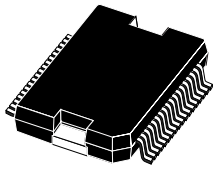
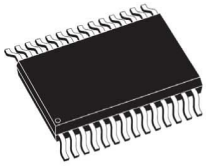
L6470

Fully integrated microstepping motor driver with motion engine and SPI

Datasheet - production data



POWERSO36



HTSSOP28

Features

- Operating voltage: 8 - 45 V
- 7.0 A out peak current (3.0 A $_{r.m.s.}$)
- Low $R_{DS(on)}$ Power MOSFETs
- Programmable speed profile and positioning
- Programmable power MOS slew rate
- Up to 1/128 microstepping
- Sensorless stall detection
- SPI interface
- Low quiescent and standby currents
- Programmable non-dissipative overcurrent protection on high and low-side
- Two-levels of overtemperature protection

Applications

- Bipolar stepper motors

Description

The L6470 device, realized in analog mixed signal technology, is an advanced fully integrated solution suitable for driving two-phase bipolar stepper motors with microstepping. It integrates a dual low $R_{DS(on)}$ DMOS full bridge with all of the power switches equipped with an accurate on-chip current sensing circuitry suitable for non-dissipative current control and overcurrent protection. Thanks to a unique control system, a true 1/128 steps resolution is achieved. The digital control core can generate user defined motion profiles with acceleration, deceleration, speed or target position, easily programmed through a dedicated registers set. All commands and data registers, including those used to set analogue values (i.e. current control value, current protection trip point, deadtime, PWM frequency, etc.) are sent through a standard 5-Mbit/s SPI. A very rich set of protections (thermal, low bus voltage, overcurrent, motor stall) allows the design of a fully protected application, as required by the most demanding motor control applications.

Table 1. Device summary

Order codes	Package	Packaging
L6470H	HTSSOP28	Tube
L6470HTR	HTSSOP28	Tape and reel
L6470PD	POWERSO36	Tube
L6470PDTR	POWERSO36	Tape and reel

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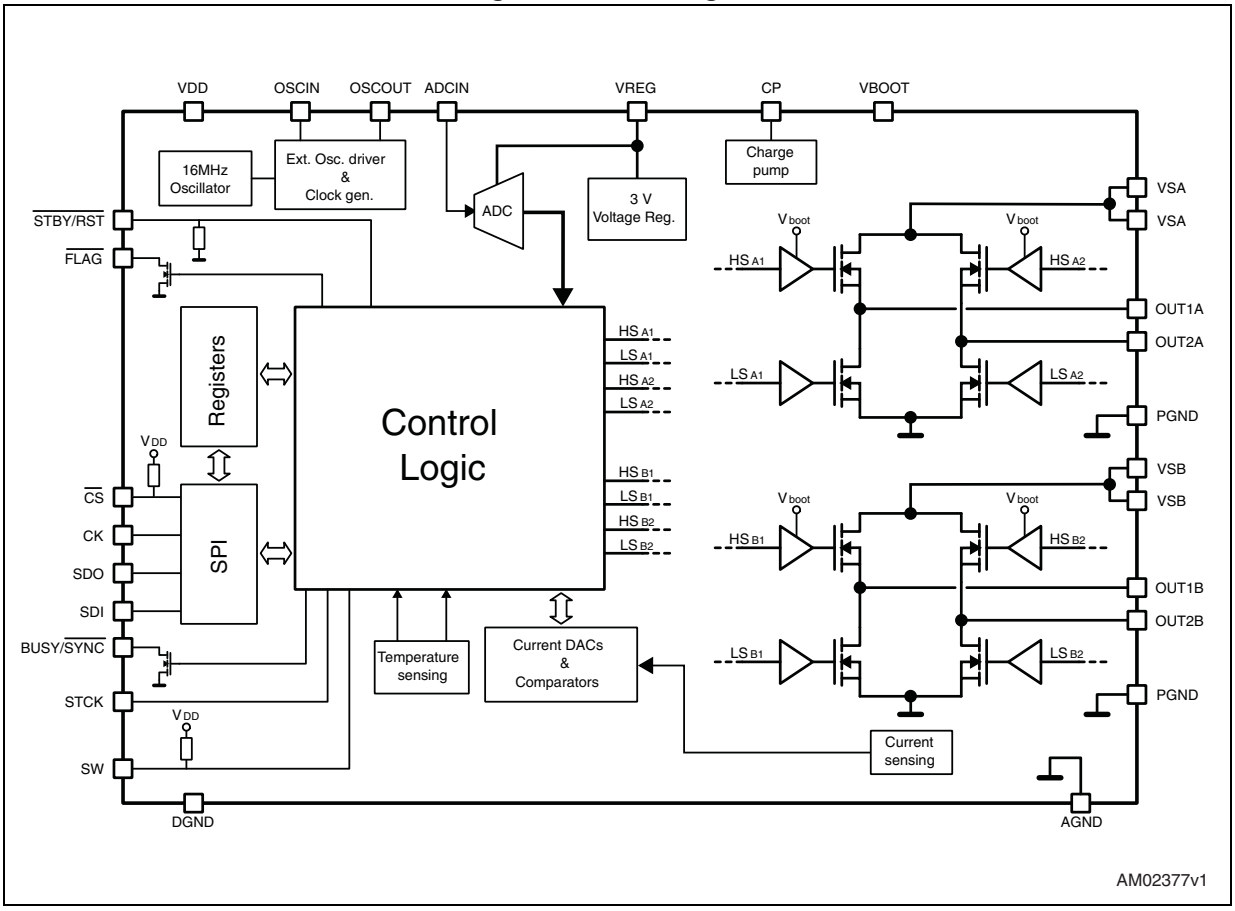
1 Block diagram

Figure 1. Block diagram



2 Electrical data

2.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Test condition	Value	Unit
V_{DD}	Logic interface supply voltage		5.5	V
V_S	Motor supply voltage	$V_{SA} = V_{SB} = V_S$	48	V
$V_{GND, diff}$	Differential voltage between AGND, PGND and DGND		±0.3	V
V_{boot}	Bootstrap peak voltage		55	V
V_{REG}	Internal voltage regulator output pin and logic supply voltage		3.6	V
V_{ADCIN}	Integrated ADC input voltage range (ADCIN pin)		-0.3 to +3.6	V
V_{OSC}	OSCIN and OSCOUT pin voltage range		-0.3 to +3.6	V
V_{out_diff}	Differential voltage between V_{SA}, $OUT1_A$, $OUT2_A$, PGND and V_{SB}, $OUT1_B$, $OUT2_B$, PGND pins	$V_{SA} = V_{SB} = V_S$	48	V
V_{LOGIC}	Logic inputs voltage range		-0.3 to +5.5	V
I_{out} (1)	R.m.s. output current		3	A
I_{out_peak} (1)	Pulsed output current	T_{PULSE} < 1 ms	7	A
T_{OP}	Operating junction temperature		-40 to 150	°C
T_s	Storage temperature range		-55 to 150	°C
P_{tot}	Total power dissipation (T_A = 25 ºC)	(2)	5	W

1. Maximum output current limit is related to metal connection and bonding characteristics. Actual limit must satisfy maximum thermal dissipation constraints.
2. HTSSOP28 mounted on EVAL6470H.

2.2 Recommended operating conditions

Table 3. Recommended operating conditions

Symbol	Parameter	Test condition	Value			Unit
V_{DD}	Logic interface supply voltage	3.3 V logic outputs		3.3		V
		5 V logic outputs		5		
V_S	Motor supply voltage	$V_{SA} = V_{SB} = V_S$	8		45	V
V_{out_diff}	Differential voltage between V_{SA}, $OUT1_A$, $OUT2_A$, PGND and V_{SB}, $OUT1_B$, $OUT2_B$, PGND pins	$V_{SA} = V_{SB} = V_S$			45	V
$V_{REG,in}$	Logic supply voltage	V_{REG} voltage imposed by external source	3.2	3.3		V
V_{ADC}	Integrated ADC input voltage (ADCIN pin)		0		V_{REG}	V

2.3 Thermal data

Table 4. Thermal data

Symbol	Parameter	Package	Typ.	Unit
R_{thJA}	Thermal resistance junction ambient	HTSSOP28(1)	22	°C/W
		POWERSO36(2)	12	

1. HTSSOP28 mounted on the EVAL6470H rev 1.0 board: a four-layer FR4 PCB with a dissipating copper surface of about 40 cm^2 on each layer and 15 via holes below the IC.
2. POWERSO36 mounted on the EVAL6470PD rev 1.0 board: a four-layer FR4 PCB with a dissipating copper surface of about 40 cm^2 on each layer and 22 via holes below the IC.

3 Electrical characteristics

V_{SA} = V_{SB} = 36 V; V_{DD} = 3.3 V; internal 3 V regulator; T_J = 25 °C, unless otherwise specified.

Table 5. Electrical characteristics

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
General						
V_{SthOn}	V_S UVLO turn-on threshold		7.5	8.2	8.9	V
V_{SthOff}	V_S UVLO turn-off threshold		6.6	7.2	7.8	V
$V_{SthHyst}$	V_S UVLO threshold hysteresis		0.7	1	1.3	V
I_q	Quiescent motor supply current	Internal oscillator selected; VREG = 3.3 V ext.; CP floating		0.5	0.65	mA
$T_{j(WRN)}$	Thermal warning temperature			130		°C
$T_{j(SD)}$	Thermal shutdown temperature			160		°C
Charge pump						
V_{pump}	Voltage swing for charge pump oscillator			10		V
$f_{pump,min}$	Minimum charge pump oscillator frequency(1)			660		kHz
$f_{pump,max}$	Maximum charge pump oscillator frequency(1)			800		kHz
I_{boot}	Average boot current	$f_{sw,A}$ = $f_{sw,B}$ = 15.6 kHz POW_SR = '10'		1.1	1.4	mA
Output DMOS transistor						
$R_{DS(on)}$	High-side switch on-resistance	T_j = 25 °C, I_{out} = 3 A		0.37		Ω
		T_j = 125 °C, (2) I_{out} = 3 A		0.51		
$R_{DS(on)}$	Low-side switch on-resistance	T_j = 25 °C, I_{out} = 3 A		0.18		
		T_j = 125 °C, (2) I_{out} = 3 A		0.23		
I_{DSS}	Leakage current	OUT = V_S			3.1	mA
		OUT = GND	-0.3			
t_r	Rise time(3)	POW_SR = '00', I_{out} = +1 A		100		ns
		POW_SR = '00', I_{out} = -1 A		80		
		POW_SR = '11', I_{out} = ±1 A		100		
		POW_SR = '10', I_{lout} = ±1 A		200		
		POW_SR = '01', I_{out} = ±1 A		300		

Table 5. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
t_f	Fall time(3)	POW_SR = '00'; I_{out} = +1 A		90		ns
		POW_SR = '00'; I_{out} = -1 A		110		
		POW_SR = '11', I_{out} = ±1 A		110		
		POW_SR = '10', I_{out} = ±1 A		260		
		POW_SR = '01', I_{load}= ±1 A		375		
SR_{out_r}	Output rising slew rate	POW_SR = '00', I_{out} = +1 A		285		V/µs
		POW_SR = '00', I_{out} = -1 A		360		
		POW_SR = '11', I_{out} = ±1 A		285		
		POW_SR = '10', I_{out} = ±1 A		150		
		POW_SR = '01', I_{out} = ±1 A		95		
SR_{out_f}	Output falling slew rate	POW_SR = '00', I_{out} = +1 A		320		V/µs
		POW_SR = '00', I_{out} = -1 A		260		
		POW_SR = '11', I_{out} = ±1 A		260		
		POW_SR = '10', I_{out} = ±1 A		110		
		POW_SR = '01', I_{out} = ±1 A		75		
Deadtime and blanking						
t_{DT}	Deadtime(1)	POW_SR = '00'		250		ns
		POW_SR = '11', f_{OSC} = 16 MHz		375		
		POW_SR = '10', f_{OSC} = 16 MHz		625		
		POW_SR = '01', f_{OSC} = 16 MHz		875		
t_{blank}	Blanking time(1)	POW_SR = '00'		250		ns
		POW_SR = '11', f_{OSC} = 16 MHz		375		
		POW_SR = '10', f_{OSC} = 16 MHz		625		
		POW_SR = '01', f_{OSC} = 16 MHz		875		
Source-drain diodes						
$V_{SD,HS}$	High-side diode forward ON voltage	I_{out} = 1 A		1	1.1	V
$V_{SD,LS}$	Low-side diode forward ON voltage	I_{out} = 1 A		1	1.1	V
t_{rrHS}	High-side diode reverse recovery time	I_{out} = 1 A		30		ns
t_{rrLS}	Low-side diode reverse recovery time	I_{out} = 1 A		100		ns
Logic inputs and outputs						
V_{IL}	Low logic level input voltage				0.8	V
V_{IH}	High logic level input voltage		2			V
I_{IH}	High logic level input current(4)	V_{IN} = 5 V			1	µA
I_{IL}	Low logic level input current(5)	V_{IN} = 0 V	-1			µA

Table 5. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
V_{OL}	Low logic level output voltage(6)	V_{DD} = 3.3 V, I_{OL} = 4 mA			0.3	V
		V_{DD} = 5 V, I_{OL} = 4 mA			0.3	
V_{OH}	High logic level output voltage	V_{DD} = 3.3 V, I_{OH} = 4 mA	2.4			V
		V_{DD} = 5 V, I_{OH} = 4 mA	4.7			
R_{PU} R_{PD}	CS pull-up and STBY pull-down resistors	$\overline{CS}$ = GND; $\overline{STBY/RST}$ = 5 V	335	430	565	kΩ
I_{logic}	Internal logic supply current	3.3 V V_{REG} externally supplied, internal oscillator		3.7	4.3	mA
$I_{logic,STBY}$	Standby mode internal logic supply current	3.3 V V_{REG} externally supplied		2	2.5	µA
f_{STCK}	Step-clock input frequency				2	MHz
Internal oscillator and external oscillator driver						
$f_{osc,i}$	Internal oscillator frequency	T_j = 25 °C, V_{REG} = 3.3 V	-3%	16	+3%	MHz
$f_{osc,e}$	Programmable external oscillator frequency		8		32	MHz
$V_{OSCOUTH}$	OSCOUT clock source high level voltage	Internal oscillator 3.3 V V_{REG} externally supplied; I_{OSCOUT} = 4 mA	2.4			V
$V_{OSCOUTL}$	OSCOUT clock source low level voltage	Internal oscillator 3.3 V V_{REG} externally supplied; I_{OSCOUT} = 4 mA			0.3	V
$t_{rOSCOUT}$ $t_{fOSCOUT}$	OSCOUT clock source rise and fall time	Internal oscillator			20	ns
t_{extosc}	Internal to external oscillator switching delay			3		ms
t_{intosc}	External to internal oscillator switching delay			1.5		µs
SPI						
$f_{CK,MAX}$	Maximum SPI clock frequency(7)		5			MHz
t_{rCK} t_{fCK}	SPI clock rise and fall time(7)	C_L = 30 pF			25	ns
t_{hCK} t_{lCK}	SPI clock high and low time(7)		75			ns
t_{setCS}	Chip select setup time(7)		350			ns
t_{holCS}	Chip select hold time(7)		10			ns
t_{disCS}	Deselect time(7)		800			ns
t_{setSDI}	Data input setup time(7)		25			ns
t_{holSDI}	Data input hold time(7)		20			ns

Table 5. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
t_{enSDO}	Data output enable time(7)				38	ns
t_{disSDO}	Data output disable time(7)				47	ns
t_{vSDO}	Data output valid time(7)				57	ns
t_{holSDO}	Data output hold time(7)		37			ns
Switch input (SW)						
R_{PUSW}	SW input pull-up resistance	SW = GND	60	85	110	kΩ
PWM modulators						
f_{PWM}	Programmable PWM frequency(1)	f_{osc} = 16 MHz	2.8		62.5	kHz
		f_{osc} = 32 MHz	5.6		125	
N_{PWM}	PWM resolution			8		bit
Stall detection						
$I_{STALL,MAX}$	Maximum programmable stall threshold	STALL_TH = '1111111'		4		A
$I_{STALL,MIN}$	Minimum programmable stall threshold	STALL_TH = '0000000'		31.25		mA
$I_{STALL,RES}$	Programmable stall threshold resolution			31.25		mA
Overcurrent protection						
$I_{OCD,MAX}$	Maximum programmable overcurrent detection threshold	OCD_TH = '1111'		6		A
$I_{OCD,MIN}$	Minimum programmable overcurrent detection threshold	OCD_TH = '0000'		0.375		A
$I_{OCD,RES}$	Programmable overcurrent detection threshold resolution			0.375		A
$t_{OCD,Flag}$	OCD to flag signal delay time	dI_{out}/dt = 350 A/µs		650	1000	ns
$t_{OCD,SD}$	OCD to shutdown delay time	dI_{out}/dt = 350 A/µs POW_SR = '10'		600		ns
Standby						
I_{qSTBY}	Quiescent motor supply current in standby conditions	V_S = 8 V		26	34	µA
		V_S = 36 V		30	36	
$t_{STBY,min}$	Minimum standby time			10		µs
$t_{logicwu}$	Logic power-on and wake-up time			38	45	µs
t_{cpwu}	Charge pump power-on and wake-up time	Power bridges disabled, C_p = 10 nF, C_{boot} = 220 nF		650		µs
Internal voltage regulator						
V_{REG}	Voltage regulator output voltage		2.9	3	3.2	V
I_{REG}	Voltage regulator output current				40	mA

Table 5. Electrical characteristics (continued)

Symbol	Parameter	Test condition	Min.	Typ.	Max.	Unit
$V_{REG, drop}$	Voltage regulator output voltage drop	I_{REG} = 40 mA		50		mV
$I_{REG,STBY}$	Voltage regulator standby output current				10	mA
Integrated analog-to-digital converter						
N_{ADC}	Analog-to-digital converter resolution			5		bit
$V_{ADC,ref}$	Analog-to-digital converter reference voltage			V_{REG}		V
f_S	Analog-to-digital converter sampling frequency			f_{PWM}		kHz

1. Accuracy depends on oscillator frequency accuracy.
2. Tested at 25 °C in a restricted range and guaranteed by characterization.
3. Rise and fall time depends on motor supply voltage value. Refer to SR_{out} values in order to evaluate the actual rise and fall time.
4. Not valid for $\overline{\text{STBY/RST}}$ pin which has internal pull-down resistor.
5. Not valid for SW and CS pins which have internal pull-up resistors.
6. $\overline{\text{FLAG}}$, $\overline{\text{BUSY}}$ and SYNC open drain outputs included.
7. See *Figure 17 on page 38* – SPI timings diagram for details.

