

## Features

- AEC-Q100 qualified
- Fully monolithic design
- High side field driver
- Self-start function
- Regulated voltage thermally compensated
- Configurable parameters through OTP cells
- Lamp driver (wake up and warning detection)
- Load response control (LRC)


## Protections

- Thermal shutdown
- Field short circuit protection
- Protected high side relay driver


## Description

The L9916 is a smart alternator voltage regulator intended to be used in automotive application for both 12 V and 24 V systems. The presence of OTP cells for parameters programmability makes it suitable for a wide range of charging application.

Table 1. Device summary

| Order code | Package | Packing |
| :---: | :---: | :---: |
| L9916 | Multiwatt8 | Tube |
| L9916BDTR | Bare die | Tape and Reel |

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## 1 Functional description

The L9916 is an alternator voltage regulator with high side power output for field driving. It can be used in both 12 V and 24 V system. It supplies a current through a power high side MOSFET to the excitation coil of the alternator and provides an internal freewheeling diode.

The L9916 is suitable for multi-phase-current alternators and its target function is to keep the battery at its nominal value, whatever the vehicle demand is. The set point control is achieved using an internal voltage reference thermally compensated. Continuous feedback to the ECU is provided through the field monitor output.

Slew rate control and filtering of the interface lines provide electromagnetic compatibility.
Figure 1. Simplified application diagram


### 1.1 State diagram

The state diagram is a description of the L9916 possible working conditions.
Below it is represented a state diagram for L9916 describing its Operation Modes depending on the Speed Phase (RPM) applied to Phase pin (PH), the battery level on Battery pin (VB), the duty cycle applied by the device on Field pin (F), and the Alarm function status. Alarm function will turn-on the lamp in case of fault.

The states represented in the diagram are:

- Standby:

It is the reference state of the FSM. It's the starting state and the final state for any regulator operation cycle (turn-on, turn-off).

- Pre-excitation:

L9916 is ready to operate (when reached from standby), or it checks external conditions to move to a different state.

- Regulation:

L9916 is working in order to regulate the battery voltage to the required voltage set point VBSP.

- Phase regulation:
the L9916 controls the field driver to keep phase signal amplitude not falling below defined values.
- LRC:

This state can be reached from Regulation in case there is a load variation that leads to meet some conditions on the expected field duty cycle value.

- Crank:

This state is necessary to manage SDT (Start Delay Time) timer before entering normal regulation.

The assumptions considered in the below diagram are that the Self-start function is enabled and the wake-up source (IGN or Key) status is kept constant during the AVR working operation.

Colored tags define the condition for jumping to the next state and non-colored tags represent configurations.

Figure 2. State diagram


Note: $\quad L R C$ procedure firstly increases $D F$ by $D F_{L R C B Z}$ and then it continues slowly increasing the DF using a fixed slope $D F_{\text {LRCUP }}$ After LRC is finished (due to VB $=$ VBSP condition reached), the new regulation duty cycle DFreg will be DF' reached by the LRC procedure.

### 1.2 Pin description

Table 2. Pin out description

| $\mathbf{N}^{\circ}$ | PIN | Function |
| :---: | :---: | :--- |
| 1 | PH | Phase sense input |
| 2 | FM | Field Monitor (PWM signal going to ECU) |
| 3 | IGN | Ignition input |
| 4 | LAMP | Key sensing and Warning Lamp terminal output |
| 5 | GND | Regulator ground |
| 6 | SENSE | Remote battery sense input |
| 7 | FIELD | High side driver output to control the Field current |
| 8 | B+ | Device power supply and Battery voltage sensing |

Figure 3. Pin out diagram


### 1.3 External component required

The only component strictly required is the capacitor C 1 ( $2.2 \mu \mathrm{~F}$ suggested) to suppress radio frequency injection and has to be connected as close as possible to $B$ and GND pins. If in the application some pins are not used, it is recommended to connect them to ground directly or with a 10 nF capacitor.

## 2 Electrical specifications

### 2.1 Absolute maximum ratings

Table 3. Absolute maximum ratings

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{VB}_{\text {LD }}$ | Transient supply voltage (load dump) $\mathrm{t}<500 \mathrm{~ms}$ | 58 | V |
| $\mathrm{VB}_{\mathrm{MAX}}$ | Transient supply voltage (low energy spikes) ISO7637-2 pulse <br> $1,2,3 /$ /SO7637-3 | 58 | V |
| $\mathrm{~T}_{\mathrm{j}}$ | Junction temperature range | -40 to 175 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\text {stg }}, \mathrm{T}_{\text {case }}$ | Storage and case temperature range | -40 to 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{P}_{\mathrm{TOT}}$ | Total power dissipation (@ $\left.\mathrm{T}_{\text {case }} \leq 150^{\circ} \mathrm{C}, \mathrm{I}_{\text {field }} \leq 5 \mathrm{~A}\right)$ | 4 | W |
| $\mathrm{VB}_{\mathrm{R}}$ | Reverse battery voltage @ $25^{\circ} \mathrm{C}, \mathrm{T}=15$ sec | -2.5 | V |
| $\mathrm{VPH}_{\text {min }}$ | Normal working condition reverse voltage (PH vs. GND) | -1.5 | V |
| $\mathrm{I}_{\text {Bond }}$ | DC pin current on F, B, GND (bonding limitation) | 15 | A |
| $\mathrm{ESD}_{\text {HBM }}$ | ESD HBM (All pins vs.GND) | $\pm 4$ | kV |

### 2.2 Internal clamping structure

Table 4. Internal clamping structure

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\mathrm{~V}_{\mathrm{cl}}$ | Internal central clamp voltage | $\mathrm{B}+$ current injected $\mathrm{I}=20 \mathrm{~mA}$ | 58 | - | 64 | V |

Table 5. Absolute maximum ratings and operative range

| Pin \# | Pin name | ABS max rating |  | Operative range |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| 1 | PH | -25 | Vcl | -3 | 36 | V |
| 2 | FM | -25 | Vcl | -0.3 | 36 | V |
| 3 | IGN | -25 | Vcl | -0.3 | 36 | V |
| 4 | LAMP | -0.3 | $\mathrm{~B}+$ | -0.3 | 36 | V |
| 5 | GND | - | - | - | - | - |
| 6 | SENSE | -25 | Vcl | -0.3 | 36 | V |
| 7 | FIELD | -1.5 | $\mathrm{~B}+$ | -1.5 | $\mathrm{~B}+$ | V |
| 8 | B+ | -2.5 | Vcl | 6 | 36 | V |

### 2.3 Thermal data

Table 6. Thermal data

| Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{R}_{\text {th } \text { j-case }}$ | Thermal resistance junction-to-case | Related to Multiwatt8 | - | - | 1.5 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{T}_{\mathrm{j} \text {-sd }}$ | Thermal shutdown threshold | Temperature to disable F, FM, L drivers. | 160 | 175 | 190 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{j} \text {-sdhy }}$ | Thermal shut-down hysteresis | L, F, FM from OFF STATE (due to thermal shutdown) to ON STATE | $\mathrm{T}_{\mathrm{j} \text {-sd }}$-10 | - | $\mathrm{T}_{\mathrm{j} \text {-sd }}-2$ | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{j}}$ | Operation temperature | - | -40 | - | 150 | ${ }^{\circ} \mathrm{C}$ |
|  |  | Fully functional. <br> Parameter <br> Deviations permissible | 150 | - | $\mathrm{T}_{\mathrm{j} \text {-sd }}$ |  |

### 2.4 Electrical characteristics

$T_{j}=-40$ to $150^{\circ} \mathrm{C}$, unless otherwise specified.

### 2.4.1 Pin "B"

The L9916 is supplied by the B+. This voltage is also used as the feedback voltage by the regulation loop.

