

Inverter for motor control

600V IGBT Intelligent Power Module (IPM) for high speed switching drive

BM63767S-VA BM63767S-VC

General Description

BM63767S-VA /-VC is an Intelligent Power Module composed of gate drivers, bootstrap diodes, IGBTs, fly wheel diodes.

Features

- 3phase DC/AC Inverter
- 600V/30A
- Low Side IGBT Open Emitter
- Built -in Bootstrap Diode
- High Side IGBT Gate Driver(HVIC): SOI (Silicon On Insulator) Process, Drive Circuit, High Voltage Level Shifting, Current Limit for Bootstrap Diode, Control Supply Under-Voltage Locked Out (UVLO)
- Low Side IGBT Gate Driver(LVIC): Drive Circuit, Short Circuit Current Protection (SCP), Control Supply Under Voltage Locked Out (UVLO), Thermal Shutdown (TSD)
- Fault Signal(LVIC) Corresponding to SCP (Low Side IGBT), TSD, UVLO Fault
- Input Interface 3.3V, 5V Line
- UL Recognized: File E468261

Application

- High Speed Switching Drive of AC100 to 240Vrms(DC Voltage: Less Than 400V) Class Motor
- High Speed Switching Drive of Motor for Washing Machine, Fan

Typical Application Circuit

Key Specifications

- IGBT Collector-Emitter Voltage V_{CESAT}: 1.7V(Typ)
 - FWD Forward Voltage V_F: 1.5V(Typ)
- FWD Reverse Recovery Time trr: 100ns(Typ)
- Module Case Temperature T_C: -25 to +100°C 150°C
- Junction Temperature T_{jmax}:

Package

HSDIP25 HSDIP25VC

W(Typ) x D(Typ) x H(Typ) 38.0mm x 24.0mm x 3.5mm 38.0mm x 24.0mm x 3.5mm



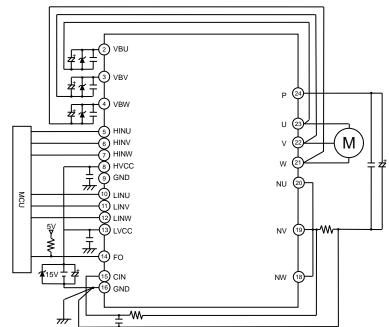
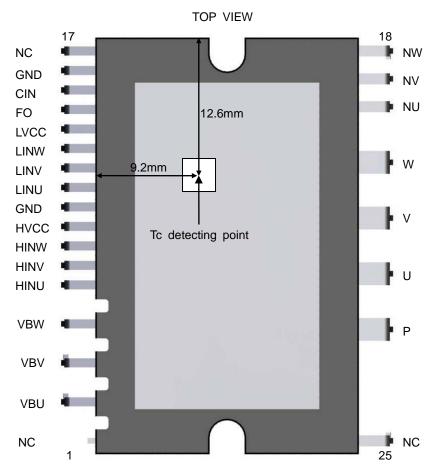
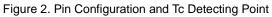


Figure 1. Example of Application Circuit

OProduct structure: Semiconductor IC OThis product is not designed for protection against radioactive rays

Pin Configuration





Pin Description

Pin No.	Pin Name	Function	Pin No.	Pin Name	Function
1	NC	No connection(GND potential)	14	FO	Alarm output
2	VBU	U phase floating control supply	15	CIN	Detecting of short circuit current trip voltage
3	VBV	V phase floating control supply	16	GND	Ground (Note 1)
4	VBW	W phase floating control supply	17	NC	No connection (Note 2)
5	HINU	U phase high side IGBT control	18	NW	W phase low side IGBT emitter
6	HINV	V phase high side IGBT control	19	NV	V phase low side IGBT emitter
7	HINW	W phase high side IGBT control	20	NU	U phase low side IGBT emitter
8	HVCC	Control supply for HVIC	21	W	W phase output
9	GND	Ground (Note 1)	22	V	V phase output
10	LINU	U phase low side IGBT control	23	U	U phase output
11	LINV	V phase low side IGBT control	24	Р	Inverter supply
12	LINW	W phase low side IGBT control	25	NC	No connection (Note 2)
13	LVCC	Control supply for LVIC			

(Note 1) Two GND pins (9 & 16pin) are connected inside IPM, please connect one pin (16pin is recommended) to the 15V power supply GND outside and leave the other open.

(Note 2) NC pins (17 & 25pin) are not electrically connected to any other potential inside.

Block Diagram

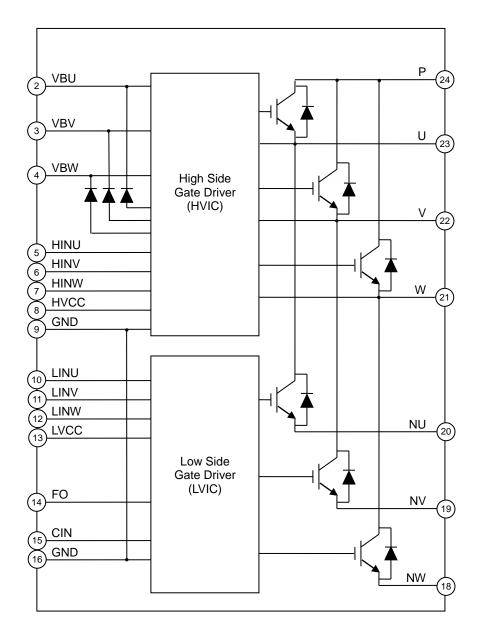


Figure 3. Block Diagram

Description of Block

- 1) High Side IGBT Drive (HVIC, Bootstrap Diode)
 - High voltage level shifting circuit drives high side IGBT.

