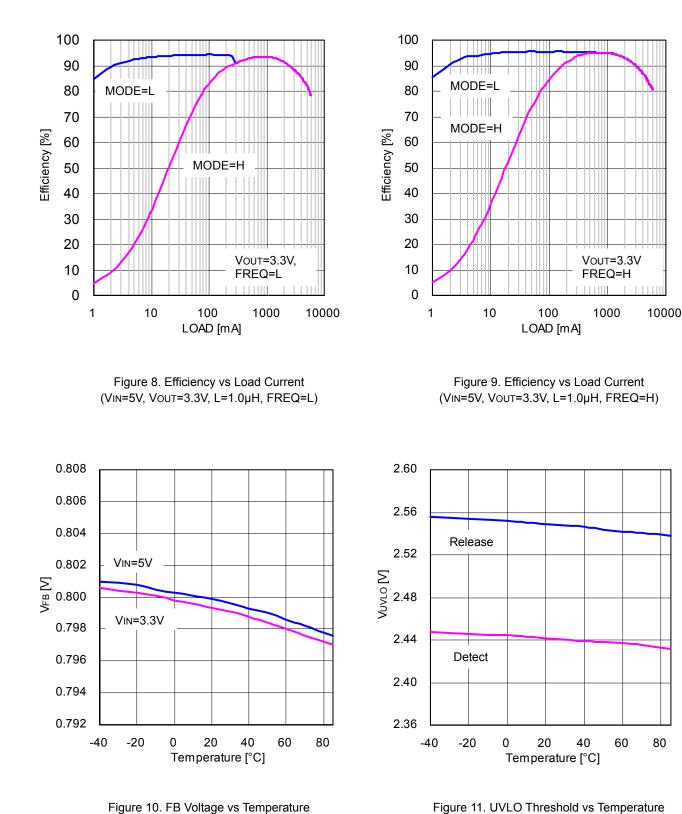


Figure 11. UVLO Threshold vs Temperature

## Datasheet

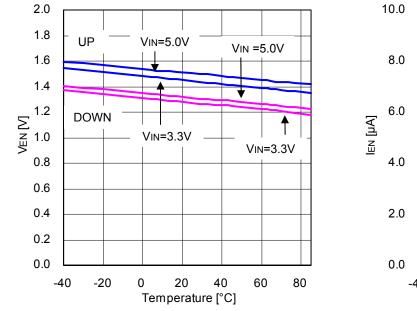
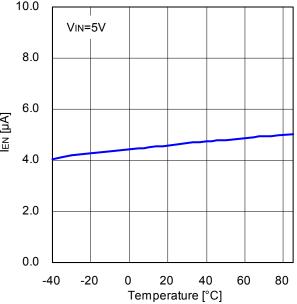


Figure 12. EN Threshold vs Temperature





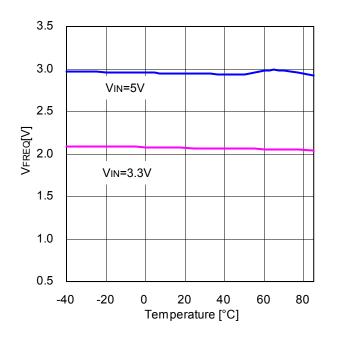


Figure 14. FREQ Threshold vs Temperature

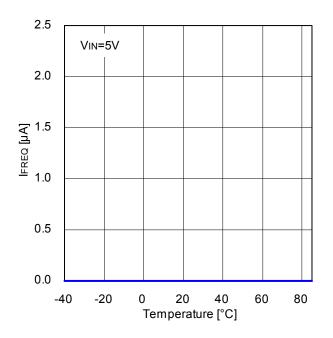


Figure 15. FREQ Input Current vs Temperature

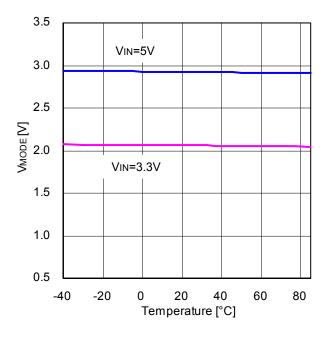
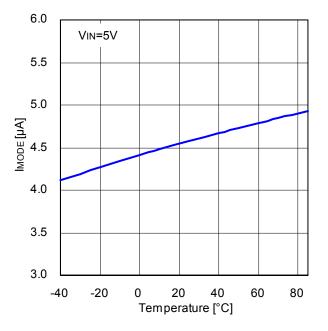
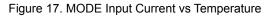


