

2 A max constant current LED driver

Features

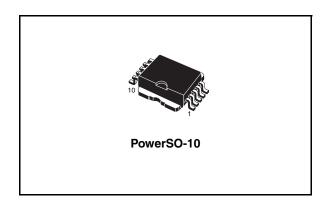
- Up to 40 V input voltage
- Less than 0.5 V voltage overhead
- Up to 2 A output current
- PWM dimming pin
- Shutdown pin
- LED disconnection diagnostic
- Slope control with external cap

Applications

- LED constant current supplying for varying input voltages
- Low voltage lighting
- Small appliances LED lighting
- Car LED lights

Description

The STCS2A is a BiCMOS constant current source designed to provide a precise constant current starting from a varying input voltage source. The main target is to replace discrete components



solution for driving LEDs in low voltage applications such as 5 V, 12 V or 24 V giving benefits in terms of precision, integration and reliability.

The current is set with external resistor up to 2 A with a \pm 10 % precision; a dedicated pin allows implementing PWM dimming. An external capacitor allows setting the slope for the current rise from tens of microseconds to tens of milliseconds allowing reduction of EMI.

An open-drain pin output provides information on load disconnection condition.

Table 1. Device summary

Order code	Package	Packaging	
STCS2ASPR	PowerSO-10	600 parts per reel	

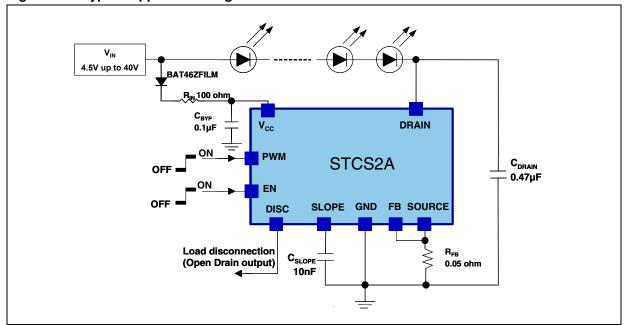
Contents STCS2A

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1 Application diagram

Figure 1. Typical application diagram for 2 A LED current



Pin configuration STCS2A

2 Pin configuration

Figure 2. Pin connections (top view)

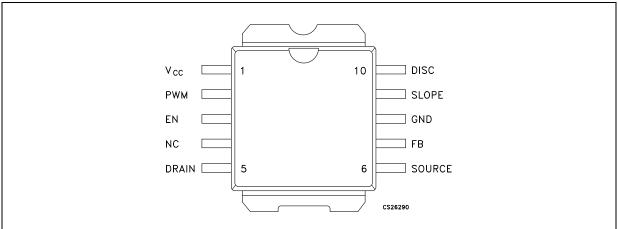


Table 2. Pin description

Pin n°	Symbol	Note
1	V _{CC}	Supply voltage
2	PWM	PWM dimming input
3	EN	Shutdown
4	NC	No connect
5	DRAIN	Internal N-MOSFET drain
6	SOURCE	Internal N-MOSFET source. Reference voltage is 100 mV. An external resistor between SOURCE and GND pins sets different current levels for different application needs
7	FB	Feedback input. The control loop regulates the current in such a way that the average voltage at the FB input is 100 mV (nominal). The cathode of the LED and a resistor to ground to set the LED current should be connected at this point
8	GND	Ground
9	SLOPE	Capacitor for slope control
10	DISC	Load disconnection flag (open drain)
	exp-pad	Internally connected to ground

STCS2A Maximum ratings

3 Maximum ratings

Table 3. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V _{CC}	DC supply voltage	-0.3 to +45	V
DRAIN	Drain pin	-0.3 to +45	V
SOURCE	Source pin	-0.3 to + 3.3	V
PWM, EN, DISC	Logic pins	-0.3 to + V _{CC} + 3.3	V
SLOPE, FB	Configuration pins	-0.3 to + 3.3	٧
ESD	Human body model (all pins)	±2	kV
T _J ⁽¹⁾	Junction temperature	-40 to 150	°C
T _{STG}	Storage temperature range	-55 to 150	°C

^{1.} T_J is calculated from the ambient temperature T_A and the power dissipation P_D according the following formula: $T_J = T_A + (P_D \ x \ R_{thJA})$. See *Figure 16.* for details of max power dissipation for ambient temperatures higher than 25 °C.

