

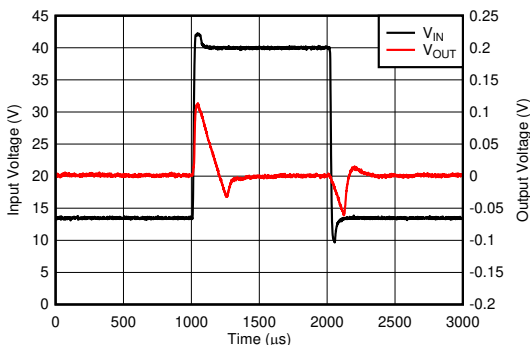
## 具有电源正常状态指示和集成电压监控功能的 TPS7B85-Q1 150mA、40V 低压降稳压器

### 1 特性

- 符合面向汽车应用的 AEC-Q100 标准：
  - 温度等级 1：-40°C 至 +125°C， $T_A$
  - 结温：-40°C 至 +150°C， $T_J$
- 输入电压范围：3V 至 40V (最大 42V)
- 输出电压范围：3.3V 和 5V (固定)
- 输出电流：高达 150mA
- 输出电压精度： $\pm 0.75\%$  (最大值)
- 低压降：
  - 150mA 时为 225mV (最大值) ( $V_{OUT} \geq 3.3V$ )
- 低静态电流：
  - 18 $\mu A$  (典型值)
  - 禁用时最大值为 4 $\mu A$
- 出色的线路瞬态响应：
  - 冷启动时出现  $\pm 2\%$   $V_{OUT}$  偏差
  - $\pm 2\%$   $V_{OUT}$  偏差 ( $V_{IN}$  压摆率 1V/ $\mu s$ )
- 集成电压检测
- 具有可调阈值和可编程延迟周期的电源正常状态指示
- 与 2.2 $\mu F$  或更高的电容器搭配使用时可保持稳定
- 提供功能安全
  - 可帮助进行功能安全系统设计的文档
- 封装：10 引脚 VSON (带散热焊盘)
  - 低热阻 ( $R_{\theta JA}$ )：50.3°C/W

### 2 应用

- 可重新配置仪表组
- 车身控制模块 (BCM)
- 常开型电池连接应用：
  - 汽车网关
  - 远程免钥匙进入 (RKE)



线路瞬态响应 ( $V_{IN}$  压摆率 3V/ $\mu s$ )

### 3 说明

TPS7B85-Q1 是一款低压降线性稳压器，专用于连接汽车应用中的电池。该器件的输入电压范围高达 40V，因此可承受汽车系统可能发生的瞬变（如负载突降）。此器件的静态电流仅为 18 $\mu A$ ，是为备用系统中微控制器 (MCU) 和控制器局域网 (CAN) 收发器等常开型器件供电的出色解决方案。

该器件具有先进的瞬态响应，因此输出端可对负载或线路变化（例如，在冷启动条件下）作出迅速响应。此外，该器件架构新颖，可在从电压跌落恢复过程中最大限度降低输出过冲幅度。正常运行时，该器件可在整个线路、负载和温度范围内维持  $\pm 0.75\%$  的直流精度。

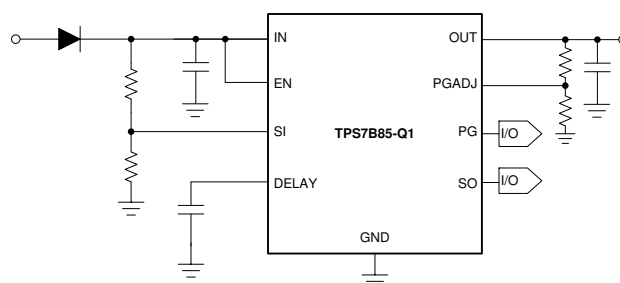
TPS7B85-Q1 具有电源正常状态指示和集成的电压监控功能。电源正常状态指示延迟和电压阈值可通过外部组件进行调整。集成式电压检测器可用于监控输入电压，并在电池电压开始降低时提醒下游组件（如 MCU）。

该器件采用小型 VSON 封装，可实现紧凑的印刷电路板 (PCB) 设计。即使整个器件散热较多，低热阻也有助于持久运行。

#### 器件信息(1)

器件型号	封装	封装尺寸 (标称值)
TPS7B85-Q1	VSON (10)	3.00mm × 3.00mm

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。



典型应用原理图



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## 4 Revision History

注：以前版本的页码可能与当前版本的页码不同

### Changes from Revision \* (February 2020) to Revision A (November 2020)

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## 5 Pin Configuration and Functions

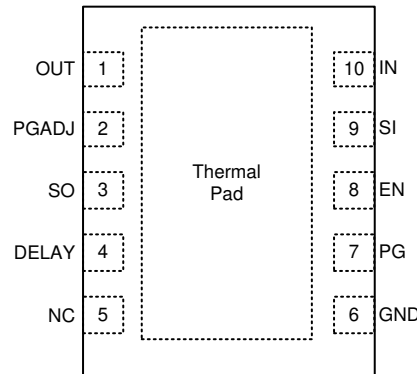


图 5-1. DRC Package, 10-Pin VSON, Top View

表 5-1. Pin Functions

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NAME	DRC		
DELAY	4	O	Power-good delay adjustment pin. Connect a capacitor from this pin to GND to set the PG reset delay. Leave this pin floating for a default ( $t_{(DLY\_FIX)}$ ) delay. See the <a href="#">Power-Good</a> section for more information. If this functionality is not desired, leave this pin floating because connecting this pin to GND causes a permanent increase in the GND current.
EN	8	I	Enable pin. The device is disabled when the enable pin becomes lower than the enable logic input low level ( $V_{IL}$ ). To ensure the device is enabled, the EN pin must be driven above the logic high level ( $V_{IH}$ ). This pin should not be left floating as this pin is high impedance if it is left floating the part may enable or disable.
GND	6	G	Ground pin. Connect this pin to the thermal pad with a low-impedance connection.
NC	5	—	No internal connection. Connect this pin to GND for the best thermal resistance.
PGADJ	2	I	Power-good threshold-adjustment pin. Connect a resistor divider between the PGADJ and OUT pins to set the power-good threshold. Connect this pin to ground to set the threshold to $V_{PG(TH\_FALLING)}$ . See <a href="#">Power-Good</a> for more information.
PG	7	O	Power-good pin. This pin has an internal pullup resistor. Do not connect this pin to $V_{OUT}$ or any other biased voltage rail. $V_{PG}$ is logic level high when $V_{OUT}$ is above the power-good threshold. See <a href="#">Power-Good</a> for more information.
SI	9	I	Sense input pin. Connect via an external voltage divider to the supply voltage to be monitored.
SO	3	O	Sense output pin. This pin has an internal pullup resistor. Do not connect this pin to $V_{OUT}$ or any other biased voltage rail. $V_{SO}$ is logic level low when $V_{SI}$ falls below the sense-low threshold.
IN	10	P	Input power-supply voltage pin. For best transient response and to minimize input impedance, use the recommended value or larger ceramic capacitor from IN to ground as listed in the <a href="#">Recommended Operating Conditions</a> table and the <a href="#">Input Capacitor</a> section. Place the input capacitor as close to the input of the device as possible.
OUT	1	O	Regulated output voltage pin. A capacitor is required from OUT to ground for stability. For best transient response, use the nominal recommended value or larger ceramic capacitor from OUT to ground; see the <a href="#">Recommended Operating Conditions</a> table and the <a href="#">Output Capacitor</a> section. Place the output capacitor as close to the output of the device as possible. If using a high ESR capacitor, decouple the output with a 100-nF ceramic capacitor.
Thermal pad		—	Thermal pad. Connect the pad to GND for the best possible thermal performance. See the <a href="#">Layout</a> section for more information.

(1) I = input; O = output; P = power; G = ground.

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
IN	Unregulated input	- 0.3	42	V
EN	Enable input	- 0.3	42	V
OUT	Regulated output	- 0.3	$V_{IN} + 0.3$ <sup>(2)</sup>	V
FB	Feedback	- 0.3	20	V
SI	Sense input	- 0.3	42	V
Delay	Reset delay input	- 0.3	6	V
SO, PG, PGADJ	Sense output, power-good, power-good adjustable threshold	- 0.3	20	V
T <sub>A</sub>	Operating ambient temperature	- 40	125	°C
T <sub>J</sub>	Operating junction temperature	- 40	150	°C
T <sub>stg</sub>	Storage temperature	- 65	150	°C

- (1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) The absolute maximum rating is  $V_{IN} + 0.3$  V or 20 V, whichever is smaller.

### 6.2 ESD Ratings

			VALUE	UNIT	
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±2000	V	
		Charged-device model (CDM), per AEC Q100-011	All pins		±500
			Corner pins		±750

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	TYP	MAX	UNIT
V <sub>IN</sub>	Input voltage	3		40	V
V <sub>OUT</sub>	Output voltage	1.2		18	V
I <sub>OUT</sub>	Output current	0		150	mA
V <sub>EN</sub> , V <sub>SI</sub>	High voltage (I/O)	0		40	V
V <sub>Delay</sub>	Delay pin voltage	0		5.5	V
V <sub>PG</sub> , V <sub>SO</sub> , V <sub>PGADJ</sub>	Low voltage (I/O), power-good adjustable threshold	0		18	V
C <sub>OUT</sub>	Output capacitor <sup>(2)</sup>	2.2		220	μF
ESR	Output capacitor ESR requirements <sup>(3)</sup>	0.001		2	Ω
C <sub>IN</sub>	Input capacitor <sup>(1)</sup>	0.1	1		μF
C <sub>Delay</sub>	Power-good delay capacitor	0		1	μF
T <sub>J</sub>	Operating junction temperature	- 40		150	°C

- (1) For robust EMI performance the minimum input capacitance is 500 nF.
- (2) Effective output capacitance of 1 μF minimum required for stability.
- (3) If using a large ESR capacitor it is recommended to decouple this with a 100-nF ceramic capacitor to improve transient performance.

## 6.4 Thermal Information

THERMAL METRIC <sup>(1) (2)</sup>		TPS7B85-Q1	UNIT
		DRC (VSON)	
		10 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance <sup>(3)</sup>	50.3	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	54.5	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	23.9	°C/W
$\psi_{JT}$	Junction-to-top characterization parameter	1.2	°C/W
$\psi_{JB}$	Junction-to-board characterization parameter	23.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	7.5	°C/W

- (1) The thermal data is based on the JEDEC standard high K profile, JESD 51-7. Two-signal, two-plane, four-layer board with 2-oz. copper. The copper pad is soldered to the thermal land pattern. Also, correct attachment procedure must be incorporated.
- (2) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.
- (3) The 1s0p  $R_{\theta JA}$  is 202.5°C/W for the DRC package.

## 6.5 Electrical Characteristics

specified at  $T_J = -40^\circ\text{C}$  to  $+150^\circ\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 0\text{ mA}$ ,  $C_{OUT} = 2.2\text{ }\mu\text{F}$ ,  $1\text{ m}\Omega < C_{OUT}\text{ ESR} < 2\text{ }\Omega$ ,  $C_{IN} = 1\text{ }\mu\text{F}$ , and  $V_{EN} = 2\text{ V}$  (unless otherwise noted); typical values are at  $T_J = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
$V_{OUT}$	Regulated output accuracy	$V_{IN} = V_{OUT} + 500\text{ mV}$ to $40\text{ V}$ , $I_{OUT} = 100\text{ }\mu\text{A}$ to $150\text{ mA}$ <sup>(1)</sup>	$T_J = 25^\circ\text{C}$	-0.5		0.5	%
			$T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$	-		0.75	
$\Delta V_{OUT(\Delta V_{IN})}$	Line regulation	Change in percent of output voltage	$V_{IN} = V_{OUT} + 500\text{ mV}$ to $40\text{ V}$ , $I_{OUT} = 100\text{ }\mu\text{A}$			0.2	%
$\Delta V_{OUT(\Delta I_{OUT})}$	Load regulation	Change in percent of output voltage	$V_{IN} = V_{OUT} + 500\text{ mV}$ , $I_{OUT} = 100\text{ }\mu\text{A}$ to $150\text{ mA}$			0.2	
$\Delta V_{OUT}$	Load transient response settling time <sup>(2) (3)</sup>	$C_{OUT} = 10\text{ }\mu\text{F}$	$C_{OUT} = 10\text{ }\mu\text{F}$			100	$\mu\text{s}$
	Load transient response overshoot, undershoot <sup>(3)</sup>		$I_{OUT} = 45\text{ mA}$ to $105\text{ mA}$	-2%		10%	% $V_{OUT}$
			$I_{OUT} = 0\text{ mA}$ to $150\text{ mA}$	-		10%	
$I_Q$	Quiescent current	$V_{IN} = V_{OUT} + 500\text{ mV}$ to $40\text{ V}$ , $I_{OUT} = 0\text{ mA}$	$T_J = 25^\circ\text{C}$		18	21	$\mu\text{A}$
			$T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$			26	
			$I_{OUT} = 500\text{ }\mu\text{A}$	$T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$			
$I_{SHUTDOWN}$	Shutdown supply current ( $I_{GND}$ )	$V_{EN} = 0\text{ V}$	$T_J = 25^\circ\text{C}$			2.5	$\mu\text{A}$
			$T_J = -40^\circ\text{C}$ to $+150^\circ\text{C}$			4	
$V_{DO}$	Dropout voltage	$I_{OUT} \leq 1\text{ mA}$ , $V_{OUT} \geq 3.3\text{ V}$ , $V_{IN} = V_{OUT(NOM)} \times 0.95$				43	mV
			$I_{OUT} = 105\text{ mA}$ , $V_{OUT} \geq 3.3\text{ V}$ , $V_{IN} = V_{OUT(NOM)}$		125	175	
			$I_{OUT} = 150\text{ mA}$ , $V_{OUT} \geq 3.3\text{ V}$ , $V_{IN} = V_{OUT(NOM)}$		155	225	
$V_{UVLO(RISING)}$	Rising input supply UVLO	$V_{IN}$ rising		2.6	2.7	2.82	V
$V_{UVLO(FALLING)}$	Falling input supply UVLO	$V_{IN}$ falling		2.38	2.5	2.6	V
$V_{UVLO(HYST)}$	$V_{UVLO}$ hysteresis				230		mV
$V_{IL}$	Enable logic input low level					0.7	V

## 6.5 Electrical Characteristics (continued)

specified at  $T_J = -40^\circ\text{C}$  to  $+150^\circ\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 0\text{ mA}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\text{ m}\Omega < C_{OUT}\text{ ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\text{ V}$  (unless otherwise noted); typical values are at  $T_J = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{IH}$	Enable logic input high level		2			V
$I_{EN}$	EN pin current	$V_{EN} = V_{IN} = 13.5\text{ V}$			50	nA
$I_{CL}$	Output current limit	$V_{IN} = V_{OUT(nom)} + 1\text{ V}$ , $V_{OUT}$ short to $90\% \times V_{OUT(NOM)}$	180	220	260	mA
PSRR	Power-supply ripple rejection	$V_{IN} - V_{OUT} = 500\text{ mV}$ , frequency = 1 kHz, $I_{OUT} = 150\text{ mA}$		55		dB
$V_n$	Output noise voltage	$V_{OUT} = 3.3\text{ V}$ , BW = 10 Hz to 100 kHz		280		$\mu\text{V}_{RMS}$
$V_{SI(HIGH)}$	Sense input threshold high	$V_{SI}$ rising	1.17	1.21	1.25	V
$V_{SI(LOW)}$	Sense input threshold low	$V_{SI}$ falling	1.07	1.12	1.15	V
$V_{SI(HYST)}$	Sense input switching hysteresis			90		mV
$I_{SI}$	Sense input current	$V_{SI} = 40\text{ V}$		0.015	1.5	$\mu\text{A}$
$R_{SO}$	Sense output internal pullup resistor		10	30	50	$\text{k}\Omega$
$V_{SO(OL)}$	Sense output low voltage	$V_{SI} \leq 1.07\text{ V}$ , $V_{IN} \geq 3\text{ V}$			0.4	V
$R_{PG}$	Power-good internal pullup resistor		10	30	50	$\text{k}\Omega$
$V_{PG(OL)}$	PG pin low level output voltage	$V_{OUT} \leq 0.83 \times V_{OUT}$			0.4	V
$V_{PG(TH,RISING)}$	Default power-good threshold	$V_{OUT}$ rising, PGADJ pin shorted to ground	85		95	% $V_{OUT}$
$V_{PG(TH,FALLING)}$		$V_{OUT}$ falling, PGADJ pin shorted to ground	83		93	
$V_{PG(HYST)}$	Power-good hysteresis			2.5		% $V_{OUT}$
$V_{PGADJ(TH,FALLING)}$	Switching voltage for the power-good adjust pin	$V_{OUT}$ falling, PGADJ falling	0.97	1	1.030	V
$V_{PGADJ(HYST)}$	PGADJ hysteresis		5	35	50	mV
$V_{DLY(TH)}$	Threshold to release power-good high	Voltage at delay pin rising	1.17	1.21	1.25	V
$I_{DLY(CHARGE)}$	Delay capacitor charging current	$V_{DLY} = 1\text{ V}$	1	1.5	2	$\mu\text{A}$
$T_{SD(SHUTDOWN)}$	Junction shutdown temperature			175		$^\circ\text{C}$
$T_{SD(HYST)}$	Hysteresis of thermal shutdown			20		$^\circ\text{C}$

- Power dissipation is limited to 2W for IC production testing purposes. The power dissipation can be higher during normal operation. Please see the thermal dissipation section for more information on how much power the device can dissipate while maintaining a junction temperature below  $150^\circ\text{C}$ .
- The settling time is measured from when  $I_{OUT}$  is stepped from 45mA to 105 mA to when the output voltage recovers to  $V_{OUT} = V_{OUT(nom)} - 5\text{ mV}$ .
- This specification is specified by design.