Figure 16. MODE Threshold Voltage vs Temperature





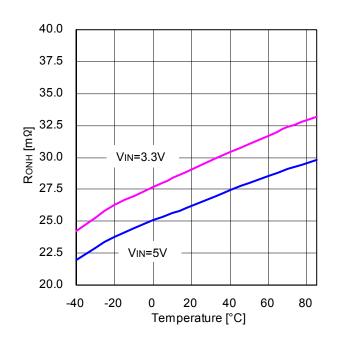


Figure 18. High Side ON-Resistance

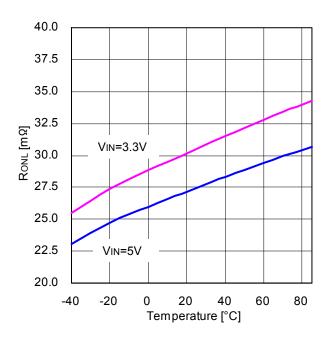


Figure 19. Low Side ON-Resistance

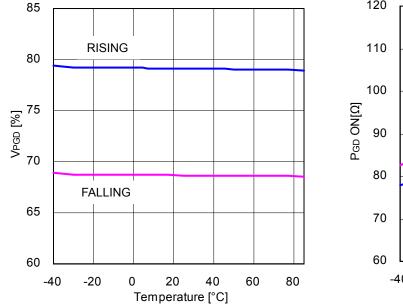
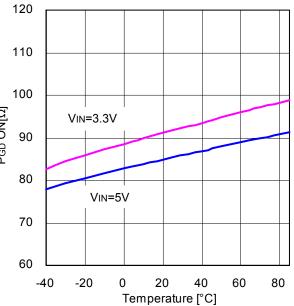


Figure 20. PGD Threshold vs Temperature





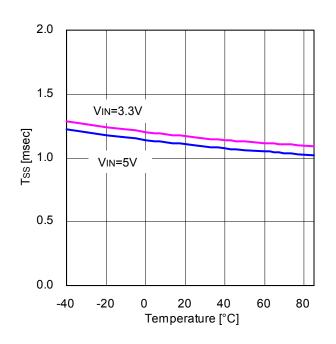


Figure 22. Soft Start Time vs Temperature

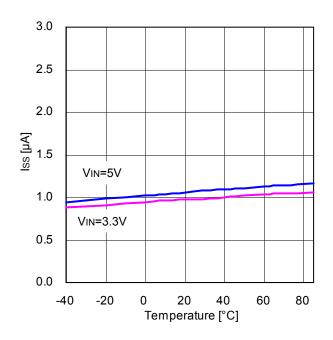


Figure 23. SS Terminal Current vs Temperature

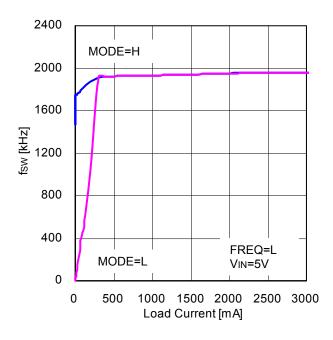
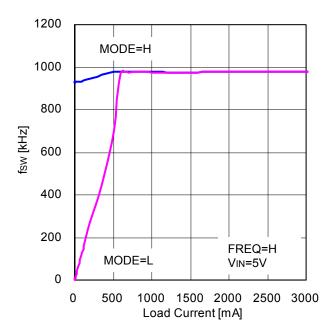
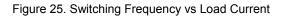


Figure 24. Switching Frequency vs Load Current





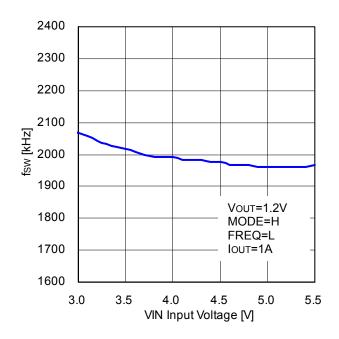


Figure 26. Switching Frequency vs Input Voltage

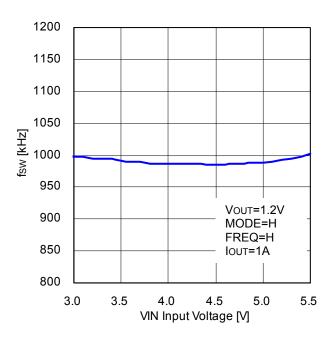
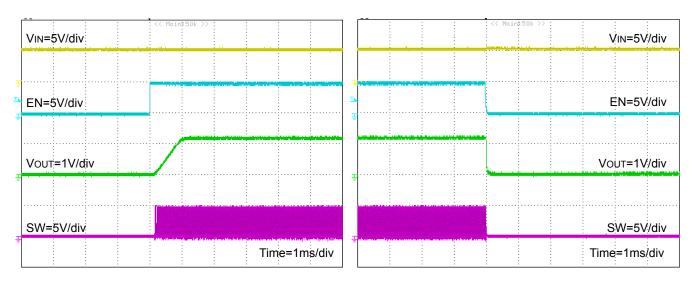


Figure 27. Switching Frequency vs Input Voltage

## Datasheet

## Typical Performance Curves - continued



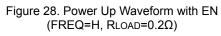
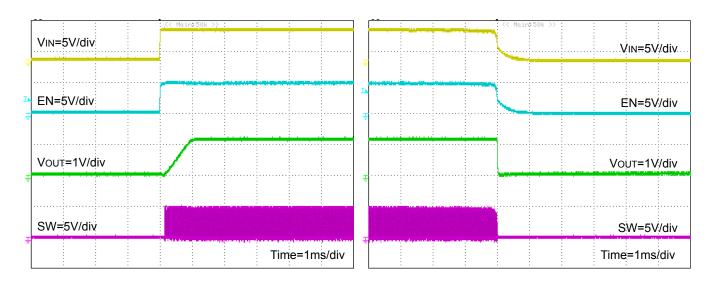


