

### 1.5MHz, 1.3A Synchronous Step-Down Converter

### GENERAL DESCRIPTION

The SD6411 are high-efficiency, high frequency synchronous step-down DC-DC regulator ICs capable of delivering up to 1.3A output currents. The SD6411 can operate over a wide input voltage range from 2.3V to 6V and integrates main switch and synchronous switch with very low RDS(ON) to minimize the conduction loss.

It is ideal for powering portable equipment that runs from a single cell Lithium-Ion (Li+) battery. The output voltage can be regulated as low as 0.6V. The SD6411 can also run at 100% duty cycle for low dropout operation, extending battery life in portable system. This device offers two operation modes, PWM mode and PFM Mode switching control, which allows a high efficiency over the wider range of the load.

The SD6411 is offered in a low profile (1mm) 5-pin, thin SOT package, and is available in an adjustable version.

### **FEATURES**

- High Efficiency: Up to 96%
- 1.5MHz Constant Frequency Operation
- 1.3A Output Current
- No Schottky Diode Required
- 2.3V to 6V Input Voltage Range
- Adjustable Output Voltage Range Options from 0.6V to Vin
- 100% Duty Cycle Low Dropout Operation
- Low Quiescent Current: 35μA
- Slope Compensated Current Mode Control for Excellent Line and Load Transient Response
- Short Circuit Protection
- Thermal Fault Protection
- Inrush Current Limit and Soft Start
- <1µA Shutdown Current</li>
- Tiny SOT23-5 Package

### **APPLICATIONS**

- Cellular and Smart Phones
- Wireless and DSL Modems
- PDAs
- Digital Still and Video Cameras
- MP3 Players

## **Typical Application**

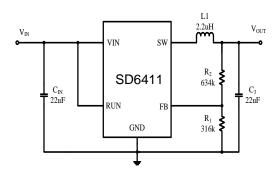
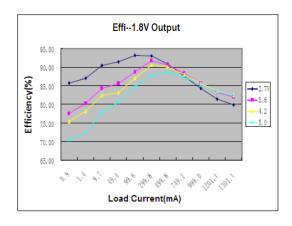


Figure 1. Typical Application Circuit

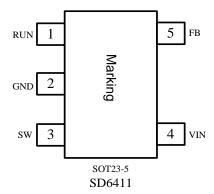




# Absolute Maximum Ratings (Note 1)

Input Supply Voltage0.3V to 6.5V	Operating Temperature Range40°C to +85°C
RUN,FB Voltages0.3V to 6.5V	Junction Temperature(Note2)125°C
SW Voltage0.3V to (Vin+0.3V)	Storage Temperature Range65°C to 150°C
Peak SW Sink and Source Current2.0A	Lead Temperature(Soldering,10s)+300°C

# Package/Order Information



Dort Number/Marking	SWICHING	Temperature	OUTPUT	OUTPUT
Part Number/Marking	FREQUENCY	Range	VOLTAGE (V)	CURRENT (A)
SD6411 /AS11D	1.5MHz	-40°C to +85°C	ADJ	1.3

# **Pin Description**

PIN	NAME	FUNCTION	
1	RUN	Chip Enable Pin. Drive RUN above 1.5V to turn on the part. Drive RUN below 0.3V to turn it off. Do not leave RUN floating.	
2	GND	Ground Pin	
3	SW	Power Switch Output. It is the switch node connection to Inductor. This pin connects to the drains of the internal P-ch and N-ch MOSFE switches.	
4	VIN	Power Supply Input. Must be closely decoupled to GND with a 10µF or greater ceramic capacitor.	
5	FB	FB Feedback Input. An internal resistive divider divides the output voltage down for comparison to the internal reference voltage.	

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## **Electrical Characteristics** (Note 3)

( $V_{IN}=V_{RUN}=3.6V$ ,  $V_{OUT}=1.8V$ ,  $T_A=25$ °C, unless otherwise noted.)