## 6.6 Switching Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>TIMING FOR SENSE INPUT AND OUTPUT (SI, SO)</b>						
$t_{(PD\_SO\_HL)}$	Sense high reaction time			25		$\mu\text{s}$
$t_{(PD\_SO\_LH)}$	Sense low reaction time			30		$\mu\text{s}$
<b>TIMING POWER-GOOD</b>						
$t_{(DLY\_FIX)}$	Power-good propagation delay	No capacitor connected at DELAY pin		100		$\mu\text{s}$
$t_{(Deglitch)}$	Power-good deglitch time	No capacitor connected at DELAY pin		90		$\mu\text{s}$
$t_{(DLY)}$	Power-good propagation delay	Delay capacitor value: $C_{(DELAY)} = 100\text{ nF}$		80		ms

### 6.7 Typical Characteristics

specified at  $T_J = -40^\circ\text{C}$  to  $+150^\circ\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)

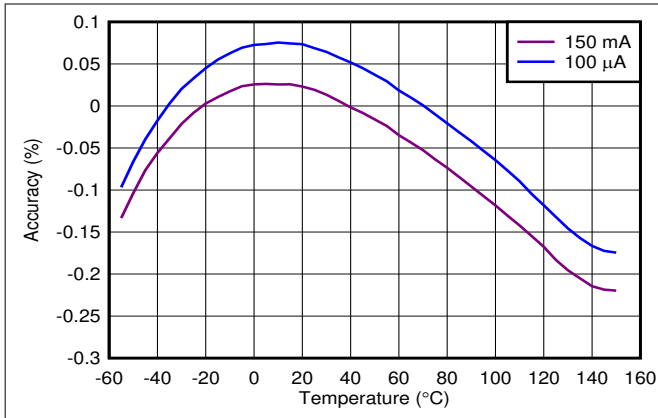
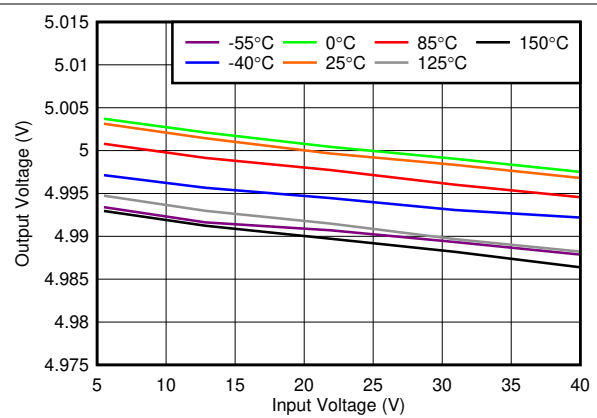
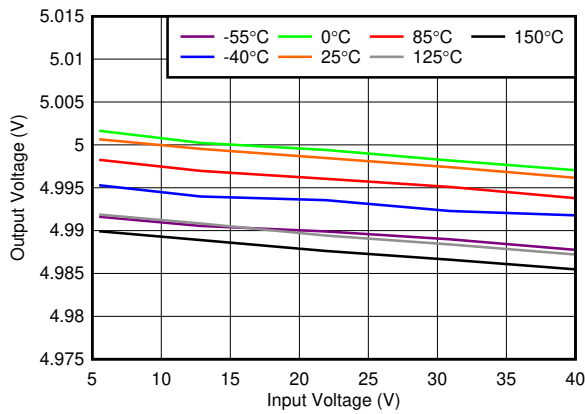


图 6-1. Accuracy vs Temperature



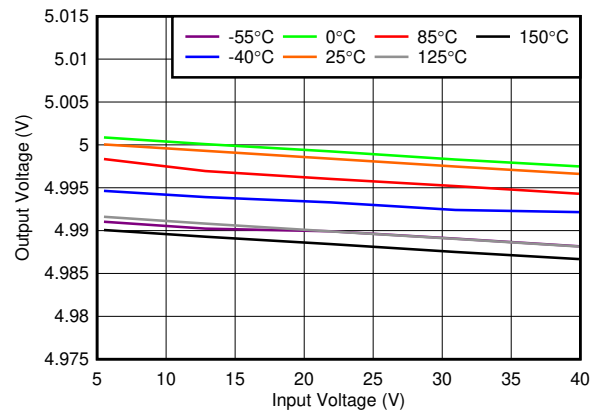
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 150\ \text{mA}$

图 6-2. Line Regulation vs  $V_{IN}$



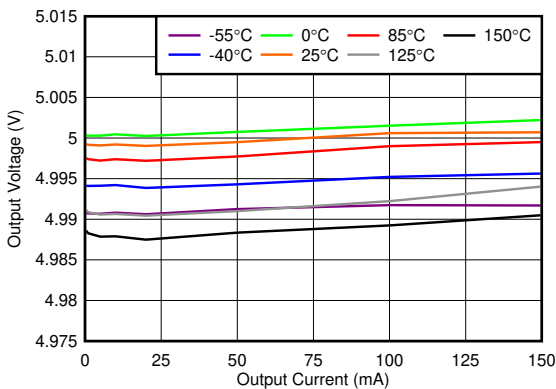
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 5\ \text{mA}$

图 6-3. Line Regulation vs  $V_{IN}$



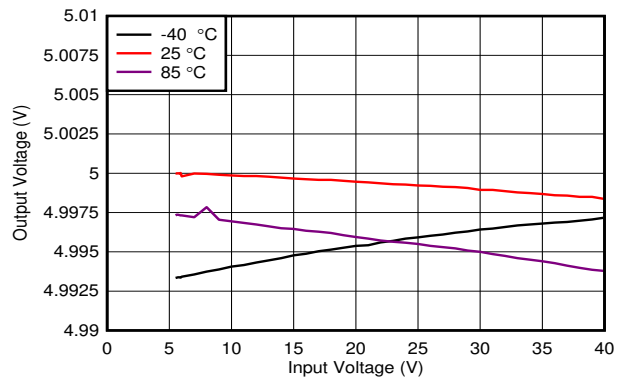
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 1\ \text{mA}$

图 6-4. Line Regulation vs  $V_{IN}$



$V_{OUT} = 5\ \text{V}$

图 6-5. Load Regulation vs  $I_{OUT}$



$C_{OUT} = 10\ \mu\text{F}$

图 6-6. Line Regulation at 50 mA

### 6.7 Typical Characteristics (continued)

specified at  $T_J = -40^\circ\text{C}$  to  $+150^\circ\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)

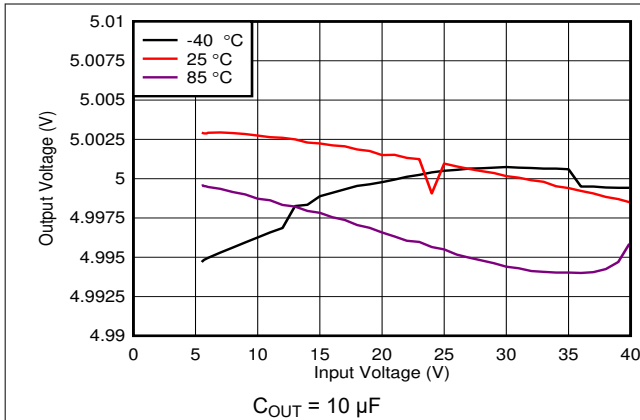


图 6-7. Line Regulation at 100 mA

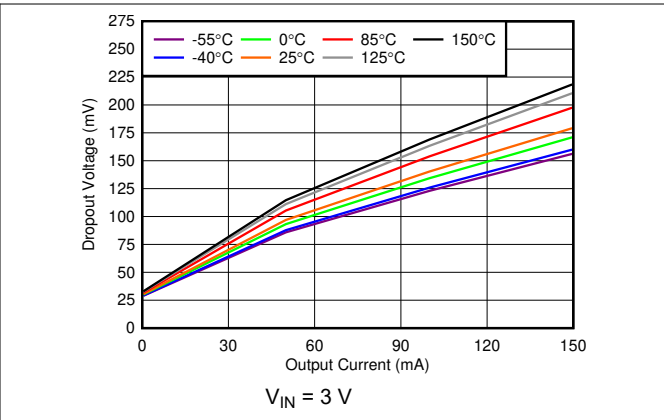


图 6-8. Dropout Voltage ( $V_{DO}$ ) vs  $I_{OUT}$

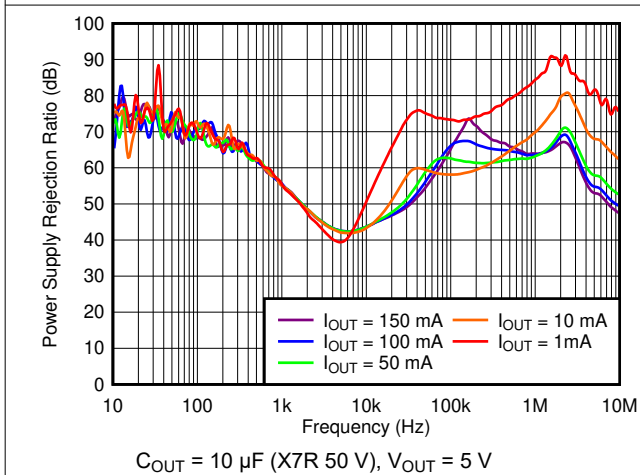


图 6-9. PSRR vs Frequency and  $I_{OUT}$

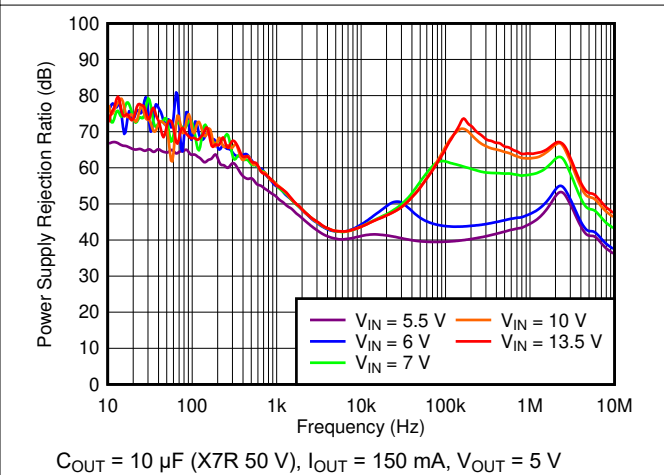


图 6-10. PSRR vs Frequency and  $V_{IN}$

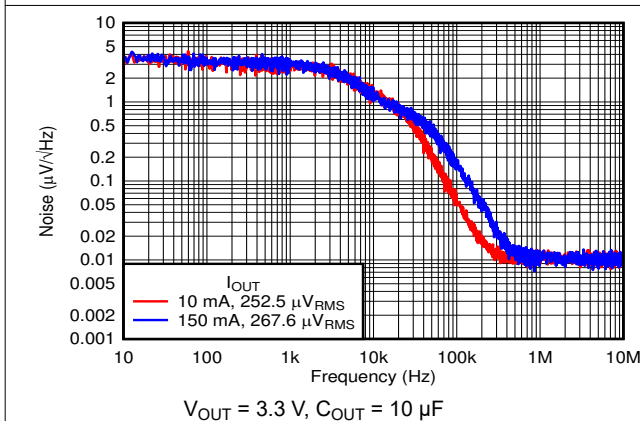


图 6-11. Noise vs Frequency at 3.3 V

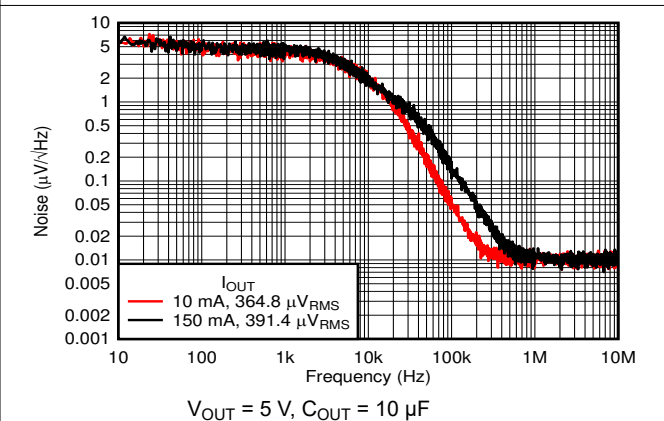
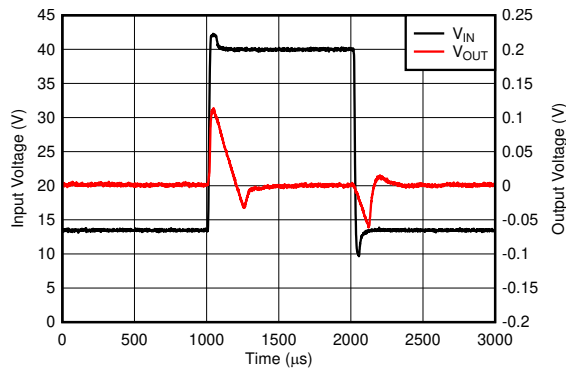


图 6-12. Noise vs Frequency at 5.0 V



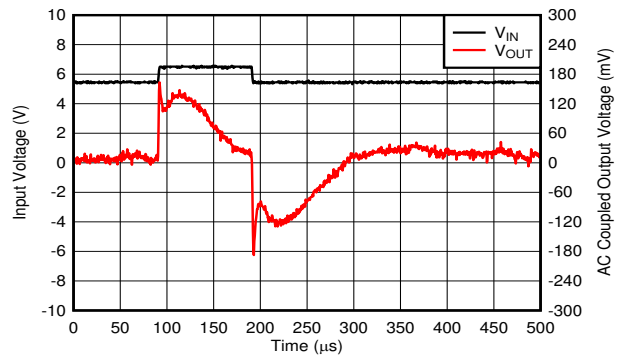
## 6.7 Typical Characteristics (continued)

specified at  $T_J = -40^\circ\text{C}$  to  $+150^\circ\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)



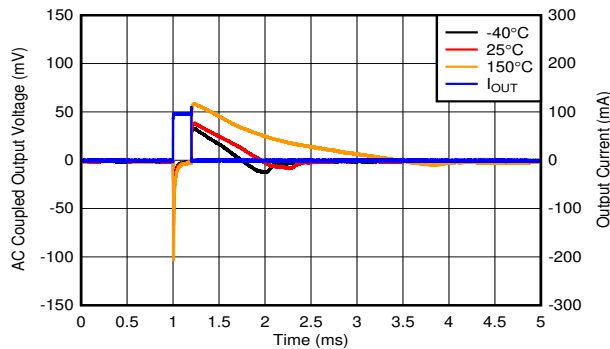
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 1\ \text{mA}$ ,  $V_{IN} = 13.5\ \text{V}$  to  $40\ \text{V}$ ,  
slew rate =  $2.7\ \text{V}/\mu\text{s}$ ,  $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-13. Line Transients at 13.5 V to 40 V



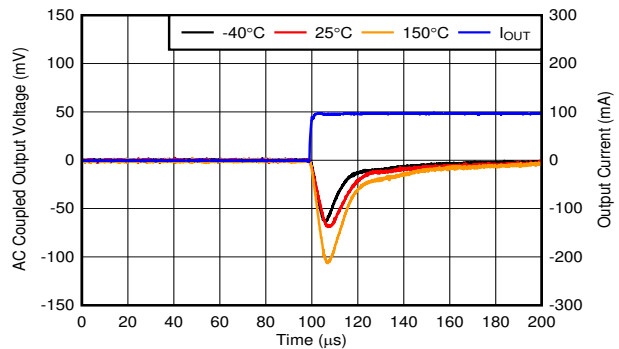
$V_{OUT} = 5\ \text{V}$ ,  $V_{IN} = 5.5\ \text{V}$  to  $6.5\ \text{V}$ ,  $t_{\text{rise}} = 1\ \mu\text{s}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-14. Line Transients at 5.5 V to 6.5 V



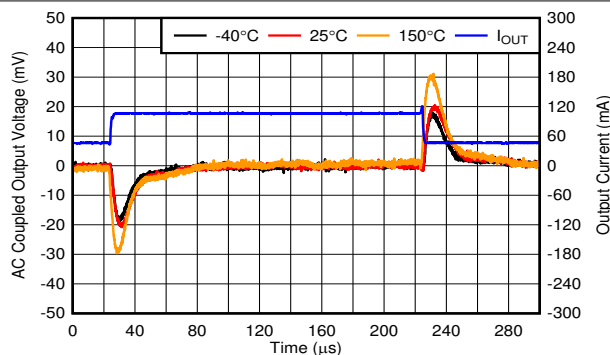
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 0\ \text{mA}$  to  $100\ \text{mA}$ , slew rate =  $1\ \text{A}/\mu\text{s}$ ,  
 $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-15. Load Transient, No Load to 100 mA



