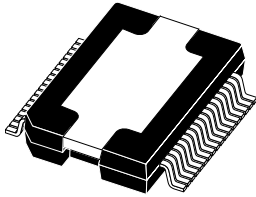


## 500 W FFX digital amplifier power stage



PowerSO-36 with exposed pad up

### Features

- Output Power at 56 V supply voltage
  - 2 x 250 W at 10% THD + N into 6  $\Omega$  BTL
  - 2 x 200 W at 10% THD + N into 8  $\Omega$  BTL
  - 4 x 130 W at 10% THD + N into 3  $\Omega$  SE
  - 4 x 100 W at 10% THD + N into 4  $\Omega$  SE
  - 1 x 480 W at 10% THD + N into 3  $\Omega$  PBTL
  - 1 x 380 W at 10% THD + N into 2  $\Omega$  PBTL
- Output Power at 52 V supply voltage
  - 2 x 200 W at 10% THD + N into 6  $\Omega$  BTL
  - 4 x 100 W at 10% THD + N into 3  $\Omega$  SE
  - 1 x 400 W at 10% THD + N into 2  $\Omega$  PBTL
- < 0.1% THD + N at 1 W
- PSO-36 thermally enhanced package
- Minimum input / output pulse width distortion
- High efficiency power stage (> 90%) with 190 m $\Omega$  RdsON
- CMOS compatible logic inputs
- Integrated self protection circuits including overtemperature, undervoltage, overvoltage, overload, short-circuit
- EMI compliant when used with recommended system design
- Automatic recovery mode after fault conditions

#### Maturity status link

[STA516BE](#)

#### Device summary

<b>Order code</b>	STA516BETR
<b>Operating temperature range</b>	0 to 90 °C
<b>Package</b>	PowerSO-36 (EPU)
<b>Packing</b>	Tape and reel

### Applications

- Home theater
- DVD receiver
- Mini / Micro Audio systems

### Description

The **STA516BE** is a monolithic quad half-bridge stage in multipower BCD technology. The device can be used as dual-bridge or reconfigured, by connecting pin CONFIG to pins VDD, as a single-bridge with double-current capability or as a half-bridge (binary mode) with half-current capability.

A cost-effective, high fidelity audio system can be designed using ST chipset, including a modulator (e.g. the STA309A or STA321) and the **STA516BE**. This system only requires a simple passive LC demodulation filter to deliver high-quality, high efficiency audio amplification with prove EMI compliance. The efficiency of this digital amplifier is greater than 90% into 8  $\Omega$  speakers, enabling the use of smaller power supplies and heatsinks.

The **STA516BE** has an innovative integrated protection system, safeguarding the device against different fault conditions that could damage the overall system.

## 1 General information

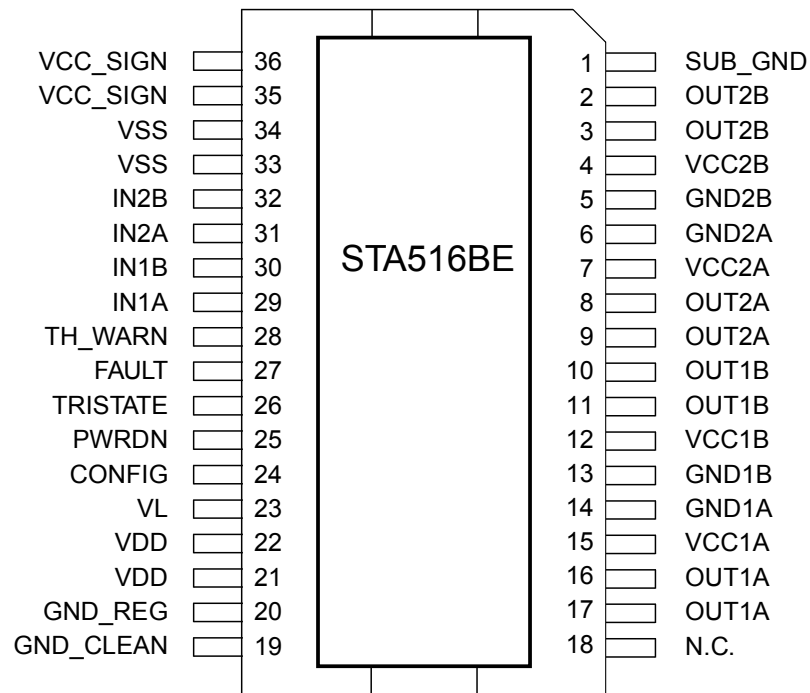
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The STA516BE is a second generation, high performance, integrated stereo digital amplifier power stage with improved protection system. It is capable of driving a 6 W bridge tied load (BTL) at up to 250 W per channel with very low noise at the output, low THD+N and low idle power dissipation.

The STA516BE is available in PowerSO-36 slug up package.

The package contains a heat slug that is located on the top side of the device for convenient thermal coupling to the heatsink.