Table 7. Electrical characteristics - Pin "B"

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| 1 | $\mathrm{VB}_{\text {OVR }}$ | Operating Voltage <br> Range | Application info, refer to Figure 7 | 6 | - | 36 | V |
| 2 | $\mathrm{IB}_{\text {stby }}$ | Standby current <br> consumption 12 V | $\mathrm{VB}=12.5 \mathrm{~V} ; \mathrm{VPH}=0 ; \mathrm{VL}=0 \mathrm{~V} ;$ <br> $\mathrm{VIGN}=$ open; VSENSE $=0 \mathrm{~V}$ <br> $\mathrm{~T}=25^{\circ} \mathrm{C}$ | - | - | 120 | $\mu \mathrm{~A}$ |
| 3 | $\mathrm{IB}_{\text {stby }}$ | Standby current <br> consumption 12 V | $\mathrm{VB}=12.5 \mathrm{~V} ; \mathrm{VPH}=0 ; \mathrm{VL}=0 \mathrm{~V} ;$ <br> $\mathrm{VIGN}=$ open; VSENSE $=12.5 \mathrm{~V}$ <br> $\mathrm{~T}=25^{\circ} \mathrm{C}$ | - | - | 170 | $\mu \mathrm{~A}$ |
| 4 | $\mathrm{IB}_{\text {stby }}$ | Standby current <br> consumption 24 V | $\mathrm{VB}=24 \mathrm{~V} ; \mathrm{VPH}=0 ; \mathrm{VL}=0 \mathrm{~V} ;$ <br> $\mathrm{VIGN}=$ open; VSENSE $=0 \mathrm{~V}$ <br> $\mathrm{~T}=25^{\circ} \mathrm{C}$ | - | - | 160 | $\mu \mathrm{~A}$ |
| 6 | $\mathrm{IB}_{\text {stby }}$ | Standby current <br> consumption 24 V | $\mathrm{VB}=24 \mathrm{~V} ; \mathrm{VPH}=0 ; \mathrm{VL}=0 \mathrm{~V} ;$ <br> $\mathrm{VIGN}=$ open; VSENSE $=24 \mathrm{~V}$ <br> $\mathrm{~T}=25^{\circ} \mathrm{C}$ | - | - | 200 | $\mu \mathrm{~A}$ |
| 6 | $\mathrm{IB}_{\text {active }}$ | Current <br> consumption out of <br> standby | $\mathrm{VB}=12.5 \mathrm{~V} ; \mathrm{VSENSE}=12.5 \mathrm{~V}$ <br> $\mathrm{VB}=24 \mathrm{~V} ; \mathrm{VSENSE}=24 \mathrm{~V}$ <br> No current load on FIELD and <br> LAMP | 5 | 11 | 30 | mA |

Table 7. Electrical characteristics - Pin "B" (continued)

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | $\mathrm{VB}_{\text {SP12,1 }}$ | Set-point voltage 12 V Option | VPH=10 Vpp square wave; L pin connected to B pin with 100 ohm; @ $\mathrm{T}_{\mathrm{j}}=30^{\circ} \mathrm{C}$ F duty cycle $=50 \%$ | -0.15 | 13.5 | +0.15 | V |
| 8 | $\mathrm{VB}_{\text {SP12,2 }}$ |  |  |  | 13.6 |  |  |
| 9 | $\mathrm{VB}_{\text {SP } 12,3}$ |  |  |  | 13.7 |  |  |
| 10 | $\mathrm{VB}_{\text {SP12,4 }}$ |  |  |  | 13.8 |  |  |
| 11 | $\mathrm{VB}_{\text {SP12,5 }}$ |  |  |  | 13.9 |  |  |
| 12 | $\mathrm{VB}_{\text {SP12,6 }}$ |  |  |  | 14 |  |  |
| 13 | $\mathrm{VB}_{\text {SP12,7 }}$ |  |  |  | 14.1 |  |  |
| 14 | VB ${ }_{\text {SP12,8 }}$ |  |  |  | 14.2 |  |  |
| 15 | $\mathrm{VB}_{\text {SP12,9 }}$ |  |  |  | 14.3 |  |  |
| 16 | $\mathrm{VB}_{\mathrm{SP} 12,10}$ |  |  |  | 14.4 |  |  |
| 17 | $\mathrm{VB}_{\mathrm{SP} 12,11}$ |  |  |  | 14.5 |  |  |
| 18 | $\mathrm{VB}_{\mathrm{SP} 12,12}$ |  |  |  | 14.6 |  |  |
| 19 | $\mathrm{VB}_{\text {SP12,13 }}$ |  |  |  | 14.7 |  |  |
| 20 | $\mathrm{VB}_{\mathrm{SP} 12,14}$ |  |  |  | 14.8 |  |  |
| 21 | $\mathrm{VB}_{\text {SP12,15 }}$ |  |  |  | 14.9 |  |  |
| 22 | $\mathrm{VB}_{\text {SP12,16 }}$ |  |  |  | 15 |  |  |
| 23 | $\mathrm{VB}_{\text {SP24,1 }}$ | Set-point voltage 24 V Option | VPH=20 Vpp square wave; L pin connected to B pin with 100 ohm; @ $\mathrm{Tj}=30^{\circ} \mathrm{C}$ <br> F duty cycle=50\% | -0.25 | 27 | +0.25 | V |
| 24 | $\mathrm{VB}_{\text {SP24,2 }}$ |  |  |  | 27.2 |  |  |
| 25 | $\mathrm{VB}_{\text {SP24,3 }}$ |  |  |  | 27.4 |  |  |
| 26 | $\mathrm{VB}_{\text {SP24,4 }}$ |  |  |  | 27.6 |  |  |
| 27 | $\mathrm{VB}_{\text {SP24,5 }}$ |  |  |  | 27.8 |  |  |
| 28 | $\mathrm{VB}_{\text {SP24,6 }}$ |  |  |  | 28 |  |  |
| 29 | $\mathrm{VB}_{\text {SP24,7 }}$ |  |  |  | 28.2 |  |  |
| 30 | VB ${ }_{\text {SP24,8 }}$ |  |  |  | 28.4 |  |  |
| 31 | $\mathrm{VB}_{\text {SP24,9 }}$ |  |  |  | 28.6 |  |  |
| 32 | $\mathrm{VB}_{\mathrm{SP} 24,10}$ |  |  |  | 28.8 |  |  |
| 33 | $\mathrm{VB}_{\text {SP24,11 }}$ |  |  |  | 29 |  |  |
| 34 | $\mathrm{VB}_{\mathrm{SP} 24,12}$ |  |  |  | 29.2 |  |  |
| 35 | $\mathrm{VB}_{\mathrm{SP} 24,13}$ |  |  |  | 29.4 |  |  |
| 36 | $\mathrm{VB}_{\text {SP24,14 }}$ |  |  |  | 29.6 |  |  |
| 37 | VB ${ }_{\text {SP24,15 }}$ |  |  |  | 29.8 |  |  |
| 38 | $\mathrm{VB}_{\text {SP24,16 }}$ |  |  |  | 30 |  |  |

Table 7. Electrical characteristics - Pin "B" (continued)

