# 4-output DVD System Video Driver 

## - Description

The BH76071FJ is a system video driver that integrated three output wideband video drivers with built-in LPF corresponding to the high-definition television and one output video driver with built-in LPF corresponding to the composite video signal in 1chip. It is suitable for the set not equipped with type $S$ terminal.

## - Features

1) Built-in 4-output video driver, supporting $P Y, P B, P R$, and CVBS
2) Supports D4 broadband standard
3) Built-in 6 dB amp
4) Built-in LPF for noise elimination (component: $f=30 \mathrm{MHz}\left(8^{\text {th }}\right) / 13.5 \mathrm{MHz}\left(6^{\text {th }}\right)$, composite: $f=6.75 \mathrm{MHz}\left(6^{\text {th }}\right)$ )
5) Built-in mute function
6) Enables two load drivers per driver channel
7) Usable for output capacitor less drivers (only 1 channel at same time)

## -Applications

DVD/BD players, DVD/BD recorders, and other video devices such as DSC, DVC, STB, TV

- Absolute Maximum Ratings

| Parameter | Symbol | Ratings | Unit |
| :--- | :---: | :---: | :---: |
| Supply voltage | VCCmax | 7.0 | V |
| Power dissipation | Pd | $820^{* 1}$ | mW |
| Input voltage | VIN | -0.3 to (VCC +0.3$)$ | V |
| Storage temperature | Tstg | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |

1 When mounted on a $70 \mathrm{~mm} \times 70 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ glass epoxy substrate(1layer).
Reduced by 8.2 mW per $1^{\circ} \mathrm{C}$ when $\mathrm{Ta}=25^{\circ} \mathrm{C}$ or higher.

## -Operation Range

| Parameter | Symbol | Ratings | Unit |
| :--- | :--- | :--- | :---: |
| Supply voltage | VCC | 4.5 to 5.5 | V |
| Operating temperature | Topr | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |

- Electrical Characteristics (unless otherwise noted, $\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{VCC}=5 \mathrm{~V}$ )

| Parameter | Symbol | Limits |  |  | Unit | Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN | TYP | MAX |  |  |
| VCC current 1 | ICC1 | 32 | 45 | 58 | mA | No signal, 30 MHz LPF is selected during 4ch ACT |
| VCC current 2 *2 | ICC2 | - | 45 | - | mA | No signal, 30 MHz LPF is selected during CVBS_OUT MUTE |
| VCC current 3 * | ICC3 | - | 45 | - | mA | No signal, 30 MHz LPF is selected during component MUTE |
| VCC current 4 *2 | ICC4 | - | 45 | - | mA | No signal, 13.5 MHz LPF is selected during 4ch ACT |
| CVBS_OUT voltage gain | Gv1 | 5.5 | 6.0 | 6.5 | dB | Vin=1.0Vp-p, f=100kHz |
| PY/PB/PR_OUT voltage gain | GV2 | 5.5 | 6.0 | 6.5 | dB | $\mathrm{Vin}=0.7 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=100 \mathrm{kHz}$ |
| Maximum output level | Vomv | 2.6 | 2.9 | - | Vp-p | THD=1 \% f=10kHz |
| 6.75 MHz LPF frequency characteristics 1 | Gf1675 | -1.5 | -0.5 | 0.5 | dB | $\mathrm{Vin}=1.0 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=6.75 \mathrm{MHz} / 100 \mathrm{kHz}$ |
| 13.5 MHz LPF frequency characteristics 1 | Gf1135 | -1.5 | -0.5 | 0.5 | dB | Vin $=0.7 \mathrm{~V}$ p-p, $\mathrm{f}=13.5 \mathrm{M} / \mathrm{Hz} 100 \mathrm{kHz}$ |
| 30 MHz LPF frequency characteristics 1 | Gf1300 | -3.0 | -1.0 | 1.0 | dB | Vin $=0.7 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=30 \mathrm{MHz} / 100 \mathrm{kHz}$ |
| 6.75 MHz LPF frequency characteristics 2 | Gf2675 | - | -48 | -30 | dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=27 \mathrm{MHz} / 100 \mathrm{kHz}$ |
| 13.5 MHz LPF frequency characteristics 2 | Gf2135 | - | -48 | -30 | dB | Vin $=0.7 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=54 \mathrm{MHz} / 100 \mathrm{kHz}$ |
| 30 MHz LPF frequency characteristics 2 | Gf2300 | - | -35 | -20 | dB | Vin $=0.7 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=74.25 \mathrm{MHz} / 100 \mathrm{kHz}$ |
| MUTE attenuation | MT | - | -65 | -55 | dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}, \mathrm{f}=4.43 \mathrm{MHz}$ |
| Cross talk between channels | CT | - | -65 | -55 | dB | Vin=1.0Vp-p, f=4.43MHz |
| PB/PR_IN input impedance | Zin | 100 | 150 | 200 | $\mathrm{k} \Omega$ |  |
| Control pin input voltage $=\mathrm{H}$ | VthH | 2.0 | - | VCC | V |  |
| Control pin input voltage $=\mathrm{L}$ | VthL | 0.0 | - | 0.8 | V |  |
| Control pin input impedance | Rin | 100 | 150 | 200 | k $\Omega$ |  |
| Differential gain *2 | Dg | - | 0.5 | - | \% | $\mathrm{Vin}=1.0 \mathrm{Vp}-\mathrm{p}$, <br> Standard stair-step signal input |
| Differential phase *2 | Dp | - | 0.5 | - | deg | Vin=1.0 Vp-p, <br> Standard stair-step signal input |
| $\mathrm{S} / \mathrm{N}^{* 2}$ | SN | - | 75 | - | dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, band: 100 kHz to 6 MHz 100\% white video signal input |

