80 watt + 80 watt dual BTL class-D audio amplifier
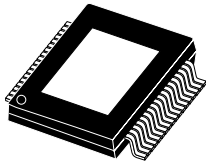
**PowerSSO-36
exposed pad up**

Features

- 80 W + 80 W output power at THD = 10% with R_L = 6 Ω and V_{CC} = 32 V
- 70 W + 70 W output power at THD = 10% with R_L = 8 Ω and V_{CC} = 34 V
- Wide-range single-supply operation (14 - 36 V)
- High efficiency (η = 90%)
- Four selectable, fixed gain settings of nominally 25.6 dB, 31.6 dB, 35.1 dB and 37.6 dB
- Differential inputs minimize common-mode noise
- Standby and mute features
- Short-circuit protection
- Thermal overload protection
- Externally synchronizable

Description

The TDA7498L is a dual BTL class-D audio amplifier with single power supply designed for home systems and active speaker applications.
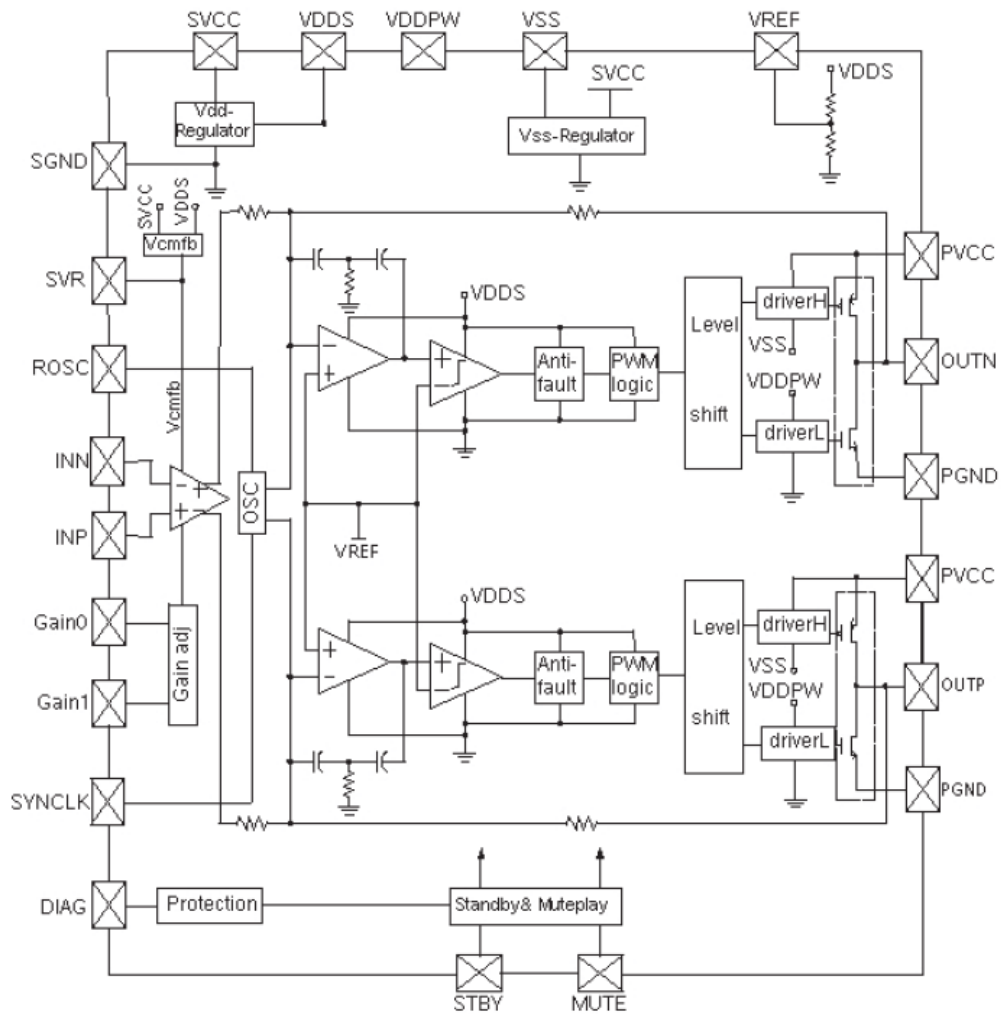
It comes in a 36-pin PowerSSO package with exposed pad up (EPU) to facilitate mounting a separate heatsink.

Product status link
TDA7498L

Product summary	
Order code	TDA7498LTR
Temperature range	-40 to 85 °C
Package	PowerSSO36 (EPU)
Packing	Tape and reel

DS6539 - Rev 6 - June 2020
For further information contact your local STMicroelectronics sales office.

www.st.com

1 Device block diagram

Figure 1. Internal block diagram (showing one channel only) shows the block diagram of one of the two identical channels of the TDA7498L.
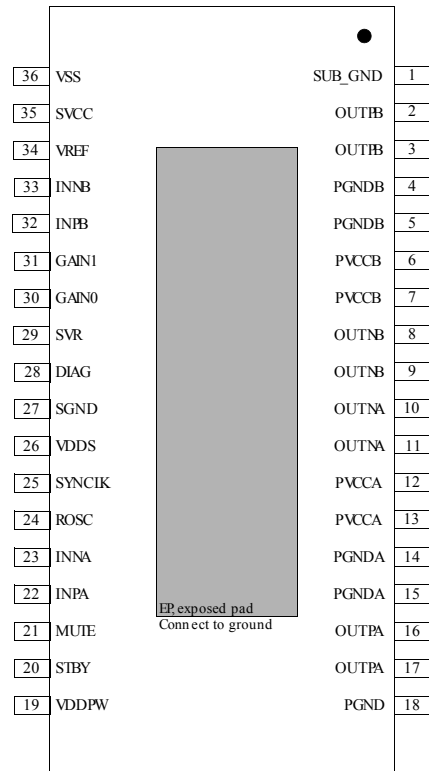
Figure 1. Internal block diagram (showing one channel only)

2 Pin description

2.1 Pinout

Figure 2. Pin connections (top view, PCB view)

Left pin	Signal		Right signal	Pin
36	VSS		SUB_GND	1
35	SVCC		OUTPB	2
34	VREF		OUTPB	3
33	INNB		PGNDB	4
32	INPB		PGNDB	5
31	GAIN1		PVCCB	6
30	GAIN0		PVCCB	7
29	SVR		OUTNB	8
28	DIAG		OUTNB	9
27	SGND		OUTNA	10
26	VDDS		OUTNA	11
25	SYNCIK		PVCCA	12
24	ROSC		PVCCA	13
23	INNA		PGNDA	14
22	INPA		PGNDA	15
21	MUTE		OUTPA	16
20	STBY		OUTPA	17
19	VDDPW		PGND	18