## 2 Pin description

**Figure 1. Pin out (top view)**

**Table 1. Pin function**

Pin	Name	Type	Description
1	GND_SUB	PWR	Substrate ground
2, 3	OUT2B	O	Output half bridge 2B
4	VCC2B	PWR	Positive supply
5	GND2B	PWR	Negative supply
6	GND2A	PWR	Negative supply
7	VCC2A	PWR	Positive supply
8, 9	OUT2A	O	Output half bridge 2A
10, 11	OUT1B	O	Output half bridge 1B
12	VCC1B	PWR	Positive supply
13	GND1B	PWR	Negative supply
14	GND1A	PWR	Negative supply
15	VCC1A	PWR	Positive supply
16, 17	OUT1A	O	Output half bridge 1A
18	N.C.	-	No internal connection
19	GND_CLEAN	PWR	Logical ground
20	GND_REG	PWR	Ground for regulator $V_{DD}$
21, 22	VDD	PWR	5-V regulator referred to ground
23	VL	PWR	High logical state setting voltage, $V_L$

Pin	Name	Type	Description
24	CONFIG	I	Configuration pin: 0: normal operation 1: bridges in parallel (OUT1A = OUT1B, OUT2A = OUT2B (if IN1A = IN1B, IN2A = IN2B))
25	PWRDN	I	Standby pin: 0: low-power mode 1: normal operation
26	TRISTATE	I	Hi-Z pin: 0: all power amplifier outputs in high impedance state 1: normal operation
27	FAULT	O	Fault pin advisor (open-drain device, needs pull-up resistor): 0: fault detected (short circuit or thermal, for example) 1: normal operation
28	TH_WARN	O	Thermal warning advisor (open-drain device, needs pull-up resistor): 0: temperature of the IC >130 °C 1: normal operation
29	IN1A	I	Input of half bridge 1A
30	IN1B	I	Input of half bridge 1B
31	IN2A	I	Input of half bridge 2A
32	IN2B	I	Input of half bridge 2B
33, 34	VSS	PWR	5-V regulator referred to +V <sub>CC</sub>
35, 36	VCC_SIGN	PWR	Signal positive supply

### 3 Electrical characteristics

**Table 2. Absolute maximum ratings**

Symbol	Parameter	Value	Unit
$V_{CC\_MAX}$	DC supply voltage (pins 4, 7, 12, 15)	65	V
$V_{max}$	Maximum voltage on pins 23 to 32	5.5	V
$T_{j\_MAX}$	Operating junction temperature	0 to 150	°C
$T_{stg}$	Storage temperature	-40 to 150	°C

**Caution:** Stresses beyond those listed under “Absolute maximum ratings” make cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “Recommended operating condition” are not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. In the real application, power supply with nominal value rated inside recommended operating conditions, may experience some rising beyond the maximum operating condition for short time when no or very low current is sinked (amplifier in mute state). In this case the reliability of the device is guaranteed, provided that the absolute maximum rating is not exceeded.

**Table 3. Thermal data**

Symbol	Parameter	Min.	Typ.	Max.	Unit
$T_{j\_case}$	Thermal resistance junction to case (thermal pad)	-	1	2.5	°C/W
$T_{warn}$	Thermal warning temperature	-	130	-	°C
$T_{jSD}$	Thermal shut-down junction temperature	-	150	-	°C
$t_{hSD}$	Thermal shut-down hysteresis	-	25	-	°C

**Table 4. Recommended operating conditions**

Symbol	Parameter	Min.	Typ.	Max.	Unit
$V_{CC}$	Supply voltage for pins PVCCA, PVCCB	10	-	60	V
$T_{amb}$	Ambient operating temperature	0	-	90	°C

Unless otherwise stated, the test conditions for [Table 5. Electrical characteristics](#) below are  $V_L = 3.3\text{ V}$ ,  $V_{CC} = 50\text{ V}$  and  $T_{amb} = 25\text{ °C}$

**Table 5. Electrical characteristics**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$R_{dsON}$	Power P-channel/N-channel MOSFET $R_{dsON}$	$I_{dd} = 1\text{ A}$	-	190	240	mΩ
$I_{dss}$	Power P-channel/N-channel leakage $I_{dss}$	-	-	-	50	μA
$g_N$	Power P-channel $R_{dsON}$ matching	$I_{dd} = 1\text{ A}$	95	-	-	%
$g_P$	Power N-channel $R_{dsON}$ matching	$I_{dd} = 1\text{ A}$	95	-	-	%
$Dt_s$	Low current dead time (static)	see <a href="#">Figure 2. Test circuit</a>	-	10	20	ns
$Dt_d$	High current dead time (dynamic)	$L = 22\text{ μH}$ , $C = 470\text{ nF}$ , $R_L = 8\text{ Ω}$ , $I_{dd} = 4.5\text{ A}$ see <a href="#">Figure 3. Current dead-time test circuit</a>	-	-	50	ns
$t_{d ON}$	Turn-on delay time	Resistive load	-	-	100	ns
$t_{d OFF}$	Turn-off delay time	Resistive load	-	-	100	ns
$t_r$	Rise time	Resistive load see <a href="#">Figure 2. Test circuit</a>	-	-	25	ns
$t_f$	Fall time	Resistive load see <a href="#">Figure 2. Test circuit</a>	-	-	25	ns
$V_{IN-High}$	High level input voltage	-	-	-	$V_L / 2 + 300\text{ mV}$	V
$V_{IN-Low}$	Low level input voltage	-	$V_L / 2 - 300\text{ mV}$	-	-	V
$I_{IN-H}$	High level input current	$V_{IN} = V_L$	-	1	-	μA
$I_{IN-L}$	Low level input current	$V_{IN} = 0.3\text{ V}$	-	1	-	μA
$I_{PWRDN-H}$	High level PWRDN pin input current	$V_L = 3.3\text{ V}$	-	35	-	μA
$V_{Low}$	Low logical state voltage (pins PWRDN, TRISTATE) (see <a href="#">Table 6. Threshold switching voltage variation with voltage on pin VL</a> )	$V_L = 3.3\text{ V}$	0.8	-	-	V
$V_{High}$	High logical state voltage (pins PWRDN, TRISTATE) (see <a href="#">Table 6. Threshold switching voltage variation with voltage on pin VL</a> )	$V_L = 3.3\text{ V}$	-	-	1.7	V
$I_{VCC.PWRDN}$	Supply current from $V_{CC}$ in power down	$V_{PWRDN} = 0\text{ V}$	-	-	2.4	mA