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Note:

Table 4.

Absolute maximum ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied.

Symbol	Parameter	PowerSO-10	Unit
R_{thJC}	Thermal resistance junction-case	2	°C/W
R _{thJA}	Thermal resistance junction-ambient ⁽¹⁾	50	°C/W
R _{thJA}	Thermal resistance junction-ambient (2)	35	°C/W
R _{thJA}	Thermal resistance junction-ambient (3)	12	°C/W

^{1.} FR4 with using the recommended pad-layout

Thermal data

^{2.} FR4 with heat sink on board (6 cm²).

^{3.} FR4 with copper-filled through holes and external heat sink applied.

Electrical characteristics STCS2A

4 Electrical characteristics

Table 5. Electrical characteristics (V_{CC} = 12 V; I_{O} = 100 mA; T_{J} = -40 °C to 125 °C; V_{DRAIN} = 1 V; C_{DRAIN} = 1 μ F; C_{BYP} = 100 nF typical values are at T_{A} = 25 °C, unless otherwise specified)

VCC Supply voltage range Supply voltage range A.5 A.5 A.5 VCC	Symbol	Parameter	eter Test conditions M		Тур.	Max.	Unit			
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	V _{CC}	Supply voltage range		4.5		40	V			
Regulation (percentage with respect to V _{CC} = 12 V)		Output current range		1		2000	mA			
respect to V _{CC} = 12 V) I _O = 100 mA; V _{DRAIN} = 1 V -1	Io	Output current	$R_{FB} = 50 \text{ m}\Omega$		2		Α			
CC				-1		+1	%			
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	V_{FB}	Feedback voltage	I _O = 0 to 2A	90	100	110	mV			
V _{CC} = 5 to 12V Shutdown Mode; V _{CC} = 12 to 40V Shutdown; V _{DRAIN} = 40 V Shutdown; V _{DRAIN} = 10 PwM frising, V _{CC} = 12 V PwM frising, V _{CC} = 10 PwM frising, V _{CC} = 10 PwM frising, V _{CC} = 12 V PwM frisin			On Mode		450	750				
$V_{DROP} \begin{tabular}{ c c c c c c c c c c c c c c c c c c c$	I _{CC}	I ·	1			1	μΑ			
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		- CC F - 1	The state of the s			3				
Leak_Drain Drain leakage current Shutdown; V_Drain 40 V 10	V	Dropout voltage (V to CND)	I _O = 100 mA		0.12	0.16	W			
$T_{P}/T_{F} \begin{array}{ l l l } \hline Rise/Fall time of the current on PWM transition & $C_{SLOPE} = 10 \text{ nF}, \\ T_{J} = -40 ^{\circ}\text{C} \text{ to } 105 ^{\circ}\text{C} & & 800 & & \mus \\ \hline \\ T_{D} \\ \hline \\ Delay on PWM signal (\textit{Figure 1}) & & & & & & & & \\ \hline \\ V_{PWM} rising, V_{CC} = 12 V \\ \hline \\ V_{CSLOPE} = floating & & & & & & \\ \hline \\ V_{PWM} falling, V_{CC} = 12 V \\ \hline \\ V_{CSLOPE} = floating & & & & & \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{CSLOPE} = floating & & & & \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DWM} falling, V_{CC} = 12 V \\ \hline \\ V_{DISC} = 5 V & & & & & & \\ \hline \\ DISC Turn-ON & & & & & \\ \hline \\ DISC Turn-OFF & & & & & \\ \hline \\ DISC Turn-OFF & & & & & \\ \hline \\ DISC Turn-OFF & & & & & \\ \hline \\ DISC Turn-OFF & & & & & \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & & \\ \hline \\ \hline \\ DISC Turn-OFF & & & & & \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & & \\ \hline \\ \hline \\ DISC Turn-OFF & & & & \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & \\ \hline \\ \hline \\ DISC Turn-OFF & & & & \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & \\ \hline \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & \\ \hline \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & \\ \hline \\ TI_{D} = -40 ^{\circ}\text{C} & & & \\ \hline \\ TI_{D} = -12 V \\ \hline \\ TI_{D} =$	V DROP	Dropout voitage (V _{DRAIN} to GND)	I _O = 2 A		0.58	0.9	7 V			
