

## AUTOSWITCHING POWER MUX

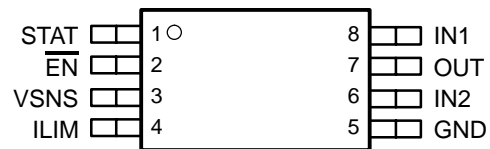
### FEATURES

- Two-Input, One-Output Power Multiplexer With Low  $r_{DS(on)}$  Switches:
  - 84 m $\Omega$  Typ (TPS2113)
  - 120 m $\Omega$  Typ (TPS2112)
- Reverse and Cross-Conduction Blocking
- Wide Operating Voltage Range . . . . 2.8 V to 5.5 V
- Low Standby Current . . . . 0.5- $\mu$ A Typ
- Low Operating Current . . . . 55- $\mu$ A Typ
- Adjustable Current Limit
- Controlled Output Voltage Transition Times, Limits Inrush Current and Minimizes Output Voltage Hold-Up Capacitance
- CMOS and TTL Compatible Control Inputs
- Auto-Switching Operating Mode
- Thermal Shutdown
- Available in a TSSOP-8 Package

### APPLICATIONS

- PCs
- PDAs
- Digital Cameras
- Modems
- Cell phones
- Digital Radios
- MP3 Players

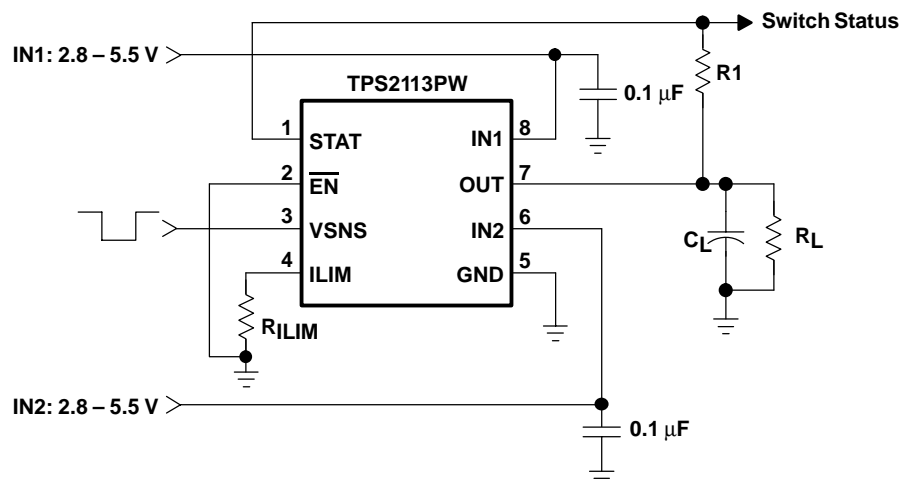
PW PACKAGE  
(TOP VIEW)



### DESCRIPTION

The TPS211x family of power multiplexers enables seamless transition between two power supplies, such as a battery and a wall adapter, each operating at 2.8–5.5 V and delivering up to 1 A. The TPS211x family includes extensive protection circuitry, including user-programmable current limiting, thermal protection, inrush current control, seamless supply transition, cross-conduction blocking, and reverse-conduction blocking. These features greatly simplify designing power multiplexer applications.

### TYPICAL APPLICATION



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### AVAILABLE OPTIONS

FEATURE		TPS2110	TPS2111	TPS2112	TPS2113	TPS2114	TPS2115
Current Limit Adjustment Range		0.31–0.75A	0.63–1.25A	0.31–0.75A	0.63–1.25A	0.31–0.75A	0.63–1.25A
Switching modes	Manual	Yes	Yes	No	No	Yes	Yes
	Automatic	Yes	Yes	Yes	Yes	Yes	Yes
Switch Status Output		No	No	Yes	Yes	Yes	Yes
Package		TSSOP-8	TSSOP-8	TSSOP-8	TSSOP-8	TSSOP-8	TSSOP-8

### ORDERING INFORMATION

T <sub>A</sub>	PACKAGE	ORDERING NUMBER(1)	MARKINGS
–40°C to 85°C	TSSOP-8 (PW)	TPS2112PW	2112
		TPS2113PW	2113

(1) The PW package is available taped and reeled. Add an R suffix to the device type (e.g., TPS2112PWR) to indicate tape and reel.

### PACKAGE DISSIPATION RATINGS

PACKAGE	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> ≤ 25°C POWER RATING	T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 85°C POWER RATING
TSSOP-8 (PW)	3.87 mW/°C	386.84 mW	212.76 mW	154.73 mW

### ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted(1)

	TPS2112, TPS2113	
Input voltage range at pins IN1, IN2, $\overline{\text{EN}}$ , VSNS, ILIM(2)	–0.3 V to 6 V	
Output voltage range, V <sub>O(OUT)</sub> , V <sub>O(STAT)</sub> (2)	–0.3 V to 6 V	
Output sink current, I <sub>O(STAT)</sub>	5 mA	
Continuous output current, I <sub>O</sub>	TPS2112	0.9 A
	TPS2113	1.5 A
Continuous total power dissipation	See Dissipation Rating Table	
Operating virtual junction temperature range, T <sub>J</sub>	–40°C to 125°C	
Storage temperature range, T <sub>stg</sub>	–65°C to 150°C	
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds	260°C	

(1) Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages are with respect to GND.

### RECOMMENDED OPERATING CONDITIONS

		MIN	MAX	UNIT
Input voltage at IN1, V <sub>I(IN1)</sub>	V <sub>I(IN2)</sub> ≥ 2.8 V	1.5	5.5	V
	V <sub>I(IN2)</sub> < 2.8 V	2.8	5.5	
Input voltage at IN2, V <sub>I(IN2)</sub>	V <sub>I(IN1)</sub> ≥ 2.8 V	1.5	5.5	V
	V <sub>I(IN1)</sub> < 2.8 V	2.8	5.5	
Input voltage, V <sub>I(EN)</sub> , V <sub>I(VSNS)</sub>		0	5.5	V
Current limit adjustment range, I <sub>O(OUT)</sub>	TPS2112	0.31	0.75	A
	TPS2113	0.63	1.25	
Operating virtual junction temperature, T <sub>J</sub>		–40	125	°C

### ELECTROSTATIC DISCHARGE (ESD) PROTECTION

	MIN	MAX	UNIT
Human body model		2	kV
CDM		500	V

## ELECTRICAL CHARACTERISTICS

over recommended operating junction temperature range,  $V_I(IN1) = V_I(IN2) = 5.5\text{ V}$ ,  $R_{ILIM} = 400\ \Omega$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TPS2112			TPS2113			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	
<b>POWER SWITCH</b>								
$r_{DS(on)}^{(1)}$ Drain-source on-state resistance (INx–OUT)	$T_J = 25^\circ\text{C}$ , $I_L = 500\text{ mA}$	$V_I(IN1) = V_I(IN2) = 5.0\text{ V}$	120	140	84	110	m $\Omega$	
		$V_I(IN1) = V_I(IN2) = 3.3\text{ V}$	120	140	84	110		
		$V_I(IN1) = V_I(IN2) = 2.8\text{ V}$	120	140	84	110		
	$T_J = 125^\circ\text{C}$ , $I_L = 500\text{ mA}$	$V_I(IN1) = V_I(IN2) = 5.0\text{ V}$		220		150	m $\Omega$	
		$V_I(IN1) = V_I(IN2) = 3.3\text{ V}$		220		150		
		$V_I(IN1) = V_I(IN2) = 2.8\text{ V}$		220		150		

(1) The TPS211x can switch a voltage as low as 1.5 V as long as there is a minimum of 2.8 V at one of the input power pins. In this specific case, the lower supply voltage has no effect on the IN1 and IN2 switch on-resistances.

