

STWLC38

Datasheet

Qi-compliant dual mode wireless power receiver for up to 15W applications

Features

- Up to 15 W output power
- Up to 5W output power in Tx mode
- Qi 1.3 inductive wireless standard communication protocol compliant
- High efficiency (98% typical) synchronous rectifier operating up to 800 kHz
- Low drop-out linear regulator with output current limit and input voltage control loop
- Adaptive Rectifier Configuration (ARC) Mode for enhanced spatial freedom
- 4 V to 12 V programmable output voltage
- Above 85% overall system efficiency
- 32-bit, 64 MHz ARM Cortex M0+ core with 32kB RRAM,16 kB SRAM,64kB ROM
- 10-bit A/D Converter
- Configurable GPIOs
- **I2C Slave interface**
- Multi-level ASK modulator, Enhanced FSK demodulator
- Output Over-Voltage clamping protection
- Accurate voltage/current measurement for Foreign Object Detection (FOD)
- On-chip thermal management and protections
- Flip chip 40 bumps (2.12mm x 3.32mm)

Applications

- **Smartphones**
- Wearable/Hearables TWS
- Asset tracking devices
- Medical and healthcare equipment

Description

The STWLC38 is an integrated Wireless Power Receiver suitable for wearable/ hearable and smart phone applications and can supply up to 15 W of output power. The chip has been designed to support Qi 1.3 specifications for inductive communication protocol, 5W Baseline Power Profile and 15W Extended Power Profile.

STWLC38 shows excellent efficiency performance thanks to the integrated low-loss synchronous rectifier and the low drop-out linear regulator: both elements are dynamically managed by the digital core to minimize the overall power dissipation over a wide range of output load conditions.

Through the I2C interface the user can access and modify different configuration parameters, tailoring the operation of the device to the needs of custom applications. The configuration parameters can be saved in the embedded Resistive RAM (RRAM) and automatically retrieved at power-up.

STWLC38 is also capable of operating in Tx mode to transmit power to charge other devices. The device can provide power up to 5W Output power in this mode. The STWLC38 is housed in a Chip-Scale Package to fit real-estate sensitive solutions in wearable devices.

1 Introduction

STWLC38 is a Wireless Power Receiver that rectifies the AC voltage developed across the receiving coil and provides a regulated DC voltage at the output.

The 32-bit core MCU is the supervisor of the whole device and manages all the functional blocks to

- establish and maintain communication with the transmitter,
- ensure adherence to Qi standard specifications (wherever required),
- optimize the efficiency by properly adjusting the operating point
- guarantee reliability by monitoring and protecting both the load and the device itself.

In order to execute the above mentioned (and many others) tasks, the MCU core relies on a resident firmware stored in ROM . In addition, some configuration parameters (e.g. output voltage, FOD tuning parameters, etc.) can be saved in the internal few times programmable Resistive RAM (RRAM) and retrieved at power-up, allowing the STWLC38 to operate as a fully autonomous stand-alone chip.

For applications in which the host system directly monitors or controls the power transfer, the I2C interface provides access to the internal registers of the STWLC38.

The device is also equipped with three programmable general-purpose I/O pins (GPIOs) to implement specific functions (e.g. driving status LEDs, enabling the output on request, informing the host system about faulty conditions, etc.).

Figure below shows the block diagram of the device with simplified interconnections among the functional blocks. The synchronous rectifier converts the AC voltage from the receiving coil into a DC voltage at the VRECT pin. The four switches of the rectifier (that is basically an H-bridge) are controlled by the digital core in order to minimize both conduction and switching losses as a function of the output voltage and current, both monitored by two channels of the ADC. Two bootstrap capacitors are externally connected to the BOOT1-BOOT2 pins to correctly drive the high-side switches of the rectifier.

The output of the rectifier, filtered by an external capacitor, is also the input rail for the main LDO linear regulator and for the auxiliary linear regulators in charge of deriving the 1.1 V and 2.5 V supply voltages.

The digital core has full control of the main LDO linear regulator in order to manage the output voltage, the output current and the drop-out voltage.