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 39 | VB ${ }_{\text {ITD1 }}$ | Internal Thermal Drift Options $30^{\circ} \mathrm{C}<\mathrm{T}<150^{\circ} \mathrm{C}$ | - | $\begin{aligned} & -1(12 \mathrm{~V}) \\ & -2(24 \mathrm{~V}) \end{aligned}$ | 0 | $\begin{aligned} & +1(12 \mathrm{~V}) \\ & +2(24 \mathrm{~V}) \end{aligned}$ | $\mathrm{mV} /{ }^{\circ} \mathrm{C}$ |
| 40 | VB ${ }_{\text {ITD2 }}$ |  | - |  | -2.5 |  |  |
| 41 | VB ${ }_{\text {ITD3 }}$ |  | - |  | -3.5 |  |  |
| 42 | VB ${ }_{\text {ITD4 }}$ |  | - |  | -7 |  |  |
| 43 | VB ITD5 |  | - |  | -10 |  |  |
| 39bis | VB ${ }_{\text {ITD1 }}$ | Internal Thermal Drift Options $-40^{\circ} \mathrm{C}<\mathrm{T}<30^{\circ} \mathrm{C}$ | - | $\begin{aligned} & -2(12 \mathrm{~V}) \\ & -4(24 \mathrm{~V}) \end{aligned}$ | 0 | $\begin{aligned} & +2(12 \mathrm{~V}) \\ & +4(24 \mathrm{~V}) \end{aligned}$ | $\mathrm{mV} /{ }^{\circ} \mathrm{C}$ |
| 40bis | VB ${ }_{\text {ITD2 }}$ |  | - |  | -2.5 |  |  |
| 41bis | VB ${ }_{\text {ITD3 }}$ |  | - |  | -3.5 |  |  |
| 42bis | VB ${ }_{\text {ITD4 }}$ |  | - |  | -7 |  |  |
| 43bis | VB ${ }_{\text {ITD }}$ |  | - |  | -10 |  |  |
| 44 | $\Delta \mathrm{VB}_{\text {load }, 1}$ | Regulated Voltage variation with the load $12 \mathrm{~V}(24 \mathrm{~V})$ system | Difference between regulated voltage when $F$ duty cycle is $10 \%$ and regulated voltage when $F$ duty cycle is $90 \%$ | - | - | $\begin{gathered} 200 \\ (300) \end{gathered}$ | mV |
| 45 | $\Delta \mathrm{VB}_{\text {load, } 2}$ | Regulated Voltage variation with the load $12 \mathrm{~V}(24 \mathrm{~V})$ system | Difference between regulated voltage when $F$ duty cycle is $10 \%$ and regulated voltage when $F$ duty cycle is $90 \%$ | - | - | $\begin{gathered} 300 \\ (400) \end{gathered}$ | mV |
| 46 | $\underset{(1)}{\Delta \mathrm{VB}_{\text {speed }}}$ | Regulated Voltage variation with the speed $12 \mathrm{~V}(24 \mathrm{~V})$ | Difference between regulated voltage when $\mathrm{I}_{\text {gen }}=10 \mathrm{~A}$ 2000rpm<Alt speed<18000rpm | - | - | $\begin{gathered} 100 \\ (150) \end{gathered}$ | mV |
| 47 | VBwb12 ${ }^{(1)}$ | Regulation without battery 12 V system | $\begin{aligned} & I_{\text {alt }}=5 \text { A resistive; } \\ & \mathrm{T}_{\text {case }}=25^{\circ} \mathrm{C} ; \\ & 2000<\mathrm{rpm}<18000 \end{aligned}$ | 12 | - | 16 | V |
| 48 | VBwb24 ${ }^{(1)}$ | Regulation without battery 24 V system | $\begin{aligned} & I_{\text {alt }}=5 \text { A resistive; } \\ & \mathrm{T}_{\text {case }}=25^{\circ} \mathrm{C} ; \\ & 2000<\mathrm{rpm}<18000 \end{aligned}$ | 22 | - | 34 | V |
| 49 | VBSP_CL12 | Set Point Voltage clamp at low temperature $\left(-40^{\circ} \mathrm{C}\right)$ <br> F duty cycle=50\% 12 V system | - | 14.7 | 15 | 15.3 | V |
| 50 | VBSP_CL24 | Set Point Voltage clamp at low temperature $\left(-40^{\circ} \mathrm{C}\right)$ <br> F duty cycle $=50 \%$ <br> 24 V system | - | 28.4 | 29 | 29.6 | V |

1. Alternator dependent parameter not tested.

Figure 4. Regulated Voltage variation with the load


Figure 5. Application 12 V - Default setpoint with $-3.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$


Regulation Curves Versus Temperature @ F Duty Cycle = 50\%
(Example with VBSP $=14.5 \mathrm{~V}$, VBITD $=-3.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ )

Figure 6. Application 24 V - Default setpoint with $-3.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$


Regulation Curves Versus Temperature @ F Duty Cycle $=50 \%$
(Example with VBSP $=29 \mathrm{~V}$, VBITD $=-3.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ ).

## Cranking security function

In order to avoid unpredictable regulation regions during the engine start, the regulator implements a Cranking Security Function.

At the ignition key on (i.e. "key" switch closed in the application schematic) the device starts in pre-excitation; in this phase the battery voltage is over both $\mathrm{VB}_{\mathrm{UV}}$ and $\mathrm{VB}_{\mathrm{Low}}$ and the pin $F$ provides a fixed duty cycle. When the starter is engaged, the battery sources a big current and then the applied voltage on the device decreases, as reported in ISO 7637-1 pulse 4 specification. If the voltage on pin $B$ decreases under $V^{\text {LOW }}$ threshold then the device will be frozen: the device will be in low current consumption with no activity on both F and FM pins.

The device will remain in frozen conditions until the voltage on B overcomes the $\mathrm{VB}_{\text {low }}$ threshold. After that the device start again to evaluate the input coming from PH pin.

Table 8. VB over and under voltage parameters

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: | :---: |
| 1 | VB $_{\text {Ovp12V }}$ | Over-voltage protections <br> threshold 12 V system | - | 16 |  | 16.9 | V |
| 2 | $\mathrm{VB}_{\text {Ovp24V }}$ | Over-voltage protections <br> threshold 24 V system | - | 32 | 33 | 34 | V |
| 3 | $\mathrm{VB}_{\text {OVft }}$ | Over-voltage filter time | - | 0.5 | 1.25 | 2 | ms |
| 4 | $\mathrm{VB}_{\text {UV12V }}$ | Under Voltage <br> 12 V system | - | 8 | 9 | 10 | V |
| 5 | $\mathrm{VB}_{\text {UV24V }}$ | Under Voltage <br> 24 V system | - | 16 | 18 | 20 | V |

Table 8. VB over and under voltage parameters (continued)

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| 6 | $\mathrm{VB}_{\text {low }}$ | Low Voltage | - | 5.6 | 6.35 | 7 | V |
| 7 | VBLV $_{\text {filt }}$ | Low-voltage filter time | Guaranteed down to $\mathrm{VB}_{\text {low }}$ | 100 | 150 | 200 | us |

## Overvoltage

For safety reason this function is implemented with independent circuitry with respect to the ones used by the field driver in order to guarantee that in case of overvoltage the field driver is securely switched off.

## Undervoltage

When detected the L9916 switches on the lamp.

## Low voltage

When detected the L9916 enters in stand by condition. All drivers are OFF including the lamp driver.

Figure 7. VB over and under voltage


### 2.4.2 Pin "SENSE"

The system battery can be directly sensed using the SENSE input. If the voltage on SENSE pin goes below the SENSE $_{\text {disc }}$ threshold a sense disconnection is detected. If voltage of ${ }^{+}+$ is SENSE $_{\text {diff }}$ greater than voltage of SENSE, the L9916 will use the B+ value to control the set point voltage, otherwise use SENSE voltage.

The device detects a 'sense loss' when either the sense disconnection or the voltage difference comparator is active.

