

EFR32FG13 Flex Gecko Proprietary Protocol SoC Family Data Sheet



The Flex Gecko proprietary protocol family of SoCs is part of the Wireless Gecko portfolio. Flex Gecko SoCs are ideal for enabling energy-friendly proprietary protocol networking for IoT devices.

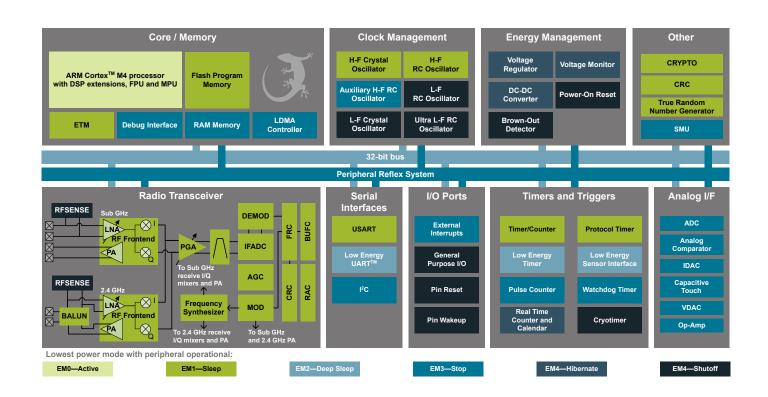
The single-die solution provides industry-leading energy efficiency, ultra-fast wakeup times, a scalable power amplifier, an integrated balun and no-compromise MCU features.

Flex Gecko applications include:

- · Home and Building Automation and Security
- Metering
- · Electronic Shelf Labels
- · Industrial Automation
- · Commercial and Retail Lighting and Sensing

KEY FEATURES

- 32-bit ARM® Cortex®-M4 core with 40 MHz maximum operating frequency
- · 512 kB of flash and 64 kB of RAM
- Pin-compatible across EFR32FG families (exceptions apply for 5V-tolerant pins)
- 12-channel Peripheral Reflex System enabling autonomous interaction of MCU peripherals
- Autonomous Hardware Crypto Accelerator and True Random Number Generator
- Integrated PA with up to 19 dBm (2.4 GHz) or 20 dBm (Sub-GHz) tx power
- · Integrated balun for 2.4 GHz
- · Robust peripheral set and up to 32 GPIO



1. Feature List

The EFR32FG13 highlighted features are listed below.

· Low Power Wireless System-on-Chip

- High Performance 32-bit 40 MHz ARM Cortex®-M4 with DSP instruction and floating-point unit for efficient signal processing
- Embedded Trace Macrocell (ETM) for advanced debugging
- 512 kB flash program memory
- · 64 kB RAM data memory
- · 2.4 GHz and Sub-GHz radio operation
- · Transmit power:
 - · 2.4 GHz radio: Up to 19 dBm
 - · Sub-GHz radio: Up to 20 dBm

· Low Energy Consumption

- 8.4 mA RX current at 38.4 kbps, GFSK, 169 MHz
- 9.5 mA RX current at 1 Mbps, GFSK, 2.4 GHz
- · 10.3 mA RX current at 250 kbps, DSSS-OQPSK, 2.4 GHz
- 8.5 mA TX current at 0 dBm output power at 2.4 GHz
- 35.3 mA TX current at 14 dBm output power at 868 MHz
- 69 µA/MHz in Active Mode (EM0)
- 1.3 µA EM2 DeepSleep current (16 kB RAM retention and RTCC running from LFRCO)
- Wake on Radio with signal strength detection, preamble pattern detection, frame detection and timeout

· High Receiver Performance

- · -94.8 dBm sensitivity at 1 Mbit/s GFSK, 2.4 GHz
- -102.7 dBm sensitivity at 250 kbps DSSS-OQPSK, 2.4 GHz
- -126.2 dBm sensitivity at 600 bps, GFSK, 915 MHz
- -120.6 dBm sensitivity at 2.4 kbps, GFSK, 868 MHz
- · -107.4 dBm sensitivity at 4.8 kbps, OOK, 433 MHz
- · -112.2 dBm sensitivity at 38.4 kbps, GFSK, 169 MHz

· Supported Modulation Formats

- · 2/4 (G)FSK with fully configurable shaping
- BPSK / DBPSK TX
- OOK / ASK
- · Shaped OQPSK / (G)MSK
- · Configurable DSSS and FEC

· Supported Protocols

- Proprietary Protocols
- · Wireless M-Bus
- Selected IEEE 802.15.4g SUN-FSK PHYs
- · Low Power Wide Area Networks

Suitable for Systems Targeting Compliance With:

- FCC Part 90.210 Mask D, FCC part 15.247, 15.231, 15.249
- ETSI Category I Operation, EN 300 220, EN 300 328
- ARIB T-108, T-96
- China regulatory

· Wide selection of MCU peripherals

- 12-bit 1 Msps SAR Analog to Digital Converter (ADC)
 - 2 × Analog Comparator (ACMP)
 - 2 × Digital to Analog Converter (VDAC)
 - 3 × Operational Amplifier (Opamp)
 - · Digital to Analog Current Converter (IDAC)
 - Low-Energy Sensor Interface (LESENSE)
 - Multi-channel Capacitive Sense Interface (CSEN)
 - Up to 32 pins connected to analog channels (APORT) shared between analog peripherals
 - Up to 32 General Purpose I/O pins with output state retention and asynchronous interrupts
 - · 8 Channel DMA Controller
 - 12 Channel Peripheral Reflex System (PRS)
 - 2 × 16-bit Timer/Counter
 - · 3 or 4 Compare/Capture/PWM channels
 - 1 × 32-bit Timer/Counter
 - · 3 Compare/Capture/PWM channels
 - 32-bit Real Time Counter and Calendar
 - 16-bit Low Energy Timer for waveform generation
 - 32-bit Ultra Low Energy Timer/Counter for periodic wake-up from any Energy Mode
 - · 16-bit Pulse Counter with asynchronous operation
 - · 2 × Watchdog Timer with dedicated RC oscillator
 - 3 × Universal Synchronous/Asynchronous Receiver/ Transmitter (UART/SPI/SmartCard (ISO 7816)/IrDA/I²S)
 - Low Energy UART (LEUART[™])
 - 2 × I²C interface with SMBus support and address recognition in EM3 Stop

· Wide Operating Range

- 1.8 V to 3.8 V single power supply
- Integrated DC-DC, down to 1.8 V output with up to 200 mA load current for system
- Standard (-40 °C to 85 °C) and Extended (-40 °C to 125 °C) temperature grades available

Support for Internet Security

- · General Purpose CRC
- · True Random Number Generator
- 2 × Hardware Cryptographic Acceleration for AES 128/256, SHA-1, SHA-2 (SHA-224 and SHA-256) and ECC
- QFN32 5x5 mm Package
- QFN48 7x7 mm Package

2. Ordering Information

Table 2.1. Ordering Information

Ordering Code	Protocol Stack	Frequency Band @ Max TX Power	Flash (kB)	RAM (kB)	GPIO	Package	Temp Range
EFR32FG13P233F512GM48-C	Proprietary	2.4 GHz @ 19 dBm	512	64	28	QFN48	-40 to +85°C
		Sub-GHz @ 20 dBm					
EFR32FG13P232F512GM48-C	Proprietary	2.4 GHz @ 19 dBm	512	64	31	QFN48	-40 to +85°C
EFR32FG13P232F512GM32-C	Proprietary	2.4 GHz @ 19 dBm	512	64	16	QFN32	-40 to +85°C
EFR32FG13P231F512GM48-C	Proprietary	Sub-GHz @ 20 dBm	512	64	32	QFN48	-40 to +85°C
EFR32FG13P231F512IM48-C	Proprietary	Sub-GHz @ 20 dBm	512	64	32	QFN48	-40 to +125°C
EFR32FG13P231F512GM32-C	Proprietary	Sub-GHz @ 20 dBm	512	64	16	QFN32	-40 to +85°C
EFR32FG13P231F512IM32-C	Proprietary	Sub-GHz @ 20 dBm	512	64	16	QFN32	-40 to +125°C

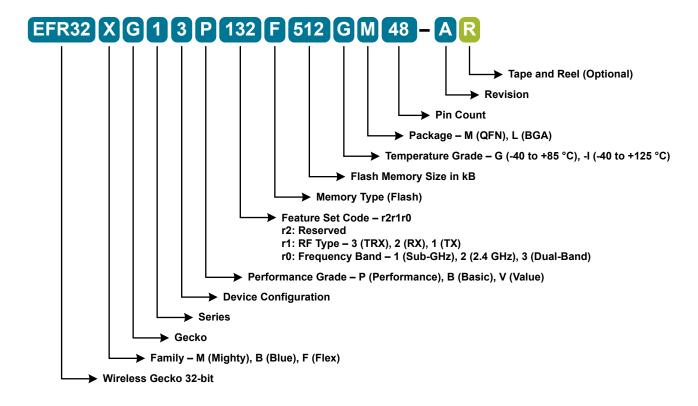


Figure 2.1. Ordering Code Key

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3. System Overview

3.1 Introduction

The EFR32 product family combines an energy-friendly MCU with a highly integrated radio transceiver. The devices are well suited for any battery operated application as well as other systems requiring high performance and low energy consumption. This section gives a short introduction to the full radio and MCU system. The detailed functional description can be found in the EFR32xG13 Reference Manual.

A block diagram of the EFR32FG13 family is shown in Figure 3.1 Detailed EFR32FG13 Block Diagram on page 7. The diagram shows a superset of features available on the family, which vary by OPN. For more information about specific device features, consult Ordering Information.

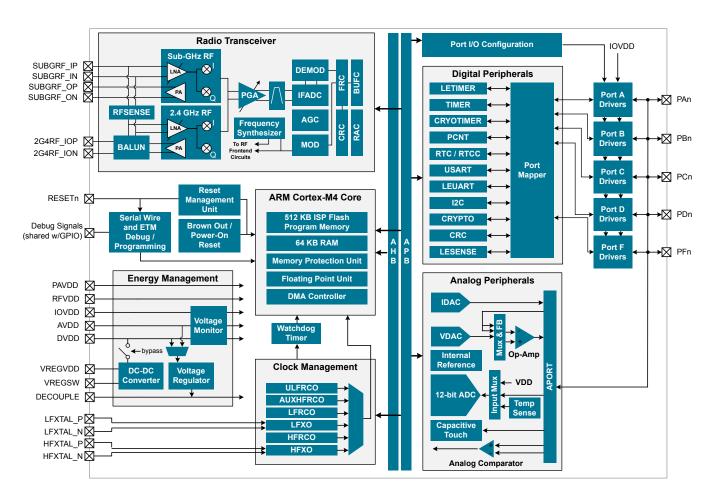


Figure 3.1. Detailed EFR32FG13 Block Diagram

3.2 Radio

The Flex Gecko family features a radio transceiver supporting proprietary wireless protocols.

3.2.1 Antenna Interface

The EFR32FG13 family includes devices which support both single-band and dual-band RF communication over separate physical RF interfaces.

The 2.4 GHz antenna interface consists of two pins (2G4RF_IOP and 2G4RF_ION) that interface directly to the on-chip BALUN. The 2G4RF_ION pin should be grounded externally.

The sub-GHz antenna interface consists of a differential transmit interface (pins SUBGRF_OP and SUBGRF_ON) and a differential receive interface (pinsSUBGRF_IP and SUBGRF_IN).

The external components and power supply connections for the antenna interface typical applications are shown in the RF Matching Networks section.

3.2.2 Fractional-N Frequency Synthesizer

The EFR32FG13 contains a high performance, low phase noise, fully integrated fractional-N frequency synthesizer. The synthesizer is used in receive mode to generate the LO frequency used by the down-conversion mixer. It is also used in transmit mode to directly generate the modulated RF carrier.

The fractional-N architecture provides excellent phase noise performance combined with frequency resolution better than 100 Hz, with low energy consumption. The synthesizer has fast frequency settling which allows very short receiver and transmitter wake up times to optimize system energy consumption.

3.2.3 Receiver Architecture

The EFR32FG13 uses a low-IF receiver architecture, consisting of a Low-Noise Amplifier (LNA) followed by an I/Q down-conversion mixer, employing a crystal reference. The I/Q signals are further filtered and amplified before being sampled by the IF analog-to-digital converter (IFADC).

The IF frequency is configurable from 150 kHz to 1371 kHz. The IF can further be configured for high-side or low-side injection, providing flexibility with respect to known interferers at the image frequency.

The Automatic Gain Control (AGC) block adjusts the receiver gain to optimize performance and avoid saturation for excellent selectivity and blocking performance. The 2.4 GHz radio is calibrated at production to improve image rejection performance. The sub-GHz radio can be calibrated on-demand by the user for the desired frequency band.

Demodulation is performed in the digital domain. The demodulator performs configurable decimation and channel filtering to allow receive bandwidths ranging from 0.1 to 2530 kHz. High carrier frequency and baud rate offsets are tolerated by active estimation and compensation. Advanced features supporting high quality communication under adverse conditions include forward error correction by block and convolutional coding as well as Direct Sequence Spread Spectrum (DSSS) for 2.4 GHz and sub-GHz bands.

A Received Signal Strength Indicator (RSSI) is available for signal quality metrics, for level-based proximity detection, and for RF channel access by Collision Avoidance (CA) or Listen Before Talk (LBT) algorithms. An RSSI capture value is associated with each received frame and the dynamic RSSI measurement can be monitored throughout reception.

The EFR32FG13 features integrated support for antenna diversity to mitigate the problem of frequency-selective fading due to multipath propagation and improve link budget. Support for antenna diversity is available for specific PHY configurations in 2.4 GHz and sub-GHz bands. Internal configurable hardware controls an external switch for automatic switching between antennae during RF receive detection operations.

Note: Due to the shorter preamble of 802.15.4 and BLE packets, RX diversity is not supported.

3.2.4 Transmitter Architecture

The EFR32FG13 uses a direct-conversion transmitter architecture. For constant envelope modulation formats, the modulator controls phase and frequency modulation in the frequency synthesizer. Transmit symbols or chips are optionally shaped by a digital shaping filter. The shaping filter is fully configurable, including the BT product, and can be used to implement Gaussian or Raised Cosine shaping.

Carrier Sense Multiple Access - Collision Avoidance (CSMA-CA) or Listen Before Talk (LBT) algorithms can be automatically timed by the EFR32FG13. These algorithms are typically defined by regulatory standards to improve inter-operability in a given bandwidth between devices that otherwise lack synchronized RF channel access.

3.2.5 Wake on Radio

The Wake on Radio feature allows flexible, autonomous RF sensing, qualification, and demodulation without required MCU activity, using a subsystem of the EFR32FG13 including the Radio Controller (RAC), Peripheral Reflex System (PRS), and Low Energy peripherals.

3.2.6 RFSENSE

The RFSENSE peripheral generates a system wakeup interrupt upon detection of wideband RF energy at the antenna interface, providing true RF wakeup capabilities from low energy modes including EM2, EM3 and EM4.

RFSENSE triggers on a relatively strong RF signal and is available in the lowest energy modes, allowing exceptionally low energy consumption. RFSENSE does not demodulate or otherwise qualify the received signal, but software may respond to the wakeup event by enabling normal RF reception.

Various strategies for optimizing power consumption and system response time in presence of false alarms may be employed using available timer peripherals.

3.2.7 Flexible Frame Handling

EFR32FG13 has an extensive and flexible frame handling support for easy implementation of even complex communication protocols. The Frame Controller (FRC) supports all low level and timing critical tasks together with the Radio Controller and Modulator/Demodulator:

- · Highly adjustable preamble length
- Up to 2 simultaneous synchronization words, each up to 32 bits and providing separate interrupts
- · Frame disassembly and address matching (filtering) to accept or reject frames
- · Automatic ACK frame assembly and transmission
- · Fully flexible CRC generation and verification:
 - · Multiple CRC values can be embedded in a single frame
 - 8, 16, 24 or 32-bit CRC value
 - · Configurable CRC bit and byte ordering
- · Selectable bit-ordering (least significant or most significant bit first)
- · Optional data whitening
- Optional Forward Error Correction (FEC), including convolutional encoding / decoding and block encoding / decoding
- Half rate convolutional encoder and decoder with constraint lengths from 2 to 7 and optional puncturing
- Optional symbol interleaving, typically used in combination with FEC
- · Symbol coding, such as Manchester or DSSS, or biphase space encoding using FEC hardware
- · UART encoding over air, with start and stop bit insertion / removal
- · Test mode support, such as modulated or unmodulated carrier output
- · Received frame timestamping

3.2.8 Packet and State Trace

The EFR32FG13 Frame Controller has a packet and state trace unit that provides valuable information during the development phase. It features:

- · Non-intrusive trace of transmit data, receive data and state information
- Data observability on a single-pin UART data output, or on a two-pin SPI data output
- · Configurable data output bitrate / baudrate
- · Multiplexed transmitted data, received data and state / meta information in a single serial data stream

3.2.9 Data Buffering

The EFR32FG13 features an advanced Radio Buffer Controller (BUFC) capable of handling up to 4 buffers of adjustable size from 64 bytes to 4096 bytes. Each buffer can be used for RX, TX or both. The buffer data is located in RAM, enabling zero-copy operations.

3.2.10 Radio Controller (RAC)

The Radio Controller controls the top level state of the radio subsystem in the EFR32FG13. It performs the following tasks:

- · Precisely-timed control of enabling and disabling of the receiver and transmitter circuitry
- · Run-time calibration of receiver, transmitter and frequency synthesizer
- · Detailed frame transmission timing, including optional LBT or CSMA-CA

3.2.11 Random Number Generator

The Frame Controller (FRC) implements a random number generator that uses entropy gathered from noise in the RF receive chain. The data is suitable for use in cryptographic applications.

Output from the random number generator can be used either directly or as a seed or entropy source for software-based random number generator algorithms such as Fortuna.

3.3 Power

The EFR32FG13 has an Energy Management Unit (EMU) and efficient integrated regulators to generate internal supply voltages. Only a single external supply voltage is required, from which all internal voltages are created. An optional integrated DC-DC buck regulator can be utilized to further reduce the current consumption. The DC-DC regulator requires one external inductor and one external capacitor.

The EFR32FG13 device family includes support for internal supply voltage scaling, as well as two different power domains groups for peripherals. These enhancements allow for further supply current reductions and lower overall power consumption.

AVDD and VREGVDD need to be 1.8 V or higher for the MCU to operate across all conditions; however the rest of the system will operate down to 1.62 V, including the digital supply and I/O. This means that the device is fully compatible with 1.8 V components. Running from a sufficiently high supply, the device can use the DC-DC to regulate voltage not only for itself, but also for other PCB components, supplying up to a total of 200 mA.

3.3.1 Energy Management Unit (EMU)

The Energy Management Unit manages transitions of energy modes in the device. Each energy mode defines which peripherals and features are available and the amount of current the device consumes. The EMU can also be used to turn off the power to unused RAM blocks, and it contains control registers for the DC-DC regulator and the Voltage Monitor (VMON). The VMON is used to monitor multiple supply voltages. It has multiple channels which can be programmed individually by the user to determine if a sensed supply has fallen below a chosen threshold.

3.3.2 DC-DC Converter

The DC-DC buck converter covers a wide range of load currents and provides up to 90% efficiency in energy modes EM0, EM1, EM2 and EM3, and can supply up to 200 mA to the device and surrounding PCB components. Patented RF noise mitigation allows operation of the DC-DC converter without degrading sensitivity of radio components. Protection features include programmable current limiting, short-circuit protection, and dead-time protection. The DC-DC converter may also enter bypass mode when the input voltage is too low for efficient operation. In bypass mode, the DC-DC input supply is internally connected directly to its output through a low resistance switch. Bypass mode also supports in-rush current limiting to prevent input supply voltage droops due to excessive output current transients.

3.3.3 Power Domains

The EFR32FG13 has two peripheral power domains for operation in EM2 and EM3. If all of the peripherals in a peripheral power domain are configured as unused, the power domain for that group will be powered off in the low-power mode, reducing the overall current consumption of the device.

Peripheral Power Domain 2

ACMP0
ACMP1
PCNT0
CSEN
ADC0
VDAC0
LETIMER0
LESENSE
I2C0
APORT
IDAC

Table 3.1. Peripheral Power Subdomains

3.4 General Purpose Input/Output (GPIO)

EFR32FG13 has up to 32 General Purpose Input/Output pins. Each GPIO pin can be individually configured as either an output or input. More advanced configurations including open-drain, open-source, and glitch-filtering can be configured for each individual GPIO pin. The GPIO pins can be overridden by peripheral connections, like SPI communication. Each peripheral connection can be routed to several GPIO pins on the device. The input value of a GPIO pin can be routed through the Peripheral Reflex System to other peripherals. The GPIO subsystem supports asynchronous external pin interrupts.

3.5 Clocking

3.5.1 Clock Management Unit (CMU)

The Clock Management Unit controls oscillators and clocks in the EFR32FG13. Individual enabling and disabling of clocks to all peripherals is performed by the CMU. The CMU also controls enabling and configuration of the oscillators. A high degree of flexibility allows software to optimize energy consumption in any specific application by minimizing power dissipation in unused peripherals and oscillators.

3.5.2 Internal and External Oscillators

The EFR32FG13 supports two crystal oscillators and fully integrates four RC oscillators, listed below.

- A high frequency crystal oscillator (HFXO) with integrated load capacitors, tunable in small steps, provides a precise timing reference for the MCU. Crystal frequencies in the range from 38 to 40 MHz are supported. An external clock source such as a TCXO can also be applied to the HFXO input for improved accuracy over temperature.
- A 32.768 kHz crystal oscillator (LFXO) provides an accurate timing reference for low energy modes.
- An integrated high frequency RC oscillator (HFRCO) is available for the MCU system, when crystal accuracy is not required. The HFRCO employs fast startup at minimal energy consumption combined with a wide frequency range.
- An integrated auxilliary high frequency RC oscillator (AUXHFRCO) is available for timing the general-purpose ADC and the Serial Wire Viewer port with a wide frequency range.
- An integrated low frequency 32.768 kHz RC oscillator (LFRCO) can be used as a timing reference in low energy modes, when crystal accuracy is not required.
- An integrated ultra-low frequency 1 kHz RC oscillator (ULFRCO) is available to provide a timing reference at the lowest energy consumption in low energy modes.

3.6 Counters/Timers and PWM

3.6.1 Timer/Counter (TIMER)

TIMER peripherals keep track of timing, count events, generate PWM outputs and trigger timed actions in other peripherals through the PRS system. The core of each TIMER is a 16-bit counter with up to 4 compare/capture channels. Each channel is configurable in one of three modes. In capture mode, the counter state is stored in a buffer at a selected input event. In compare mode, the channel output reflects the comparison of the counter to a programmed threshold value. In PWM mode, the TIMER supports generation of pulse-width modulation (PWM) outputs of arbitrary waveforms defined by the sequence of values written to the compare registers, with optional dead-time insertion available in timer unit TIMER_0 only.

3.6.2 Wide Timer/Counter (WTIMER)

WTIMER peripherals function just as TIMER peripherals, but are 32 bits wide. They keep track of timing, count events, generate PWM outputs and trigger timed actions in other peripherals through the PRS system. The core of each WTIMER is a 32-bit counter with up to 4 compare/capture channels. Each channel is configurable in one of three modes. In capture mode, the counter state is stored in a buffer at a selected input event. In compare mode, the channel output reflects the comparison of the counter to a programmed threshold value. In PWM mode, the WTIMER supports generation of pulse-width modulation (PWM) outputs of arbitrary waveforms defined by the sequence of values written to the compare registers, with optional dead-time insertion available in timer unit WTIMER 0 only.

3.6.3 Real Time Counter and Calendar (RTCC)

The Real Time Counter and Calendar (RTCC) is a 32-bit counter providing timekeeping in all energy modes. The RTCC includes a Binary Coded Decimal (BCD) calendar mode for easy time and date keeping. The RTCC can be clocked by any of the on-board oscillators with the exception of the AUXHFRCO, and it is capable of providing system wake-up at user defined instances. When receiving frames, the RTCC value can be used for timestamping. The RTCC includes 128 bytes of general purpose data retention, allowing easy and convenient data storage in all energy modes down to EM4H.

A secondary RTC is used by the RF protocol stack for event scheduling, leaving the primary RTCC block available exclusively for application software.

3.6.4 Low Energy Timer (LETIMER)

The unique LETIMER is a 16-bit timer that is available in energy mode EM2 Deep Sleep in addition to EM1 Sleep and EM0 Active. This allows it to be used for timing and output generation when most of the device is powered down, allowing simple tasks to be performed while the power consumption of the system is kept at an absolute minimum. The LETIMER can be used to output a variety of waveforms with minimal software intervention. The LETIMER is connected to the Real Time Counter and Calendar (RTCC), and can be configured to start counting on compare matches from the RTCC.

3.6.5 Ultra Low Power Wake-up Timer (CRYOTIMER)

The CRYOTIMER is a 32-bit counter that is capable of running in all energy modes. It can be clocked by either the 32.768 kHz crystal oscillator (LFXO), the 32.768 kHz RC oscillator (LFRCO), or the 1 kHz RC oscillator (ULFRCO). It can provide periodic Wakeup events and PRS signals which can be used to wake up peripherals from any energy mode. The CRYOTIMER provides a wide range of interrupt periods, facilitating flexible ultra-low energy operation.

3.6.6 Pulse Counter (PCNT)

The Pulse Counter (PCNT) peripheral can be used for counting pulses on a single input or to decode quadrature encoded inputs. The clock for PCNT is selectable from either an external source on pin PCTNn_S0IN or from an internal timing reference, selectable from among any of the internal oscillators, except the AUXHFRCO. The peripheral may operate in energy mode EM0 Active, EM1 Sleep, EM2 Deep Sleep, and EM3 Stop.

3.6.7 Watchdog Timer (WDOG)

The watchdog timer can act both as an independent watchdog or as a watchdog synchronous with the CPU clock. It has windowed monitoring capabilities, and can generate a reset or different interrupts depending on the failure mode of the system. The watchdog can also monitor autonomous systems driven by PRS.

3.7 Communications and Other Digital Peripherals

3.7.1 Universal Synchronous/Asynchronous Receiver/Transmitter (USART)

The Universal Synchronous/Asynchronous Receiver/Transmitter is a flexible serial I/O interface. It supports full duplex asynchronous UART communication with hardware flow control as well as RS-485, SPI, MicroWire and 3-wire. It can also interface with devices supporting:

- ISO7816 SmartCards
- IrDA
- I²S

3.7.2 Low Energy Universal Asynchronous Receiver/Transmitter (LEUART)

The unique LEUARTTM provides two-way UART communication on a strict power budget. Only a 32.768 kHz clock is needed to allow UART communication up to 9600 baud. The LEUART includes all necessary hardware to make asynchronous serial communication possible with a minimum of software intervention and energy consumption.

3.7.3 Inter-Integrated Circuit Interface (I²C)

The I²C interface enables communication between the MCU and a serial I²C bus. It is capable of acting as both a master and a slave and supports multi-master buses. Standard-mode, fast-mode and fast-mode plus speeds are supported, allowing transmission rates from 10 kbit/s up to 1 Mbit/s. Slave arbitration and timeouts are also available, allowing implementation of an SMBus-compliant system. The interface provided to software by the I²C peripheral allows precise timing control of the transmission process and highly automated transfers. Automatic recognition of slave addresses is provided in active and low energy modes.

3.7.4 Peripheral Reflex System (PRS)

The Peripheral Reflex System provides a communication network between different peripherals without software involvement. Peripherals producing Reflex signals are called producers. The PRS routes Reflex signals from producers to consumer peripherals, which in turn perform actions in response. Edge triggers and other functionality such as simple logic operations (AND, OR, NOT) can be applied by the PRS to the signals. The PRS allows peripheral to act autonomously without waking the MCU core, saving power.

3.7.5 Low Energy Sensor Interface (LESENSE)

The Low Energy Sensor Interface LESENSETM is a highly configurable sensor interface with support for up to 16 individually configurable sensors. By controlling the analog comparators, ADC, and DAC, LESENSE is capable of supporting a wide range of sensors and measurement schemes, and can for instance measure LC sensors, resistive sensors and capacitive sensors. LESENSE also includes a programmable finite state machine which enables simple processing of measurement results without CPU intervention. LESENSE is available in energy mode EM2, in addition to EM0 and EM1, making it ideal for sensor monitoring in applications with a strict energy budget.

3.8 Security Features

3.8.1 GPCRC (General Purpose Cyclic Redundancy Check)

The GPCRC block implements a Cyclic Redundancy Check (CRC) function. It supports both 32-bit and 16-bit polynomials. The supported 32-bit polynomial is 0x04C11DB7 (IEEE 802.3), while the 16-bit polynomial can be programmed to any value, depending on the needs of the application.

3.8.2 Crypto Accelerator (CRYPTO)

The Crypto Accelerator is a fast and energy-efficient autonomous hardware encryption and decryption accelerator. EFR32 devices support AES encryption and decryption with 128- or 256-bit keys, ECC over both GF(P) and GF(2^m), SHA-1 and SHA-2 (SHA-224 and SHA-256).

Supported block cipher modes of operation for AES include: ECB, CTR, CBC, PCBC, CFB, OFB, GCM, CBC-MAC, GMAC and CCM.

Supported ECC NIST recommended curves include P-192, P-224, P-256, K-163, K-233, B-163 and B-233.

The CRYPTO1 block is tightly linked to the Radio Buffer Controller (BUFC) enabling fast and efficient autonomous cipher operations on data buffer content. It allows fast processing of GCM (AES), ECC and SHA with little CPU intervention.

CRYPTO also provides trigger signals for DMA read and write operations.

3.8.3 True Random Number Generator (TRNG)

The TRNG is a non-deterministic random number generator based on a full hardware solution. The TRNG is validated with NIST800-22 and AIS-31 test suites as well as being suitable for FIPS 140-2 certification (for the purposes of cryptographic key generation).

3.8.4 Security Management Unit (SMU)

The Security Management Unit (SMU) allows software to set up fine-grained security for peripheral access, which is not possible in the Memory Protection Unit (MPU). Peripherals may be secured by hardware on an individual basis, such that only priveleged accesses to the peripheral's register interface will be allowed. When an access fault occurs, the SMU reports the specific peripheral involved and can optionally generate an interrupt.

3.9 Analog

3.9.1 Analog Port (APORT)

The Analog Port (APORT) is an analog interconnect matrix allowing access to many analog peripherals on a flexible selection of pins. Each APORT bus consists of analog switches connected to a common wire. Since many clients can operate differentially, buses are grouped by X/Y pairs.

3.9.2 Analog Comparator (ACMP)

The Analog Comparator is used to compare the voltage of two analog inputs, with a digital output indicating which input voltage is higher. Inputs are selected from among internal references and external pins. The tradeoff between response time and current consumption is configurable by software. Two 6-bit reference dividers allow for a wide range of internally-programmable reference sources. The ACMP can also be used to monitor the supply voltage. An interrupt can be generated when the supply falls below or rises above the programmable threshold.

3.9.3 Analog to Digital Converter (ADC)

The ADC is a Successive Approximation Register (SAR) architecture, with a resolution of up to 12 bits at up to 1 Msps. The output sample resolution is configurable and additional resolution is possible using integrated hardware for averaging over multiple samples. The ADC includes integrated voltage references and an integrated temperature sensor. Inputs are selectable from a wide range of sources, including pins configurable as either single-ended or differential.

3.9.4 Capacitive Sense (CSEN)

The CSEN peripheral is a dedicated Capacitive Sensing block for implementing touch-sensitive user interface elements such a switches and sliders. The CSEN peripheral uses a charge ramping measurement technique, which provides robust sensing even in adverse conditions including radiated noise and moisture. The peripheral can be configured to take measurements on a single port pin or scan through multiple pins and store results to memory through DMA. Several channels can also be shorted together to measure the combined capacitance or implement wake-on-touch from very low energy modes. Hardware includes a digital accumulator and an averaging filter, as well as digital threshold comparators to reduce software overhead.

3.9.5 Digital to Analog Current Converter (IDAC)

The IDAC can source or sink a configurable constant current. This current can be driven on an output pin or routed to the selected ADC input pin for capacitive sensing. The full-scale current is programmable between $0.05~\mu A$ and $64~\mu A$ with several ranges consisting of various step sizes.

3.9.6 Digital to Analog Converter (VDAC)

The Digital to Analog Converter (VDAC) can convert a digital value to an analog output voltage. The VDAC is a fully differential, 500 ksps, 12-bit converter. The opamps are used in conjunction with the VDAC, to provide output buffering. One opamp is used per single-ended channel, or two opamps are used to provide differential outputs. The VDAC may be used for a number of different applications such as sensor interfaces or sound output. The VDAC can generate high-resolution analog signals while the MCU is operating at low frequencies and with low total power consumption. Using DMA and a timer, the VDAC can be used to generate waveforms without any CPU intervention. The VDAC is available in all energy modes down to and including EM3.

3.9.7 Operational Amplifiers

The opamps are low power amplifiers with a high degree of flexibility targeting a wide variety of standard opamp application areas, and are available down to EM3. With flexible built-in programming for gain and interconnection they can be configured to support multiple common opamp functions. All pins are also available externally for filter configurations. Each opamp has a rail to rail input and a rail to rail output. They can be used in conjunction with the VDAC peripheral or in stand-alone configurations. The opamps save energy, PCB space, and cost as compared with standalone opamps because they are integrated on-chip.

3.10 Reset Management Unit (RMU)

The RMU is responsible for handling reset of the EFR32FG13. A wide range of reset sources are available, including several power supply monitors, pin reset, software controlled reset, core lockup reset, and watchdog reset.

3.11 Core and Memory

3.11.1 Processor Core

The ARM Cortex-M processor includes a 32-bit RISC processor integrating the following features and tasks in the system:

- ARM Cortex-M4 RISC processor achieving 1.25 Dhrystone MIPS/MHz
- · Memory Protection Unit (MPU) supporting up to 8 memory segments
- Up to 512 kB flash program memory
- Up to 64 kB RAM data memory
- · Configuration and event handling of all peripherals
- · 2-pin Serial-Wire debug interface

3.11.2 Memory System Controller (MSC)

The Memory System Controller (MSC) is the program memory unit of the microcontroller. The flash memory is readable and writable from both the Cortex-M and DMA. The flash memory is divided into two blocks; the main block and the information block. Program code is normally written to the main block, whereas the information block is available for special user data and flash lock bits. There is also a read-only page in the information block containing system and device calibration data. Read and write operations are supported in energy modes EM0 Active and EM1 Sleep.

3.11.3 Linked Direct Memory Access Controller (LDMA)

The Linked Direct Memory Access (LDMA) controller allows the system to perform memory operations independently of software. This reduces both energy consumption and software workload. The LDMA allows operations to be linked together and staged, enabling sophisticated operations to be implemented.

3.12 Memory Map

The EFR32FG13 memory map is shown in the figures below. RAM and flash sizes are for the largest memory configuration.

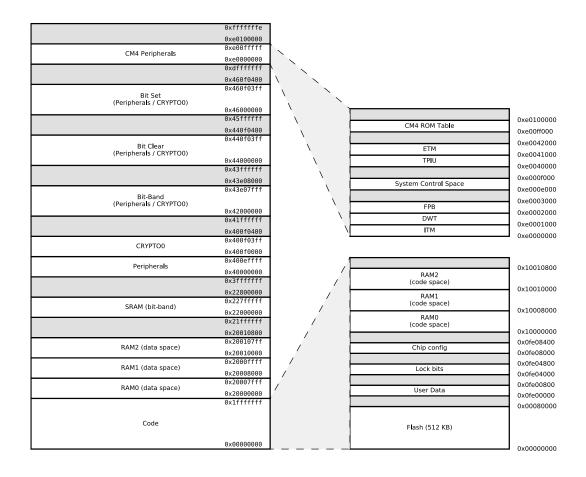


Figure 3.2. EFR32FG13 Memory Map — Core Peripherals and Code Space

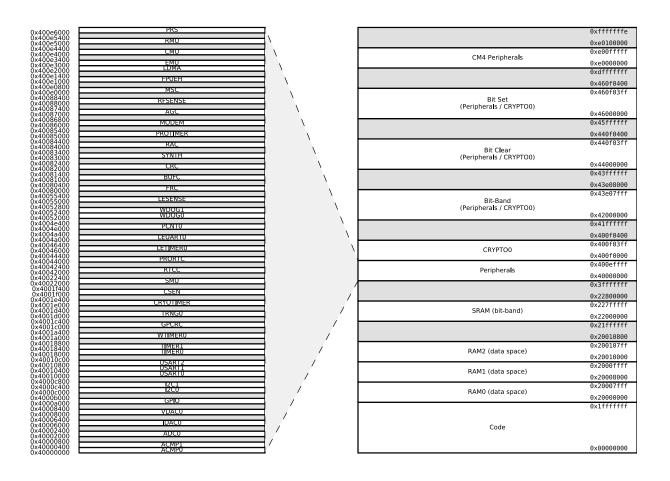


Figure 3.3. EFR32FG13 Memory Map — Peripherals

3.13 Configuration Summary

The features of the EFR32FG13 are a subset of the feature set described in the device reference manual. The table below describes device specific implementation of the features. Remaining peripherals support full configuration.

Table 3.2. Configuration Summary

Peripheral	Configuration	Pin Connections
USART0	IrDA	US0_TX, US0_RX, US0_CLK, US0_CS
	SmartCard	
USART1	l ² S	US1_TX, US1_RX, US1_CLK, US1_CS
	SmartCard	
USART2	IrDA	US2_TX, US2_RX, US2_CLK, US2_CS
	SmartCard	
TIMER0	with DTI	TIM0_CC[2:0], TIM0_CDTI[2:0]
TIMER1	-	TIM1_CC[3:0]
WTIMER0	with DTI	WTIM0_CC[2:0], WTIM0_CDTI[2:0]

4. Electrical Specifications

4.1 Electrical Characteristics

All electrical parameters in all tables are specified under the following conditions, unless stated otherwise:

- Typical values are based on T_{AMB} =25 °C and V_{DD} = 3.3 V, by production test and/or technology characterization.
- Radio performance numbers are measured in conducted mode, based on Silicon Laboratories reference designs using output power-specific external RF impedance-matching networks for interfacing to a 50 Ω source or load.
- Minimum and maximum values represent the worst conditions across supply voltage, process variation, and operating temperature, unless stated otherwise.

Refer to 4.1.2.1 General Operating Conditions for more details about operational supply and temperature limits.

4.1.1 Absolute Maximum Ratings

Stresses above those listed below may cause permanent damage to the device. This is a stress rating only and functional operation of the devices at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. For more information on the available quality and reliability data, see the Quality and Reliability Monitor Report at http://www.silabs.com/support/quality/pages/default.aspx.

Table 4.1. Absolute Maximum Ratings

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Storage temperature range	T _{STG}		-50	_	150	°C
Voltage on any supply pin	V_{DDMAX}		-0.3	_	3.8	V
Voltage ramp rate on any supply pin	V _{DDRAMPMAX}		_	_	1	V / µs
DC voltage on any GPIO pin	V _{DIGPIN}	5V tolerant GPIO pins ^{1 2 3}	-0.3	_	Min of 5.25 and IOVDD +2	V
		Standard GPIO pins	-0.3	_	IOVDD+0.3	V
Voltage on HFXO pins	V _{HFXOPIN}		-0.3	_	1.4	V
Input RF level on pins 2G4RF_IOP and 2G4RF_ION	P _{RFMAX2G4}		_	_	10	dBm
Voltage differential between RF pins (2G4RF_IOP - 2G4RF_ION)	V _{MAXDIFF2G4}		-50	_	50	mV
Absolute voltage on RF pins 2G4RF_IOP and 2G4RF_ION	V _{MAX2G4}		-0.3	_	3.3	V
Absolute voltage on Sub- GHz RF pins	V _{MAXSUBG}	Pins SUBGRF_OP and SUBGRF_ON	-0.3	_	3.3	V
		Pins SUBGRF_IP and SUBGRF_IN,	-0.3	_	0.3	V
Total current into VDD power lines	I _{VDDMAX}	Source	_	_	200	mA
Total current into VSS ground lines	I _{VSSMAX}	Sink	_	_	200	mA
Current per I/O pin	I _{IOMAX}	Sink	_	_	50	mA
		Source	_	_	50	mA
Current for all I/O pins	I _{IOALLMAX}	Sink			200	mA
		Source	_	_	200	mA
Junction temperature	TJ	-G grade devices	-40	_	105	°C
		-I grade devices	-40	_	125	°C

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit

Note:

- 1. When a GPIO pin is routed to the analog block through the APORT, the maximum voltage = IOVDD.
- 2. Valid for IOVDD in valid operating range or when IOVDD is undriven (high-Z). If IOVDD is connected to a low-impedance source below the valid operating range (e.g. IOVDD shorted to VSS), the pin voltage maximum is IOVDD + 0.3 V, to avoid exceeding the maximum IO current specifications.
- 3. To operate above the IOVDD supply rail, over-voltage tolerance must be enabled according to the GPIO_Px_OVTDIS register. Pins with over-voltage tolerance disabled have the same limits as Standard GPIO.