EP, exposed pad
Connect to ground

2.2 Pin list

Table 1. Pin description list

Number	Name	Type	Description
1	SUB_GND	PWR	Connect to the frame
2,3	OUTPB	O	Positive PWM for right channel
4,5	PGNDB	PWR	Power stage ground for right channel
6,7	PVCCB	PWR	Power supply for right channel
8,9	OUTNB	O	Negative PWM output for right channel
10,11	OUTNA	O	Negative PWM output for left channel
12,13	PVCCA	PWR	Power supply for left channel
14,15	PGNDA	PWR	Power stage ground for left channel
16,17	OUTPA	O	Positive PWM output for left channel
18	PGND	PWR	Power stage ground
19	VDDPW	O	3.3-V (nominal) regulator output referred to ground for power stage
20	STBY	I	Standby mode control
21	MUTE	I	Mute mode control
22	INPA	I	Positive differential input of left channel
23	INNA	I	Negative differential input of left channel
24	ROSC	O	Master oscillator frequency-setting pin
25	SYNCLK	I/O	Clock in/out for external oscillator
26	VDDS	O	3.3-V (nominal) regulator output referred to ground for signal blocks
27	SGND	PWR	Signal ground
28	DIAG	O	Open-drain diagnostic output
29	SVR	O	Supply voltage rejection
30	GAIN0	I	Gain setting input 1
31	GAIN1	I	Gain setting input 2
32	INPB	I	Positive differential input of right channel
33	INNB	I	Negative differential input of right channel
34	VREF	O	Half VDDS (nominal) referred to ground
35	SVCC	PWR	Signal power supply decoupling
36	VSS	O	3.3-V (nominal) regulator output referred to power supply
-	EP	-	Exposed pad for heatsink, to be connected to ground

3 Electrical specifications

3.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

Symbol	Parameter	Value	Unit
V_{CC_MAX}	DC supply voltage for pins PVCCA, PVCCB, SVCC	45	V
V_{L_MAX}	Voltage limits for input pins STBY, MUTE, INNA, INPA, INNB, INPB, GAIN0, GAIN1	-0.3 to 3.6	V
T_{j_MAX}	Operating junction temperature	0 to 150	°C
T_{stg}	Storage temperature	-40 to 150	°C

Note: *warning: stresses beyond those listed under "Absolute maximum ratings" make cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "Recommended operating condition" are not implied. Exposure to absolute maximum rated conditions for extended periods may affect device reliability. In the real application, the power supply with the nominal value rated in the recommended operating conditions may rise beyond the maximum operating condition for a short time when no or very low current is sunk (amplifier in mute state). In this case the reliability of the device is guaranteed, provided that the absolute maximum rating is not exceeded.*

3.2 Thermal data

Table 3. Thermal data

Symbol	Parameter	Min.	Typ.	Max.	Unit
$R_{th\ j\text{-}case}$	Thermal resistance, junction to case	-	2	3	°C/W

3.3 Recommended operating conditions

Table 4. Recommended operating conditions

Symbol	Parameter	Min.	Typ.	Max.	Unit
V_{CC}	Supply voltage for pins PVCCA, PVCCB	14	-	36	V
Tamb	Ambient operating temperature	-20	-	85	°C

3.4 Electrical characteristics

Unless otherwise stated, the values in the table below are specified for the conditions: V_{CC} = 32 V, R_L = 6 Ω, R_{OSC} = R3 = 39 kΩ, C8 = 100 nF, f = 1 kHz, G_V = 25.6 dB Tamb = 25 °C.

Table 5. Electrical characteristics

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_q	Total quiescent current	No LC filter, no load	-	40	60	mA
I_{qSTBY}	Quiescent current in standby	-	-	1	10	µA
V_{OS}	Output offset voltage	Play mode	-100	-	100	mV
		Mute mode	-60	-	60	
I_{OCP}	Overcurrent protection threshold	R_L = 0 Ω	5.0	6.0	-	A
T_{jS}	Junction temperature at thermal shutdown	-	-	150	-	°C
R_i	Input resistance	Differential input	48	60	-	kΩ
V_{OVP}	Overvoltage protection threshold	-	42	43	-	V
V_{UVP}	Undervoltage protection threshold	-	-	-	8	V
R_{dsON}	Power transistor on-resistance	High side	-	0.2	-	Ω
		Low side	-	0.2	-	
P_o	Output power	THD = 10%	-	80	-	W
		THD = 1%	-	65	-	
P_o	Output power	R_L = 8 Ω, THD = 10%, V_{CC} = 32 V	-	65	-	W
P_D	Dissipated power	P_o = 80 W + 80 W, THD = 10%	-	16	-	W
η	Efficiency	P_o = 80 W + 80 W	-	90	-	%
THD	Total harmonic distortion	P_o = 1 W	-	0.1	-	%
G_V	Closed-loop gain	GAIN0 = L, GAIN1 = L	24.6	25.6	26.6	dB
		GAIN0 = L, GAIN1 = H	30.6	31.6	32.6	
		GAIN0 = H, GAIN1 = L	34.1	35.1	36.1	
		GAIN0 = H, GAIN1 = H	36.6	37.6	38.6	
ΔG_V	Gain matching	-	-1	-	1	dB
C_T	Crosstalk	f = 1 kHz, P_o = 1 W	50	70	-	dB
eN	Total input noise	A Curve, G_V = 20 dB	-	15	-	µV
		f = 22 Hz to 22 kHz	-	25	50	
SVRR	Supply voltage rejection ratio	fr = 100 Hz, Vr = 0.5 Vpp, C_{SVR} = 10 µF	-	70	-	dB
T_r, T_f	Rise and fall times	-	-	50	-	ns
f_{SW}	Switching frequency	Internal oscillator	290	310	330	kHz
f_{SWR}	Output switching frequency range	With internal oscillator [1]	250	-	400	kHz
		With external oscillator [2]	250	-	400	
V_{inH}	Digital input high (H)	-	2.3	-	-	V
V_{inL}	Digital input low (L)		-	-	0.8	

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{STBY}	Pin STBY voltage high (H)	-	2.7	-	-	V
	Pin STBY voltage low (L)		-	-	0.5	
V_{MUTE}	Pin MUTE voltage high (H)	-	2.5	-	-	V
	Pin MUTE voltage low (L)		-	-	0.8	
A_{MUTE}	Mute attenuation	V_{MUTE} < 0.8 V	-	70	-	dB

1. $f_{SW} = 10^6 / ((16 * R_{OSC} + 182) * 4)$ kHz, $f_{SYNCLK} = 2 * f_{SW}$ with R3 = 39 kΩ (see Figure 20. Application circuit).