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>LOGIC INPUTS (EN)</b>					
$V_{IH}$ High-level input voltage		2			V
$V_{IL}$ Low-level input voltage				0.7	V
Input current	$\overline{\text{EN}}$ = High, sink current			1	$\mu\text{A}$
	$\overline{\text{EN}}$ = Low, source current	0.5	1.4	5	
<b>SUPPLY AND LEAKAGE CURRENTS</b>					
Supply current from IN1 (operating)	$V_I(VSNS) = 1.5\text{ V}$ , $\overline{\text{EN}}$ = Low (IN1 active), $V_I(IN1) = 5.5\text{ V}$ , $V_I(IN2) = 3.3\text{ V}$ , $I_O(OUT) = 0\text{ A}$		55	90	$\mu\text{A}$
	$V_I(VSNS) = 1.5\text{ V}$ , $\overline{\text{EN}}$ = Low (IN1 active), $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5.5\text{ V}$ , $I_O(OUT) = 0\text{ A}$		1	12	
	$V_I(VSNS) = 0\text{ V}$ , $\overline{\text{EN}}$ = Low (IN2 active), $V_I(IN1) = 5.5\text{ V}$ , $V_I(IN2) = 3.3\text{ V}$ , $I_O(OUT) = 0\text{ A}$			75	
	$V_I(VSNS) = 0\text{ V}$ , $\overline{\text{EN}}$ = Low (IN2 active), $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5.5\text{ V}$ , $I_O(OUT) = 0\text{ A}$			1	
Supply current from IN2 (operating)	$V_I(VSNS) = 1.5\text{ V}$ , $\overline{\text{EN}}$ = Low (IN1 active), $V_I(IN1) = 5.5\text{ V}$ , $V_I(IN2) = 3.3\text{ V}$ , $I_O(OUT) = 0\text{ A}$			1	$\mu\text{A}$
	$V_I(VSNS) = 1.5\text{ V}$ , $\overline{\text{EN}}$ = Low (IN1 active), $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5.5\text{ V}$ , $I_O(OUT) = 0\text{ A}$			75	
	$V_I(VSNS) = 0\text{ V}$ , $\overline{\text{EN}}$ = Low (IN2 active), $V_I(IN1) = 5.5\text{ V}$ , $V_I(IN2) = 3.3\text{ V}$ , $I_O(OUT) = 0\text{ A}$		1	12	
	$V_I(VSNS) = 0\text{ V}$ , $\overline{\text{EN}}$ = Low (IN2 active), $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5.5\text{ V}$ , $I_O(OUT) = 0\text{ A}$		55	90	
Quiescent current from IN1 (STANDBY)	$\overline{\text{EN}}$ = High (inactive), $V_I(IN1) = 5.5\text{ V}$ , $V_I(IN2) = 3.3\text{ V}$ , $I_O(OUT) = 0\text{ A}$		0.5	2	$\mu\text{A}$
	$\overline{\text{EN}}$ = High (inactive), $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5.5\text{ V}$ , $I_O(OUT) = 0\text{ A}$			1	
Quiescent current from IN2 (STANDBY)	$\overline{\text{EN}}$ = High (inactive), $V_I(IN1) = 5.5\text{ V}$ , $V_I(IN2) = 3.3\text{ V}$ , $I_O(OUT) = 0\text{ A}$			1	$\mu\text{A}$
	$\overline{\text{EN}}$ = High (inactive), $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5.5\text{ V}$ , $I_O(OUT) = 0\text{ A}$		0.5	2	
Forward leakage current from IN1 (measured from OUT to GND)	$\overline{\text{EN}}$ = High (inactive), $V_I(IN1) = 5.5\text{ V}$ , IN2 open, $V_O(OUT) = 0\text{ V}$ (shorted), $T_J = 25^\circ\text{C}$		0.1	5	$\mu\text{A}$
Forward leakage current from IN2 (measured from OUT to GND)	$\overline{\text{EN}}$ = High (inactive), $V_I(IN2) = 5.5\text{ V}$ , IN1 open, $V_O(OUT) = 0\text{ V}$ (shorted), $T_J = 25^\circ\text{C}$		0.1	5	$\mu\text{A}$
Reverse leakage current to INx (measured from INx to GND)	$\overline{\text{EN}}$ = High (inactive), $V_I(INx) = 0\text{ V}$ , $V_O(OUT) = 5.5\text{ V}$ , $T_J = 25^\circ\text{C}$		0.3	5	$\mu\text{A}$
<b>STAT OUTPUT</b>					
Leakage current	$V_O(STAT) = 5.5\text{ V}$		0.01	1	$\mu\text{A}$
Saturation voltage	$I_I(STAT) = 2\text{ mA}$ , IN1 switch is on		0.13	0.4	V
Deglitch time (falling edge only)			150		$\mu\text{s}$

**ELECTRICAL CHARACTERISTICS (Continued)**

over recommended operating junction temperature range,  $V_{I(IN1)} = V_{I(IN2)} = 5.5\text{ V}$ ,  $R_{ILIM} = 400\ \Omega$  (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>CURRENT LIMIT CIRCUIT</b>						
Current limit accuracy	TPS2112	$R_{ILIM} = 400\ \Omega$	0.51	0.63	0.80	A
		$R_{ILIM} = 700\ \Omega$	0.30	0.36	0.50	
	TPS2113	$R_{ILIM} = 400\ \Omega$	0.95	1.25	1.56	
		$R_{ILIM} = 700\ \Omega$	0.47	0.71	0.99	
$t_d$	Current limit settling time <sup>(1)</sup>	Time for short-circuit output current to settle within 10% of its steady state value.	1			ms
Input current at ILIM		$V_{I(ILIM)} = 0\text{ V}$ , $I_{O(OUT)} = 0\text{ A}$	-15	0		$\mu\text{A}$

(1) Not tested in production.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>VSNS COMPARATOR</b>						
VSNS threshold voltage	$V_{I(VSNS)} \uparrow$		0.78	0.8	0.82	V
	$V_{I(VSNS)} \downarrow$		0.735	0.755	0.775	
VSNS comparator hysteresis <sup>(1)</sup>			30	60		mV
Deglitch of VSNS comparator (both $\uparrow\downarrow$ ) <sup>(1)</sup>			90	150	220	$\mu\text{s}$
Input current		$0\text{ V} \leq V_{I(VSNS)} \leq 5.5\text{ V}$	-1	1		$\mu\text{A}$
<b>UVLO</b>						
IN1 and IN2 UVLO	Falling edge		1.15	1.25		V
	Rising edge		1.30		1.35	
IN1 and IN2 UVLO hysteresis <sup>(1)</sup>			30	57	65	mV
Internal $V_{DD}$ UVLO (the higher of IN1 and IN2)	Falling edge		2.4	2.53		V
	Rising edge		2.58		2.8	
Internal $V_{DD}$ UVLO hysteresis <sup>(1)</sup>			30	50	75	mV
UVLO deglitch for IN1, IN2 <sup>(1)</sup>		Falling edge	110			$\mu\text{s}$