Of course the minimization of the drop-out voltage requires a closed loop regulation of the voltage at the VRECT pin, i.e. a feedback information that is sent to the transmitter (via ASK modulation) which, in turn, adjusts the delivered power by acting on the supply voltage, the switching frequency or the switching duty-cycle (or a combination of the three) of its own power stage, depending on the adopted technique.

This regulation loop involving the transmitter is an essential part of the wireless power transmission and is extensively described in Qi specifications.

STWLC38 is also capable to function as a transmitter to provide power up to 5W.

2 Block diagram

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Figure 1. Simplified block diagram

3 Device pin out

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Figure 2. Pin assignment (through top view)

Table 1. Pin description

4 Electrical and thermal specifications

4.1 Absolute maximum ratings

Stresses beyond those listed in Table 2 may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other condition above those indicated in Table 2 is not implied. Exposure to absolute maximum rated conditions for extended periods may affect device reliability.

Table 2. Absolute maximum ratings

4.2 Thermal characteristics

Table 3. Thermal characteristics

1. TA,OP -40°C to +85°C, limits over the operating range guaranteed by design and characterization, if not otherwise specified.

2. Device mounted on a standard JESD51-5 test board

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4.3 Electrical characteristics

0 °C < T_A < 85 °C; V_{VRECT} = 5 V to 10 V. Typical values are at T_J = 25 °C, if not otherwise specified.

Table 4. Electrical characteristics

4.4 Recommended operating conditions

4.5 Typical characteristics

Measurements performed at room temperature (25°C) using evaluation boards STEVAL-WBC86 TX (MP-A11a) and STEVAL-WLC38RX (8uH Coil).

Figure 3. System efficiency and Thermal Performance 5V,1A

4.6 Start up waveforms

This section shows typical waveforms measured on AC1,AC2,VRECT,VOUT pins during receiver power up and ARC mode operation.

Figure 4. STWLC38 Power Up and ARC mode waveforms

5 Device description

5.1 Chip-Enable Pin

When Chip-enable pin is set to HIGH , the device is in reset mode. Both rectifier low-side switches are turned-on while high-side switches are turned-off. After releasing the enable pin, STWLC38 can resume normal operational mode.

5.2 Synchronous rectifier

The synchronous rectifier of the STWLC38 is a key block in charge of converting the AC input power from the receiving coil into a DC supply rail for the following linear regulator.

In principle it consists of four N-channel MOSFETs arranged in an H-bridge, conveniently driven by a control block that monitors the voltage at the AC1 and AC2 pins to optimize the commutations and to charge the external bootstrap capacitors for the high-side switches.

Different driving schemes are possible for the switches of the rectifier and the MCU core dynamically selects the optimal one to maximize the overall efficiency as a function of the operating point.

When designing the filtering capacitor at the output of the synchronous rectifier, it must be taken into account that it has to minimize the AC residual ripple and to provide energy storage to sustain load transients, without impacting on the ASK communication with the transmitter.

5.3 ASK and FSK communication

Robust and reliable in-band ASK modulation is critical to the operation of any Qi compliant devices.

STWLC38 has dedicated hardware on top of the firmware algorithm to improve the performance of the in-band communication.

STWLC38 allows for two sets of modulation capacitors, namely COMA1/2 and COMB1/2. These 2 sets of ASK communication capacitors can be used in parallel or individually according to the load condition of the device. This allows for high level of flexibility to cater for a wide range of wireless transmitters.

STWLC38 comes with an advanced FSK demodulation filter which is able to remove any glitches present in the rectifier input.

The filter comes with a programmable masking pulse which is controlled by the firmware. This allows the firmware to adaptively control the filter to cater to different rectifier modes thereby ensuring robust FSK demodulation across the operating range.

5.4 ARC (Adaptive Rectifier Configuration) Mode

ARC (Adaptive Rectifier Configuration) mode improves the ping up and power transfer spatial freedom of the system in both X and Y direction.