$ \begin{array}{c} T_{D} \\ T_{D} \\ \end{array} \begin{array}{c} Delay \ on \ PWM \ signal \ (\textit{Figure 1}) \\ \end{array} \begin{array}{c} V_{PWM} \ rising, V_{CC} = 12 \ V \\ C_{SLOPE} = \ floating \\ \end{array} \begin{array}{c} 3 \\ V_{PWM} \ rising, V_{CC} = 12 \ V \\ C_{SLOPE} = \ floating \\ \end{array} \begin{array}{c} 1.2 \\ \end{array} \begin{array}{c} \mu_{S} \\ \end{array} \\ \end{array} \\ DISC \\ \end{array} \begin{array}{c} Low \ level \ voltage \\ Leakage \ current \\ Load \ disconnection \ threshold \\ (V_{DRAIN}\text{-GND}) \\ \end{array} \begin{array}{c} DISC \ Turn-ON \\ DISC \ Turn-OFF \\ \end{array} \begin{array}{c} 110 \\ Thermal \\ Protection \\ \end{array} \begin{array}{c} Shutdown \ temperature \\ Hysteresis \\ \end{array} \begin{array}{c} DISC \ Turn-OFF \\ \end{array} \begin{array}{c} 110 \\ \end{array} \begin{array}{c} \circ C \\ \end{array} \\ \end{array} \begin{array}{c} \circ C \\ \end{array} \\ \end{array} \\ \begin{array}{c} C \\ \end{array} \\ \begin{array}{c} Logic \ inputs \ (PWM \ and \ EN) \\ \end{array} \begin{array}{c} V_{L} \\ U_{R} \ Input \ low \ level \\ U_{R} \ PWM \ leakage \ current \\ \end{array} \begin{array}{c} V_{EN} = 5 \ V; \ V_{PWM} = 5 \ V \\ V_{EN} = 40 \ V \\ \end{array} \begin{array}{c} 0.4 \\ V_{EN} = 40 \ V \\ \end{array} \begin{array}{c} V_{D} \\ \end{array} \begin{array}{c} 0.4 \\ V_{D} \\ \end{array} \begin{array}{c} \mu_{A} \\ \end{array} \end{array} $	LEAK _{DRAIN}	Drain leakage current	Shutdown; V _{DRAIN} = 40 V			10	μA			
$T_{D} = \begin{array}{c} D_{elay} \text{ on PWM signal (} \textit{Figure 1}) \\ \hline D_{elay} \text{ on PWM signal (} \text{ on PWM signal (} \textit{Figure 1}) \\ \hline D_{elay} \text{ on PWM signal (} on PWM $	T _R /T _F				800		μs			
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	Ŧ	Delay on DIMM signed (Figure 1)			3		- µs			
$DISC \begin{tabular}{ c c c c c c c c c c c c c c c c c c c$	'D	Delay of Pww Signal (Pigure 1)			1.2					
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$		Low level voltage	I _{SINK} = 5 mA		0.2	0.5	V			
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	DICC	Leakage current	V _{DISC} = 5 V			1	μA			
	DISC	Load disconnection threshold	DISC Turn-ON		75					
		(V _{DRAIN} -GND)	DISC Turn-OFF		110		IIIV			
Protection Hysteresis 25 Logic inputs (PWM and EN) V _L Input low level 0.4 V V _H Input high level 1.2 V EN, PWM leakage current V _{EN} = 5 V; V _{PWM} = 5 V 2 EN input leakage current V _{EN} = 40 V 60 μA	Thermal	Shutdown temperature			155		°C			
V_L Input low level 0.4 V V_H Input high level 1.2 V EN, PWM leakage current $V_{EN} = 5 \text{ V}; V_{PWM} = 5 \text{ V}$ 2 EN input leakage current $V_{EN} = 40 \text{ V}$ 60 μA	Protection	Hysteresis			25					
V_H Input high level 1.2 V EN, PWM leakage current $V_{EN} = 5 \text{ V}$; $V_{PWM} = 5 \text{ V}$ 2 EN input leakage current $V_{EN} = 40 \text{ V}$ 60 μA	Logic inputs (PWM and EN)									
EN, PWM leakage current $V_{EN} = 5 \text{ V}; V_{PWM} = 5 \text{ V}$ 2 EN input leakage current $V_{EN} = 40 \text{ V}$ 60 μ A	V _L	Input low level				0.4	V			
EN input leakage current V _{EN} = 40 V 60 μA	V _H	Input high level		1.2			V			
		EN, PWM leakage current	V _{EN} = 5 V; V _{PWM} = 5 V			2				
PWM input leakage current V _{PWM} = 40 V 120		EN input leakage current	V _{EN} = 40 V			60	μА			
		PWM input leakage current	V _{PWM} = 40 V			120				