(1) Not tested in production.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>REVERSE CONDUCTION BLOCKING</b>						
$\Delta V_{O(I\_block)}$	Minimum output-to-input voltage difference to block switching	$\overline{EN} = \text{high}$ , $V_{I(IN1)} = 3.3\text{ V}$ and $V_{I(IN2)} = V_{I(VSNS)} = 0\text{ V}$ . Connect OUT to a 5 V supply through a series 1-k $\Omega$ resistor. Let $\overline{EN} = \text{low}$ . Slowly decrease the supply voltage until OUT connects to IN1.	80	100	120	mV

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>THERMAL SHUTDOWN</b>						
Thermal shutdown threshold <sup>(1)</sup>		TPS211x is in current limit.	135			$^{\circ}\text{C}$
Recovery from thermal shutdown <sup>(1)</sup>		TPS211x is in current limit.	125			
Hysteresis <sup>(1)</sup>			10			
<b>IN2-IN1 COMPARATORS</b>						
Hysteresis of IN2-IN1 comparator			0.1	0.2		V
Deglitch of IN2-IN1 comparator, (both $\uparrow\downarrow$ ) <sup>(1)</sup>			90	150	220	$\mu\text{s}$

(1) Not tested in production.

## SWITCHING CHARACTERISTICS

over recommended operating junction temperature range,  $V_I(IN1) = V_I(IN2) = 5.5\text{ V}$ ,  $R_{ILIM} = 400\ \Omega$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS	TPS2112			TPS2113			UNIT			
		MIN	TYP	MAX	MIN	TYP	MAX				
<b>POWER SWITCH</b>											
$t_r$	Output rise time from an enable(1)	$V_I(IN1) = V_I(IN2) = 5\text{ V}$ , $V_I(VSNS) = 1.5\text{ V}$	$T_J = 25^\circ\text{C}$ , $C_L = 1\ \mu\text{F}$ , $I_L = 500\text{ mA}$ , See Figure 1(a)	0.5	1.0	1.5	1	1.8	3	ms	
$t_f$	Output fall time from a disable(1)	$V_I(IN1) = V_I(IN2) = 5\text{ V}$ , $V_I(VSNS) = 1.5\text{ V}$	$T_J = 25^\circ\text{C}$ , $C_L = 1\ \mu\text{F}$ , $I_L = 500\text{ mA}$ , See Figure 1(a)	0.35	0.5	0.7	0.5	1	2	ms	
$t_t$	Transition time(1)	IN1 to IN2 transition, $V_I(IN1) = 3.3\text{ V}$ , $V_I(IN2) = 5\text{ V}$ , $V_I(EN) = 0\text{ V}$	$T_J = 125^\circ\text{C}$ , $C_L = 10\ \mu\text{F}$ , $I_L = 500\text{ mA}$ [Measure transition time as 10–90% rise time or from 3.4 V to 4.8 V on $V_O(OUT)$ ], See Figure 1(b)		40	60		40	60	$\mu\text{s}$	
$t_{PLH1}$	Turn-on propagation delay from enable(1)	$V_I(IN1) = V_I(IN2) = 5\text{ V}$ Measured from enable to 10% of $V_O(OUT)$ , $V_I(VSNS) = 1.5\text{ V}$	$T_J = 25^\circ\text{C}$ , $C_L = 10\ \mu\text{F}$ , $I_L = 500\text{ mA}$ , See Figure 1(a)		0.5			1		ms	
$t_{PHL1}$	Turn-off propagation delay from a disable(1)	$V_I(IN1) = V_I(IN2) = 5\text{ V}$ , Measured from disable to 90% of $V_O(OUT)$ , $V_I(VSNS) = 1.5\text{ V}$	$T_J = 25^\circ\text{C}$ , $C_L = 10\ \mu\text{F}$ , $I_L = 500\text{ mA}$ , See Figure 1(a)		3			5		ms	
$t_{PLH2}$	Switch-over rising propagation delay(1)	Logic 1 to Logic 0 transition on VSNS, $V_I(IN1) = 1.5\text{ V}$ , $V_I(IN2) = 5\text{ V}$ , $V_I(EN) = 0\text{ V}$ , Measured from VSNS to 10% of $V_O(OUT)$	$T_J = 25^\circ\text{C}$ , $C_L = 10\ \mu\text{F}$ , $I_L = 500\text{ mA}$ , See Figure 1(c)		0.17	1		0.17	1	ms	
$t_{PHL2}$	Switch-over falling propagation delay(1)	Logic 0 to Logic 1 transition on VSNS $V_I(IN1) = 1.5\text{ V}$ , $V_I(IN2) = 5\text{ V}$ , $V_I(EN) = 0\text{ V}$ , Measured from VSNS to 90% of $V_O(OUT)$	$T_J = 25^\circ\text{C}$ , $C_L = 10\ \mu\text{F}$ , $I_L = 500\text{ mA}$ , See Figure 1(c)		2	3	10	2	5	10	ms

(1) Not tested in production.