Without any change in hardware or optimization of the coil, the ping up distance is enhanced by up to 50% in all directions by enabling ARC mode. This transforms the whole surface of the Tx to a usable area. Further enhancement is possible by customization of the coil.

Coil parameter tolerance requirements are widely relaxed due to ARC mode ping up feature. This is critical to wearables and hearables application where coils are of smaller and thinner dimensions, and it is relatively costlier to keep coil parameters within tight tolerances.

5.5 Protections

Over-voltage protection

The STWLC38 integrates different Over-Voltage Protection circuits to protect itself, the load connected to its output rail and the external components from damage due to overheating and/or exceeding AMR condition.

Under normal operating conditions the voltage at the output of the synchronous rectifier is slightly higher than the output one thanks to the communication with the transmitter.

A sudden change in the coupling factor between transmitting and receiving coils, for example due to abrupt reciprocal repositioning of the coils, easily leads to unpredictable voltage peaks at the AC input terminals: the TX-RX regulation loop is not fast enough to prevent such an event and additional precautions must be taken.

There are 3 over voltage protections, which are POVP, SOVP and HOVP.

- POVP (Ping Over Voltage Protection) During power-up when VAA and VDD are lower than UVLO threshold, POVP is triggered when VRECT > 14V, both rectifier low-side switches are turned-on while high-side switches are turned-off. POVP is released when VRECT < 11V.
- SOVP (Soft over voltage protection) When VRECT > VOUT + SOVP threshold, IEXT switch will be turned-on. SOVP threshold is programmable and can be set by PC GUI tool. SOVP is released when VRECT < SOVP threshold - 1V.
- HOVP (Hard Over voltage Protection) When VRECT > HOVP threshold, both rectifier low-side switches are turned on while high-side switches are turned-off. HOVP threshold is set by internal register. HOVP is released when VRECT < VOUT + SOVP release threshold, which is SOVP threshold - 1V.

Over-temperature protection

The STWLC38 is equipped with over-temperature detection circuits based on different sources:

- Internal temperature sensor
- external NTC temperature sensor
- TSHUT (hardware)

Over temperature protection(Software) The signals coming from the internal temperature sensors are conditioned and routed to the ADC. The temperature can be monitored in dedicated register. When the temperature exceeds set threshold level , it can turn off Main Voltage regulator (VOUT) , EPT can be sent to TX to stop power transfer.

The external sensor (NTC), typically placed very close to the coil to detect the over temperature of the coil , low-sided NTC of a resistor divider whose center tap is connected to the NTC pin(analog input), while the high-side resistor is connected to the VAA pin.

The temperature threshold is programmable by GUI.

If this function is not used, the pin must be pulled-up to VAA through a 10 kΩ resistor to prevent triggering the coil over-temperature protection.

TSHUT comparator monitor die temperature and turns off Main voltage regulator (VOUT) when temperature exceed set threshold level. The temperature threshold is programmable by GUI from 105 °C to 135 °C with 10 °C step (10 °C hysteresis).

When TSHUT is triggered both rectifier low-side switches are turned on while high-side switches are turned-off.

Over current Protection

The current limit is programmable (1.25A, 1.5A, 1.75A, 1.93A) , when output current exceed the set current limit , it can turn off Main Voltage regulator (VOUT) , EPT can be sent to TX to stop power transfer.

5.6 GPIOx and INTB pins

The GPIO0 through GPIO2 pins are programmable general-purpose I/O pins whose functions can be assigned in NVM memory. These pins can be configured both as inputs and outputs (either push-pull or open-drain) according to the selected function.

The INTB (GPIO3) pin is an interrupt output line that can be associated to any internal interrupt condition and used to inform the host system about specific events. The INTB pin is programmed as open-drain type.

5.7 RRAM (Resistive Random Access Memory)

STWLC38 has a 32kB RRAM which allows for multiple erase/re-write cycles.

This provides flexibility for custom firmware needed for various applications like proprietary protocols or field firmware upgrades.

RRAM also offers design in flexibility during preproduction optimization.

RRAM programming can be done in standalone mode, applying 5V (USB-VBUS 5V) to VOUT pin. Using USB-I2C interface to PC GUI tool.

