Shouding

## Features

- Simple low parts count
- Internal 40 V NDMOS switch
- 1A output current
- Single pin on/off and brightness control

Using DC voltage or PWM

- Soft-start
- High efficiency (up to 97\%)
- Wide input voltage range: 6 V to 40 V
- Output shutdown
- Open LED Protection
- Short LED Protection
- Up to 1 MHz switching frequency
- Inherent open-circuit LED protection
- Typical 5\% output current accuracy
- Pb-free SOT23-5 and SOT89-5 Packages


## Applications

- Low voltage halogen replacement LEDs
- Low voltage industrial lighting
- LED back-side lighting
- Illuminated signs


## Description

The SD3361 is a continuous mode inductive step-down converter, designed for driving single or multiple series connected LEDs efficiently from a voltage source higher than the LED voltage. The device operates from an input supply between 6V and 40 V and provides an externally adjustable output current of up to 1 A . Depending upon supply voltage and external components, this can provide up to 30 watts of output power.

The SD3361 1 includes the output switch and a high-side output current sensing circuit, which uses an external resistor to set the nominal average output current.

Output current can be adjusted below the set value, by applying an external control signal to the VSET pin.

The VSET pin will accept either a DC voltage or a PWM waveform.

The soft-start time can be increased using an external capacitor from the VSET pin to ground. Applying a voltage of 0.2 V or lower to the VSET pin turns the output off and switches the device into a low current standby state.

## Typical Application



## Block Diagram



## Pin Configuration



EQ: Product Code of
X: Internal Code
Y: Year
W: Week

## Pin Descriptions

| Pin Number | Name | Description |
| :---: | :---: | :---: |
| SOT23-5/ SOT89-5 |  |  |
| 1 | LX | Drain of NDMOS switch. |
| 2 | GND | Ground (0V) |
| 3 | VSET | - Multi-function On/Off and brightness control pin: <br> - Leave floating for normal operation. <br> - Drive to voltage below 0.2 V to turn off output current <br> - Drive with DC voltage ( $0.3 \mathrm{~V}<\mathrm{VSET}<2.5 \mathrm{~V}$ ) to adjust output current from $12 \%$ to $100 \%$ of $\mathrm{I}_{\text {OUT }}$ nom <br> - Drive with PWM signal from open-collector or open-drain transistor, to adjust output current. Adjustment range $1 \%$ to $100 \%$ of $\mathrm{I}_{\text {OuT }}$ nom for $\mathrm{f}<500 \mathrm{~Hz}$ <br> - Connect a capacitor from this pin to ground to increase soft-start time. (Default soft-start time $=0.1 \mathrm{~ms}$. Additional soft-start time is approx. $1.5 \mathrm{~ms} / 1 \mathrm{nF}$ ) |
| 4 | ISENSE | Connect resistor $\mathrm{R}_{\mathrm{S}}$ from this pin to VIN to define nominal average output current loutnom = 0.1/Rs |
| 5 | VIN | Input voltage ( 6 V to 40 V ). Decouple to ground with $10 \mu \mathrm{~F}$ or higher X 7 R ceramic capacitor close to device. |

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## Absolute Maximum Ratings

These are stress ratings only and functional operation is not implied. Exposure to absolute maximum ratings for prolonged time periods may affect device reliability. All voltages are with respect to ground.

Input Voltage Range $\qquad$ -0.3V to 40V Maximum Junction Temperature $170^{\circ} \mathrm{C}$
LX, ISENSE Pin voltage
0.3 V to 40 V Storage Temperature

VSET Pin voltage
.-0.3 V to 6 V Soldering Temperature. $65^{\circ} \mathrm{C}$ to $170^{\circ} \mathrm{C}$ $300^{\circ} \mathrm{C}, 5 \mathrm{sec}$

## Recommended Operating Conditions

Input Voltage Range. $\qquad$ 6 V to 40 V

Ambient Temperature Range. $\qquad$ .$-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ Junction Temperature Range $-40^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$

