

LED Driver IC with Average-Mode Constant Current Control

Features

- · Fast Average Current Control
- Correction for Propagation Delay and Offset Voltage
- · Fixed Offtime Switching Mode
- Linear Dimming Input
- PWM Dimming Input
- Output Short-Circuit Protection with Programmable Skip Mode
- · Input Undervoltage Shutdown

Applications

- · Backlighting of LCD Panels
- · General Lighting

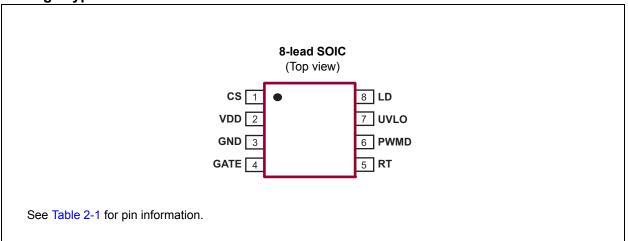
General Description

The HV9803 is an open-loop, Average-mode current control LED driver IC operating in a constant Offtime mode. The IC features ±2% current accuracy and tight line and load regulation of the LED current without any need for loop compensation or high-side current sensing. Its auto-zero circuit cancels the effect of both the input offset voltage and the propagation delay in the current sense comparator.

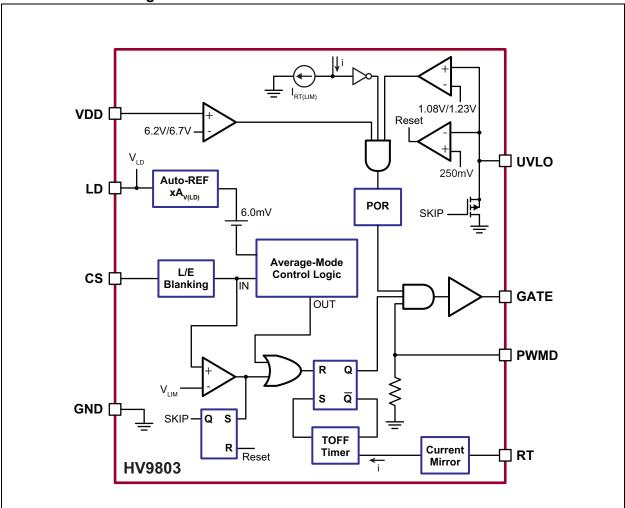
The HV9803 can be powered from a 7V to 13.2V supply. The IC features fast PWM dimming response. The linear dimming input LD can accept a reference voltage of up to 2.5V.

The IC is equipped with a current limit comparator for Hiccup-mode output short-circuit protection. It also features a programmable input undervoltage shutdown.

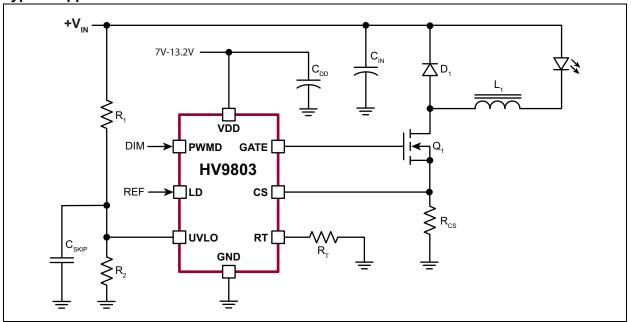
Package Type



Functional Block Diagram



Typical Application Circuit



1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings†

V _{DD} , GATE, CS	
LD, RT, PWMD, UVLO	0.3V to +6V
Junction Temperature, T _{.1}	
Storage Temperature, T _S	
Power Dissipation (at 25 °C):	
8-lead SOIC	650 mW

† Notice: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

Damana atau	0	Min	T	Mare	11	Conditions			
+125°C. Unless otherwise noted, T_A = 25°C, V_{DD} = 12V and V_{PWMD} = 5V.									
Electrical Specifications : Specifications apply over the full operating ambient temperature range, $T_A = -40$ °C to									