Parameter	Conditions	MIN	TYP	MAX	unit
Input Voltage Range		2.3		6	V
UVLO Threshold		1.7	1.9	2.1	V
Input DC Supply Current	(Note 4)				μΑ
PWM Mode	Vout = 90%, ILoad=0mA		140	300	μΑ
PFM Mode	Vout = 105%, ILoad=0mA		35	70	μΑ
Shutdown Mode	$V_{RUN} = 0V$ , $V_{IN}=4.2V$		0.1	1.0	μΑ
Dec lete I Fee II and	T <sub>A</sub> = 25°C	0.588	0.600	0.612	V
Regulated Feedback Voltage	$T_A = 0$ °C $\leq T_A \leq 85$ °C	0.586	0.600	0.613	V
Vollage	$T_A = -40$ °C $\leq T_A \leq 85$ °C	0.585	0.600	0.615	V
Reference Voltage Line Regulation	Vin=2.5V to 6V		0.04	0.40	%/V
Output Voltage Line Regulation	V <sub>IN</sub> = 2.5V to 6V		0.04	0.4	%
Output Voltage Load Regulation			0.5		%
Oscillation Frequency	Vout=100%		1.5		MHz
, ,	Vout=0V		300		KHz
On Resistance of PMOS	I <sub>SW</sub> =100mA		0.13	0.2	Ω
ON Resistance of NMOS	I <sub>SW</sub> =-100mA		0.1	0.2	Ω
Peak Current Limit	V <sub>IN</sub> = 3V, Vout=90%		2.0		Α
RUN Threshold		0.30	1.0	1.50	V
RUN Leakage Current			±0.01	±1.0	μΑ
SW Leakage Current	V <sub>RUN</sub> =0V,V <sub>IN</sub> =Vsw=5V		±0.01	±1.0	μΑ

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

**Note 2:**  $T_J$  is calculated from the ambient temperature  $T_A$  and power dissipation  $P_D$  according to the following formula:  $T_J = T_A + (P_D) \times (250^{\circ}C/W)$ .

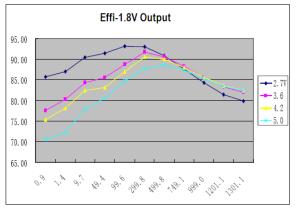
**Note3:** 100% production test at +25°C. Specifications over the temperature range are guaranteed by design and characterization.

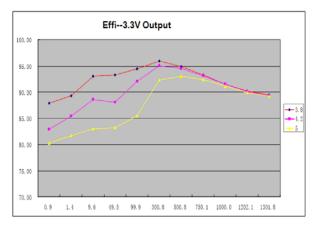
**Note 4:** Dynamic supply current is higher due to the gate charge being delivered at the switching frequency

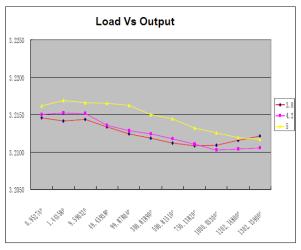
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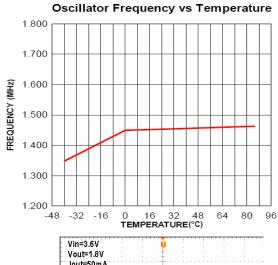


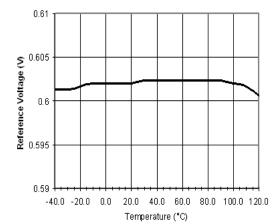
## **Typical Performance Characteristics**

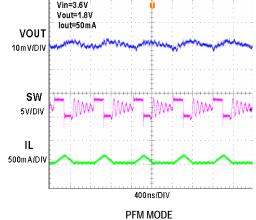














### **Functional Block Diagram**

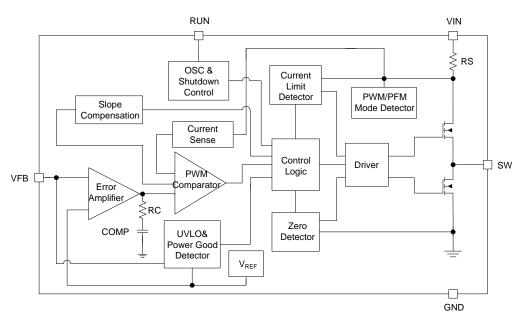


Figure 2. SD6411 Block Diagram

## **Functional Description**

SD6411 is a synchronous buck regulator IC that integrates the PWM/PFM control, high-side and low-side MOSFETs on the same die to minimize the switching transition loss and conduction loss. With ultra low RDS(ON) power switches and proprietary PWM control, this regulator IC can achieve the highest efficiency and the highest switch frequency simultaneously to minimize the external inductor and capacitor size, and thus achieving the minimum solution footprint.