4.1.2 Operating Conditions

When assigning supply sources, the following requirements must be observed:

- VREGVDD must be greater than or equal to AVDD, DVDD, RFVDD, PAVDD and all IOVDD supplies.
- VREGVDD = AVDD
- DVDD ≤ AVDD
- IOVDD ≤ AVDD
- RFVDD ≤ AVDD
- PAVDD ≤ AVDD

4.1.2.1 General Operating Conditions

Table 4.2. General Operating Conditions

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Operating ambient tempera-	T _A	-G temperature grade	-40	25	85	°C
ture range ¹		-I temperature grade	-40	25	125	°C
AVDD supply voltage ²	V _{AVDD}		1.8	3.3	3.8	V
VREGVDD operating supply	V _{VREGVDD}	DCDC in regulation	2.4	3.3	3.8	V
voltage ^{2 3}		DCDC in bypass, 50mA load	1.8	3.3	3.8	V
		DCDC not in use. DVDD externally shorted to VREGVDD	1.8	3.3	3.8	V
VREGVDD current	I _{VREGVDD}	DCDC in bypass, T ≤ 85 °C	_	_	200	mA
		DCDC in bypass, T > 85 °C	_	_	100	mA
RFVDD operating supply voltage	V _{RFVDD}		1.62	_	V _{VREGVDD}	V
DVDD operating supply voltage	V _{DVDD}		1.62	_	V _{VREGVDD}	V
PAVDD operating supply voltage	V _{PAVDD}		1.62	_	V _{VREGVDD}	V
IOVDD operating supply voltage	V _{IOVDD}	All IOVDD pins ⁴	1.62	_	V _{VREGVDD}	V
DECOUPLE output capacitor ^{5 6}	C _{DECOUPLE}		0.75	1.0	2.75	μF
Difference between AVDD and VREGVDD, ABS(AVDD-VREGVDD) ²	dV_{DD}		_	_	0.1	V
HFCORECLK frequency	f _{CORE}	VSCALE2, MODE = WS1	_	_	40	MHz
		VSCALE0, MODE = WS0	_	_	20	MHz
HFCLK frequency	f _{HFCLK}	VSCALE2	_	_	40	MHz
		VSCALE0	_	_	20	MHz

Note:

- 1. The maximum limit on T_A may be lower due to device self-heating, which depends on the power dissipation of the specific application. T_A (max) = T_J (max) (THETA_{JA} x PowerDissipation). Refer to the Absolute Maximum Ratings table and the Thermal Characteristics table for T_J and THETA_{JA}.
- 2. VREGVDD must be tied to AVDD. Both VREGVDD and AVDD minimum voltages must be satisfied for the part to operate.
- 3. The minimum voltage required in bypass mode is calculated using R_{BYP} from the DCDC specification table. Requirements for other loads can be calculated as $V_{DVDD\ min}+I_{LOAD}*R_{BYP\ max}$.
- 4. When the CSEN peripheral is used with chopping enabled (CSEN_CTRL_CHOPEN = ENABLE), IOVDD must be equal to AVDD.
- 5. The system designer should consult the characteristic specs of the capacitor used on DECOUPLE to ensure its capacitance value stays within the specified bounds across temperature and DC bias.
- 6. VSCALE0 to VSCALE2 voltage change transitions occur at a rate of 10 mV / usec for approximately 20 usec. During this transition, peak currents will be dependent on the value of the DECOUPLE output capacitor, from 35 mA (with a 1 μ F capacitor) to 70 mA (with a 2.7 μ F capacitor).

4.1.3 Thermal Characteristics

Table 4.3. Thermal Characteristics

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Thermal resistance, QFN48	THETA _{JA_QFN48}	2-Layer PCB, Air velocity = 0 m/s	_	75.7	_	°C/W
Package		2-Layer PCB, Air velocity = 1 m/s	_	61.5	_	°C/W
		2-Layer PCB, Air velocity = 2 m/s	_	55.4	_	°C/W
		4-Layer PCB, Air velocity = 0 m/s	_	30.2	_	°C/W
		4-Layer PCB, Air velocity = 1 m/s	_	26.3	_	°C/W
		4-Layer PCB, Air velocity = 2 m/s	_	24.9	_	°C/W
Thermal resistance, QFN32	THETA _{JA_QFN32}	4-Layer PCB, Air velocity = 0 m/s	_	30.1	_	°C/W
Package		4-Layer PCB, Air velocity = 1 m/s	_	24.9	_	°C/W
		4-Layer PCB, Air velocity = 2 m/s	_	23.9	_	°C/W

4.1.4 DC-DC Converter

Test conditions: L_DCDC=4.7 μ H (Murata LQH3NPN4R7MM0L), C_DCDC=4.7 μ F (Samsung CL10B475KQ8NQNC), V_DCDC_I=3.3 V, V_DCDC_O=1.8 V, I_DCDC_LOAD=50 mA, Heavy Drive configuration, F_DCDC_LN=7 MHz, unless otherwise indicated.

Table 4.4. DC-DC Converter

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Input voltage range	V _{DCDC_I}	Bypass mode, I _{DCDC_LOAD} = 50 mA	1.8	_	V _{VREGVDD} _ MAX	V
		Low noise (LN) mode, 1.8 V output, I _{DCDC_LOAD} = 100 mA, or Low power (LP) mode, 1.8 V output, I _{DCDC_LOAD} = 10 mA	2.4	_	V _{VREGVDD} MAX	V
		Low noise (LN) mode, 1.8 V output, I _{DCDC_LOAD} = 200 mA	2.6	_	V _{VREGVDD} _	V
Output voltage programma- ble range ¹	V _{DCDC_O}		1.8	_	V _{VREGVDD}	V
Regulation DC accuracy	ACC _{DC}	Low Noise (LN) mode, 1.8 V target output	1.7	_	1.9	V
Regulation window ²	WIN _{REG}	Low Power (LP) mode, LPCMPBIASEMxx³ = 0, 1.8 V target output, I _{DCDC_LOAD} ≤ 75 μA	1.63	_	2.2	V
		Low Power (LP) mode, LPCMPBIASEMxx ³ = 3, 1.8 V target output, I _{DCDC_LOAD} ≤ 10 mA	1.63	_	2.1	V
Steady-state output ripple	V _R	Radio disabled	_	3	_	mVpp
Output voltage under/over- shoot	Vov	CCM Mode (LNFORCECCM ³ = 1), Load changes between 0 mA and 100 mA	_	25	60	mV
		DCM Mode (LNFORCECCM ³ = 0), Load changes between 0 mA and 10 mA	_	45	90	mV
		Overshoot during LP to LN CCM/DCM mode transitions compared to DC level in LN mode	_	200	_	mV
		Undershoot during BYP/LP to LN CCM (LNFORCECCM ³ = 1) mode transitions compared to DC level in LN mode	_	40	_	mV
		Undershoot during BYP/LP to LN DCM (LNFORCECCM ³ = 0) mode transitions compared to DC level in LN mode	_	100	_	mV
DC line regulation	V _{REG}	Input changes between V _{VREGVDD_MAX} and 2.4 V	_	0.1	_	%
DC load regulation	I _{REG}	Load changes between 0 mA and 100 mA in CCM mode	_	0.1	_	%

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Max load current	I _{LOAD_MAX}	Low noise (LN) mode, Heavy Drive ⁴ , T ≤ 85 °C	_	_	200	mA
		Low noise (LN) mode, Heavy Drive ⁴ , T > 85 °C	_	_	100	mA
		Low noise (LN) mode, Medium Drive ⁴	_	_	100	mA
		Low noise (LN) mode, Light Drive ⁴	_	_	50	mA
		Low power (LP) mode, LPCMPBIASEMxx ³ = 0	_	_	75	μA
		Low power (LP) mode, LPCMPBIASEMxx ³ = 3	_	_	10	mA
DCDC nominal output capacitor ⁵	C _{DCDC}	25% tolerance	1	4.7	4.7	μF
DCDC nominal output inductor	L _{DCDC}	20% tolerance	4.7	4.7	4.7	μH
Resistance in Bypass mode	R _{BYP}		_	1.2	2.5	Ω

Note:

- 1. Due to internal dropout, the DC-DC output will never be able to reach its input voltage, $V_{VREGVDD}$.
- 2. LP mode controller is a hysteretic controller that maintains the output voltage within the specified limits.
- 3. LPCMPBIASEMxx refers to either LPCMPBIASEM234H in the EMU_DCDCMISCCTRL register or LPCMPBIASEM01 in the EMU_DCDCLOEM01CFG register, depending on the energy mode.
- 4. Drive levels are defined by configuration of the PFETCNT and NFETCNT registers. Light Drive: PFETCNT=NFETCNT=3; Medium Drive: PFETCNT=NFETCNT=7; Heavy Drive: PFETCNT=NFETCNT=15.
- 5. Output voltage under/over-shoot and regulation are specified with C_{DCDC} 4.7 μ F. Different settings for DCDCLNCOMPCTRL must be used if C_{DCDC} is lower than 4.7 μ F. See Application Note AN0948 for details.

4.1.5 Current Consumption

4.1.5.1 Current Consumption 3.3 V without DC-DC Converter

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = DVDD = RFVDD = PAVDD = 3.3 V. T = 25 °C. DCDC is off. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 °C.

Table 4.5. Current Consumption 3.3 V without DC-DC Converter

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in EM0 mode with all peripherals disabled	I _{ACTIVE}	38.4 MHz crystal, CPU running while loop from flash ¹	_	128	_	µA/MHz
abieu		38 MHz HFRCO, CPU running Prime from flash	_	97	_	µA/MHz
		38 MHz HFRCO, CPU running while loop from flash	_	98	107	µA/MHz
		38 MHz HFRCO, CPU running CoreMark from flash	_	119	_	µA/MHz
		26 MHz HFRCO, CPU running while loop from flash	_	100	109	µA/MHz
		1 MHz HFRCO, CPU running while loop from flash	_	246	430	µA/MHz
Current consumption in EM0 mode with all peripherals dis-	I _{ACTIVE_VS}	19 MHz HFRCO, CPU running while loop from flash	_	86	_	µA/MHz
abled and voltage scaling enabled		1 MHz HFRCO, CPU running while loop from flash	_	209	_	µA/MHz
Current consumption in EM1	І _{ЕМ1}	38.4 MHz crystal ¹	_	76	_	μΑ/MHz
mode with all peripherals disabled		38 MHz HFRCO	_	47	51	μΑ/MHz
		26 MHz HFRCO	_	49	55	µA/MHz
		1 MHz HFRCO	_	195	374	µA/MHz
Current consumption in EM1	I _{EM1_VS}	19 MHz HFRCO	_	43	_	µA/MHz
mode with all peripherals dis- abled and voltage scaling enabled		1 MHz HFRCO	_	167	_	µA/MHz
Current consumption in EM2 mode, with voltage scaling	I _{EM2_VS}	Full 64 kB RAM retention and RTCC running from LFXO	_	1.9	_	μА
enabled		Full 64 kB RAM retention and RTCC running from LFRCO	_	2.2	_	μА
		1 bank (16 kB) RAM retention and RTCC running from LFRCO ²	_	1.9	3.3	μА
Current consumption in EM3 mode, with voltage scaling enabled	I _{EM3_VS}	Full 64 kB RAM retention and CRYOTIMER running from ULFR-CO	_	1.53	3.0	μА
Current consumption in EM4H mode, with voltage	I _{EM4H_VS}	128 byte RAM retention, RTCC running from LFXO	<u> </u>	0.93	_	μА
scaling enabled		128 byte RAM retention, CRYO- TIMER running from ULFRCO	_	0.45	_	μА
		128 byte RAM retention, no RTCC	_	0.44	0.9	μA

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in EM4S mode	I _{EM4S}	No RAM retention, no RTCC	_	0.04	0.085	μA

Note:

- 1. CMU_HFXOCTRL_LOWPOWER=0.
- 2. CMU_LFRCOCTRL_ENVREF = 1, CMU_LFRCOCTRL_VREFUPDATE = 1

4.1.5.2 Current Consumption 3.3 V using DC-DC Converter

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD = 1.8 V DC-DC output. $T = 25 \,^{\circ}C$. Minimum and maximum values in this table represent the worst conditions across process variation at $T = 25 \,^{\circ}C$.

Table 4.6. Current Consumption 3.3 V using DC-DC Converter

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in EM0 mode with all peripherals disabled, DCDC in Low Noise	I _{ACTIVE} _DCM	38.4 MHz crystal, CPU running while loop from flash ²	_	87	_	μΑ/MHz
DCM mode ¹		38 MHz HFRCO, CPU running Prime from flash	_	69	_	μA/MHz
		38 MHz HFRCO, CPU running while loop from flash	_	70	_	μA/MHz
		38 MHz HFRCO, CPU running CoreMark from flash	_	82	_	μA/MHz
		26 MHz HFRCO, CPU running while loop from flash	_	76	_	μA/MHz
		1 MHz HFRCO, CPU running while loop from flash	_	615	_	μA/MHz
Current consumption in EM0 mode with all peripherals dis-	I _{ACTIVE} CCM	38.4 MHz crystal, CPU running while loop from flash ²	_	97	_	μA/MHz
abled, DCDC in Low Noise CCM mode ³		38 MHz HFRCO, CPU running Prime from flash	_	80	_	μΑ/MHz
		38 MHz HFRCO, CPU running while loop from flash	_	81	_	μΑ/MHz
		38 MHz HFRCO, CPU running CoreMark from flash	_	92	_	μA/MHz
		26 MHz HFRCO, CPU running while loop from flash	_	94	_	μA/MHz
		1 MHz HFRCO, CPU running while loop from flash	_	1145	_	μA/MHz
Current consumption in EM0 mode with all peripherals dis-	IACTIVE_CCM_VS	19 MHz HFRCO, CPU running while loop from flash	_	101	_	μA/MHz
abled and voltage scaling enabled, DCDC in Low Noise CCM mode ³		1 MHz HFRCO, CPU running while loop from flash	_	1124	_	μΑ/MHz
Current consumption in EM1	I _{EM1_DCM}	38.4 MHz crystal ²	_	56	_	μA/MHz
mode with all peripherals disabled, DCDC in Low Noise		38 MHz HFRCO	_	39	_	μA/MHz
DCM mode ¹		26 MHz HFRCO	_	46	_	μA/MHz
		1 MHz HFRCO	_	588	_	μA/MHz
Current consumption in EM1	I _{EM1_DCM_VS}	19 MHz HFRCO	_	50	_	μA/MHz
mode with all peripherals dis- abled and voltage scaling enabled, DCDC in Low Noise DCM mode ¹		1 MHz HFRCO	_	572	_	μA/MHz

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in EM2 mode, with voltage scaling enabled, DCDC in LP mode ⁴	I _{EM2_VS}	Full 64 kB RAM retention and RTCC running from LFXO	_	1.4	_	μА
		Full 64 kB RAM retention and RTCC running from LFRCO	_	1.5	_	μА
		1 bank RAM retention and RTCC running from LFRCO ⁵	_	1.3	_	μА
Current consumption in EM3 mode, with voltage scaling enabled	I _{EM3_VS}	Full 64 kB RAM retention and CRYOTIMER running from ULFR-CO	_	1.14	_	μA
Current consumption in EM4H mode, with voltage scaling enabled	I _{EM4H_VS}	128 byte RAM retention, RTCC running from LFXO	_	0.75	_	μА
		128 byte RAM retention, CRYO- TIMER running from ULFRCO	_	0.44	_	μА
		128 byte RAM retention, no RTCC	_	0.42	_	μA
Current consumption in EM4S mode	I _{EM4S}	No RAM retention, no RTCC	_	0.07	_	μА

Note:

- 1. DCDC Low Noise DCM Mode = Light Drive (PFETCNT=NFETCNT=3), F=3.0 MHz (RCOBAND=0), ANASW=DVDD.
- 2. CMU_HFXOCTRL_LOWPOWER=0.
- 3. DCDC Low Noise CCM Mode = Light Drive (PFETCNT=NFETCNT=3), F=6.4 MHz (RCOBAND=4), ANASW=DVDD.
- 4. DCDC Low Power Mode = Medium Drive, LPOSCDIV=1, LPCMPBIASEM234H=0, LPCLIMILIMSEL=1, ANASW=DVDD.
- 5. CMU_LFRCOCTRL_ENVREF = 1, CMU_LFRCOCTRL_VREFUPDATE = 1

4.1.5.3 Current Consumption 1.8 V without DC-DC Converter

Unless otherwise indicated, typical conditions are: $VREGVDD = AVDD = DVDD = RFVDD = PAVDD = 1.8 \text{ V. T} = 25 ^{\circ}\text{C}$. DCDC is off. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 $^{\circ}\text{C}$.

Table 4.7. Current Consumption 1.8 V without DC-DC Converter

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in EM0 mode with all peripherals disabled	I _{ACTIVE}	38.4 MHz crystal, CPU running while loop from flash ¹	_	128	_	μA/MHz
abica		38 MHz HFRCO, CPU running Prime from flash	_	97	_	µA/MHz
		38 MHz HFRCO, CPU running while loop from flash	_	98	_	μA/MHz
		38 MHz HFRCO, CPU running CoreMark from flash	_	119	_	μA/MHz
		26 MHz HFRCO, CPU running while loop from flash	_	100	_	μA/MHz
		1 MHz HFRCO, CPU running while loop from flash	_	243	_	μA/MHz
Current consumption in EM0 mode with all peripherals dis-	I _{ACTIVE_VS}	19 MHz HFRCO, CPU running while loop from flash	_	86	_	μA/MHz
abled and voltage scaling enabled		1 MHz HFRCO, CPU running while loop from flash	_	206	_	μA/MHz
Current consumption in EM1	I _{EM1}	38.4 MHz crystal ¹	_	76	_	μA/MHz
mode with all peripherals disabled		38 MHz HFRCO	_	47	_	μA/MHz
		26 MHz HFRCO	_	48	_	μA/MHz
		1 MHz HFRCO	_	191	_	μA/MHz
Current consumption in EM1	I _{EM1_VS}	19 MHz HFRCO	_	43	_	μA/MHz
mode with all peripherals dis- abled and voltage scaling enabled		1 MHz HFRCO	_	163	_	μA/MHz
Current consumption in EM2 mode, with voltage scaling	I _{EM2_VS}	Full 64 kB RAM retention and RTCC running from LFXO	_	1.8	_	μА
enabled		Full 64 kB RAM retention and RTCC running from LFRCO	_	2.0	_	μА
		1 bank (16 kB) RAM retention and RTCC running from LFRCO ²	_	1.6	_	μА
Current consumption in EM3 mode, with voltage scaling enabled	I _{EM3_VS}	Full 64 kB RAM retention and CRYOTIMER running from ULFR-CO	_	1.43	_	μА
Current consumption in EM4H mode, with voltage scaling enabled	I _{EM4H_VS}	128 byte RAM retention, RTCC running from LFXO	_	0.83	_	μА
		128 byte RAM retention, CRYO- TIMER running from ULFRCO	<u> </u>	0.37	_	μА
		128 byte RAM retention, no RTCC	_	0.36	_	μA
Current consumption in EM4S mode	I _{EM4S}	no RAM retention, no RTCC	_	0.05	_	μА

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Note:						
1. CMU_HFXOCTRL_LOW	POWER=0.					
2. CMU LFRCOCTRL ENV	VREF = 1, CMU	LFRCOCTRL VREFUPDATE = 1				

4.1.5.4 Current Consumption Using Radio 3.3 V with DC-DC

Unless otherwise indicated, typical conditions are: VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD = 1.8 V. T = 25 °C. Minimum and maximum values in this table represent the worst conditions across process variation at T = 25 °C.

Table 4.8. Current Consumption Using Radio 3.3 V with DC-DC

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in receive mode, active packet reception (MCU in EM1 @ 38.4 MHz, peripheral clocks disabled), T ≤ 85 °C	I _{RX_ACTIVE}	500 kbit/s, 2GFSK, F = 915 MHz, Radio clock prescaled by 4	_	9.3	10.2	mA
		38.4 kbit/s, 2GFSK, F = 868 MHz, Radio clock prescaled by 4	_	8.6	10.2	mA
		38.4 kbit/s, 2GFSK, F = 490 MHz, Radio clock prescaled by 4	_	8.6	10.2	mA
		50 kbit/s, 2GFSK, F = 433 MHz, Radio clock prescaled by 4		8.6	10.2	mA
		38.4 kbit/s, 2GFSK, F = 315 MHz, Radio clock prescaled by 4	_	8.6	10.2	mA
		38.4 kbit/s, 2GFSK, F = 169 MHz, Radio clock prescaled by 4	_	8.4	10.2	mA
		125 kbit/s, 2GFSK, F = 2.4 GHz, Radio clock prescaled by 4	_	9.2	_	mA
		500 kbit/s, 2GFSK, F = 2.4 GHz, Radio clock prescaled by 4	_	9.3	_	mA
		1 Mbit/s, 2GFSK, F = 2.4 GHz, Radio clock prescaled by 4	_	9.5	_	mA
		2 Mbit/s, 2GFSK, F = 2.4 GHz, Radio clock prescaled by 4	_	10.6	_	mA
		802.15.4 receiving frame, F = 2.4 GHz, Radio clock prescaled by 3	_	10.3	_	mA
Current consumption in receive mode, active packet	I _{RX_ACTIVE_HT}	500 kbit/s, 2GFSK, F = 915 MHz, Radio clock prescaled by 4	_		13	mA
reception (MCU in EM1 @ 38.4 MHz, peripheral clocks disabled), T > 85 °C		38.4 kbit/s, 2GFSK, F = 868 MHz, Radio clock prescaled by 4	_		13	mA
		38.4 kbit/s, 2GFSK, F = 490 MHz, Radio clock prescaled by 4	_	_	13	mA
		50 kbit/s, 2GFSK, F = 433 MHz, Radio clock prescaled by 4	_	_	13	mA
		38.4 kbit/s, 2GFSK, F = 315 MHz, Radio clock prescaled by 4	_	_	13	mA
		38.4 kbit/s, 2GFSK, F = 169 MHz, Radio clock prescaled by 4	_	_	13	mA

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in receive mode, listening for packet (MCU in EM1 @ 38.4 MHz, peripheral clocks disabled), T ≤ 85 °C	I _{RX_LISTEN}	500 kbit/s, 2GFSK, F = 915 MHz, No radio clock prescaling	_	10.2	11	MA M
		38.4 kbit/s, 2GFSK, F = 868 MHz, No radio clock prescaling	_	9.5	11	mA
		38.4 kbit/s, 2GFSK, F = 490 MHz, No radio clock prescaling	_	9.5	11	mA
		50 kbit/s, 2GFSK, F = 433 MHz, No radio clock prescaling		9.5	11	mA
		38.4 kbit/s, 2GFSK, F = 315 MHz, No radio clock prescaling		9.4	11	mA
		38.4 kbit/s, 2GFSK, F = 169 MHz, No radio clock prescaling		9.3	11	mA
		125 kbit/s, 2GFSK, F = 2.4 GHz, No radio clock prescaling	_	10.4	_	mA
		500 kbit/s, 2GFSK, F = 2.4 GHz, No radio clock prescaling		10.4	_	mA
		1 Mbit/s, 2GFSK, F = 2.4 GHz, No radio clock prescaling	_	10.5	_	mA
		2 Mbit/s, 2GFSK, F = 2.4 GHz, No radio clock prescaling		11.3	_	mA
		802.15.4, F = 2.4 GHz, No radio clock prescaling	_	11.9	_	mA
Current consumption in receive mode, listening for	I _{RX_LISTEN_HT}	500 kbit/s, 2GFSK, F = 915 MHz, No radio clock prescaling		_	14	mA
packet (MCU in EM1 @ 38.4 MHz, peripheral clocks disabled), T > 85 °C		38.4 kbit/s, 2GFSK, F = 868 MHz, No radio clock prescaling		_	14	mA
		38.4 kbit/s, 2GFSK, F = 490 MHz, No radio clock prescaling	_	_	14	mA
		50 kbit/s, 2GFSK, F = 433 MHz, No radio clock prescaling	_	<u> </u>	14	mA
		38.4 kbit/s, 2GFSK, F = 315 MHz, No radio clock prescaling	_	_	14	mA
		38.4 kbit/s, 2GFSK, F = 169 MHz, No radio clock prescaling	_	_	14	mA

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in transmit mode (MCU in EM1	I _{TX}	F = 915 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	90.2	134.3	mA
@ 38.4 MHz, peripheral clocks disabled), T ≤ 85 °C		F = 915 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	36	42.5	mA
		F = 868 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	79.7	106.7	mA
		F = 868 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	35.3	41	mA
		F = 490 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	93.8	125.4	mA
		F = 433 MHz, CW, 10 dBm match, External PA supply con- nected to DC-DC output	_	20.3	24	mA
		F = 433 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	34	41.5	mA
		F = 315 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	33.5	42	mA
		F = 169 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	88.6	116.7	mA
		F = 2.4 GHz, CW, 0 dBm output power, Radio clock prescaled by 3	_	8.5	_	mA
		F = 2.4 GHz, CW, 0 dBm output power, Radio clock prescaled by 1	_	9.5	_	mA
		F = 2.4 GHz, CW, 3 dBm output power	_	16.5	_	mA
		F = 2.4 GHz, CW, 8 dBm output power	_	26.0	_	mA
		F = 2.4 GHz, CW, 10.5 dBm output power	_	34.0	_	mA
		F = 2.4 GHz, CW, 16.5 dBm output power, PAVDD connected directly to external 3.3V supply	_	86.0	_	mA
		F = 2.4 GHz, CW, 19.5 dBm output power, PAVDD connected directly to external 3.3V supply	_	131.0	_	mA

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Current consumption in transmit mode (MCU in EM1 @ 38.4 MHz, peripheral clocks disabled), T > 85 °C	Ітх_нт	F = 915 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	_	134.3	mA
		F = 915 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	_	42.5	mA
		F = 868 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	_	109.8	mA
		F = 868 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	_	41.3	mA
		F = 490 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	_	130.8	mA
		F = 433 MHz, CW, 10 dBm match, External PA supply con- nected to DC-DC output	_	_	24.4	mA
		F = 433 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	_	41.5	mA
		F = 315 MHz, CW, 14 dBm match, External PA supply con- nected to DCDC output	_	_	42	mA
		F = 169 MHz, CW, 20 dBm match, External PA supply = 3.3V	_	_	122.8	mA

4.1.6 Wake Up Times

Table 4.9. Wake Up Times

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Wake up time from EM1	t _{EM1_WU}		_	3	_	AHB Clocks
Wake up from EM2	t _{EM2_WU}	Code execution from flash	_	10.9	_	μs
		Code execution from RAM	_	3.8	_	μs
Wake up from EM3	t _{EM3_WU}	Code execution from flash	_	10.9	_	μs
		Code execution from RAM	_	3.8	_	μs
Wake up from EM4H ¹	t _{EM4H_WU}	Executing from flash	_	90	_	μs
Wake up from EM4S ¹	t _{EM4S_WU}	Executing from flash	_	300	_	μs
Time from release of reset	t _{RESET}	Soft Pin Reset released	_	51	_	μs
source to first instruction execution		Any other reset released	_	358	_	μs
Power mode scaling time	^t SCALE	VSCALE0 to VSCALE2, HFCLK = 19 MHz ^{2 3}	_	31.8	_	μs
		VSCALE2 to VSCALE0, HFCLK = 19 MHz ⁴	_	4.3	_	μs

- 1. Time from wake up request until first instruction is executed. Wakeup results in device reset.
- 2. Scaling up from VSCALE0 to VSCALE2 requires approximately 30.3 µs + 28 HFCLKs.
- 3. VSCALE0 to VSCALE2 voltage change transitions occur at a rate of 10 mV/ μ s for approximately 20 μ s. During this transition, peak currents will be dependent on the value of the DECOUPLE output capacitor, from 35 mA (with a 1 μ F capacitor) to 70 mA (with a 2.7 μ F capacitor).
- 4. Scaling down from VSCALE2 to VSCALE0 requires approximately 2.8 µs + 29 HFCLKs.

4.1.7 Brown Out Detector (BOD)

Table 4.10. Brown Out Detector (BOD)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
DVDD BOD threshold	$V_{DVDDBOD}$	DVDD rising	_	_	1.62	V
		DVDD falling (EM0/EM1)	1.35	_	_	V
		DVDD falling (EM2/EM3)	1.3	_	_	V
DVDD BOD hysteresis	V _{DVDDBOD_HYST}		_	18	_	mV
DVDD BOD response time	t _{DVDDBOD_DELAY}	Supply drops at 0.1V/µs rate	_	2.4	_	μs
AVDD BOD threshold	V _{AVDDBOD}	AVDD rising	_	_	1.8	V
		AVDD falling (EM0/EM1)	1.62	_	_	V
		AVDD falling (EM2/EM3)	1.53	_	_	V
AVDD BOD hysteresis	V _{AVDDBOD_HYST}		_	20	_	mV
AVDD BOD response time	tavddbod_delay	Supply drops at 0.1V/µs rate	_	2.4	_	μs
EM4 BOD threshold	V _{EM4DBOD}	AVDD rising	_	_	1.7	V
		AVDD falling	1.45	_	_	V
EM4 BOD hysteresis	V _{EM4BOD_HYST}		_	25	_	mV
EM4 BOD response time	t _{EM4BOD_DELAY}	Supply drops at 0.1V/µs rate	_	300	_	μs

4.1.8 Frequency Synthesizer

Table 4.11. Frequency Synthesizer

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF synthesizer frequency	f _{RANGE}	2400 - 2483.5 MHz	2400	_	2483.5	MHz
range		779 - 956 MHz	779	_	956	MHz
		584 - 717 MHz	584	_	717	MHz
		358 - 574 MHz	358	_	574	MHz
		191 - 358 MHz	191	_	358	MHz
		110 - 191 MHz	110	_	191	MHz
LO tuning frequency resolution with 38.4 MHz crystal	f _{RES}	2400 - 2483.5 MHz	_	_	73	Hz
		779 - 956 MHz	_	_	24	Hz
		584 - 717 MHz	_	_	18.3	Hz
		358 - 574 MHz	_	_	12.2	Hz
		191 - 358 MHz	_	_	7.3	Hz
		110 - 191 MHz	_	_	4.6	Hz
Frequency deviation resolu-	df _{RES}	2400 - 2483.5 MHz	_	_	73	Hz
tion with 38.4 MHz crystal		779 - 956 MHz	_	_	24	Hz
		584 - 717 MHz	_	_	18.3	Hz
		358 - 574 MHz	_	_	12.2	Hz
		191 - 358 MHz	_	_	7.3	Hz
		110 - 191 MHz	_	_	4.6	Hz
Maximum frequency devia-	df _{MAX}	2400 - 2483.5 MHz	_	_	1677	kHz
tion with 38.4 MHz crystal		779 - 956 MHz	_	_	559	kHz
		584 - 717 MHz	_	_	419	kHz
		358 - 574 MHz	_	_	280	kHz
		191 - 358 MHz	_	_	167	kHz
		110 - 191 MHz	_	_	105	kHz

4.1.9 2.4 GHz RF Transceiver Characteristics

4.1.9.1 RF Transmitter General Characteristics for 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz.

Table 4.12. RF Transmitter General Characteristics for 2.4 GHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Maximum TX power ¹	POUT _{MAX}	19 dBm-rated part numbers. PAVDD connected directly to external 3.3V supply	_	19.5	_	dBm
Minimum active TX Power	POUT _{MIN}	CW		-30	_	dBm
Output power step size	POUT _{STEP}	-5 dBm< Output power < 0 dBm	_	1	_	dB
		0 dBm < output power < POUT _{MAX}	_	0.5	_	dB
Output power variation vs supply at POUT _{MAX}	POUT _{VAR_V}	1.8 V < V _{VREGVDD} < 3.3 V, PAVDD connected directly to ex- ternal supply, for output power > 10 dBm.	_	4.5	_	dB
		1.8 V < V _{VREGVDD} < 3.3 V using DC-DC converter	_	2.2	_	dB
Output power variation vs temperature at POUT _{MAX}	POUT _{VAR_T}	From -40 to +85 °C, PAVDD connected to DC-DC output	_	1.5	_	dB
		From -40 to +125 °C, PAVDD connected to DC-DC output	_	2.2	_	dB
		From -40 to +85 °C, PAVDD connected to external supply	_	1.5	_	dB
		From -40 to +125 °C, PAVDD connected to external supply	_	3.4	_	dB
Output power variation vs RF frequency at POUT _{MAX}	POUT _{VAR_F}	Over RF tuning frequency range	_	0.4	_	dB
RF tuning frequency range	F _{RANGE}		2400	_	2483.5	MHz

^{1.} Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.9.2 RF Receiver General Characteristics for 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz.

Table 4.13. RF Receiver General Characteristics for 2.4 GHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		2400	_	2483.5	MHz
Receive mode maximum spurious emission	SPUR _{RX}	30 MHz to 1 GHz	_	-57	_	dBm
		1 GHz to 12 GHz	_	-47	_	dBm
Max spurious emissions during active receive mode, per	SPUR _{RX_FCC}	216 MHz to 960 MHz, Conducted Measurement	_	-55.2	_	dBm
FCC Part 15.109(a)		Above 960 MHz, Conducted Measurement	_	-47.2	_	dBm
Level above which RFSENSE will trigger ¹	RFSENSE _{TRIG}	CW at 2.45 GHz	_	-24	_	dBm
Level below which RFSENSE will not trigger ¹	RFSENSE _{THRES}	CW at 2.45 GHz	_	-50	_	dBm
1% PER sensitivity	SENS _{2GFSK}	2 Mbps 2GFSK signal	_	-89.6	_	dBm
		250 kbps 2GFSK signal	_	-100.7	_	dBm

^{1.} RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.9.3 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 85%.

Table 4.14. RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Transmit 6dB bandwidth	TXBW	10 dBm	_	763	_	kHz
Power spectral density limit	PSD _{LIMIT}	Per FCC part 15.247 at 10 dBm	_	-9.1	_	dBm/ 3kHz
		Per FCC part 15.247 at 20 dBm	_	-2	_	dBm/ 3kHz
		Per ETSI 300.328 at 10 dBm/1 MHz	_	10	_	dBm
Occupied channel bandwidth per ETSI EN300.328	OCP _{ETSI328}	99% BW at highest and lowest channels in band, 10 dBm	_	1.1	_	MHz
Emissions of harmonics out- of-band, per FCC part 15.247	SPUR _{HRM_FCC}	2nd,3rd, 5, 6, 8, 9,10 harmonics; continuous transmission of modulated carrier	_	-47	_	dBm
Spurious emissions out-of- band, excluding harmonics captured in SPUR _{HARM,FCC} . Emissions taken at	SPUR _{OOB_FCC}	Per FCC part 15.205/15.209, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Restricted Bands ¹ ²	_	-47	_	dBm
POUT _{MAX} , PAVDD connected to external 3.3 V supply		Per FCC part 15.247, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Non-Restricted Bands	_	-26	_	dBc
Spurious emissions out-of- band; per ETSI 300.328	SPUR _{ETSI328}	[2400-BW to 2400] MHz, [2483.5 to 2483.5+BW] MHz	_	-16	_	dBm
		[2400-2BW to 2400-BW] MHz, [2483.5+BW to 2483.5+2BW] MHz per ETSI 300.328	_	-26	_	dBm
Spurious emissions per ETSI EN300.440	SPUR _{ETSI440}	47-74 MHz,87.5-108 MHz, 174-230 MHz, 470-862 MHz	_	-60	_	dBm
		25-1000 MHz	_	-42	_	dBm
		1-12 GHz	_	-36	_	dBm

^{1.} For 2476 MHz, 1.5 dB of power backoff is used to achieve this value.

^{2.} For 2478 MHz, 4.2 dB of power backoff is used to achieve this value.

4.1.9.4 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz.

Table 4.15. RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Max usable receiver input level, 0.1% BER	SAT	Signal is reference signal ¹ . Packet length is 20 bytes.	_	10	_	dBm
Sensitivity, 0.1% BER	SENS	Signal is reference signal ¹ . Using DC-DC converter.	_	-94.8	_	dBm
Signal to co-channel interferer, 0.1% BER	C/I _{CC}	Desired signal 3 dB above reference sensitivity.	_	10.4	_	dB
N+1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -67 dBm	C/I ₁₊	Interferer is reference signal at +1 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-1.7	_	dB
N-1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -67 dBm	C/I ₁₋	Interferer is reference signal at -1 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-0.5	_	dB
Alternate selectivity, 0.1% BER, with allowable excep- tions. Desired is reference signal at -67 dBm	C/I ₂	Interferer is reference signal at ± 2 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-40.8	_	dB
Alternate selectivity, 0.1% BER, with allowable excep- tions. Desired is reference signal at -67 dBm	C/I ₃	Interferer is reference signal at ± 3 MHz offset. Desired frequency 2404 MHz ≤ Fc ≤ 2480 MHz	_	-45.1	_	dB
Selectivity to image frequency, 0.1% BER. Desired is reference signal at -67 dBm	C/I _{IM}	Interferer is reference signal at image frequency with 1 MHz precision	_	-38.2	_	dB
Selectivity to image frequency ± 1 MHz, 0.1% BER. Desired is reference signal at -67 dBm	C/I _{IM+1}	Interferer is reference signal at image frequency ± 1 MHz with 1 MHz precision	_	-45.7	_	dB
Blocking, less than 0.1% BER. Desired is -67dBm	BLOCK _{OOB}	Interferer frequency 30 MHz ≤ f ≤ 2000 MHz	-5	_	_	dBm
BLE reference signal at 2426MHz. Interferer is CW in OOB range ²		Interferer frequency 2003 MHz ≤ f ≤ 2399 MHz	-24	_	_	dBm
		Interferer frequency 2484 MHz ≤ f ≤ 2997 MHz	-10	_	_	dBm
		Interferer frequency 3 GHz ≤ f ≤ 6 GHz	-10	_	_	dBm
		Interferer frequency 6 GHz ≤ f ≤ 12.75 GHz	-17	_	_	dBm

- 1. Reference signal is defined 2GFSK at -67 dBm, Modulation index = 0.5, BT = 0.5, Bit rate = 1 Mbps, desired data = PRBS9; interferer data = PRBS15; frequency accuracy better than 1 ppm.
- 2. Interferer max power limited by equipment capabilities and path loss. Minimum specified at 25 °C.

4.1.9.5 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 85%.