Note: All devices 100 % production tested at $T_A = 25$ °C. Limits over the operating temperature range are guaranteed by design.

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STCS2A Timing

5 Timing

Figure 3. PWM and output current timing

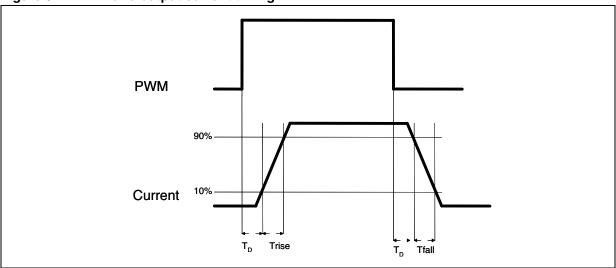
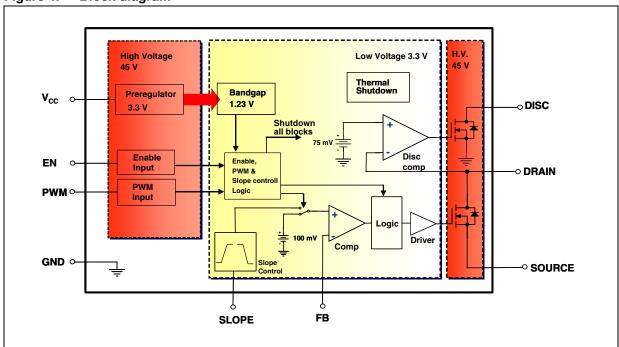


Figure 4. Block diagram



6 Typical performance characteristics

Figure 5. I_{DRAIN} vs V_{CC} , $T_A = 25^{\circ}C$

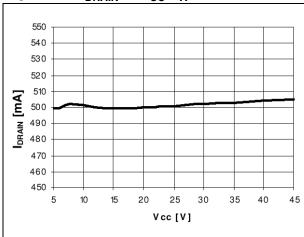


Figure 6. I_{DRAIN} vs R_{SET}

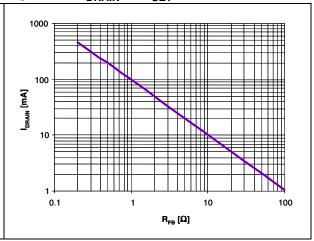
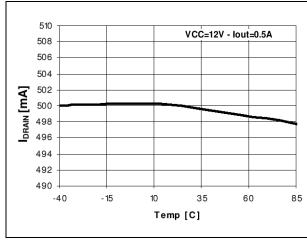