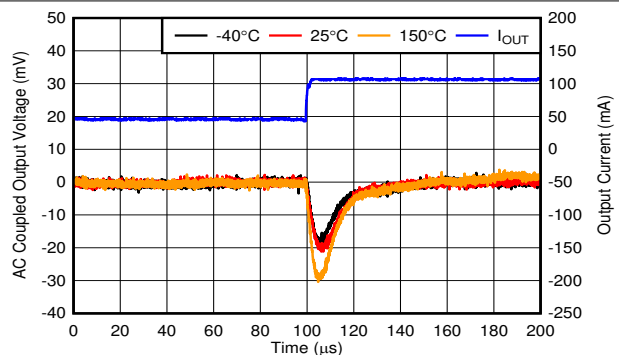
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 0\ \text{mA}$  to  $100\ \text{mA}$ , slew rate =  $1\ \text{A}/\mu\text{s}$ ,  
 $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-16. Load Transient, No Load to 100-mA Rising Edge



$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 45\ \text{mA}$  to  $105\ \text{mA}$ , slew rate =  $0.1\ \text{A}/\mu\text{s}$ ,  
 $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-17. Load Transient, 45 mA to 105 mA

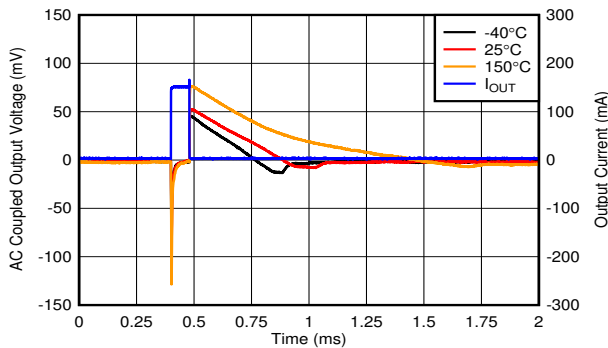


$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 45\ \text{mA}$  to  $105\ \text{mA}$ , slew rate =  $0.1\ \text{A}/\mu\text{s}$ ,  
 $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-18. Load Transient, 45-mA to 105-mA Rising Edge

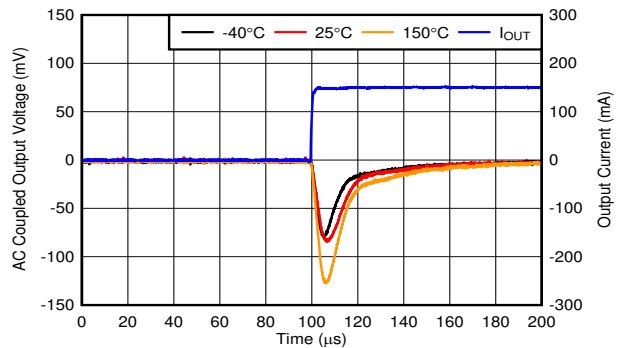
### 6.7 Typical Characteristics (continued)

specified at  $T_J = -40^\circ\text{C}$  to  $+150^\circ\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)



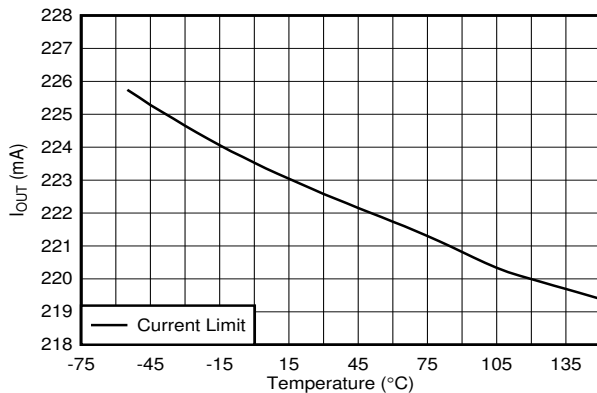
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 0\ \text{mA}$  to  $150\ \text{mA}$ , slew rate =  $1\ \text{A}/\mu\text{s}$ ,  $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-19. Load Transient, No Load to 150 mA



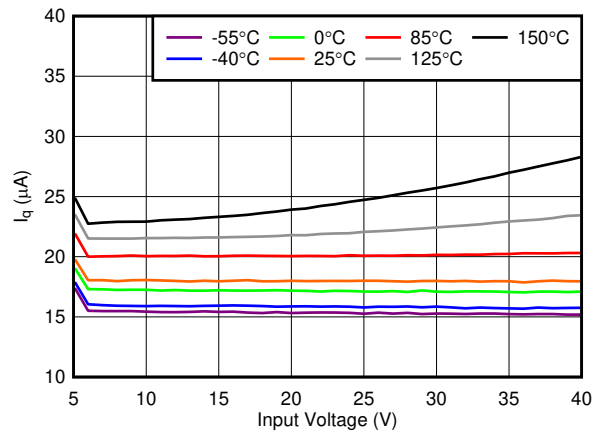
$V_{OUT} = 5\ \text{V}$ ,  $I_{OUT} = 0\ \text{mA}$  to  $150\ \text{mA}$ , slew rate =  $1\ \text{A}/\mu\text{s}$ ,  $V_{EN} = 3.3\ \text{V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-20. Load Transient, No Load to 150-mA Rising Edge



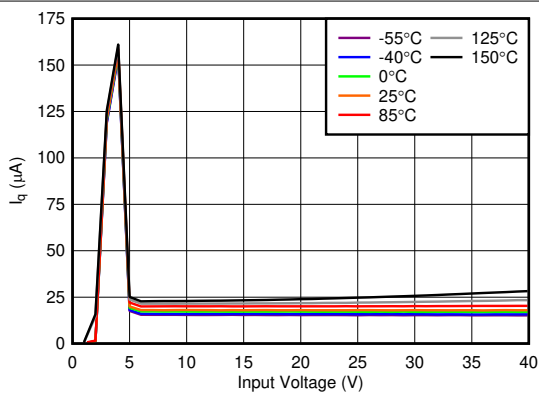
$V_{IN} = V_{OUT} + 1\ \text{V}$ ,  $V_{OUT} = 90\% \times V_{OUT(NOM)}$

图 6-21. Output Current Limit vs Temperature



$V_{OUT} = 5\ \text{V}$

图 6-22. Quiescent Current ( $I_Q$ ) vs  $V_{IN}$



$V_{OUT} = 5\ \text{V}$

图 6-23. Quiescent Current ( $I_Q$ ) vs  $V_{IN}$

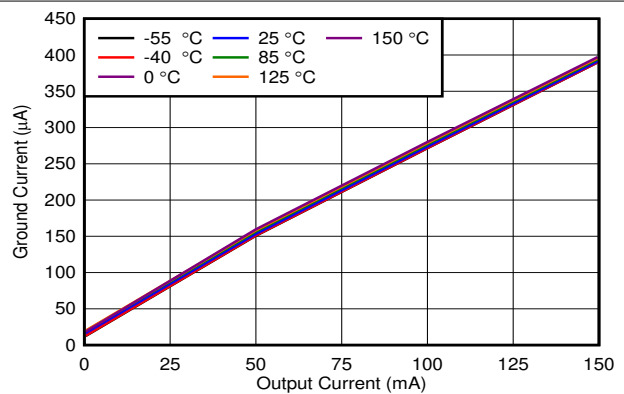


图 6-24. Ground Current ( $I_{GND}$ ) vs  $I_{OUT}$

### 6.7 Typical Characteristics (continued)

specified at  $T_J = -40^{\circ}\text{C}$  to  $+150^{\circ}\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)

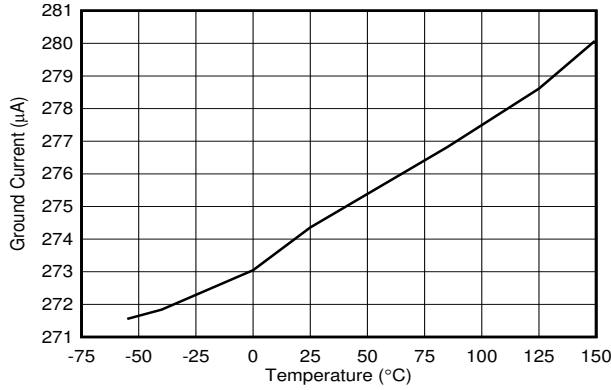


图 6-25. Ground Current

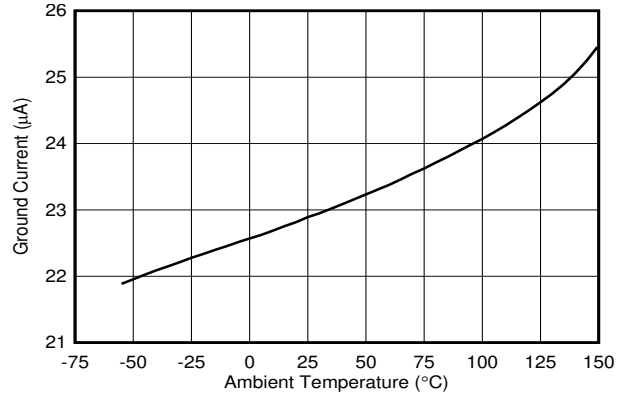


图 6-26. Ground Current

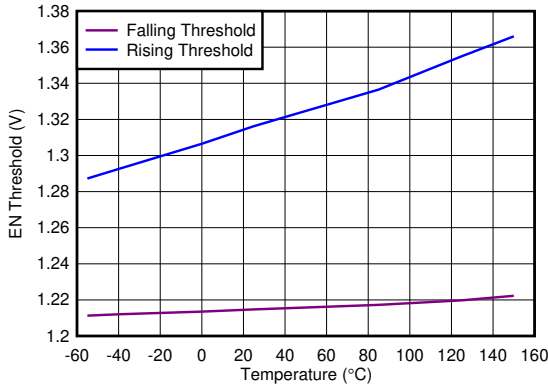


图 6-27. EN Threshold vs Temperature

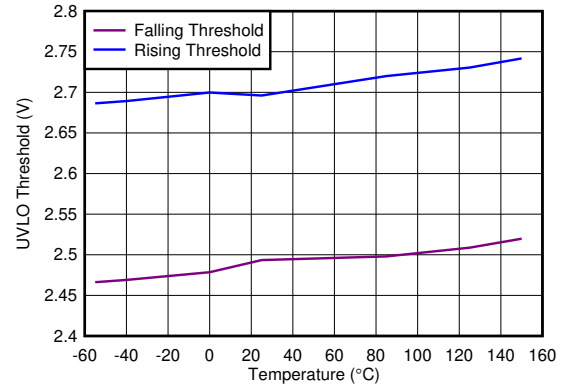


图 6-28. Undervoltage Lockout (UVLO) Threshold vs Temperature

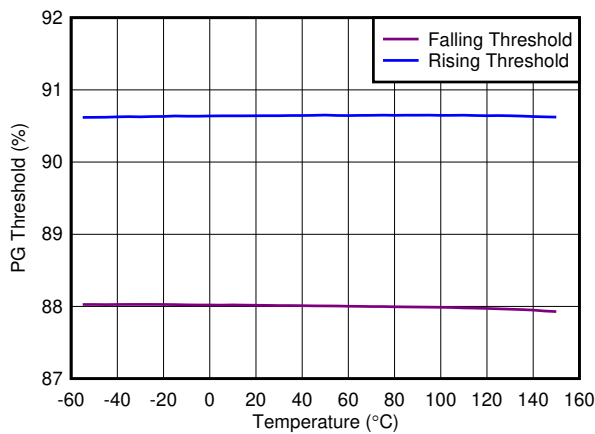


图 6-29. PG Threshold vs Temperature

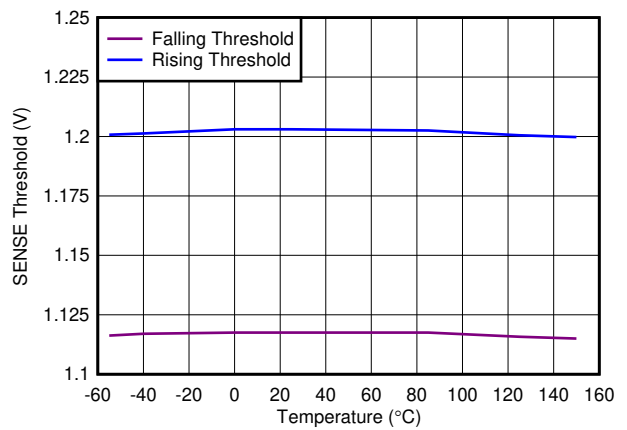


图 6-30. Sense Input Threshold vs Temperature

### 6.7 Typical Characteristics (continued)

specified at  $T_J = -40^{\circ}\text{C}$  to  $+150^{\circ}\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)

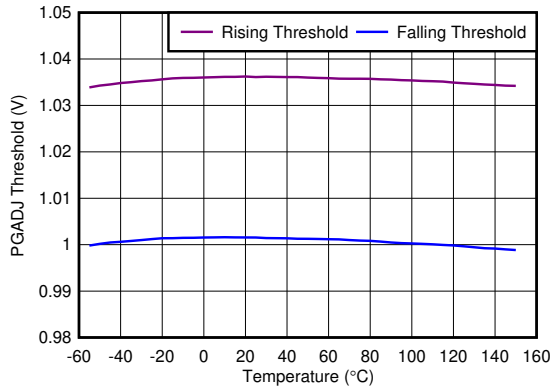
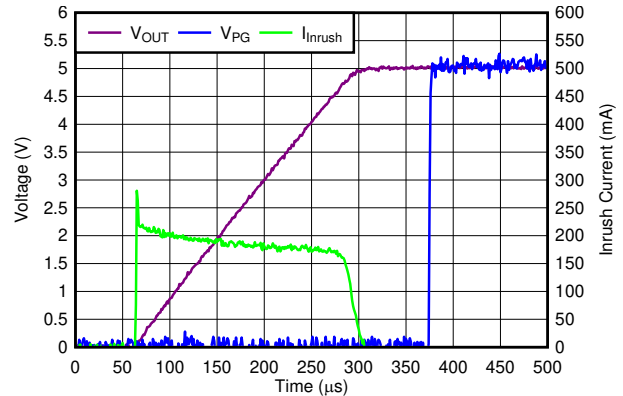
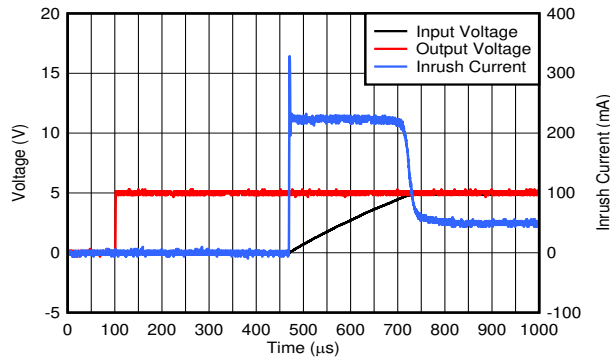


图 6-31. PGADJ Threshold vs Temperature



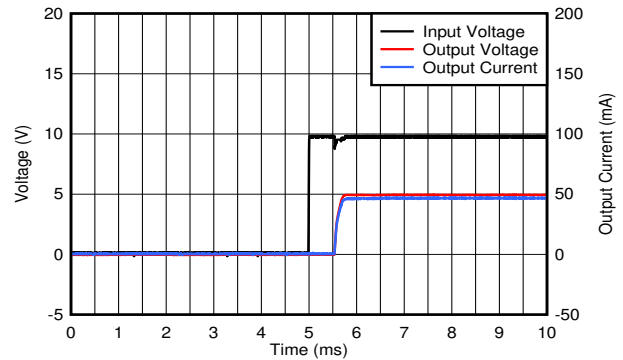
$C_{OUT} = 10\ \mu\text{F}$

图 6-32. Startup Plot Inrush Current



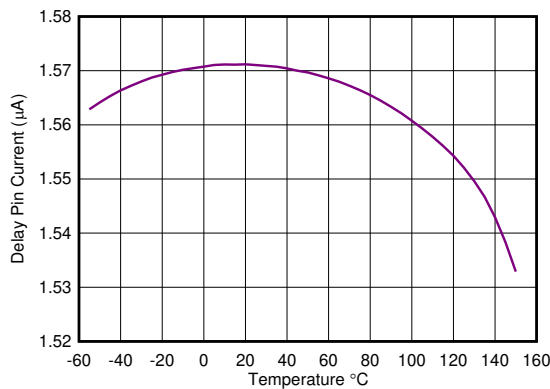
$V_{IN} = 13.5\text{ V}$ ,  $C_{OUT} = 10\ \mu\text{F}$