Figure 29. Power Down Waveform with EN (FREQ=H, RLOAD=0.2 $\Omega$ )



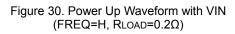


Figure 31. Power Down Waveform with VIN (FREQ=H, RLOAD=0.2 $\Omega$ )

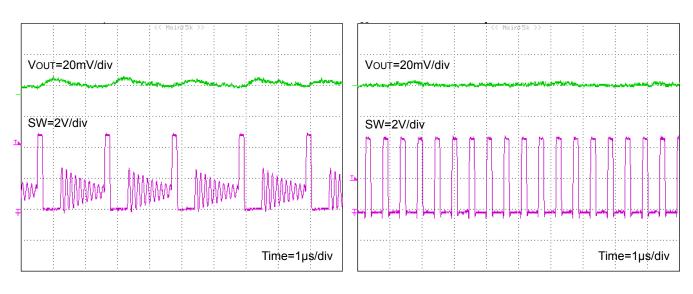




Figure 33. Switching Waveform (VIN=5V, VOUT=1.2V, FREQ=L, IOUT=6A)

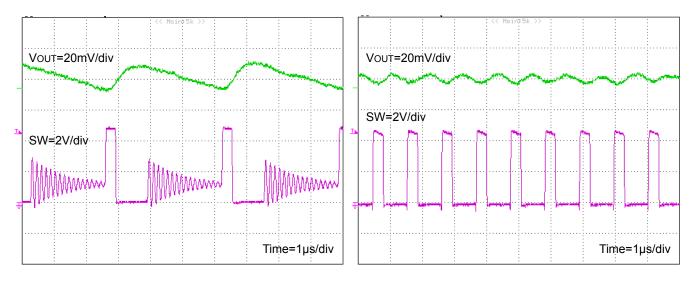


Figure 34. Switching Waveform (VIN=5V, VOUT=1.2V, FREQ=H, IOUT=0.1A)

Figure 35. Switching Waveform (VIN=5V, VOUT=1.2V, FREQ=H, IOUT=6A)

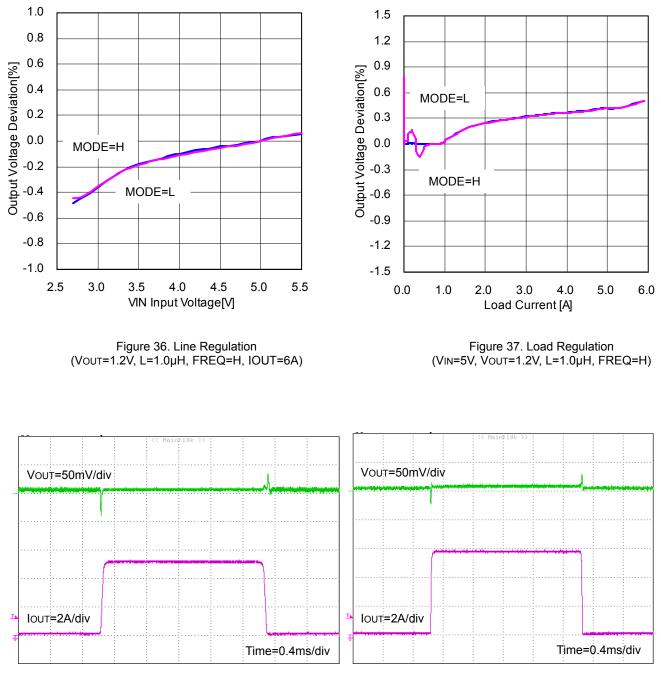


Figure 38. Load Transient Response IOUT=0.1A to 5A (VIN=5V, VOUT=1.2V, FREQ=L, MODE=L, COUT=Ceramic 44µF)

Figure 39. Load Transient Response IOUT=0.1A to 6A (VIN=5V, VOUT=1.2V, FREQ=L, MODE=H, COUT=Ceramic 44 $\mu$ F)

## Application Example(s)

#### 1. Basic Operation

## (1) DC/DC Converter operation

BD9B600MUV is a synchronous rectifying step-down switching regulator that achieves faster transient response by employing constant on-time control system. It utilizes switching operation in PWM (Pulse Width Modulation) mode for heavier load, while it utilizes Deep-SLLM (Deep\_Simple Light Load Mode) control for lighter load to improve efficiency.