Figure 7. I_{DRAIN} vs temperature

Figure 8. V_{DROP} (including V_{FB}) vs temperature



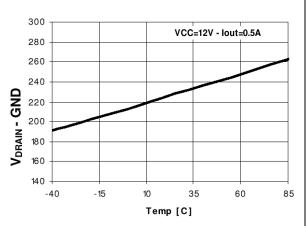


Figure 9. I_{CC} vs temperature

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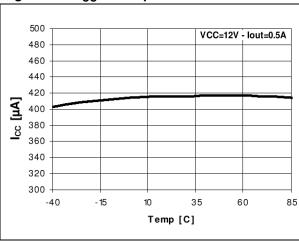


Figure 10. I_{CC} vs V_{CC}

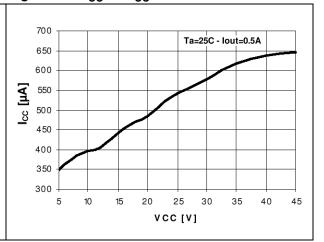


Figure 11. T_{rise}/T_{fall} vs C_{SLOPE} Figure 12. **Turn-on time** 10000 Trise 1000 Tfall Time [usec] 100 10 V_{EN} 1V/DIV I_{DRAIN} 0.2A/DIV VCC=12V - lout=0.5A DRAIN 1V/DIV Time 10µs/DIV 0,001 0,01 10 100 0,1 CSET [nF] M 10.0μs A Ch2 F 2.32 V 1.00 V Ch3 1.00 V Ch4 200mAΩ

Figure 13. Dimming operation: Rise

Figure 14. Dimming operation: Fall

V_{CC}=5V - I_{DRAIN}=0.5A - T_A=25 °C

V_{CC}=5V - I_{DRAIN}=0.5A - T_A=25 °C

V_{PWM} 1V/DIV
I_{DRAIN} 0.2A/DIV
V_{DRAIN} 1V/DIV
Time 20µs/DIV
Time 20µs/DIV

Ch3 1.00 V

C_{SLOPE} = 1 nF

Ch2 1.00 V Ch4 200mA Ω

Ch3 1.00 V

 $C_{SLOPE} = 1 \text{ nF}$

M 20.0µs A Ch2 J 1.30 V

Detail description STCS2A

7 Detail description

The STCS2A is a BiCMOS constant current source designed to provide a precise constant current starting from a varying input voltage source. The main target is to replace discrete components solution for driving LEDs in low voltage applications such as 5 V, 12 V or 24 V giving benefits in terms of precision, integration and reliability.

7.1 Current setting

The current is set with an external sensing resistor connected to the FB pin. The feedback voltage is 100 mV, then a low resistor value can be chosen reducing power dissipation. A value between 1 mA and 2 A can be set according to the resistor value the resulting output current has a tolerance of \pm 10%.

For instance, should one need a 700 mA LEDs current, R_F should be selected according to the following equation:

 $R_F = V_{FB} / I_{LEDs} = 100 \text{ mV} / 700 \text{ mA} = 142 \text{ m}\Omega$

7.2 Enable

When the enable pin is low the device completely off thus reducing current consumption to less than 1 μ A. When in shutdown mode, the internal main switch is off.

7.3 PWM dimming

The PWM input allows implementing PWM dimming on the LED current; when the PWM input is high the main switch will be on and vice versa. A typical frequency range for the input is from few Hertz to 50 kHz. The maximum dimming frequency is limited by the minimum rise/fall time of the current (obtained with $C_{SLOPE}=0$) which is around 4 μ s each. Above 50 kHz the current waveforms starts assuming a triangular shape.

While the PWM input is switching, the overall circuitry remains on, this is needed in order to implement two important features: short delay time and controlled slope for the current.

Since the PWM pin is controlling just the main switch, the overall circuitry is always on and it is able to control the delay time between the PWM input signal and the output current in the range of few μ s, this is important to implement synchronization among several light LED sources.

