

Operational Amplifiers Series

Ground Sense Low Power General Purpose Operational Amplifiers

LMR321G, LMR358xxx, LMR324xxx

General Description

LMR321, LMR358 and LMR324 are single, dual and quad low voltage operational amplifier with output full swing.

LMR321, LMR358 and LMR324 are the most effective solutions for applications where low supply current consumption and low voltage operation.

Features

- Operable with low voltage
- Input Ground Sense, Output Full Swing
- High open loop voltage gain
- Low supply current
- Low input offset voltage

Packages	W(Typ.) x D(Typ.) x H(Max.)
SSOP5	2.90mm x 2.80mm x 1.25mm
SOP8	5.00mm x 6.20mm x 1.71mm
SOP-J8	4.90mm x 6.00mm x 1.65mm
SSOP-B8	3.00mm x 6.40mm x 1.35mm
TSSOP-B8	3.00mm x 6.40mm x 1.20mm
MSOP8	2.90mm x 4.00mm x 0.90mm
TSSOP-B8J	3.00mm x 4.90mm x 1.10mm
SOP14	8.70mm x 6.20mm x 1.71mm
SOP-J14	8.65mm x 6.00mm x 1.65mm
SSOP-B14	5.00mm x 6.40mm x 1.35mm
TSSOP-B14J	5.00mm x 6.40mm x 1.20mm

Applications

- Portable equipment
- Low voltage application
- Active filter

Key Specifications

■ Operable with low voltage (single supply):

+2.7V to +5.5V

■ Low Supply Current:

LMR321	<u>130μA(Typ.)</u>
LMR358	210µA(Typ.)
LMR324	410µA(Typ.)
High Slew Rate:	1.0V/µs(Typ.)
Wide Temperature Range:	-40°C to +85°C
Low Input Offset Current:	5nA (Typ.)
Low Input Bias Current:	15nA (Typ.)

Simplified schematic

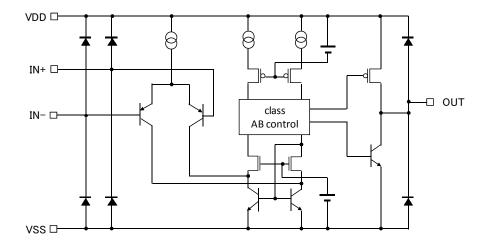
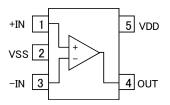


Figure 1. Simplified schematic

OProduct structure: Silicon monolithic integrated circuit OThis product is not designed protection against radioactive rays.

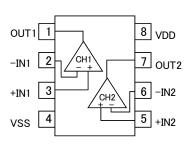
●Pin Configuration

SSOP5



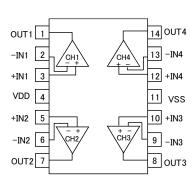
Pin No.	Symbol
1	+IN
2	VSS
3	-IN
4	OUT
5	VDD

SOP8, SOP-J8, SSOP-B8, TSSOP-B8, MSOP8, TSSOP-B8J



Pin No.	Symbol
1	OUT1
2	-IN1
3	+IN1
4	VSS
5	+IN2
6	-IN2
7	OUT2
8	VDD

SOP14, SOP-J14, SSOP-B14, TSSOP-B14J



Pin No.	Symbol		
1	OUT1		
2	-IN1		
3	+IN1		
4	VDD		
5	+IN2		
6	-IN2		
7	OUT2		
8	OUT3		
9	-IN3		
10	+IN3		
11	VSS		
12	+IN4		
13	-IN4		
14	OUT4		

Package								
SSOP5	SOP8	SOP-J8	SSOP-B8	TSSOP-B8	MSOP8			
LMR321G	LMR358F	LMR358FJ	LMR358FV	LMR358FVT	LMR358FVM			
	Package							
TSSOP-B8J	SOP14	SOP-J14	SSOP-B14	TSSOP-B14J	-			
LMR358FVJ	LMR324F	LMR324FJ	LMR324FV	LMR324FVJ	-			

Ordering Information

L M R 3 x x x x x - xx

Part Number LMR321G LMR358xxx LMR324xxx Package G : SSOP5

F : SOP8, SOP14 FV : SSOP-B8

SSOP-B14

FVM: MSOP8 FJ: SOP-J8

SOP-J14
FVJ: TSSOP-B8J
TSSOP-B14J

FVT: TSSOP-B8

Packaging and forming specification

E2: Embossed tape and reel

(SOP8/SOP-J8/SSOP-B8/TSSOP-B8/ TSSOP-B8J/SOP14/SOP-J14/SSOP-B14

TSSOP-B14J)

TR: Embossed tape and reel

(SSOP5/MSOP8)

●Line-up

Topr	Input type	V _{DD} (Min.)	Supply Current (Typ.)	Input Offset Voltage (Max.)	Pack	kage	Orderable Part Number
			130µA	±4mV	SSOP5	Reel of 3000	LMR321G-TR
					SOP8	Reel of 2500	LMR358F-E2
					MSOP8	Reel of 3000	LMR358FVM-TR
			210µA	±5mV	SOP-J8	Reel of 2500	LMR358FJ-E2
					SSOP-B8	Reel of 2500	LMR358FV-E2
-40°C to + 85°C	Ground Sense	2.7V			TSSOP-B8	Reel of 3000	LMR358FVT-E2
	Selise				TSSOP-B8J	Reel of 2500	LMR358FVJ-E2
					SOP14	Reel of 2500	LMR324F-E2
			4404	. 0 \ /	SOP-J14	Reel of 2500	LMR324FJ-E2
			410µA	±9mV	SSOP-B14	Reel of 2500	LMR324FV-E2
l					TSSOP-B14J	Reel of 2500	LMR324FVJ-E2

● Absolute Maximum Ratings(Ta=25°C)

	Symbol VDD-VSS			Rating			
Parameter			LMR321G	LMR358	LMR324	Unit	
Supply Voltage				+7		V	
		SSOP5	675 ^{*1*9}	-	-		
		SOP-J8	-	675 ^{*1*9}	-		
		SOP8	-	690 ^{*2*9}	-		
		SSOP-B8	-	625 ^{*3*9}	-		
		TSSOP-B8	-	625 ^{*3*9}	-		
Power dissipation	Pd	MSOP8	-	587 ^{*4*9}	-	mW	
		TSSOP-B8J	-	587 ^{*4*9}	-		
		SOP-J14	-	-	1025 ^{*5*9}	7	
		SSOP-B14	-	-	875 ^{*6*9}		
		TSSOP-B14J	-	-	850 ^{*7*9}		
		SOP14	-	-	562 ^{*8*9}		
Differential Input Voltage*10		Vid	VDD - VSS			V	
Input Common-mode Voltage Range		Vicm	(VSS-0.3) to (VDD+0.3)			V	
Operable with low voltage	Vopr		+2.7 to +5.5			V	
Operating Temperature	Topr		-40 to +85			°C	
Storage Temperature	Tstg		-55 to +150			°C	
Maximum Junction Temperature		Tjmax		+150		°C	

Note: Absolute maximum rating item indicates the condition which must not be exceeded.

Application of voltage in excess of absolute maximum rating or use out absolute maximum rated temperature environment may cause deterioration of characteristics.

- *1 To use at temperature above Ta=25°C reduce 5.4mW/°C.
- To use at temperature above Ta=25°C reduce 5.52mW/°C.
- *3 To use at temperature above Ta=25°C reduce 5mW/°C.
- *4 To use at temperature above Ta=25°C reduce 4.7mW/°C.
- *5 To use at temperature above Ta=25°C reduce 8.2mW/°C.
- *6 To use at temperature above Ta=25°C reduce 7mW/°C.
- *7 To use at temperature above Ta=25°C reduce 6.8mW/°C.
 *8 To use at temperature above Ta=25°C reduce 4.5mW/°C.
- *9 Mounted on a glass epoxy PCB(70mm×70mm×1.6mm).
- *10 The voltage difference between inverting input and non-inverting input is the differential input voltage.

Then input terminal voltage is set to more than VSS.

● Electrical Characteristics

OLMR321 (Unless otherwise specified VDD=+5V, VSS=0V)

Parameter	Cumbal	Temperature	Limits			Lloit	0 111
Parameter	Symbol	Range	Min.	Тур.	Max.	Unit	Condition
Input Offset Voltage *11	Vio	25°C	-	0.1	4	mV	VDD=2.7V to 5V
input Offset Voltage	VIO	Full range	-	-	5	IIIV	VDD=2.7 V to 5 V
Input Offset Voltage drift	ΔVio/ΔΤ	25°C	-	3	-	μV/°C	-
Input Offset Current *11	lio	25°C	-	5	50	nA	-
Input Bias Current *11	lb	25°C	-	15	100	nA	-
		25°C	-	107	180		VDD=2.7V, Av=0dB
Supply Current *12	IDD	Full range	-	-	260		VIN=0.95V
Supply Current	טטו	25°C	-	130	200	μA	VDD=5V, Av=0dB
		Full range	-	-	280		VIN=2.1V
Maximum Output Voltage(High)	VOH	25°C	VDD-0.1	VDD-0.04	-	V	RL=2kΩ to 2.5V
Maximum Output Voltage(Low)	VOL	25°C	-	VSS+0.08	VSS+0.16	V	RL=2kΩ to 2.5V
Large Signal Voltage Gain	Av	25°C	78	110	-	dB	RL=2kΩ
Input Common-mode Voltage Range	Vicm	25°C	0	-	4.2	V	VSS to VDD-0.8V
Common-mode Rejection Ratio	CMRR	25°C	65	90	-	dB	-
Power Supply Rejection Ratio	PSRR	25°C	65	90	-	dB	-
		0-	6	13	-		OUT=VDD-0.4V
Output Source Current *13	Isource	25°C	-	70	-	mA	OUT=0V, short current
*13		0-	30	60	-		OUT=VSS+0.4V
Output Sink Current *13	Isink	25°C	-	180	-	mA	OUT=5V, short current
Slew Rate	SR	25°C	-	1.0	-	V/µs	CL=25pF
Linite Donal width		05°0	-	2	-	N 41 1-	CL=25pF, Av=40dB
Unity Band width	f⊤	25°C	-	1	-	MHz	CL=200pF
Gain Band Width	GBW	25°C	-	3	-	MHz	f=100kHz
Phase Margin	θ	25°C	-	45	-	deg	CL=25pF, Av=40dB
Gain Margin	GM	25°C	-	10	-	dB	-
Input Referred Noise	Via	05°0	-	5.5	-	μVrms	Av=40dB
Voltage	Vn	25°C	-	39	-	nV/(Hz) ^{1/2}	Av=40dB, f=1kHz
Total Harmonic Distortion + Noise	THD+N	25°C	-	0.0015	-	%	OUT=0.4V _{P-P} f=1kHz