It's possible to select if the voltage setpoint is increased by 1 V in case of sense loss.
Table 9. Electrical characteristics - Pin "SENSE"

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | SENSE $_{\text {disc,12 }}$ | Sense disconnected when Sense voltage is below | - | 7.5 | 8.5 | 9.5 | V |
| 2 | SENSE ${ }_{\text {disc, } 24}$ |  | - | 11 | 12 | 13 | V |
| 3 | SENSE $_{\text {diff,1 }}$ | Voltage difference (B+) - Sense | - | 1.2 | 1.6 | 2.1 | V |
| 4 | SENSE $_{\text {diff,2 }}$ |  | - | 1.8 | 2.3 | 2.9 | V |
| 5 | Delta VB ${ }_{\text {SP }}$ | Voltage setpoint increase in case of SENSE loss | - | 0.9 | 1 | 1.1 | V |
| 6 | Filter time S to B+ | Digital filter time to switch from $S$ to $B+$ regulation | - | 0.8 | 1 | 1.2 | ms |
| 7 | Filter time $\mathrm{B}+\text { to } \mathrm{S}$ | Digital filter to switch from B+ to $S$ regulation | - | 160 | 200 | 240 | ms |
| 8 | $\mathrm{I}_{\text {SENSE_PD }}$ | Pull down current | - | 10 | 45 | 80 | $\mu \mathrm{A}$ |

### 2.4.3 Pin "IGN"

The IGN pin is an input that can be used to wake-up the device in place of the Lamp and the selection is performed by OTP. The usage of one wake-up input (L or IGN) excludes the other one. The signal applied to IGN is compared to a threshold VIGN the result is used to turn the device on. If not used this pin must be left open.

Table 10. Electrical characteristics - Pin "IGN"

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| 1 | VIGN $_{\text {ON }}$ | Voltage threshold input <br> comparator to switch on | - | 6 | 8 | 10 | V |
| 2 | VIGN $_{\text {OFF }}$ | Voltage threshold input <br> comparator to switch off | - | 1.3 | 1.6 | 1.9 | V |
| 3 | VIGN $_{\text {OFFHYS }}$ | Voltage threshold input <br> comparator to go in logic <br> state $=$ High Z | - | VIGN $_{\text {OFF }}$ <br> +0.25 | VIGN $_{\text {OFF }}$ <br> +0.35 | VIGN $_{\text {OFF }}$ <br> +0.45 | V |
| 4 | $\mathrm{R}_{\text {pu }}$ | Pull-up resistance | - | 26 | 52 | 80 | $\mathrm{k} \Omega$ |

Figure 8. Pin "IGN" diagrams


### 2.4.4 Wake-up behaviour

In this chapter it is described the device wake-up behaviour depending on the signal on IGN, L, PH pin and the selection between Mode1 and Mode2 done through OTP programming.

The device wakes-up through L pin or IGN pin.
Note: $\quad$ The IGN input pin is able to read 3 level voltages (see Figure 8), the below values for IGN are intended as follows:

1 = input voltage higher then VIGNIT $T_{\text {ON }}$
0 = input voltage lower then VIGNIT OFF
High-Z = input voltage between VIGNIT OFF and VIGNIT $_{\text {ON }}$

IGN input not used (see Section 2.4.3).
Table 11. Self-start function enabled

| IGN | $\mathbf{L}$ | PH | Regulator |
| :---: | :---: | :---: | :---: |
| $X$ | 0 | Not present | OFF |
| $X$ | 0 | Present | ON |
| $X$ | 1 | Not present | ON |
| $X$ | 1 | Present | ON |

IGN input not used (see Section 2.4.3).
Table 12. Self-start function disabled

| IGN | L | PH | Regulator |
| :---: | :---: | :---: | :---: |
| $X$ | 0 | Not present | OFF |
| $X$ | 0 | Present | OFF |
| $X$ | 1 | Not present | ON |
| $X$ | 1 | Present | ON |

In case of condition $L=1, P H=$ present, regulator $=O N$ the $L$ signal becomes $=0$, the regulator goes in OFF condition.
IGN input used (see Section 2.4.3).
Table 13. Self-start function enabled (Mode 1)

| IGN | $\mathbf{L}$ | PH | Regulator |
| :---: | :---: | :---: | :---: |
| 0 | $X$ | $X$ | OFF |
| 1 | $X$ | $X$ | ON |
| High-Z | $X$ | Not present | OFF |
| High-Z | $X$ | Present | ON |

IGN input used (see Section 2.4.3).
Table 14. Self-start function disabled (Mode 1)

| IGN | $\mathbf{L}$ | PH | Regulator |
| :---: | :---: | :---: | :---: |
| 0 | $X$ | $X$ | OFF |
| 1 | $X$ | $X$ | ON |
| High-Z | $X$ | Not present | OFF |
| High-Z | $X$ | Present | OFF |

IGN input used (see Section 2.4.3).
Table 15. Self-start function enabled (Mode 2)

| IGN | $\mathbf{L}$ | PH | Regulator |
| :---: | :---: | :---: | :---: |
| 0 | $X$ | Not present | OFF |
| 0 | $X$ | Present | ON |

Table 15. Self-start function enabled (Mode 2) (continued)

| IGN | $\mathbf{L}$ | PH | Regulator |
| :---: | :---: | :---: | :---: |
| 1 | $X$ | $X$ | ON |
| High-Z | $X$ | Not present | OFF |
| High-Z | $X$ | Present | ON |

Note: $\quad X=$ Don't care .

### 2.4.5 Pin "FM"

The pin Field Monitor is used to communicate to the ECU the information about the activity on the field.

Figure 9. Pin "FM" circuit and waveform


Table 16. Electrical characteristics - Pin "FM"

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | VM ${ }_{\text {LVs }}$ | Low voltage saturation | I-sink $=14 \mathrm{~mA}$ | 0.9 | 1.2 | 1.5 | V |
| 2 | $\mathrm{fM}_{\text {SW,1 }}$ | Field monitor frequency switch Option1 | - | - | $\mathrm{fF}_{\text {SW }}$ | - | Hz |
|  | $\mathrm{fM}_{\text {Sw,2 }}$ | Field monitor frequency switch Option2 | Not available if the Field switching freq $=125 \mathrm{~Hz}$ | - | $\mathrm{fF}_{\text {Sw }} / 2$ | - | Hz |
| 3 | $\mathrm{DM}_{\text {DCR }}$ | Field monitor duty cycle range | In case of duty cycle clamp selected | 5 | - | 95 | \% |
| 4 | $\mathrm{IM}_{\text {lim }}$ | Field monitor current limitation | - | 25 | 50 | 75 | mA |
| 5 | $\mathrm{T}_{\text {lim }}$ | Duration of current limitation | - | 50 | - | 100 | $\mu \mathrm{s}$ |
| 6 | $\mathrm{TF}_{\text {retry }}$ | Retry time in case of overcurrent |  | 30 | 40 | 50 | ms |
| 7 | $\mathrm{l}_{\text {leak_FM }}$ | Leakage current | - | - | - | 3 | $\mu \mathrm{A}$ |

The Field Monitor switching frequency can be selected to be the same as the field driver switching frequency or one half of it, using the OTP cells
The polarity of the Field Monitor signal can be selected "direct" or "reverse" respect to the field driver signal, using the OTP cells

The duty cycle of the Field Monitor signal can be selected to have maximum and minimum values of $95 \%$ and $5 \%$ in case the field signal has a duty cycle higher than $95 \%$ or lower than $5 \%$, using the OTP cells.

In case of short to $V B$ the current output is limited to $I M_{\text {lim }}$ for a time equal to $T_{\text {lim }}$ then the output is switched off and turned on again after $\mathrm{TF}_{\text {retry, }}$

The duty cycle presented on FM is a replica, of a frequency depending on the OTP selection, of the duty cycle into the Field in order to provide to the ECU the information of the load connected to the alternator.

The duty cycle presented on FM is different from the signal on the Field in the following cases:

In the pre-excitation state FM always it has a fixed duty cycle equal to DFPreex (12.5\%) whatever the battery voltage is.
When the device exits the pre-excitation, for all the time in which the phase regulation is active, on pin FM a signal with a duty cycle of $6 \%$ is presented. If the phase regulation persists over time the device continues to show a 6\% duty cycle. Elapsed the phase regulation, the duty cycle of FM becomes equal to the duty cycle on the Field.
In the following requests of phase regulation, the Field executes the request while the FM continues to send the information of the entity of the current load. The activation of the field needed to implement the phase regulation is excluded from the account of the duty cycle: for example, if the battery voltage rises above the set point as a consequence of a load disconnection and a phase regulation is required, the field executes while FM continues to have a duty cycle of $0 \%$ (or its clamp value) that is the real load on the alternator in that moment.