## Thermal Information

| Parameter | Package | Symbol | Maximum | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Thermal Resistance (Junction to Ambient) | SOT23-5 | $\theta_{\mathrm{j}} \mathrm{A}$ | 250 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
|  | SOT89-5 |  | 160 |  |
| Thermal Resistance (Junction to Case) | SOT23-5 | $\theta_{\mathrm{jc}}$ | 130 |  |
|  | SOT89-5 |  | 45 |  |

1A LED Driver with Internal Switch

## Electrical Characteristic

$\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{IN}}=16 \mathrm{~V}$, unless otherwise noted.

| Parameter | Symbol | Condition | Min | Typ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input voltage | Vin |  | 6 |  | 40 | V |
| Output current | $\mathrm{I}_{\text {Led }}$ | $\mathrm{R}_{\mathrm{S}}=0.3 \Omega$ |  | 333 |  | mA |
|  |  | $\mathrm{R}_{\mathrm{S}}=0.1 \Omega$ |  | 1 |  | A |
| Shutdown current | $\mathrm{I}_{\text {SD }}$ | VSET pin grounded |  | 20 | 40 | $\mu \mathrm{A}$ |
| Quiescent current without switching | $\mathrm{I}_{\mathrm{Q}}$ | VSET pin floating, Vin=16V |  | 0.6 |  | mA |
| Mean current sense threshold voltage | Vsense | Measured on Isense pin with respect to VIN | 95 | 100 | 105 | mV |
| Sense threshold hysteresis | Vsense_hys |  |  | $\pm 13$ |  | \% |
| Isense pin input current | Isense | Vsense $=$ Vin - 0.1 |  | 8 |  | $\mu \mathrm{A}$ |
| Vset range on VSET pin | Ven | For DC dimming | 0.3 |  | 2.5 | V |
| DC voltage on VSET pin to enable | Venon | Ven rising |  | 0.25 |  | V |
| DC voltage on VSET pin to disable | venoff | Ven falling |  | 0.2 |  | V |
| LX switch on resistance | $\mathrm{R}_{\text {LX }}$ | @ $\mathrm{L}_{\mathrm{Lx}}=100 \mathrm{~mA}$ |  | 0.3 |  | $\Omega$ |
| LX switch leakage current | $\mathrm{I}_{\text {LX(leak) }}$ |  |  |  | 5 | $\mu \mathrm{A}$ |
| Soft start time | Tss | Vin=16V, Cen $=1 \mathrm{nF}$ |  | 1.5 |  | ms |
| Operating frequency | $\mathrm{F}_{\text {LX }}$ | $\begin{aligned} & \mathrm{Vi}=16 \mathrm{~V}, \mathrm{Vo}=9.6 \mathrm{~V}(3 \mathrm{LEDS}), \\ & \mathrm{L}=47 \mu \mathrm{H}, \Delta \mathrm{l}=0.25 \mathrm{~A}\left(\mathrm{I}_{\mathrm{LED}}=1 \mathrm{~A}\right) \end{aligned}$ |  | 233 |  | kHz |
| Recommended minimum switch ON time | Ton_rec | For 4\% accuracy |  | 500 |  | ns |
| Recommended maximum switch frequency | $F_{\text {LXmax }}$ |  |  |  | 1.0 | MHz |
| Max duty circle |  |  |  | 98 |  | \% |
| Recommended duty cycle range | $\mathrm{D}_{\text {LX }}$ |  | 25 |  | 75 | \% |
| Internal comparator propagation delay | TPD* |  |  | 45 |  | ns |
| Over temperature protection | T OTP |  |  | 150 |  | ${ }^{\circ} \mathrm{C}$ |
| Temp protection hysteresis | TotP_hys |  |  | 40 |  | ${ }^{\circ} \mathrm{C}$ |
| Current limit | IXLmax | Peak inductor current | 1.5 |  |  | A |

*parameters are not tested at production, but guaranteed by design.
Note: Vsense: level A: 95 mV to 100 mV ; level B:100mV to 105 mV

## Typical Operating Characteristics

$T_{A}=25^{\circ} \mathrm{C}, \mathrm{V}_{\text {IN }}=16 \mathrm{~V}$, unless otherwise noted.