Table 4.16. RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Transmit 6dB bandwidth	TXBW	10 dBm	_	1395	_	kHz
Power spectral density limit	PSD _{LIMIT}	Per FCC part 15.247 at 10 dBm		-12.7	_	dBm/ 3kHz
		Per FCC part 15.247 at 20 dBm	_	-4.7	_	dBm/ 3kHz
		Per ETSI 300.328 at 10 dBm/1 MHz	_	10	_	dBm
Occupied channel bandwidth per ETSI EN300.328	OCP _{ETSI328}	99% BW at highest and lowest channels in band, 10 dBm	_	2.1	_	MHz
Emissions of harmonics out- of-band, per FCC part 15.247	SPUR _{HRM_FCC}	2nd,3rd, 5, 6, 8, 9,10 harmonics; continuous transmission of modulated carrier	_	-47	_	dBm
Spurious emissions out-of- band, excluding harmonics captured in SPUR _{HARM,FCC} . Emissions taken at POUT _{MAX} , PAVDD connec-	SPUR _{OOB_FCC}	Per FCC part 15.205/15.209, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Restricted Bands ^{1 2 3}	_	-47	_	dBm
ted to external 3.3 V supply		Per FCC part 15.247, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Non-Restricted Bands	_	-26	_	dBc
Spurious emissions out-of- band; per ETSI 300.328	SPUR _{ETSI328}	[2400-BW to 2400] MHz, [2483.5 to 2483.5+BW] MHz	_	-16	_	dBm
		[2400-2BW to 2400-BW] MHz, [2483.5+BW to 2483.5+2BW] MHz per ETSI 300.328	_	-26	_	dBm
Spurious emissions per ETSI EN300.440	SPUR _{ETSI440}	47-74 MHz,87.5-108 MHz, 174-230 MHz, 470-862 MHz	_	-60	_	dBm
		25-1000 MHz	_	-42	_	dBm
		1-12 GHz	_	-36	_	dBm

- 1. For 2472 MHz, 1.3 dB of power backoff is used to achieve this value.
- $2.\,\mbox{For 2474 MHz},\,3.8\mbox{ dB}$ of power backoff is used to achieve this value.
- 3. For 2476 MHz, 7 dB of power backoff is used to achieve this value.
- 4. For 2478 MHz, 11.2 dB of power backoff is used to achieve this value.

4.1.9.6 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz¹.

Table 4.17. RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Max usable receiver input level, 0.1% BER	SAT	Signal is reference signal ² . Packet length is 20 bytes.	_	10	_	dBm
Sensitivity, 0.1% BER	SENS	Signal is reference signal ² . Using DC-DC converter.	_	-91.5	_	dBm
Signal to co-channel interferer, 0.1% BER	C/I _{CC}	Desired signal 3 dB above reference sensitivity.	_	7.3	_	dB
N+1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -67 dBm	C/I ₁₊	Interferer is reference signal at +2 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-10.5	_	dB
N-1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -67 dBm	C/I ₁₋	Interferer is reference signal at -2 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-14.3	_	dB
Alternate selectivity, 0.1% BER, with allowable excep- tions. Desired is reference signal at -67 dBm	C/I ₂	Interferer is reference signal at ± 4 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-40.3	_	dB
Alternate selectivity, 0.1% BER, with allowable excep- tions. Desired is reference signal at -67 dBm	C/I ₃	Interferer is reference signal at ± 6 MHz offset. Desired frequency 2404 MHz ≤ Fc ≤ 2480 MHz	_	-42.2	_	dB
Selectivity to image frequency, 0.1% BER. Desired is reference signal at -67 dBm	C/I _{IM}	Interferer is reference signal at image frequency with 1 MHz precision	_	-10.5	_	dB
Selectivity to image frequency ± 2 MHz, 0.1% BER. Desired is reference signal at -67 dBm	C/I _{IM+1}	Interferer is reference signal at image frequency ± 2 MHz with 2 MHz precision	_	-39	_	dB

- 1. For the BLE 2Mbps in-band blocking performance, there may be up to 5 spurious response channels where the requirement of 30.8% PER is not met and therefore an exception will need to be taken for each of these frequencies to meet the requirements of the BLE standard.
- 2. Reference signal is defined 2GFSK at -67 dBm, Modulation index = 0.5, BT = 0.5, Bit rate = 2 Mbps, desired data = PRBS9; interferer data = PRBS15; frequency accuracy better than 1 ppm.

4.1.9.7 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 500 kbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 85%.

Table 4.18. RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 500 kbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Transmit 6dB bandwidth	TXBW	10 dBm	_	761	_	kHz
Power spectral density limit	PSD _{LIMIT}	Per FCC part 15.247 at 10 dBm	_	-8.9	_	dBm/ 3kHz
		Per FCC part 15.247 at 20 dBm ¹	_	_	8	dBm/ 3kHz
Occupied channel bandwidth per ETSI EN300.328	OCP _{ETSI328}	99% BW at highest and lowest channels in band, 10 dBm	_	1.1	_	MHz
Emissions of harmonics out- of-band, per FCC part 15.247	SPUR _{HRM_FCC}	2nd,3rd, 5, 6, 8, 9,10 harmonics; continuous transmission of modulated carrier	_	-47	_	dBm
Spurious emissions out-of- band, excluding harmonics captured in SPUR _{HARM,FCC} . Emissions taken at	SPUR _{OOB_FCC}	Per FCC part 15.205/15.209, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Restricted Bands ²	_	-47	_	dBm
POUT _{MAX} , PAVDD connected to external 3.3 V supply		Per FCC part 15.247, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Non-Restricted Bands	_	-26	_	dBc
Spurious emissions out-of-band; per ETSI 300.328	SPUR _{ETSI328}	[2400-BW to 2400] MHz, [2483.5 to 2483.5+BW] MHz	_	-16	_	dBm
		[2400-2BW to 2400-BW] MHz, [2483.5+BW to 2483.5+2BW] MHz per ETSI 300.328	_	-26	_	dBm
Spurious emissions per ETSI EN300.440	SPUR _{ETSI440}	47-74 MHz,87.5-108 MHz, 174-230 MHz, 470-862 MHz	_	-60	_	dBm
		25-1000 MHz	_	-42	_	dBm
		1-12 GHz	_	-36	_	dBm

- 1. Output power limited to 14 dBm to ensure compliance with FCC specifications.
- 2. For 2476 MHz, 1.2 dB of power backoff is used to achieve this value.
- 3. For 2478 MHz, 5.8 dB of power backoff is used to achieve this value.

4.1.9.8 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 500 kbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz.

Table 4.19. RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 500 kbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Max usable receiver input level, 0.1% BER	SAT	Signal is reference signal ¹ . Packet length is 20 bytes.	_	10	_	dBm
Sensitivity, 0.1% BER	SENS	Signal is reference signal ¹ . Using DC-DC converter.	_	-99	_	dBm
N+1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -72 dBm	C/I ₁₊	Interferer is reference signal at +1 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-9	_	dB
N-1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -72 dBm	C/I ₁₋	Interferer is reference signal at -1 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-9	_	dB
Alternate selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -72 dBm	C/I ₂	Interferer is reference signal at ± 2 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-50.8	_	dB
Selectivity to image frequency, 0.1% BER. Desired is reference signal at -72 dBm	C/I _{IM}	Interferer is reference signal at image frequency with 1 MHz precision	_	-46.2	_	dB
Selectivity to image frequency ± 1 MHz, 0.1% BER. Desired is reference signal at -72 dBm	C/I _{IM+1}	Interferer is reference signal at image frequency ± 1 MHz with 1 MHz precision	_	-56.1	_	dB

^{1.} Reference signal is defined 2GFSK at -72 dBm, Modulation index = 0.5, BT = 0.5, Bit rate = 500 kbps, desired data = PRBS9; interferer data = PRBS15; frequency accuracy better than 1 ppm.

4.1.9.9 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 125 kbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 85%.

Table 4.20. RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 125 kbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Transmit 6dB bandwidth	TXBW	10 dBm	_	756	_	kHz
Power spectral density limit	PSD _{LIMIT}	Per FCC part 15.247 at 10 dBm	_	-9	_	dBm/ 3kHz
		Per FCC part 15.247 at 20 dBm ¹	_	_	8	dBm/ 3kHz
Occupied channel bandwidth per ETSI EN300.328	OCP _{ETSI328}	99% BW at highest and lowest channels in band, 10 dBm	_	1.1	_	MHz
Emissions of harmonics out- of-band, per FCC part 15.247	SPUR _{HRM_FCC}	2nd,3rd, 5, 6, 8, 9,10 harmonics; continuous transmission of modu- lated carrier	_	-47	_	dBm
Spurious emissions out-of-band, excluding harmonics captured in SPUR _{HARM,FCC} . Emissions taken at POUT _{MAX} , PAVDD connected to external 3.3 V supply	SPUR _{OOB_FCC}	Per FCC part 15.205/15.209, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Restricted Bands ²	_	-47	_	dBm
		Per FCC part 15.247, Above 2.483 GHz or below 2.4 GHz; continuous transmission of CW carrier, Non-Restricted Bands	_	-26	_	dBc
Spurious emissions out-of- band; per ETSI 300.328	SPUR _{ETSI328}	[2400-BW to 2400] MHz, [2483.5 to 2483.5+BW] MHz	_	-16	_	dBm
		[2400-2BW to 2400-BW] MHz, [2483.5+BW to 2483.5+2BW] MHz per ETSI 300.328	_	-26	_	dBm
Spurious emissions per ETSI EN300.440	SPUR _{ETSI440}	47-74 MHz,87.5-108 MHz, 174-230 MHz, 470-862 MHz	_	-60	_	dBm
		25-1000 MHz	_	-42	_	dBm
		1-12 GHz	_	-36	_	dBm

- 1. Output power limited to 14 dBm to ensure compliance with FCC specifications.
- 2. For 2476 MHz, 1.2 dB of power backoff is used to achieve this value.
- 3. For 2478 MHz, 5.8 dB of power backoff is used to achieve this value.

4.1.9.10 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 125 kbps Data Rate

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4MHz. RF center frequency 2.45 GHz.

Table 4.21. RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 125 kbps Data Rate

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Max usable receiver input level, 0.1% BER	SAT	Signal is reference signal ¹ . Packet length is 20 bytes.	_	10	_	dBm
Sensitivity, 0.1% BER	SENS	Signal is reference signal ¹ . Using DC-DC converter.	_	-103.3	_	dBm
N+1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -79 dBm	C/I ₁₊	Interferer is reference signal at +1 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-13.6	_	dB
N-1 adjacent channel selectivity, 0.1% BER, with allowable exceptions. Desired is reference signal at -79 dBm	C/I ₁₋	Interferer is reference signal at -1 MHz offset. Desired frequency 2402 MHz ≤ Fc ≤ 2480 MHz	_	-13.1	_	dB
Selectivity to image frequency, 0.1% BER. Desired is reference signal at -79 dBm	C/I _{IM}	Interferer is reference signal at image frequency with 1 MHz precision	_	-49.7	_	dB
Selectivity to image frequency ± 1 MHz, 0.1% BER. Desired is reference signal at -79 dBm	C/I _{IM+1}	Interferer is reference signal at image frequency ± 1 MHz with 1 MHz precision	_	-59.6	_	dB

^{1.} Reference signal is defined 2GFSK at -79 dBm, Modulation index = 0.5, BT = 0.5, Bit rate = 125 kbps, desired data = PRBS9; interferer data = PRBS15; frequency accuracy better than 1 ppm.

4.1.9.11 RF Transmitter Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz. Maximum duty cycle of 66%.

Table 4.22. RF Transmitter Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Error vector magnitude (off-set EVM), per 802.15.4-2011	EVM	Average across frequency. Signal is DSSS-OQPSK reference packet ¹	_	3.8	_	% rms
Power spectral density limit	PSD _{LIMIT}	Relative, at carrier ± 3.5 MHz, output power at POUT _{MAX}	_	-26	_	dBc/ 100kHz
		Absolute, at carrier ± 3.5 MHz, output power at POUT _{MAX} ²	_	-36	_	dBm/ 100kHz
		Per FCC part 15.247, output power at POUT _{MAX}	_	-4.0	_	dBm/ 3kHz
		ETSI	_	12.1	_	dBm
Occupied channel bandwidth per ETSI EN300.328	OCP _{ETSI328}	99% BW at highest and lowest channels in band		2.25	_	MHz
Spurious emissions of harmonics in restricted bands per FCC Part 15.205/15.209, Emissions taken at POUT _{MAX} , PAVDD connected to external 3.3 V supply, Test Frequency is 2450 MHz	SPUR _{HRM_FCC_} R	Continuous transmission of modulated carrier	_	-45.8	_	dBm
Spurious emissions of harmonics in non-restricted bands per FCC Part 15.247/15.35, Emissions taken at POUT _{MAX} , PAVDD connected to external 3.3 V supply, Test Frequency is 2450 MHz	SPUR _{HRM_FCC_} NRR	Continuous transmission of modulated carrier	_	-26	_	dBc
Spurious emissions out-of- band (above 2.483 GHz or below 2.4 GHz) in restricted	SPUR _{OOB_FCC_}	Restricted bands 30-88 MHz; continuous transmission of modulated carrier	_	-61	_	dBm
bands, per FCC part 15.205/15.209, Emissions taken at POUT _{MAX} , PAVDD connected to external 3.3 V supply, Test Frequency = 2450 MHz		Restricted bands 88-216 MHz; continuous transmission of modulated carrier	_	-58	_	dBm
		Restricted bands 216-960 MHz; continuous transmission of modulated carrier	_	-55	_	dBm
		Restricted bands >960 MHz; continuous transmission of modulated carrier ^{3 4}	_	-47	_	dBm

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Spurious emissions out-of- band in non-restricted bands per FCC Part 15.247, Emis- sions taken at POUT _{MAX} , PAVDD connected to exter- nal 3.3 V supply, Test Fre- quency = 2450 MHz	SPUR _{OOB_FCC_} NR	Above 2.483 GHz or below 2.4 GHz; continuous transmission of modulated carrier	_	-26	_	dBc
Spurious emissions out-of- band; per ETSI 300.328 ⁵	SPUR _{ETSI328}	[2400-BW to 2400], [2483.5 to 2483.5+BW];	_	-16	_	dBm
		[2400-2BW to 2400-BW], [2483.5+BW to 2483.5+2BW]; per ETSI 300.328	_	-26	_	dBm
Spurious emissions per ETSI EN300.440 ⁵	SPUR _{ETSI440}	47-74 MHz,87.5-108 MHz, 174-230 MHz, 470-862 MHz	_	-60	_	dBm
		25-1000 MHz, excluding above frequencies	_	-42	_	dBm
		1G-14G	_	-36	_	dBm

- 1. Reference packet is defined as 20 octet PSDU, modulated according to 802.15.4-2011 DSSS-OQPSK in the 2.4GHz band, with pseudo-random packet data content.
- 2. For 2415 MHz, 2 dB of power backoff is used to achieve this value.
- 3. For 2475 MHz, 2 dB of power backoff is used to achieve this value.
- 4. For 2480 MHz, 13 dB of power backoff is used to achieve this value.
- 5. Specified at maximum power output level of 10 dBm.

4.1.9.12 RF Receiver Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = PAVDD. RFVDD and PAVDD path is filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 2.45 GHz.

Table 4.23. RF Receiver Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Max usable receiver input level, 1% PER	SAT	Signal is reference signal ¹ . Packet length is 20 octets.	_	10	_	dBm
Sensitivity, 1% PER	SENS	Signal is reference signal. Packet length is 20 octets. Using DC-DC converter.	_	-102.7	_	dBm
		Signal is reference signal. Packet length is 20 octets. Without DC-DC converter.	_	-102.7	_	dBm
Co-channel interferer rejection, 1% PER	CCR	Desired signal 3 dB above sensitivity limit	_	-4.6	_	dB
High-side adjacent channel rejection, 1% PER. Desired is reference signal at 3dB	ACR _{P1}	Interferer is reference signal at +1 channel-spacing.	_	40.7	_	dB
above reference sensitivity level ²		Interferer is filtered reference signal ³ at +1 channel-spacing.	_	47	_	dB
		Interferer is CW at +1 channel-spacing ⁴ .	_	54.3	_	dB
Low-side adjacent channel rejection, 1% PER. Desired	ACR _{M1}	Interferer is reference signal at -1 channel-spacing.	_	40.8	_	dB
is reference signal at 3dB above reference sensitivity level ²		Interferer is filtered reference signal ³ at -1 channel-spacing.	_	47.5	_	dB
		Interferer is CW at -1 channel-spacing.	_	56.5	_	dB
Alternate channel rejection, 1% PER. Desired is refer-	ACR ₂	Interferer is reference signal at ± 2 channel-spacing	_	51.5	_	dB
ence signal at 3dB above reference sensitivity level ²		Interferer is filtered reference signal ³ at ± 2 channel-spacing		53.7	_	dB
		Interferer is CW at ± 2 channel- spacing	_	62.4	_	dB
Image rejection , 1% PER, Desired is reference signal at 3dB above reference sensi- tivity level ²	IR	Interferer is CW in image band ⁴	_	50.4	_	dB
Blocking rejection of all other channels. 1% PER, Desired	BLOCK	Interferer frequency < Desired frequency - 3 channel-spacing	_	58.5	_	dB
is reference signal at 3dB above reference sensitivity level ² . Interferer is reference signal		Interferer frequency > Desired frequency + 3 channel-spacing	_	56.4	_	dB
Blocking rejection of 802.11g signal centered at +12MHz or -13MHz ⁵	BLOCK _{80211G}	Desired is reference signal at 6dB above reference sensitivity level ²	_	50	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	over RSSI _{MIN} to RSSI _{MAX}	_	0.25	_	dB
RSSI accuracy in the linear region as defined by 802.15.4-2003	RSSI _{LIN}		_	+/-6	_	dB

- 1. Reference signal is defined as O-QPSK DSSS per 802.15.4, Frequency range = 2400-2483.5 MHz, Symbol rate = 62.5 ksymbols/s.
- 2. Reference sensitivity level is -85 dBm.
- 3. Filter is characterized as a symmetric bandpass centered on the adjacent channel having a 3dB bandwidth of 4.6 MHz and stop-band rejection better than 26 dB beyond 3.15 MHz from the adjacent carrier.
- 4. Due to low-IF frequency, there is some overlap of adjacent channel and image channel bands. Adjacent channel CW blocker tests place the Interferer center frequency at the Desired frequency ± 5 MHz on the channel raster, whereas the image rejection test places the CW interferer near the image frequency of the Desired signal carrier, regardless of the channel raster.
- 5. This is an IEEE 802.11b/g ERP-PBCC 22 MBit/s signal as defined by the IEEE 802.11 specification and IEEE 802.11g addendum.

4.1.10 Sub-GHz RF Transceiver Characteristics

4.1.10.1 Sub-GHz RF Transmitter characteristics for 915 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 915 MHz.

Table 4.24. Sub-GHz RF Transmitter characteristics for 915 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		902	_	930	MHz
Maximum TX Power ¹	POUT _{MAX}	External PA supply = 3.3V, 20 dBm output power setting	18	19.8	23.3	dBm
		External PA supply connected to DC-DC output, 14 dBm output power setting	12.6	14.2	16.1	dBm
Minimum active TX Power	POUT _{MIN}		_	-45.5	_	dBm
Output power step size	POUT _{STEP}	output power > 0 dBm	_	0.5	_	dB
Output power variation vs supply at POUT _{MAX}	POUT _{VAR_V}	1.8 V < V _{VREGVDD} < 3.3 V, External PA supply = 3.3 V, T = 25 °C	_	4.8	_	dB
		1.8 V < V _{VREGVDD} < 3.3 V, External PA supply connected to DC-DC output, T = 25 °C	_	1.9	_	dB
Output power variation vs temperature, peak to peak	POUT _{VAR_T}	-40 to +85 °C with External PA supply = 3.3 V	_	0.6	1.3	dB
		-40 to +125 °C with External PA supply = 3.3 V	_	0.8	1.6	dB
		-40 to +85 °C with External PA supply connected to DC-DC output	_	0.7	1.4	dB
		-40 to +125 °C with External PA supply connected to DC-DC output	_	1.0	1.9	dB
Output power variation vs RF frequency	POUT _{VAR_F}	External PA supply = 3.3 V, T = 25 °C	_	0.2	0.6	dB
		External PA supply connected to DC-DC output, T = 25 °C	_	0.3	0.6	dB
Spurious emissions of harmonics at 20 dBm output	SPUR _{HARM_FCC} _20	In restricted bands, per FCC Part 15.205 / 15.209	_	-45	-42	dBm
power, Conducted measure- ment, 20dBm match, Exter- nal PA supply = 3.3V, Test Frequency = 915 MHz		In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Spurious emissions out-of- band at 20 dBm output pow-	SPUR _{OOB_FCC_}	In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
er, Conducted measurement, 20dBm match, External PA supply = 3.3V, Test Frequen-		In restricted bands (30-88 MHz), per FCC Part 15.205 / 15.209	_	-52	-46	dBm
cy = 915 MHz		In restricted bands (88-216 MHz), per FCC Part 15.205 / 15.209	_	-61	-56	dBm
		In restricted bands (216-960 MHz), per FCC Part 15.205 / 15.209	_	-58	-52	dBm
		In restricted bands (>960 MHz), per FCC Part 15.205 / 15.209	_	-47	-42	dBm
Spurious emissions of har- monics at 14 dBm output	SPUR _{HARM_FCC}	In restricted bands, per FCC Part 15.205 / 15.209	_	-47	-42	dBm
power, Conducted measure- ment, 14dBm match, Exter- nal PA supply connected to DC-DC output, Test Fre- quency = 915 MHz		In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
Spurious emissions out-of- band at 14 dBm output pow-	SPUR _{OOB_FCC_} 14	In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
er, Conducted measurement, 14dBm match, External PA supply connected to DC-DC		In restricted bands (30-88 MHz), per FCC Part 15.205 / 15.209	_	-52	-46	dBm
output, Test Frequency = 915 MHz		In restricted bands (88-216 MHz), per FCC Part 15.205 / 15.209	_	-61	-56	dBm
		In restricted bands (216-960 MHz), per FCC Part 15.205 / 15.209	_	-58	-52	dBm
		In restricted bands (>960 MHz), per FCC Part 15.205 / 15.209		-45	-42	dBm
Error vector magnitude (offset EVM), per 802.15.4-2011	EVM	Signal is DSSS-OQPSK reference packet. Modulated according to 802.15.4-2011 DSSS-OQPSK in the 915MHz band, with pseudorandom packet data content. External PA supply = 3.3V.	_	1.0	2.8	%rms
Power spectral density limit	PSD	Relative, at carrier ± 1.2 MHz. Average spectral power shall be measured using a 100kHz resolution bandwidth. The reference level shall be the highest average spectral power measured within ± 600kHz of the carrier frequency. External PA supply = 3.3V.	_	-37.1	-24.8	dBc/ 100kHz
		Absolute, at carrier ± 1.2 MHz. Average spectral power shall be measured using a 100kHz resolution bandwidth. External PA supply = 3.3V.	_	-24.2	-20	dBm/ 100kHz

1. Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.2 Sub-GHz RF Receiver Characteristics for 915 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 915 MHz.

Table 4.25. Sub-GHz RF Receiver Characteristics for 915 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Tuning frequency range	F _{RANGE}		902	_	930	MHz
Max usable input level, 0.1% BER	SAT _{500K}	Desired is reference 500 kbps GFSK signal ¹	_	10	_	dBm
Sensitivity	SENS	Desired is reference 4.8 kbps OOK signal ² , 20% PER, T ≤ 85 °C	_	-105.2	-100.7	dBm
		Desired is reference 4.8 kbps OOK signal ² , 20% PER, T > 85 °C	_	_	-99.5	dBm
		Desired is reference 600 bps GFSK signal ³ , 0.1% BER	_	-126.2	_	dBm
		Desired is reference 50 kbps GFSK signal ⁴ , 0.1% BER, T ≤ 85 °C	_	-108.2	-104.2	dBm
		Desired is reference 50 kbps GFSK signal ⁴ , 0.1% BER, T > 85 °C	_	_	-103.1	dBm
		Desired is reference 100 kbps GFSK signal ⁵ , 0.1% BER, T ≤ 85 °C	_	-105.1	-101.5	dBm
		Desired is reference 100 kbps GFSK signal ⁵ , 0.1% BER, T > 85 °C	_	_	-101.3	dBm
		Desired is reference 500 kbps GFSK signal ¹ , 0.1% BER, T ≤ 85 °C	_	-98.2	-93.2	dBm
		Desired is reference 500 kbps GFSK signal ¹ , 0.1% BER, T > 85 °C	_	_	-93.1	dBm
		Desired is reference 400 kbps 4GFSK signal ⁶ , 1% PER, T ≤ 85 °C	_	-95.2	-91	dBm
		Desired is reference 400 kbps 4GFSK signal ⁶ , 1% PER, T > 85 °C	_	_	-91	dBm
		Desired is reference O-QPSK DSSS signal ⁷ , 1% PER, Payload length is 20 octets	_	-100.1	_	dBm
Level above which RFSENSE will trigger ⁸	RFSENSE _{TRIG}	CW at 915 MHz	_	-28.1	_	dBm
Level below which RFSENSE will not trigger ⁸	RFSENSE _{THRES}	CW at 915 MHz	_	-50	_	dBm

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Adjacent channel selectivity, Interferer is CW at ± 1 × channel-spacing	C/I ₁	Desired is 4.8 kbps OOK signal ² at 3dB above sensitivity level, 20% PER	_	48.1	_	dB
		Desired is 600 bps GFSK signal ³ at 3dB above sensitivity level, 0.1% BER	_	71.4	_	dB
		Desired is 50 kbps GFSK signal ⁴ at 3dB above sensitivity level, 0.1% BER	_	49.8	_	dB
		Desired is 100 kbps GFSK signal ⁵ at 3dB above sensitivity level, 0.1% BER	_	51.1	_	dB
		Desired is 500 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	48.1	_	dB
		Desired is 400 kbps 4GFSK signal ⁶ at 3dB above sensitivity level, 0.1% BER	_	41.4	_	dB
		Desired is reference O-QPSK DSSS signal ⁷ at 3dB above sensi- tivity level, 1% PER	_	49.1	_	dB
Alternate channel selectivity, Interferer is CW at ± 2 × channel-spacing	C/I ₂	Desired is 4.8 kbps OOK signal ² at 3dB above sensitivity level, 20% PER	_	56.3	_	dB
		Desired is 600 bps GFSK signal ³ at 3dB above sensitivity level, 0.1% BER	_	74.7	_	dB
		Desired is 50 kbps GFSK signal ⁴ at 3dB above sensitivity level, 0.1% BER	_	55.8	_	dB
		Desired is 100 kbps GFSK signal ⁵ at 3dB above sensitivity level, 0.1% BER	_	56.4	_	dB
		Desired is 500 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	51.8	_	dB
		Desired is 400 kbps 4GFSK signal ⁶ at 3dB above sensitivity level, 0.1% BER	_	46.8	_	dB
		Desired is reference O-QPSK DSSS signal ⁷ at 3dB above sensitivity level, 1% PER	_	57.7	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Image rejection, Interferer is CW at image frequency	C/I _{IMAGE}	Desired is 4.8 kbps OOK signal ² at 3dB above sensitivity level, 20% PER	_	48.4	_	dB
		Desired is 50 kbps GFSK signal ⁴ at 3dB above sensitivity level, 0.1% BER	_	54.9	_	dB
		Desired is 100 kbps GFSK signal ⁵ at 3dB above sensitivity level, 0.1% BER	_	49.1	_	dB
		Desired is 500 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	47.9	_	dB
		Desired is 400 kbps 4GFSK signal ⁶ at 3dB above sensitivity level, 0.1% BER	_	42.8	_	dB
	ſ	Desired is reference O-QPSK DSSS signal ⁷ at 3dB above sensi- tivity level, 1% PER	_	48.9	_	dB
Blocking selectivity, 0.1%	C/I _{BLOCKER}	Interferer CW at Desired ± 1 MHz	_	58.7	_	dB
BER. Desired is 100 kbps GFSK signal at 3dB above		Interferer CW at Desired ± 2 MHz	_	62.5	_	dB
sensitivity level		Interferer CW at Desired ± 10 MHz	_	76.4	_	dB
Intermod selectivity, 0.1% BER. CW interferers at 400 kHz and 800 kHz offsets	C/I _{IM}	Desired is 100 kbps GFSK signal ⁵ at 3dB above sensitivity level	_	45	_	dB
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	Over RSSI _{MIN} to RSSI _{MAX} range	_	0.25	_	dBm
Max spurious emissions dur-	SPUR _{RX_FCC}	216-960 MHz	_	-55	-49.2	dBm
ing active receive mode, per FCC Part 15.109(a)		Above 960 MHz	_	-47	-41.2	dBm
Max spurious emissions dur-	SPUR _{RX_ARIB}	Below 710 MHz, RBW=100kHz	_	-60	-54	dBm
ing active receive mode,per ARIB STD-T108 Section 3.3		710-900 MHz, RBW=1MHz	_	-61	-55	dBm
		900-915 MHz, RBW=100kHz	_	-61	-55	dBm
		915-930 MHz, RBW=100kHz	_	-61	-55	dBm
		930-1000 MHz, RBW=100kHz		-60	-54	dBm
		Above 1000 MHz, RBW=1MHz	_	-53	-47	dBm

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit

- 1. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 175 kHz, RX channel BW = 835.076 kHz, channel spacing = 1 MHz.
- 2. Definition of reference signal is 4.8 kbps OOK, RX channel BW = 306.036 kHz, channel spacing = 500 kHz.
- 3. Definition of reference signal is 600 bps 2GFSK, BT=0.5, Δf = 0.3 kHz, RX channel BW = 1.2 kHz, channel spacing = 300 kHz.
- 4. Definition of reference signal is 50 kbps 2GFSK, BT=0.5, Δf = 25 kHz, RX channel BW = 99.012 kHz, channel spacing = 200 kHz.
- 5. Definition of reference signal is 100 kbps 2GFSK, BT=0.5, Δf = 50 kHz, RX channel BW = 198.024 kHz, channel spacing = 400 kHz.
- 6. Definition of reference signal is 400 kbps 4GFSK, BT=0.5, inner deviation = 33.3 kHz, RX channel BW = 368.920 kHz, channel spacing = 600 kHz.
- 7. Definition of reference signal is O-QPSK DSSS per 802.15.4, Frequency Range = 902-928 MHz, Data rate = 250 kbps, 16-chip PN sequence mapping.
- 8. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.3 Sub-GHz RF Transmitter characteristics for 868 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 868 MHz.

Table 4.26. Sub-GHz RF Transmitter characteristics for 868 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		863	_	876	MHz
Maximum TX Power ¹	POUT _{MAX}	External PA supply = 3.3V, 20 dBm output power setting	17.1	19.3	22.9	dBm
		External PA supply connected to DC-DC output, 14 dBm output power setting	11.4	13.7	16.5	dBm
Minimum active TX Power	POUT _{MIN}		_	-43.5	_	dBm
Output power step size	POUT _{STEP}	output power > 0 dBm	_	0.5	_	dB
Output power variation vs supply at POUT _{MAX}	POUT _{VAR_V}	1.8 V < V _{VREGVDD} < 3.3 V, External PA supply = 3.3 V, T = 25 °C	_	5	_	dB
		1.8 V < V _{VREGVDD} < 3.3 V, External PA supply connected to DC-DC output, T = 25 °C	_	2	_	dB
Output power variation vs temperature, peak to peak	POUT _{VAR_T}	-40 to +85 °C with External PA supply = 3.3 V	_	0.6	0.9	dB
		-40 to +125 °C with External PA supply = 3.3 V	_	0.8	1.3	dB
		-40 to +85 °C with External PA supply connected to DC-DC output	_	0.5	1.2	dB
		-40 to +125 °C with External PA supply connected to DC-DC output	_	0.7	1.5	dB
Output power variation vs RF frequency	POUT _{VAR_F}	External PA supply = 3.3 V, T = 25 °C	_	0.2	0.6	dB
		External PA supply connected to DC-DC output, T = 25 °C	_	0.2	0.8	dB
Spurious emissions of harmonics, Conducted measurement, External PA supply connected to DC-DC output, Test Frequency = 868 MHz	SPUR _{HARM_ETSI}	Per ETSI EN 300-220, Section 7.8.2.1	_	-35	-30	dBm
Spurious emissions out-of- band, Conducted measure- ment, External PA supply connected to DC-DC output, Test Frequency = 868 MHz	SPUR _{OOB_ETSI}	Per ETSI EN 300-220, Section 7.8.2.1 (47-74 MHz, 87.5-118 MHz, 174-230 MHz, and 470-862 MHz)	_	-59	-54	dBm
		Per ETSI EN 300-220, Section 7.8.2.1 (other frequencies below 1 GHz)	_	-42	-36	dBm
		Per ETSI EN 300-220, Section 7.8.2.1 (frequencies above 1 GHz)	_	-36	-30	dBm

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Error vector magnitude (offset EVM), per 802.15.4-2015	EVM	Signal is DSSS-BPSK reference packet. Modulated according to 802.15.4-2015 DSSS-BPSK in the 868MHz band, with pseudo-random packet data content. External PA supply connected to external 3.3V supply	_	5.7	_	%rms

^{1.} Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.4 Sub-GHz RF Receiver Characteristics for 868 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 868 MHz.

Table 4.27. Sub-GHz RF Receiver Characteristics for 868 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Tuning frequency range	F _{RANGE}		863	_	876	MHz
Max usable input level, 0.1% BER	SAT _{2k4}	Desired is reference 2.4 kbps GFSK signal ¹	_	10	_	dBm
Max usable input level, 0.1% BER	SAT _{38k4}	Desired is reference 38.4 kbps GFSK signal ²	_	10	_	dBm
Sensitivity	SENS	Desired is reference 2.4 kbps GFSK signal ¹ , 0.1% BER	_	-120.6	_	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T ≤ 85 °C	_	-109.5	-105.4	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T > 85 °C	_	_	-105.2	dBm
		Desired is reference 500 kbps GFSK signal ³ , 0.1% BER	_	-96.4	_	dBm
Level above which RFSENSE will trigger ⁴	RFSENSE _{TRIG}	CW at 868 MHz	_	-28.1	_	dBm
Level below which RFSENSE will not trigger ⁴	RFSENSE _{THRES}	CW at 868 MHz	_	-50	_	dBm
Adjacent channel selectivity, Interferer is CW at ± 1 × channel-spacing	C/I ₁	Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	44.5	56.9	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	35.4	43	_	dB
Alternate channel selectivity, Interferer is CW at ± 2 × channel-spacing	C/I ₂	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	56.8	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	48.2	_	dB
Image rejection, Interferer is CW at image frequency	C/I _{IMAGE}	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	50.2	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	48.7	_	dB
Blocking selectivity, 0.1%	C/I _{BLOCKER}	Interferer CW at Desired ± 1 MHz	_	72.1	_	dB
BER. Desired is 2.4 kbps GFSK signal ¹ at 3 dB above		Interferer CW at Desired ± 2 MHz	_	77.5	_	dB
sensitivity level		Interferer CW at Desired ± 10 MHz	_	90.4	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	Over RSSI _{MIN} to RSSI _{MAX} range	_	0.25	_	dBm
	SPUR _{RX}	30 MHz to 1 GHz	_	-63	-57	dBm
ing active receive mode		1 GHz to 12 GHz	_	-53	-47	dBm

- 1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.797 kHz, channel spacing = 12.5 kHz.
- 2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.
- 3. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 125 kHz, RX channel BW = 753.320 kHz.
- 4. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.5 Sub-GHz RF Transmitter characteristics for 490 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 490 MHz.

Table 4.28. Sub-GHz RF Transmitter characteristics for 490 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		470	_	510	MHz
Maximum TX Power ¹	POUT _{MAX}	External PA supply = 3.3V	18.1	20.3	23.7	dBm
Minimum active TX Power	POUT _{MIN}			-44.9	_	dBm
Output power step size	POUT _{STEP}	output power > 0 dBm	_	0.5	_	dB
Output power variation vs supply, peak to peak	POUT _{VAR_V}	at 20 dBm;1.8 V < V _{VREGVDD} < 3.3 V, External PA supply connected directly to external supply, T = 25 °C	_	4.3	_	dB
Output power variation vs	POUT _{VAR_T}	-40 to +85 °C at 20 dBm	_	0.2	0.9	dB
temperature, peak to peak		-40 to +125 °C at 20 dBm	_	0.3	1.3	dB
Output power variation vs RF frequency	POUT _{VAR_F}	T = 25 °C	_	0.2	0.4	dB
Harmonic emissions, 20 dBm output power setting, 490 MHz	SPUR _{HARM_CN}	Per China SRW Requirement, Section 2.1, frequencies below 1GHz	_	-40	-36	dBm
		Per China SRW Requirement, Section 2.1, frequencies above 1GHz	_	-36	-30	dBm
Spurious emissions, 20 dBm output power setting, 490 MHz	SPUR _{OOB_CN}	Per China SRW Requirement, Section 3 (48.5-72.5MHz, 76-108MHz, 167-223MHz, 470-556MHz, and 606-798MHz)	_	-54	_	dBm
		Per China SRW Requirement, Section 2.1 (other frequencies below 1GHz)	_	-42	_	dBm
		Per China SRW Requirement, Section 2.1 (frequencies above 1GHz)	_	-36	_	dBm

^{1.} Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.6 Sub-GHz RF Receiver Characteristics for 490 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 490 MHz.

Table 4.29. Sub-GHz RF Receiver Characteristics for 490 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Tuning frequency range	F _{RANGE}		470	_	510	MHz
Max usable input level, 0.1% BER	SAT _{2k4}	Desired is reference 2.4 kbps GFSK signal ¹	_	10	_	dBm
Max usable input level, 0.1% BER	SAT _{38k4}	Desired is reference 38.4 kbps GFSK signal ²	_	10	_	dBm
Sensitivity	SENS	Desired is reference 2.4 kbps GFSK signal ¹ , 0.1% BER	_	-122.2	_	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T ≤ 85 °C	_	-111.4	-108.9	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T > 85 °C	_	_	-107.9	dBm
		Desired is reference 10 kbps GFSK signal ³ , 0.1% BER, T ≤ 85 °C	_	-116.8	-113.9	dBm
		Desired is reference 10 kbps GFSK signal ³ , 0.1% BER, T > 85 °C	_	_	-113.2	dBm
		Desired is reference 100 kbps GFSK signal ⁴ , 0.1% BER, T ≤ 85 °C	_	-107.3	-104.7	dBm
		Desired is reference 100 kbps GFSK signal ⁴ , 0.1% BER, T > 85 °C	_	_	-104	dBm
Level above which RFSENSE will trigger ⁵	RFSENSE _{TRIG}	Desired is reference 100 kbps GFSK signal ⁴ , 0.1% BER	_	-28.1	_	dBm
Level below which RFSENSE will not trigger ⁵	RFSENSE _{THRES}	CW at 490 MHz	_	-50	_	dBm
Adjacent channel selectivity, Interferer is CW at ± 1 × channel-spacing	C/I ₁	Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	48	60.3	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	38.3	45.6	_	dB
Alternate channel selectivity, Interferer is CW at ± 2 × channel-spacing	C/I ₂	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	60.4	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	52.6	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Image rejection, Interferer is CW at image frequency	C/I _{IMAGE}	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	56.5	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	54.1	_	dB
Blocking selectivity, 0.1%	C/I _{BLOCKER}	Interferer CW at Desired ± 1 MHz	_	73.9	_	dB
BER. Desired is 2.4 kbps GFSK signal ¹ at 3 dB above		Interferer CW at Desired ± 2 MHz	_	75.4	_	dB
sensitivity level		Interferer CW at Desired ± 10 MHz	_	90.2	_	dB
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	Over RSSI _{MIN} to RSSI _{MAX} range	_	0.25	_	dBm
Max spurious emissions dur-	SPUR _{RX}	30 MHz to 1 GHz	_	-53	-47	dBm
ing active receive mode		1 GHz to 12 GHz	_	-53	-47	dBm

- 1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.
- 2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.
- 3. Definition of reference signal is 10 kbps 2GFSK, BT=0.5, Δf = 5 kHz, RX channel BW = 20.038 kHz.
- 4. Definition of reference signal is 100 kbps 2GFSK, BT=0.5, Δf = 50 kHz, RX channel BW = 198.024 kHz.
- 5. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.7 Sub-GHz RF Transmitter characteristics for 433 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 433 MHz.