*2 Indicates items with design certification (shipment inspections are not performed for these items.)

- Reference Data (1/10)


Fig.1 6.75MHzLPF Vcc - Freq. Characteristics


Fig. 3 6.75MHz LPF Temp. - Freq. Characteristics


Fig.5 6.75MHzLPF Vcc - Group Delay


Fig.2 6.75MHzLPF Vcc - Freq. Characteristics Magnification pass-band


Fig. 4 6.75MHz LPF Temp. - Freq. Characteristics Magnification pass-band


Fig. 6 6.75MHzLPF Temp. - Group Delay

## - Reference Data (2/10)



Fig.7 13.5MHzLPF Vcc - Freq. Characteristics


Fig. 9 13.5MHzLPF Temp. - Freq. Characteristics


Fig. 11 13.5MHzLPF Vcc - Group Delay


Fig. 8 13.5MHzLPF Vcc - Freq. Characteristics Magnification pass-band


Fig. 10 13.5MHzLPF Temp. - Freq. Characteristics Magnification pass-band


Fig. 12 13.5MHzLPF Temp. - Group Delay

## - Reference Data (3/10)



Fig. 13 30MHzLPF Vcc - Freq. Characteristics


Fig. 15 30MHzLPF Temp. Freq. Characteristics


Fig. 17 30MHzLPF Vcc - Group Delay


Fig. 14 30MHzLPF Vcc - Freq. Characteristics Magnification pass-band


Fig. 16 30MHzLPF Temp. - Freq. Characteristics Magnification pass-band


Fig. 18 30MHzLPF Temp. - Group Delay

## - Reference Data (4/10)



Fig. 19 6.75MHzLPF Vcc - MUTE attenuation


Fig. 21 13.5MHzLPF Vcc - MUTE attenuation


Fig. 23 30MHzLPF vcc - MUTE attenuation


Fig. 20 6.75MHzLPF Temp. - MUTE attenuation


Fig. 22 13.5MHzLPF Temp. - MUTE attenuation


Fig. 24 30MHzLPF Temp. - MUTE attenuation

## -Reference Data (5/10)



Fig. 25 Crosstalk CVBS_IN


Fig. 27 Crosstalk PB_IN


Fig. 26 Crosstalk PY_IN


Fig. 28 Crosstalk PR_IN


Fig. 29 ICC30MHz LPF


Fig. $32 \operatorname{Vomv}(\mathrm{PY}-13.5 \mathrm{MHz}$ )


Fig. 30 Vomv(CVBS)


Fig. 33 Vomv (PY-13.5MHz)


Fig.31Vomv(CVBS)


Fig. 34 Vomv (PY-30MHz)

## - Reference Data (6/10)



Fig. 35 Vomv (PY-30MHz)


Fig. 38 Vomv (PB-30MHz)


Fig. 41 Vomv (PR-13.5MHz)


Fig. 44 PB/PR_IN input impedance


Fig. 36 Vomv (PB-13.5MHz)


Fig. 39 Vomv (PB-30MHz)


Fig. 42 Vomv (PR-30MHz)


Fig. 45 PB/PR_IN input impedance


Fig. 37 Vomv (PB-13.5MHz)


Fig. 40 Vomv (PR-13.5MHz)