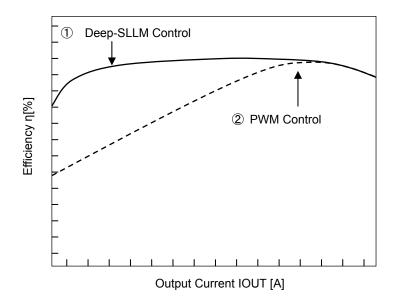
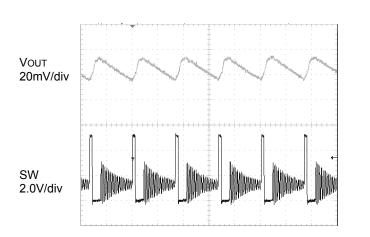


Figure 40. Efficiency (Deep-SLLM Control and PWM Control)



Deep-SLLM Control

Figure 41. Switching Waveform at Deep-SLLM Control (VIN=5.0V, VOUT=1.2V, IOUT=100mA) 2 PWM Control

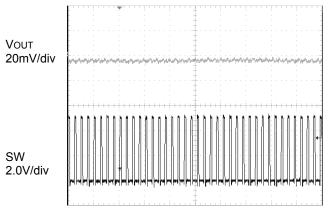


Figure 42. Switching Waveform at PWM Control (VIN=5.0V, VOUT=1.2V, IOUT=6A)

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#### (2) Enable Control

The IC shutdown can be controlled by the voltage applied to the EN terminal. When VEN reaches 2.0 V(Typ), the internal circuit is activated and the IC starts up. To enable shutdown control with the EN terminal, the shutdown interval (Low level interval of EN) must be set to 100  $\mu$ s or longer. Startup by EN must be at the same time or after the input of power supply voltage.

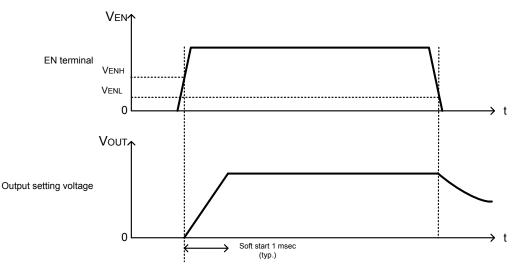


Figure 43. Start Up and Down with Enable

#### (3) Power Good

When the output voltage reaches more than 80% of the voltage setting, the open drain NMOSFET, internally connected to the PGD terminal, turns off and the PGD terminal turns to Hi-z condition. Also when the output voltage falls below 70% of voltage setting, the open drain NMOS FET turns on and PGD terminal pulls down with 100 $\Omega$ . Connecting a pull up resistor (10K $\Omega$  to 100K $\Omega$ ) is recommended.

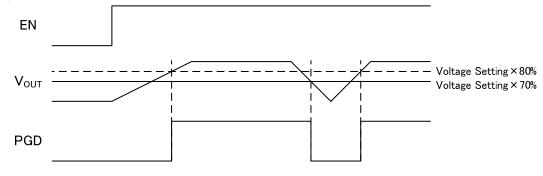


Figure 44. Power Good Timing Chart

#### (4) Soft Start

When EN terminal is turned High, Soft Start operates and output voltage gradually rises. With the Soft Start Function, over shoot of output voltage and rush current can be prevented. Rising time of output voltage when SS terminal is open is 1msec (Typ). Capacitor connected to SS terminal makes rising time more than 1msec. Please refer to page 23 for the method of setting rising time.

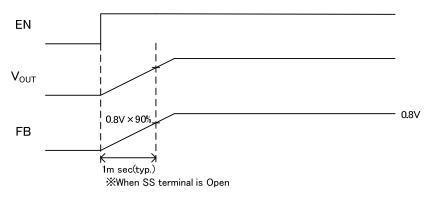


Figure 45. Soft Start Timing Chart

#### 2. Protection

## The protective circuits are intended for prevention of damage caused by unexpected accidents. Do not use them for continuous protective operation

#### (1) Over Current Protection (OCP) / Short Circuit Protection (SCP)

Setting (Typ) of over current protection are 8A (lower MOSFET) and 11A (upper MOSFET). When OCP is triggered, over current protection is realized by restricting On / Off Duty of current flowing in upper and lower MOSFET by each switching cycle. Also, if Over current protection operates 512 cycles in a condition where FB terminal voltage reaches below 70% of internal standard voltage, Short Circuit protection (SCP) operates and stops switching for 1msec (Typ) before it initiates restart. However, during startup, Short circuit protection will not operate even if the IC is still in the SCP condition.

EN terminal	PGD	Startup	Over current protection	Short circuit protection
		While start up	Valid	Invalid
More than 2.0V	L	Startup completed	Valid	Valid
	Н	*	Valid	Invalid
Less than 0.8V	*	*	Invalid	Invalid