4 Pin connection

Figure 2. HTSSOP28 pin connection (top view)

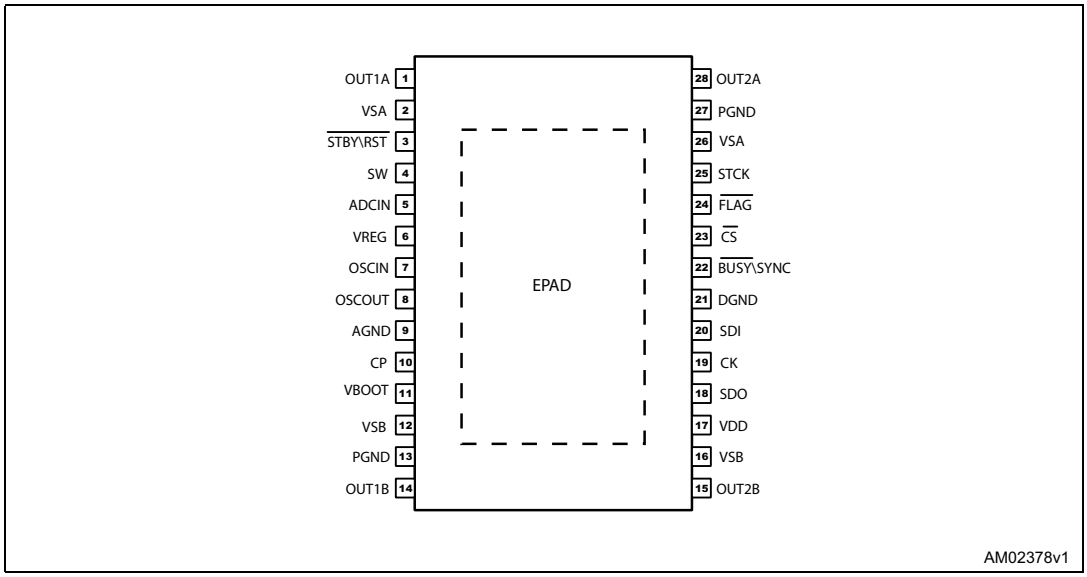
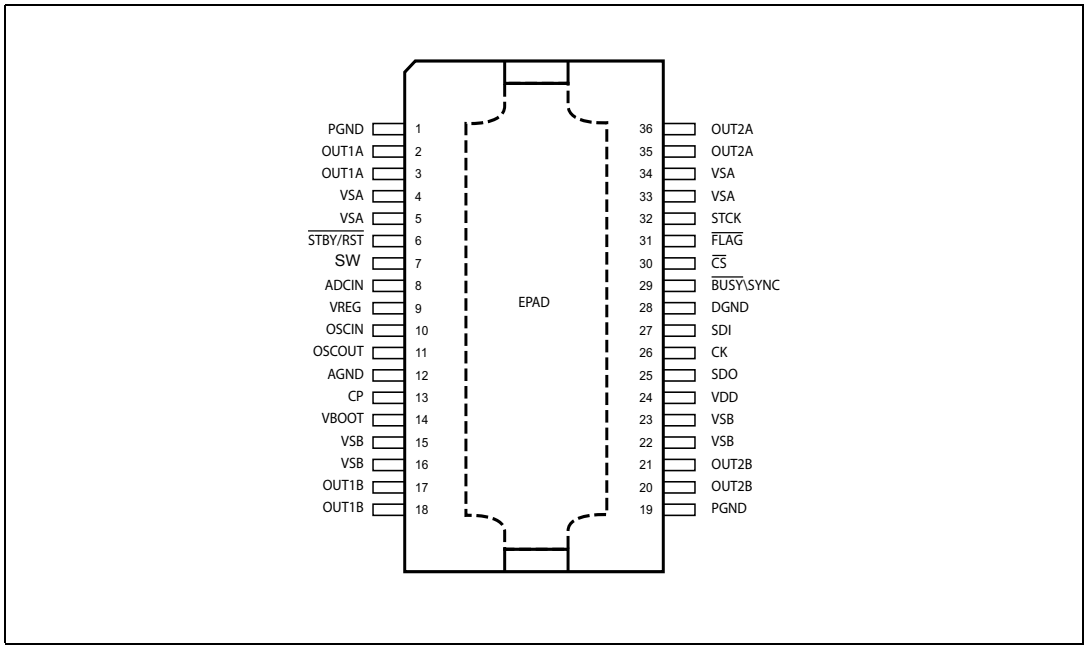


Figure 3. POWERSO36 pin connection (top view)



Pin list

Table 6. Pin description

No.		Name	Type	Function
HTSSOP	POWERSO			
17	24	VDD	Power	Logic outputs supply voltage (pull-up reference)
6	9	VREG	Power	Internal 3 V voltage regulator output and 3.3 V external logic supply
7	10	OSCIN	Analog input	Oscillator pin 1. To connect an external oscillator or clock source. If this pin is unused, it should be left floating.
8	11	OSCOUT	Analog output	Oscillator pin 2. To connect an external oscillator. When the internal oscillator is used this pin can supply 2/4/8/16 MHz. If this pin is unused, it should be left floating.
10	13	CP	Output	Charge pump oscillator output
11	14	VBOOT	Supply voltage	Bootstrap voltage needed for driving the high-side power DMOS of both bridges (A and B)
5	8	ADCIN	Analog input	Internal analog-to-digital converter input
2, 26	4, 5, 33, 34	VSA	Power supply	Full bridge A power supply pin. It must be connected to VSB.
12, 16	15, 16, 22, 23	VSB	Power supply	Full bridge B power supply pin. It must be connected to VSA.
27, 13	1, 19	PGND	Ground	Power ground pin
1	2, 3	OUT1A	Power output	Full bridge A output 1
28	35, 36	OUT2A	Power output	Full bridge A output 2
14	17, 18	OUT1B	Power output	Full bridge B output 1
15	20, 21	OUT2B	Power output	Full bridge B output 2
9	12	AGND	Ground	Analog ground.
4	7	SW	Logical input	External switch input pin. If not used the pin should be connected to VDD.
21	28	DGND	Ground	Digital ground
22	29	$\overline{\text{BUSY}}$\SYNC	Open drain output	By default, this BUSY pin is forced low when the device is performing a command. Otherwise the pin can be configured to generate a synchronization signal.
18	25	SDO	Logic output	Data output pin for serial interface
20	27	SDI	Logic input	Data input pin for serial interface
19	26	CK	Logic input	Serial interface clock
23	30	$\overline{\text{CS}}$	Logic input	Chip select input pin for serial interface

Table 6. Pin description (continued)

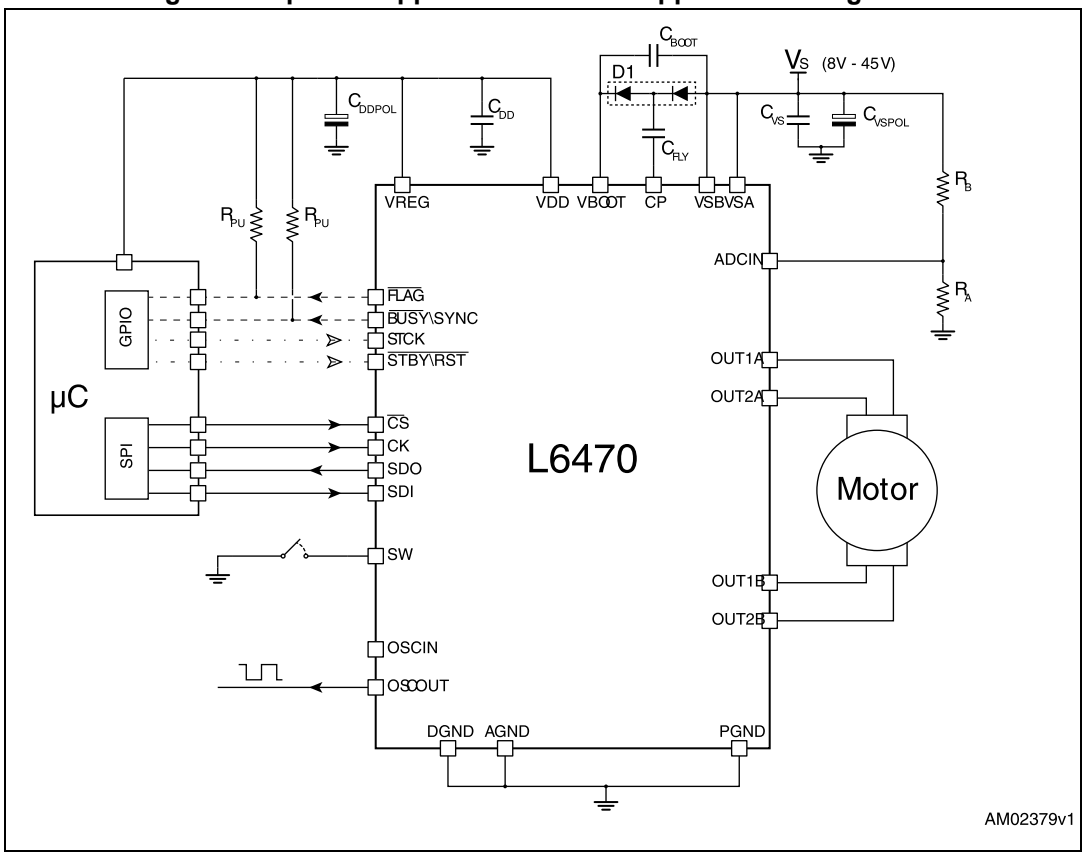
No.		Name	Type	Function
HTSSOP	POWERSO			
24	31	$\overline{\text{FLAG}}$	Open drain output	Status flag pin. An internal open drain transistor can pull the pin to GND when a programmed alarm condition occurs (step loss, OCD, thermal pre-warning or shutdown, UVLO, wrong command, non-performable command)
3	6	$\overline{\text{STBY\RST}}$	Logic input	Standby and reset pin. LOW logic level resets the logic and puts the device into Standby mode. If not used, it should be connected to VDD.
25	32	STCK	Logic input	Step-clock input
EPAD	EPAD	Exposed pad	Ground	Internally connected to PGND, AGND and DGND pins

5 Typical applications

Table 7. Typical application values

Name	Value
C_{VS}	220 nF
C_{VSPOL}	100 µF
C_{REG}	100 nF
C_{REGPOL}	47 µF
C_{DD}	100 nF
C_{DDPOL}	10 µF
D1	Charge pump diodes
C_{BOOT}	220 nF
C_{FLY}	10 nF
R_{PU}	39 kΩ
R_{SW}	100 Ω
C_{SW}	10 nF
R_A	2.7 kΩ (V_S = 36 V)
R_B	62 kΩ (V_S = 36 V)

Figure 4. Bipolar stepper motor control application using L6470



6 Functional description

6.1 Device power-up

At power-up end, the device state is the following:

- Registers are set to default
- Internal logic is driven by internal oscillator and a 2 MHz clock is provided by the OSCOUT pin
- Bridges are disabled (High Z)
- UVLO bit in the STATUS register is forced low (fail condition)
- FLAG output is forced low.

During power-up, the device is under reset (all logic IOs disabled and power bridges in high impedance state) until the following conditions are satisfied:

- V_S is greater than V_{SthOn}
- V_{REG} is greater than V_{REGth} = 2.8 V typical
- Internal oscillator is operative.

Any motion command makes the device exit from High Z state (HardStop and SoftStop included).

6.2 Logic I/O

Pins $\overline{CS}$, CK, SDI, STCK, SW and $\overline{STBY}$\$\overline{RST}$ are TTL/CMOS 3.3 V - 5 V compatible logic inputs.

Pin SDO is a TTL/CMOS compatible logic output. VDD pin voltage sets the logic output pin voltage range; when it is connected to VREG or 3.3 V external supply voltage, the output is 3.3 V compatible. When VDD is connected to a 5 V supply voltage, SDO is 5 V compatible.

VDD is not internally connected to V_{REG}, an external connection is always needed.

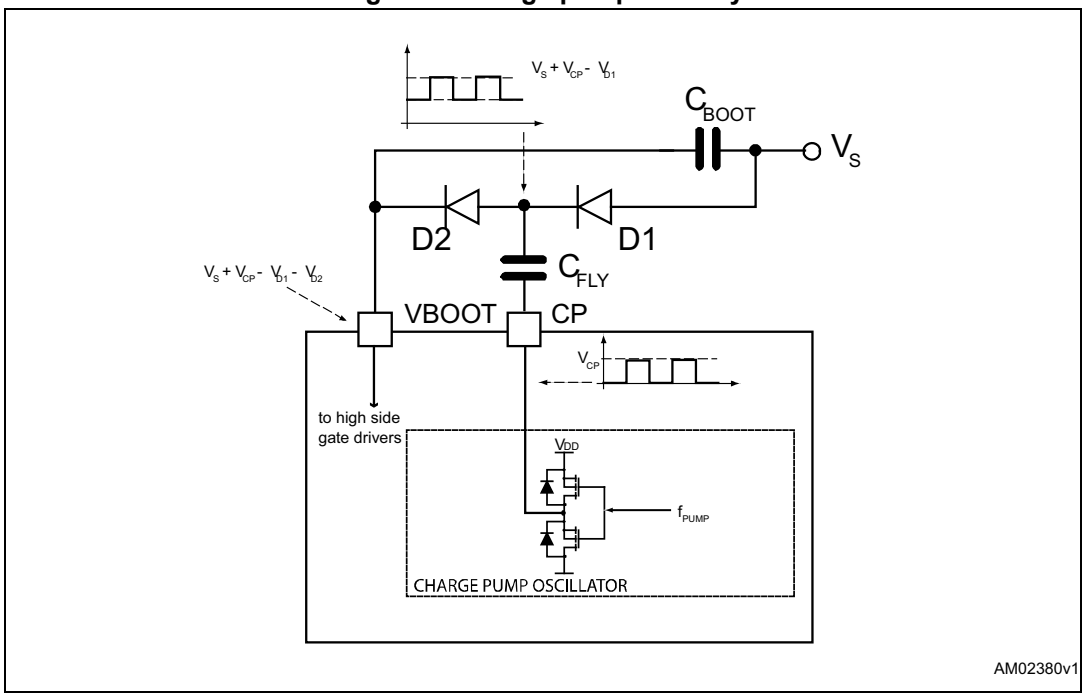
A 10 µF capacitor should be connected to the VDD pin in order to obtain a proper operation.

Pins $\overline{FLAG}$ and $\overline{BUSY}$\SYNC are open drain outputs.

6.3 Charge pump

To ensure the correct driving of the high-side integrated MOSFETs, a voltage higher than the motor power supply voltage needs to be applied to the VBOOT pin. The high-side gate driver supply voltage, V_{boot}, is obtained through an oscillator and a few external components realizing a charge pump (see *Figure 5*).