Fig. 43 Vomv (PR-30MHz)


Fig. 46 Control pin input voltage $=\mathrm{H}$

## -Reference Data (7/10)



Fig. 47 Control pin input voltage $=\mathrm{H}$


Fig. 50 Control pin input impedance


Fig. 53 DG(CVBS)


Fig. $56 \mathrm{DG}(\mathrm{PY}-30 \mathrm{MHz})$


Fig. 48 Control pin input voltage $=$ L


Fig. 51 Control pin input impedance


Fig. 54 DG(PY-13.5MHz)


Fig. 57 DG(PY-30MHz)


Fig. 49 Control pin input voltage= $L$


Fig.52DG(CVBS)


Fig. 55 DG(PY-13.5MHz)


Fig. 58 DG(PB-13.5MHz)

## -Reference Data (8/10)



Fig. 59 DG(PB-13.5MHz)


Fig. 62 DG(PR-13.5MHz)


Fig. 65 DG(PR-30MHz)


Fig. $68 \mathrm{DP}(\mathrm{PY}-13.5 \mathrm{MHz})$


Fig. 60 DG(PB-30MHz)


Fig. 63 DG(PR-13.5MHz)


Fig. 66 DP(CVBS)


Fig. 69 DP(PY-13.5MHz)


Fig. 61 DG(PB-30MHz)


Fig. 64 DG(PR-30MHz)


Fig. 67 DP(CVBS)


Fig. 70 DP(PY-30MHz)

## - Reference Data (9/10)



Fig. 71 DP(PY-30MHz)


Fig. 74 DP(PB-30MHz)


Fig. 77 DP(PR-13.5MHz)


Fig. 80 S/N(CVBS)


Fig. $72 \mathrm{DP}(\mathrm{PB}-13.5 \mathrm{MHz})$


Fig. 75 DP(PB-30MHz)


Fig. 78 DP(PR-30MHz)


Fig. 81 S/N(CVBS)


Fig. 73 DP(PB-13.5MHz)


Fig. 76 DP(PR-13.5MHz)


Fig. $79 \mathrm{DP}(\mathrm{PR}-30 \mathrm{MHz})$


Fig. $82 \mathrm{~S} / \mathrm{N}(\mathrm{PY}-13.5 \mathrm{MHz})$

## - Reference Data (10/10)



Fig. $83 \mathrm{~S} / \mathrm{N}$ ( $\mathrm{PY}-13.5 \mathrm{MHz}$ )


Fig. 86 S/N (PB-13.5MHz)


Fig. 89 S/N (PB-30MHz)


Fig. 92 S/N (PR-30MHz)


Fig. $84 \mathrm{~S} / \mathrm{N}(\mathrm{PY}-30 \mathrm{MHz})$
 Fig. 87 S/N (PB-13.5MHz)


Fig. $90 \mathrm{~S} / \mathrm{N}$ (PR-13.5MHz)



Fig. 85 S/N (PY-30MHz)


Fig. 88 S/N (PB-30MHz)


Fig. 91 S/N (PR-13.5MHz)

## -Block Diagram



Fig. 94 Block Diagram

- Control Specifications

| Pin | Function |
| :--- | :--- |
| 6pin (MUTE1) | Mute control of composite output (CVBS_OUT) *3 <br> L: MUTE <br> H: Normal operation |
| 7pin (MUTE2) | Mute control of component outputs (PY/PB/PR_OUT) *3 <br> L: MUTE <br> H: Normal operation |
| 8pin (LPF_SEL) | LPF selector for component <br> L: 13.5 MHz LPF <br> H:30 MHz LPF |

*3 When operating in mute mode, each output pin normally outputs the bias voltage when there is no signal.