Table 4.30. Sub-GHz RF Transmitter characteristics for 433 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		426	_	445	MHz
Maximum TX Power ¹	POUT _{MAX}	External PA supply connected to DC-DC output, 14dBm output power	12.5	15.1	17.4	dBm
		External PA supply connected to DC-DC output, 10dBm output power	8.3	10.6	13.3	dBm
Minimum active TX Power	POUT _{MIN}		_	-42	_	dBm
Output power step size	POUT _{STEP}	output power > 0 dBm	_	0.5	_	dB
Output power variation vs supply, peak to peak, Pout = 10dBm	POUT _{VAR_V}	At 10 dBm;1.8 V < V _{VREGVDD} < 3.3 V, External PA supply = DC-DC output, T = 25 °C	_	1.7	_	dB
Output power variation vs	POUT _{VAR_T}	-40 to +85C at 10dBm	_	0.5	1.2	dB
temperature, peak to peak, Pout= 10dBm		-40 to +125C at 10dBm	_	0.7	1.7	dB
Output power variation vs RF frequency, Pout = 10dBm	POUT _{VAR_F}	T = 25 °C	_	0.1	0.2	dB
Spurious emissions of harmonics FCC, Conducted	SPUR _{HARM_FCC}	In restricted bands, per FCC Part 15.205 / 15.209	_	-47	-42	dBm
measurement, 14dBm match, External PA supply connected to DC-DC output, Test Frequency = 434 MHz		In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
Spurious emissions out-of- band FCC, Conducted	SPUR _{OOB_FCC}	In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
measurement, 14dBm match, External PA supply connected to DC-DC output,		In restricted bands (30-88 MHz), per FCC Part 15.205 / 15.209	_	-52	-46	dBm
Test Frequency = 434 MHz		In restricted bands (88-216 MHz), per FCC Part 15.205 / 15.209	_	-61	-56	dBm
		In restricted bands (216-960 MHz), per FCC Part 15.205 / 15.209	_	-58	-52	dBm
		In restricted bands (>960 MHz), per FCC Part 15.205 / 15.209	_	-47	-42	dBm
Spurious emissions of harmonics ETSI, Conducted	SPUR _{HARM_ETSI}	Per ETSI EN 300-220, Section 7.8.2.1 (frequencies below 1Ghz)	_	-42	-36	dBm
measurement, 14dBm match, External PA supply connected to DC-DC output, Test Frequency = 434 MHz		Per ETSI EN 300-220, Section 7.8.2.1 (frequencies above 1Ghz)	_	-36	-30	dBm

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Spurious emissions out-of- band ETSI, Conducted measurement, 14dBm match, External PA supply connected to DC-DC output, Test Frequency = 434 MHz	SPUR _{OOB_ETSI}	Per ETSI EN 300-220, Section 7.8.2.1 (47-74 MHz, 87.5-118 MHz, 174-230 MHz, and 470-862 MHz)	_	-60	-54	dBm
	7. G P 7.	Per ETSI EN 300-220, Section 7.8.2.1 (other frequencies below 1 GHz)	_	-42	-36	dBm
		Per ETSI EN 300-220, Section 7.8.2.1 (frequencies above 1 GHz)	_	-36	-30	dBm

^{1.} Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.8 Sub-GHz RF Receiver Characteristics for 433 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 433 MHz.

Table 4.31. Sub-GHz RF Receiver Characteristics for 433 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Tuning frequency range	F _{RANGE}		426	_	445	MHz
Max usable input level, 0.1% BER	SAT _{2k4}	Desired is reference 2.4 kbps GFSK signal ¹	_	10	_	dBm
Max usable input level, 0.1% BER	SAT _{50k}	Desired is reference 50 kbps GFSK signal ²	_	10	_	dBm
Sensitivity	SENS	Desired is reference 4.8 kbps OOK signal ³ , 20% PER	_	-107.4	_	dBm
		Desired is reference 100 kbps GFSK signal ⁴ , 0.1% BER, T ≤ 85 °C	_	-107.3	-105	dBm
		Desired is reference 100 kbps GFSK signal ⁴ , 0.1% BER, T > 85 °C	_	_	-104	dBm
		Desired is reference 50 kbps GFSK signal ² , 0.1% BER, T ≤ 85 °C	_	-110.3	-107.2	dBm
		Desired is reference 50 kbps GFSK signal ² , 0.1% BER, T > 85 °C	_	_	-106.6	dBm
		Desired is reference 2.4 kbps GFSK signal ¹ , 0.1% BER	_	-123.1	_	dBm
		Desired is reference 9.6 kbps GFSK signal ⁵ , 1% PER, T ≤ 85 °C	_	-112.6	-109	dBm
		Desired is reference 9.6 kbps GFSK signal ⁵ , 1% PER, T > 85 °C	_	_	-108	dBm
Level above which RFSENSE will trigger ⁶	RFSENSETRIG	CW at 433 MHz	_	-28.1	_	dBm
Level below which RFSENSE will not trigger ⁶	RFSENSETHRES	CW at 433 MHz	_	-50	_	dBm

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Adjacent channel selectivity, Interferer is CW at ± 1 × channel-spacing	C/I ₁	Desired is 4.8 kbps OOK signal ³ at 3dB above sensitivity level, 20% PER	_	51.6	_	dB
		Desired is 100 kbps GFSK signal ⁴ at 3dB above sensitivity level, 0.1% BER	35	44.1	_	dB
		Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	47	61.5	_	dB
		Desired is 50 kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	45.6	53.1	_	dB
		Desired is 9.6 kbps 4GFSK signal ⁵ at 3dB above sensitivity level, 1% PER	_	35.7	_	dB
Alternate channel selectivity, Interferer is CW at ± 2 × channel-spacing	C/I ₂	Desired is 4.8 kbps OOK signal ³ at 3dB above sensitivity level, 20% PER		57.8	_	dB
		Desired is 100 kbps GFSK signal ⁴ at 3dB above sensitivity level, 0.1% BER	_	54.6	_	dB
		Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	62.4	_	dB
		Desired is 50 kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	58.1	_	dB
		Desired is 9.6 kbps 4GFSK signal ⁵ at 3dB above sensitivity level, 1% PER	_	50.6	_	dB
Image rejection, Interferer is CW at image frequency	C/I _{IMAGE}	Desired is 4.8 kbps OOK signal ³ at 3dB above sensitivity level, 20% PER	_	46.5	_	dB
		Desired is 100 kbps GFSK signal ⁴ at 3dB above sensitivity level, 0.1% BER	_	51.7	_	dB
		Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	57.5	_	dB
		Desired is 50 kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	54.4	_	dB
		Desired is 9.6 kbps 4GFSK signal ⁵ at 3dB above sensitivity level, 1% PER	_	48	_	dB
Blocking selectivity, 0.1% BER. Desired is 2.4 kbps	C/I _{BLOCKER}	Interferer CW at Desired ± 1 MHz	_	75.7	_	dB
GFSK signal ¹ at 3dB above		Interferer CW at Desired ± 2 MHz	_	77.2	_	dB
sensitivity level		Interferer CW at Desired ± 10 MHz	_	92	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Intermod selectivity, 0.1% BER. CW interferers at 12.5 kHz and 25 kHz offsets	C/I _{IM}	Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level	_	58.8	_	dB
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	Over RSSI _{MIN} to RSSI _{MAX} range	_	0.25	_	dBm
Max spurious emissions dur-	SPUR _{RX_FCC}	216-960 MHz	_	-55	-49	dBm
ing active receive mode, per FCC Part 15.109(a)		Above 960 MHz	_	-47	-41	dBm
Max spurious emissions dur-	SPUR _{RX_ETSI}	Below 1000 MHz	_	-63	-57	dBm
ing active receive mode, per ETSI 300-220 Section 8.6		Above 1000 MHz	_	-53	-47	dBm
Max spurious emissions during active receive mode, per ARIB STD T67 Section 3.3(5)	SPUR _{RX_ARIB}	Below 710 MHz, RBW=100kHz	_	-60	-54	dBm

- 1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.
- 2. Definition of reference signal is 50 kbps 2GFSK, BT=0.5, Δf = 25 kHz, RX channel BW = 99.012 kHz, channel spacing = 200 kHz.
- 3. Definition of reference signal is 4.8 kbps OOK, RX channel BW = 306.036 kHz, channel spacing = 500 kHz.
- 4. Definition of reference signal is 100 kbps 2GFSK, BT=0.5, Δf = 50 kHz, RX channel BW = 198.024 kHz, channel spacing = 200 kHz.
- 5. Definition of reference signal is 9.6 kbps 4GFSK, BT=0.5, inner deviation = 0.8 kHz, RX channel BW = 8.5 kHz, channel spacing = 12.5 kHz.
- 6. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.9 Sub-GHz RF Transmitter characteristics for 315 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 315 MHz.

Table 4.32. Sub-GHz RF Transmitter characteristics for 315 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		195	_	358	MHz
Maximum TX Power ¹	POUT _{MAX}	External PA supply connected to DC-DC output	13.8	17.2	21.1	dBm
Minimum active TX Power	POUT _{MIN}			-43.9	_	dBm
Output power step size	POUT _{STEP}	output power > 0 dBm	_	0.5	_	dB
Output power variation vs supply	POUT _{VAR_V}	1.8 V < V _{VREGVDD} < 3.3 V, External PA supply = DC-DC output, T = 25 °C	_	1.8	_	dB
Output power variation vs temperature	POUT _{VAR_T}	-40 to +85C	_	0.5	1.2	dB
		-40 to +125C	_	0.7	1.5	dB
Output power variation vs RF frequency	POUT _{VAR_F}	T = 25 °C	_	0.1	0.7	dB
Spurious emissions of harmonics at 14 dBm output	SPUR _{HARM_FCC}	In restricted bands, per FCC Part 15.205 / 15.209	_	-47	-42	dBm
power, Conducted measure- ment, 14dBm match, Exter- nal PA supply connected to DC-DC output, Test Fre- quency = 303 MHz		In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
Spurious emissions out-of- band at 14 dBm output pow-	SPUR _{OOB_FCC}	In non-restricted bands, per FCC Part 15.231	_	-26	-20	dBc
er, Conducted measurement, 14dBm match, External PA supply connected to DC-DC		In restricted bands (30-88 MHz), per FCC Part 15.205 / 15.209	_	-52	-46	dBm
output, Test Frequency = 303 MHz		In restricted bands (88-216 MHz), per FCC Part 15.205 / 15.209	_	-61	-56	dBm
		In restricted bands (216-960 MHz), per FCC Part 15.205 / 15.209	_	-58	-52	dBm
		In restricted bands (>960 MHz), per FCC Part 15.205 / 15.209	_	-47	-42	dBm

^{1.} Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.

4.1.10.10 Sub-GHz RF Receiver Characteristics for 315 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 315 MHz.

Table 4.33. Sub-GHz RF Receiver Characteristics for 315 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Tuning frequency range	F _{RANGE}		195	_	358	MHz
Max usable input level, 0.1% BER	SAT _{2k4}	Desired is reference 2.4 kbps GFSK signal ¹	_	10	_	dBm
Max usable input level, 0.1% BER	SAT _{38k4}	Desired is reference 38.4 kbps GFSK signal ²	_	10	_	dBm
Sensitivity	SENS	Desired is reference 2.4 kbps GFSK signal ¹ , 0.1% BER, T ≤ 85 °C	_	-123.2	-120.7	dBm
		Desired is reference 2.4 kbps GFSK signal ¹ , 0.1% BER, T > 85 °C	_	_	-120	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T ≤ 85 °C	_	-111.4	-108.6	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T > 85 °C	_	_	-107.9	dBm
		Desired is reference 500 kbps GFSK signal ³ , 0.1% BER, T ≤ 85 °C	_	-98.8	-95.5	dBm
		Desired is reference 500 kbps GFSK signal ³ , 0.1% BER, T > 85 °C	_	_	-94.5	dBm
Level above which RFSENSE will trigger ⁴	RFSENSE _{TRIG}	CW at 315 MHz	_	-28.1	_	dBm
Level below which RFSENSE will not trigger ⁴	RFSENSE _{THRES}	CW at 315 MHz	_	-50	_	dBm
Adjacent channel selectivity, Interferer is CW at ± 1 × channel-spacing	C/I ₁	Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	54.1	63.6	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	49.9	_	dB
Alternate channel selectivity, Interferer is CW at ± 2 × channel-spacing	C/I ₂	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	64.2	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level ² , 0.1% BER	_	56.2	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Image rejection, Interferer is CW at image frequency	C/I _{IMAGE}	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	-	53	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	51.4	_	dB
Blocking selectivity, 0.1% BER. Desired is 2.4 kbps GFSK signal ¹ at 3 dB above	C/I _{BLOCKER}	Interferer CW at Desired ± 1 MHz	_	75	_	dB
		Interferer CW at Desired ± 2 MHz	_	76.5	_	dB
sensitivity level		Interferer CW at Desired ± 10 MHz	72.6	91.9	_	dB
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	Over RSSI _{MIN} to RSSI _{MAX} range	_	0.25	_	dBm
Max spurious emissions dur-	SPUR _{RX_FCC}	216-960 MHz	_	-63	-57	dBm
ing active receive mode, per FCC Part 15.109(a)		Above 960MHz	_	-53	-47	dBm

- 1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz
- 2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.
- 3. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 125 kHz, RX channel BW = 753.320 kHz.
- 4. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.10.11 Sub-GHz RF Transmitter Characteristics for 169 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 169 MHz.

Table 4.34. Sub-GHz RF Transmitter Characteristics for 169 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
RF tuning frequency range	F _{RANGE}		169	_	170	MHz
Maximum TX Power ¹	POUT _{MAX}	External PA supply = 3.3 V	18.1	19.7	22.4	dBm
Minimum active TX Power	POUT _{MIN}			-42.6	_	dBm
Output power step size	POUT _{STEP}	output power > 0 dBm	_	0.5	_	dB
Output power variation vs supply, peak to peak	POUT _{VAR_V}	1.8 V < V _{VREGVDD} < 3.3 V, External PA supply = 3.3 V, T = 25 °C	_	4.8	5.0	dB
Output power variation vs temperature, peak to peak	POUT _{VAR_T}	-40 to +85 °C at 20 dBm	_	0.6	1.2	dB
		-40 to +125 °C at 20 dBm	_	0.8	1.5	dB
Spurious emissions of har- monics, Conducted meas- urement, External PA supply = 3.3 V, Test Frequency =	SPUR _{HARM_ETSI}	Per ETSI EN 300-220, Section 7.8.2.1 (47-74 MHz, 87.5-118 MHz, 174-230 MHz, and 470-862 MHz)	_	-42	_	dBm
169 MHz		Per ETSI EN 300-220, Section 7.8.2.1 (other frequencies below 1 GHz) ²	_	-38	_	dBm
		Per ETSI EN 300-220, Section 7.8.2.1 (frequencies above 1 GHz) ²	_	-36	_	dBm
Spurious emissions out-of- band, Conducted measure- ment, External PA supply = 3.3 V, Test Frequency = 169 MHz	SPUR _{OOB_ETSI}	Per ETSI EN 300-220, Section 7.8.2.1 (47-74 MHz, 87.5-118 MHz, 174-230 MHz, and 470-862 MHz)	_	-42	-36	dBm
		Per ETSI EN 300-220, Section 7.8.2.1 (other frequencies below 1 GHz)	_	-42	-36	dBm
		Per ETSI EN 300-220, Section 7.8.2.1 (frequencies above 1 GHz)	_	-36	-30	dBm

- 1. Supported transmit power levels are determined by the ordering part number (OPN). Transmit power ratings for all devices covered in this datasheet can be found in the Max TX Power column of the Ordering Information Table.
- 2. Typical value marginally passes specification. Additional margin can be obtained by increasing the order of the harmonic filter.

4.1.10.12 Sub-GHz RF Receiver Characteristics for 169 MHz Band

Unless otherwise indicated, typical conditions are: T = 25 °C, VREGVDD = AVDD = IOVDD = 3.3 V, DVDD = RFVDD = External PA Supply. RFVDD and external PA supply paths filtered using ferrites. Crystal frequency = 38.4 MHz. RF center frequency 169 MHz.

Table 4.35. Sub-GHz RF Receiver Characteristics for 169 MHz Band

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Tuning frequency range	F _{RANGE}		169	_	170	MHz
Max usable input level, 0.1% BER	SAT _{2k4}	Desired is reference 2.4 kbps GFSK signal ¹	_	10	_	dBm
Max usable input level, 0.1% BER	SAT _{38k4}	Desired is reference 38.4 kbps GFSK signal ²	_	10	_	dBm
Sensitivity	SENS	Desired is reference 2.4 kbps GFSK signal ¹ , 0.1% BER	_	-124	_	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T ≤ 85 °C	_	-112.2	-108	dBm
		Desired is reference 38.4 kbps GFSK signal ² , 0.1% BER, T > 85 °C	_	_	-107	dBm
		Desired is reference 500 kbps GFSK signal ³ , 0.1% BER, T ≤ 85 °C	_	-99.2	-96	dBm
		Desired is reference 500 kbps GFSK signal ³ , 0.1% BER, T > 85 °C	_	_	-95	dBm
Level above which RFSENSE will trigger ⁴	RFSENSE _{TRIG}	CW at 169 MHz	_	-28.1	_	dBm
Level below which RFSENSE will not trigger ⁴	RFSENSE _{THRES}	CW at 169 MHz	_	-50	_	dBm
Adjacent channel selectivity, Interferer is CW at ± 1 x channel-spacing	C/I ₁	Desired is 2.4 kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	64.8	_	dB
		Desired is 38.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	43.3	51.4	_	dB
Alternate channel selectivity, Interferer is CW at ± 2 x channel-spacing	C/I ₂	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	67.4	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	60.6	_	dB
Image rejection, Interferer is CW at image frequency	C/I _{IMAGE}	Desired is 2.4kbps GFSK signal ¹ at 3dB above sensitivity level, 0.1% BER	_	47.1	_	dB
		Desired is 38.4kbps GFSK signal ² at 3dB above sensitivity level, 0.1% BER	_	47.1	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Blocking selectivity, 0.1% BER. Desired is 2.4 kbps GFSK signal ¹ at 3 dB above sensitivity level	C/I _{BLOCKER}	Interferer CW at Desired ± 1 MHz	_	73.4	_	dB
		Interferer CW at Desired ± 2 MHz	_	75	_	dB
		Interferer CW at Desired ± 10 MHz	80	90.1	_	dB
Upper limit of input power range over which RSSI resolution is maintained	RSSI _{MAX}		_	_	5	dBm
Lower limit of input power range over which RSSI resolution is maintained	RSSI _{MIN}		-98	_	_	dBm
RSSI resolution	RSSI _{RES}	Over RSSI _{MIN} to RSSI _{MAX} range	_	0.25	_	dBm
Max spurious emissions dur-	SPUR _{RX}	30 MHz to 1 GHz	_	-63	-57	dBm
ing active receive mode		1 GHz to 12 GHz	_	-53	-47	dBm

- 1. Definition of reference signal is 2.4 kbps 2GFSK, BT=0.5, Δf = 1.2 kHz, RX channel BW = 4.798 kHz, channel spacing = 12.5 kHz.
- 2. Definition of reference signal is 38.4 kbps 2GFSK, BT=0.5, Δf = 20 kHz, RX channel BW = 74.809 kHz, channel spacing = 100 kHz.
- 3. Definition of reference signal is 500 kbps 2GFSK, BT=0.5, Δf = 125 kHz, RX channel BW = 753.320 kHz.
- 4. RFSENSE performance is only valid from 0 to 85 °C. RFSENSE should be disabled outside this temperature range.

4.1.11 Modem

Table 4.36. Modem

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Receive bandwidth	BW _{RX}	Configurable range with 38.4 MHz crystal	0.1	_	2530	kHz
IF frequency	f _{IF}	Configurable range with 38.4 MHz crystal. Selected steps available.	150	_	1371	kHz
DSSS symbol length	SL _{DSSS}	Configurable in steps of 1 chip	2	_	32	chips
DSSS bits per symbol	BPS _{DSSS}	Configurable	1	_	4	bits/ symbol

4.1.12 Oscillators

4.1.12.1 Low-Frequency Crystal Oscillator (LFXO)

Table 4.37. Low-Frequency Crystal Oscillator (LFXO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Crystal frequency	f _{LFXO}		_	32.768	_	kHz
Supported crystal equivalent series resistance (ESR)	ESR _{LFXO}		_	_	70	kΩ
Supported range of crystal load capacitance ¹	C _{LFXO_CL}		6	_	18	pF
On-chip tuning cap range ²	C _{LFXO_T}	On each of LFXTAL_N and LFXTAL_P pins	8	_	40	pF
On-chip tuning cap step size	SS _{LFXO}		_	0.25	_	pF
Current consumption after startup ³	I _{LFXO}	ESR = 70 kOhm, C_L = 7 pF, GAIN ⁴ = 2, AGC ⁴ = 1	_	273	_	nA
Start- up time	t _{LFXO}	ESR = 70 kOhm, $C_L = 7 pF$, $GAIN^4 = 2$	_	308	_	ms

- 1. Total load capacitance as seen by the crystal.
- 2. The effective load capacitance seen by the crystal will be C_{LFXO_T} /2. This is because each XTAL pin has a tuning cap and the two caps will be seen in series by the crystal.
- 3. Block is supplied by AVDD if ANASW = 0, or DVDD if ANASW=1 in EMU_PWRCTRL register.
- 4. In CMU_LFXOCTRL register.

4.1.12.2 High-Frequency Crystal Oscillator (HFXO)

Table 4.38. High-Frequency Crystal Oscillator (HFXO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Crystal frequency	f _{HFXO}	38.4 MHz required for radio transciever operation	38	38.4	40	MHz
Supported crystal equivalent series resistance (ESR)	ESR _{HFXO_38M4}	Crystal frequency 38.4 MHz	_	_	60	Ω
Supported range of crystal load capacitance ¹	C _{HFXO_CL}		6	_	12	pF
On-chip tuning cap range ²	C _{HFXO_T}	On each of HFXTAL_N and HFXTAL_P pins	9	20	25	pF
On-chip tuning capacitance step	SS _{HFXO}		_	0.04	_	pF
Startup time	t _{HFXO}	38.4 MHz, ESR = 50 Ohm, C _L = 10 pF	_	300	_	μs
Frequency tolerance for the crystal	FT _{HFXO}	38.4 MHz, ESR = 50 Ohm, C _L = 10 pF	-40	_	40	ppm

Note:

- 1. Total load capacitance as seen by the crystal.
- 2. The effective load capacitance seen by the crystal will be C_{HFXO_T} /2. This is because each XTAL pin has a tuning cap and the two caps will be seen in series by the crystal.

4.1.12.3 Low-Frequency RC Oscillator (LFRCO)

Table 4.39. Low-Frequency RC Oscillator (LFRCO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Oscillation frequency	f _{LFRCO}	ENVREF ¹ = 1, T ≤ 85 °C	31.3	32.768	33.6	kHz
		ENVREF ¹ = 1, T > 85 °C	31.6	32.768	36.8	kHz
		ENVREF ¹ = 0, T ≤ 85 °C	31.3	32.768	33.4	kHz
		ENVREF ¹ = 0, T > 85 °C	30	32.768	33.4	kHz
Startup time	t _{LFRCO}		_	500	_	μs
Current consumption ²	I _{LFRCO}	ENVREF = 1 in CMU_LFRCOCTRL	_	342	_	nA
		ENVREF = 0 in CMU_LFRCOCTRL	_	494	_	nA

- 1. In CMU_LFRCOCTRL register.
- 2. Block is supplied by AVDD if ANASW = 0, or DVDD if ANASW=1 in EMU_PWRCTRL register.

4.1.12.4 High-Frequency RC Oscillator (HFRCO)

Table 4.40. High-Frequency RC Oscillator (HFRCO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Frequency accuracy	f _{HFRCO_ACC}	At production calibrated frequencies, across supply voltage and temperature	-2.5	_	2.5	%
Start-up time	t _{HFRCO}	f _{HFRCO} ≥ 19 MHz	_	300	_	ns
		4 < f _{HFRCO} < 19 MHz	_	1	_	μs
		f _{HFRCO} ≤ 4 MHz	_	2.5	_	μs
Current consumption on all	I _{HFRCO}	f _{HFRCO} = 38 MHz	_	267	299	μA
supplies		f _{HFRCO} = 32 MHz	_	224	248	μA
		f _{HFRCO} = 26 MHz	_	189	211	μA
		f _{HFRCO} = 19 MHz	_	154	172	μA
		f _{HFRCO} = 16 MHz	_	133	148	μA
		f _{HFRCO} = 13 MHz	_	118	135	μA
		f _{HFRCO} = 7 MHz	_	89	100	μA
		f _{HFRCO} = 4 MHz	_	34	44	μA
		f _{HFRCO} = 2 MHz	_	29	40	μA
		f _{HFRCO} = 1 MHz	_	26	36	μA
Coarse trim step size (% of period)	SS _{HFRCO_COARS}		_	0.8	_	%
Fine trim step size (% of period)	SS _{HFRCO_FINE}		_	0.1	_	%
Period jitter	PJ _{HFRCO}		_	0.2	_	% RMS
Frequency limits	f _{HFRCO_BAND}	FREQRANGE = 0, FINETUNIN- GEN = 0	3.47	_	6.15	MHz
		FREQRANGE = 3, FINETUNIN- GEN = 0	6.24	_	11.45	MHz
		FREQRANGE = 6, FINETUNIN- GEN = 0	11.3	_	19.8	MHz
		FREQRANGE = 7, FINETUNIN- GEN = 0	13.45	_	22.8	MHz
		FREQRANGE = 8, FINETUNIN- GEN = 0	16.5	_	29.0	MHz
		FREQRANGE = 10, FINETUNIN- GEN = 0	23.11	_	40.63	MHz
		FREQRANGE = 11, FINETUNIN- GEN = 0	27.27	_	48	MHz
		FREQRANGE = 12, FINETUNIN- GEN = 0	33.33	_	54	MHz

4.1.12.5 Auxiliary High-Frequency RC Oscillator (AUXHFRCO)

Table 4.41. Auxiliary High-Frequency RC Oscillator (AUXHFRCO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Frequency accuracy	f _{AUXHFRCO_ACC}	At production calibrated frequencies, across supply voltage and temperature	-3	_	3	%
Start-up time	t _{AUXHFRCO}	f _{AUXHFRCO} ≥ 19 MHz	_	400	_	ns
		4 < f _{AUXHFRCO} < 19 MHz	_	1.4	_	μs
		f _{AUXHFRCO} ≤ 4 MHz	_	2.5	_	μs
Current consumption on all supplies	I _{AUXHFRCO}	f _{AUXHFRCO} = 38 MHz	_	187	207	μA
		f _{AUXHFRCO} = 32 MHz	_	152	168	μA
		f _{AUXHFRCO} = 26 MHz	_	131	145	μA
		f _{AUXHFRCO} = 19 MHz	_	106	118	μA
		f _{AUXHFRCO} = 16 MHz	_	98	110	μA
		f _{AUXHFRCO} = 13 MHz	_	75	86	μA
		f _{AUXHFRCO} = 7 MHz	_	52	61	μA
		f _{AUXHFRCO} = 4 MHz	_	29	37	μA
		f _{AUXHFRCO} = 2 MHz	_	26	35	μA
		f _{AUXHFRCO} = 1 MHz	_	25	33	μA
Coarse trim step size (% of period)	SS _{AUXHFR} - CO_COARSE		_	0.8	_	%
Fine trim step size (% of period)	SS _{AUXHFR} - CO_FINE		_	0.1	_	%
Period jitter	PJ _{AUXHFRCO}		_	0.2	_	% RMS

4.1.12.6 Ultra-low Frequency RC Oscillator (ULFRCO)

Table 4.42. Ultra-low Frequency RC Oscillator (ULFRCO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Oscillation frequency	f _{ULFRCO}		0.95	1	1.07	kHz

4.1.13 Flash Memory Characteristics¹

Table 4.43. Flash Memory Characteristics¹

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Flash erase cycles before failure	EC _{FLASH}		10000	_	_	cycles
Flash data retention	RET _{FLASH}	T ≤ 85 °C	10	_	_	years
		T ≤ 125 °C	10	_	_	years
Word (32-bit) programming time	t _{W_PROG}	Burst write, 128 words, average time per word	20	26.3	30	μs
		Single word	62	68.9	80	μs
Page erase time ²	t _{PERASE}		20	29.5	40	ms
Mass erase time ³	t _{MERASE}		20	30	40	ms
Device erase time ^{4 5}	t _{DERASE}	T ≤ 85 °C	_	56.2	70	ms
		T ≤ 125 °C	_	56.2	75	ms
Erase current ⁶	I _{ERASE}	Page Erase	_	_	2.0	mA
Write current ⁶	I _{WRITE}		_	_	3.5	mA
Supply voltage during flash erase and write	V _{FLASH}		1.62	_	3.6	V

- 1. Flash data retention information is published in the Quarterly Quality and Reliability Report.
- 2. From setting the ERASEPAGE bit in MSC_WRITECMD to 1 until the BUSY bit in MSC_STATUS is cleared to 0. Internal setup and hold times for flash control signals are included.
- 3. Mass erase is issued by the CPU and erases all flash.
- 4. Device erase is issued over the AAP interface and erases all flash, SRAM, the Lock Bit (LB) page, and the User data page Lock Word (ULW).
- 5. From setting the DEVICEERASE bit in AAP_CMD to 1 until the ERASEBUSY bit in AAP_STATUS is cleared to 0. Internal setup and hold times for flash control signals are included.
- 6. Measured at 25 °C.

4.1.14 General-Purpose I/O (GPIO)

Table 4.44. General-Purpose I/O (GPIO)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Input low voltage	V _{IL}	GPIO pins	_	_	IOVDD*0.3	V
Input high voltage	V _{IH}	GPIO pins	IOVDD*0.7	_	_	V
Output high voltage relative	V _{OH}	Sourcing 3 mA, IOVDD ≥ 3 V,	IOVDD*0.8	_	_	V
to IOVDD		DRIVESTRENGTH ¹ = WEAK				
		Sourcing 1.2 mA, IOVDD ≥ 1.62 V,	IOVDD*0.6	_	_	V
		DRIVESTRENGTH ¹ = WEAK				
		Sourcing 20 mA, IOVDD ≥ 3 V,	IOVDD*0.8	_	_	V
		DRIVESTRENGTH ¹ = STRONG				
		Sourcing 8 mA, IOVDD ≥ 1.62 V,	IOVDD*0.6	_	_	V
		DRIVESTRENGTH ¹ = STRONG				
Output low voltage relative to	V _{OL}	Sinking 3 mA, IOVDD ≥ 3 V,	_	_	IOVDD*0.2	V
IOVDD		DRIVESTRENGTH ¹ = WEAK				
		Sinking 1.2 mA, IOVDD ≥ 1.62 V,	_	_	IOVDD*0.4	V
		DRIVESTRENGTH ¹ = WEAK				
		Sinking 20 mA, IOVDD ≥ 3 V,	_	_	IOVDD*0.2	V
		DRIVESTRENGTH ¹ = STRONG				
		Sinking 8 mA, IOVDD ≥ 1.62 V,	_	_	IOVDD*0.4	V
		DRIVESTRENGTH ¹ = STRONG				
Input leakage current	I _{IOLEAK}	All GPIO except LFXO pins, GPIO ≤ IOVDD, T ≤ 85 °C	_	0.1	30	nA
		LFXO Pins, GPIO ≤ IOVDD, T ≤ 85 °C	_	0.1	50	nA
		All GPIO except LFXO pins, GPIO ≤ IOVDD, T > 85 °C	_	<u> </u>	110	nA
		LFXO Pins, GPIO ≤ IOVDD, T > 85 °C	_	_	250	nA
Input leakage current on 5VTOL pads above IOVDD	I _{5VTOLLEAK}	IOVDD < GPIO ≤ IOVDD + 2 V	_	3.3	15	μΑ
I/O pin pull-up/pull-down resistor	R _{PUD}		30	40	65	kΩ
Pulse width of pulses removed by the glitch suppression filter	t _I OGLITCH		15	25	45	ns

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Output fall time, From 70%	t _{IOOF}	C _L = 50 pF,	_	1.8	_	ns
to 30% of V _{IO}		DRIVESTRENGTH ¹ = STRONG,				
		SLEWRATE ¹ = 0x6				
		C _L = 50 pF,	_	4.5	_	ns
		DRIVESTRENGTH ¹ = WEAK,				
		SLEWRATE ¹ = 0x6				
Output rise time, From 30%		C _L = 50 pF,	_	2.2	_	ns
to 70% of V _{IO}		DRIVESTRENGTH ¹ = STRONG,				
		SLEWRATE = 0x6 ¹				
		C _L = 50 pF,	_	7.4	_	ns
		DRIVESTRENGTH ¹ = WEAK,				
		SLEWRATE ¹ = 0x6				

1. In GPIO_Pn_CTRL register.

4.1.15 Voltage Monitor (VMON)

Table 4.45. Voltage Monitor (VMON)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Supply current (including I_SENSE)	I _{VMON}	In EM0 or EM1, 1 active channel, T ≤ 85 °C	_	6.3	8	μA
		In EM0 or EM1, 1 active channel, T > 85 °C	_	_	10	μA
		In EM0 or EM1, All channels active, T ≤ 85 °C	_	12.5	15	μA
		In EM0 or EM1, All channels active, T > 85 °C	_	_	18	μA
		In EM2, EM3 or EM4, 1 channel active and above threshold	_	62	_	nA
		In EM2, EM3 or EM4, 1 channel active and below threshold	_	62	_	nA
		In EM2, EM3 or EM4, All channels active and above threshold	<u>—</u>	99	_	nA
		In EM2, EM3 or EM4, All channels active and below threshold	_	99	_	nA
Loading of monitored supply	I _{SENSE}	In EM0 or EM1	_	2	_	μA
		In EM2, EM3 or EM4	_	2	_	nA
Threshold range	V _{VMON_RANGE}		1.62	_	3.4	V
Threshold step size	N _{VMON_STESP}	Coarse	_	200	_	mV
		Fine	_	20	_	mV
Response time	t _{VMON_RES}	Supply drops at 1V/µs rate	_	460	_	ns
Hysteresis	V _{VMON_HYST}		_	26	_	mV

4.1.16 Analog to Digital Converter (ADC)

Specified at 1 Msps, ADCCLK = 16 MHz, BIASPROG = 0, GPBIASACC = 0, unless otherwise indicated.

Table 4.46. Analog to Digital Converter (ADC)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Resolution	V _{RESOLUTION}		6	_	12	Bits
Input voltage range ¹	V _{ADCIN}	Single ended	_	_	V _{FS}	V
		Differential	-V _{FS} /2	_	V _{FS} /2	V
Input range of external reference voltage, single ended and differential	V _{ADCREFIN_P}		1	_	V _{AVDD}	V
Power supply rejection ²	PSRR _{ADC}	At DC	_	80	_	dB
Analog input common mode rejection ratio	CMRR _{ADC}	At DC	_	80	_	dB
Current from all supplies, using internal reference buffer.	I _{ADC_CONTINU} - OUS_LP	1 Msps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 1 ⁴	_	270	290	μA
Continuous operation. WAR- MUPMODE ³ = KEEPADC- WARM		250 ksps / 4 MHz ADCCLK, BIA- SPROG = 6, GPBIASACC = 1 ⁴	_	125	_	μA
		62.5 ksps / 1 MHz ADCCLK, BIA- SPROG = 15, GPBIASACC = 1 ⁴	_	80	_	μA
Current from all supplies, using internal reference buffer.	I _{ADC_NORMAL_LP}	35 ksps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 1 ⁴	_	45	_	μA
Duty-cycled operation. WAR- MUPMODE ³ = NORMAL		5 ksps / 16 MHz ADCCLK BIA- SPROG = 0, GPBIASACC = 1 ⁴	_	8	_	μA
Current from all supplies, using internal reference buffer. Duty-cycled operation.	I _{ADC_STAND} - BY_LP	125 ksps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 1 ⁴	_	105	_	μA
AWARMUPMODE ³ = KEEP-INSTANDBY or KEEPIN-SLOWACC		35 ksps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 1 ⁴	_	70	_	μА
Current from all supplies, using internal reference buffer. Continuous operation. WAR-	I _{ADC_CONTINU} - OUS_HP	1 Msps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 0 ⁴	_	325	_	μA
MUPMODE ³ = KEEPADC- WARM		250 ksps / 4 MHz ADCCLK, BIA- SPROG = 6, GPBIASACC = 0 ⁴	_	175	_	μA
		62.5 ksps / 1 MHz ADCCLK, BIA- SPROG = 15, GPBIASACC = 0 ⁴	_	125	_	μA
Current from all supplies, using internal reference buffer.	IADC_NORMAL_HP	35 ksps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 0 ⁴	_	85	_	μA
Duty-cycled operation. WAR-MUPMODE ³ = NORMAL		5 ksps / 16 MHz ADCCLK BIA- SPROG = 0, GPBIASACC = 0 ⁴	_	16	_	μА
Current from all supplies, using internal reference buffer. Duty-cycled operation. AWARMUPMODE ³ = KEEP-INSTANDBY or KEEPIN-SLOWACC	I _{ADC_STAND} - BY_HP	125 ksps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 0 ⁴	_	160	_	μА
		35 ksps / 16 MHz ADCCLK, BIA- SPROG = 0, GPBIASACC = 0 ⁴	_	125	_	μА
Current from HFPERCLK	I _{ADC_CLK}	HFPERCLK = 16 MHz	_	140	_	μA

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
ADC clock frequency	f _{ADCCLK}		_	_	16	MHz
Throughput rate	fADCRATE		_	_	1	Msps
Conversion time ⁵	t _{ADCCONV}	6 bit	_	7	_	cycles
		8 bit	_	9	_	cycles
		12 bit	_	13	_	cycles
Startup time of reference generator and ADC core	tADCSTART	WARMUPMODE ³ = NORMAL	_	_	5	μs
generator and ADC core		WARMUPMODE ³ = KEEPIN- STANDBY	_	_	2	μs
		WARMUPMODE ³ = KEEPINSLO- WACC	_	_	1	μs
SNDR at 1Msps and f _{IN} = 10kHz	SNDR _{ADC}	Internal reference ⁶ , differential measurement	58	67	_	dB
		External reference ⁷ , differential measurement	_	68	_	dB
Spurious-free dynamic range (SFDR)	SFDR _{ADC}	1 MSamples/s, 10 kHz full-scale sine wave	_	75	_	dB
Differential non-linearity (DNL)	DNL _{ADC}	12 bit resolution, No missing codes	-1	_	2	LSB
Integral non-linearity (INL), End point method	INL _{ADC}	12 bit resolution	-6	_	6	LSB
Offset error	V _{ADCOFFSETERR}		-3	0	3	LSB
Gain error in ADC	V _{ADCGAIN}	Using internal reference	_	-0.2	3.5	%
		Using external reference	_	-1	_	%
Temperature sensor slope	V _{TS_SLOPE}		_	-1.84	_	mV/°C

- 1. The absolute voltage allowed at any ADC input is dictated by the power rail supplied to on-chip circuitry, and may be lower than the effective full scale voltage. All ADC inputs are limited to the ADC supply (AVDD or DVDD depending on EMU_PWRCTRL_ANASW). Any ADC input routed through the APORT will further be limited by the IOVDD supply to the pin.
- 2. PSRR is referenced to AVDD when ANASW=0 and to DVDD when ANASW=1 in EMU PWRCTRL.
- 3. In ADCn_CNTL register.
- 4. In ADCn BIASPROG register.
- 5. Derived from ADCCLK.
- 6. Internal reference option used corresponds to selection 2V5 in the SINGLECTRL_REF or SCANCTRL_REF register field. The differential input range with this configuration is ± 1.25 V. Typical value is characterized using full-scale sine wave input. Minimum value is production-tested using sine wave input at 1.5 dB lower than full scale.
- 7. External reference is 1.25 V applied externally to ADCnEXTREFP, with the selection CONF in the SINGLECTRL_REF or SCANCTRL_REF register field and VREFP in the SINGLECTRLX_VREFSEL or SCANCTRLX_VREFSEL field. The differential input range with this configuration is \pm 1.25 V.