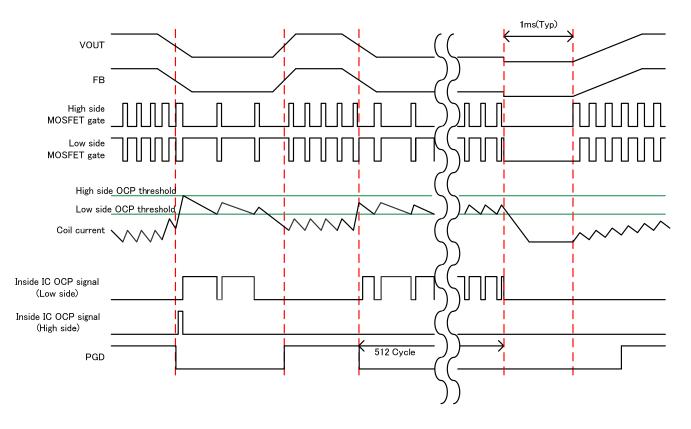


Figure 46. Short Circuit Protection (SCP) Timing Chart

## (2) Under Voltage Lockout Protection (UVLO) The Under Voltage Lockout Protection circuit monitors the AVIN terminal voltage. The operation enters standby when the AVIN terminal voltage is 2.45V (Typ) or lower. The operation starts when the AVIN terminal voltage is 2.55V (Typ) or higher. $V_{IN}$ UVLO OFF ∬hys UVLO ON 0V VOUT Soft start FB terminal High side MOSFET gate Low side MOSFET gate UVLO Normal operation Normal operation

Figure 47. UVLO Timing Chart

#### (3) Thermal Shutdown

When the chip temperature exceeds  $Tj=175^{\circ}C$  (Typ), the DC/DC converter output is stopped. The circuits are automatically restored to normal operation when the chip temperature falls. It has a hysteresis of  $25^{\circ}C$  (Typ). The thermal shutdown circuit is intended for shutting down the IC from thermal runaway in an abnormal state with the temperature exceeding Tjmax=150^{\circ}C. It is not meant to protect or guarantee the soundness of the application. Do not use the function of this circuit for application protection design.

## **Application Example**

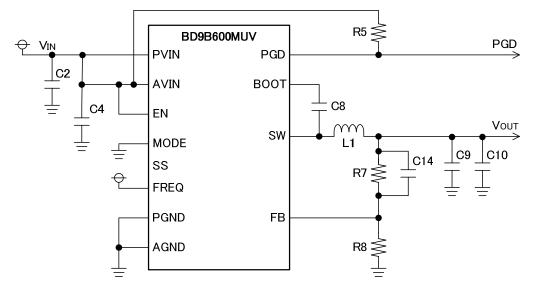


Figure 48. Application Circuit

Table 2. Recommended Component Values (VIN=5V, FREQ=H)
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Reference			Vout			Description
Designator	1.0V	1.2V	1.5V	1.8V	3.3V	Description
R5	100kΩ	100kΩ	100kΩ	100kΩ	100kΩ	-
R7	51kΩ	75kΩ	160kΩ	150kΩ	75kΩ	-
R8	20Ω	150kΩ	180kΩ	120kΩ	24kΩ	-
C2 <sup>(Note 3)</sup>	22µF	22µF	22µF	22µF	22µF	10V, X5R, 3216
C4	0.1µF	0.1µF	0.1µF	0.1µF	0.1µF	25V, X5R, 1608
C8 <sup>(Note 4)</sup>	0.1µF	0.1µF	0.1µF	0.1µF	0.1µF	-
C9	22µF	22µF	22µF	22µF	22µF	6.3V, X5R, 3225
C10	22µF	22µF	22µF	22µF	22µF	6.3V, X5R, 3225
C14	120pF	120pF	150pF	150pF	180pF	-
L1	1.0µH	1.0µH	1.0µH	1.0µH	1.0µH	TOKO, FDSD0630

Table 3. Recommended Component Values (VIN=5V, FREQ=L)

Reference		Description				
Designator	1.0V	1.2V	1.5V	1.8V	3.3V	Description
R5	100kΩ	100kΩ	100kΩ	100kΩ	100kΩ	-
R7	51kΩ	75kΩ	160kΩ	150kΩ	75kΩ	-
R8	200kΩ	150kΩ	180kΩ	120kΩ	24kΩ	-
C2 <sup>(Note 3)</sup>	22µF	22µF	22µF	22µF	22µF	10V, X5R, 3216
C4	0.1µF	0.1µF	0.1µF	0.1µF	0.1µF	25V, X5R, 1608
C8 <sup>(Note 4)</sup>	0.1µF	0.1µF	0.1µF	0.1µF	0.1µF	-
C9	22µF	22µF	22µF	22µF	22µF	6.3V, X5R, 3225
C10	22µF	22µF	22µF	22µF	22µF	6.3V, X5R, 3225
C14	100pF	100pF	120pF	120pF	120pF	-
L1	1.0µH	1.0µH	1.0µH	1.0µH	1.0µH	TOKO, FDSD0630

(Note 3) For capacitance of input capacitor take temperature characteristics, DC bias characteristics, etc. into consideration to set minimum value to no less than 10µF.

(Note 4) For capacitance of bootstrap capacitor take temperature characteristics, DC bias characteristics, etc. into consideration to set minimum value to no less than 0.047 µF.

\* Evaluation using the actual machine must be done for above constant is only a value on our evaluation board.

## Selection of Components Externally Connected

#### 1. Output LC Filter Constant

In order to supply a continuous current to the load, the DC/DC converter requires an LC filter for smoothing the output voltage. It is recommended to use inductors of values  $0.47\mu$ H to  $1.0\mu$ H when FREQ=L or  $1.0\mu$ H to  $1.5\mu$ H at FREQ=H.

