

TPRT9266 Step-up DC/DC converter

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General Description

The TPRT9266 sasmall,highefficiency,andlowvoltage step-up DC/DC converter with an Adaptive Current Mode

PWM control loop, includes an error amplifier, ramp generator, comparator, switch pass element and driver in which providing a stable and high efficient operation over a wide range of load currents. It operates in stable waveforms without external compensation.

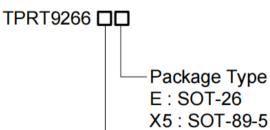
Thelowstart-upinputvoltagebelow1VmakesTPRT9266 suitable for 1 to 4 battery cells applications of providing up to 300mA output current. The 450KHz high switching rate minimized the size of external components. Besides, the 17μ A low quiescent current together with high efficiency maintains long battery lifetime.

TheTPRT9266 isavailableinsmallpackageSOT-26and SOT89-5

Applications

- LCD Panel
- DSC
- ♦ MP3
- Wireless Equipment
- Portable Instrument
- PDA

Ordering Information



- Operating Temperature Range

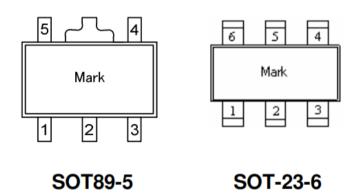
G : Green (Halogen Free and Pb Free)

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Features

- 90% Efficiency
- High Supply Capability to Deliver 3.3V 100mA with 1 Alkaline cell
- ♦ 17µA Quiescent (switch-off) Supply Current
- ♦ 450KHz Fixed Switching Frequency
- Zero Shutdown Mode Supply Current
- Providing Flexibility for Using Internal and External Power Switches
- ◆ 1.0V Low Start-up Input Voltage
- Small SOT-26 & SOT89-5 Package

Package Information



Pin Configurations

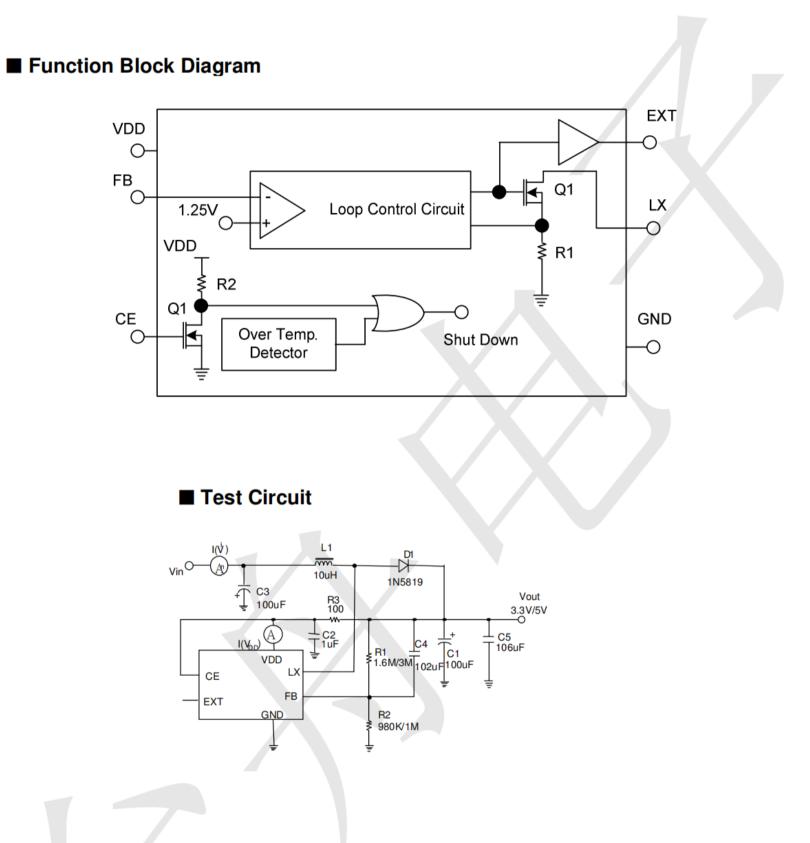
SOT-26	SOT-89-5	Pin Name	Pin Function
1	1	CE	Chip enable
2		EXT	Output
3	5	GND	Ground
4	4	LX	Switching
5	2	VDD	Input
6	3	FB	Feedback input 1.25V

P : P : Pb Free



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■ Absolute Maximum Ratings (Ta=25°C)

Paramenter	Symbol	Ratings	Unit	
Supply Voltage	Vin	- 0.3V ~ 7V	V	
LX Pin Switch Voltage	VLX	- 0.3 ~ (VDD + 0.8V)	V	
Other I/O Pin Voltages	Vss	- 0.3V ~ (VDD + 0.3V)	V	
Lx Pin Switch Current	ILX	2.5	Α	
EXT Pin Driver Current	IEXT	200	mA	
Package Thermal Resistance	SOT-26	Pd	145	°C/W
	SOT-89-5		45	°C/W
Operating Junction Temperature	Tj	125	°C	
Storage Temperature Range	Tstg	- 65 ~ +150	°C	