6 TX Mode

STWLC38 can be configured to Tx mode, it is capable of delivering output power up to 5W (coil dependent).

Input Voltage

The device can support wide range of input voltage of 5V up to 12V. The power is applied to VOUT pin (VIN), which is the same VOUT pin when in Receiver Mode

Tx Inverter

The power transmitter uses a Full Bridge inverter, four N-channel MOSFETs arranged in an H-bridge. The inverter converts the DC input into AC waveform that drives resonant circuit, which consists of Primary coil plus series capacitor.

The power transmitted to the coil is regulated by varying the switching frequency of the bridge. A higher the operating frequency (example 200kHZ) for lower transmitted power, while a lower operating frequency (example 110kHZ) for higher transmitted power.

ASK Demodulation

Using ASK (Amplitude Shift Keying) modulation, the power receiver regularly sends Control error Packets to tune operating point to match Load requirements.

Tx receives this modulated signal from AC1 input, the coil signal is conditioned using discrete filter circuit as shown below and fed into VS pin for demodulation.

Figure 5. ASK Demodulator Circuit

7 Wireless power interface

The blocks that refer to the wireless power interface are the synchronous rectifier, the main LDO linear regulator and the ASK modulator, as well as the digital core as supervisor. The power transfer from the transmitter to the receiver is established as a result of a procedure which consists of several distinct stages.

The power transfer begins after the transmitter has properly detected a valid receiver and a specific communication has been established between the two parts.

Power transfer phases

The flow-chart in Figure 6 reports the whole process of power transfer in Baseline Power Profile (BPP up to 1A@5V)

Figure 6. Power transfer phases for Baseline Power Profile

BPP power transfer phases

- Digital ping: this phase is an interrogation session based on a more energetic AC burst during which the potential receiver is expected to reply through amplitude shift-keying (ASK) modulation, the receiver device sends Signal strength packet.
- Identification & configuration: this phase is aiming to identify the receiver and to gather information about its power transfer capability. The transmitter generates a so-called "Power Transfer Contract" tailoring some parameters that will characterize the following power transfer phase.
- Power transfer: this is the final step, where the transmitter initially increases and subsequently modulates the transmitted power in response to the control (feedback) data from the receiver.

The flow-chart in below shows the whole process leading to a power transfer in Extended Power Profile (EPP) up to 1.25A@12V

Figure 7. Power transfer phases for Extended Power Profile

EPP power transfer phases

Without entering the details of the different phases, the basic sequence of events taking place when a receiver is properly placed on the transmitting coil are summarized as:

- Digital ping: this phase is an interrogation session based on a more energetic AC burst during which the potential receiver is expected to reply through amplitude shift-keying (ASK) modulation, the receiver device sends Signal strength packet.
- Identification & configuration: this phase is aiming to identify the receiver and to gather information about its power transfer capability. The transmitter generates a so-called "Power Transfer Contract" tailoring some parameters that will characterize the following power transfer phase.
- Negotiation: in this phase the Power Receiver negotiates with the Power Transmitter to fine tune the Power Transfer Contract.
- Power transfer: this is the final step, where the transmitter initially increases and subsequently modulates the transmitted power in response to the control (feedback) data from the receiver

STWLC38 goes autonomously through Selection, Ping, Identification & Configuration phases, entering Power Transfer phase if no error occurs.

During the Power Transfer phase, the device sends Received-Power and Control-Error packets periodically as feedback information for the transmitter.

If a critical event like over-voltage, over-current or over-temperature occurs, the STWLC38 automatically sends the End-Power-Transfer packet.

When the Power Transfer is up and running, the End-Power-Transfer packet (with any response value) or any custom packet (e.g. Proprietary packet or Charge-Status packet) can be sent to the transmitter simply through commands via I2C interface.