TRUTH TABLE

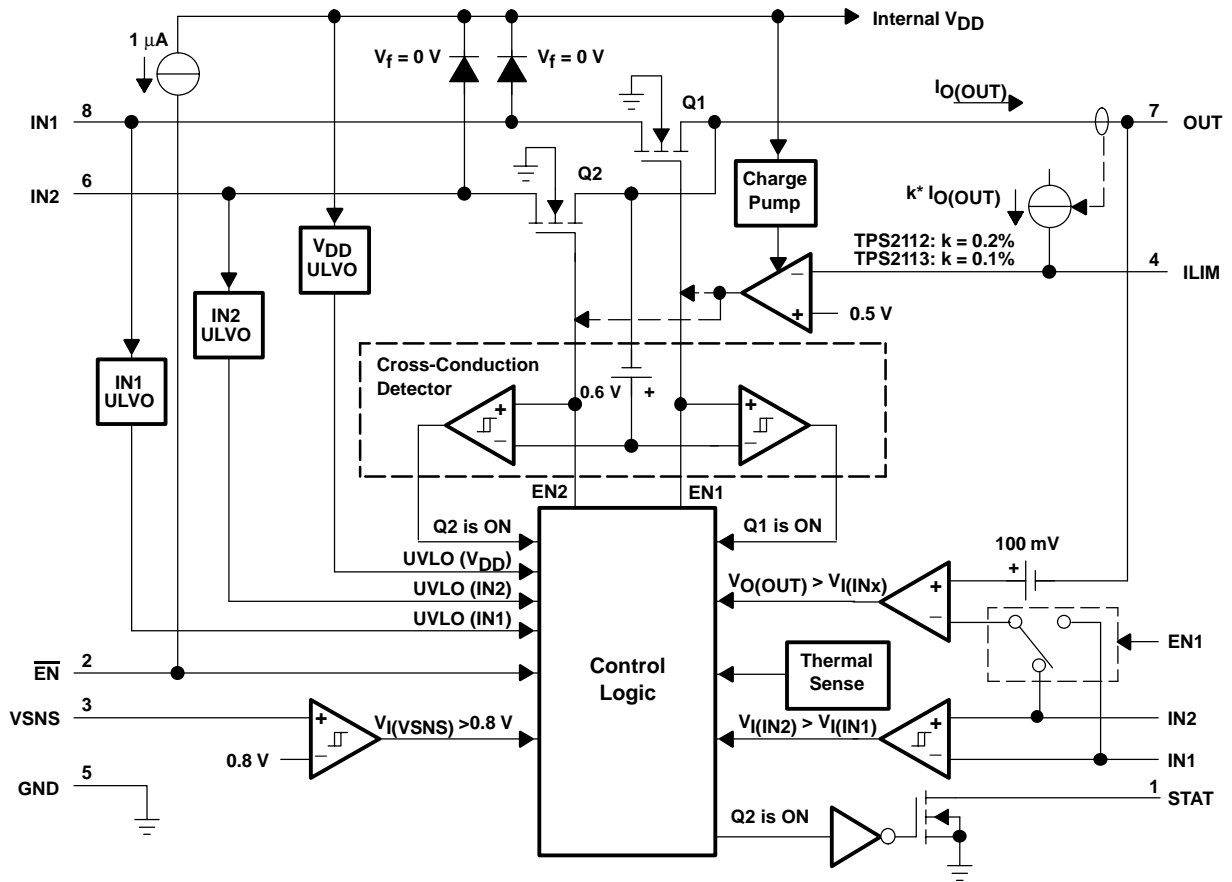
$\overline{\text{EN}}$	$V_{\text{I}}(\text{VSNS}) > 0.8\text{V}$	$V_{\text{I}}(\text{IN2}) > V_{\text{I}}(\text{IN1})$	STAT	OUT(1)
0	Yes	X	0	IN1
	No	No	0	IN1
	No	Yes	Hi-Z	IN2
1	X	X	0	Hi-Z

(1)The under-voltage lockout circuit causes the output OUT to go Hi-Z if the selected power supply does not exceed the IN1/IN2 UVLO, or if neither of the supplies exceeds the internal  $V_{\text{DD}}$  UVLO.

### Terminal Functions

TERMINAL NAME	NO.	I/O	DESCRIPTION
$\overline{\text{EN}}$	2	I	$\overline{\text{EN}}$ is a TTL and CMOS compatible input with a 1- $\mu\text{A}$ pull-up. The truth table shown above illustrates the functionality of $\overline{\text{EN}}$ .
GND	5	I	Ground
IN1	8	I	Primary power switch input. The IN1 switch can be enabled only if the IN1 supply is above the UVLO threshold and at least one supply exceeds the internal $V_{\text{DD}}$ UVLO.
IN2	6	I	Secondary power switch input. The IN2 switch can be enabled only if the IN2 supply is above the UVLO threshold and at least one supply exceeds the internal $V_{\text{DD}}$ UVLO.
ILIM	4	I	A resistor $R_{\text{ILIM}}$ from ILIM to GND sets the current limit $I_{\text{L}}$ to $250/R_{\text{ILIM}}$ and $500/R_{\text{ILIM}}$ for the TPS2112 and TPS2113, respectively.
OUT	7	O	Power switch output
STAT	1	O	STAT is an open-drain output that is Hi-Z if the IN2 switch is ON. STAT pulls low if the IN1 switch is ON or if OUT is Hi-Z (i.e., $\overline{\text{EN}}$ is equal to logic 0)
VSNS	3	I	An internal power FET connects OUT to IN1 if the VSNS voltage is greater than 0.8 V. Otherwise, the FET connects OUT to the higher of IN1 and IN2. The truth table shown above illustrates the functionality of VSNS.

**FUNCTIONAL BLOCK DIAGRAM**



PARAMETER MEASUREMENT INFORMATION

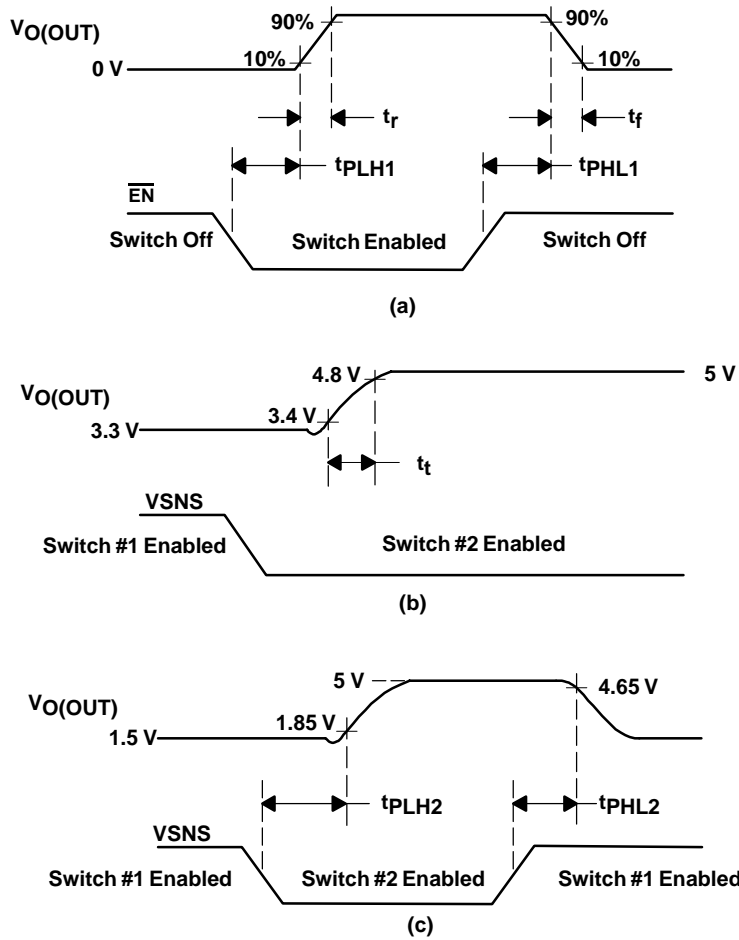
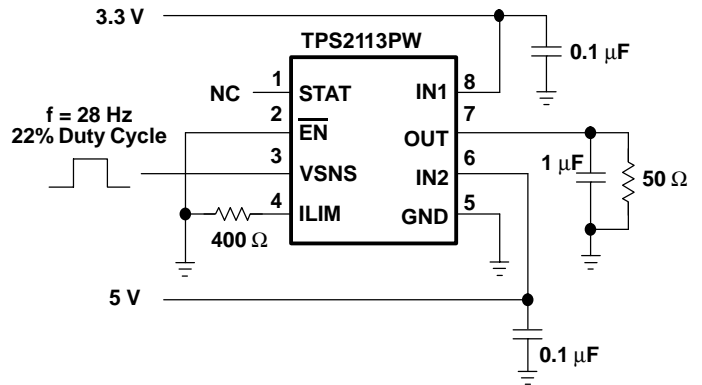
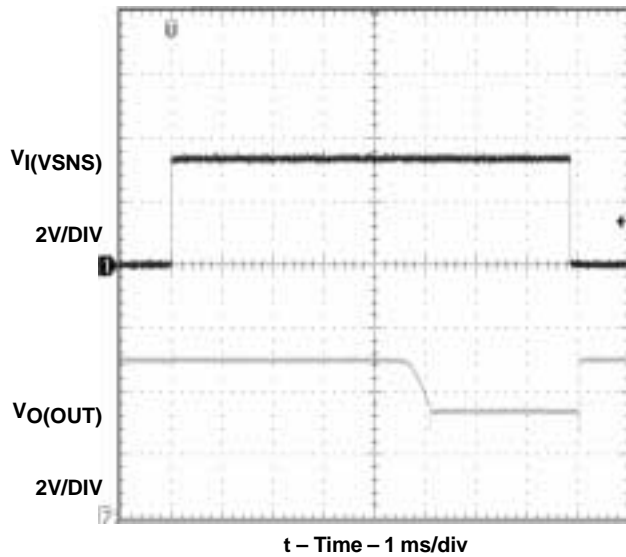