1. Efficiency vs Input Voltage (Rs=0.3,$L=100 \mu \mathrm{H}$ )

2. Efficiency vs Input Voltage
(Rs=0.1 $\Omega, L=33 \mu \mathrm{H}$ )

3. Operating Frequency vs Input Voltage (Rs $=0.15 \Omega, L=47 \mu \mathrm{H}$ )

4. Efficiency vs Input Voltage
(Rs=0.15 $\Omega, L=47 \mu \mathrm{H}$ )

5. Operating Frequency vs Input Voltage (Rs $=0.3 \Omega, L=100 \mu \mathrm{H}$ )

6. Operating Frequency vs Input Voltage (Rs=0.1 $\Omega, L=33 \mu \mathrm{H}$ )


## Typical Operating Characteristics

$T_{A}=25^{\circ} \mathrm{C}, \mathrm{V}_{\text {IN }}=16 \mathrm{~V}$, unless otherwise noted.
7. Quiescent Current VS. Input Voltage

9. Feedback Voltage VS Vin

11. $\mathrm{I}_{\text {LED }} \mathrm{Vs} \mathrm{V}_{\text {SET }}$
$\left(\mathrm{V}_{\text {IN }}=16 \mathrm{~V}, 3\right.$ LEDs, $\left.100 \mu \mathrm{H}, \mathrm{Rs}=0.3 \Omega\right)$

8. Shutdown Current VS. Input Voltage

10. Duty Cycle VS LED Current (PWM $=200 \mathrm{~Hz}$ )

12. $I_{\text {LED }}$ Vs $V_{\text {SET }}$
$\left(\mathrm{V}_{\text {IN }}=16 \mathrm{~V}, 3\right.$ LEDs, $\left.47 \mu \mathrm{H}, \mathrm{Rs}=0.15 \Omega\right)$


## Typical Operating Characteristics

$T_{A}=25^{\circ} \mathrm{C}, \mathrm{V}_{\text {IN }}=16 \mathrm{~V}$, unless otherwise noted.

15. Dimming Waveforms (PWM=50\%)

14. Start up Waveforms

16. Pulse skip mode


## Application Information

## Setting nominal average output current with external resistor $R_{s}$

The nominal average output current in the LED(s) is determined by the value of the external current sense resistor ( $\mathrm{R}_{\mathrm{s}}$ ) connected between VIN and ISENSE and is given by:

$$
\mathrm{I}_{\text {OuTnom }}=\frac{0.1}{\mathrm{R}_{\mathrm{s}}}
$$

The table below gives values of nominal average output current for several preferred values of current setting resistor $\left(R_{s}\right)$ in the typical application circuit shown on page 1.

| $\mathbf{R}_{\mathbf{S}}(\boldsymbol{\Omega})$ | Nominal average output current (mA) |
| :---: | :---: |
| 0.1 | 1000 |
| 0.13 | 760 |
| 0.15 | 667 |
| 0.3 | 333 |

The above values assume that the VSET pin is floating and at a nominal voltage of VREF (1.25V). Note that $R_{s}=0.1 \Omega$ is the minimum allowed value of sense resistor under these conditions to maintain switch current below the specified maximum value. It is possible to use different values of $R_{s}$ if the VSET pin is driven from an external voltage.

## Capacitor selection

A low ESR capacitor should be used for input decoupling, as the ESR of this capacitor appears in series with the supply source impedance and lowers overall efficiency. This capacitor has to supply the relatively high peak current to the coil and smooth the current ripple on the input supply. A minimum value of $4.7 \mu \mathrm{~F}$ is acceptable if the input source is close to the device, but higher values will improve performance at lower input voltages, especially when the source impedance is high. The input capacitor should be placed as close as possible to the IC.

For maximum stability over temperature and voltage, capacitors with X7R, X5R, or better dielectric are recommended. Capacitors with Y5V dielectric are not suitable for decoupling in this application and should NOT be used.

## Inductor selection

Recommended inductor values for the SD3361 are in the range $33 \mu \mathrm{H}$ to $100 \mu \mathrm{H}$. Higher values of inductance are recommended at higher supply voltages in order to minimize errors due to switching delays, which result in increased ripple and lower efficiency. Higher values of inductance also result in a smaller change in output current
over the supply voltage range. The inductor should be mounted as close to the device as possible with low resistance connections to the LX and VIN pins. The chosen coil should have a saturation current higher than the peak output current and a continuous current rating above the required mean output current.