120 C. Chiloco Galerinico Hetea, 1A	.o o, v _{DD}	_ ,	PWWD	• • • • • • • • • • • • • • • • • • • •				
Parameter	Sym.	Min.	Тур.	Max	Unit	Conditions		
INPUT								
Input DC Supply Voltage Range	V_{DD}	_	_	13.2	V	DC input voltage		
Quiescent V _{DD} Supply Current	I _{DD}	_	1.5	2.5	mA	V _{CS} = 0V		
V _{DD} UNDERVOLTAGE LOCKOUT			•					
V _{DD} Undervoltage Lockout Upper Threshold	V _{DDUV,R}	6.45	6.7	6.95	٧	V _{DD} Rising		
V _{DD} Undervoltage Lockout Hysteresis	ΔV_{DDUV}		500	_	mV	V _{DD} Falling		
PWM DIMMING	•							
PWMD Input Low Voltage	V _{PWMD(LO)}	_	_	1	V			
PWMD Input High Voltage	V _{PWMD(HI)}	2.6	_	_	V			
Internal Pull-Down Resistance at PWMD	R _{PWMD}	50	100	150	kΩ			
CURRENT SENSE COMPARATOR			•					
External Reference Voltage	V_{LD}	0	_	3	V			
CS Threshold Voltage	V _{CST}	762	778	794	mV	V _{LD} = 1.6V		
CS Threshold Voltage		955	975	995	IIIV	V _{LD} = 2V		
LD to CS Voltage Ratio	A _{V(LD)}	_	0.49	_	_			
Current Sense Blanking Interval	T _{BLANK}	150	_	280	ns			
Minimum On-Time	T _{ON(MIN)}	_	_	760	ns	$V_{CS} = 0.5 V_{LD} + 30 \text{ mV}$		
Maximum Steady State Duty Cycle	D _{MAX}	80	_	_	%	Reduction in output LED current may occur beyond this duty cycle.		
SHORT-CIRCUIT PROTECTION								
Current Limit Threshold Voltage	V_{LIM}	1.57	1.75	1.93	V			
Current Limit Delay CS-to-GATE	T _{DELAY}	_	_	150	ns	V _{CS} = V _{LIM} + 30 mV		
UVLO Skip Timer Reset Switch Resistance	R _{UVRST}	_	_	500	Ω			

DC ELECTRICAL CHARACTERISTICS (CONTINUED)

Electrical Specifications: Specifications apply over the full operating ambient temperature range, $T_A = -40^{\circ}\text{C}$ to +125°C. Unless otherwise noted, $T_A = 25^{\circ}\text{C}$, $V_{DD} = 12\text{V}$ and $V_{PWMD} = 5\text{V}$.

Parameter	Sym.	Min.	Тур.	Max	Unit	Conditions	
UVLO Skip Timer Reset Voltage	V _{UVRST}	200	_	300	mV		
Minimum On-Time (Short Circuit)	T _{ON(MIN)}	_	_	430	ns	V _{CS} = V _{LIM} + 30 mV	
T _{OFF} TIMER			•	•			
Offtime	т.	6.7	9	11.3	μs	R _T = 250 kΩ	
Online	T _{OFF}	0.8	1	1.2	μs	R _T = 25 kΩ	
RT Overcurrent Threshold	I _{RT(LIM)}	_	2.8	_	mA		
GATE DRIVER							
Gate Sourcing Current	I _{SOURCE}	0.165	_	_	Α	V _{GATE} = 0V	
Gate Sinking Current	I _{SINK}	0.165	_	_	Α	V _{GATE} = V _{DD}	
GATE Output Rise Time	t _{RISE}	_	30	50	ns	C _{GATE} = 500 pF	
GATE Output Fall Time	t _{FALL}	_	30	50	ns	C _{GATE} = 500 pF	
UVLO	•						
Undervoltage Upper Threshold Voltage	V _{UVLO,R}	1.17	1.23	1.29	V	V _{UVLO} rising	
Undervoltage Threshold Voltage Hysteresis	ΔV _{UVLO}	_	150	_	mV	V _{UVLO} falling	