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{\text{FAULT}}$	Output current on pins FAULT, TH_WARN with fault condition	$V_{\text{pin}} = 3.3 \text{ V}$	-	1	-	mA
$I_{\text{VCC-HIZ}}$	Supply current from $V_{\text{CC}}$ in tristate	$V_{\text{TRISTATE}} = 0 \text{ V}$	-	22	-	mA
$I_{\text{VCC}}$	Supply current from $V_{\text{CC}}$ in operation, both channels switching)	Input pulse width = 50% duty, switching frequency = 384 kHz, no LC filters	-	70	-	mA
$I_{\text{OCP}}$	Overcurrent protection threshold $I_{\text{sc}}$ (short-circuit current limit) <sup>(1)</sup>	-	8.5	9.5	11	A
$V_{\text{UVP}}$	Undervoltage protection threshold	-	-	7	-	V
$V_{\text{OVP}}$	Overvoltage protection threshold	-	61	62.5	-	V
$t_{\text{pw\_min}}$	Output minimum pulse width	No load	50	-	110	ns

1. See specific application note number: AN1994

**Table 6. Threshold switching voltage variation with voltage on pin VL**

Voltage on pin VL, $V_{\text{L}}$	$V_{\text{LOW max.}}$	$V_{\text{HIGH min.}}$	Unit
2.7	1.05	1.65	V
3.3	1.4	1.95	V
5.0	2.2	2.8	V

**Table 7. Logic truth table**

Pin TRISTATE	Inputs as per Figure 3. Current dead-time test circuit		Transistors as per Figure 3. Current dead-time test circuit				Output mode
	INxA	INxB	Q1	Q2	Q3	Q4	
0	x	x	Off	Off	Off	Off	Hi Z
1	0	0	Off	Off	On	On	Dump
1	0	1	Off	On	On	Off	Negative
1	1	0	On	Off	Off	On	Positive
1	1	1	On	On	Off	Off	Not used

### 3.1 Test circuits

Figure 2. Test circuit

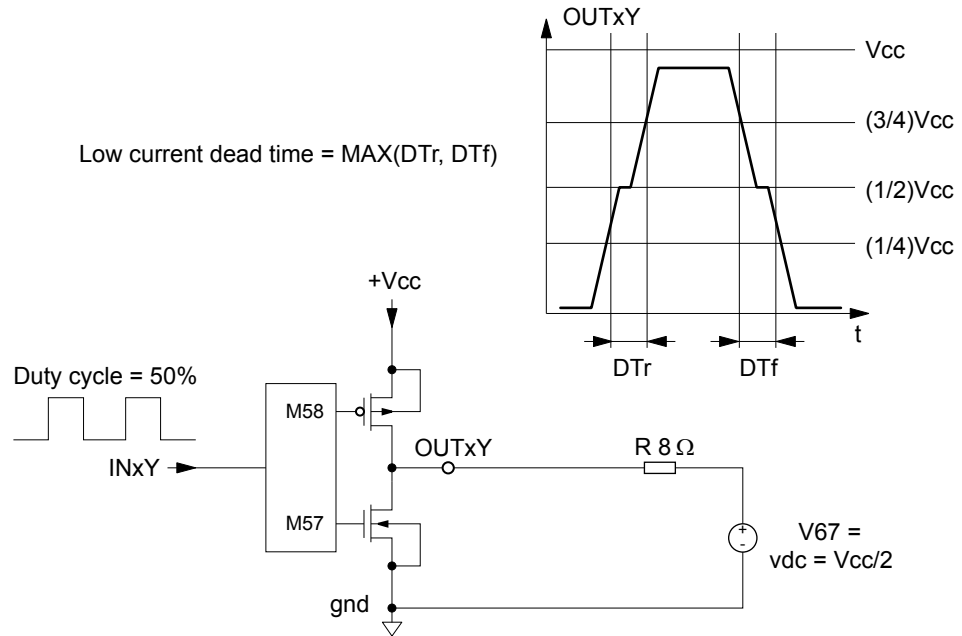
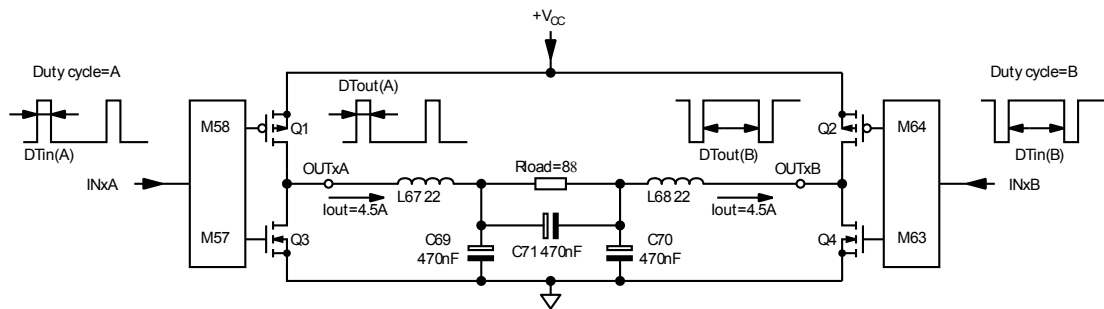