图 6-33. Startup Plot With EN



$C_{OUT} = 10\ \mu\text{F}$

图 6-34. Startup Plot



$V_{DELAY} = 1\ \text{V}$

图 6-35. Delay Pin Current vs Temperature

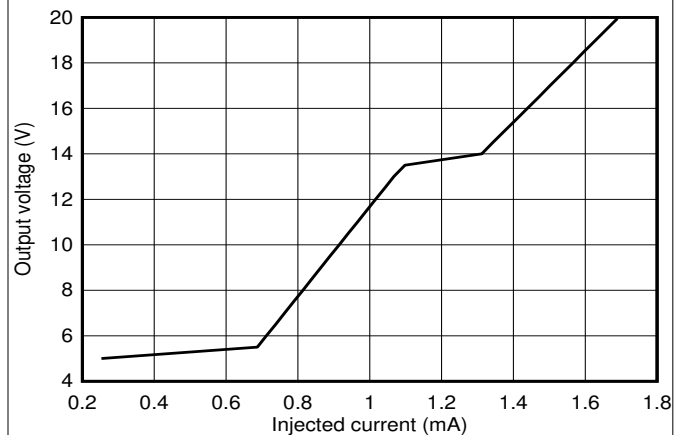
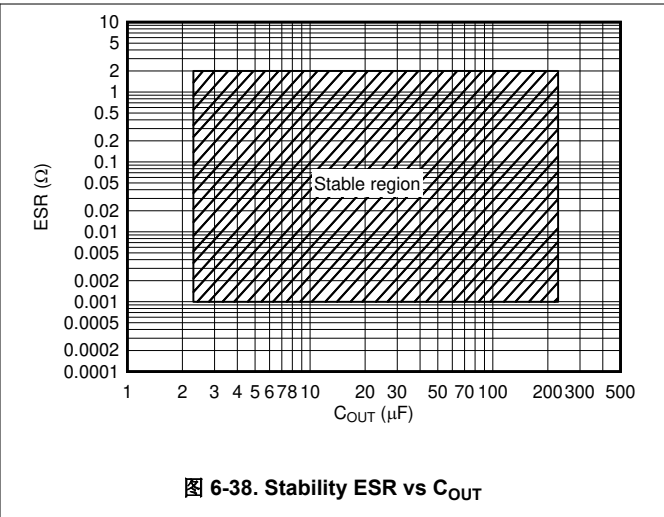
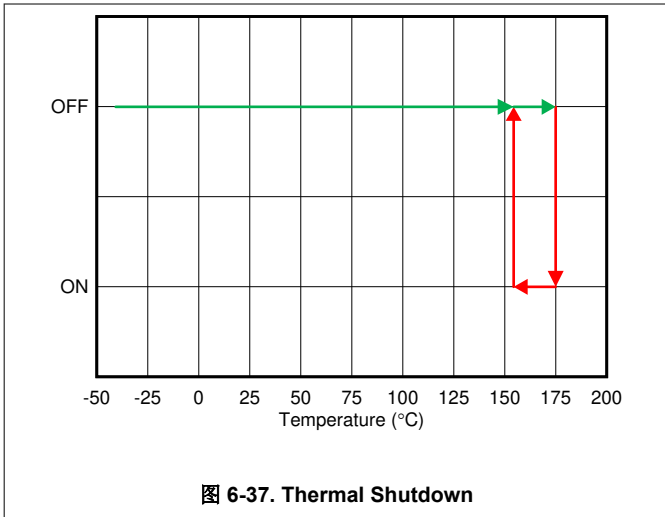


图 6-36. Output Voltage vs Injected Current

### 6.7 Typical Characteristics (continued)

specified at  $T_J = -40^{\circ}\text{C}$  to  $+150^{\circ}\text{C}$ ,  $V_{IN} = 13.5\text{ V}$ ,  $I_{OUT} = 100\ \mu\text{A}$ ,  $C_{OUT} = 2.2\ \mu\text{F}$ ,  $1\ \text{m}\Omega < C_{OUT}\ \text{ESR} < 2\ \Omega$ ,  $C_{IN} = 1\ \mu\text{F}$ , and  $V_{EN} = 2\ \text{V}$  (unless otherwise noted)



## 7 Detailed Description

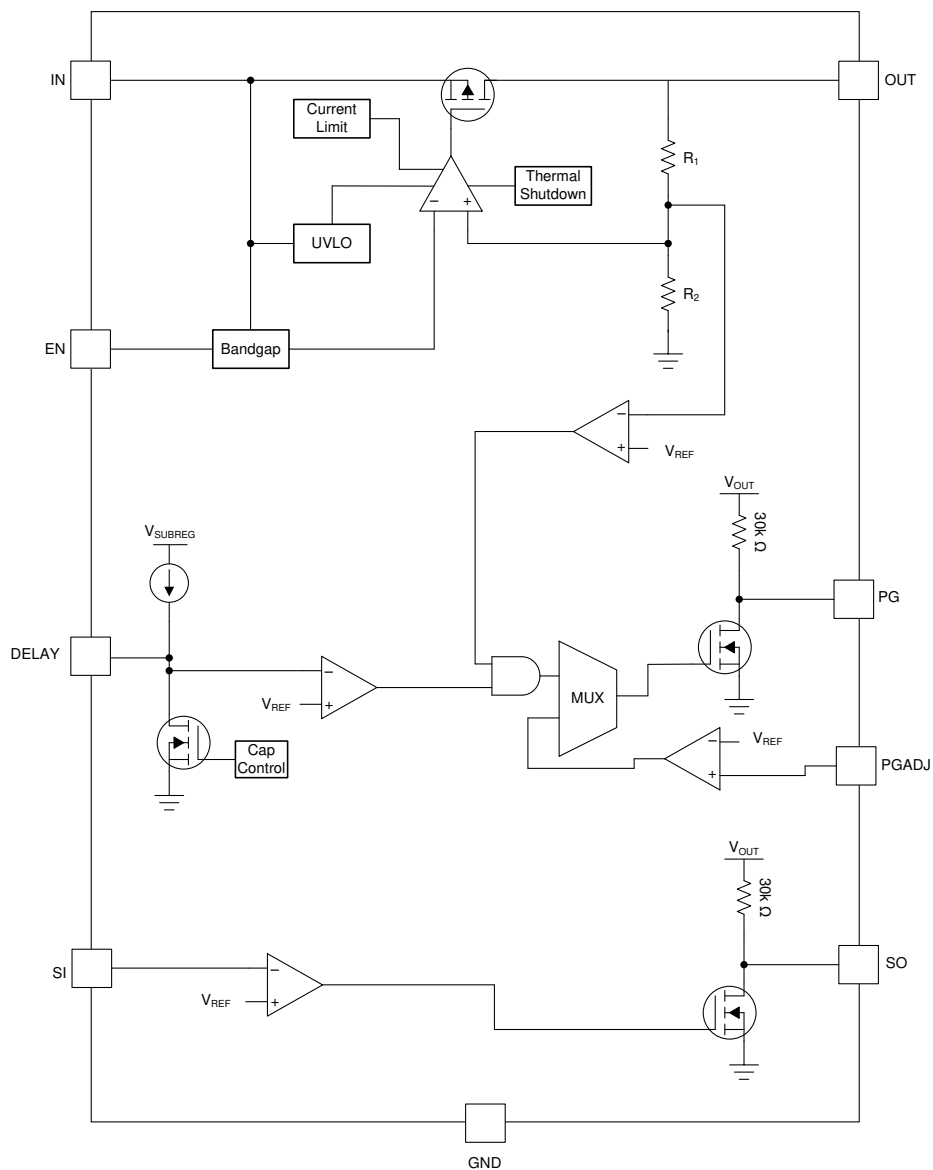
### 7.1 Overview

The TPS7B85-Q1 is a low-dropout linear regulator (LDO) designed to connect to the battery in automotive applications. The device has an input voltage range extending to 40 V, which allows the device to withstand transients (such as load dumps) that are anticipated in automotive systems. With only a 18- $\mu$ A quiescent current at light loads, the device is an optimal solution for powering always-on components.

The device has a state-of-the-art transient response that allows the output to quickly react to changes in the load or line (for example, during cold-crank conditions). Additionally, the device has a novel architecture that minimizes output overshoot when recovering from dropout. During normal operation, the device has a tight DC accuracy of  $\pm 0.75\%$  over line, load, and temperature.

The TPS7B85-Q1 is equipped with power-good and integrated voltage monitoring. The power-good delay and voltage threshold can be adjusted by external components. The integrated voltage detector can be used to monitor the input voltage and alert downstream components (such as MCUs) when the battery voltage begins to fall.

### 7.2 Functional Block Diagram



## 7.3 Feature Description

### 7.3.1 Enable (EN)

The enable pin for the device is an active-high pin. The output voltage is enabled when the voltage of the enable pin is greater than the high-level input voltage of the EN pin and disabled with the enable pin voltage is less than the low-level input voltage of the EN pin. If independent control of the output voltage is not needed, connect the enable pin to the input of the device.

### 7.3.2 Power-Good (PG)

The PG signal provides an easy solution to meet demanding sequencing requirements because PG alerts when the output nears its nominal value. PG can be used to signal other devices in a system when the output voltage is near, at, or above the set output voltage ( $V_{OUT(nom)}$ ). 图 7-1 shows a simplified schematic. The PG signal has an internal pullup resistor to the nominal output voltage and is active high. The PG circuit sets the PG pin into a high-impedance state to indicate that the power is good.

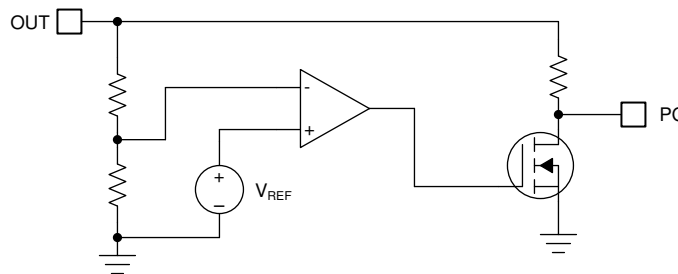


图 7-1. Simplified Power-Good Schematic

#### 7.3.2.1 Adjustable Power-Good (PGADJ)

One unique feature of this LDO, as shown in 图 7-2, is the ability to adjust the power-good threshold through the use of a resistor divider. The adjustable power-good threshold allows the PG threshold to be set to the desired level to further assist in transient detection or sequencing requirements. If this feature is not desired, then tie the PGADJ pin to GND and the default PG threshold is used. For more information on how to calculate the power-good threshold, see the [Setting the Adjustable Power-Good Delay](#) section.

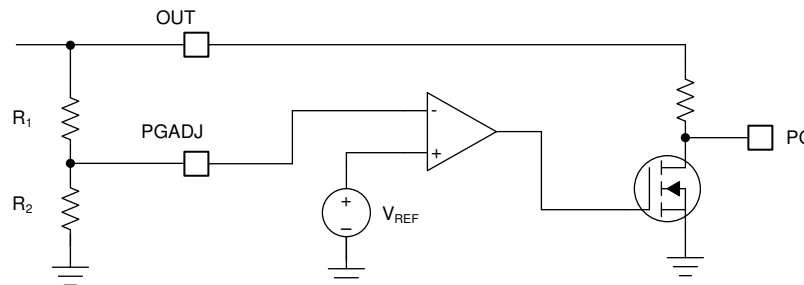


图 7-2. Typical Use of Power-Good Adjust Pin

#### 7.3.3 Adjustable Power-Good Delay Timer (DELAY)

The power-good delay period is a function of the external capacitor on the DELAY pin. The adjustable delay configures the amount of time required before the PG pin becomes high. This delay is configured by connecting an external capacitor from this pin to GND. 图 7-3 illustrates the typical timing diagram for the power-good delay pin. If the DELAY pin is left floating, the power-good delay is  $t_{(DLY\_FIX)}$ . For more information on how to program the PG delay, see the [Setting the Adjustable Power-Good Delay](#) section.

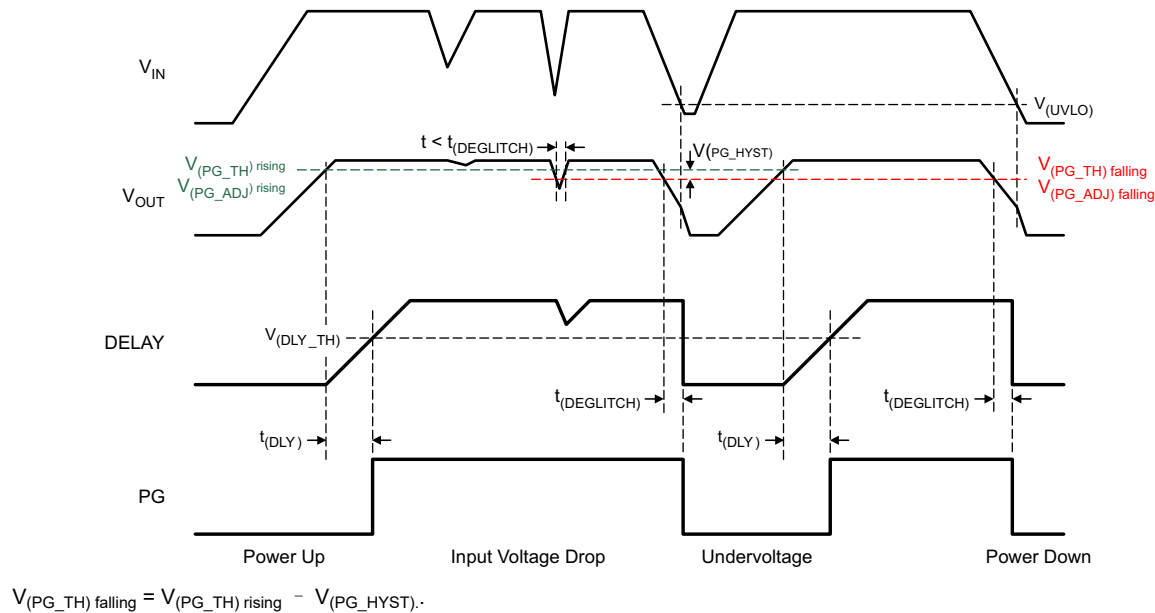


图 7-3. Typical Power-Good Timing Diagram

### 7.3.4 Sense Comparator

The sense comparator compares the input signal with an internal voltage reference of 1.223 V for a rising threshold and 1.123 V for a falling threshold. Using an external voltage divider makes this comparator very flexible in the application.

The device can supervise the input voltage either before or after the protection diode and provides additional information to the microprocessor (such as low-voltage warnings).

### 7.3.5 Undervoltage Lockout

The device has an independent undervoltage lockout (UVLO) circuit that monitors the input voltage, allowing a controlled and consistent turn on and off of the output voltage. To prevent the device from turning off if the input drops during turn on, the UVLO has hysteresis as specified in the *Electrical Characteristics* table.

### 7.3.6 Thermal Shutdown

The device contains a thermal shutdown protection circuit to disable the device when the junction temperature ( $T_J$ ) of the pass transistor rises to  $T_{SD(\text{shutdown})}$  (typical). Thermal shutdown hysteresis assures that the device resets (turns on) when the temperature falls to  $T_{SD(\text{reset})}$  (typical).

The thermal time-constant of the semiconductor die is fairly short, thus the device may cycle on and off when thermal shutdown is reached until power dissipation is reduced. Power dissipation during startup can be high from large  $V_{IN} - V_{OUT}$  voltage drops across the device or from high inrush currents charging large output capacitors. Under some conditions, the thermal shutdown protection disables the device before startup completes.

For reliable operation, limit the junction temperature to the maximum listed in the *Recommended Operating Conditions* table. Operation above this maximum temperature causes the device to exceed its operational specifications. Although the internal protection circuitry of the device is designed to protect against thermal overall conditions, this circuitry is not intended to replace proper heat sinking. Continuously running the device into thermal shutdown or above the maximum recommended junction temperature reduces long-term reliability.

### 7.3.7 Current Limit

The device has an internal current limit circuit that protects the regulator during transient high-load current faults or shorting events. The current limit is a brickwall scheme. In a high-load current fault, the brickwall scheme limits the output current to the current limit ( $I_{CL}$ ).  $I_{CL}$  is listed in the *Electrical Characteristics* table.



The output voltage is not regulated when the device is in current limit. When a current limit event occurs, the device begins to heat up because of the increase in power dissipation. When the device is in brickwall current limit, the pass transistor dissipates power  $[(V_{IN} - V_{OUT}) \times I_{CL}]$ . If thermal shutdown is triggered, the device turns off. After the device cools down, the internal thermal shutdown circuit turns the device back on. If the output current fault condition continues, the device cycles between current limit and thermal shutdown. For more information on current limits, see the [Know Your Limits application report](#).

图 7-4 shows a diagram of the current limit.