2. $f_{SW} = f_{SYNCLK} / 2$ with the external oscillator.

4 Characterization curves

Figure 20. Application circuit shows the test circuit with which the characterization curves, shown in the next sections, were measured. Figure 3. Test board below shows the PCB layout.

4.1 PCB layout

Figure 3. Test board
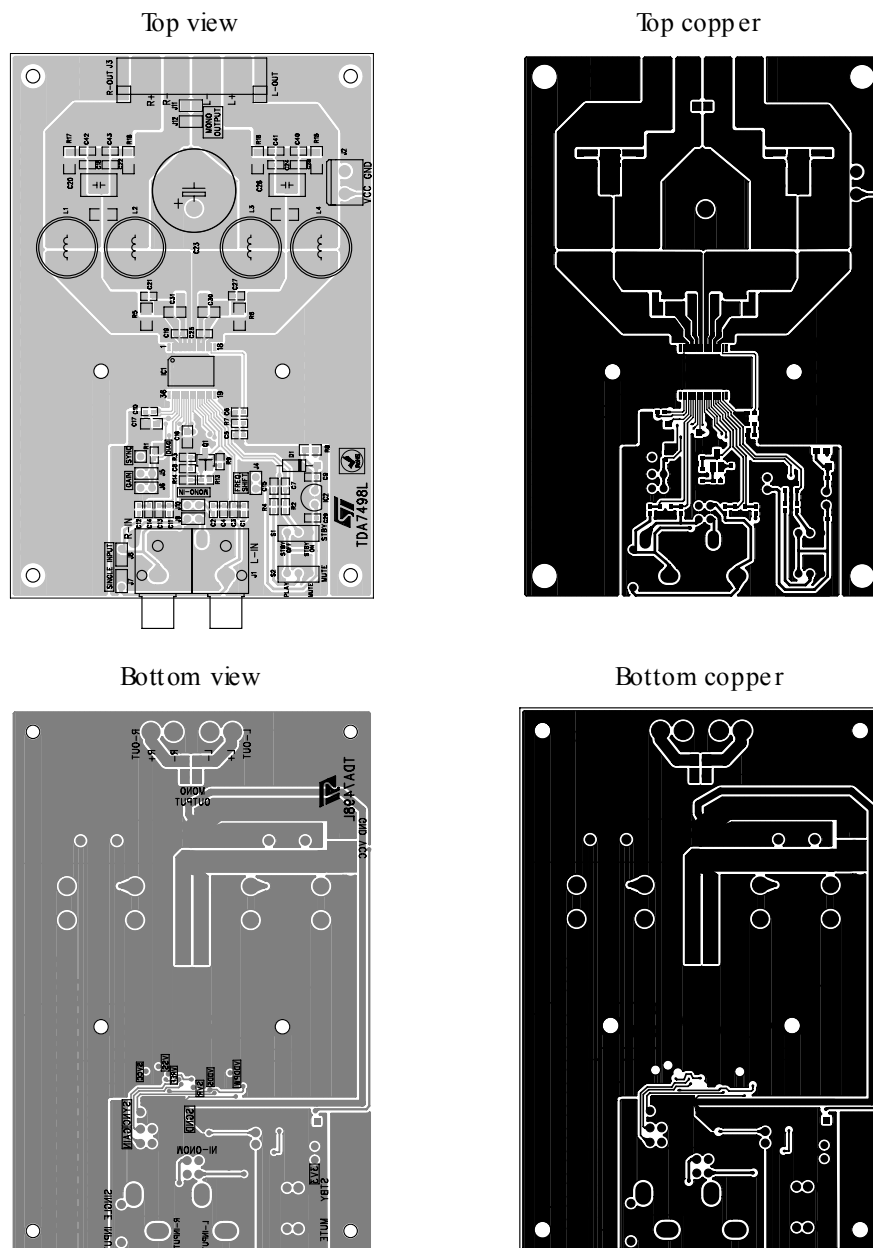
4.2 Characterization curves

Unless otherwise stated the measurements were made under the following conditions:

V_{CC} = 32 V, f = 1 kHz, G_V = 25.6 dB, R_{OSC} = 39 kΩ, C_{OSC} = 100 nF, Tamb = 25 °C

