

# **Application Note: SY8823**

### Ultra Low Quiescent Current, 2.0MHz, 3.5A Synchronous Step Down Regulator

### **General Description**

SY8823 is a high efficiency ultra low quiescent current, 2.0MHz synchronous step-down DC-DC regulator IC capable of delivering up to 3.5A output current.

SY8823 operates over a wide input voltage range from 2.5V to 5.5V and integrates main switch and synchronous switch with very low  $R_{DS\;(ON)}$  to minimize the conduction loss.

Temperature Code

Package Type

QFN2x1.5-8

Package Code Optional Spec Code

### **Ordering Information**

SY8823 □(□,□)□

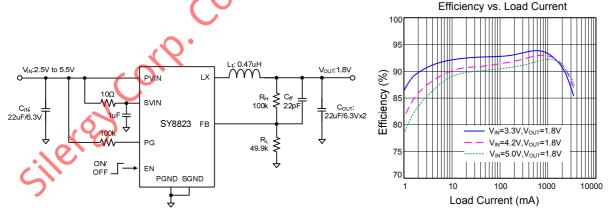
**Ordering Number** 

SY8823QUC

### Features

Input voltage range: 2.5V to 5.5V 2.0 MHz switching frequency 3.5A maximum output current Peak current mode control for the fast transient speed 100% drop out function Typical 18uA quiescent current Low R<sub>DS(ON)</sub> for internal switches (PFET/NFET):  $55m\Omega/35m\Omega$ Hic-cup mode protection for hard short condition RoHS Compliant and Halogen Free Compact package: QFN2x1.5-8 pplications Set Top Box Net PC Mini-Notebook PC Access Point Router





Note

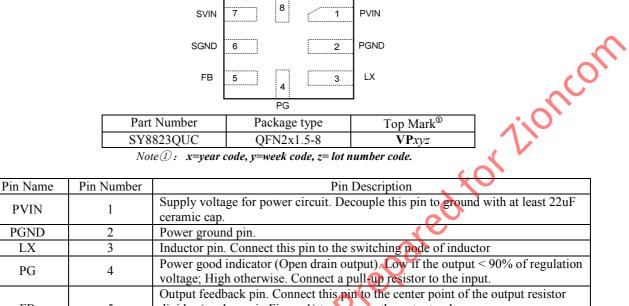
3.5A

#### Figure 1. Schematic Diagram

Figure 2. Efficiency Figure

## SY8823

**Pinout** (top view)



EN

FB	5	divider (as shown in Figure 1) to program the output voltage:
		$V_{OUT} = 0.6V*(1+R_H/R_L).$
SGND	6	Signal ground pin.
SVIN	7	Supply voltage for control circuit. Decouple this pin to ground with at least 1uF
5 111	/	ceramic cap.
EN	8	Enable input pin. Integrated $4M\Omega$ pull down resistor.

## Absolute Maximum Ratings (Note 1)

l pins 6V	All pins
l pins $6V$ wer Dissipation, PD @ TA = 25°C QFN2x1.5-81W	Power Dissipation,
ckage Thermal Resistance (Note 2)	Package Thermal F
θ JA 100°C/W	θ ја
θ JC 15°C/W	Ө лс
nction Temperature Range 150°C	Junction Temperat
ad Temperature (Soldering, 10 sec.) 260°C	
orage Temperature Range	Storage Temperatu
namic LX voltage in 10ns duration	