Sending a custom packet may result in a reply (either a data packet or a pattern response from the transmitter) or no reply at all: if a response is received, the content is captured and stored in specific ${}^{12}C$ registers. Important notes:

- Changing the output voltage must respect the overall system design (selected coil, transmitter type, etc.).
- Output load transient response strongly depends on a correct design of the output capacitors. Severe load transients may lead to temporary output voltage collapse due to the overall TX-RX response time.
- A minimal output load significantly helps in increasing the signal-to-noise ratio during digital ping and is advisable to ensure interoperability with all transmitters. For this purpose, the STWLC38 allows the user to set a dummy load (reservoir current) which is dynamically managed to fade-out when an output load is applied.
- The initial load at power-up should not exceed 2.5 W, smoothly ramping-up to full power subsequently.

8 I2C interface

The STWLC38 can operate fully independently, i.e. without being interfaced with a host system. In applications in which the STWLC38 has to be a part of peripherals managed by the host system, the two SDA

and SCL pins could be connected to the existing I²C bus.

The device works as an I²C slave and supports standard (100 kbps) fast (400 kbps) data transfer modes. The STWLC38 has been assigned 0x61 7-bit hardware address. The pins are up to 3.3 V tolerant and the pull-up resistors should be selected as a trade-off between communication speed (lower resistors lead to faster edges) and data integrity (the input logic levels have to be guaranteed to preserve communication reliability). When the bus is idle, both SDA and SCL lines are pulled HIGH.

Data Validity

The data on the SDA line must be stable during the high period of the clock. The high and low states of the SDA line can only change when the SCL clock signal is low.

Start and Stop Conditions

Both the SDA and the SCL lines remain high when the I²C bus is not busy. A START condition is a high-to-low transition of the SDA line when SCL is HIGH, while the STOP condition is a low-to-high transition of the SDA line when SCL is HIGH. A STOP condition must be sent before each START condition.

Byte format

Every byte transferred over the SDA line must contain 8 bits. Each byte received by the STWLC38 generally is followed by an acknowledge (ACK) bit. The MSB is transferred first. One data bit is transferred during each clock pulse. The data on the SDA line must remain stable during the high state of each SCL clock pulse.

The device generates the ACK pulse (by pulling-down the SDA line during the acknowledge clock pulse) to confirm the correct device address or received data bytes.

Interface protocol

The interface protocol consists of

- Start condition (START)
- 7-bit device address $(0x61) + R/W$ bit (read = 1 / write = 0)
- Register pointer, high-byte
- Register pointer, low-byte
- Data sequence: N x (data byte + ACK)
- Stop condition (STOP)

The register pointer (or address) byte defines the destination register to which the read or write operation applies. When the read or write operation is finished, the register pointer is automatically incremented.

Writing to a single register

Writing to a single register begins with a START condition followed by device address 0xC2 (7-bit device address plus R/W bit cleared), two bytes of the register pointer and the data byte to be written in the destination register. Each transmitted byte is acknowledged by the STWLC38 through an ACK pulse.

Figure 9. Writing to single register byte

Writing to multiple registers (page write)

The STWLC38 supports writing to multiple registers with auto-incremental addressing. When data is written into a register, the register pointer is automatically incremented, therefore transferring data to a set of subsequent registers (also know as page write) is a straightforward operation.

Figure 10. Writing multiple register bytes

Reading from a single register

Reading from a single register begins with a START condition followed by the device address byte 0xC2 (7-bit device address plus R/W bit cleared) and two bytes of register pointer, then a re-START condition is generated and the device address 0xC3 (7-bit device address plus R/W bit asserted) is sent, followed by data reading. ACK pulse is generated by the STWLC38 at the end of each byte, but not for data bytes retrieved from the register. A STOP condition is finally generated to terminate the operation.

Figure 11. Reading single register byte

Reading from multiple registers (page reading)

Similarly to multiple (page) writing, reading from subsequent registers relies on an auto-increment of the register: the master can extend data reading to the following registers by generating and an ACK pulse at the end of each byte. Data reading starts immediately and the stream is terminated by a NMAK pulse at the end of the last data byte, followed by a STOP condition.