Figure 5. Charge pump circuitry

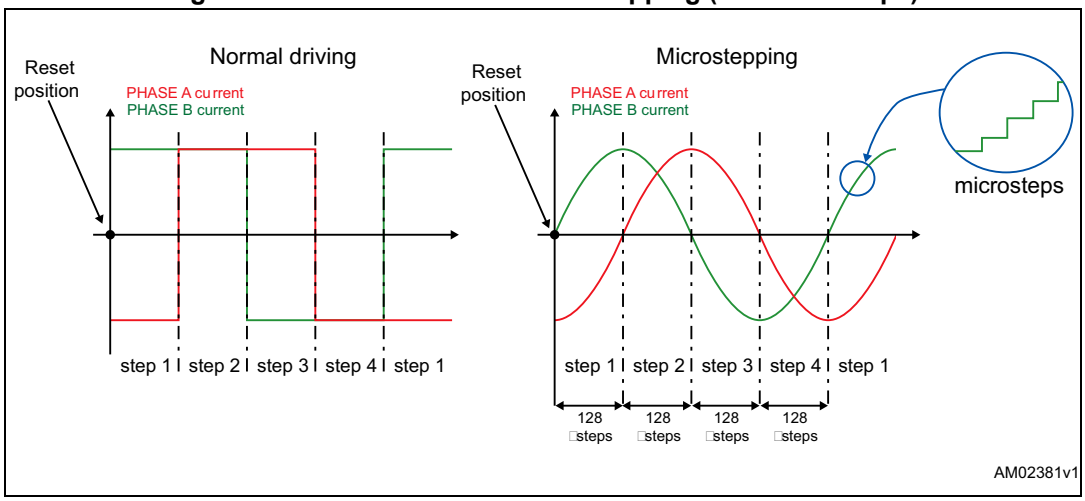


6.4 Microstepping

The driver is able to divide the single step into up to 128 microsteps. Stepping mode can be programmed by the STEP_SEL parameter in the STEP_MODE register (see *Table 18 on page 48*).

Step mode can only be changed when bridges are disabled. Every time the step mode is changed the electrical position (i.e. the point of microstepping sinewave that is generated) is reset to zero and the absolute position counter value (see *Section 6.5*) becomes meaningless.

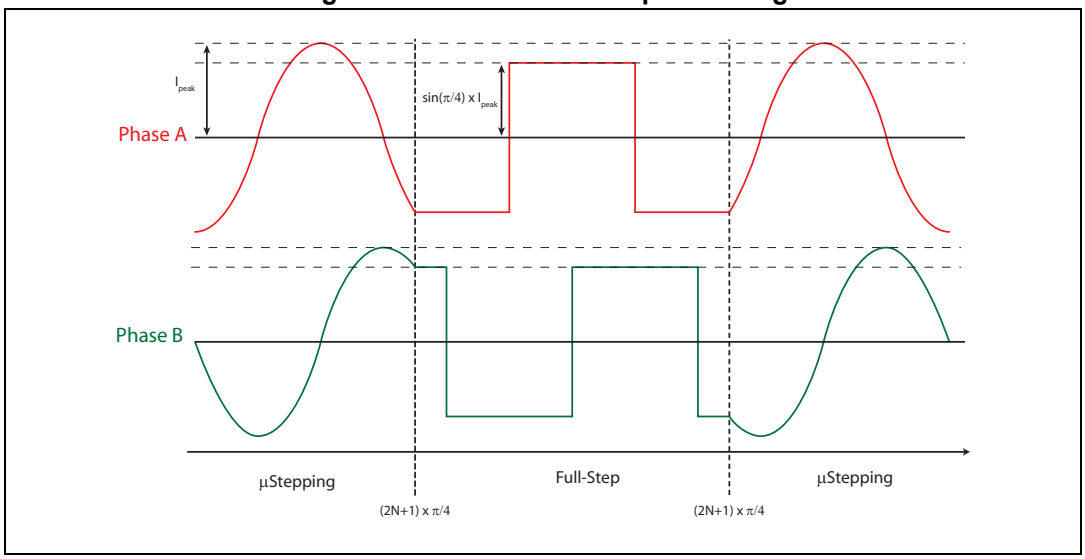
Figure 6. Normal mode and microstepping (128 microsteps)



Automatic full-step mode

When motor speed is greater than a programmable full-step speed threshold, the L6470 device switches automatically to Full-step mode (see *Figure 7*); the driving mode returns to microstepping when motor speed decreases below the full-step speed threshold. The full-step speed threshold is set through the FS_SPD register (see *Section 9.1.9 on page 44*).

Figure 7. Automatic full-step switching



6.5 Absolute position counter

An internal 22-bit register (ABS_POS) records the motor motion according to the selected step mode; the stored value unit is equal to the selected step mode (full, half, quarter, etc.). The position range is from -2^{21} to $+2^{21}-1$ (µ)steps (see *Section 9.1.1 on page 41*).

6.6 Programmable speed profiles

The user can easily program a customized speed profile defining independently acceleration, deceleration, maximum and minimum speed values by the ACC, DEC, MAX_SPEED and MIN_SPEED registers respectively (see *Section 9.1.5 on page 42*, *Section 9.1.6 on page 43*, *Section 9.1.7 on page 43* and *Section 9.1.8 on page 43*).

When a command is sent to the device, the integrated logic generates the microstep frequency profile that performs a motor motion compliant to speed profile boundaries.

All acceleration parameters are expressed in step/tick^2 and all speed parameters are expressed in step/tick; the unit of measurement does not depend on the selected step mode. Acceleration and deceleration parameters range from 2^{-40} to $(2^{12} - 2) \cdot 2^{-40}$ step/tick^2 (equivalent to 14.55 to 59590 step/s^2).

The minimum speed parameter ranges from 0 to $(2^{12}-1) \cdot 2^{-24}$ step/tick (equivalent to 0 to 976.3 step/s).

The maximum speed parameter ranges from 2^{-18} to $(2^{10} - 1) \cdot 2^{-18}$ step/tick (equivalent to 15.25 to 15610 step/s).

6.7 Motor control commands

The L6470 device can accept different types of commands:

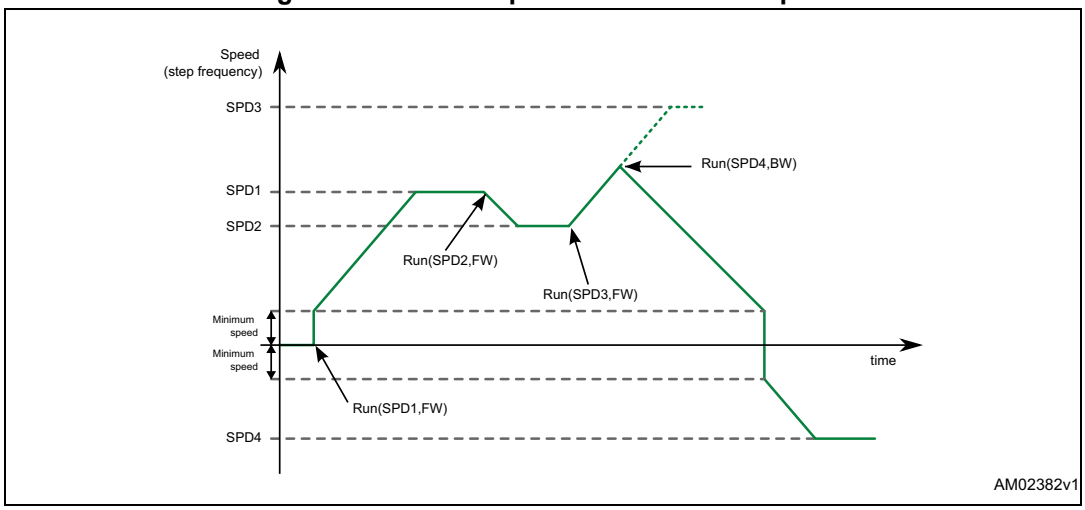
- constant speed commands (Run, GoUntil, ReleaseSW)
- absolute positioning commands (GoTo, GoTo_DIR, GoHome, GoMark)
- motion commands (Move)
- stop commands (SoftStop, HardStop, SoftHiz, HardHiz).

For detailed command descriptions refer to *Section 9.2 on page 56*.

6.7.1 Constant speed commands

A constant speed command produces a motion in order to reach and maintain a user-defined target speed starting from the programmed minimum speed (set in the MIN_SPEED register) and with the programmed acceleration/deceleration value (set in the ACC and DEC registers). A new constant speed command can be requested anytime.

Figure 8. Constant speed command examples



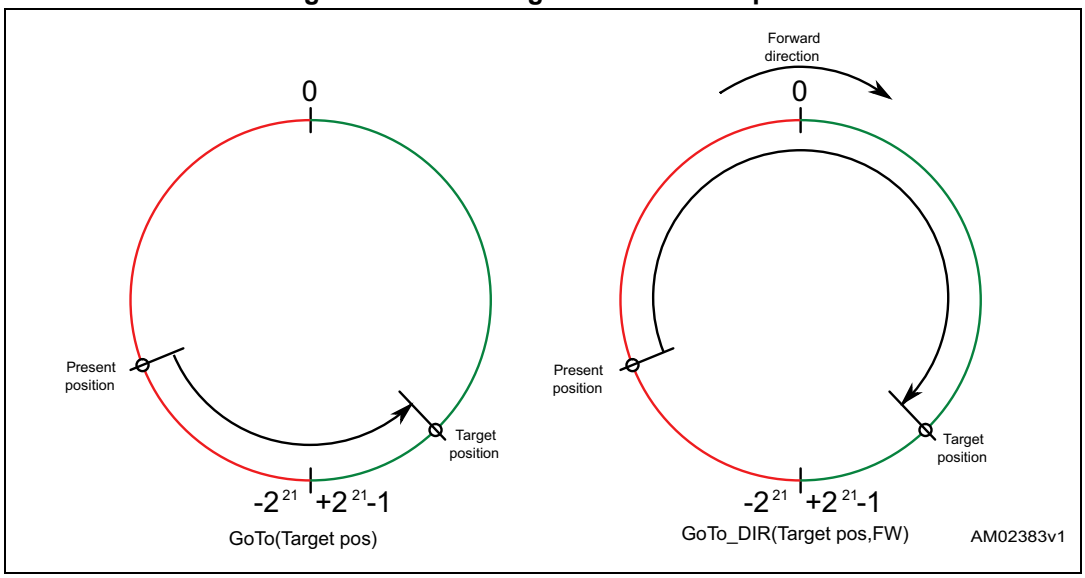
6.7.2 Positioning commands

An absolute positioning command produces a motion in order to reach a user-defined position that is sent to the device together with the command. The position can be reached performing the minimum path (minimum physical distance) or forcing a direction (see *Figure 9*).

The performed motor motion is compliant to programmed speed profile boundaries (acceleration, deceleration, minimum and maximum speed).

Note that with some speed profiles or positioning commands, the deceleration phase can start before the maximum speed is reached.

Figure 9. Positioning command examples



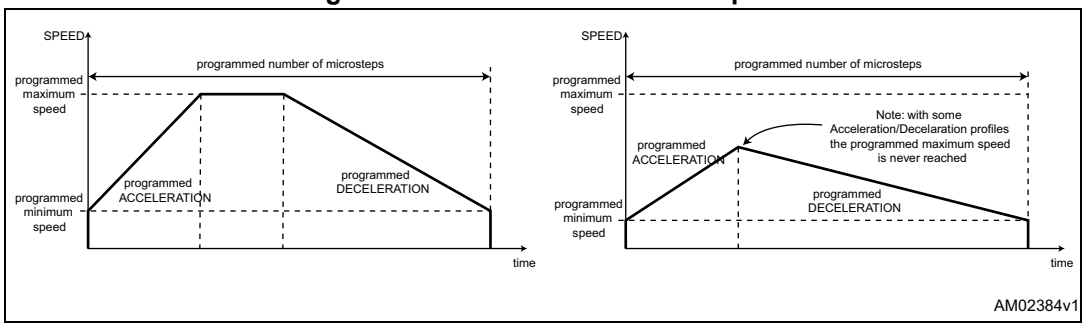
6.7.3 Motion commands

Motion commands produce a motion in order to perform a user-defined number of microsteps in a user-defined direction that are sent to the device together with the command (see *Figure 10*).

The performed motor motion is compliant to programmed speed profile boundaries (acceleration, deceleration, minimum and maximum speed).

Note that with some speed profiles or motion commands, the deceleration phase can start before the maximum speed is reached.

Figure 10. Motion command examples



6.7.4 Stop commands

A stop command forces the motor to stop. Stop commands can be sent anytime.

The SoftStop command causes the motor to decelerate with programmed deceleration value until the MIN_SPEED value is reached and then stops the motor keeping the rotor position (a holding torque is applied).

The HardStop command stops the motor instantly, ignoring deceleration constraints and keeping the rotor position (a holding torque is applied).

The SoftHiZ command causes the motor to decelerate with programmed deceleration value until the MIN_SPEED value is reached and then forces the bridges in high impedance state (no holding torque is present).

The HardHiZ command instantly forces the bridges into high impedance state (no holding torque is present).

6.7.5 Step-clock mode

In Step-clock mode the motor motion is defined by the step-clock signal applied to the STCK pin. At each step-clock rising edge, the motor is moved one microstep in the programmed direction and the absolute position is consequently updated.

When the system is in Step-clock mode, the SCK_MOD flag in the STATUS register is raised, the SPEED register is set to zero and motor status is considered stopped whatever the STCK signal frequency (MOT_STATUS parameter in STATUS register equal to “00”).

6.7.6 GoUntil and ReleaseSW commands

In most applications the power-up position of the stepper motor is undefined, so an initialization algorithm driving the motor to a known position is necessary.

The GoUntil and ReleaseSW commands can be used in combination with external switch input (see *Section 6.13 on page 31*) to easily initialize the motor position.

The GoUntil command makes the motor run at the constant target speed until the SW input is forced low (falling edge). When this event occurs, one of the following actions can be performed:

- The ABS_POS register is set to zero (home position) and the motor decelerates to zero speed (as a SoftStop command)
- The ABS_POS register value is stored in the MARK register and the motor decelerates to zero speed (as a SoftStop command).

If the SW_MODE bit of the CONFIG register is set to ‘0’, the motor does not decelerate but it immediately stops (as a HardStop command).

The ReleaseSW command makes the motor run at the programmed minimum speed until the SW input is forced high (rising edge). When this event occurs, one of the following actions can be performed:

- The ABS_POS register is set to zero (home position) and the motor immediately stops (as a HardStop command)
- The ABS_POS register value is stored in the MARK register and the motor immediately stops (as a HardStop command).

If the programmed minimum speed is less than 5 step/s, the motor is driven at 5 step/s.

6.8 Internal oscillator and oscillator driver

The control logic clock can be supplied by the internal 16-MHz oscillator, an external oscillator (crystal or ceramic resonator) or a direct clock signal.

These working modes can be selected by EXT_CLK and OSC_SEL parameters in the CONFIG register (see *Table 23 on page 50*).

At power-up the device starts using the internal oscillator and provides a 2-MHz clock signal on the OSCOUT pin.

Attention: In any case, before changing clock source configuration, a hardware reset is mandatory. Switching to different clock configurations during operation may cause unexpected behavior.

6.8.1 Internal oscillator

In this mode the internal oscillator is activated and OSCIN is unused. If the OSCOUT clock source is enabled, the OSCOUT pin provides a 2, 4, 8 or 16-MHz clock signal (according to OSC_SEL value); it is otherwise unused (see *Figure 11*).

6.8.2 External clock source

Two types of external clock source can be selected: crystal/ceramic resonator or direct clock source. Four programmable clock frequencies are available for each external clock source: 8, 16, 24 and 32 MHz.

When an external crystal/resonator is selected, the OSCIN and OSCOUT pins are used to drive the crystal/resonator (see *Figure 11*). The crystal/resonator and load capacitors (C_L) must be placed as close as possible to the pins. Refer to *Table 8* for the choice of load capacitor values according to the external oscillator frequency.