4.2.1 For R_L = 6 Ω

Figure 4. Output power vs. supply voltage
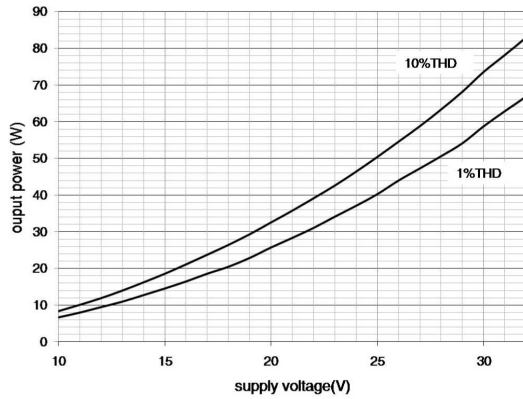
Figure 5. THD vs. output power (1 kHz)
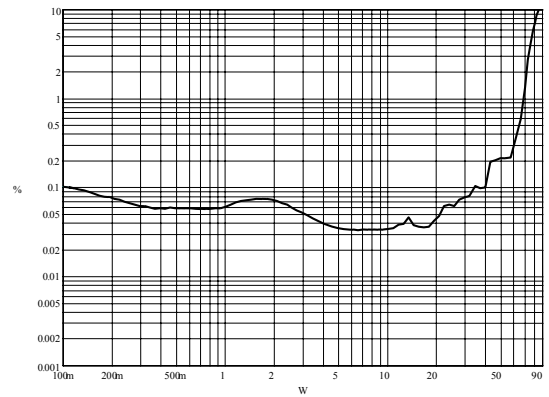
Figure 6. THD vs. output power (100 Hz)
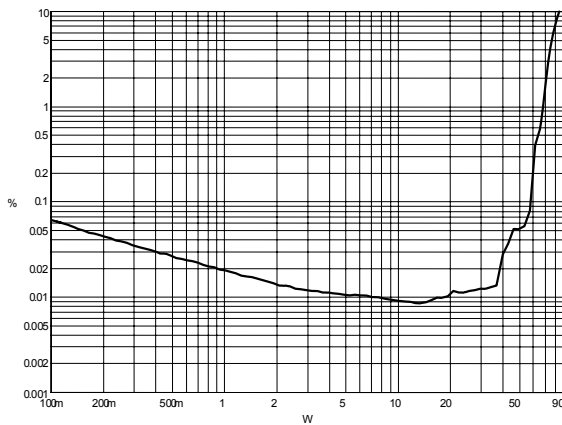
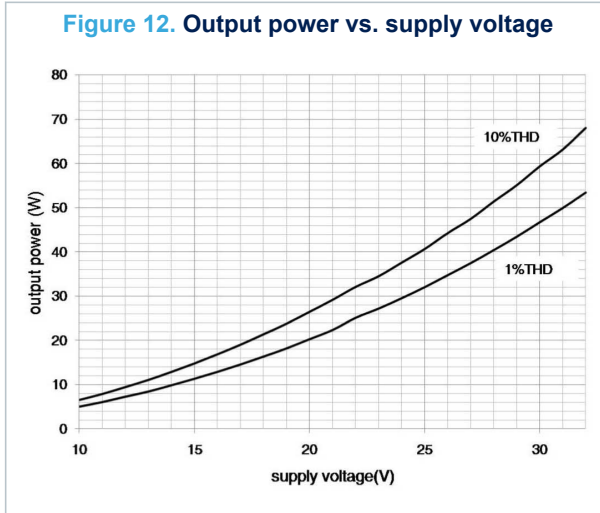**Figure 7. THD vs. frequency (1 W)**
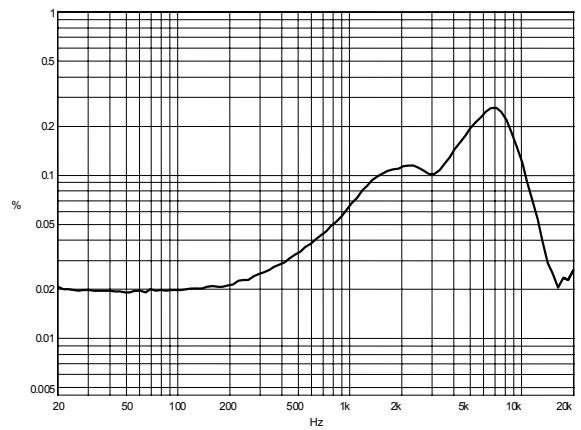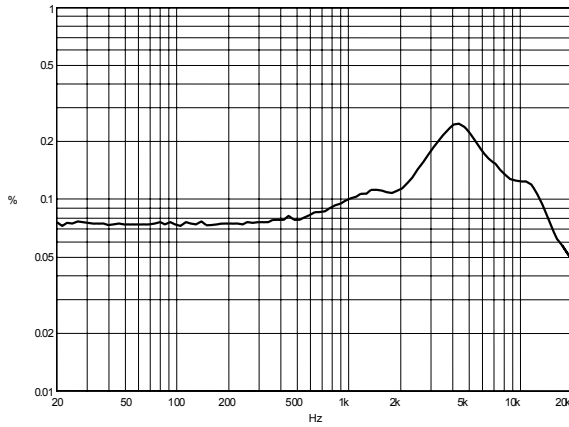
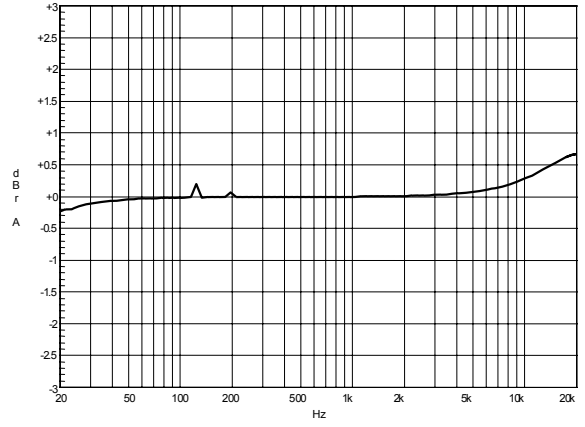**Figure 8. THD vs. frequency (100 mW)**
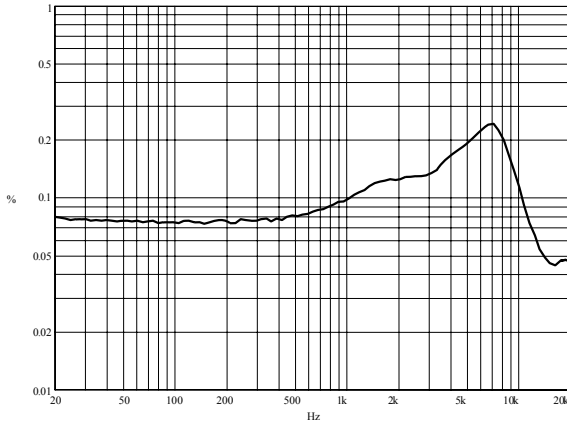
Figure 9. Frequency response
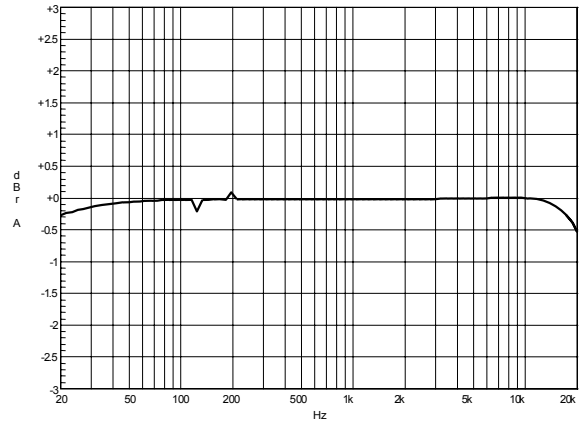
Figure 10. FFT performance (0 dBFS)
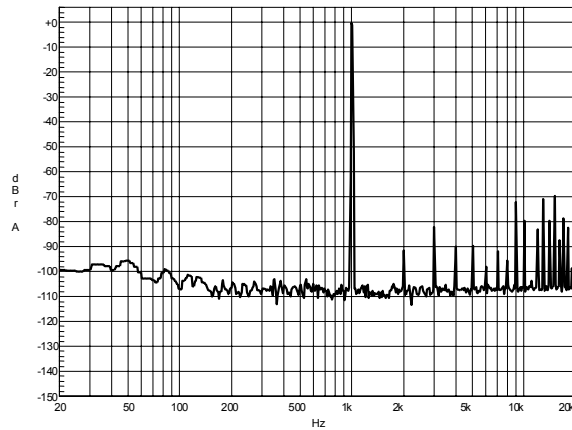
Figure 11. FFT performance (0 dBFS)
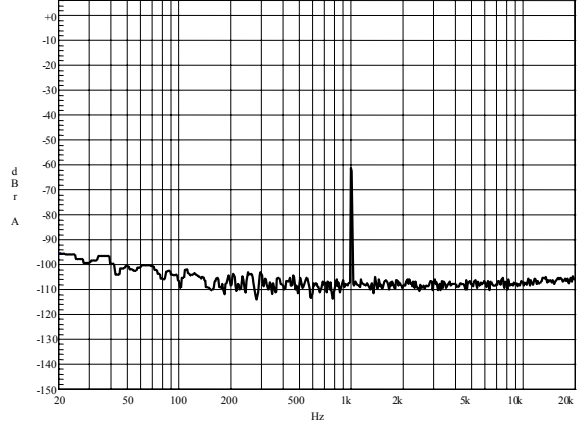
4.2.2 For $R_L = 8 \, \Omega$