^{*11} Absolute value

 ^{*12} Full range: LMR321: Ta=-40°C to +85°C
 *13 Under the high temperature environment, consider the power dissipation of IC when selecting the output current. When the terminal short circuits are continuously output, the output current is reduced to climb to the temperature inside IC.

OLMR358 (Unless otherwise specified VDD=+5V, VSS=0V)

Danamatan	0	Temperature	Limits			l leit	0 1111
Parameter	Symbol	Range	Min.	Min. Typ. Max.		Unit	Condition
		25°C	-	0.1	5		
Input Offset Voltage *14	Vio	Full range	-	-	5	mV	VDD=2.7V to 5.0V
Input Offset Voltage drift	ΔVio/ΔT	25°C	-	3	-	μV/°C	-
Input Offset Current *14	lio	25°C	-	5	50	nA	-
Input Bias Current *14	lb	25°C	-	15	100	nA	-
		25°C	-	210	360		VDD=2.7V, Av=0dB
Supply Current *15	IDD	Full range	-	-	520		VIN=0.95V
Supply Current	טטו	25°C	-	210	380	μA	VDD=5V, Av=0dB
		Full range	-	-	540		VIN=2.1V
Maximum Output Voltage(High)	VOH	25°C	VDD-0.1	VDD-0.04	-	V	RL=2kΩ to 2.5V
Maximum Output Voltage(Low)	VOL	25°C	-	VSS+0.08	VSS+0.16	V	RL= $2k\Omega$ to $2.5V$
Large Signal Voltage Gain	Av	25°C	78	110	-	dB	RL=2kΩ
Input Common-mode Voltage Range	Vicm	25°C	0	-	4.2	V	VSS to VDD-0.8V
Common-mode Rejection Ratio	CMRR	25°C	65	90	-	dB	-
Power Supply Rejection Ratio	PSRR	25°C	65	90	-	dB	-
Output Source Current *16	Isource	25°C	6	13	-	mA	OUT=VDD-0.4V
Output Source Current	isource	25 0	-	70	-	IIIA	OUT=0V, short current
Output Sink Current *16	Isink	25°C	30	60	-	mA	OUT=VSS+0.4V
Output Sink Current	ISIIIK	25 C	-	180	-	IIIA	OUT=5V, short current
Slew Rate	SR	25°C	-	1.0	-	V/µs	CL=25pF
Unity Dand Width	4	25°C	-	2	-	MHz	CL=25F, Av=40dB
Unity Band Width	f⊤	25 C	-	1	-	IVITZ	CL=200pF
Gain Band Width	GBW	25°C	-	3	-	MHz	f=100kHz
Phase Margin	θ	25°C	-	45	-	0	CL=25pF, Av=40dB
Gain Margin	GM	25°C	-	10	-	dB	-
Input Referred Noise	\/-	05°0	-	5.5	-	μVrms	Av=40dB
Voltage	Vn	25°C	-	39	-	nV/(Hz) ^{1/2}	Av=40dB, f=1kHz
Total Harmonic Distortion + Noise	THD+N	25°C	-	0.0015	-	%	OUT=0.4V _{P-P} f=1kHz
Channel Separation	CS	25°C	-	100	-	dB	Av=40dB

^{*14} Absolute value

^{*15} Full range: LMR358: Ta=-40°C to +85°C

^{*16} Under the high temperature environment, consider the power dissipation of IC when selecting the output current.

When the terminal short circuits are continuously output, the output current is reduced to climb to the temperature inside IC.

OLMR324 (Unless otherwise specified VDD=+5V, VSS=0V)

Parameter	Symbol	Temperature	Limits			Unit	Condition
. a.a.neter	O J 2	Range	Min.	Тур.	Max.		
Input Offset Voltage *17	Vio	25°C	-	1.0	9	mV	VDD=2.7V to 5.0V
input Onset voltage	VIO	Full range	-	-	9	IIIV	VDD=2.7 V to 5.0 V
Input Offset Voltage drift	ΔVio/ΔΤ	25°C	-	3	-	μV/°C	-
Input Offset Current *17	lio	25°C	-	5	50	nA	-
Input Bias Current *17	lb	25°C	-	15	100	nA	-
		25°C	-	410	720		VDD=2.7V, Av=0dB
Supply Current *18	IDD	Full range	-	-	880		VIN=0.95V
Supply Current	טטו	25°C	-	410	800	μA	VDD=5V, Av=0dB
		Full range	-	-	900		VIN=2.1V
Maximum Output Voltage(High)	VOH	25°C	VDD-0.1	VDD-0.04	-	V	RL=2kΩ to 2.5V
Maximum Output Voltage(Low)	VOL	25°C	-	VSS+0.08	VSS+0.16	V	RL=2kΩ to 2.5V
Large Signal Voltage Gain	Av	25°C	78	110	-	dB	RL=2kΩ
Input Common-mode Voltage Range	Vicm	25°C	0	-	4.2	V	VSS to VDD-0.8V
Common-mode Rejection Ratio	CMRR	25°C	65	90	-	dB	-
Power Supply Rejection Ratio	PSRR	25°C	65	90	-	dB	-
Output Source Current *19	Isource	25°C	6	13	-	mA	OUT=VDD-0.4V
Catput Course Carront	1000100	200	-	70	-	1117 (OUT=0V, short currer
Output Sink Current *19	Isink	25°C	30	60	-	mA	OUT=VSS+0.4V
- Catput Clint Carront	ionnt	200	-	180	-		OUT=5V, short currer
Slew Rate	SR	25°C	-	1.0	-	V/µs	CL=25pF
Unity Gain Frequency	f⊤	25°C	-	2	-	MHz	CL=25pF, Av=40dB CL=200pF
Gain Band width	GBW	25°C	-	3	-	MHz	f=100kHz
Phase Margin	θ	25°C	-	45	-	deg	CL=25pF, Av=40dB
Gain Margin	GM	25°C	-	10	-	dB	-
Input Referred Noise	Vn	25°C	-	5.5	-	μVrms	Av=40dB
Voltage	• • • • • • • • • • • • • • • • • • • •		-	39	-	nV/(Hz) ^{1/2}	Av=40dB, f=1kHz
Total Harmonic Distortion + Noise	THD+N	25°C	-	0.0015	-	%	OUT=0.4V _{P-P} f=1kHz
Channel Separation	CS	25°C	-	100	-	dB	Av=40dB

^{*17} Absolute value

^{*18} Full range: LMR324: Ta=-40°C to +85°C

^{*19} Under the high temperature environment, consider the power dissipation of IC when selecting the output current. When the terminal short circuits are continuously output, the output current is reduced to climb to the temperature inside IC.

Description of electrical characteristics

Described here are the terms of electric characteristics used in this datasheet. Items and symbols used are also shown. Note that item name and symbol and their meaning may differ from those on another manufacture's document or general document.

1. Absolute maximum ratings

Absolute maximum rating item indicates the condition which must not be exceeded. Application of voltage in excess of absolute maximum rating or use out of absolute maximum rated temperature environment may cause deterioration of characteristics.

1.1 Power supply voltage (VDD/VSS)

Indicates the maximum voltage that can be applied between the positive power supply terminal and negative power supply terminal without deterioration or destruction of characteristics of internal circuit.

1.2 Differential input voltage (Vid)

Indicates the maximum voltage that can be applied between non-inverting terminal and inverting terminal without deterioration and destruction of characteristics of IC.

1.3 Input common-mode voltage range (Vicm)

Indicates the maximum voltage that can be applied to non-inverting terminal and inverting terminal without deterioration or destruction of characteristics. Input common-mode voltage range of the maximum ratings not assures normal operation of IC. When normal Operation of IC is desired, the input common-mode voltage of characteristics item must be followed.

1.4 Power dissipation (Pd)

Indicates the power that can be consumed by specified mounted board at the ambient temperature 25°C(normal temperature). As for package product, Pd is determined by the temperature that can be permitted by IC chip in the package (maximum junction temperature) and thermal resistance of the package.

2. Electrical characteristics item

2.1 Input offset voltage (Vio)

Indicates the voltage difference between non-inverting terminal and inverting terminal. It can be translated into the input voltage difference required for setting the output voltage at 0 V.