Figure 3. Current dead-time test circuit

$$\text{High Current Dead time for Bridge application} = \text{ABS}(\text{DT}_{\text{out}}(\text{A}) - \text{DT}_{\text{in}}(\text{A})) + \text{ABS}(\text{DT}_{\text{out}}(\text{B}) - \text{DT}_{\text{in}}(\text{B}))$$



Duty cycle A and B: Fixed to have DC output current of 4.5A in the direction shown in figure

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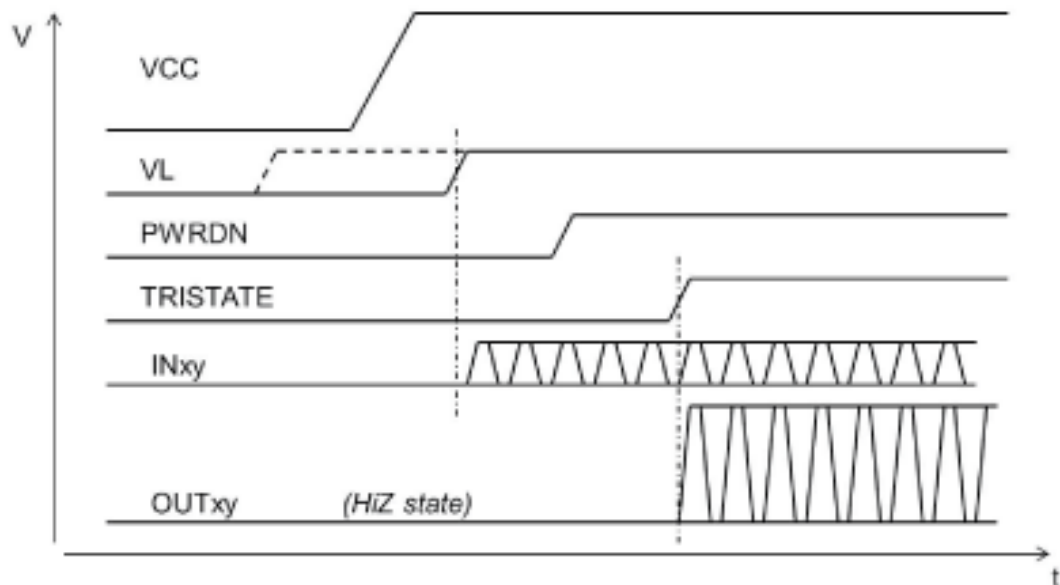
## 4 Power supply and control sequencing

To guarantee correct operation and reliability, the recommended power-on sequence as given below should be followed:

- Apply  $V_{CC}$  and  $V_L$ , in any order, keeping PWRDN low in this phase
- Release PWRDN from low to high, keeping TRISTATE low (until  $V_{DD}$  and  $V_{SS}$  are stable)
- Release TRISTATE from low to high

Always maintain PWM inputs  $IN_{xy} < V_L$ .

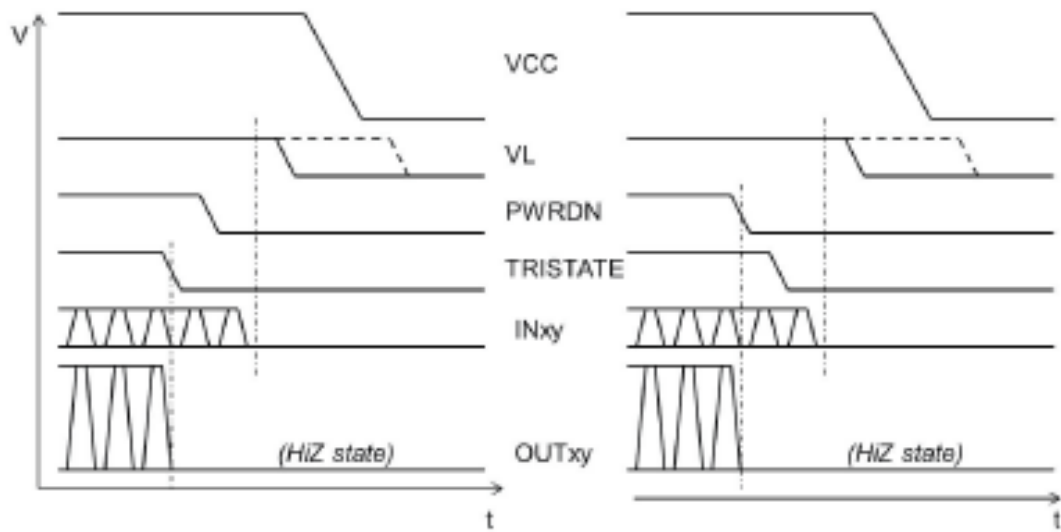
Figure 4. Power-ON sequence



$V_{CC}$  should be turned on before  $V_L$ . This prevents uncontrolled current flowing through the internal protection diode connected between  $V_L$  (logic supply) and  $V_{CC}$  (high power supply), which could result in damage to the device.

PWRDN must be released after  $V_L$  is switched on. An input signal can then be sent to the power stage.