Figure 12. **Output power vs. supply voltage**

Figure 13. **THD vs. output power (1 kHz)**

Figure 14. **THD vs. output power (100 Hz)**
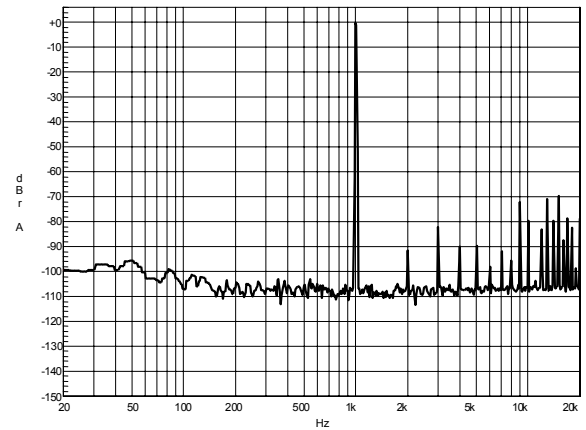
Figure 15. **THD vs. frequency (1 W)**
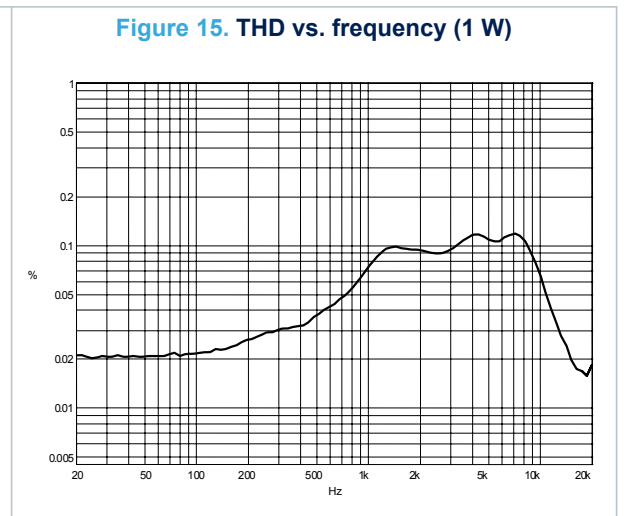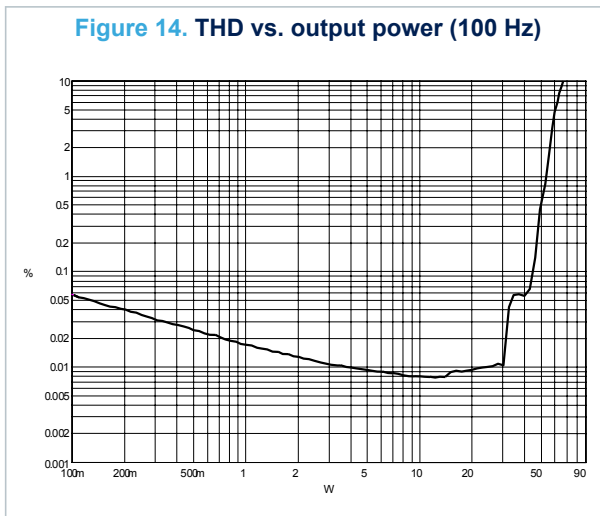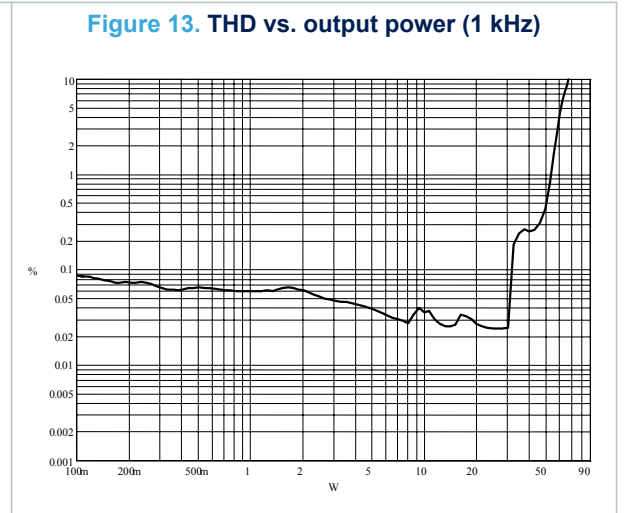
Figure 16. THD vs. frequency (100 mW)