2.2 Input offset voltage drift (ΔVio/ΔT)

Denotes the ratio of the input offset voltage fluctuation to the ambient temperature fluctuation.

2.3 Input offset current (lio)

Indicates the difference of input bias current between non-inverting terminal and inverting terminal.

2.4 Input bias current (lb)

Indicates the current that flows into or out of the input terminal. It is defined by the average of input bias current at non-inverting terminal and input bias current at inverting terminal.

2.5 Circuit current (IDD)

Indicates the IC current that flows under specified conditions and no-load steady status.

2.6 Maximum Output Voltage(High) / Maximum Output Voltage(Low) (VOH/VOL)

Indicates the voltage range that can be output by the IC under specified load condition. It is typically divided into maximum output voltage High and low. Maximum output voltage high indicates the upper limit of output voltage. Maximum output voltage low indicates the lower limit.

2.7 Large signal voltage gain (Av)

Indicates the amplifying rate (gain) of output voltage against the voltage difference between non-inverting terminal and inverting terminal. It is normally the amplifying rate (gain) with reference to DC voltage.

Av = (Output voltage fluctuation) / (Input offset fluctuation)

2.8 Input common-mode voltage range (Vicm)

Indicates the input voltage range where IC operates normally.

2.9 Common-mode rejection ratio (CMRR)

Indicates the ratio of fluctuation of input offset voltage when in-phase input voltage is changed. It is normally the fluctuation of DC.

CMRR = (Change of Input common-mode voltage)/(Input offset fluctuation)

2.10 Power supply rejection ratio (PSRR)

Indicates the ratio of fluctuation of input offset voltage when supply voltage is changed. It is normally the fluctuation of DC. PSRR= (Change of power supply voltage)/(Input offset fluctuation)

2.11 Output source current/ output sink current (Isource/Isink)

The maximum current that can be output under specific output conditions, it is divided into output source current and output sink current. The output source current indicates the current flowing out of the IC, and the output sink current the current flowing into the IC.

2.12 Channel separation (CS)

Indicates the fluctuation of output voltage with reference to the change of output voltage of driven channel.

2.13 Slew Rate (SR)

SR is a parameter that shows movement speed of operational amplifier. It indicates rate of variable output voltage as unit time.

2.14 Unity gain frequency (f_T)

Indicates a frequency where the voltage gain of Op-Amp is 1.

2.15 Gain Band Width (GBW)

Indicates to multiply by the frequency and the gain where the voltage gain decreases 6dB/octave.

2.16 Phase Margin (θ)

Indicates the margin of phase from 180 degree phase lag at unity gain frequency.

2.17 Gain Margin (GM)

Indicates the difference between 0dB and the gain where operational amplifier has 180 degree phase delay.

2.18 Total harmonic distortion + Noise (THD+N)

Indicates the fluctuation of input offset voltage or that of output voltage with reference to the change of output voltage of driven channel.

2.19 Input referred noise voltage (Vn)

Indicates a noise voltage generated inside the operational amplifier equivalent by ideal voltage source connected in series with input terminal.

●Typical Performance Curves

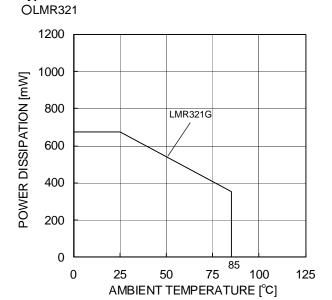


Figure 2. Derating curve

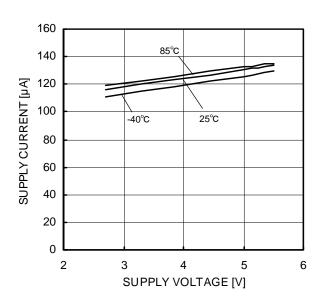


Figure 3.
Supply Current – Supply Voltage

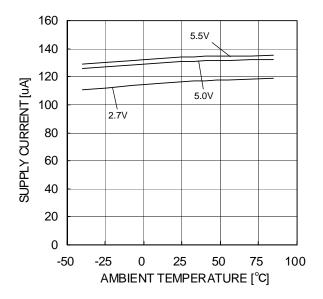


Figure 4.
Supply Current – Ambient Temperature

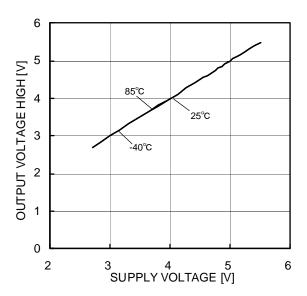


Figure 5. Maximum Output Voltage(High) – Supply Voltage (RL=2kΩ)

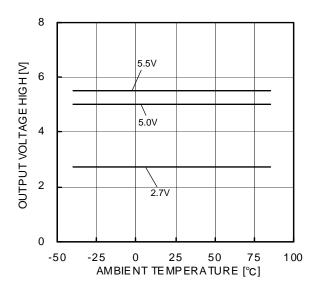


Figure 6.
Maximum Output Voltage(High)
– Ambient Temperature
(RL=2kΩ)

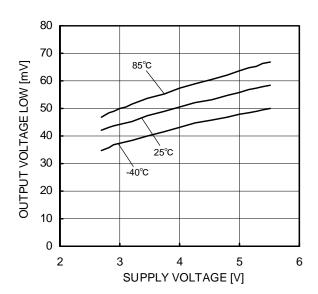


Figure 7.
Maximum Output Voltage(Low)
– Supply Voltage
(RL=2kΩ)

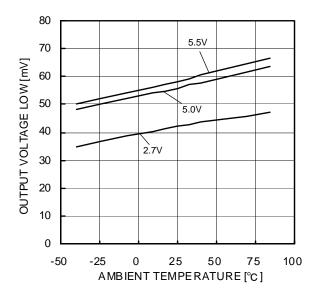


Figure 8.

Maximum Output Voltage(Low)

– Ambient Temperature

(RL= $2k\Omega$)

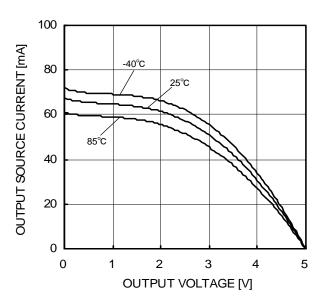


Figure 9.
Output Source Current – Output Voltage (VDD=5V)

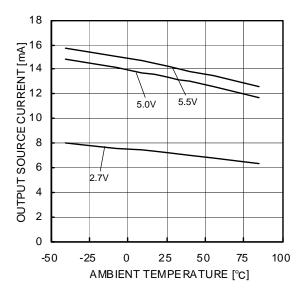


Figure 10.
Output Source Current – Ambient Temperature
(OUT=VDD-0.4V)

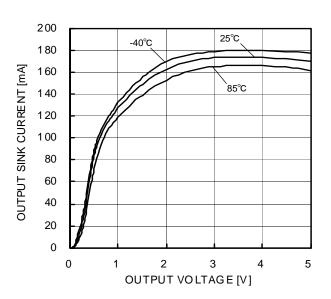


Figure 11.
Output Sink Current – Output Voltage
(VDD=5V)

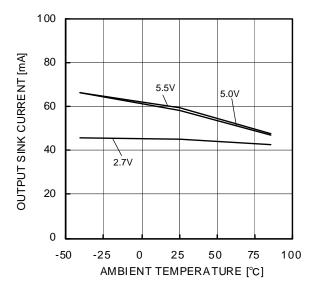


Figure 12.
Output Sink Current – Ambient Temperature (OUT=VSS+0.4V)

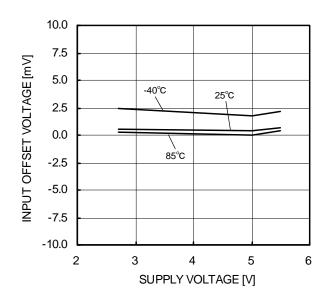


Figure 13.
Input Offset Voltage – Supply Voltage (Vicm= VDD, OUT= 0.1V)

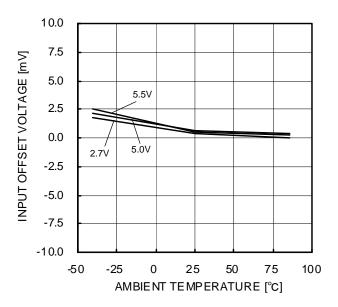


Figure 14.
Input Offset Voltage – Ambient Temperature
(Vicm= VDD, OUT= 0.1V)

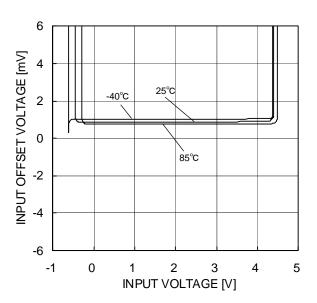


Figure 15.
Input Offset Voltage – Input Voltage (VDD=5V)

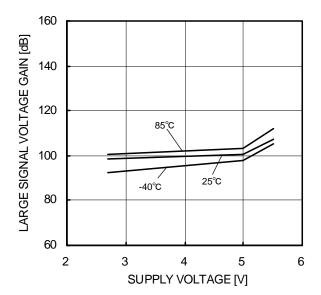


Figure 16. Large Signal Voltage Gain – Supply Voltage

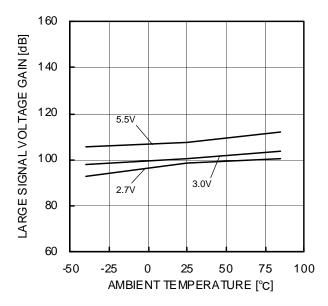
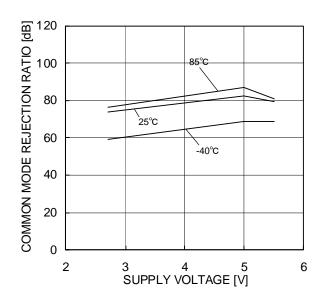


Figure 17.
Large Signal Voltage Gain – Ambient Temperature



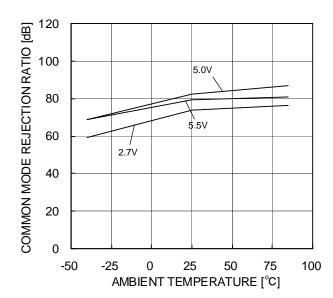
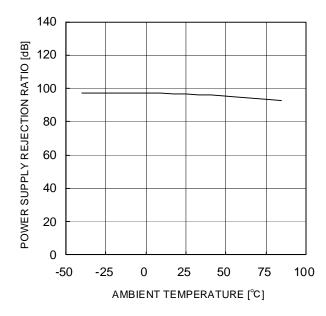


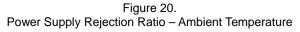
Figure 18.