4.1.17 Analog Comparator (ACMP)

Table 4.47. Analog Comparator (ACMP)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Input voltage range	V _{ACMPIN}	ACMPVDD = ACMPn_CTRL_PWRSEL ¹	_	_	V _{ACMPVDD}	V
Supply voltage	VACMPVDD	BIASPROG ² \leq 0x10 or FULL- BIAS ² = 0	1.8	_	V _{VREGVDD} _ MAX	V
		$0x10 < BIASPROG^2 \le 0x20$ and FULLBIAS ² = 1	2.1	_	V _{VREGVDD} _	V
Active current not including	I _{ACMP}	BIASPROG ² = 1, FULLBIAS ² = 0	_	50	_	nA
voltage reference ³		BIASPROG ² = 0x10, FULLBIAS ² = 0	_	306	_	nA
		BIASPROG ² = 0x02, FULLBIAS ² = 1	_	6.1	11	μΑ
		BIASPROG ² = 0x20, FULLBIAS ² = 1	_	74	92	μΑ
Current consumption of internal voltage reference ³	I _{ACMPREF}	VLP selected as input using 2.5 V Reference / 4 (0.625 V)	_	50	_	nA
		VLP selected as input using VDD	_	20	_	nA
		VBDIV selected as input using 1.25 V reference / 1	_	4.1	_	μΑ
		VADIV selected as input using VDD/1	_	2.4	_	μΑ

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Hysteresis (V _{CM} = 1.25 V,	V _{ACMPHYST}	HYSTSEL ⁴ = HYST0	-3	0	3	mV
BIASPROG ² = $0x10$, FULL- BIAS ² = 1)		HYSTSEL ⁴ = HYST1	5	18	27	mV
,		HYSTSEL ⁴ = HYST2	12	33	50	mV
		HYSTSEL ⁴ = HYST3	17	46	67	mV
		HYSTSEL ⁴ = HYST4	23	57	86	mV
		HYSTSEL ⁴ = HYST5	26	68	104	mV
		HYSTSEL ⁴ = HYST6	30	79	130	mV
		HYSTSEL ⁴ = HYST7	34	90	155	mV
		HYSTSEL ⁴ = HYST8	-3	0	3	mV
		HYSTSEL ⁴ = HYST9	-27	-18	-5	mV
		HYSTSEL ⁴ = HYST10	-50	-33	-12	mV
		HYSTSEL ⁴ = HYST11	-67	-45	-17	mV
		HYSTSEL ⁴ = HYST12	-86	-57	-23	mV
		HYSTSEL ⁴ = HYST13	-104	-67	-26	mV
		HYSTSEL ⁴ = HYST14	-130	-78	-30	mV
		HYSTSEL ⁴ = HYST15	-155	-88	-34	mV
Comparator delay ⁵	t _{ACMPDELAY}	$BIASPROG^2 = 1$, $FULLBIAS^2 = 0$	_	30	95	μs
		BIASPROG ² = 0x10, FULLBIAS ² = 0	_	3.7	10	μs
		BIASPROG ² = 0x02, FULLBIAS ² = 1	_	360	1000	ns
		BIASPROG ² = 0x20, FULLBIAS ² = 1	_	35	_	ns
Offset voltage	V _{ACMPOFFSET}	BIASPROG ² =0x10, FULLBIAS ² = 1	-35	_	35	mV
Reference voltage	V _{ACMPREF}	Internal 1.25 V reference	1	1.25	1.47	V
		Internal 2.5 V reference	1.98	2.5	2.8	V
Capacitive sense internal resistance	R _{CSRES}	CSRESSEL ⁶ = 0	_	infinite	_	kΩ
Sisterior		CSRESSEL ⁶ = 1	_	15	_	kΩ
		CSRESSEL ⁶ = 2	_	27	_	kΩ
		CSRESSEL ⁶ = 3	_	39	_	kΩ
		CSRESSEL ⁶ = 4	_	51	_	kΩ
		CSRESSEL ⁶ = 5	_	102	_	kΩ
		CSRESSEL ⁶ = 6	_	164	_	kΩ
		CSRESSEL ⁶ = 7	_	239	_	kΩ

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
	_					

- 1. ACMPVDD is a supply chosen by the setting in ACMPn_CTRL_PWRSEL and may be IOVDD, AVDD or DVDD.
- 2. In ACMPn_CTRL register.
- 3. The total ACMP current is the sum of the contributions from the ACMP and its internal voltage reference. $I_{ACMPTOTAL} = I_{ACMP} + I_{ACMPREF}$.
- 4. In ACMPn_HYSTERESIS registers.
- 5. ± 100 mV differential drive.
- 6. In ACMPn_INPUTSEL register.

4.1.18 Digital to Analog Converter (VDAC)

DRIVESTRENGTH = 2 unless otherwise specified. Primary VDAC output.

Table 4.48. Digital to Analog Converter (VDAC)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Output voltage	V _{DACOUT}	Single-Ended	0	_	V _{VREF}	V
		Differential ¹	-V _{VREF}	_	V _{VREF}	V
Current consumption including references (2 channels) ²	I _{DAC}	500 ksps, 12-bit, DRIVES- TRENGTH = 2, REFSEL = 4	_	396	_	μA
		44.1 ksps, 12-bit, DRIVES- TRENGTH = 1, REFSEL = 4	_	72	_	μА
		200 Hz refresh rate, 12-bit Sample-Off mode in EM2, DRIVES-TRENGTH = 2, BGRREQTIME = 1, EM2REFENTIME = 9, REFSEL = 4, SETTLETIME = 0x0A, WARMUPTIME = 0x02	_	1.2	_	μА
Current from HFPERCLK ³	I _{DAC_CLK}		_	5.8	_	μA/MHz
Sample rate	SR _{DAC}		_	_	500	ksps
DAC clock frequency	f _{DAC}		_	_	1	MHz
Conversion time	t _{DACCONV}	f _{DAC} = 1MHz	2	_	_	μs
Settling time	t _{DACSETTLE}	50% fs step settling to 5 LSB	_	2.5	_	μs
Startup time	t _{DACSTARTUP}	Enable to 90% fs output, settling to 10 LSB	_	_	12	μs
Output impedance	R _{OUT}	DRIVESTRENGTH = 2, 0.4 V \leq V _{OUT} \leq V _{OPA} - 0.4 V, -8 mA $<$ I _{OUT} $<$ 8 mA, Full supply range	_	2	_	Ω
		DRIVESTRENGTH = 0 or 1, 0.4 V \leq V _{OUT} \leq V _{OPA} - 0.4 V, -400 μ A $<$ I _{OUT} $<$ 400 μ A, Full supply range	_	2	_	Ω
		DRIVESTRENGTH = 2, 0.1 V \leq V _{OUT} \leq V _{OPA} - 0.1 V, -2 mA $<$ I _{OUT} $<$ 2 mA, Full supply range	_	2	_	Ω
		DRIVESTRENGTH = 0 or 1, 0.1 V \leq V _{OUT} \leq V _{OPA} - 0.1 V, -100 μ A $<$ I _{OUT} $<$ 100 μ A, Full supply range	_	2	_	Ω
Power supply rejection ratio ⁴	PSRR	Vout = 50% fs. DC	_	65.5	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Signal to noise and distortion ratio (1 kHz sine wave),	SNDR _{DAC}	500 ksps, single-ended, internal 1.25V reference	_	60.4	_	dB
Noise band limited to 250 kHz		500 ksps, single-ended, internal 2.5V reference	_	61.6	_	dB
		500 ksps, single-ended, 3.3V VDD reference		64.0	_	dB
		500 ksps, differential, internal 1.25V reference	_	63.3	_	dB
		500 ksps, differential, internal 2.5V reference	_	64.4	_	dB
		500 ksps, differential, 3.3V VDD reference	_	65.8	_	dB
Signal to noise and distortion ratio (1 kHz sine wave), Noise band limited to 22 kHz	SNDR _{DAC_BAND}	500 ksps, single-ended, internal 1.25V reference	_	65.3	_	dB
		500 ksps, single-ended, internal 2.5V reference	_	66.7	_	dB
		500 ksps, single-ended, 3.3V VDD reference		70.0	_	dB
		500 ksps, differential, internal 1.25V reference	_	67.8	_	dB
		500 ksps, differential, internal 2.5V reference	_	69.0	_	dB
		500 ksps, differential, 3.3V VDD reference	_	68.5	_	dB
Total harmonic distortion	THD		_	70.2	_	dB
Differential non-linearity ⁵	DNL _{DAC}		-0.99	_	1	LSB
Intergral non-linearity	INL _{DAC}		-4	_	4	LSB
Offset error ⁶	V _{OFFSET}	T = 25 °C	-8	_	8	mV
		Across operating temperature range	-25	_	25	mV
Gain error ⁶	V _{GAIN}	T = 25 °C, Low-noise internal reference (REFSEL = 1V25LN or 2V5LN)	-2.5	_	2.5	%
		T = 25 °C, Internal reference (RE-FSEL = 1V25 or 2V5)	-5	_	5	%
		T = 25 °C, External reference (REFSEL = VDD or EXT)	-1.8	_	1.8	%
		Across operating temperature range, Low-noise internal reference (REFSEL = 1V25LN or 2V5LN)	-3.5	_	3.5	%
		Across operating temperature range, Internal reference (RE-FSEL = 1V25 or 2V5)	-7.5	_	7.5	%
		Across operating temperature range, External reference (RE-FSEL = VDD or EXT)	-2.0	_	2.0	%

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
External load capactiance, OUTSCALE=0	C _{LOAD}		_	_	75	pF

- 1. In differential mode, the output is defined as the difference between two single-ended outputs. Absolute voltage on each output is limited to the single-ended range.
- 2. Supply current specifications are for VDAC circuitry operating with static output only and do not include current required to drive the load.
- 3. Current from HFPERCLK is dependent on HFPERCLK frequency. This current contributes to the total supply current used when the clock to the DAC peripheral is enabled in the CMU.
- 4. PSRR calculated as 20 * $log_{10}(\Delta VDD / \Delta V_{OUT})$, VDAC output at 90% of full scale
- 5. Entire range is monotonic and has no missing codes.
- 6. Gain is calculated by measuring the slope from 10% to 90% of full scale. Offset is calculated by comparing actual VDAC output at 10% of full scale to ideal VDAC output at 10% of full scale with the measured gain.

4.1.19 Current Digital to Analog Converter (IDAC)

Table 4.49. Current Digital to Analog Converter (IDAC)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Number of ranges	N _{IDAC_RANGES}		_	4	_	ranges
Output current	I _{IDAC_OUT}	RANGSEL ¹ = RANGE0	0.05	_	1.6	μA
		RANGSEL ¹ = RANGE1	1.6	_	4.7	μA
		RANGSEL ¹ = RANGE2	0.5	_	16	μA
		RANGSEL ¹ = RANGE3	2	_	64	μA
Linear steps within each range	N _{IDAC_STEPS}		_	32	_	steps
Step size	SS _{IDAC}	RANGSEL ¹ = RANGE0	_	50	_	nA
		RANGSEL ¹ = RANGE1	_	100	_	nA
		RANGSEL ¹ = RANGE2	_	500	_	nA
		RANGSEL ¹ = RANGE3	_	2	_	μA
Total accuracy, STEPSEL ¹ = A	ACC _{IDAC}	EM0 or EM1, AVDD=3.3 V, T = 25 °C	-3	_	3	%
		EM0 or EM1, Across operating temperature range	-18	_	22	%
		EM2 or EM3, Source mode, RANGSEL ¹ = RANGE0, AVDD=3.3 V, T = 25 °C	_	-2	_	%
		EM2 or EM3, Source mode, RANGSEL ¹ = RANGE1, AVDD=3.3 V, T = 25 °C	_	-1.7	_	%
		EM2 or EM3, Source mode, RANGSEL ¹ = RANGE2, AVDD=3.3 V, T = 25 °C	_	-0.8	_	%
		EM2 or EM3, Source mode, RANGSEL ¹ = RANGE3, AVDD=3.3 V, T = 25 °C	_	-0.5	_	%
		EM2 or EM3, Sink mode, RANG- SEL ¹ = RANGE0, AVDD=3.3 V, T = 25 °C	_	-0.7	_	%
		EM2 or EM3, Sink mode, RANG- SEL ¹ = RANGE1, AVDD=3.3 V, T = 25 °C	_	-0.6	_	%
		EM2 or EM3, Sink mode, RANG- SEL ¹ = RANGE2, AVDD=3.3 V, T = 25 °C	_	-0.5	_	%
		EM2 or EM3, Sink mode, RANG- SEL ¹ = RANGE3, AVDD=3.3 V, T = 25 °C	_	-0.5	_	%
Start up time	t _{IDAC_SU}	Output within 1% of steady state value	_	5	_	μs

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Settling time, (output settled	t _{IDAC_SETTLE}	Range setting is changed	_	5	_	μs
within 1% of steady state value),		Step value is changed	_	1	_	μs
Current consumption ²	I _{IDAC}	EM0 or EM1 Source mode, excluding output current, Across operating temperature range	_	11	15	μА
		EM0 or EM1 Sink mode, excluding output current, Across operating temperature range	_	13	18	μА
		EM2 or EM3 Source mode, excluding output current, T = 25 °C	_	0.023	_	μA
		EM2 or EM3 Sink mode, excluding output current, T = 25 °C	_	0.041	_	μA
		EM2 or EM3 Source mode, excluding output current, T ≥ 85 °C	_	11	_	μA
		EM2 or EM3 Sink mode, excluding output current, T ≥ 85 °C	_	13	_	μA
Output voltage compliance in source mode, source current change relative to current	I _{COMP_SRC}	RANGESEL1=0, output voltage = min(V _{IOVDD} , V _{AVDD} ² -100 mV)	_	0.11	_	%
sourced at 0 V		RANGESEL1=1, output voltage = min(V _{IOVDD} , V _{AVDD} ² -100 mV)	_	0.06	_	%
		RANGESEL1=2, output voltage = min(V _{IOVDD} , V _{AVDD} ² -150 mV)	_	0.04	_	%
		RANGESEL1=3, output voltage = min(V _{IOVDD} , V _{AVDD} ² -250 mV)	_	0.03	_	%
Output voltage compliance in sink mode, sink current	I _{COMP_SINK}	RANGESEL1=0, output voltage = 100 mV	_	0.12	_	%
change relative to current sunk at IOVDD		RANGESEL1=1, output voltage = 100 mV	_	0.05	_	%
		RANGESEL1=2, output voltage = 150 mV	_	0.04	_	%
		RANGESEL1=3, output voltage = 250 mV	_	0.03	_	%

- 1. In IDAC_CURPROG register.
- 2. The IDAC is supplied by either AVDD, DVDD, or IOVDD based on the setting of ANASW in the EMU_PWRCTRL register and PWRSEL in the IDAC_CTRL register. Setting PWRSEL to 1 selects IOVDD. With PWRSEL cleared to 0, ANASW selects between AVDD (0) and DVDD (1).

4.1.20 Capacitive Sense (CSEN)

Table 4.50. Capacitive Sense (CSEN)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Single conversion time (1x	t _{CNV}	12-bit SAR Conversions	_	20.2	_	μs
accumulation)		16-bit SAR Conversions	_	26.4	_	μs
		Delta Modulation Conversion (single comparison)	_	1.55	_	μs
Maximum external capacitive load	C _{EXTMAX}	IREFPROG=7 (Gain = 1x), including routing parasitics	_	68	_	pF
		IREFPROG=0 (Gain = 10x), including routing parasitics	_	680	_	pF
Maximum external series impedance	R _{EXTMAX}		_	1	_	kΩ
Supply current, EM2 bonded conversions, WARMUP-MODE=NORMAL, WAR-MUPCNT=0	CSEN_BOND	12-bit SAR conversions, 20 ms conversion rate, IREFPROG=7 (Gain = 1x), 10 channels bonded (total capacitance of 330 pF) ¹	_	326	_	nA
		Delta Modulation conversions, 20 ms conversion rate, IRE-FPROG=7 (Gain = 1x), 10 channels bonded (total capacitance of 330 pF) ¹	_	226	_	nA
		12-bit SAR conversions, 200 ms conversion rate, IREFPROG=7 (Gain = 1x), 10 channels bonded (total capacitance of 330 pF) ¹	_	33	_	nA
		Delta Modulation conversions, 200 ms conversion rate, IRE- FPROG=7 (Gain = 1x), 10 chan- nels bonded (total capacitance of 330 pF) ¹	_	25	_	nA
Supply current, EM2 scan conversions, WARMUP-MODE=NORMAL, WAR-MUDCNT-0	I _{CSEN_EM2}	12-bit SAR conversions, 20 ms scan rate, IREFPROG=0 (Gain = 10x), 8 samples per scan ¹	_	690	_	nA
MUPCNT=0		Delta Modulation conversions, 20 ms scan rate, 8 comparisons per sample (DMCR = 1, DMR = 2), IREFPROG=0 (Gain = 10x), 8 samples per scan ¹	_	515	_	nA
		12-bit SAR conversions, 200 ms scan rate, IREFPROG=0 (Gain = 10x), 8 samples per scan ¹	_	79	_	nA
		Delta Modulation conversions, 200 ms scan rate, 8 comparisons per sample (DMCR = 1, DMR = 2), IREFPROG=0 (Gain = 10x), 8 samples per scan ¹	_	57	_	nA

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Supply current, continuous conversions, WARMUP-MODE=KEEPCSENWARM	ICSEN_ACTIVE	SAR or Delta Modulation conversions of 33 pF capacitor, IRE-FPROG=0 (Gain = 10x), always on	_	90.5	_	μА
HFPERCLK supply current	ICSEN_HFPERCLK	Current contribution from HFPERCLK when clock to CSEN block is enabled.	_	2.25	_	μA/MHz

^{1.} Current is specified with a total external capacitance of 33 pF per channel. Average current is dependent on how long the peripheral is actively sampling channels within the scan period, and scales with the number of samples acquired. Supply current for a specific application can be estimated by multiplying the current per sample by the total number of samples per period (total_current = single_sample_current * (number_of_channels * accumulation)).

4.1.21 Operational Amplifier (OPAMP)

Unless otherwise indicated, specified conditions are: Non-inverting input configuration, VDD = 3.3 V, DRIVESTRENGTH = 2, MAIN-OUTEN = 1, C_{LOAD} = 75 pF with OUTSCALE = 0, or C_{LOAD} = 37.5 pF with OUTSCALE = 1. Unit gain buffer and 3X-gain connection as specified in table footnotes^{1 2}.

Table 4.51. Operational Amplifier (OPAMP)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Supply voltage (from AVDD)	V _{OPA}	HCMDIS = 0, Rail-to-rail input range	2	_	3.8	V
		HCMDIS = 1	1.62	_	3.8	V
Input voltage	V _{IN}	HCMDIS = 0, Rail-to-rail input range	V _{VSS}	_	V _{OPA}	V
		HCMDIS = 1	V _{VSS}	_	V _{OPA} -1.2	V
Input impedance	R _{IN}		100	_	_	МΩ
Output voltage	V _{OUT}		V _{VSS}	_	V _{OPA}	V
Load capacitance ³	C _{LOAD}	OUTSCALE = 0	_	_	75	pF
		OUTSCALE = 1	_	_	37.5	pF
Output impedance	R _{OUT}	DRIVESTRENGTH = 2 or 3, 0.4 V \leq V _{OUT} \leq V _{OPA} - 0.4 V, -8 mA < I _{OUT} < 8 mA, Buffer connection, Full supply range	_	0.25	_	Ω
		DRIVESTRENGTH = 0 or 1, 0.4 V \leq V _{OUT} \leq V _{OPA} - 0.4 V, -400 μ A $<$ I _{OUT} $<$ 400 μ A, Buffer connection, Full supply range	_	0.6	_	Ω
		DRIVESTRENGTH = 2 or 3, 0.1 V \leq V _{OUT} \leq V _{OPA} - 0.1 V, -2 mA $<$ I _{OUT} $<$ 2 mA, Buffer connection, Full supply range	_	0.4	_	Ω
		DRIVESTRENGTH = 0 or 1, 0.1 V \leq V _{OUT} \leq V _{OPA} - 0.1 V, -100 μ A $<$ I _{OUT} $<$ 100 μ A, Buffer connection, Full supply range	_	1	_	Ω
Internal closed-loop gain	G _{CL}	Buffer connection	0.99	1	1.01	-
		3x Gain connection	2.93	2.99	3.05	-
		16x Gain connection	15.07	15.7	16.33	-
Active current ⁴	I _{OPA}	DRIVESTRENGTH = 3, OUT- SCALE = 0	_	580	_	μA
		DRIVESTRENGTH = 2, OUT- SCALE = 0	_	176	_	μA
		DRIVESTRENGTH = 1, OUT- SCALE = 0	_	13	_	μΑ
		DRIVESTRENGTH = 0, OUT- SCALE = 0	_	4.7	_	μΑ

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Open-loop gain	G _{OL}	DRIVESTRENGTH = 3	_	135	_	dB
		DRIVESTRENGTH = 2	_	137	_	dB
		DRIVESTRENGTH = 1	_	121	_	dB
		DRIVESTRENGTH = 0	_	109	_	dB
Loop unit-gain frequency ⁵	UGF	DRIVESTRENGTH = 3, Buffer connection	_	3.38	_	MHz
		DRIVESTRENGTH = 2, Buffer connection	_	0.9	_	MHz
		DRIVESTRENGTH = 1, Buffer connection	_	132	_	kHz
		DRIVESTRENGTH = 0, Buffer connection	_	34	_	kHz
		DRIVESTRENGTH = 3, 3x Gain connection	_	2.57	_	MHz
		DRIVESTRENGTH = 2, 3x Gain connection	_	0.71	_	MHz
		DRIVESTRENGTH = 1, 3x Gain connection	_	113	_	kHz
		DRIVESTRENGTH = 0, 3x Gain connection	_	28	_	kHz
Phase margin	PM	DRIVESTRENGTH = 3, Buffer connection	_	67	_	0
		DRIVESTRENGTH = 2, Buffer connection	_	69	_	0
		DRIVESTRENGTH = 1, Buffer connection	_	63	_	o
		DRIVESTRENGTH = 0, Buffer connection	_	68	_	o
Output voltage noise	N _{OUT}	DRIVESTRENGTH = 3, Buffer connection, 10 Hz - 10 MHz	_	146	_	μVrms
		DRIVESTRENGTH = 2, Buffer connection, 10 Hz - 10 MHz	_	163	_	μVrms
		DRIVESTRENGTH = 1, Buffer connection, 10 Hz - 1 MHz	_	170	_	μVrms
		DRIVESTRENGTH = 0, Buffer connection, 10 Hz - 1 MHz	_	176	_	μVrms
		DRIVESTRENGTH = 3, 3x Gain connection, 10 Hz - 10 MHz	_	313	_	μVrms
		DRIVESTRENGTH = 2, 3x Gain connection, 10 Hz - 10 MHz	_	271	_	μVrms
		DRIVESTRENGTH = 1, 3x Gain connection, 10 Hz - 1 MHz	_	247	_	μVrms
		DRIVESTRENGTH = 0, 3x Gain connection, 10 Hz - 1 MHz	_	245	_	μVrms

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Slew rate ⁶	SR	DRIVESTRENGTH = 3, INCBW=1 ⁷	_	4.7	_	V/µs
		DRIVESTRENGTH = 3, INCBW=0	_	1.5	_	V/µs
		DRIVESTRENGTH = 2, INCBW=1 ⁷	_	1.27	_	V/µs
		DRIVESTRENGTH = 2, INCBW=0	_	0.42	_	V/µs
		DRIVESTRENGTH = 1, INCBW=1 ⁷	_	0.17	_	V/µs
		DRIVESTRENGTH = 1, INCBW=0	_	0.058	_	V/µs
		DRIVESTRENGTH = 0, INCBW=1 ⁷	_	0.044	_	V/µs
		DRIVESTRENGTH = 0, INCBW=0	_	0.015	_	V/µs
Startup time ⁸	T _{START}	DRIVESTRENGTH = 2	_	_	12	μs
Input offset voltage	V _{OSI}	DRIVESTRENGTH = 2 or 3, T = 25 °C	-2	_	2	mV
		DRIVESTRENGTH = 1 or 0, T = 25 °C	-2	_	2	mV
		DRIVESTRENGTH = 2 or 3, across operating temperature range	-12	_	12	mV
		DRIVESTRENGTH = 1 or 0, across operating temperature range	-30	_	30	mV
DC power supply rejection ratio ⁹	PSRR _{DC}	Input referred	_	70	_	dB
DC common-mode rejection ratio ⁹	CMRR _{DC}	Input referred	_	70	_	dB
Total harmonic distortion	THD _{OPA}	DRIVESTRENGTH = 2, 3x Gain connection, 1 kHz, V _{OUT} = 0.1 V to V _{OPA} - 0.1 V	_	90	_	dB
		DRIVESTRENGTH = 0, 3x Gain connection, 0.1 kHz, V _{OUT} = 0.1 V to V _{OPA} - 0.1 V	_	90	_	dB

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
	_					

- 1. Specified configuration for Unit gain buffer configuration is: INCBW = 0, HCMDIS = 0, RESINSEL = DISABLE. $V_{INPUT} = 0.5 \text{ V}$, $V_{OUTPUT} = 0.5 \text{ V}$.
- 2. Specified configuration for 3X-Gain configuration is: INCBW = 1, HCMDIS = 1, RESINSEL = VSS, V_{INPUT} = 0.5 V, V_{OUTPUT} = 1.5 V. Nominal voltage gain is 3.
- 3. If the maximum C_{I OAD} is exceeded, an isolation resistor is required for stability. See AN0038 for more information.
- 4. Current into the load resistor is excluded. When the OPAMP is connected with closed-loop gain > 1, there will be extra current to drive the resistor feedback network. The internal resistor feedback network has total resistance of 143.5 kOhm, which will cause another ~10 μA current when the OPAMP drives 1.5 V between output and ground.
- 5. In unit gain connection, UGF is the gain-bandwidth product of the OPAMP. In 3x Gain connection, UGF is the gain-bandwidth product of the OPAMP and 1/3 attenuation of the feedback network.
- 6. Step between 0.2V and V_{OPA}-0.2V, 10%-90% rising/falling range.
- 7. When INCBW is set to 1 the OPAMP bandwidth is increased. This is allowed only when the non-inverting close-loop gain is ≥ 3, or the OPAMP may not be stable.
- 8. From enable to output settled. In sample-and-off mode, RC network after OPAMP will contribute extra delay. Settling error < 1mV.
- 9. When HCMDIS=1 and input common mode transitions the region from V_{OPA}-1.4V to V_{OPA}-1V, input offset will change. PSRR and CMRR specifications do not apply to this transition region.

4.1.22 Pulse Counter (PCNT)

Table 4.52. Pulse Counter (PCNT)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Input frequency	F _{IN}	Asynchronous Single and Quadrature Modes	_	_	10	MHz
		Sampled Modes with Debounce filter set to 0.	_	_	8	kHz

4.1.23 Analog Port (APORT)

Table 4.53. Analog Port (APORT)

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
Supply current ^{1 2}	I _{APORT}	Operation in EM0/EM1	_	7	_	μA
		Operation in EM2/EM3	_	63	_	nA

- 1. Supply current increase that occurs when an analog peripheral requests access to APORT. This current is not included in reported peripheral currents. Additional peripherals requesting access to APORT do not incur further current.
- 2. Specified current is for continuous APORT operation. In applications where the APORT is not requested continuously (e.g. periodic ACMP requests from LESENSE in EM2), the average current requirements can be estimated by mutiplying the duty cycle of the requests by the specified continuous current number.

4.1.24 I2C

4.1.24.1 I2C Standard-mode (Sm)¹

Table 4.54. I2C Standard-mode (Sm)¹

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
SCL clock frequency ²	f _{SCL}		0	_	100	kHz
SCL clock low time	t _{LOW}		4.7	_	_	μs
SCL clock high time	t _{ніGн}		4	_	_	μs
SDA set-up time	t _{SU_DAT}		250	_	_	ns
SDA hold time ³	t _{HD_DAT}		100	_	3450	ns
Repeated START condition set-up time	t _{SU_STA}		4.7	_	_	μs
(Repeated) START condition hold time	t _{HD_STA}		4	_	_	μs
STOP condition set-up time	t _{SU_STO}		4	_	_	μs
Bus free time between a STOP and START condition	t _{BUF}		4.7	_	_	μs

- 1. For CLHR set to 0 in the I2Cn_CTRL register.
- 2. For the minimum HFPERCLK frequency required in Standard-mode, refer to the I2C chapter in the reference manual.
- 3. The maximum SDA hold time ($t_{HD\ DAT}$) needs to be met only when the device does not stretch the low time of SCL (t_{LOW}).

4.1.24.2 I2C Fast-mode (Fm)¹

Table 4.55. I2C Fast-mode (Fm)¹

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
SCL clock frequency ²	f _{SCL}		0	_	400	kHz
SCL clock low time	t _{LOW}		1.3	_	_	μs
SCL clock high time	t _{HIGH}		0.6	_	_	μs
SDA set-up time	t _{SU_DAT}		100	_	_	ns
SDA hold time ³	t _{HD_DAT}		100	_	900	ns
Repeated START condition set-up time	t _{SU_STA}		0.6	_	_	μs
(Repeated) START condition hold time	t _{HD_STA}		0.6	_	_	μs
STOP condition set-up time	t _{SU_STO}		0.6	_	_	μs
Bus free time between a STOP and START condition	t _{BUF}		1.3	_	_	μs

- 1. For CLHR set to 1 in the I2Cn_CTRL register.
- 2. For the minimum HFPERCLK frequency required in Fast-mode, refer to the I2C chapter in the reference manual.
- 3. The maximum SDA hold time (t_{HD.DAT}) needs to be met only when the device does not stretch the low time of SCL (t_{LOW}).

4.1.24.3 I2C Fast-mode Plus (Fm+)¹

Table 4.56. I2C Fast-mode Plus (Fm+)¹

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
SCL clock frequency ²	f _{SCL}		0	_	1000	kHz
SCL clock low time	t _{LOW}		0.5	_	_	μs
SCL clock high time	t _{HIGH}		0.26	_	_	μs
SDA set-up time	t _{SU_DAT}		50	_	_	ns
SDA hold time	t _{HD_DAT}		100	_	_	ns
Repeated START condition set-up time	t _{SU_STA}		0.26	_	_	μs
(Repeated) START condition hold time	t _{HD_STA}		0.26	_	_	μs
STOP condition set-up time	t _{SU_STO}		0.26	_	_	μs
Bus free time between a STOP and START condition	t _{BUF}		0.5	_	_	μs

- 1. For CLHR set to 0 or 1 in the I2Cn_CTRL register.
- 2. For the minimum HFPERCLK frequency required in Fast-mode Plus, refer to the I2C chapter in the reference manual.

4.1.25 **USART SPI**

SPI Master Timing

Table 4.57. SPI Master Timing

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
SCLK period ^{1 2 3}	t _{SCLK}		2 * t _{HFPERCLK}	_	_	ns
CS to MOSI 1 2	t _{CS_MO}		-12.5	_	14	ns
SCLK to MOSI 1 2	t _{SCLK_MO}		-8.5	_	10.5	ns
MISO setup time ^{1 2}	t _{SU_MI}	IOVDD = 1.62 V	90	_	_	ns
		IOVDD = 3.0 V	42	_	_	ns
MISO hold time ^{1 2}	t _{H_MI}		-9	_	_	ns

- 1. Applies for both CLKPHA = 0 and CLKPHA = 1 (figure only shows CLKPHA = 0).
- 2. Measurement done with 8 pF output loading at 10% and 90% of V_{DD} (figure shows 50% of V_{DD}).
- $3.\,t_{\mbox{\scriptsize HFPERCLK}}$ is one period of the selected HFPERCLK.

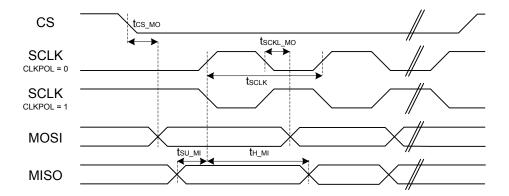


Figure 4.1. SPI Master Timing Diagram

SPI Slave Timing

Table 4.58. SPI Slave Timing

Parameter	Symbol	Test Condition	Min	Тур	Max	Unit
SCLK period ^{1 2 3}	t _{SCLK}		6 * ^t HFPERCLK	_	_	ns
SCLK high time ^{1 2 3}	t _{SCLK_HI}		2.5 * the the thick the th	_	_	ns
SCLK low time ^{1 2 3}	t _{SCLK_LO}		2.5 * the three th	_	_	ns
CS active to MISO ^{1 2}	tcs_act_mi		4	_	70	ns
CS disable to MISO ^{1 2}	tcs_dis_mi		4	_	50	ns
MOSI setup time ^{1 2}	tsu_mo		12.5	_	_	ns
MOSI hold time ^{1 2 3}	t _{H_MO}		13	_	_	ns
SCLK to MISO ^{1 2 3}	t _{SCLK_MI}		6 + 1.5 * thfperclk	_	45 + 2.5 * t _{HFPERCLK}	ns

Note:

- 1. Applies for both CLKPHA = 0 and CLKPHA = 1 (figure only shows CLKPHA = 0).
- 2. Measurement done with 8 pF output loading at 10% and 90% of V_{DD} (figure shows 50% of V_{DD}).
- 3. t_{HFPERCLK} is one period of the selected HFPERCLK.

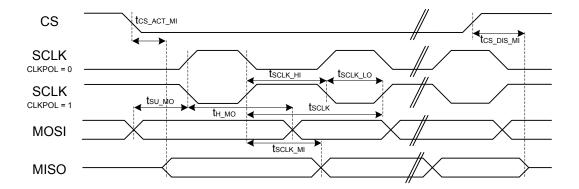


Figure 4.2. SPI Slave Timing Diagram

4.2 Typical Performance Curves

Typical performance curves indicate typical characterized performance under the stated conditions.

4.2.1 Supply Current

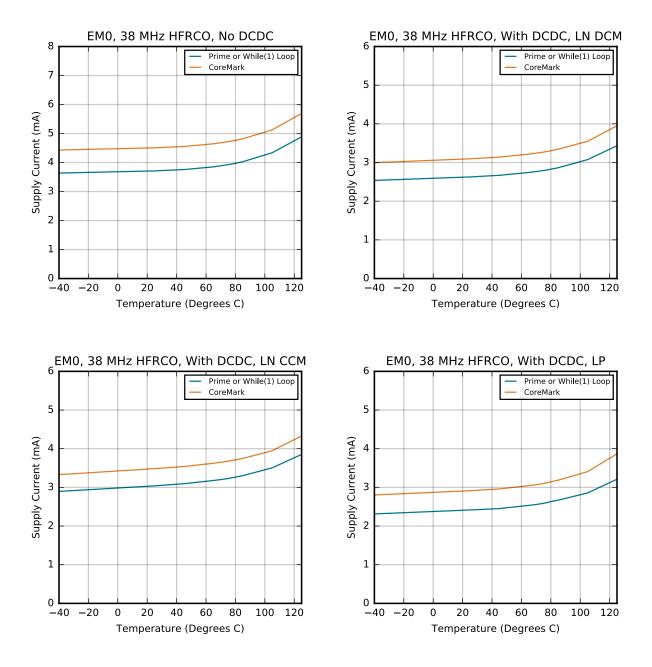


Figure 4.3. EM0 Active Mode Typical Supply Current vs. Temperature

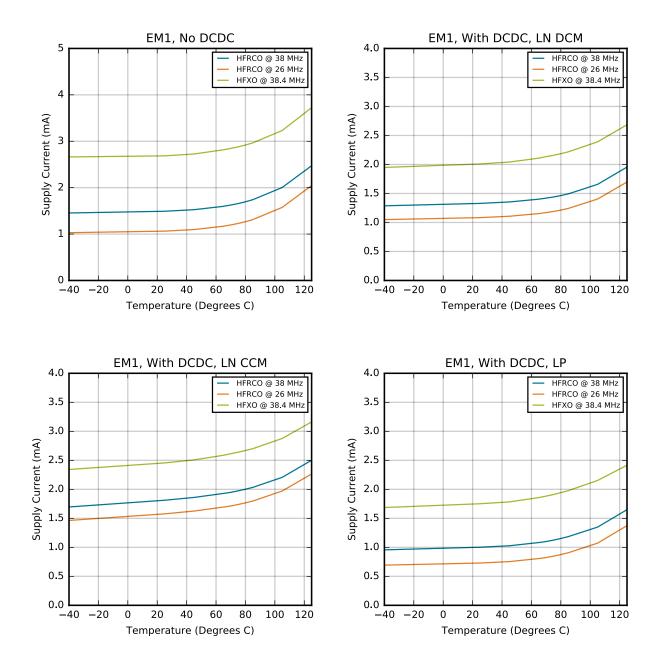


Figure 4.4. EM1 Sleep Mode Typical Supply Current vs. Temperature

Typical supply current for EM2, EM3 and EM4H using standard software libraries from Silicon Laboratories.

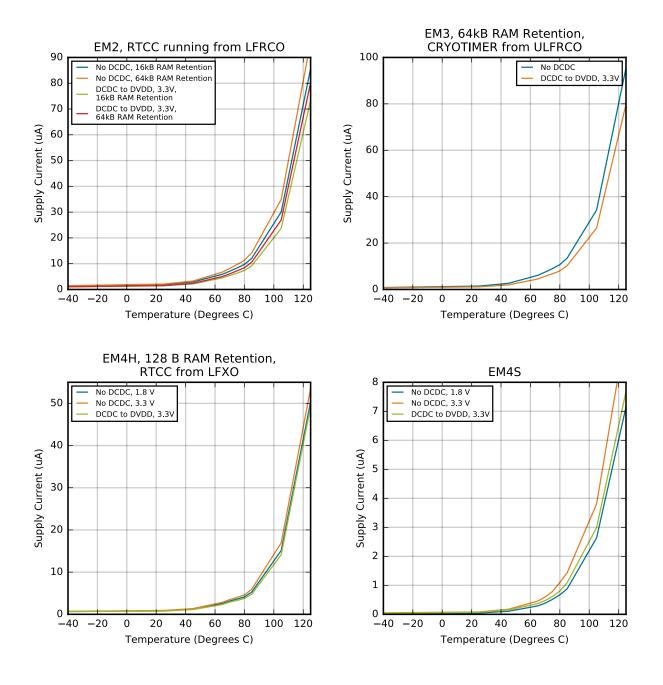


Figure 4.5. EM2, EM3, EM4H and EM4S Typical Supply Current vs. Temperature

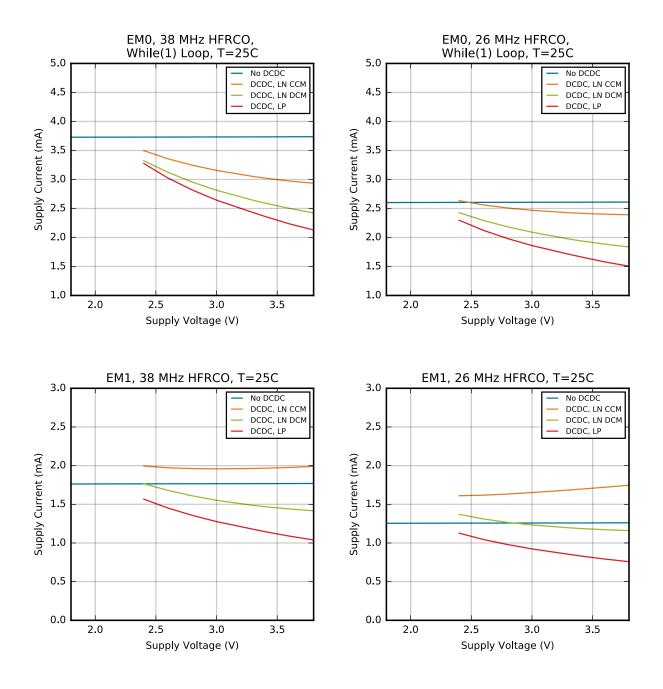


Figure 4.6. EM0 and EM1 Mode Typical Supply Current vs. Supply

Typical supply current for EM2, EM3 and EM4H using standard software libraries from Silicon Laboratories.