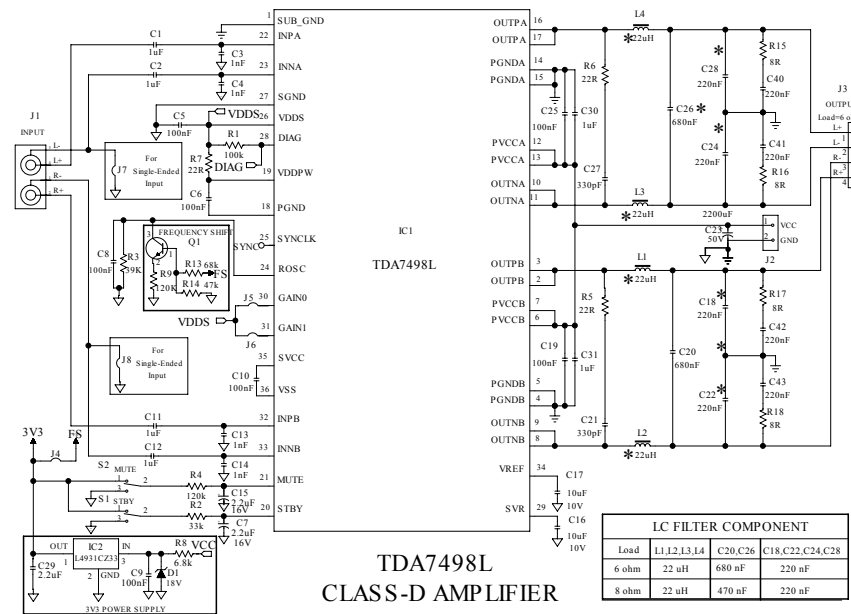
Figure 17. Frequency response

Figure 18. FFT performance (0 dB)
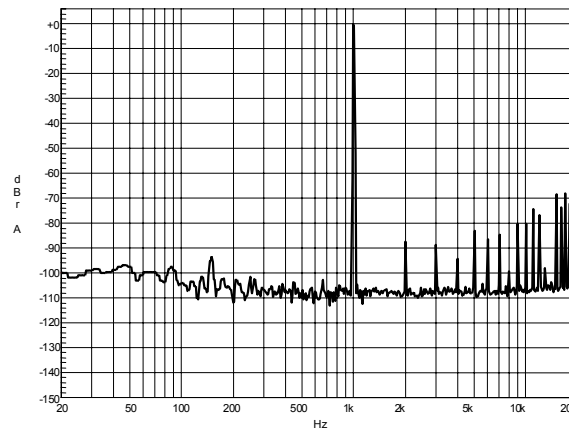
Figure 19. FFT performance (-60 dB)

5 Application information

5.1 Application circuit

Figure 20. Application circuit

	LC FILTER COMPONENT		
Load	L1,L2,L3,L4	C20,C26	C18,C22,C24,C28
6 ohm	22 uH	680 nF	220 nF
8 ohm	22 uH	470 nF	220 nF

5.2 Mode selection

The three operating modes of the TDA7498L are set by the two inputs, STBY (pin 20) and MUTE (pin 21).

- Standby mode: all circuits are turned off, very low current consumption.
- Mute mode: inputs are connected to ground and the positive and negative PWM outputs are at 50% duty cycle.
- Play mode: the amplifiers are active.

The protection functions of the TDA7498L are enabled by pulling down the voltages of the STBY and MUTE inputs shown in Figure 21. Standby and mute circuits. The input current of the corresponding pins must be limited to 200 µA.

Table 6. Mode settings

Mode	STBY	MUTE
Standby	L [1]	X (don't care)
Mute	H [1]	L
Play	H	H

1. Drive levels defined in Table 5. Electrical characteristics

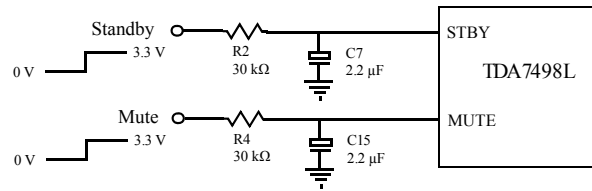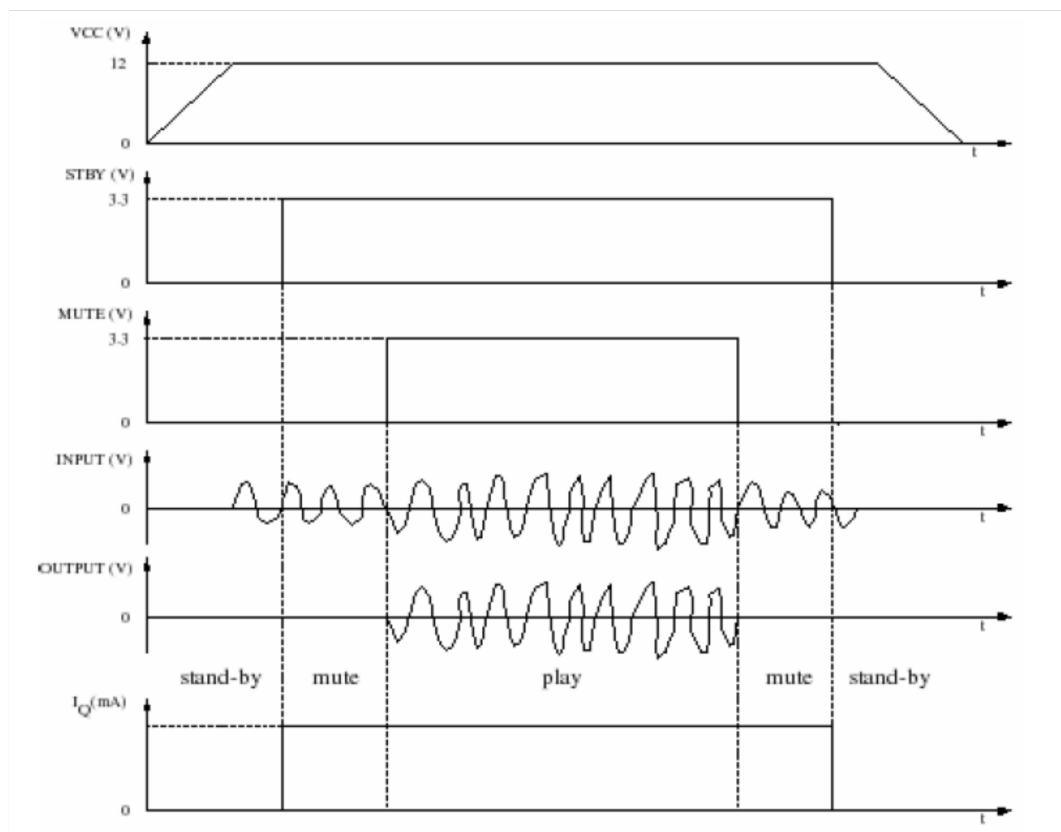**Figure 21. Standby and mute circuits**

Figure 22. Turn on/off sequence for minimizing speaker "pop"

5.3 Gain settings

The gain of the TDA7498L is set by the two inputs, GAIN0 (pin 30) and GAIN1 (pin31). Internally, the gain is set by changing the feedback resistors of the amplifier.

Table 7. Gain settings

| GAIN0	GAIN1	Nominal gain, G_v (dB)
L	L	25.6
L	H	31.6
H	L	35.6
H	H	37.6

5.4 Input resistance and capacitance

The input impedance is set by an internal resistor Ri = 60 kΩ (typical). An input capacitor (Ci) is required to couple the AC input signal.

