SD6420

## 2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter

## GENERAL DESCRIPTION

The SD6420 is a 1.5 MHz constant frequency, current mode step-down converter. It is ideal for portable equipment requiring very high current up to 2A from single-cell Lithium-ion batteries while still achieving over $90 \%$ efficiency during peak load conditions. The SD6420 also can run at 100\% duty cycle for low dropout operation, extending battery life in portable systems while light load operation provides very low output ripple for noise sensitive applications. The SD6420 can supply up to 2 A output load current from a 2.3 V to 6 V input voltage and the output voltage can be regulated as low as 0.6 V . The high switching frequency minimizes the size of external components while keeping switching losses low. The internal slope compensation setting allows the device to operate with smaller inductor values to optimize size and provide efficient operation. The SD6420 is offered in a low profile 6-pin, SOT package, and is available in an adjustable version. This device offers two operation modes, PWM control and PFM Mode switching control, which allows a high efficiency over the wider range of the load.

## feATURES

- High Efficiency: Up to $95 \%$
- 1.5 MHz Constant Frequency Operation
- 2A Output Current
- No Schottky Diode Required
- 2.4 V to 6 V Input Voltage Range
- Output Voltage as Low as 0.6 V
- PFM Mode for High Efficiency in Light Load
- 100\% Duty Cycle in Dropout Operation
- Low Quiescent Current: $40 \mu \mathrm{~A}$
- Short Circuit Protection
- Thermal Fault Protection
- Inrush Current Limit and Soft Start
- $\quad<1 \mu \mathrm{~A}$ Shutdown Current
- SOT23-6 package


## APPLICATIONS

- Cellular and Smart Phones
- Wireless and DSL Modems
- PDAs
- Portable Instruments
- Digital Still and Video Cameras
- PC Cards

TYPICAL APPLICATION


Figure 1. Basic Application Circuit


SD6420

## 2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter ABSOLUTE MAXIMUM RATINGS (Note 1)

Input Supply Voltage
EN,FB Voltages
SW Voltage. -0.3 V to 6.5 V
Operating Temperature Range........ $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$
-0.3 V to $\left(\mathrm{V}_{\text {IN }}+0.3 \mathrm{~V}\right)$
Power Dissipation
Thermal Resistance $\theta_{\mathrm{Jc}}$
Thermal Resistance $\theta_{\mathrm{JA}}$ $\qquad$ $250^{\circ} \mathrm{C} / \mathrm{W}$

Lead Temperature(Soldering,10s)
Storage Temperature Range $.300^{\circ} \mathrm{C}$ Storage Temperature Range........ $65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$ ESD HBM(Human Body Mode) .2kV

## PACKAGE/ORDER INFORMATION

|  | Order Part Number | Package | Top Marking |
| :---: | :---: | :---: | :---: |
|  | SD6420 | SOT23-6 | AS20BW |

## PIN DESCRIPTION

| Pin Name | Pin Number | Description |
| :---: | :---: | :--- |
| EN | 1 | Chip Enable Pin. Drive EN above 1.5V to turn on the part. Drive EN below <br> $0.3 V$ to turn it off. Do not leave EN floating. |
| GND | 2 | Analog ground pin. |

## SD6420

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## ELECTRICAL CHARACTGRISTICS (Note 3)