Common Mode Rejection Ratio – Supply Voltage (VDD=5V)

Figure 19.

Common Mode Rejection Ratio – Ambient Temperature
(VDD=3V)





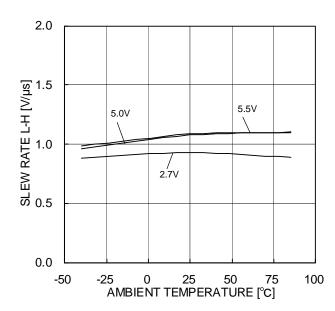


Figure 21.
Slew Rate L-H – Ambient Temperature

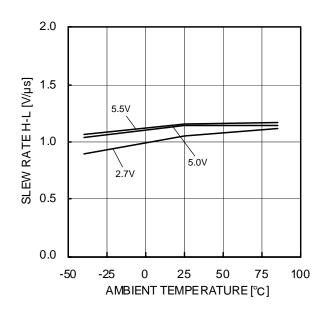


Figure 22.
Slew Rate H-L – Ambient Temperature

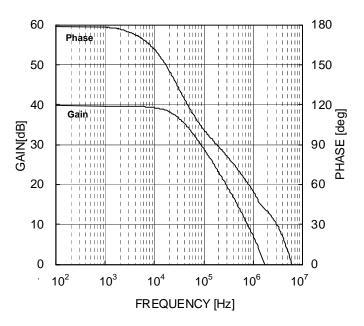


Figure 23.
Voltage Gain • Phase – Frequency

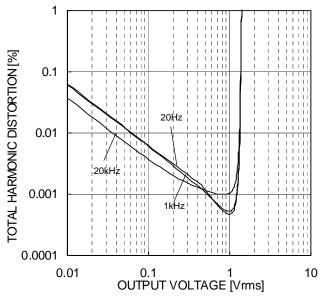


Figure 24.

Total Harmonic Distortion—Output Voltage
(VDD/VSS=+2.5V/-2.5V, Av=0dB,
RL=2kΩ, DIN-AUDIO, Ta=25°C)

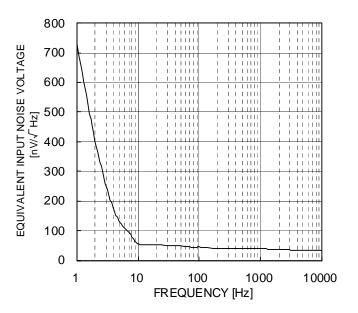
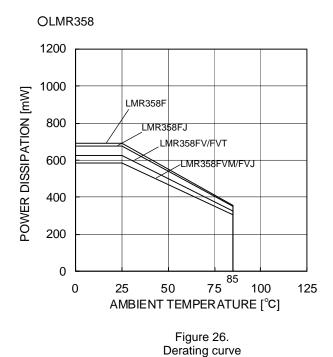


Figure 25.
Input Referred Noise Voltage—Frequency (VDD/VSS=+2.5V/-2.5V, Av=0dB, Ta=25°C)



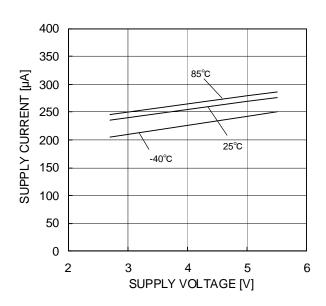


Figure 27.
Supply Current – Supply Voltage

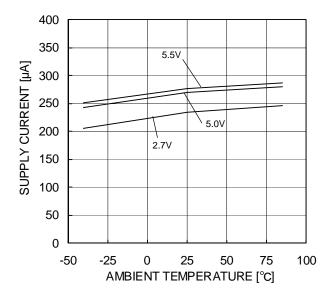


Figure 28.
Supply Current – Ambient Temperature

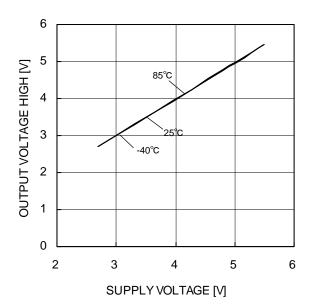


Figure 29.

Maximum Output Voltage(High)

– Supply Voltage
(RL=2kΩ)

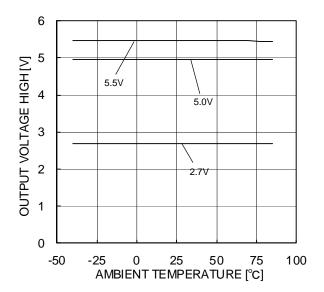


Figure 30.
Maximum Output Voltage(High)
– Ambient Temperature
(RL=2kΩ)

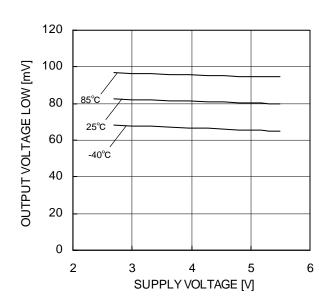


Figure 31.
Maximum Output Voltage(Low)
– Supply Voltage
(RL=2kΩ)

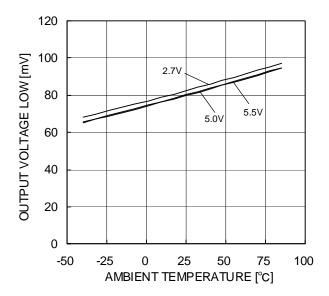


Figure 32.

Maximum Output Voltage(Low)

– Ambient Temperature
(RL=2kΩ)

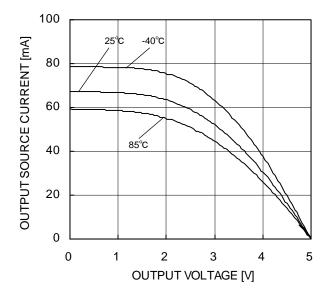


Figure 33.
Output Source Current – Output Voltage (VDD=5V)

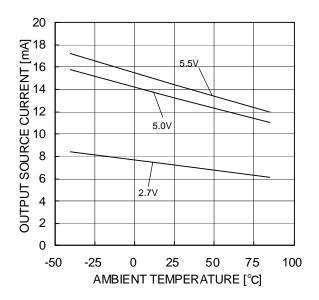


Figure 34.

Output Source Current – Ambient Temperature
(OUT=VDD-0.4V)

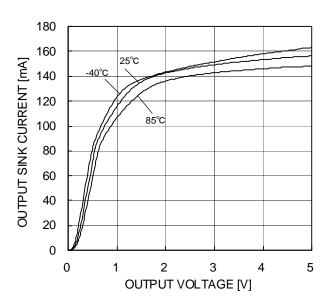


Figure 35.
Output Sink Current – Output Voltage (VDD=5V)

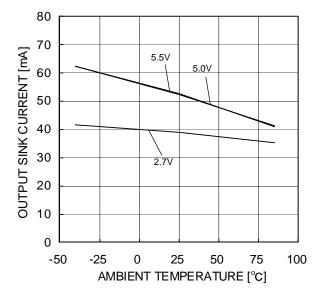


Figure 36.
Output Sink Current – Ambient Temperature (OUT=VSS+0.4V)

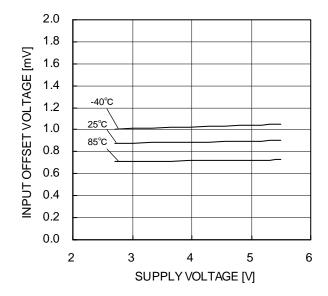


Figure 37.
Input Offset Voltage – Supply Voltage (Vicm= VDD, OUT= 0.1V)

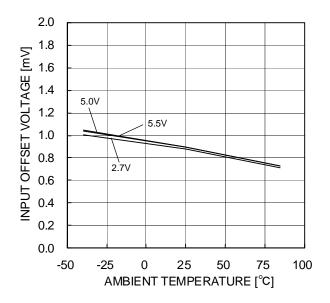


Figure 38.
Input Offset Voltage – Ambient Temperature (Vicm= VDD, OUT= 0.1V)