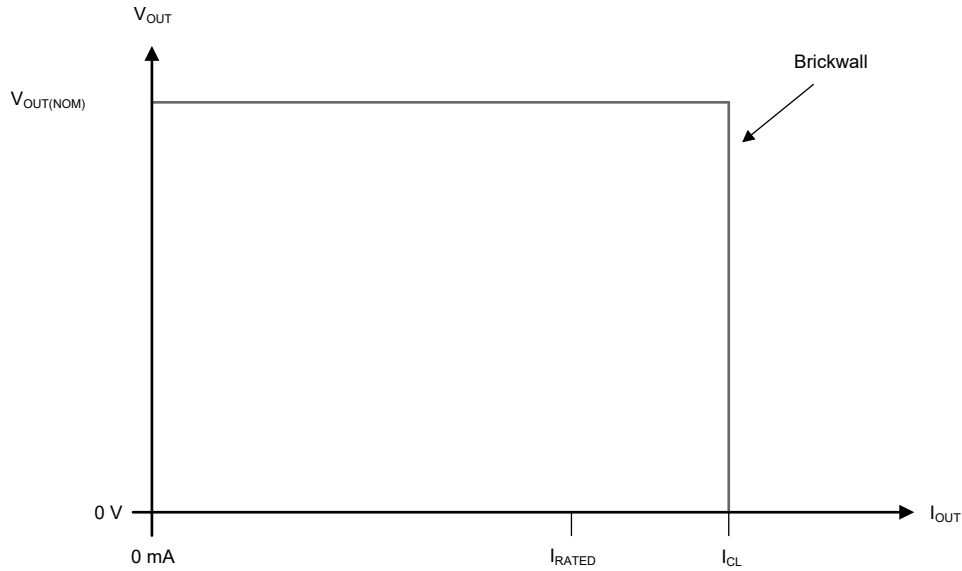


图 7-4. Current Limit

## 7.4 Device Functional Modes

### 7.4.1 Device Functional Mode Comparison

The *Device Functional Mode Comparison* table shows the conditions that lead to the different modes of operation. See the *Electrical Characteristics* table for parameter values.

表 7-1. Device Functional Mode Comparison

OPERATING MODE	PARAMETER			
	$V_{IN}$	$V_{EN}$	$I_{OUT}$	$T_J$
Normal operation	$V_{IN} > V_{OUT(nom)} + V_{DO}$ and $V_{IN} > V_{IN(min)}$	$V_{EN} > V_{EN(HI)}$	$I_{OUT} < I_{OUT(max)}$	$T_J < T_{SD(shutdown)}$
Dropout operation	$V_{IN(min)} < V_{IN} < V_{OUT(nom)} + V_{DO}$	$V_{EN} > V_{EN(HI)}$	$I_{OUT} < I_{OUT(max)}$	$T_J < T_{SD(shutdown)}$
Disabled (any true condition disables the device)	$V_{IN} < V_{UVLO}$	$V_{EN} < V_{EN(LOW)}$	Not applicable	$T_J > T_{SD(shutdown)}$

### 7.4.2 Normal Operation

The device regulates to the nominal output voltage when the following conditions are met:

- The input voltage is greater than the nominal output voltage plus the dropout voltage ( $V_{OUT(nom)} + V_{DO}$ )
- The output current is less than the current limit ( $I_{OUT} < I_{CL}$ )
- The device junction temperature is less than the thermal shutdown temperature ( $T_J < T_{SD}$ )
- The enable voltage has previously exceeded the enable rising threshold voltage and has not yet decreased to less than the enable falling threshold

### 7.4.3 Dropout Operation

If the input voltage is lower than the nominal output voltage plus the specified dropout voltage, but all other conditions are met for normal operation, the device operates in dropout mode. In this mode, the output voltage tracks the input voltage. During this mode, the transient performance of the device becomes significantly degraded because the pass transistor is in the ohmic or triode region, and acts as a switch. Line or load transients in dropout can result in large output-voltage deviations.

When the device is in a steady dropout state (defined as when the device is in dropout,  $V_{IN} < V_{OUT(NOM)} + V_{DO}$ , directly after being in a normal regulation state, but *not* during startup), the pass transistor is driven into the ohmic or triode region. When the input voltage returns to a value greater than or equal to the nominal output voltage plus the dropout voltage ( $V_{OUT(NOM)} + V_{DO}$ ), the output voltage can overshoot for a short period of time while the device pulls the pass transistor back into the linear region.

### 7.4.4 Disabled

The output of the device can be shutdown by forcing the voltage of the enable pin to less than the maximum EN pin low-level input voltage (see the *Electrical Characteristics* table). When disabled, the pass transistor is turned off, internal circuits are shutdown, and the output voltage is actively discharged to ground by an internal discharge circuit from the output to ground.

## 8 Application and Implementation

### Note

以下应用部分的信息不属于 TI 组件规范，TI 不担保其准确性和完整性。客户应负责确定 TI 组件是否适用于其应用。客户应验证并测试其设计，以确保系统功能。

### 8.1 Application Information

#### 8.1.1 Input and Output Capacitor Selection

The TPS7B85-Q1 requires an output capacitor of 2.2  $\mu\text{F}$  or larger (1  $\mu\text{F}$  or larger capacitance) for stability and an equivalent series resistance (ESR) between 0.001  $\Omega$  and 2  $\Omega$ . For the best transient performance, use X5R- and X7R-type ceramic capacitors because these capacitors have minimal variation in value and ESR over temperature. When choosing a capacitor for a specific application, be mindful of the DC bias characteristics for the capacitor. Higher output voltages cause a significant derating of the capacitor. For best performance, the maximum recommended output capacitance is 220  $\mu\text{F}$ .

Although an input capacitor is not required for stability, good analog design practice is to connect a capacitor from IN to GND. Some input supplies have a high impedance, thus placing the input capacitor on the input supply helps reduce the input impedance. This capacitor counteracts reactive input sources and improves transient response, input ripple, and PSRR. If the input supply has a high impedance over a large range of frequencies, several input capacitors can be used in parallel to lower the impedance over frequency. Use a higher-value capacitor if large, fast, rise-time load transients are anticipated, or if the device is located several inches from the input power source.

#### 8.1.2 Dropout Voltage

Dropout voltage ( $V_{\text{DO}}$ ) is defined as the input voltage minus the output voltage ( $V_{\text{IN}} - V_{\text{OUT}}$ ) at the rated output current ( $I_{\text{RATED}}$ ), where the pass transistor is fully on.  $I_{\text{RATED}}$  is the maximum  $I_{\text{OUT}}$  listed in the *Recommended Operating Conditions* table. The pass transistor is in the ohmic or triode region of operation, and acts as a switch. The dropout voltage indirectly specifies a minimum input voltage greater than the nominal programmed output voltage at which the output voltage is expected to stay in regulation. If the input voltage falls to less than the nominal output regulation, then the output voltage falls as well.

For a CMOS regulator, the dropout voltage is determined by the drain-source on-state resistance ( $R_{\text{DS(ON)}}$ ) of the pass transistor. Therefore, if the linear regulator operates at less than the rated current, the dropout voltage for that current scales accordingly. The following equation calculates the  $R_{\text{DS(ON)}}$  of the device.

$$R_{\text{DS(ON)}} = \frac{V_{\text{DO}}}{I_{\text{RATED}}} \quad (1)$$

#### 8.1.3 Reverse Current

Excessive reverse current can damage this device. Reverse current flows through the intrinsic body diode of the pass transistor instead of the normal conducting channel. At high magnitudes, this current flow degrades the long-term reliability of the device.

Conditions where reverse current can occur are outlined in this section, all of which can exceed the absolute maximum rating of  $V_{\text{OUT}} \leq V_{\text{IN}} + 0.3 \text{ V}$ .

- If the device has a large  $C_{\text{OUT}}$  and the input supply collapses with little or no load current
- The output is biased when the input supply is not established
- The output is biased above the input supply

If reverse current flow is expected in the application, external protection is recommended to protect the device. Reverse current is not limited in the device, so external limiting is required if extended reverse voltage operation is anticipated.

### 8.1.4 Power Dissipation ( $P_D$ )

Circuit reliability requires consideration of the device power dissipation, location of the circuit on the printed circuit board (PCB), and correct sizing of the thermal plane. The PCB area around the regulator must have few or no other heat-generating devices that cause added thermal stress.

To first-order approximation, power dissipation in the regulator depends on the input-to-output voltage difference and load conditions. The following equation calculates power dissipation ( $P_D$ ).

$$P_D = (V_{IN} - V_{OUT}) \times I_{OUT} \quad (2)$$

---

#### Note

Power dissipation can be minimized, and therefore greater efficiency can be achieved, by correct selection of the system voltage rails. For the lowest power dissipation use the minimum input voltage required for correct output regulation.

---

For devices with a thermal pad, the primary heat conduction path for the device package is through the thermal pad to the PCB. Solder the thermal pad to a copper pad area under the device. This pad area must contain an array of plated vias that conduct heat to additional copper planes for increased heat dissipation.

The maximum power dissipation determines the maximum allowable ambient temperature ( $T_A$ ) for the device. According to the following equation, power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance ( $R_{\theta JA}$ ) of the combined PCB and device package and the temperature of the ambient air ( $T_A$ ).

$$T_J = T_A + (R_{\theta JA} \times P_D) \quad (3)$$

Thermal resistance ( $R_{\theta JA}$ ) is highly dependent on the heat-spreading capability built into the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The junction-to-ambient thermal resistance listed in the *Thermal Information* table is determined by the JEDEC standard PCB and copper-spreading area, and is used as a relative measure of package thermal performance.

#### 8.1.4.1 Thermal Performance Versus Copper Area

The most used thermal resistance parameter,  $R_{\theta JA}$ , is highly dependent on the heat-spreading capability built into the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The  $R_{\theta JA}$  recorded in the *Thermal Information* table in the [Specifications](#) section is determined by the JEDEC standard (see [Figure 8-1](#)), PCB, and copper-spreading area, and is only used as a relative measure of package thermal performance. For a well-designed thermal layout,  $R_{\theta JA}$  is actually the sum of the package junction-to-case (bottom) thermal resistance ( $R_{\theta JCBot}$ ) plus the thermal resistance contribution by the PCB copper.

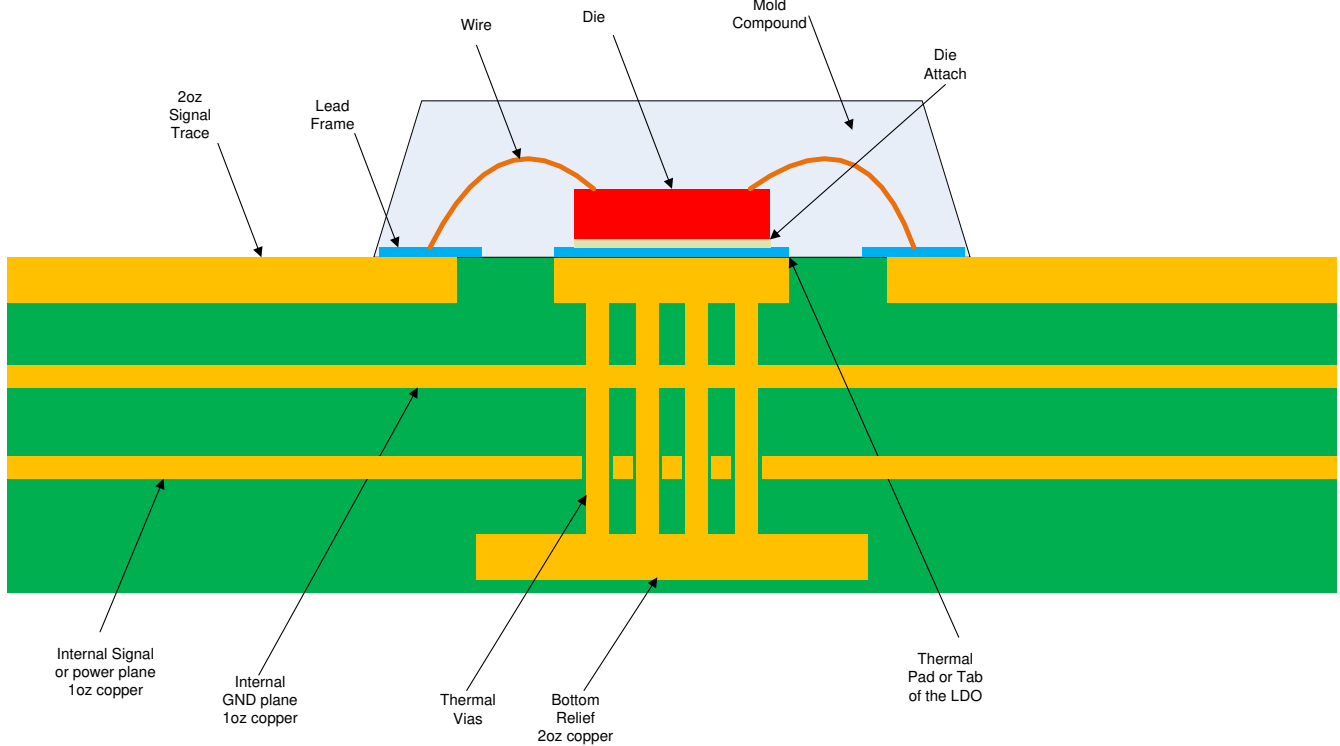


图 8-1. JEDEC Standard 2s2p PCB

图 8-2 和 图 8-3 描绘了  $R_{\theta JA}$  和  $\psi_{JB}$  相对于铜面积和厚度的函数。这些图是在一个 101.6-mm x 101.6-mm x 1.6-mm 的两层和四层 PCB 上生成的。对于四层板，内层使用 1-oz 铜厚度。外层模拟了 1-oz 和 2-oz 铜厚度。在器件的热垫下方有一个 4 x 4 的铜通孔阵列，孔径为 300- $\mu$ m，铜镀层厚度为 25- $\mu$ m。铜通孔将顶层、底层和（对于四层板）第一个内层 GND 平面连接起来。每一层都有一个面积相等的铜平面。

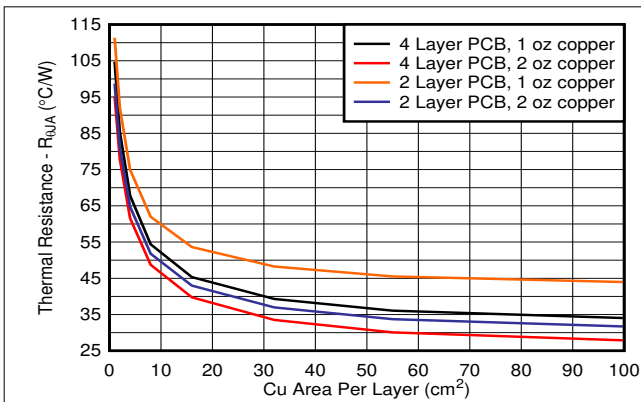


图 8-2.  $R_{\theta JA}$  vs Copper Area 2s2p DRC Package

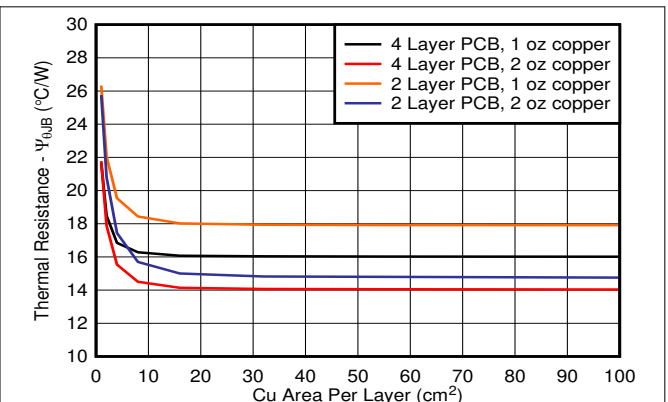


图 8-3.  $R_{\psi JB}$  vs Copper Area 2s2p DRC Package

### 8.1.5 Estimating Junction Temperature

The JEDEC standard now recommends the use of psi ( $\Psi$ ) thermal metrics to estimate the junction temperatures of the linear regulator when in-circuit on a typical PCB board application. These metrics are not thermal resistance parameters and instead offer a practical and relative way to estimate junction temperature. These psi metrics are determined to be significantly independent of the copper area available for heat-spreading. The *Thermal Information* table lists the primary thermal metrics, which are the junction-to-top characterization parameter ( $\psi_{JT}$ ) and junction-to-board characterization parameter ( $\psi_{JB}$ ). These parameters provide two methods for calculating the junction temperature ( $T_J$ ), as described in the following equations. Use the junction-to-top characterization parameter ( $\psi_{JT}$ ) with the temperature at the center-top of device package ( $T_T$ ) to calculate the junction temperature. Use the junction-to-board characterization parameter ( $\psi_{JB}$ ) with the PCB surface temperature 1 mm from the device package ( $T_B$ ) to calculate the junction temperature.

$$T_J = T_T + \psi_{JT} \times P_D \quad (4)$$

where:

- $P_D$  is the dissipated power
- $T_T$  is the temperature at the center-top of the device package

$$T_J = T_B + \psi_{JB} \times P_D \quad (5)$$

where

- $T_B$  is the PCB surface temperature measured 1 mm from the device package and centered on the package edge

For detailed information on the thermal metrics and how to use them, see the [Semiconductor and IC Package Thermal Metrics application report](#).