$\left(\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {EN }}=3.6 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=1.8 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right.$, unless otherwise noted.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Input Voltage Range |  | 2.3 |  | 6 | V |
| UVLO Threshold |  | 1.7 | 1.9 | 2.1 | V |
| Input DC Supply Current PWM Mode PFM Mode Shutdown Mode | (Note 4) $\begin{aligned} & \mathrm{V}_{\text {OUT }}=90 \%, \mathrm{I}_{\text {LOAD }}=0 \mathrm{~mA} \\ & \mathrm{~V}_{\text {OUT }}=105 \%, \mathrm{I}_{\text {LOAD }}=0 \mathrm{~mA} \\ & \mathrm{~V}_{\text {EN }}=0 \mathrm{~V}, \mathrm{~V}_{\text {IN }}=4.2 \mathrm{~V} \end{aligned}$ |  | $\begin{aligned} & 150 \\ & 40 \\ & 0.1 \end{aligned}$ | $\begin{gathered} 300 \\ 75 \\ 1.0 \end{gathered}$ | $\begin{aligned} & \mu \mathrm{A} \\ & \mu \mathrm{~A} \\ & \mu \mathrm{~A} \\ & \mu \mathrm{~A} \end{aligned}$ |
| Regulated Feedback Voltage $\mathrm{V}_{\mathrm{FB}}$ | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ | 0.588 | 0.600 | 0.612 | V |
|  | $\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C} \leqslant \mathrm{T}_{\mathrm{A}} \leqslant 85^{\circ} \mathrm{C}$ | 0.586 | 0.600 | 0.613 | V |
|  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \leqslant \mathrm{T}_{\mathrm{A}} \leqslant 85^{\circ} \mathrm{C}$ | 0.585 | 0.600 | 0.615 | V |
| Reference Voltage Line Regulation | $\mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$ to 5.5 V |  | 0.1 |  | \%/V |
| Output Voltage Accuracy | $\begin{aligned} & \mathrm{V}_{\mathrm{IN}}=2.5 \mathrm{~V} \text { to } 5.5 \mathrm{~V}, \\ & \mathrm{I}_{\text {OUT }}=10 \mathrm{~mA} \text { to } 2000 \mathrm{~mA} \end{aligned}$ | -3 |  | +3 | \% $\mathrm{V}_{\text {OUt }}$ |
| Output Voltage Load Regulation | $\mathrm{I}_{\text {Out }}=10 \mathrm{~mA}$ to 2000 mA |  | 0.2 |  | \%/A |
| Oscillation Frequency | $\mathrm{V}_{\text {OUT }}=100 \%$ |  | 1.5 |  | MHz |
|  | $\mathrm{V}_{\text {Out }}=0 \mathrm{~V}$ |  | 300 |  | kHz |
| On Resistance of PMOS | $\mathrm{I}_{\text {SW }}=100 \mathrm{~mA}$ |  | 100 | 150 | $\mathrm{m} \Omega$ |
| On Resistance of NMOS | $\mathrm{I}_{\text {sw }}=-100 \mathrm{~mA}$ |  | 80 | 150 | $\mathrm{m} \Omega$ |
| Peak Current Limit | $\mathrm{V}_{\text {IN }}=3 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=90 \%$ |  | 4 |  | A |
| EN Threshold |  | 0.30 | 1.0 | 1.50 | V |
| EN Leakage Current |  |  | $\pm 0.01$ | $\pm 1.0$ | $\mu \mathrm{A}$ |
| SW Leakage Current | $\mathrm{V}_{\text {EN }}=0 \mathrm{~V}, \mathrm{~V}_{1 N}=\mathrm{V}_{\text {SW }}=5 \mathrm{~V}$ |  | $\pm 0.01$ | $\pm 1.0$ | $\mu \mathrm{A}$ |

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.
Note 2: $T_{\mathrm{J}}$ is calculated from the ambient temperature $\mathrm{T}_{\mathrm{A}}$ and power dissipation $\mathrm{P}_{\mathrm{D}}$ according to the following formula: $T_{J}=T_{A}+\left(P_{D}\right) \times\left(250^{\circ} \mathrm{C} / \mathrm{W}\right)$.
Note 3: $100 \%$ production test at $+25^{\circ} \mathrm{C}$. Specifications over the temperature range are guaranteed by design and characterization.
Note 4: Dynamic supply current is higher due to the gate charge being delivered at the switching frequency.