Figure 5. Power-OFF sequence



## 5 Technical information

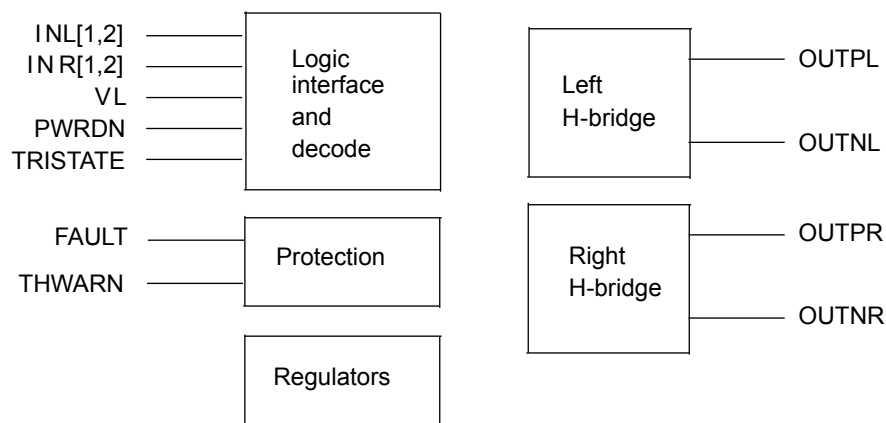
The STA516BE is a dual channel H-bridge that is able to deliver 200 W per channel (into  $R_L = 6 \Omega$  with THD = 10% and  $V_{CC} = 51 \text{ V}$ ) of audio output power very efficiently. It operates in conjunction with a pulse-width modulator driver such as the STA321 or STA309A.

The STA516BE converts ternary, phase-shift or binary-controlled PWM signals into audio power at the load. It includes a logic interface, integrated bridge drivers, high efficiency MOSFET outputs and thermal and short-circuit protection circuitry.

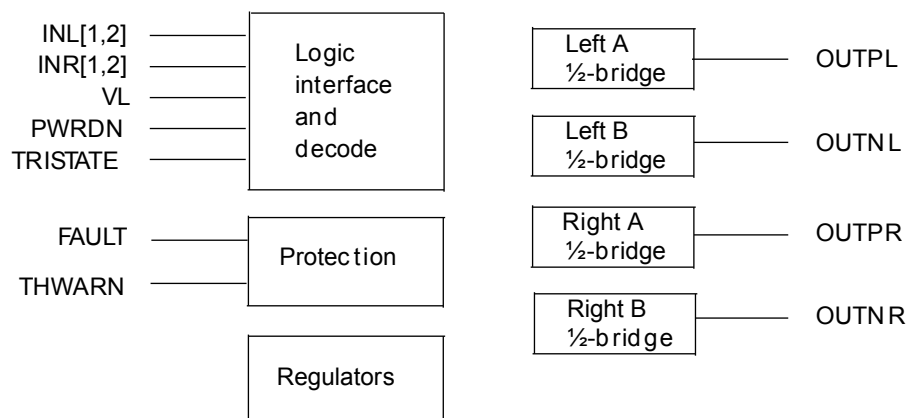
In differential mode (ternary, phase-shift or binary differential), two logic level signals per channel are used to control high-speed MOSFET switches to connect the speaker load to the input supply or to ground in a bridge configuration, according to the damped ternary modulation operation.

In binary mode, both full bridge and half bridge modes are supported. The STA516BE includes overcurrent and thermal protection as well as an undervoltage lockout with automatic recovery. A thermal warning status is also provided.

**Figure 6. Block diagram of full-bridge FFX<sup>®</sup> or binary mode**



**Figure 7. Block diagram of binary half-bridge mode**



### 5.1 Logic interface and decode

The STA516BE power outputs are controlled using one or two logic-level timing signals. In order to provide a proper logic interface, the VL input must operate at the same voltage as the FFX control logic supply.

## 5.2 Protection circuitry

The STA516BE includes protection circuitry for overcurrent and thermal overload conditions. A thermal warning pin (THWARN, pin 28, open drain MOSFET) is activated low when the IC temperature exceeds 130 °C, just in advance of thermal shutdown. When a fault condition is detected an internal fault signal immediately disables the output power MOSFETs, placing both H-bridges in a high-impedance state. At the same time the opendrain MOSFET of pin FAULT (pin 27) is switched on.

There are two possible modes subsequent to activating a fault.

- **Shutdown mode:** with pins FAULT (with pull-up resistor) and TRISTATE separate, an activated fault disables the device, signaling a low at pin FAULT output.
- The device may subsequently be reset to normal operation by toggling pin TRISTATE from high to low to high using an external logic signal.
- **Automatic recovery mode:** This is shown in the applications circuits below where pins FAULT and TRISTATE are connected together to a timeconstant circuit (R59 and C58).
- An activated fault forces a reset on pin TRISTATE causing normal operation to resume following a delay determined by the time constant of the circuit.
- If the fault condition persists, the circuit operation repeats until the fault condition is cleared.
- An increase in the time constant of the circuit produces a longer recovery interval. Care must be taken in the overall system design not to exceed the protection thresholds under normal operation.

## 5.3 Power outputs

The STA516BE power and output pins are duplicated to provide a low-impedance path for the device bridged outputs. All duplicate power, ground and output pins must be connected for proper operation.

The PWRDN or TRISTATE pin should be used to set all power MOSFETs to the highimpedance state during power-up until the logic power supply,  $V_L$ , has settled.

## 5.4 Parallel output / high current operation

When using the FFX mode output, the STA516BE outputs can be connected in parallel in order to increase the output current capability to a load. In this configuration the STA516BE can provide up to 240 W into a 3  $\Omega$  load.