TEMPERATURE SPECIFICATIONS

Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions		
TEMPERATURE RANGE								
Operating Ambient Temperature	T _A	-40	_	+125	°C			
Maximum Junction Temperature	TJ	_	_	+150	°C			
Storage Temperature	T _S	-65	_	+150	°C			
PACKAGE THERMAL RESISTANCE								
8-lead SOIC	θ_{JA}	_	101	_	°C/W			

2.0 PIN DESCRIPTION

The details on the pins of HV9803 are listed in Table 2-1. See location of pins in **Package Type**.

TABLE 2-1: PIN FUNCTION TABLE

Pin Number	Pin Name	Description
1	CS	This pin is the current sense pin used to detect the MOSFET source current by means of an external sense resistor.
2	VDD	This is the power supply input for the GATE output and input of the low-voltage regulator powering the internal logic. It must be bypassed with a low-ESR capacitor to GND (at least $0.1~\mu F$).
3	GND	Ground return for all internal circuitry. This pin must be electrically connected to the ground of the power train.
4	GATE	This pin is the output of gate driver for an external N-channel power MOSFET.
5	RT	A resistor connected between RT and GND programs the GATE offtime.
6	PWMD	This is the PWM dimming input of the IC. When this pin is pulled to GND, the gate driver is turned off. When the pin is pulled high, the gate driver operates normally.
7	UVLO	This pin is the undervoltage comparator input. It is also used to program a short-circuit protection skip delay.
8	LD	This pin is the reference voltage input for programming the LED current.

3.0 FUNCTIONAL DESCRIPTION

3.1 General

The peak-current control of a buck converter is the most economical and simplest way to regulate its output current. However, it suffers accuracy and regulation problems that arise from peak-to-average current error due to the current ripple in the output inductor and the propagation delay in the current sense comparator. The full inductor current signal is unavailable for switch current sensing with current-sensing resistor in the ground path between the low-side switch and ground in a buck converter with low-side main switch configuration when the switch is turned-on. While it is very simple to detect the peak current in the switch, controlling the average inductor current is usually implemented by level-translating the current sense signal from the positive input supply rail. While this is practical for relatively low-input voltage, this type of average-current control may become excessively complex and expensive in the case of input voltages above 100V.

The HV9803 uses a control scheme that senses only the switch current to quickly and accurately control the average current in the buck inductor. No compensation of the current control loop is required. The inductor current ripple amplitude does not affect this control scheme significantly. The LED current is independent of the variation in inductance, switching frequency and output voltage. Constant offtime control of the buck converter is used for stability and to improve the LED current regulation over a wide range of input voltages. The IC features excellent PWM dimming response.

3.2 OFF Timer

In the HV9803, the timing resistor connected at the RT pin determines the offtime of the gate driver, and the resistor must be wired to GND. The equation governing the offtime of the GATE output is derived with Equation 3-1.

EQUATION 3-1:

$$T_{OFF} = R_T \times 40pF$$

The RT input is protected from short circuit. Overcurrent condition at RT inhibits the IC.

3.3 Current Sense Comparator and Timer Circuits

The function of the HV9803's current sense comparator is similar to that of a peak current controller. However, the GATE pulse is not terminated immediately as the CS threshold is met. The GATE turn-off in the nth cycle is delayed by a time $T_{2,n}$ determined by a timer circuit as shown in Equation 3-2.

EQUATION 3-2:

$$T_{2,n} = \frac{1}{2} \times (T_{1,n} + T_{1,n-1})$$

Where $T_{1,n}$ and $T_{1,n-1}$ are the times to the CS threshold in any two consecutive switching cycles.

This iterative control law is needed for damping sub-harmonic oscillation. Note that the control law is only valid up to a maximum switching duty cycle, $D_{MAX} = 0.8.$ Exceeding D_{MAX} will cause a reduction in the LED current.