Built-in bootstrap diode and current limit function for bootstrap diode enable HVIC to drive high side IGBT without external component (bootstrap diode, resistor). There is under-voltage-locked-out (UVLO) function for floating control power supply.

2) Low Side IGBT Drive (LVIC)

LVIC drives low side IGBT.

There is short circuit current protection (SCP), under-voltage locked out (UVLO) for control power supply LVCC, thermal shutdown (TSD) function. Alarm signal (FO) will output when these protection circuits work.

Absolute Maximum Ratings (Unless otherwise specified, Tj=25°C)

Inverter Part

Item	Item		Ratings	Unit	Conditions
Supply Voltage		VP	450	V	Applied between P-NU,NV,NW
Supply Voltage(Su	ırge)	VP(surge)	surge) 500 V Applied between P-NU,NV,NW		Applied between P-NU,NV,NW
Collector-Emitter Voltage		VCES	600	V	
Collector Current	DC	lc	±30 (Note 1)	А	Tc=25°C
Collector Current	PEAK	ICP	±60 (Note 1)	А	Tc=25°C, less than 1ms
Collector Power Di	Collector Power Dissipation		59	W	T _c =25°C, per 1 chip
Junction Temperat	ure	Tjmax	150	°C	
				Ŭ	

(Note 1) Do not, however exceed P_C , ASO.

Control part

Item	Symbol	Ratings	Unit	Conditions
Control Power Supply	Vcc	20	V	Applied between HVCC-GND, LVCC-GND
Floating Control Power Supply	V _{BS}	20	V	Applied between VBU-U, VBV-V, VBW-W
Control Input Voltage	V _{IN}	-0.5 to V _{CC} +0.5	V	Applied between HINX, LINX-GND (X=U,V,W)
Fault Output Supply Voltage	VFO	-0.5 to V _{cc} +0.5	V	Applied between FO-GND
Fault Output Current	I _{FO}	1	mA	Sink current at FO pin
Current Sensing Input Voltage	Vcin	-0.5 to +7.0	V	Applied between CIN-GND

Bootstrap diode part

Item	Symbol	Ratings	Unit	Conditions
Reverse Voltage	V _{RB}	600	V	
Junction Temperature	T _{jmaxD}	150	°C	

Total system

Item	Symbol	Ratings	Unit	Conditions		
Self Protection Supply Voltage (SCP Capability)	VP(PROT)	400	V	V _{cc} =13.5 to 16.5V, Inverter part $T_j=125^{\circ}C$, non-repetitive, less than 2µs		
Module Case Temperature	Tc	-25 to +100	°C	Measurement point of T_C is provided in Figure 2		
Storage Temperature	T _{stg}	-40 to +125	°C			
Isolation Voltage	V _{iso}	1500	Vrms	Sinusoidal, 60Hz, AC 1minute, between connected all pins and heat sink plate		

Thermal resistance

ltem	Svmbol		Limit			Conditions
item	Symbol	Min	Тур	Max	Unit	Conditions
Junction to Case Thermal	Rth(j-c)_IGBT	-	-	2.1	°C /W	Inverter IGBT(1/6 module)
Resistance (Note 2)	R _{th(j-c)_FWD}	-	-	3.0	°C /W	Inverter FWD(1/6 module)

(Note 2) Grease with good conductivity and high reliability should be applied evenly with +100 to +200µm on the contacting surface of IPM and heat sink. Use a torque wrench to fasten up to the specified torque rating. The contacting thermal resistance between IPM case and heat sink is determined by the thickness and the thermal conductivity of the applied grease.

Caution: Operating the IPM over the absolute maximum ratings may damage the IPM. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IPM is operated over the absolute maximum ratings.

Recommended Operating Conditions

Item	Symbol		Limit		Unit	Conditions
lieni	Symbol	Min	Тур	Max	Unit	Conditions
Supply Voltage	VP	0	300	400	V	Applied between P-NU,NV,NW
Control Power Supply	Vcc	13.5	15.0	16.5	V	Applied between HVCC-GND, LVCC-GND
Floating Control Power Supply	V_{BS}	13.0	15.0	18.5	V	Applied between VBU-U, VBV-V, VBW-W
Control Power Supply Variation	⊿Vcc ⊿Vвs	-1	-	+1	V/µs	
Control Input Voltage	Vin	0	-	5.5	V	
Current Sensing Input Voltage	V _{CIN}	0	-	5.5	V	
Blanking Time for Preventing Arm-short	t _{dead}	1.5	-	-	μs	For each input signal
PWM Input Frequency	fрwм	-	-	20	kHz	T _C ≤ 100°C, T _j ≤ 125°C
High Side IGBT	PWONH	0.8	-	-	μs	
Minimum Input Pulse Width ^(Note1)	PWOFFH	0.8	-	-	μs	
Low Side IGBT	PWONL	2.5	-	-	μs	
Minimum Input Pulse Width ^(Note1)	PWOFFL	0.8	-	-	μs	
Voltage Variation Between GND- NU, NV, NW	VN	-5	-	+5	V	Between GND-NU, NV, NW (Including surge voltage)
Junction Temperature	Tj	-25	-	+125	°C	

(Note 1) IPM might not respond if the input signal pulse width is less than PW_{ON} , PW_{OFF} .