The SD6411 requires only three external power components (Cin, Cout and L). The adjustable version can be programmed with external feedback to any voltage, ranging from 0.6V to the input voltage.

At dropout operation, the converter duty cycle increases to 100% and the output voltage tracks the input voltage minus the Rdson drop of the high-side MOSFET.

The internal error amplifier and compensation provides excellent transient response, load, and line regulation. Soft start function prevents input inrush current and output overshoot during start up.

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### **APPLICATIONS INFORMATION**

#### **Setting the Output Voltage**

The internal reference VREF is 0.6V(Typical). The output voltage is divided by a resistor, R1 and R2 to the FB pin. The output voltage is given by:

$$V_{OUT} = V_{REF} \times (1 + \frac{R_2}{R_1})$$

#### **Inductor Selection**

For most designs, the SD6411 operates with inductors of  $1\mu H$  to  $4.7\mu H$ . Low inductance values are physically smaller but require faster switching, which results in some efficiency loss. The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_{L} \times f_{OSC}}$$

Where  $\Delta I_L$  is inductor Ripple Current.

Large value inductors result in lower ripple current and small value inductors result in high ripple current. For optimum voltage-positioning load transients, choose an inductor with DC series resistance in the  $50m\Omega$  to  $150m\Omega$  range.

#### **Input Capacitor Selection**

With the maximum load current at 1.3A, the maximum ripple current through input capacitor is about 0.6Arms. A typical X7R or better grade ceramic capacitor with 6V rating and greater than 10uF capacitance can handle this ripple current well. To minimize the potential noise problem, place this ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by CIN, and IN/GND pins.

#### **Output Capacitor Selection**

The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current ratings. The output ripple  $\Delta V_{OUT}$  is determined by:

$$\Delta V_{OUT} \leq \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{OSC} \times L} \times \left(ESR + \frac{1}{8 \times f_{osc} \times C3}\right)$$

A 10µF ceramic Capacitor can satisfy most applications.

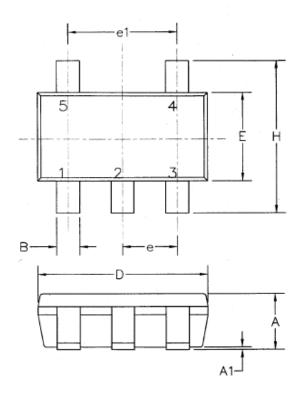
#### **PC Board Layout Checklist**

When laying out the printed circuit board, the following checking should be used to ensure proper operation of the SD6411. Check the following in your layout:

- The power traces, consisting of the GND trace, the SW trace and the VIN trace should be kept short, direct and wide.
- Does the (+) plates of Cin connect to Vin as closely as possible. This capacitor provides the AC current to the internal power MOSFETs.
- 3. Keep the switching node, SW, away from the sensitive VOUT node.
- Keep the (-) plates of Cin and Cout as close as possible.

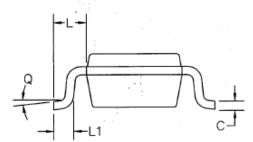


# **Package Description**



5LD SOT-23 PACKAGE OUTLINE DIMENSIONS ALL DIMENSIONS IN MM.

Dimension	Min.	Max.	
Α	0.9	1.10	
A1	0.01	0.13	
В	0.3	0.5	
C	0.09	0.2	
۵	2.8	3.0	
H	2.5	3.1	
Ε	1.5	1.7	
е	0.95 REF.		
e1	1.90 REF.		
L1	0.2	0.55	
L	0.35	0.8	
Q	0.	10°	



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