### 2.4.6 Pin "PH"

The PH pin is the input for the phase signal coming from the alternator.
The $\mathrm{VP}_{\mathrm{HTh}}, \mathrm{VP}_{\mathrm{LTh}}$ are the voltage thresholds for the phase input detection used when the L pin or the IGN pin is connected. In case of self-start there are 4 couples of thresholds that can be selected and that are used for the first turn-on, once the speed exceeds the SPSS, $x$ threshold the device will switch to the $\mathrm{VP}_{\mathrm{HTh}}, \mathrm{VP}_{\mathrm{LTh}}$ thresholds.

Table 17. Electrical characteristics - Pin "PH"

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $\mathrm{VP}_{\text {HTh }}$ | High voltage threshold of hysteresis input buffer in case L or IGN connected and after self-start | - | 1.3 | 1.45 | 1.6 | V |
| 2 | VP LTh | Low voltage threshold of hysteresis input buffer in case L or IGN connected and after self-start | - | 0.85 | 1 | 1.15 | V |
| 3 | VP ${ }_{\text {HTh_SS1 }}$ | High voltage threshold of hysteresis input buffer for self-start switch-on | - | 0.25 | 0.37 | 0.45 | V |
| 4 | VPLTh_ss1 | Low voltage threshold of hysteresis input buffer for self-start switch-on | - | 0.15 | 0.24 | 0.35 | V |
| 5 | VP ${ }_{\text {HTh_SS2 }}$ | High voltage threshold of hysteresis input buffer for self-start switch-on | - | 0.6 | 0.76 | 0.9 | V |
| 6 | VP ${ }_{\text {LTh_S }}$ S 2 | Low voltage threshold of hysteresis input buffer for self-start switch-on | - | 0.35 | 0.45 | 0.6 | V |
| 7 | VP ${ }_{\text {HTh_Ss3 }}$ | High voltage threshold of hysteresis input buffer for self-start switch-on | - | 1.1 | 1.22 | 1.35 | V |
| 8 | VP ${ }_{\text {LTh__S }}$ | Low voltage threshold of hysteresis input buffer for self-start switch-on | - | 0.85 | 1 | 1.15 | V |
| 9 | VPHTh_SS4 | High voltage threshold of hysteresis input buffer for self-start switch-on | - | 1.3 | 1.45 | 1.6 | V |
| 10 | VP LTh_SS4 | Low voltage threshold of hysteresis input buffer for self-start switch-on | - | 0.85 | 1 | 1.15 | V |
| 11 | R $\mathrm{P}_{\mathrm{pd}}$ | Pull-down resistor (in addition to $\mathrm{IP}_{\text {pull-dw }}$ ) | - | 7 | 15 | 22 | k $\Omega$ |
| 12 | TP ${ }_{\text {SR }}$ | Spike rejection time | - | 70 | 125 | 180 | us |
| 13 | $I P_{\text {pull-dw }}$ | Pull-down current (in addition to $\mathrm{RP}_{\text {pd }}$ ) | - | 1.5 | 2 | 3.5 | mA |

Table 17. Electrical characteristics - Pin "PH" (continued)

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 14 | $\mathrm{SP}_{\text {HPrex, } 1}$ | High rpm Threshold to exit pre-excitation Option1 ${ }^{(1)}$ | - | Typ-10\% | 900 | Typ+10\% | rpm |
| 15 | SP ${ }_{\text {HPrex,2 }}$ | High rpm Threshold to exit pre-excitation Option2 ${ }^{(1)}$ | - | Typ-10\% | 1200 | Typ+10\% |  |
| 16 | $\mathrm{SP}_{\text {HPrex,3 }}$ | High rpm Threshold to exit pre-excitation Option3 ${ }^{(1)}$ | - | Typ-10\% | 1500 | Typ+10\% |  |
| 17 | SP ${ }_{\text {HPrex, } 4}$ | High rpm Threshold to exit pre-excitation Option4 ${ }^{(1)}$ | - | Typ-10\% | 1800 | Typ+10\% |  |
| 18 | $\mathrm{SP}_{\mathrm{HYS}}$ | Speed hysteresis | - | Typ-10\% | 200 | Typ+10\% | rpm |
| 19 | SP ${ }_{\text {LPrex }}$ | Low speed threshold to enter pre excitation | - | $\mathrm{SP}_{\text {HPrex }}-$ SPhys |  |  | rpm |
| 20 | $\mathrm{SP}_{\mathrm{SS}, 1}$ | Self-start rpm Threshold options ${ }^{(1)}$ | - | Typ-10\% | 1200 | Typ+10\% | rpm |
| 21 | $\mathrm{SP}_{\mathrm{SS}, 2}$ |  | - | Typ-10\% | 1500 | Typ+10\% |  |
| 22 | $\mathrm{SP}_{\mathrm{SS}, 3}$ |  | - | Typ-10\% | 2800 | Typ+10\% |  |
| 23 | $\mathrm{SP}_{\mathrm{SS}, 4}$ |  | - | Typ-10\% | 3200 | Typ+10\% |  |
| 24 | $\mathrm{SP}_{\mathrm{LRC}, 1}$ | rpm Threshold to exit/enter in LRC options | - | Typ-10\% | 1500 | Typ+10\% | rpm |
| 25 | $\mathrm{SP}_{\text {LRC, } 2}$ |  | - | Typ-10\% | 2800 | Typ+10\% |  |
| 26 | $\mathrm{SP}_{\text {LRC, } 3}$ |  | - | Typ-10\% | 3000 | Typ+10\% |  |
| 27 | SP ${ }_{\text {LRC, } 4}$ |  | - | Typ-10\% | 3200 | Typ+10\% |  |
| 28 | VP ${ }_{\text {prHTh12V }}$ | High phase regulation voltage threshold 12 Vapp | - | 6.7 | 7.7 | 8.7 | V |
| 29 | $\mathrm{VP}_{\text {prLTh12V }}$ | Low phase regulation voltage threshold 12 Vapp | - | 3 | 4 | 5 | V |
| 30 | $\mathrm{VP}_{\text {prHTh24V }}$ | High phase regulation voltage threshold 24 Vapp | - | 14 | 15.5 | 17 | V |
| 31 | $\mathrm{VP}_{\text {prLTh24V }}$ | Low phase regulation voltage threshold 24 Vapp | - | 5.2 | 6.2 | 7.2 | V |
| 32 | TP ${ }_{\text {OFF }}$ | Max windows time to detect 4 phase periods | - | 57 | 64 | 71 | ms |

1. Recommended: $S P_{S S} \geq S P_{\text {HPrex }}$.

To convert phase frequency $(\mathrm{Hz})$ to rotation speed (rpm) according to alternator poles pair number ( N ), use the following equation:

$$
\text { rotation speed }(\mathrm{rpm})=\text { phase frequency }(\mathrm{Hz}) * 60 / \mathrm{N}(\mathrm{rpm} \text { spread: } \pm 10 \%)
$$

## Principle of phase regulation

When VB is above the set-point voltage, the field driver is controlled to keep the phase peak voltage from falling below $\mathrm{VP}_{\text {prTh. }}$. If the phase peak voltage drops below $\mathrm{VP}_{\mathrm{H} T \mathrm{~h}}$, the phase regulation does not work. During the Phase regulation the field is driven with a fixed dutycycle and frequency, $T_{\text {on }}=40 \mathrm{~ms}$ and $T_{\text {off }}=40 \mathrm{~ms}$. If during the $\mathrm{T}_{\text {on }}$ phase the $\mathrm{VP}_{\mathrm{prTh}}$ threshold is reached the field is switched off even if the 40 ms have not elapsed.