# Recommended Operating Conditions (Note 3)

Supply Input Voltage	2.5V to 5.5	5V
Ambient Temperature Range	e40°C to 85°	°C



### **Electrical Characteristics**

(VIN = 5V, VOUT = 2.5V, L = 0.47uH, COUT = 22uF, TA = 25°C, unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage Range	V <sub>IN</sub>		2.5		5.5	V
Quiescent Current	IQ	$I_{OUT}=0, V_{FB}=V_{REF} \cdot 105\%$		18		μΑ 🗸
Shutdown Current	I <sub>SHDN</sub>	EN=0		0.1	1	μΑ
Feedback Reference Voltage	$V_{REF}$		0.591	0.6	0.609	V
PFET RON	R <sub>DS(ON)</sub> ,P		30	55	80	mΩ
NFET RON	R <sub>DS(ON)</sub> ,N		15	35	55	mΩ
Peak Current Limit	I <sub>LIM</sub>		4.5		7.5	А
EN rising threshold	V <sub>ENH</sub>		1.5			V
EN falling threshold	V <sub>ENL</sub>			O'	0.4	V
Input UVLO threshold	V <sub>UVLO</sub>				2.5	V
UVLO hysteresis	$V_{HYS}$			0.15		V
Oscillator Frequency	F <sub>OSC</sub>	I <sub>OUT</sub> =500mA	.0.	2		MHz
PGOOD Under-voltage Threshold	V <sub>FB,LV</sub>			0.55		V
Short Circuit Protection Threshold	V <sub>SCP</sub>			0.26		V
Min ON Time				80		ns
Soft Start time	T <sub>SS</sub>			1		ms
Output Discharge Switch On Resistance	R <sub>DSC</sub>			50		Ω
Thermal Shutdown Temperature	T <sub>SD</sub>			150		°C
Thermal Shutdown Hysteresis	T <sub>HYS</sub>			15		°C

**Note 1**: Stresses beyond the "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 2:  $\theta$  JA is measured in the natural convection at T<sub>A</sub> = 25°C on a low effective single layer thermal conductivity test board of JEDEC 51-3 thermal measurement standard.

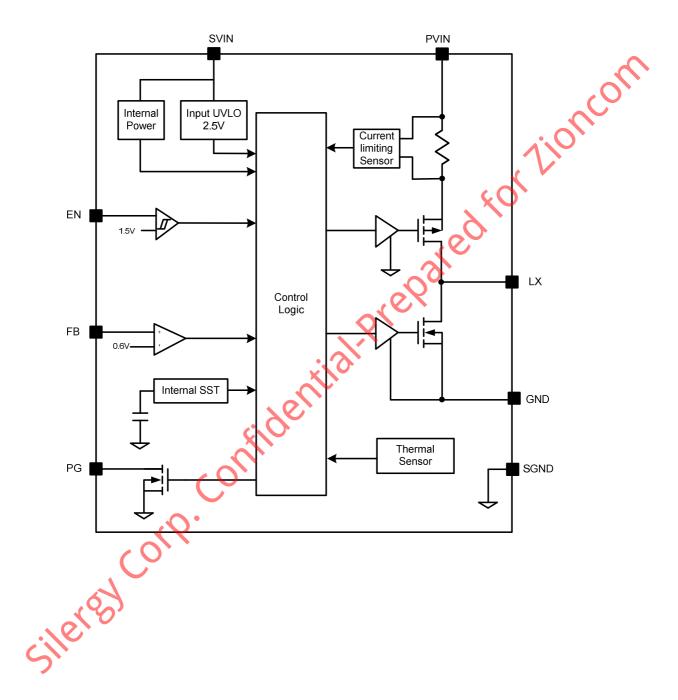
Note 3: The device is not guaranteed to function outside its operating conditions.



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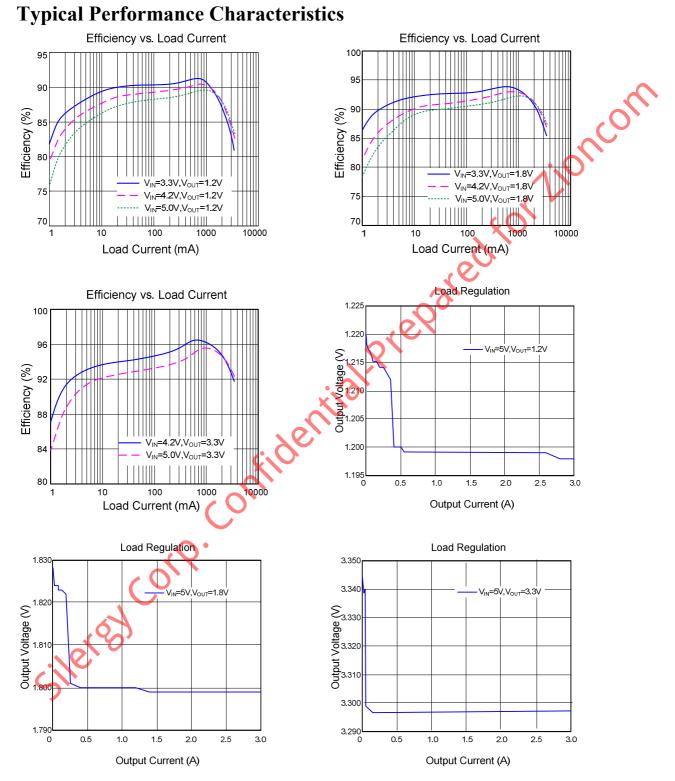