This mode of operation is enabled with the pin CONFIG (pin 24) connected to pin VDD. The inputs are joined so that IN1A = IN1B, IN2A = IN2B and similarly the outputs OUT1A = OUT1B, OUT2A = OUT2B as shown in Figure 8. Typical audio application circuit (dual BTL).

## 5.5 Output filtering

A passive 2nd-order filter is used on the STA516BE power outputs to reconstruct the analog audio signal. System performance can be significantly affected by the output filter design and choice of passive components. A filter design for 6 or 8  $\Omega$  loads is shown in the application circuit of Figure 8. Typical audio application circuit (dual BTL), and for 3 or 4  $\Omega$  loads in Figure 9. Typical Mono-BTL (PBTL) configuration and Figure 10. Typical quad half-bridge configuration (Quad Single Ended).

## 6 Audio application circuits

Figure 8. Typical audio application circuit (dual BTL) shows a stereo-BTL configuration capable of giving 210 W per channel into a 6 Ω load at 10% THD with  $V_{CC} = 52$  V. This result was obtained using the STA309A+STA516B demo board.

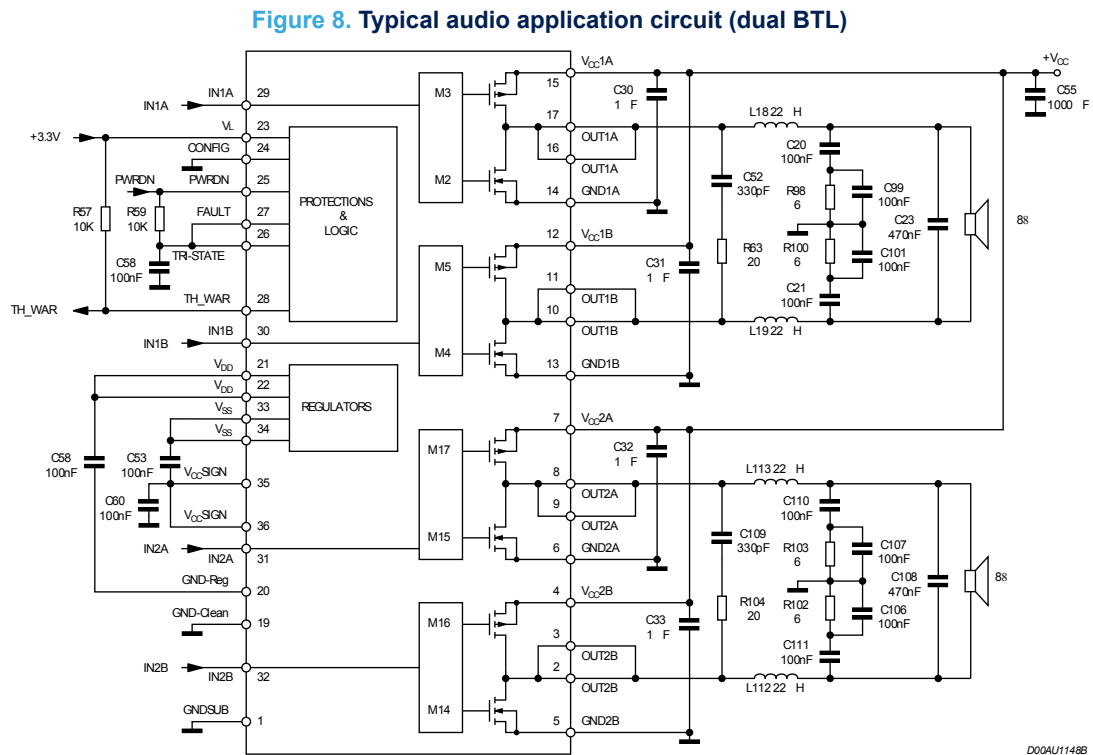
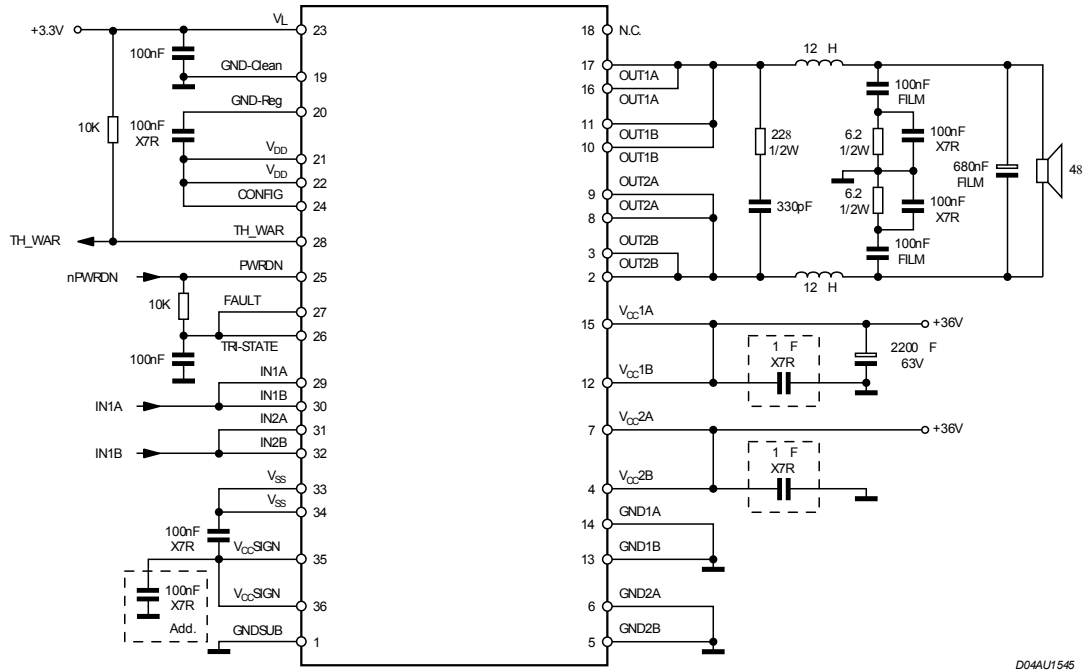