SD6420
2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter TYPICAL PERFORMANCE CHARACTERISTICS


Efficiency vs. Load Current
$V_{\text {OUT }}=3.3 \mathrm{~V}$


Oscillator Frequency vs Temperature


PFM MODE


2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter


SD6420
2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter fUNCTIONAL BLOCK DIAGRAM


Figure 2. SD6420 Block Diagram

## 2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter

## FUNCTIONAL DESCRIPTION

The SD6420 is a high output current monolithic switch mode step-down DC-DC converter. The device operates at a fixed 1.5 MHz switching frequency, and uses a slope compensated current mode architecture. This step-down DC-DC converter can supply up to 2A output current at $\mathrm{V}_{\mathbb{N}}=3.6 \mathrm{~V}$ and has an input voltage range from 2.3 V to 6 V . It minimizes external component size and optimizes efficiency at the heavy load range. The slope compensation allows the device to remain stable over a wider range of inductor values so that smaller values ( $1 \mu \mathrm{H}$ to $4.7 \mu \mathrm{H}$ ) with lower DCR can be used to achieve higher efficiency. Only a small bypass input capacitor is required at the output. The adjustable output voltage can be programmed with external feedback to any voltage, ranging from 0.6 V to near the input voltage. It uses internal MOSFETs to achieve high efficiency and can generate very low output voltages by using an internal reference of 0.6 V . At dropout operation, the converter duty cycle increases to $100 \%$ and the output voltage tracks the input voltage minus the low $\mathrm{R}_{\mathrm{DS}(0 \mathrm{O})}$ drop of the P-channel high-side MOSFET and the inductor DCR. The internal error amplifier and compensation provides excellent transient response, Ioad and line regulation. Internal soft start eliminates any output voltage overshoot when the enable or the input voltage is applied.

## Setting the Output Voltage

Figure 1 shows the basic application circuit for the SD6420. The SD6420 can be externally programmed. Resistors R1 and R2 in Figure 1 program the output to regulate at a voltage higher than 0.6 V . To limit the bias current required for the external feedback resistor string while
maintaining good noise immunity, the minimum suggested value for R 2 is $59 \mathrm{k} \Omega$. Although a larger value will further reduce quiescent current, it will also increase the impedance of the feedback node, making it more sensitive to external noise and interference. Table 1 summarizes the resistor values for various output voltages with R2 set to either $59 \mathrm{k} \Omega$ for good noise immunity or $316 \mathrm{k} \Omega$ for reduced no load input current.
The external resistor sets the output voltage according to the following equation:

$$
\begin{aligned}
& \mathrm{V}_{\text {OUT }}=0.6 \times\left(1+\frac{\mathrm{R} 1}{\mathrm{R} 2}\right) \\
& \mathrm{R} 1=\left(\frac{\mathrm{V}_{\text {OUT }}}{0.6}-1\right) \times \mathrm{R} 2
\end{aligned}
$$

Table 1 shows the resistor selection for different output voltage settings.

| $V_{\text {out }}(\mathrm{V})$ | $R 2=59 \mathrm{k} \Omega$ <br> $\mathrm{R} 1(\mathrm{k} \Omega)$ | $\mathrm{R} 2=316 \mathrm{k} \Omega$ <br> $\mathrm{R} 1(\mathrm{k} \Omega)$ |
| :---: | :---: | :---: |
| 0.8 | 19.6 | 105 |
| 0.9 | 29.4 | 158 |
| 1.0 | 39.2 | 210 |
| 1.1 | 49.9 | 261 |
| 1.2 | 59.0 | 316 |
| 1.3 | 68.1 | 365 |
| 1.4 | 78.7 | 422 |
| 1.5 | 88.7 | 475 |
| 1.8 | 118 | 634 |
| 1.85 | 124 | 655 |
| 2.0 | 137 | 732 |
| 2.5 | 187 | 1000 |
| 3.3 | 267 | 1430 |

Table 1: Resistor selections for different output voltage settings (standard 1\% resistors substituted for calculated values).