### **Block Diagram**



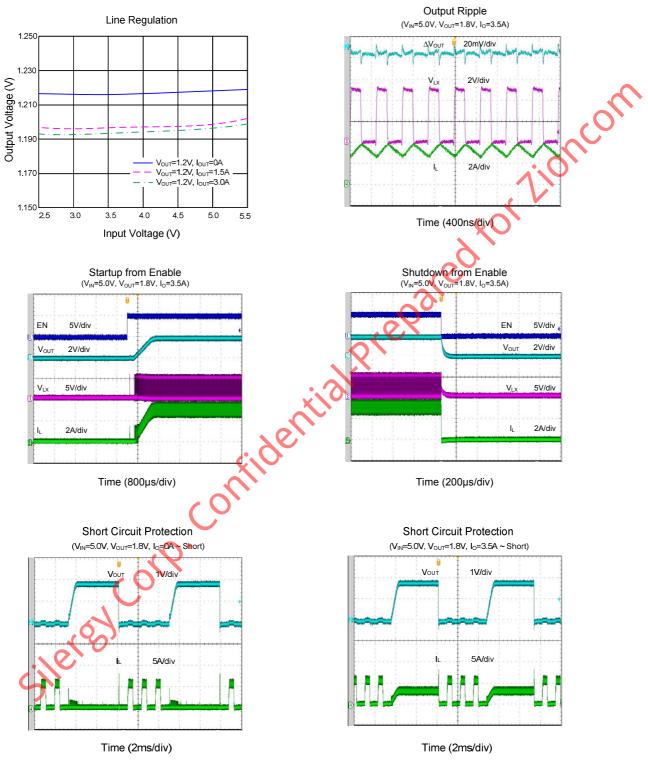
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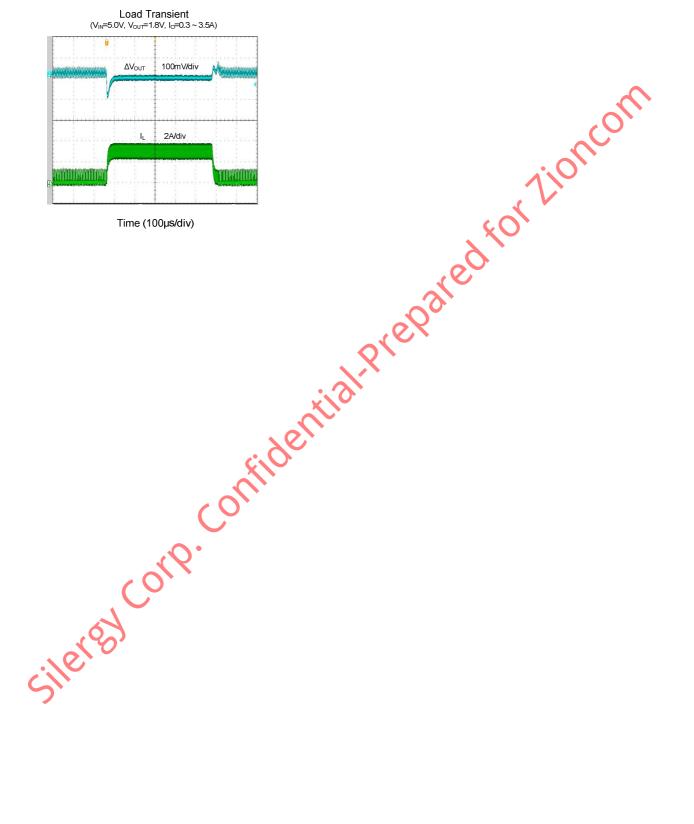




# SY8823







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### Operation

SY $\overline{8823}$  is an ultra low quiescent current synchronous buck regulator IC that integrates the PWM control, top and bottom switches on the same die to minimize the switching losses and conduction losses. With low  $R_{DS(ON)}$  power switches and proprietary PWM control, this regulator IC achieves a higher efficiency with high switching frequency to minimize the external inductor and capacitor size, and thus achieving the minimum solution footprint.