## -Pin Number / Pin Name

| No. | Pin Name | I/O | DESCRIPTION |
| :---: | :---: | :---: | :---: |
| 1 | CVBS_IN | 1 | Video signal input pin <br> Sync tip clamp input LPF=6.75 MHz |
| 2 | VREF | I | Bias capacitor connection pin |
| 3 | PY_IN | 1 | Video signal input pin <br> Sync tip clamp input $\quad L P F=13.5 \mathrm{MHz} / 30 \mathrm{MHz}$ |
| 4 | PB_IN | I | Video signal input pin <br> Bias input LPF=13.5 MHz/30 MHz |
| 5 | PR_IN | 1 | Video signal input pin <br> Bias input LPF=13.5 MHz/30 MHz |
| 6 | MUTE1 | 1 | Mute control of CVBS_OUT L: MUTE H:Normal operation |
| 7 | MUTE2 | I | Mute control of PY/PB/PR_OUT L: MUTE H:Normal operation |
| 8 | LPF_SEL | 1 | LPF selector for component output <br> L: 13.5 MHz LPF H:30 MHz LPF |
| 9 | PR_OUT | 0 | Video signal output pin Enable load driving up to $75 \Omega$ (2 drives) |
| 10 | PB_OUT | 0 | Video signal output pin <br> Enable load driving up to $75 \Omega$ (2 drives) |
| 11 | GND | 1 | GND pin |
| 12 | PY_OUT | 0 | Video signal output pin <br> Enable load driving up to $75 \Omega$ (2 drives) |
| 13 | VCC | 1 | Power supply pin |
| 14 | CVBS_OUT | 0 | Video signal output pin Enable load driving up to $75 \Omega$ (2 drives) |

## -Selection of Application Parts

Method for determining capacity of input coupling capacitor

| Input pin | Input impedance <br> Zin | Capacity of input coupling <br> capacitor (recommended value) | Capacity of output coupling <br> capacitor (recommended value) |
| :--- | :--- | :--- | :--- |
| CVBS/PY_IN | Approximately $10 \mathrm{M} \Omega$ | $0.1 \mu \mathrm{~F}$ | $470 \mu \mathrm{~F}$ to $1000 \mu \mathrm{~F}$ |
| PB/PR_IN | $150 \mathrm{k} \Omega$ | $1.0 \mu \mathrm{~F}$ |  |

The HPF includes an input coupling capacitor and an Internal input impedance Zin of the IC. Since the fc value of the HPF is calculated using the following equation (a), above recommended value of capacity for the input capacitor is derived. Usually, the cutoff frequency fc is set several Hz .

$$
\mathrm{fc}=1 /(2 \pi \times \mathrm{C} \times \mathrm{Zin}) \cdots \cdot(\mathrm{a})
$$

A horizontal stripe signal called an "H bar signal" (shown in Fig. 95) is suitable when evaluating sag characteristics and determining the capacity of the capacitor during video signal input, and this type of signal is used instead of a color bar signal to evaluate characteristics and determine capacity.


Fig. 95 Example of Screen with Obvious Sag (H-bar Signal)

## Method for determining capacity of output coupling capacitor

The output pins have an HPF that includes an output coupling capacitor and load resistance $R_{L}(=150 \Omega)$. When fc is set within the range of 1 Hz to 2 Hz , the capacity of the output coupling capacitor must be within the range of $470 \mu \mathrm{~F}$ to $1000 \mu \mathrm{~F}$.
With this model, up to two monitors (loads) can be connected (see the connection example in Fig. 96). When there are multiple loads, the number of output coupling capacitors must be increased, or a larger capacitance must be set, According to the table shown below.


Fig.96(a) Application Circuit 1 (2 Drives)


Fig.96(b) Application Circuit 2 (2 Drives)

| Application circuit | Number of output <br> capacitors | Capacitance per output capacitor (recommended values) |
| :--- | :---: | :---: |
| Fig.96(a) | 2 | $470 \mu \mathrm{~F}$ to $1000 \mu \mathrm{~F}($ Same as with 1 drive) |
| Fig.96(b) | 1 | $(2 \times 470$ to 1000) $\mu \mathrm{F}$ |

- Evaluation Board Layout (2 layers)