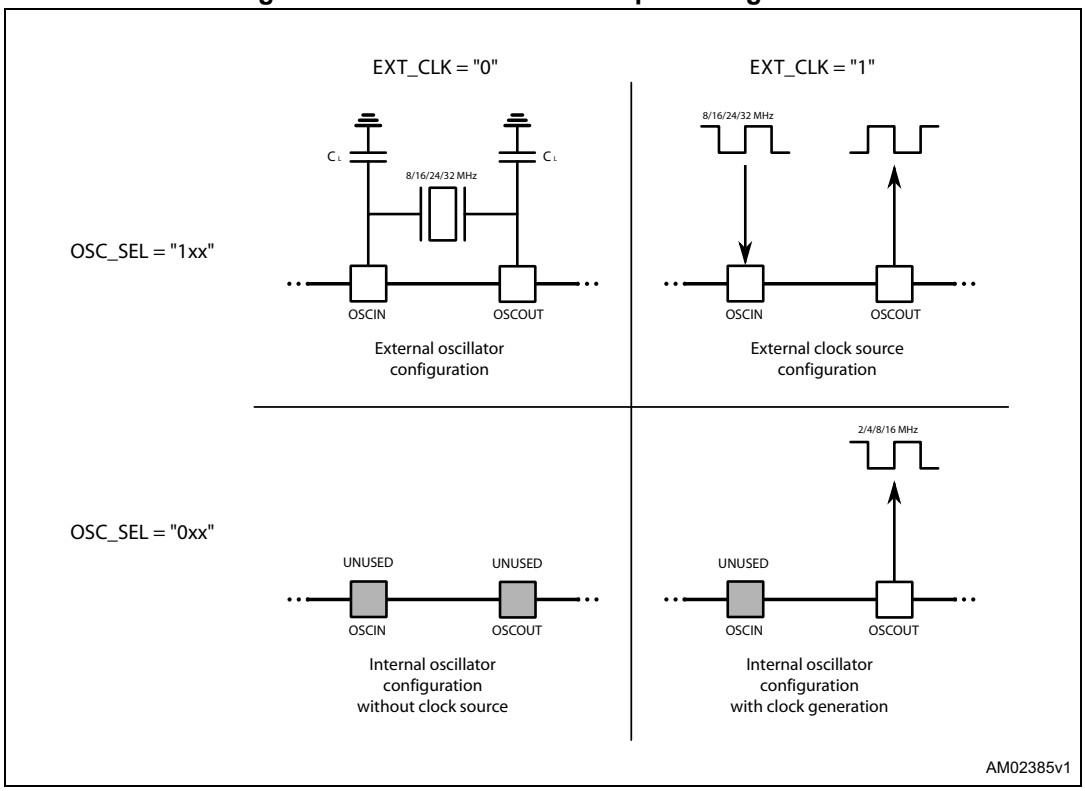
Table 8. CL values according to external oscillator frequency

Crystal/resonator freq. (1)	C_L (2)
8 MHz	25 pF (ESR_{max} = 80 Ω)
16 MHz	18 pF (ESR_{max} = 50 Ω)
24 MHz	15 pF (ESR_{max} = 40 Ω)
32 MHz	10 pF (ESR_{max} = 40 Ω)

1. First harmonic resonance frequency.
2. Lower ESR value allows the driving of greater load capacitors.

If a direct clock source is used, it must be connected to the OSCIN pin and the OSCOUT pin supplies the inverted OSCIN signal (see *Figure 11*).

Figure 11. OSCIN and OSCOUT pin configuration



Note: When OSCIN is UNUSED, it should be left floating.

When OSCOUT is UNUSED, it should be left floating.

6.9 Overcurrent detection

When the current in any of the Power MOSFETs exceeds a programmed overcurrent threshold, the STATUS register OCD flag is forced low until the overcurrent event has expired and a GetStatus command is sent to the IC (see *Section 9.1.22 on page 55* and *Section 9.1.17 on page 47*). The overcurrent event expires when all the Power MOSFET currents fall below the programmed overcurrent threshold.

The overcurrent threshold can be programmed through the OCD_TH register in one of 16 available values ranging from 375 mA to 6 A with steps of 375 mA (see *Table 9 on page 40*, *Section 9.1.17*).

It is possible to set whether an overcurrent event causes or not the MOSFET turn-off (bridges in high impedance status) acting on the OC_SD bit in the CONFIG register (see *Section 9.1.21*). The OCD flag in the STATUS register is raised anyway (see *Table 34 on page 55*, *Section 9.1.22*).

When the IC outputs are turned off by an OCD event, they cannot be turned on until the OCD flag is released by a GetStatus command.

Attention: **The overcurrent shutdown is a critical protection feature. It is not recommended to disable it.**

6.10 Undervoltage lockout (UVLO)

The L6470 device provides a motor supply UVLO protection. When the motor supply voltage falls below the V_{SthOff} threshold voltage, the STATUS register UVLO flag is forced low. When a GetStatus command is sent to the IC, and the undervoltage condition has expired, the UVLO flag is released (see *Section 9.1.22 on page 55* and *9.2.20 on page 66*). The undervoltage condition expires when the motor supply voltage goes over the V_{SthOn} threshold voltage. When the device is in undervoltage condition, no motion command can be performed. The UVLO flag is forced low by logic reset (power-up included) even if no UVLO condition is present.

6.11 Thermal warning and thermal shutdown

An internal sensor allows the L6470 to detect when the device internal temperature exceeds a thermal warning or an overtemperature threshold.

When the thermal warning threshold ($T_{j(WRN)}$) is reached, the TH_WRN bit in the STATUS register is forced low (see *Section 9.1.22*) until the temperature decreases below $T_{j(WRN)}$ and a GetStatus command is sent to the IC (see *Section 9.1.22* and *9.2.20*).

When the thermal shutdown threshold ($T_{j(OFF)}$) is reached, the device goes into thermal shutdown condition: the TH_SD bit in the STATUS register is forced low, the power bridges are disabled bridges in high impedance state and the HiZ bit in the STATUS register is raised (see *Section 9.1.22*).

The thermal shutdown condition only expires when the temperature goes below the thermal warning threshold ($T_{j(WRN)}$).

On exiting thermal shutdown condition, the bridges are still disabled (HiZ flag high); any motion command makes the device exit from High Z state (HardStop and SoftStop included).

6.12 Reset and standby

The device can be reset and put into Standby mode through a dedicated pin. When the STBY\RST pin is driven low, the bridges are left open (High Z state), the internal charge pump is stopped, the SPI interface and control logic are disabled and the internal 3 V voltage regulator maximum output current is reduced to $I_{REG,STBY}$; as a result, the L6470 heavily reduces the power consumption. At the same time the register values are reset to default and all protection functions are disabled. STBY\RST input must be forced low at least for $t_{STBY,min}$ in order to ensure the complete switch to Standby mode.

On exiting Standby mode, as well as for IC power-up, a delay of up to $t_{logicwu}$ must be given before applying a new command to allow proper oscillator and logic startup and a delay of up to t_{cpwu} must be given to allow the charge pump startup.

On exiting Standby mode, the bridges are disabled (HiZ flag high) and any motion command makes the device exit High Z state (HardStop and SoftStop included).

Attention: It is not recommended to reset the device when outputs are active. The device should be switched to high impedance state before being reset.

6.13 External switch (SW pin)

The SW input is internally pulled up to V_{DD} and detects if the pin is open or connected to ground (see *Figure 12*).

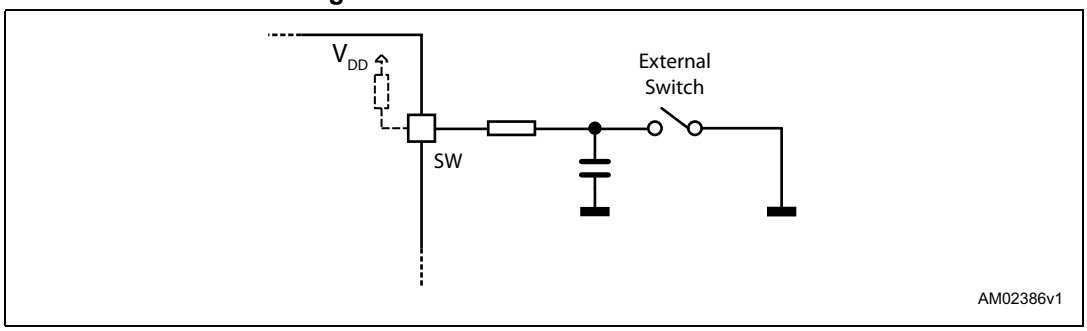
The SW_F bit of the STATUS register indicates if the switch is open (‘0’) or closed (‘1’) (see *Section 9.1.22 on page 55*); the bit value is refreshed at every system clock cycle (125 ns). The SW_EVN flag of the STATUS register is raised when a switch turn-on event (SW input falling edge) is detected (see *Section 9.1.22*). A GetStatus command releases the SW_EVN flag (see *Section 9.2.20 on page 66*).

By default, a switch turn-on event causes a HardStop interrupt (SW_MODE bit of CONFIG register set to ‘0’). Otherwise (SW_MODE bit of CONFIG register set to ‘1’), switch input events do not cause interrupts and the switch status information is at the user’s disposal (see *Table 34 on page 55*, *Section 9.1.22*).

The switch input may be used by the GoUntil and ReleaseSW commands as described in *Section 9.2.10 on page 62* and *9.2.11 on page 63*.

If the SW input is not used, it should be connected to VDD.

Figure 12. External switch connection



6.14 Programmable DMOS slew rate, deadtime and blanking time

Using the POW_SR parameter in the CONFIG register, it is possible to set the commutation speed of the power bridges output (see *Table 26 on page 51*, *Section 9.1.21 on page 49*).

6.15 Integrated analog-to-digital converter

The L6470 device integrates an N_{ADC} bit ramp-compare analog-to-digital converter with a reference voltage equal to VREG. The analog-to-digital converter input is available through the ADCIN pin and the conversion result is available in the ADC_OUT register (see *Section 9.1.16 on page 46*).

Sampling frequency is equal to the programmed PWM frequency.

The ADC_OUT value can be used for motor supply voltage compensation or can be at the user's disposal.

6.16 Internal voltage regulator

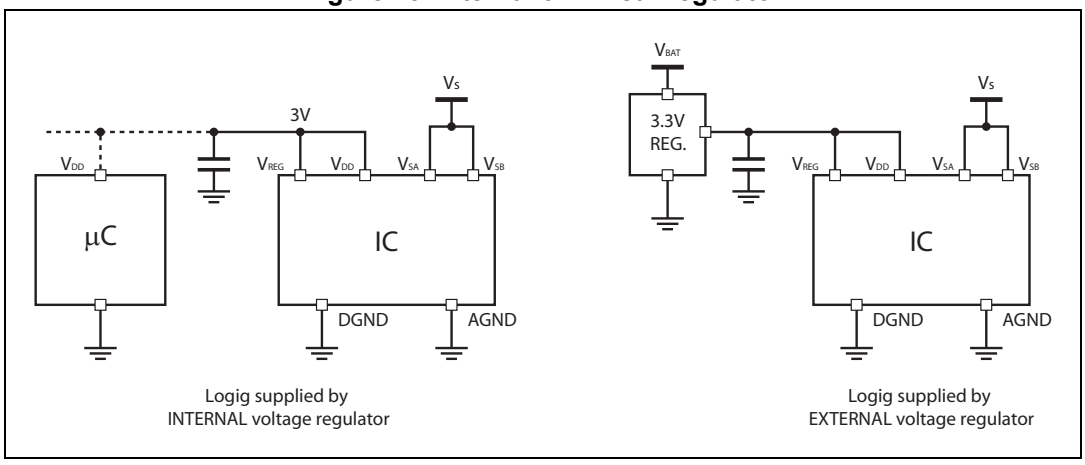
The L6470 integrates a voltage regulator which generates a 3 V voltage starting from the motor power supply (VSA and VSB). In order to make the voltage regulator stable, at least 22 µF should be connected between the VREG pin and ground (suggested value is 47 µF).

The internal voltage regulator can be used to supply the VDD pin in order to make the device digital output range 3.3 V compatible (*Figure 13*). A digital output range, 5 V compatible, may be obtained connecting the VDD pin to an external 5 V voltage source. In both cases, a 10 µF capacitance should be connected to the VDD pin in order to obtain a correct operation.

The internal voltage regulator is able to supply a current up to $I_{REG,MAX}$, internal logic consumption included (I_{logic}). When the device is in Standby mode, the maximum current that can be supplied is $I_{REG, STBY}$, internal consumption included ($I_{logic, STBY}$).

If an external 3.3 V regulated voltage is available, it can be applied to the VREG pin in order to supply all the internal logic and to avoid power dissipation of the internal 3 V voltage regulator (*Figure 13*). The external voltage regulator should never sink current from the VREG pin.

Figure 13. Internal 3 V linear regulator



6.17 BUSY\SYNC pin

This pin is an open drain output which can be used as the busy flag or synchronization signal according to the SYNC_EN bit value (STEP_MODE register).

6.17.1 BUSY operation mode

The pin works as busy signal when the SYNC_EN bit is set low (default condition). In this mode the output is forced low while a constant speed, absolute positioning or motion command is under execution. The BUSY pin is released when the command has been executed (target speed or target position reached). The STATUS register includes a BUSY flag that is the BUSY pin mirror (see *Section 9.1.22 on page 55*).

In the case of daisy chain configuration, BUSY pins of different ICs can be hard-wired to save host controller GPIOs.

6.17.2 SYNC operation mode

The pin works as synchronization signal when the SYNC_EN bit is set high. In this mode a step-clock signal is provided on the output according to a SYNC_SEL and STEP_SEL parameter combination (see *Section 9.1.19 on page 47*).

6.18 FLAG pin

By default, an internal open drain transistor pulls the FLAG pin to ground when at least one of the following conditions occurs:

- Power-up or standby/reset exit
- Stall detection on A bridge
- Stall detection on B bridge
- Overcurrent detection
- Thermal warning
- Thermal shutdown
- UVLO
- Switch turn-on event
- Wrong command
- Non-performable command.

It is possible to mask one or more alarm conditions by programming the ALARM_EN register (see *Section 9.1.20 on page 49*, *Table 21 on page 49*). If the corresponding bit of the ALARM_EN register is low, the alarm condition is masked and it does not cause a FLAG pin transition; all other actions imposed by alarm conditions are performed anyway. In the case of daisy chain configuration, FLAG pins of different ICs can be or-wired to save host controller GPIOs.

7 Phase current control

The L6470 controls the phase current applying a sinusoidal voltage to motor windings. Phase current amplitude is not directly controlled but depends on phase voltage amplitude, load torque, motor electrical characteristics and rotation speed. Sinewave amplitude is proportional to the motor supply voltage multiplied by a coefficient (K_{VAL}). K_{VAL} ranges from 0 to 100% and the sinewave amplitude can be obtained through the following formula:

Equation 1

$$V_{OUT} = V_S \cdot K_{VAL}$$

Different K_{VAL} values can be programmed for acceleration, deceleration and constant speed phases and when the motor is stopped (HOLD phase) through the KVAL_ACC, KVAL_DEC, KVAL_RUN and KVAL_HOLD registers (see *Section 9.1.10 on page 44*). KVAL value is calculated according to the following formula:

Equation 2

$$K_{VAL} = [(K_{VAL\_X} + BEMF\_COMP) \times VSCOMP \times K\_THERM] \times microstep$$

where K_{VAL_X} is the starting K_{VAL} value programmed for present motion phase (KVAL_ACC, KVAL_DEC, KVAL_RUN or KVAL_HOLD), *BEMF_COMP* is the BEMF compensation curve value, *VSCOMP* and *K_THERM* are the motor supply voltage and winding resistance compensation factors and microstep is the current microstep value (fraction of target peak current).

The L6470 device offers various methods to guarantee a stable current value, allowing the compensation of:

- low speed optimization (*Section 7.3*)
- back electromotive force value (*Section 7.4*)
- motor supply voltage variation (*Section 7.5*)
- windings resistance variation (*Section 7.6*).

7.1 PWM sinewave generators

The two voltage sinewaves applied to the stepper motor phases are generated by two PWM modulators.

The PWM frequency (f_{PWM}) is proportional to the oscillator frequency (f_{OSC}) and can be obtained through the following formula:

Equation 3

$$f_{PWM} = \frac{f_{OSC}}{512 \cdot N} \cdot m$$

'*N*' is the integer division factor and '*m*' is the multiplication factor. 'N' and 'm' values can be programmed by the F_PWM_INT and F_PWM_DEC parameters in the CONFIG register (see *Table 28 on page 52* and *Table 29 on page 52*, *Section 9.1.21 on page 49*).

Available PWM frequencies are listed in *Section 9.1.21* from *Table 30 on page 53* to *Table 33 on page 54*.

7.2 Sensorless stall detection

Depending on motor speed and load angle characteristics, the L6470 offers a motor stall condition detection using a programmable current comparator.