Electrical Characteristics (Unless otherwise specified, Tj=25°C, V_{CC}=V_{BS}=15V, V_P=300V) In

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										-

ltom	Symbol		Limit		Unit	Conditions	
Item	Symbol	Min	Тур	Max	Unit	Conditions	
Collector Emitter Seturation Voltage	Manada	-	1.70	2.10	V	Ic=30A	
Collector-Emitter Saturation Voltage	VCESAT	-	0.90	1.20	V	Ic=3A	
Collector-Emitter Cut-off Current	ICES	-	-	100	μA	V _{CE} =V _{CES}	
FWD Forward Voltage	VF	-	1.50	2.00	V	IF=30A	
FWD Reverse Recovery Time	t _{rr}	-	100	-	ns	Inductive Load, Ic=30A	
High Side IGBT Turn on Delay Time	t _{onH}	0.50	0.95	1.50	μs	Inductive Load, Ic=30A	
High Side IGBT Turn on Switching Time	t _{c(on)H}	-	0.20	-	μs	Inductive Load, Ic=30A	
High Side IGBT Turn off Delay Time	t _{offH}	-	1.40	2.40	μs	Inductive Load, Ic=30A	
High Side IGBT Turn off Switching Time	t _{c(off)} H	-	0.25	-	μs	Inductive Load, Ic=30A	
Low Side IGBT Turn on Delay Time	t _{onL}	0.90	1.70	2.70	μs	Inductive Load, Ic=30A	
Low Side IGBT Turn on Switching Time	t _{c(on)L}	-	0.25	-	μs	Inductive Load, Ic=30A	
Low Side IGBT Turn off Delay Time	t _{offL}	-	0.90	1.50	μs	Inductive Load, I _C =30A	
Low Side IGBT Turn off Switching Time	t _{c(off)L}	-	0.20	-	μs	Inductive Load, Ic=30A	

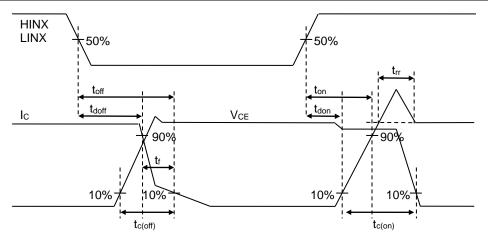


Figure 4. Switching Time Definition

Control part

ltom	Symbol Limit					Conditions
Item	Symbol	Min	Тур	Max	Unit	Conditions
Whole						
VCC Circuit Current 1	Icc1	-	0.70	1.60	mA	VIN=0V
VCC Circuit Current 2	I _{CC2}	-	0.90	2.00	mA	VIN=5V
VBS Circuit Current 1	I _{BS1}	-	0.15	0.35	mA	VIN=0V
VBS Circuit Current 2	I _{BS2}	-	0.15	0.35	mA	VIN=5V
Control Input(HINU,HINV,HINW,	LINU,LINV	,LINW)				
H Level Input Current	linh	0.7	1.0	1.5	mA	V _{IN} =5V
L Level Input Current	IINL	-10	-	-	μΑ	V _{IN} =0V
H Level Input Threshold Voltage	VINH	-	-	2.6	V	
L Level Input Threshold Voltage	V _{INL}	0.8	-	-	V	
Input Hysteresis Voltage	V _{HYS}	-	0.25	-	V	
Short Circuit Current Protection	n					
CIN Input Bias Current	ICIN	-2	-	-	μA	CIN=0V
Trip Voltage	Vsc	0.43	0.48	0.53	V	
Under Voltage Locked Out						
VCC Trip Voltage	Vccuvt	10.5	11.5	12.5	V	
VCC Release Voltage	V _{CCUVR}	11	12	13	V	
VBS Trip Voltage	VBSUVT	10	11	12	V	
VBS Release Voltage	VBSUVR	10.5	11.5	12.5	V	
Thermal Shutdown						
Trip Temperature	TSDT	100	130	-	°C	Monitor LVIC temperature
Hysteresis Temperature	T _{SDHYS}	-	20	-	°C	Monitor LVIC temperature
Fault Output(FO)			·			
Output low Voltage	V_{FO}	-	-	0.95	V	I _{FO} =1mA
Leak Current	IFOLEAK	-	-	10	μA	V _{FO} =5V
Output Pulse Width	t _{FO}	20	-	-	μs	

Bootstrap Diode Part

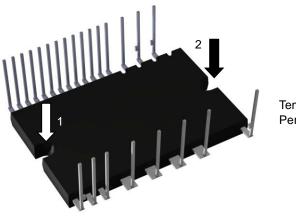
ltem	Symbol		Limit		Unit	Conditions
ltern	Symbol	Min	Тур	Max	Unit	
Forward Voltage	Vfb1	0.3	0.6	0.9	V	I _{FB} =1mA Voltage drop between HVCC-VBX (X=U,V,W)
Forward Voltage	Vfb2	1.1	2.0	2.9	V	I _{FB} =100mA Voltage drop between HVCC-VBX (X=U,V,W)
Reverse Current	I _{RB}	-	-	10	μA	V _{RB} =600V
Reverse Recovery Time	t _{rrB}	-	80	-	ns	I _{FB} =0.1A