### 8.1.6 SI Pin

#### 8.1.6.1 Calculating the Sense Input (SI) Pin Threshold

To use the SI pin, connect this pin to the rail being monitored through a resistor divider. This input can be configured as an undervoltage supervisor that can monitor voltage rails greater than 1.2 V or used as an overvoltage supervisor with an inverted output. 表 8-1 lists typical 1% resistor values for undervoltage monitoring where the trip point is a 5% threshold. The resistor values can be scaled to decrease the amount of current flowing through the resistor divider, but increasing the resistor values also decreases the accuracy of the resistor divider. General practice is for the current flowing through the resistor divider to be 100 times greater than the current going into the SI pin. This practice ensures the highest possible accuracy. 方程式 6 can be used to calculate the resistors required in the resistor divider for any desired falling threshold. 图 8-4 depicts the typical timing for this comparator and 图 8-5 illustrates a block diagram for the adjustable operation.

$$V_{\text{mon(falling)}} = V_{\text{SI(LOW)}} \times \left( 1 + \frac{R1}{R2} \right) \quad (6)$$

表 8-1. SI Resistor Divider Values

INPUT VOLTAGE (V)	5% THRESHOLD		
	R1 (kΩ)	R2 (kΩ)	THRESHOLD VOLTAGE (V)
3.3	18.2	10	3.13
5	32.4	10	4.71
6	41.2	10	5.68
7	49.9	10	6.65
8	59	10	7.66
9	66.5	10	8.49
10	75.5	10	9.49
11	80.6	10	10.06
12	93.1	10	11.44
13.5	105	10	12.77

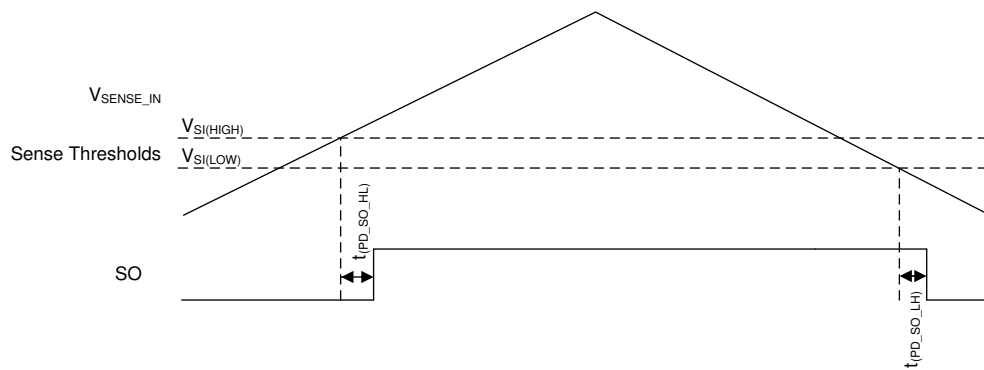


图 8-4. SI Timing Diagram

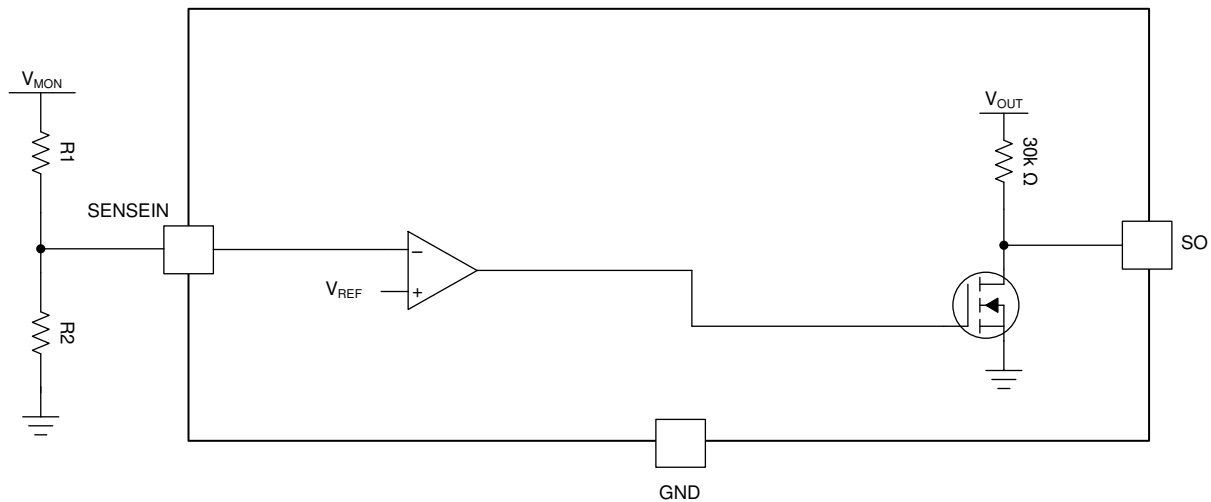


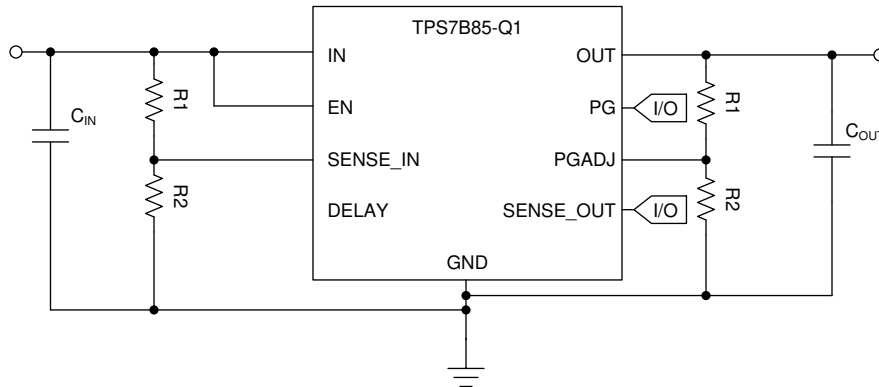
图 8-5. SI Basic Block Diagram

### 8.1.6.2 Different Uses for the Sense Input Pin

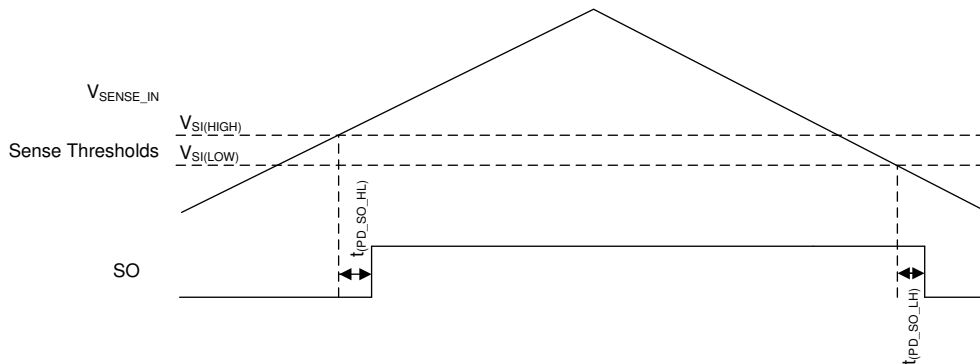
The sense input pin incorporates a comparator with hysteresis into the LDO. The SI pin can help replace a supervisor in the system by connecting the SI pin to rails that need to be monitored. The three most common uses for this supervisor are described in the [Monitoring Input Voltage](#), [Creating OV and UV Power-Good](#), and [Monitoring a Separate Supply Voltage](#) sections.

### 8.1.6.2.1 Monitoring Input Voltage

Monitoring the input voltage of the LDO, as shown in [图 8-6](#), is the most common way that the SI pin is used. The device has a built-in precision comparator that allows the device to compare a divided-down version of the input to the internal reference of the LDO. When the voltage on the sense pin is below  $V_{SI(Low)}$ , the output of the SO pin is low. However, when  $V_{SI}$  crosses  $V_{SI(HIGH)}$  the voltage on the SO pin gets pulled up to  $V_{OUT}$  through a pullup resistor,  $R_{SO}$ . This pin also has built-in hysteresis to keep the pin from toggling between the two states from small changes on the SENSE voltage. [图 8-7](#) shows a typical timing diagram for the SENSE pin.



**图 8-6. Monitoring the Device Input Voltage**



**图 8-7. SENSE Pin Timing Diagram**

### 8.1.6.2.2 Creating OV and UV Power-Good

Another feature that is often desired is the ability to monitor the output voltage for overvoltage (OV) or undervoltage (UV) events. Because the integrated power-good pin only detects undervoltage events, a separate solution must be implemented to monitor for overvoltage issues. This monitoring can be done by using the integrated SI pin and connecting this pin to the output through a resistor divider. Then place the rising threshold of the SI pin where the output voltage is going to be flagged as overvoltage. [方程式 7](#) depicts how to calculate the resistor divider for this application based on the desired overvoltage threshold. If this method is used for creating an overvoltage detection, the output of the overvoltage signal has inverted logic. [图 8-8](#) shows the typical configuration for using the device as an overvoltage monitor.

$$V_{\text{mon(rising)}} = V_{SI(HIGH)} \times \left( 1 + \frac{R1}{R2} \right) \quad (7)$$



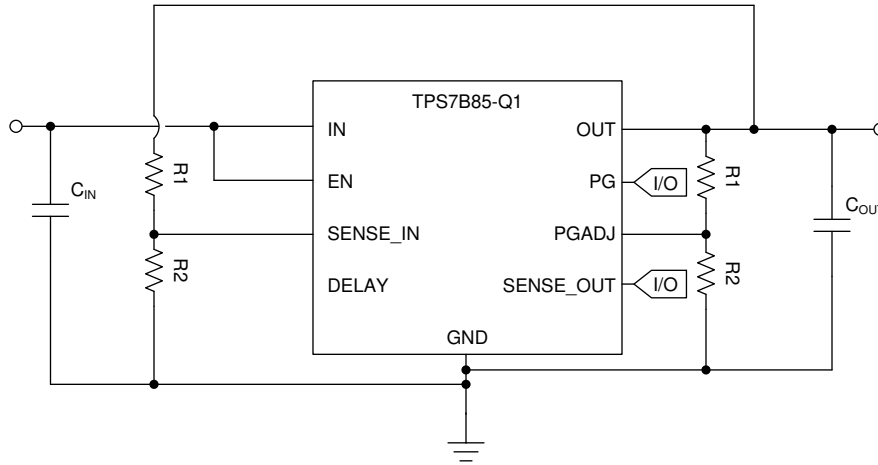


图 8-8. Creating an Overvoltage Detector on the Output

### 8.1.6.2.3 Monitoring a Separate Supply Voltage

One of the final applications for the SI pin is monitoring a separate supply. This method can be implemented as either an overvoltage detection or undervoltage detection for the externally monitored supply. 方程式 6 和 方程式 7 可以用于计算所需的电阻分压器，以实现对外部电源的监控。图 8-9 显示了此监控应用的块图。

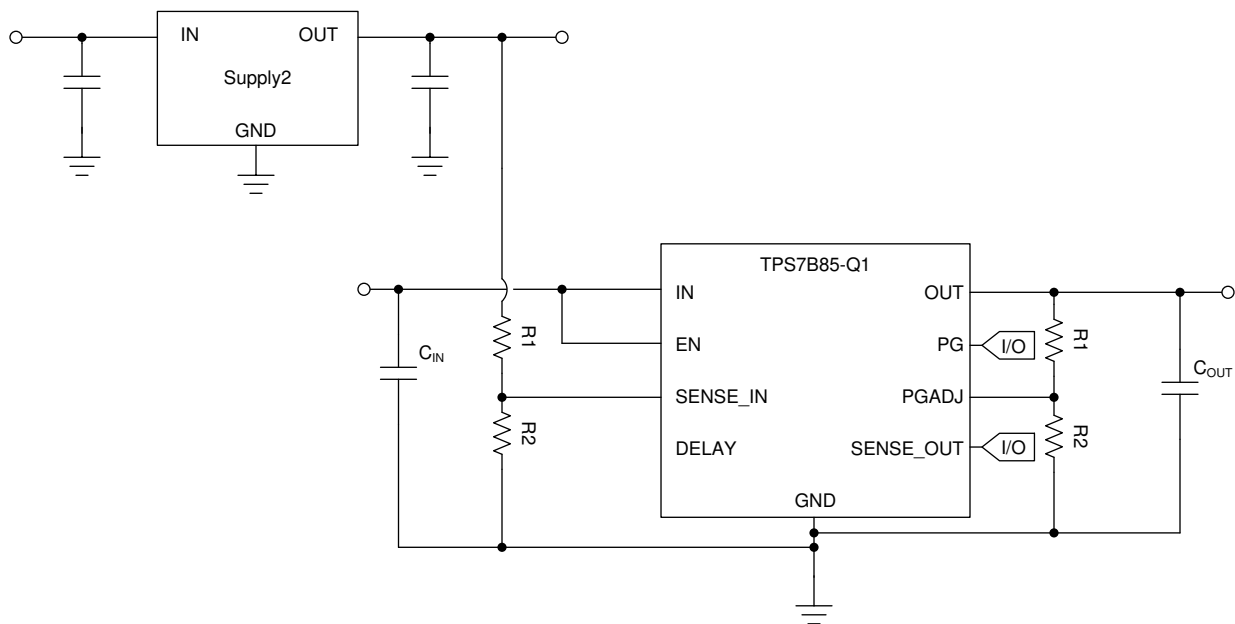


图 8-9. Monitoring a Separate Power Supply

### 8.1.7 Pulling Up the SO and PG Pins to a Different Voltage

Because the sense out (SO) and power-good (PG) pins are pulled up internally to the output rail, they cannot be pulled up to any voltage or wire AND'd like a typical open-drain PG output can be. If these signals must be pulled up to another logic level then an external circuit can be implemented using a PMOS transistor and a pullup resistor. Implementing the circuit shown in 图 8-10 允许输出被拉到任何逻辑轨。此实现也允许输出像传统的电源好引脚一样被 AND 在一起。

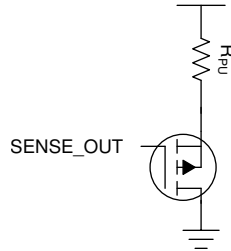


图 8-10. Additional Components for the SO and PG Pins to be Pulled Up to Another Rail

### 8.1.8 Power-Good

#### 8.1.8.1 Setting the Adjustable Power-Good Threshold

The power-good threshold is also adjustable from 1 V to 18 V with an external resistor divider between PGADJ and OUT. Use 方程式 8 to calculate this threshold:

$$V_{(PG\_ADJ)falling} = V_{(PGADJ\_TH)falling} \times \left( \frac{R_1 + R_2}{R_2} \right)$$

$$V_{(PG\_ADJ)rising} = \left[ V_{(PGADJ\_TH)falling} + V_{PGADJ(HYST)} \right] \times \left( \frac{R_1 + R_2}{R_2} \right) \quad (8)$$

where

- $V_{(PG\_ADJ)rising}$ ,  $V_{(PG\_ADJ)falling}$  is the adjustable power-good threshold
- $V_{(PGADJ\_TH)falling}$  is the internal comparator reference voltage of the PGADJ pin

By setting the power-good threshold  $V_{(PG\_ADJ)rising}$ , when  $V_{OUT}$  exceeds this threshold, the PG output turns high after the power-good delay period has expired. When  $V_{OUT}$  falls below  $V_{(PG\_ADJ)falling}$ , the PG output turns low after a short deglitch time. 图 8-11 shows a diagram of the PG threshold.

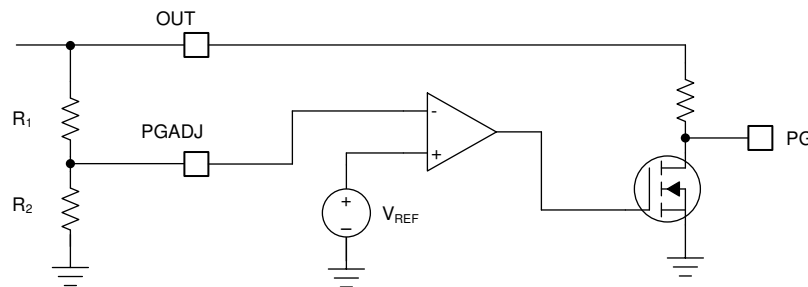


图 8-11. Adjustable Power-Good Threshold

#### 8.1.8.2 Setting the Adjustable Power-Good Delay

The power-good delay time can be set in two ways: either by floating the delay pin or by connecting a capacitor from this pin to GND. When the DELAY pin is floating, the time defaults to  $t_{(DLY\_FIX)}$ . The delay time is set by 方程式 9 if a capacitor is connected between the DELAY pin and GND.

$$t = t_{(DLY\_FIX)} + C_{DELAY} \left( \frac{V_{DLY(TH)}}{I_{DLY(CHARGE)}} \right) \quad (9)$$

## 8.2 Typical Application

图 8-12 shows a typical application circuit for the TPS7B85-Q1. Use different values of external components, depending on the end application. An application may require a larger output capacitor during fast load steps in order to prevent a reset from occurring. TI recommends a low-ESR ceramic capacitor with a dielectric of type X5R or X7R.