Propagation delay in the current sense comparator is one of the most significant contributors to the LED current error. It must be noted that the control scheme described above does not improve this deficiency of the peak-current control scheme by itself. Moreover, it samples the propagation delay during T_1 and replicates it during T_2 , essentially doubling the error introduced by this delay. To eliminate this error, the reference voltage is corrected by an auto-zero circuit. In essence, the HV9803 samples its CS signal when the current sense comparator triggers and detects the difference between the sampled CS level and the reference input of the current sense comparator. The resulting difference is subtracted from the reference level to generate a new reference in the next switching cycle.

3.4 GATE Output

The GATE output of the HV9803 is used to drive an external MOSFET. It is recommended that the gate charge Q_G of the external MOSFET should be less than 25 nC for switching frequencies \leq 100 kHz and less than 15 nC for switching frequencies \geq 100 kHz.

The resulting LED current is calculated using Equation 3-3.

EQUATION 3-3:

$$I_{LED} = \frac{0.49 \times V_{LD} - 6mV}{R_{CS}}$$

3.5 Short-Circuit Protection

The HV9803 is equipped with a short-circuit protection comparator having another CS threshold V_{LIM} . When this second threshold is triggered, the GATE output shuts off for the duration of a restart delay, determined by the RC constant set at UVLO pin. Meanwhile, the capacitor C_{SKIP} is discharged below 200 mV. The restart delay due to charging C_{SKIP} to the UVLO upper threshold is calculated as shown in Equation 3-4.

EQUATION 3-4:

$$T_{SKIP} = k \times R_1 \times C_{SKIP} \times In \times \left(\frac{k \times V_{IN} - 0.30 \, V}{k \times V_{IN} - 1.17 \, V}\right)$$
 Where:
$$k = \frac{R_2}{R_1 + R_2}$$

3.6 Undervoltage Shutdown

Undervoltage comparator input is provided to disable the IC when the UVLO input is below the UVLO lower threshold. Hysteresis is provided to avoid oscillation.

3.7 Failure Modes and Effects Analysis (FMEA)

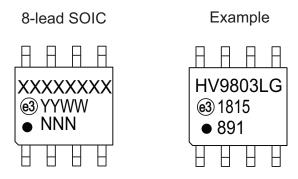
The HV9803 is designed to withstand short circuit between its adjacent pins without damage. Table 3-1 describes the effect of such incidental short-circuit conditions.

TABLE 3-1: SHORT-CIRCUIT MODES AND EFFECTS

Short-Circuit Mode	Effect
CS to V _{DD}	The IC triggers the short-circuit protection and operates in the Auto-restart mode continuously.
V _{DD} to GND	Short circuit across the 12V should cause the external bias supply overcurrent protection.
GND to GATE	This should cause the external bias supply overcurrent protection. The power MOSFET Q1 is off.
RT to PWMD	Case 1-PWMD = Lo: The RT pin sources its maximum current. GATE = 0V and Q1 is off. Case 2-PWMD = Hi: The RT pin is pulled up, shutting off the timer. GATE is off.
PWMD to UVLO	This overdrives the undervoltage threshold. However, since the V_{IN} UV condition is harmless to the IC, there is no effect.
UVLO to LD	LD overdrives the UVLO. If LD is lower than the UVLO lower threshold, the IC shuts off. No effect otherwise.

4.0 PACKAGING INFORMATION

4.1 Package Marking Information



Legend: XX...X Product Code or Customer-specific information

Y Year code (last digit of calendar year)
YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

e3 Pb-free JEDEC® designator for Matte Tin (Sn)

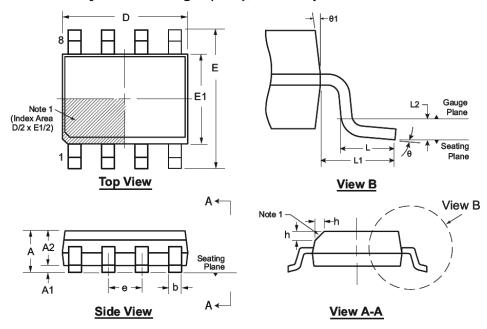
This package is Pb-free. The Pb-free JEDEC designator (e3)

can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for product code or customer-specific information. Package may or not include the corporate logo.