Mechanical Characteristics And Ratings

ltom		Limit		Unit	Following Standard	Conditions	
Item	Min	Тур	Max	Unit	Following Standard		
Mounting Torque	0.59	0.69	0.78	N∙m	-	Mounting Screw M3 ^(Note 1) Recommended 0.69N • m ^(Note 2)	
Pin Pulling Strength	10	-	-	S	EIAJ-ED-4701/400	Load Control Pin:4.9N Power Pin:9.8N	
Pin Bending Strength	2	-	-	times	EIAJ-ED-4701/400	Load Control Pin: 2.45N Power Pin:4.9N 90deg. Bend	
Weight	-	10	-	g	-		
Heat Sink Flatness	0	-	+200	μm	-	Measurement point is provided in Figure 6-1.	

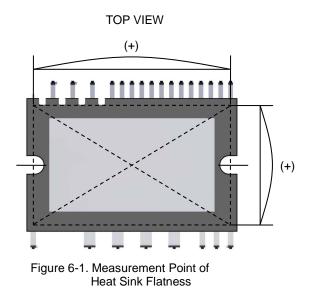
(Note 1) Plain washers of 8mm outside diameter (ISO 7089 to 7094) are recommended.

(Note 2) When installing a module to a heat sink, excessive uneven fastening force might apply stress to inside chips or ceramic of heat sink plate, which will break or crack or degrade a module. An example of recommended fastening sequence is shown in Figure 5. The temporary fastening torque is set to 20 to 30% of the maximum torque rating. Evenly apply thermally-conductive grease with 100µm to 200µm thickness over the contact surface between the module and the heat sink. Also, pay attention not to have any dirt left on the contact surface between the module and the heat sink. It is recommended to install a module directly to a heat sink after applying grease. When installing a module to a heat sink, inserting a heat radiation sheet between a module and a heat sink might apply stress depending on thickness and elastic modulus of the sheet to inside chips or ceramic of heat sink plate, which will break or crack or degrade a module. When using a heat radiation sheet, it is needed to prevent IPM from bending into + side of Figure 6-2.



Temporary fastening : $1 \rightarrow 2$ Permanent fastening : $1 \rightarrow 2$





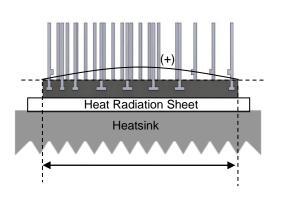
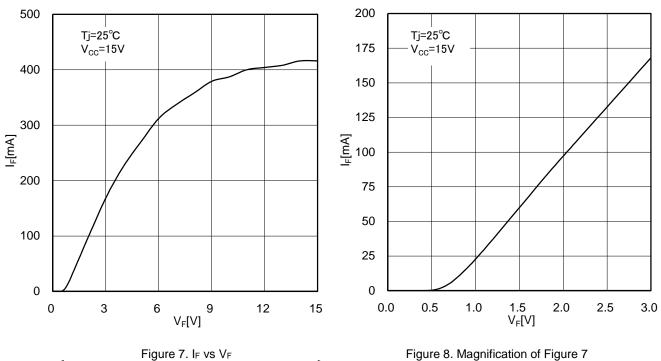
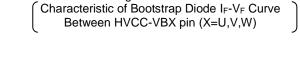


Figure 6-2. Flatness after Installing to a Heat Sink (When Using a Heat Radiation Sheet)

Typical Performance Curve







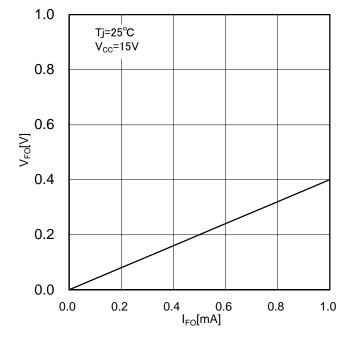


Figure 9. VFO vs IFO (Characteristic of FO pin VFO-IFO Curve)

Timing Chart

- 1) Short Circuit Current Protection (protection with the external shunt resistor and RC filter)
 - a1. Normal operation: IGBT ON and outputs current $I_{\text{C}.}$
 - a2. Short circuit current detection (SCP trigger)
 - It is recommended to set RC time constant of 1.0 to 2.0µs so that IGBT shuts down within 2.0µs when SCP is triggered. a3. All low side IGBT's gates are shut down (soft turn off).
 - a4. All low side IGBTs turn off.
 - a5. FO outputs for t_{FO}=20µs (Min).
 - a6. LIN=L
 - a7. LIN=H, but all IGBTs keep OFF during SCP=H.
 - a8. FO finishes output , but IGBTs don't turn on until inputting the next ON signal(LIN=L \rightarrow H) IGBT of each phase can return to normal state by inputting ON signal to each phase.
 - a9. Normal operation: IGBT ON and outputs current Ic.

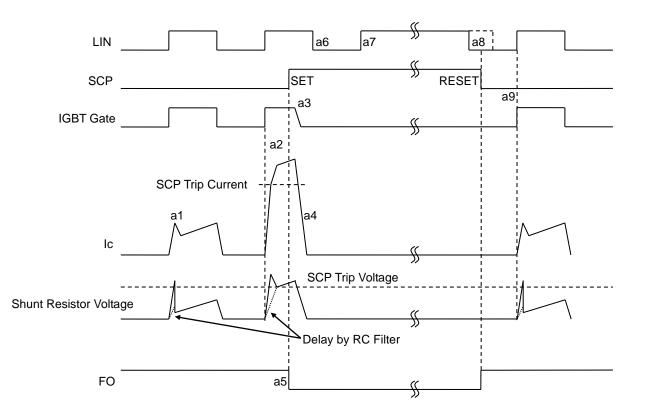


Figure 10. SCP Timing Chart

Notice

SCP works only for low side IGBT only.