The phase regulation is performed in both cases, either the high level phase signal does not cross the $\mathrm{VP}_{\text {prTh }}$ threshold or the low level phase signal does not cross the VP LTh threshold.

Figure 10. Phase regulation in case VPH signal does not cross VP ${ }_{\text {prTh }}$


GAPG1803160852PS

Figure 11. Phase regulation in case VPH signal does not cross VP LTh


### 2.4.7 Pin "F"

The pin $F$ is the output of the high side PowerMOS used to drive the field load.
Table 18. Electrical characteristics - Pin "F"

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | Ron | Ron field driver | $\mathrm{T}_{\mathrm{j}}=13{ }^{\circ} \mathrm{C}$; $\mathrm{I}_{\text {sunk }}=4.5 \mathrm{~A}$ |  |  | 130 | $\mathrm{m} \Omega$ |
| 2 | VF ${ }_{\text {diode }}$ | Freewheeling diode | $\mathrm{I}_{\text {sourced }}=5 \mathrm{~A}$ | -2 |  |  | V |
| 3 | $1 \mathrm{~F}_{\text {leak }}$ | Field leakage current | $\mathrm{VB}=50 \mathrm{~V}$; VF $=0$ | -5 |  | 5 | $\mu \mathrm{A}$ |
| 4 | IFovp | Field driver overcurrent protection | $\mathrm{T}_{\mathrm{j}}=-40^{\circ} \mathrm{C}$ | 9 |  | 18 | A |
|  |  |  | $\mathrm{T}_{\mathrm{j}}=25^{\circ} \mathrm{C}$ | 8.5 |  | 18 | A |
|  |  |  | $\mathrm{T}_{\mathrm{j}}=130{ }^{\circ} \mathrm{C}$ | 8 |  | 18 | A |
| 5 | $\mathrm{l}_{\text {ocft }}$ | Over-current filter time | - | 2.5 | 5 | 7.5 | $\mu \mathrm{s}$ |
| 5 | TF ${ }_{\text {retry }}$ | Retry time in case of over-current |  | 30 | 40 | 50 | ms |
| 6 | VFondet | Voltage threshold ON-state detection | - | 0.9 | 1.1 | 1.3 | V |
| 7 | $\mathrm{fF}_{\text {SW,1 }}$ | Field switching frequency options | - | -10\% | 125 | +10\% | Hz |
| 8 | $\mathrm{fF}_{\text {SW,2 }}$ |  | - | -10\% | 250 | +10\% |  |
| 9 | $\mathrm{fF}_{\text {SW, }}$ |  | - | -10\% | 333 | +10\% |  |
| 10 | $\mathrm{fF}_{\text {SW,4 }}$ |  | - | -10\% | 400 | +10\% |  |
| 11 | $\mathrm{T}_{\text {fall }}$ | Voltage slew rate for field driver <br> (Measurement is performed between $80 \%$ and $20 \%$ of the slope | $\begin{aligned} & \mathrm{VB}=14 \mathrm{~V} \\ & \mathrm{R}=270 \Omega \end{aligned}$ | 1 | 4 | 10 | $\mu \mathrm{s}$ |
| 12 | $\mathrm{T}_{\text {rise }}$ | Voltage slew rate for field driver <br> (Measurement is performed between $20 \%$ and $80 \%$ of the slope) | $\begin{aligned} & \mathrm{VB}=14 \mathrm{~V} \\ & \mathrm{R}=270 \Omega \end{aligned}$ | 1 | 4 | 10 | $\mu \mathrm{s}$ |
| 13 | $\mathrm{DF}_{\text {Preex }}$ | Field duty cycle in pre-excitation | - | 11 | 12.5 | 14 | \% |
| 14 | $\mathrm{DF}_{\text {SS }}$ | Field duty cycle in self start | Applied when only "ph" signal is present and speed< $\mathrm{SP}_{\mathrm{SS}}$ | 4 | 6 | 8 | \% |

Table 18. Electrical characteristics - Pin "F" (continued)

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 15 | TF ${ }_{\text {LRCUP, } 1}$ | Load Response Control Time (0\% to $100 \%$ DC) | - | 2.12 | 2.5 | 2.88 | s |
| 16 | TF ${ }_{\text {LRCUP,2 }}$ |  | - | 2.55 | 3 | 3.45 |  |
| 17 | TF ${ }_{\text {LRCUP,3 }}$ |  | - | 4.25 | 5 | 5.75 |  |
| 18 | TF ${ }_{\text {LRCUP,4 }}$ |  | - | 5.1 | 6 | 6.9 |  |
| 19 | TF ${ }_{\text {LRCUP, } 5}$ |  | - | 6.37 | 7.5 | 8.63 |  |
| 20 | TF ${ }_{\text {LRCUP, } 6}$ |  | - | 7.65 | 9 | 10.35 |  |
| 21 | TF ${ }_{\text {LRCUP,7 }}$ |  | - | 8.5 | 10 | 11.5 |  |
| 22 | TF ${ }_{\text {LRCUP, }}$ |  | - | 10.2 | 12 | 13.8 |  |
| 23 | DF ${ }_{\text {LRCUP }}$ | Positive Duty Cycle vs. time variation | - | 100 / TF LRCUP |  |  | \%/s |
| 24 | TF ${ }_{\text {LRCDW }}$ | Load Response Control Time (100\% to 0\% DC) | - | TF ${ }_{\text {LRCUP }}$ * 3 / 8 |  |  | s |
| 25 | DF ${ }_{\text {LRCDW }}$ | Negative Duty Cycle vs. time variation | - | - 100 / TF LRCDW |  |  | \%/s |
| 26 | DF ${ }_{\text {LRCEnab }}$ | Current vs. Previous Duty Cycle Variation to Enable LRC Function (internal duty cycle sampled every 1 ms ) | - | 4.1 | 6.7 | 9.2 | \% |
| 27 | DF ${ }_{\text {LRCBZ }}$ | Blind Zone | Test conditions: <br> Battery: VB = SetPoint - 0.3; <br> SDT: Disabled <br> Key/Ignition: Key-ON (VIGN $\geq$ VIGNON) <br> Phase: VPH $\geq$ VPRHT <br> Test: <br> RPM change from SPLPrex -100 <br> to SPHPrex +100 ; <br> F Duty-cycle has to change from $12.5 \%$ to $12.5 \%+B Z$; | 2.6 | 5.3 | 7.8 | \% |
| 28 | TF ${ }_{\text {SDT }, 1}$ | Start Delay Time options | - | 0.4 | 0.5 | 0.6 | s |
|  | $\mathrm{TF}_{\text {SDT, } 2}$ |  | - | 2.12 | 2.5 | 2.88 |  |
|  | TF ${ }_{\text {SDT, } 3}$ |  | - | 4.25 | 5 | 5.75 |  |
|  | TF ${ }_{\text {SDT, } 4}$ |  | - | 8.5 | 10 | 11.5 |  |

## Self-start function

In case no other wake up source is detected except the phase signal the device wakes up by self-start. During the self-start phase, the device will apply the $\mathrm{DF}_{S S}$ field duty cycle on the field if the speed is $<\mathrm{SP}_{S S}$. Above $\mathrm{SP}_{S S}$ the device will work with the normal regulation.

## Load response control function (LRC)

Figure 12. Load response control function (LRC)


When an electrical load is applied in the system application, a drop in the regulated voltage (VB) occurs and the alternator reacts by increasing the output current. If the LRC function is active then the alternator output current is controlled by the Field current variation strategy, that is directly linked to the duty cycle on Field signal.
The LRC function can operate when the alternator runs at low speed (the PH signal frequency has to be lower than $\mathrm{fP}_{\mathrm{LRC}}$ ) and it is activated when a positive variation between current duty cycle on Field and its previous duty cycle values is higher than the DF LRCEnab When the LRC function is required, the duty cycle increases slowly with the defined slope $D F_{\text {LRCUP }}$ starting from the previous duty cycle increased by the fixed value $D F_{\text {LRCBZ }}$. The actual duty cycle management during a LRC insertion is shown in the Figure 12. Once the LRC function is started it completes the required ramp even if the alternator speed becomes higher than the $\mathrm{SP}_{\mathrm{LRC}}$.