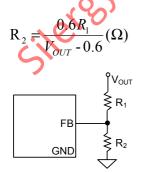
SY8823 senses the output voltage conditions for the fault protection. If the DC output voltage is about 3% over the regulation level, both switches turn off and remain in the off state. If the DC output voltage is below 42% of the regulation level, the IC will enter hic-up short protection mode. When the output voltage is below 42% of the regulation, the frequency is folded back to 25% of the normal frequency and the current limit is decreased to 55% of the normal current limit to prevent the inductor current runaway and to reduce the power dissipation within the IC under true short circuit conditions.

### **Applications Information**

Because of the high integration in SY8823, the application circuit based on this regulator IC is rather simple. Only input capacitor  $C_{IN}$ , output capacitor  $C_{OUT}$ , inductor L and feedback resistors ( $R_1$  and  $R_2$ ) need to be selected for the targeted applications.

#### Feedback resistor divider R1 and R2

Choose  $R_1$  and  $R_2$  to program the proper output voltage. To minimize the power consumption under light load, it is desirable to choose large resistance values for both  $R_1$  and  $R_2$ . A value between 10k and 1M is recommended for both resistors. If  $R_1$ =100k is chosen, then  $R_2$  can be calculated to be:



#### Input capacitor CIN

This ripple current through input capacitor is calculated as:

$$I_{CIN_{RMS}} = I_{OUT} \times \sqrt{D(1-D)}$$
 (A)

This formula has a maximum at  $V_{IN}=2 \times V_{OUT}$  condition, where  $I_{CIN RMS}=I_{OUT}/2$ .

With the maximum load current at 3.5A, a typical X5R or better grade ceramic capacitor with 6.3V rating and greater than 22uF capacitance can handle this ripple current well. To minimize the potential noise problem, place this ceramic capacitor really close to the PVIN and GND pins. Care should be taken to minimize the loop area formed by  $C_{IN}$ , PVIN and GND pins.

A  $1\mu$ F ceramic capacitor needs to be added across SVIN and GND.

#### Output capacitor Cour

Both steady state ripple and transient requirements must be taken into account when selecting this capacitor. For the best performance, it is recommended to use X5R or better grade ceramic capacitor with 6.3V rating and more than two pcs  $22\mu$ F capacitors.

#### **Output inductor L:**

There are several considerations in choosing this inductor.

1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum average input current. The inductance is calculated as:

$$L = \frac{V_{\text{OUT}}(1 - V_{\text{OUT}}/V_{\text{IN}_{\text{MAX}}})}{F_{\text{SW}} \times I_{\text{OUT} \text{ MAX}} \times 40\%} (H)$$

Where  $F_{SW}$  is the switching frequency and Iout\_MAX is the maximum load current.

SY8823 is less sensitive to the ripple current variations. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

2) The saturation current rating of an inductor must be selected to guarantee an adequate margin to the peak inductor current under full load conditions.





 $I_{\text{SAT, MIN}} > I_{\text{OUT, MAX}} + \frac{V_{\text{OUT}}(1 \text{-} V_{\text{OUT}}/V_{\text{IN, MAX}})}{2 \cdot F_{\text{SW}} \cdot L}$ 

3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with DCR<15m  $\Omega$  to achieve a good overall efficiency.

#### **Enable Operation**

Pulling the EN pin low (<0.4V) will shut down the device. During shutdown, the SY8823 shutdown current drops to lower than  $0.1\mu$ A, Driving the EN pin high (>1.5V) will turn on the IC again.

#### **Power Good Indication**

PG is an open-drain output pin. Connect an above 100k pull-up resistor to  $V_{IN}$ . PG pin will output high after the output voltage exceeds 90% of normal output voltage.

#### **Load Transient Considerations:**

SY8823 integrates the compensation components to achieve good stability and fast transient responses. Adding a  $10pF\sim22pF$  ceramic capacitor in parallel with R<sub>1</sub> may further speed up the load transient responses.

#### Layout Design:

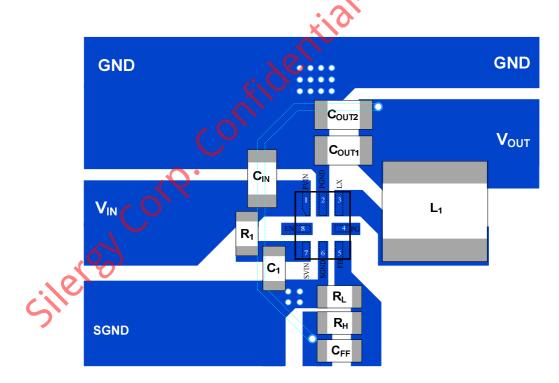
To achieve a higher efficiency and better noise immunity, following components should be placed close to the IC:  $C_{IN}$ , L,  $R_L$ ,  $R_H$ ,  $C_{FF}$ .