In case of SCP trip and FO output, please stop controlling IPM quickly to avoid the abnormal state.

- 2) Control Supply (LVCC) Under Voltage Locked Out (UVLO)
 - b1. Control supply(LVCC) voltage exceeds UVLO release level (V_{CCUVR}), but IGBT turns on by the next ON signal (LIN=L→ H).IGBT of each phase can return to normal state by inputting ON signal to each phase.
 - b2. Normal operation: IGBT ON and outputs current Ic.
 - b3. LVCC drops to UVLO trip level (V_{CCUVT}).
 - b4. All low side IGBTs turn off in spite of control input condition.
 - b5. FO outputs for t_{FO}=minimum 20µs, but output is extended while LVCC is below V_{CCUVR}.
 - b6. LVCC reaches V_{CCUVR} .
 - b7. Even if LVCC reaches V_{CCUVR} during LIN=H, IGBTs don't turn on until inputting the next ON signal (LIN=L \rightarrow H).
 - b8. Normal operation: IGBT ON and outputs current Ic.

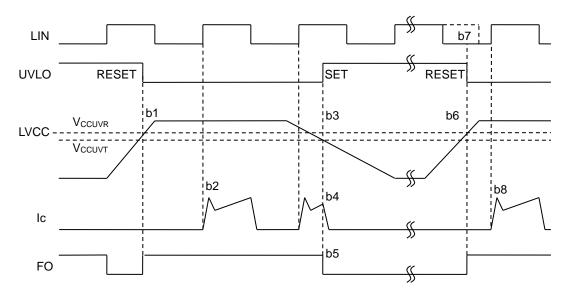


Figure 11. LVCC UVLO Timing Chart

- 3) Control supply (VBS) Under Voltage Locked Out (UVLO)
 - c1. Control supply(VBS) voltage exceeds UVLO release level (V_{BSUVR}), but IGBT turns on by the next ON signal (HIN=L \rightarrow H).
 - c2. Normal operation: IGBT ON and outputs current Ic.
 - c3. VBS drops to UVLO trip level (VBSUVT).
 - c4. Only IGBT of the corresponding phase turns off in spite of control input signal, there is no FO signal output.
 - c5. VBS reaches V_{BSUVR} .
 - c6. Even if VBS reaches V_{BSUVR} during HIN=H, IGBTs don't turn on until inputting the next ON signal (HIN=L \rightarrow H).
 - c7. Normal operation: IGBT ON and outputs current Ic.

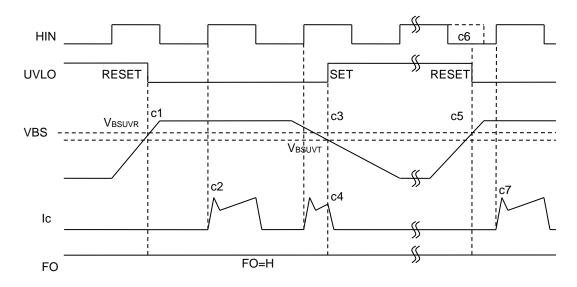


Figure 12. VBS UVLO Timing Chart

- 4) Thermal Shutdown (TSD), monitoring LVIC temperature
 - d1. Normal operation: IGBT ON and outputs current $I_{C.}$
 - d2. LVIC temperature (T_j) exceeds thermal shutdown trip level (T_{SDT}).
 - d3. All low side IGBTs turn off in spite of control input condition. d4. FO outputs for $t_{FO}=20\mu s$ (Min),
 - but output is extended while T_j is above thermal shutdown release level (T_{SDT} - T_{SDHYS}).
 - d5. T_i drops to T_{SDT}-T_{SDHYS}
 - d6. Even if T_j reaches T_{SDT}-T_{SDHYS} during LIN=H, IGBTs don't turn on until inputting the next ON signal (LIN=L \rightarrow H).
 - IGBT of each phase can return to normal state by inputting ON signal to each phase.
 - d7. Normal operation: IGBT ON and outputs current $\ensuremath{I_{\text{C}}}$