### 2.4.8 Pin "L"

The $L$ pin is used to detect the key presence and to drive the warning lamp. It can also drive an optional relay for auxiliary loads.

Figure 13. Pin "L" circuit diagram


The $L$ pin is used to drive the fault indicator lamp and the optional auxiliary load relay.
The device exits the stand-by mode when the switch "Key" is closed (i.e. VL>VL ${ }_{H T h}$ ).
The Lamp is driven by an internal low side N-channel MOSFET whereas the relay is driven by an internal high-side N -channel MOSFET.

The current in Low side driver is limited to $\mathrm{IL}_{\text {limLS }}$ for a time $\mathrm{TL}_{\mathrm{OC}}$, then the driver is switched off. After $\mathrm{T}_{\text {retry }}$ the low side driver is turned on again.

To reduce the power dissipation in the lamp driver while it is ON, the following strategy is implemented: after key-on, as soon as the $L$ pin voltage overcomes $V L_{\text {th }}$, after $T_{\text {filter }}$ time it is brought to $V L_{\text {sat }}$ voltage (between its drain and source), then the key status is verified every $\mathrm{T}_{\text {wait }}$ time intervals within a $\mathrm{T}_{\text {chk }}$ time window.

During this window, if the key is switched on the $L$ voltage reaches the $\mathrm{VL}_{\mathrm{HTh}}$ value and immediately returns to the $\mathrm{VL}_{\text {sat }}$ value before the window expiration whereas, if the key is no longer active, the $L$ voltage cannot increase and remains below key detection threshold $\left(\mathrm{VL}_{\mathrm{HT}}\right)$.

Figure 14. Pin "L" waveform diagrams


The high side driver is switched off if the current overcomes the $\mathrm{IL}_{\mathrm{OVCHS}}$ for $\mathrm{T}_{\mathrm{LH}}$ filter time.
Table 19. Electrical characteristics - Pin "L"

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| 1 | VL $_{\text {SAT }}$ | Lamp Driver <br> saturation <br> Voltage | $\mathrm{I}_{\text {sunk }}=300 \mathrm{~mA}$ | - | - | 0.3 | V |
| 2 | $\mathrm{R}_{\text {onL }}$ | Ron Lamp <br> driver | $\mathrm{T}_{\mathrm{j}}=175^{\circ} \mathrm{C} ; \mathrm{I}_{\text {sunk }}=300 \mathrm{~mA}$ | - | - | 1 | $\Omega$ |
| 3 | $\mathrm{I}_{\text {pulldw }}$ | Pull down <br> current | - | 0.6 | 1 | 1.4 | mA |
| 4 | $\mathrm{~T}_{\text {filt }}$ | Key-on filter <br> time | - | - | 40 | - | $\mu \mathrm{s}$ |
| 5 | $\mathrm{~T}_{\text {chk }}$ | Key presence <br> check time <br> window | Digital window time | - | - | 1 | ms |

Table 19. Electrical characteristics - Pin "L" (continued)

| \# | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 6 | $\mathrm{T}_{\text {wait }}$ | Key presence interval | - | 34 | 40 | 46 | ms |
| 7 | $\mathrm{V} \mathrm{L}_{\mathrm{HTh}}$ | High Voltage Threshold keyON detector | - | 0.8 | 0.9 | 1 | V |
| 8 | IL $\mathrm{limLS}^{\text {, } 12 \mathrm{~V}}$ | LS current limitation threshold | - | 1.2 | - | 2.4 | A |
| 9 | IL $\mathrm{limLS}^{\text {a } 24 \mathrm{~V}}$ | LS current limitation threshold | - | 0.6 | - | 1.2 | A |
| 10 | TL ${ }_{\text {delay }}$ | Turn ON delay time | - | - | - | 100 | us |
| 11 | TLoc | Maximum time duration of linear current limitation | Over-current Threshold | - | 30 | 35 | ms |
| 12 | TL ${ }_{\text {retry }}$ | Retry time in case of overcurrent duration > TLoc |  | 510 | 600 | 690 | ms |
| 13 | $\mathrm{VH}_{\text {sat }}$ | Relay Driver saturation Voltage | $\mathrm{V}(\mathrm{B}+)=12.6 \mathrm{~V}$; $\mathrm{I}_{\text {sourced }}=0.3 \mathrm{~A}$ | - |  | 1 | V |
| 14 | $\mathrm{Ron}_{\mathrm{HS}}$ | Ron HS relay driver | $\mathrm{T}_{\mathrm{j}}=175^{\circ} \mathrm{C} ; \mathrm{I}_{\text {source }}=300 \mathrm{~mA}$ | - | - | 3 | $\Omega$ |
| 15 | ILovchs | HS Overcurrent threshold | - | 1 | - | 2 | A |
| 16 | $\mathrm{TL}_{\text {HSretry }}$ | Retry time in case of overcurrent | - | 30 | 40 | 50 | ms |
| 17 | $\mathrm{TL}_{\text {Hfilter }}$ | HS Over current filter time | - | - | 5 | - | $\mu \mathrm{s}$ |
| 18 | $\mathrm{TL}_{\text {alarm, }}$ | Alarm validation time | - | 0.42 | 0.5 | 0.58 | s |
| 19 | TL ${ }_{\text {alarm,2 }}$ |  | - | 0.85 | 1 | 1.15 |  |
| 20 | TL $\mathrm{a}_{\text {alarm, }}$ |  | - | 1.27 | 1.5 | 1.73 |  |
| 21 | $\mathrm{TL}_{\text {alarm, } 4}$ |  | - | 1.7 | 2 | 2.3 |  |

Note: In case Lamp Driver exceeds safety temperature, it will be turned OFF,

### 2.5 Alarm detection

The device turns ON the lamp after the validation time ( $\mathrm{TL}_{\text {alarm }}$ ) if one of the conditions in the below table is verified.

Table 20. Alarm detection

| Detection condition | Related pin/ Description | Configurable |
| :---: | :---: | :---: |
| $\mathrm{VPH}<\mathrm{VP}_{\text {prTh }}$ and $\mathrm{VB}<\mathrm{VB}_{\text {SP }}$ | F driver or its connection degraded | No |
| IF>IF ${ }_{\text {OVP }}$ | F shortened to GND (Over-current on F driver) | No |
| $\mathrm{VB}>\mathrm{VB}_{\text {SP }}$ and $\mathrm{VF}>\mathrm{VF}_{\text {ONdet }}$ | F shortened to B | No |
| $\mathrm{VB}>\mathrm{VB}_{\text {Ovp12V }} \mid \mathrm{VB}>\mathrm{VB}_{\text {Ovp24V }}{ }^{(1)}$ | Battery sensor on B pin or F driver degraded | No |
| $V B<V B_{u V}$ | Low B Voltage (Battery under-voltage) | No |
| VSENSE< SENSE $_{\text {disc }}$ | SENSE connection loss | Yes |
| $\mathrm{VIGNIT}_{\text {OFF }}<\mathrm{VIGN}^{\text {- VIGNIT }}$ ON | VIGN open | Yes |

1. In case of IGN interface enabled with self-start function enabled, if the IGN pin is in High_z (Mode 1 and Mode 2) or at 0 level (only mode 2), the lamp and relay drivers are disabled if an overvoltage occurs.

In case of VPH<VPHTh or PH frequency $<\mathrm{f}_{\text {PLPrex }}$ the device enters in pre-excitation state and turns ON the lamp.
Also during self start when SPHprex<PH<SPss the device will enters in wakeup mode and turn on the lamp.