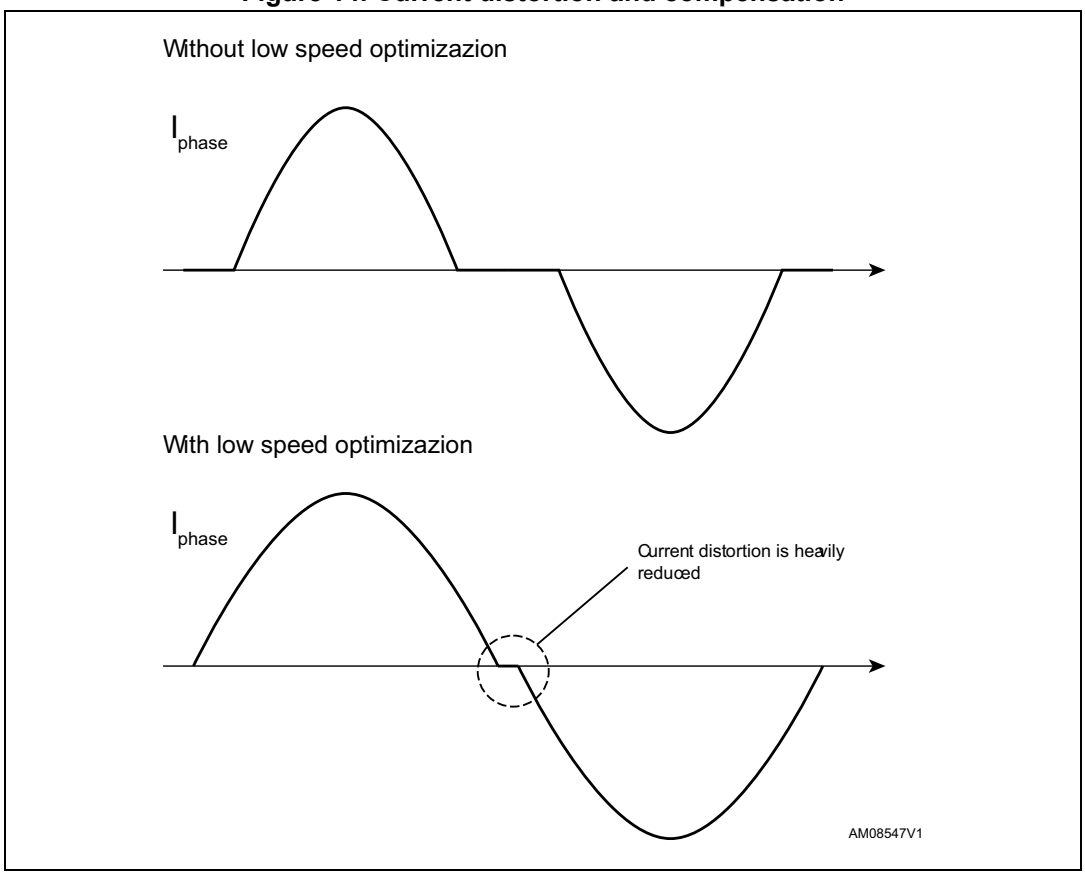
When a stall event occurs, the respective flag (STEP_LOSS_A or STEP_LOSS_B) is forced low until a GetStaus command or a system reset occurs (see *Section 9.2.20 on page 66*).

7.3 Low speed optimization

When the motor is driven at a very low speed using a small driving voltage, the resulting phase current can be distorted. As a consequence, the motor position is different from the ideal one (see *Figure 14*).

The L6470 device implements a low speed optimization in order to remove this effect.

Figure 14. Current distortion and compensation



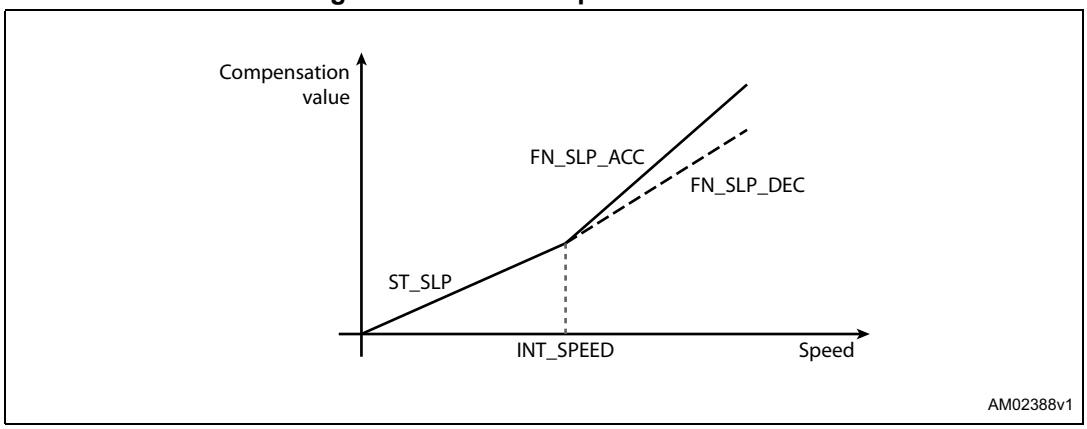
The optimization can be enabled setting high the LSPD_OPT bit in the MIN_SPEED register (see *Section 9.1.8 on page 43*) and is active in a speed range from zero to MIN_SPEED. When low speed optimization is enabled, speed profile minimum speed is forced to zero.

7.4 BEMF compensation

Using the speed information, a compensation curve is added to the amplitude of the voltage waveform applied to the motor winding in order to compensate the BEMF variations during acceleration and deceleration (see *Figure 15*).

The compensation curve is approximated by a stacked line with a starting slope (ST_SLP) when speed is lower than a programmable threshold speed (INT_SPEED) and a fine slope (FN_SLP_ACC and FN_SLP_DEC) when speed is greater than the threshold speed (see *Section 9.1.11 on page 45*, *9.1.12 on page 45*, *9.1.13 on page 45* and *Section 9.1.14 on page 45*).

Figure 15. BEMF compensation curve



To obtain different current values during acceleration and deceleration phases, two different final slope values, and consequently two different compensation curves, can be programmed.

The acceleration compensation curve is applied when the motor runs. No BEMF compensation is applied when the motor is stopped.

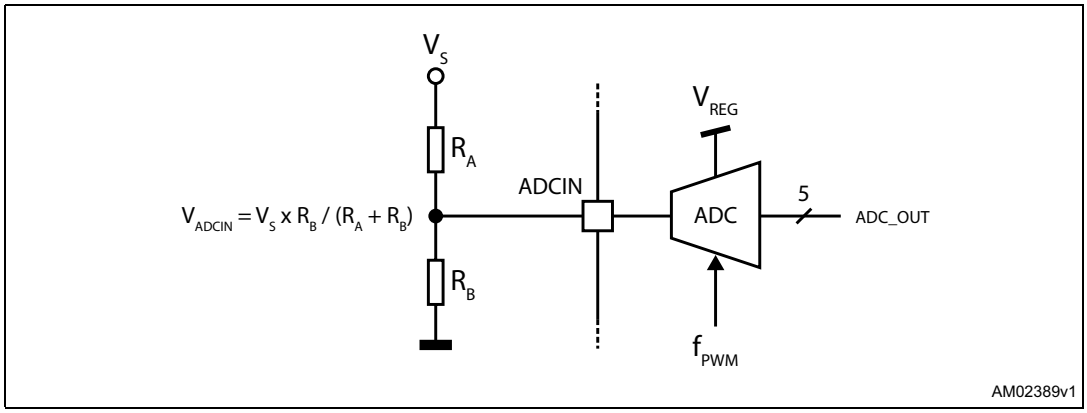
7.5 Motor supply voltage compensation

The sinewave amplitude generated by the PWM modulators is directly proportional to the motor supply voltage (V_S). When the motor supply voltage is different from its nominal value, the motor phases are driven with an incorrect voltage. The L6470 device can compensate motor supply voltage variations in order to avoid this effect.

The motor supply voltage should be connected to the integrated ADC input through a resistor divider in order to obtain $V_{REG}/2$ voltage at the ADCIN pin when V_S is at its nominal value (see *Figure 16*).

The ADC input is sampled at f_S frequency, which is equal to PWM frequency.

Figure 16. Motor supply voltage compensation circuit



Motor supply voltage compensation can be enabled setting high the EN_VSCOMP bit of the CONFIG register (see *Table 22 on page 49*, *Section 9.1.21 on page 49*). If the EN_VSCOMP bit is low, the compensation is disabled and the internal analog-to-digital converter is at the user's disposal; sampling rate is always equal to PWM frequency.

7.6 Winding resistance thermal drift compensation

The higher the winding resistance, the greater the voltage to be applied in order to obtain the same phase current.

The L6470 integrates a register (K_THERM) which can be used to compensate phase resistance increment due to temperature rising.

The value in the K_THERM register (see *Section 9.1.15 on page 46*) multiplies the duty cycle value allowing a higher phase resistance value to be faced.

The compensation algorithm and the eventual motor temperature measurement should be implemented by microcontroller firmware.

8 Serial interface

The integrated 8-bit serial peripheral interface (SPI) is used for a synchronous serial communication between the host microprocessor (always master) and the L6470 (always slave).

The SPI uses chip select ($\overline{CS}$), serial clock (CK), serial data input (SDI) and serial data output (SDO) pins. When $\overline{CS}$ is high, the device is unselected and the SDO line is inactive (high-impedance).

The communication starts when $\overline{CS}$ is forced low. The CK line is used for synchronization of data communication.

All commands and data bytes are shifted into the device through the SDI input, most significant bit first. The SDI is sampled on the rising edges of the CK.

All output data bytes are shifted out of the device through the SDO output, most significant bit first. The SDO is latched on the falling edges of the CK. When a return value from the device is not available, an all zero byte is sent.

After each byte transmission the $\overline{CS}$ input must be raised and be kept high for at least t_{disCS} in order to allow the device to decode the received command and put the return value into the SHIFT register.

All timing requirements are shown in *Figure 17* (see *Section 3: Electrical characteristics on page 11* for values).

Multiple devices can be connected in daisy chain configuration, as shown in *Figure 18*.

Figure 17. SPI timings diagram

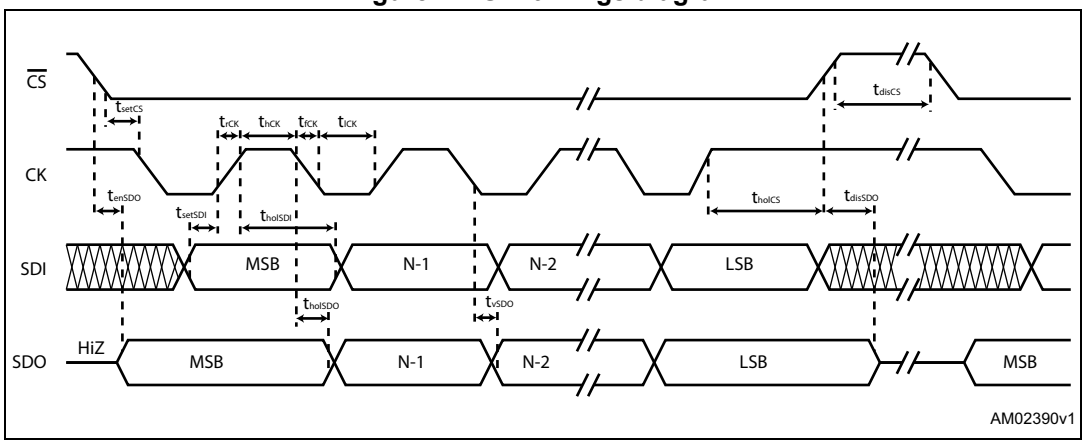
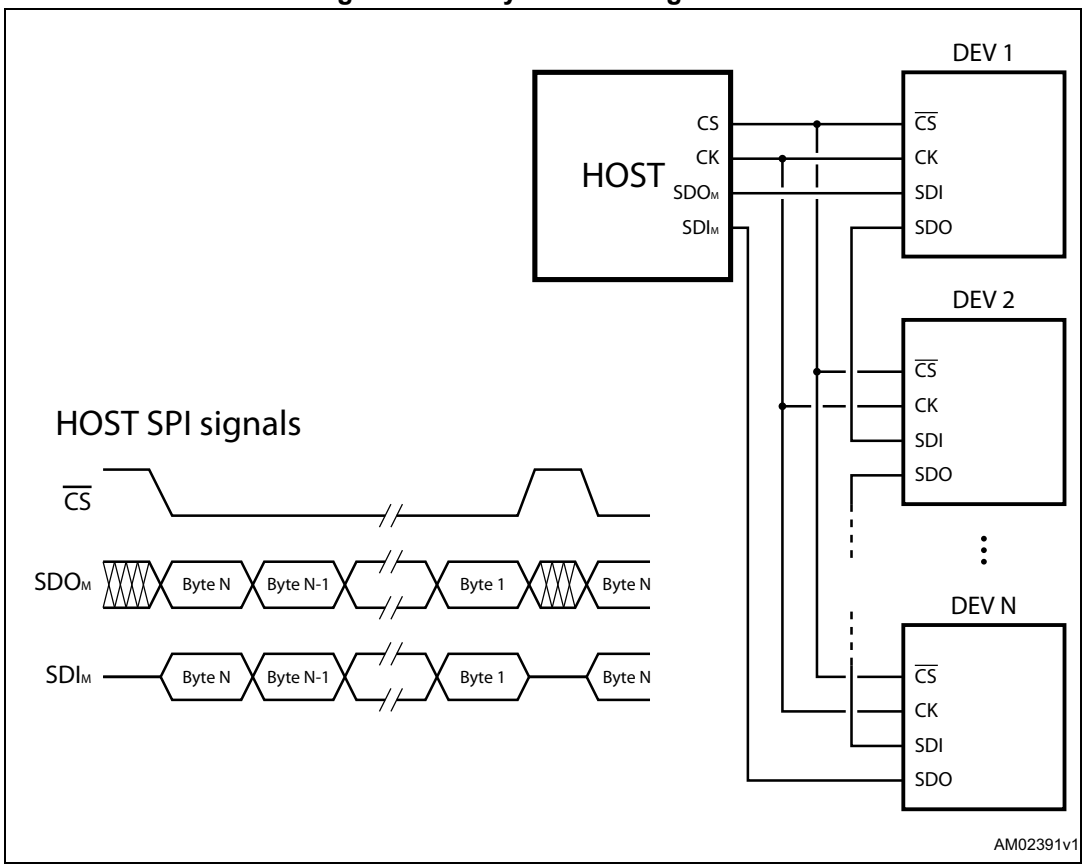


Figure 18. Daisy chain configuration



9 Programming manual

9.1 Registers and flags description

Table 9 is a map of the user registers available (detailed description in respective paragraphs):

Table 9. Register map

Address [Hex]	Register name	Register function	Len. [bit]	Reset Hex	Reset value	Remarks(1)
h01	ABS_POS	Current position	22	000000	0	R, WS
h02	EL_POS	Electrical position	9	000	0	R, WS
h03	MARK	Mark position	22	000000	0	R, WR
h04	SPEED	Current speed	20	00000	0 step/tick (0 step/s)	R
h05	ACC	Acceleration	12	08A	125.5e-12 step/tick^2 (2008 step/s^2)	R, WS
h06	DEC	Deceleration	12	08A	125.5e-12 step/tick^2 (2008 step/s^2)	R, WS
h07	MAX_SPEED	Maximum speed	10	041	248e-6 step/tick (991.8 step/s)	R, WR
h08	MIN_SPEED	Minimum speed	13	000	0 step/tick (0 step/s)	R, WS
h15	FS_SPD	Full-step speed	10	027	150.7e-6 step/tick (602.7 step/s)	R, WR
h09	KVAL_HOLD	Holding K_{VAL}	8	29	0.16·VS	R, WR
h0A	KVAL_RUN	Constant speed K_{VAL}	8	29	0.16·VS	R, WR
h0B	KVAL_ACC	Acceleration starting K_{VAL}	8	29	0.16·VS	R, WR
h0C	KVAL_DEC	Deceleration starting K_{VAL}	8	29	0.16·VS	R, WR
h0D	INT_SPEED	Intersect speed	14	0408	15.4e-6 step/tick (61.5 step/s)	R, WH
h0E	ST_SLP	Start slope	8	19	0.038% s/step	R, WH
h0F	FN_SLP_ACC	Acceleration final slope	8	29	0.063% s/step	R, WH
h10	FN_SLP_DEC	Deceleration final slope	8	29	0.063% s/step	R, WH
h11	K_THERM	Thermal compensation factor	4	0	1.0	R, WR
h12	ADC_OUT	ADC output	5	XX(2)		R
h13	OCD_TH	OCD threshold	4	8	3.38A	R, WR
h14	STALL_TH	STALL threshold	7	40	2.03A	R, WR
h16	STEP_MODE	Step mode	8	7	128 microsteps	R, WH
h17	ALARM_EN	Alarm enable	8	FF	All alarms enabled	R, WS

Table 9. Register map (continued)

Address [Hex]	Register name	Register function	Len. [bit]	Reset Hex	Reset value	Remarks(1)
h18	CONFIG	IC configuration	16	2E88	Internal oscillator, 2 MHz OSCOUT clock, supply voltage compensation disabled, overcurrent shutdown enabled, slew rate = 290 V/µs PWM frequency = 15.6 kHz.	R, WH
h19	STATUS	Status	16	XXXX(2)	High impedance state, UVLO/Reset flag set.	R
h1A	RESERVED	Reserved address				
h1B	RESERVED	Reserved address				

1. R: readable, WH: writable only when outputs are in high impedance, WS: writable only when motor is stopped, WR: always writable.
2. According to startup conditions.