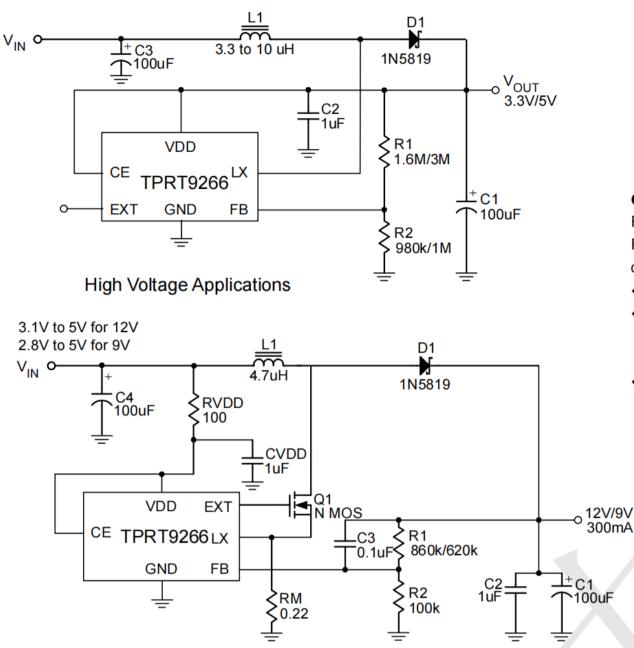
■ Electrical Characteristics (Vin =1.5V, VDD=3.3V, I=0, Ta = 25°C, unless otherwise spec

Parameter	Symbol	Test Conditions	Min	Тур	Max	Units
Start-UP Voltage	V _{ST}	l _L = 1mA		0.98	1.05	V
Operating VDD Range	V _{DD}	VDD pin voltage	2		6.5	V
No Load Current I (V _{IN})	INO LOAD	$V_{IN} = 1.5V, V_{OUT} = 3.3V$		75		μA
Switch-off Current I (VDD)	ISWITCH OFF	$V_{IN} = 6V$		17		μA
Shutdown Current I (V _{IN})	I _{OFF}	CE Pin = 0V, $V_{IN} = 4.5V$		0.01	1	μA
Feedback Reference Voltage	V _{REF}	Close Loop, VDD = 3.3V	1.225	1.25	1.275	V
Switching Frequency	Fs	VDD = 3.3V		450		KHz
Maximum Duty	D _{MAX}	VDD = 3.3V		95		%
LX ON Resistance		VDD = 3.3V		0.3		Ω
Current Limit Setting	ILIMIT	VDD = 3.3V		2		А
EXT ON Resistance to VDD		VDD = 3.3V		5		Ω
EXT ON Resistance to GND		VDD = 3.3V		5		Ω
Line Regulation	$\triangle V_{\text{LINE}}$	V _{IN} = 1.5 ∼ 2.5V, I _L = 1mA		10		mV/V
Load Regulation	$\triangle V_{LOAD}$	$V_{IN} = 2.5V, I_L = 1 \sim 100 \text{mA}$		0.25		mV/mA
CE Pin Trip Level		VDD = 3.3V	0.4	0.8	1.2	V
Temperature Stability for Vout	Ts			50		ppm /℃
Thermal Shutdown	T _{SD}			165		°C
Thermal Shutdown Hysterises	ΔT_{SD}			10		°C

* Note: The CE pin shall be tied to VDD pin and inhibit to act the ON/OFF state whenever the VDD pin voltage may reach to 5.5V or above.

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Application for Portable Instruments



Application Note

Output Voltage Setting

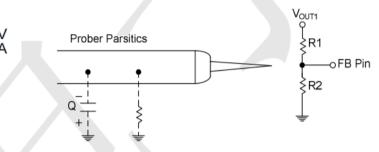
Referring to application circuits, the output voltage of the switching regulator (V_{OUT}) can be set

$$V_{OUT1} = (1 + \frac{R1}{R2}) \times 1.25V$$
A

Feedback Loop Design

Referring to application circuits, The selection of R1and R2 based on the trade-off between quiescent current consumption and interference immunity is stated below:

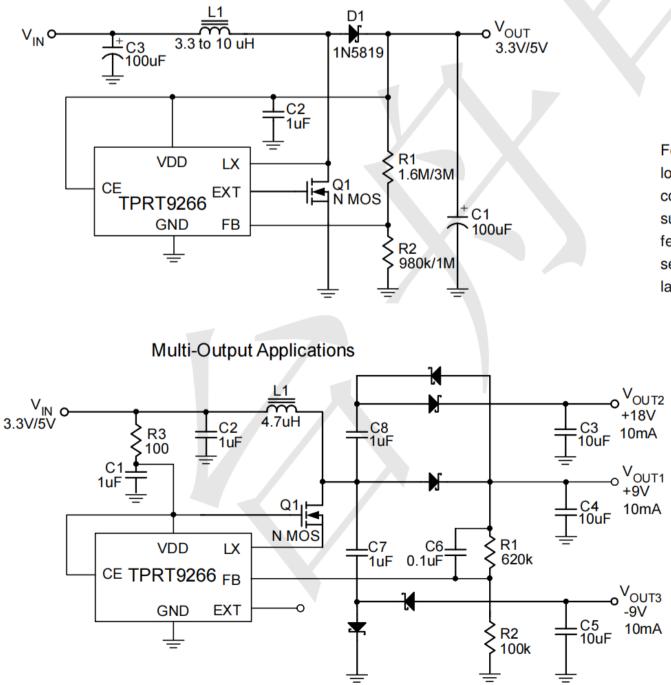
- Follow Equation A
- Higher R reduces the quiescent current (Path current =1.25V/R2), however resistors beyond 5MΩ are not recommended.
- Lower R gives better noise immunity, and is less sensitive to interference, layout parasitics, FB node leakage, and improper probing to FB pins.



 A proper value of feed forward capacitor parallel with R1 can improve the noise immunity of the feedback loops, especially in an improper layout. An empirical suggestion is around 0~33pF for feedback resistors of MΩ, and 10nF~0.1µF for feedback resistors of tens to hundreds KΩ.

For applications without standby or suspend modes, lower values of R1 and R2 are preferred. For applications concerning the current consumption in standby or suspend modes, the higher values of R1and R2 are feedback needed. Such "high impedance loops" are sensitive to any interference, which require careful layout and avoid any interference, e.g. probing to FB pin.





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