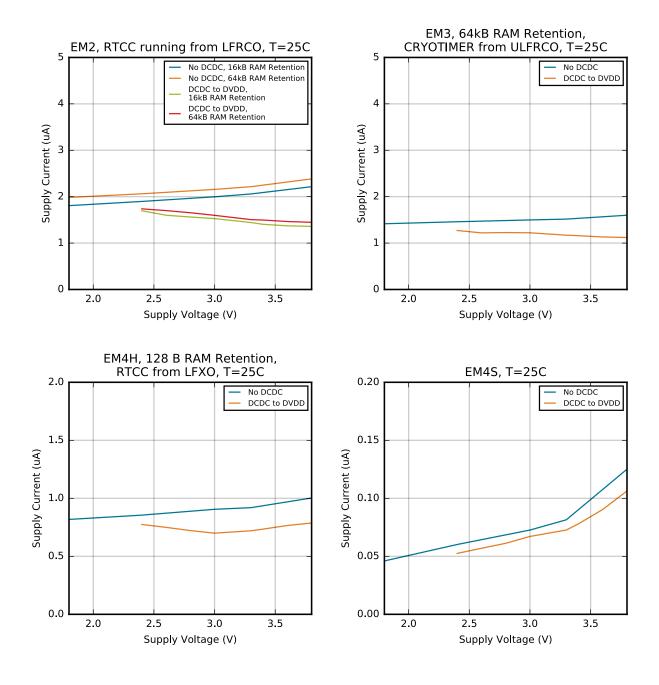


Figure 4.7. EM2, EM3, EM4H and EM4S Typical Supply Current vs. Supply

4.2.2 DC-DC Converter

Default test conditions: CCM mode, LDCDC = $4.7 \mu H$, CDCDC = $4.7 \mu F$, VDCDC_I = 3.3 V, VDCDC_O = 1.8 V, FDCDC_LN = 7 MHz

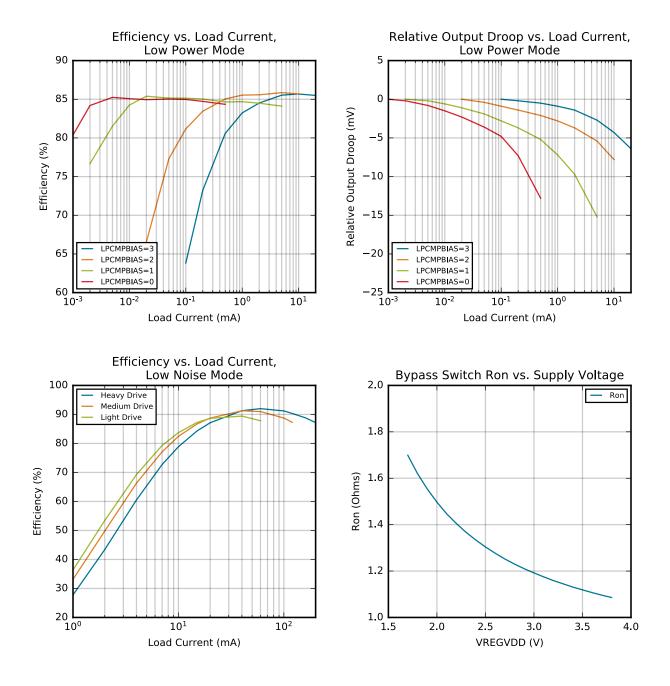


Figure 4.8. DC-DC Converter Typical Performance Characteristics

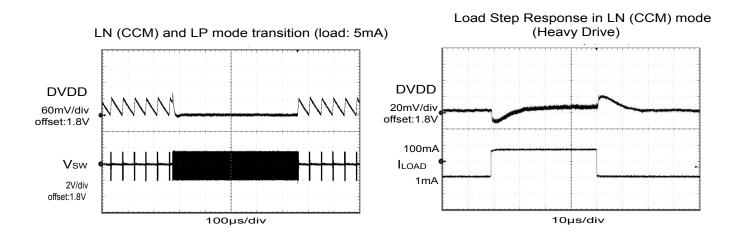
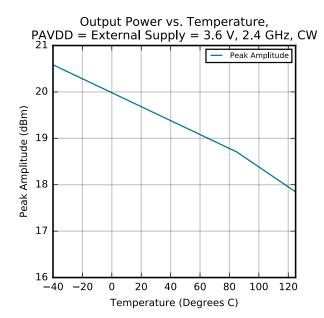
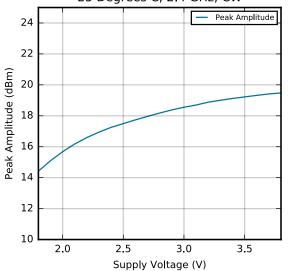


Figure 4.9. DC-DC Converter Transition Waveforms



Output Power vs. Supply, PAVDD = External Supply, 25 Degrees C, 2.4 GHz, CW



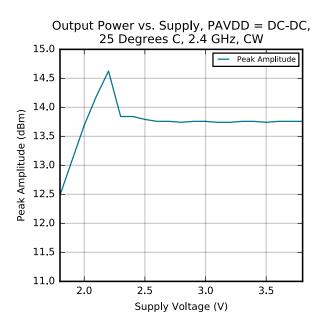


Figure 4.10. 2.4 GHz RF Transmitter Output Power

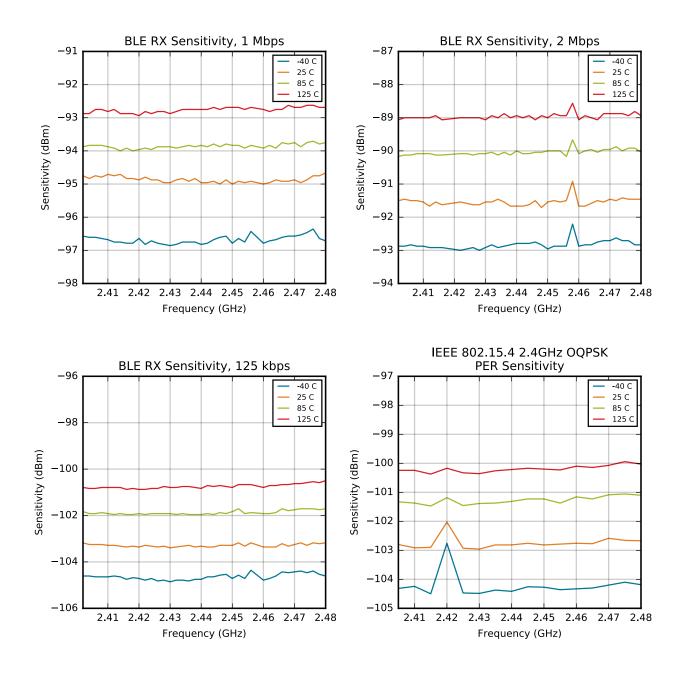


Figure 4.11. 2.4 GHz RF Receiver Sensitivity

5. Typical Connection Diagrams

5.1 Power

Typical power supply connections for direct supply, without using the internal DC-DC converter, are shown in the following figure.

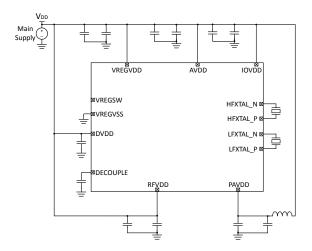


Figure 5.1. EFR32FG13 Typical Application Circuit: Direct Supply Configuration without DC-DC converter

Typical power supply circuits using the internal DC-DC converter are shown below. The MCU operates from the DC-DC converter supply. For low RF transmit power applications less than 13dBm, the RF PA may be supplied by the DC-DC converter. For OPNs supporting high power RF transmission, the RF PA must be directly supplied by VDD for RF transmit power greater than 13 dBm.

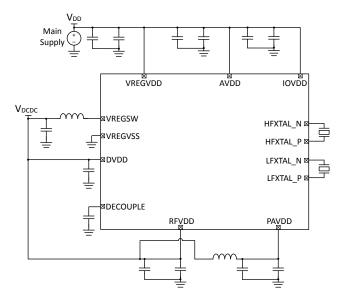


Figure 5.2. EFR32FG13 Typical Application Circuit: Configuration with DC-DC converter (PAVDD from VDCDC)

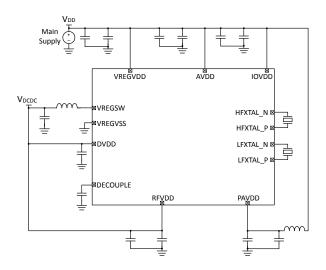


Figure 5.3. EFR32FG13 Typical Application Circuit: Configuration with DC-DC converter (PAVDD from VDD)

5.2 RF Matching Networks

Typical RF matching network circuit diagrams are shown in Figure 5.4 Typical 2.4 GHz RF impedance-matching network circuits on page 120 for applications in the 2.4GHz band, and in Figure 5.5 Typical Sub-GHz RF impedance-matching network circuits on page 120 for applications in the sub-GHz band. Application-specific component values can be found in the EFR32xG13 Reference Manual. For low RF transmit power applications less than 13dBm, the two-element match is recommended. For OPNs supporting high power RF transmission, the four-element match is recommended for high RF transmit power (> 13dBm).

2-Element Match for 2.4GHz Band

4-Element Match for 2.4GHz Band

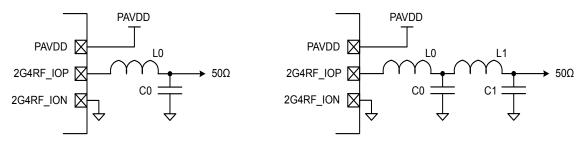
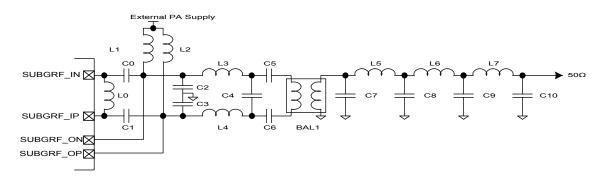


Figure 5.4. Typical 2.4 GHz RF impedance-matching network circuits

Sub-GHz Match Topology I (169-500 MHz)



Sub-GHz Match Topology 2 (500-915 MHz)

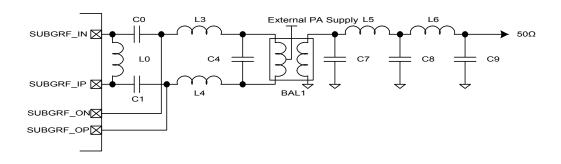


Figure 5.5. Typical Sub-GHz RF impedance-matching network circuits

5.3 Other Connections

Other components or connections may be required to meet the system-level requirements. Application Note AN0002: "Hardware Design Considerations" contains detailed information on these connections. Application Notes can be accessed on the Silicon Labs website (www.silabs.com/32bit-appnotes).

6. Pin Definitions

6.1 QFN48 2.4 GHz and Sub-GHz Device Pinout

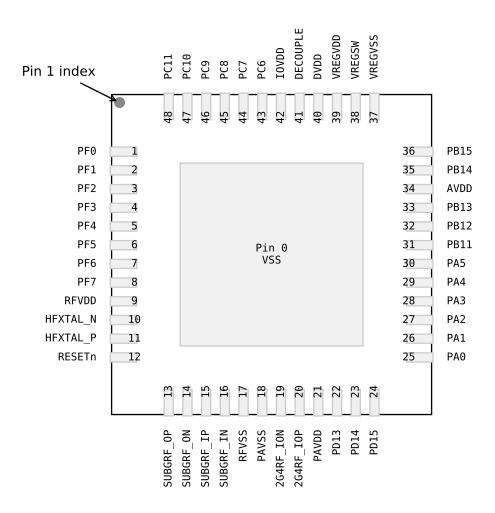


Figure 6.1. QFN48 2.4 GHz and Sub-GHz Device Pinout

Table 6.1. QFN48 2.4 GHz and Sub-GHz Device Pinout

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
VSS	0	Ground	PF0	1	GPIO (5V)	
PF1	2	GPIO (5V)	PF2	3	GPIO (5V)	
PF3	4	GPIO (5V)	PF4	5	GPIO (5V)	
PF5	6	GPIO (5V)	PF6	7	GPIO (5V)	
PF7	8	GPIO (5V)	RFVDD	9	Radio power supply	
HFXTAL_N	10	High Frequency Crystal input pin.	HFXTAL_P	11	High Frequency Crystal output pin.	

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
RESETn	12	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.	SUBGRF_OP	13	Sub GHz Differential RF output, positive path.	
SUBGRF_ON	14	Sub GHz Differential RF output, negative path.	SUBGRF_IP	15	Sub GHz Differential RF input, positive path.	
SUBGRF_IN	16	Sub GHz Differential RF input, negative path.	RFVSS	17	Radio Ground	
PAVSS	18	Power Amplifier (PA) voltage regulator VSS	2G4RF_ION	19	2.4 GHz Differential RF input/output, negative path. This pin should be externally grounded.	
2G4RF_IOP	20	2.4 GHz Differential RF input/output, positive path.	PAVDD	21	Power Amplifier (PA) voltage regulator VDD input	
PD13	22	GPIO	PD14	23	GPIO	
PD15	24	GPIO	PA0	25	GPIO	
PA1	26	GPIO	PA2	27	GPIO	
PA3	28	GPIO	PA4	29	GPIO	
PA5	30	GPIO (5V)	PB11	31	GPIO	
PB12	32	GPIO	PB13	33	GPIO	
AVDD	34	Analog power supply.	PB14	35	GPIO	
PB15	36	GPIO	VREGVSS	37	Voltage regulator VSS	
VREGSW	38	DCDC regulator switching node	VREGVDD	39	Voltage regulator VDD input	
DVDD	40	Digital power supply.	DECOUPLE	41	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.	
IOVDD	42	Digital IO power supply.	PC6	43	GPIO (5V)	
PC7	44	GPIO (5V)	PC8	45	GPIO (5V)	
PC9	46	GPIO (5V)	PC10	47	GPIO (5V)	
PC11	48	GPIO (5V)				

1. GPIO with 5V tolerance are indicated by (5V).

6.2 QFN48 2.4 GHz Device Pinout

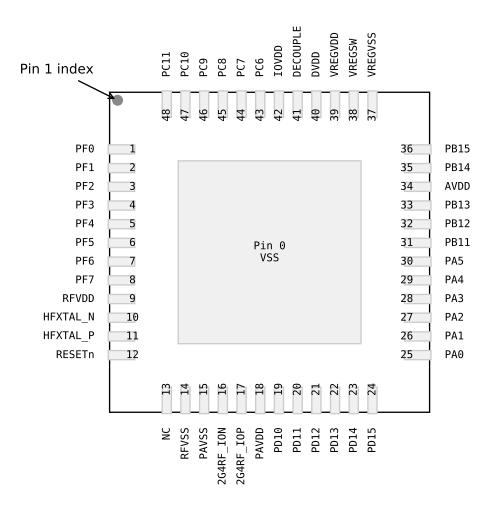


Figure 6.2. QFN48 2.4 GHz Device Pinout

Table 6.2. QFN48 2.4 GHz Device Pinout

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
VSS	0	Ground	PF0	1	GPIO (5V)	
PF1	2	GPIO (5V)	PF2	3	GPIO (5V)	
PF3	4	GPIO (5V)	PF4	5	GPIO (5V)	
PF5	6	GPIO (5V)	PF6	7	GPIO (5V)	
PF7	8	GPIO (5V)	RFVDD	9	Radio power supply	
HFXTAL_N	10	High Frequency Crystal input pin.	HFXTAL_P	11	High Frequency Crystal output pin.	

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
RESETn	12	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.	NC	13	No Connect.	
RFVSS	14	Radio Ground	PAVSS	15	Power Amplifier (PA) voltage regulator VSS	
2G4RF_ION	16	2.4 GHz Differential RF input/output, negative path. This pin should be externally grounded.	2G4RF_IOP	17	2.4 GHz Differential RF input/output, positive path.	
PAVDD	18	Power Amplifier (PA) voltage regulator VDD input	PD10	19	GPIO (5V)	
PD11	20	GPIO (5V)	PD12	21	GPIO (5V)	
PD13	22	GPIO	PD14	23	GPIO	
PD15	24	GPIO	PA0	25	GPIO	
PA1	26	GPIO	PA2	27	GPIO	
PA3	28	GPIO	PA4	29	GPIO	
PA5	30	GPIO (5V)	PB11	31	GPIO	
PB12	32	GPIO	PB13	33	GPIO	
AVDD	34	Analog power supply.	PB14	35	GPIO	
PB15	36	GPIO	VREGVSS	37	Voltage regulator VSS	
VREGSW	38	DCDC regulator switching node	VREGVDD	39	Voltage regulator VDD input	
DVDD	40	Digital power supply.	DECOUPLE	41	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.	
IOVDD	42	Digital IO power supply.	PC6	43	GPIO (5V)	
PC7	44	GPIO (5V)	PC8	45	GPIO (5V)	
PC9	46	GPIO (5V)	PC10	47	GPIO (5V)	
PC11	48	GPIO (5V)				

1. GPIO with 5V tolerance are indicated by (5V).

6.3 QFN48 Sub-GHz Device Pinout

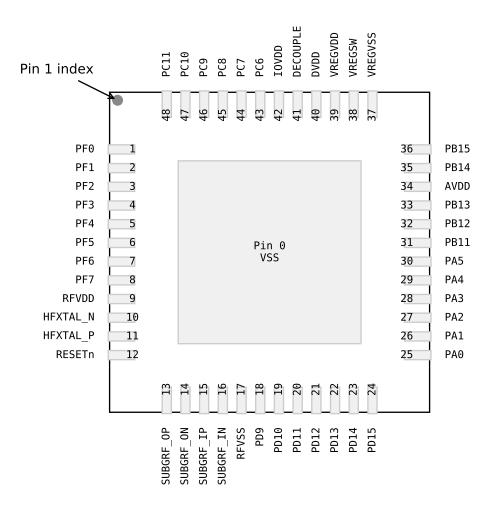


Figure 6.3. QFN48 Sub-GHz Device Pinout

Table 6.3. QFN48 Sub-GHz Device Pinout

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
VSS	0	Ground	PF0	1	GPIO (5V)	
PF1	2	GPIO (5V)	PF2	3	GPIO (5V)	
PF3	4	GPIO (5V)	PF4	5	GPIO (5V)	
PF5	6	GPIO (5V)	PF6	7	GPIO (5V)	
PF7	8	GPIO (5V)	RFVDD	9	Radio power supply	
HFXTAL_N	10	High Frequency Crystal input pin.	HFXTAL_P	11	High Frequency Crystal output pin.	

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
RESETn	12	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.	SUBGRF_OP	13	Sub GHz Differential RF output, positive path.	
SUBGRF_ON	14	Sub GHz Differential RF output, negative path.	SUBGRF_IP	15	Sub GHz Differential RF input, positive path.	
SUBGRF_IN	16	Sub GHz Differential RF input, negative path.	RFVSS	17	Radio Ground	
PD9	18	GPIO (5V)	PD10	19	GPIO (5V)	
PD11	20	GPIO (5V)	PD12	21	GPIO (5V)	
PD13	22	GPIO	PD14	23	GPIO	
PD15	24	GPIO	PA0	25	GPIO	
PA1	26	GPIO	PA2	27	GPIO	
PA3	28	GPIO	PA4	29	GPIO	
PA5	30	GPIO (5V)	PB11	31	GPIO	
PB12	32	GPIO	PB13	33	GPIO	
AVDD	34	Analog power supply.	PB14	35	GPIO	
PB15	36	GPIO	VREGVSS	37	Voltage regulator VSS	
VREGSW	38	DCDC regulator switching node	VREGVDD	39	Voltage regulator VDD input	
DVDD	40	Digital power supply.	DECOUPLE	41	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.	
IOVDD	42	Digital IO power supply.	PC6	43	GPIO (5V)	
PC7	44	GPIO (5V)	PC8	45	GPIO (5V)	
PC9	46	GPIO (5V)	PC10	47	GPIO (5V)	
PC11	48	GPIO (5V)				

1. GPIO with 5V tolerance are indicated by (5V).

6.4 QFN32 2.4 GHz Device Pinout

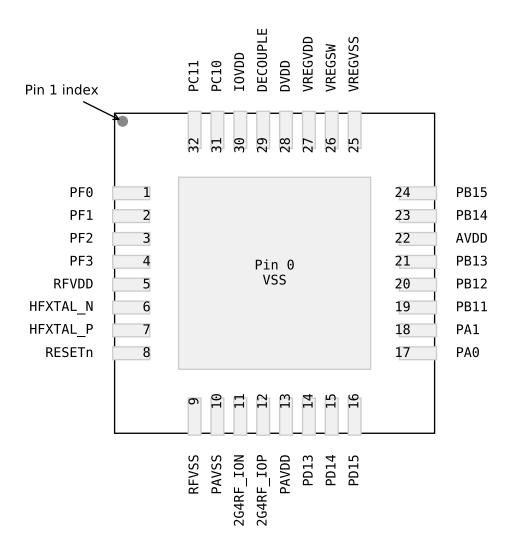


Figure 6.4. QFN32 2.4 GHz Device Pinout

Table 6.4. QFN32 2.4 GHz Device Pinout

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description
VSS	0	Ground	PF0	1	GPIO (5V)
PF1	2	GPIO (5V)	PF2	3	GPIO (5V)
PF3	4	GPIO (5V)	RFVDD	5	Radio power supply
HFXTAL_N	6	High Frequency Crystal input pin.	HFXTAL_P	7	High Frequency Crystal output pin.
RESETn	8	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.	RFVSS	9	Radio Ground

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
PAVSS	10	Power Amplifier (PA) voltage regulator VSS	2G4RF_ION	11	2.4 GHz Differential RF input/output, negative path. This pin should be externally grounded.	
2G4RF_IOP	12	2.4 GHz Differential RF input/output, positive path.	PAVDD	13	Power Amplifier (PA) voltage regulator VDD input	
PD13	14	GPIO	PD14	15	GPIO	
PD15	16	GPIO	PA0	17	GPIO	
PA1	18	GPIO	PB11	19	GPIO	
PB12	20	GPIO	PB13	21	GPIO	
AVDD	22	Analog power supply.	PB14	23	GPIO	
PB15	24	GPIO	VREGVSS	25	Voltage regulator VSS	
VREGSW	26	DCDC regulator switching node	VREGVDD	27	Voltage regulator VDD input	
DVDD	28	Digital power supply.	DECOUPLE	29	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.	
IOVDD	30	Digital IO power supply.	PC10	31	GPIO (5V)	
PC11	32	GPIO (5V)				

^{1.} GPIO with 5V tolerance are indicated by (5V).

6.5 QFN32 Sub-GHz Device Pinout

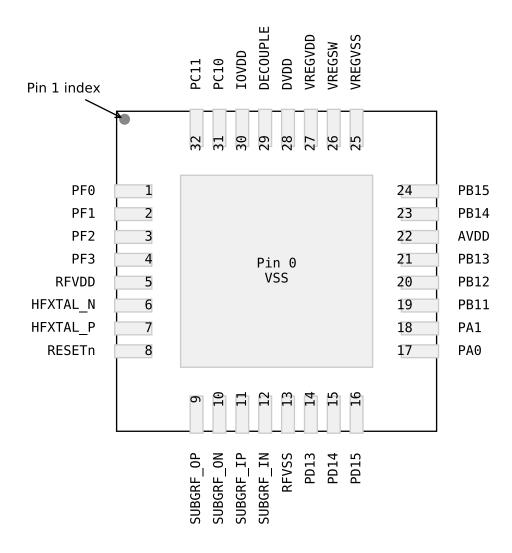


Figure 6.5. QFN32 Sub-GHz Device Pinout

Table 6.5. QFN32 Sub-GHz Device Pinout

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description
VSS	0	Ground	PF0	1	GPIO (5V)
PF1	2	GPIO (5V)	PF2	3	GPIO (5V)
PF3	4	GPIO (5V)	RFVDD	5	Radio power supply
HFXTAL_N	6	High Frequency Crystal input pin.	HFXTAL_P	7	High Frequency Crystal output pin.
RESETn	8	Reset input, active low. To apply an external reset source to this pin, it is required to only drive this pin low during reset, and let the internal pull-up ensure that reset is released.	SUBGRF_OP	9	Sub GHz Differential RF output, positive path.

Pin Name	Pin(s)	Description	Pin Name	Pin(s)	Description	
SUBGRF_ON	10	Sub GHz Differential RF output, negative path.	SUBGRF_IP	11	Sub GHz Differential RF input, positive path.	
SUBGRF_IN	12	Sub GHz Differential RF input, negative path.	RFVSS	13	Radio Ground	
PD13	14	GPIO	PD14	15	GPIO	
PD15	16	GPIO	PA0	17	GPIO	
PA1	18	GPIO	PB11	19	GPIO	
PB12	20	GPIO	PB13	21	GPIO	
AVDD	22	Analog power supply.	PB14	23	GPIO	
PB15	24	GPIO	VREGVSS	25	Voltage regulator VSS	
VREGSW	26	DCDC regulator switching node	VREGVDD	27	Voltage regulator VDD input	
DVDD	28	Digital power supply.	DECOUPLE	29	Decouple output for on-chip voltage regulator. An external decoupling capacitor is required at this pin.	
IOVDD	30	Digital IO power supply.	PC10	31	GPIO (5V)	
PC11	32	GPIO (5V)				

^{1.} GPIO with 5V tolerance are indicated by (5V).

6.6 GPIO Functionality Table

A wide selection of alternate functionality is available for multiplexing to various pins. The following table shows the name of each GPIO pin, followed by the functionality available on that pin. Refer to 6.7 Alternate Functionality Overview for a list of GPIO locations available for each function.

Table 6.6. GPIO Functionality Table

GPIO Name	Pin Alternate Functionality / Description							
	Analog	Timers	Communication	Radio	Other			
PA0	BUSDY BUSCX ADC0_EXTN	TIM0_CC0 #0 TIM0_CC1 #31 TIM0_CC2 #30 TIM0_CDTI0 #29 TIM0_CDTI1 #28 TIM0_CDTI2 #27 TIM1_CC0 #0 TIM1_CC1 #31 TIM1_CC2 #30 TIM1_CC3 #29 WTIM0_CC0 #0 LE- TIM0_OUT0 #0 LE- TIM0_OUT1 #31 PCNT0_S0IN #0 PCNT0_S1IN #31	US0_TX #0 US0_RX #31 US0_CLK #30	FRC_DCLK #0 FRC_DOUT #31 FRC_DFRAME #30 MODEM_DCLK #0 MODEM_DIN #31 MODEM_DOUT #30 MODEM_ANT0 #29 MODEM_ANT1 #28	CMU_CLK1 #0 PRS_CH6 #0 PRS_CH7 #10 PRS_CH8 #9 PRS_CH9 #8 ACMP0_O #0 ACMP1_O #0 LES_CH8			
PA1	BUSCY BUSDX ADC0_EXTP VDAC0_EXT	TIM0_CC0 #1 TIM0_CC1 #0 TIM0_CC2 #31 TIM0_CDTI0 #30 TIM0_CDTI1 #29 TIM0_CDTI2 #28 TIM1_CC0 #1 TIM1_CC1 #0 TIM1_CC3 #30 WTIM0_CC0 #1 LE- TIM0_OUT0 #1 LE- TIM0_OUT1 #0 PCNT0_S0IN #1 PCNT0_S1IN #0	US0_TX #1 US0_RX #0 US0_CLK #31	FRC_DCLK #1 FRC_DOUT #0 FRC_DFRAME #31 MODEM_DCLK #1 MODEM_DIN #0 MODEM_DOUT #31 MODEM_ANT0 #30 MODEM_ANT1 #29	CMU_CLK0 #0 PRS_CH6 #1 PRS_CH7 #0 PRS_CH8 #10 PRS_CH9 #9 ACMP0_O #1 ACMP1_O #1 LES_CH9			
PA2	VDAC0_OUT1ALT / OPA1_OUTALT #1 BUSDY BUSCX OPA0_P	TIM0_CC0 #2 TIM0_CC1 #1 TIM0_CC2 #0 TIM0_CDTI0 #31 TIM0_CDTI1 #30 TIM0_CDTI2 #29 TIM1_CC0 #2 TIM1_CC1 #1 TIM1_CC2 #0 TIM1_CC3 #31 WTIM0_CC0 #2 WTIM0_CC1 #0 LE- TIM0_OUT0 #2 LE- TIM0_OUT1 #1 PCNT0_S0IN #2 PCNT0_S1IN #1	US0_TX #2 US0_RX #1 US0_CLK #0 US0_CS #31 US0_CTS #30 US0_RTS #29 US1_TX #2 US1_RX #1 US1_CLK #0 US1_CS #31 US1_CTS #30 US1_RTS #29 LEU0_TX #2 LEU0_RX #1 I2C0_SDA #2 I2C0_SCL #1	FRC_DCLK #2 FRC_DOUT #1 FRC_DFRAME #0 MODEM_DCLK #2 MODEM_DIN #1 MODEM_DOUT #0 MODEM_ANTO #31 MODEM_ANT1 #30	PRS_CH6 #2 PRS_CH7 #1 PRS_CH8 #0 PRS_CH9 #10 ACMP0_O #2 ACMP1_O #2 LES_CH10			

GPIO Name		Pin Alter	nate Functionality / De	scription	
	Analog	Timers	Communication	Radio	Other
PA3	BUSCY BUSDX VDAC0_OUT0 / OPA0_OUT	TIM0_CC0 #3 TIM0_CC1 #2 TIM0_CC2 #1 TIM0_CDTI0 #0 TIM0_CDTI1 #31 TIM0_CDTI2 #30 TIM1_CC0 #3 TIM1_CC1 #2 TIM1_CC2 #1 TIM1_CC3 #0 WTIM0_CC0 #3 WTIM0_CC1 #1 LE- TIM0_OUT0 #3 LE- TIM0_OUT1 #2 PCNT0_S0IN #3 PCNT0_S1IN #2	US0_TX #3 US0_RX #2 US0_CLK #1 US0_CS #0 US0_CTS #31 US0_RTS #30 US1_TX #3 US1_RX #2 US1_CLK #1 US1_CS #0 US1_CTS #31 US1_RTS #30 LEU0_TX #3 LEU0_RX #2 I2C0_SDA #3 I2C0_SCL #2	FRC_DCLK #3 FRC_DOUT #2 FRC_DFRAME #1 MODEM_DCLK #3 MODEM_DIN #2 MODEM_DOUT #1 MODEM_ANT0 #0 MODEM_ANT1 #31	PRS_CH6 #3 PRS_CH7 #2 PRS_CH8 #1 PRS_CH9 #0 ACMP0_O #3 ACMP1_O #3 LES_CH11 GPIO_EM4WU8
PA4	VDAC0_OUT1ALT / OPA1_OUTALT #2 BUSDY BUSCX OPA0_N	TIM0_CC0 #4 TIM0_CC1 #3 TIM0_CC2 #2 TIM0_CDTI0 #1 TIM0_CDTI1 #0 TIM0_CDTI2 #31 TIM1_CC0 #4 TIM1_CC1 #3 TIM1_CC2 #2 TIM1_CC3 #1 WTIM0_CC0 #4 WTIM0_CC1 #2 WTIM0_CC2 #0 LE- TIM0_OUT0 #4 LE- TIM0_OUT1 #3 PCNT0_S0IN #4 PCNT0_S1IN #3	US0_TX #4 US0_RX #3 US0_CLK #2	FRC_DCLK #4 FRC_DOUT #3 FRC_DFRAME #2 MODEM_DCLK #4 MODEM_DIN #3 MODEM_DOUT #2 MODEM_ANTO #1 MODEM_ANT1 #0	PRS_CH6 #4 PRS_CH7 #3 PRS_CH8 #2 PRS_CH9 #1 ACMP0_O #4 ACMP1_O #4 LES_CH12
PA5	VDACO_OUT0ALT / OPAO_OUTALT #0 BUSCY BUSDX	TIM0_CC0 #5 TIM0_CC1 #4 TIM0_CC2 #3 TIM0_CDTI0 #2 TIM0_CDTI1 #1 TIM0_CDTI2 #0 TIM1_CC0 #5 TIM1_CC1 #4 TIM1_CC2 #3 TIM1_CC3 #2 WTIM0_CC0 #5 WTIM0_CC1 #3 WTIM0_CC1 #3 WTIM0_CC2 #1 LE- TIM0_OUT0 #5 LE- TIM0_OUT1 #4 PCNT0_S0IN #5 PCNT0_S1IN #4	US0_TX #5 US0_RX #4 US0_CLK #3	FRC_DCLK #5 FRC_DOUT #4 FRC_DFRAME #3 MODEM_DCLK #5 MODEM_DIN #4 MODEM_DOUT #3 MODEM_ANT0 #2 MODEM_ANT1 #1	CMU_CLKI0 #4 PRS_CH6 #5 PRS_CH7 #4 PRS_CH8 #3 PRS_CH9 #2 ACMP0_O #5 ACMP1_O #5 LES_CH13 ETM_TCLK #1

GPIO Name	Pin Alternate Functionality / Description							
	Analog Timers Communication		Radio	Other				
PB11	BUSCY BUSDX OPA2_P	TIM0_CC0 #6 TIM0_CC1 #5 TIM0_CC2 #4 TIM0_CDTI0 #3 TIM0_CDTI1 #2 TIM0_CDTI2 #1 TIM1_CC0 #6 TIM1_CC1 #5 TIM1_CC3 #3 WTIM0_CC0 #15 WTIM0_CC0 #15 WTIM0_CC1 #13 WTIM0_CC1 #13 WTIM0_CDTI0 #7 WTIM0_CDTI0 #7 WTIM0_CDTI1 #5 WTIM0_CDTI2 #3 LETIM0_OUT1 #5 PCNT0_S0IN #6 PCNT0_S1IN #5	US0_TX #6 US0_RX #5 US0_CLK #4 US0_CS #3 US0_CTS #2 US0_RTS #1 US1_TX #6 US1_RX #5 US1_CLK #4 US1_CTS #2 US1_RTS #1 LEU0_TX #6 LEU0_RX #5 I2C0_SDA #6 I2C0_SCL #5	FRC_DCLK #6 FRC_DOUT #5 FRC_DFRAME #4 MODEM_DCLK #6 MODEM_DIN #5 MODEM_DOUT #4 MODEM_ANTO #3 MODEM_ANT1 #2	PRS_CH6 #6 PRS_CH7 #5 PRS_CH8 #4 PRS_CH9 #3 ACMP0_O #6 ACMP1_O #6			
PB12	BUSDY BUSCX OPA2_OUT	TIMO_CC0 #7 TIMO_CC1 #6 TIMO_CC2 #5 TIMO_CDTI0 #4 TIMO_CDTI1 #3 TIMO_CDTI2 #2 TIM1_CC0 #7 TIM1_CC1 #6 TIM1_CC2 #5 TIM1_CC3 #4 WTIM0_CC0 #16 WTIM0_CC1 #14 WTIM0_CC1 #14 WTIM0_CDTI0 #8 WTIM0_CDTI0 #8 WTIM0_CDTI1 #6 WTIM0_CDTI1 #6 PCNT0_S0IN #7 PCNT0_S1IN #6	US0_TX #7 US0_RX #6 US0_CLK #5 US0_CS #4 US0_CTS #3 US0_RTS #2 US1_TX #7 US1_RX #6 US1_CLK #5 US1_CTS #4 US1_CTS #3 US1_RTS #2 LEU0_TX #7 LEU0_RX #6 I2C0_SDA #7 I2C0_SCL #6	FRC_DCLK #7 FRC_DOUT #6 FRC_DFRAME #5 MODEM_DCLK #7 MODEM_DIN #6 MODEM_DOUT #5 MODEM_ANTO #4 MODEM_ANT1 #3	PRS_CH6 #7 PRS_CH7 #6 PRS_CH8 #5 PRS_CH9 #4 ACMP0_O #7 ACMP1_O #7			
PB13	BUSCY BUSDX OPA2_N	TIM0_CC0 #8 TIM0_CC1 #7 TIM0_CC2 #6 TIM0_CDTI0 #5 TIM0_CDTI1 #4 TIM0_CDTI2 #3 TIM1_CC0 #8 TIM1_CC1 #7 TIM1_CC2 #6 TIM1_CC3 #5 WTIM0_CC0 #17 WTIM0_CC1 #15 WTIM0_CC1 #15 WTIM0_CDTI0 #9 WTIM0_CDTI0 #9 WTIM0_CDTI1 #7 WTIM0_CDTI1 #7 WTIM0_CDTI2 #5 LETIM0_OUT1 #7 PCNT0_S0IN #8 PCNT0_S1IN #7	US0_TX #8 US0_RX #7 US0_CLK #6 US0_CS #5 US0_CTS #4 US0_RTS #3 US1_TX #8 US1_RX #7 US1_CLK #6 US1_CTS #4 US1_RTS #3 LEU0_TX #8 LEU0_RX #7 I2C0_SDA #8 I2C0_SCL #7	0_TX #8 US0_RX #7 US0_CLK #6 US0_CS #5 US0_CTS #4 US0_RTS #3 1_TX #8 US1_RX #7 US1_CLK #6 US1_CS #5 US1_CS #5 US1_CTS #4 US1_RTS #3 LEU0_TX #8 LEU0_RX #7 I2C0_SDA #8				

GPIO Name	Pin Alternate Functionality / Description							
	Analog	Timers	Communication	Radio	Other			
PB14	BUSDY BUSCX LFXTAL_N	BUSDY BUSCX LFXTAL_N TIM1_CC3 #6 WTIM0_CC0 #18 WTIM0_CC1 #16 WTIM0_CC2 #14 WTIM0_CDTI0 #10 WTIM0_CDTI1 #8 WTIM0_CDTI2 #6 LETIM0_OUT0 #9 LETIM0_OUT1 #8 PCNT0_S0IN #9 PCNT0_S1IN #8 #8 US1_CLK #7 US1_CTS #6 US1_CTS #5 US1_CTS #6 US1_CTS #5 US1_CTS #5 US1_CTS #5 US1_CTS #5 US1_CTS #5 US1_CTS #6 US1		FRC_DCLK #9 FRC_DOUT #8 FRC_DFRAME #7 MODEM_DCLK #9 MODEM_DIN #8 MODEM_DOUT #7 MODEM_ANTO #6 MODEM_ANT1 #5	CMU_CLK1 #1 PRS_CH6 #9 PRS_CH7 #8 PRS_CH8 #7 PRS_CH9 #6 ACMP0_O #9 ACMP1_O #9			
PB15	BUSCY BUSDX LFXTAL_P	TIMO_CC0 #10 TIMO_CC1 #9 TIMO_CC2 #8 TIMO_CDTI0 #7 TIMO_CDTI1 #6 TIMO_CDTI2 #5 TIM1_CC0 #10 TIM1_CC1 #9 TIM1_CC3 #7 WTIM0_CC1 #17 WTIM0_CC1 #17 WTIM0_CC2 #15 WTIM0_CDTI0 #11 WTIM0_CDTI1 #9 WTIM0_CDTI1 #9 WTIM0_CDTI1 #9 WTIM0_CDTI1 #9 WTIM0_CDTI2 #7 LETIM0_OUT1 #9 PCNT0_S0IN #10 PCNT0_S1IN #9 US0_TX #10 US0_CS #7 US0_CTS #6 US1_TX #10 US1_RX #9 US1_CLK #8 US1_CTS #6 US1_CTS #6 US1_RTS #5 LEU0_TX #10 LEU0_RX #9 12C0_SDA #10 I2C0_SCL #9		FRC_DCLK #10 FRC_DOUT #9 FRC_DFRAME #8 MODEM_DCLK #10 MODEM_DIN #9 MODEM_DOUT #8 MODEM_ANT0 #7 MODEM_ANT1 #6	CMU_CLK0 #1 PRS_CH6 #10 PRS_CH7 #9 PRS_CH8 #8 PRS_CH9 #7 ACMP0_O #10 ACMP1_O #10			
PC6	BUSBY BUSAX	TIMO_CC0 #11 TIMO_CC1 #10 TIMO_CC2 #9 TIMO_CDTI0 #8 TIMO_CDTI1 #7 TIMO_CDTI2 #6 TIM1_CC0 #11 TIM1_CC1 #10 TIM1_CC2 #9 TIM1_CC3 #8 WTIM0_CC0 #26 WTIM0_CC1 #24 WTIM0_CC1 #24 WTIM0_CDTI0 #18 WTIM0_CDTI0 #18 WTIM0_CDTI1 #16 WTIM0_CDTI1 #16 WTIM0_CDTI2 #14 LETIM0_OUT0 #11 LETIM0_OUT1 #10 PCNT0_S0IN #11	US0_TX #11 US0_RX #10 US0_CLK #9 US0_CS #8 US0_CTS #7 US0_RTS #6 US1_TX #11 US1_RX #10 US1_CLK #9 US1_CS #8 US1_CTS #7 US1_RTS #6 LEU0_TX #11 LEU0_RX #10 I2C0_SDA #11 I2C0_SCL #10	FRC_DCLK #11 FRC_DOUT #10 FRC_DFRAME #9 MODEM_DCLK #11 MODEM_DIN #10 MODEM_DOUT #9 MODEM_ANT0 #8 MODEM_ANT1 #7	CMU_CLK0 #2 CMU_CLKI0 #2 PRS_CH0 #8 PRS_CH9 #11 PRS_CH10 #0 PRS_CH11 #5 ACMP0_O #11 ACMP1_O #11 ETM_TCLK #3			