9.1.1 ABS_POS

The ABS_POS register contains the current motor absolute position in agreement with the selected step mode; the stored value unit is equal to the selected step mode (full, half, quarter, etc.). The value is in 2's complement format and it ranges from -2^{21} to $+2^{21}-1$.

At power-on the register is initialized to “0” (HOME position).

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.1.2 EL_POS

The EL_POS register contains the current electrical position of the motor. The two MSbits indicate the current step and the other bits indicate the current microstep (expressed in step/128) within the step.

Table 10. EL_POS register

Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
STEP		MICROSTEP						

When the EL_POS register is written by the user, the new electrical position is instantly imposed. When the EL_POS register is written, its value must be masked in order to match with the step mode selected in the STEP_MODE register in order to avoid a wrong microstep value generation (see *Section 9.1.19 on page 47*); otherwise the resulting microstep sequence is incorrect.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.1.3 MARK

The MARK register contains an absolute position called MARK according to the selected step mode; the stored value unit is equal to the selected step mode (full, half, quarter, etc.). It is in 2's complement format and it ranges from -2^{21} to $+2^{21}-1$.

9.1.4 SPEED

The SPEED register contains the current motor speed, expressed in step/tick (format unsigned fixed point 0.28).

In order to convert the SPEED value in step/s, the following formula can be used:

Equation 4

$$[\text{step/s}] = \frac{\text{SPEED} \cdot 2^{-28}}{\text{tick}}$$

where *SPEED* is the integer number stored in the register and tick is 250 ns.

The available range is from 0 to 15625 step/s with a resolution of 0.015 step/s.

Note: The range effectively available to the user is limited by the MAX_SPEED parameter.

Any attempt to write the register causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.1.5 ACC

The ACC register contains the speed profile acceleration expressed in step/tick^2 (format unsigned fixed point 0.40).

In order to convert ACC value in step/s^2, the following formula can be used:

Equation 5

$$[\text{step/s}^2] = \frac{\text{ACC} \cdot 2^{-40}}{\text{tick}^2}$$

where *ACC* is the integer number stored in the register and tick is 250 ns.

The available range is from 14.55 to 59590 step/s^2 with a resolution of 14.55 step/s^2.

The 0xFFF value of the register is reserved and it should never be used.

Any attempt to write to the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.1.6 DEC

The DEC register contains the speed profile deceleration expressed in step/tick^2 (format unsigned fixed point 0.40).

In order to convert DEC value in step/s^2, the following formula can be used:

Equation 6

$$[\text{step/s}^2] = \frac{\text{DEC} \cdot 2^{-40}}{\text{tick}^2}$$

where *DEC* is the integer number stored in the register and tick is 250 ns.

The available range is from 14.55 to 59590 step/s2 with a resolution of 14.55 step/s2.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.1.7 MAX_SPEED

The MAX_SPEED register contains the speed profile maximum speed expressed in step/tick (format unsigned fixed point 0.18).

In order to convert it in step/s, the following formula can be used:

Equation 7

$$[\text{step/s}] = \frac{\text{MAX\_SPEED} \cdot 2^{-18}}{\text{tick}}$$

where *MAX_SPEED* is the integer number stored in the register and tick is 250 ns.

The available range is from 15.25 to 15610 step/s with a resolution of 15.25 step/s.

9.1.8 MIN_SPEED

The MIN_SPEED register contains the following parameters:

Table 11. MIN_SPEED register

Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
LSPD_OPT	MIN_SPEED											

The MIN_SPEED parameter contains the speed profile minimum speed. Its value is expressed in step/tick and to convert it in step/s, the following formula can be used:

Equation 8

$$[\text{step/s}] = \frac{\text{MIN\_SPEED} \cdot 2^{-24}}{\text{tick}}$$

where *MIN_SPEED* is the integer number stored in the register and tick is the ramp 250 ns.

The available range is from 0 to 976.3 step/s with a resolution of 0.238 step/s.

When the LSPD_OPT bit is set high, the low speed optimization feature is enabled and the MIN_SPEED value indicates the speed threshold below which the compensation works. In this case the minimum speed of the speed profile is set to zero.

An attempt to write the register when the motor is running causes the NOTPERF_CMD flag to rise.

9.1.9 FS_SPD

The FS_SPD register contains the threshold speed. When the actual speed exceeds this value, the step mode is automatically switched to full-step two-phase on. Its value is expressed in step/tick (format unsigned fixed point 0.18) and to convert it in step/s, the following formula can be used.

Equation 9

$$[\text{step/s}] = \frac{(\text{FS\_SPD} + 0.5) \cdot 2^{-18}}{\text{tick}}$$

If the FS_SPD value is set to h3FF (max.) the system always works in microstepping mode (SPEED must go beyond the threshold to switch to Full-step mode). Setting FS_SPD to zero does not have the same effect as setting Step mode to full-step two-phase on: the zero FS_SPD value is equivalent to a speed threshold of about 7.63 step/s.

The available range is from 7.63 to 15625 step/s with a resolution of 15.25 step/s.

9.1.10 KVAL_HOLD, KVAL_RUN, KVAL_ACC and KVAL_DEC

The KVAL_HOLD register contains the K_{VAL} value that is assigned to the PWM modulators when the motor is stopped (compensation excluded).

The KVAL_RUN register contains the K_{VAL} value that is assigned to the PWM modulators when the motor is running at constant speed (compensation excluded).

The KVAL_ACC register contains the starting K_{VAL} value that can be assigned to the PWM modulators during acceleration (compensation excluded).

The KVAL_DEC register contains the starting K_{VAL} value that can be assigned to the PWM modulators during deceleration (compensation excluded).

The available range is from 0 to 0.996 x V_S with a resolution of 0.004 x V_S, as shown in *Table 12*.

Table 12. Voltage amplitude regulation registers

KVAL_X [7 … 0]								Output voltage
0	0	0	0	0	0	0	0	0
0	0	0	0	0	0	0	1	V_S x (1/256)
⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮
1	1	1	1	1	1	1	0	V_S x (254/256)
1	1	1	1	1	1	1	1	V_S x (255/256)

9.1.11 INT_SPEED

The INT_SPEED register contains the speed value at which the BEMF compensation curve changes slope (see *Section 7.4 on page 36*). Its value is expressed in step/tick and to convert it in step/s, the following formula can be used:

Equation 10

$$[step/s] = \frac{INT\_SPEED \cdot 2^{-26}}{tick}$$

where *INT_SPEED* is the integer number stored in the register and tick is 250 ns.

The available range is from 0 to 976.5 step/s with a resolution of 0.0596 step/s.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.1.12 ST_SLP

The ST_SLP register contains the BEMF compensation curve slope that is used when the speed is lower than the intersect speed (see *Section 7.4*). Its value is expressed in s/step and the available range is from 0 to 0.004 with a resolution of 0.000015.

When ST_SLP, FN_SLP_ACC and FN_SLP_DEC parameters are set to zero, no BEMF compensation is performed.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.1.13 FN_SLP_ACC

The FN_SLP_ACC register contains the BEMF compensation curve slope that is used when the speed is greater than the intersect speed during acceleration (see *Section 7.4*). Its value is expressed in s/step and the available range is from 0 to 0.004 with a resolution of 0.000015.

When ST_SLP, FN_SLP_ACC and FN_SLP_DEC parameters are set to zero, no BEMF compensation is performed.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.1.14 FN_SLP_DEC

The FN_SLP_DEC register contains the BEMF compensation curve slope that is used when the speed is greater than the intersect speed during deceleration (see *Section 7.4*). Its value is expressed in s/step and the available range is from 0 to 0.004 with a resolution of 0.000015.

When ST_SLP, FN_SLP_ACC and FN_SLP_DEC parameters are set to zero, no BEMF compensation is performed.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.1.15 K_THERM

The K_THERM register contains the value used by the winding resistance thermal drift compensation system (see *Section 7.6 on page 37*).

The available range is from 1 to 1.46875 with a resolution of 0.03125, as shown in *Table 13*.

Table 13. Winding resistance thermal drift compensation coefficient

K_THERM [3 …. 0]				Compensation coefficient
0	0	0	0	1
0	0	0	1	1.03125
⋮	⋮	⋮	⋮	⋮
1	1	1	0	1.4375
1	1	1	1	1.46875

9.1.16 ADC_OUT

The ADC_OUT register contains the result of the analog-to-digital conversion of the ADCIN pin voltage; the result is available even if the supply voltage compensation is disabled.

Any attempt to write to the register causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

Table 14. ADC_OUT value and motor supply voltage compensation feature

V_S	V_{ADCIN}/V_{REG}	ADC_OUT [4 … 0]					Compensation coefficient
Greater than $V_{S,nom}$ + 50%	> 24/32	1	1	X	X	X	0.65625
$V_{S,nom}$ + 50%	24/32	1	1	0	0	0	0.65625
⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮
$V_{S,nom}$	16/32	1	0	0	0	0	1
⋮	⋮	⋮	⋮	⋮	⋮	⋮	⋮
$V_{S,nom}$ – 50%	8/32	0	1	0	0	0	1.968875
Lower than $V_{S,nom}$ – 50%	< 8/32	0	0	X	X	X	1.968875

9.1.17 OCD_TH

The OCD_TH register contains the overcurrent threshold value (see *Section 6.9 on page 29*). The available range is from 375 mA to 6 A, in steps of 375 mA, as shown in *Table 15*.

Table 15. Overcurrent detection threshold

OCD_TH [3 … 0]				Overcurrent detection threshold
0	0	0	0	375 mA
0	0	0	1	750 mA
…	…	…	…	…
1	1	1	0	5.625 A
1	1	1	1	6 A

9.1.18 STALL_TH

The STALL_TH register contains the stall detection threshold value (see *Section 7.2 on page 35*). The available range is from 31.25 mA to 4 A with a resolution of 31.25 mA.

Table 16. Stall detection threshold

STALL_th [6 … 0]							Stall detection threshold
0	0	0	0	0	0	0	31.25 mA
0	0	0	0	0	0	1	62.5 mA
…	…	…	…	…	…	…	…
1	1	1	1	1	1	0	3.969 A
1	1	1	1	1	1	1	4 A

9.1.19 STEP_MODE

The STEP_MODE register has the following structure:

Table 17. STEP_MODE register

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
SYNC_EN	SYNC_SEL			0(1)	STEP_SEL		

1. When the register is written, this bit should be set to 0.

The STEP_SEL parameter selects one of eight possible stepping modes:

Table 18. Step mode selection

STEP_SEL[2 .... 0]			Step mode
0	0	0	Full-step
0	0	1	Half-step
0	1	0	1/4 microstep
0	1	1	1/8 microstep
1	0	0	1/16 microstep
1	0	1	1/32 microstep
1	1	0	1/64 microstep
1	1	1	1/128 microstep

Every time the step mode is changed, the electrical position (i.e. the point of microstepping sinewave that is generated) is reset to the first microstep.

Warning: Every time STEP_SEL is changed, the value in the ABS_POS register loses meaning and should be reset.

Any attempt to write the register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

When the SYNC_EN bit is set low, $\overline{\text{BUSY}}$/SYNC output is forced low during command execution, otherwise, when the SYNC_EN bit is set high, $\overline{\text{BUSY}}$/SYNC output provides a clock signal according to the SYNC_SEL parameter.

Table 19. SYNC output frequency

		STEP_SEL (f_{FS} is the full-step frequency)							
		000	**001**	**010**	**011**	**100**	**101**	**110**	**111**
SYNC_SEL	**000**	f_{FS} /2	f_{FS} /2	f_{FS} /2	f_{FS} /2	f_{FS} /2	f_{FS} /2	f_{FS} /2	f_{FS} /2
	001	NA	f_{FS}	f_{FS}	f_{FS}	f_{FS}	f_{FS}	f_{FS}	f_{FS}
	010	NA	NA	2· f_{FS}	2· f_{FS}	2· f_{FS}	2· f_{FS}	2· f_{FS}	2· f_{FS}
	011	NA	NA	NA	4· f_{FS}	4· f_{FS}	4· f_{FS}	4· f_{FS}	4· f_{FS}
	100	NA	NA	NA	NA	8· f_{FS}	8· f_{FS}	8· f_{FS}	8· f_{FS}
	101	NA	NA	NA	NA	NA	16· f_{FS}	16· f_{FS}	16· f_{FS}
	110	NA	NA	NA	NA	NA	NA	32· f_{FS}	32· f_{FS}
	111	NA	NA	NA	NA	NA	NA	NA	64· f_{FS}

The synchronization signal is obtained starting from electrical position information (EL_POS register) according to *Table 20*:

Table 20. SYNC signal source

SYNC_SEL[2 …. 0]			Source
0	0	0	EL_POS[7]
0	0	1	EL_POS[6]
0	1	0	EL_POS[5]
0	1	1	EL_POS[4]
1	0	0	EL_POS[3]
1	0	1	EL_POS[2]
1	1	0	EL_POS[1]
1	1	1	EL_POS[0]

9.1.20 ALARM_EN

The ALARM_EN register allows the selection of which alarm signals are used to generate the FLAG output. If the respective bit of the ALARM_EN register is set high, the alarm condition forces the FLAG pin output down.

Table 21. ALARM_EN register

ALARM_EN bit	Alarm condition
0 (LSB)	Overcurrent
1	Thermal shutdown
2	Thermal warning
3	Undervoltage
4	Stall detection (Bridge A)
5	Stall detection (Bridge B)
6	Switch turn-on event
7 (MSB)	Wrong or non-performable command

9.1.21 CONFIG

The CONFIG register has the following structure:

Table 22. CONFIG register

Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
F_PWM_INT			F_PWM_DEC			POW_SR	
Bit 7	**Bit 6**	**Bit 5**	**Bit 4**	**Bit 3**	**Bit 2**	**Bit 1**	**Bit 0**
OC_SD	RESERVED	EN_VSCOMP	SW_MODE	EXT_CLK	OSC_SEL		

The OSC_SEL and EXT_CLK bits set the system clock source:

Table 23. Oscillator management

EXT_CLK	OSC_SEL[2 …. 0]			Clock source	OSCIN	OSCOUT
0	0	0	0	Internal oscillator: 16 MHz	Unused	Unused
0	0	0	1			
0	0	1	0			
0	0	1	1			
1	0	0	0	Internal oscillator: 16 MHz	Unused	Supplies a 2-MHz clock
1	0	0	1	Internal oscillator: 16 MHz	Unused	Supplies a 4-MHz clock
1	0	1	0	Internal oscillator: 16 MHz	Unused	Supplies an 8-MHz clock
1	0	1	1	Internal oscillator: 16 MHz	Unused	Supplies a 16-MHz clock
0	1	0	0	External crystal or resonator: 8 MHz	Crystal/resonator driving	Crystal/resonator driving
0	1	0	1	External crystal or resonator: 16 MHz	Crystal/resonator driving	Crystal/resonator driving
0	1	1	0	External crystal or resonator: 24 MHz	Crystal/resonator driving	Crystal/resonator driving
0	1	1	1	External crystal or resonator: 32 MHz	Crystal/resonator driving	Crystal/resonator driving
1	1	0	0	Ext clock source: 8 MHz (Crystal/resonator driver disabled)	Clock source	Supplies inverted OSCIN signal
1	1	0	1	Ext clock source: 16 MHz (Crystal/resonator driver disabled)	Clock source	Supplies inverted OSCIN signal
1	1	1	0	Ext clock source: 24 MHz (Crystal/resonator driver disabled)	Clock source	Supplies inverted OSCIN signal
1	1	1	1	Ext clock source: 32 MHz (Crystal/resonator driver disabled)	Clock source	Supplies inverted OSCIN signal

The SW_MODE bit sets the external switch to act as HardStop interrupt or not:

Table 24. External switch hard stop interrupt mode

SW_MODE	Switch mode
0	HardStop interrupt
1	User disposal

The OC_SD bit sets whether an overcurrent event causes or not the bridges to turn off; the OCD flag in the STATUS register is forced low anyway:

Table 25. Overcurrent event

OC_SD	Overcurrent event
1	Bridges shut down
0	Bridges do not shut down

The POW_SR bits set the slew rate value of power bridge output:

Table 26. Programmable power bridge output slew rate values

POW_SR [1 .... 0]		Output slew rate(1) [V/µs]
0	0	320
0	1	75
1	0	110
1	1	260

1. See S_{Rout_r} and S_{Rout_f} parameters in *Table 5 on page 11* for details.

The EN_VSCOMP bit sets whether the motor supply voltage compensation is enabled or not.