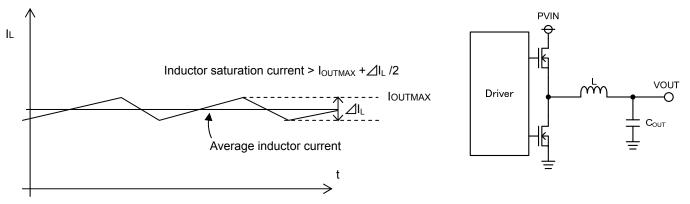


Figure 49. Waveform of current through inductor

Figure 50. Output LC filter circuit

Inductor ripple current  $\Delta IL$ 

$$\Delta I_{L} = V_{OUT} \times (V_{IN} - V_{OUT}) \times \frac{1}{V_{IN} \times F_{OSC} \times L} = 912 [\text{mA}]$$

Where: VIN= 5V VOUT= 1.2V L=1.0µH FOSC=1MHz (switching frequency)

The saturation current of the inductor must be larger than the sum of the maximum output current and 1/2 of the inductor ripple current  $\Delta$ IL.

The output capacitor COUT affects the output ripple voltage characteristics. The output capacitor COUT must satisfy the required ripple voltage characteristics.

The output ripple voltage can be represented by the following equation.

$$\Delta V_{RPL} = \Delta I_L \times (R_{ESR} + \frac{1}{8 \times C_{OUT} \times F_{OSC}}) [V]$$

where RESR is the Equivalent Series Resistance (ESR) of the output capacitor.

\* The capacitor rating must allow a sufficient margin with respect to the output voltage. The output ripple voltage can be decreased with a smaller ESR. A ceramic capacitor of about 22  $\mu$ F to 47  $\mu$ F is recommended.

\*Be careful of total capacitance value, when additional capacitor  $C_{LOAD}$  is connected in addition to output capacitor  $C_{OUT}$ . Use maximum additional capacitor  $C_{LOAD}$  (Max) condition which satisfies the following condition.

Maximum starting inductor bottom ripple current  $IL_{START}$ < Low side FET Over Current limit 6.5A(min)

Maximum starting inductor ripple bottom current ILSTART can be expressed using the following equation.

 $IL_{START} = Maximum \ starting \ output \ current(I_{OMAX}) + Charge \ current \ to \ output \ capacitor(I_{CAP}) - \frac{\Delta I_L}{2}$ 

Charge current to output capacitor I<sub>CAP</sub> can be expressed using the following equation.

$$I_{CAP} = \frac{(C_{OUT} + C_{LOAD}) \times V_{OUT}}{T_{SS}} [A]$$

For example, given VIN= 5V, VOUT= 3.3V, L=  $1.5\mu$ H, switching frequency F<sub>OSC</sub>= 1.2MHz(Max), Output capacitor COUT=  $44\mu$ F, Soft Start time Tss= 0.5ms(Min), and load current during soft start Ioss= 6A, maximum C<sub>LOAD</sub> can be computed using the following equation.

$$C_{LOAD}(max) < \frac{(6.5 - I_{OSS} + \Delta I_{L}/2) \times T_{SS}}{V_{OUT}} - C_{OUT} = 78.9 [\mu F]$$

If the value of CLOAD is large, and cannot meet the above equation, adjust the value of the capacitor Css to meet the condition below.

$$C_{LOAD}(max) < \frac{(6.5 - I_{OSS} + \Delta I_{L}/2) \times V_{FB}}{V_{OUT} \times I_{SS}} \times C_{SS} - C_{OUT}$$

(Refer to the following items (3) Soft Start Setting equation of time Tss and soft-start value of the capacitor to be connected to the Css.)

For example, given VIN = 5V, VOUT = 3.3V, L =  $1.5\mu$ H, load current during soft start IOSS = 6A, switching frequency FOSC= 1.2MHz (Max), Output capacitor COUT =  $44\mu$ F, VFB = 0.792V(Max), ISS =  $3.6\mu$ A(Max), with CLOAD =  $220\mu$ F, capacitor CSS is computed as follows.

$$C_{SS} > \frac{V_{OUT} \times I_{SS}}{(6.5 - I_{OSS} + \Delta I_L/2) \times V_{FB}} \times (C_{LOAD} + C_{OUT}) = 2713 [\text{pF}]$$

#### 2. Output Voltage Setting

The output voltage value can be set by the feedback resistance ratio.

For stable operation, it is recommended to use feedback resistance R1 of more than  $20k\Omega$ .

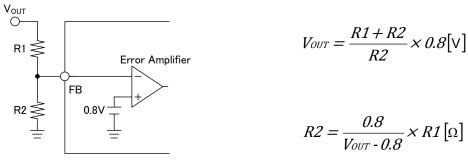


Figure 51. Feedback Resistor Circuit

#### 3. Soft Start Setting

Turning the EN terminal signal High activates the soft start function. This causes the output voltage to rise gradually while the current at startup is placed under control. This allows the prevention of output voltage overshoot and inrush current. The rise time depends on the value of the capacitor connected to the SS terminal.