Fig. 97 Top Surface Silk


Fig. 99 Top Surface Resist


Fig. 101 Top Surface Pattern


Fig. 98 Bottom Surface Silk


Fig. 100 Bottom Surface Resist


Fig. 102 Bottom Surface Pattern

## - Evaluation Board Circuit Diagram



Fig. 103 Evaluation Board Circuit Diagram

## -Description of External Parts

| Symbol | Function | Recommended <br> value | Comments |
| :--- | :---: | :---: | :--- |
| C_1, C_3 | Input coupling capacitor | $0.1 \mu \mathrm{~F}$ | Refer to *1 |
| C_2 | Bias capacitor | $10 \mu \mathrm{~F}$ | Refer to *1 |
| C_4, C_5 | Input coupling capacitor | $1 \mu \mathrm{~F}$ | Refer to *1 |
| C_9, C_10, C_12, C_14 | Output coupling capacitor | $470 \mu \mathrm{~F}$ | Recommended <br> Electrolytic Capacitor |
| C_13A | Decoupling capacitor | $22 \mu \mathrm{~F}$ | Refer to *1 |
| C_13B | Decoupling capacitor | $0.1 \mu \mathrm{~F}$ | Refer to *1 |
| R1, R3, R4, R5 | Input terminating resistor | $75 \Omega$ |  |
| R9_A, R10_A, R12_A, R13_A <br> R9_B, R10_B, R12_B, R13_B | Output terminating resistor | $75 \Omega$ |  |
| R9_C, R10_C, R12_C, R13_C | Output terminating resistor | $150 \Omega$ | For comparing number <br> of drives (for 2 drives) |
| BNC1, BNC3, BNC4, BNC5 | BNC connector for video signal input |  |  |
| BNC9, BNC10, BNC12, BNC14 | BNC connector for video signal output |  |  |
| D-CONNECTOR | Type D-connector for video signal output |  |  |

*1 : Recommended Capacitance Tolerance code B type Ceramic Capacitor

## - Test Circuit Diagram


※ Test circuits are used for shipment inspections, and differ from application circuits.
Fig. 104 Test Circuit Diagram

## - Application Circuit

(1) Use output coupling capacitor (Enable load driving up to $75 \Omega$ (2 drives))


Fig. 105 Application Circuit 1
See page $15 / 22$ for description of how to determine the capacitance of an input/output coupling capacitor. Use when the terminating impedance of the clamp input pins (1 and 3pin) is $1 \mathrm{k} \Omega$ or less.
(2) Without output coupling capacitors (Enable load driving up to $150 \Omega$ (1 drive only))


Fig. 106 Application Circuit 2
See page $15 / 22$ for description of how to determine the capacitance of an input/output coupling capacitor.
Use when the terminating impedance of the clamp input pins ( 1 and 3 pin) is $1 \mathrm{k} \Omega$ or less.
The frequency characteristic of the low area can realize the improvement of the SAG characteristic, the substrate space's and part cost's being able to be reduced without output coupling capacitors.
However, because the DC current flows to the set connected this IC without output coupling Capacitors, be careful of the specification of the connection set and so on sufficiently.
Use a substrate in equal to or more than 4 layers when mounts of the IC without output coupling capacitors from the viewpoint of the permission loss.

## - Application Circuit (Cont.)

Reference data

|  |  | With output capacitors | Without output capacitors | Conditions |
| :---: | :---: | :---: | :---: | :---: |
| ICC |  | 45 mA | 70 mA |  |
| CVBS | DG | 0.24\% | 0.26\% | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, <br> Standard stair-step signal input |
|  | DP | 0.36 deg | 0.31deg | Vin=1.0 Vp-p, Standard stair-step signal input |
|  | S/N | -74.5dB | -74.7dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, band: 100 kHz to 6 MHz $100 \%$ white video signal input |
| PY | DG | 0.23\% | 0.26\% | $\mathrm{Vin}=1.0 \mathrm{Vp}-\mathrm{p}$, <br> Standard stair-step signal input |
|  | DP | 0.33deg | 0.26 deg | Vin=1.0 Vp-p, <br> Standard stair-step signal input |
|  | S/N | -75.2dB | -75.2dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, band: 100 kHz to 6 MHz $100 \%$ white video signal input |
| PB | DG | 0.56\% | 0.81\% | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, <br> Standard stair-step signal input |
|  | DP | 0.54 deg | 0.59deg | Vin=1.0 Vp-p, <br> Standard stair-step signal input |
|  | S/N | -75.0dB | -75.1dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, band: 100 kHz to 6 MHz $100 \%$ white video signal input |
| PR | DG | 0.65\% | 0.62\% | $\mathrm{Vin}=1.0 \mathrm{Vp}-\mathrm{p}$, <br> Standard stair-step signal input |
|  | DP | 0.60deg | 0.50deg | Vin=1.0 Vp-p, <br> Standard stair-step signal input |
|  | S/N | -74.5dB | -75.3dB | Vin $=1.0 \mathrm{Vp}-\mathrm{p}$, band: 100 kHz to 6 MHz $100 \%$ white video signal input |

When mounted on a $70 \mathrm{~mm} \times 70 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ glass epoxy substrate (4layer). $\mathrm{Pd}=1.45 \mathrm{~W}$
Reduced by 14.5 mW per $1^{\circ} \mathrm{C}$ when $\mathrm{Ta}=25^{\circ} \mathrm{C}$ or higher.