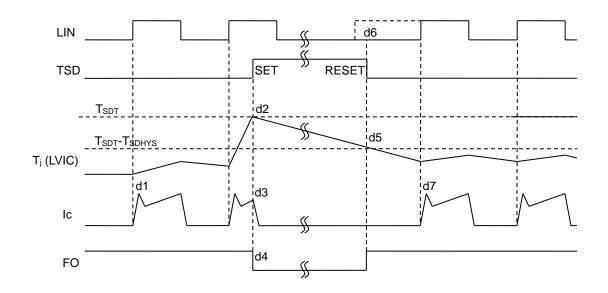
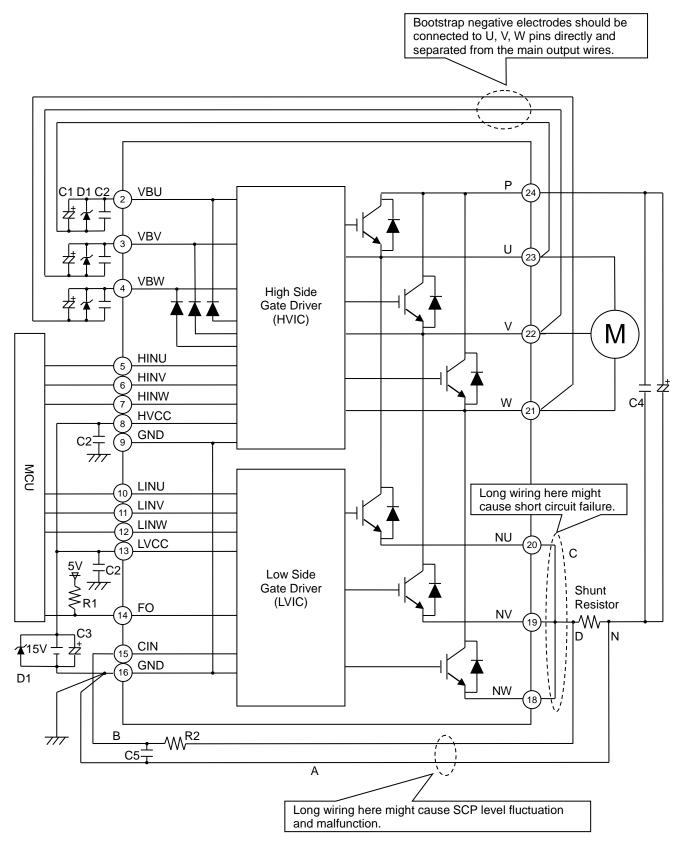


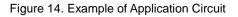
Figure 13. TSD Timing Chart

Notice

- 1) In case of TSD trip and FO output, please stop controlling IPM quickly to avoid the abnormal state.
- 2) If the cooling system is in abnormal state (e.g. heat sink comes off, fixed loosely, or cooling fan stops) when TSD trips, don't reuse IPM. This may cause the junction temperature of power chips to exceed its maximum rating of T_{jmax} (150°C).
- 3) TSD function detects LVIC temperature, so it cannot respond to rapid temperature rise of power chip. Therefore, TSD will not work properly in the case of rapid temperature rise like motor lock or over current.

Application Example(one shunt resistor drive)



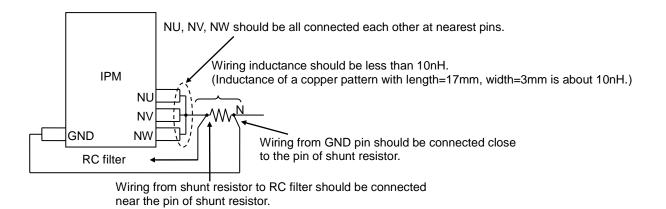


Selection of Components Externally Connected (Refer to Figure 14)

1) VBU, VBV, VBW pin

- The bypass capacitor(good temperature, frequency characteristic electrolytic type C1: 22µF to 100µF) should be mounted as close as possible to the pin in order to prevent malfunction or destruction due to switching noise and power supply ripple. In addition, for the purpose of reducing of the power supply's impedance in wide frequency bandwidth, ceramic capacitor (good temperature, frequency and DC bias characteristic ceramic type C2: 0.1µF to 0.22µF) should also be mounted.
- · Zenner diode D1(1W) should be mounted between each pair of control supply pins to prevent surge destruction.
- Line ripple voltage should meet dV/dt ≤1V/µs, V_{ripple} ≤ 2V_{p-p.}
- The wiring from U, V, W pin should be as thick and as short as possible. They should be connected directly and separated from the main output wires.
- 2) HVCC, LVCC pin
 - The bypass capacitor(good temperature, frequency characteristic electrolytic type C3) should be mounted as close as
 possible to the pin in order to prevent malfunction or destruction due to switching noise and power supply ripple. In
 addition, for the purpose of reducing of the power supply's impedance in wide frequency bandwidth, ceramic capacitor
 (good temperature, frequency and DC bias characteristic ceramic type C2: 0.1µF to 0.22µF) should also be mounted.
 - Zenner diode D1(1W) should be mounted between each pair of control supply pins to prevent surge destruction.
 - Line ripple voltage should meet $dV/dt \le 1V/\mu s$, $V_{ripple} \le 2V_{p-p}$.
- 3) P pin
 - To prevent surge destruction, the wiring between the smoothing capacitor and P, N pins should be as short as possible.
 - Snubber capacitor(C4: 0.1µF to 0.22µF) should be mounted between the P-N pin.
- 4) Control Input pin (HINU, HINV, HINW, LINU, LINV, LINW)
 - The wiring should be as short as possible to prevent malfunction.
 - Input drive is active-high type. There is a 3.3kΩ(Min) pull-down resistor in the input circuit of IPM. When using RC coupling circuit, make sure the input signal level meet the input threshold voltage.
 - · Dead time of input signal should be more than specified value.
- 5) FO pin
 - FO output is open drain type. It should be pulled up to control power supply(e.g. 5V, 15V) by a resistor that makes I_{FO} up to 1mA.I_{FO} is estimated roughly by the formula of control power supply voltage divided by pull-up resistance(R1). In the case of pulled up to 5V, R1=10k Ω is recommended.
- 6) CIN pin
 - RC filter(R2, C5) should be mounted as close as possible to the pin in order to prevent malfunction by recovery current or switching noise. It is recommended to select tight tolerance, temp-compensated type for RC filter (R2, C5). The time constant R2C5 (1.0µs to 2.0µs is recommended) should be set so that SCP current is shut down within 2µs. Please confirm operation on the actual application since SCP shutdown time changes depending on the PCB wiring pattern.
 - The point D at which the wiring to CIN filter is divided should be near the pin of shunt resistor. NU, NV, NW pin should be connected at near NU, NV, NW pin.
 - To prevent malfunction, the wiring of B should be as short as possible.
- 7) GND pin
 - Two GND pins (9 & 16 pin) are connected inside IPM. Please connect one pin (16 pin is recommended.) to the 15V power supply GND outside and leave the other open.
 - If control GND is connected with power GND by common broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect control GND and power GND at only a point N (near the pin of shunt resistor).
 - To prevent malfunction, the wiring of A should be as short as possible.
- 8) NU, NV, NW pin
 - When operating with one-shunt resistor, please short the three pins(NU, NV, NW). In addition, to prevent malfunction, the wiring of C should be as short as possible.