 $T_{SS} = (C_{SS} \times V_{FB})/I_{SS}$   $C_{SS} = (I_{SS} \times T_{SS})/V_{FB}$   $T_{SS} : \text{Soft Start Time}$   $C_{SS} : \text{Capacitor connected to Soft Start Time Terminal}$   $V_{FB} : \text{FB Terminal Voltage (0.8V (Typ))}$ 

 $I_{ss}$  : Soft Start Terminal Source Current (1.0µA(Typ))

with 
$$C_{ss} = 0.01 \mu F$$
,  
 $T_{ss} = (0.01 [\mu F] \times 0.8 [V]) / 1.0 [\mu A]$   
 $= 8.0 [msec]$ 

Turning the EN terminal signal High with the SS terminal open or with the terminal signal High (no capacitor connected) causes the output voltage to rise in 1msec (Typ).

#### 4. FB Capacitor

Generally, in fixed ON time control (hysteresis control), sufficient ripple voltage in FB voltage is needed to operate comparator stably. Regarding this IC, by injecting ripple voltage to FB voltage inside IC it is designed to correspond to low ESR output capacitor. Please set the FB capacitor within the range of the following expression to inject an appropriate ripple.

$$\frac{V_{OUT} \times (1 - \frac{V_{OUT}}{V_{IN}})}{f_{SW} \times 7.65 \times 10^{3}} < C_{FB} < \frac{V_{OUT} \times (1 - \frac{V_{OUT}}{V_{IN}})}{f_{SW} \times 3.3 \times 10^{3}}$$

VIN : Input Voltage Vour : Output Voltage fsw : Switching Frequency

## **PCB Layout Design**

In the step-down DC/DC converter, a large pulse current flows into two loops. The first loop is the one into which the current flows when the High-Side FET is turned ON. The flow starts from the input capacitor CIN, runs through the FET, inductor L and output capacitor COUT and back to GND of CIN via GND of COUT. The second loop is the one into which the current flows when the Low-Side FET is turned on. The flow starts from the Low-Side FET, runs through the inductor L and output capacitor COUT and back to GND of the Low-Side FET via GND of COUT. Route these two loops as thick and as short as possible to allow noise to be reduced for improved efficiency. It is recommended to connect the input and output capacitors directly to the GND plane. The PCB layout has a great influence on the DC/DC converter in terms of all of the heat generation, noise and efficiency characteristics.

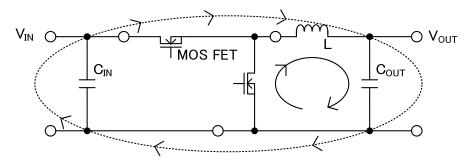


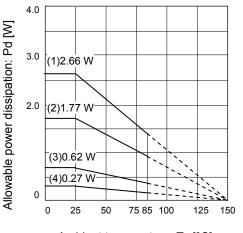
Figure 52. Current Loop of Buck Converter

Accordingly, design the PCB layout considering the following points.

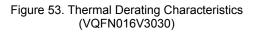
- Connect an input capacitor as close as possible to the IC PVIN terminal on the same plane as the IC.
- If there is any unused area on the PCB, provide a copper foil plane for the GND node to assist heat dissipation from the IC and the surrounding components.
- Switching nodes such as SW are susceptible to noise due to AC coupling with other nodes. Route the coil pattern as thick and as short as possible.
- Provide lines connected to FB far from the SW nodes.
- Place the output capacitor away from the input capacitor in order to avoid the effect of harmonic noise from the input.

#### **Power Dissipation**

When designing the PCB layout and peripheral circuitry, sufficient consideration must be given to ensure that the power dissipation is within the allowable dissipation curve.



Ambient temperature: Ta [°C]