* When mounted on a $70 \mathrm{~mm} \times 70 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ glass epoxy substrate (1layer). $\mathrm{Pd}=0.82 \mathrm{~W}$

Reduced by 8.2 mW per $1^{\circ} \mathrm{C}$ when $\mathrm{Ta}=25^{\circ} \mathrm{C}$ or higher.

## - I/O Equivalent Circuit Diagrams (page 1 of 2)

\begin{tabular}{|c|c|c|c|c|}
\hline \[
\begin{aligned}
\& \text { Pin } \\
\& \text { No. }
\end{aligned}
\] \& Pin name \& Standard potential \& I/O equivalent circuit diagram \& Description of pins \\
\hline 1
3 \& CVBS_IN
PY_IN \& 1.5 V \&  \& \begin{tabular}{l}
Video signal input pin \\
Sync tip clamp input
\end{tabular} \\
\hline 2 \& VREF \& 2.2V \&  \& Bias capacitor connection pin \\
\hline 4

5 \& PB_IN
PR_IN \& 2.9 V \&  \& Video signal input pin Bias input <br>

\hline 6 \& MUTE1 \& OV \&  \& | Mute control of CVBS_OUT |
| :--- |
| L : MUTE |
| H: Normal operation | <br>


\hline 7 \& MUTE2 \& OV \&  \& | Mute control of PY/PB/PR_OUT |
| :--- |
| L: MUTE |
| H: Normal operation | <br>


\hline 8 \& LPF_SEL \& OV \&  \& | Composite output LPF select |
| :--- |
| L: 13.5 MHz LPF |
| H: 30 MHz LPF | <br>

\hline
\end{tabular}

## - I/O Equivalent Circuit Diagrams (page 2 of 2)

| Pin No. | Pin name | Standard potential | I/O equivalent circuit diagram | Description of pins |
| :---: | :---: | :---: | :---: | :---: |
| 9 | PR_OUT |  |  |  |
| 10 | PB_OUT |  | - |  |
| 12 | PY_OUT |  | $\Delta$ | le load driving up to $75 \Omega$ (2 drives) |
| 14 | CVBS_OUT |  |  |  |
| 11 | GND | OV |  | GND pin |
| 13 | VCC | 5 V |  | Power supply pin |

Note 1) The above DC potential is only for when VCC $=5 \mathrm{~V}$. This is a reference value, and it is not guaranteed.
Note 2) Numerical values in the figures are design values, and standards compliance is not guaranteed.

## - Notes for use

1. Numerical values and data that are cited are representative design values and their values are not guaranteed.
2. We are confident in recommending the above application circuit example, but we ask that you carefully check the characteristics before using it. When using it with different external part constants, differences between the external part and this IC, including not just the static characteristics but also transient characteristics, should be carefully checked to determine adequate margins.
3. Absolute maximum ratings

If absolute maximum ratings such as applied voltage and operating temperature range are exceeded, the IC may be damaged. Do not apply voltages or temperatures that exceed the absolute maximum ratings. If you are considering circumstances in which an absolute maximum rating may be exceeded, try using a physical safety measure such as a fuse so that conditions that exceed the absolute maximum ratings are not applied to the IC.
4. GND potential

Even if the voltage of the GND pin is left in an operating state, make it the minimum voltage. Actually confirm that the voltage of each pin does not become a lower voltage than the GND pin, including for transient phenomena.
5. Thermal design

Thermal design should be done using an ample margin that takes into consideration the allowable dissipation under actual use conditions.
6. Shorts between pins and mounting errors

When mounting the IC on a board, be careful of the direction of the IC and of misalignment. If the IC is mounted badly and current is passed through it, it may be damaged. The IC also may be damaged if shorted by a foreign substance getting in between IC pins, or between an IC pin and the power supply or GND.
7. Operation in a strong electromagnetic field

When used within a strong electromagnetic field, evaluate carefully to avoid the risk of operation faults.
8. Place the power supply's decoupling capacitor as close as possible to the VCC pin (PIN 13).
9. Be careful to avoid inserting the IC upside down. When upside down, the electrostatic damage preventing diode may be set to operation mode, depending on the input condition of the CONTROL pin, and the IC may become damaged.
10. If any input pins that use the clamp input method are left OPEN they will oscillate, so unused input pins should instead be connected to GND via a capacitor or else directly connected to VCC.
11. Problems such as sync contraction or oscillation may occur if the terminating impedance of the clamp input pins (pins 1 and 3 ) is high. Fully evaluate the temperature characteristics as well, to keep this value to $1 \mathrm{k} \Omega$ or less.

## - Ordering part number



SOP-J14



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