The inductor value should be chosen to maintain operating duty cycle and switch 'on'/'off' times within the specified limits over the supply voltage and load current range. The following equations can be used as a guide.
LX Switch 'On' time

$$
T_{\mathrm{ON}}=\frac{\mathrm{L} \Delta \mathrm{I}}{\mathrm{~V}_{\mathrm{IN}}-\mathrm{V}_{\mathrm{LED}}-\mathrm{I}_{\mathrm{LED}}\left(\mathrm{R}_{\mathrm{S}}+\mathrm{R}_{\mathrm{L}}+\mathrm{R}_{\mathrm{LX}}\right)}
$$

LX Switch 'Off' time

$$
T_{\text {OFF }}=\frac{\mathrm{L} \Delta I}{\mathrm{~V}_{\text {LED }}+\mathrm{V}_{\mathrm{D}}+\mathrm{I}_{\mathrm{LED}}\left(\mathrm{R}_{\mathrm{S}}+\mathrm{R}_{\mathrm{L}}\right)}
$$

Where: $L$ is the coil inductance; $R_{L}$ is the coil resistance ; $R_{s}$ is the current sense resistance $I_{\text {LED }}$ is the required LED current; $\Delta l$ is the coil peakpeak ripple current (Internally set to $0.25 \times \mathrm{I}_{\text {LED }}$ ); $\mathrm{V}_{\text {IN }}$ is the supply voltage; $\mathrm{V}_{\text {LED }}$ is the total LED forward voltage; $R_{L x}$ is the switch resistance $(0.3 \Omega$ nominal); $\mathrm{V}_{\mathrm{D}}$ is the diode forward voltage at the required load current.

## Diode selection

For maximum efficiency and performance, the rectifier (D1) should be a fast low capacitance Schottky diode with low reverse leakage at the maximum operating voltage and temperature. They also provide better efficiency than silicon diodes, due to a combination of lower forward voltage and reduced recovery time.

It is important to select parts with a peak current rating above the peak coil current and a continuous current rating higher than the maximum output load current. It is very important to consider the reverse leakage of the diode when operating above $85^{\circ} \mathrm{C}$. Excess leakage will increase the power dissipation in the device and if close to the load may create a thermal runaway condition.

The higher forward voltage and overshoot due to reverse recovery time in silicon diodes will increase the peak voltage on the LX output. If a silicon diode is used, care should be taken to ensure that the total voltage appearing on the LX pin including supply ripple, does not exceed the specified maximum value.

## Ordering Information



| Package Type | Number of pins |
| :---: | :---: |
| A: SOT23 | B: 5 |
| C: SOT89 |  |


| Part Number | Marking | Package Type | MOQ/ Packing |
| :---: | :---: | :---: | :---: |
| SD3361 ABR | EQXYW | SOT23-5 | 3,000 Units/ Tape \& Reel |
| SD3361 CBR | P2 $* * *$ <br> XXXYW | SOT89-5 | 1,000 Units/ Tape \& Reel |

## Outline Dimensions

SOT23-5


| REF. | Millimeter |  |  |
| :---: | :---: | :---: | :---: |
|  | Min | Nom | Max |
| A | 1.10 MAX |  |  |
| A1 | 0 | 0.05 | 0.10 |
| A2 | 0.70 | 1.00 | 1.295 |
| c | $0.12 R E F$ |  |  |
| D | 2.70 | 2.90 | 3.10 |
| E | 2.60 | 2.80 | 3.00 |
| E1 | 1.40 | 1.60 | 1.80 |
| L | $0.45 R E F$ |  |  |
| L1 | 0.60 REF. |  |  |
| $\theta$ | $0^{\circ}$ | $5^{\circ}$ | $10^{\circ}$ |
| b | 0.30 | 0.40 | 0.50 |
| e | $0.95 R E F$ |  |  |
| e1 | 1.90 REF. |  |  |

## Outline Dimensions



$0.395 \pm 0.045$
(Unit: mm)


Recommended Land Pattern

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