Figure 12. Reading multiple bytes

9 I2C register map

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The STWLC38 can be monitored and controlled by accessing the internal registers via I²C interface. The following registers map reports the accessible addresses. Addresses not shown in the map and blank bits have to be considered reserved and not altered as well.

These CFG parameters are accessible using the GUI tool, a new memh file needs to be generated after customized changes and to be Programmed in to NVM.

Table 5. Register abbreviations

Table 6. Chip information

Table 7. System information

STWLC38 I2C register map

Table 8. Communication

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Table 9. Commands

Table 10. Mode monitor

Table 11. GPIO

GPIO configuration - alternate functions

- 0x00: GPIO configured as Input, FLOATING (Default)
- 0x01: GPIO configured as Input, Pull up resistor to internal 1.8V.
- 0x02: GPIO configured as Input ,Pull down resistor to Internal Ground.
- 0x03:GPIO configured as Interrupt pin, Open Drain output.
- 0x04:GPIO configured as Interrupt pin, Push pull output.
- 0x05: Initialization complete, FW ready (Active High).
- 0x06: Input to turn-off Main LDO block (Active High).
- 0x07: Input to Disable ASK communication (Active High).
- 0x0B: Input to Disable RX_POWEROUT after Negotiation- Push pull (Active High).
- 0x0C: Input to Disable RX_POWEROUT after Negotiation- Push pull (Active Low).
- 0x0D:Input to Disable RX_POWEROUT after Negotiation- Open Drain.
- 0x29: GPIO configured as Output, Open Drain (Active Low).
- 0x2A:GPIO configured as Output, Open Drain (Active High).
- 0x2B: GPIO configured as Output, Push pull (Active Low).
- 0x2C:GPIO configured as Output, Push pull (Active High).

Receiver mode (Rx) registers

Table 12. RX interrupts enable

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Table 13. RX interrupt Clear

STWLC38 I2C register map

Table 14. RX interrupt latch

Table 15. RX interrupt status

Table 16. RX commands

Table 17. RX Configuration

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Example :

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VOUT SET = 5V ; 0x00B2 = 2D ; 0x00B1 [7..6] = 00

VOUT SET = 5.075V ; 0x00B2 = 2D; 0x00B1[7..6] = 11

Table 18. RX LDO configuration

Table 19. RX BPP FOD configuration

Address Register name R/W Default Description RX EPP FOD CUR THR1 CFG - Bits [7..0] EPP FOD current threshold 1, in units of 10mA. RX EPP FOD CUR THR2 $\begin{array}{|c|c|c|c|c|}\n\hline\n\end{array}$ CFG $\begin{array}{|c|c|c|c|c|c|}\n\hline\n\end{array}$ Bits [7..0] EPP FOD current threshold 2, in units of 10mA. RX EPP FOD CUR THR3 $\begin{bmatrix} CFG \end{bmatrix}$ - Bits [7..0] EPP FOD current threshold 3, in units of 10mA. RX EPP FOD CUR THR4 $\begin{array}{|c|c|c|c|c|}\n\hline\n\end{array}$ CFG $\begin{array}{|c|c|c|c|c|c|}\n\hline\n\end{array}$ Bits [7..0] EPP FOD current threshold 4, in units of 10mA. RX EPP FOD CUR THR5 CFG - Bits [7..0] EPP FOD current threshold 5, in units of 10mA. RX EPP FOD OFFSET 0 CFG - Bits [7..0] EPP FOD offset at current 0, in units of 8mW. RX EPP FOD OFFSET $1 \mid CFG \mid - \mid B$ its [7..0] EPP FOD offset at current 1, in units of 8mW. RX EPP FOD OFFSET 2 $\begin{array}{|c|c|c|c|c|}\n\hline\n\end{array}$ CFG $\begin{array}{|c|c|c|c|c|c|}\n\hline\n\end{array}$ Bits [7..0] EPP FOD offset at current 2, in units of 8mW. RX EPP FOD OFFSET 3 CFG Bits [7..0] EPP FOD offset at current 3, in units of 8mW. RX EPP FOD OFFSET 4 $|CFG|$ - Bits [7..0] EPP FOD offset at current 4, in units of 8mW. RX EPP FOD OFFSET 5 \vert CFG \vert - Bits [7..0] EPP FOD offset at current 5, in units of 8mW. RX BPP FOD RSER CFG - Bits [7..0] Coil series resistance, in units of 4mOhm.