### 2.6 End of line test mode

Through the FM pin the device can enter a special "test mode" where some time consuming functions are not present.

The functions that are not present are:
LRC, SDT and TL $_{\text {alarm }}$
To enter the "test mode" it is necessary to bring the FM pin voltage to $\mathrm{V}_{\mathrm{FM}}$ _TM and the device must not be in overvoltage condition (VB> $\mathrm{VBO}_{\mathrm{vp}}$ ).

Table 21. End of line test mode

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| 1 | V FM_TM | Voltage threshold to enter in <br> EOL test mode | - | 37 | 41 | 45 | V |

### 2.7 OTP programming

The programming of the device parameters is achieved by using the $I^{2} \mathrm{C}$ interface which is implemented on two pins (PH and DFM) having also such alternative functions.
Here below some general information on the $I^{2} C$

The L9916 is $I^{2} \mathrm{C}$ slave device, so SCL (clock line from $\mathrm{I}^{2} \mathrm{C}$ Master device) is input, while SDA (data line) is bidirectional to allow transmit/receive operations to/from $I^{2} \mathrm{C}$ master device. Both SCL and SDA lines are connected to a positive power supply voltage via pullup resistor. The $I^{2} \mathrm{C}$ protocol defines the proper operations of the link. When the bus is free, both lines are High (pulled-up). The output stages of the devices connected to the bus must have an open-drain or open-collector to perform the wired-AND function. The maximum link rate is $400 \mathrm{kbit} / \mathrm{s}$.
The required voltage for the OTP programming is $15.7 \mathrm{~V} \pm 0.5 \mathrm{~V}$.
For the programming procedure refer to the dedicated document.
Figure 15. $I^{2} \mathrm{C}$ interface circuit


Table 22. OTP programming electrical characteristics

| $\#$ | Symbol | Parameter | Test condition / Note | Min | Typ | Max | Unit |
| :---: | :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| 1 | $\mathrm{f}_{\text {SCL }}$ | Clock frequency | - | 10 |  | 400 | kHz |
| 2 | VIL | Input low voltage | - |  |  | 1.5 | V |
| 3 | VIH | Input high voltage | - | 2.3 |  |  | V |

### 2.7.1 External pull up resistor sizing

Given:
$C_{p}=$ wiring (line) capacitance
f = Target frequency
$R_{p}=$ Pull up resistor
$\mathrm{T}=1 / \mathrm{f}$
$\mathrm{T}=$ Time constant
and assuming that for proper operation $T<T / 4$, if the target frequency (SCL) is 100 kHz and the line capacitance is $\mathrm{C}_{\mathrm{p}}=100 \mathrm{pF}$ we got:
$\mathrm{T}=1 / \mathrm{f}=10 \mu \mathrm{~s}$
$\mathrm{T}=\mathrm{R}_{\mathrm{p}}{ }^{*} \mathrm{C}_{\mathrm{p}}<\mathrm{T} / 4=2.5 \mathrm{~ms} \Rightarrow \mathrm{R}_{\mathrm{p}}<2.5 \mathrm{~ms} / 100 \mathrm{pF}=25 \mathrm{k} \Omega$

## 3 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK ${ }^{\circledR}$ packages, depending on their level of environmental compliance. ECOPACK ${ }^{\circledR}$ specifications, grade definitions and product status are available at: www.st.com. ECOPACK ${ }^{\circledR}$ is an ST trademark.

### 3.1 Multiwatt 8 (pin 5 GND) package information

Figure 16. Multiwatt 8 (pin 5GND) package outline


Table 23. Multiwatt 8 (pin 5GND) package mechanical drawing

| Ref | Dimensions |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Millimeters |  |  |  | Inches $^{(1)}$ |  |  |
|  | Min. | Typ. | Max. | Min. | Typ. | Max. |  |
| A | - | - | 5 | - | - | 0.1969 |  |
| B | - | - | 2.65 | - | - | 0.1043 |  |
| C | - | - | 1.6 | - | - | 0.0630 |  |
| E | 0.49 | - | 0.55 | 0.0193 | - | 0.0217 |  |
| F | 0.78 | - | 0.85 | 0.0307 | - | 0.0335 |  |
| G | 2.40 | 2.54 | 2.68 | 0.0945 | 0.1000 | 0.1055 |  |
| G1 | 17.64 | 17.78 | 17.92 | 0.6945 | 0.7000 | 0.7055 |  |
| H1 | 19.6 | - | - | 0.7717 | - | - |  |
| H2 | - | - | 20.2 | - | - | 0.7953 |  |
| L | 20.35 |  | 20.65 | 0.8012 |  | 0.8130 |  |
| L2 | 17.05 | 17.20 | 17.35 | 0.6713 | 0.6772 | 0.6831 |  |
| L3 | 17.25 | 17.5 | 17.75 | 0.6791 | 0.6890 | 0.6988 |  |
| L4 | 10.3 | 10.7 | 10.9 | 0.4055 | 0.4213 | 0.4291 |  |
| L7 | 2.65 | - | 2.9 | 0.1043 | - | 0.1142 |  |
| S | 1.9 | - | 2.6 | 0.0748 | - | 0.1024 |  |
| S1 | 1.9 | - | 2.6 | 0.0748 | - | 0.1024 |  |
| U | 0.40 | - | 0.55 | 0.0157 | - | 0.0217 |  |
| Z | 0.70 | - | 0.85 | 0.0276 | - | 0.0335 |  |
| diam1 | 3.65 | - | 3.85 | 0.1437 | - | 0.1516 |  |

1. Values in inches are converted from mm and rounded to 4 decimal digits.

### 3.2 Multiwatt 8 marking information

Figure 17. Multiwatt 8 marking information


Parts marked as 'ES' are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

## 4 Revision history

Table 24. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :---: |
| 20-Apr-2016 | 1 | Initial release. |
| 13-Jun-2016 | 2 | Modified on the Table 7: Electrical characteristics - Pin " $B$ " the test conditions for " $\Delta \mathrm{VB}_{\text {load }, 1}$ and $\Delta \mathrm{VB}_{\text {load }, 2 \text { " parameter. }}$ <br> Modified on the Table 8: VB over and under voltage parameters the values for "VB ${ }_{\text {OVft" }}$ parameter. <br> Updated: <br> Figure 2: State diagram on page 6; <br> Figure 8: Pin "IGN" diagrams on page 16. |
| 21-Aug-2017 | 3 | Updated: <br> Sections: <br> - Section 1.3: External component required on page 7, <br> - Section 2.1: Absolute maximum ratings on page 8, <br> - Section 2.4: Electrical characteristics on page 9, <br> - Section 2.5: Alarm detection on page 29. <br> Figures: <br> - Figure 4: Regulated Voltage variation with the load on page 12, <br> - Figure 5: Application 12 V - Default setpoint with $-3.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ on page 12, <br> - Figure 6: Application 24 V - Default setpoint with $-3.5 \mathrm{mV} /{ }^{\circ} \mathrm{C}$ on page 13, <br> - Figure 7: VB over and under voltage on page 14, <br> - Figure 8: Pin "IGN" diagrams on page 16, <br> - Figure 14: Pin "L" waveform diagrams on page 27. <br> Tables: <br> - Table 1: Device summary on page 1 <br> - Table 3: Absolute maximum ratings on page 8, <br> - Table 6: Thermal data on page 9; <br> - Table 7: Electrical characteristics - Pin "B" on page 9; <br> - Table 8: VB over and under voltage parameters on page 13, <br> - Table 10: Electrical characteristics - Pin "IGN" on page 16, <br> - Table 17: Electrical characteristics - Pin "PH" on page 20, <br> - Table 18: Electrical characteristics - Pin "F" on page 23, <br> - Table 19: Electrical characteristics - Pin "L" on page 27. |

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