Table 27. Motor supply voltage compensation enable

EN_VSCOMP	Motor supply voltage compensation
0	Disabled
1	Enabled

The F_PWM_INT bits set the integer division factor of PWM frequency generation.

Table 28. PWM frequency: integer division factor

F_PWM_INT [2 …. 0]			Integer division factor
0	0	0	1
0	0	1	2
0	1	0	3
0	1	1	4
1	0	0	5
1	0	1	6
1	1	0	7
1	1	1	

The F_PWM_DEC bits set the multiplication factor of PWM frequency generation.

Table 29. PWM frequency: multiplication factor

F_PWM_DEC [2 …. 0]			Multiplication factor
0	0	0	0.625
0	0	1	0.75
0	1	0	0.875
0	1	1	1
1	0	0	1.25
1	0	1	1.5
1	1	0	1.75
1	1	1	2

In the following tables all available PWM frequencies are listed according to oscillator frequency, F_PWM_INT and F_PWM_DEC values (CONFIG register OSC_SEL parameter must be correctly programmed).

Table 30. Available PWM frequencies [kHz]: 8-MHz oscillator frequency

	F_PWM_DEC							
F_PWM_INT	**000**	**001**	**010**	**011**	**100**	**101**	**110**	**111**
000	9.8	11.7	13.7	15.6	19.5	23.4	27.3	31.3
001	4.9	5.9	6.8	7.8	9.8	11.7	13.7	15.6
010	3.3	3.9	4.6	5.2	6.5	7.8	9.1	10.4
011	2.4	2.9	3.4	3.9	4.9	5.9	6.8	7.8
100	2.0	2.3	2.7	3.1	3.9	4.7	5.5	6.3
101	1.6	2.0	2.3	2.6	3.3	3.9	4.6	5.2
110	1.4	1.7	2.0	2.2	2.8	3.3	3.9	4.5

Table 31. Available PWM frequencies [kHz]: 16-MHz oscillator frequency

	F_PWM_DEC							
F_PWM_INT	**000**	**001**	**010**	**011**	**100**	**101**	**110**	**111**
000	19.5	23.4	27.3	31.3	39.1	46.9	54.7	62.5
001	9.8	11.7	13.7	15.6	19.5	23.4	27.3	31.3
010	6.5	7.8	9.1	10.4	13.0	15.6	18.2	20.8
011	4.9	5.9	6.8	7.8	9.8	11.7	13.7	15.6
100	3.9	4.7	5.5	6.3	7.8	9.4	10.9	12.5
101	3.3	3.9	4.6	5.2	6.5	7.8	9.1	10.4
110	2.8	3.3	3.9	4.5	5.6	6.7	7.8	8.9

Table 32. Available PWM frequencies [kHz]: 24-MHz oscillator frequency

	F_PWM_DEC							
F_PWM_INT	**000**	**001**	**010**	**011**	**100**	**101**	**110**	**111**
000	29.3	35.2	41.0	46.9	58.6	70.3	82.0	93.8
001	14.6	17.6	20.5	23.4	29.3	35.2	41.0	46.9
010	9.8	11.7	13.7	15.6	19.5	23.4	27.3	31.3
011	7.3	8.8	10.3	11.7	14.6	17.6	20.5	23.4
100	5.9	7.0	8.2	9.4	11.7	14.1	16.4	18.8
101	4.9	5.9	6.8	7.8	9.8	11.7	13.7	15.6
110	4.2	5.0	5.9	6.7	8.4	10.0	11.7	13.4

Table 33. Available PWM frequencies [kHz]: 32-MHz oscillator frequency

	F_PWM_DEC							
F_PWM_INT	**000**	**001**	**010**	**011**	**100**	**101**	**110**	**111**
000	39.1	46.9	54.7	62.5	78.1	93.8	109.4	125.0
001	19.5	23.4	27.3	31.3	39.1	46.9	54.7	62.5
010	13.0	15.6	18.2	20.8	26.0	31.3	36.5	41.7
011	9.8	11.7	13.7	15.6	19.5	23.4	27.3	31.3
100	7.8	9.4	10.9	12.5	15.6	18.8	21.9	25.0
101	6.5	7.8	9.1	10.4	13.0	15.6	18.2	20.8
110	5.6	6.7	7.8	8.9	11.2	13.4	15.6	17.9

Any attempt to write the CONFIG register when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.1.22 STATUS

Table 34. STATUS register

Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8
SCK_MOD	STEP_LOSS_B	STEP_LOSS_A	OCD	TH_SD	TH_WRN	UVLO	WRONG_CMD
Bit 7	**Bit 6**	**Bit 5**	**Bit 4**	**Bit 3**	**Bit 2**	**Bit 1**	**Bit 0**
NOTPERF_CMD	MOT_STATUS		DIR	SW_EVN	SW_F	BUSY	HiZ

When the HiZ flag is high, it indicates that the bridges are in high impedance state. Any motion command makes the device exit from High Z state (HardStop and SoftStop included), unless error flags forcing a High Z state are active.

The UVLO flag is active low and is set by an undervoltage lockout or reset events (power-up included).

The TH_WRN, TH_SD, OCD flags are active low and indicate, respectively, thermal warning, thermal shutdown and overcurrent detection events.

STEP_LOSS_A and STEP_LOSS_B flags are forced low when a stall is detected on bridge A or bridge B respectively.

The NOTPERF_CMD and WRONG_CMD flags are active high and indicate, respectively, that the command received by SPI cannot be performed or does not exist at all.

The SW_F flag reports the SW input status (low for open and high for closed).

The SW_EVN flag is active high and indicates a switch turn-on event (SW input falling edge).

The UVLO, TH_WRN, TH_SD, OCD, STEP_LOSS_A, STEP_LOSS_B, NOTPERF_CMD, WRONG_CMD and SW_EVN flags are latched: when the respective conditions make them active (low or high), they remain in that state until a GetStatus command is sent to the IC.

The BUSY bit reflects the $\overline{\text{BUSY}}$ pin status. The BUSY flag is low when a constant speed, positioning or motion command is under execution and is released (high) after the command has been completed.

The SCK_MOD bit is an active high flag indicating that the device is working in Step-clock mode. In this case the step-clock signal should be provided through the STCK input pin. The DIR bit indicates the current motor direction:

Table 35. STATUS register DIR bit

DIR	Motor direction
1	Forward
0	Reverse

MOT_STATUS indicates the current motor status:

Table 36. STATUS register MOT_STATUS bits

MOT_STATUS		Motor status
0	0	Stopped
0	1	Acceleration
1	0	Deceleration
1	1	Constant speed

Any attempt to write to the register causes the command to be ignored and the NOTPERF_CMD flag to rise.

9.2 Application commands

The command summary is given in *Table 37*.

Table 37. Application commands

Command mnemonic	Command binary code					Action
	[7 …. 5]	[4]	[3]	[2 …. 1]	[0]	
NOP	000	0	0	00	0	Nothing
SetParam(PARAM,VALUE)	000	[PARAM]				Writes VALUE in PARAM register
GetParam(PARAM)	001	[PARAM]				Returns the stored value in PARAM register
Run(DIR,SPD)	010	1	0	00	DIR	Sets the target speed and the motor direction
StepClock(DIR)	010	1	1	00	DIR	Puts the device into Step-clock mode and imposes DIR direction
Move(DIR,N_STEP)	010	0	0	00	DIR	Makes N_STEP (micro)steps in DIR direction (Not performable when motor is running)
GoTo(ABS_POS)	011	0	0	00	0	Brings motor into ABS_POS position (minimum path)
GoTo_DIR(DIR,ABS_POS)	011	0	1	00	DIR	Brings motor into ABS_POS position forcing DIR direction
GoUntil(ACT,DIR,SPD)	100	0	ACT	01	DIR	Performs a motion in DIR direction with speed SPD until SW is closed, the ACT action is executed then a SoftStop takes place.
ReleseSW(ACT, DIR)	100	1	ACT	01	DIR	Performs a motion in DIR direction at minimum speed until the SW is released (open), the ACT action is executed then a HardStop takes place.
GoHome	011	1	0	00	0	Brings the motor into HOME position
GoMark	011	1	1	00	0	Brings the motor into MARK position
ResetPos	110	1	1	00	0	Resets the ABS_POS register (set HOME position)

Table 37. Application commands (continued)

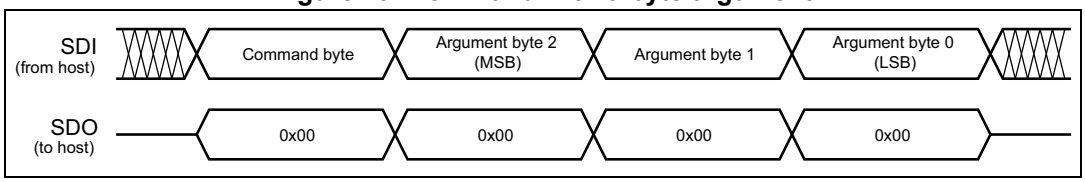
Command mnemonic	Command binary code					Action
	[7 …. 5]	[4]	[3]	[2 …. 1]	[0]	
ResetDevice	110	0	0	00	0	Device is reset to power-up conditions.
SoftStop	101	1	0	00	0	Stops motor with a deceleration phase
HardStop	101	1	1	00	0	Stops motor immediately
SoftHiZ	101	0	0	00	0	Puts the bridges into high impedance status after a deceleration phase
HardHiZ	101	0	1	00	0	Puts the bridges into high impedance status immediately
GetStatus	110	1	0	00	0	Returns the STATUS register value
RESERVED	111	0	1	01	1	RESERVED COMMAND
RESERVED	111	1	1	00	0	RESERVED COMMAND

9.2.1 Command management

The host microcontroller can control motor motion and configure the L6470 device through a complete set of commands.

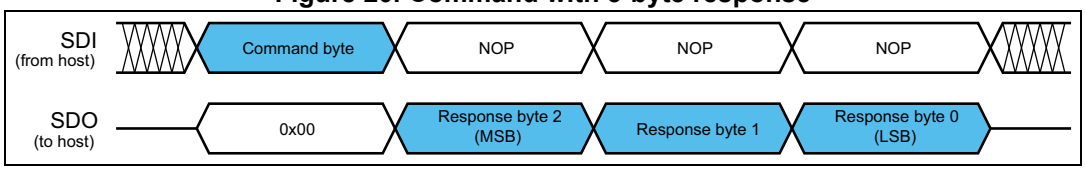
All commands are composed by a single byte. After the command byte, some bytes of arguments should be needed (see *Figure 19*). Argument length can vary from 1 to 3 bytes.

Figure 19. Command with 3-byte argument



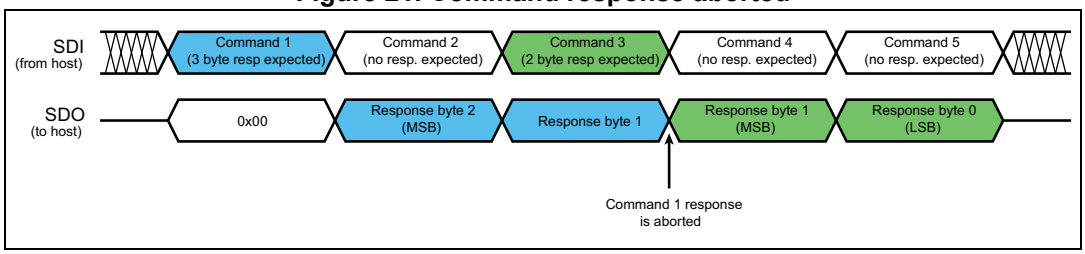
By default, the device returns an all zero response for any received byte, the only exceptions are the GetParam and GetStatus commands. When one of these commands is received, the following response bytes represent the related register value (see *Figure 20*). Response length can vary from 1 to 3 bytes.

Figure 20. Command with 3-byte response



During response transmission, new commands can be sent. If a command requiring a response is sent before the previous response is completed, the response transmission is aborted and the new response is loaded into the output communication buffer (see *Figure 21*).

Figure 21. Command response aborted



When a byte that does not correspond to a command is sent to the IC, it is ignored and the WRONG_CMD flag in the STATUS register is raised (see *Section 9.1.22*).

9.2.2 Nop

Table 38. Nop command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	0	0	0	0	0	0	0	From host

Nothing is performed.

9.2.3 SetParam (PARAM, VALUE)

Table 39. SetParam command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	0	0	PARAM					From host
VALUE Byte 2 (if needed)								
VALUE Byte 1 (if needed)								
VALUE Byte 0								

The SetParam command sets the PARAM register value equal to VALUE; PARAM is the respective register address listed in *Table 12 on page 44*.

The command should be followed by the new register VALUE (most significant byte first). The number of bytes making up the VALUE argument depends on the length of the target register (see *Table 12*).

Some registers cannot be written (see *Table 12*); any attempt to write one of those registers causes the command to be ignored and the WRONG_CMD flag to rise at the end of the command byte as if an unknown command code were sent (see *Section 9.1.22 on page 55*).

Some registers can only be written in particular conditions (see *Table 12*); any attempt to write one of those registers when the conditions are not satisfied causes the command to be ignored and the NOTPERF_CMD flag to rise at the end of the last argument byte (see *Section 9.1.22*).

Any attempt to set an inexistent register (wrong address value) causes the command to be ignored and the WRONG_CMD flag to rise at the end of the command byte as if an unknown command code were sent.

9.2.4 GetParam (PARAM)

Table 40. GetParam command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	0	1	PARAM					From host
ANS Byte 2 (if needed)								To host
ANS Byte 1 (if needed)								To host
ANS Byte 0								To host

This command reads the current PARAM register value; PARAM is the respective register address listed in *Table 12*.

The command response is the current value of the register (most significant byte first). The number of bytes making up the command response depends on the length of the target register (see *Table 12*).

The returned value is the register one at the moment of GetParam command decoding. If register values change after this moment, the response is not accordingly updated.

All registers can be read anytime.

Any attempt to read an inexistent register (wrong address value) causes the command to be ignored and the WRONG_CMD flag to rise at the end of the command byte as if an unknown command code were sent.

9.2.5 Run (DIR, SPD)

Table 41. Run command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	0	1	0	0	0	DIR	From host
X	X	X	X	SPD (Byte 2)				From host
SPD (Byte 1)								From host
SPD (Byte 0)								From host

The Run command produces a motion at SPD speed; the direction is selected by the DIR bit: '1' forward or '0' reverse. The SPD value is expressed in step/tick (format unsigned fixed point 0.28) that is the same format as the SPEED register (see *Section 9.1.4 on page 42*).

Note: The SPD value should be lower than MAX_SPEED and greater than MIN_SPEED otherwise the Run command is executed at MAX_SPEED or MIN_SPEED respectively.

This command keeps the BUSY flag low until the target speed is reached.

This command can be given anytime and is immediately executed.

9.2.6 StepClock (DIR)

Table 42. Stepclock command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	0	1	1	0	0	DIR	From host

The StepClock command switches the device in Step-clock mode (see *Section 6.7.5 on page 26*) and imposes the forward (DIR = '1') or reverse (DIR = '0') direction.

When the device is in Step-clock mode, the SCK_MOD flag in the STATUS register is raised and the motor is always considered stopped (see *Section 6.7.5* and *Section 9.1.22 on page 55*).

The device exits from Step-clock mode when a constant speed, absolute positioning or motion command is sent through SPI. Motion direction is imposed by the respective StepClock command argument and can by changed by a new StepClock command without exiting Step-clock mode.

Events that cause bridges to be forced into high impedance state (overtemperature, overcurrent, etc.) do not cause the device to leave Step-clock mode.

The StepClock command does not force the BUSY flag low. This command can only be given when the motor is stopped. If a motion is in progress, the motor should be stopped and it is then possible to send a StepClock command.

Any attempt to perform a StepClock command when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.2.7 Move (DIR, N_STEP)

Table 43. Move command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	0	0	0	0	0	DIR	From host
X	X	N_STEP (Byte 2)						From host
N_STEP (Byte 1)								From host
N_STEP (Byte 0)								From host

The Move command produces a motion of N_STEP microsteps; the direction is selected by the DIR bit ('1' forward or '0' reverse).

The N_STEP value is always in agreement with the selected step mode; the parameter value unit is equal to the selected step mode (full, half, quarter, etc.).

This command keeps the BUSY flag low until the target number of steps is performed. This command can only be performed when the motor is stopped. If a motion is in progress, the motor must be stopped and it is then possible to perform a Move command.

Any attempt to perform a Move command when the motor is running causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.2.8 GoTo (ABS_POS)

Table 44. GoTo command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	1	0	0	0	0	0	From host
X	X	ABS_POS (Byte 2)						From host
ABS_POS (Byte 1)								From host
ABS_POS (Byte 0)								From host

The GoTo command produces a motion to ABS_POS absolute position through the shortest path. The ABS_POS value is always in agreement with the selected step mode; the parameter value unit is equal to the selected step mode (full, half, quarter, etc.).