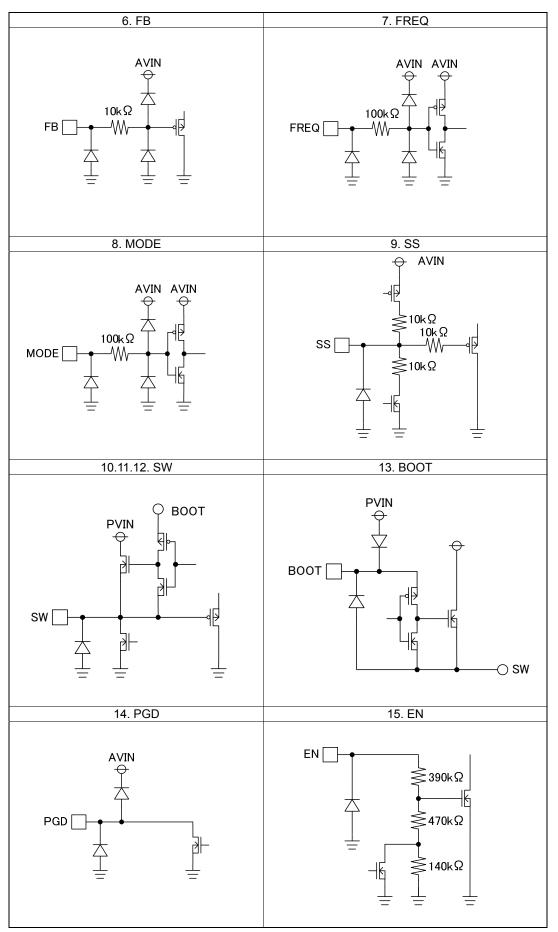


- (1) 4-layer board (surface heat dissipation copper foil 5505 mm<sup>2</sup>) (copper foil laminated on each layer)  $\theta_{JA} = 47.0^{\circ}C/W$
- (2) 4-layer board (surface heat dissipation copper foil 6.28 mm<sup>2</sup>) (copper foil laminated on each layer) θJA = 70.62°C/W
- (3) 1-layer board (surface heat dissipation copper foil 6.28 mm<sup>2</sup>)  $\theta_{JA} = 201.6^{\circ}C/W$

(4

Board specification: Glass-Epoxy, 70mm x 70mm x 1.6mm Copper foil thickness: Front side and reverse side 70µm be used, 2nd and 3rd 35µm be used.

## I/O equivalence circuit(s)



## **Operational Notes**

#### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

#### 2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

#### 3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

#### 4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

#### 5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded, the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

#### 6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

#### 7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

#### 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

## 9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

#### 10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

## **Operational Notes – continued**

#### 11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

#### 12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

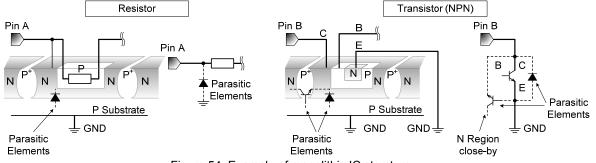


Figure 54. Example of monolithic IC structure

#### 13. Ceramic Capacitor

When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

## 14. Area of Safe Operation (ASO)

Operate the IC such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

#### 15. Thermal Shutdown Circuit(TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (Tj) will rise which will activate the TSD circuit that will turn OFF all output pins. When the Tj falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

## 16. Over Current Protection Circuit (OCP)

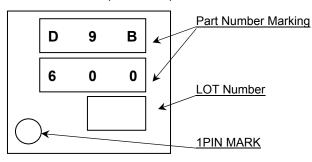
This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

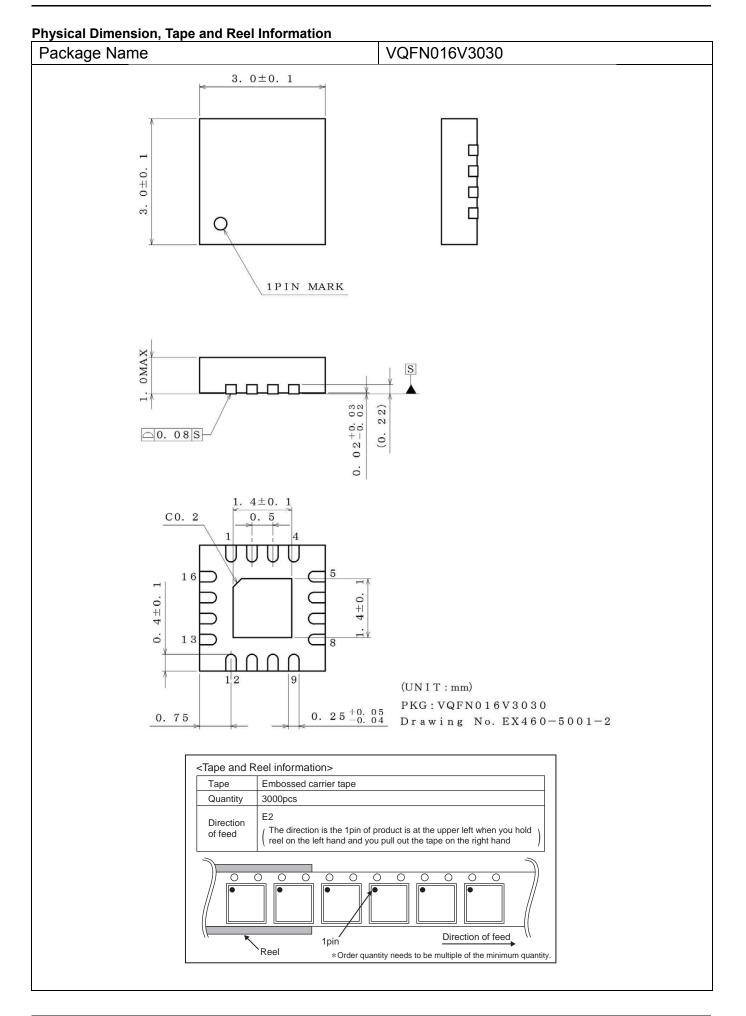
## **Ordering Information**



## **Marking Diagrams**

VQFN016V3030 (TOP VIEW)





## **Revision History**

Date	Revision	Changes
18.Jun.2015	001	New Release

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