Figure 9. Typical Mono-BTL (PBTL) configuration below shows a single-BTL configuration capable of giving 400 W into a 3 Ω load at 10% THD with  $V_{CC} = 52$  V. STA516BE can also drive 2 Ω speakers as single-BTL configuration, to provide up to 280 W per channel at 10% THD with  $V_{CC} = 37$  V.

Figure 10. Typical quad half-bridge configuration (Quad Single Ended) below shows a quad-SE configuration capable of giving 110 W into a 3 Ω load at 10% THD with  $V_{CC} = 54$  V. STA516BE can also drive 2 Ω speakers as quad-SE configuration, to provide up to 80 W per channel at 10% THD with  $V_{CC} = 38$  V.

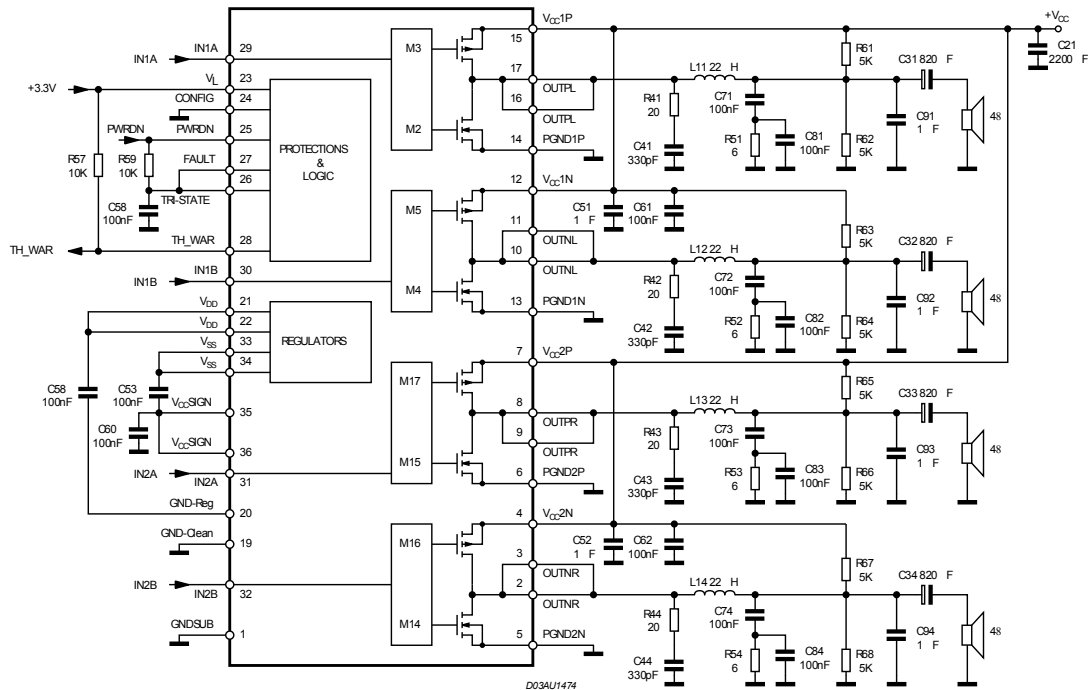
All results were obtained using the STA309A+STA516B demo board. Note that a PWM modulator as driver is required to feed the STA516BE.

Figure 9. Typical Mono-BTL (PBTL) configuration



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Figure 10. Typical quad half-bridge configuration (Quad Single Ended)