The GoTo command keeps the BUSY flag low until the target position is reached.

This command can be given only when the previous motion command has been completed (BUSY flag released).

Any attempt to perform a GoTo command when a previous command is under execution (BUSY low) causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22*).

9.2.9 GoTo_DIR (DIR, ABS_POS)

Table 45. GoTo_DIR command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	1	0	1	0	0	DIR	From host
X	X	ABS_POS (Byte 2)						From host
ABS_POS (Byte 1)								From host
ABS_POS (Byte 0)								From host

The GoTo_DIR command produces a motion to ABS_POS absolute position imposing a forward (DIR = '1') or a reverse (DIR = '0') rotation. The ABS_POS value is always in agreement with the selected step mode; the parameter value unit is equal to the selected step mode (full, half, quarter, etc.).

The GoTo_DIR command keeps the BUSY flag low until the target speed is reached. This command can be given only when the previous motion command has been completed (BUSY flag released).

Any attempt to perform a GoTo_DIR command when a previous command is under execution (BUSY low) causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.2.10 GoUntil (ACT, DIR, SPD)

Table 46. GoUntil command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	0	0	0	ACT	0	1	DIR	From host
X	X	X	X	SPD (Byte 2)				From host
SPD (Byte 1)								From host
SPD (Byte 0)								From host

The GoUntil command produces a motion at SPD speed imposing a forward (DIR = '1') or a reverse (DIR = '0') direction. When an external switch turn-on event occurs (see *Section 6.13 on page 31*), the ABS_POS register is reset (if ACT = '0') or the ABS_POS register value is copied into the MARK register (if ACT = '1'); then the system performs a SoftStop command.

The SPD value is expressed in step/tick (format unsigned fixed point 0.28) that is the same format as the SPEED register (see *Section 9.1.4 on page 42*).

The SPD value should be lower than MAX_SPEED and greater than MIN_SPEED, otherwise the target speed is imposed at MAX_SPEED or MIN_SPEED respectively.

If the SW_MODE bit of the CONFIG register is set low, the external switch turn-on event causes a HardStop interrupt instead of the SoftStop one (see *Section 6.13* and *Section 9.1.21 on page 49*).

This command keeps the BUSY flag low until the switch turn-on event occurs and the motor is stopped. This command can be given anytime and is immediately executed.

9.2.11 ReleaseSW (ACT, DIR)

Table 47. ReleaseSW command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	0	0	1	ACT	0	1	DIR	From host

The ReleaseSW command produces a motion at minimum speed imposing a forward (DIR = '1') or reverse (DIR = '0') rotation. When SW is released (opened), the ABS_POS register is reset (ACT = '0') or the ABS_POS register value is copied into the MARK register (ACT = '1'); the system then performs a HardStop command.

Note that resetting the ABS_POS register is equivalent to setting the HOME position.

If the minimum speed value is less than 5 step/s or low speed optimization is enabled, the motion is performed at 5 step/s.

The ReleaseSW command keeps the BUSY flag low until the switch input is released and the motor is stopped.

9.2.12 GoHome

Table 48. GoHome command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	1	1	0	0	0	0	From host

The GoHome command produces a motion to the HOME position (zero position) via the shortest path.

Note that this command is equivalent to the “GoTo(0…0)” command. If a motor direction is mandatory, the GoTo_DIR command must be used (see *Section 9.2.9*).

The GoHome command keeps the BUSY flag low until the home position is reached. This command can be given only when the previous motion command has been completed. Any attempt to perform a GoHome command when a previous command is under execution (BUSY low) causes the command to be ignored and the NOTPERF_CMD to rise (see *Section 9.1.22 on page 55*).

9.2.13 GoMark

Table 49. GoMark command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0	1	1	1	1	0	0	0	From host

The GoMark command produces a motion to the MARK position performing the minimum path.

Note that this command is equivalent to the “GoTo (MARK)” command. If a motor direction is mandatory, the GoTo_DIR command must be used.

The GoMark command keeps the BUSY flag low until the MARK position is reached. This command can be given only when the previous motion command has been completed (BUSY flag released).

Any attempt to perform a GoMark command when a previous command is under execution (BUSY low) causes the command to be ignored and the NOTPERF_CMD flag to rise (see *Section 9.1.22 on page 55*).

9.2.14 ResetPos

Table 50. ResetPos command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	1	0	1	1	0	0	0	From host

The ResetPos command resets the ABS_POS register to zero. The zero position is also defined as HOME position (see *Section 6.5 on page 23*).

9.2.15 ResetDevice

Table 51. ResetDevice command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	1	0	0	0	0	0	0	From host

The ResetDevice command resets the device to power-up conditions (see *Section 6.1 on page 21*).

Note: At power-up the power bridges are disabled.

9.2.16 SoftStop

Table 52. SoftStop command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	0	1	1	0	0	0	0	From host

The SoftStop command causes an immediate deceleration to zero speed and a consequent motor stop; the deceleration value used is the one stored in the DEC register (see *Section 9.1.6 on page 43*).

When the motor is in high impedance state, a SoftStop command forces the bridges to exit from high impedance state; no motion is performed.

This command can be given anytime and is immediately executed. This command keeps the BUSY flag low until the motor is stopped.

9.2.17 HardStop

Table 53. HardStop command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	0	1	1	1	0	0	0	From host

The HardStop command causes an immediate motor stop with infinite deceleration.

When the motor is in high impedance state, a HardStop command forces the bridges to exit from high impedance state; no motion is performed.

This command can be given anytime and is immediately executed. This command keeps the BUSY flag low until the motor is stopped.

9.2.18 SoftHiZ

Table 54. SoftHiZ command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	0	1	0	0	0	0	0	From host

The SoftHiZ command disables the power bridges (high impedance state) after a deceleration to zero; the deceleration value used is the one stored in the DEC register (see *Section 9.1.6 on page 43*). When bridges are disabled, the HiZ flag is raised.

When the motor is stopped, a SoftHiZ command forces the bridges to enter into high impedance state.

This command can be given anytime and is immediately executed. This command keeps the BUSY flag low until the motor is stopped.

9.2.19 HardHiZ

Table 55. HardHiZ command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	0	1	0	1	0	0	0	From host

The HardHiZ command immediately disables the power bridges (high impedance state) and raises the HiZ flag.

When the motor is stopped, a HardHiZ command forces the bridges to enter into high impedance state.

This command can be given anytime and is immediately executed. This command keeps the BUSY flag low until the motor is stopped.

9.2.20 GetStatus

Table 56. GetStatus command structure

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
1	1	0	1	0	0	0	0	From host
STATUS MSByte								To host
STATUS LSByte								To host

The GetStatus command returns the STATUS register value.

The GetStatus command resets the STATUS register warning flags. The command forces the system to exit from any error state. The GetStatus command DOES NOT reset the HiZ flag.

10 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

10.1 HTSSOP28 package information

Figure 22. HTSSOP28 package outline

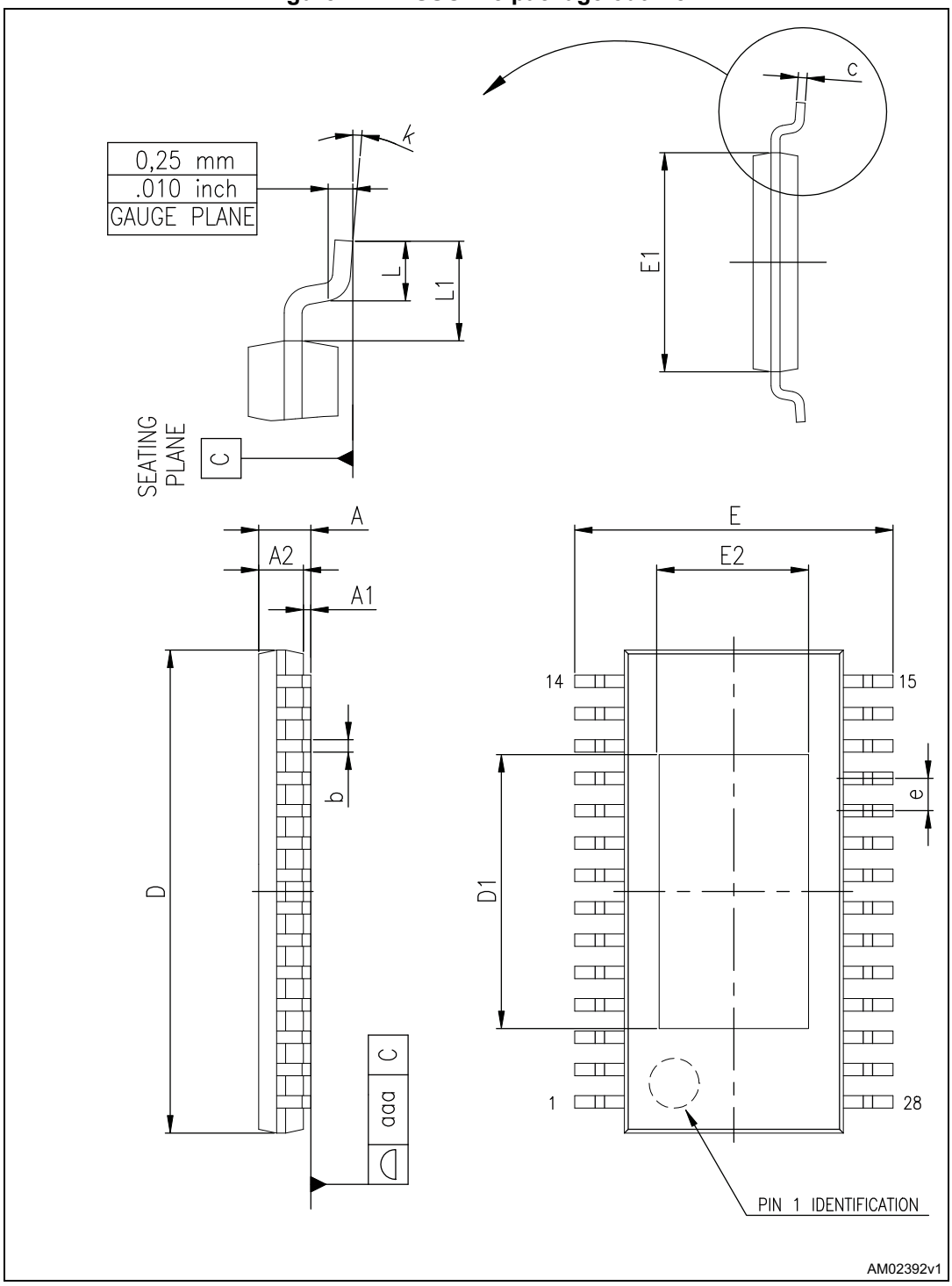


Table 57. HTSSOP28 package mechanical data

Symbol	Dimensions (mm)		
	Min.	Typ.	Max.
A			1.2
A1			0.15
A2	0.8	1.0	1.05
b	0.19		0.3
c	0.09		0.2
D(1)	9.6	9.7	9.8
D1		5.5	
E	6.2	6.4	6.6
E1(2)	4.3	4.4	4.5
E2		2.8	
e		0.65	
L	0.45	0.6	0.75
L1		1.0	
K	0°		8°
aaa		0.1	

1. Dimension “D” does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs must not exceed 0.15 mm per side.
2. Dimension “E1” does not include interlead flash or protrusions. Interlead flash or protrusions must not exceed 0.25 mm per side.

10.2 POWERSO36 package information

Figure 23. POWERSO36 package outline

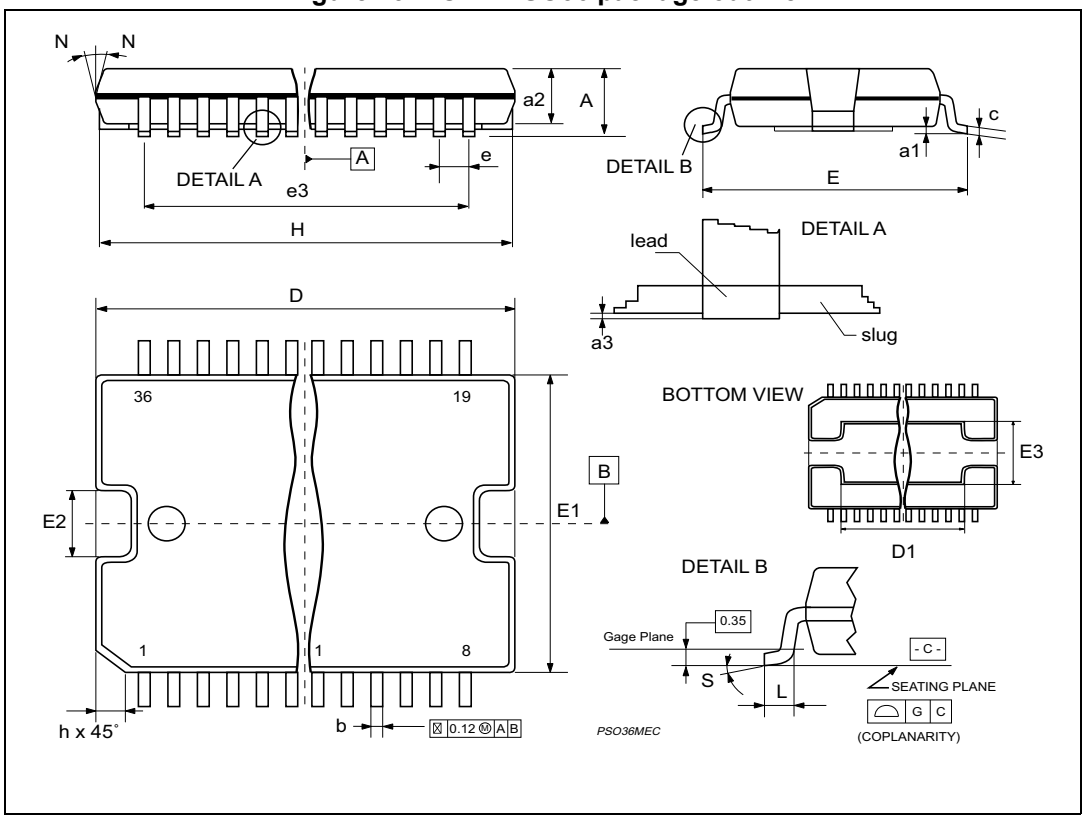


Table 58. POWERSO36 package mechanical data

Symbol	Dimensions (mm)		
	Min.	Typ.	Max.
A			3.60
a1	0.10		0.30
a2			3.30
a3	0		0.10
b	0.22		0.38
c	0.23		0.32
D(1)	15.80		16.00
D1	9.40		9.80
E	13.90		14.50
E1(1)	10.90		11.10
E2			2.90
E3	5.8		6.2
e		0.65	
e3		11.05	
G	0		0.10
H	15.50		15.90
h			1.10
L	0.80		1.10
N			10°
S	0°		8°

1. Dimension “D/E1” does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs must not exceed 0.15 mm per side.

11 Revision history

Table 59. Revision history

Date	Revision	Changes
06-Nov-2009	1	Initial release
05-Nov-2010	2	Document status promoted from preliminary data to datasheet
18-May-2011	3	Updated: *Table 4*, *Table 5* Added: *Section 6.7.6*, *Section*
19-Jun-2012	4	Added device in POWERSO36 and *Figure 3* Updated: *Table 2*, *Table 3*, *Table 4*, *Table 5*, *Table 6*, *Table 9* and *Section 9.1.11*. Minor text changes.
20-Dec-2012	5	Changed the title. Changed T_{OP} value in *Table 2* Removed T_j value in *Table 3*. Updated HTSSOP28 mechanical data.
19-May-2014	6	Updated *Figure 3 on page 16* (replaced “DIR” pin by “SW” pin). Updated *Section 6.4 on page 22* (replaced “the first microstep” by “zero”). Removed Section “Infinite acceleration/deceleration mode” from page 23. Updated *Section 9.1.5 on page 42* (replaced “When the ACC value is set to 0xFFF, the device works in infinite acceleration mode.” by “The 0xFFF value of the register is reserved and it should never be used.”). Updated *Section 9.1.6 on page 43* (removed “When the device is working in infinite acceleration mode, this value is ignored.”). Updated title of *Table 36 on page 56* (replaced “MOT_STATE” by “MOT_STATUS”). Updated *Section 10: Package information on page 67* (updated titles, reversed order of *Figure 22* and *Table 57*, *Figure 23* and *Table 58*, added note *1* below *Table 58*). Updated cross-references throughout document. Minor modifications throughout document.
12-Mar-2015	7	Removed “dSPIN™” from the main title *on page 1*. Minor modifications throughout document.

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