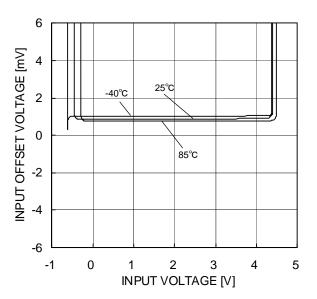


Figure 39.
Input Offset Voltage – Input Voltage (VDD=5V)

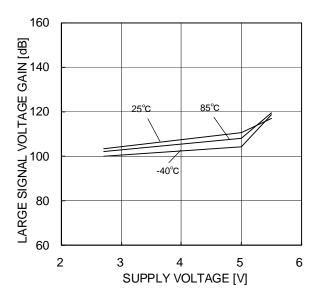


Figure 40. Large Signal Voltage Gain – Supply Voltage

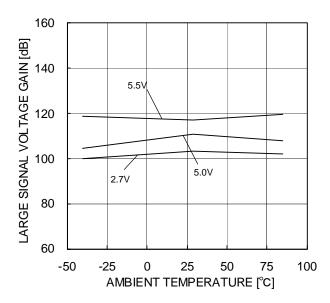
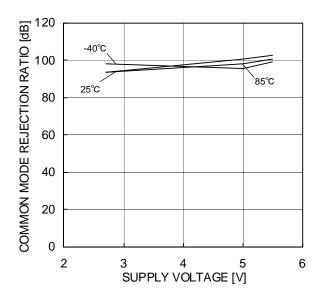


Figure 41.
Large Signal Voltage Gain – Ambient Temperature



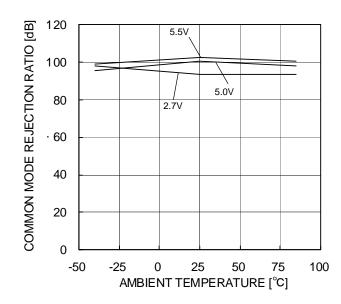
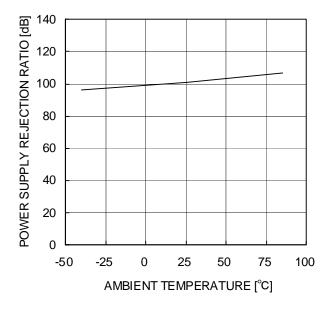


Figure 42.

Common Mode Rejection Ratio – Supply Voltage (VDD=5V)

Figure 43.

Common Mode Rejection Ratio – Ambient Temperature
(VDD=3V)



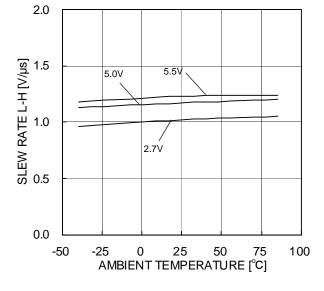


Figure 44.
Power Supply Rejection Ratio – Ambient Temperature

Figure 45.
Slew Rate L-H – Ambient Temperature

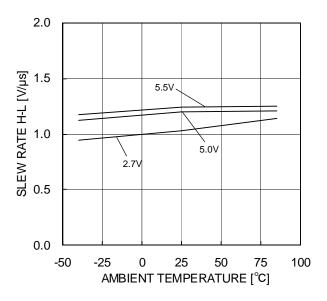


Figure 46.
Slew Rate H-L – Ambient Temperature

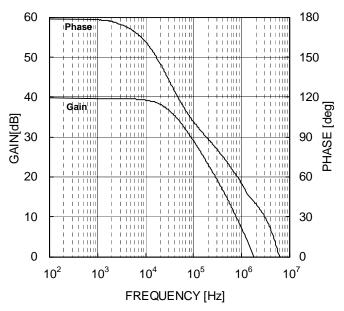


Figure 47.
Voltage Gain • Phase – Frequency

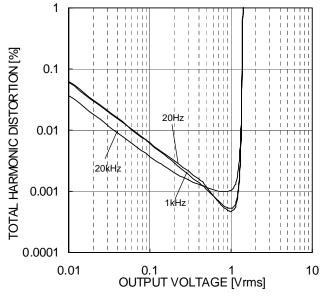


Figure 48.

Total Harmonic Distortion — Output Voltage
(VDD/VSS=+2.5V/-2.5V, Av=0dB,
RL=2kΩ, DIN-AUDIO, Ta=25°C)

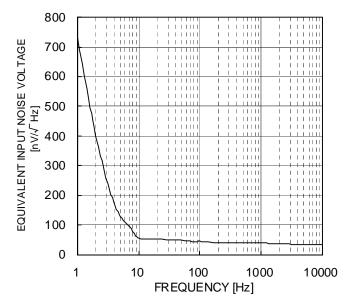
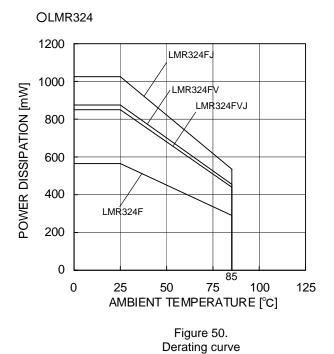


Figure 49.
Input Referred Noise Voltage—Frequency (VDD/VSS=+2.5V/-2.5V, Av=0dB, Ta=25°C)



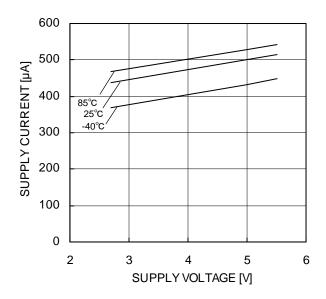
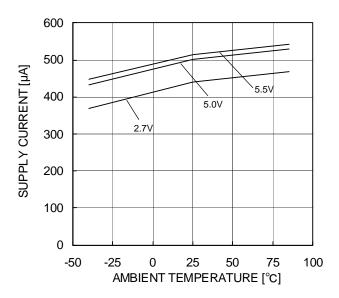
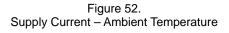


Figure 51.
Supply Current – Supply Voltage





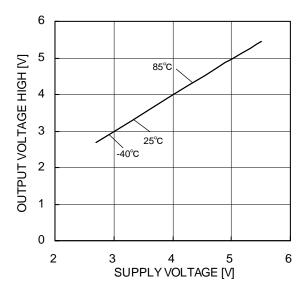


Figure 53.

Maximum Output Voltage(High)

– Supply Voltage
(RL=2 kΩ)

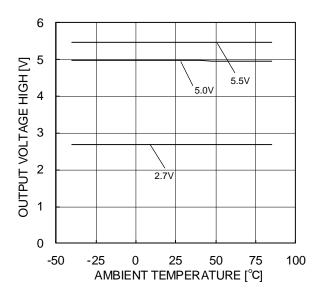


Figure 54.
Maximum Output Voltage(High)
– Ambient Temperature
(RL=2kΩ)

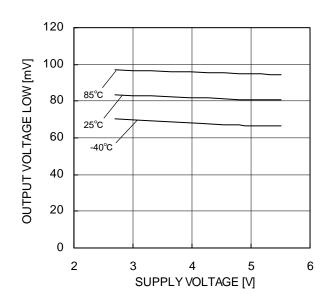


Figure 55.

Maximum Output Voltage(Low)

– Supply Voltage
(RL=2kΩ)

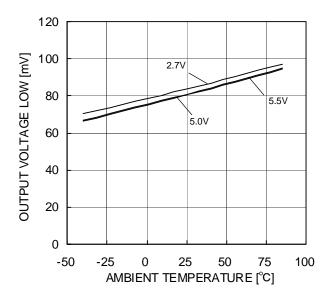


Figure 56.

Maximum Output Voltage(Low)

– Ambient Temperature

(RL= $2k\Omega$)

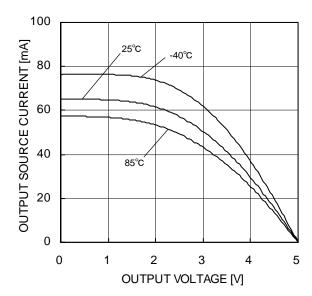


Figure 57.
Output Source Current – Output Voltage (VDD=5V)

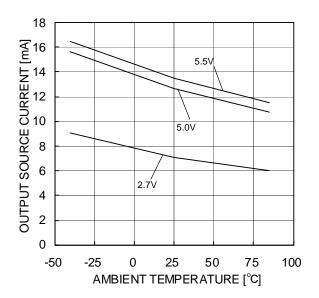


Figure 58.