1) It is desirable to maximize the PCB copper area connecting to GND pin to achieve the best thermal and noise performance. Reasonable vias are suggested to be placed underneath the ground pad to enhance the soldering quality and thermal performance

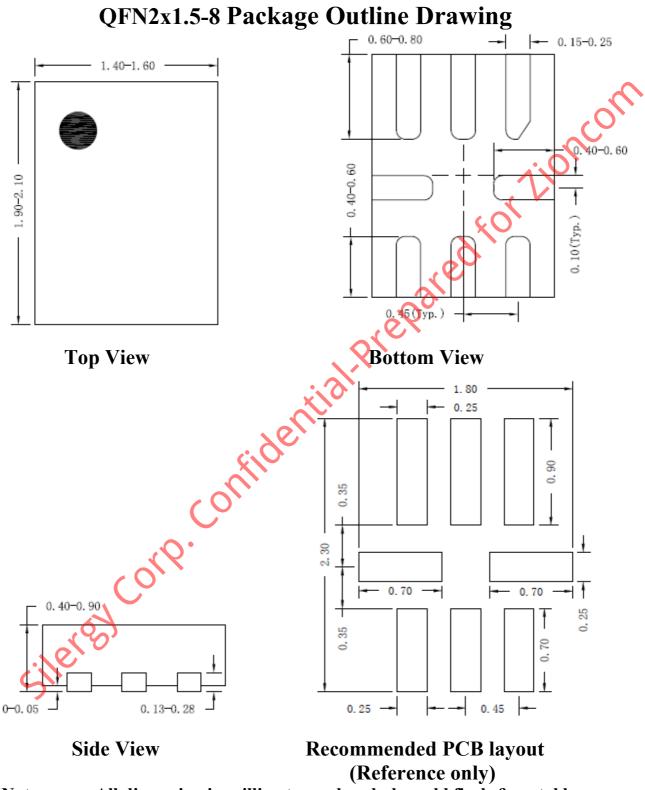
2) The decoupling capacitor of PVIN and SVIN must be placed as close as possible to the pins. The loop area formed by the capacitors and GND must be minimized.

3) The PCB copper area associated with LX pin must be minimized to improve the noise immunity.

4) The components  $R_L$ ,  $R_H$  and the trace connecting to the FB pin must NOT be adjacent to the LX node on the PCB layout to minimize the noise coupling to FB pin.





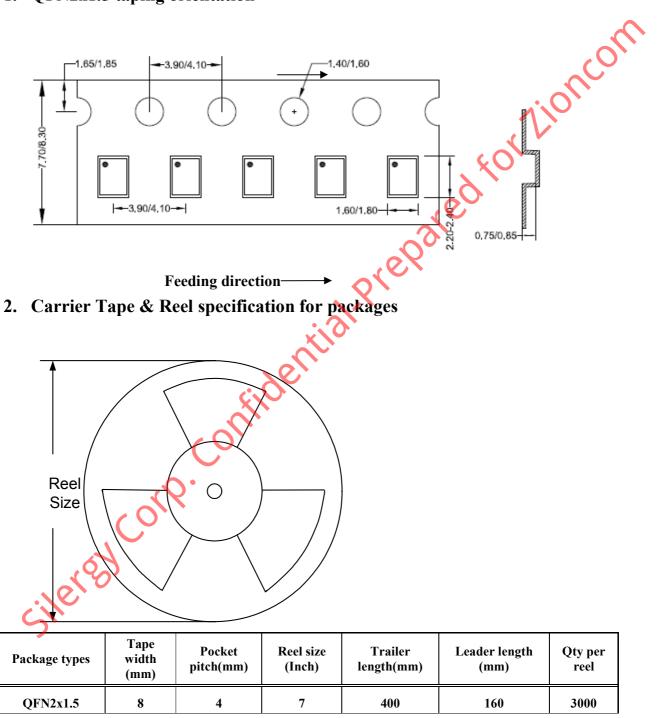


Notes: All dimension in millimeter and exclude mold flash & metal burr



## **Taping & Reel Specification**

### 1. QFN2x1.5 taping orientation



3. Others: NA

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