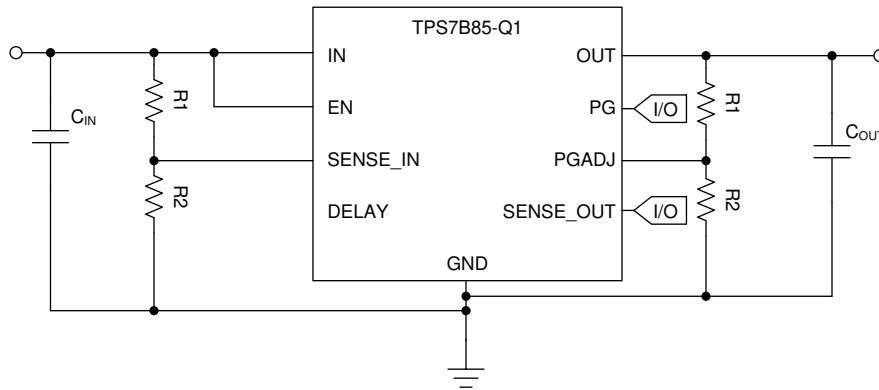


图 8-12. Typical Application Schematic for the TPS7B85-Q1

### 8.2.1 Design Requirements

For this design example, use the parameters listed in 表 8-2 as the input parameters.

表 8-2. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Input voltage range	6 V to 40 V
Output voltage	5 V
Output current	100 mA
Output capacitor	10 $\mu$ F
Power-good delay capacitor	100 nF
Sense input trip threshold	5.5 V

### 8.2.2 Detailed Design Procedure

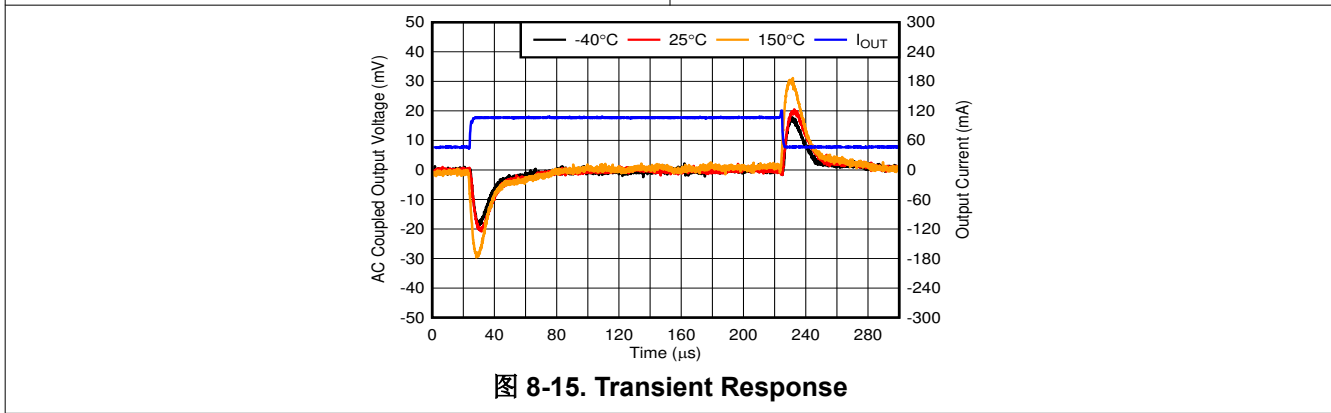
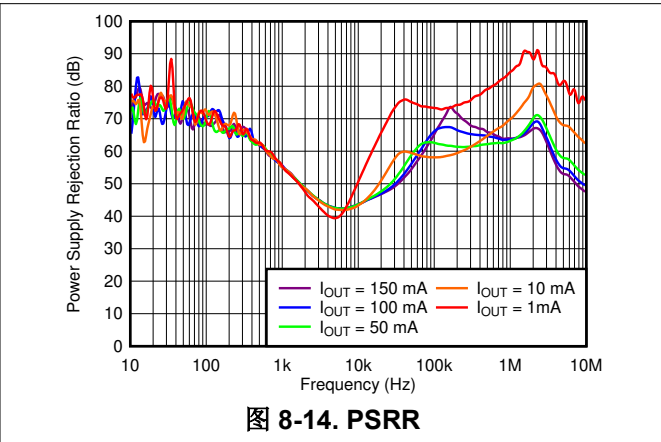
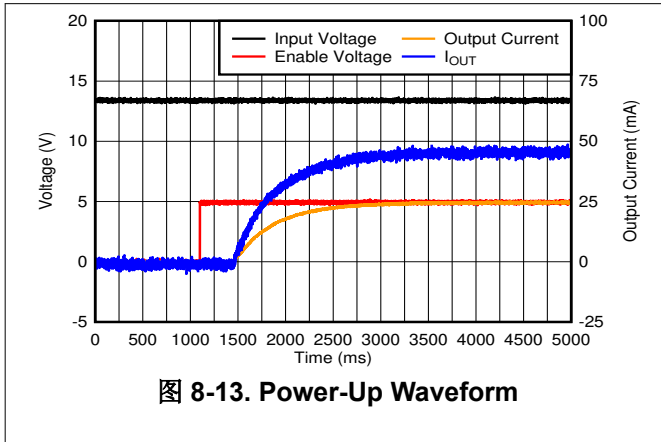
#### 8.2.2.1 Input Capacitor

The device requires an input decoupling capacitor, the value of which depends on the application. The typical recommended value for the decoupling capacitor is 1  $\mu$ F. The voltage rating must be greater than the maximum input voltage.

#### 8.2.2.2 Output Capacitor

The device requires an output capacitor to stabilize the output voltage. The capacitor value must be between 2.2  $\mu$ F and 200  $\mu$ F and the ESR range must be between 1 m $\Omega$  and 2  $\Omega$ . For this design a low ESR, 10- $\mu$ F ceramic capacitor was used to improve transient performance.

### 8.2.3 Application Curves



## 9 Power Supply Recommendations

This device is designed for operation from an input voltage supply with a range between 3 V and 40 V. This input supply must be well regulated. If the input supply is located more than a few inches from the TPS7B85-Q1, add an electrolytic capacitor and a ceramic bypass capacitor at the input.

## 10 Layout


### 10.1 Layout Guidelines

For best overall performance, place all circuit components on the same side of the circuit board and as near as practical to the respective LDO pin connections. Place ground return connections to the input and output capacitor, and to the LDO ground pin as close as possible to each other, connected by a wide, component-side, copper surface. The use of vias and long traces to the input and output capacitors is strongly discouraged and negatively affects system performance. TI also recommends a ground reference plane either embedded in the PCB itself or located on the bottom side of the PCB opposite the components. This reference plane serves to assure accuracy of the output voltage, shield noise, and behaves similarly to a thermal plane to spread (or sink) heat from the LDO device when connected to the thermal pad. In most applications, this ground plane is necessary to meet thermal requirements.

#### 10.1.1 Package Mounting

Solder pad footprint recommendations for the TPS7B85-Q1 are available at the end of this document and at [www.ti.com](http://www.ti.com).

#### 10.1.2 Board Layout Recommendations to Improve PSRR and Noise Performance

As depicted in  10-1, place the input and output capacitors close to the device for the layout of the TPS7B85-Q1. In order to enhance the thermal performance, place as many vias as possible around the device. These vias improve the heat transfer between the different GND planes in the PCB.

To improve ac performance such as PSRR, output noise, and transient response, TI recommends a board design with separate ground planes for IN and OUT, with each ground plane connected only at the GND pin of the device. In addition, the ground connection for the output capacitor must connect directly to the GND pin of the device.

Minimize equivalent series inductance (ESL) and ESR in order to maximize performance and ensure stability. Place each capacitor as close as possible to the device and on the same side of the PCB as the regulator itself.

Do not place any of the capacitors on the opposite side of the PCB from where the regulator is installed. TI strongly discourages the use of vias and long traces to connect the capacitors because these can negatively impact system performance and may even cause instability.

If possible, and to ensure the maximum performance specified in this document, use the same layout pattern used for the TPS7B85-Q1 evaluation board, available at [www.ti.com](http://www.ti.com).

## 10.2 Layout Example

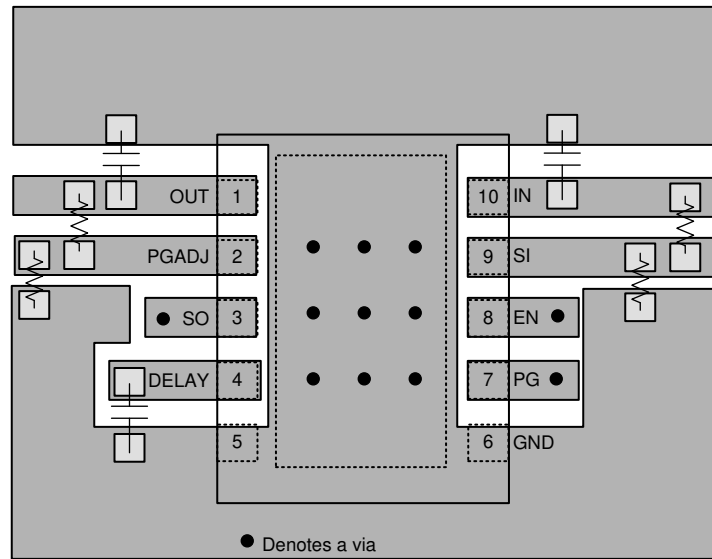


图 10-1. VSON (DRC) Layout

## 11 Device and Documentation Support

### 11.1 Device Support

#### 11.1.1 Device Nomenclature

表 11-1. Device Nomenclature<sup>(1)</sup>

PRODUCT	V <sub>OUT</sub>
TPS7B85xxQWyyyRQ1	<p><b>xx</b> is the nominal output voltage (for example, 33 = 3.3 V; 50 = 5.0 V).</p> <p><b>Q</b> indicates that this device is a grade-1 device in accordance with the AEC-Q100 standard.</p> <p><b>W</b> indicates the package has wettable flanks.</p> <p><b>yyy</b> is the package designator.</p> <p><b>R</b> is the package quantity. R is for reel (3000 pieces).</p> <p><b>Q1</b> indicates that this device is an automotive grade (AEC-Q100) device.</p>

(1) For the most current package and ordering information see the Package Option Addendum at the end of this document, or visit the device product folder on [www.ti.com](http://www.ti.com).

#### 11.2 接收文档更新通知

要接收文档更新通知，请导航至 [ti.com](http://ti.com) 上的器件产品文件夹。点击 [订阅更新](#) 进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

#### 11.3 支持资源

**TI E2E™ 支持论坛**是工程师的重要参考资料，可直接从专家获得快速、经过验证的解答和设计帮助。搜索现有解答或提出自己的问题可获得所需的快速设计帮助。

链接的内容由各个贡献者“按原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的《[使用条款](#)》。

#### 11.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.

所有商标均为其各自所有者的财产。

#### 11.5 静电放电警告



静电放电 (ESD) 会损坏这个集成电路。德州仪器 (TI) 建议通过适当的预防措施处理所有集成电路。如果不遵守正确的处理和安装程序，可能会损坏集成电路。

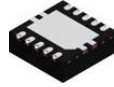
ESD 的损坏小至导致微小的性能降级，大至整个器件故障。精密的集成电路可能更容易受到损坏，这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

#### 11.6 术语表

**TI 术语表** 本术语表列出并解释了术语、首字母缩略词和定义。

### Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

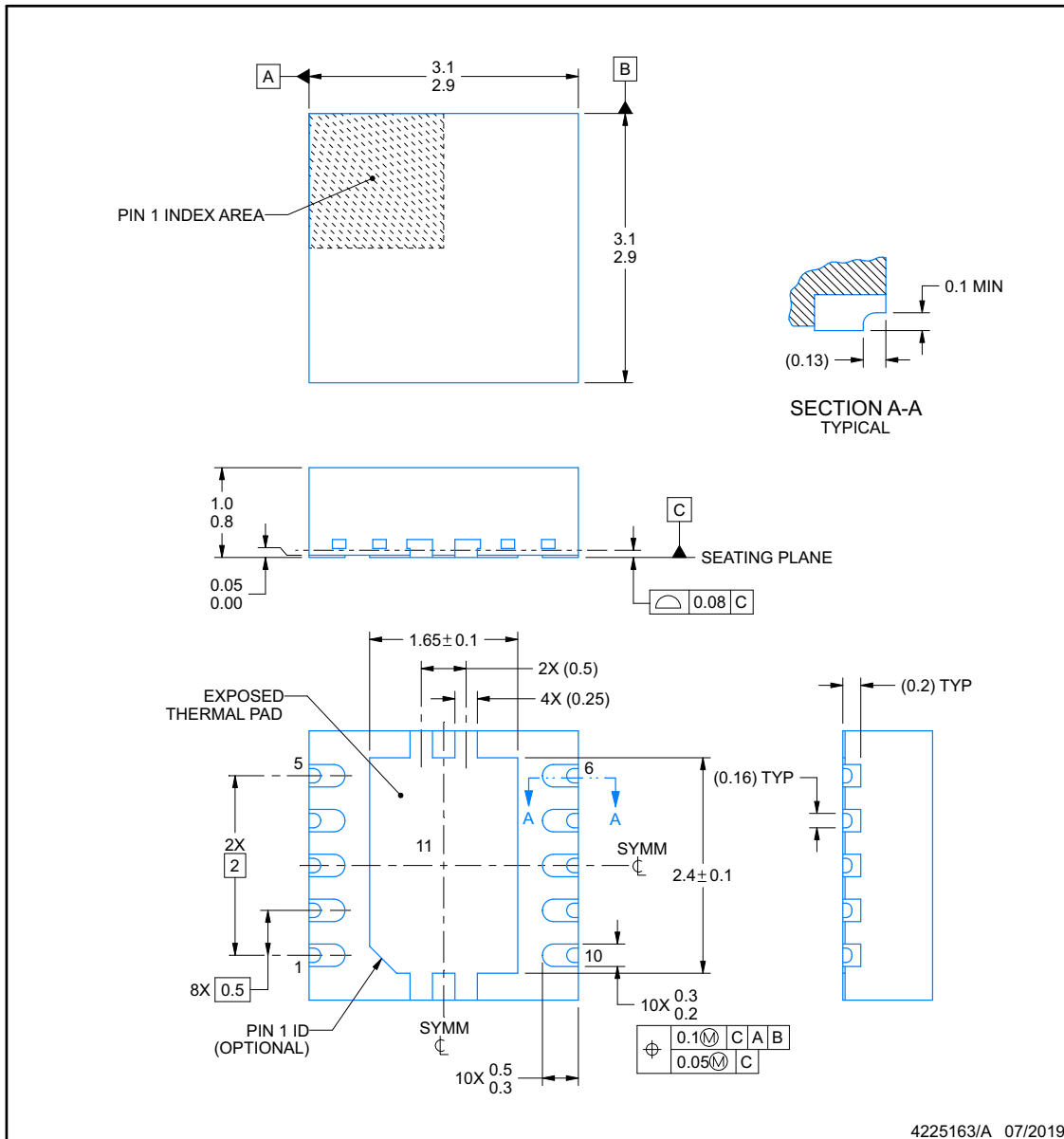


## PACKAGE OUTLINE

### DRC0010U

### VSON - 1 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



**NOTES:**

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.