The equivalent circuit and frequency response of the input components are shown in Figure 23. Input circuit and frequency response. For Ci = 470 nF the high-pass filter cutoff frequency is below 20 Hz:

$f_C = 1 / (2 * \pi * Ri * Ci)$

Figure 23. Input circuit and frequency response
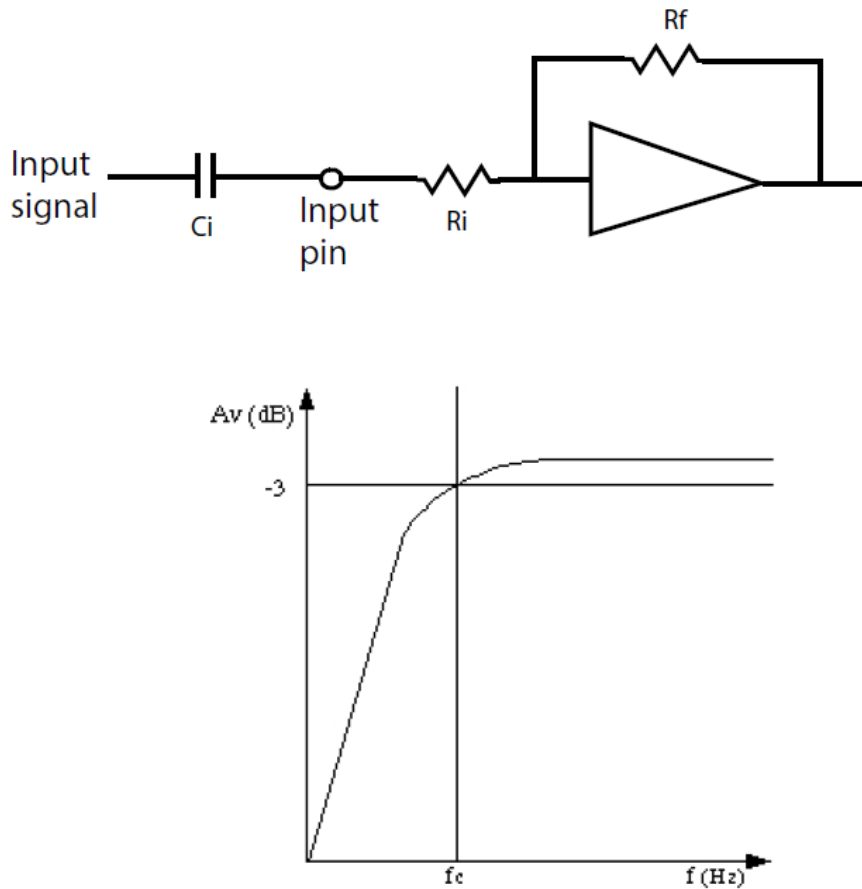
5.5 Internal and external clocks

The clock of the class-D amplifier can be generated internally or can be driven by an external source.

If two or more class-D amplifiers are used in the same system, it is recommended that all devices operate at the same clock frequency. This can be implemented by using one TDA7498L as master clock, while the other devices are in slave mode, that is, externally clocked. The clock interconnect is via pin SYNCLK of each device. As explained below, SYNCLK is an output in master mode and an input in slave mode.

5.5.1 Master mode (internal clock)

Using the internal oscillator, the output switching frequency, f_{SW}, is controlled by the resistor, R_{OSC}, connected to pin ROSC:

$f_{SW} = 10^6 / [(R_{OSC} * 16 + 182) * 4]$ kHz

where R_{OSC} is in kΩ.

In master mode, pin SYNCLK is used as a clock output pin whose frequency is:

$f_{SYNCLK} = 2 * f_{SW}$

For master mode to operate correctly then resistor R_{OSC} must be less than 60 kΩ as given below in Table 8. How to set up SYNCLK.

5.5.2 Slave mode (external clock)

In order to accept an external clock input the pin ROSC must be left open, that is, floating. This forces pin SYNCLK to be internally configured as an input as given in Table 8. How to set up SYNCLK.
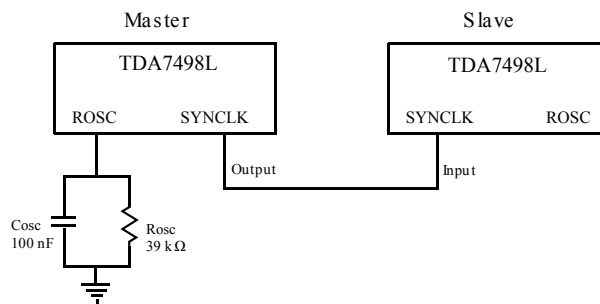
The output switching frequency of the slave devices is:

$f_{SW} = f_{SYNCLK} / 2$

Table 8. How to set up SYNCLK

Mode	ROSC	SYNCLK
Master	R_{OSC} < 60 kΩ	Output
Slave	Floating (not connected)	Input

Figure 24. Master and slave connection

5.6 Output low-pass filter

To avoid EMI problems, it may be necessary to use a low-pass filter before the speaker. The cutoff frequency should be larger than 22 kHz and much lower than the output switching frequency. It is necessary to choose the L and C component values depending on the loudspeaker impedance. Some typical values, which give a cutoff frequency of 27 kHz, are shown in Figure 25. Typical LC filter for an 8 Ω speaker and Figure 26. Typical LC filter for a 6 Ω speaker below.
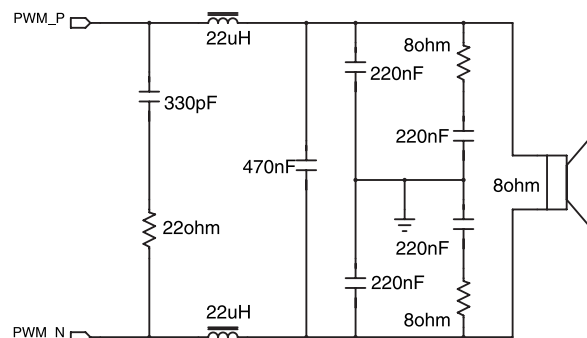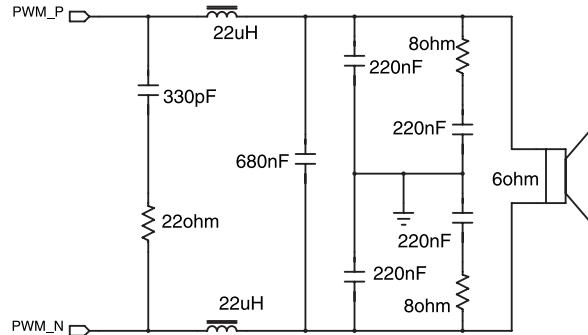
Figure 25. Typical LC filter for an 8 Ω speaker

Figure 26. Typical LC filter for a 6 Ω speaker

5.7 Protection functions

The TDA7498L is fully protected against overvoltages, undervoltages, overcurrents and thermal overloads as explained here.

Overvoltage protection (OVP)

If the supply voltage exceeds the value for V_{OVP} given in Table 5. Electrical characteristics , the overvoltage protection is activated which forces the outputs to the high-impedance state. When the supply voltage falls back to within the operating range, the device restarts.

Undervoltage protection (UVP)

If the supply voltage drops below the value for V_{UVP} given in Table 5. Electrical characteristics , the undervoltage protection is activated which forces the outputs to the high-impedance state. When the supply voltage falls back to within the operating range, the device restarts.