Output Source Current – Ambient Temperature
(OUT=VDD-0.4V)

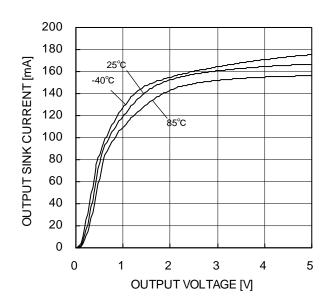


Figure 59.
Output Sink Current – Output Voltage (VDD=5V)

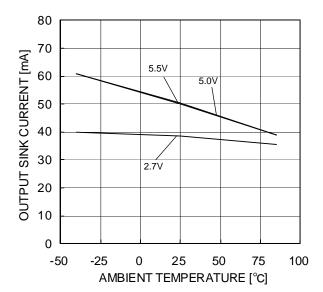


Figure 60.
Output Sink Current – Ambient Temperature (OUT=VSS+0.4V)

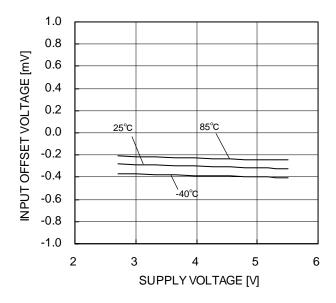


Figure 61.
Input Offset Voltage – Supply Voltage (Vicm= VDD, OUT= 0.1V)

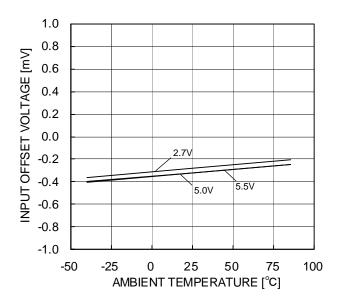


Figure 62.
Input Offset Voltage – Ambient Temperature
(Vicm= VDD, OUT= 0.1V)

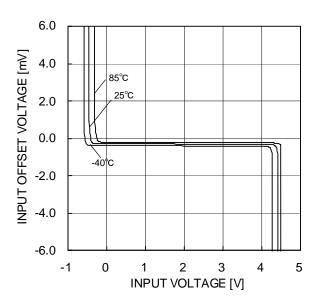


Figure 63.
Input Offset Voltage – Input Voltage (VDD=5V)

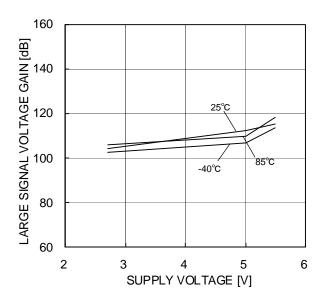


Figure 64. Large Signal Voltage Gain – Supply Voltage

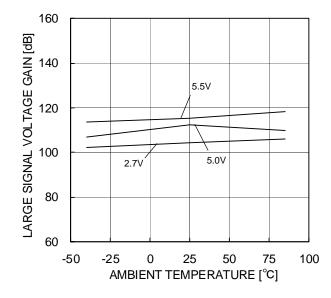
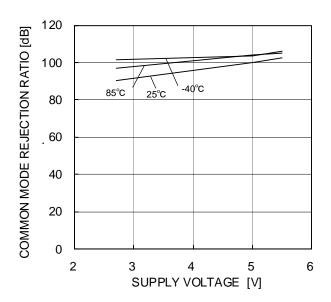


Figure 65.
Large Signal Voltage Gain – Ambient Temperature



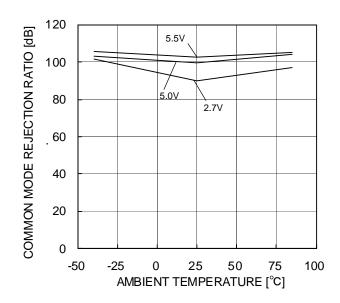
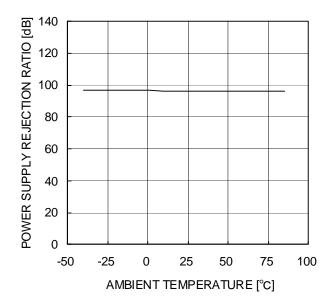


Figure 66.

Common Mode Rejection Ratio – Supply Voltage (VDD=5V)

Figure 67.

Common Mode Rejection Ratio – Ambient Temperature
(VDD=3V)



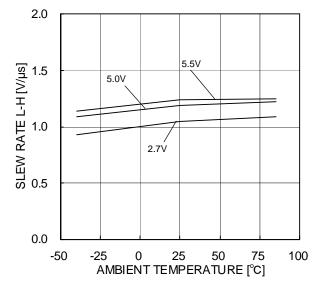


Figure 68.
Power Supply Rejection Ratio – Ambient Temperature

Figure 69. Slew Rate L-H – Ambient Temperature

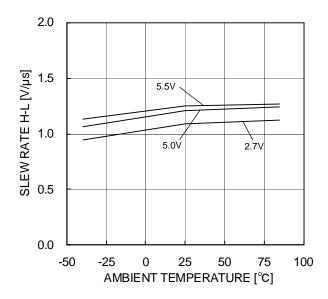


Figure 70.
Slew Rate H-L – Ambient Temperature

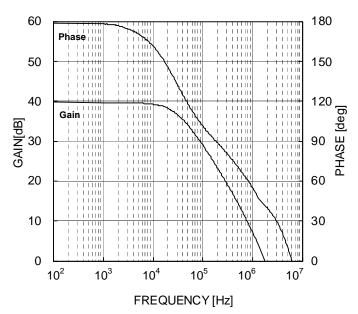


Figure 71.
Voltage Gain • Phase – Frequency

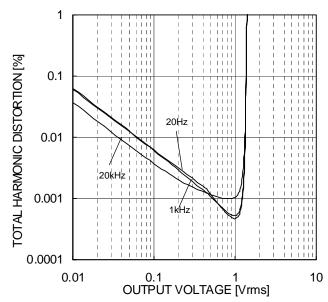


Figure 72.

Total Harmonic Distortion—Output Voltage
(VDD/VSS=+2.5V/-2.5V, Av=0dB,
RL=2kΩ, DIN-AUDIO, Ta=25°C)

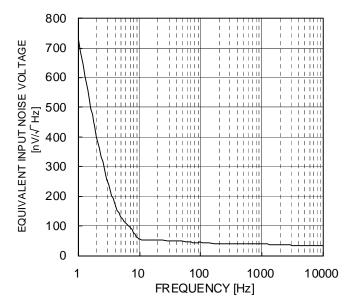


Figure 73.
Input Referred Noise Voltage—Frequency (VDD/VSS=+2.5V/-2.5V, Av=0dB, Ta=25°C)

● Application Information NULL method condition for Test Circuit 1

VDD	1/00	$\sqsubseteq \nu$	\/icm	Unit:V
VDD.	voo.	LIV.	VICILI	OHIL. V

Parameter	VF	S1	S2	S3	VDD	VSS	EK	Vicm	Calculation
Input Offset Voltage	VF1	ON	ON	OFF	5	0	-2.5	2.1	1
Larga Signal Valtaga Cain	VF2	ON	ON	ON	5	0	-1.5	2.1	2
Large Signal Voltage Gain	VF3 ON	ON	ON	5	U	-3.5	2.1		
Common-mode Rejection Ratio	VF4	ON	ON	OFF	5	0	-1.5	0	3
(Input Common-mode Voltage Range)	VF5	VF5		OFF	3	U	-1.5	1.8	
Power Supply Rejection Ratio	VF6	ON	ON	OFF	3	0	-2.9	4	4
Power Supply Rejection Ratio	VF7	ON	ON	OFF	5		-2.9	4	4

- Calculation-

1. Input Offset Voltage (Vio)
$$Vio = \frac{|VF1|}{4 + PF/PS}$$

2. Large Signal Voltage Gain(Av)
$$Av = 20Log \quad \frac{2 \times (1 + RF/RS)}{|VF2 - VF3|} \quad [dB]$$

3. Common-mode Rejection Ratio (CMRR)

CMRR=20Log
$$\frac{1.8 \times (1+RF/RS)}{|VF4 - VF5|}$$
 [dB]

4. Power Supply Rejection Ratio (PSRR)

$$PSRR = 20Log \quad \frac{3.8 \times (1 + RF/RS)}{|VF6 - VF7|} [dB]$$

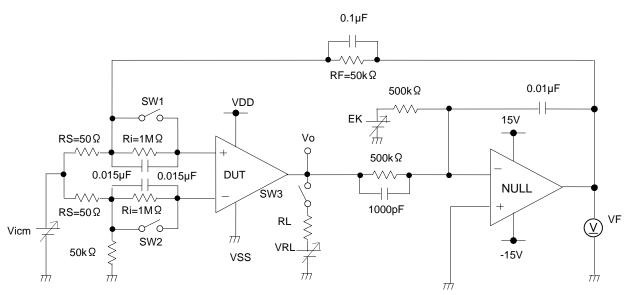
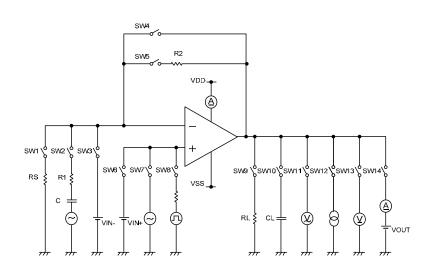


Figure 74. Test circuit 1 (one channel only)

Switch Condition for Test Circuit 2

SW No.	SW1	SW2	SW3	SW4	SW5	SW6	SW7	SW8	SW9	SW10	SW11	SW12	SW13	SW14
Supply Current	OFF	OFF	OFF	ON	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	OFF	OFF
Maximum Output Voltage(High)	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	ON	OFF
Maximum Output Voltage(Low)	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	ON	OFF
Output Source Current	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON
Output Sink Current	OFF	OFF	ON	OFF	OFF	ON	OFF	OFF	OFF	OFF	OFF	OFF	OFF	ON
Slew Rate	OFF	OFF	OFF	ON	OFF	OFF	OFF	ON	ON	ON	OFF	OFF	OFF	OFF
Gain Bandwidth Product	OFF	ON	OFF	OFF	ON	ON	OFF	OFF	ON	ON	OFF	OFF	OFF	OFF
Equivalent Input Noise Voltage	ON	OFF	OFF	OFF	ON	ON	OFF	OFF	OFF	OFF	ON	OFF	OFF	OFF



Input voltage

VH

Input wave

Output voltage

VH

90% SR=\Delta V/\Delta t

Output wave

Figure 75. Test Circuit 2 (each Op-Amp)

Figure 76. Slew Rate Input Waveform

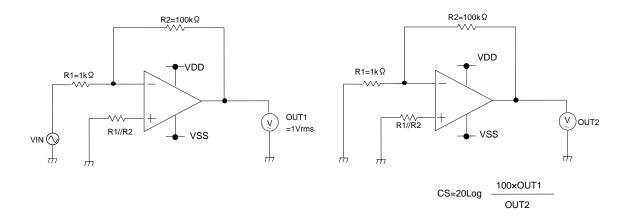


Figure 77. Test circuit 3(Channel Separation)

Application example

OVoltage Follower

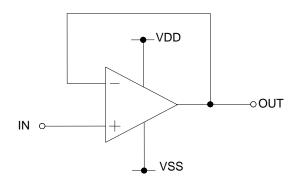


Figure 78. Voltage follower

Voltage gain is 0 dB.