9) One-shunt Resistor Drive



Low inductance shunt resistor like surface mounted (SMD) type is recommended.

Figure 15. Wiring Pattern around the Shunt Resistor when Operating with One-shunt Resistor

- 10) Three-shunt Resistors Drive
 - It is not recommended to input the voltage of each shunt resistor directly to the CIN pin when IPM is operated with three shunt resistor. In that case, it is necessary to use the external protection circuit as below.
 - It is necessary to set the time constant Rufr(1.0µs to 2.0µs is recommended) of external comparator input so that IGBT stops within 2µs when short circuit occurs. Please confirm operation on the actual application since SCP shutdown time changes depending on the PCB wiring pattern.
 - It is recommended for the threshold voltage VREF to be set to the same rating of short circuit trip level(Vsc=0.48V(Typ))
 - To prevent malfunction, the wiring of A, B, C should be as short as possible.
 - OR output high level when protection works should be 0.53V(maximum Vsc rating) to 7V(CIN absolute maximum rating).

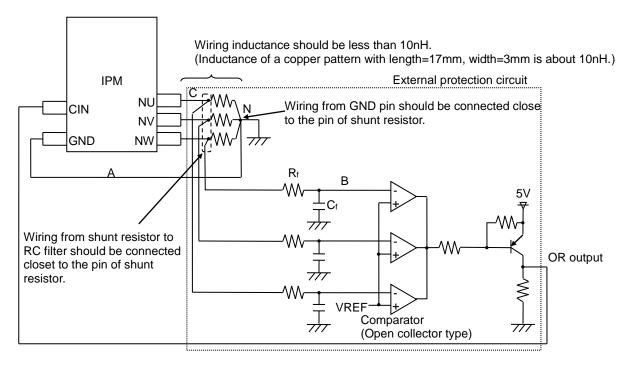
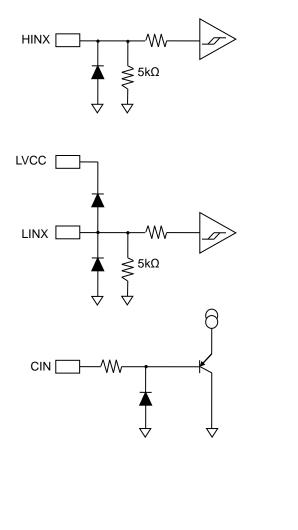


Figure 16. Wiring Pattern around the Shunt Resistor when operating with Three-shunt Resistors

I/O Equivalence Circuit



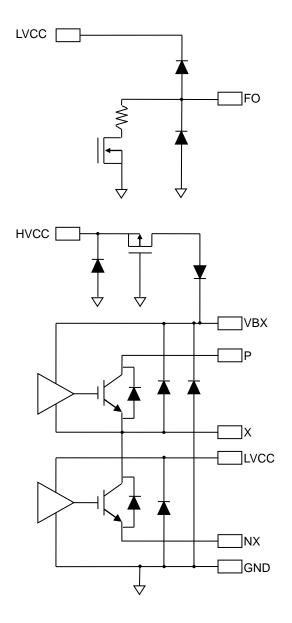


Figure 17. Input / Output Equivalent Circuit (X=U, V, W)

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IPM. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IPM's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition. However, pins that drive inductive loads (e.g. motor driver outputs, DC-DC converter outputs) may inevitably go below ground due to back EMF or electromotive force. In such cases, the user should make sure that such voltages going below ground will not cause the IPM and the system to malfunction by examining carefully all relevant factors and conditions such as motor characteristics, supply voltage, operating frequency and PCB wiring to name a few.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IPM can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IPM, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IPM has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IPM in the presence of a strong electromagnetic field may cause the IPM to malfunction.

9. Testing on Application Boards

When testing the IPM on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IPM to stress. Always discharge capacitors completely after each process or step. The IPM's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IPM during assembly and use similar precautions during transport and storage.