Figure 1. Propagation Delays and Transition Timing Waveforms



TYPICAL CHARACTERISTICS

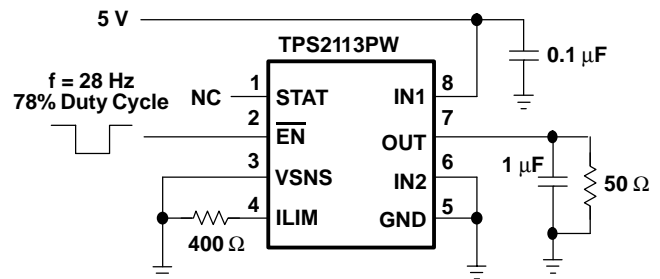
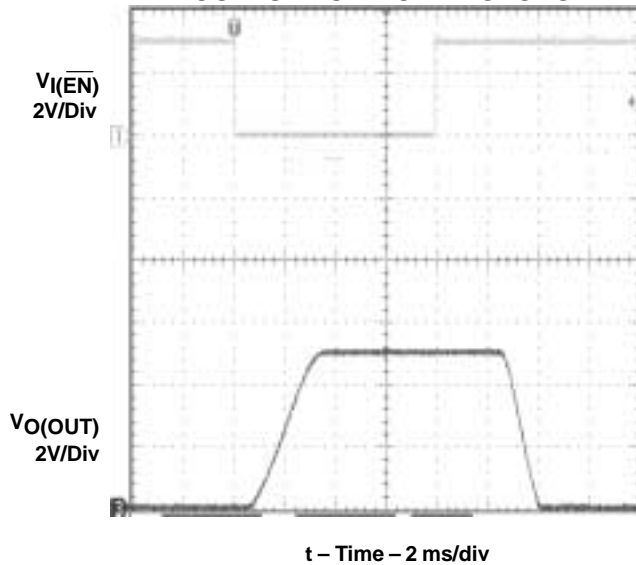
OUTPUT SWITCHOVER RESPONSE



Output Switchover Response Test Circuit

Figure 2

OUTPUT TURN-ON RESPONSE

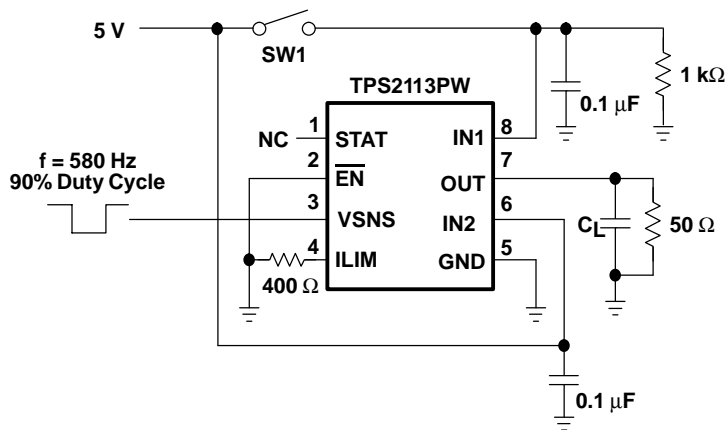
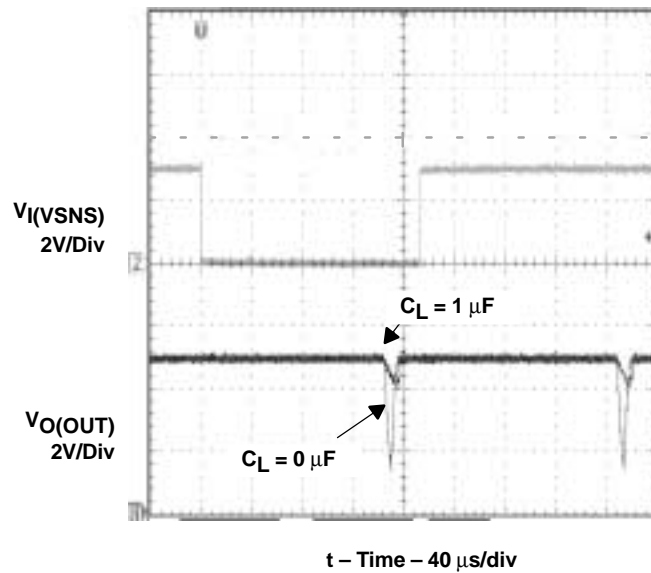


Output Turn-On Response Test Circuit

Figure 3

TYPICAL CHARACTERISTICS

OUTPUT SWITCHOVER VOLTAGE DROOP



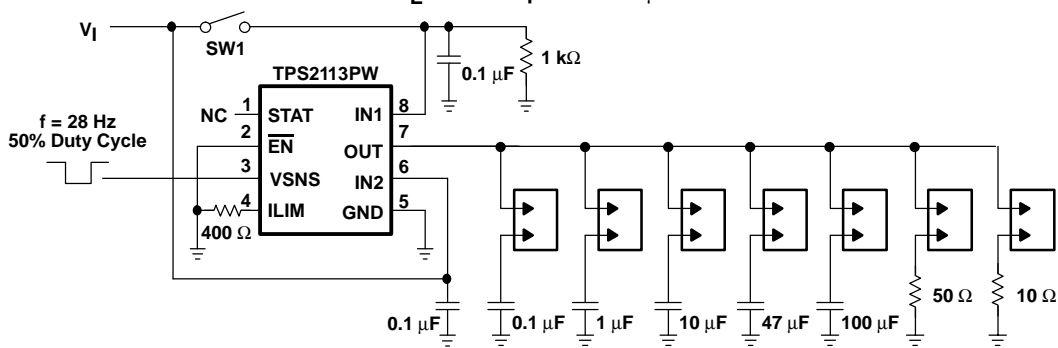
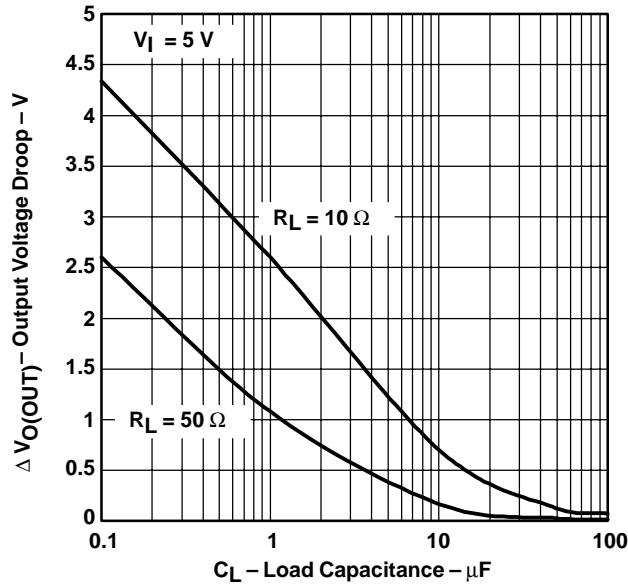
Output Switchover Voltage Droop Test Circuit