GPIO Name	Pin Alternate Functionality / Description						
	Analog	Timers	Communication	Radio	Other		
PC7	BUSAY BUSBX	BUSAY BUSBX TIM1_CC2 #10 TIM1_CC3 #9 WTIM0_CC0 #27 WTIM0_CC1 #25 WTIM0_CC2 #23 WTIM0_CDTI0 #19 WTIM0_CDTI1 #17 WTIM0_CDTI2 #15 LETIM0_OUT0 #12 LETIM0_OUT1 #11 PCNT0_S0IN #12 PCNT0_S1IN #11 US1_TX #12 US1_CK #10 US1_CS #9 WS1_CTS #8 US1_CTS #8 US1_CTS #8 US1_CTS #8 I US1_RTS #7 LEU0_TX #12 LEU0_RX #11 I2C0_SDA #12 I2C0_SCL #11		FRC_DCLK #12 FRC_DOUT #11 FRC_DFRAME #10 MODEM_DCLK #12 MODEM_DIN #11 MODEM_DOUT #10 MODEM_ANT0 #9 MODEM_ANT1 #8	CMU_CLK1 #2 PRS_CH0 #9 PRS_CH9 #12 PRS_CH10 #1 PRS_CH11 #0 ACMP0_O #12 ACMP1_O #12 ETM_TD0		
PC8	BUSBY BUSAX	TIMO_CC0 #13 TIMO_CC1 #12 TIMO_CC2 #11 TIMO_CDTI0 #10 TIMO_CDTI1 #9 TIMO_CDTI2 #8 TIM1_CC0 #13 TIM1_CC1 #12 TIM1_CC3 #10 WTIM0_CC0 #28 WTIM0_CC1 #26 WTIM0_CC1 #26 WTIM0_CC1 #26 WTIM0_CDTI0 #20 WTIM0_CDTI0 #20 WTIM0_CDTI0 #20 WTIM0_CDTI1 #18 WTIM0_CDTI2 #16 LETIM0_OUT0 #13 LETIM0_OUT1 #12 PCNT0_S0IN #13 PCNT0_S1IN #12	US0_TX #13 US0_RX #12 US0_CLK #11 US0_CS #10 US0_CTS #9 US0_RTS #8 US1_TX #13 US1_RX #12 US1_CLK #11 US1_CS #10 US1_CTS #9 US1_RTS #8 LEU0_TX #13 LEU0_RX #12 I2C0_SDA #13 I2C0_SCL #12	FRC_DCLK #13 FRC_DOUT #12 FRC_DFRAME #11 MODEM_DCLK #13 MODEM_DIN #12 MODEM_DOUT #11 MODEM_ANT0 #10 MODEM_ANT1 #9	PRS_CH0 #10 PRS_CH9 #13 PRS_CH10 #2 PRS_CH11 #1 ACMP0_O #13 ACMP1_O #13 ETM_TD1		
PC9	BUSAY BUSBX	TIMO_CC0 #14 TIMO_CC1 #13 TIMO_CC2 #12 TIMO_CDTI0 #11 TIMO_CDTI1 #10 TIMO_CDTI2 #9 TIM1_CC0 #14 TIM1_CC1 #13 TIM1_CC2 #12 TIM1_CC3 #11 WTIM0_CC0 #29 WTIM0_CC1 #27 WTIM0_CC1 #27 WTIM0_CC1 #27 WTIM0_CDTI0 #21 WTIM0_CDTI0 #21 WTIM0_CDTI1 #19 WTIM0_CDTI1 #19 WTIM0_CDTI2 #17 LETIM0_OUT1 #13 PCNT0_S0IN #14 PCNT0_S1IN #13	US0_TX #14 US0_RX #13 US0_CLK #12 US0_CS #11 US0_CTS #10 US0_RTS #9 US1_TX #14 US1_RX #13 US1_CLK #12 US1_CS #11 US1_CTS #10 US1_RTS #9 LEU0_TX #14 LEU0_RX #13 I2C0_SDA #14 I2C0_SCL #13	FRC_DCLK #14 FRC_DOUT #13 FRC_DFRAME #12 MODEM_DCLK #14 MODEM_DIN #13 MODEM_DOUT #12 MODEM_ANT0 #11 MODEM_ANT1 #10	PRS_CH0 #11 PRS_CH9 #14 PRS_CH10 #3 PRS_CH11 #2 ACMP0_O #14 ACMP1_O #14 ETM_TD2		

GPIO Name	Pin Alternate Functionality / Description						
	Analog	Timers	Communication	Radio	Other		
PC10	BUSBY BUSAX	TIMO_CC0 #15 TIMO_CC1 #14 TIMO_CC2 #13 TIMO_CDTI0 #12 TIMO_CDTI1 #11 TIMO_CDTI2 #10 TIM1_CC0 #15 TIM1_CC1 #14 TIM1_CC2 #13 TIM1_CC3 #12 WTIM0_CC0 #30 WTIM0_CC1 #28 WTIM0_CC1 #28 WTIM0_CC1 #28 WTIM0_CDTI0 #22 WTIM0_CDTI0 #22 WTIM0_CDTI1 #20 WTIM0_CDTI1 #20 WTIM0_CDTI1 #14 PCNT0_S0IN #15 PCNT0_S1IN #14	US0_TX #15 US0_RX #14 US0_CLK #13 US0_CS #12 US0_CTS #11 US0_RTS #10 US1_TX #15 US1_RX #14 US1_CLK #13 US1_CS #12 US1_CTS #11 US1_RTS #10 LEU0_TX #15 LEU0_RX #14 I2C0_SDA #15 I2C0_SCL #14 I2C1_SDA #19 I2C1_SCL #18	FRC_DCLK #15 FRC_DOUT #14 FRC_DFRAME #13 MODEM_DCLK #15 MODEM_DIN #14 MODEM_DOUT #13 MODEM_ANT0 #12 MODEM_ANT1 #11	CMU_CLK1 #3 PRS_CH0 #12 PRS_CH9 #15 PRS_CH10 #4 PRS_CH11 #3 ACMP0_O #15 ACMP1_O #15 ETM_TD3 GPIO_EM4WU12		
PC11	BUSAY BUSBX	TIMO_CC0 #16 TIMO_CC1 #15 TIMO_CC2 #14 TIMO_CDTI0 #13 TIMO_CDTI1 #12 TIMO_CDTI2 #11 TIM1_CC0 #16 TIM1_CC1 #15 TIM1_CC3 #13 WTIM0_CC0 #31 WTIM0_CC1 #29 WTIM0_CC1 #29 WTIM0_CC1 #29 WTIM0_CDTI0 #23 WTIM0_CDTI0 #23 WTIM0_CDTI1 #21 WTIM0_CDTI1 #21 WTIM0_CDTI2 #19 LETIM0_OUT1 #15 PCNT0_S0IN #16 PCNT0_S1IN #15	US0_TX #16 US0_RX #15 US0_CLK #14 US0_CS #13 US0_CTS #12 US0_RTS #11 US1_TX #16 US1_RX #15 US1_CLK #14 US1_CS #13 US1_CTS #12 US1_RTS #11 LEU0_TX #16 LEU0_RX #15 I2C0_SDA #16 I2C0_SCL #15 I2C1_SDA #20 I2C1_SCL #19	FRC_DCLK #16 FRC_DOUT #15 FRC_DFRAME #14 MODEM_DCLK #16 MODEM_DIN #15 MODEM_DOUT #14 MODEM_ANT0 #13 MODEM_ANT1 #12	CMU_CLK0 #3 PRS_CH0 #13 PRS_CH9 #16 PRS_CH10 #5 PRS_CH11 #4 ACMP0_O #16 ACMP1_O #16 DBG_SWO #3		
PD9	BUSCY BUSDX	TIMO_CC0 #17 TIMO_CC1 #16 TIMO_CC2 #15 TIMO_CDTI0 #14 TIMO_CDTI1 #13 TIMO_CDTI2 #12 TIM1_CC0 #17 TIM1_CC1 #16 TIM1_CC2 #15 TIM1_CC3 #14 WTIM0_CC1 #31 WTIM0_CC1 #31 WTIM0_CC1 #31 WTIM0_CDTI0 #25 WTIM0_CDTI0 #25 WTIM0_CDTI1 #23 WTIMO_CDTI1 #23 WTIMO_CDTI2 #21 LETIMO_OUT0 #17 LETIMO_OUT1 #16 PCNT0_S0IN #17 PCNT0_S1IN #16	US0_TX #17 US0_RX #16 US0_CLK #15 US0_CS #14 US0_CTS #13 US0_RTS #12 US1_TX #17 US1_RX #16 US1_CLK #15 US1_CS #14 US1_CTS #13 US1_RTS #12 LEU0_TX #17 LEU0_RX #16 I2C0_SDA #17 I2C0_SCL #16	FRC_DCLK #17 FRC_DOUT #16 FRC_DFRAME #15 MODEM_DCLK #17 MODEM_DIN #16 MODEM_DOUT #15 MODEM_ANT0 #14 MODEM_ANT1 #13	CMU_CLK0 #4 PRS_CH3 #8 PRS_CH4 #0 PRS_CH5 #6 PRS_CH6 #11 ACMP0_O #17 ACMP1_O #17 LES_CH1		

GPIO Name	Pin Alternate Functionality / Description							
	Analog	Timers	Communication	Radio	Other			
PD10	BUSDY BUSCX	TIMO_CCO #18 TIMO_CC1 #17 TIMO_CC2 #16 TIMO_CDTI0 #15 TIMO_CDTI1 #14 TIMO_CDTI2 #13 TIM1_CC0 #18 TIM1_CC0 #18 TIM1_CC1 #17 TIM1_CC2 #16 TIM1_CC2 #16 TIM1_CC2 #30 WTIM0_CDTI0 #26 WTIM0_CDTI0 #26 WTIM0_CDTI1 #24 WTIM0_CDTI1 #24 WTIM0_CDTI1 #24 WTIM0_CDTI1 #24 WTIM0_CDTI1 #24 WTIM0_CDTI2 #22 LETIM0_OUT0 #18 LETIM0_OUT1 #17 PCNT0_S0IN #18 PCNT0_S1IN #17		FRC_DCLK #18 FRC_DOUT #17 FRC_DFRAME #16 MODEM_DCLK #18 MODEM_DIN #17 MODEM_DOUT #16 MODEM_ANT0 #15 MODEM_ANT1 #14	CMU_CLK1 #4 PRS_CH3 #9 PRS_CH4 #1 PRS_CH5 #0 PRS_CH6 #12 ACMP0_O #18 ACMP1_O #18 LES_CH2			
PD11	BUSCY BUSDX	TIMO_CC0 #19 TIMO_CC1 #18 TIMO_CC2 #17 TIMO_CDTI0 #16 TIMO_CDTI1 #15 TIMO_CDTI2 #14 TIM1_CC0 #19 TIM1_CC1 #18 TIM1_CC2 #17 TIM1_CC3 #16 WTIM0_CDTI0 #27 WTIM0_CDTI1 #25 WTIMO_CDTI2 #23 LETIMO_OUT0 #19 LETIMO_OUT1 #18 PCNT0_S0IN #19 PCNT0_S1IN #18	US0_TX #19 US0_RX #18 US0_CLK #17 US0_CS #16 US0_CTS #15 US0_RTS #14 US1_TX #19 US1_RX #18 US1_CLK #17 US1_CS #16 US1_CTS #15 US1_RTS #14 LEU0_TX #19 LEU0_RX #18 I2C0_SDA #19 I2C0_SCL #18	FRC_DCLK #19 FRC_DOUT #18 FRC_DFRAME #17 MODEM_DCLK #19 MODEM_DIN #18 MODEM_DOUT #17 MODEM_ANT0 #16 MODEM_ANT1 #15	PRS_CH3 #10 PRS_CH4 #2 PRS_CH5 #1 PRS_CH6 #13 ACMP0_O #19 ACMP1_O #19 LES_CH3			
PD12	VDAC0_OUT1ALT / OPA1_OUTALT #0 BUSDY BUSCX	TIMO_CC0 #20 TIMO_CC1 #19 TIMO_CC2 #18 TIMO_CDTI0 #17 TIMO_CDTI1 #16 TIMO_CDTI2 #15 TIM1_CC0 #20 TIM1_CC1 #19 TIM1_CC2 #18 TIM1_CC3 #17 WTIM0_CDTI0 #28 WTIM0_CDTI1 #26 WTIM0_CDTI2 #24 LETIMO_OUT0 #20 LETIMO_OUT1 #19 PCNT0_S0IN #20 PCNT0_S1IN #19	US0_TX #20 US0_RX #19 US0_CLK #18 US0_CS #17 US0_CTS #16 US0_RTS #15 US1_TX #20 US1_RX #19 US1_CLK #18 US1_CS #17 US1_CTS #16 US1_RTS #15 LEU0_TX #20 LEU0_RX #19 I2C0_SDA #20 I2C0_SCL #19	FRC_DCLK #20 FRC_DOUT #19 FRC_DFRAME #18 MODEM_DCLK #20 MODEM_DIN #19 MODEM_DOUT #18 MODEM_ANT0 #17 MODEM_ANT1 #16	PRS_CH3 #11 PRS_CH4 #3 PRS_CH5 #2 PRS_CH6 #14 ACMP0_O #20 ACMP1_O #20 LES_CH4			

GPIO Name	Pin Alternate Functionality / Description							
	Analog	Timers	Communication	Radio	Other			
PD13	VDAC0_OUT0ALT / OPA0_OUTALT #1 BUSCY BUSDX OPA1_P	TIM0_CC0 #21 TIM0_CC1 #20 TIM0_CC2 #19 TIM0_CDTI0 #18 TIM0_CDTI1 #17 TIM0_CDTI2 #16 TIM1_CC0 #21 TIM1_CC1 #20 TIM1_CC2 #19 TIM1_CC3 #18 WTIM0_CDTI0 #29 WTIM0_CDTI1 #27 WTIM0_CDTI2 #25 LETIM0_OUT0 #21 LETIM0_OUT1 #20 PCNT0_S0IN #21 PCNT0_S1IN #20	US0_TX #21 US0_RX #20 US0_CLK #19 US0_CS #18 US0_CTS #17 US0_RTS #16 US1_TX #21 US1_RX #20 US1_CLK #19 US1_CS #18 US1_CTS #17 US1_RTS #16 LEU0_TX #21 LEU0_RX #20 I2C0_SDA #21 I2C0_SCL #20	FRC_DCLK #21 FRC_DOUT #20 FRC_DFRAME #19 MODEM_DCLK #21 MODEM_DIN #20 MODEM_DOUT #19 MODEM_ANT0 #18 MODEM_ANT1 #17	PRS_CH3 #12 PRS_CH4 #4 PRS_CH5 #3 PRS_CH6 #15 ACMP0_O #21 ACMP1_O #21 LES_CH5			
PD14	BUSDY BUSCX VDAC0_OUT1 / OPA1_OUT	TIM0_CC0 #22 TIM0_CC1 #21 TIM0_CC2 #20 TIM0_CDTI0 #19 TIM0_CDTI1 #18 TIM0_CDTI2 #17 TIM1_CC0 #22 TIM1_CC1 #21 TIM1_CC2 #20 TIM1_CC3 #19 WTIM0_CDTI0 #30 WTIM0_CDTI1 #28 WTIM0_CDTI2 #26 LETIM0_OUT0 #22 LETIM0_OUT1 #21 PCNT0_S0IN #22 PCNT0_S1IN #21	US0_TX #22 US0_RX #21 US0_CLK #20 US0_CS #19 US0_CTS #18 US0_RTS #17 US1_TX #22 US1_RX #21 US1_CLK #20 US1_CS #19 US1_CTS #18 US1_RTS #17 LEU0_TX #22 LEU0_RX #21 I2C0_SDA #22 I2C0_SCL #21	FRC_DCLK #22 FRC_DOUT #21 FRC_DFRAME #20 MODEM_DCLK #22 MODEM_DIN #21 MODEM_DOUT #20 MODEM_ANT0 #19 MODEM_ANT1 #18	CMU_CLK0 #5 PRS_CH3 #13 PRS_CH4 #5 PRS_CH5 #4 PRS_CH6 #16 ACMP0_O #22 ACMP1_O #22 LES_CH6 GPIO_EM4WU4			
PD15	VDACO_OUT0ALT / OPA0_OUTALT #2 BUSCY BUSDX OPA1_N	TIM0_CC0 #23 TIM0_CC1 #22 TIM0_CC2 #21 TIM0_CDTI0 #20 TIM0_CDTI1 #19 TIM0_CDTI2 #18 TIM1_CC0 #23 TIM1_CC1 #22 TIM1_CC2 #21 TIM1_CC3 #20 WTIM0_CDTI0 #31 WTIM0_CDTI1 #29 WTIM0_CDTI1 #29 WTIM0_CDTI2 #27 LETIM0_OUT0 #23 LETIM0_OUT1 #22 PCNT0_S0IN #23 PCNT0_S1IN #22	US0_TX #23 US0_RX #22 US0_CLK #21 US0_CS #20 US0_CTS #19 US0_RTS #18 US1_TX #23 US1_RX #22 US1_CLK #21 US1_CS #20 US1_CTS #19 US1_RTS #18 LEU0_TX #23 LEU0_RX #22 I2C0_SDA #23 I2C0_SCL #22	FRC_DCLK #23 FRC_DOUT #22 FRC_DFRAME #21 MODEM_DCLK #23 MODEM_DIN #22 MODEM_DOUT #21 MODEM_ANT0 #20 MODEM_ANT1 #19	CMU_CLK1 #5 PRS_CH3 #14 PRS_CH4 #6 PRS_CH5 #5 PRS_CH6 #17 ACMP0_O #23 ACMP1_O #23 LES_CH7 DBG_SWO #2			

GPIO Name	Pin Alternate Functionality / Description							
	Analog	Timers	Communication	Radio	Other			
PF0	BUSBY BUSAX	TIM0_CC0 #24 TIM0_CC1 #23 TIM0_CC2 #22 TIM0_CDTI0 #21 TIM0_CDTI1 #20 TIM0_CDTI2 #19 TIM1_CC0 #24 TIM1_CC1 #23 TIM1_CC2 #22 TIM1_CC3 #21 WTIM0_CDTI1 #30 WTIM0_CDTI2 #28 LETIM0_OUT0 #24 LETIM0_OUT1 #23 PCNT0_S0IN #24 PCNT0_S1IN #23	US0_TX #24 US0_RX #23 US0_CLK #22 US0_CS #21 US0_CTS #20 US0_RTS #19 US1_TX #24 US1_RX #23 US1_CLK #22 US1_CS #21 US1_CTS #20 US1_RTS #19 US2_TX #14 US2_RX #13 US2_CLK #12 US2_CS #11 US2_CTS #10 US2_RTS #9 LEU0_TX #24 LEU0_RX #23 I2C0_SDA #24 I2C0_SCL #23	FRC_DCLK #24 FRC_DOUT #23 FRC_DFRAME #22 MODEM_DCLK #24 MODEM_DIN #23 MODEM_DOUT #22 MODEM_ANT0 #21 MODEM_ANT1 #20	PRS_CH0 #0 PRS_CH1 #7 PRS_CH2 #6 PRS_CH3 #5 ACMP0_O #24 ACMP1_O #24 DBG_SWCLKTCK BOOT_TX			
PF1	BUSAY BUSBX	TIMO_CC0 #25 TIMO_CC1 #24 TIMO_CC2 #23 TIMO_CDTI0 #22 TIMO_CDTI1 #21 TIMO_CDTI2 #20 TIM1_CC0 #25 TIM1_CC1 #24 TIM1_CC2 #23 TIM1_CC3 #22 WTIM0_CDTI1 #31 WTIM0_CDTI2 #29 LETIMO_OUT0 #25 LETIMO_OUT1 #24 PCNT0_S0IN #25 PCNT0_S1IN #24	US0_TX #25 US0_RX #24 US0_CLK #23 US0_CS #22 US0_CTS #21 US0_RTS #20 US1_TX #25 US1_RX #24 US1_CLK #23 US1_CS #22 US1_CTS #21 US1_RTS #20 US2_TX #15 US2_RX #14 US2_CLK #13 US2_CLK #13 US2_CS #12 US2_CTS #11 US2_RTS #10 LEU0_TX #25 LEU0_RX #24 I2C0_SDA #25 I2C0_SCL #24	FRC_DCLK #25 FRC_DOUT #24 FRC_DFRAME #23 MODEM_DCLK #25 MODEM_DIN #24 MODEM_DOUT #23 MODEM_ANT0 #22 MODEM_ANT1 #21	PRS_CH0 #1 PRS_CH1 #0 PRS_CH2 #7 PRS_CH3 #6 ACMP0_O #25 ACMP1_O #25 DBG_SWDIOTMS BOOT_RX			
PF2	BUSBY BUSAX	TIMO_CC0 #26 TIMO_CC1 #25 TIMO_CC2 #24 TIMO_CDTI0 #23 TIMO_CDTI1 #22 TIMO_CDTI2 #21 TIM1_CC0 #26 TIM1_CC1 #25 TIM1_CC2 #24 TIM1_CC3 #23 WTIM0_CDTI2 #30 LETIM0_OUT0 #26 LETIM0_OUT1 #25 PCNT0_S0IN #26 PCNT0_S1IN #25	US0_TX #26 US0_RX #25 US0_CLK #24 US0_CS #23 US0_CTS #22 US0_RTS #21 US1_TX #26 US1_RX #25 US1_CLK #24 US1_CS #23 US1_CTS #22 US1_RTS #21 LEU0_TX #26 LEU0_RX #25 I2C0_SDA #26 I2C0_SCL #25	FRC_DCLK #26 FRC_DOUT #25 FRC_DFRAME #24 MODEM_DCLK #26 MODEM_DIN #25 MODEM_DOUT #24 MODEM_ANT0 #23 MODEM_ANT1 #22	CMU_CLK0 #6 PRS_CH0 #2 PRS_CH1 #1 PRS_CH2 #0 PRS_CH3 #7 ACMP0_O #26 ACMP1_O #26 DBG_TDO DBG_SWO #0 GPIO_EM4WU0			

GPIO Name	Pin Alternate Functionality / Description						
	Analog	Timers	Communication	Radio	Other		
PF3	BUSAY BUSBX	TIMO_CC0 #27 TIMO_CC1 #26 TIMO_CC2 #25 TIMO_CDTI0 #24 TIMO_CDTI1 #23 TIMO_CDTI2 #22 TIM1_CC0 #27 TIM1_CC1 #26 TIM1_CC2 #25 TIM1_CC2 #25 TIM1_CC3 #24 WTIM0_CDTI2 #31 LETIM0_OUT0 #27 LETIM0_OUT1 #26 PCNT0_S0IN #27 PCNT0_S1IN #26 US0_CX #27 US0_CS #24 US0_CTS #23 US0_RTS #22 US1_TX #27 US1_RX #26 US1_CLK #25 US1_CS #24 US1_CTS #23 US1_RTS #22 US2_TX #16 US2_RX #15 US2_CLK #14 US2_CS #13 US2_CTS #12 US2_RTS #11 LEU0_TX #27 LEU0_RX #26 I2C0_SDA #27 I2C0_SCL #26		FRC_DCLK #27 FRC_DOUT #26 FRC_DFRAME #25 MODEM_DCLK #27 MODEM_DIN #26 MODEM_DOUT #25 MODEM_ANT0 #24 MODEM_ANT1 #23	CMU_CLK1 #6 PRS_CH0 #3 PRS_CH1 #2 PRS_CH2 #1 PRS_CH3 #0 ACMP0_O #27 ACMP1_O #27 DBG_TDI		
PF4	BUSBY BUSAX	TIM0_CC0 #28 TIM0_CC1 #27 TIM0_CC2 #26 TIM0_CDTI0 #25 TIM0_CDTI1 #24 TIM0_CDTI2 #23 TIM1_CC0 #28 TIM1_CC1 #27 TIM1_CC2 #26 TIM1_CC3 #25 LE- TIM0_OUT0 #28 LE- TIM0_OUT0 #28 LE- TIM0_OUT1 #27 PCNT0_S0IN #28 PCNT0_S1IN #27	US0_TX #28 US0_RX #27 US0_CLK #26 US0_CS #25 US0_CTS #24 US0_RTS #23 US1_TX #28 US1_RX #27 US1_CLK #26 US1_CS #25 US1_CTS #24 US1_RTS #23 US2_TX #17 US2_RX #16 US2_CLK #15 US2_CLK #15 US2_CS #14 US2_CTS #13 US2_RTS #12 LEU0_TX #28 LEU0_RX #27 I2C0_SDA #28 I2C0_SCL #27	FRC_DCLK #28 FRC_DOUT #27 FRC_DFRAME #26 MODEM_DCLK #28 MODEM_DIN #27 MODEM_DOUT #26 MODEM_ANT0 #25 MODEM_ANT1 #24	PRS_CH0 #4 PRS_CH1 #3 PRS_CH2 #2 PRS_CH3 #1 ACMP0_O #28 ACMP1_O #28		

GPIO Name		Pin Alternate Functionality / Description						
	Analog	Timers	Communication	Radio	Other			
PF5	BUSAY BUSBX	TIM0_CC0 #29 TIM0_CC1 #28 TIM0_CC2 #27 TIM0_CDTI0 #26 TIM0_CDTI1 #25 TIM0_CDTI2 #24 TIM1_CC0 #29 TIM1_CC1 #28 TIM1_CC2 #27 TIM1_CC3 #26 LE- TIM0_OUT0 #29 LE- TIM0_OUT1 #28 PCNT0_S0IN #29 PCNT0_S1IN #28	US0_TX #29 US0_RX #28 US0_CLK #27 US0_CS #26 US0_CTS #25 US0_RTS #24 US1_TX #29 US1_RX #28 US1_CLK #27 US1_CS #26 US1_CTS #25 US1_RTS #24 US2_TX #18 US2_RX #17 US2_CLK #16 US2_CS #15 US2_CTS #14 US2_RTS #13 LEU0_TX #29 LEU0_RX #28 I2C0_SDA #29 I2C0_SCL #28	FRC_DCLK #29 FRC_DOUT #28 FRC_DFRAME #27 MODEM_DCLK #29 MODEM_DIN #28 MODEM_DOUT #27 MODEM_ANT0 #26 MODEM_ANT1 #25	PRS_CH0 #5 PRS_CH1 #4 PRS_CH2 #3 PRS_CH3 #2 ACMP0_O #29 ACMP1_O #29			
PF6	BUSBY BUSAX	TIM0_CC0 #30 TIM0_CC1 #29 TIM0_CC2 #28 TIM0_CDTI0 #27 TIM0_CDTI1 #26 TIM0_CDTI2 #25 TIM1_CC0 #30 TIM1_CC1 #29 TIM1_CC2 #28 TIM1_CC3 #27 LE- TIM0_OUT0 #30 LE- TIM0_OUT1 #29 PCNT0_S0IN #30 PCNT0_S1IN #29	US0_TX #30 US0_RX #29 US0_CLK #28 US0_CS #27 US0_CTS #26 US0_RTS #25 US1_TX #30 US1_RX #29 US1_CLK #28 US1_CS #27 US1_CTS #26 US1_RTS #25 US2_TX #19 US2_RX #18 US2_CLK #17 US2_CS #16 US2_CTS #15 US2_CTS #15 US2_RTS #14 LEU0_TX #30 LEU0_RX #29 I2C0_SDA #30 I2C0_SCL #29	FRC_DCLK #30 FRC_DOUT #29 FRC_DFRAME #28 MODEM_DCLK #30 MODEM_DIN #29 MODEM_DOUT #28 MODEM_ANT0 #27 MODEM_ANT1 #26	CMU_CLK1 #7 PRS_CH0 #6 PRS_CH1 #5 PRS_CH2 #4 PRS_CH3 #3 ACMP0_O #30 ACMP1_O #30			

GPIO Name	Pin Alternate Functionality / Description							
	Analog	Timers	Communication	Radio	Other			
PF7	BUSAY BUSBX	TIM0_CC0 #31 TIM0_CC1 #30 TIM0_CC2 #29 TIM0_CDTI0 #28 TIM0_CDTI1 #27 TIM0_CDTI2 #26 TIM1_CC0 #31 TIM1_CC1 #30 TIM1_CC2 #29 TIM1_CC3 #28 LE- TIM0_OUT0 #31 LE- TIM0_OUT0 #31 PCNT0_S0IN #31 PCNT0_S1IN #30	US0_TX #31 US0_RX #30 US0_CLK #29 US0_CS #28 US0_CTS #27 US0_RTS #26 US1_TX #31 US1_RX #30 US1_CLK #29 US1_CS #28 US1_CTS #27 US1_RTS #26 US2_TX #20 US2_RX #19 US2_CLK #18 US2_CS #17 US2_CTS #16 US2_RTS #15 LEU0_TX #31 LEU0_RX #30 I2C0_SDA #31 I2C0_SCL #30	FRC_DCLK #31 FRC_DOUT #30 FRC_DFRAME #29 MODEM_DCLK #31 MODEM_DIN #30 MODEM_DOUT #29 MODEM_ANT0 #28 MODEM_ANT1 #27	CMU_CLKI0 #1 CMU_CLK0 #7 PRS_CH0 #7 PRS_CH1 #6 PRS_CH2 #5 PRS_CH3 #4 ACMP0_O #31 ACMP1_O #31 GPIO_EM4WU1			

6.7 Alternate Functionality Overview

A wide selection of alternate functionality is available for multiplexing to various pins. The following table shows the name of the alternate functionality in the first column, followed by columns showing the possible LOCATION bitfield settings and the associated GPIO pin. Refer to 6.6 GPIO Functionality Table for a list of functions available on each GPIO pin.

Note: Some functionality, such as analog interfaces, do not have alternate settings or a LOCATION bitfield. In these cases, the pinout is shown in the column corresponding to LOCATION 0.

Table 6.7. Alternate Functionality Overview

Alternate	LOCATION								
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
101100	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Analog comparator
ACMP0_O	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	ACMP0, digital output.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
A CAMPA C	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Analog comparator
ACMP1_O	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	ACMP1, digital output.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
ADC0_EXTN	0: PA0								Analog to digital converter ADC0 external reference input negative pin.
ADC0_EXTP	0: PA1								Analog to digital converter ADC0 external reference input positive pin.
BOOT_RX	0: PF1								Bootloader RX.
BOOT_TX	0: PF0								Bootloader TX.
	0: PA1	4: PD9							
CMIL CLKO	1: PB15	5: PD14							Clock Management
CMU_CLK0	2: PC6	6: PF2							Unit, clock output number 0.
	3: PC11	7: PF7							
	0: PA0	4: PD10							
OMILL OLIKA	1: PB14	5: PD15							Clock Management
CMU_CLK1	2: PC7	6: PF3							Unit, clock output number 1.
	3: PC10	7: PF6							
	0: PB13	4: PA5							Clock Management
CMU_CLKI0	1: PF7								Unit, clock input
	2: PC6								number 0.

Alternate				LOCA	ATION				
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PF0								Debug-interface Serial Wire clock input and JTAG Test Clock.
DBG_SWCLKTCK									Note that this function is enabled to the pin out of reset, and has a built-in pull down.
DBC SWDIOTMS	0: PF1								Debug-interface Serial Wire data in- put / output and JTAG Test Mode Select.
DBG_SWDIOTMS									Note that this function is enabled to the pin out of reset, and has a built-in pull up.
	0: PF2 1: PB13								Debug-interface Serial Wire viewer Output.
DBG_SWO	2: PD15 3: PC11								Note that this function is not enabled after reset, and must be enabled by software to be used.
DBG_TDI	0: PF3								Debug-interface JTAG Test Data In. Note that this function becomes available after the first valid JTAG command is received, and has a built-in pull up when JTAG is active.
DBG_TDO	0: PF2								Debug-interface JTAG Test Data Out. Note that this func- tion becomes avail- able after the first valid JTAG com-
ETM_TCLK	1: PA5 3: PC6								mand is received. Embedded Trace Module ETM clock .
ETM_TD0	3: PC7								Embedded Trace Module ETM data 0.

Alternate									
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
ETM_TD1	3: PC8								Embedded Trace Module ETM data 1.
ETM_TD2	3: PC9								Embedded Trace Module ETM data 2.
ETM_TD3	3: PC10								Embedded Trace Module ETM data 3.
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
EDC DCLK	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Frame Controller,
FRC_DCLK	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	Data Sniffer Clock.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
	0: PA2	4: PB11	8: PB15	12: PC9	16: PD10	20: PD14	24: PF2	28: PF6	
EDC DEBAME	1: PA3	5: PB12	9: PC6	13: PC10	17: PD11	21: PD15	25: PF3	29: PF7	Frame Controller, Data Sniffer Frame
FRC_DFRAME	2: PA4	6: PB13	10: PC7	14: PC11	18: PD12	22: PF0	26: PF4	30: PA0	active
	3: PA5	7: PB14	11: PC8	15: PD9	19: PD13	23: PF1	27: PF5	31: PA1	
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	
EDC DOUT	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	Frame Controller, Data Sniffer Out-
FRC_DOUT	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	put.
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	
GPIO_EM4WU0	0: PF2								Pin can be used to wake the system up from EM4
GPIO_EM4WU1	0: PF7								Pin can be used to wake the system up from EM4
GPIO_EM4WU4	0: PD14								Pin can be used to wake the system up from EM4
GPIO_EM4WU8	0: PA3								Pin can be used to wake the system up from EM4
GPIO_EM4WU9	0: PB13								Pin can be used to wake the system up from EM4
GPIO_EM4WU12	0: PC10								Pin can be used to wake the system up from EM4
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	
1300 801	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	I2C0 Serial Clock
I2C0_SCL	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	Line input / output.
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	

Alternate	LOCATION								
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
1000 004	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	I2C0 Serial Data in-
I2C0_SDA	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	put / output.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
1204 001					18: PC10				I2C1 Serial Clock
I2C1_SCL					19: PC11				Line input / output.
I2C1_SDA					19: PC10	20: PC11			I2C1 Serial Data input / output.
LES_CH1	0: PD9								LESENSE channel 1.
LES_CH2	0: PD10								LESENSE channel 2.
LES_CH3	0: PD11								LESENSE channel 3.
LES_CH4	0: PD12								LESENSE channel 4.
LES_CH5	0: PD13								LESENSE channel 5.
LES_CH6	0: PD14								LESENSE channel 6.
LES_CH7	0: PD15								LESENSE channel 7.
LES_CH8	0: PA0								LESENSE channel 8.
LES_CH9	0: PA1								LESENSE channel 9.
LES_CH10	0: PA2								LESENSE channel 10.
LES_CH11	0: PA3								LESENSE channel 11.
LES_CH12	0: PA4								LESENSE channel 12.
LES_CH13	0: PA5								LESENSE channel 13.
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
LETIMO_OUT0	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Low Energy Timer LETIM0, output
LETIMO_GGTO	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	channel 0.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	
LETIM0_OUT1	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	Low Energy Timer LETIM0, output
	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	channel 1.
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	

Alternate				LOCA	ATION				
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	
	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	LEUART0 Receive
LEU0_RX	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	input.
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	LEUART0 Transmit
1 = 110 = TV	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	output. Also used
LEU0_TX	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	as receive input in half duplex commu-
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	nication.
LFXTAL_N	0: PB14								Low Frequency Crystal (typically 32.768 kHz) nega- tive pin. Also used as an optional ex- ternal clock input pin.
LFXTAL_P	0: PB15								Low Frequency Crystal (typically 32.768 kHz) posi- tive pin.
	0: PA3	4: PB12	8: PC6	12: PC10	16: PD11	20: PD15	24: PF3	28: PF7	
MODEM ANTO	1: PA4	5: PB13	9: PC7	13: PC11	17: PD12	21: PF0	25: PF4	29: PA0	MODEM antenna control output 0,
WODEW_ANTO	2: PA5	6: PB14	10: PC8	14: PD9	18: PD13	22: PF1	26: PF5	30: PA1	used for antenna diversity.
	3: PB11	7: PB15	11: PC9	15: PD10	19: PD14	23: PF2	27: PF6	31: PA2	
	0: PA4	4: PB13	8: PC7	12: PC11	16: PD12	20: PF0	24: PF4	28: PA0	MODEM
MODEM ANT1	1: PA5	5: PB14	9: PC8	13: PD9	17: PD13	21: PF1	25: PF5	29: PA1	MODEM antenna control output 1,
MODEM_7 ((V)	2: PB11	6: PB15	10: PC9	14: PD10	18: PD14	22: PF2	26: PF6	30: PA2	used for antenna diversity.
	3: PB12	7: PC6	11: PC10	15: PD11	19: PD15	23: PF3	27: PF7	31: PA3	,
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
MODEM_DCLK	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	MODEM data clock
<u>_</u>	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	out.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	
MODEM DIN	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	MODEM data in.
<u>-</u>	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	
	0: PA2	4: PB11	8: PB15	12: PC9	16: PD10	20: PD14	24: PF2	28: PF6	
MODEM DOUT	1: PA3	5: PB12	9: PC6	13: PC10	17: PD11	21: PD15	25: PF3	29: PF7	MODEM data out.
MODEM_DOUT 2	2: PA4	6: PB13	10: PC7	14: PC11	18: PD12	22: PF0	26: PF4	30: PA0	INIODEINI dala dul.
	3: PA5	7: PB14	11: PC8	15: PD9	19: PD13	23: PF1	27: PF5	31: PA1	

Alternate	LOCATION								
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
OPA0_N	0: PA4								Operational Amplifier 0 external negative input.
OPA0_P	0: PA2								Operational Amplifier 0 external positive input.
OPA1_N	0: PD15								Operational Amplifier 1 external negative input.
OPA1_P	0: PD13								Operational Amplifier 1 external positive input.
OPA2_N	0: PB13								Operational Amplifier 2 external negative input.
OPA2_OUT	0: PB12								Operational Amplifier 2 output.
OPA2_P	0: PB11								Operational Amplifier 2 external positive input.
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	
DONTO COIN	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Pulse Counter PCNT0 input num-
PCNT0_S0IN	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	ber 0.
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	
PCNT0 S1IN	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	Pulse Counter PCNT0 input num-
01410_01114	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	ber 1.
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	
	0: PF0	4: PF4	8: PC6	12: PC10					
PRS_CH0	1: PF1	5: PF5	9: PC7	13: PC11					Peripheral Reflex System PRS, chan-
1110_0110	2: PF2	6: PF6	10: PC8						nel 0.
	3: PF3	7: PF7	11: PC9						
	0: PF1	4: PF5							
PRS_CH1	1: PF2	5: PF6							Peripheral Reflex System PRS, chan-
FRS_CITI	2: PF3	6: PF7							nel 1.
	3: PF4	7: PF0							
	0: PF2	4: PF6							
DDS CH3	1: PF3	5: PF7							Peripheral Reflex System PRS, chan-
PRS_CH2	2: PF4	6: PF0							nel 2.
	3: PF5	7: PF1							