8-Lead SOIC (Narrow Body) Package Outline (LG/TG)

4.90x3.90mm body, 1.75mm height (max), 1.27mm pitch



Note: For the most current package drawings, see the Microchip Packaging Specification at www.microchip.com/packaging.

Note

 This chamfer feature is optional. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.

Symbo	I	Α	A1	A2	b	D	E	E 1	е	h	L	L1	L2	θ	θ1
	MIN	1.35*	0.10	1.25	0.31	4.80*	5.80*	3.80*		0.25	0.40			0 o	5º
Dimension (mm)	NOM	-	-	-	-	4.90	6.00	3.90	1.27 BSC	-	-	1.04 REF	0.25 BSC	-	-
()	MAX	1.75	0.25	1.65*	0.51	5.00*	6.20*	4.00*	500	0.50	1.27		500	8º	15º

JEDEC Registration MS-012, Variation AA, Issue E, Sept. 2005.

Drawings are not to scale.

^{*} This dimension is not specified in the JEDEC drawing.

APPENDIX A: REVISION HISTORY

Revision A (August 2018)

- Converted Supertex Doc# DSFP-HV9803 to Microchip DS20005641A
- Changed the packaging quantity for the 8-lead SOIC LG package from 2500/Reel to 3300/Reel
- Added a maximum junction temperature to the Temperature Specifications Table
- Made minor text changes throughout the document

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, contact your local Microchip representative or sales office.

XX		- х - х	Ex	ample:	
		Environmental Media Type	a)	HV9803LG-G:	LED Driver IC with Average-Mode Constant Current Control, 8-lead SOIC Package, 3300/Reel
HV9803	=	LED Driver IC with Average-Mode Constant Current Control			
LG	=	8-lead SOIC			
G	=	Lead (Pb)-free/RoHS-compliant Package			
(blank)	=	3300/Reel for an LG Package			
	Package Options HV9803 LG G	Package Options HV9803 = LG = G =	Package Options Environmental Media Type HV9803 = LED Driver IC with Average-Mode Constant Current Control LG = 8-lead SOIC G = Lead (Pb)-free/RoHS-compliant Package	Package Options Environmental Media Type a) HV9803 = LED Driver IC with Average-Mode Constant Current Control LG = 8-lead SOIC G = Lead (Pb)-free/RoHS-compliant Package	Package Options Environmental Media Type a) HV9803LG-G: HV9803 = LED Driver IC with Average-Mode Constant Current Control LG = 8-lead SOIC G = Lead (Pb)-free/RoHS-compliant Package

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AP5725WUG-7 STP4CMPQTR NCL30086BDR2G CAT4004BHU2-GT3 LV52207AXA-VH AP1694AS-13 TLE4242EJ AS3688

IS31LT3172-GRLS4-TR TLD2311EL KTD2694EDQ-TR KTZ8864EJAA-TR IS32LT3174-GRLA3-TR ZXLD1374QESTTC MP2488DN-LF-Z NLM0010XTSA1 AL1676-20BS7-13 MPQ7220GF-AEC1-P MPQ4425BGJ-AEC1-P MPQ7220GF-AEC1-Z MPQ4425BGJ-AEC1-Z IS31FL3737B-QFLS4-TR IS31FL3239-QFLS4-TR KTD2058EUAC-TR KTD2037EWE-TR DIO5662ST6 IS31BL3508A-TTLS2-TR

MAX20052CATC/V+ MAX25606AUP/V+ BD6586MUV-E2 BD9206EFV-E2 BD9416FS-E2 LYT4227E LYT6079C-TL MP3394SGF-P MP4689AGN-P