Figure 4

NOTE: To initialize the TPS2113 for this test, set input VSNS equal to 0 V, turn on the 5 V supply, and then turn on switch SW1.

TYPICAL CHARACTERISTICS

OUTPUT SWITCHOVER VOLTAGE DROOP  
vs  
LOAD CAPACITANCE



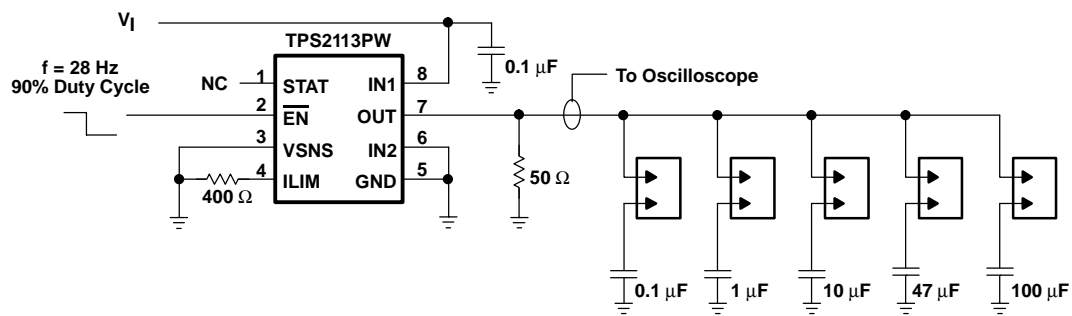
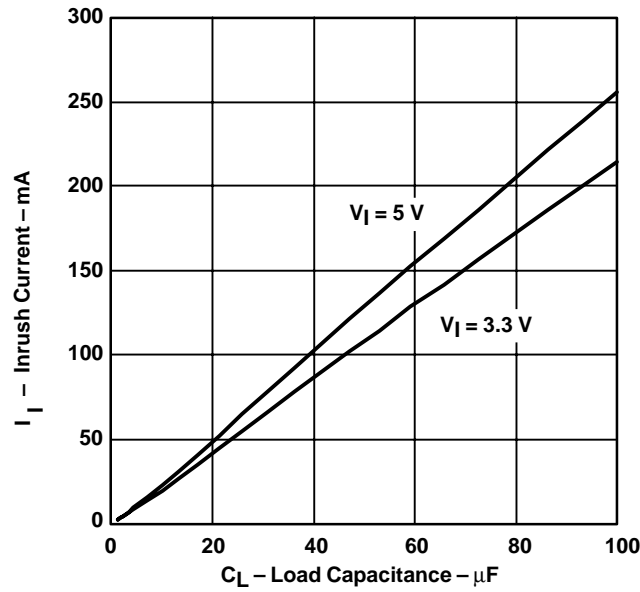
Output Swicover Voltage Droop Test Circuit

Figure 5

NOTE: To initialize the TPS2113 for this test, set input VSNS equal to 0 V, turn on the supply  $V_I$ , and then turn on switch SW1.