Table 20. RX EPP FOD configuration

Table 21. RX protections

Table 22. RX Power transfer contract

Table 23. RX configuration Qi

Table 24. RX ASK configuration

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The default settings are optimized for light and high load conditions.

Table 25. RX Protection threshold

Table 26. RX Qi chip ID

Transmitter mode (TX) registers

Table 27. TX Commands

Table 28. TX Interrupt enable

STWLC38 I2C register map

Table 29. TX Interrupt clear

STWLC38 I2C register map

Table 30. TX interrupt latch

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Table 31. TX Interrupt status

Table 32. TX configuration

Table 33. TX EPT reason

STWLC38 I2C register map

Table 34. TX power transfer contract

STWLC38 I2C register map

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Table 35. TX protections

10 Application information

This chapter is aimed to provide some application hints. The reference schematic, the PCB layout guidelines, the minimum components to properly run the application and other aspects.

10.1 Reference application

STWLC38 typical application schematic for Rx and Tx modes are shown below.

Figure 13. STWLC38 basic application diagram in Rx mode

Figure 14. STWLC38 basic application diagram in RTx mode

10.2 Reference BOM

Typical key component values are listed in the tables below.

Values for a 2.5W wireless power receiver application are listed in the table below.

Note: All of the components listed above refer to a typical application. Operation in the application may be limited by a choice of these external components (voltage ratings, current and power dissipation capability, etc.).

The basic application schematic is relatively simple, since STWLC38 does not require many external parts to operate. Anyway, there are different aspects that must be carefully considered to properly design a customized application. In most cases the main constrains are limited PCB size/room and thickness, that unavoidably lead to crowded solutions with a far-from-optimal electrical and thermal performance.

10.3 PCB routing guidelines

- 1. Auxiliary 2V5 (VAA) and 1V1 (VDD) LDO filtering capacitors should be placed as close as possible to the STWLC38. Connection traces should be short and placed in the top layer. Capacitors ground can be connected directly to the GND plane.
- 2. CRECT and C_{OUT} capacitors should be placed close to STWLC38 with higher priority than R_{CL} resistor.
- 3. Power traces (AC1, AC2, VRECT, VOUT) should be kept wide enough to sustain high current. Duplicating these traces in inner layers is advisable wherever possible.
- 4. AC1 and AC2 tracks should be routed closely to minimize the area of the resulting loop.
- 5. Thermal performance and grounding should be always optimized by preserving bottom layer (usually assigned to ground) integrity.

10.4 External components selection

RX series resonant circuit components

Series resonant circuit, both Cs and Cd should show excellent quality factor, relatively high RMS current capability and superior capacitance stability in the frequency range of interest. Multi-Layer Ceramic capacitors (MLCCs) are inherently good devices in terms of RMS current capability and quality factor. Capacitance tolerance and stability strongly depend on the dielectric type.

Dielectrics such as X5R ,X7R,COG are used to achieve higher capacitance per volume at the cost of lower accuracy and undesired dependencies (e.g. DC-biasing, temperature, etc.). In practice, the $C_S(most critical)$ usually consists of a few smaller, low-profile X5R/X7R/COG capacitors connected in parallel. The parallel connection also helps to increase RMS current capability and mitigate the effect of capacitance tolerance due to production spread. The voltage rating for these capacitors is usually maximized to take into account the voltage developed in proximity to resonance: 50V-rated capacitors are generally a good choice.