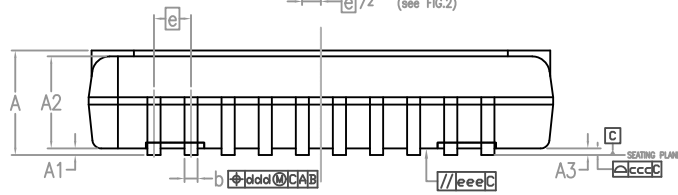
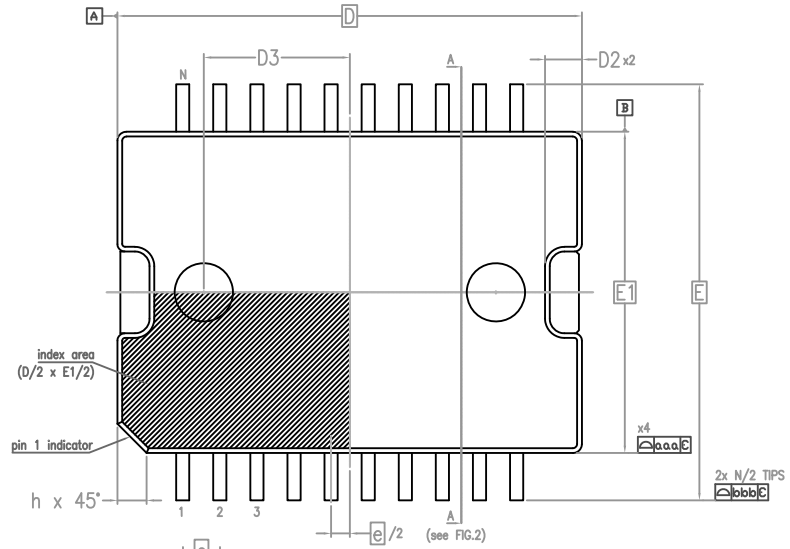
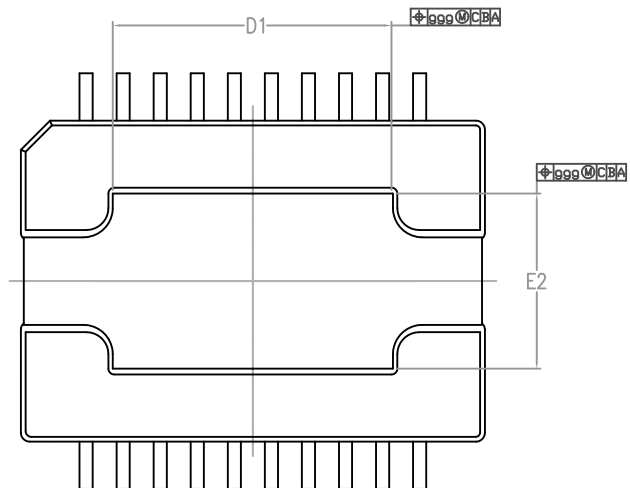
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For more information, refer to the application note AN1994.

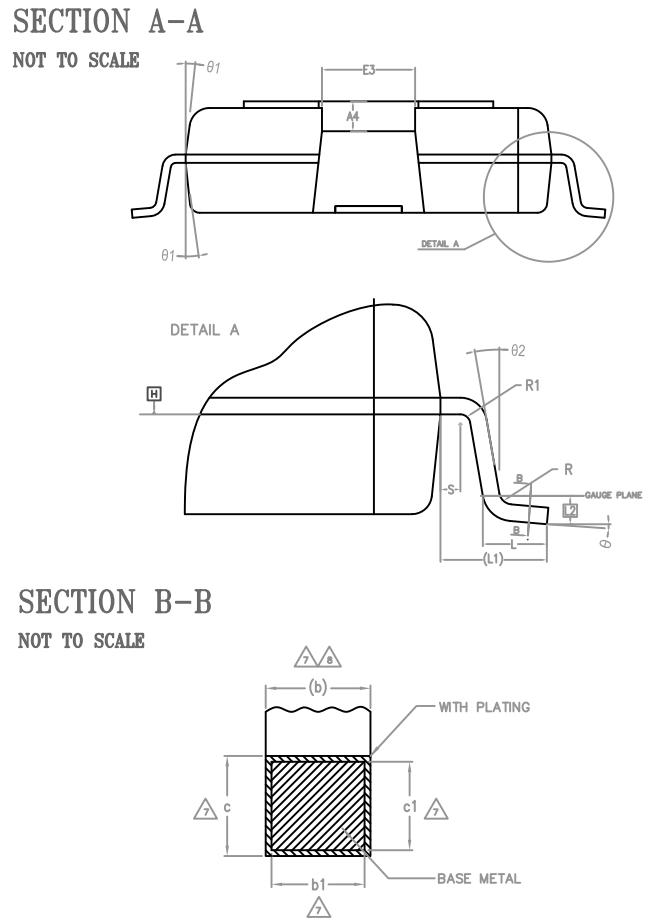
## 7 Package information

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In order to meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK is an ST trademark.

**7.1 PowerSO-36 exposed pad up package information**
**Figure 11. PowerSO-36 exposed pad up outline drawing**
**BOTTOM VIEW**

**TOP VIEW**




**Figure 12. PowerSO-36 section A-A and B-B package outline**

**Table 8. PowerSO-36 package mechanical data**

Symbol	mm		
	Min.	Typ.	Max.
$\theta$	0°		8°
$\theta 1$	5°		10°
$\theta 2$	0°		
A			3.41
A1	0.30		0.40
A2	3.10	3.14	3.18
A3		0.2	
A4	0.80		1.00
b	0.22		0.41
b1	0.22		0.38
c	0.23		0.32
c1	0.23	0.25	0.29

Symbol	mm		
	Min.	Typ.	Max.
D	15.90 BSC		
D1	VARIATIONS		
D2			1.00
D3		5.00	
e	0.65 BSC		
E	14.20 BSC		
E1	11.00 BSC		
E2	VARIATIONS		
E3			2.90
h			1.10
L	0.80		1.10
L1	1.60 REF		
L2	0.35 BSC		
N	36		
R	0.20		
R1	0.20		
s	0.25		

**Table 9. Tolerance of form and position**

Symbol	Databook
aaa	0.10
bbb	0.30
ccc	0.075
ddd	0.25
eee	0.10
ggg	0.25
Note	1.2

**Table 10. Variations**

Symbol	Databook			Opt.
	Min.	Typ.	Max.	
D1	9.40		9.80	A
E2	5.80		6.20	

## Revision history

**Table 11. Document revision history**

Date	Revision	Changes
02-Apr-2014	1	Initial release.
05-Jun-2020	2	Updated part number in device summary on the cover page.
16-Nov-2020	3	Updated <a href="#">Section 7.1</a> PowerSO-36 exposed pad up package information.

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