This circuit controls output voltage (OUT) equal input voltage (IN), and keeps OUT with stable because of high input impedance and low output impedance.

OUT is shown next expression.

OUT=IN

OInverting amplifier

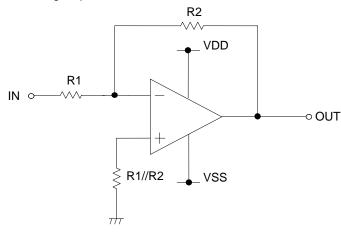


Figure 79. Inverting amplifier

For inverting amplifier, IN is amplified by voltage gain decided R1 and R2, and phase reversed voltage is output

OUT is shown next expression.

OUT=-(R2/R1) • IN

Input impedance is R1.

ONon-inverting amplifier

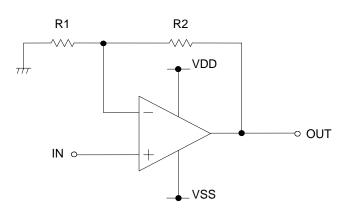


Figure 80. Non-inverting amplifier

For non-inverting amplifier, IN is amplified by voltage gain decided R1 and R2, and phase is same with Vin. OUT is shown next expression.

OUT=(1+R2/R1) • IN

This circuit performes high input impedance because Input impedance is operational amplifier's input Impedance.

OAdder circuit

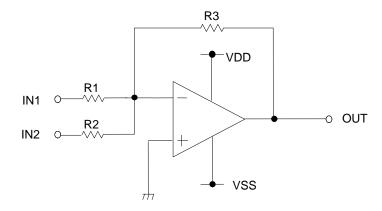


Figure 81. Adder circuit

Adder circuit output the voltage that added up Input voltage. A phase of the output voltage turns over, because non-inverting circuit is used.

OUT is shown next formula.

OUT = -R3(IN1/R1+IN2/R2)

When three input voltage is as above, it connects with input through resistance like R1 and R2.

ODifferential amplifier

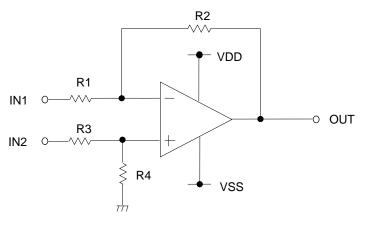


Figure 82. Differential amplifier

Differential amplifier output the voltage that amplified a difference of input voltage. In the case of R1=R3=Ra, R2=R4=Rb OUT is shown next formula. OUT = -Rb/Ra(IN1-IN2)

■Power Dissipation

Power dissipation (total loss) indicates the power that can be consumed by IC at Ta=25°C(normal temperature). IC is heated when it consumed power, and the temperature of IC ship becomes higher than ambient temperature. The temperature that can be accepted by IC chip depends on circuit configuration, manufacturing process, and consumable power is limited. Power dissipation is determined by the temperature allowed in IC chip (maximum junction temperature) and thermal resistance of package (heat dissipation capability). The maximum junction temperature is typically equal to the maximum value in the storage package (heat dissipation capability). The maximum junction temperature is typically equal to the maximum value in the storage temperature range. Heat generated by consumed power of IC radiates from the mold resin or lead frame of the package. The parameter which indicates this heat dissipation capability (hardness of heat release) is called thermal resistance, represented by the symbol θ ja°C/W. The temperature of IC inside the package can be estimated by this thermal resistance. Figure 83. (a) shows the model of thermal resistance of the package. Thermal resistance θ ja, ambient temperature Ta, maximum junction temperature Tjmax, and power dissipation Pd can be calculated by the equation below: θ ja = (Tjmax-Ta) / Pd °C/W ····· (I)

Derating curve in Figure 83. (b) indicates power that can be consumed by IC with reference to ambient temperature. Power that can be consumed by IC begins to attenuate at certain ambient temperature. This gradient iis determined by thermal resistance θja. Thermal resistance θja depends on chip size, power consumption, package, ambient temperature, package condition, wind velocity, etc even when the same of package is used. Thermal reduction curve indicates a reference value measured at a specified condition. Figure 84 (c)-(e) show a derating curve for an example LMR321, LMR358, LMR324.

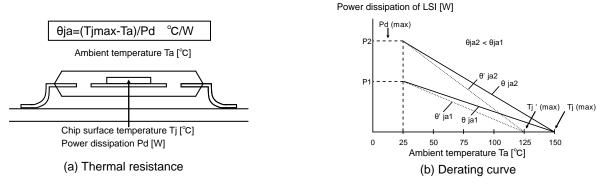
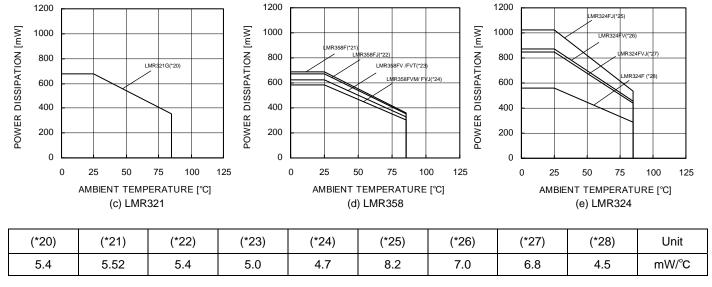


Figure 83. Thermal resistance and derating



When using the unit above Ta=25°C, subtract the value above per degree°C. Permissible dissipation is the value. When FR4 glass epoxy board 70mm×70mm×1.6mm (cooper foil area below 3%) is mounted.

Figure 84. Thermal resistance and derating

Operational Notes

1) Processing of unused circuit

It is recommended to apply connection (see the Figure 85.) and set the non inverting input terminal at the potential within input common-mode voltage range (Vicm), for any unused circuit.

2) Applied voltage to the input terminal

For normal circuit operation of voltage comparator, please input voltage for its input terminal within input common mode voltage VDD + 0.3V. Then, regardless of power supply voltage, VSS-0.3V can be applied to input terminals without deterioration or destruction of its characteristics.

3) Short-circuit of output terminal

When output terminal and VDD or VSS terminal are shorted, excessive Output current may flow under some conditions, and heating may destroy IC. It is necessary to connect a resistor as shown in Figure 86, thereby protecting against load shorting.

4) Operating power supply (split power supply/single power supply) The voltage comparator operates if a given level of voltage is applied between VDD and VSS. Therefore, the operational amplifier can be operated under single power supply or split power supply.

5) Power dissipation (pd)

If the IC is used under excessive power dissipation. An increase in the chip temperature will cause deterioration of the radical characteristics of IC. For example, reduction of current capability. Take consideration of the effective power dissipation and thermal design with a sufficient margin. Pd is reference to the provided power dissipation curve.

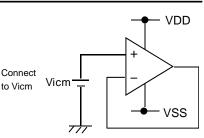


Figure 85. The example of application circuit for unused op-amp

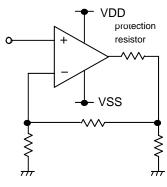


Figure 86. The example of output short protection

6) Short circuits between pins and incorrect mounting

Short circuits between pins and incorrect mounting when mounting the IC on a printed circuits board, take notice of the direction and positioning of the IC. If IC is mounted erroneously, It may be damaged. Also, when a foreign object is inserted between output, between output and VDD terminal and VSS terminal which causes short circuit, the IC may be damaged.

7) Using under strong electromagnetic field

Be careful when using the IC under strong electromagnetic field because it may malfunction.

8) Usage of IC

When stress is applied to the IC through warp of the printed circuit board, The characteristics may fluctuate due to the piezo effect. Be careful of the warp of the printed circuit board.

9) Testing IC on the set board

When testing IC on the set board, in cases where the capacitor is connected to the low impedance, make sure to discharge per fabrication because there is a possibility that IC may be damaged by stress. When removing IC from the set board, it is essential to cut supply voltage. As a countermeasure against the static electricity, observe proper grounding during fabrication process and take due care when carrying and storage it.

10) The IC destruction caused by capacitive load

The transistors in circuits may be damaged when VDD terminal and VSS terminal is shorted with the charged output terminal capacitor. When IC is used as a operational amplifier or as an application circuit, where oscillation is not activated by an output capacitor, the output capacitor must be kept below 0.1 µF in order to prevent the damage mentioned above.

11) Latch up

Be careful of input voltage that exceed the VDD and VSS. When CMOS device have sometimes occur latch up operation. And protect the IC from abnormaly noise

12) Decupling capacitor

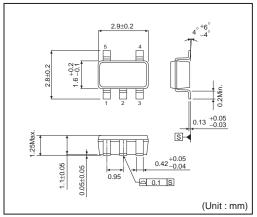
Insert the decupling capacitance between VDD and VSS, for stable operation of operational amplifier.