ASK modulation capacitors

The capacitors at the COMAx/COMBx pins are connected to the AC1-AC2 terminals through controlled switches (ASK modulator): the de-tuning effect of closing these switches results in an amplitude modulation detected by the transmitter and also visible at the rectified voltage. Positive or negative modulation may occur, depending on the operating frequency and other factors. The ASK modulation index clearly depends on the capacitance value of these capacitors, whose value has to be adjusted in case of a heavy negative modulation at VRECT (that is generally undesirable). The same considerations made above for the resonant capacitors is also applicable here, where capacitance tolerance is less critical: X5R dielectric-type are a good choice and an initial value of 22 nF and 10nF is typically doing the job.

VRECT over-voltage clamping resistor

The voltage at the VRECT pin is primarily dictated by the transmitter, whose operating point is linked to the feedback information received via ASK modulation. Unexpected conditions, however, may increase the VRECT voltage to dangerous values (close to AMR levels). A sudden change in relative alignment between the transmitting and receiving coils, for example, could result in a dramatic change in coupling factor and, in turn, a fast-rising voltage. Since the reaction of the transmitter is relatively slow, the STWLC38 protects itself by closing the switch internally connected to the IEXT pin. This switch is externally connected to VRECT via a resistor (R_{CL}) to implement an active clamper. The value of R_{CL} is selected so that most of the power is dissipated in the clamper circuit rather than inside of the chip. Special resistors (surge resistors) capable of withstanding higher energy pulses are recommended.

ESD protection diodes

Since the receiving coil is a easy entry point for ESD (relatively large area with remarkable capacitive coupling), a good application design should consider protections for the most exposed pins: AC1 and AC2. Uni-directional Transient Voltage Suppression (TVS) diodes at both pins are recommended. ESDs have essentially a commonmode nature and, although the receiving coil has low DC-resistance, its AC impedance may appear quite high to fast voltage spikes: independent clamping at AC1 and AC2 pins is thus mandatory. The knee-voltage of the TVS diodes should be selected considering the maximum VRECT voltage plus some margin to avoid non-negligible leakage current at higher temperature, while their energy dissipation capability should be maximized considering the size of the package.

Coil thermal protection

Maximizing the amount of transferred power is ofter desirable, but also limited by operating conditions : applications in which the wireless power receiver has poor power dissipation capability and/or the power loss in the receiving coil is relevant (e.g. very tiny and slim coils with relatively high DC-resistance) may require some thermal protection. This feature is implemented with STWLC38 thanks to its NTC pin. A channel of the internal ADC is routed to the NTC pin, allowing the user to read the voltage across an external NTC thermistor.

11 Reference schematic

Figure 8. shows typical application schematic for the STWLC38.

Figure 15. STWLC38 basic schematic diagram

Note: All of the components listed above refer to a typical application. Operation in the application may be limited by a choice of these external components (voltage ratings, current and power dissipation capability, etc.)

12 Glossary/Abbreviations

 $\sqrt{1}$

- ADC: Auxiliary Data Control
- ADT : Auxiliary Data Transport
- ASK: Amplitude Shift Keying
- BPP : Baseline Power Profile
- CHS : Charge Status
- CEP: Control Error Packets
- CFG : Configuration
- DSR : Data Stream Response
- EPP: Extended Power Profile
- EPT: End Power Transfer
- FOD : Foreign Object Detection
- FSK: Frequency Shift Keying
- ID: Identification
- PCH: Power Control Hold-off
- PP: Proprietary Packet
- RP/RP8: Received Power Packet
- SRQ : Specific Request
- SS/SIG : Signal Strength Packet
- XID : Extended Identification

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13 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

13.1 Package information

WAFER BACK/ MARKING SIDE

BUMP SIDE

Figure 16. WLCSP40 2.126x3.327x0.546 0.4 Pitch 0.25 Ball package outline

Table 38. WLCSP40 2.126x3.327x0.546 0.4 Pitch 0.25 Ball mechanical data

Figure 17. Recommended footprint

Figure 18. Device marking , Tape and reel information

ST

Device Marking Information

- A Stripe (PIN 1 Identifier)
B Standard ST Logo
-
-
-
- B Standard ST Logo
C Second Level Interconnect
D Marking Area max 8 characters
E Assembly Year (Y) 1 character
F Assembly Week (WW) 2 characters

Revision history

Table 39. Document revision history

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