The rise and fall slope of the current is controlled by the C_{SLOPE} capacitor. The rise and fall time are linear dependent from the C_{SLOPE} capacitor value (see graph in typical characteristics). A controlled rise time has two main benefits: reducing EMI noise and avoid current spike at turn on.

When C_{SLOPE} is left floating, the internal switch is turned on at maximum speed, in this condition an overshoot can be present on the LED current before the system goes into regulation.

STCS2A Detail description

7.4 Diagnostic

When STCS2A is in on mode (EN is high), the device is able to detect disconnection or fail of the LED string monitoring V_{DRAIN} pin. If V_{DRAIN} is lower than 75 mV the DISC pin is pulled low regardless the PWM pin status. This information can be used by the system to inform that some problem happens in the LEDs.

8 Application information

8.1 Reverse polarity protection

STCS2A must be protected from reverse connection of the supply voltage. Since the current sunk from V_{CC} pin is in the range of 450 μ A a small diode connected to V_{CC} is able to protect the chip. Care must be taken for the whole application circuit, especially for the LEDs, in fact, in case a negative voltage is applied between V_{IN} and GND, a negative voltage will be applied to the LED string that must have a total breakdown voltage higher than the negative applied voltage in order to avoid any damage.

BAT46 or similar

V_{CC}

DRAIN

DISC

PWM

SLOPE GND FB SOURCE

R_{SENSE}

Figure 15. Reverse polarity condition

8.2 Thermal considerations

The STCS2A is able to control a LED current up to 2 A and able to sustain a voltage on the drain pin up to 40 V. Those operating conditions are however limited by thermal constraints.

The poor thermal conduction of epoxy FR4 boards does not permit to benefit of the outstanding thermal performance of the PowerSO-10.

In any case one way to improve the thermal conduction is the use of large heat spreader areas at the copper layer of the PC board. This leads to a reduction of thermal resistance to 30 - 36 °C/W for 3 to 6 cm^2 on-board heatsink.

Use of copper-filled through holes on conventional FR4 techniques increases the metallization and decreases thermal resistance accordingly. Using a configuration with 16 holes under the spreader of the package with a pitch of 1.8 mm and a diameter of 0.7 mm, the thermal resistance (junction - heatsink) can be reduced to 12 °C/W.

The thermal resistances shown in the Error! Reference source not found. section are the typical ones.

The power dissipation in the device can be calculated as follow:

$$P_D = (V_{DRAIN} - V_{FB}) \times I_{LED} + (V_{CC} \times I_{CC})$$

basing on this and on the thermal resistance and ambient temperature, the junction temperature can be calculated as:

$$T_{J} = R_{thJA} \times P_{D} + T_{A}$$

A typical application could be:

- Input voltage: 12 V;
- 3 white LEDs with an typical V_F = 3.6 V;
- LEDs current: 1000 mA;
- Package: PowerSO-10;
- T_A = 50 °C;

In this case the drain voltage is given by:

$$V_{DRAIN} = 12 - 3 \times 3.6 = 1.2 \text{ V}$$

end the power dissipated in the IC is the following:

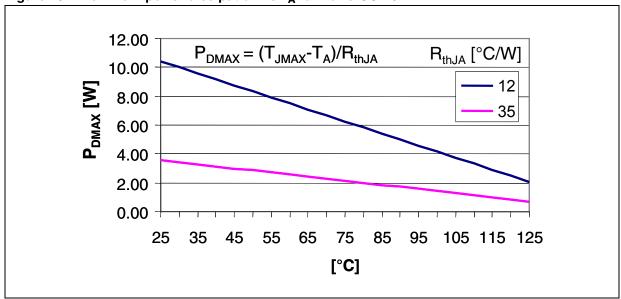
$$P_D = (1.2 - 0.1) \times 1 + 12 \times 0.5 \times 10^{-3} = 1.1 \text{ W}$$

With a thermal resistance junction-ambient equal to 12 °C/W the junction temperature is:

$$T_J = 12 \times 1.1 + 50 = 63 \,^{\circ}\text{C}.$$

The following pictures show the maximum power dissipation according to the ambient temperature:



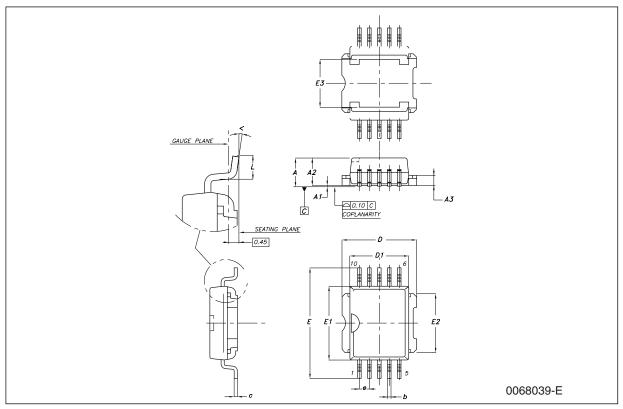


9 Package mechanical data

In order to meet environmental requirements, ST offers these devices in ECOPACK[®] packages. These packages have a lead-free second level interconnect. The category of second Level Interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com.

PowerSO-10 Mechanical Data

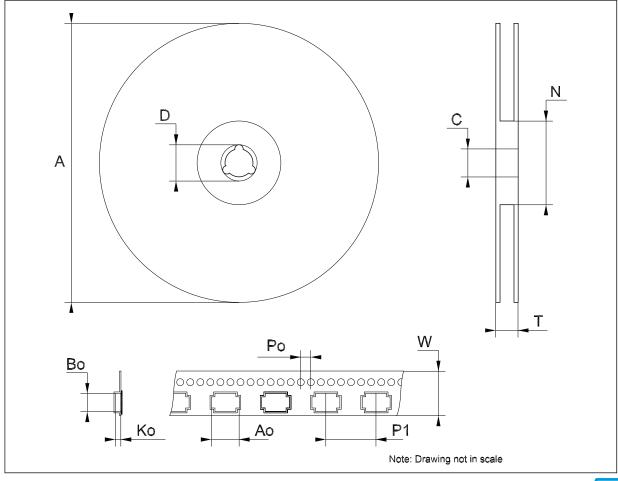
Dim.		mm.		inch.			
Dilli.	Min.	Тур.	Max.	Min.	Тур.	Max.	
А			3.70			0.146	
A1			0.10			0.004	
A2	3.40		3.60	0.134		0.142	
A3	1.25		1.35	0.049		0.053	
b	0.40		0.53	0.016		0.021	
С	0.35		0.55	0.014		0.022	
D	9.40		9.60	0.370		0.378	
D1	7.40		7.60	0.291		0.299	
Е	13.80		14.40	0.543		0.567	
E1	9.30		9.50	0.366		0.374	
E2	7.20		7.60	0.283		0.299	
E3	5.90		6.10	0.232		0.240	
е		1.27			0.050		
L	0.95		1.65	0.037		0.065	
α	0°		8°	0°		8°	



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Tana	Q.	Pool	PowerSO-	10	Mechanical	Data
i abe	α	Reei	Power5U-	ΙU	wechanical	Dala

Dim		mm.		inch.			
Dim.	Min.	Тур.	Max.	Min.	Тур.	Max.	
А			330			12.992	
С	12.8		13.2	0.504		0.519	
D	20.2			0.795			
N	60			2.362			
Т			30.4			1.197	
Ao	14.9		15.1	0.587		0.594	
Во	9.9		10.1	0.390		0.398	
Ko	4.15		4.35	0.163		0.171	
Ро	3.9		4.1	0.153		0.161	
Р	23.9		24.1	0.941		0.949	
W	23.7		24.3	0.933		0.957	



STCS2A Revision history

10 Revision history

Table 6. Document revision history

Date	Revision	Changes	
19-Feb-2008	1	Initial release.	
28-May-2008	2	Modified: Table 3 on page 5.	
02-Jul-2008	3	Modified: Table 5 on page 6.	

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