TYPICAL CHARACTERISTICS

INRUSH CURRENT  
VS  
LOAD CAPACITANCE



Output Capacitor Inrush Current Test Circuit

Figure 6

TYPICAL CHARACTERISTICS

SWITCH ON-RESISTANCE  
vs  
JUNCTION TEMPERATURE

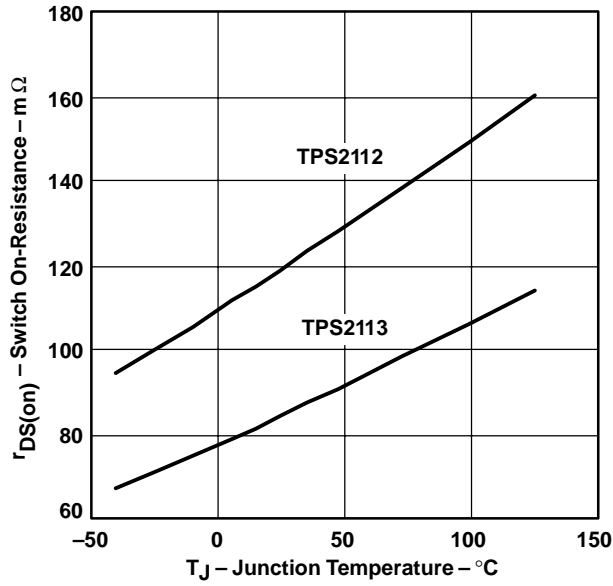


Figure 7

SWITCH ON-RESISTANCE  
vs  
SUPPLY VOLTAGE

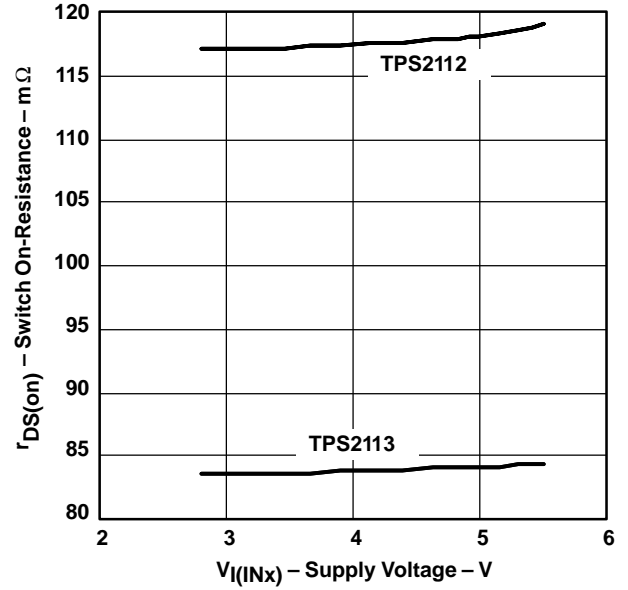


Figure 8

IN1 SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

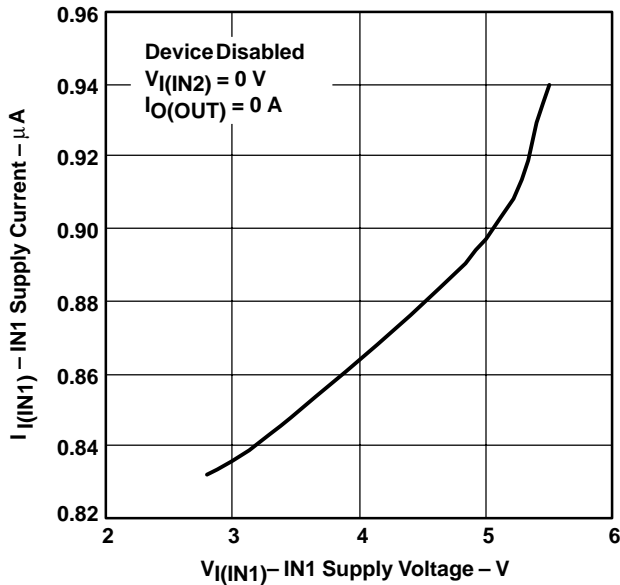


Figure 9

IN1 SUPPLY CURRENT  
vs  
SUPPLY VOLTAGE

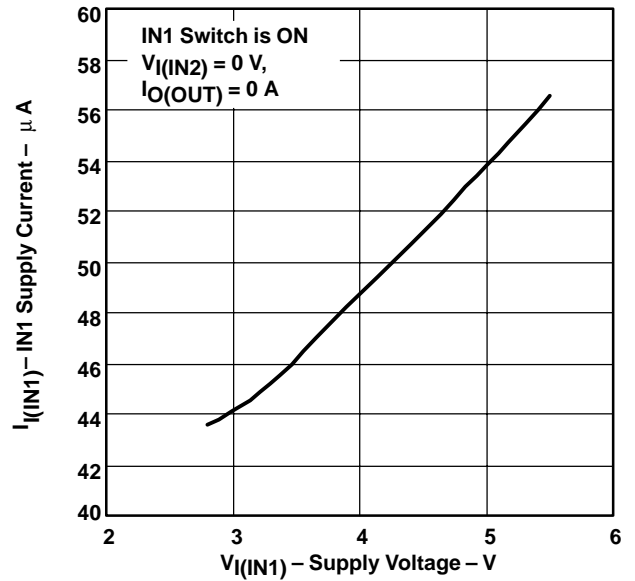


Figure 10

TYPICAL CHARACTERISTICS

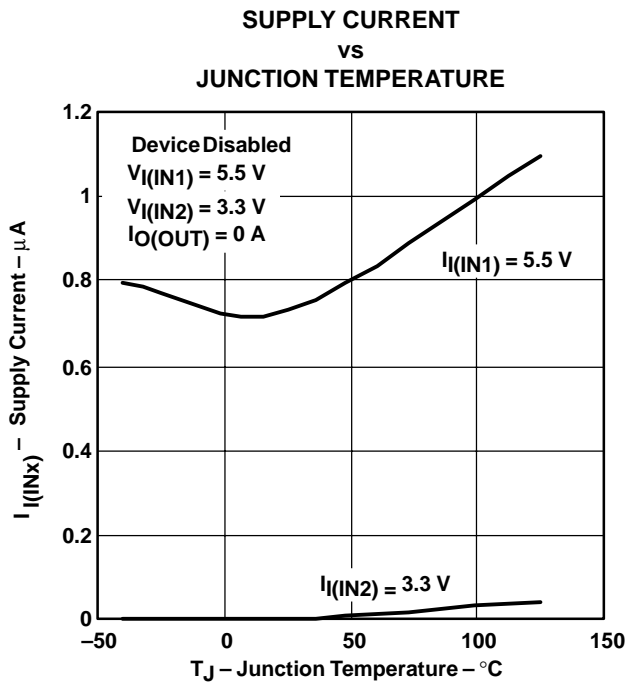


Figure 11

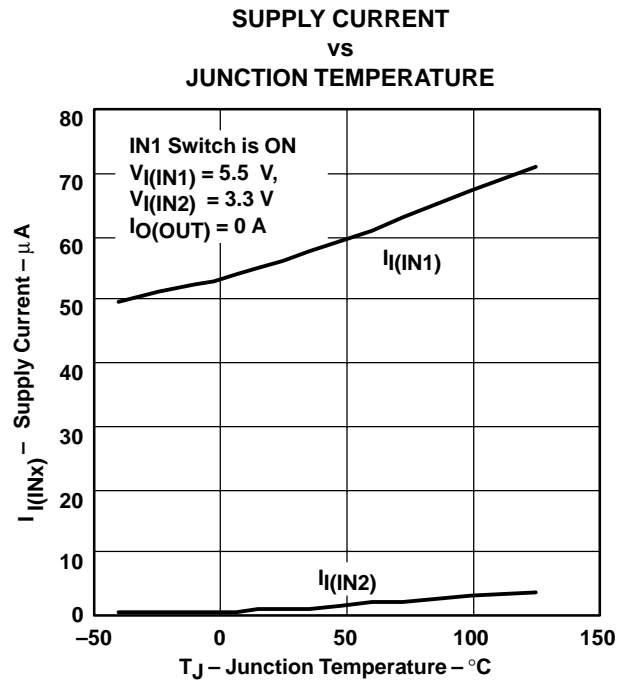


Figure 12

APPLICATION INFORMATION

Some applications have two energy sources, one of which should be used in preference to another. Figure 13 shows a circuit that will connect IN1 to OUT until the voltage at IN1 falls below a user-specified threshold. Once the voltage on IN1 falls below this threshold, the TPS2112/3 will select the higher of the two supplies. This usually means that the TPS2112/3 will swap to IN2.

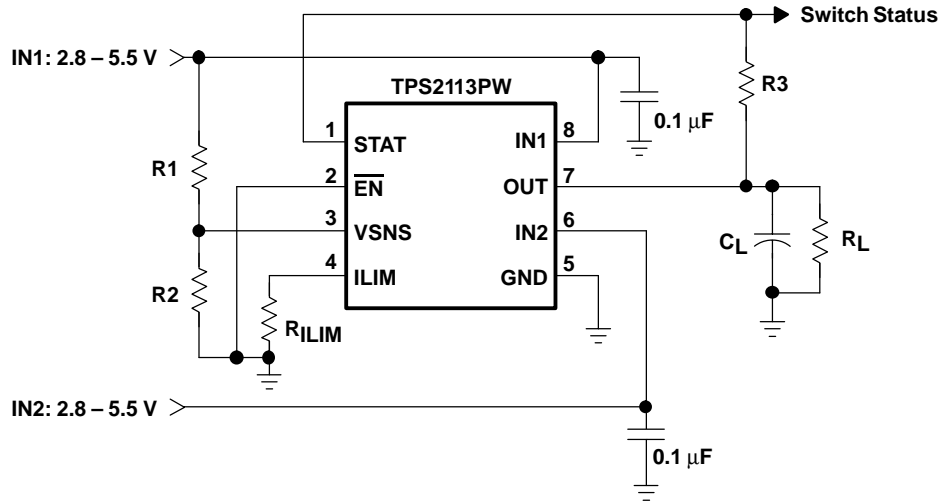


Figure 13. Auto-Selecting for a Dual Power Supply Application

In Figure 14, the multiplexer selects between two power supplies based upon the  $\overline{\text{EN}}$  logic signal. OUT connects to IN1 if  $\overline{\text{EN}}$  is logic 1, otherwise OUT connects to IN2. The logic thresholds for the  $\overline{\text{EN}}$  terminal are compatible with both TTL and CMOS logic.