Alternate	Alternate LOCATION									
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description	
	0: PF3	4: PF7	8: PD9	12: PD13						
DDC CH2	1: PF4	5: PF0	9: PD10	13: PD14					Peripheral Reflex	
PRS_CH3	2: PF5	6: PF1	10: PD11	14: PD15					System PRS, channel 3.	
	3: PF6	7: PF2	11: PD12							
	0: PD9	4: PD13								
PRS_CH4	1: PD10	5: PD14							Peripheral Reflex System PRS, chan-	
PR3_CH4	2: PD11	6: PD15							nel 4.	
	3: PD12									
	0: PD10	4: PD14								
PRS_CH5	1: PD11	5: PD15							Peripheral Reflex System PRS, chan-	
PRS_CHS	2: PD12	6: PD9							nel 5.	
	3: PD13									
	0: PA0	4: PA4	8: PB13	12: PD10	16: PD14					
PRS_CH6	1: PA1	5: PA5	9: PB14	13: PD11	17: PD15				Peripheral Reflex System PRS, chan-	
FK3_CH0	2: PA2	6: PB11	10: PB15	14: PD12					nel 6.	
	3: PA3	7: PB12	11: PD9	15: PD13						
	0: PA1	4: PA5	8: PB14							
PRS_CH7	1: PA2	5: PB11	9: PB15						Peripheral Reflex System PRS, chan-	
1113_011	2: PA3	6: PB12	10: PA0						nel 7.	
	3: PA4	7: PB13								
	0: PA2	4: PB11	8: PB15							
PRS CH8	1: PA3	5: PB12	9: PA0						Peripheral Reflex System PRS, chan-	
1110_0110	2: PA4	6: PB13	10: PA1						nel 8.	
	3: PA5	7: PB14								
	0: PA3	4: PB12	8: PA0	12: PC7	16: PC11					
PRS_CH9	1: PA4	5: PB13	9: PA1	13: PC8					Peripheral Reflex System PRS, chan-	
1110_0110	2: PA5	6: PB14	10: PA2	14: PC9					nel 9.	
	3: PB11	7: PB15	11: PC6	15: PC10						
	0: PC6	4: PC10								
PRS_CH10	1: PC7	5: PC11							Peripheral Reflex System PRS, chan-	
1110_01110	2: PC8								nel 10.	
	3: PC9									
	0: PC7	4: PC11								
PRS CH11	1: PC8	5: PC6							Peripheral Reflex System PRS, chan-	
PRS CH11	2: PC9								nel 11.	
	3: PC10									

Alternate	Alternate LOCATION									
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description	
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4		
TIMO CCO	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Timer 0 Capture	
TIM0_CC0	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	Compare input / output channel 0.	
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7		
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5		
TIMO CC1	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	Timer 0 Capture Compare input /	
TIM0_CC1	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	output channel 1.	
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0		
	0: PA2	4: PB11	8: PB15	12: PC9	16: PD10	20: PD14	24: PF2	28: PF6		
TIM0_CC2	1: PA3	5: PB12	9: PC6	13: PC10	17: PD11	21: PD15	25: PF3	29: PF7	Timer 0 Capture Compare input /	
TIMO_CC2	2: PA4	6: PB13	10: PC7	14: PC11	18: PD12	22: PF0	26: PF4	30: PA0	output channel 2.	
	3: PA5	7: PB14	11: PC8	15: PD9	19: PD13	23: PF1	27: PF5	31: PA1		
	0: PA3	4: PB12	8: PC6	12: PC10	16: PD11	20: PD15	24: PF3	28: PF7		
TIM0_CDTI0	1: PA4	5: PB13	9: PC7	13: PC11	17: PD12	21: PF0	25: PF4	29: PA0	Timer 0 Compli- mentary Dead Time	
TIMO_CDTIO	2: PA5	6: PB14	10: PC8	14: PD9	18: PD13	22: PF1	26: PF5	30: PA1	Insertion channel 0.	
	3: PB11	7: PB15	11: PC9	15: PD10	19: PD14	23: PF2	27: PF6	31: PA2		
	0: PA4	4: PB13	8: PC7	12: PC11	16: PD12	20: PF0	24: PF4	28: PA0		
TIM0_CDTI1	1: PA5	5: PB14	9: PC8	13: PD9	17: PD13	21: PF1	25: PF5	29: PA1	Timer 0 Compli- mentary Dead Time	
TIMO_OBTIT	2: PB11	6: PB15	10: PC9	14: PD10	18: PD14	22: PF2	26: PF6	30: PA2	Insertion channel 1.	
	3: PB12	7: PC6	11: PC10	15: PD11	19: PD15	23: PF3	27: PF7	31: PA3		
	0: PA5	4: PB14	8: PC8	12: PD9	16: PD13	20: PF1	24: PF5	28: PA1		
TIM0_CDTI2	1: PB11	5: PB15	9: PC9	13: PD10	17: PD14	21: PF2	25: PF6	29: PA2	Timer 0 Compli- mentary Dead Time	
111110_02112	2: PB12	6: PC6	10: PC10	14: PD11	18: PD15	22: PF3	26: PF7	30: PA3	Insertion channel 2.	
	3: PB13	7: PC7	11: PC11	15: PD12	19: PF0	23: PF4	27: PA0	31: PA4		
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4		
TIM1_CC0	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	Timer 1 Capture Compare input /	
111111_000	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	output channel 0.	
	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7		
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5		
TIM1_CC1	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	Timer 1 Capture Compare input /	
	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	output channel 1.	
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0		
	0: PA2	4: PB11	8: PB15	12: PC9	16: PD10	20: PD14	24: PF2	28: PF6		
TIM1_CC2	1: PA3	5: PB12	9: PC6	13: PC10	17: PD11	21: PD15	25: PF3	29: PF7	Timer 1 Capture Compare input /	
002	2: PA4 6: PB13 10: PC7 14: PC11 18: PD12 22: PF0	26: PF4	30: PA0	output channel 2.						
	3: PA5	7: PB14	11: PC8	15: PD9	19: PD13	23: PF1	27: PF5	31: PA1		

Alternate				LOCA	ATION				
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PA3	4: PB12	8: PC6	12: PC10	16: PD11	20: PD15	24: PF3	28: PF7	
TIMA 000	1: PA4	5: PB13	9: PC7	13: PC11	17: PD12	21: PF0	25: PF4	29: PA0	Timer 1 Capture
TIM1_CC3	2: PA5	6: PB14	10: PC8	14: PD9	18: PD13	22: PF1	26: PF5	30: PA1	Compare input / output channel 3.
	3: PB11	7: PB15	11: PC9	15: PD10	19: PD14	23: PF2	27: PF6	31: PA2	
	0: PA2	4: PB11	8: PB15	12: PC9	16: PD10	20: PD14	24: PF2	28: PF6	
1100 0116	1: PA3	5: PB12	9: PC6	13: PC10	17: PD11	21: PD15	25: PF3	29: PF7	USART0 clock in-
US0_CLK	2: PA4	6: PB13	10: PC7	14: PC11	18: PD12	22: PF0	26: PF4	30: PA0	put / output.
	3: PA5	7: PB14	11: PC8	15: PD9	19: PD13	23: PF1	27: PF5	31: PA1	
	0: PA3	4: PB12	8: PC6	12: PC10	16: PD11	20: PD15	24: PF3	28: PF7	
1100 00	1: PA4	5: PB13	9: PC7	13: PC11	17: PD12	21: PF0	25: PF4	29: PA0	USART0 chip se-
US0_CS	2: PA5	6: PB14	10: PC8	14: PD9	18: PD13	22: PF1	26: PF5	30: PA1	lect input / output.
	3: PB11	7: PB15	11: PC9	15: PD10	19: PD14	23: PF2	27: PF6	31: PA2	
	0: PA4	4: PB13	8: PC7	12: PC11	16: PD12	20: PF0	24: PF4	28: PA0	
LICO CTC	1: PA5	5: PB14	9: PC8	13: PD9	17: PD13	21: PF1	25: PF5	29: PA1	USARTO Clear To
US0_CTS 2:	2: PB11	6: PB15	10: PC9	14: PD10	18: PD14	22: PF2	26: PF6	30: PA2	Send hardware flow control input.
	3: PB12	7: PC6	11: PC10	15: PD11	19: PD15	23: PF3	27: PF7	31: PA3	
	0: PA5	4: PB14	8: PC8	12: PD9	16: PD13	20: PF1	24: PF5	28: PA1	
LICO DTC	1: PB11	5: PB15	9: PC9	13: PD10	17: PD14	21: PF2	25: PF6	29: PA2	USART0 Request
US0_RTS	2: PB12	6: PC6	10: PC10	14: PD11	18: PD15	22: PF3	26: PF7	30: PA3	To Send hardware flow control output.
	3: PB13	7: PC7	11: PC11	15: PD12	19: PF0	23: PF4	27: PA0	31: PA4	
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	USART0 Asynchro-
	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	nous Receive.
US0_RX	2: PA3	6: PB12	10: PC6	14: PC10	18: PD11	22: PD15	26: PF3	30: PF7	USART0 Synchro- nous mode Master
	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	Input / Slave Output (MISO).
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	USART0 Asynchro-
	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	nous Transmit. Also used as receive
LICO TV	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	input in half duplex communication.
US0_TX	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	USART0 Synchro-
									nous mode Master Output / Slave In- put (MOSI).
	0: PA2	4: PB11	8: PB15	12: PC9	16: PD10	20: PD14	24: PF2	28: PF6	
US1_CLK	1: PA3	5: PB12	9: PC6	13: PC10	17: PD11	21: PD15	25: PF3	29: PF7	USART1 clock in-
OSI_CLK	2: PA4	6: PB13	10: PC7	14: PC11	18: PD12	22: PF0	26: PF4	30: PA0	put / output.
	3: PA5	7: PB14	11: PC8	15: PD9	19: PD13	23: PF1	27: PF5	31: PA1	

Alternate				LOCA	ATION				
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PA3	4: PB12	8: PC6	12: PC10	16: PD11	20: PD15	24: PF3	28: PF7	
1104 00	1: PA4	5: PB13	9: PC7	13: PC11	17: PD12	21: PF0	25: PF4	29: PA0	USART1 chip se-
US1_CS	2: PA5	6: PB14	10: PC8	14: PD9	18: PD13	22: PF1	26: PF5	30: PA1	lect input / output.
	3: PB11	7: PB15	11: PC9	15: PD10	19: PD14	23: PF2	27: PF6	31: PA2	
	0: PA4	4: PB13	8: PC7	12: PC11	16: PD12	20: PF0	24: PF4	28: PA0	
LICA CTC	1: PA5	5: PB14	9: PC8	13: PD9	17: PD13	21: PF1	25: PF5	29: PA1	USART1 Clear To
US1_CTS	2: PB11	6: PB15	10: PC9	14: PD10	18: PD14	22: PF2	26: PF6	30: PA2	Send hardware flow control input.
	3: PB12	7: PC6	11: PC10	15: PD11	19: PD15	23: PF3	27: PF7	31: PA3	
	0: PA5	4: PB14	8: PC8	12: PD9	16: PD13	20: PF1	24: PF5	28: PA1	
LICA DTC	1: PB11	5: PB15	9: PC9	13: PD10	17: PD14	21: PF2	25: PF6	29: PA2	USART1 Request
US1_RTS	2: PB12	6: PC6	10: PC10	14: PD11	18: PD15	22: PF3	26: PF7	30: PA3	To Send hardware flow control output.
	3: PB13	7: PC7	11: PC11	15: PD12	19: PF0	23: PF4	27: PA0	31: PA4	
	0: PA1	4: PA5	8: PB14	12: PC8	16: PD9	20: PD13	24: PF1	28: PF5	USART1 Asynchro-
	1: PA2	5: PB11	9: PB15	13: PC9	17: PD10	21: PD14	25: PF2	29: PF6	nous Receive.
US1_RX	2: PA3 6: PB12 10: PC6 14: PC10 18: PD11 22: PD15 26: PF3 30: PF7 r	USART1 Synchro- nous mode Master							
3	3: PA4	7: PB13	11: PC7	15: PC11	19: PD12	23: PF0	27: PF4	31: PA0	Input / Slave Output (MISO).
	0: PA0	4: PA4	8: PB13	12: PC7	16: PC11	20: PD12	24: PF0	28: PF4	USART1 Asynchronous Transmit. Al-
	1: PA1	5: PA5	9: PB14	13: PC8	17: PD9	21: PD13	25: PF1	29: PF5	so used as receive
US1 TX	2: PA2	6: PB11	10: PB15	14: PC9	18: PD10	22: PD14	26: PF2	30: PF6	input in half duplex communication.
001_17	3: PA3	7: PB12	11: PC6	15: PC10	19: PD11	23: PD15	27: PF3	31: PF7	USART1 Synchro- nous mode Master Output / Slave In- put (MOSI).
				12: PF0	16: PF5			30: PA5	
1100 0116				13: PF1	17: PF6				USART2 clock in-
US2_CLK				14: PF3	18: PF7				put / output.
				15: PF4					
			11: PF0	12: PF1	16: PF6			29: PA5	
1100 00				13: PF3	17: PF7				USART2 chip se-
US2_CS				14: PF4					lect input / output.
				15: PF5					
			10: PF0	12: PF3	16: PF7			28: PA5	
LICA CTO			11: PF1	13: PF4					USART2 Clear To
US2_CTS				14: PF5					Send hardware flow control input.
				15: PF6					

Alternate	LOCATION								
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
			9: PF0	12: PF4			27: PA5		
LICO DTO			10: PF1	13: PF5					USART2 Request
US2_RTS			11: PF3	14: PF6					To Send hardware flow control output.
				15: PF7					
				13: PF0	16: PF4			31: PA5	USART2 Asynchro-
				14: PF1	17: PF5				nous Receive.
US2_RX				15: PF3	18: PF6				USART2 Synchro- nous mode Master
					19: PF7				Input / Slave Output (MISO).
	0: PA5			14: PF0	16: PF3	20: PF7			USART2 Asynchro- nous Transmit. Al-
				15: PF1	17: PF4				so used as receive
US2_TX					18: PF5				input in half duplex communication.
032_1X					19: PF6				USART2 Synchro-
									nous mode Master Output / Slave In- put (MOSI).
	0: PA1								Digital to analog
VDAC0_EXT									converter VDAC0 external reference input pin.
VDAGO OUTO /	0: PA3								Digital to Analog
VDAC0_OUT0 / OPA0_OUT									Converter DAC0 output channel number 0.
VDAC0_OUT0AL	0: PA5								Digital to Analog
T / OPA0_OUT-	1: PD13								Converter DAC0 al- ternative output for
ALT	2: PD15								channel 0.
	0: PD14								Digital to Analog
VDAC0_OUT1 / OPA1_OUT									Converter DAC0 output channel number 1.
VDAC0_OUT1AL	0: PD12								Digital to Analog
T / OPA1_OUT-	1: PA2								Converter DAC0 alternative output for
ALT	2: PA4								channel 1.
	0: PA0	4: PA4		15: PB11	16: PB12		26: PC6	28: PC8	
MATINA COS	1: PA1	5: PA5			17: PB13		27: PC7	29: PC9	Wide timer 0 Cap- ture Compare in-
WTIM0_CC0	2: PA2				18: PB14			30: PC10	put / output channel 0.
	3: PA3				19: PB15			31: PC11	O.

Alternate									
Functionality	0 - 3	4 - 7	8 - 11	12 - 15	16 - 19	20 - 23	24 - 27	28 - 31	Description
	0: PA2			13: PB11	16: PB14		24: PC6	28: PC10	
W/TIMO 004	1: PA3			14: PB12	17: PB15		25: PC7	29: PC11	Wide timer 0 Cap- ture Compare in-
WTIM0_CC1	2: PA4			15: PB13			26: PC8	31: PD9	put / output channel
	3: PA5						27: PC9		
	0: PA4		11: PB11	12: PB12		22: PC6	24: PC8	29: PD9	
WTIM0 CC2	1: PA5			13: PB13		23: PC7	25: PC9	30: PD10	Wide timer 0 Cap- ture Compare in-
WTIMO_CC2				14: PB14			26: PC10	31: PD11	put / output channel 2.
				15: PB15			27: PC11		
		7: PB11	8: PB12		18: PC6	20: PC8	25: PD9	28: PD12	Mida timan O Cam
WITING CDTIG			9: PB13		19: PC7	21: PC9	26: PD10	29: PD13	Wide timer 0 Com- plimentary Dead
WTIM0_CDTI0			10: PB14			22: PC10	27: PD11	30: PD14	Time Insertion channel 0.
			11: PB15			23: PC11		31: PD15	Citatino Ci
		5: PB11	8: PB14		16: PC6	20: PC10	24: PD10	28: PD14	
WITING CDTIA		6: PB12	9: PB15		17: PC7	21: PC11	25: PD11	29: PD15	Wide timer 0 Com- plimentary Dead
WTIM0_CDTI1		7: PB13			18: PC8	23: PD9	26: PD12	30: PF0	Time Insertion channel 1.
					19: PC9		27: PD13	31: PF1	Citatillor 1:
	3: PB11	4: PB12		14: PC6	16: PC8	21: PD9	24: PD12	28: PF0	
WITING COTIO		5: PB13		15: PC7	17: PC9	22: PD10	25: PD13	29: PF1	Wide timer 0 Com- plimentary Dead
WTIM0_CDTI2		6: PB14			18: PC10	23: PD11	26: PD14	30: PF2	Time Insertion channel 2.
		7: PB15			19: PC11		27: PD15	31: PF3	Gridiffici Z.

Certain alternate function locations may have non-interference priority. These locations will take precedence over any other functions selected on that pin (i.e. another alternate function enabled to the same pin inadvertently).

Some alternate functions may also have high speed priority on certain locations. These locations ensure the fastest possible paths to the pins for timing-critical signals.

The following table lists the alternate functions and locations with special priority.

Table 6.8. Alternate Functionality Priority

Alternate Functionality	Location	Priority
CMU_CLKI0	1: PF7	High Speed

6.8 Analog Port (APORT) Client Maps

The Analog Port (APORT) is an infrastructure used to connect chip pins with on-chip analog clients such as analog comparators, ADCs, DACs, etc. The APORT consists of a set of shared buses, switches, and control logic needed to configurably implement the signal routing. Figure 6.6 APORT Connection Diagram on page 156 shows the APORT routing for this device family (note that available features may vary by part number). A complete description of APORT functionality can be found in the Reference Manual.

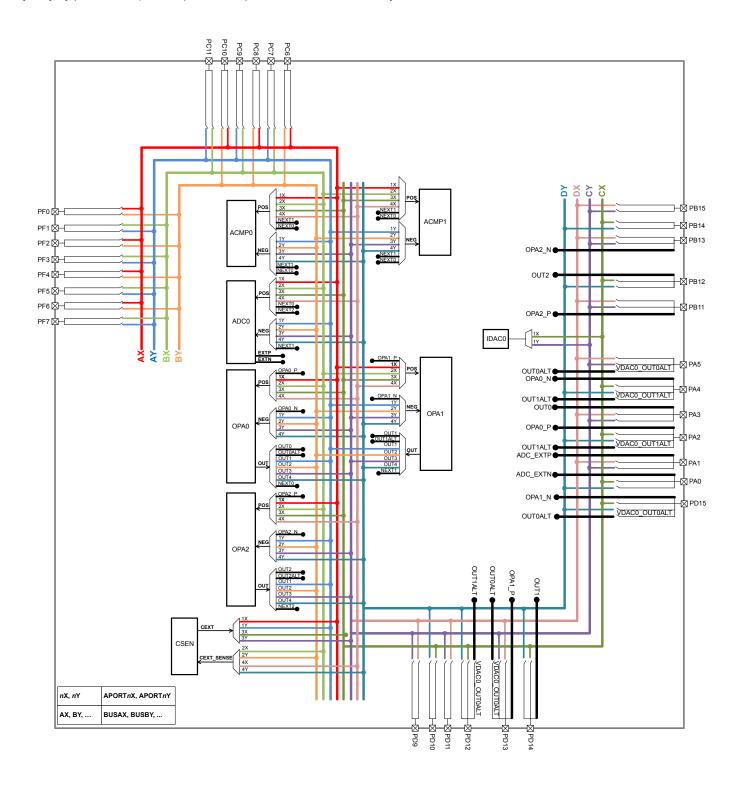


Figure 6.6. APORT Connection Diagram

Client maps for each analog circuit using the APORT are shown in the following tables. The maps are organized by bus, and show the peripheral's port connection, the shared bus, and the connection from specific bus channel numbers to GPIO pins.

In general, enumerations for the pin selection field in an analog peripheral's register can be determined by finding the desired pin connection in the table and then combining the value in the Port column (APORT__), and the channel identifier (CH__). For example, if pin PF7 is available on port APORT2X as CH23, the register field enumeration to connect to PF7 would be APORT2XCH23. The shared bus used by this connection is indicated in the Bus column.

Table 6.9. ACMP0 Bus and Pin Mapping

Port	Bus	CH31	СН30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	СН8	CH7	СН6	CH5	CH4	СНЗ	CH2	CH1	СНО
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		LC7							
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

Table 6.10. ACMP1 Bus and Pin Mapping

Port	Bus	CH31	СН30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	CH8	CH7	СН6	СН5	CH4	СНЗ	CH2	CH1	СНО
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

Table 6.11. ADC0 Bus and Pin Mapping

Port	Bus	CH31	СН30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	6НЭ	CH8	CH7	9НО	CH5	CH4	СНЗ	CH2	CH1	СНО
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

Table 6.12. CSEN Bus and Pin Mapping

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	СН8	CH7	9Н2	CH5	CH4	СНЗ	CH2	CH1	СНО
CE																																	
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
CE	XT_	SEN	ISE																														
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

Table 6.13. IDAC0 Bus and Pin Mapping

Port	Bus	CH31	СН30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	CH8	CH7	9НЭ	CH5	CH4	СНЗ	CH2	CH1	СНО
APORT1X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT1Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	

Table 6.14. VDAC0 / OPA Bus and Pin Mapping

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	СН8	CH7	СН6	СН5	CH4	СНЗ	CH2	CH1	СНО
ОР	A0_	N																															
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
ОР	A0_	Р	•			•		•			•											•				•			•		'	,	
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	CH8	CH7	СН6	CH5	CH4	СНЗ	CH2	CH1	СНО
ОР	A1_	N		ı									ı			ı					ı						ı						
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
ОР	A1_	Р			•								•		•	•										•	•					'	
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
ОР	A2_	N															ı								ı			ı					
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	CH8	CH7	СН6	CH5	CH4	СНЗ	CH2	CH1	СНО
ОР	A2_	OU.	Т																														
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
OP	A2_	Р																															
APORT1X	BUSAX										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT2X	BUSBX									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT3X	BUSCX		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		
APORT4X	BUSDX	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
VD	AC	0_0	JT0	/ OI	PA0	_οι	JT																		ı								
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		9OA						
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

Port	Bus	CH31	CH30	CH29	CH28	CH27	CH26	CH25	CH24	CH23	CH22	CH21	CH20	CH19	CH18	CH17	CH16	CH15	CH14	CH13	CH12	CH11	CH10	СНЭ	СН8	CH7	СН6	СН5	CH4	СНЗ	CH2	CH1	СНО
VD	ACC	_ol	JT1	/ OF	PA1	_ou	JT																										
APORT1Y	BUSAY									PF7		PF5		PF3		PF1						PC11		PC9		PC7							
APORT2Y	BUSBY										PF6		PF4		PF2		PF0						PC10		PC8		PC6						
APORT3Y	BUSCY	PB15		PB13		PB11														PA5		PA3		PA1		PD15		PD13		PD11		PD9	
APORT4Y	BUSDY		PB14		PB12																PA4		PA2		PA0		PD14		PD12		PD10		

7. QFN48 Package Specifications

7.1 QFN48 Package Dimensions

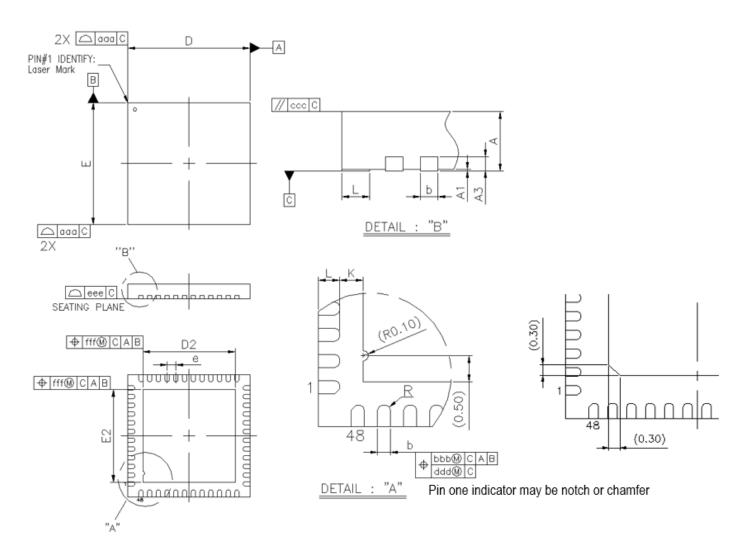


Figure 7.1. QFN48 Package Drawing

Table 7.1. QFN48 Package Dimensions

Dimension	Min	Тур	Max
A	0.80	0.85	0.90
A1	0.00	0.02	0.05
A3		0.20 REF	
b	0.18	0.25	0.30
D	6.90	7.00	7.10
Е	6.90	7.00	7.10
D2	5.15	5.30	5.45
E2	5.15	5.30	5.45
е		0.50 BSC	
L	0.30	0.40	0.50
К	0.20	_	_
R	0.09	_	_
aaa		0.15	
bbb		0.10	
ccc		0.10	
ddd		0.05	
eee		0.08	
fff		0.10	

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
- 3. This drawing conforms to the JEDEC Solid State Outline MO-220, Variation VKKD-4.
- 4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

7.2 QFN48 PCB Land Pattern

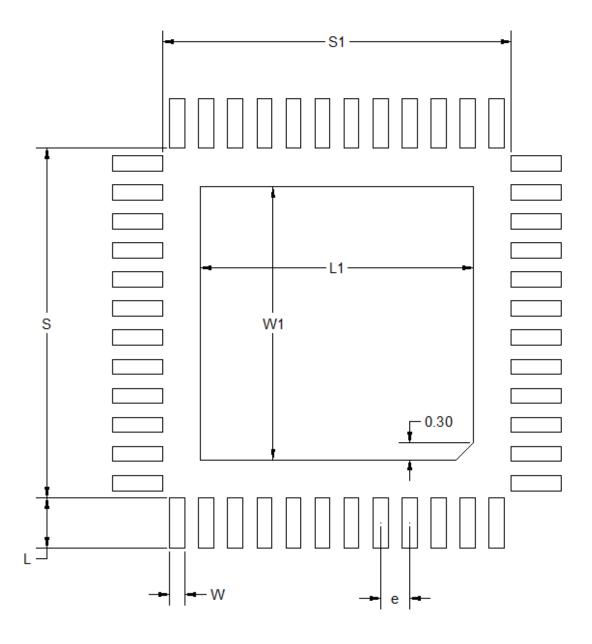


Figure 7.2. QFN48 PCB Land Pattern Drawing

Table 7.2. QFN48 PCB Land Pattern Dimensions

Dimension	Тур
S1	6.01
S	6.01
L1	4.70
W1	4.70
e	0.50
W	0.26
L	0.86

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. This Land Pattern Design is based on the IPC-7351 guidelines.
- 3. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be $60 \mu m$ minimum, all the way around the pad.
- 4. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
- 5. The stencil thickness should be 0.125 mm (5 mils).
- 6. The ratio of stencil aperture to land pad size can be 1:1 for all perimeter pads.
- 7. A 4x4 array of 0.75 mm square openings on a 1.00 mm pitch can be used for the center ground pad.
- 8. A No-Clean, Type-3 solder paste is recommended.
- 9. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

7.3 QFN48 Package Marking



Figure 7.3. QFN48 Package Marking

The package marking consists of:

- PPPPPPPP The part number designation.
 - 1. Family Code (B | M | F)
 - 2. G (Gecko)
 - 3. Series (1, 2,...)
 - 4. Device Configuration (1, 2,...)
 - 5. Performance Grade (P | B | V)
 - 6. Feature Code (1 to 7)
 - 7. TRX Code (3 = TXRX | 2= RX | 1 = TX)
 - 8. Band (1 = Sub-GHz | 2 = 2.4 GHz | 3 = Dual-band)
 - 9. Flash (J = 1024K | H = 512K | G = 256K | F = 128K | E = 64K | D = 32K)
 - 10. Temperature Grade (G = -40 to 85 | I = -40 to 125)
- · YY The last 2 digits of the assembly year.
- WW The 2-digit workweek when the device was assembled.
- TTTTTT A trace or manufacturing code. The first letter is the device revision.

8. RQFN32 Package Specifications

8.1 RQFN32 Package Dimensions

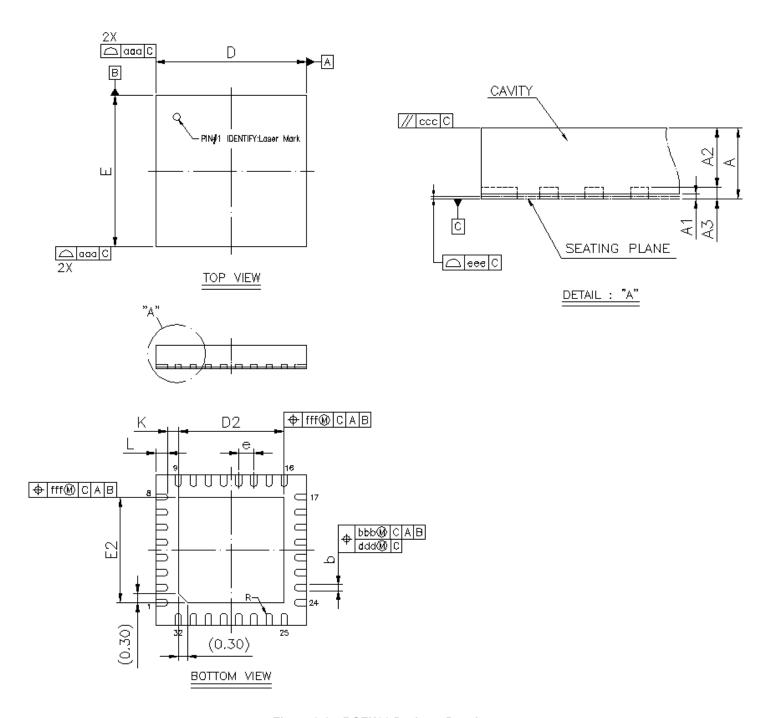


Figure 8.1. RQFN32 Package Drawing

Table 8.1. RQFN32 Package Dimensions

Dimension	Min	Тур	Max
A	0.75	0.85	0.90
A1	_	_	0.05
A2	0.70	0.75	0.80
A3	_	0.10 REF	_
D/E	4.90	5.00	5.10
D2/E2	3.40	3.50	3.60
е		0.50 BSC	
b	0.150	0.200	0.250
L	0.350	0.400	0.450
R	0.075	0.10	0.125
К	0.20	_	_
aaa		0.150	
bbb		0.100	
ccc		0.100	
ddd		0.050	
eee		0.080	
fff		0.100	
	•		

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. Dimensioning and Tolerancing per ANSI Y14.5M-1994.
- 3. This drawing conforms to the JEDEC Solid State Outline MO-220, Variation VKKD-4.
- 4. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

8.2 RQFN32 PCB Land Pattern

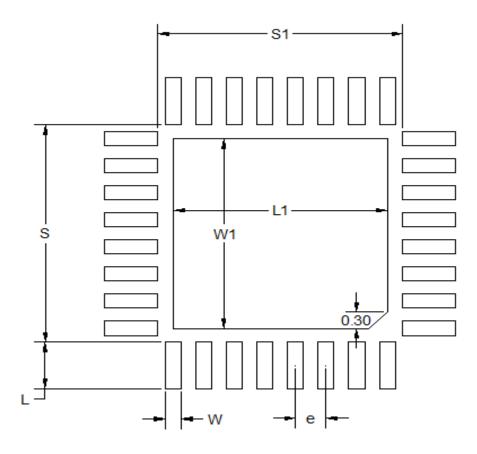


Figure 8.2. RQFN32 PCB Land Pattern Drawing

Table 8.2. RQFN32 PCB Land Pattern Dimensions

Dimension	Тур
S1	4.01
S	4.01
L1	3.50
W1	3.50
е	0.50
W	0.26
L	0.86

- 1. All dimensions shown are in millimeters (mm) unless otherwise noted.
- 2. This Land Pattern Design is based on the IPC-7351 guidelines.
- 3. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be $60 \mu m$ minimum, all the way around the pad.
- 4. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
- 5. The stencil thickness should be 0.125 mm (5 mils).
- 6. The ratio of stencil aperture to land pad size can be 1:1 for all perimeter pads.
- 7. A 3x3 array of 0.85 mm square openings on a 1.00 mm pitch can be used for the center ground pad.
- 8. A No-Clean, Type-3 solder paste is recommended.
- 9. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

8.3 RQFN32 Package Marking



Figure 8.3. RQFN32 Package Marking

The package marking consists of:

- PPPPPPPP The part number designation.
 - 1. Family Code (B | M | F)
 - 2. G (Gecko)
 - 3. Series (1, 2,...)
 - 4. Device Configuration (1, 2,...)
 - 5. Performance Grade (P | B | V)
 - 6. Feature Code (1 to 7)
 - 7. TRX Code (3 = TXRX | 2= RX | 1 = TX)
 - 8. Band (1 = Sub-GHz | 2 = 2.4 GHz | 3 = Dual-band)
 - 9. Flash (J = 1024K | H = 512K | G = 256K | F = 128K | E = 64K | D = 32K)
 - 10. Temperature Grade (G = -40 to 85 | I = -40 to 125)
- YY The last 2 digits of the assembly year.
- WW The 2-digit workweek when the device was assembled.
- TTTTTT A trace or manufacturing code. The first letter is the device revision.

9. Revision History

Revision 1.3

November 2018

- Updated the current consumed in receive mode under 802.15.4 receiving frame test condition at 2.4 GHz frequency with and without radio pre-scaling - Table 4.8 Current Consumption Using Radio 3.3 V with DC-DC on page 34.
- "PAVDD" references in Sub-GHz specification tables clarified to refer to "External PA Supply". The PAVDD pin is for the 2.4 GHz radio.
- Table 6.6 GPIO Functionality Table on page 132 re-sorted to list pins alphabetically by GPIO Name column.
- · Reworded or removed mentions of "modules" in reference to device peripherals in system overview.
- Ordered table footnotes in the order in which their references appear in the table.
- Table 4.25 Sub-GHz RF Receiver Characteristics for 915 MHz Band on page 58 Clarified that 400 kbps reference signal is 4GFSK.
- Corrected the reference signal value in Table 4.19 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 500 kbps Data Rate on page 48 and Table 4.21 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 125 kbps Data Rate on page 50.
- Corrected the Error vector magnitude parameter in Table 4.22 RF Transmitter Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band on page 51.
- · Corrected minor typos throughout the document.

Revision 1.2

March 2018

- · Added new orderable part numbers in QFN32 packaging.
- Figure 6.6 APORT Connection Diagram on page 156: Corrected OPA output connections to route through "Y" buses.

Revision 1.1

January 2017

- Updated 3.12 Memory Map with latest formatting and low-energy peripherals.
- 4.1.1 Absolute Maximum Ratings:
 - Changed "Non-5V tolerant GPIO pins" to "Standard GPIO pins".
 - Added footnotes to clarify V_{DIGPIN} specification for 5V tolerant GPIO.
- Table 4.2 General Operating Conditions on page 24: Added footnote for additional information on peak current during voltage scaling operations.
- 4.1.9.4 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 1 Mbps Data Rate: BLOCK_{OOB} specifications changed to show Min values instead of Typ, and footnote added.
- 4.1.9.6 RF Receiver Characteristics for 2GFSK in the 2.4GHz Band, 2 Mbps Data Rate: BLOCK_{OOB} specifications removed (not part of BLE 2 Mbps specification).
- 4.1.9.7 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 500 kbps Data Rate and 4.1.9.9 RF Transmitter Characteristics for 2GFSK in the 2.4GHz Band, 125 kbps Data Rate: PSD_{LIMIT} changed to specify maximum instead of typ, with footnote added for FCC output power limit.
- 4.1.9.12 RF Receiver Characteristics for 802.15.4 DSSS-OQPSK in the 2.4 GHz Band: Footnote added to BLOCK_{80211G} specification to clarify blocker signal definition.
- 4.1.10.2 Sub-GHz RF Receiver Characteristics for 915 MHz Band: Added O-QPSK DSSS phy specifications.
- 4.1.18 Digital to Analog Converter (VDAC): Widened V_{GAIN} Gain error limits.
- 4.1.25 USART SPI:
 - SPI Slave Timing: Corrected t_{SU MO}, t_{H MO} and t_{SCLK MIN} limits.
 - Updated remainder of specifications to match formatting and common specs in all EFR32xG1x product families.
- · 4.2.3 2.4 GHz Radio:
 - Extended temperature plots to 125 degrees C.
 - · Updated RX sensitivity plots with latest phy characterization data.
- Added Figure 6.6 APORT Connection Diagram on page 156.

Revision 1.0

2017-Aug-02

- Updated spcification tables with latest characterization values and production test min/max limits.
- · Added high-temperature OPNs and associated specifications.
- Added performance specifications and supported radio modulations/protocols to Feature List.
- Clarified / corrected energy mode mentions in RTCC and Opamp sections of the System Overview.
- Sub-GHz RF Transmitter characteristics for 868 MHz Band Electrical Specifications Table:
 - POUT_{VAR V NODCDC} specification symbol corrected to POUT_{VAR V}
 - POUT_{VAR F NODCDC} specification symbol corrected to POUT_{VAR F}
- Sub-GHz RF Transmitter characteristice for 433 MHz Band Electrical Specifications Table: POT_{MIN} specification symbol corrected to POUT_{MIN}
- Analog to Digital Converter (ADC) Electrical Specifications Table: Added footnote for clarification of input voltage limits.
- Typical Sub-GHz Impedance-matching network circuits Figure: Corrected split between two examples from 450 MHz to 500 MHz.
- RF Transmitter General Characteristics for 2.4 GHz Band Electrical Specifications Table:
 - Test Conditions changed from "19.5 dBm" to "19 dBm" and from "10.5 dBm" to "10 dBm"
- RF Receiver Characteristics for Bluetooth Low Energy in the 2.4GHz Band, 1 Mbps Data Rate Electrical Specifications Table:
 - Sensitivity (SENS) for Reference Signal changed from "-95.8 dBm" to "-94.8 dBm"

Revision 0.5

2017-Apr-25

- · Added RFSENSE section to System Overview.
- · Updated specification tables with latest characterization results.
- Split 2.4 GHz 2GFSK tables into separate tables for 1 Mbps, 2 Mbps, 500kbps, and 125kbps data rates.
- · Added typical performance graphs.
- · Condensed pin function tables with new formatting.
- · Added APORT Connection Diagram.
- · Corrected package marking flash size designator.
- · Removed OPNs for QFN32 package options.

Revision 0.1

2016-Nov-15

Initial release.





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