Overcurrent protection (OCP)

If the output current exceeds the value for I_{OCP} given in Table 5. Electrical characteristics , the overcurrent protection is activated which forces the outputs to the high-impedance state. Periodically, the device attempts to restart. If the overcurrent condition is still present then the OCP remains active. The restart time, T_{OC}, is determined by the RC components connected to pin STBY.

Thermal protection (OTP)

If the junction temperature, T_j, reaches 145 °C (nominally), the device goes to mute mode and the positive and negative PWM outputs are forced to 50% duty cycle. If the junction temperature reaches the value for T_j given in Table 5. Electrical characteristics , the device shuts down and the output is forced to the high-impedance state. When the device cools sufficiently, the device restarts.

5.8 Diagnostic output

The output pin DIAG is an open drain transistor. When any protection is activated it switches to the high-impedance state. The pin can be connected to a power supply (< 36 V) by a pull-up resistor whose value is limited by the maximum sinking current (200 µA) of the pin.

Figure 27. Behavior of pin DIAG for various protection conditions
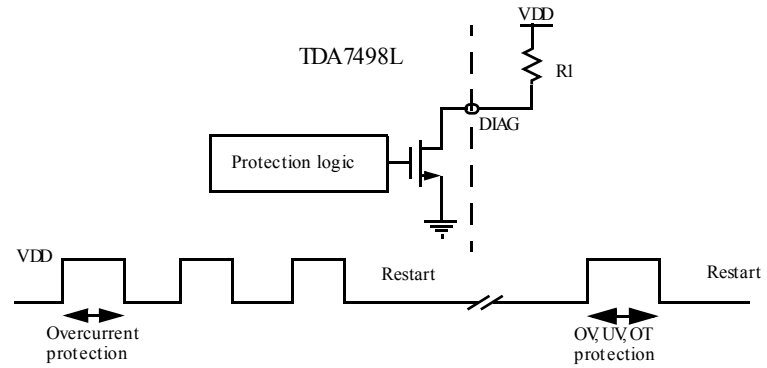
6   Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

6.1   PowerSSO-36 EPU package information
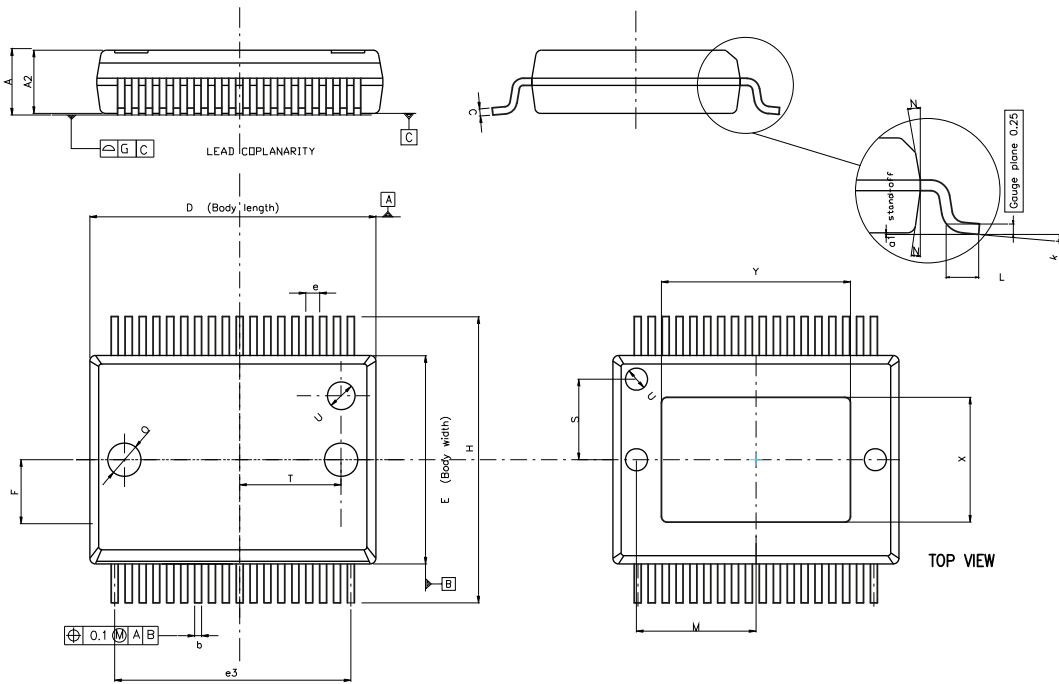
Figure 28. PowerSSO-36 EPU package outline

Table 9. PowerSSO-36 EPU package mechanical data

Symbol	mm			inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	2.15	-	2.45	0.085	-	0.096
A2	2.15	-	2.35	0.085	-	0.093
a1	0	-	0.10	0	-	0.004
b	0.18	-	0.36	0.007	-	0.014
c	0.23	-	0.32	0.009	-	0.013
D	10.10	-	10.50	0.398	-	0.413
E	7.40	-	7.60	0.291	-	0.299
e	-	0.5	-	-	0.020	-
e3	-	8.5	-	-	0.335	-
F	-	2.3	-	-	0.091	-
G	-	-	0.10	-	-	0.004
H	10.10	-	10.50	0.398	-	0.413
h	-	-	0.40	-	-	0.016
k	0	-	8 degrees	0	-	8 degrees
L	0.55	-	0.85	0.022	-	0.033
M	-	4.30	-	-	0.169	-
N	-	-	10 degrees	-	-	10 degrees
O	-	1.20	-	-	0.047	-
Q	-	0.80	-	-	0.031	-
S	-	2.90	-	-	0.114	-
T	-	3.65	-	-	0.144	-
U	-	1.00	-	-	0.039	-
X	4.10	-	4.70	0.161	-	0.185
Y	6.50	-	7.10	0.193	-	0.280

Revision history

Table 10. Document revision history

Date	Revision	Changes
04-Dec-2009	1	Initial release.
02-Jul-2010	2	Removed datasheet preliminary status, updated Section Features list and Table 1. Device summary Updated minimum supply voltage and temperature range in Table 5. Recommended operating conditions Updated typical power output for 8 Ω to 32 V in Table 6. Electrical specifications
12-Sep-2011	3	Updated OUTNA in Table 2. Pin description list; minor textual updates
09-Sep-2015	4	Updated VCC_MAX in Table 3. Absolute maximum ratings and dimension L in Table 10. PowerSSO-36 EPU package mechanical data
24-Jan-2019	5	Updated product summary table in cover page
26-Jul-2020	6	Updated Figure 28 and Y min. value in Table 9

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