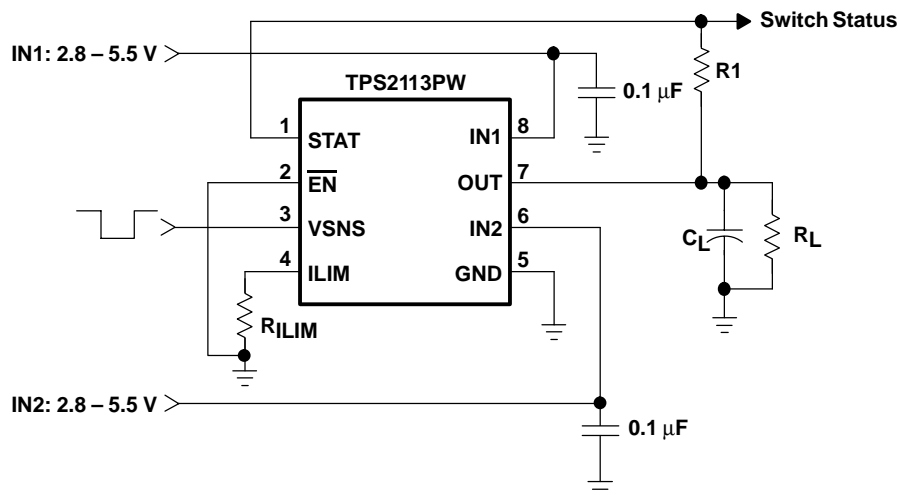


Figure 14. Manually Switching Power Sources

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## DETAILED DESCRIPTION

### AUTO-SWITCHING MODE

The TPS2112/3 only supports the auto-switching mode. In this mode, OUT connects to IN1 if  $V_{I(VSNS)}$  is greater than 0.8 V, otherwise OUT connects to the higher of IN1 and IN2.

The VSNS terminal includes hysteresis equal to 3.75–7.5% of the threshold selected for transition from the primary supply to the higher of the two supplies. This hysteresis helps avoid repeated switching from one supply to the other due to resistive drops.

### N-CHANNEL MOSFETs

Two internal high-side power MOSFETs implement a single-pole double-throw (SPDT) switch. Digital logic selects the IN1 switch, IN2 switch, or no switch (Hi-Z state). The MOSFETs have no parallel diodes so output-to-input current cannot flow when the FET is off. An integrated comparator prevents turn-on of a FET switch if the output voltage is greater than the input voltage.

### CROSS-CONDUCTION BLOCKING

The switching circuitry ensures that both power switches will never conduct at the same time. A comparator monitors the gate-to-source voltage of each power FET and allows a FET to turn on only if the gate-to-source voltage of the other FET is below the turn-on threshold voltage.

### REVERSE-CONDUCTION BLOCKING

When the TPS211x switches from a higher-voltage supply to a lower-voltage supply, current can potentially flow back from the load capacitor into the lower-voltage supply. To minimize such reverse conduction, the TPS211x will not connect a supply to the output until the output voltage has fallen to within 100 mV of the supply voltage. Once a supply has been connected to the output, it will remain connected regardless of output voltage.

### CHARGE PUMP

The higher of supplies IN1 and IN2 powers the internal charge pump. The charge pump provides power to the current limit amplifier and allows the output FET gate voltage to be higher than the IN1 and IN2 supply voltages. A gate voltage that is higher than the source voltage is necessary to turn on the N-channel FET.

### CURRENT LIMITING

A resistor  $R_{ILIM}$  from ILIM to GND sets the current limit to  $250/R_{ILIM}$  and  $500/R_{ILIM}$  for the TPS2112 and TPS2113, respectively. Setting resistor  $R_{ILIM}$  equal to zero is not recommended as that disables current limiting.

### OUTPUT VOLTAGE SLEW-RATE CONTROL

The TPS2112/3 slews the output voltage at a slow rate when OUT switches to IN1 or IN2 from the Hi-Z state (see *Truth Table*). A slow slew rate limits the inrush current into the load capacitor. High inrush currents can glitch the voltage bus and cause a system to hang up or reset. It can also cause reliability issues—like pit the connector power contacts, when hot plugging a load like a PCI card. The TPS2112/3 slews the output voltage at a much faster rate when OUT switches between IN1 and IN2. The fast rate minimizes the output voltage droop and reduces the output voltage hold-up capacitance requirement.



**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS2112PW	ACTIVE	TSSOP	PW	8	150	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2112	<a href="#">Samples</a>
TPS2112PWR	ACTIVE	TSSOP	PW	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2112	<a href="#">Samples</a>
TPS2113PW	ACTIVE	TSSOP	PW	8	150	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2113	<a href="#">Samples</a>
TPS2113PWR	ACTIVE	TSSOP	PW	8	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	2113	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**

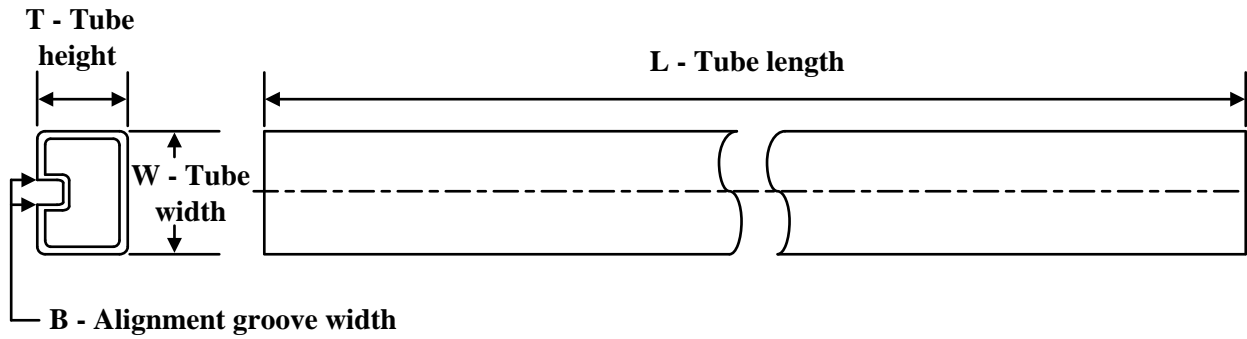

\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2112PWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1
TPS2113PWR	TSSOP	PW	8	2000	330.0	12.4	7.0	3.6	1.6	8.0	12.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2112PWR	TSSOP	PW	8	2000	356.0	356.0	35.0
TPS2113PWR	TSSOP	PW	8	2000	356.0	356.0	35.0

**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
TPS2112PW	PW	TSSOP	8	150	530	10.2	3600	3.5
TPS2113PW	PW	TSSOP	8	150	530	10.2	3600	3.5

PW0008A



PACKAGE OUTLINE  
TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



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NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-153, variation AA.

# EXAMPLE BOARD LAYOUT

PW0008A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
SCALE:10X



SOLDER MASK DETAILS  
NOT TO SCALE

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NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

PW0008A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:10X

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NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.



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