## 2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter

## APPLICATIONS INFORMATION

## Inductor Selection

For most designs, the SD6420 operates with inductors of $1 \mu \mathrm{H}$ to $4.7 \mu \mathrm{H}$. Low inductance values are physically smaller but require faster switching, which results in some efficiency loss. The inductor value can be derived from the following equation:

$$
\mathrm{L}=\frac{\mathrm{V}_{\text {OUT }} \times\left(\mathrm{V}_{\text {IN }}-\mathrm{V}_{\text {OUT }}\right)}{\mathrm{V}_{\text {IN }} \times \Delta \mathrm{I}_{\mathrm{L}} \times \mathrm{f}_{\text {OSC }}}
$$

Where $\Delta I_{L}$ is inductor Ripple Current. Large value inductors result in lower ripple current and small value inductors result in high ripple current. For optimum voltage-positioning load transients, choose an inductor with DC series resistance in the $50 \mathrm{~m} \Omega$ to $150 \mathrm{~m} \Omega$ range.

| PART NUMBER | VALUE (uH) | DCR <br> $(\Omega$ MAX | MAX DC <br> CURRENT <br> $(A)$ | SIZE <br> $L * W * H\left(\mathrm{~mm}^{3}\right)$ |
| :---: | :---: | :---: | :---: | :---: |
| Sumida <br> CDRH5D16 | 2.2 | 28.7 | 3 | $5.8 \times 5.8 \times 1.8$ |
|  | 4.7 | 35.6 | 2.6 |  |
|  | 2.2 | 23 | 3.3 | $8.3 \times 8.3 \times 3.0$ |
|  | 3.3 | 29 | 2.6 |  |
|  | 4.7 | 39 | 2.1 |  |

Table2.Recommend Surface Mount Inductors

## Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency should be less than input source impedance to prevent high frequency switching current passing to the input. A low ESR input capacitor sized for maximum RMS current must be used. Ceramic capacitors with X 5 R or $\mathrm{X7R}$ dielectrics are highly recommended because of their low ESR and small temperature coefficients. A $22 \mu \mathrm{~F}$ ceramic capacitor for most applications is sufficient. A large value may be used for improved input voltage filtering.

## Output Capacitor Selection

The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current ratings. The output ripple $\Delta \mathrm{V}_{\text {out }}$ is determined by:
$\Delta V_{\text {OUT }} \leq \frac{V_{\text {out }} \times\left(V_{\text {W }}-V_{\text {OUT }}\right)}{V_{\text {IN }} \times f_{\text {OSC }} \times L} \times\left(E S R+\frac{1}{8 \times f_{\text {osC }} \times C_{\text {out }}}\right)$
A $22 \mu \mathrm{~F}$ ceramic can satisfy most applications.

## 2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter

## PCB Layout Recommendations

When laying out the printed circuit board, the following checking should be used to ensure proper operation of the SD6420. Check the following in your layout:
$>$ The power traces, consisting of the GND trace, the SW trace and the VIN trace should be kept short, direct and wide
$>$ Does the $(+)$ plates of $\mathrm{C}_{\text {N上 }}$ connect to VIN as closely as possible. This capacitor provides the $A C$ current to the internal power MOSFETs.
> Keep the switching node, SW, away from the sensitive $\mathrm{V}_{\text {out }}$ node.
$>$ Keep the (-) plates of $\mathrm{C}_{\mathbb{N}}$ and $\mathrm{C}_{\text {out }}$ as close as possible.


Figure 3. SD6420 Suggested Layout

SD6420
2A,2.4V-6V Input, 1.5MHz Synchronous Step-Down Converter PACKAGE DESCRIPTION

SOT23-6


NOTE:
1.DIMENSIONS ARE IN MILLIMETERS
2.DRAWING NOT TO SCALE.
3.DIMENSIONS ARE INCLUSIVE OF PLATING.
4.DIMENSIONS ARE EXCLUSIVE OF MOLD FLASH AND METAL BURR

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