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IPM on the PCB. Incorrect mounting may result in damaging the IPM. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IPM are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IPM. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

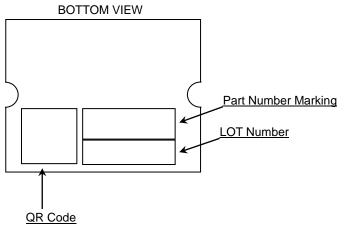
12. Area of Safe Operation (ASO)

Operate the IPM such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

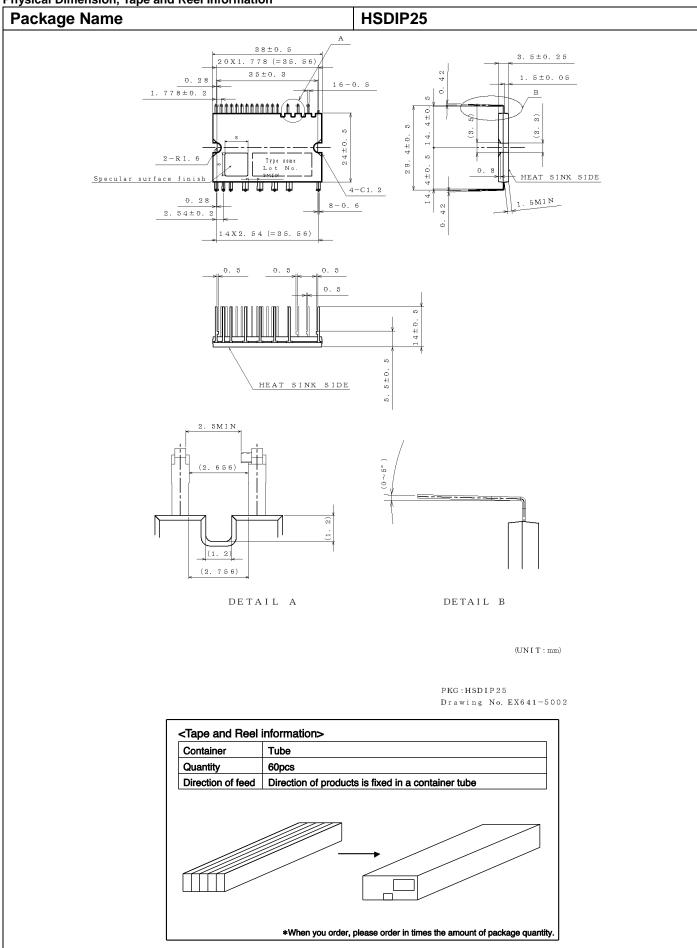
Ordering Information



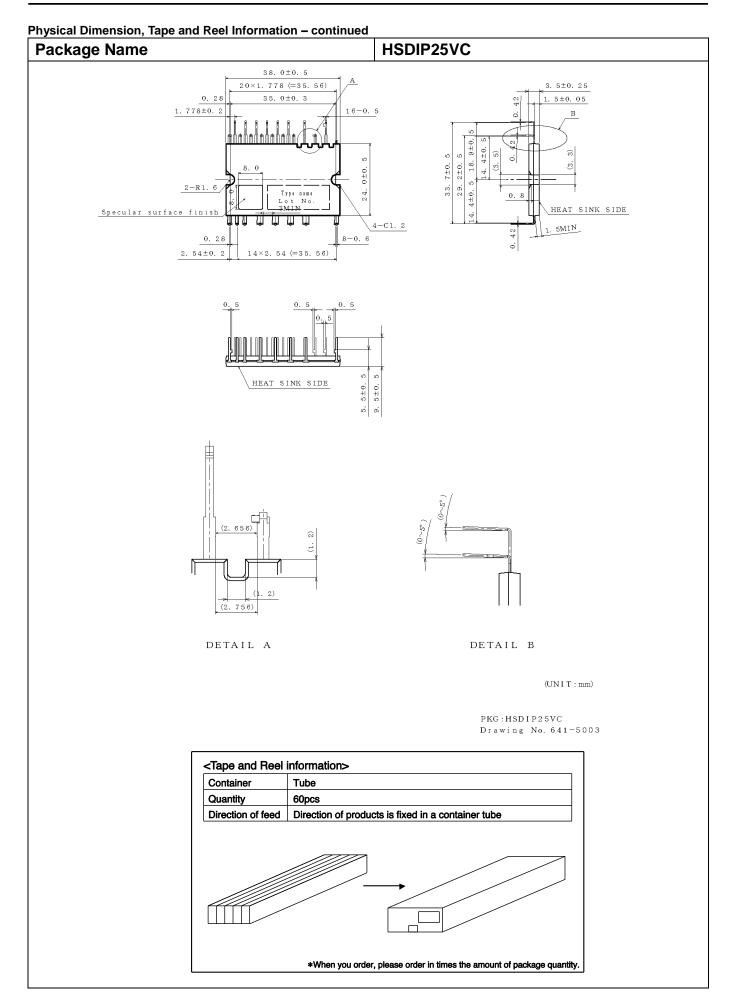
Marking Diagram



Physical Dimension, Tape and Reel Information



Datasheet



Revision History

Date	Revision	Changes		
15.Apr.2016	001	New Release		
24.Jun.2016	002	Page 1: "UL Recognized" added Page 3: Description of Block changed Page 16: Operational Notes changed Page 17: Marking Diagram changed		
20.Jun.2017	003	Page 7: (Note 1), (Note 2) changed, Figure6-2 added		
30.Oct.2017	004	Page 17: Description modified		

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CLASSⅣ		CLASSⅢ	

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 - [c] the Products are exposed to direct sunshine or condensation
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