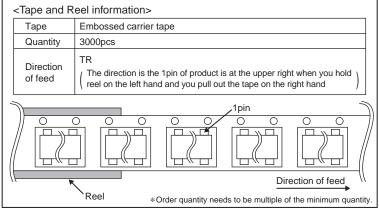
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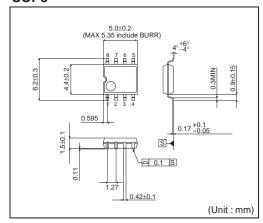
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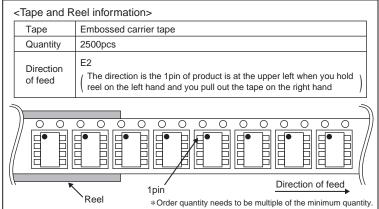
Physical Dimensions Tape and Reel Information SSOP5



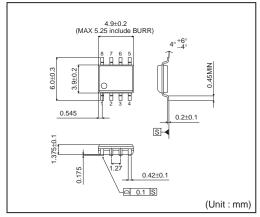


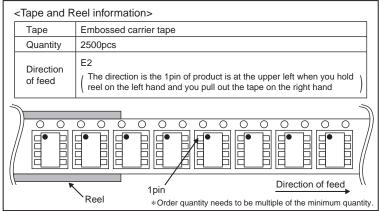
SOP8



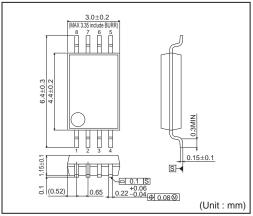


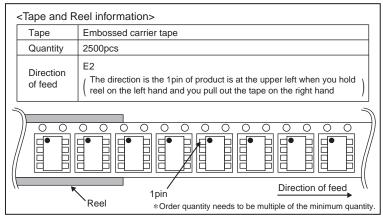
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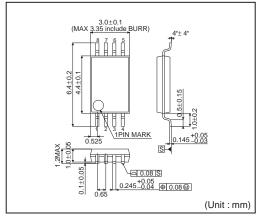


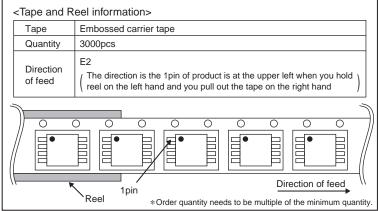
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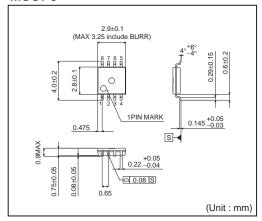


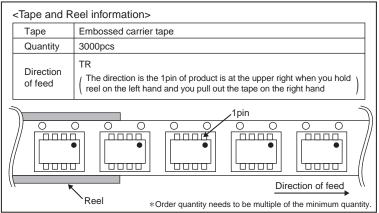
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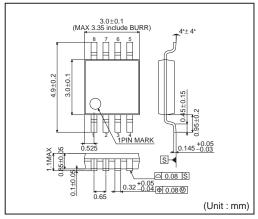


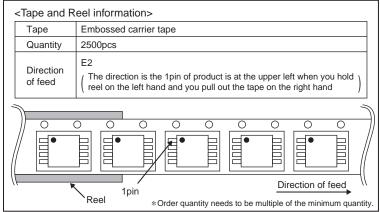
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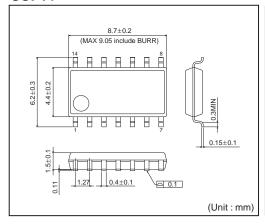


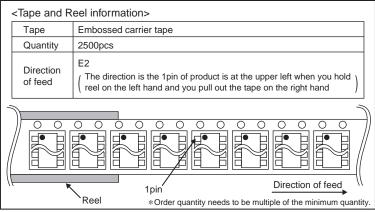
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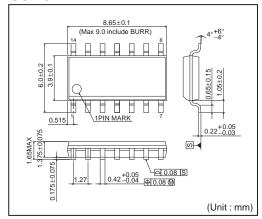


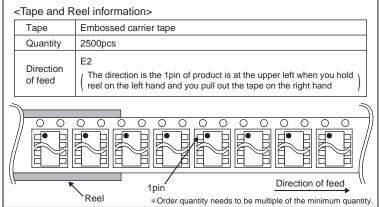
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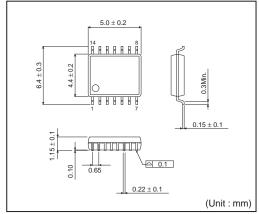


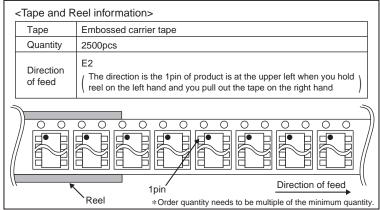
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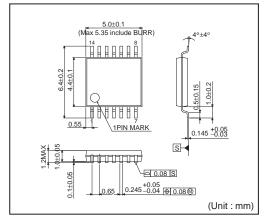


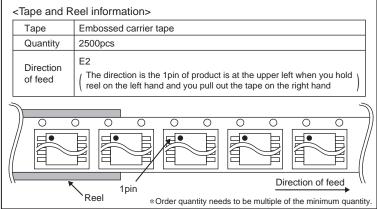
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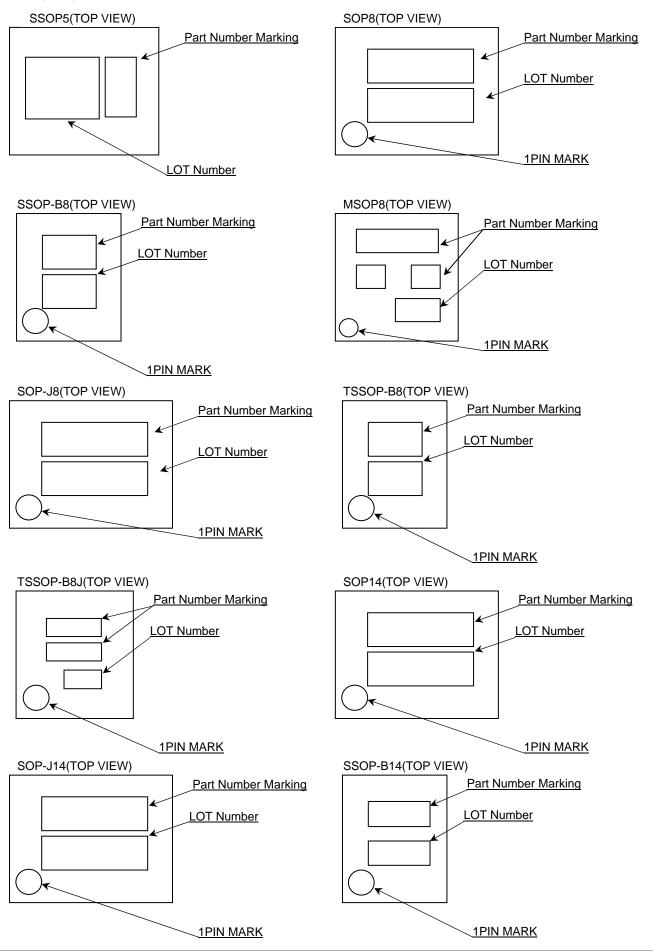


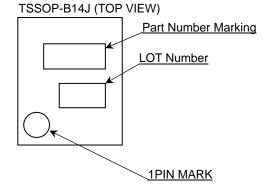
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Marking Diagrams

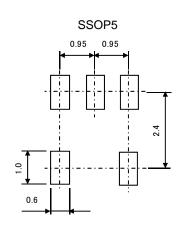


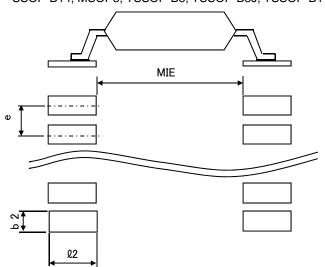


Product Na	Product Name		Product Name Marking		
LMR321 G		SSOP5	L2		
	F	SOP8	L358		
	FJ	SOP-J8	R358		
LMR358	FV	SSOP-B8	L358		
LIVINSSO	FVT	TSSOP-B8	R358		
	FVM	MSOP8	L358		
	FVJ	TSSOP-B8J	R358		
	F	SOP14	LMR324F		
LMR324	FJ	SOP-J14	LMR324FJ		
LIVIR 324	FV	SSOP-B14	L324		
	FVJ	TSSOP-B14J	R324		

●Land pattern data

SOP8, SOP14, SOP-J8, SOP-J14, SSOP-B8 SSOP-B14, MSOP8, TSSOP-B8, TSSOP-B8J, TSSOP-B14J





all dimensions in mm

all differsions in the							
PKG	PKG Land pitch e		Land length ≧ℓ 2	Land width b2			
SSOP5	0.95	2.4	1.0	0.6			
SOP8 SOP14	1.27	4.60	1.10	0.76			
SOP-J8 SOP-J14	1.27	3.90	1.35	0.76			
SSOP-B8 SSOP-B14	0.65	4.60	1.20	0.35			
MSOP8	0.65	2.62	0.99	0.35			
TSSOP-B8	0.65	4.60	1.20	0.35			
TSSOP-B8J	0.65	3.20	1.15	0.35			
TSSOP-B14J	0.65	4.60	1.20	0.35			

Revision History

Date	Revision	Changes
30.NOV.2012	001	New Release

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