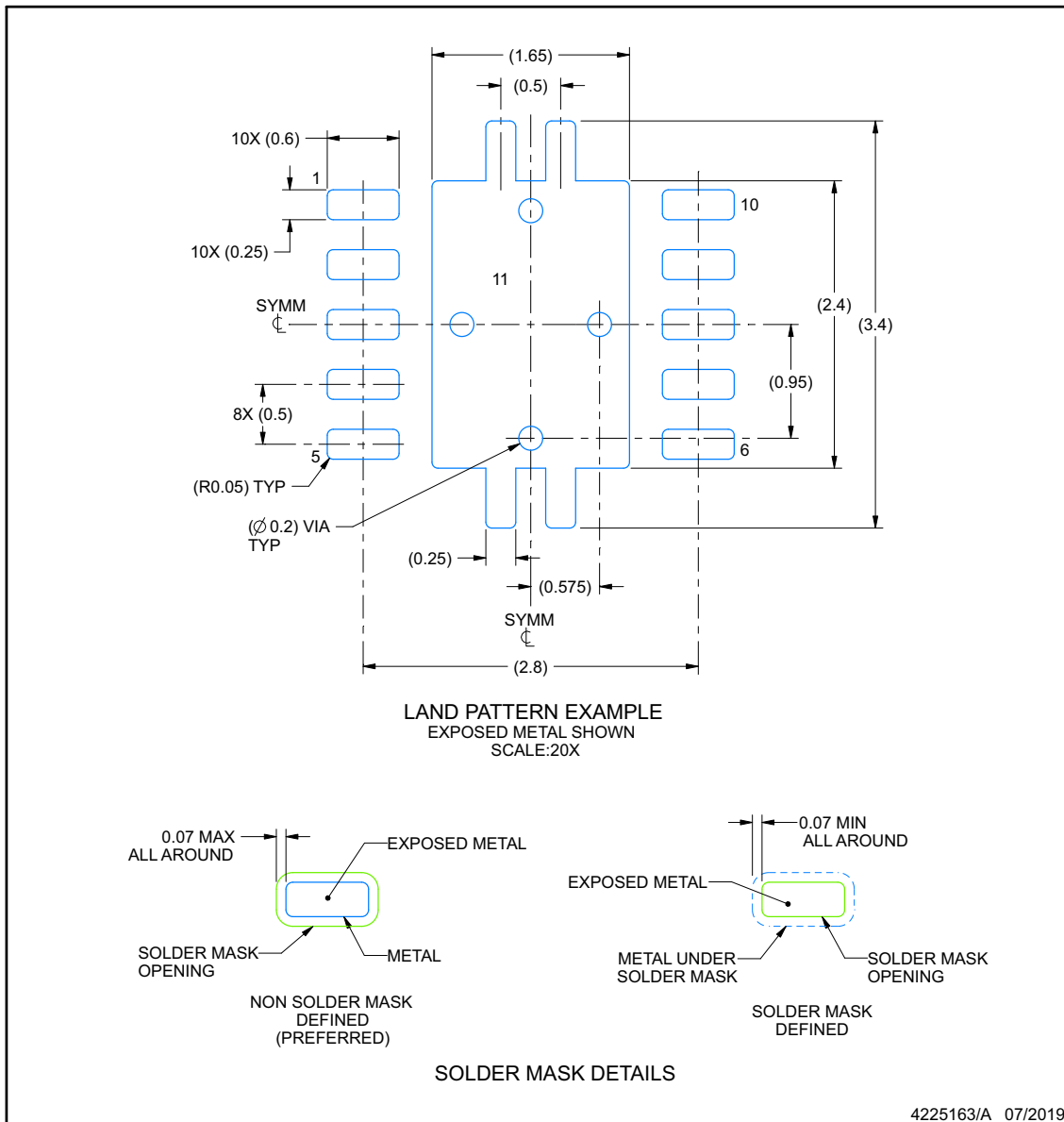


## EXAMPLE BOARD LAYOUT

**DRC0010U**

**VSON - 1 mm max height**

PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

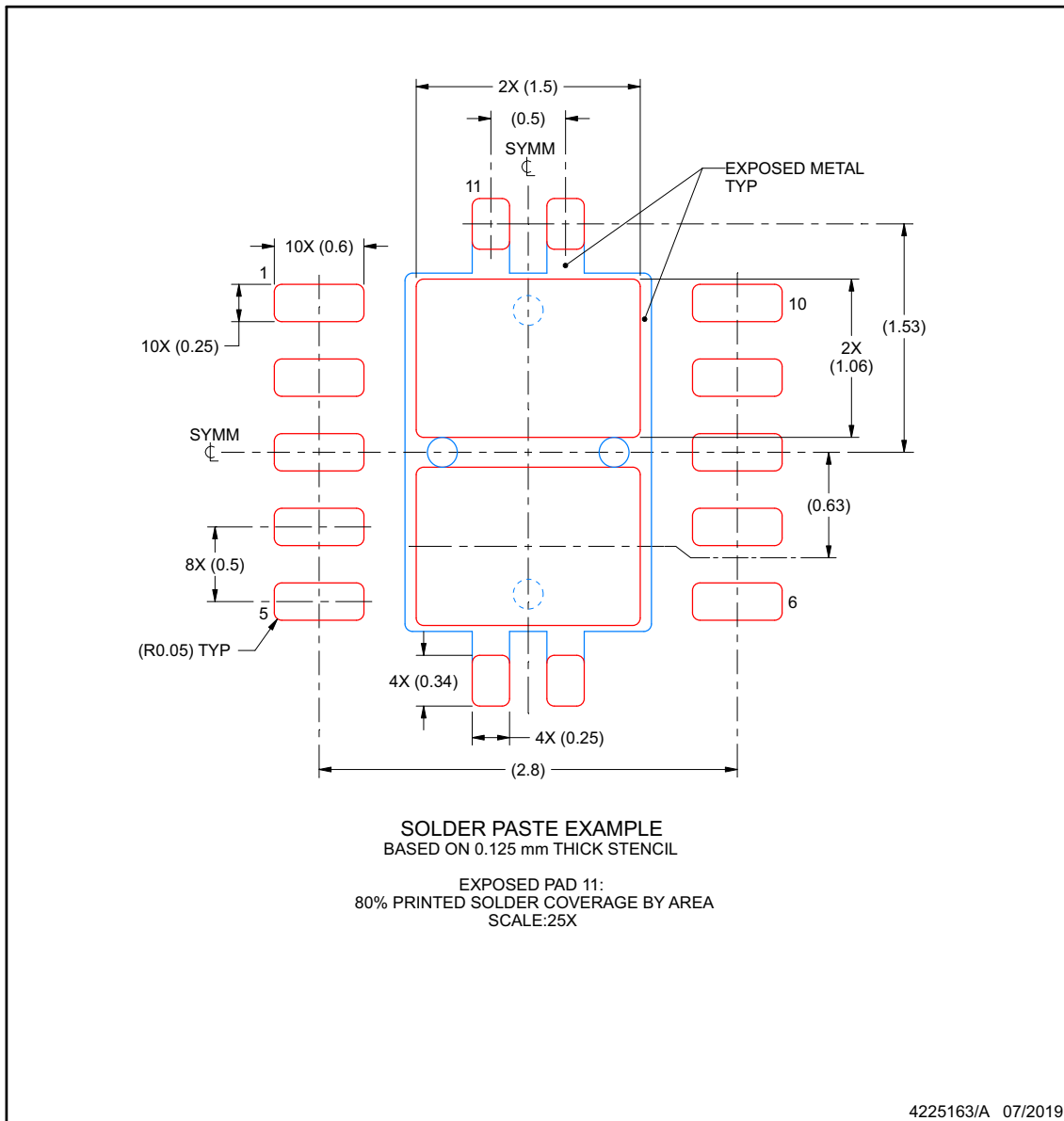
- This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/slua271](http://www.ti.com/lit/slua271)).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

**EXAMPLE STENCIL DESIGN**

**DRC0010U**

**VSON - 1 mm max height**

PLASTIC SMALL OUTLINE - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS7B8533QWDRCRQ1	ACTIVE	VSON	DRC	10	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 150	7B8533	<a href="#">Samples</a>
TPS7B8550QWDRCRQ1	ACTIVE	VSON	DRC	10	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 150	7B8550	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS7B8533QWDRCRQ1	VSON	DRC	10	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
TPS7B8550QWDRCRQ1	VSON	DRC	10	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS7B8533QWDRCRQ1	VSON	DRC	10	3000	367.0	367.0	35.0
TPS7B8550QWDRCRQ1	VSON	DRC	10	3000	367.0	367.0	35.0

## GENERIC PACKAGE VIEW

**DRC 10**

**VSON - 1 mm max height**

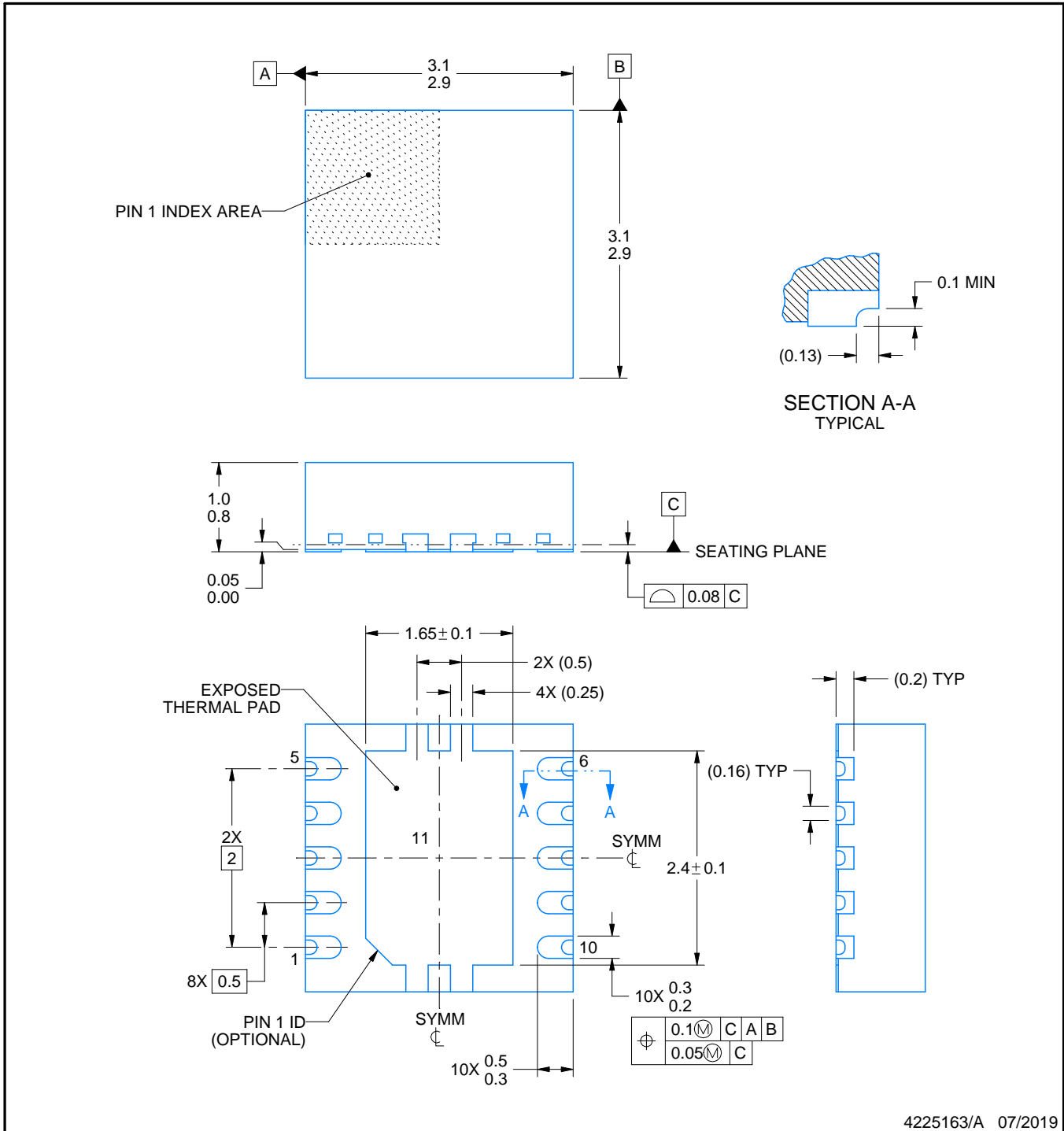
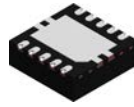
3 x 3, 0.5 mm pitch

PLASTIC SMALL OUTLINE - NO LEAD

This image is a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.



4226193/A



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NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.

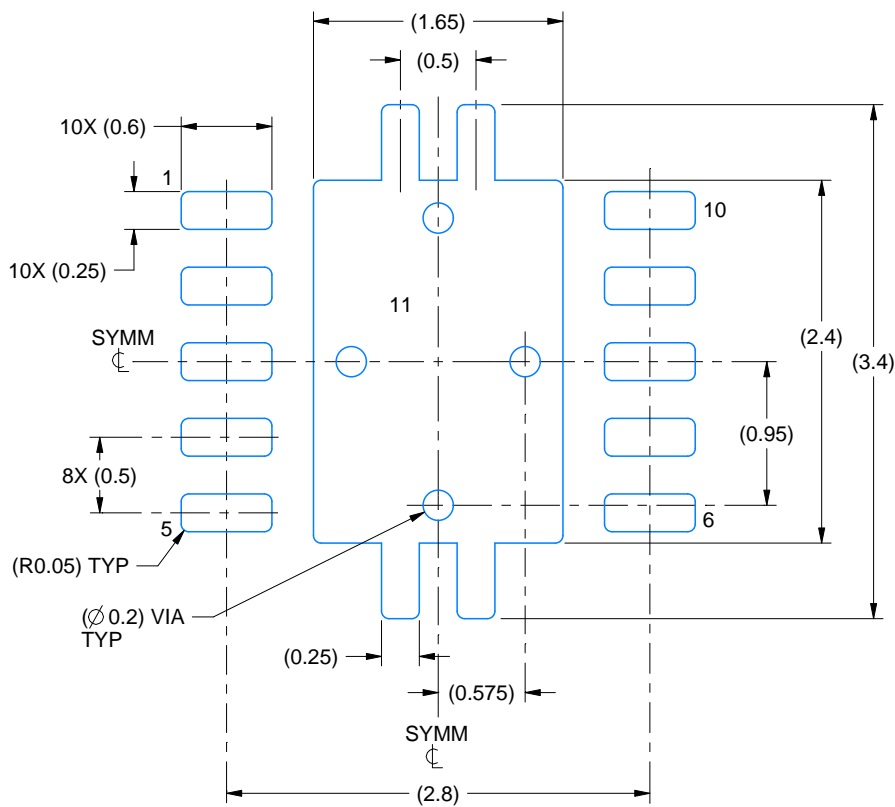


# EXAMPLE BOARD LAYOUT

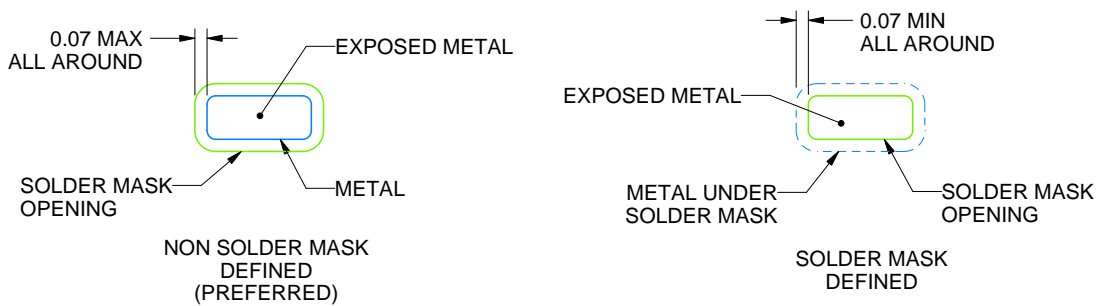
DRC0010U

VSON - 1 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:20X



SOLDER MASK DETAILS

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NOTES: (continued)

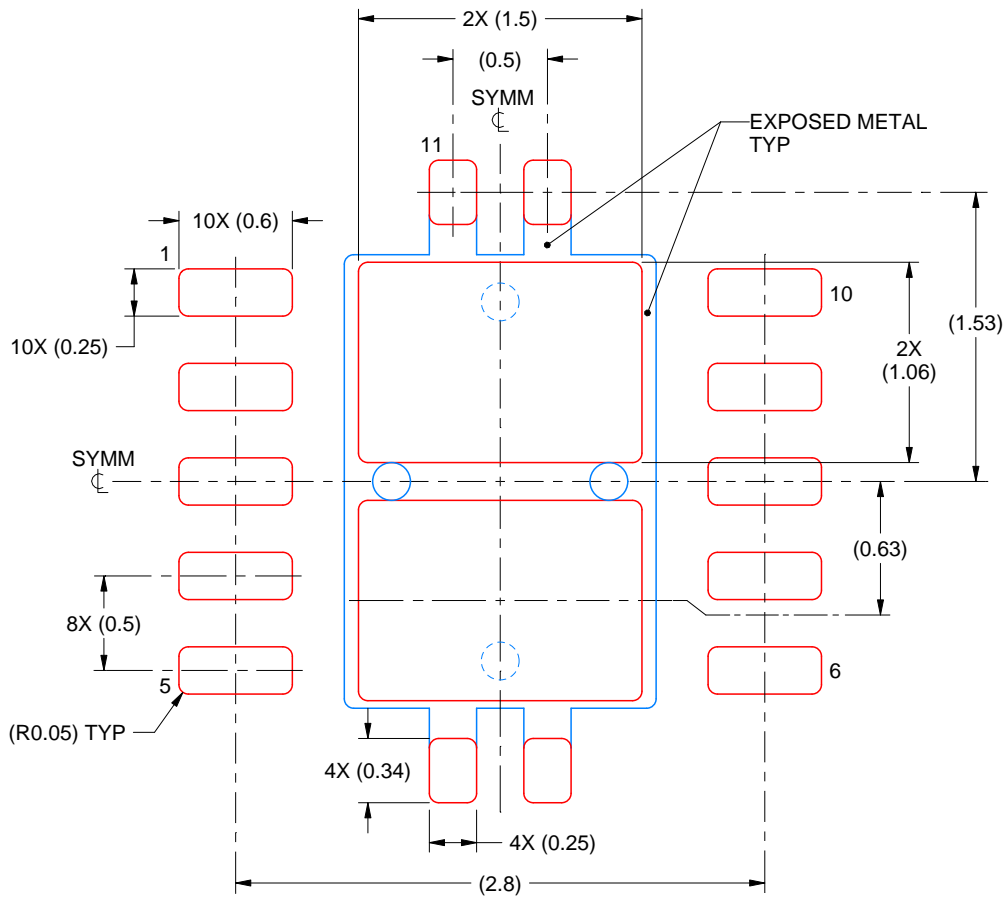
4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/sluea271](http://www.ti.com/lit/sluea271)).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

DRC0010U

VSON - 1 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 11:  
80% PRINTED SOLDER COVERAGE BY AREA  
SCALE:25X

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NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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